AN-452 MICROWIRE Serial Interface



Literature Number: SNOA743

# MICROWIRE<sup>™</sup> Serial Interface

### INTRODUCTION

MICROWIRE is a simple three-wire serial communications interface. Built into COPSTM, this standardized protocol handles serial communications between controller and peripheral devices. In this application note are some clarifications of MICROWIRE logical operation and of hardware and software considerations.

### LOGICAL OPERATION

The MICROWIRE interface is essentially the serial I/O port on COPS microcontrollers, the SIO register in the shift register mode. SI is the shift register input, the serial input line to the microcontroller. SO is the shift register output, the serial output line to the peripherals. SK is the serial clock; data is clocked into or out of peripheral devices with this clock.

It is important to examine the logical diagram of the SIO and SK circuitry to fully understand the operation of this I/O port (Figure 1).

National Semiconductor Application Note 452 Abdul Aleaf January 1992

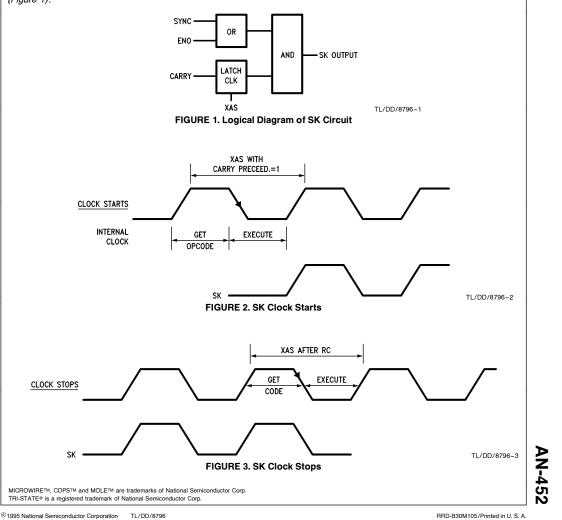


MICROWIRE Serial Interface

The output at SK is a function of SYNC, ENO, CARRY, and the XAS instruction. If CARRY had been set and propagated to the SKL latch by the execution of an XAS instruction, SYNC is enabled to SK and can only be overridden by ENO (*Figure 2*). Trouble could arise if the user changes the state of ENO without paying close attention to the state of the latch in the SK circuit.

If the latch is set to a logical high and the SIO register enabled as a binary counter, SK is driven high. From this state, if the SIO register is enabled as a serial shift register, SK will output the SYNC pulse immediately, without any intervening XAS instruction.

The SK clock (SYNC pulse) can be terminated by issuing an XAS instruction with CARRY = 0 (*Figure 3*).



The SIO register can be compared to four master-slave flipflops shown in *Figure 4*. The masters are clocked by the rising edges of the internal clock. The slaves are clocked by the falling edges of the internal clock. Upon execution of an XAS, the outputs of the masters are exchanged with the contents of the accumulator (read and overdrive) in such a way that the new data are present at the inputs of the four slaves when the falling edge of the internal clock occurs. The content of the accumulator is, therefore, latched respectively in the four slave flip-flops and bit 3 appears directly on SO.

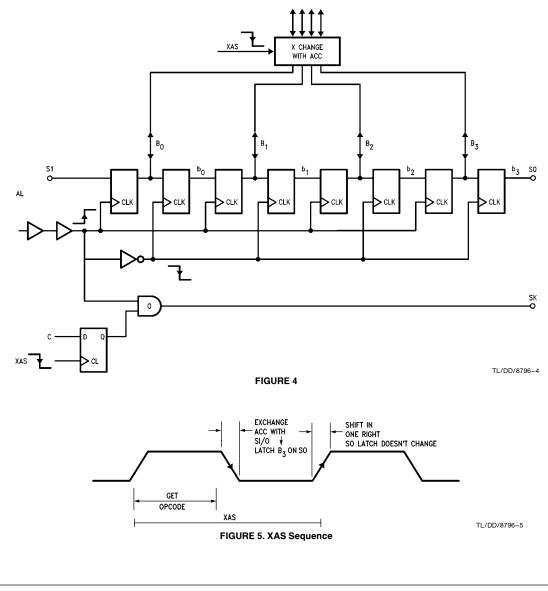
## This means that:

