











SLOS094G - NOVEMBER 1970-REVISED JANUARY 2018

uA741

µA741 General-Purpose Operational Amplifiers

Features

- **Short-Circuit Protection**
- Offset-Voltage Null Capability
- Large Common-Mode and Differential Voltage Ranges
- No Frequency Compensation Required
- No Latch-Up

Applications

- **DVD Recorders and Players**
- **Pro Audio Mixers**

3 Description

The µA741 device is a general-purpose operational amplifier featuring offset-voltage null capability.

The high common-mode input voltage range and the absence of latch-up make the amplifier ideal for voltage-follower applications. The device short-circuit protected and the internal frequency compensation ensures stability without external components. A low-value potentiometer may be connected between the offset null inputs to null out the offset voltage as shown in Figure 12.

The µA741C device is characterized for operation from 0°C to 70°C.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
μA741CD	SOIC (8)	4.90 mm × 3.91 mm
μA741CP	PDIP (8)	9.81 mm × 6.35 mm
μA741CPS	SO (8)	6.20 mm × 5.30 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

Simplified Schematic

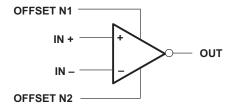




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4 Revision History

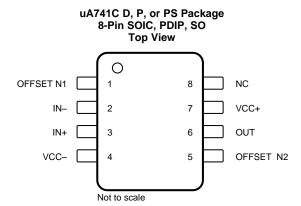
Cł	Inges from Revision E (January 2015) to Revision F Updated data sheet text to the latest documentation and translation standards Deleted text regarding μΑ741M device (obsolete package) from Description section			
•	Changed supply voltage unit from "°C" to "V" in Absolute Maximum Ratings table	5		
Cł	nanges from Revision E (January 2015) to Revision F	је		
•	Updated data sheet text to the latest documentation and translation standards	1		
•	Deleted text regarding µA741M device (obsolete package) from Description section	1		
•	Added µA741CD, µA741CP, and µA741CPS devices to Device Information table	1		
•	Deleted µA741x device from <i>Device Information</i> table	1		
•	Updated pinout diagrams and Pin Functions tables in the Pin Configurations and Functions section	4		
•	Deleted µA741M pinout drawings information from <i>Pin Configurations and Functions</i> section	4		
•	Deleted Electrical Characteristics: µA741M table from Specifications section	5		
•	Added operating junction temperature (T _J) and values to <i>Absolute Maximum Ratings</i> table	5		
•	Deleted text regarding µA741M from Absolute Maximum Ratings table	5		
•	Deleted text regarding µA741M device from Recommended Operating Conditions table	5		
•	Deleted Dissipation Ratings table	5		
•	Added Thermal Information table and values	5		
•	Deleted µA741M in Switching Characteristics table	7		
•	Correct typo in Figure 1	8		
•	Deleted text regarding µA741M device from <i>Detailed Description</i> section	10		
•	Updated text in Overview section	10		
•	Added 2017 copyright to Functional Block Diagram 1	10		
•	Added caption to Figure 11 in Device Functional Modes section	11		
•	Changed pins 1 and 5 from "NC" to "Offset N1" and "Offset N2" in Figure 18	15		



CI	hanges from Revision D (February 2014) to Revision E	Page
•	Added Applications, Device Information table, Pin Functions table, ESD Ratings table, Thermal Information table, Feature Description section, Device Functional Modes, Application and Implementation section, Power Supply Recommendations section, Layout section, Device and Documentation Support section, and Mechanical, Packaging, and Orderable Information section.	1
•	Moved Typical Characteristics into Specifications section.	8
CI	hanges from Revision C (January 2014) to Revision D	Page
<u>.</u>	Fixed Typical Characteristics graphs to remove extra lines.	8
CI	hanges from Revision B (September 2000) to Revision C	Page
•	Updated document to new TI data sheet format - no specification changes.	1
•	Deleted Ordering Information table.	1



