











**TRS3122E** 

ZHCSEZ3C -MAY 2016-REVISED MAY 2016

# TRS3122E 1.8V 低功耗双路 RS-232 收发器

# 1 特性

- 扩展 V<sub>CC</sub> 工作节点: 1.8V、3.3V 或 5.0V
  - 独特的三倍频器电荷泵架构,与 3.3V 和 5V 电源保持兼容的同时实现低至 1.8V 的 V<sub>CC</sub>
- 集成电平转换功能,连接低电压 MCU 时无需使用 外部电源或附加电平转换器
- RIN 输入和 DOUT 输出端的增强型 ESD 保护
  - ±15kV IEC 61000-4-2 气隙放电
  - ±8kV IEC 61000-4-2 接触放电
  - ±15kV 人体放电模式
- 数据速率指定为 1000Kbps
- 自动断电增强特性
- 关断电源电流低至 0.5µA
- 满足甚至超过 RS-232 接口的兼容性要求
- 对于 2.5V 单电源 应用,可以考虑选用 TRS3318E 作为优化解决方案

### 2 应用

- 远程射频单元 (RRU)
- 基带装置 (BBU)
- 电子销售点 (EPOS)
- 诊断和数据传输
- 电池供电类设备

# 3 说明

TRS3122E 是一款具有双驱动器、双接收器的 RS-232 接口器件,配有分离电源引脚以支持混合电压运行。使用 IEC 61000-4-2 气隙放电方法,IEC 61000-4-2 接触放电方法和人体放电模型分别保护全部 RS-232 输入和输出不受 ±15kV、±8kV 和 ±15kV 电压的影响。

电荷泵需要使用 5 个 0.1μF 小电容,以便器件能够在低至 1.8V 的电源供电下运行。TRS3122E 能够以高达 1000kbps 的数据速率运行,同时保持与 RS-232 相兼容的输出水平。

TRS3122E 具有一个独特的 V<sub>L</sub> 引脚,可实现在混合逻辑电压系统内运行。可通过 V<sub>L</sub> 引脚设定驱动器输入 (DIN) 和接收器输出 (ROUT) 逻辑电平。当连接低电压微控制器或通用异步收发器 (UART) 时,不再需要使用附加电压电平转换器。

该器件具备"自动省电增强"(Auto Powerdown Plus) 功能,如果连续 30 秒未收发任何数据,则自动进入低功耗模式。该功能使得该器件备受电池供电类应用或其他功率敏感型应用 的青睐。

### 器件信息(1)

	BB 11 1B 00	
器件型号	封装 (引脚)	封装尺寸 (标称值)
TRS3122ERGER	RGE (24)	4.00mm x 4.00mm

(1) 如需了解所有可用封装,请见数据表末尾的可订购产品附录。

### 功能图 Logic Supply-VL 1.8V, 3.3V, 5V POWER AUTO-**FORCEON** POWERDOWN PLUS **FORCEOFF** DOUT DIN-**RS232** 1000 kb/s RIN ROUT RX**RS232** INVALID STATUS

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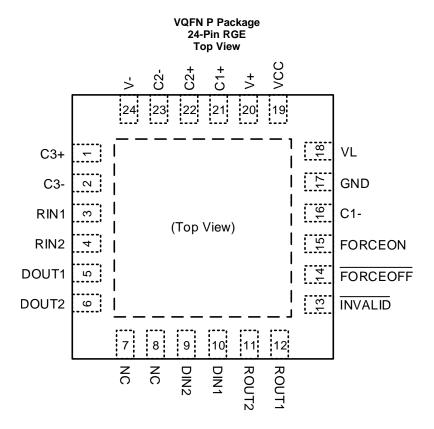
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# 4 修订历史记录

Changes from Revision B (May 2016) to Revision C	Page
已按优先级对特性部分的各要点进行了重新排序	1
Changes from Revision A (May 2016) to Revision B	Page
Updated ESD ratings values to reflect current device specifications	4
• 己添加 all Typical Characteristic graphs and schematics to the <i>Typical Characteristics</i> section	8
• 己添加 Application Curve image to Application Curves section	18
Changes from Original (June 2014) to Revision A	Page
Added Pin Functions table.	3



# 5 Pin Configuration and Functions



**Pin Functions** 

Pin		1/0	DESCRIPTION						
NAME	NO.	., 0	DECOMM MON						
C1+, C2+	21, 22	-	Positive terminals of voltage-doubler charge-pump capacitors (required)						
C3+	1	-	Positive terminal of voltage-tripler charge-pump capacitor (Not needed for VCC 3V to 5.5V)						
C1-, C2-	16, 23	-	Negative terminals of voltage-doubler charge-pump capacitors (required)						
C3-	2	-	Negative terminal of voltage-tripler charge-pump capacitor (Not needed for VCC 3V to 5.5V)						
V+	20	-	Positive charge pump storage capacitor (required)						
V-	24	-	Negative charge pump storage capacitor (required)						
GND	17	-	Ground						
$V_{CC}$	19	-	1.8-V or 3-V to 5-V supply voltage						
$V_L$	18	-	Logic-level supply. All CMOS inputs (DIN) and outputs (ROUT) are referenced to this supply.						
FORCEOFF	14	1	Auto Powerdown Control input (Refer to Truth Table)						
FORCEON	15	1	Auto Powerdown Control input (Refer to Truth Table)						
INVALID	13	0	Invalid Output Pin						
DIN1, DIN2	10,9	I	Driver inputs						
DOUT1, DOUT2	5, 6	0	RS-232 driver outputs						
RIN1, RIN2	3, 4	1	RS-232 receiver inputs						
ROUT1, ROUT2	12, 11	0	Receiver outputs; swing between 0 and V <sub>L</sub>						
NC	7, 8	I	Factory pins, can be unconnected or connected to GND						

# TEXAS INSTRUMENTS

# 6 Specifications

# 6.1 Absolute Maximum Ratings<sup>(1)</sup>

over operating free-air temperature range (unless otherwise noted)

