Using the ADS8382 With the TMS320C6713 DSP

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Data Acquisition Applications

ABSTRACT

This application report presents a solution to interfacing the ADS8382 18-bit, 600-KSPS serial interface converter to the TMS320C6713 DSP. The hardware solution is made up of existing hardware: the ADS8382EVM, 'C6713 DSK, and 5-6K Interface Board. The software demonstrates how to use an EDMA and Timer peripheral to collect data at 598 kHz. Discussed also are some of the key points to remember when designing the hardware and writing the software. The software developed is available for download to involve the user in the discussion and as sample code to use in system development. Project collateral discussed in this application report can be downloaded from the following URL: www.ti.com/lit/zip/SLAA239.

Contents

1 Introduction ................................................................................................................... 1
2 Hardware...................................................................................................................... 1
3 Software Interface ........................................................................................................ 3
4 Conclusion .................................................................................................................... 7
5 References .................................................................................................................. 7
Appendix A Main.C ........................................................................................................... 8
Appendix B Functions.C ................................................................................................... 9

List of Figures

1 Hardware Connection .................................................................................................... 3
2 Data Transfer Block Diagram ........................................................................................ 3
3 Analog Input Signal Range ............................................................................................ 4
4 Config1.cdb ................................................................................................................ 6
5 Screen Shot .................................................................................................................. 6

List of Tables

1 Jumper Settings for ADS8382EVM........................................................................................ 2

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1 Introduction

The ADS8382 converter is the first high-resolution, high-speed, serial interface SAR converter to come from Texas Instruments' Burr Brown line of products. It is an analog-to-digital converter with an 18-bit resolution and a 600-kHz sample rate. This application report presents a hardware and software solution to interfacing and using the converter with the TMS320C6713 digital signal processor (DSP). The software developed uses the EDMA, in combination with Timer1, to collect 2048 samples. Discussed also are some key points to remember when interfacing the converter to host processors.

2 Hardware

The hardware solution involves theTMS320C6713 DSP Starter Kit (DSK), 5-6K Interface Board, and the ADS8382EVM. The hardware used in this report is available from Texas Instruments.
2.1 TMS320C6713 DSK

The TMS320C6713 DSK (‘C6713 DSK) not only provides an introduction to ‘C6000 technology, but is powerful enough to use for fast development of networking, communications, imaging, and other applications like data acquisition. For more information, search for part number TMDSDSK6713 on the TI Web site at www.ti.com.

2.2 ADS8382EVM

The ADS8382 evaluation module (EVM), or ADS8382EVM, is an easy way to test both the functional and dynamic performance of this 18-bit, analog-to-digital converter (ADC). The EVM includes those circuits essential to showing the performance of the converter and interfacing it to a serial bus. These circuits include the analog input, reference, power, and digital buffer circuits. Table 1 describes solder and pin jumper settings used in this application report.

<table>
<thead>
<tr>
<th>REFERENCE DESIGNATOR</th>
<th>DESCRIPTION</th>
<th>JUMPER</th>
</tr>
</thead>
<tbody>
<tr>
<td>SJP1</td>
<td>Buffer onboard reference, REF1004-2.5</td>
<td>Installed (2)</td>
</tr>
<tr>
<td></td>
<td>Buffer user reference voltage applied at P1 pin 20</td>
<td>Installed</td>
</tr>
<tr>
<td>SJP2</td>
<td>Apply external reference directly to SJP4</td>
<td>Installed (2)</td>
</tr>
<tr>
<td></td>
<td>Apply buffered external reference to SJP4</td>
<td>Installed</td>
</tr>
<tr>
<td>SJP3</td>
<td>Set negative supply of U3 to ground</td>
<td>Installed</td>
</tr>
<tr>
<td></td>
<td>Select negative supply of U3 to –VCC.</td>
<td>Installed (2)</td>
</tr>
<tr>
<td>SJP4</td>
<td>Apply internal reference to REFIN</td>
<td>Installed (2)</td>
</tr>
<tr>
<td></td>
<td>Apply external reference to REFIN</td>
<td>Installed</td>
</tr>
<tr>
<td>SJP5</td>
<td>Apply common mode voltage to THS4131</td>
<td>Installed (2)</td>
</tr>
<tr>
<td>W1</td>
<td>Set BVDD to +5VD</td>
<td>Installed (2)</td>
</tr>
<tr>
<td></td>
<td>Set BVDD to +3.3VD</td>
<td>Installed</td>
</tr>
<tr>
<td>W3</td>
<td>Set signal PD High</td>
<td>Installed</td>
</tr>
<tr>
<td>W4</td>
<td>Set signal FS High</td>
<td>Installed</td>
</tr>
<tr>
<td>W5</td>
<td>Set signal CS LOW</td>
<td>Installed (2)</td>
</tr>
</tbody>
</table>