5 Pin Configurations and Functions



NC- no internal connection

Pin Functions

PIN ,,		1/0	DECORIDATION	
NAME	NO.	I/O	DESCRIPTION	
IN+	3	I	Noninverting input	
IN-	2	I	erting input	
NC	8	_	No internal connection	
OFFSET N1	1	I	External input offset voltage adjustment	
OFFSET N2	5	I	External input offset voltage adjustment	
OUT	6	0	Output	
VCC+	7	_	Positive supply	
VCC-	4	_	Negative supply	



6 Specifications

6.1 Absolute Maximum Ratings

over virtual junction temperature range (unless otherwise noted)⁽¹⁾

over virtual junicilon temperature range (unit			MIN	MAX	UNIT
Supply voltage, V _{CC} ⁽²⁾	μA741C		-18	18	V
Differential input voltage, V _{ID} ⁽³⁾	μA741C		-15	15	V
Input voltage, V _I (any input) (2)(4)	μA741C		-15	15	V
Voltage between offset null (either OFFSET N1 or OFFSET N2) and $\rm V_{CC-}$	μΑ741C		-15	15	V
Duration of output short circuit ⁽⁵⁾			Unlimited		
Continuous total power dissipation			See Thermal Information		
Case temperature for 60 seconds		μΑ741C	N/A	N/A	°C
Lead temperature 1.6 mm (1/16 inch) from case for 60 s	econds	μΑ741C	N/A	N/A	°C
Lead temperature 1.6 mm (1/16 inch) from case for 10 seconds	D, P, or PS package	μA741C		260	°C
Operating junction temperature, T _J				150	°C
Storage temperature range, T _{stg}		μA741C	-65	150	°C

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- (2) All voltage values, unless otherwise noted, are with respect to the midpoint between V_{CC+} and V_{CC-}.
- (3) Differential voltages are at IN+ with respect to IN -.
- (4) The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
- (5) The output may be shorted to ground or either power supply.

6.2 Recommended Operating Conditions

			MIN	MAX	UNIT
V_{CC+}	Cumply voltage		5	15	\/
V _{CC} -	Supply voltage		-5	-15	V
T _A	Operating free-air temperature	μΑ741C	0	70	°C

6.3 Thermal Information

			μΑ741				
	THERMAL METRIC ⁽¹⁾	D (SOIC)	P (PDIP)	PS (SO)	UNIT		
		8 PINS	8 PINS	8 PINS			
$R_{\theta JA}$	Junction-to-ambient thermal resistance	129.2	87.4	119.7	°C/W		
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	73.6	89.3	66	°C/W		
$R_{\theta JB}$	Junction-to-board thermal resistance	72.4	64.4	70	°C/W		
ΨЈТ	Junction-to-top characterization parameter	25.9	49.8	27.2	°C/W		
ΨЈВ	Junction-to-board characterization parameter	71.7	64.1	69	°C/W		

(1) For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

Product Folder Links: uA741



6.4 Electrical Characteristics: μΑ741C

at specified virtual junction temperature, $V_{CC\pm} = \pm 15 \text{ V}$ (unless otherwise noted)

