

SBVS026D - JULY 2001 - REVISED SEPTEMBER 2005

DMOS 100mA Low-Dropout Regulator

FEATURES

- NEW DMOS TOPOLOGY:
 Ultra Low Dropout Voltage:
 60mV typ at 100mA
 Output capacitor NOT required for stability
- FAST TRANSIENT RESPONSE
- VERY LOW NOISE: 23μVrms
- HIGH ACCURACY: ±1.5% max
- HIGH EFFICIENCY:
 I_{GND} = 500μA at I_{OUT} = 100mA
 Not Enabled: I_{GND} = 10nA
- 2.5V, 2.8V, 2.85V, 3.0V, 3.3V, 5.0V, AND ADJUSTABLE OUTPUT VERSIONS
- OTHER OUTPUT VOLTAGES AVAILABLE UPON REQUEST
- FOLDBACK CURRENT LIMIT
- THERMAL PROTECTION
- SMALL SURFACE-MOUNT PACKAGES: SOT23-5 and SO-8

APPLICATIONS

- PORTABLE COMMUNICATION DEVICES
- BATTERY-POWERED EQUIPMENT
- PERSONAL DIGITAL ASSISTANTS
- MODEMS
- BAR-CODE SCANNERS
- BACKUP POWER SUPPLIES

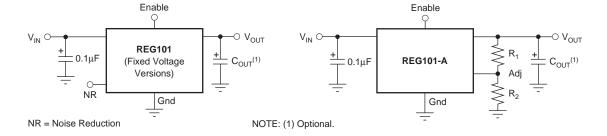
DESCRIPTION

The REG101 is a family of low-noise, low-dropout linear regulators with low ground pin current. Its new DMOS topology provides significant improvement over previous designs, including low dropout voltage (only 60mV typ at full load), and better transient performance. In addition, no output capacitor is required for stability, unlike conventional low-dropout regulators that are difficult to compensate and require expensive low ESR capacitors greater than 1µF.

Typical ground pin current is only $500\mu A$ (at $I_{OUT}=100mA$) and drops to 10nA when not in enabled mode. Unlike regulators with PNP pass devices, quiescent current remains relatively constant over load variation and under dropout conditions.

The REG101 has very low output noise (typically $23\mu Vrms$ for $V_{OUT} = 3.3V$ with $C_{NR} = 0.01\mu F$), making it ideal for use in portable communications equipment. Accuracy is maintained over temperature, line, and load variations. Key parameters are tested over the specified temperature range ($-40^{\circ}C$ to $+85^{\circ}C$).

The REG101 is well protected—internal circuitry provides a current limit that protects the load from damage. Thermal protection circuitry keeps the chip from being damaged by excessive temperature. The REG101 is available in the SOT23-5 and the SO-8 packages.





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ABSOLUTE MAXIMUM RATINGS(1)

0.3V to 12V
0.3V to 6.0V
0.3V to 6.0V
Indefinite
55°C to +125°C
65°C to +150°C
-8)+240°C

NOTE: (1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability.

ELECTROSTATIC DISCHARGE SENSITIVITY

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

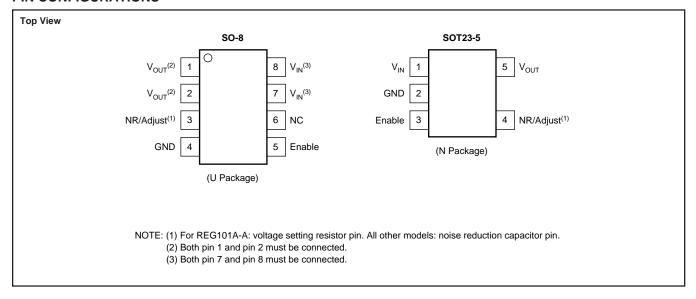
ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

PACKAGE/ORDERING INFORMATION(1)

PRODUCT	V _{OUT} ⁽²⁾				
REG101xx-yyyy/zzz	XX is package designator.				
	YYYY is typical output voltage (5 = 5.0V, 2.85 = 2.85V, A = Adjustable).				
	ZZZ is package quantity.				

- (1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI website at www.ti.com.
- (2) Output voltages from 2.5V to 5.1V in 50mV increments are available; minimum order quantities apply. Contact factory for details and availability.

PIN CONFIGURATIONS





ELECTRICAL CHARACTERISTICS

Boldface limits apply over the specified temperature range, $T_J = -40^{\circ}C$ to +85°C.