			MIN	MAX	UNIT
V <sub>CC</sub>	Charge pump power su	pply	-0.3	6	٧
$V_L$	Logic power supply		-0.3	6	V
V+	Positive storage capacit	tor voltage	-0.3	7	V
V <sub>L</sub> I V+ I V- I V <sub>1</sub> I V <sub>0</sub> 0	Negative storage capac	itor voltage	0.3	-7	V
	V+ +  V-  <sup>(2)</sup>			13	V
		FORCEOFF, FORCEON	-0.3	6	.,
		DIN	-0.3	V <sub>L</sub> + 0.3	
VI	Input voltage	RIN (0Ω series resistance)		±20	V
		RIN (≥250Ω series resistance)		±25	İ
.,	Outract walts we	DOUT		±13.2	
VO	Output voltage	ROUT	-0.3	V <sub>L</sub> + 0.3	V
TJ	Junction temperature			150	ô
T <sub>stg</sub>	Storage temperature ra	nge	-65	150	ô

<sup>(1)</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

# 6.2 ESD Ratings

				VALUE	UNIT
		Human body model (HBM), per ANSI/ESDA/JEDEC JS-001,	All pins except RS-232 bus	±2000	
V <sub>(ESD)</sub> Electrostatic discharge	all pi	all pins <sup>(1)</sup>	RS-232 bus pins	±15000	
		Charged device model (CDM), per JEDEC specification JESD22-C101, all pins (2)	All pins	±500	V
		IEC 61000-4-2 Air-Gap Discharge	DC 222 hua nina	±15000	
		IEC 61000-4-2 Contact Discharge	RS-232 bus pins	±8000	

<sup>(1)</sup> JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

<sup>(2)</sup> V+ and V- can have maximum magnitudes of 7 V, but their absolute difference cannot exceed 13 V.

<sup>2)</sup> JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



# 6.3 Recommended Operating Conditions

				MIN	TYP	MAX	UNIT
			Tripler Mode	1.65	1.8	2	
V <sub>L</sub> IRIN IDOUT I	Charge pump power supply		Doubler Made	3	3.3	3.6	V
			Doubler Mode	4.5	5	5.5	
V <sub>L</sub>	Logic power supply			1.65		$V_{CC}$	V
RIN	RS-232 Receiver int	erface		-15		15	V
DOUT	RS-232 Transmitter	interface		-12		12	V
			V <sub>L</sub> = 5.0 V	0		1.7	
$V_{IL}$	GPIO Input logic threshold low	DIN, FORCEOFF, FORCEON	$V_{L} = 3.3V$	0		1.1	V
	unconoid low		V <sub>L</sub> = 1.8 V	0		0.6	
			V <sub>L</sub> = 5.0V	3.3		$V_L$	
$V_{IH}$	GPIO Input logic threshold high	DIN, FORCEOFF, FORCEON	$V_L = 3.3V$	2.2		$V_{L}$	V
	tillesiloid filgir		$V_L = 1.8V$	1.2		$V_{L}$	
V <sub>OZ</sub>	ROUT disabled		FORCEOFF = 0V	0		$V_{L}$	V
	Operating temperatu	ire		-40		85	°C

# 6.4 Thermal Characteristics

	THERMAL METRIC	TRS3122E	LINIT
	THERMAL METRIC	RGE	UNIT
$R_{\theta JA}$	Junction-to-ambient thermal resistance	34.2	
$R_{\theta JCtop}$	Junction-to-case (top) thermal resistance	27.2	
$R_{\theta JB}$	Junction-to-board thermal resistance	11.4	9 <b>0</b> AA
ΨЈТ	Junction-to-top characterization parameter	0.4	°C/W
ΨЈВ	Junction-to-board characterization parameter	11.4	
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	3.6	

# 6.5 Power and Status Electrical Characteristics

 $V_{CC} = V_L = (1.65 \text{ V to } 2.0 \text{ V}) \& (3.0 \text{V to } 5.5 \text{V}), T_A = -40 ^{\circ}\text{C} \text{ to } 85 ^{\circ}\text{C} \text{ (unless otherwise noted)}.$  Typical data is  $T_A = 25 ^{\circ}\text{C}, V_{CC} = V_L = 3.3 \text{V}$  unless otherwise noted.

	PARAMETER		TEST CONDITIONS	S	MIN	TYP	MAX	UNIT
		DIN1 = GND or $V_L$ ;		V <sub>CC</sub> = 1.65V to 2.0V		1.0	1.9	
Icc (Statio	c)	$\frac{DIN2 = GND}{FORCEOFF} = V_1$	No load	V <sub>CC</sub> = 3.0V to 3.6V		0.7	1.4	mA
		FORCEON = V <sub>L</sub>		V <sub>CC</sub> = 4.5V to 5.5V		0.8	1.9	
Icc (off)		FORCEOFF = GND				0.4	10	μΑ
V <sub>IT+</sub>	RIN postive voltage threshold for INVALID output change	DINIA DINIO			0.3		2.4	V
V <sub>IT-</sub>	RIN negative voltage threshold for INVALID output change	KINI = KINZ	N1 = RIN2				-0.3	V
V <sub>OH</sub>	INVALID high-level output voltage	I <sub>OH</sub> = -1 mA, FORCE	OH = -1 mA, FORCEON = GND, FORCEOFF = VL		V <sub>L</sub> -0.4	V <sub>L</sub> -0.08	$V_{L}$	٧
V <sub>OL</sub>	INVALID low-level output voltage	I <sub>OL</sub> = 1.6 mA, FORCE	EON = GND, FORC	EOFF = VL	0	0.06	0.4	٧



### 6.6 Driver Electrical Characteristics

 $V_{CC} = V_L = (1.65 \text{ V to } 2.0 \text{ V})$  & (3.0V to 5.5V),  $T_A = -40^{\circ}\text{C}$  to 85°C (unless otherwise noted). Typical data is  $T_A = 25^{\circ}\text{C}$ ,  $V_{CC} = V_L = 3.3 \text{V}$  unless otherwise noted.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	T Output voltage swing	All driver outputs loaded with 3 k $\Omega$ to ground C3 = 100 nF, V <sub>CC</sub> = 1.8 V	±4.25	±4.7		V
V <sub>OUT</sub>		All driver outputs loaded with 3 k $\Omega$ to ground C3 = 0 F, V <sub>CC</sub> = 3.3 V or 5 V	±5	±5.4		V
$r_{O}$	Output resistance	$(V_{CC} = V + = V - = 0)$ ; Driver output = $\pm 2 V$	300	10M		Ω
Ios	Output short-circuit current	V <sub>DOUT</sub> = 0			±60	mA
l <sub>OZ</sub>	Output leakage current	V <sub>DOUT</sub> = ±12 V, FORCEOFF = GND	0		±25	μΑ
	Driver input hysteresis			0.5	1	V
	Input leakage current	DIN = GND to $V_L$ ; FORCEOFF = GND to $V_L$ ; FORCEON = GND to $V_L$		0	±5	μΑ

# 6.7 Receiver Electrical Characteristics

 $V_{CC} = V_L = (1.65 \text{ V to } 2.0 \text{ V})$  & (3.0V to 5.5V),  $T_A = -40^{\circ}\text{C}$  to 85°C (unless otherwise noted). Typical data is  $T_A = 25^{\circ}\text{C}$ ,  $V_{CC} = V_L = 3.3 \text{V}$  unless otherwise noted.