	PARAMETER	TEST CO	ONDITIONS ⁽¹⁾	MIN	TYP	MAX	UNIT	
\	Innuit offect velters	V 0	25°C		1	6	mV	
V _{IO}	Input offset voltage	V _O = 0	Full range			7.5	mv	
$\Delta V_{IO(adj)}$	Offset voltage adjust range	V _O = 0	25°C		±15		mV	
1	Input offset current	V 0	25°C		20	200	nA	
Ю	input onset current	V _O = 0	Full range			300	ΠA	
1	Input bias current	V _O = 0	25°C		80	500	nA	
I _{IB}	input bias current	v _O = 0	Full range			800	IIA	
\/	Common-mode input voltage range	25°C		±12	±13		V	
V _{ICR}	Common-mode input voltage range	Full range		±12			V	
		$R_L = 10 \text{ k}\Omega$	25°C	±12	±14			
\	Maximum pools output voltage avring	R _L ≥ 10 kΩ	Full range	±12			V	
V _{OM}	Maximum peak output voltage swing	$R_L = 2 k\Omega$	25°C	±10			V	
		$R_L \ge 2 k\Omega$	Full range	±10				
^	Large-signal differential voltage	$R_L \ge 2 k\Omega$	25°C	20	200		V/mV	
A _{VD}	amplification	V _O = ±10 V	Full range	15			V/IIIV	
r _i	Input resistance	25°C		0.3	2		$M\Omega$	
r _o	Output resistance	$V_0 = 0$; see ⁽²⁾	25°C		75		Ω	
C _i	Input capacitance	25°C			1.4		pF	
CMRR	Common-mode rejection ratio	V - V	25°C	70	90		dB	
CIVIKK	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	Full range	70			ав	
ما	Supply voltage consitivity (AV /AV)	$V_{CC} = \pm 9 \text{ V to } \pm 15 \text{ V}$	25°C		30	150	μV/V	
k _{SVS}	Supply voltage sensitivity ($\Delta V_{IO}/\Delta V_{CC}$)	V _{CC} = ±9 V (0 ±15 V	Full range			150 µ	μν/ν	
os	Short-circuit output current	25°C			±25	±40	mA	
	Supply ourrent	V = 0: no load	25°C		1.7	2.8	m ^	
cc	Supply current	pply current $V_0 = 0$; no load	Full range			3.3	mA	
-	Total names discination	V O. no load	25°C		50	85	m\\\/	
P _D	Total power dissipation	$V_O = 0$; no load	Full range			100	mW	

All characteristics are measured under open-loop conditions with zero common-mode input voltage unless otherwise specified.
 Full range for the μA741C is 0°C to 70°C.

⁽²⁾ This typical value applies only at frequencies above a few hundred hertz because of the effects of drift and thermal feedback.



6.5 Electrical Characteristics: μΑ741Υ

at specified virtual junction temperature, $V_{CC\pm} = \pm 15 \text{ V}$, $T_A = 25^{\circ}\text{C}$ (unless otherwise noted)⁽¹⁾

	PARAMETER	TEST CONDITIONS (2)	MIN	TYP	MAX	UNIT
V _{IO}	Input offset voltage	V _O = 0		1	5	mV
$\Delta V_{IO(adj)}$	Offset voltage adjust range	V _O = 0		±15		mV
I _{IO}	Input offset current	V _O = 0		20	200	nA
I _{IB}	Input bias current	V _O = 0		80	500	nA
V _{ICR}	Common-mode input voltage range		±12	±13		V
	Management and a section of the section of	$R_L = 10 \text{ k}\Omega$	±12	±14		
V_{OM}	Maximum peak output voltage swing	$R_L = 2 k\Omega$	±10	±13		V
A _{VD}	Large-signal differential voltage amplification	$R_L \ge 2 k\Omega$	20	200		V/mV
ri	Input resistance		0.3	2		ΜΩ
r _o	Output resistance	V _O = 0; see ⁽¹⁾		75		Ω
Ci	Input capacitance			1.4		pF
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	70	90		dB
k _{SVS}	Supply voltage sensitivity (ΔV _{IO} /ΔV _{CC})	V _{CC} = ±9 V to ±15 V		30	150	μV/V
I _{OS}	Short-circuit output current			±25	±40	mA
I _{CC}	Supply current	V _O = 0; no load		1.7	2.8	mA
P _D	Total power dissipation	V _O = 0; no load		50	85	mW

⁽¹⁾ This typical value applies only at frequencies above a few hundred hertz because of the effects of drift and thermal feedback.