At T_J = +25°C, V_{IN} = V_{OUT} + 1V (V_{OUT} = 2.5V for REG101-A), V_{ENABLE} = 1.8V, I_{OUT} = 2mA, C_{NR} = 0.01 μ F, and C_{OUT} = 0.1 μ F(1), unless otherwise noted.

				REG101NA REG101UA		
PARAMETER		CONDITION	MIN	TYP	MAX	UNITS
OUTPUT VOLTAGE						
Output Voltage	V_{OUT}					
REG101-2.5				2.5		V
REG101-2.8				2.8		V
REG101-2.85				2.85		V
REG101-3.0				3.0		V
REG101-3.3				3.3		V
REG101-5				5		V
REG101-A			2.5		5.5	V
Reference Voltage	V_{REF}			1.267		V
Adjust Pin Current	I _{ADJ}			0.2	1	μА
Accuracy	·ADJ			±0.5	±1.5	%
Over Temperature				±0.0	±2.2	%
vs Temperature	dV _{OUT} /dT			50		ppm/°C
Includes Line and Load	a v out a i	$I_{OUT} = 2mA$ to 100mA, $V_{IN} = (V_{OUT} + 0.4V)$ to 10V		±0.8	±2.0	%
Over Temperature		$V_{IN} = (V_{OUT} + 0.6V)$ to 10V		±0.0	±2.7	%
•				_		
DC DROPOUT VOLTAGE(2)	V_{DROP}	I _{OUT} = 2mA		4	10	mV
For all models		I _{OUT} = 100mA		60	100	mV
Over Temperature		I _{OUT} = 100mA			130	mV
VOLTAGE NOISE	V_n	f = 10Hz to 100kHz				
Without C _{NR}		$C_{NR} = 0, C_{OUT} = 0$	2:	3μVrms/V • V _{OL}	ı T	μVrms
With C _{NR} (all fixed voltage models)	C _{NR} = 0.01μF, C _{OUT} = 10μF	7	μVrms/V • V _{OU}	г	μVrms
OUTPUT CURRENT	-					
Current Limit ⁽³⁾	I _{CL}		130	170	220	mA
Over Temperature	'CL		110	170	240	mA
Short-Circuit Current			110	60	240	mA
	I _{sc}			00		IIIA
RIPPLE REJECTION						
f = 120Hz		I _{OUT} = 100mA		65		dB
ENABLE CONTROL						
V _{ENABLE} High (output enabled)	V_{ENABLE}		1.8		V _{IN}	V
V _{ENABLE} Low (output disabled)			-0.2		0.5	V
I _{ENABLE} High (output enabled)	I _{ENABLE}	$V_{ENABLE} = 1.8V \text{ to } V_{IN}, V_{IN} = 1.8V \text{ to } 6.5^{(4)}$		1	100	nA
I _{ENABLE} Low (output disabled)		$V_{ENABLE} = 0V \text{ to } 0.5V$		2	100	nA
Output Disable Time		$C_{OUT} = 1.0 \mu F$, $R_{LOAD} = 33 \Omega$		200		μs
Output Enable Time		$C_{OUT} = 1.0 \mu F$, $R_{LOAD} = 33 \Omega$		1.5		ms
THERMAL SHUTDOWN						
Junction Temperature						
Shutdown				160		°C
Reset from Shutdown				140		°C
				140		
GROUND PIN CURRENT				466	500	
Ground Pin Current	I_{GND}	$I_{OUT} = 2mA$		400	500	μΑ
		I _{OUT} = 100mA		500	650	μΑ
Enable Pin Low		V _{ENABLE} ≤ 0.5V		0.01	0.2	μΑ
INPUT VOLTAGE	V_{IN}					
Operating Input Voltage Range ⁽⁵⁾			1.8		10	V
Specified Input Voltage Range		V _{IN} > 1.8V	V _{OUT} + 0.4		10	V
Over Temperature		V _{IN} > 1.8V	V _{OUT} + 0.6		10	V
TEMPERATURE RANGE						
Specified Range	T_J		-40		+85	°C
Operating Range	-		-40 -55			°C
	T _J				+125	°C
Storage Range	T_A		-65		+150	"
Thermal Resistance						
SOT23-5 Surface Mount	$ heta_{\sf JA}$	Junction-to-Ambient		200		°C/W

