

Sample &

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LF412-N

SNOSBH7F-APRIL 1999-REVISED SEPTEMBER 2014

LF412-N Low Offset, Low Drift Dual JFET Input Operational Amplifier

Technical

Documents

1 Features

- Internally Trimmed Offset Voltage: 1 mV (Max)
- Input Offset Voltage Drift: 7 µV/°C (Typ)
- Low Input Bias Current: 50 pA
- Low Input Noise Current: 0.01 pA / VHz
- Wide Gain Bandwidth: 3 MHz (Min)
- High Slew Rate: 10V/µs (Min)
- Low Supply Current: 1.8 mA/Amplifier
- High Input Impedance: 10¹²Ω
- Low Total Harmonic Distortion: ≤0.02%
- Low 1/f Noise Corner: 50 Hz
- Fast Settling Time to 0.01%: 2 µs

Applications 2

- **High Speed Integrators**
- Fast D/A Converters
- Sample and Hold Circuits

3 Description

Tools &

Software

These devices are low cost, high speed, JFET input operational amplifiers with very low input offset voltage and input offset voltage drift. They require low supply current yet maintain a large gain bandwidth product and fast slew rate. In addition, well matched high voltage JFET input devices provide very low input bias and offset currents. The LF412-N dual is pin compatible with the LM1558, allowing designers to immediately upgrade the overall performance of existing designs.

Support &

Community

20

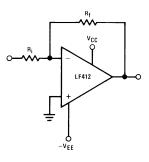
These amplifiers may be used in applications such as high speed integrators, fast D/A converters, sample and hold circuits and many other circuits requiring low input offset voltage and drift, low input bias current, high input impedance, high slew rate and wide bandwidth.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
LF412ACN	PDIP	9.59 mm x 6.35 mm
LF412CN	PDIP	9.59 mm x 6.35 mm
LF412MH	ТО	9.14 mm diameter

(1) For all available packages, see the orderable addendum at the end of the datasheet.

Inverting Amplifier





TEXAS INSTRUMENTS

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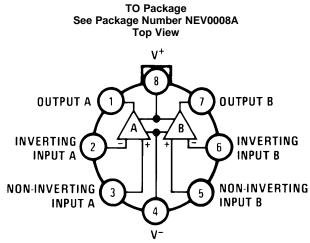
4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision E (March 2014) to Revision F	Page
Updated datasheet to new TI layout	1
Deleted note.	
- Deleted $\Delta V_{OS}/\Delta T$ Max specification for LF412A.	
- Deleted $\Delta V_{OS}/\Delta T$ Max specification for LF412.	
Added Application Note	14
Changes from Revision D (March 2013) to Revision E	Page

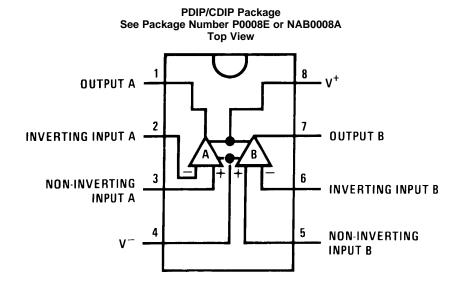
•	Changed layout of National Data Sheet to TI format	14
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5 Pin Configuration and Functions



Note. Pin 4 connected to case.





Pin Functions

PI	PIN I/O		DESCRIPTION
NAME	NO.	1/0	DESCRIPTION
Output A	1	0	Amplifier A Output
Inverting Input A	2	I	Amplifier A Inverting Input
Non-Inverting Input A	3	I	Amplifier A Non-Inverting Input
V-	4	Р	Negative Supply
Non-Inverting Input B	5	I	Amplifier B Non-Inverting Input
Inverting Input B	6	I	Amplifier B Inverting Input
Output B	7	0	Amplifier B Output
V+	8	Р	Positive Supply