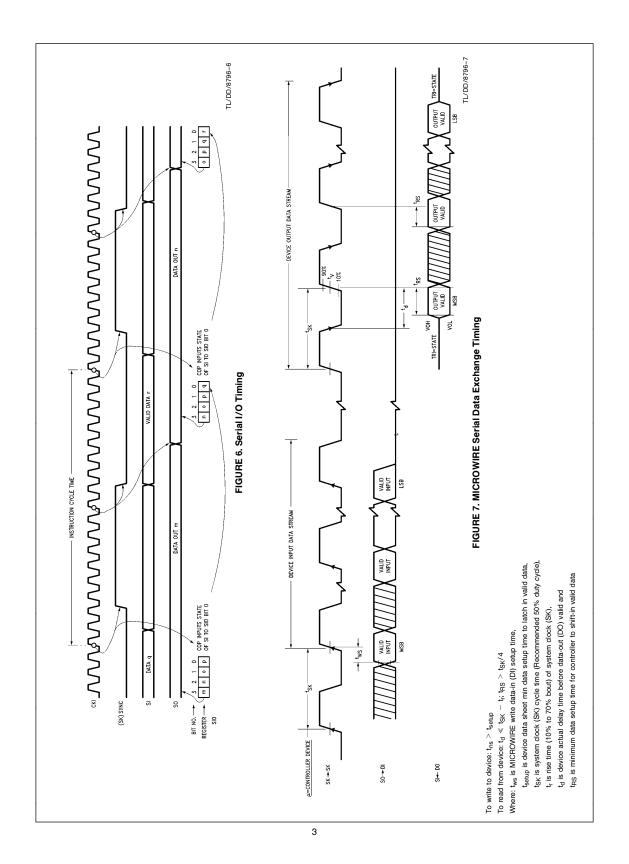
a) SO will be shifted out upon the falling edges of SK and will be stable during rising edges of SK.

b) SI will be shifted in upon the rising edge of SK, and will be stable when executing, i.e., an XAS instruction.

The shifting function is automatically performed on each of the four instruction cycles that follow an XAS instruction *(Figure 5)*.

When the SIO register is in the shift register mode (EN0 = 0), it left shifts its contents once each instruction cycle. The data present on the SI input is shifted into the least significant bit (bit 0) of the serial shift register. SO will output the most significant bit of the SIO register (bit 3) if EN3 = 1. Otherwise, SO is held low. The SK is a logic controlled clock which issues a pulse each instruction cycle. To ensure that the serial data stream is continuous, an XAS instruction must be executed every fourth time. Serial I/O timing is related to instruction cycle timing in the following way:





The first clock rising edge of the instruction cycle triggers the low-to-high transition of SYNC output via SK. At this time, the processor reads the state of SI into SIO bit 0, shifting the current bits 0-2 left. Halfway through the cycle (shown in *Figure 6* as the eight clock rising edge), SK is reset low and the new SIO bit 3 is outputted via SO.

# INTERFACING CONSIDERATIONS

To ensure data exchange, two aspects of interfacing have to be considered: 1) serial data exchange timing; 2) fan-out/ fan-in requirements. Theoretically, infinite devices can access the same interface and be uniquely enabled sequentially in time. In practice, however, the actual number of devices that can access the same serial interface depends on the following: system data transfer rate, system supply requirement capacitive loading on SK and SO outputs, the fan-in requirements of the logic families or discrete devices to be interfaced.