6.6 Switching Characteristics: μΑ741C

over operating free-air temperature range, $V_{CC\pm} = \pm 15 \text{ V}$, $T_A = 25^{\circ}\text{C}$ (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _r	Rise time	$V_{l} = 20 \text{ mV}, R_{L} = 2 \text{ k}\Omega$		0.3		μs
	Overshoot factor	C _L = 100 pF; see Figure 1		5%		
SR	Slew rate at unity gain	V_I = 10 V, R_L = 2 k Ω C_L = 100 pF; see Figure 1		0.5		V/µs

6.7 Switching Characteristics: µA741Y

over operating free-air temperature range, $V_{CC\pm}$ = ±15 V, T_A = 25°C (unless otherwise noted)

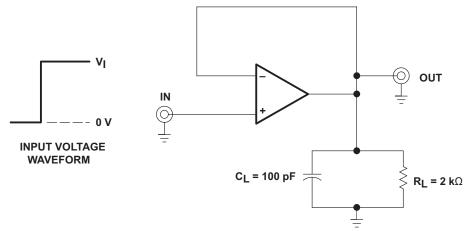
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _r	Rise time	V_{I} = 20 mV, R_{L} = 2 k Ω C_{L} = 100 pF; see Figure 1		0.3		μs
	Overshoot factor			5%		
SR	Slew rate at unity gain	$V_I = 10 \text{ V}, R_L = 2 \text{ k}\Omega$ $C_L = 100 \text{ pF}; \text{ see Figure 1}$		0.5		V/µs

⁽²⁾ All characteristics are measured under open-loop conditions with zero common-mode voltage unless otherwise specified.

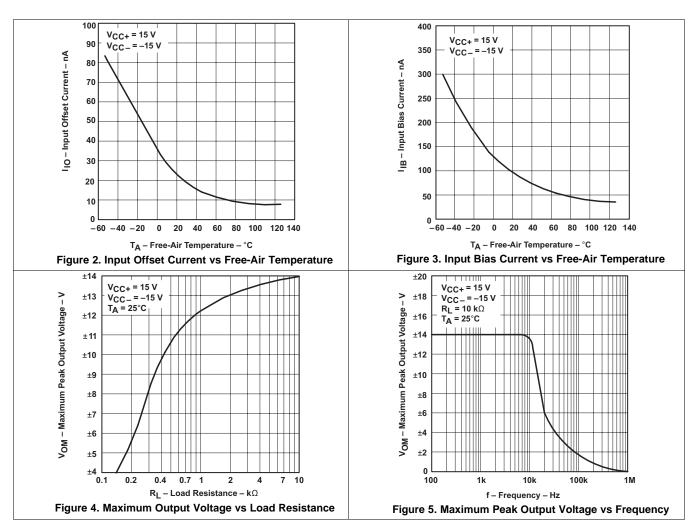


6.8 Typical Characteristics

Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



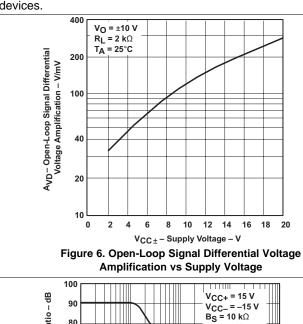
TEST CIRCUIT Figure 1. Rise Time, Overshoot, and Slew Rate





Typical Characteristics (continued)

Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



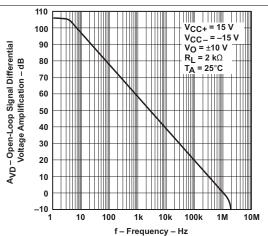
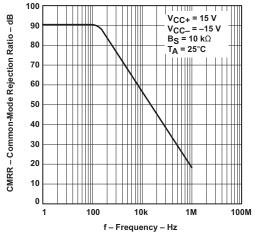


Figure 7. Open-Loop Large-Signal Differential Voltage Amplification vs Frequency



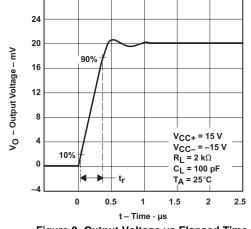
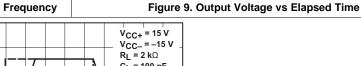


Figure 8. Common-Mode Rejection Ratio vs Frequency



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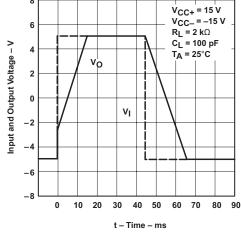