(1) Reference designator W2 is not used.
(2) Set jumpers as shown.

2.3 5-6K Interface Evaluation Module

Many TI data acquisition evaluation modules (EVM) are built to have a common set of connectors and signals at these connectors. The 5-6K Interface Board allows designers to easily connect these EVMs to the ‘C5000 and ‘C6000 family of digital signal processor starter kits (DSKs).

The 5-6K Interface Board consists of two serial, two signal conditioning areas, and one set of parallel interface connectors. The ADS8382EVM plugs onto connectors J10 (analog), J16 (serial control) and JP5 (Power). See TI literature number SLAU104 for more information on the 5-6k Interface Board, or search for keyword 5-6K Interface on the TI Web site.

2.4 Hardware Connections

The hardware connections used for this application report are shown in Figure 1. Normally, the ‘C6713 DSP signals CLXX0, FSR0, and DRR0 are connected to ADS8382 signals SCLK, FS, and SDO, respectively. This particular converter has a number of timing restrictions call quiet times (explained later). In order to meet those restrictions with a DSP, the BUSY signal is used as the frame sync to the DSP and the converter. ADS8382 timing specification, tH8, calls for a 15-ns hold time between BUSY falling and frame sync. This is accomplished on the ADS8382EVM by using the BUSY signal buffer (U11) and FS
buffer (U6). Each of these buffers will delay the signal about 8ns, giving us a total of 16ns which exceeds the tH8. Since there isn’t a jumper to short BUSY to FS on the EVM, a small wire was used to short buffered BUSY signal at connector P2 pin 15 to connector P2 pin 7. This means the DSP has to be programmed for clock-stop mode and the converter for Read Frame controlled by FS. This interface is one way of meeting all the timing restrictions required to achieve optimal performance.

**Figure 1. Hardware Connection**

### 3 Software Interface

The objective here is to collect a block of samples as quickly as possible, while freeing up the processor to perform other tasks. The processor should be alerted only when the samples are ready for processing. The most efficient way of doing this is with an EDMA channel, one of the timers, and an interrupt service routine. For this discussion, it is assumed that the reader is familiar with the DSP and its peripherals. If not, the reader should study the relevant documentation listed in the References section at the end of this application report.

**Figure 2. Data Transfer Block Diagram**

Figure 2 provides a visual representation of the data flow through the hardware. Timer1 is programmed to generate a pulse at 598 kHz and wired to the convert-start pin of the ADS8382. The sampling frequency can be modified by writing to the Timer period register. McBSP0 is programmed for clock-stop mode, sometimes call SCI mode. As previously mentioned, the BUSY signal is used to frame each serial transfer. The McBSP0 transmit clock is configured as an output and is wired to SCLK. Serial data output signal from the converter is shorted to data receive register. EDMA channel 13 is synchronized to the McBSP0 receive event. Such that each time a receive event occurs, the EDMA transfers data from the receive data register to a data array called ad_buffer. Once all 2048 samples are captured, the EDMA transfers future data points to the location temp. Forcing samples 2049+ to temp prevents data in ad_buffer from being corrupted.
The ADS8382 is a high-performance, 18-bit, 600-kHz ADC with fully differential, pseudo-bipolar input. The term fully differential, pseudo-bipolar may be new to some readers. This phase refers to the characteristics of the analog input pins. Fully differential means that each input pin must be 180 degrees out of phase. Pseudo-bipolar means the input cannot go below ground reference or zero volts (see Figure 3). The input span(+IN - (-IN)) is from -VREF to +VREF. The maximum differential voltage between +IN and -IN is 2*VREF. The converter outputs code in 2's complement format.