### HARDWARE INTERFACE

Provided an output can switch between a HIGH level and a LOW level, it must do so in a predetermined amount of time for the data transfer to occur. Since the transfer is strictly synchronous, the timing is related to the system clock (SK) (*Figure 7*). For example, if a COPS controller outputs a value at the falling edge of the clock and is latched in by the peripheral device at the rising edge, then the following relationship has to be satisfied:

### $t_{DELAY} + t_{SETUP} \le t_{CK}$

where  $t_{CK}$  is the time from data output starts to switch to data being latched into the peripheral chip,  $t_{\text{SETUP}}$  is the setup time for the peripheral device where the data has to be at a valid level, and  $t_{DELAY}$  is the time for the output to read the valid level.  $t_{CK}$  is related to the system clock provided by the SK pin of the COPS controller and can be increased by increasing the COPS instruction cycle time.

The maximum  $t_{\text{SETUP}}$  is specified in the peripheral chip data sheets. The maximum  $t_{\text{DELAY}}$  allowed may then be derived from the above relationship.

Most of the delay time before the output becomes valid comes from charging the capacitive load connected to the output. Each integrated circuit pin has a maximum load of 7 pF. Other sources come from connecting wires and connection from PC boards. The total capacitive load may then be estimated. The propagation delay values given in data sheets assume particular capacitive loads (e.g.  $V_{CC} = 5V$ ,  $V_{OH} = 0.4V$ , loading = 50 pF, etc.).

If the calculated load is less than the given load, those values should be used. Otherwise, a conservative estimate is to assume that the delay time is proportional to the capacitive load.

If the capacitive load is too large to satisfy the delay time criterion, then three choices are available. An external buffer may be used to drive the large load. The COPS instruction cycle may be slowed down. An external pullup resistor may be added to speed up the LOW level to HIGH level transition. The resistor will also increase the output LOW level and increase the HIGH level to LOW level transition time, but the increased time is negligible as long as the output LOW level changes by less than 0.3V. For a 100 pF load, the standard COPS controller may use a 4.7k external resistor, with the output LOW level increased by less than

0.2V. For the same load, the low power COPS controller may use a 22k resistor, with the SO and SK LOW levels increased by less than 0.1V.

Besides the timing requirements, system supply and fanout/fan-in requirements also have to be considered when interfacing with MICROWIRE. For the following discussion, we assume single supply push-pull outputs for system clock (SK) and serial output (SO), high-impedance input for serial input (SI).

To drive multi-devices on the same MICROWIRE, the output drivers of the controller need to source and sink the total maximum leakage current of all the inputs connected to it and keep the signal level within the valid logic "1" or logic "0". However, in general, different logic families have different valid "1" and "0" input voltage levels. Thus, if devices of different types are connected to the same serial interface, the output driver of the controller must satisfy all the input requirements of each device. Similarly, devices with TRI-STATE® outputs, when connected to the SI input, must satisfy the minimum valid input level of the controller and the maximum TRI-STATE leakage current of all outputs.

So, for devices that have incompatible input levels or source/sink requirements, external pull-up resistors or buffers are necessary to provide level-shifting or driving.

# SOFTWARE INTERFACE

The existing MICROWIRE protocol is very flexible, basically divided into two groups:

### 1) 1AAA.....ADDD.....D

where leading 1 is the start bit and leading zeroes are ignored.

AAA.....A is device variable instruction/address word.

DDD.....D is variable data stream between controller and device.

2) No start bit, just bit stream, i.e., bbb.....b

where b is a variable bit stream. Thus, device has to decode various fields within the bit stream by counting exact bit position.

### SERIAL I/O ROUTINES

Routines for handling serial I/O are provided below. The routines are written for 16-bit transmissions, but are trivially expandable up to 64-bit transmissions by merely changing the initial LBI instruction. The routines arbitrarily select register 0 as the I/O register. It is assumed that the external device requires a logic low chip select. It is further assumed that chip select is high and SK and SO are low on entry to the routines. The routines exit with chip select high, SK and SO low. GO is arbitrarily chosen as the chip select for the external device.

