Li-Ion Battery Charger Solution Using an MSP430™ MCU

ABSTRACT

Rechargeable batteries are used as the power supply source in many portable electronic equipment such as laptops, cell phones, and digital cameras. The most popular rechargeable batteries are nickel-cadmium (NiCd), nickel-metal-hydride (NiMH), and lithium-ion (Li-ion), and charging circuits depend on the battery chemistry. This application report describes a method to charge a Li-ion battery using an MSP430™ microcontroller.

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1 Introduction

Li-ion is the chemistry of choice for portable applications, because of its high capacity-to-size ratio and a low self-discharge characteristic. Methods to implement battery charging solutions include options such as power management ICs, MCU controlled, and even logic devices. Advantages of the MCU-controlled charging method include safe charging, time efficiency, and low cost.

Battery capacity (C), expressed in milliamp-hours (mAh), is a measure of battery life between charges. Battery current has the units of C-rate. For example, for a 500 mA-h battery, the current corresponding to 1C is 500 mA and the current corresponding to 0.1C is 50 mA.

A Li-ion battery charging process can include three stages:

- Slow charge: Precharging stage using a current of 0.1C
- Fast charge: Constant current charging stage using a current of 1C
- Constant voltage charging stage

During the slow charge stage, the battery is charged with a constant low charge current of 0.1C if the battery voltage is below 2.5 V. If some batteries, like NiCd, are recharged without fully discharging, they suffer from a phenomenon called memory effect, which causes a reduction in the battery capacity. Li-ion batteries do not suffer from memory effect and do not need to be fully discharged before recharging. The slow charge stage is rarely used during the charging process of a Li-ion battery.

The fast charge (constant current) and constant voltage charging are the most important stages during a recharge process. Most Li-ion batteries have a fully charged voltage of 4.1 V or 4.2 V. The battery is first charged with a constant current of 1C until the battery voltage reaches 4.1 V or 4.2 V. The firmware continuously checks the charging current by sensing the voltage at the current sense resistor (R_{sense}) and adjusting the duty cycle of the PWM output from the MCU. The battery voltage is checked frequently. When the battery's voltage reaches 4.1 V or 4.2 V, the charger switches to constant voltage charging mode.

The battery is then charged with a constant voltage source at a fixed battery voltage of 4.1 or 4.2 V. The battery voltage is checked and maintained at 4.1 V by controlling the duty cycle of the PWM output. During this process the charging current starts to fall due to internal cell resistance. When the charging current falls below 0.1C, the charging process must stop.
If charging continues when the battery is fully charged, most of the electrical energy is converted to thermal energy. Overcharging batteries can cause overheating or explosion due to outgassing of the electrolyte and can severely reduce battery life. Li-ion batteries are extremely sensitive to overcharging, and it is critical that the final voltage be controlled to within ±50 mV of 4.1 V or 4.2 V. A battery charger design must be able to determine a fully charged battery to avoid overcharging. A few methods to determine a fully charged condition are:

- When the current drops to 0.1C during the constant voltage charging stage, the battery can be considered as fully charged.
- Measure the battery temperature to avoid overheating.
- Safe timing method: If the charging time is longer than a predetermined time, the battery can be considered as fully charged.
2 Measurement Circuit

Figure 2 shows block-level schematic of a charger.

A buck converter is used during the constant current and voltage charge stages. A buck converter is a step-down voltage converter that uses the inductor as a current source to the output load impedance, which is the battery. The PNP and NPN transistors form a switch that is controlled by a PWM signal. The Timer_A3 on the MSP430 MCU can control the current for charging the battery using the PWM feature. When this switch is closed (on), current flows through the inductor and the capacitor is charged (see Figure 3). When the switch is open (off), the inductor tries to maintain its current flow by inducing a voltage, because an inductor cannot have an instantaneous change in current. The current now flows through the diode and the inductor charges the capacitor (see Figure 4). The LC network acts as a low-pass filter, and if the PWM frequency is much higher than the cut-off frequency of the LC network, the capacitor voltage is constant and equal to the mean value of the input voltage to the buck converter.

Figure 2. Block Level Schematic

Figure 3. Buck Converter Switch On

Figure 4. Buck Converter Switch Off
The value of the inductor can be calculated by Equation 1.

\[ L = \text{Duty cycle} \times T \times (V_I - V_{\text{sat}} - V_O) / 2I_O \]

where
- Duty Cycle = the duty cycle of the PWM
- \( T \) = Period of the PWM
- \( V_O \) = Voltage output
- \( V_{\text{sat}} \) = Voltage loss on the switch
- \( V_I \) = Voltage input to the switch
- \( I_O \) = Current for constant current stage

