# Implementing Fast Fourier Transform Algorithms of Real-Valued Sequences With the TMS320 DSP Platform 


#### Abstract

The Fast Fourier Transform (FFT) is an efficient computation of the Discrete Fourier Transform (DFT) and one of the most important tools used in digital signal processing applications. Because of its well-structured form, the FFT is a benchmark in assessing digital signal processor (DSP) performance.

The development of FFT algorithms has assumed an input sequence consisting of complex numbers. This is because complex phase factors, or twiddle factors, result in complex variables. Thus, FFT algorithms are designed to perform complex multiplications and additions. However, the input sequence consists of real numbers in a large number of real applications.

This application report discusses the theory and usage of two algorithms used to efficiently compute the DFT of real-valued sequences as implemented on the Texas Instruments TMS320C6000™.

The first algorithm performs the DFT of two $N$-point real-valued sequences using one $N$-point complex DFT and additional computations.

The second algorithm performs the DFT of a 2 N -point real-valued sequence using one N -point complex DFT and additional computations. Implementations of these additional computations, referred to as the split operation, are presented both in C and C6000™ assembly language. For implementation on the C6000, optimization techniques in both C and assembly are covered.


## Contents

$\qquad$
1 Introduction3
2 Basics of the DFT and FFT ..... 3
3 Efficient Computation of the DFT of Real Sequences ..... 5
3.1 Efficient Computation of the DFT of Two Real Sequences ..... 5
3.2 Efficient Computation of the DFT of a 2 N -Point Real Sequence ..... 7
4 TMS320C62xE Architecture and Tools Overview ..... 11
5 Implementation and Optimization of Real-Valued DFTs ..... 14
6 Summary ..... 17
7 References ..... 17
TMS320C6000 and C6000 are trademarks of Texas Instruments.
Trademarks are the property of their respective owners.
Appendix A Derivation of Equation Used to Compute the DFT/IDFT of Two Real Sequences ..... 18
A. 1 Forward Transform ..... 18
A. 2 Inverse Transform ..... 20
Appendix B Derivation of Equations Used to Compute the DFT/IDFT of a Real Sequence ..... 21
B. 1 Forward Transform ..... 21
B. 2 Inverse Transform ..... 23
Appendix C C Implementations of the DFT of Real Sequences ..... 27
C. 1 Implementation Notes ..... 27
Appendix D Optimized C Implementation of the DFT of Real Sequences ..... 42
D. 1 Implementation Notes ..... 42
D. 2 Description ..... 42
Appendix E Optimized C-Callable 'C62xx Assembly Language Functions Used to Implement the DFT of Real Sequences ..... 54
E. 1 Implementation Notes ..... 54
List of Figures
Figure 1. TMS320C6201 DSP Block Diagram ..... 11
Figure 2. Code Development Flow Chart ..... 13
List of Tables
Table 1. Comparison of Computational Complexity for Direct Computationof the DFT Versus the Radix-2 FFT Algorithm ..... 4
List of Examples
Example 1. Efficient Computation of the DFT of a $2 N$-Point Real Sequence ..... 15
Example 2. Efficient Computation of the DFT of Two Real Sequences ..... 15
Example 3. Efficient Computation of the DFT of a 2 N -Point Real Sequence ..... 16
Example 4. Efficient Computation of the DFT of Two Real Sequences ..... 16
Example 5. Efficient Computation of the DFT of a 2N-Point Real Sequence ..... 16
Example 6. Efficient Computation of the DFT of Two Real Sequences ..... 16
Example C-1. realdft1.c File ..... 27
Example C-2. split1.c File ..... 31
Example C-3. data1.c File ..... 32
Example C-4. params1.h File ..... 32
Example C-5. realdft2.c File ..... 33
Example C-6. split2.c File ..... 36
Example C-7. data2.c File ..... 37
Example C-8. params2.h File ..... 37
Example C-9. dft.c File ..... 37
Example C-10. params.h File ..... 39
Example C-11. vectors.asm ..... 40
Example C-12. Ink.cmd ..... 40
Example D-1. realdft3.c File ..... 42
Example D-2. realdft4.c File ..... 47
Example D-3. radix4.c File ..... 50
Example D-4. digit.c File ..... 51
Example D-5. digitgen.c File ..... 52
Example D-6. splitgen.c File ..... 53
Example E-1. split1.asm File ..... 54
Example E-2. split2.asm File ..... 61
Example E-3. radix4.asm File ..... 67
Example E-4. digit.asm File ..... 72

## 1 Introduction

Tl's C6000 platform of high-performance, fixed-point DSPs provides architectural and speed improvements that makes the FFT computation faster and easier to program than other fixed-point DSPs.

The C6000 platform devices are based on an advanced Very Long Instruction Word (VLIW) central processing unit (CPU) with eight functional units that include two multipliers and six arithmetic logic units (ALUs). The CPU can execute up to eight instructions per cycle.
Complementing the architecture is a very efficient C compiler that increases performance and reduces code development time. The C6000 architecture and development tools featured are discussed in this application report along with the following topics:

- Theory of DFTs of real-valued sequences
- Algorithm implementation
- C6000 CPU features
- C6000 development tools
- Optimizing C code for the C6000
- C-callable assembly language functions for the C6000


## 2 Basics of the DFT and FFT

Methods of performing the DFT of real sequences involve complex-valued DFTs. This section reviews the basics of the DFT and FFT.

The DFT is viewed as a frequency domain representation of the discrete-time sequence $x(n)$. The $N$-point DFT of finite-duration sequence $x(n)$ is defined as

$$
\begin{equation*}
X(k)=\sum_{n=0}^{N-1} x(n) W \stackrel{k n}{N} \quad I=0,1, \ldots, N-1 \tag{1}
\end{equation*}
$$

and the inverse DFT (IDFT) is defined as

$$
\begin{equation*}
X(n)=\frac{1}{N} \sum_{k=0}^{N-1} X(k) W-k n \quad n=0,1, \ldots, N-1 \tag{2}
\end{equation*}
$$

where

$$
\begin{equation*}
W_{N}^{n k}=e^{-j 2 \pi n k / N} \tag{3}
\end{equation*}
$$

The $W_{N} k n$ factor is also referred to as the twiddle factor. Observation of the above equations shows that the computational requirements of the DFT increase rapidly as the number of samples in the sequence $N$ increases. Because of the large computational requirements, direct implementation of the DFT of large sequences has not been practical for real-time applications. However, the development of fast algorithms, known as FFTs, has made implementation of DFT practical in real-time applications.

The definition of FFT is the same as DFT, but the method of computation differs. The basics of FFT algorithms involve a divide-and-conquer approach in which an $N$-point DFT is divided into successively smaller DFTs. Many FFT algorithms have been developed, such as radix-2, radix-4, and mixed radix; in-place and not-in-place; and decimation-in-time and decimation-in-frequency.

In most FFT algorithms, restrictions may apply. For example, a radix-2 FFT restricts the number of samples in the sequence to a power of two.

In addition, some FFT algorithms require the input or output to be re-ordered. For example, the radix-2 decimation-in-frequency algorithm requires the output to be bit-reversed. It is up to implementers to choose the FFT algorithm that best fits their application.

Table 1 compares the number of math computations involved in direct computation of the DFT versus the radix-2 FFT algorithm. As you can see, the speed improvement of the FFT increases as $N$ increases. Detailed descriptions of the DFT and FFT can be found in the references. ${ }^{123}$ The following sections describe methods of efficiently computing the DFT of real-valued sequences using complex-valued DFTs/IDFTs.

Table 1. Comparison of Computational Complexity for Direct Computation of the DFT Versus the Radix-2 FFT Algorithm

|  | Direct Computation of the DFT |  |  | Radix-2 FFT |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Number <br> of Points | Complex <br> Multiplies | Complex <br> Additions | $N^{2}-N$ |  | Complex <br> Multiplies |
| $N$ | 12 | 12 |  | $(N / 2) \log _{2} N$ | Complex <br> Additions |
| 4 | 16 | 240 | 4 | $N^{2} \log _{2} N$ |  |
| 16 | 256 | 4032 | 32 | 8 |  |
| 64 | 4096 | 65280 | 192 | 64 |  |
| 256 | 65536 | 1047552 | 1024 | 384 |  |
| 1024 | 1048576 |  | 5120 | 2048 |  |

## 3 Efficient Computation of the DFT of Real Sequences

In many real applications, the data sequences to be processed are real-valued. Even though the data is real, complex-valued DFT algorithms can still be used. One simple approach creates a complex sequence from the real sequence; that is, real data for the real components and zeros for the imaginary components. The complex DFT can then be applied directly. However, this method is not efficient. This section shows you how to use the complex-valued DFT algorithms to efficiently process real-valued sequences.

### 3.1 Efficient Computation of the DFT of Two Real Sequences

Suppose $x_{1}(n)$ and $x_{2}(n)$ are real-valued sequences of length $N$, and $x(n)$ is a complex-valued sequence defined as

$$
\begin{equation*}
x(n)=x_{1}(n)+j x_{2}(n) \quad 0 \leq n \leq N-1 \tag{4}
\end{equation*}
$$

The DFT of the two $N$-length sequences $x_{1}(n)$ and $x_{2}(n)$ can be found by performing a single $N$-length DFT on the complex-valued sequence and some additional computation. These additional computations are referred to as the split operation, and are shown below.

$$
\begin{align*}
& X_{1}(k)=\frac{1}{2}\left[X(k)+X^{*}(N-k)\right]  \tag{5}\\
& X_{2}(k)=\frac{1}{2 j}\left[X(k)-X^{*}(N-k)\right]
\end{align*}
$$

$$
k=0,1, \ldots, N-1
$$

As you can see from the above equations, the transforms of $x_{1}(n)$ and $x_{2}(n), X_{1}(k)$ and $X_{2}(k)$, respectively, are solved by computing one complex-valued DFT, $X(k)$, and some additional computations.

Now assume we want to get back $x_{1}(n)$ and $x_{2}(n)$ from $X_{1}(k)$ and $X_{2}(k)$, respectively. As with the forward DFT, the IDFT of $X_{1}(k)$ and $X_{2}(k)$ is found using a single complex-valued DFT. Because the DFT operation is linear, the DFT of equation (4) can be expressed as

$$
\begin{equation*}
X(k)=X_{1}(k)+j X_{2}(k) \tag{6}
\end{equation*}
$$

This shows that $X(k)$ can be expressed in terms of $X_{1}(k)$ and $X_{2}(k)$; thus, taking the inverse DFT of $X(k)$, we get $x(n)$, which gives us $x_{1}(n)$ and $x_{2}(n)$.

The above equations require complex arithmetic not directly supported by DSPs; thus, to implement these complex-valued equations, it is helpful to express the real and imaginary terms in real arithmetic. The forward DFT of the equations shown in (5) can be written as follows:

$$
\begin{align*}
X_{1} r(k)=\frac{1}{2}[X r(k)+X r(N-k)] \text { and } X_{1} i(k)= & \frac{1}{2}[X i(k)-X i(N-k)] \\
& k=0,1, \ldots, N-1  \tag{7}\\
X_{2} r(k)=\frac{1}{2}[X i(k)+X i(N-k)] \text { and } X_{2} i(k)= & \frac{-1}{2}[X r(k)-X r(N-k)]
\end{align*}
$$

In addition, because the DFT of real-valued sequences has the properties of complex conjugate symmetry and periodicity, the number of computations in (7) can be reduced. Using the properties, the equations in (7) can be rewritten as follows:

$$
\begin{array}{ll}
X_{1} r(0)=X_{r}(0) & X_{1} i(0)=0 \\
X_{2} r(0)=X_{i}(0) & X_{2} i(0)=0 \\
X_{1} r(N / 2)=X r(N / 2) & X_{1} i(N / 2)=0 \\
X_{2} r(N / 2)=X i(N / 2) & X_{2} i(N / 2)=0 \\
X_{1} r(k)=\frac{1}{2}\left[X r(k)+X_{r}(N-k)\right] & X_{1} i(k)=\frac{1}{2}[X i(k)-X i(N-k)]  \tag{8}\\
X_{2} r(k)=\frac{1}{2}[X i(k)+X i(N-k)] & X_{2} i(k)=\frac{-1}{2}[X r(k)-X r(N-k)] \\
X_{1} r(N-k)=X_{1} r(k) & X_{1} i(N-k)=-X_{1} i(k) \\
X_{2} r(N-k)=X_{2} r(k) & X_{2} i(N-k)=-X_{2} i(k)
\end{array}
$$

Similarly, the additional computation involved in computing the IDFT can be written as follows:

$$
\begin{array}{ll}
X_{r}(k)=X_{1} r(k)-X_{2} i(k) & \\
X_{i}(k)=X_{1} i(k)+X_{2} r(k) & \tag{9}
\end{array}
$$

See Appendix A for a detailed derivation of these equations.
Now that we have the equations for the split operation used in computing the DFT of two real-valued sequences, we turn to the following steps, which outline how to use the equations. The forward DFT is outlined as follows.

Step 1: Form the $N$-point complex-valued sequence $x(n)$ from the two $N$-length sequences $x_{1}(n)$ and $x_{2}(n)$.
for $n=0, \ldots, N-1$

$$
\begin{aligned}
& x_{\mathrm{r}}(n)=x_{1}(n) \\
& x_{\mathrm{i}}(n)=x_{2}(n)
\end{aligned}
$$

Step 2: Compute the $N$-length complex DFT of $x(n)$.

$$
X(k)=\operatorname{DFT}[x(n)]
$$

NOTE: The DFT can be any efficient DFT algorithm (such as one of the various FFT algorithms), but the output must be in normal order.

Step 3: Compute the split operation equations.

$$
\begin{array}{ll}
X_{1} r(0)=X r(0) & X_{1} i(0)=0 \\
X_{2} r(0)=X i(0) & X_{2} i(0)=0 \\
X_{1} r(N / 2)=X i(N / 2) & X_{1} i(N / 2)=0 \\
X_{2} r(N / 2)=X i(N / 2) & X_{2} i(N / 2)=0
\end{array}
$$

for $k=1, \ldots, N / 2-1$
$X_{1} r(k)=0.5^{*}[\operatorname{Xr}(k)+X r(N-k)]$
$X_{2} r(k)=0.5^{*}[X i(k)+X i(N-k)]$
$X_{1} r(N-k)=X_{1} r(k)$
$X_{2} r(N-k)=X_{2} r(k)$
$X_{1} i(k)=0.5^{*}[X i(k)-X i(N-k)]$
$X_{2} i(k)=0.5^{*}[X r(k)-X r(N-k)]$
$X_{1} i(N-k)=-X_{1} i(k)$
$X_{2} i(N-k)=-X_{2} i(k)$

For two frequency domain sequences, $X_{1}(k)$ and $X_{2}(k)$, derived from real-valued sequences, perform the following steps to take the IDFT of $X_{1}(k)$ and $X_{2}(k)$.

Step 1: Form a single complex-valued sequence $X(k)$ from $X_{1}(k)$ and $X_{2}(k)$ using the IDFT split equations.
for $\mathrm{k}=0, \ldots, \mathrm{~N}-1$

$$
\begin{aligned}
& X r(k)=X_{1} r(k)-X_{2} i(k) \\
& X i(k)=X_{1} i(k)+X_{2} r(k)
\end{aligned}
$$

Step 2: Compute the $N$-length IDFT of $X(k)$.

$$
x(n)=\operatorname{IDFT}[X(k)]
$$

As with the forward DFT, the IDFT can be any efficient IDFT algorithm. The IDFT can be computed using the forward DFT and some conjugate operations.
$x(n)=\left[\operatorname{DFT}\left\{X^{*}(k)\right\}\right]^{*}$
where:

* is the complex conjugate operator

Step 3: From $x(n)$, form $x_{1}(n)$ and $x_{2}(n)$.
for $n=0,1, \ldots . N-1$

$$
\begin{aligned}
& x_{1}(n)=x r(n) \\
& x_{2}(n)=x i(n)
\end{aligned}
$$

Appendix C contains C implementation of the outlined DFT and IDFT algorithms.

### 3.2 Efficient Computation of the DFT of a 2 N -Point Real Sequence

Assume $g(n)$ is a real-valued sequence of $2 N$ points. We outline the equations involved in obtaining the $2 N$-point DFT of $g(n)$ from the computation of one $N$-point complex-valued DFT. First, we subdivide the $2 N$-point real sequence into two $N$-point sequences as follows:

And define $x(n)$ to be the $N$-point complex-valued sequence:

$$
\begin{array}{ll}
x_{1}(n)=g(2 n) & \\
x_{2}(n)=g(2 n+1) & 0 \leq n \leq N-1
\end{array}
$$

The DFT of $g(n), G(k)$, can be computed using

$$
\begin{equation*}
x(n)=x_{1}(n)+j x_{2}(n) \quad 0 \leq n \leq N-1 \tag{11}
\end{equation*}
$$

where

$$
\begin{array}{rr}
G(k)=X(k) A(k)+X^{*}(N-k) B(k) \quad & k=0,1, \ldots, N-1 \\
& \text { with } X(N)=X(0) \\
A(k)=\frac{1}{2}\left(1-j W_{2 N}^{k}\right) \text { and } B(k)=\frac{1}{2}\left(1+j W_{2 N}^{k}\right) \tag{13}
\end{array}
$$

As you can see, we have computed the DFT of a $2 N$-point sequence from one $N$-point DFT and additional computations, which we call the split operation.

Similarly, if we have a frequency domain $2 N$-point sequence, which was derived from a real-valued sequence, we can use an $N$-point IDFT to obtain the time domain $2 N$-point real-valued sequence using the following equation:

$$
\begin{array}{ll}
X(k)=G(k) A^{*}(k)+G^{*}(N-k) B^{*}(k) \quad & k=0,1, \ldots, N-1  \tag{14}\\
& \text { with } G(N)=G(0)
\end{array}
$$

The equations shown in (12) and (14) are of the same form. Equation (14) can be obtained from equation (12) if $G(k)$ is swapped with $X(k)$, and $A(k)$ and $B(k)$ are complex conjugated. Thus, equations (12) and (14) can be implemented with one common split function.

NOTE: In implementing these equations, $A(k), A^{*}(k), B(k)$, and $B^{\star}(k)$ can be pre-computed and stored in a table. Their values can thus be obtained by table look-up as opposed to arithmetic computation. The result is a large computational savings because the sine and cosine functions required by twiddle factors do not need to be computed when performing the split. (A detailed derivation of these equations is provided in Appendix B.)

As in the previous section, Efficient Computation of the DFT of Two Real Sequences, when implementing the above equations, it is useful to express them in their real and imaginary terms.

Only $N$ points of $G(k)$ are computed in equation (12) because other $N$ points can be found using the complex conjugate property. This is applied to the following equation.

$$
\begin{array}{ll}
\operatorname{Gr}(k)=X r(k) \operatorname{Ar}(k)-X i(k) \operatorname{Ai}(k)+X r(N-k) \operatorname{Br}(k)+X i(N-k) \operatorname{Bi}(k) \\
\operatorname{Gi}(k)=X i(k) \operatorname{Ar}(k)+X r(k) \operatorname{Ai}(k)+X r(N-k) & \operatorname{Bi}(k)-X i(N-k) \operatorname{Br}(k) \\
\operatorname{Gr}(2 N-k)=\operatorname{Gr}(k) & k=0,1, \ldots, N-1 \\
\operatorname{Gi}(2 N-k)=-G i(k) & \text { with } X(N)=X(0)  \tag{15}\\
\operatorname{Gr}(N)=\operatorname{Gr}(0)-\operatorname{Gi}(0) & \\
\operatorname{Gi}(N)=0 &
\end{array}
$$

As with the forward DFT, the equations for the IDFT can be expressed in their real and imaginary terms as follows.

$$
\begin{align*}
& X r(k)=\operatorname{Gr}(k) \operatorname{Ar}(k)-G i(k)(-A i(k))+G r(N-k) \operatorname{Br}(k)+G i(N-k) \operatorname{Bi}(k) \\
& X i(k)=G i(k) \operatorname{Ar}(k)+G X r(k)(-A i(k))+G r(N-k)(-\operatorname{Bi}(k))-G i(n-k) \operatorname{Br}(k)  \tag{16}\\
& \quad k=0,1, \ldots, N-1 \\
& \text { with } G(N)=G(0)
\end{align*}
$$

Now that we have the equations for the split operation to compute the DFT of a real-valued sequence, the steps for using these equations are outlined. The forward DFT is outlined first.