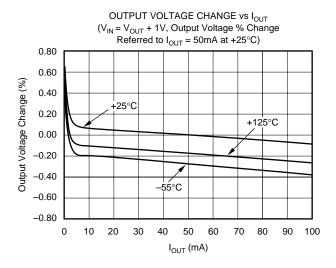
 ⁽³⁾ Current limit is the output current that produces a 10% change in output voltage from V_{IN} = V_{OUT} + 1V and I_{OUT} = 2mA.
 (4) For V_{ENABLE} > 6.5V, see typical characteristic "I_{ENABLE} vs V_{ENABLE}".
 (5) The REG101 no longer regulates when V_{IN} < V_{OUT} + V_{DROP (MAX)}. In drop-out, the impedance from V_{IN} to V_{OUT} is typically less than 1Ω at T_J = +25°C.

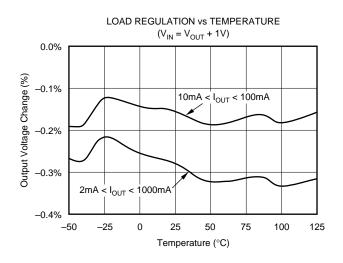


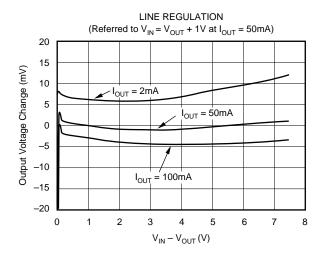
NOTES: (1) The REG101 does not require a minimum output capacitor for stability. However, transient response can be improved with proper capacitor selection.

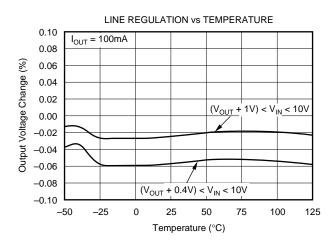
(2) Dropout voltage is defined as the input voltage minus the output voltage that produces a 2% change in the output voltage from the value at V_{IN}= V_{OUT}+1V at fixed load.

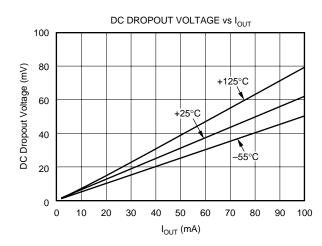
TYPICAL CHARACTERISTICS

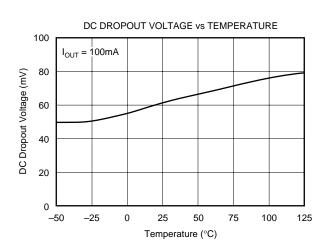






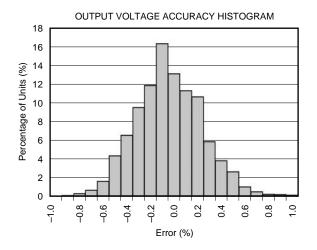


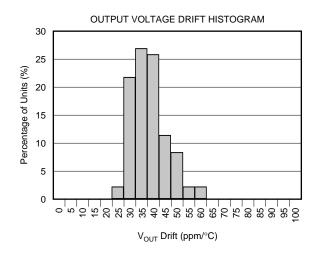


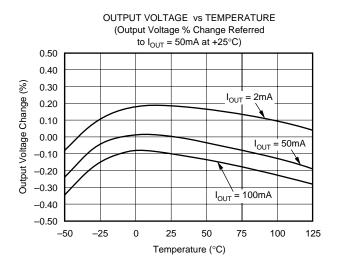


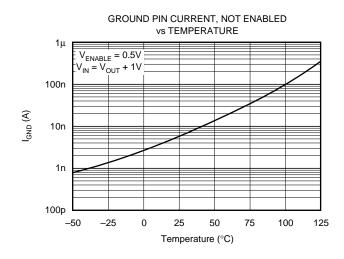


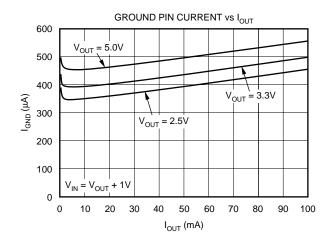
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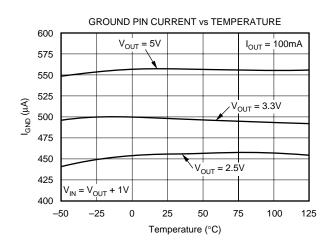