#### SERIAL DATA OUTPUT

This routine outputs the data under the conditions specified above. The transmitted data is preserved in the microcontroller.

0UT2:	LBI	0,12	; point to start of data word
	SC OGI	14	; select the external
			device

Features GENERAL Chip Function Process			Pa	Part Number		
GENERAL Chip Function Process	<u>I</u>	DS8908	MM545X	COP472-3	COP430 (ADC83X)	NM93C06A
Chip Function Process						
Process		AM/PM PLL	LED Display Driver	LCD Display Driver	A/D	E <sup>2</sup> PROM
		ECL	SOMN	CMOS	CMOS	NMOS
V <sub>CC</sub> Range		4.75V-5.25V	4.5V-11V	3.0V-5.5V	4.5V-0.3V	4.5V-5.5V
Pinout		20	40	20	8/14/20	14
HARDWARE INTERFACE						
Min V <sub>IH</sub> / Max V <sub>IL</sub>		2.1V/0.7V	2.2V/0.8V	0.7V <sub>CC</sub> /0.8V	2.0V/0.8V	2.0V/0.8V
SK Clock Range		0-625 kHz	0-500 kHz	4–250 kHz	10–200 kHz	0-250 kHz
Write Setup	tup	0.3 µ.s	0.3 µs	1 JuS	0.2 µs	0.4 µs
Data DI Min	bi ci	0.8 µs	(3)	100 ns	0.2 µs	0.4 µs
Read Data Prop Delay		(Note 4)	(Note 3)	(Note 3)	(Note 3)	
Chip	dn	Sπl 0.3	0.4 µS	1 μs (Note 1)	0.2 µs	0.2 µs
Enable HOLD	LD LD	0.8 µ.s	(Note 3)	1 μs (Note 2)	0.2 µs	0
Max AM	5	8 MHz	(Note 3)	(Note 3)	(Note 3)	(Note 3)
Frequency Range FM	5	120 MHz	(Note 3)	(Note 3)	(Note 3)	(Note 3)
SOFT						
Serial I/O Protocol		11D1D20	1D1D35	b1b40	1xxx	1 AADD
Instruction/Address Word		None	None	None	(Note 4)	(Note 4)
Note 1: Reference to SK rising edge. Note 2: Reference to SK falling edge. Note 3: Not defined. Note 4: See data sheet for different modes of operation.	s of operation.					

```
LEI
            8
                    : enable shift
                      register mode
        JP
             SEND2
SEND1:
        XAS
SEND2:
        LD
                    : data output loop
        XIS
        JP
             SEND1
        XAS
                    ; send last data
        RC
        CLRA
        NOP
        XAS
                    ; turn SK clock off
        OGI
             15
                    ; deselect the device
        LEI
                    ; turn SO low
             0
        RET
```

The code for reading serial data is almost the same as the serial output code. This should be expected because of the nature of the SIO register and the XAS instruction.

# MICROWIRE STANDARD FAMILY

A whole family of off-the-shelf devices exists that is directly compatible with MICROWIRE serial data exchange standard. This allows direct interface with the COPS family of microcontrollers.

Table I provides a summary of the existing devices and their functions and specifications.

# TYPICAL APPLICATION

Figure  $\theta$  shows pin connection involved in interfacing an E<sup>2</sup>PROM with the COP420 microcontroller.