6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾⁽²⁾

	LF4	12A	LF412		UNIT
	MIN MAX MIN -22 22 -18 -38 38 -30 Continuous Continuous Continuous TO Package PDIP P See ⁽⁶⁾ 6 150 1	MAX	UNIT		
Supply Voltage	-22	22	-18	18	V
Differential Input Voltage	-38	38	-30	30	V
Input voltage Range ⁽³⁾					
Output Short Circuit Duration ⁽⁴⁾	Continuous		Continuous		
	TO Package		PDIP P	ackage	
Power Dissipation ⁽⁵⁾	Se	e ⁽⁶⁾	670		mW
T _j max	150		1	15	°C
Operating Temp. Range	Se	e ⁽⁷⁾	Se	e ⁽⁷⁾	
Lead Temp. (Soldering, 10 sec.)	2	60	2	60	°C

(1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) Refer to RETS412X for LF412MH and LF412MJ military specifications.

(3) Unless otherwise specified the absolute maximum negative input voltage is equal to the negative power supply voltage.

(4) Any of the amplifier outputs can be shorted to ground indefinitely, however, more than one should not be simultaneously shorted as the maximum junction temperature will be exceeded.

(5) Max. Power Dissipation is defined by the package characteristics. Operating the part near the Max. Power Dissipation may cause the part to operate outside guaranteed limits.

(6) For operating at elevated temperature, these devices must be derated based on a thermal resistance of θ_{iA} .

(7) These devices are available in both the commercial temperature range 0°C≤T_A≤70°C and the military temperature range -55°C≤T_A≤125°C. The temperature range is designated by the position just before the package type in the device number. A "C" indicates the commercial temperature range and an "M" indicates the military temperature range. The military temperature range is available in TO package only. In all cases the maximum operating temperature is limited by internal junction temperature T_i max.

6.2 Handling Ratings

			TO and PD	IP Package		
			MIN			
T _{stg}	Storage temperature rar	ge	-65	150	°C	
	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins ⁽¹⁾	-1700	1700 ⁽²⁾	Ň	
V _(ESD)		Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽³⁾				

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) Human body model, 1.5 k Ω in series with 100 pF.

(3) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	NOM	MAX	UNIT
Supply Voltage LF412A			±20	V
Supply Voltage LF412			±15	V

6.4 Thermal Information

	THERMAL METRIC ⁽¹⁾	TO Package	PDIP Package	UNIT
R_{\thetaJA}	Junction-to-ambient thermal resistance (Typical)	152	115	
R _{0JC(top)}	Junction-to-case (top) thermal resistance			
$R_{\theta JB}$	Junction-to-board thermal resistance			°C/W
Ψ _{JT}	Junction-to-top characterization parameter			°C/VV
Ψ_{JB}	Junction-to-board characterization parameter			
R _{0JC(bot)}	Junction-to-case (bottom) thermal resistance			

(1) For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.

6.5 DC Electrical Characteristics

over operating free-air temperature range (unless otherwise noted)

				L	.F412A ⁽¹⁾		I	_F412 ⁽¹⁾		
	PARAMETER	TEST CON	IDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	UNIT
V _{OS}	Input Offset Voltage	$R_S=10 \text{ k}\Omega, T_A=25^{\circ}$	С		0.5	1.0		1.0	3.0	mV
$\Delta V_{OS}/\Delta$ T	Average TC of Input Offset Voltage	R _S =10 kΩ			7			7		µV/°C
			T _j =25°C		25	100		25	100	pА
I _{OS}	Input Offset Current	$V_{S}=\pm 15V^{(1)(2)}$ $V_{S}=\pm 15V^{(1)(2)}$ $T_{j}=25^{\circ}C$ $R_{L}=2k, T_{A}=25^{\circ}C, V_{S}$	T _j =70°C			2			2	nA
I _B Inp			T _j =125°C			25			25	nA
			T _j =25°C		50	200		50	200	pА
I _B	Input Bias Current	$V_{S}=\pm 15V^{(1)(2)}$	T _j =70°C			4			4	nA
			T _j =125°C			50			50	nA
R _{IN}	Input Resistance	T _j =25°C			10 ¹²			10 ¹²		Ω
A _{VOL}	Large Signal	R _L =2k, T _A =25°C, V V _O =±10V	R _L =2k, T _A =25°C, V _S =±15V,		200		25	200		V/mV
	Voltage Gain	Over Temperature		25	200		15	200		
Vo	Output Voltage Swing	$V_S=\pm 15V, R_L=10k$		±12	±13.5		±12	±13.5		V
V	Input Common-Mode			±16	+19.5		±11	+14.5		V
V _{CM}	Voltage Range				-16.5			-11.5		V
CMRR	Common-Mode Rejection Ratio	R _S ≤10k		80	100		70	100		dB
PSRR	Supply Voltage Rejection Ratio	See ⁽³⁾		80	100		70	100		dB
ls	Supply Current	$V_0 = 0V, R_L = \infty$			3.6	5.6		3.6	6.5	mA