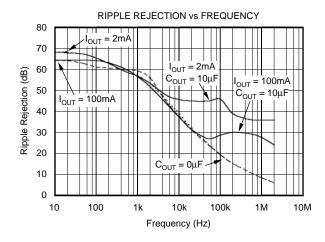


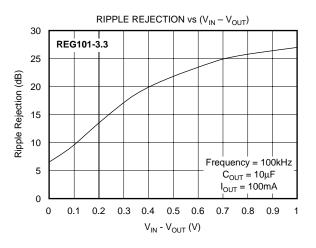


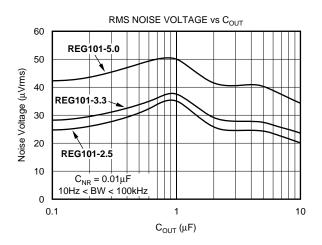


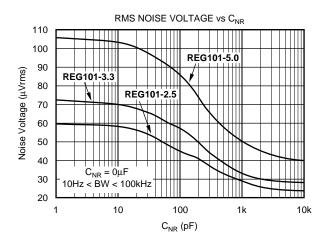


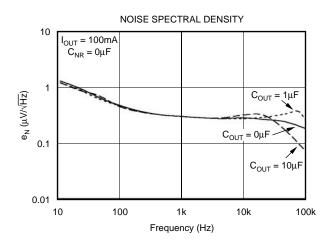
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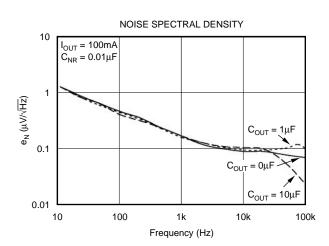






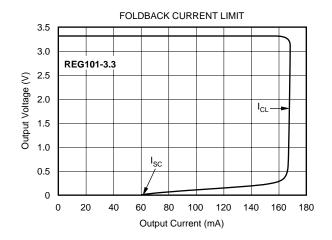


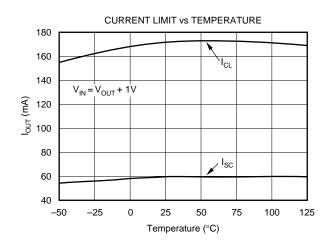


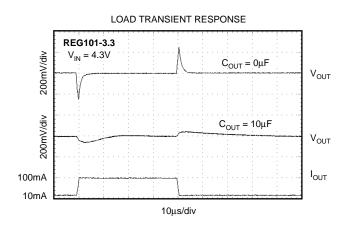


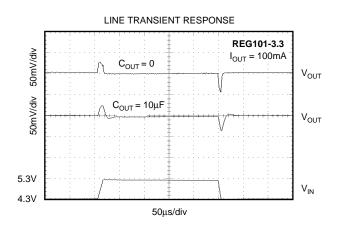


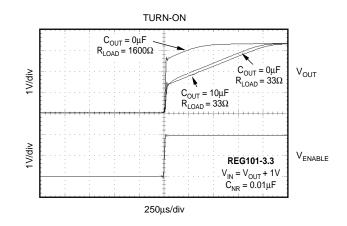
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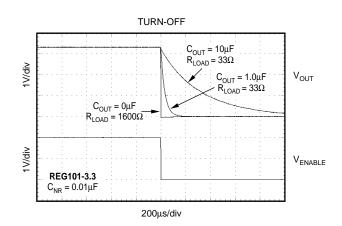




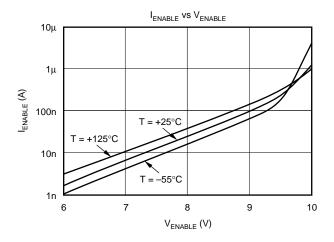


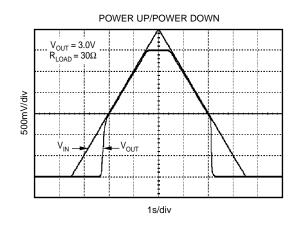


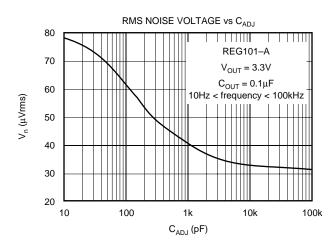


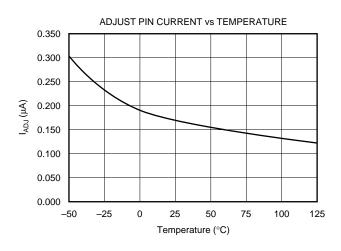


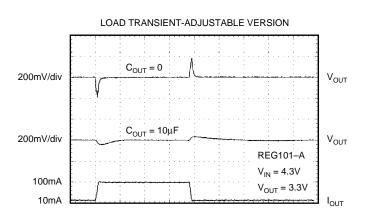
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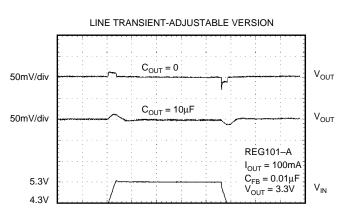