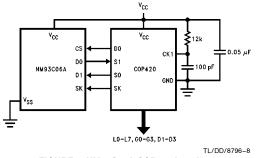
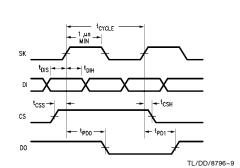


FIGURE 8. NM93C06A COP420 Interface

The following points have to be considered:

- 1. For NM93C06A the SK clock frequency should be in the 0 kHz-250 kHz range. This is easily achieved with COP420 running at 4  $\mu$ s-10  $\mu$ s instruction cycle time (SK period is the COP420 instruction cycle time). Since the minimum SK clock high time is greater than 1  $\mu$ s, the duty cycle is not a critical factor as long as the frequency does not exceed the 250 kHz max.
- CS low period following an E/W instruction must not exceed 30 ms maximum. It should be set at typical or minimum spec of 10 ms. This is easily done in software using the SKT timer on COP420.



### FIGURE 9. NM93C06A Timing

- 3. As shown in WRITE timing diagram, the start bit on DI must be set by a "0" to "1" transition following a CS enable ("0" to "1") when executing any instruction. One CS enable transition can only execute one instruction.
- 4. In the read mode, following an instruction and data train, the DI can be a "don't care," while the data is being outputted, i.e., for the next 17 bits or clocks. The same is true for other instructions after the instruction and data has been fed in.
- 5. The data-out train starts with a dummy bit 0 and is terminated by chip deselect. Any extra SK cycle after 16 bits is not essential. If CS is held on after all 16 of the data bits have been outputted, the DO will output the state of DI until another CS LO to HI transition starts a new instruction cycle.
- After a read cycle, the CS must be brought low for one SK clock cycle before another instruction cycle starts.

### INSTRUCTION SET

Commands	Opcode	Comments
READ	10000A3A2A1A0	Read Register 0-15
WRITE	11000A3A2A1A0	Write Register 0-15
ERASE	10100A3A2A1A0	Erase Register 0-15
EWEN	111000 0 0 1	Write/Erase Enable
ENDS	111000 0 1 0	Write/Erase Disable
***WRAL	111000 1 0 0	Write All Registers
ERAL	111000 1 0 1	Erase All Registers

All commands, data in, and data out are shifted in/out on rising edge of SK clock.

Write/erase is then done by pulsing CS low for 10 ms. All instructions are initiated by a LO-HI transition on CS followed by a LO-HI transition on DI.

- READ— After read command is shifted in DI becomes don't care and data can be read out on data out, starting with dummy bit zero.
- WRITE— Write command shifted in followed by data in (16 bits) then CS pulsed low for 10 ms minimum.