	PARAMETER	TEST CO	ONDITIONS	MIN	TYP	MAX	UNIT
I <sub>off</sub>	Output leakage current	ROUT, receivers disable	ed		±0.01	±10	μΑ
$V_{OL}$	Output voltage low	I <sub>OUT</sub> = 2.0 mA			0.04	0.3	V
$V_{OH}$	Output voltage high	I <sub>OUT</sub> = -2.0mA		V <sub>L</sub> -0.3	V <sub>L</sub> -0.04		V
			V <sub>L</sub> = 5 V	0.8	1.5		V
V <sub>IT-</sub>	Input threshold low	T <sub>A</sub> =25°C	V <sub>L</sub> = 3.3 V	0.7	1.1		
			V <sub>L</sub> = 1.8 V	0.6	0.7		
			V <sub>L</sub> = 5 V		2.0	2.4	V
$V_{IT+}$	Input threshold high	T <sub>A</sub> =25°C	V <sub>L</sub> = 3.3 V		1.5	2.4	
			V <sub>L</sub> = 1.8 V		0.9	1.4	
			V <sub>L</sub> = 5 V		0.45		V
V <sub>hys</sub>	Input hysteresis	T <sub>A</sub> =25°C	$V_{L} = 3.3 \text{ V}$		0.35		
			V <sub>L</sub> = 1.8 V		0.26		
	Input resistance	T <sub>A</sub> =-40 to 85°C		3	5	7	kΩ

# 6.8 Driver Switching Characteristics

 $V_{CC} = V_L = (1.65 \text{ V to } 2.0 \text{ V})$  & (3.0V to 5.5V),  $T_A = -40^{\circ}\text{C}$  to 85°C (unless otherwise noted). Typical data is  $T_A = 25^{\circ}\text{C}$ ,  $V_{CC} = V_L = 3.3\text{V}$  unless otherwise noted.

PARAMETER			MIN	TYP	MAX	UNIT
Maximum data sata	$R_L = 3 \text{ k}\Omega$ , $C_L = 500 \text{ pF}$ (one	e driver)	1000			labara a
Maximum data rate	$R_L = 3 \text{ k}\Omega, C_L = 1000 \text{ pF (or}$	ne driver)	500			kbps
Time-to-exit powerdo	own  V <sub>DOUT</sub>   > 3.7 V	V <sub>DOUT</sub>   > 3.7 V		30	150	μs
t <sub>PHL</sub> - t <sub>PLH</sub>   Driver skew <sup>(1)</sup>	$R_L = 3 \text{ k}\Omega$		0	50	100	ns
		VCC = 1.8V, C <sub>L</sub> = 200 pF		33		
	$R_1 = 3 \text{ k}\Omega \text{ to } 7 \text{ k}\Omega$	VCC = 1.8V, C <sub>L</sub> = 1000 pF		25		
Transition region als	T _ 25°C	$VCC = 3.3 \text{ V}, C_L = 200 \text{ pF}$		38		
Transition-region sle	Measured from 3 V to -3 V or -3 V to 3 V	VCC = 3.3 V, C <sub>L</sub> = 1000 pF		28		V/µs
		VCC = 5 V, C <sub>L</sub> = 200 pF		41		
		VCC = 5 V, C <sub>L</sub> = 1000 pF		30		

<sup>(1)</sup> Driver skew is measured at the driver zero crosspoint.



# 6.9 Receiver Switching Characteristics

 $V_{CC} = V_L =$  (1.65 V to 2.0 V) & (3.0V to 5.5V),  $T_A = -40^{\circ}C$  to 85°C (unless otherwise noted). Typical data is  $T_A = 25^{\circ}C$ ,  $V_{CC} = V_L = 3.3V$  unless otherwise noted.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PHL</sub>	Receiver propagation delay, high to low	Receiver input to receiver output		0.15	0.4	
t <sub>PLH</sub>	Receiver propagation delay, low to high	C <sub>L</sub> = 150 pF		0.15	0.4	μs
t <sub>PHL</sub> – t <sub>PLH</sub>	Receiver skew			50	300	ns
t <sub>en</sub>	Receiver output enable time	From FORCEOFF to ROUT= V <sub>L</sub> /2		200	400	ns
t <sub>dis</sub>	Receiver output disable time	$CL = 150 \text{ pF}, RL = 3 \text{ k}\Omega$		200	400	ns

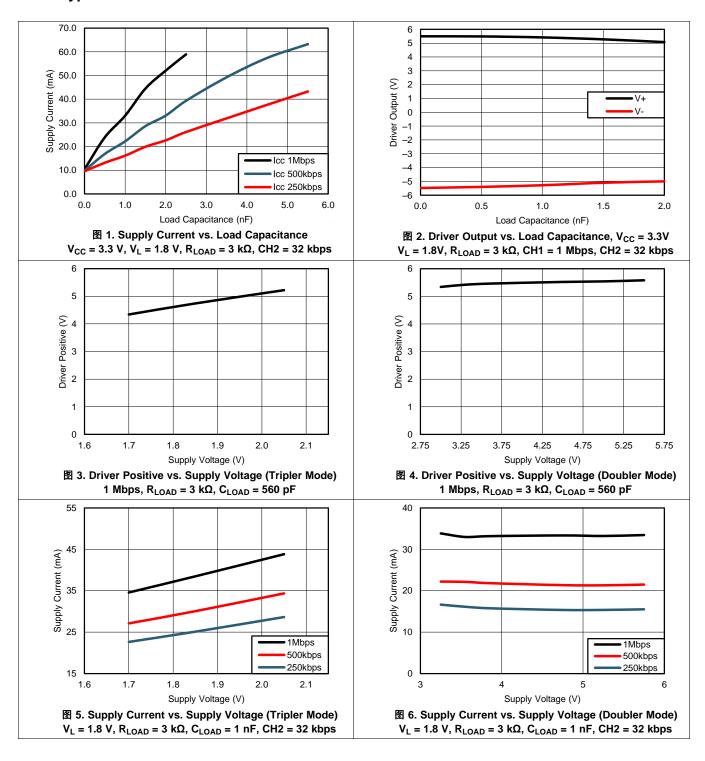
# 6.10 Power and Status Switching Characteristics

 $V_{CC} = V_L = (1.65 \text{ V to } 2.0 \text{ V})$  & (3.0V to 5.5V),  $T_A = -40^{\circ}\text{C}$  to 85°C (unless otherwise noted). Typical data is  $T_A = 25^{\circ}\text{C}$ ,  $V_{CC} = V_L = 3.3 \text{V}$  unless otherwise noted.