Figure 10. Voltage-Follower Large-Signal Pulse Response

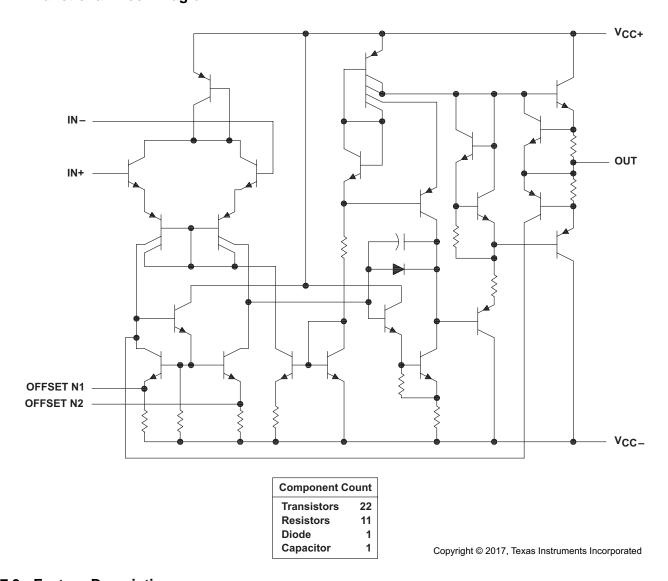


7 Detailed Description

7.1 Overview

The μ A741 has been a popular operational amplifier for over four decades. Typical open loop gain is 106 dB while driving a 2000- Ω load. Short circuit tolerance, offset voltage trimming, and unity-gain stability makes the μ A741 useful for many applications.

7.2 Functional Block Diagram



7.3 Feature Description

7.3.1 Offset-Voltage Null Capability

The input offset voltage of operational amplifiers (op amps) arises from unavoidable mismatches in the differential input stage of the op-amp circuit caused by mismatched transistor pairs, collector currents, current-gain betas (β), collector or emitter resistors and so forth. The input offset pins allow the designer to adjust for mismatches caused by external circuitry. See *Application and Implementation* for more details on design techniques.



Feature Description (continued)

7.3.2 Slew Rate

The slew rate is the rate at which an operational amplifier can change an output when there is a change on the input. The µA741 device has a 0.5-V/µs slew rate. Parameters that vary significantly with operating voltages or temperature are shown in Typical Characteristics.

7.4 Device Functional Modes

The µA741 device is powered on when the power supply is connected. The device can operate as a singlesupply or dual-supply operational amplifier depending on the application.

7.5 µA741Y Chip Information

When properly assembled, this chip displays characteristics similar to the µA741C device. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips can be mounted with conductive epoxy or a gold-silicon preform.

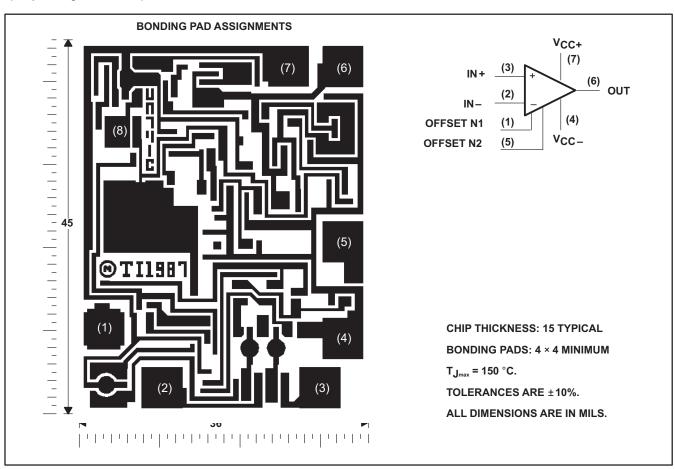


Figure 11. Bonding Pad Assignments

Product Folder Links: uA741



8 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

The input offset voltage of operational amplifiers (op amps) arises from unavoidable mismatches in the differential input stage of the op-amp circuit caused by mismatched transistor pairs, collector currents, current-gain betas (β), collector or emitter resistors and so forth. The input offset pins allow the designer to adjust for mismatches resulting from external circuitry. These input mismatches can be adjusted by placing resistors or a potentiometer between the inputs as shown in Figure 12. A potentiometer can fine-tune the circuit during testing or for applications which require precision offset control. For more information about designing using the input-offset pins, see *Nulling Input Offset Voltage of Operational Amplifiers*.