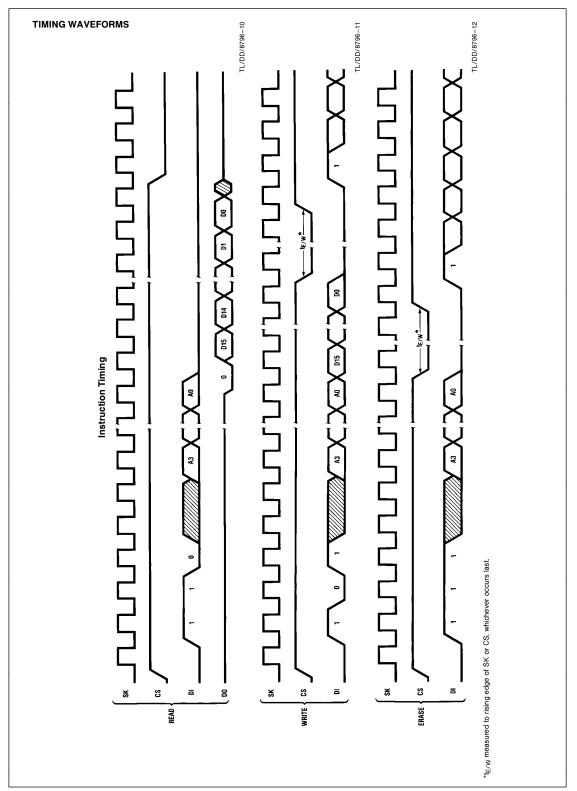
ERASE

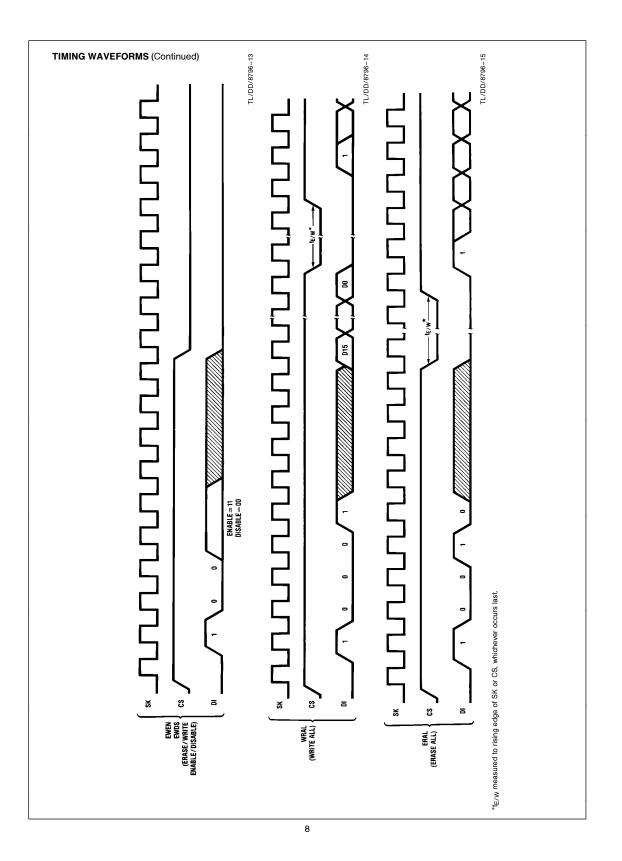
ERASE ALL—Command shifted in followed by CS low. WRITE ALL—Pulsing CS low for 10 ms.

WRITE

ENABLE/DISABLE—Command shifted in.

\*\*\*(This instruction is not speced on Data sheet.)





1				.TITLE		"I/O ROUTINE TO EVALUATE COP494"
2		01A4		.CHIP	420	
3	(	0000		.PAGE	0	
4						
5			;THIS I	S I/O RO	UTINE TO EV	ALUATE COP494
6			;			
7			;			
8						
9			;RAM VA		DECLARATION	
LO		000E			= 0, 14	;494 8BITS INST/ADDR WORD
11	(	0010		RWDATA	= 1, 12	;494 16BITS R/W DATA BUFFER
L2						
	000 (		PON:	CLRA		;POWER-ON INIT
	001 3			RC		RESET SK CLOCK
	002 4		at 5 · · ·	XAS	7 0	
	003		CLRAM:		3, 0	CLEAR RAM FROM 7, 0 TO 0, 15;
	004 (		CLR:			;
	005 (			XIS	at D	;
	006 (			JP	CLR	;CONTI CLEAR REG
	007 ]			XABR		;(A) TO BR
	008 5			AISC	15	REG O CLEARED?
		600F	DONE:		C494DR	Y, DONE CLEAR RAM, CALL 494 D
	00B 1			XABR	<b>47</b> D	;N, DEC BR
	000 (			JP	CLR	CONTI CLEAR REG TILL DONE
	00D 4			NOP		
	00E 4	44		NOP		
27			. * * *		OA DETWEE O	
28 29			***	START 4	94 DRIVER S	AMPLE CALLING SEQUENCE ***
59 30			C494DR:			TNIT GALLING CROUPNOR
	00 7 7	3350	0494DN;	OGI	0	;INIT CALLING SEQUENCE ;GO=L TO DESELECT 494
		3368		LEI	8	;GOEL TO DESELECT 494 ;ENABLE SIO AS S.R.
33		0000	ERASE:		0	,ENADLE SIO AS S.N.
	013 (	מח	DIADD.	LBI	COMMAND	;PRELOAD 494 ERASE REG A3-A0
	014 1			STII	OC	;PRELOAD 494 ERASE INST
	015 1			STII	0	SELECT REG A3-A0
	016 6			JSR	WI4P4	;SEND IT
38	、		WEEN:			· · · · · ·
	01B (	OD		LBI	COMMAND	;LOAD 494 WHEN REG A3-AO
	019 1			STII	3	PRELOAD 494 WREN INST
39	OlA 1			STII	0	SELECT REG A3-A0
39 10				JSR	WI494	;SEND IT
39 40 41	01B 6		WRITE:		-	·
39 40 41 42					COMMAND	;PRELOAD WR REG A3-A0
39 40 41 42 43		OD		LBI		PRELOAD 494 WRITE INST
39 40 41 42 43 44	01B 6			LBI STII	4	FREDORD 494 WRITE INST
39 40 41 42 43 44 45	01B 0	74			4 0	;SELECT REG A3-A0
39 40 41 42 43 44 45 46	01B 0 01D 0 01E 7	74 70		STII		-
39 40 41 42 43 44 45 46 47	01B 0 01D 0 01E 7 01F 7	74 70 1B		STII STII	0	;SELECT REG A3-A0
39 10 11 12 13 14 15 16 17 18	01B 0 01D 0 01E 7 01F 7 020 1	74 70 1B 75		STII STII LBI	0 RWDATA	;SELECT REG A3-A0
39 40 41 42 43 44 45 46 47 48 49	01B ( 01D ( 01E ( 01F ( 020 1 021 (	74 70 1B 75 7A		STII STII LBI STII	0 RWDATA 5	;SELECT REG A3-A0