	PARAMETER	MIN	TYP	MAX	UNIT
t <sub>valid</sub>	Propagation delay time, low- to high-level output		1		μs
t <sub>invalid</sub>	Propagation delay time, high- to low-level output		30		μs
t <sub>dis</sub>	Receiver or driver edge to auto-powerdown plus	15	30	60	S

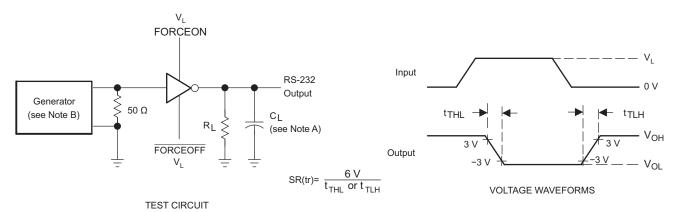


# 6.11 Typical Characteristics



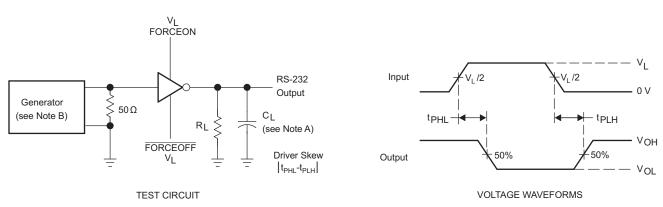


# 7 Parameter Measurement Information



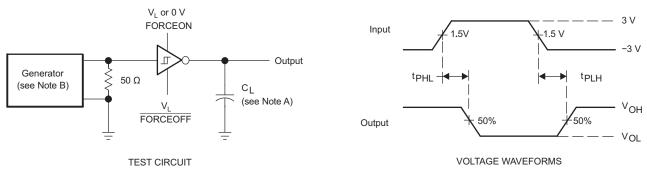
- A. C<sub>L</sub> includes probe and jig capacitance.
- B. The pulse generator has the following characteristics: PRR = 1000 kbit/s,  $Z_0$  = 50  $\Omega$ , 50% duty cycle,  $t_r \le$  10 ns,  $t_f \le$  10 ns.

# 图 7. Driver Slew Rate



- A. C<sub>L</sub> includes probe and jig capacitance.
- B. The pulse generator has the following characteristics: PRR = 1000 kbit/s,  $Z_0$  = 50  $\Omega$ , 50% duty cycle,  $t_r \le$  10 ns,  $t_f \le$  10 ns.

### 图 8. Driver Pulse Skew

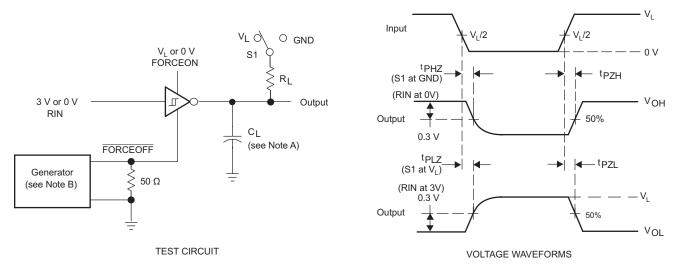


- A. C<sub>L</sub> includes probe and jig capacitance.
- B. The pulse generator has the following characteristics:  $Z_0 = 50 \Omega$ , 50% duty cycle,  $t_r \le 10$  ns,  $t_f \le 10$  ns.

# 图 9. Receiver Propagation Delay Times



# Parameter Measurement Information (接下页)

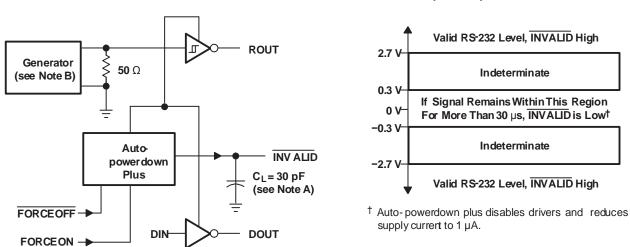


- A. C<sub>L</sub> includes probe and jig capacitance.
- B. The pulse generator has the following characteristics:  $Z_0 = 50 \Omega$ , 50% duty cycle,  $t_r \le 10$  ns,  $t_f \le 10$  ns.
- C. t<sub>PLZ</sub> and t<sub>PHZ</sub> are the same as t<sub>dis</sub>.
- D.  $t_{PZL}$  and  $t_{PZH}$  are the same as  $t_{en}$ .