Step 1: Initialize $A(k) s$ and $B(k) s$.
Real applications usually perform this only once during a power-up or initialization sequence. These values can be pre-stored in a boot ROM or computed. In either case, once they are generated, this step is no longer needed when performing the DFT. The pseudo code for generating them is given below.
for $k=0,1, \ldots, N-$

$$
\begin{aligned}
& \operatorname{Ai}(k)=-\cos (\pi k / N) \\
& \operatorname{Ar}(k)=-\sin (\pi k / N) \\
& \operatorname{Bi}(k)=\cos (\pi k / N) \\
& \operatorname{Br}(k)=\sin (\pi k / N)
\end{aligned}
$$

Step 2: Let $g(n)$ be a $2 N$-point real sequence. From $g(n)$, form the $N$-point complex-valued sequence.
$x(n)=x_{1}(n)+j x_{2}(n)$
where
$x_{1}(n)=g(2 n)$
$x_{2}(n)=g(2 n+1)$
for $n=0,1, \ldots ., N-1$

$$
\begin{aligned}
& x r(n)=g(2 n) \\
& x i(n)=g(2 n+1)
\end{aligned}
$$

Step 3: Perform an $N$-point complex FFT on the complex-valued sequence $x(n)$.
$\mathrm{X}(\mathrm{k})=\operatorname{DFT}[\mathrm{x}(\mathrm{n})]$
NOTE: The FFT can be any DFT method, such as radix-2, radix-4, mixed radix, direct implementation of the DFT, etc. However, the DFT output must be in normal order.

Step 4: Implement the split operation equations.

$$
\begin{aligned}
& X(N)=X(0) \\
& \operatorname{Gr}(N)=\operatorname{Gr}(0)-\operatorname{Gr}(0) \\
& \operatorname{Gi}(N)=0
\end{aligned}
$$

for $\mathrm{k}=0,1, \ldots, N-1$

$$
\begin{aligned}
& G r(k)=X r(k) \operatorname{Ar}(k)-X i(k) A i(k)+X r(N-k) \operatorname{Br}(k)+X i(N-k) \operatorname{Bi}(k) \\
& G i(k)=X i(k) \operatorname{Ar}(k)+X r(k) A i(k)+X r(N-k) B i(k)-X i(N-k) \operatorname{Br}(k) \\
& G r(2 N-k)=\operatorname{Gr}(k) \\
& G i(2 N-k)=-G i(k)
\end{aligned}
$$

For a $2 N$-point frequency domain sequences $G(k)$ derived from a $2 N$-point real-valued sequences, perform the following steps for the IDFT of $G(k)$.

Step 1: Initialize $A^{*}(k) \mathrm{s}$ and $B^{\star}(k) \mathrm{s}$.
As with the forward DFT, this step is usually performed only once during a power-up or initialization sequence. The values can be pre-stored in a boot ROM or computed. In either case, once the values are generated, this step is no longer needed when performing the DFT.

Because $A^{*}(k)$ and $B^{*}(k)$ are the complex conjugates of $A(k)$ and $B(k)$, respectively, each can be derived from the $A(k) \mathrm{s}$ and $B(k)$ s. The following pseudo code is used to generate them.

$$
\text { for } \begin{aligned}
k=0,1, \ldots . & , N-1 \\
A^{*} i(k) & =\cos (\pi k / N) \\
A^{*} r(k) & =1-\sin (\pi k / N) \\
B^{*} i(k) & =-\cos (\pi k / N) \\
B^{*} r(k) & =1+\sin (\pi k / N)
\end{aligned}
$$

or, if $A(k)$ and $B(k)$ are already generated, you can use the following pseudo code:

$$
\text { for } \begin{aligned}
k=0,1, \ldots . & , N-1 \\
A^{*} i(k) & =-\operatorname{Ai}(k) \\
A^{*} r(k) & =\operatorname{Ar}(k) \\
B^{* i}(k) & =-\operatorname{Bi}(k) \\
B^{*} r(k) & =\operatorname{Br}(k)
\end{aligned}
$$

Step 2: Let $G(k)$ be a $2 N$-point complex-valued sequence derived from a real-valued sequence $g(n)$.
We want to get back $g(n)$ from $G(k) \rightarrow g(n)=\operatorname{IDFT}[G(k)]$. However, we want to apply the same techniques we applied with the forward DFT, that is, use an $N$-point IFFT. This can be accomplished using the following equations.

```
\(G(N)=G(0)\)
for \(\mathrm{k}=0,1, \ldots, \mathrm{~N}-1\)
\(X r(k)=\operatorname{Gr}(k) \operatorname{Ar}(k)-\operatorname{Gi}(k)(-A i(k))+\operatorname{Gr}(N-k) \operatorname{Br}(k)+\operatorname{Gi}(N-k)(-B i(k))\)
\(X i(k)=\operatorname{Gi}(k) \operatorname{Ar}(k)+\operatorname{Gr}(k)(-A i(k))+\operatorname{Gr}(N-k)(-B i(k))-\operatorname{Gi}(N-k) \operatorname{Br}(k)\)
```

Step 3: Perform the $N$-point inverse DFT of $X(k)$.

$$
x(n)=x_{1}(n)+j x_{2}(n)=\operatorname{IDFT}[X(k)]
$$

NOTE: The IDFT can be any method but must have an output in normal order.
Step 4: $g(n)$ can then be found from $x(n)$.

$$
\text { for } \begin{aligned}
n=0, & 1, \ldots ., N \\
& g(2 n)=x_{1}(n) \\
& g(2 n+1)=x_{2}(n)
\end{aligned}
$$

Appendix C contains C implementations of the outlined DFT and IDFT algorithms.

## 4 TMS320C62x ${ }^{\text {TM }}$ Architecture and Tools Overview

Before we discuss how to efficiently implement the real-valued FFT algorithms on the C62xTM, it is helpful to take a brief look at the C62x architecture and code development tools. The TMS320C62x fixed-point processors are based on a 256 -bit advanced VLIW CPU with eight functional units, including two multipliers and six arithmetic and logic units (ALUs). The CPU can execute up to eight 32-bit instructions per cycle. With an instruction clock frequency of 200 MHz and greater, C62x peak performance starts at 1600 million instructions per second (MIPs).

The C62x processor consists of three main parts:

- CPU
- Peripherals
- Memory

Figure 1 shows a block diagram of the first device in this generation, the TMS320C6201 DSP.


Figure 1. TMS320C6201 DSP Block Diagram

TMS320C62x and C62x are trademarks of Texas Instruments.

This discussion focuses on the CPU, or core, of the device. The C62x CPU is the central building block of all TMS320C62x devices and features two data paths where processing occurs. Each data path has four functional units (.L, .S, .M, .D), along with a register file containing 16 32-bit general-purpose registers.

The C62x is a load-store architecture in which all functional units obtain operands from a register file, rather than directly from memory.

The .D units load/store data from and to memory from the register file with an address reach of 32 bits. The C62x architecture is also byte addressable: the .D units can load or store data in either 8 bits (byte), 16 bits (half-word), or 32 bits (word). In addition, the .D units can perform 32-bit addition and subtraction and address calculations.

The .M units perform multiplication, featuring a 16-bit by 16-bit multiplier that produces a 32 -bit result. Additional multiplier features include the ability to select either the 16 most significant bits (MSBs) or 16 least significant bits (LSBs) of a register operand, and optionally left-shift the multiplier output by one with saturation.

The .S units perform branches and shifting primarily, but also perform bit field operations such as extract, set and clear bit fields, as well as 32-bit logical operations and 32-bit addition and subtraction. Another advanced feature of each .S unit is the ability to split its ALU to perform two 16 -bit adds or subtracts in a single cycle.

Although the .S and .D units perform ALU functions, the .L unit is the main ALU for the CPU, performing both 32-bit and 40-bit integer arithmetic. The .L unit also features saturation logic, comparison instructions, and bit counting and can perform 32-bit logical operations. In support of the eight functional units, the CPU has a program fetch unit; instruction dispatch unit; instruction decode unit; control registers; control logic; and test, emulation and interrupt logic.

The C62x features a state of the art software development environment. A very efficient C compiler, along with a linear assembly optimizer, allows fast time to market through ease of use. Its orthogonal reduced instruction set computing (RISC)-like CPU architecture makes the C62x CPU a very good C-compiler target. Combined with TI's compiler expertise, these features make the C62x compiler the most efficient DSP compiler on the market today.

Because of its efficiency, most C62x coding can be done in C. However, as with many other DSPs, some tasks or routines require assembly coding to achieve the highest performance possible.

As a result, TI has developed a new tool called the assembly optimizer that makes assembly language coding easier and faster. The assembly optimizer allows you to write linear assembly code (no parallel instructions) without assigning registers to operands.

The assembly optimizer accepts this input syntax and generates an assembly language output that parallelizes the linear instructions and assigns registers to operands. This relieves the assembly language programmer of the following responsibilities:

- Determining which instructions can be executed in parallel
- Knowing how to position code to avoid delay slot conflicts
- Keeping track of which registers are live or free

The C62x assembler is also included in the code development tool set.

Figure 2 shows the process flow to develop code for the C62x.


Figure 2. Code Development Flow Chart
In phase 1 of the code development process, TI recommends that the algorithm first be implemented in C , which serves the following purposes:

- Provides an easy way to verify the functionality of an algorithm
- Provides a working model to verify results of optimized versions
- May meet your efficiency requirements, and thus completes your implementation

If phase 1 fails to meet your performance requirements, you may need to proceed to phase 2 to refine and optimize your C code. The process includes modifying your C code for efficiency using the C code optimization methods. This section offers a brief overview of the C-code optimization methods. For a more detailed explanation, see the TMS320C6000 Programmers Guide (SPRU198).

One of the easiest methods used to optimize your C code is the C compilers' optimizer, evoked using compiler options. Some of the most commonly used optimizer options are: $-03,-p m,-m t$, and $-x 2$. See the TMS320C6000 Optimizing C Compiler User's Guide (SPRU187) for a list of available compiler optimization options and usage.

Other methods to optimize C code for efficiency involve modifying your C code. One very effective method uses compiler intrinsics - special functions that map directly to inlined C62x instructions. Intrinsic functions typically allow you to use a C62x specific feature that is not directly expressible in C , such as .L unit saturation.

Other effective optimization techniques include:

- Loop unrolling
- Software pipelining
- Trip count specification
- Using the const keyword to eliminate memory dependencies

All of these methods produce very efficient C code. Nevertheless, the compiler still may not produce the efficiency required. In this case, phase 3 may be required. Phase 3 uses the assembly optimizer and/or the assembler to generate C62x assembly code. By far, the easiest and recommended route is the assembly optimizer. The assembly optimizer usage is outlined in detail in the TMS320C6000 Optimizing C Compiler User's Guide (SPRU187). In addition, the TMS320C6000 Assembly Language Tools User's Guide (SPRU186) outlines assembler usage.

A recommended approach for using either assembly method is to implement assembly routines as C-callable assembly functions.

NOTE: Use caution when implementing C-callable assembly routines so that you do not disrupt the C environment and cause a program to fail. The TI TMS320C6000 Optimizing C Compiler User's Guide (SPRU187) details the register, stack, calling, and return requirements of the C62x run-time environment. TI recommends that you read the material covering these requirements before implementing a C-callable assembly language function.

## 5 Implementation and Optimization of Real-Valued DFTs

Appendix C contains the source code listings for C implementation of the two efficient methods for performing the DFT of real-valued sequences outlined in this application report. Each implementation fits into phase 1 of the code development flow chart shown in Figure 2.

The primary purpose of this particular implementation is to verify the functionality of split operation algorithm implementations and provide a known good model to compare against optimized versions. Another benefit is that this implementation is generic C code and thus can be easily ported to other DSPs or CPUs featuring C compilers.

Because the primary focus of this application report is the split operations used in the efficient computation of DFTs, the C implementation is not efficient with respect to the other operations involved in the computation of real-valued DFTs. For example, the direct form of the DFT is implemented rather than a more computationally efficient FFT. Optimizing the C code to yield better performance is addressed in Appendix D .

Example 1 and Example 2 show the compiler usage for building the executable files for these implementations.

## Example 1. Efficient Computation of the DFT of a 2N-Point Real Sequence

```
cl6x -g vectors.asm realdft1.c split1.c data1.c dft.c -z -o test1.out -l
```

rts6201.lib lnk.cmd

## Example 2. Efficient Computation of the DFT of Two Real Sequences

```
cl6x -g vectors.asm realdft2.c split2.c data2.c dft.c -z -o test2.out -l
rts6201.lib lnk.cmd
```

The example compiler usage results in two executable files that can be loaded into the C62x device simulator and run:

- test1.out
- test2.out

The -g option used in the above compiler usage tells the compiler to build the code with debug information. This means that the compiler does not use the optimizer but allows the code to be easily viewed by the debugger.

First-time users of the C62x are encouraged to try different compiler options and compare the effects of each on code performance. For benchmarking code on the C62x debugger, see the TMS320C6x C Source Debugger User's Guide (SPRU188).

Appendix D contains the source code listings for optimized C implementations of the two efficient methods for performing the DFT of real-valued sequences outlined in this application report. These implementations apply to phase 2 of the code development flowchart shown in .

For this implementation, the C code is refined to yield better performance. Not all C optimization techniques outlined in this application report have been implemented. This is so that the C code remains generic and can be ported easily to other DSPs. However, you can easily apply other C62x C optimization techniques to increase performance. The following optimizations are implemented in Appendix D.

- The DFT is replaced with a radix-4 FFT, yielding a large computational savings as the number of data samples to be transformed increases. The radix-4 FFT restricts the size to a power of 4 .
- Split operation tables and FFT twiddle factors are generated using pre-generated look-up tables instead of the run-time support functions $\sin ()$ and $\cos ()$. This reduces the number of cycles required for the setup code.
- The code is organized as a series of functions to separate the independent tasks so they could be easily and independently optimized.

Example 3 and Example 4 show the compiler usage for building executable files for these implementations.

## Example 3. Efficient Computation of the DFT of a 2N-Point Real Sequence

```
cl6x -g vectors.asm datal.c digitgen.c digit.c radix4.c realdft3.c split1.c
splitgen.c -z -o test3c.out -l rts6201.lib lnk.cmd
```


## Example 4. Efficient Computation of the DFT of Two Real Sequences

```
cl6x -g vectors.asm data2.c realdft4.c split2.c radix4.c digit.c digitgen.c -z
-o test4c.out -l rts6201.lib lnk.cmd
```

The result of the above compiles is two executables:

- test3c.out
- test4c.out

The executables can be loaded into the C62x device simulator and run. The same restrictions that apply to the executables in Appendix C apply to these executables.

Appendix E contains C62x assembly language source code listings. These implementations fit into phase 3 of the code development flowchart shown in Figure 2. Each assembly listing contains a C62x C-callable assembly language function that replaces an equivalent C function shown in Appendix D. The following list includes functions implemented in assembly.
split1.asm The C-callable assembly language function that implements the split routine for the efficient computation of the DFT of two real sequences algorithm.
split2.asm The-callable assembly language function that implements the split routine for the efficient computation of the DFT of 2 N -point real sequence.
radix4.asm Replaces radix4.c. A C-callable assembly language function that implements the radix-4 FFT.
digit.asm Replaces digit.c. A C-callable assembly language function that implements the digit reversal for the radix-4 FFT

Because each of the above routines is functionally equivalent in C and assembly, no modification of other functions in Appendix $D$ is required to use them. All that must be changed to use these functions is the way in which we build the executables. Example 5 and Example 6 show how to build the executables with the assembly versions.

## Example 5. Efficient Computation of the DFT of a 2N-Point Real Sequence

```
cl6x -g vectors.asm data1.c digitgen.c digit.asm radix4.asm realdft3.c
split1.asm splitgen.c -z -o test3a.out -l rts6201.lib lnk.cmd
```


## Example 6. Efficient Computation of the DFT of Two Real Sequences

```
cl6x -g vectors.asm data2.c realdft4.c split2.asm radix4.asm digit.asm digit-
gen.c -z -o test4a.out -l rts6201.lib lnk.cmd
```

The result of the compiles shown in Example 5 and Example 6 is two executables:

- test3a.out
- test4a.out

These can be loaded into the C62x device simulator and run. The same restrictions that apply to the executables in Appendix $C$ apply to these executables.

## 6 Summary

This application report examined the theory and implementation of two efficient methods for computing the DFT of real-valued sequences. The implementation was presented in both C and C62x assembly language. As this application report reveals, a large computational savings can be achieved using these methods on real-valued sequences rather using complex-valued DFTs or FFTs. Moreover, the TMS320C62x CPU performs well when implementing these algorithms in either C or assembly.

## 7 References

1. Burrus, C.S., and Parks, T.W. DFT/FFT and Convolution Algorithms, John Wiley and Sons, New York, 1985.
2. Manolakis, D.G., and Proakis, J.G. Introduction to Digital Signal Processing, Macmillan Publishing Company, 1988.
3. Digital Signal Processing Applications with the TMS320 Family, Theory, Algorithms and Implementations, Volume 3 (SPRA017).
4. TMS320C6000 Technical Brief (SPRU197).
5. Burrus, C.S., Heideman, M.T., Jones, D.L., Sorensen, H.V. "Real-Valued Fast Fourier Transform Algorithms", IEEE Transactions on Acoustics, Speech, and Signal Processing, Vol. ASSP-35, No. 6,pp. 849-863, June 1987.
6. TMS320C6000 Programmer's Guide (SPRU198).
7. TMS320C6000 Optimizing C Compiler User's Guide (SPRU187).
8. TMS320C6000 Assembly Language Tools User's Guide (SPRU186).
9. TMS320C6x C Source Debugger User's Guide (SPRU188).

## Appendix A Derivation of Equation Used to Compute the DFT/IDFT of Two Real Sequences

This appendix provides a detailed derivation of the equations used to compute the FFT/IDFT of two real sequences using one complex DFT/IDFT.