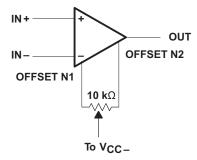


Figure 12. Input Offset Voltage Null Circuit

8.2 Typical Application

The voltage follower configuration of the operational amplifier is used for applications where a weak signal drives a relatively high current load. This circuit is also called a buffer amplifier or unity-gain amplifier. The inputs of an operational amplifier have a very high resistance which puts a negligible current load on the voltage source. The output resistance of the operational amplifier is almost negligible, so the resistance can provide as much current as necessary to the output load.

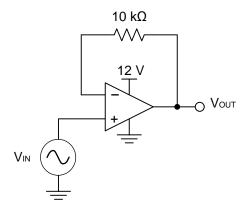


Figure 13. Voltage Follower Schematic



Typical Application (continued)

8.2.1 Design Requirements

- Output range from 2 V to 11.5 V
- Input range from 2 V to 11.5 V
- Resistive feedback to negative input

8.2.2 Detailed Design Procedure

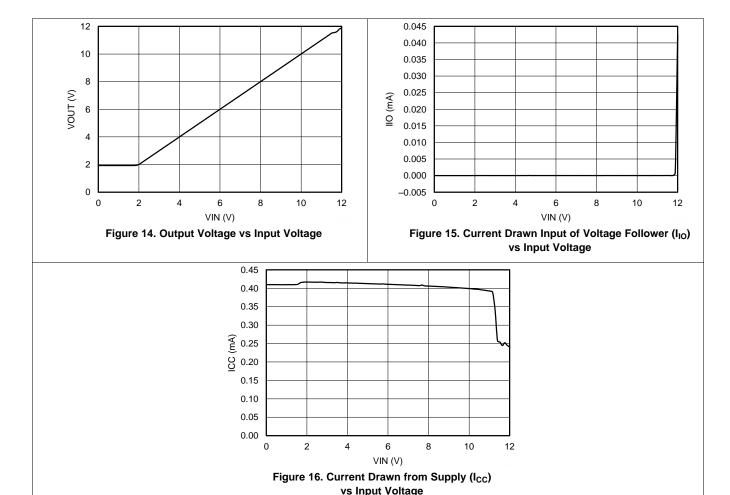
8.2.2.1 Output Voltage Swing

The output voltage of an operational amplifier is limited by the internal circuitry to some level below the supply rails. For this amplifier, the output voltage swing is within ±12 V, which accommodates the input and output voltage requirements.

8.2.2.2 Supply and Input Voltage

For correct operation of the amplifier, neither input must be higher than the recommended positive supply rail voltage or lower than the recommended negative supply rail voltage. The selected amplifier must be able to operate at the supply voltage that accommodates the inputs. Because the input for this application goes up to 11.5 V, the supply voltage must be 12 V. Using a negative voltage on the lower rail rather than ground allows the amplifier to maintain linearity for inputs below 2 V.

8.2.3 Application Curves for Output Characteristics





9 Power Supply Recommendations

The μ A741 device is specified for operation from ±5 to ±15 V; many specifications apply from 0°C to 70°C. *Typical Characteristics* presents parameters that can exhibit significant variance with regard to operating voltage or temperature.

Place 0.1-μF bypass capacitors close to the power-supply pins to reduce errors coupling in from noisy or high impedance power supplies. For more detailed information on bypass capacitor placement, see *Layout Guidelines*.

CAUTION

Supply voltages larger than ±18 V can permanently damage the device (see *Absolute Maximum Ratings*).