BASIC OPERATION

The REG101 series of LDO (Low Drop-Out) linear regulators offers a wide selection of fixed output voltage versions and an adjustable output version. The REG101 belongs to a family of new generation LDO regulators that utilize a DMOS pass transistor to achieve ultra-low dropout performance and freedom from output capacitor constraints. Ground pin current remains under 650µA over all line, load, and temperature conditions. All versions have thermal and overcurrent protection, including foldback current limit.

The REG101 does not require an output capacitor for regulator stability and is stable over most output currents and with almost any value and type of output capacitor up to $10\mu F$ or more. For applications where the regulator output current drops below several milliamps, stability can be enhanced by: adding a $1k\Omega$ to $2k\Omega$ load resistor; using capacitance values less than $10\mu F$; or keeping the effective series resistance greater than 0.05Ω including the capacitor's ESR and parasitic resistance in printed circuit board traces, solder joints, and sockets.

Although an input capacitor is not required, it is good analog design practice to connect a 0.1µF low ESR capacitor across

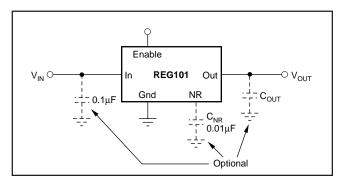


FIGURE 1. Fixed Voltage Nominal Circuit for REG101.

the input supply voltage. This is recommended to improve ripple rejection by reducing input voltage ripple.

Figure 1 shows the basic circuit connections for the fixed voltage models. Figure 2 gives the connections for the adjustable output version (REG101A) and example resistor values for some commonly used output voltages. Values for other voltages can be calculated from the equation shown in Figure 2.

INTERNAL CURRENT LIMIT

The REG101 internal current limit has a typical value of 170mA. A foldback feature limits the short-circuit current to a typical short-circuit value of 60mA. This helps to protect the regulator from damage under all load conditions. A characteristic of V_{OUT} versus I_{OUT} is given in Figure 3 and in the Typical Characteristics section.

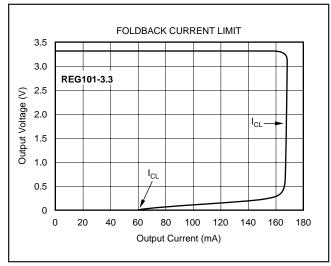
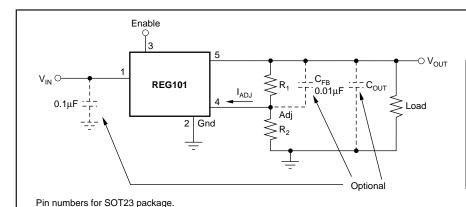


FIGURE 3. Foldback Current Limit of the REG101-3.3 at 25°C.



EXAMPLE RESISTOR VALUES

V _{OUT} (V)	R ₁ (W) ⁽¹⁾	R ₂ (Ω) ⁽¹⁾
2.5	11.3k 1.13k	11.5k 1.15k
3.0	15.8k 1.58k	11.5k 1.15k
3.3	18.7k 1.87k	11.5k 1.15k
5.0	34.0k 3.40k	11.5k 1.15k

NOTE: (1) Resistors are standard 1% values.

 $V_{OUT} = (1 + R_1/R_2) \cdot 1.267V$

To reduce current through divider, increase resistor values (see table at right).

As the impedance of the resistor divider increases,

 I_{ADJ} (~200nA) may introduce an error.

C_{FR} improves noise and transient response.

FIGURE 2. Adjustable Voltage Circuit for REG101A.