## A. 1 Forward Transform

Assume $x_{1}(n)$ and $x_{2}(n)$ are real-valued sequences of length $N$, and let $x(n)$ be a complex-valued sequence defined as

$$
\begin{equation*}
x(n)=x_{1}(n)+j x_{2}(n) \quad 0 \leq n \leq N-1 \tag{17}
\end{equation*}
$$

The DFT operation is linear, thus the DFT of $x(n)$ may be expressed as:

$$
\begin{equation*}
X(k)=X_{1}(k)+j X_{2}(k) \quad 0 \leq k \leq N-1 \tag{18}
\end{equation*}
$$

We can express the sequences $x_{1}(n)$ and $x_{2}(n)$ in terms of $x(n)$ as follows:

$$
\begin{equation*}
x_{1}(n)=\frac{x(n)+x^{*}(n)}{2} \tag{19}
\end{equation*}
$$

where * is the complex conjugate operator

$$
x_{2}(n)=\frac{x(n)-x^{*}(n)}{2 j}
$$

The following shows that these equalities are true:

$$
\begin{align*}
& \frac{x(n)+x^{*}(n)}{2}=\frac{x_{1}(n)+j x_{2}(n)+x_{1}(n)-j x_{2}(n)}{2}=x_{1}(n) \\
& \frac{x(n)-x^{*}(n)}{2 j}=\frac{x_{1}(n)+j x_{2}(n)-x_{1}(n)+j x_{2}(n)}{2 j}=x_{2}(n) \tag{20}
\end{align*}
$$

Therefore, we can express the DFT of $x_{1}(n)$ and $x_{2}(n)$ in terms of $x(n)$ as shown below:

$$
\begin{align*}
& X_{1}(k)=\operatorname{DFT}\left[x_{1}(n)\right]=\frac{1}{2}\left\{D F T[x(n)]+\operatorname{DFT}\left[x^{*}(n)\right]\right\}  \tag{21}\\
& X_{2}(k)=\operatorname{DFT}\left[x_{2}(n)\right]=\frac{1}{2 j}\left\{\operatorname{DFT}[x(n)]-\operatorname{DFT}\left[x^{*}(n)\right]\right\}
\end{align*}
$$

From the complex property of the DFT, we know the following is true:

$$
\text { If } x(n) \stackrel{D F T}{\longleftrightarrow} X(k) \text {, then } x^{*}(n) \underset{N}{\stackrel{D F T}{\leftrightarrows}} X^{*}(N-k)
$$

Thus, we can express $X_{1}(k)$ and $X_{2}(k)$ as follows:

$$
\begin{align*}
& X_{1}(k)=\frac{1}{2}\left[x(k)+X^{*}(N-k)\right]  \tag{22}\\
& X_{2}(k)=\frac{1}{2 j}\left[X(k)-X^{*}(N-k)\right]
\end{align*}
$$

From the above equations, we can see that by performing a single DFT on the complex-valued sequence $x(n)$, we have obtained the DFT of two real-valued sequences with only a small amount of additional computation in calculating $X_{1}(k)$ and $X_{2}(k)$ from $X(k)$.

In addition, because $x_{1}(n)$ and $x_{2}(n)$ are real-valued sequences, $X_{1}(k)$ and $X_{2}(k)$ has complex conjugate symmetry $-X_{1}(N-k)=X_{1}{ }^{*}(k)$ and $X_{2}(N-k)=X_{2}{ }^{*}(k)$; thus, we only need to compute $X_{1}(k)$ and $X_{2}(k)$ for $k=0,1,2, \ldots, N / 2$.

$$
\begin{array}{ll}
X_{1}(k)=\frac{1}{2}\left\{X(k)+X^{*}(N-k)\right\} & \\
X_{1}(N-k)=X_{1}^{*}(k) & k=0,1, \ldots, N / 2 \\
& \text { with } X(N)=X(0) \\
X_{2}(k)=\frac{1}{2 j}\left\{X(k)-X^{*}(N-k)\right\} &  \tag{23}\\
X_{2}(N-k)=X_{2}^{*}(k) &
\end{array}
$$

To implement these equations, it is helpful if we express them in terms of their real and imaginary terms.

$$
\begin{array}{ll}
\begin{aligned}
X_{1}(k) & =\frac{1}{2}\{X r(k)+j X i(k)+X r(N-k)-j X i(N-k)\} \\
& =\frac{1}{2}\{(X r(k)+X r(N-k))+j(X i(k)-X i(N-k))\}
\end{aligned} \\
\text { or } & k=0,1, \ldots, N / 2 \\
X_{1} r(k)=\frac{1}{2}\{X r(k)+X r(N-k)\} & \text { with } X(N)=X(0) \\
X_{1} i(k)=\frac{1}{2}\{X i(k)-X i(N-k)\} & \tag{24}
\end{array}
$$

Similarly, it can be shown that

$$
\begin{array}{ll}
X_{2} r(k)=\frac{1}{2}\{X i(k)+X i(N-k)\} & k=0,1, \ldots, N / 2 \\
X_{2} i(k)=\frac{-1}{2}\left\{X_{r}(k)-X r(N-k)\right\} & \text { with } X(N)=X(0) \tag{25}
\end{array}
$$

There are two special cases with the above equations, $k=0$ and $k=N / 2$. For $k=0$ :

$$
\begin{align*}
& X_{1} r(0)=\frac{1}{2}\left\{X_{r}(0)+X_{r}(N)\right\} \\
& X_{1} i(0)=\frac{1}{2}\{X i(0)-X i(N)\}  \tag{26}\\
& X_{2} r(0)=\frac{1}{2}\{X i(0)+X i(N)\} \\
& X_{2} i(0)=\frac{-1}{2}\left\{X_{r}(0)-X_{r}(N)\right\}
\end{align*}
$$

Because of the periodicity property of the DFT, we know $X(k+N)=X(k)$. Therefore, $X_{r}(0)=$ $X_{r}(N)$ and $X_{i}(0)=X_{i}(N)$. Using this property, the above equations can be expressed as follows:

$$
\begin{aligned}
& x_{1} r(0)=X r(0) \\
& X_{1} i(0)=0 \\
& X_{2} r(0)=X i(0) \\
& X_{2} i(0)
\end{aligned}
$$

For $k=N / 2$ :

$$
\begin{align*}
& X_{1} r(N / 2)=\frac{1}{2}\left\{X_{r}(N / 2)+X_{r}(N / 2)\right\} \\
& X_{1} i(N / 2)=\frac{1}{2}\left\{X_{i}(N / 2)-X_{i}(N / 2)\right\}  \tag{28}\\
& X_{2} r(N / 2)=\frac{1}{2}\left\{X_{i}(N / 2)+X_{i}(N / 2)\right\} \\
& X_{2} i(N / 2)=\frac{-1}{2}\left\{X_{r}(N / 2)-X_{r}(N / 2)\right\}
\end{align*}
$$

or

$$
\begin{align*}
& X_{1} r(N / 2)=\operatorname{Xr}(N / 2) \\
& X_{1} i(N / 2)=0  \tag{29}\\
& X_{2} r(N / 2)=\operatorname{Xr}(N / 2) \\
& X_{2} i(N / 2)=0
\end{align*}
$$

Thus, (24) and (25) must be computed only for $k=1,2, \ldots N / 2-1$.

## A. 2 Inverse Transform

We can use a similar method to obtain the IDFT. We know $X_{1}(k)$ and $X_{2}(k)$. We want to express $X(k)$ in terms of $X_{1}(k)$ and $X_{2}(k)$. Recall, the relationship between $x_{1}(n), x_{2}(n)$ and $x(n)$ is $x(n)=x_{1}(n)+\mathrm{j} x_{2}(n)$. Since the DFT operator is linear, $X(k)=X_{1}(k)+\mathrm{j} X_{2}(k)$. Thus, $X(k)$ can be found by the following equations:

$$
\begin{aligned}
& X r(k)=X_{1} r(k)-X_{2} i(k) \\
& X i(k)=X_{1} i(k)+X_{2} r(k)
\end{aligned}
$$

$x(n)$ can then be found by taking the inverse transform of $X(k)$.
$x(n)=\operatorname{IDFT}[X(k)]$
From $x(n)$, we can get $x_{1}(n)$ and $x_{2}(n)$.

$$
\begin{aligned}
& X_{1}(n)=x r(n) \\
& x_{2}(n)=x i(n)
\end{aligned}
$$

## Appendix B Derivation of Equations Used to Compute the DFT/IDFT of a Real Sequence

This appendix details the derivation of the equations used to compute the DFT/IDFT of a $2 N$-length real-valued sequence using an $N$-length complex DFT/IDFT.

## B. 1 Forward Transform

Assume $g(n)$ is a real-valued sequence of $2 N$ points. The following shows how to obtain the $2 N$-point DFT of $g(n)$ using an $N$-point complex DFT.

Let

$$
\begin{align*}
& X_{1}(n)=g(2 n) \\
& X_{2}(n)=g(2 n+1) \tag{30}
\end{align*}
$$

We have subdivided a $2 N$-point real sequence into two $N$-point sequences. We now can apply the same method shown in Appendix A.
Let $x(n)$ be the $N$-point complex-valued sequence.

$$
\begin{equation*}
x(n)=x_{1}(n)+j x_{2}(n) \quad 0 \leq n \leq N-1 \tag{31}
\end{equation*}
$$

From the results shown in Appendix A, we have

$$
\begin{align*}
& X_{1}(k)=\frac{1}{2}\left\{X(k)+X^{*}(N-k)\right\}  \tag{32}\\
& X_{2}(k)=\frac{1}{2 j}\left\{X(k)-X^{*}(N-k)\right\}
\end{align*}
$$

$$
k=0,1, \ldots, N-1
$$

We now express the $2 N$-point DFT in terms of two $N$-point DFTs.

$$
\begin{align*}
&G(k)=\operatorname{DFT}[g(n)]=\operatorname{DFT}[g(2 n)+g(2 n+1)]=D F T g(2 n)]+D F T g(2 n+1)] \\
&=\sum_{n=0}^{N-1} g(2 n) W_{2 N}^{2 n k}+\sum_{n=0}^{N-1} g(2 n+1) W_{2 N}^{(2 n+1) k} \quad k=0,1  \tag{33}\\
&=\sum_{n=0}^{N-1} x_{1}(n) W_{N}^{n k}+W_{2 N}^{k} \sum_{n=0}^{N-1} x_{2}(n) W_{N}^{n k}
\end{align*}
$$

Thus,

$$
\begin{equation*}
G(k)=X_{1}(k)+W_{2 N}^{k} X_{2}(k) \quad k=0,1, \ldots, N-1 \tag{34}
\end{equation*}
$$

Using equation (32), we can express $G(k)$ in terms of $X(k)$.

$$
\begin{align*}
G(k)= & \frac{1}{2}\left\{X(k)+X^{*}(N-k)\right\}+W_{2 N}^{k} \frac{1}{2 j}\left\{X(k)-X^{*}(N-k)\right\} \\
& k=0,1, \ldots, N-1  \tag{35}\\
= & X(k)\left[\frac{1}{2}\left(1-j W_{2 N}^{k}\right)\right]+X^{*}(N-k)\left[\frac{1}{2}\left(1+j W_{2 N}^{k}\right)\right]
\end{align*}
$$

Let

$$
\begin{align*}
& A(k)=\left[\frac{1}{2}\left(1-j W_{2 N}^{k}\right)\right]  \tag{36}\\
& B(k)=\left[\frac{1}{2}\left(1+j W_{2 N}^{k}\right)\right]
\end{align*}
$$

$$
k=0,1, \ldots, N-1
$$

Thus, $G(k)$ can be expressed as follows:

$$
\begin{equation*}
G(k)=X(k) A(k)+X^{*}(N-k) B(k) \quad k=0,1, \ldots, N-1 \tag{37}
\end{equation*}
$$

Because $x(n)$ is a real-valued sequence, we know that the DFT transform results will have complex conjugate symmetry. Also, because of the periodicity property of the DFT, we know $X(k+N)=X(k)$; therefore, $X(N)=X(0)$. Using these properties, we can find the other half of the DFT result.

Thus, we have computed the DFT of a $2 N$-point real sequence using one $N$-point complex DFT and additional computations.

To implement these equations, it is helpful to express them in terms of their real and imaginary terms.

$$
\begin{array}{r}
G(k)=(X r(k)+j X i(k))(\operatorname{Ar}(k)+j A i(k))+(X r(N-k)-j X i(N-k))(\operatorname{Br}(k)+j B i(k))  \tag{38}\\
k=0,1, \ldots, N-1
\end{array}
$$

Carrying out the multiplication, separating the real and imaginary terms, and applying the periodicity and complex conjugate properties, we have the following:

$$
\begin{align*}
& \operatorname{Gr}(k)=\operatorname{Xr}(k) \operatorname{Ar}(k)-X i(k) A i(k)+X r(N-k) \operatorname{Br}(k)+X i(N-k) B i(k) \\
& k=0,1, \ldots, N-1 \\
& \text { with } X(N)=X(0) \\
& \operatorname{Gi}(k)=X i(k) \operatorname{Ar}(k)+X r(k) A i(k)+X r(N-k) B i(k)-X i(N-k) \operatorname{Br}(k) \\
& \operatorname{Gr}(k)=X r(0)-X i(0) \\
& k=N  \tag{39}\\
& G i(k)=0 \\
& \operatorname{Gr}(2 N-k)=\operatorname{Gr}(k) \\
& G i(2 N-k)=G i(k)
\end{align*}
$$

## B. 2 Inverse Transform

We will now derive the equations for the IDFT of a 2 N -point complex sequence derived from a real sequence using an $N$-point complex IDFT. We express the $N$-point complex sequence, $X(k)$, in terms of the $2 N$-point complex sequence $G(k)$. Once $X(k)$ is known, $x(n)$ can be found by taking the IDFT of $X(k)$. Once $x(n)$ is known, $g(n)$ follows.

Equation (37) can be rewritten as follows:

$$
\begin{array}{ll}
G(k)=X(k) A(k)+X^{*}(N-k) B(k) & k=0,1, \ldots, N-2  \tag{40}\\
G(N-k)=X(N-k) A(N-k)+X^{*}(k) B(N-k) & k=0,1, \ldots, N / 2-1
\end{array}
$$

where

$$
\begin{align*}
& A(N-k)=\left[\frac{1}{2}\left(1-j W_{2 N}^{N-k}\right)\right]=\left[\frac{1}{2}\left(1+j W_{2 N}^{-k}\right)\right]=A^{*}(k) \\
& B(N-k)=\left[\frac{1}{2}\left(1+j W_{2 N}^{N-k}\right)\right]=\left[\frac{1}{2}\left(1-j W_{2 N}^{-k}\right)\right]=B^{*}(k) \tag{41}
\end{align*}
$$

The above equalities can be shown to be true by recalling the following definition and substituting appropriately for $k$.

$$
\begin{equation*}
W_{2 N}^{k}=e^{-j 2 \pi k / 2 N}=\cos (2 \pi k / 2 N)-j \sin (2 \pi k / 2 N) \tag{42}
\end{equation*}
$$

We would like to make the ranges of $k$ for $G(k)$ and $G(N-k)$ the same. Look at $G(N / 2)$ :

$$
\begin{align*}
G(N / 2) & =X(N / 2) A(N / 2)+X^{\star}(N / 2) B(N / 2) \\
& =X(N / 2)\left[\frac{1}{2}\left(1-j W_{2 N}^{N / 2}\right)\right]+X^{*}(N / 2)\left[\frac{1}{2}\left(1+j W_{2 N}^{N / 2}\right)\right] \tag{43}
\end{align*}
$$

But from (42), we see that

$$
\begin{equation*}
w_{2 N}^{N / 2}=-j \tag{44}
\end{equation*}
$$

Therefore:

$$
\begin{equation*}
G(N / 2)=X^{*}(N / 2) \tag{45}
\end{equation*}
$$

Now we can express $G(k)$ and $G(N-k)$ with the same ranges of $k$, and along with (41) and (43) we have

$$
\begin{array}{ll}
G(k)=X(k) A(k)+X^{*}(N-k) B(k) & k=0,1, \ldots, N / 2-1 \\
G(N-k)=X(N-k) A^{*}(k)+X^{*}(k) B(k) & k=0,1, \ldots, N / 2-1  \tag{47}\\
G(N / 2)=X^{*}(N / 2) &
\end{array}
$$

From (46) and (47) you can see we have two equations and two unknowns, $X(k)$ and $X(N-k)$. We can use some algebra tricks to come up with equations for $X(k)$ and $X(N-k)$. If we multiply both sides of (46) with $A(k)$, complex conjugate both sides of (47), then multiply both sides by $B(k)$, we have the following:

$$
\begin{aligned}
& G(k) A(k)=X(k) A(k)+X^{*}(N-k) B(k) A(k) \\
& \frac{-G^{*}(N-k) B(k)=X(k) B(k) B(k)+X^{*}(N-k) B(k) A(k)}{G(k) A(k)-G^{*}(N-k) B(k)=X(k)\{A(k) A(k)-B(k) A(k)\}} \\
& \quad k=0,1, \ldots, N / 2-1
\end{aligned}
$$

Solving for $X(k)$ :

$$
\begin{equation*}
X(k)=\frac{G(k) A(k)-G^{*}(N-k) B(k)}{A(k) A(k)-B(k) B(k)} \quad k=0,1, \ldots, N / 2-1 \tag{49}
\end{equation*}
$$

Equation (49) can be simplified as follows:

$$
\begin{gather*}
A(k) A(k)=\left[\frac{1}{2}\left(1-j W_{2 N}^{k}\right)\right]\left[\frac{1}{2}\left(1-j W_{2 N}^{k}\right)\right]=\frac{1}{4}\left[1-2 j W_{2 N}^{k}-W_{2 N}^{k}\right]  \tag{50}\\
B(k) B(k)=\left[\frac{1}{2}\left(1+j W_{2 N}^{k}\right)\right]\left[\frac{1}{2}\left(1+j W_{2 N}^{k}\right)\right]=\frac{1}{4}\left[1+2 j W_{2 N}^{k}-W_{2 N}^{k}\right]  \tag{51}\\
A(k) k A(k)-B(k) B(k)=-j W_{2 N}^{k} \\
X(k)=\frac{G(k) A(k)-G^{*}(N-k) B(k)}{-j W_{2 N}^{k}} \quad k=0,1, \ldots, N / 2-1 \tag{52}
\end{gather*}
$$

Similarly, if we multiply both sides of (47) with $A^{*}(k)$, conjugate both sides of (46), then multiply both sides by $B^{\star}(k)$, we have the following:

$$
\begin{align*}
& G^{*}(k) B^{*}(k)=X^{*}(k) A^{*}(k) B^{*}(k)+X(N-k) B^{*}(k) B^{*}(k) \\
& \frac{-G^{*}(N-k) A^{*}(k)=X^{*}(k) A^{*}(k) B^{*}(k)+X(N-k) A^{*}(k) A^{*}(k)}{G^{*}(k) B^{*}(k)-G(N-k) A^{*}(k)=X(N-k)\left\{A^{*}(k) A^{*}(k)-B^{*}(k) B^{*}(k)\right\}}  \tag{53}\\
& k=0,1, \ldots, N / 2-1
\end{align*}
$$

Solving for $X(N-k)$ :

$$
\begin{equation*}
X(N-k)=\frac{G^{*}(k) B^{*}(k)-G(N-k) A^{*}(k)}{A^{*}(k) A^{*}(k)-B^{*}(k) B^{*}(k)} \quad k=0,1, \ldots, N / 2-1 \tag{54}
\end{equation*}
$$

Equation (54) can be simplified as follows:

$$
\begin{align*}
& A^{*}(k) A^{*}(k)=\left[\frac{1}{2}\left(1+j W_{2 N}^{-k}\right)\right]\left[\frac{1}{2}\left(1+j W_{2 N}^{-k}\right)\right]=\frac{1}{4}\left[1+2 j W_{2 N}^{-k}-W_{2 N}^{-k}\right]  \tag{55}\\
& B^{*}(k) B^{*}(k)=\left[\frac{1}{2}\left(1-j W_{2 N}^{-k}\right)\right]\left[\frac{1}{2}\left(1-j W_{2 N}^{-k}\right)\right]=\frac{1}{4}\left[1-2 j W_{2 N}^{-k}-W_{2 N}^{-k}\right]  \tag{56}\\
& A^{*}(k) A^{*}(k)-B^{*}(k) B^{*}(k)=j W_{2 N}^{-k}  \tag{57}\\
& X(N-k)=\frac{G^{*}(k) B^{*}(k)-G(N-k) A^{*}(k)}{-j W_{2 N}^{-k}} \quad k=0,1, \ldots, N / 2-1 \tag{58}
\end{align*}
$$

It can be shown that

$$
\begin{align*}
& \frac{A(k)}{-j W_{2 N}^{k}}=A^{*}(k) \quad \text { and } \quad \frac{B(k)}{-j W_{2 N}^{k}}=B(k)  \tag{59}\\
& \frac{A^{*}(k)}{j W_{2 N}^{k}}=A^{*}(k) \quad \text { and } \quad \frac{B^{*}(k)}{j W_{2 N}^{k}}=B(k) \tag{60}
\end{align*}
$$

Making these substitutions, we get:

$$
\begin{align*}
& X(k)=G(k) A^{*}(k)+G^{*}(N-k) B^{*}(k) \\
& X(N-k)=G^{*}(k) B(k)+G(N-k) A(k)  \tag{61}\\
& X(N / 2)=G^{*}(N / 2)
\end{align*}
$$

For equation (61), if we make the following substitutions, along with replacing $k$ with $N / 2-k$,

$$
\begin{equation*}
A(N-k)=A^{*}(k) \quad \text { and } \quad B(N-k)=B^{*}(k) \tag{62}
\end{equation*}
$$

we can see that $X(k)$ can be expressed as a single equation.

$$
\begin{array}{ll}
X(k)=G(k) A^{*}(k)+G^{*}(N-k) B^{*}(k) \quad & k=0,1, \ldots, N-1  \tag{63}\\
& G(N)=G(0)
\end{array}
$$

Now, in terms of implementing these equations, it is helpful to express them in terms of their real and imaginary terms.

$$
\begin{array}{r}
X(k)=(\operatorname{Gr}(k)+j G i(k))(A r(k)-j A i(k))+(\operatorname{Gr}(N-k)-j G i(N-k))(B r(k)-j B i(k)) \\
k=0,1, \ldots, N-1  \tag{64}\\
\text { with } G(N)=G(0)
\end{array}
$$

Carrying out the multiplication and separating the real and imaginary terms, we have the following:

$$
\begin{aligned}
X_{r}(k)=\operatorname{Gr}(k) \operatorname{Ar}(k)+G i(k) \operatorname{Ai}(k)+\operatorname{Gr}(N-k) \operatorname{Br}(k)-G i(N-k) & B i(k) \\
& \\
& k=0,1, \ldots, N-1 \\
& \text { with } G(N)=G(0)
\end{aligned}
$$

$$
X i(k)=\operatorname{Gi}(k) \operatorname{Ar}(k)-\operatorname{Gr}(k) \operatorname{Ai}(k)-\operatorname{Gr}(N-k) \operatorname{Bi}(k)-\operatorname{Gi}(N-k) \operatorname{Br}(k)
$$

Now we have formed the complex sequence with which we can use an $N$-point complex DFT to obtain $x(n)$, which we then can use to get $g(n)$.

$$
\begin{array}{ll}
x(n)=x r(n)+j x i(n)=\operatorname{IDFT}[X(k)] & \\
g(2 n)=x r(n) & n=0,1, \ldots, N-1 \\
g(2 n+1)=x i(n) &
\end{array}
$$

## Appendix C C Implementations of the DFT of Real Sequences

This appendix contains C implementations of the efficient methods for performing the DFT of real-valued sequences.