51	024	7A		STII	0A	
		6900		JSR	WD494	;SEND THEM TO 494
53			READ:			
54	027	0D		LBI	COMMAND	;PRELOAD READ REG A3-A0
55	028	78		STII	8	PRELOAD 494 READ INST
56	029	70		STII	0	SELECT REG A3-A0
57	02A	6908		JSR	RD494	READ 494 DATA BACK VIA SI
58	02C	44		NOP		
59	02D	44		NOP		
30						
31		0080		.PAGE	2	;SUBROUTINE PAGE
32	080	32	SETUP:	RC		RESET SK BEFORE SELECT 494
33	081	4F		XAS		
34	082	3351		OGI	1	;GO=1 TO SELECT 494
35	084	00		CLRA		ENSURE SO=L BEFORE GEN START B
66	085	22		SC		;
	086			XAS		TURN ON SK CLOCK
38	087	00		CLRA		;GENERATE 494 START BIT
39	088	51		AISC	1	;
70	089	22		SC		;
71	08A	4F		XAS		;SEND IT AS MSB VIA SO
72	08B	OD		LBI	COMMAND	;FETCH 1ST INST/ADDR WORD
73	08C	05		LD		
74	08D	44		NOP		;
75	08E	4F		XAS		;SEND IT (MSB OF INST FIRST)
76	08F	0E		LBI	COMMAND+1	;FETCH 2ND INST/ADDR NIBBLE
77	090	05		LD		
78	091	44		NOP		
79	092	4F		XAS		;SEND IT
30	093	1B		LBI	RWDATA	;POINT TO READ/WRITE DATA BUFFER
31	094	48		RET		;RET OF SETUP
32						
33	095	00	TWEDLY:	CLRA		;VPP WIDTH, TWE>20MS @ 4Us/INST
34	096	5B	TWECONT:	AISC	11	;5 SKT LOOPS?
35	097	99		JP	. + 2	;N,CONTI
36	098	48	TWEDONE:	RET		;Y,DONE
37	099	41		SKT		;
38	09A	99		JP	1	;
39	09B	96		JP	TWECONT	;CONTI TWE TIME
90						
91	09C	48	RET;	RET		;2 CYCLES DELAY
92						
93		0100		.PAGE	4	;
94						
95			•*** 9	START	494 I/O DRIVER	SUBROUTINE ***
96						
97	100	80	WD494:	JSRP	SETUP	;ENTRY TO WRITE 494 REG A3-A0
98	101	05	RWLOOP:	LD		;R/W 494 16 DATA BITS
99	102	4F		XAS		;
00	103	04		XIS		;