图 10. Receiver Enable and Disable Times



# Parameter Measurement Information (接下页)



**TEST CIRCUIT** 

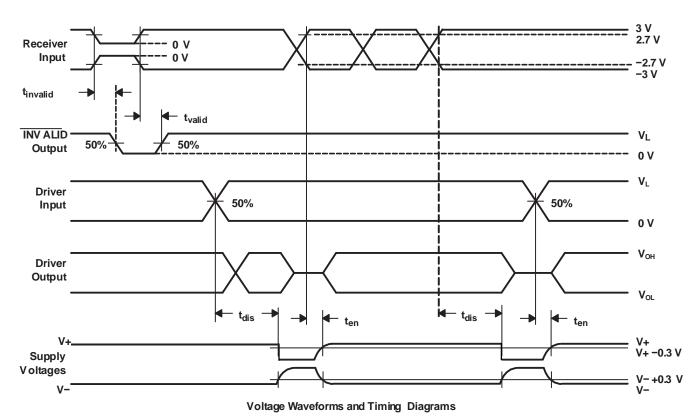


图 11. INVALID Propagation-Delay Times and Supply-Enabling Time

# 8 Detailed Description

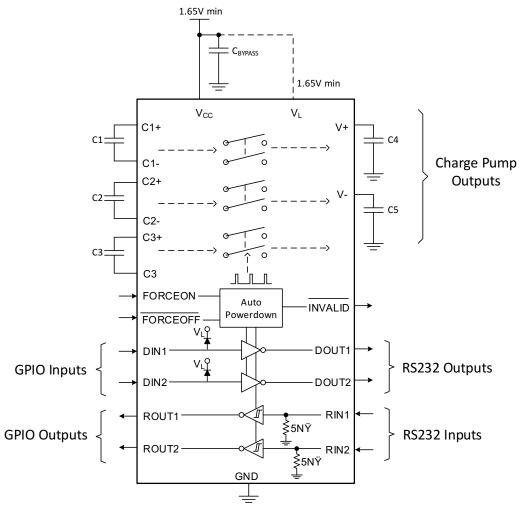
### 8.1 Overview

The TRS3122E is an upgrade to standard RS232 transceivers, offering compatibility with modern system needs like 1.8-V GPIO capability, enhanced ESD & ultra low stand-by current. The majority of RS-232 transceivers with 1.8-V GPIO compatibility require a logic supply pin for the I/O translation, in addition to a minimum 3.3 V V<sub>CC</sub> for all of the other active circuitry on the chip. Unlike these transceivers, TRS3122E can operate with both V<sub>L</sub> and V<sub>CC</sub> equal to 1.8 V. When V<sub>CC</sub>= 3.0 V to 5.5 V, the charge pump will sense V<sub>CC</sub> and switch to doubler mode. C1 & C2 are the necessary flying capacitors, C3 is not needed, and the charge pump outputs V+ & V- will regulate to ~+/-5.4 V. When V<sub>CC</sub>= 1.65 V to 2.0 V, the charge pump will sense V<sub>CC</sub> and switch to tripler mode. C1, C2 & C3 are all necessary, and the charge pump outputs V+ & V- will regulate to ~+/-2.65\*V<sub>CC</sub> from V<sub>CC</sub>= 1.65 V to 2.0 V

With many modern applications expanding into products that use RS232 as a backup communication protocol, it is important for the transceiver to have efficient standby operation. In order to accommodate this, Auto Powerdown Plus has been integrated to shut-off all active circuitry, allowing TRS3122E to achieve an I<sub>off</sub> of 1 uA.

In order to comply with common interface system needs and environments, the RS-232 receive and transmit I/O pins comply with IEC 61000-4-2 ratings.

### 8.2 Functional Block Diagram



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图 12. Schematic



### 8.3 Feature Description

### 8.3.1 Charge Pump

The internal power supply consists of a regulated auto-sensing charge pump that provides RS-232 compatible output voltages, over the 1.65 V to 2.0 V and 3.0 V to 5.5 V  $V_{CC}$  ranges. The charge pump operates in two modes to efficiently accommodate low voltage (1.8 V) and higher voltage (3.3 V & 5.0 V) supplies.

### 8.3.1.1 Doubler Mode

The charge pump requires two flying capacitors (C1, C2) and reservoir capacitors (C4, C5) to generate the V+ and V- supplies of approximately  $\pm 5.4$  V when V<sub>CC</sub> is greater than 3 V. When V<sub>CC</sub> is >2.9V, TRS3122E will sense the supply voltage level and switch the charge pump to a doubler. Hence, no need for a third flying capacitor. C3+ & C3- pins can be left open for proper operation. If a capacitor is placed between C3+ & C3-, the charge pump will ignore this capacitor and still behave as a doubler.

For capacitor choice recommendations, please refer to 表 1.

### 8.3.1.2 Tripler Mode

The charge pump requires three flying capacitors (C1, C2 & C3) and reservoir capacitors (C4, C5) to generate the V+ and V- supplies of approximately  $\pm 2.65$  \*  $V_{CC}$  when  $V_{CC}$  is greater than 1.65 V. When  $V_{CC}$  is <2.1 V, TRS3122E will sense the supply voltage level and switch the charge pump to a tripler.

For capacitor choice recommendations, please refer to 表 1.

### 8.3.2 Drivers

The drivers are inverting level transmitters that convert TTL or CMOS logic levels to RS-232 levels. For  $V_{CC}$ =3.0 V to 5.0 V, the RS-232 output voltage swing is typically ±5.4 V fully loaded and ±5 V minimum fully loaded. For Vcc = 1.8 V, the RS-232 output voltage swing is typically ±.4.7 V fully loaded and ±4.25 V minimum fully loaded.

The driver outputs are protected against indefinite short-circuits to ground without degradation in reliability. These drivers are compatible with RS-232 logic levels and all previous RS-232 versions. Unused driver inputs should be connected to GND or VCC.

### 8.3.3 Receivers

The receivers convert EIA/TIA-232 levels to TTL or CMOS logic output levels. Receivers have an inverting output that can be disabled by using the  $\overline{FORCEOFF}$  pin. Receivers remain active when the Auto Powerdown Plus circuitry autonomously enters a low power state. See Auto Powerdown Plus for more information on the Auto Powerdown mode. If the  $\overline{FORCEOFF}$  pin is manually set low, the receivers will be disabled and put into 3-state mode. In either of these powerdown modes, the device will typically consume about 0.5 uA. The truth table logic of the TRS3122E driver and receiver outputs can be found in Device Functional Modes. Since receiver input is usually from a transmission line where long cable lengths and system interference can degrade the signal, the inputs have a typical hysteresis margin of 300 mV. This ensures that the receiver is virtually immune to noisy transmission lines. Should an input be left unconnected, an internal  $5k\Omega$  pull-down resistor to ground will commit the output of the receiver to a HIGH state.

### 8.3.4 ESD Protection

ESD protection structures are incorporated on all pins to protect against electrostatic discharges encountered during handling and assembly. The bus pins (driver outputs and receiver inputs) have extra protection structures, which have been tested up to  $\pm 15$  kV.