## C. 1 Implementation Notes

The following lists usage, assumptions, and limitations of the code.

| Data format | All data and state variables are 16 -bit signed integers (shorts). In this example, the decimal point is assumed to be between bits 15 and 14, thus the Q15 data format. For complex data and variables, the real and imaginary components are both Q15 numbers. From this data format, you can see that this code was developed for a fixed-point processor. |
| :---: | :---: |
| Memory | Complex data is stored in memory in imaginary/real pairs. The imaginary component is stored in the most significant halfword ( 16 bits) and the real component is stored in the least significant halfword, unless otherwise noted. |
| Endianess | The code is presented and tested in little endian format. Some modification to the code is necessary for big endian format. |
| Overflow | No overflow protection or detection is performed. |
| File | Description |
| realdft1.c | DFT of a 2 N -point real sequence main program |
| split1.c | Split function for the DFT of a $2 N$-point real sequence |
| data1.c | Sample data |
| params1.h | Header file, for example |
| realdft2.c | DFT of a two N -point real sequence main program |
| split2.c | Split function for the DFT of two N -point real sequence |
| data2.c | Sample data |
| params2.h | Header file, for example |
| dft.c | Direct implementation of the DFT function |
| params.h | Header file |
| vectors.asm | Reset vector assembly source |
| Ink.cmd | Example linker command file |

## Example C-1. realdft1.c File

```
/**********************************************************************************
    FILE
    realdft1.c - C source for an example implementation of the DFT/IDFT
    of a 2N-point real sequences using one N-point complex DFT/IDFT.
```


## Description

This program is an example implementation of an efficient way of computing the DFT/IDFT of a real-valued sequence.

In many applications, the input is a sequence of real numbers. If this condition is taken into consideration, additional computational savings can be achieved because the FFT of a real sequence has some symmetrical properties. The DFT of a 2 N -point real sequence can be efficiently computed using a N -point complex DFT and some additional computations.

The following steps are required in the computation of the FFT of a real-valued sequence using the split function:

1. Let $g(n)$ be a $2 N$-point real sequence. From $g(n)$, form the the $N$-point complex valued sequence, $x(n)=x 1(n)+j x 2(n)$, where $x 1(n)=g(2 n)$ and $x 2(n)=g(2 n+1)$.
2. Perform an N-point complex FFT on the complex valued sequence $x(n)->X(k)=$ DFT $\{x(\mathrm{n})\}$. Note that the FFT can be any DFT method, such as radix-2, radix-4, mixed radix, direct implementation of the DFT, etc. However, the DFT output must be in normal order.
3. The following additional computation are used to get $\mathrm{G}(\mathrm{k})$ from

$$
\begin{aligned}
& X(k) \operatorname{Gr}(\mathrm{k})=\mathrm{Xr}(\mathrm{k}) \operatorname{Ar}(\mathrm{k})-\mathrm{Xi}(\mathrm{k}) \operatorname{Ai}(\mathrm{k})+\mathrm{Xr}(\mathrm{~N}-\mathrm{k}) \operatorname{Br}(\mathrm{k})+\mathrm{Xi}(\mathrm{~N}-\mathrm{k}) \operatorname{Bi}(\mathrm{k}) \\
& \mathrm{k}=0,1, \ldots, \mathrm{~N}-1 \\
& \text { and } \mathrm{X}(\mathrm{~N})=\mathrm{X}(0) \\
& \mathrm{Gi}(\mathrm{k})=\mathrm{Xi}(\mathrm{k}) \operatorname{Ar}(\mathrm{k})+\mathrm{Xr}(\mathrm{k}) \operatorname{Ai}(\mathrm{k})+\mathrm{Xr}(\mathrm{~N}-\mathrm{k}) \operatorname{Bi}(\mathrm{k})-\mathrm{Xi}(\mathrm{~N}-\mathrm{k}) \operatorname{Br}(\mathrm{k})
\end{aligned}
$$

Note that only N -points of the 2 N -point sequence of $\mathrm{G}(\mathrm{k})$ are computed in the above equations. Because the DFT of a real-sequence has symmetric properties, we can easily compute the remaining $N$ points of $G(k)$ with the following equations.

$$
\begin{aligned}
& \operatorname{Gr}(\mathrm{N})=\operatorname{Xr}(0)-\mathrm{Xi}(0) \\
& \operatorname{Gi}(\mathrm{N})=0 \\
& \operatorname{Gr}(2 N-\mathrm{N})=\operatorname{Gr}(\mathrm{k}) \\
& \mathrm{Gi}(2 N-k)=-\operatorname{Gi}(\mathrm{k}
\end{aligned} \quad \mathrm{k}=1,2, \ldots, \mathrm{~N}-1 .
$$

As you can see, the above equations assume that $A(k)$ and $B(k)$, which are sine and cosine coefficients, are pre-computed. The C-code can be used to initialize $A(k)$ and $B(k)$.

```
for(k=0; k<N; k++)
{
    A[k].imag = (short)(16383.0*(-cos(2*PI/(double) (2*N)*(double)k)));
    A[k].real = (short)(16383.0*(1.0 - sin(2*PI/(double)(2*N)*(double)k)));
    B[k].imag = (short) (16383.0*(cos(2*PI/(double) (2*N)*(double)k)));
    B[k].real = (short) (16383.0*(1.0 + sin(2*PI/(double) (2*N)*(double)k)));
}
```

The following steps are required in the computation of the IFFT of a complex valued frequency domain sequence that was derived from a real sequence:

1. Let $G(k)$ be a $2 N$-point complex valued sequence derived from a real valued sequence $g(n)$. We want to get back $g(n)$ from $G(k) \rightarrow g(n)=\operatorname{IDFT}\{G(k)\}$. However, we want to apply the same techniques as we did with the forward FFT. Use a N-point IFFT. This can be accomplished by the following equations.
2. Perform the $N$-point inverse DFT of $X(k) \rightarrow x(n)=x 1(n)+j x 2(n)=\operatorname{IDFT}\{X(k)\}$. Note that the IDFT can be any method, but must have an output that is in normal order.
3. $g(n)$ can then be found from $x(n)$.
$g(2 n)=x 1(n)$

$$
\mathrm{n}=0,1, \ldots, \mathrm{~N}-1
$$

$$
g(2 n+1)=x 2(n)
$$

As you can see, the above equations can be used for both the forward and inverse FFTs, however, the pre-computed coefficients are slightly different. The following C-code can be used to initialize $\mathrm{IA}(\mathrm{k})$ and $\mathrm{IB}(\mathrm{k})$.

```
for(k=0; k<N; k++)
{
    IA[k].imag = - (short) (16383.0*(-cos(2*PI/(double) (2*N)*(double)k)));
    IA[k].real = (short) (16383.0*(1.0 - sin(2*PI/(double) (2*N)*(double)k)));
    IB[k].imag = - (short) (16383.0*(cos(2*PI/(double) (2*N)*(double)k)));
    IB[k].real = (short) (16383.0*(1.0 + sin(2*PI/(double) (2*N)*(double)k)));
    }
Note, IA(k) is the complex conjugate of A(k) and IB(k) is the complex conjugate of
B(k).
```



```
#include <math.h>
#include "params1.h"
#include "params.h"
extern short g[];
void dft(int, COMPLEX *);
void split(int, COMPLEX *, COMPLEX *, COMPLEX *, COMPLEX *);
main()
{
    int n, k;
    COMPLEX x[NUMPOINTS+1]; /* array of complex DFT data */
    COMPLEX A[NUMPOINTS]; /* array of complex A coefficients */
    COMPLEX B[NUMPOINTS]; /* array of complex B coefficients */
    COMPLEX IA[NUMPOINTS]; /* array of complex A* coefficients */
    COMPLEX IB[NUMPOINTS]; /* array of complex B* coefficients */
    COMPLEX G[2*NUMPOINTS]; /* array of complex DFT result */
```

$$
\begin{aligned}
& \operatorname{Xr}(\mathrm{k})=\operatorname{Gr}(\mathrm{k}) \operatorname{IAr}(\mathrm{k})-\operatorname{Gi}(\mathrm{k}) \operatorname{IAi}(\mathrm{k})+\operatorname{Gr}(\mathrm{N}-\mathrm{k}) \operatorname{IBr}(\mathrm{k})+\mathrm{Gi}(\mathrm{~N}-\mathrm{k}) \operatorname{IBi}(\mathrm{k}) \\
& \mathrm{k}=0,1, \ldots, \mathrm{~N}-1 \\
& \text { and } G(N)=G(0) \\
& X i(k)=\operatorname{Gi}(k) \operatorname{IAr}(k)+\operatorname{Gr}(k) \operatorname{IAi}(k)+\operatorname{Gr}(N-k) \operatorname{IBi}(k)-\operatorname{Gi}(N-k) \operatorname{IBr}(k)
\end{aligned}
$$

```
/* Initialize A,B, IA, and IB arrays */
```

```
    for(k=0; k<NUMPOINTS; k++)
    {
    A[k].imag = (short) (16383.0*(-cos (2*PI/ (double)(2*NUMPOINTS)* (double)k)));
    A[k].real = (short) (16383.0*(1.0 - sin(2*PI/(double) (2*NUMPOINTS)* (double)k)));
    B[k].imag = (short) (16383.0*(cos (2*PI/(double) (2*NUMPOINTS)* (double)k)));
    B[k].real = (short) (16383.0*(1.0 + sin(2*PI/(double) (2*NUMPOINTS)* (double)k)));
    IA[k].imag = -A[k].imag;
    IA[k].real = A[k].real;
    IB[k].imag = -B[k].imag;
    IB[k].real = B[k].real;
    }
```

/* Forward DFT */
/* From the $2 N$ point real sequence, $g(n)$, for the $N$-point complex sequence, $x(n) * /$
for ( $n=0 ; n<N U M P O I N T S ; ~ n++$ )
\{
$\mathrm{x}[\mathrm{n}] . \operatorname{imag}=\mathrm{g}[2 * \mathrm{n}+1] ; / * \mathrm{x} 2(\mathrm{n})=\mathrm{g}(2 \mathrm{n}+1) * /$
$x[n] . r e a l=g[2 * n] ; \quad / * x 1(n)=g(2 n) \quad * /$
\}
/* Compute the DFT of $x(n)$ to get $X(k)->X(k)=\operatorname{DFT}\{x(n)\} \quad * /$
dft (NUMPOINTS, x) ;
/* Because of the periodicity property of the DFT, we know that $X(N+k)=X(k) . ~ * /$
$x[N U M P O I N T S] . r e a l=x[0] . r e a l ;$
$\mathrm{x}[\mathrm{NUMPOINTS}] . i m a g=x[0] . i m a g ;$
/* The split function performs the additional computations required to get
G(k) from X(k). */
split (NUMPOINTS, $x, A, B, G)$;
/* Use complex conjugate symmetry properties to get the rest of $G(k)$ */
$G[N U M P O I N T S] . r e a l=x[0] . r e a l-x[0] . i m a g ;$
$\mathrm{G}[$ NUMPOINTS $]$. imag $=0$;
for $(k=1 ; k<N U M P O I N T S ; ~ k++)$
\{
$G[2 * N U M P O I N T S-k] . r e a l=G[k]$.real;
$G[2 * N U M P O I N T S-k]$.imag $=-G[k]$.imag;
\}
/* Inverse DFT - We now want to get back g(n). */

```
/* The split function performs the additional computations required to get
    X(k) from G(k). */
        split(NUMPOINTS, G, IA, IB, x);
/* Take the inverse DFT of X(k) to get x(n). Note the inverse DFT could be any
    IDFT implementation, such as an IFFT. */
/* The inverse DFT can be calculated by using the forward DFT algorithm directly by
complex conjugation - x(n) = (1/N) (DFT{X*(k)})*, where * is the complex conjugate
operator. */
/* Compute the complex conjugate of X(k). */
    for (k=0; k<NUMPOINTS; k++)
    {
        x[k].imag = -x[k].imag; /* complex conjugate X(k) */
    }
/* Compute the DFT of X*(k). */
    dft(NUMPOINTS, x);
/* Complex conjugate the output of the DFT and divide by N to get x(n). */
    for (n=0; n<NUMPOINTS; n++)
        {
            x[n].real = x[n].real/16;
            x[n].imag = (-x[n].imag)/16;
        }
/* g(2n) = xr(n) and g(2n + 1) = xi(n) */
        for (n=0; n<NUMPOINTS; n++)
        {
            g[2*n] = x[n].real;
            g[2*n + 1] = x[n].imag;
    }
    return(0);
}
```

Example C-2. split1.c File

```
FILE
    split1.c - This is the C source code for the implementation of the
    split routine, which is the additional computation in computing the
    DFT of an 2N-point real-valued sequences using a N-point complex DFT.
```


## Description

Computation of the DFT of 2N-point real-valued sequences can be efficiently computed using one N-point complex DFT and some additional computations. This function implements these additional computations, which are shown below.

$$
\begin{array}{r}
\operatorname{Gr}(\mathrm{k})=\mathrm{Xr}(\mathrm{k}) \operatorname{Ar}(\mathrm{k})-\mathrm{Xi}(\mathrm{k}) \operatorname{Ai}(\mathrm{k})+\mathrm{Xr}(\mathrm{~N}-\mathrm{k}) \operatorname{Br}(\mathrm{k})+\mathrm{Xi}(\mathrm{~N}-\mathrm{k}) \mathrm{Bi}(\mathrm{k}) \\
\mathrm{k}=0,1, \ldots, \mathrm{~N}-1 \\
\text { and } \mathrm{X}(\mathrm{~N})=\mathrm{X}(0) \\
\mathrm{Gi}(\mathrm{k})=\mathrm{Xi}(\mathrm{k}) \operatorname{Ar}(\mathrm{k})+\mathrm{Xr}(\mathrm{k}) \operatorname{Ai}(\mathrm{k})+\mathrm{Xr}(\mathrm{~N}-\mathrm{k}) \operatorname{Bi}(\mathrm{k})-\mathrm{Xi}(\mathrm{~N}-\mathrm{k}) \operatorname{Br}(\mathrm{k})
\end{array}
$$

```
*****************************************************************************/
#include "params1.h"
#include "params.h"
void split(int N, COMPLEX *X, COMPLEX *A, COMPLEX *B, COMPLEX *G)
{
    int k;
    int Tr, Ti;
    for (k=0; k<N; k++)
    {
        Tr = (int)X[k].real * (int)A[k].real - (int)X[k].imag * (int)A[k].imag +
        (int)X[N-k].real * (int)B[k].real + (int)X[N-k].imag * (int)B[k].imag;
        G[k].real = (short) (Tr>>15);
        Ti = (int)X[k].imag * (int)A[k].real + (int)X[k].real * (int)A[k].imag +
        (int)X[N-k].real * (int)B[k].imag - (int)X[N-k].imag * (int)B[k].real;
        G[k].imag = (short) (Ti>>15);
    }
}
```


## Example C-3. data1.c File

```
/**********************************************************************************
    FILE
        datal.c - Sample data used in realdftl.c
/* array of real-valued input sequence, g(n) */
short g[] = {255, -35, 255, -35, 255, 255, 255, 255,
    255, 255, 255, 20, 255, 255, 255, 255,
    0, 0, 0, 0, 0, 0, 0, 0,
    0, 0, 0, 0, 0, 0, 0, 0};
```


## Example C-4. params1.h File

```
/**********************************************************************************
    FILE
    params1.h - This is the C header file for example real FFT
    implementations.
#define NUMDATA 32 /* number of real data samples */
#define NUMPOINTS NUMDATA/2 /* number of point in the DFT */
```


## Example C-5. realdft2.c File

```
/**************************************************************************************
    FILE
        realdft2.c - C source for an example implementation of the DFT/IDFT of two
        N-point real sequences using one N-point complex DFT/IDFT.
```


## Description

This program is an example implementation of an efficient way of computing the DFT/IDFT of two real-valued sequences.
Assume we have two real-valued sequences of length $N-x 1[n]$ and $x 2[n]$. The DFT of $x 1[n]$ and $x 2[n]$ can be computed with one complex-valued DFT of length N, as shown above, by following this algorithm.