Assuming \( V_I \) is 6 V, \( V_{\text{sat}} \) is 0.5 V, \( I_O \) is 500 mA, \( V_O \) is 4.575 V, \( 1/T \) is 15 kHz, duty cycle is 50%:

\[ V_O = V_{\text{bat}} + I_{\text{bat}} \times R_{\text{sense}} = 4.2 + 500 \text{ mA} \times 0.75 = 4.575 \text{ V} \]  

(2)

The inductor should be at least 62 µH. For this implementation, a value of 75 µH is used. When the timer is clocked from the DCO with a frequency of 3.84 MHz, the TACCR0 value must be 255 to achieve a PWM frequency of 15 kHz (3.84 MHz/256). The timer runs in up mode, and the timer output switches in toggle/set mode. The duty cycle of the timer output (TA1) can be controlled by adjusting the value of TACCR1. A PWM resolution of 8 bits is enough to control the constant current flow in the battery during the constant current charging stage and maintain a constant voltage on the battery during the constant voltage charging stage. If the capacitor is 220 µF and the inductor is 75 µH, the cutoff frequency of the LC network is 1.2 kHz \((1 / (2 \times \pi \times \sqrt{L \times C}))\), which is much lower than the PWM frequency. This helps the capacitor to effectively reduce the output voltage ripple and maintain a DC voltage level.

Three channels (A0, A1, A2) on the 10-bit ADC on the MSP430 MCU can be used to monitor the battery voltage, battery temperature, and battery current. 1 LSB is equal to \( V_{\text{ref}} / (N - 1) \), where \( V_{\text{ref}} \) is the reference voltage and \( N \) is the resolution in bits of the ADC. With a 1.5-V on chip reference, 1 LSB is \( 1.5 / 1023 = 1.47 \text{ mV} \).

The range of voltages that the ADC needs to detect can be calculated as follows:

The highest voltage, \( V_O \), during the constant current charging stage is 4.575 V. The voltage to the ADC input due to the voltage divider \((R1 = 2.1 \times R2)\) is 4.575 / 3.1 = 1.5 V. This is within \( V_{\text{ref}} \) and can be resolved by the ADC.

The lowest voltage that the ADC needs to detect is during the constant voltage stage to monitor the battery current and stop the charging process. The ADC needs to detect the voltage drop created by a 0.1C current through the battery. In this case it is 50 mA \( \times 0.75 = 37.5 \text{ mV} \). This is approximately 25 LSB of resolution and can be resolved by the ADC.

For this application, a thermistor (RT1) is connected to the negative pole of the battery. The thermistor resistance decreases with temperature and so does the thermistor voltage. A 10k thermistor was used.

An abnormally low voltage indicates overheating, and the charging process must stop. This voltage can be detected by the ADC input.
3 Software

The following sections describe the division of the software provided with this application report into a main routine and two ISRs.

3.1 Main

This routine configures the DCO to run SMCLK at 3.84 MHz. The Timer_A3 registers are configured to output a PWM on TA1. The timer is also configured to trigger ADC conversions. The ADC is configured in repeat sequence of channels mode along with the DTC, which is configured to continuously transfer data from channels A2, A1, and A0 to a RAM array.

Channel A2 samples the thermistor voltage (battery temperature), A1 monitors the voltage on Rsense (battery current), and A0 monitors the voltage on R2 (battery voltage).

The 1.5-V reference is output on the V_REF+ pin on the MCU to be used as a reference for thermistor voltage measurements.

3.2 ADC ISR

The ADC ISR is triggered when the DTC completes the transfer of one block (3 words of data) to ADCSA, which is a RAM array that stores conversion results from channel A2, A1, and A0.

If the battery temperature exceeds 40°C, the charging process is stopped. The timer, ADC, and the watchdog timer (interval timer mode) are halted.

In the beginning of the charge cycle, if the battery voltage is less than 1 V or greater than 4.3 V, a LED is set to indicate a short circuit or battery not detected status.

During the constant current and constant voltage charging stages, TACCR1 controls the duty cycle of PWM output on TA1 to maintain constant current and voltage.

3.3 Watchdog Timer ISR

When the charge current drops to 0.1C during the constant voltage charging stage, the watchdog timer is set as an interval timer clocked from ACLK/8. The watchdog timer ISR is entered every 8 seconds to create a 15-minute delay before ending the charging process.

The MSP430 MCU stays in LPM0 during the entire charging process and enters active mode only when set by the interrupt service routines.
3.4 Software Flowchart

Figure 5 shows a software flowchart of the charging algorithm.

![Software Flowchart Diagram]

**Figure 5. Software Flowchart**
4 Conclusion

The MSP430 microcontrollers are a good fit for a Li-ion battery charger solution, because of integrated peripherals like the high-resolution ADC and watchdog timer. The example application for this report has been implemented and tested to be functional for the operations described.

5 References

1. MSP430x1xx Family User's Guide
2. MSP430F11x2, MSP430F12x2 Mixed-Signal Microcontrollers
## Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

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