ENABLE

The Enable pin is active HIGH and compatible with standard TTL-CMOS levels. Inputs below 0.5V (max) turn the regulator off and all circuitry is disabled. Under this condition, ground pin current drops to approximately 10nA. When a pull-up resistor is used, and operation down to $V_{\rm IN}=1.8V$ is required, use values $<50k\Omega.$

OUTPUT NOISE

A precision band-gap reference is used for the internal reference voltage, V_{REF} . This reference is the dominant noise source within the REG101 and it generates approximately 29 μ Vrms in the 10Hz to 100kHz bandwidth at the reference output. The regulator control loop gains up the reference noise, so that the noise voltage of the regulator is approximately given by:

$$V_N = 29 \mu V \text{rms} \frac{R_1 + R_2}{R2} = 29 \mu V \text{rms} \cdot \frac{V_{OUT}}{V_{REF}}$$

Since the value of V_{REF} is 1.267V, this relationship reduces to:

$$V_{\rm N} = 23 \frac{\mu \rm Vrms}{\rm V} \cdot V_{\rm OUT}$$

Connecting a capacitor, C_{NR} , from the Noise Reduction (NR) pin to ground, as shown in Figure 4, forms a low-pass filter for the voltage reference. For $C_{NR}=10$ nF, the total noise in the 10Hz to 100kHz bandwidth is reduced by approximately a factor of 2.8 for $V_{O}=3.3$ V. This noise reduction effect is shown in Figure 5 and as "RMS Noise Voltage vs CNR" in the Typical Characteristics section.

Noise can be further reduced by carefully choosing an output capacitor, C_{OUT} . Best overall noise performance is

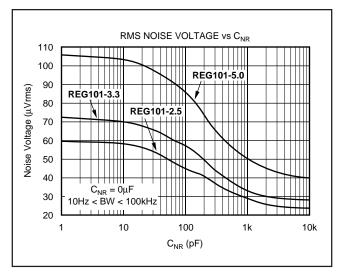


FIGURE 5. Output Noise versus Noise Reduction Capacitor.

achieved with very low (< $0.22\mu F$) or very high (> $2.2\mu F$) values of C_{OUT} . See "RMS Noise Voltage vs C_{OUT} " in the Typical Characteristics section.

The REG101 utilizes an internal charge pump to develop an internal supply voltage sufficient to drive the gate of the DMOS pass element above $V_{\rm IN}$. The charge-pump switching noise (nominal switching frequency = 2MHz) is not measurable at the output of the regulator over most values of $C_{\rm OUT}$ and $I_{\rm OUT}$.

The REG101 adjustable version does not have the noise-reduction pin available, however, the adjust pin is the summing junction of the error amplifier. A capacitor, C_{FB} , connected from the output to the adjust pin will reduce both the output noise and the peak error from a load transient. See the typical characteristics for output noise performance.

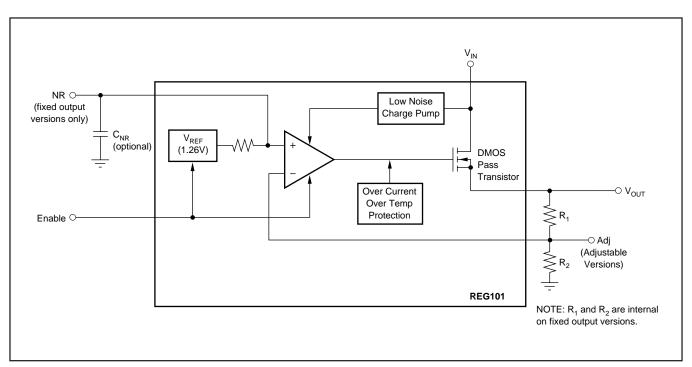


FIGURE 4. Block Diagram.



DROP-OUT VOLTAGE

The REG101 uses an N-channel DMOS as the "pass" element. When the input voltage is within a few tens of millivolts of the output voltage, the DMOS device behaves like a resistor. Therefore, for low values of $V_{\rm IN}$ to $V_{\rm OUT}$, the regulator's input-to-output resistance is the Rds_ON of the DMOS pass element (typically $600 {\rm m}\Omega$). For static (DC) loads, the REG101 will typically maintain regulation down to $V_{\rm IN}$ to $V_{\rm OUT}$ voltage drop of $60 {\rm mV}$ at full rated output current. In Figure 6, the bottom line (DC dropout) shows the minimum $V_{\rm IN}$ to $V_{\rm OUT}$ voltage drop required to prevent drop-out under DC load conditions.

For large step changes in load current, the REG101 requires a larger voltage drop across it to avoid degraded transient response. The boundary of this "transient drop-out" region is shown as the top line in Figure 6. Values of $V_{\rm IN}$ to $V_{\rm OUT}$ voltage drop above this line insure normal transient response.