![Figure 3. Analog Input Signal Range](image)

Traditionally, SAR-type converters have only two modes of operation. The device is either sampling the input voltage or converting the analog input to its digital representation. The ADS8382 analog-to-digital converter has two additional modes, Wait and NAP. Figure 40 of the ADS8382 data sheet (SLAS416) shows a graphical representation of the states. The device requires a certain combination of signals to enter into sample mode and another to enter into convert mode. If the convert-start signal is HIGH and chip select is LOW, the device enters sample mode. The device has a minimum sample time and a maximum sample time. The minimum sample is 500 ns and maximum of 1 ms. If the device is held in sample mode for more than 1 ms, the conversion result is adversely affected. Therefore, if continuous sampling is not wanted, place the device in either NAP or WAIT mode.

The ADS8382 has a CMOS serial interface that can be clocked at up to 40 MHz. This high rate allows the user to read the data out of the device within 450 ns after conversion is complete. The remaining time can therefore be used to allow the input to settle to half an 18-bit LSB level.

For best performance, this converter requires the user satisfy three quiet times. The quiet times state the digital bus must be inactive 1) 30 ns before the convert-start signal goes low, 2) 10 ns after the convert-start signal goes low, and 3) 600 ns before the BUSY signal goes low. The quiet time around convert start going low is to ensure that the ground reference plane is settled and quiet when the device goes to capture or hold the input voltage. Any digital glitch or ground bounce during this critical time will cause the voltage captured on the internal sample and hold circuit to vary. The last 600 ns of the conversion process is also critical because it is then that the device resolves the last few bits. A glitch in the ground plane, or substrate of the device, will show itself as an offset error in the digital word because the minus reference changed during the critical bit decision time. It is therefore very important the timing specifications be met.

The ADS8382 product data sheet (SLAS416) provides detailed explanation of this device or any of the aforementioned items.

### 3.1.2 C6713 DSP

The TMS320C6713 DSP and its respective peripherals (i.e., timer1, EMIF, and EDMA) need to be configured properly to work with the converter. It is assumed the reader has a working understanding of the host processor; therefore, the DSP setup is not covered in detail in this application report. For more information, see the downloadable code and references listed in section 5.

The settings for the various peripherals can be found and modified in the config1.cdb file. The seed file for the DSP/BIOS configuration file was the dskC6713.cdb file. All the peripheral register settings and interrupt control were handled within this configuration file, allowing the user code to be simple, containing only statements to enable the peripherals and format data.

The register settings for the Multi-Channel Buffered Serial Port (McBSP) follow:

```c
MCBSP_Config mcbspCfg0 = ( 0x00303800, /* Serial Port Control Reg. (SPCR) */ 0x00020060, /*
```
Receiver Control Reg. (RCR) */ 0x00010060, /* Transmitter Control Reg. (XCR) */ 0x201F0002, /*
Sample-
Rate Generator Reg. (SRGR) */ 0x00000000, /* Multichannel Control Reg. (MCR) */ 0x00000000, /*
Receiver Channel Enable(RCER) */ 0x00000000, /* Transmitter Channel Enable(XCER) */ 0x0000006D /*
Pin Control Reg. (PCR) */};

The serial port is set up for a 20-bit, clock-stop mode with a delayed clock. CLKX0 is an output pin, and
the clock signal was generated by the sample rate generator. The data shifted out is delayed by one
clock. In clock-stop mode, CLKX0 and CLKR0 signals are tied together internally. The serial shift clock
rate was set 37.5 MHz, because it's the highest frequency allowed that's less than 40MHz. The calculation
is as shown below:

\[
SCLK = \frac{\text{CPU CLOCK}}{2} / \text{CLKDIV} = \frac{225 \text{ MHz}}{2} / 3 = 37.5 \text{ MHz}
\]