ESD protection is tested in various ways. TI uses the following standards to qualify the ESD structures designed into TRS3122E:

- ±8 kV using IEC 61000-4-2 Contact Discharge (on RINx and DOUTx pins)
- ±15 kV using IEC 61000-4-2 Airgap Discharge (on RINx and DOUTx pins)
- ±15 kV using the Human Body Model (HBM) (on RINx and DOUTx pins)
- ±2 kV using the Human Body Model (HBM) (on all pins except RINx and DOUTx pins)
- ±0.5 kV using the Charged Device Model (CDM) (on all pins)



# Feature Description (接下页)

The IEC 61000-4-2 standard is more rigorous than HBM, resulting in lower voltage levels compared with HBM for the same level of ESD protection. Because IEC 61000-4-2 specifies a lower series resistance, the peak current is higher than HBM. The TRS3122E has passed both HBM and IEC 61000-4-2 testing.

### 8.3.5 Auto Powerdown Plus

Powerdown is engaged in two separate cases: automatically, when no activity has occurred for a period of time, and manually, using the FORCEOFF device pin.

### 8.3.5.1 Automatic Powerdown

Auto Powerdown Plus is enabled when FORCEON is set LOW and FORCEOFF is set HIGH. Using TRS3122E's integrated edge detection circuitry and timer, the device can sense when there is no activity on the driver or receiver inputs for 30 seconds. When this condition is sensed by the device, it automatically shuts the charge pump off, reducing supply current to 0.5 uA. When a valid transition is sensed on one of the driver or receiver inputs, the charge pump turns back on and TRS3122E exits powerdown. The typical time to exit powerdown is typically in 30 us, but can be as long as 150 us. As a result, the system saves power without requiring any software control. Device Functional Modes summarizes the operating modes in truth table form.

While in the low power mode with Automatic Powerdown enabled (FORCEOFF = HIGH and FORCEON = LOW), the receiver inputs are still enabled.

### 8.3.5.2 Manual Powerdown

The device can be manually powered down by externally setting FORCEOFF pin to low logic level. Both the drivers and receivers will be powered off. Device Functional Modes summarizes the operating modes in truth table form.

### 8.3.5.3 Forced On

If the FORCEOFF and FORCEON pins are both set HIGH, the device will power on with Auto Powerdown Plus disabled. Both the drivers and receiver will be active regardless of inactivity. Because powerdown is autonomous, FORCEON can be used ensure drivers are ready for new data transmission if the time since last transmission (or receive data) was more than 15 seconds. Device Functional Modes summarizes the operating modes in truth table form.



### 8.4 Device Functional Modes

# 8.4.1 Each Driver<sup>(1)</sup>

		INPUTS		OUTPUT			
DIN	FORCEON	FORCEOFF	TIME ELAPSED SINCE LAST RIN OR DIN TRANSITION	DOUT	DRIVER STATUS		
Х	Χ	L	X	Z	Powered off		
L	Н	Н	X	Н	Normal operation with		
Н	Н	Н	X	L	auto-powerdown plus disabled		
L	L	Н	<30 s	Н	Normal operation with		
Н	L	Н	<30 s	L	auto-powerdown plus enabled		
L	L	Н	>30 s	Z	Powered off by		
Н	L	Н	>30 s	Z	auto-powerdown plus feature		

<sup>(1)</sup> H = high level, L = low level, X = irrelevant, Z = high impedance (off), 30s is typical inactivity time

# 8.4.2 Each Receiver(1)

	INPL	ITS	OUTPUTS	
RIN	FORCEOFF	TIME ELAPSED SINCE LAST RIN OR DIN TRANSITION	ROUT	RECEIVER STATUS
X	L	X	Z	Powered off
L	Н	X	Н	Normal operation with
Н	Н	X	L	auto-powerdown plus
Open	Н	X	Н	disabled/enabled

<sup>(1)</sup> H = high level, L = low level, X = irrelevant, Z = high impedance (off), Open = input disconnected or connected driver off

# 8.4.3 INVALID Status Truth Table<sup>(1)</sup>

	OUTPUT			
RIN1, RIN2	FORCEON	FORCEOFF	TIME ELAPSED SINCE LAST RIN OR DIN TRANSITION	INVALID
Any L or H	X	X	X	Н
All Open	X	X	X	L

<sup>(1)</sup> H = high level, L = low level, X = irrelevant, Z = high impedance (off), Open = input disconnected or connected driver off

### 8.4.4 Capacitor Selection Table

# 表 1. Capacitor Selection

V <sub>CC</sub> = V <sub>L</sub>	C1 Capacitor Value	C2 Capacitor Value	C3 Capacitor Value	C4 Capacitor Value	C5 Capacitor Value		
1.65 V to 2 V <sup>(1)</sup>			100 nF				
3.0 V to 3.6 V <sup>(1)</sup>	100	) nF	100 nF or open	100 nF			
4.5 V to 5.5 V <sup>(1)</sup>	47 nF	330 nF	100 nF or open	330	) nF		
3 V to 5.5 V <sup>(2)</sup>	47 nF	470 nF	100 nF or open	470	nF		

<sup>(1)</sup> For optimized performance, we recommend using these configurations.

<sup>(2)</sup> For applications where the Vcc variation is larger, this configuration is acceptable.



# 9 Application and Implementation

注

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 9.1 Application Information

RS232 is used to communicate between two electrical units on separate PCBs across cables <40 ft. Common RS232 cables are RJ45, DB9 & DB25.

# 9.2 Typical 1.8-V Application

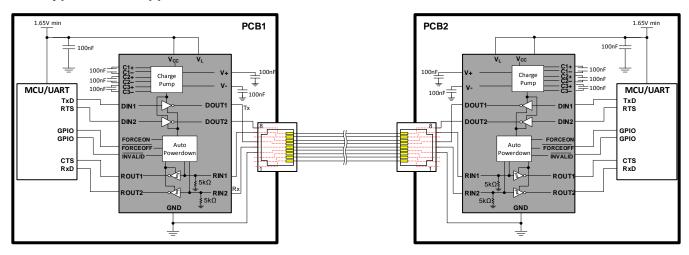


图 13. TRS3122E Typical Application

### 9.2.1 Design Requirements

For this design example, use the parameters listed in 表 2 as the input parameters.