In the transient dropout region between "DC" and "Transient", transient response recovery time increases. The time required to recover from a load transient is a function of both the magnitude and rate of the step change in load current and the available "headroom" $V_{\rm IN}$ to $V_{\rm OUT}$ voltage drop. Under worst-case conditions (full-scale load change with $V_{\rm IN}$ to $V_{\rm OUT}$ voltage drop close to DC dropout levels), the REG101 can take several hundred microseconds to re-enter the specified window of regulation.

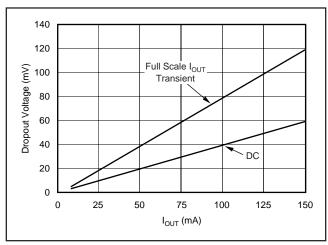


FIGURE 6. Transient and DC Dropout.

TRANSIENT RESPONSE

The REG101 response to transient line and load conditions improves at lower output voltages. The addition of a capacitor (nominal value $0.47\mu F$) from the output pin to ground may improve the transient response. In the adjustable version, the addition of a capacitor, C_{FB} (nominal value 10nF), from the output to the adjust pin will also improve the transient response.

THERMAL PROTECTION

The REG101 has thermal shutdown circuitry that protects the regulator from damage. The thermal protection circuitry disables the output when the junction temperature reaches approximately 160°C, allowing the device to cool. When the junction temperature cools to approximately 140°C, the output circuitry is again enabled. Depending on various conditions, the thermal protection circuit may cycle on and off. This limits the dissipation of the regulator, but may have an undesirable effect on the load.

Any tendency to activate the thermal protection circuit indicates excessive power dissipation or an inadequate heat sink. For reliable operation, junction temperature should be limited to 125°C, maximum. To estimate the margin of safety in a complete design (including heat sink), increase the ambient temperature until the thermal protection is triggered. Use worst-case loads and signal conditions. For good reliability, thermal protection should trigger more than 35°C above the maximum expected ambient condition of your application. This produces a worst-case junction temperature of 125°C at the highest expected ambient temperature and worst-case load.

The internal protection circuitry of the REG101 has been designed to protect against overload conditions. It was not intended to replace proper heat sinking. Continuously running the REG101 into thermal shutdown will degrade reliability.



POWER DISSIPATION

The REG101 is available in two different package configurations. The ability to remove heat from the die is different for each package type and, therefore, presents different considerations in the printed circuit board (PCB) layout. The PCB area around the device that is free of other components moves the heat from the device to the ambient air. While it is difficult to impossible to quantify all of the variables in a thermal design of this type, performance data for several configurations are shown in Figure 7.

Power dissipation depends on input voltage, load condition, and duty cycle. Power dissipation is equal to the product of the average output current times the voltage across the output element, $V_{\rm IN}$ to $V_{\rm OUT}$ voltage drop.

$$P_{D} = (V_{IN} - V_{OUT}) \bullet I_{OUT(AVG)}$$

Power dissipation can be minimized by using the lowest possible input voltage necessary to assure the required output voltage.

REGULATOR MOUNTING

Solder pad footprint recommendations for the various REG101 devices are presented in the Application Bulletin AB-132, "Solder Pad Recommendations for Surface-Mount Devices" (SBFA015), available from the Texas Instruments web site (www.ti.com).

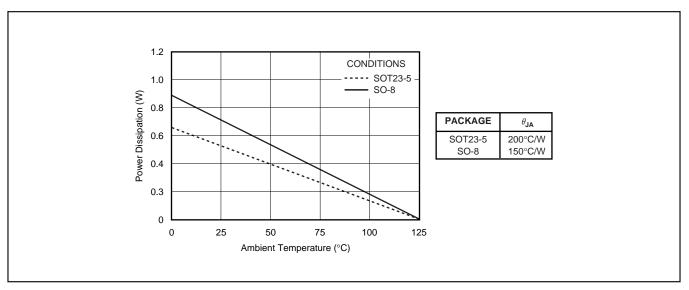


FIGURE 7. Maximum Power Dissipation versus Ambient Temperature for the Various Packages.