Register settings for EDMA channel 13 follow:

```c
EDMA_Config edmaCfg13 = { 0x40360003, /* Option */ 0x00000000, /* Source Address -
Numeric */ 0x00000000, /* Transfer Counter -
Numeric */ (Uint32) ad_buffer, /* Destination Address -
Extern Decl. Obj */ 0x00000000, /* Index register -
Numeric */ 0x00010000 /* Element Count Reload and Link Address */};
```

The EDMA channel is configured to transfer 2048 data samples from the McBSP0 receive data register.
Each of those transfers are triggered by a receive event from McBSP0. The EDMA channel is linked to the
EDMA configuration as shown in the following code listing. After 2048 samples are captured, the EDMA
loads these settings, takes any additional data samples, and copies them over to the location `temp`. This is
done to ensure that the EDMA does not overwrite the data stored in the array `ad_buffer`.

```c
EDMA_Config edmaCfg0 = { 0x20160003, /* Option */ 0x00000000, /* Source Address -
Numeric */ 0x00000000, /* Transfer Counter - Numeric */ (Uint32) &temp, /* Destination Address -
Extern Decl. Obj */ 0x00000000, /* Index register -
Numeric */ 0x00010000 /* Element Count Reload and Link Address */};
```

After 2048 samples have been gathered, the EDMA interrupts the CPU. After some time, the CPU
executes function `hwIDMA_isr()`. In this function, the CPU stops the timer, which initiates new conversions.
Before resetting the various interrupts and exiting the function, the DSP clears out the last two bits of the
20-bit transfer, thereby leaving only the 18-bit digital representations of the input. The data is MSB-
mapped in the array `ad_buffer[]` and `data[]`. This means that the 18-bit data word is mapped MSB to MSB
in a 32-bit word.

The following is the register setting for timer1 which sets the sampling rate:

```c
TIMER_Config timerCfg1 = { 0x000002D3, /* Control Register (CTL) */ 0x0000005E, /* Period
Register (PRD) */ 0x00000000 /* Counter Register (CNT) */};
```

Timer1 input clock source is the CPU CLOCK divided by 4.

\[
\text{Timer1 Input Clock} = \frac{225\text{E6}}{4} = 56.25\text{MHz}
\]

The formula for setting the sampling frequency (Fs) is

\[
Fs = \frac{56.25\text{e6}}{(\text{Period Value})}
\]

OR

\[
\text{Period Value} = \frac{56.25\text{e6}}{(Fs)}
\]

The timer1 output frequency is set to achieve the highest possible sample rate of 598.4 kHz. Timer1 is set
to CPU clock divided by 4, pulse mode, Inverted, and two clocks per pulse. The period register is set to
94.

To change the timer frequency, open file `Config1.cdb` in Code Composer Studio™ and expand as shown in
Figure 4. Right click `timerCfg1`, select properties, and select counter control tab.
Figure 4. Config1.cdb

Figure 5. Screen Shot
4 Conclusion

This application report presented one solution to interfacing the ADS8382 converter to the 'C6713 DSP. The ADS8382EVM plugs onto the 5-6k Interface Board, which in turn plugs onto the 'C6713 DSK. All the hardware used for this application report can be ordered from Texas Instruments. The software solution involves using DSP/BIOS and configuration tool (i.e., config1.cdb file) to visually set up the EDMA, Timer, and interrupt. The EDMA and Timer1 are used to trigger a conversion cycle at 598 kHz. The C-language code developed with this report is available for download.