1. Form the complex-valued sequence $x[n]$ from $x 1[n]$ and $x 2[n]$
$x r[n]=x 1[n]$ and $x i[n]=x 2[n], \quad 0,1, \ldots, N-1$
Note, if the sequences $x 1[n]$ and $x 2[n]$ are coming from another algorithm or a data acquisition driver, this step may be eliminated if these put the data in the complex-valued format correctly.
2. Compute $\mathrm{X}[\mathrm{k}]=\mathrm{DFT}\{x[\mathrm{n}]\}$

This can be the direct-form DFT algorithm or an FFT algorithm. If using an FFT algorithm, make sure the output is in normal order - bit reversal is performed.
3. Compute the following equations to get the DFTs of $x 1[n]$ and $x 2[n]$.

$$
\begin{aligned}
& \mathrm{X} 1 \mathrm{r}[0]=\mathrm{Xr}[0] \\
& \mathrm{X} 1 \mathrm{i}[0]=0 \\
& \mathrm{X} 2 \mathrm{r}[0]=\mathrm{Xi}[0] \\
& \mathrm{X} 2 \mathrm{i}[0]=0 \\
& \mathrm{X} 1 \mathrm{r}[\mathrm{~N} / 2]=\mathrm{Xr}[\mathrm{~N} / 2] \\
& \mathrm{X} 1 \mathrm{i}[\mathrm{~N} / 2]=0 \\
& \mathrm{X} 2 \mathrm{r}[\mathrm{~N} / 2]=\mathrm{Xi}[\mathrm{~N} / 2] \\
& \text { X2i[ } \mathrm{N} / 2]=0 \\
& \text { for } k=1,2,3, \ldots ., N / 2-1 \\
& \mathrm{X} 1 \mathrm{r}[\mathrm{k}]=(\mathrm{Xr}[\mathrm{k}]+\mathrm{Xr}[\mathrm{~N}-\mathrm{k}]) / 2 \\
& \mathrm{X} 1 \mathrm{i}[\mathrm{k}]=(\mathrm{Xi}[\mathrm{k}]-\mathrm{Xi}[\mathrm{~N}-\mathrm{k}]) / 2 \\
& \mathrm{X} 1 \mathrm{r}[\mathrm{~N}-\mathrm{k}]=\mathrm{X} 1 \mathrm{r}[\mathrm{k}] \\
& \mathrm{X} 1 \mathrm{i}[\mathrm{~N}-\mathrm{k}]=\mathrm{X} 1 \mathrm{i}[\mathrm{k}] \\
& \mathrm{X} 2 \mathrm{r}[\mathrm{k}]=(\mathrm{Xi}[\mathrm{k}]+\mathrm{Xi}[\mathrm{~N}-\mathrm{k}]) / 2 \\
& \mathrm{X} 2[\mathrm{k}]=(\mathrm{Xr}[\mathrm{~N}-\mathrm{k}]-\mathrm{Xr}[\mathrm{k}]) / 2 \\
& \mathrm{X} 2 \mathrm{r}[\mathrm{~N}-\mathrm{k}]=\mathrm{X} 2 \mathrm{r}[\mathrm{k}] \\
& \text { X2i }[\mathrm{N}-\mathrm{k}]=\mathrm{X} 2 \mathrm{i}[\mathrm{k}]
\end{aligned}
$$

4. Form $\mathrm{X}[\mathrm{k}]$ from $\mathrm{X} 1[\mathrm{k}]$ and $\mathrm{X} 2[\mathrm{k}]$

$$
\text { for } \begin{aligned}
\mathrm{k}= & 0,1, \ldots, \mathrm{~N}-1 \\
& \mathrm{Xr}[\mathrm{k}]=\mathrm{X} 1 \mathrm{r}[\mathrm{k}]-\mathrm{X} 2 \mathrm{i}[\mathrm{k}] \\
& \mathrm{Xi}[\mathrm{k}]=\mathrm{X} 1 \mathrm{i}[\mathrm{k}]+\mathrm{X} 2 \mathrm{r}[\mathrm{k}]
\end{aligned}
$$

5. Compute $\mathrm{x}[\mathrm{n}]=\operatorname{IDFT}\{\mathrm{X}[\mathrm{k}]\}$

This can be the direct form IDFT algorithm, or an IFFT algorithm. If using an IFFT algorithm, make sure the output is in normal order - bit reversal is performed

```
#include <math.h> /* include the C RTS math library */
#include "params2.h" /* include file with parameters */
#include "params.h" /* include file with parameters */
extern short x1[];
extern short x2[];
void dft(int, COMPLEX *);
extern void split2(int, COMPLEX *, COMPLEX *, COMPLEX *);
main()
{
    int n, k;
    COMPLEX X1[NUMDATA]; /* array of real-valued DFT output sequence, X1(k) */
    COMPLEX X2[NUMDATA]; /* array of real-valued DFT output sequence, X2(k) */
    COMPLEX x[NUMPOINTS+1]; /* array of complex DFT data, X(k) */
/* Forward DFT */
/* From the two N-point real sequences, x1(n) and x2(n), form the N-point complex
    sequence, x(n) = x1(n) + jx2(n) */
        for (n=0; n<NUMDATA; n++)
        {
            x[n].real = x1[n];
            x[n].imag = x2[n];
        }
/* Compute the DFT of x (n), X(k) = DFT{x(n)}. Note, the DFT can be any
    DFT implementation such as FFTs. */
        dft(NUMPOINTS, x);
/* Because of the periodicity property of the DFT, we know that X(N+k)=X(k). */
    x[NUMPOINTS].real = x[0].real;
    x[NUMPOINTS].imag = x[0].imag;
```

```
/* The split function performs the additional computations required to get
    X1(k) and X2(k) from X(k). */
        split2(NUMPOINTS, x, X1, X2);
/* Inverse DFT - We now want to get back x1(n) and x2(n) from X1(k) and X2(k) using
    one complex DFT */
/* Recall that }x(n)=x1(n) + jx2(n). Since the DFT operator is linear
    X(k) = X1(k) + jX2(k). Thus we can express X(k) in terms of X1(k) and X2(k). */
        for (k=0; k<NUMPOINTS; k++)
        {
            x[k].real = X1[k].real - X2[k].imag;
            x[k].imag = X1[k].imag + X2[k].real;
            }
/* Take the inverse DFT of X(k) to get x(n). Note the inverse DFT could be any
    IDFT implementation, such as an IFFT. */
/* The inverse DFT can be calculated by using the forward DFT algorithm directly
    by complex conjugation - x(n) = (1/N) (DFT{X*(k)})*, where * is the complex
    conjugate operator. */
/* Compute the complex conjugate of X(k). */
    for (k=0; k<NUMPOINTS; k++)
    {
        x[k].imag = -x[k].imag;
    }
/* Compute the DFT of X*(k). */
    dft(NUMPOINTS, x);
/* Complex conjugate the output of the DFT and divide by N to get x(n). */
    for (n=0; n<NUMPOINTS; n++)
    {
        x[n].real = x[n].real/16;
        x[n].imag = (-x[n].imag)/16;
    }
/* x1(n) is the real part of x(n), and x2(n) is the imaginary part of x(n). */
    for (n=0; n<NUMDATA; n++)
    {
        x1[n] = x[n].real;
        x2[n] = x[n].imag;
    }
    return(0);
}
```

INSTRUMENTS

## Example C-6. split2.c File

```
/******************************************************************************
    FILE
        split2.c - This is the C source code for the implementation of the
        split routine, which is the additional computations in computing the
        DFT of two N-point real-valued sequences using one N-point complex DFT.
```


## Description

Computation of the DFT of two N-point real-valued sequences can be efficiently computed using one N-point complex DFT and some additional computations. This function implements these additional computations, which are shown below.

```
X1r[0] = Xr[0]
X1i[0] = 0
X2r[0] = Xi[0]
X2i[0] = 0
X1r[N/2] = Xr[N/2]
X1i[N/2] = 0
X2r[N/2] = Xi[N/2]
X2i[N/2] = 0
for k = 1,2,3, ...., N/2-1
        X1r[k] = (Xr[k] + Xr[N-k])/2
        X1i[k] = (Xi[k] - Xi[N-k])/2
        X1r[N-k] = X1r[k]
        X1i[N-k] = X1i[k]
        X2r[k] = (Xi[k] + Xi[N-k])/2
        X2i[k] = (Xr[N-k] - Xr[k])/2
        X2r[N-k] = X2r[k]
        X2i[N-k] = X2i[k]
```

    ***************************************************************************************)
    \#include "params.h"
void split2(int $N$, COMPLEX *X, COMPLEX *X1, COMPLEX *X2)
\{
int k;
X1[0].real $=$ X[0].real;
X1[0].imag = 0;
$\mathrm{X} 2[0]$. real $=\mathrm{X}[0]$. imag;
$\mathrm{X} 2[0]$. imag $=0$;
X1[N/2].real $=X[N / 2] . r e a l ;$
$\mathrm{X} 1[\mathrm{~N} / 2]$. imag $=0$;
$\mathrm{X} 2[\mathrm{~N} / 2]$. real $=\mathrm{X}[\mathrm{N} / 2]$. imag;
X2[N/2].imag $=0$;
for ( $k=1$; $k<N / 2 ; k++$ )
\{
X1[k].real $=(X[k] . r e a l+X[N-k] . r e a l) / 2 ;$

```
        X1[k].imag = (X[k].imag - X[N-k].imag)/2;
        X2[k].real = (X[k].imag + X[N-k].imag)/2;
        X2[k].imag = (X[N-k].real - X[k].real)/2;
        X1[N-k].real = X1[k].real;
        X1[N-k].imag = -X1[k].imag;
        X2[N-k].real = X2[k].real;
        X2[N-k].imag = -X2[k].imag;
    }
}
```

Example C-7. data2.c File

```
/*************************************************************************************
    FILE
    data2.c - Sample data used in realdft2.c
***************************************************************************************
/* array of real-valued input sequence, x1(n) */
short x1[] = {255, 255, 255, 255, 255, 255, 255, 255,
    0, 0, 0, 0, 0, 0, 0, 0};
/* array of real-valued input sequence, x2(n) */
short x2[] = {-35, -35, -35, -35, -35, -35, -35, -35,
    0, 0, 0, 0, 0, 0, 0, 0};
```


## Example C-8. params2.h File

```
/**************************************************************************************
    FILE
        params2.h - This is the C header file for example real FFT
        implementations.
***************************************************************************************/
\begin{tabular}{llll} 
\#define & NUMDATA & 16 & /* number of real data samples */ \\
\#define & NUMPOINTS & NUMDATA /* number of point in the DFT */
\end{tabular}
```


## Example C-9. dft.c File

```
/******************************************************************************
    FILE
        dft.c - This is the C source code for the direct implementation of the
    Discrete Fourier Transform (DFT) algorithm.
```


## Description

This function computes the DFT of an N-length complex-valued sequence. Note, N cannot exceed 1024 without modification to this code.
The $N$ point DFT of a finite-duration sequence $x(n)$ of length $L<-N$ is defined as

$$
X(k)={\underset{\sim}{\text { SUM }}}_{\mathrm{n}=0}^{\mathrm{N}-1} \mathrm{x}(\mathrm{n}) \quad * \exp (-j 2 \operatorname{pikn} / \mathrm{N}) \quad \mathrm{k}=0,1,2, \ldots, \mathrm{~N}-1
$$

It is always helpful to express the above equation in its real and imaginary terms for implementation.

```
    exp(-j2*pi*n*k/N) = cos(2*pi*n*k/N) - jsin(2*pi*n*k/N) -> several
    identities used here
            e(jb) = cos(b) + j sin(b)
            e(-jb) = cos(-b) + j sin(-b)
            cos(-b) = cos(b) and sin(-b) = - sin(b)
            e(-jb) = cos(b) - j sin(b)
        N-1
    X(k) = SUM {[xr(n) + j xi(n)][cos(2*pi*n*k/N) - jsin(2*pi*n*k/N)]}
        n=0
                                    k=0,1,2, ... ,N-1
                                    OR
            N-1
    Xr(k) = SUM {[xr(n) * cos(2*pi*n*k/N)] + [xi(n) * sin(2*pi*n*k/N)]}
            n=0
                k=0,1,2, ... ,N-1
            N-1
Xi(k) = SUM {[xi(n) * cos(2*pi*n*k/N)] - [xr(n) * sin(2*pi*n*k/N)]}
            n=0
***************************************************************************************
#include <math.h>
#include "params.h"
void dft(int N, COMPLEX *X)
{
    int n, k;
    double arg;
    int Xr[1024];
    int Xi[1024];
    short Wr, Wi;
```

```
        for(k=0; k<N; k++)
        {
            Xr[k] = 0;
            Xi[k] = 0;
        for(n=0; n<N; n++)
        {
            arg =(2*PI*k*n)/N;
            Wr = (short)((double) 32767.0 * cos(arg));
            Wi = (short)((double) 32767.0 * sin(arg));
            Xr[k] = Xr[k] + X[n].real * Wr + X[n].imag * Wi;
            Xi[k] = Xi[k] + X[n].imag * Wr - X[n].real * Wi;
        }
    }
    for (k=0;k<N;k++)
    {
        X[k].real = (short)(Xr[k]>>15);
        X[k].imag = (short)(Xi[k]>>15);
    }
}
```


## Example C-10. params.h File

```
/**********************************************************************************
    FILE
        params.h - This is the C header file for example real FFT
        implementations.
    ***********************************************************************************
#define TRUE 1
#define FALSE 0
#define BE TRUE
#define LE FALSE
#define ENDIAN LE /* selects proper endianess. If
                                building code in Big Endian,
                                use BE, else use LE */
#define PI 3.141592653589793 /* definition of pi */
/* Some functions used in the example implementations use word loads which make
    the code endianess dependent. Thus, one of the below definitions need to be
    used depending on the endianess you are using to build your code */
/* BIG Endian */
#if ENDIAN == TRUE
    typedef struct {
        short imag;
```

```
    short real;
    } COMPLEX;
#else
/* LITTLE Endian */
    typedef struct {
    short real;
    short imag;
    } COMPLEX;
#endif
```


## Example C-11. vectors.asm

```
/*****************************************************************************
/* vectors.asm - reset vector assembly */
/**********************************************************************/
    .def RESET
    .ref _c_int00
    .sect ".vectors"
RESET:
    mvk .s2 _c_int00, B2
    mvkh .s2 _c_int00, B2
    b .s2 B2
    nop
    nop
    nop
    nop
    nop
```


## Example C-12. Ink.cmd

```
/********************************************************************************
/* lnk.cmd - example linker command file */
/**************************************************************************
-c
-heap 0x2000
-stack 0x8000
MEMORY
{
    VECS: O = 00000000h l=00200h /* reset & interrupt vectors*/
    IPRAM: O = 00000200h l=0FEOOh /* internal program memory */
    IDRAM: O = 80000000h l=10000h /* internal data memory */
```

\}

## SECTIONS

\{

| vectors | $>$ | VECS |
| :--- | :--- | :--- |
| .text | $>$ | IPRAM |
| .tables | $>$ | IDRAM |
| .data | $>$ | IDRAM |
| . stack | $>$ | IDRAM |
| .bss | $>$ | IDRAM |
| . sysmem | $>$ | IDRAM |
| .cinit | $>$ | IDRAM |
| .const | $>$ | IDRAM |
| .cio | $>$ | IDRAM |
| .far | $>$ | IDRAM |

## Appendix D Optimized C Implementation of the DFT of Real Sequences

This appendix contains optimized C implementations of the efficient methods for performing the DFT of real-valued sequences outlined in this application report.

## D. 1 Implementation Notes

The following lists usage, assumption, and limitations of the code.

| Data format | All data and state variables are 16-bit signed integers (shorts). In this example, the decimal point is assumed to be between bits 15 and 14, thus the Q15 data format. For complex data and variables, the real and imaginary components are both Q15 numbers. From this data format, you can see that this code was developed for a fixed-point processor. |
| :---: | :---: |
| Memory | Complex data is stored in memory in imaginary/r,eal pairs. The imaginary component is stored in the most significant halfword (16 bits) and the real component is stored in the least significant halfword, unless otherwise noted. |
| Endianess | The code is presented and tested in little endian format. Some modification to the code is necessary for big endian format. |
| Overflow | No overflow protection or detection is performed. |
| File | Description |
| realdft3.c | DFT of a $2 N$-point real sequence main program |
| realdft4.c | DFT of a two N -point real sequence main program |
| radix4.c | Radix-4 FFT C function |
| digit.c | Radix-4 digit reversal C function |
| digitgen.c | C function used to initialize digit reversal table used by the function in digit .c |
| splitgen.c | C function used to initialize the split tables used by the split1 routines |

## Example D-1. realdft3.c File

```
/***********************************************************************************
    FILE
    realdft3.c - C source for an example implementation of the DFT/IDFT
    of a 2N-point real sequence, using one N-point complex DFT/IDFT.
```


## D. 2 Description

This program is an example implementation of an efficient way of computing the DFT/IDFT of a real-valued sequence.
In many applications, the input is a sequence of real numbers. If this condition is taken into consideration, additional computational savings can be achieved because the FFT of a real sequence has some symmetrical properties. The DFT of a 2 N -point real sequence can be efficiently computed using a N-point complex DFT and some additional computations.

The following steps are required in the computation of the FFT of a real-valued sequence using the split function:

1. Let $g(n)$ be a $2 N$-point real sequence. From $g(n)$, form the the $N$-point complex-valued sequence, $x(n)=x 1(n)+j x 2(n)$, where $x 1(n)=g(2 n)$ and $x 2(n)=g(2 n+1)$.
2. Perform an $N$-point complex FFT on the complex valued sequence $x(n) \rightarrow X(k)=$ $\operatorname{DFT}\{\mathrm{x}(\mathrm{n})$ \}. Note that the FFT can be any DFT method, such as radix-2, radix-4, mixed radix, direct implementation of the DFT, etc. However, the DFT output must be in normal order.
3. The following additional computation are used to get $G(k)$ from $X(k)$

$$
\begin{array}{r}
\operatorname{Gr}(\mathrm{k})=\mathrm{Xr}(\mathrm{k}) \operatorname{Ar}(\mathrm{k})-\mathrm{Xi}(\mathrm{k}) \operatorname{Ai}(\mathrm{k})+\mathrm{Xr}(\mathrm{~N}-\mathrm{k}) \operatorname{Br}(\mathrm{k})+\mathrm{Xi}(\mathrm{~N}-\mathrm{k}) \mathrm{Bi}(\mathrm{k}) \\
\mathrm{k}=0,1, \ldots, \mathrm{~N}-1 \\
\text { and } \mathrm{X}(\mathrm{~N})=\mathrm{X}(0) \\
\mathrm{Gi}(\mathrm{k})=\mathrm{Xi}(\mathrm{k}) \operatorname{Ar}(\mathrm{k})+\mathrm{Xr}(\mathrm{k}) \operatorname{Ai}(\mathrm{k})+\mathrm{Xr}(\mathrm{~N}-\mathrm{k}) \mathrm{Bi}(\mathrm{k})-\mathrm{Xi}(\mathrm{~N}-\mathrm{k}) \operatorname{Br}(\mathrm{k})
\end{array}
$$

Note that only $N$-points of the $2 N$-point sequence of $G(k)$ are computed in the above equations. Because the DFT of a real-sequence has symmetric properties, we can easily compute the remaining $N$ points of $G(k)$ with the following equations.
$\operatorname{Gr}(\mathrm{N})=\operatorname{Xr}(0)-\mathrm{Xi}(0)$
$\mathrm{Gi}(\mathrm{N})=0$
$\operatorname{Gr}(2 \mathrm{~N}-\mathrm{k})=\operatorname{Gr}(\mathrm{k})$

$$
\mathrm{k}=1,2, \ldots, \mathrm{~N}-1
$$

$\mathrm{Gi}(2 \mathrm{~N}-\mathrm{k})=-\mathrm{Gi}(\mathrm{k})$
As you can see, the above equations assume that $A(k)$ and $B(k)$, which are sine and cosine coefficients, are pre-computed. The $C$-code can be used to initialize $A(k)$ and $B(k)$.

```
for(k=0; k<N; k++)
{
    A[k].imag = (short)(16383.0*(-cos(2*PI/(double) (2*N)*(double)k)));
    A[k].real = (short) (16383.0*(1.0 - sin(2*PI/(double) (2*N)*(double)k)));
    B[k].imag = (short) (16383.0*(cos(2*PI/(double) (2*N)* (double)k)));
    B[k].real = (short)(16383.0*(1.0 + sin(2*PI/(double)(2*N)*(double)k)));
}
```

The following steps are required in the computation of the IFFT of a complex-valued frequency domain sequence that was derived from a real sequence:

1. Let $G(k)$ be a $2 N$-point complex valued sequence derived from a real-valued sequence $g(n)$. We want to get back $g(n)$ from $G(k) \rightarrow g(n)=\operatorname{IDFT}\{G(k)\}$. However, we want to apply the same techniques as we did with the forward FFT, using an N-point IFFT. This can be accomplished by the following equations.