7-Apr-2024 www.ti.com

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
REG101NA-2.5/250	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	RO1G	Samples
REG101NA-2.5/3K	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	RO1G	Samples
REG101NA-2.8/250	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	RO1E	Samples
REG101NA-2.85/250	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	RO1N	Samples
REG101NA-2.85/3K	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	RO1N	Samples
REG101NA-3.3/250	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	RO1C	Samples
REG101NA-3.3/250G4	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	RO1C	Samples
REG101NA-3.3/3K	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	RO1C	Samples
REG101NA-3.3/3KG4	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	RO1C	Samples
REG101NA-3/250	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	R01D	Samples
REG101NA-3/3K	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	R01D	Samples
REG101NA-5/250	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	RO1B	Samples
REG101NA-5/3K	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	RO1B	Samples
REG101NA-A/250	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	RO1A	Samples
REG101NA-A/3K	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	RO1A	Samples
REG101NA-A/3KG4	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	RO1A	Samples
REG101UA-2.5	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	REG 101U25	Samples
REG101UA-3	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	REG 101U30	Samples
REG101UA-3.3	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	REG 101U33	Samples



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Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
REG101UA-3.3/2K5	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	REG 101U33	Samples
REG101UA-5	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	REG 101U50	Samples
REG101UA-5/2K5	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	REG 101U50	Samples
REG101UA-A	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	REG 101UA	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE OPTION ADDENDUM

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In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.



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TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
REG101NA-2.5/250	SOT-23	DBV	5	250	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
REG101NA-2.5/3K	SOT-23	DBV	5	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
REG101NA-2.8/250	SOT-23	DBV	5	250	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
REG101NA-2.85/250	SOT-23	DBV	5	250	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
REG101NA-2.85/3K	SOT-23	DBV	5	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
REG101NA-3.3/250	SOT-23	DBV	5	250	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
REG101NA-3.3/3K	SOT-23	DBV	5	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
REG101NA-3/250	SOT-23	DBV	5	250	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
REG101NA-3/3K	SOT-23	DBV	5	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
REG101NA-5/250	SOT-23	DBV	5	250	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
REG101NA-5/3K	SOT-23	DBV	5	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
REG101NA-A/250	SOT-23	DBV	5	250	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
REG101NA-A/3K	SOT-23	DBV	5	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
REG101UA-3.3/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
REG101UA-5/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1



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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
REG101NA-2.5/250	SOT-23	DBV	5	250	203.0	203.0	35.0
REG101NA-2.5/3K	SOT-23	DBV	5	3000	200.0	183.0	25.0
REG101NA-2.8/250	SOT-23	DBV	5	250	200.0	183.0	25.0
REG101NA-2.85/250	SOT-23	DBV	5	250	200.0	183.0	25.0
REG101NA-2.85/3K	SOT-23	DBV	5	3000	200.0	183.0	25.0
REG101NA-3.3/250	SOT-23	DBV	5	250	200.0	183.0	25.0
REG101NA-3.3/3K	SOT-23	DBV	5	3000	200.0	183.0	25.0
REG101NA-3/250	SOT-23	DBV	5	250	200.0	183.0	25.0
REG101NA-3/3K	SOT-23	DBV	5	3000	200.0	183.0	25.0
REG101NA-5/250	SOT-23	DBV	5	250	200.0	183.0	25.0
REG101NA-5/3K	SOT-23	DBV	5	3000	200.0	183.0	25.0
REG101NA-A/250	SOT-23	DBV	5	250	200.0	183.0	25.0
REG101NA-A/3K	SOT-23	DBV	5	3000	200.0	183.0	25.0
REG101UA-3.3/2K5	SOIC	D	8	2500	356.0	356.0	35.0
REG101UA-5/2K5	SOIC	D	8	2500	356.0	356.0	35.0

PACKAGE MATERIALS INFORMATION

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TUBE



*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
REG101UA-2.5	D	SOIC	8	75	506.6	8	3940	4.32
REG101UA-3	D	SOIC	8	75	506.6	8	3940	4.32
REG101UA-3.3	D	SOIC	8	75	506.6	8	3940	4.32
REG101UA-5	D	SOIC	8	75	506.6	8	3940	4.32
REG101UA-A	D	SOIC	8	75	506.6	8	3940	4.32



SMALL OUTLINE TRANSISTOR



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
 2. This drawing is subject to change without notice.
 3. Reference JEDEC MO-178.

- 4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.
- 5. Support pin may differ or may not be present.



SMALL OUTLINE TRANSISTOR



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE TRANSISTOR



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.





SMALL OUTLINE INTEGRATED CIRCUIT



NOTES:

- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- 4. This dimension does not include interlead flash.
- 5. Reference JEDEC registration MS-012, variation AA.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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