10 Layout

10.1 Layout Guidelines

For best operational performance of the device, use good PCB layout practices, including:

- Noise can propagate into analog circuitry through the power pins of the circuit as a whole and the operational
 amplifier. Bypass capacitors reduce the coupled noise by providing low impedance power sources local to the
 analog circuitry.
 - Connect low-ESR, 0.1-μF ceramic bypass capacitors between each supply pin and ground, placed as close as possible to the device. A single bypass capacitor from V+ to ground is applicable for singlesupply applications.
- Separate grounding for analog and digital portions of circuitry is one of the simplest and most-effective
 methods of noise suppression. One or more layers on multilayer PCBs are usually devoted to ground planes.
 A ground plane helps distribute heat and reduces EMI noise pickup. Make sure to physically separate digital
 and analog grounds, paying attention to the flow of the ground current. For more detailed information, see
 Circuit Board Layout Techniques.
- To reduce parasitic coupling, run the input traces as far away as possible from the supply or output traces. If
 it is not possible to keep them separate, it is much better to cross the sensitive trace perpendicular as
 opposed to in parallel with the noisy trace.
- Place the external components as close as possible to the device. Keeping RF and RG close to the inverting input minimizes parasitic capacitance, as shown in *Layout Example*.
- Keep the length of input traces as short as possible. Always remember that the input traces are the most sensitive part of the circuit.
- Consider a driven, low-impedance guard ring around the critical traces. A guard ring can significantly reduce leakage currents from nearby traces that are at different potentials.

10.2 Layout Example

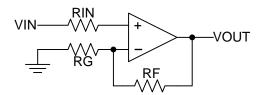


Figure 17. Operational Amplifier Schematic for Noninverting Configuration

Product Folder Links: uA741



Layout Example (continued)

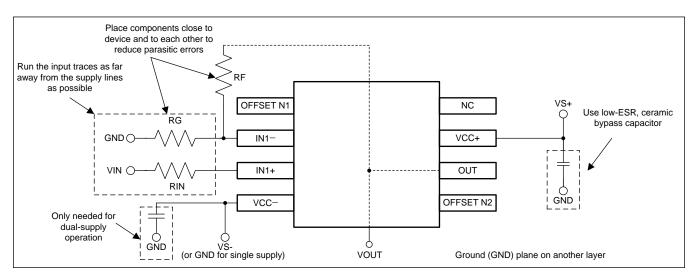


Figure 18. Operational Amplifier Board Layout for Noninverting Configuration



11 Device and Documentation Support

11.1 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

11.2 Trademarks

All trademarks are the property of their respective owners.

11.3 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

11.4 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser based versions of this data sheet, refer to the left hand navigation.



www.ti.com 30-Jul-2024

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
UA741CD	OBSOLETE	SOIC	D	8		TBD	Call TI	Call TI	0 to 70	UA741C	
UA741CDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	UA741C	Samples
UA741CP	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	UA741CP	Samples
UA741CPE4	ACTIVE	PDIP	Р	8	50	TBD	Call TI	Call TI	0 to 70		Samples
UA741CPSR	ACTIVE	SO	PS	8	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	U741	Samples
UA741CPSRE4	ACTIVE	SO	PS	8	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	U741	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.



PACKAGE OPTION ADDENDUM

www.ti.com 30-Jul-2024

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PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	U	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
UA741CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
UA741CPSR	SO	PS	8	2000	330.0	16.4	8.35	6.6	2.4	12.0	16.0	Q1



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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins SPQ		Length (mm)	Width (mm)	Height (mm)	
UA741CDR	SOIC	D	8	2500	340.5	338.1	20.6	
UA741CPSR	SO	PS	8	2000	356.0	356.0	35.0	

PACKAGE MATERIALS INFORMATION

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TUBE



*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)	
UA741CP	Р	PDIP	8	50	506	13.97	11230	4.32	



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES:

- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- 4. This dimension does not include interlead flash.
- 5. Reference JEDEC registration MS-012, variation AA.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.





NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Body dimensions do not include mold flash or protrusion, not to exceed 0,15.



PS (R-PDSO-G8)

PLASTIC SMALL OUTLINE



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-001 variation BA.



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