5 References

1. TMS320C621x/TMS320C671x EDMA Architecture application report (SPRA996)
2. Applications Using the TMS320C6000 Enhanced DMA application report (SPRA636)
4. 16-Bit, 600-kHz Fully Differential, Pseudo-Bipolar Input, Micropower Sampling Analog-to-Digital Converter with Serial Interface and Reference data sheet (SLAS416)
5. TMS320C6713, TMS320C6713B Floating-Point Digital Signal Processors data sheet (SPRS186)
6. ADS8382EVM User's Guide (SLAU143)
7. 5-6K Interface Board EVM User's Guide (SLAU104)
/**
 * @file main.c
 *
 * @brief This program uses McBSP0 of C6713 DSK to read 2048 samples continuously from the ADS8382 18-bit, 598KSPS Analog-to-Digital Converter. The 20-bit words read during each data frame is in "ad_buffer". The 18-bit digital sample is stored in array Data. The ADS8382EVM plugs onto the 5-6k interface card. The 5-6k interface card plugs onto the C6713 DSK. Hardware connections: 6713DSK (DSP/5-6k interface card) => ADS8382EVM * CNTLb => CS# (CS# can be tied LOW permanently) * CLKX(J16.3) => SCLK (P2.3) * FSX(J16.7) => BUSY (short BUSY P2.15 to P2.7 on ADS8382EVM.) * DRR(J16.13) => SDO (P2.13) * INT6 -> BUSY * Timer1 => CONVST# * AUTHOR: DAP Application Group, L. Philipose, Dallas * CREATED 2005 © BY TEXAS INSTRUMENTS INCORPORATED. * VERSION: 1.0
 **/

#include "Config1cfg.h"
#include "dsk6713.h"
#include "dc_conf.h"

void main() { int I=0; /* Initialize the board support library, must be first BSL call */
    DSK6713_init(); /* Set McBSP0 for use with daughtercard */
    /* Initialize the ad_buffer */
    for (i=0;i<BLOCK_SZ;i++)
        ad_buffer[i]= 0x0; data[i]= 0x0;
    /* Enable the EDMA controller interrupt */
    IRQ_enable(IRQ_EVT_EDMAINT); /*Reset EDMA interrupt */
    EDMA_enableChannel(hEdmaCha13); /*Enable EDMA interrupt */
    /*Configure EDMA Channel and McBSP0*/
    MCBSP_config(hMcbsp0,mcbspCFG0);
    /*Start McBSP0*/
    }
B.1 Functions.C

/* Include Header File */
#include "Config1cfg.h"
#include "dsk6713.h"
#include "dc_conf.h"

extern int ad_buffer[BLOCK_SZ];
extern int data[BLOCK_SZ];

void hwiDMA_isr() {
  int i = 0;

  /* Hardware Interrupt Function disables EDMA channels and Timer0. Then post software interrupt. */
  IRQ_reset(IRQ_EVT_EDMAINT);
  EDMA_intDisable(TCCINTNUM6);
  EDMA_intClear(TCCINTNUM6);
  EDMA_intEnable(TCCINTNUM6);
  IRQ_enable(IRQ_EVT_EDMAINT);

  for (i = 0; i < BLOCK_SZ; i++) {
    data[i] = (ad_buffer[i] & 0x000FFFFF) >> 2; /* LSB mapped in 20-bit xfer mode */
  }
}

/* To continuously collect samples. Link EDMA config. edmaConf13 instead of hEdmatbl13 instead of edmaConf13. Also comment out */
/* Timer_pause() function call below. */
for (i = 0; i < BLOCK_SZ; i++) {
  data[i] = (ad_buffer[i] & 0x000000FF) >> 2; /* LSB mapped in 20-bit xfer mode */
  data[i] = (ad_buffer[i] & 0xFFFFFFF0); /* MSB mapped in 20-bit xfer mode */
  IRQ_resetIRQ_EVT_EDMAINT();
  EDMA_intDisable(TCCINTNUM6);
  EDMA_intClear(TCCINTNUM6);
  EDMA_intEnable(TCCINTNUM6);
  IRQ_enable(IRQ_EVT_EDMAINT);
  // TIMER_resume(hTimer1); }
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