$$
\begin{array}{r}
\mathrm{Xr}(\mathrm{k})=\mathrm{Gr}(\mathrm{k}) \operatorname{IAr}(\mathrm{k})-\mathrm{Gi}(\mathrm{k}) \mid \mathrm{Ai}(\mathrm{k})+\mathrm{Gr}(\mathrm{~N}-\mathrm{k}) \operatorname{IBr}(\mathrm{k})+\mathrm{Gi}(\mathrm{~N}-\mathrm{k}) \operatorname{IBi}(\mathrm{k}) \\
\mathrm{k}=0,1, \ldots, \mathrm{~N}-1 \\
\text { and } \mathrm{G}(\mathrm{~N})=\mathrm{G}(0) \\
\mathrm{Xi}(\mathrm{k})=\mathrm{Gi}(\mathrm{k}) \operatorname{IAr}(\mathrm{k})+\mathrm{Gr}(\mathrm{k}) \operatorname{IAi}(\mathrm{k})+\mathrm{Gr}(\mathrm{~N}-\mathrm{k}) \operatorname{IBi}(\mathrm{k})-\mathrm{Gi}(\mathrm{~N}-\mathrm{k}) \operatorname{IBr}(\mathrm{k})
\end{array}
$$

2. Perform the $N$-point inverse DFT of $X(k) \rightarrow x(n)=x 1(n)+j x 2(n)=I D F T\{X(k)\}$. Note that the IDFT can be any method, but must have an output that is in normal order.
3. $g(n)$ can then be found from $x(n)$.
$g(2 n)=x 1(n)$

$$
\mathrm{n}=0,1, \ldots, \mathrm{~N}-1
$$

$g(2 n+1)=x 2(n)$

As you can see, the above equations can be used for both the forward and inverse FFTs; however, the pre-computed coefficients are slightly different. The following C code can be used to initialize $\mathrm{IA}(\mathrm{k})$ and $\mathrm{IB}(\mathrm{k})$.

```
for(k=0; k<N; k++)
{
    IA[k].imag = - (short) (16383.0*(-cos(2*PI/(double) (2*N)*(double)k)));
    IA[k].real = (short) (16383.0*(1.0 - sin(2*PI/(double) (2*N)*(double)k)));
    IB[k].imag = - (short) (16383.0*(cos(2*PI/(double) (2*N)* (double)k)));
    IB[k].real = (short) (16383.0*(1.0 + sin(2*PI/(double) (2*N)*(double)k)));
}
```

Note that $\mathrm{IA}(\mathrm{k})$ is the complex conjugate of $\mathrm{A}(\mathrm{k})$, and $\mathrm{IB}(\mathrm{k})$ is the complex conjugate of $\mathrm{B}(\mathrm{k})$.
typedef struct $\{1 *$ define the data type for the radix-4 twiddle factors */ short imag; short real; \} COEFF;

```
#include "params1.h" /* header files with parameters */
#include "params.h"
#include "splittbl.h" /* header file that contains tables used to generate
    the split tables */
#include "sinestbl.h" /* header file that contains the FFT twiddle factors */
#pragma DATA_ALIGN(x,64); /* radix-4 routine requires x to be
    aligned to a 4*NUMPOINTS boundry */
COMPLEX x[NUMPOINTS+1]; /* array of complex DFT data */
extern short g[]; /* real-valued input sequence */
/* functions defined externally */
void FftSplitTableGen(int N, COMPLEX *W, COMPLEX *A, COMPLEX *B);
void R4DigitRevIndexTableGen(int, int *, unsigned short *, unsigned short *);
void split1(int, COMPLEX *, COMPLEX *, COMPLEX *, COMPLEX *);
void digit_reverse(int *, unsigned short *, unsigned short *, int);
void radix4(int, short[], short[]);
main()
```

\{

```
int n, k;
    COMPLEX A[NUMPOINTS]; /* array of complex A coefficients */
    COMPLEX B[NUMPOINTS]; /* array of complex B coefficients */
    COMPLEX IA[NUMPOINTS]; /* array of complex A* coefficients */
    COMPLEX IB[NUMPOINTS]; /* array of complex B* coefficients */
    COMPLEX G[2*NUMPOINTS]; /* array of complex DFT result */
    unsigned short IIndex[NUMPOINTS], JIndex[NUMPOINTS];
```

int count;
/* Initialize A, B, IA, and IB arrays */
FftSplitTableGen(NUMPOINTS, W, A, B);

```
/* Split tables for the IDFT are the complex conjugate of the split
    tables of the DFT */
    for(k=0; k<NUMPOINTS; k++)
    {
        IA[k].imag = -A[k].imag;
        IA[k].real = A[k].real;
        IB[k].imag = -B[k].imag;
        IB[k].real = B[k].real;
    }
/* Initialize tables for FFT digit reversal function */
    R4DigitRevIndexTableGen(NUMPOINTS, &Count, IIndex, JIndex);
```

/* Forward DFT */
/* From the 2N point real sequence, $g(n)$, for the $N$-point complex sequence, $x(n)$ */
for ( $\mathrm{n}=0$; $\mathrm{n}<$ NUMPOINTS; $\mathrm{n}++$ )
\{
$\mathrm{x}[\mathrm{n}]$. imag $=\mathrm{g}\left[2 *_{\mathrm{n}}+1\right] ; \quad / * \mathrm{x} 2(\mathrm{n})=\mathrm{g}(2 \mathrm{n}+1) \mathrm{t} /$
$x[n] . r e a l=g[2 * n] ; \quad / * x 1(n)=g(2 n) \quad * /$
\}
/* Compute the DFT of $x(n)$ to get $X(k)->X(k)=\operatorname{DFT}\{x(n)\}$ */
radix4 (NUMPOINTS, (short *)x, (short *)W4);
digit_reverse((int *)x, IIndex, JIndex, count);
/* Because of the periodicity property of the DFT, we know that $X(N+k)=X(k)$. */
x[NUMPOINTS].real $=$ x[O].real;
$\mathrm{x}[\mathrm{NUMPOINTS}] . i m a g=\mathrm{x}[0] . i m a g ;$
/* The split function performs the additional computations required to get
G(k) from X(k). */
split1 (NUMPOINTS, $x, A, B, G) ;$
/* Use complex conjugate symmetry properties to get the rest of $G(k)$ */
G[NUMPOINTS].real $=x[0] . r e a l ~-~ x[0] . i m a g ; ~$
G[NUMPOINTS].imag $=0$;
for (k=1; $\mathrm{k}<$ NUMPOINTS; $\mathrm{k}++$ )
\{

```
        G[2*NUMPOINTS-k].real = G[k].real;
        G[2*NUMPOINTS-k].imag = -G[k].imag;
    }
/* Inverse DFT - We now want to get back g(n). */
/* The split function performs the additional computations required to get
    X(k) from G(k). */
    split1(NUMPOINTS, G, IA, IB, x);
/* Take the inverse DFT of X(k) to get x(n). Note the inverse DFT could be any
    IDFT implementation, such as an IFFT. */
/* The inverse DFT can be calculated by using the forward DFT algorithm directly
    by complex conjugation - x(n) = (1/N)(DFT{X*(k)})*, where * is the complex
    conjugate operator. */
/* Compute the complex conjugate of X(k). */
    for (k=0; k<NUMPOINTS; k++)
    {
            x[k].imag = -x[k].imag; /* complex conjugate X(k) */
    }
/* Compute the DFT of X* (k). */
        radix4(NUMPOINTS, (short *)x, (short *)W4);
        digit_reverse((int *)x, IIndex, JIndex, count);
/* Complex conjugate the output of the DFT and divide by N to get x(n). */
    for (n=0; n<NUMPOINTS; n++)
    {
            x[n].real = x[n].real/16;
            x[n].imag = (-x[n].imag)/16;
    }
/* g(2n) = xr (n) and g(2n + 1) = xi(n) */
    for (n=0; n<NUMPOINTS; n++)
    {
        g[2*n] = x[n].real;
        g[2*n + 1] = x[n].imag;
    }
    return(0);
}
```


## Example D-2. realdft4.c File

```
/**********************************************************************************
    FILE
    realdft4.c - C source for an example implementation of the DFT/IDFT
    of two N-point real sequences using one N-point complex DFT/IDFT.
```


## Description

This program is an example implementation of an efficient way of computing the DFT/IDFT of two real-valued sequences.
Assume we have two real-valued sequences of length $N-x 1[n]$ and $x 2[n]$. The DFT of $x 1[n]$ and x2[n] can be computed with one complex-valued DFT of length N, as shown above, by following this algorithm.

1. Form the complex-valued sequence $x[n]$ from $x 1[n]$ and $x 2[n]$
$r[n]=x 1[n]$ and $x[n]=x 2[n], 0,1, \ldots, N-1$
Note, if the sequences $\mathrm{x} 1[\mathrm{n}]$ and $\mathrm{x} 2[\mathrm{n}]$ are coming from another algorithm or a data acquisition driver, this step may be eliminated if these put the data in the complex-valued format correctly.
2. Compute $\mathrm{X}[\mathrm{k}]=\operatorname{DFT}\{\mathrm{x}[\mathrm{n}]\}$

This can be the direct form DFT algorithm, or an FFT algorithm. If using an FFT algorithm, make sure the output is in normal order - bit reversal is performed.
3. Compute the following equations to get the DFTs of $x 1[n]$ and $x 2[n]$.

$$
\begin{aligned}
& \mathrm{X} 1 \mathrm{r}[0]=\mathrm{Xr}[0] \\
& \mathrm{X} 1 \mathrm{i}[0]=0 \\
& \mathrm{X} 2 \mathrm{r}[0]=\mathrm{Xi}[0] \\
& \text { X2i[0] = } 0 \\
& \mathrm{X} 1 \mathrm{r}[\mathrm{~N} / 2]=\mathrm{Xr}[\mathrm{~N} / 2] \\
& \text { X1i[N/2] = } 0 \\
& \text { X2r[N/2] }=\mathrm{Xi}[\mathrm{~N} / 2] \\
& \text { X2i[ } \mathrm{N} / 2]=0 \\
& \text { for } k=1,2,3, \ldots ., N / 2-1 \\
& \mathrm{X} 1 \mathrm{r}[\mathrm{k}]=(\mathrm{Xr}[\mathrm{k}]+\mathrm{Xr}[\mathrm{~N}-\mathrm{k}] / 2 \\
& \mathrm{X} 1 \mathrm{i}[\mathrm{k}]=(\mathrm{Xi}[\mathrm{k}]-\mathrm{Xi}[\mathrm{~N}-\mathrm{k}]) / 2 \\
& \mathrm{X} 1 \mathrm{r}[\mathrm{~N}-\mathrm{k}]=\mathrm{X} 1 \mathrm{r}[\mathrm{k}] \\
& \text { X1i[N-k] = X1i[k] }
\end{aligned}
$$

$$
\begin{aligned}
& \mathrm{X} 2 \mathrm{r}[\mathrm{k}]=(\mathrm{Xi}[\mathrm{k}]+\mathrm{Xi}[\mathrm{~N}-\mathrm{k}]) / 2 \\
& \mathrm{X} 2 \mathrm{i}[\mathrm{k}]=(\mathrm{Xr}[\mathrm{~N}-\mathrm{k}]-\mathrm{Xr}[\mathrm{k}]) / 2 \\
& \mathrm{X} 2 \mathrm{r}[\mathrm{~N}-\mathrm{k}]=\mathrm{X} 2 \mathrm{r}[\mathrm{k}] \\
& \mathrm{X} 2 \mathrm{i}[\mathrm{~N}-\mathrm{k}]=\mathrm{X} 2[\mathrm{i}[\mathrm{k}]
\end{aligned}
$$

4. Form $X[k]$ from $X 1[k]$ and $X 2[k]$

$$
\text { for } \begin{aligned}
\mathrm{k}=0,1 & \ldots, \mathrm{~N}-1 \\
& \mathrm{Xr}[\mathrm{k}]=\mathrm{X} 1 \mathrm{r}[\mathrm{k}]-\mathrm{X} 2 \mathrm{i}[\mathrm{k}] \\
& \mathrm{Xi}[\mathrm{k}]=\mathrm{X} 1 i[\mathrm{k}]+\mathrm{X} 2 \mathrm{r}[\mathrm{k}]
\end{aligned}
$$

5. Compute $\mathrm{x}[\mathrm{n}]=\operatorname{IDFT}\{\mathrm{X}[\mathrm{k}]\}$

This can be the direct form IDFT algorithm or an IFFT algorithm. If using an IFFT algorithm, make sure the output is in normal order - bit reversal is performed.

```
typedef struct { /* define the data type for the radix-4 twiddle factors */
    short imag;
    short real;
    } COEFF;
#include "params2.h" /* include file with parameters */
#include "params.h" /* include file with parameters */
#include "sinestbl.h" /* header file that contains the FFT twiddle factors */
#pragma DATA_ALIGN(x,64); /* radix-4 routine requires x to be
    aligned to a 4*NUMPOINTS boundary */
COMPLEX x[NUMPOINTS+1]; /* array of complex DFT data, X(k) */
extern short x1[];
extern short x2[];
void R4DigitRevIndexTableGen(int, int *, unsigned short *, unsigned short *);
extern void split2(int, COMPLEX *, COMPLEX *, COMPLEX *);
void digit_reverse(int *, unsigned short *, unsigned short *, int);
void radix4(int, short[], short[]);
main()
{
    int n, k;
```

    COMPLEX X1[NUMDATA]; /* array of real-valued DFT output sequence, X1(k) */
    COMPLEX X2[NUMDATA]; /* array of real-valued DFT output sequence, \(\mathrm{X} 2(\mathrm{k})\) */
    unsigned short IIndex[NUMPOINTS], JIndex[NUMPOINTS];
    int count;
    /* Initialize tables for FFT digit reversal function */
R4DigitRevIndexTableGen (NUMPOINTS, \&count, IIndex, JIndex);
/* Forward DFT */
/* From the two N-point real sequences, $x 1(n)$ and $x 2(n)$, form the $N$-point complex
sequence, $x(n)=x 1(n)+j x 2(n)$ */

```
for (n=0; n<NUMDATA; n++)
{
        x[n].real = x1[n];
        x[n].imag = x2[n];
}
```

```
/* Compute the DFT of x(n), X(k) = DFT{x(n)}. Note, the DFT can be any
```

    DFT implementation such as FFTs. */
    radix4 (NUMPOINTS, (short *)x, (short *)W4);
    digit_reverse((int *)x, IIndex, JIndex, count);
    /* Because of the periodicity property of the DFT, we know that $X(N+k)=X(k)$. */
$\mathrm{x}[$ NUMPOINTS].real $=\mathrm{x}[0] . r e a l$;
$\mathrm{x}[\mathrm{NUMPOINTS}] . i m a g=x[0] . i m a g ;$
/* The split function performs the additional computations required to get
X1 (k) and X2(k) from X(k). */
split2 (NUMPOINTS, x, X1, X2);
/* Inverse DFT - We now want to get back $\mathrm{x} 1(\mathrm{n})$ and $\mathrm{x} 2(\mathrm{n})$ from $\mathrm{X} 1(\mathrm{k})$ and $\mathrm{X} 2(\mathrm{k})$ using
one complex DFT */
/* Recall that $x(n)=x 1(n)+j x 2(n)$. Since the DFT operator is linear,
$X(k)=X 1(k)+j X 2(k)$. Thus we can express $X(k)$ in terms of $X 1(k)$ and $X 2(k)$. */
for $(\mathrm{k}=0$; $\mathrm{k}<$ NUMPOINTS; $\mathrm{k}++$ )
\{
$\mathrm{x}[\mathrm{k}] . \mathrm{real}=\mathrm{X} 1[\mathrm{k}]$. real $-\mathrm{X} 2[\mathrm{k}]$. imag;
x[k].imag $=\mathrm{X} 1[k] . i m a g+X 2[k] . r e a l ;$
\}
/* Take the inverse DFT of $\mathrm{X}(\mathrm{k})$ to get $\mathrm{x}(\mathrm{n})$. Note the inverse DFT could be any
IDFT implementation, such as an IFFT. */
/* The inverse DFT can be calculated by using the forward DFT algorithm directly
by complex conjugation $-x(n)=(1 / N)(D F T\{X *(k)\}) *$, where * is the complex
conjugate operator. */
/* Compute the complex conjugate of $\mathrm{X}(\mathrm{k})$. */
for (k=0; k<NUMPOINTS; k++)
\{
$x[k] . i m a g=-x[k] . i m a g ;$
\}
/* Compute the DFT of $\mathrm{X}^{*}(\mathrm{k})$. */
radix4 (NUMPOINTS, (short *)x, (short *)W4);
digit_reverse((int *)x, IIndex, JIndex, count);
/* Complex conjugate the output of the DFT and divide by N to get $\mathrm{x}(\mathrm{n})$. */
for ( $\mathrm{n}=0$; $\mathrm{n}<$ NUMPOINTS; $\mathrm{n}++$ )

```
    {
        x[n].real = x[n].real/16;
        x[n].imag = (-x[n].imag)/16;
    }
/* x1(n) is the real part of x(n), and x2(n) is the imaginary part of x(n). */
    for (n=0; n<NUMDATA; n++)
    {
        x1[n] = x[n].real;
        x2[n] = x[n].imag;
    }
    return(0);
}
```


## Example D-3. radix4.c File

```
/******************************************************************************
    FILE
        radix4.c - radix-4 FFT function based on Burrus, Parks p . 113
void radix4(int n, short x[], short w[])
{
    int n1, n2, ie, ia1, ia2, ia3, i0, i1, i2, i3, j, k;
    short t, r1, r2, s1, s2, co1, co2, co3, si1, si2, si3;
    n2 = n;
    ie = 1;
    for (k = n; k > 1; k >>= 2) {
        n1 = n2;
        n2 >>= 2;
        ia1 = 0;
        for (j = 0; j < n2; j++) {
        ia2 = ia1 + ia1;
        ia3 = ia2 + ia1;
        co1 = w[ia1 * 2 + 1];
        si1 = w[ia1 * 2];
        co2 = w[ia2 * 2 + 1];
        si2 = w[ia2 * 2];
        co3 = w[ia3 * 2 + 1];
        si3 = w[ia3 * 2];
        ia1 = ia1 + ie;
        for (iO = j; iO < n; iO += n1) {
            i1 = i0 + n2;
            i2 = i1 + n2;
```

```
                i3 = i2 + n2;
                r1 = x[2 * i0] + x[2 * i2];
                r2 = x[2 * i0] - x[2 * i2];
                t = x[2 * i1] + x[2 * i3];
                x[2 * i0] = r1 + t;
                r1 = r1 - t;
                s1 = x[2 * i0 + 1] + x[2 * i2 + 1];
                    s2 = x[2 * i0 + 1] - x[2 * i2 + 1];
                t = x[2 * i1 + 1] + x[2 * i3 + 1];
                x[2 * i0 + 1] = s1 + t;
                s1 = s1 - t;
                x[2 * i2] = (r1 * co2 + s1 * si2) >> 15;
                x[2 * i2 + 1] = (s1 * co2-r1 * si2)>>15;
                t = x[2 * i1 + 1] - x[2 * i3 + 1];
                r1 = r2 + t;
                r2 = r2 - t;
                    t = x[2 * i1] - x[2 * i3];
                    s1 = s2 - t;
                    s2 = s2 + t;
                x[2 * i1] = (r1 * co1 + s1 * si1) >>15;
                x[2 * i1 + 1] = (s1 * co1-r1 * si1)>>15;
                x[2 * i3] = (r2 * co3 + s2 * si3) >>15;
                x[2 * i3 + 1] = (s2 * co3-r2 * si3)>>15;
            }
        }
        ie <<= 2;
    }
}
```

Example D-4. digit.c File

```
/*********************************************************************************
    FILE
        digit.c - This is the C source code for a digit reversal function for
        a radix-4 FFT.
void digit_reverse(int *yx, unsigned short *JIndex, unsigned short *IIndex, int
count)
{
    int i;
    unsigned short I, J;
    int YXI, YXJ;
```

```
    for (i = 0; i<count; i++)
    {
    I = IIndex[i];
    J = JIndex[i];
        YXI = Yx[I];
        YXJ = YX[J];
        Yx[J] = YXI;
        YX[I] = YXJ;
    }
}
```


## Example D-5. digitgen.c File

```
/********************************************************************************
    FILE
        digitgen.c - This is the C source code for a function used to generate
        index tables for a digit reversal function for a radix-4 FFT.
void R4DigitRevIndexTableGen(int n, int *count, unsigned short *IIndex, unsigned
short *JIndex)
{
    int j, n1, k, i;
    j = 1;
    n1 = n - 1;
    *count = 0;
    for(i=1; i<=n1; i++)
    {
        if(i < j)
        {
            IIndex[*count] = (unsigned short)(i-1);
            JIndex[*count] = (unsigned short)(j-1);
            *count = *count + 1;
        }
        k = n >> 2;
        while(k*3 < j)
        {
            j = j - k*3;
            k = k >> 2;
        }
        j = j + k;
    }
}
```


## Example D-6. splitgen.c File

```
/*************************************************************************************
    FILE
        splitgen.c - This is the C source code for a function used to generate
        tables for a split routine used to efficiently compute the DFT of a 2N-point
        real-valued sequence.
    ****************************************************************************************)
#include "params.h"
void FftSplitTableGen(int N, COMPLEX *W, COMPLEX *A, COMPLEX *B)
{
    int k;
    for(k=0; k<N/2; k++)
    {
        A[k].real = 16383 - W[k].imag;
        A[k].imag = -W[k].real;
        A[k + N/2].real = 16383 - W[k].real;
        A[k + N/2].imag = W[k].imag;
        B[k].real = 16383 + W[k].imag;
        B[k].imag = W[k].real;
        B[k + N/2].real = 16383 + W[k].real;
        B[k + N/2].imag = -W[k].imag;
    }
}
```


## Appendix E Optimized C-Callable 'C62xx Assembly Language Functions Used to Implement the DFT of Real Sequences

This appendix contains optimized C-callable 'C62xx assembly language functions used to implement the DFT of real sequences.