# DESIGN PARAMETER EXAMPLE VALUE MCU GPIO Supple Voltage 1.8 V Transmission Voltage +/-4.7 V Data-rate 1 Mbps Number of Transmitters / Receivers 2 Charge Pump Capacitor Values 100nF (see 表 3)

表 2. Design Parameters

# 9.2.2 Detailed Design Procedure

When using TRS3122E, determine the following:

- All DIN, FORCEOFF, and FORCEON inputs must be connected to valid low or high logic levels.
- Select capacitor values based on V<sub>CC</sub> level for best performance. (see 表 3)



### 9.2.2.1 Data-Rate and Cable Length

RS-232 intended is for short range data transmission. The rise time for RS-232 driver edges is slow enough that the data cable appears as a capacitor instead of a transmission line impedance. The elapsed time for one bit of data far exceeds the transit time of any practical RS-232 cable length. The capacitance of the cable is the limiting factor. Therefore the capacitance per foot (or meter) of the cable is important if long data cables are used. Capacitance slows the rise and fall time of the signal. For low data rates, the delay is insignificant. However, high data rates will have reduced percentage of time that the output is at  $V_{OL}$  or  $V_{OH}$  and more time in the transitions. The timing of the UART (universal asynchronous receiver/transmitter) must sample the signal at the right time to coincide with  $V_{OL}$  and  $V_{OH}$  plateaus. At some point data reliability will be impacted. There are no hard limits for cable capacitance and data rate.

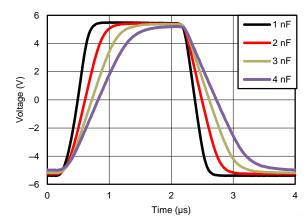


图 14. Typical Waveform with Capacitive Load  $V_{CC} = 3.3 \text{ V}$ ,  $R_{LOAD} = 3 \text{ k}\Omega$ , Date Rate = 500kbps

The maximum cable length depends on the cable used (pf/ft), data rate, timing of receiving UART, system tolerance to data errors.

### 9.2.2.2 Capacitor Selection

The capacitor type used for C1–C5 is not critical for proper operation; polarized or non-polarized capacitors can be used, though lower ESR capacitors are preferred. The charge pump requires 0.1  $\mu$ F capacitors for V<sub>CC</sub> = 1.8-V or V<sub>CC</sub> = 3.3-V operation. For other supply voltages, see  $\frac{1}{8}$  1 for required capacitor values. Do not use values smaller than those listed in  $\frac{1}{8}$  1. Increasing the capacitor values(e.g., by a factor of 2), except for C1, reduces ripple on the transmitter outputs and slightly reduces power consumption. C2, C3, C4 and C5 can be increased without changing C1's value. However, do not increase C1 without also increasing the values of C2, C3, C4, C5, C<sub>BYPASS1</sub>, and C<sub>BYPASS2</sub> to maintain the proper ratios (C1 to the other capacitors). When using the minimum required capacitor values, make sure the capacitor value does not degrade excessively with temperature. If in doubt, use capacitors with a larger nominal value. The capacitor's equivalent series resistance (ESR) usually increases at low temperatures.

For best charge pump efficiency locate the charge pump and bypass capacitors as close as possible to the IC. Surface mount capacitors are best for this purpose. Using capacitors with lower equivalent series resistance (ESR) and self-inductance, along with minimizing parasitic PCB trace inductance will optimize charge pump operation. Designers are also advised to consider that capacitor values may shift over time and operating temperature.

表 3. Capacitor Selection

$V_{CC} = V_{L}$	C1 Capacitor Value	C2 Capacitor Value	C3 Capacitor Value	C4 Capacitor Value	C5 Capacitor Value	
1.65 V to 2 V <sup>(1)</sup>			100 nF			
3.0 V to 3.6 V <sup>(1)</sup>	100	) nF	100 nF or open	100 nF		
4.5 V to 5.5 V <sup>(1)</sup>	47 nF	330 nF	100 nF or open	330	nF	
3 V to 5.5 V <sup>(2)</sup>	47 nF	470 nF	100 nF or open	470	nF	

<sup>(1)</sup> For optimized performance, we recommend using these configurations.

<sup>2)</sup> For applications where the Vcc variation is larger, this configuration is acceptable.

### 9.2.3 Application Curves

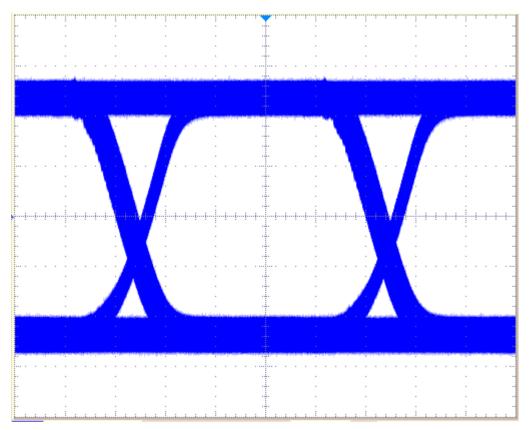


图 15. 1 Mbps Eye Diagram, 2 V/div, 200 ns/ div  $V_{CC}$  = 1.8 V,  $C_{LOAD}$  = 500 pF,  $R_{LOAD}$  = 3 k $\Omega$ 

# 10 Power Supply Recommendations

In most circumstances, a 0.1- $\mu$ F V<sub>CC</sub> bypass capacitor and a 1- $\mu$ F V<sub>L</sub> bypass capacitor are adequate. In applications that are sensitive to power-supply noise, use larger value V<sub>CC</sub> bypass capacitor. There is no maximum limit for bypass capacitor. Place bypass capacitors as close to the IC as possible.

It is not recommended to use this device when  $V_{CC}$  is powered and  $V_{L}$ = 0 V or floating for an extended period of time because operation is undefined.  $V_{CC}$  and  $V_{L}$  must be powered to guarantee charge pump operation.

Also, to achieve full functionality as described in Specifications, it is recommended to not use a higher voltage on  $V_L$  than  $V_{CC}$ . Full functionality can be achieved when  $V_{CC}$  is greater than or equal to  $V_L$ .



# 11 Layout

# 11.1 Layout Guidelines

Minimize the length of all capacitor traces to ensure the device can maintain quick rising and falling transitions. Vias are recommended to accommodate layouts for the capacitors.