I/O ROUTINE TO EVALUATE COP494 (Continued)						
101	104	Cl		JP	RWLOOP	
102	105	3350		OGI	0	;DESELECT 494 AFTER R/W DATA
103	107	Dl		JP	FINI	;
104	108	80	RD494:	JSRP	SETUP	;ENTRY TO RD 494 REG A3-A0
105	109	00		CLRA		;FINISH SEND OUT A3-AO VIA SO
106	10A	44		NOP		;
107	10B	44		NOP		;WAIT 1BIT TIME FOR VALID D15
108	100	44		NOP		
109	10D	Cl		JP	RWLOOP	;
110	10E	80	WI494:	JSRP	SETUP	;ENTRY TO WRITE INST TO 494
111	10F	00		CLRA		;ENSURE SO = L
112	110	4F		XAS		;
113	111	00	FINI:	CLRA		;ENSURE SO = L BETWEEN INST
114	112	3350		OGI	0	;DESELECT 494 BETWEEN INST WRIT
115	114	32		RC		;
116	115	4F		XAS		;TURN OFF SK CLOCK
117	116	95		JSRP	TWEDLY	;DELAY TWE >20MS TO PULSE VPP=21
118	117	48		RET		;RET OF WD494 OR RD494 OR WI494
119						
120				.END		

### SOFTWARE DEBUG OF SERIAL REGISTER FUNCTIONS

In order to understand the method of software debug when dealing with the SIO register, one must first become familiar with the method in which the COPS MOLETM (Development System) BREAKPOINT and TRACE operations are carried out. Once these operations are explained, the difficulties which could arise when interrogating the status of the SIO register should become apparent.

### SERIAL OUT DURING BREAKPOINT

When the MOLE BREAKPOINTs, the COPS user program execution is stopped and execution of a monitor-type program, within the COP device, is started. At no time does the COP part "idle." The monitor program loads the development system with the information contained in the COP registers.

Note also that single-step is simply a BREAKPOINT on every instruction.

If the COP chip is BREAKPOINTed while a serial function is in progress, the contents of the SIO register will be destroyed.

By the time the monitor program dumps the SIO register to the MOLE, the contents of the SIO register will have been written over by clocking in SI. To inspect the SIO register using BREAKPOINT, an XAS must be executed prior to BREAKPOINT; therefore, the SIO register will be saved in the accumulator. An even more severe consequence is that the monitor program executes an XAS instruction to get the contents of the SIO register to the MOLE. Therefore, the SK latch is dependent on the state of the CARRY prior to the BREAKPOINT. In order to guarantee the integrity of the SIO register, one must carefully choose the position of the BREAKPOINT address.

As can be seen, it is impossible to single-step or BREAK-POINT through a serial operation in the SIO register.

### SERIAL OUT DURING TRACE

In the TRACE mode, the user's program execution is never stopped. This mode is a real-time description of the program counter and the external event lines; therefore, the four external event lines can be used as logic analyzers to monitor the state of any input or output on the COPS device. The external event lines must be tied to the I/O which is to be monitored.

The state of these I/O (external event lines) is displayed along with the TRACE information. The safest way to monitor the real-time state of SO is to use the TRACE function in conjunction with the external event lines.

#### CONCLUSIONS

National's super-sensible MICROWIRE serial data exchange standard allows interfacing to any number of specialized peripherals using an absolute minimum number of valuable I/O pins; this leaves more I/O lines available for system interfacing and may permit the COPS controller to be packaged in a smaller package.

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