## E. 1 Implementation Notes

The following lists usage, assumption, and limitations of the code.

| Data format | All data and state variables are 16 -bit signed integers (shorts). In this example, the decimal point is assumed to be between bits 15 and 14, thus the Q15 data format. For complex data and variables, the real and imaginary components are both Q15 numbers. From this data format, you can see that this code was developed for a fixed-point processor. |
| :---: | :---: |
| Memory | Complex data is stored in memory in imaginary/real pairs. The imaginary componen is stored in the most significant halfword (16 bits) and the real component is stored in the least significant halfword, unless otherwise noted. |
| Endianess | The code is presented and tested in little endian format. Some modification to the code is necessary for big endian format. |
| Overflow | No overflow protection or detection is performed. |
| File | Description |
| split1.asm | C-callable 'C62xx assembly version of the split function for the DFT of a $2 N$-point real sequence |
| split2.asm | C-callable 'C62xx assembly version of the split function for the DFT of a $2 N$-point real sequences. |
| radix4.asm | Radix-4 FFT C-callable 'C62xx assembly function. |
| digit.asm | Radix-4 digit reversal C-callable 'C62xx assembly function. |

## Example E-1. split1.asm File

$\star$
TEXAS INSTRUMENTS, INC.
Real FFT/IFFT split operation
Revision Date: 5/15/97
USAGE This routine is C-callable, and can be called as:
void split1 (int $N$, COMPLEX *X, COMPLEX *A, COMPLEX *B, COMPLEX *G)
$N=1 / 2$ the number of samples of the real valued sequence
$X=$ pointer to complex input array
$A=$ pointer to complex coefficients
$B=$ pointer to complex coefficients
G = pointer to complex output array

```
* int Tr, Ti;
``` guide).
    \{
        \{
            \}

If routine is not to be used as a C-callable function, then all instructions relating to stack should be removed. Refer to comments of individual instructions. You will also need to initialize values for all of the values passed, as these are assumed to be in registers as defined by the calling convention of the compiler, (refer to the Compiler reference

This is the \(C\) equivalent of the assembly code without restrictions. Note that the assembly code is hand-optimized, and restrictions may apply.

One small, but important note. The split functions uses word loads to read imaginary/real pairs from memory. Because of this, some \(C\) definitions may need to be endianess-dependent. Below are the type definitions for COMPLEX for both big and little endian. Also, the split function, as shown below, is written for big endian. See comments in the code to see how to modify, if little endian is desired.

LITTLE ENDIAN
typedef struct \{ typedef struct \{
short real;
short imag;
\} COMPLEX;
BIG ENDIAN short imag;
short real;
\} COMPLEX;
```

void split(int N, COMPLEX *X, COMPLEX *A, COMPLEX *B, COMPLEX *G)

```
```

        for (k=0; k<N; k++)
    ```
            Tr \(=\) (int) X[k].real * (int)A[k].real -
            (int) X[k].imag * (int)A[k].imag +
            (int)X[N-k].real * (int)B[k].real +
            (int) X[N-k].imag * (int)B[k].imag;
            G[k].real \(=(\) short \()(\operatorname{Tr} \gg 15)\);
            Ti \(=\) (int)X[k].imag * (int)A[k].real +
            (int)X[k].real * (int)A[k].imag +
            (int) X[N-k].real * (int)B[k].imag -
            (int)X[N-k].imag * (int)B[k].real;
            G[k].imag \(=(\) short \()(T i \gg 15) ;\)

DESCRIPTION
In many applications, the input is a sequence of real numbers. If this condition is taken into consideration, additional computational savings can be achieved because the FFT of a real sequence has some symmetrical properties. The DFT of a 2 N -point real sequence can be efficiently computed using a N-point complex DFT and some additional computations which have been implemented in this split function. Note this split function can be used in the computation of FFTs and IFFTs.

The following steps are required in the computation of the FFT of a real-valued sequence using the split function:
1. Let \(g(n)\) be a \(2 N\)-point real sequence. From \(g(n)\), form the the \(N\)-point complex-valued sequence, \(x(n)=x 1(n)+j x 2(n)\), where \(x 1(n)=g(2 n)\) and \(x 2(n)=g(2 n+1)\).
2. Perform an \(N\)-point complex FFT on the complex-valued sequence, \(x(n) \rightarrow X(k)=\operatorname{DFT}\{x(n)\}\). Note that the FFT can be any DFT method, such as radix-2, radix-4, mixed radix, direct implementation of the DFT, etc. However, the DFT output must be in normal order.
3. The following additional computations are used to get \(G(k)\) from \(X(k)\), and are implemented by the split function.
\(\operatorname{Gr}(\mathrm{k})=\mathrm{Xr}(\mathrm{k}) \operatorname{Ar}(\mathrm{k})-\mathrm{Xi}(\mathrm{k}) \mathrm{Ai}(\mathrm{k})+\mathrm{Xr}(\mathrm{N}-\mathrm{k}) \operatorname{Br}(\mathrm{k})+\mathrm{Xi}(\mathrm{N}-\mathrm{k}) \mathrm{Bi}(\mathrm{k})\)
\(\mathrm{k}=0,1, \ldots, \mathrm{~N}-1\)
and \(X(N)=X(0)\)
\(G i(k)=X i(k) A r(k)+X r(k) A i(k)+X r(N-k) B i(k)-X i(N-k) B r(k)\)

Note that only \(N\)-points of the \(2 N\)-point sequence of \(G(k)\) are computed in the above equations. Because the DFT of a real-sequence has symmetric properties, we can easily compute the remaining \(N\) points of \(G(k)\) with the following equations.
```

$\mathrm{Gr}(\mathrm{N})=\mathrm{Gr}(0)-\mathrm{Gi}(0)$
$G i(N)=0$
$\mathrm{Gr}(2 \mathrm{~N}-\mathrm{k})=\mathrm{Gr}(\mathrm{k})$
$\mathrm{k}=1,2, \ldots, \mathrm{~N}-1$
$\mathrm{Gi}(2 \mathrm{~N}-\mathrm{k})=-\mathrm{Gi}(\mathrm{k})$

```

As you can see, the split function assumes that \(A(k)\) and \(B(k)\), which are sine and cosine coefficient, are pre-computed. The \(C\)-code can be used to initialize \(A(k)\) and \(B(k)\).
```

for(k=0; k<N; k++)
{
A[k].imag = (short)(16383.0*(-cos (2*PI/(double) (2*N)* (double) k)));
A[k].real = (short) (16383.0*(1.0 - sin(2*PI/(double) (2*N)* (double)k)));
B[k].imag = (short) (16383.0* (cos (2*PI/ (double) (2*N)* (double)k)));
B[k].real = (short) (16383.0*(1.0 + sin(2*PI/(double) (2*N)* (double)k)));
}

```
    different.
    The following \(C\) code can be used to initialize IA(k) and IB(k).
* \(\{\)
* \(\}\)

The following steps are required in the computation of the IFFT of a real sequence using the split function: valued sequence \(g(n)\). We want to get back \(g(n)\) from \(G(k)\)-> \(g(n)=\operatorname{IDFT}\{G(k)\}\). However, we want to apply the same techniques as we did with the forward FFT, use a N-point IFFT. This can be
\[
\mathrm{n}=0,1, \ldots, \mathrm{~N}-1
\]
\[
g(2 n+1)=x 2(n)
\]
```

* for (k=0; k<N; k++)
for $(k=0 ; k<N ; k++)$

```
\(\star \quad\) IA [k].real \(=(\) short \()(16383.0 *(1.0-\sin (2 * P I /(\operatorname{double})(2 * N) *(\) double)k)));
\(\star \quad I B[k]\). imag \(=-(\) short \()(16383.0 *(\cos (2 * P I /(\) double \()(2 * N) *(d o u b l e) k)))\);
* \(\operatorname{IB}[k] . r e a l=(\operatorname{short})(16383.0 *(1.0+\sin (2 * P I /(\operatorname{double})(2 * N) *(d o u b l e) k)))\); a complex-valued frequency domain sequence that was derived from
1. Let \(G(k)\) be a \(2 N\)-point complex-valued sequence derived from a real accomplished by the following equations.
```

        Xr(k) = Gr(k)IAr(k) - Gi(k)IAi(k) + Gr(N-k)IBr(k) + Gi(N-k)IBi(k)
                                k = 0, 1, ..., N-1
                                and G(N) = G(0)
        Xi(k) = Gi(k)IAr(k) + Gr(k)IAi(k) + Gr(N-k)IBi(k) - Gi(N-k)IBr(k)
    ```
    2. Perform the \(N\)-point inverse DFT of \(X(k) \rightarrow x(n)=x 1(n)+j x 2(n)=\)
        \(\operatorname{IDFT}\{\mathrm{X}(\mathrm{k})\}\). Note that the IDFT can be any method, but must have an
        output that is in normal order.
    3. \(g(n)\) can then be found from \(x(n)\).
\[
g(2 n)=x 1(n)
\]

As you can see, the split function can be used for both the forward and different.
The following \(C\) code can be used to initialize \(I A(k)\) and \(I B(k)\).
\{
IA [k].imag \(=-\) (short) \((16383.0 *(-\cos (2 * P I /(\) double \()(2 * N) *(\) double \() k))\);
\(\operatorname{IB}[k]\). imag \(=-(\) short \()(16383.0 *(\cos (2 * P I /(\) double \()(2 * N) *(\) double \() k))\);
\}
Note that \(I A(k)\) is the complex conjugate of \(A(k)\), and \(I B(k)\) is the complex conjugate of \(B(k)\).

\section*{TECHNIQUES}

32-bit loads are used to load two 16-bit loads.

A, B, X, and G are stored as imaginary/real pairs.
Big endian is used. If little endian is desired, modification to the code is required. See comments in the code for which instructions

\section*{MEMORY NOTE}

A, B, X and G arrays should be aligned to word boundaries. Also, \(A\) and \(B\) should be aligned such that they do not generate a memory hit with X .

A, B, X, and G are all complex data and are required to be stored as imaginary/real pairs in memory, regardless of endianess. In other words, a load word from any of these arrays should result in the imaginary component in the upper 16 bits of a register, and the real component in the lower 16 bits.
```

CYCLES 4*N + 32

```
\(\star\)


|| add . 11 x N, XPtr, XNPtr ; XNPtr -> yx[N]
; Because there are delay slots in loads, we will begin by
; SW pipelining the split operations - in other words, while
; we are finishing the current loop iteration, we will be
; beginning the next.
ldw .d1 *APtr++[1],aI_aR ; Load a coefficient pointed by APtr.
|| ldw .d2 *XPtr++[1],xI_xR ; Load a data value pointed by XPtr.
nop ; Fill a delay slot.
ldw .d1 *XNPtr--[1],x2I_x2R ; Load a data value pointed by XNPtr.
|| ldw .d2 *BPtr++[1],bI_bR ; Load a coefficient pointed by BPtr.
nop ; Fill a delay slot.
ldw .d1 *APtr++[1],aI_aR ; Load the next value pointed by APtr.
; (Note that it will not overwrite
; the current value of aI_aR until
|| ldw .d2 *XPtr++[1],xI_xR ; Load the next value pointed by XPtr.
; for performing the multiplies, we take advantage of the feature
; feature that allows you to choose the operands from either the upper
; or lower halves of the register.
\begin{tabular}{|c|c|c|c|c|}
\hline mpy & .m1x & xI_xR, aI_aR, xRaR & & xRaR = xR * aR - mpy lower * lower \\
\hline mpyhl & .m2x & xI_xR, aI_aR, xIaR & & \(x I a R=x I *\) aR - mpy upper * lower \\
\hline mpylh & .m2x & xI_xR, aI_aR, xRaI & & xRaI = xR * aI - mpy lower * upper \\
\hline mpyh & .m1x & xI_xR, aI_aR,xIaI & & xIaI = xI * aI - mpy upper * upper \\
\hline ldw & . d 1 & *XNPtr--[1],x2I_x2R & & load a data value pointed by XNPtr \\
\hline ldw & . d 2 & *BPtr++[1],bI_bR & & load a coefficient pointed by BPtr \\
\hline mpy & .m1x & x2I_x2R,bI_bR, x 2 RbR & & \(x 2 \mathrm{RbR}=\mathrm{x} 2 \mathrm{R}\) * bR - mpy lower * lower \\
\hline mpyhl & . m 2 x & x2I_x2R,bI_bR, x2IbR & & \(x 2 I b R=x 2 I ~ * ~ b R ~-~ m p y ~ u p p e r ~ * ~ l o w e r ~\) \\
\hline mpylh & .m2x & x2I_x2R,bI_bR, x2RbI & & \(x 2 \mathrm{RbI}=\mathrm{x} 2 \mathrm{R}\) * bI - mpy lower * upper \\
\hline mpyh & .m1x & x2I_x2R,bI_bR, x2IbI & & x2IbI = x2I * bI - mpy upper * upper \\
\hline sub & . 11 & xRaR, xIaI, rel & & re1 = xRaR - xIaI \\
\hline add & . 12 & xRaI, xIaR, im1 & & im1 = xRaI + xIaR \\
\hline ldw & . d 1 & *APtr++[1],aI_aR & & 3rd load of aI_aR \\
\hline ldw & . d 2 & *XPtr++[1],xI_xR & & 3rd load of \(x I \_x R\) \\
\hline
\end{tabular}
; the second loads of \(x I \_x R\) and aI_aR are now avaiable, thus we can use
; them to begin the 2nd iteration of \(X^{\prime} s\) and \(A^{\prime} s\) multiplies

\begin{tabular}{llll} 
mpy &.\(m 1\) & \(x 2 I \_x 2 R, b I \_b R, x 2 R b R\) & ; \(x 2 R b R=x 2 R * b R-m p y\) lower * lower \\
mpyhl &.\(m 2 x\) & \(x 2 I \_x 2 R, b I \_b R, x 2 I b R\) & ; x2IbR \(=x 2 I *\) bR - mpy upper * lower \\
add &.\(l 1\) & re1,re2,real & ; real \(=r e 1+r e 2\) \\
add &.\(l 2\) & im1,im2,imag & ;imag \(=i m 1+i m 2\)
\end{tabular}
|| b .s2 LOOP
\begin{tabular}{lll} 
mpylh &.\(m 2\) & \(x 2 I \_x 2 R, b I \_b R, x 2 R b I\) \\
mpyh &.\(m 1 x\) & \(x 2 I \_x 2 R, b I \_b R, x 2 I b I\) \\
sub & .11 & xRaR,xIaI,re1 \\
add & .12 & \(x R a I, x I a R, i m 1\) \\
shr &.\(s 1\) & real,15,real \\
shr &.\(s 2\) & imag,15,imag \\
ldw &.\(d 1\) & *APtr++[1],aI_aR \\
ldw &.\(d 2\) & *XPtr++[1],xI_xR
\end{tabular}
; Branch to LOOP - Note that this is the ; branch for the first time through ; the loop. Because of this, we only ; need to do count-1 branches to LOOP ; within LOOP. ; \(x 2 R b I=x 2 R\) * bI - mpy lower * upper ; \(x 2 I b I=x 2 I\) * bI - mpy upper * upper ; rel = xRaR - xIaI ; im1 = xRaI + xIaR ; real = real >> 15 ; imag = imag >> 15 ; 4th load of aI_aR ; 4th load of xI_xR CAUTION - because of SW pipelining, we actually load more values of aI_aR, \(x I \_x R, b I \_b R\), and \(x 2 I \_x 2 R\) than we actually use. Thus, make sure these arrays are NOT aligned to a boundary close to the edge of illegal memory.

\section*{LOOP: ; this loop is executed \(N\) times}
mpy .m1 xI_xR,aI_aR,xRaR
; \(x R a R=x R\) * \(a R\) - mpy lower * lower mpyhl .m2x xI_xR,aI_aR,xIaR
; xIaR = xI * aR - mpy upper * lower
sth .d1 imag,*GPtr++[1]
\begin{tabular}{|c|c|c|}
\hline \begin{tabular}{l}
sth \\
[CNT]
\end{tabular} & \[
\begin{aligned}
& . d 1 \\
& \text { sub }
\end{aligned}
\] & \[
\begin{aligned}
& \text { real, *GPtr++[1] } \\
& .12 \quad \mathrm{CNT}, 1, \mathrm{CNT}
\end{aligned}
\] \\
\hline mpylh & . m 2 & xI_xR, aI_aR, xRaI \\
\hline mpyh & .m1x & xI_xR, aI_aR,xIaI \\
\hline add & . 11 & x2RbR, x 2 Ib I, re2 \\
\hline sub & . 12 & x2RbI, x2IbR,im2 \\
\hline ldw & . d 1 & *XNPtr--[1],x2I_x2R \\
\hline ldw & . d 2 & *BPtr++[1],bI_bR \\
\hline mpy & . m 1 & x2I_x2R,bI_bR, x2RbR \\
\hline mpyhl & .m2x & x2I_x2R,bI_bR, x2IbR \\
\hline add & . 11 & re1,re2,real \\
\hline add & . 12 & im1,im2,imag \\
\hline sth & . d 1 & real,*GPtr++[1] \\
\hline sth & . d 1 & imag, *GPtr++[1] \\
\hline [CNT] & b & .s2 LOOP \\
\hline mpylh & . m 2 x & x2I_x2R,bI_bR, x2RbI \\
\hline mpyh & .m1x & x2I_x2R,bI_bR, x2IbI \\
\hline sub & . 11 & xRaR,xIaI, re1 \\
\hline add & . 12 & xRaI, xIaR, im1 \\
\hline shr & . 11 & real,15,real \\
\hline shr & . s2 & imag, 15,imag \\
\hline ldw & . d 1 & *APtr++[1],aI_aR \\
\hline
\end{tabular}
```

; If (CNT != 0), CNT = CNT - 1.