# 11.2 Layout Example

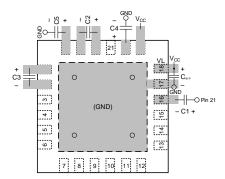


图 16. TRS3122E Typical Layout



# 12 器件和文档支持

### 12.1 器件支持

### 12.1.1 Third-Party Products Disclaimer

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### 12.4 静电放电警告



这些装置包含有限的内置 ESD 保护。 存储或装卸时,应将导线一起截短或将装置放置于导电泡棉中,以防止 MOS 门极遭受静电损伤。

### 12.5 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

### 13 机械、封装和可订购信息

以下页中包括机械、封装和可订购信息。这些信息是针对指定器件可提供的最新数据。这些数据会在无通知且不对本文档进行修订的情况下发生改变。欲获得该数据表的浏览器版本,请查阅左侧的导航栏。

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### PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
TRS3122ERGER	ACTIVE	VQFN	RGE	24	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TRS3122	Samples
TRS3122ERGET	ACTIVE	VQFN	RGE	24	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TRS3122	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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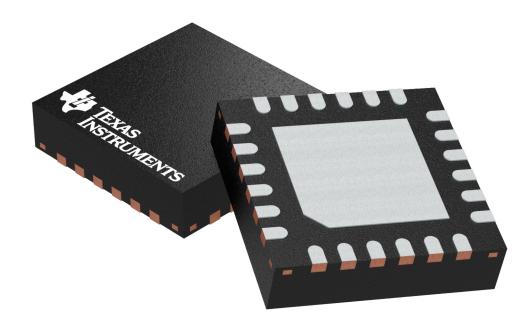
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# **PACKAGE OPTION ADDENDUM**

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PLASTIC QUAD FLATPACK - NO LEAD

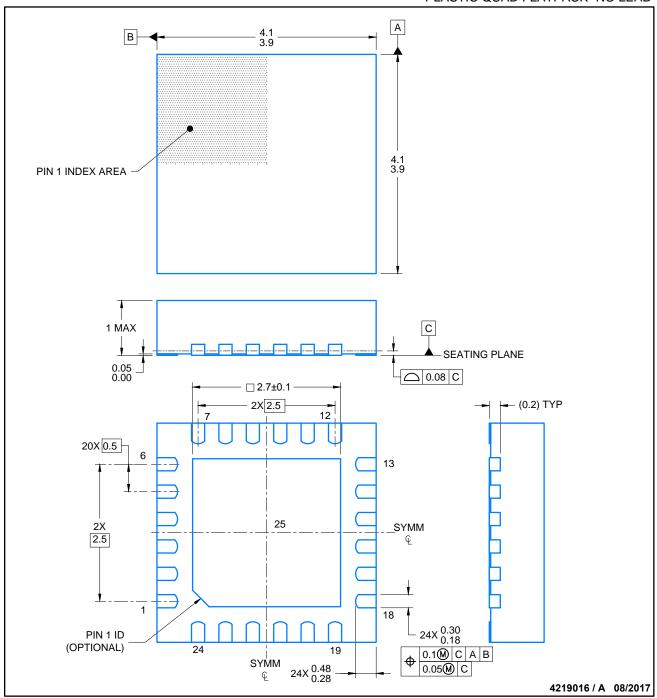


Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.

4204104/H



PLASTIC QUAD FLATPACK- NO LEAD

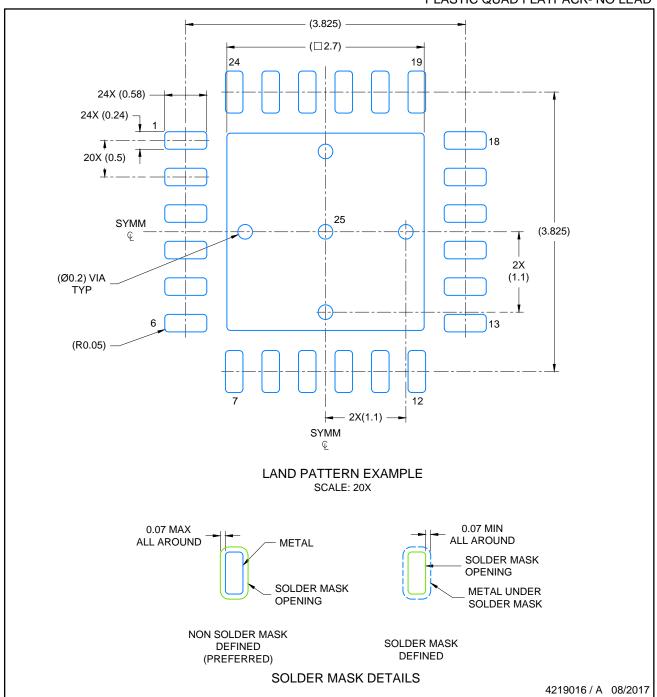


NOTES:

- All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.



PLASTIC QUAD FLATPACK- NO LEAD

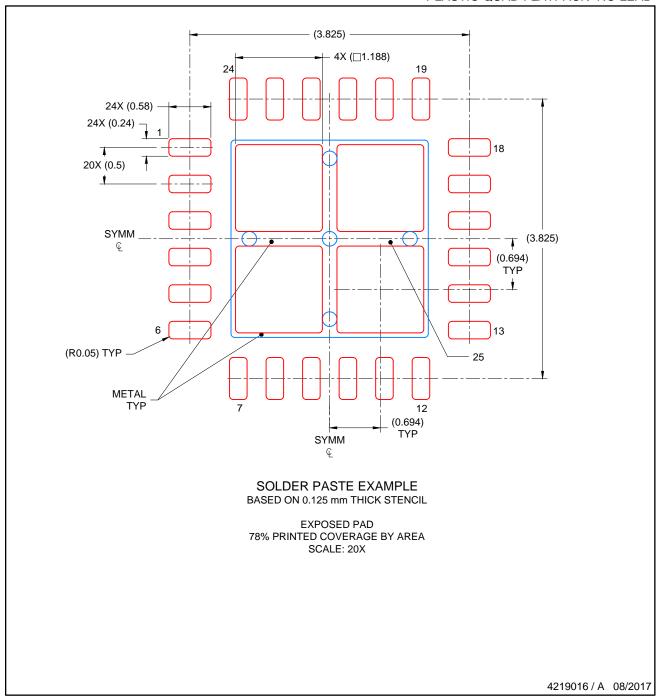


NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- Solder mask tolerances between and around signal pads can vary based on board fabrication site.



PLASTIC QUAD FLATPACK- NO LEAD



NOTES: (continued)

Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations..



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