```
; xRaI = xR * aI - mpy lower * upper
; xIaI = xI * aI - mpy upper * upper
; re2 \(=x 2 R b R+x 2 I b I\)
; im2 = x2RbI - x2IbR
; Next load of \(x 2 I \_x 2 R\)
; Next load of bI_bR
```

; x2RbR = x2R * bR - mpy lower * lower

```
; \(x 2 I b R=x 2 I\) * bR - mpy upper * lower
; real \(=\) re1 + re2
; imag = im1 + im2
; Store real in output buffer.
; CAUTION - Big Endian specific code
; If Little Endian is desired,
; replace this line with:
; If (CNT ! = 0), branch to LOOP.
; \(x 2 R b I=x 2 R\) * bI - mpy lower * upper
; x2IbI = x2I * bI - mpy upper * upper
; rel = xRaR - xIaI
; im1 = xRaI + xIaR
; real = real >> 15
; imag = imag >> 15
; next load of aI_aR
```

| ldw .d2 *XPtr++[1],xI_xR ; next load of xI_xR
; end of LOOP
ldw .d2 *--B15[1],B12 ; Pop B12 from the stack.
ldw .d2 *--B15[1],B11 ; Pop B11 from the stack.
ldw .d2 *--B15[1],B10 ; Pop B10 from the stack.
ldw .d2 *--B15[1],A12 ; Pop A12 from the stack.
ldw .d2 *--B15[1],A11 ; Pop A11 from the stack.
ldw .d2 *--B15[1],A10 ; Pop A10 from the stack.
b .s2 B3 ; Function return add .d2
B15,24,B15 ; Deallocate space from the stack.
nop 4 ; Fill delay slots.

```

\section*{Example E-2. split2.asm File}

*
* TEXAS INSTRUMENTS, INC.
\(\star\)
* Real FFT/IFFT split operation
*
* Revision Date: 6/4/97
*
\(\star\)
\(\star\)
\(\star\)
\(\star\)
* \(N=\) the number of samples of each real valued sequence
* \(X=\) pointer to complex input array
*
*
*
*
\(\star\)
*
*
\(\star\)
*
*
\(\star\)
\(\star\)
\(\star\)
\(\star\)
\(\star\)
\(\star\)
\(\star\)
\(\star\)
\(\star\)
*
*
\(\star\)
* typedef struct \{
* short real;
* short imag;
* \} COMPLEX;
\(\star\)
\(\star\)
* \(\{\)
\(\star\)
    void split2(int N, COMPLEX *X, COMPLEX *X1, COMPLEX *X2)
    \{
```

int k;
X1[0].real = X[0].real;
X1[0].imag = 0;
X2[0].real = X[0].imag;
X2[0].imag = 0;
X1[N/2].real = X[N/2].real;
X1[N/2].imag = 0;
X2[N/2].real = X[N/2].imag;
X2[N/2].imag = 0;
for (k=1; k<N/2; k++)
{
X1[k].real = (X[k].real + X[N-k].real)/2;
X1[k].imag = (X[k].imag - X[N-k].imag)/2;
X2[k].real = (X[k].imag + X[N-k].imag)/2;
X2[k].imag = (X[N-k].real - X[k].real)/2;
X1[N-k].real = X1[k].real;
X1[N-k].imag = -X1[k].imag;
X2[N-k].real = X2[k].real;
X2[N-k].imag = -X2[k].imag;
}
}
DESCRIPTION
In many applications, the input is a sequence of real numbers.
If this condition is taken into consideration, additional computational
savings can be achieved because the FFT of a real sequence has some
symmetrical properties. The DFT of a two N-point real sequence can be
efficiently computed using one N-point complex DFT and some additional
computations which have been implemented in this split function.
Note that this split function can be used in the computation of FFTs and IFFTs.
The following steps are required in the computation of the FFT of
two real-valued sequence using the split function:
Assume we have two real-valued sequences of length N - x1[n] and x2[n].
The DFT of x1[n] and x2[n] can be computed with one complex-valued DFT of
length N, as shown above, by following this algorithm.
1. Form the complex-valued sequence x[n] from x1[n] and x2[n]
xr[n] = x1[n] and xi[n] = x2[n], 0,1, ..., N-1

```
* Note that if the sequences \(x 1[n]\) and \(x 2[n]\) are coming from another algorithm * or a data acquisition driver, this step may be eliminated, if these put the
for \(k=0,1, \ldots, N-1\)
\(\mathrm{Xr}[\mathrm{k}]=\mathrm{X} 1 \mathrm{r}[\mathrm{k}]-\mathrm{X} 2 \mathrm{i}[\mathrm{k}]\)
\(\mathrm{Xi}[k]=\mathrm{X} 1 i[k]+\mathrm{X} 2 r[k]\)
5. Compute \(\mathrm{x}[\mathrm{n}]=\operatorname{IDFT}\{\mathrm{X}[\mathrm{k}]\}\)

This can be the direct-form IDFT algorithm, or an IFFT algorithm. If using an IFFT algorithm, make sure the output is in normal order - bit reversal is performed.

\section*{TECHNIQUES}

32-bit loads are used to load two 16-bit loads.

\section*{ASSUMPTIONS}

X, X1, and \(X 2\) are stored as imaginary/real pairs.
Little endian is used. If little endian is desired, modification to the code is required.

MEMORY NOTE
X must be aligned to a 32 -bit boundary.
CYCLES 5* (N/2-1) + 29
\(\star\)

N .set a4
XPtr .set b4
X1Ptr .set a6
X2Ptr .set b6
CNT .set b0
XNmkPtr .set a3
N4 .set a0
XiXr .set b2
XNiXNr .set a2
X1NmkPtr .set a1
X2NmkPtr .set b1
X2rX1r .set a8
X1iX2i .set b8
X1r .set a9
X1i .set b9
X2r .set a10
X2i .set b10
\(\mathrm{X1Nr}\).set a12
\(\mathrm{X1Ni}\).set b12
X2Nr .set a13
X2Ni .set b13
nullA .set a14
nullB .set b5
.global_split2
_split2:
subaw .d2 B15,10,B15 ; Allocate space on the stack
ldh .d2 *XPtr,X1r ; X1r = Xr[0]
    add .l B15,4,A15 ; A15 points to the stack as well
|| ldh .d2 *+XPtr[1],X2r \(\quad\); X2r = Xi[0]
    stw .d1 A10,*A15++[2] ; Push A10 onto the stack.
|| stw .d2 B10,*B15++[2] ; Push B10 onto the stack.
    stw .d1 A11,*A15++[2] ; Push A11 onto the stack.
|| stw .d2 B11,*B15++[2] ; Push B11 onto the stack.
    stw .d1 A12,*A15++[2] ; Push A12 onto the stack.
|| stw .d2 B12,*B15++[2] ; Push B12 onto the stack.
    stw .d1 A13,*A15++[2] ; Push A13 onto the stack.
|| stw .d2 B13,*B15++[2] ; Push B13 onto the stack.
    stw .d1 A14,*A15++[2] ; Push A14 onto the stack.



\section*{Example E-3. radix4.asm File}
```

*********************************************************************************
*

* TEXAS INSTRUMENTS INC.
* 
* 
* 
* 
* 
* 
* 
* 
* 
* n --- FFT size (power of 4) (input)
* 
* 
* 
* 
* 
* 
* 
* 
* 
* 
* 
* SOURCE:Burrus, Parks p . }11
* 
* void radix4(int n, short x[], short w[])
* 
* 
* 
* 
* n2 = n;
* ie = 1;
* for (k=n; k > 1; k >>= 2) {
* 

C CODE
This is the C equivalent of the assembly code, without the
assumptions listed below. Note that the assembly code is hand-
optimized and assumptions apply.
{
int n1, n2, ie, ia1, ia2, ia3, i0, i1, i2, i3, j, k;
short t, r1, r2, s1, s2, co1, co2, co3, si1, si2, si3;
n1 = n2;
n2 >>= 2;
ial = 0;
for (j = 0; j < n2; j++) {
ia2 = ia1 + ia1;
ia3 = ia2 + ial;
co1 = w[ial * 2 + 1];
sil = w[ial * 2];
co2 = w[ia2 * 2 + 1];
si2 = w[ia2 * 2];
co3 = w[ia3 * 2 + 1];
si3 = w[ia3 * 2];
ial = ial + ie;

```
```

                for (i0 = j; i0 < n; i0 += n1) {
                    i1 = i0 + n2;
                i2 = i1 + n2;
                i3 = i2 + n2;
                r1 = x[2 * i0] + x[2 * i2];
                r2 = x[2 * i0] - x[2 * i2];
                t = x[2 * i1] + x[2 * i3];
                x[2 * i0] = r1 + t;
                r1 = r1 - t;
                s1 = x[2 * i0 + 1] + x[2 * i2 + 1];
                s2 = x[2 * i0 + 1] - x[2 * i2 + 1];
                t = x[2 * i1 + 1] + x[2 * i3 + 1];
                x[2 * i0 + 1] = s1 + t;
                s1 = s1 - t;
                        x[2 * i2] = (r1 * co2 + s1 * si2) >> 15;
                                x[2 * i2 + 1] = (s1 * co2-r1 * si2)>>15;
                                t = x[2 * i1 + 1] - x[2 * i3 + 1];
                                r1 = r2 + t;
                                r2 = r2 - t;
                                t = x[2 * i1] - x[2 * i3];
                                s1 = s2 - t;
                s2 = s2 + t;
                x[2 * il] = (r1 * co1 + s1 * sil) >>15;
                x[2 * i1 + 1] = (s1 * co1-r1 * si1)>>15;
                                x[2 * i3] = (r2 * co3 + s2 * si3) >>15;
                                x[2 * i3 + 1] = (s2 * co3-r2 * si3)>>15;
                    }
                }
                ie <<= 2;
    }
    }
DESCRIPTION
This routine is used to compute FFT of a complex sequence of size $n$, a power of 4, with "decimation-in-frequency decomposition" method. The output is in digit-reversed order. Each complex value is with interleaved 16-bit real and imaginary parts.

```

\section*{TECHNIQUES}
```

1. Loading input $x$ as well as coefficient $w$ in word.
2. Both loops j and iO shown in the C code are placed in the INNERLOOP of the assembly code.
ASSUMPTIONS
$4<=\mathrm{n}<=65536$
Both input $x$ and coefficient $w$ should be aligned on word boundary.
```
* MEMORY NOTE
```

* Align x and w on different word boundaries to minimize
* memory bank hits. There are N/4 memory bank hits total
* CYCLES
* CYCLES
* 

```

B_START:
```

; A1 = 32
; 31 - log2(n)
; n2 = n / 4
; Mask
; Push A11 on dummy.
; Push B11 on dummy.
; log2(n)+1 (circ buff size in bytes)
; 2 * n2 = n / 2, a-side
; 2 * n2 = n / 2, b-side
; n / 4
; Push A12 on dummy.
; Push B12 on dummy.
; Shift into BKO field.
; Save off x.
; Push A13 on dummy.
; Push B13 on dummy.
; A5, B5 set circular mode on BK0
; ie = 1
; Push A14 on dummy.
; Push B14 on dummy.
; Load AMR.
; Push A15 on dummy.
; Push B15 on dummy.
; Loop coutner = n / 4 - 1
; Reset X load pointer.
; Reset X store pointer.
; i = loop counter + 1
; Setup twiddle factor pointer
; j = 0
; Setup for first preincrement
; J loop twiddle reload test
; Setup for first preincrement

```


\begin{tabular}{|c|c|c|c|c|c|}
\hline ZERO & . L2 & B2 & & & \\
\hline LDW & & . D1 & * + A0 [1], & A10 & ; Pop A10 off dummy. \\
\hline LDW & & . D2 & * + B0 [2], & B10 & ; Pop B10 off dummy. \\
\hline MVC & . S 2 & B2, & AMR & & ; Reset AMR. \\
\hline LDW & & . D1 & *+A0 [3], & A11 & ; Pop A11 off dummy. \\
\hline LDW & & . D2 & * + B0 [4], & B11 & ; Pop B11 off dummy. \\
\hline LDW & & . D1 & * + A0 [5], & A12 & ; Pop A12 off dummy. \\
\hline LDW & & . D2 & * + B0 [6], & B12 & ; Pop B12 off dummy. \\
\hline LDW & & . D1 & * + A0 [7], & A13 & ; Pop A13 off dummy. \\
\hline LDW & & . D2 & * + B0 [8], & B13 & ; Pop B13 off dummy. \\
\hline LDW & & . D1 & *+A0[9], & A14 & ; Pop A14 off dummy. \\
\hline LDW & & . D2 & * + B0 [10], & B14 & ; Pop B14 off dummy. \\
\hline B . S2 & B3 & & & & \\
\hline LDW & & . D1 & *+A0[11], & A15 & ; Pop A15 off dummy. \\
\hline LDW & & . D2 & * + B0 [12], & B15 & ; Pop B15 off dummy. \\
\hline NOP & & 4 & & & ; Wait 4 cycles for the last pop ; to occur before returning. \\
\hline
\end{tabular}

\section*{Example E-4. digit.asm File}
```

; ********************************************************************************
; FILE
digit.asm - C62xx assembly source for a C-callable FFT digit reversal
function.
;
; ********************************************************************************
DESCRIPTION
This functions implements, by table lookup, digit/bit reversal for FFT
algorithms. The function assumes that index tables which contain the
indexes of data pairs that get swapped are pre-computed, and stored as
two separate arrays. Since this is a table lookup method, this is a
generic routine. It can be used for bit-reversal of radix-2 FFTs, or
digit-reversal of radix-4 FFTs etc.
;
**********************************************************************************
POTOTYPE
void digit_reverse(int *yx, unsigned short *IIndex,
unsigned short *JIndex, int count)
;
;********************************************************************************
IMPLEMENTATION
The following C code is functional equivalent to this assembly version.
void digit_reverse(int *yx, unsigned short *JIndex,
unsigned short *IIndex, int count)
{
int i;
unsigned short I, J;
int YXI, YXJ;

```

INSTRUMENTS
```

; for (i = 0; i<count; i++)
; {
; I = IIndex[i];
; J = JIndex[i];
; YXI = Yx[I];
; YXJ = Yx[J];
; Yx[J] = YXI;
; Yx[I] = YXJ;
; }
;
; }
;
; **********************************************************************************
.global __digit_reverse
AXPtr .set a4 ; arg1 passed by calling function
; Pointer to FFT data. This is a static
; pointer. Data to be reversed is accessed
; using indexes. Also this an A register,
JIndexPtr .set b4 ; thus it is used in the .d1 unit
; pointer to digit reversal index
IIndexPtr .set a6 ; arg3 passed by calling function
count .set b6 ; arg4 passed by calling function
; Number of points to reverse
J .set a0 ; index loaded using JIndex pointer
I .set b0 ; Index loaded using IIndex pointer
TJ .set a7 ; Temporary copy of J. This is needed
; because the next value of J is loaded
; before the current one is finished being used.
; It is used to store the data value
; loaded by the I index.
TI .set b7 ; Temporary copy of I. This is needed
; because the next value of I is loaded
; before the current one is finished being used.
; It is used to store the data value
; loaded by the J index.
XI .set a5 ; Data value loaded using the I index.
XJ .set b5 ; Data value loaded using the J index.
BXPtr .set b2 ; Pointer to FFT data, points to the same
; memory location as AXPtr. It is a B register,
; so it can be used in the .d2 unit.
CNT .set b1 ; Count register, used for looping
.text
_digit_reverse:
ldh .d1 *IIndexPtr++[1], I ; Load an I index.
ldh .d2 *JIndexPtr++[1], J ; Load a J index.
mv.l2x AXPtr, BXPtr ; Copy AXPtr to BXPtr.
nop 2 ; Fill the delay slots.
ldh .d1 *IIndexPtr++[1], I ; Load the next I index.
| ldh .d2 *JIndexPtr++[1], J ; Load the next J index.

```
\begin{tabular}{|c|c|c|c|}
\hline || sub & . 12 & count, 1, CNT & \begin{tabular}{l}
; Decrement the count by \\
; one, and put into a register \\
; that can be used as a \\
; condition register.
\end{tabular} \\
\hline nop & & 1 & ; Fill a delay slot. \\
\hline 1 dw & . d 1 & *+AXPtr[J], XJ & ; Load the value pointed by \\
\hline || ldw & . d 2 & *+BXPtr[I], XI & \begin{tabular}{l}
; the first \(J\) index loaded. \\
; Load the value pointed by \\
; the first I index loaded.
\end{tabular} \\
\hline | b & LOOP & & \begin{tabular}{l}
; Branch for the first time \\
; through the loop.
\end{tabular} \\
\hline nop & & & ; Fill a delay slot. \\
\hline mv.l1 & J, TJ & & \begin{tabular}{l}
; Make a copy of \(J\) so that \\
; the value is not lost due \\
; to the reloading of J.
\end{tabular} \\
\hline || mv & . 12 & I, TI & \begin{tabular}{l}
; Make a copy of I so that \\
; the value is not lost due \\
; to the reloading of I.
\end{tabular} \\
\hline \multicolumn{4}{|l|}{LOOP :} \\
\hline \(1 d w\) & . d 1 & *+AXPtr[J], XJ & ; load the value pointed by J \\
\hline 1 dw & . d 2 & *+BXPtr[I], XI & ; load the value pointed by I \\
\hline [CNT] & b & .s1 LOOP & ; conditional branch, branch \\
\hline & & & ; if CNT ! = 0 \\
\hline ; || [! CNT]b & . s2 & B3 & \begin{tabular}{l}
; having the return \\
; here may be a bug, \\
; we can try it when \\
; we get everything else \\
; working
\end{tabular} \\
\hline 1 dh & . \({ }^{\text {d }}\) & *IIndexPtr++[1], I & ; Load the next I index. \\
\hline 1 dh & . d2 & *JIndexPtr++[1], J & ; Load the next \(J\) index. \\
\hline [CNT] & sub & . 12 CNT, 1, CNT & ; Decrement the loop counter. \\
\hline stw & . \({ }^{\text {d }}\) & XI, * + AXPtr [TJ] & \begin{tabular}{l}
; Data loaded from the I index \\
; is stored at the location \\
; pointed by the J index.
\end{tabular} \\
\hline || stw & . d 2 & XJ, *+BXPtr[TI] & \begin{tabular}{l}
; Data loaded from the \(J\) index \\
; is stored at the location \\
; pointed by the I index. \\
; Note that TJ and TI have the I \\
; and J values, 3 iterations back.
\end{tabular} \\
\hline || mv.l1 & J, TJ & & \begin{tabular}{l}
; Make a copy of \(J\) so that \\
; the value is not lost due \\
; to the reloading of J.
\end{tabular} \\
\hline || mv & . 12 & I, TI & \begin{tabular}{l}
; Make a copy of I so that \\
; the value is not lost due \\
; to the reloading of \(I\).
\end{tabular} \\
\hline \multicolumn{4}{|l|}{; loop end} \\
\hline \[
\begin{aligned}
& \mathrm{b} . \mathrm{s}^{2} \\
& \text { nop }
\end{aligned}
\] & B3 & 5 & ; Function return \\
\hline
\end{tabular}

\section*{IMPORTANT NOTICE}

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, modifications, enhancements, improvements, and other changes to its products and services at any time and to discontinue any product or service without notice. Customers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All products are sold subject to Tl's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its hardware products to the specifications applicable at the time of sale in accordance with TI's standard warranty. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by government requirements, testing of all parameters of each product is not necessarily performed.

TI assumes no liability for applications assistance or customer product design. Customers are responsible for their products and applications using TI components. To minimize the risks associated with customer products and applications, customers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any TI patent right, copyright, mask work right, or other TI intellectual property right relating to any combination, machine, or process in which TI products or services are used. Information published by TI regarding third-party products or services does not constitute a license from TI to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. Reproduction of this information with alteration is an unfair and deceptive business practice. TI is not responsible or liable for such altered documentation.

Resale of TI products or services with statements different from or beyond the parameters stated by TI for that product or service voids all express and any implied warranties for the associated TI product or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

\author{
Mailing Address: \\ Texas Instruments \\ Post Office Box 655303 \\ Dallas, Texas 75265
}```