(1) Unless otherwise specified, the specifications apply over the full temperature range and for $V_S=\pm 20V$ for the LF412A and for $V_S=\pm 15V$ for the LF412. V_{OS} , I_B , and I_{OS} are measured at $V_{CM}=0$.

(2) The input bias currents are junction leakage currents which approximately double for every 10°C increase in the junction temperature, T_j. Due to limited production test time, the input bias currents measured are correlated to junction temperature. In normal operation the junction temperature rises above the ambient temperature as a result of internal power dissipation, P_D. T_j=T_A+θ_{jA} P_D where θ_{jA} is the thermal resistance from junction to ambient. Use of a heat sink is recommended if input bias current is to be kept to a minimum.

(3) Supply voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously in accordance with common practice. $V_S = \pm 6V$ to $\pm 15V$.

6.6 AC Electrical Characteristics

over operating free-air temperature range (unless otherwise noted)

PARAMETER		LF412A ⁽¹⁾			L	UNIT		
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	UNIT
Amplifier to Amplifier Coupling	T _A =25°C, f=1 Hz-20 kHz (Input Referred)		-120			-120		dB

(1) Unless otherwise specified, the specifications apply over the full temperature range and for $V_S=\pm 20V$ for the LF412A and for $V_S=\pm 15V$ for the LF412. V_{OS} , I_B , and I_{OS} are measured at $V_{CM}=0$.



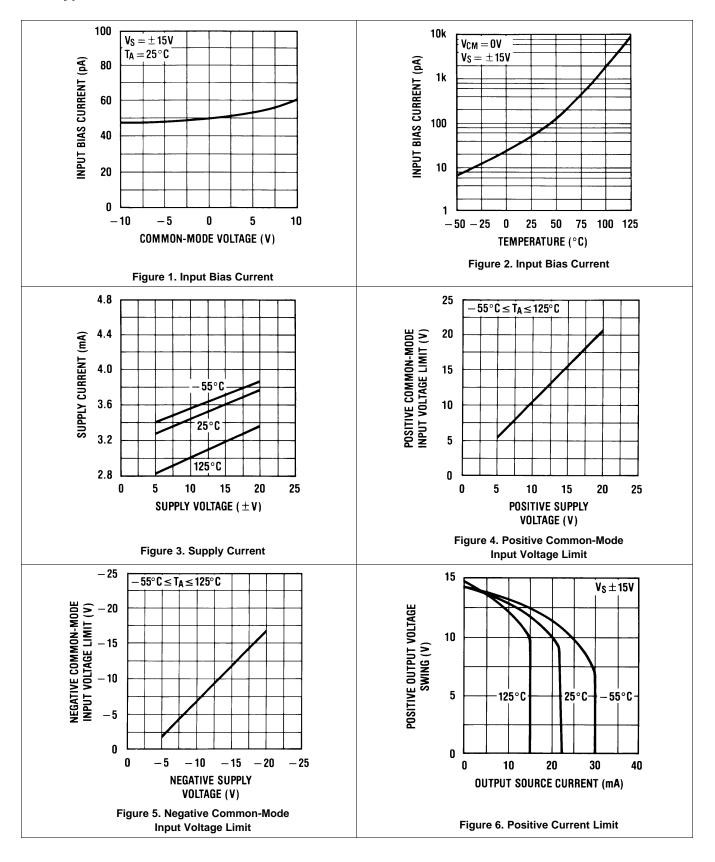
AC Electrical Characteristics (continued)

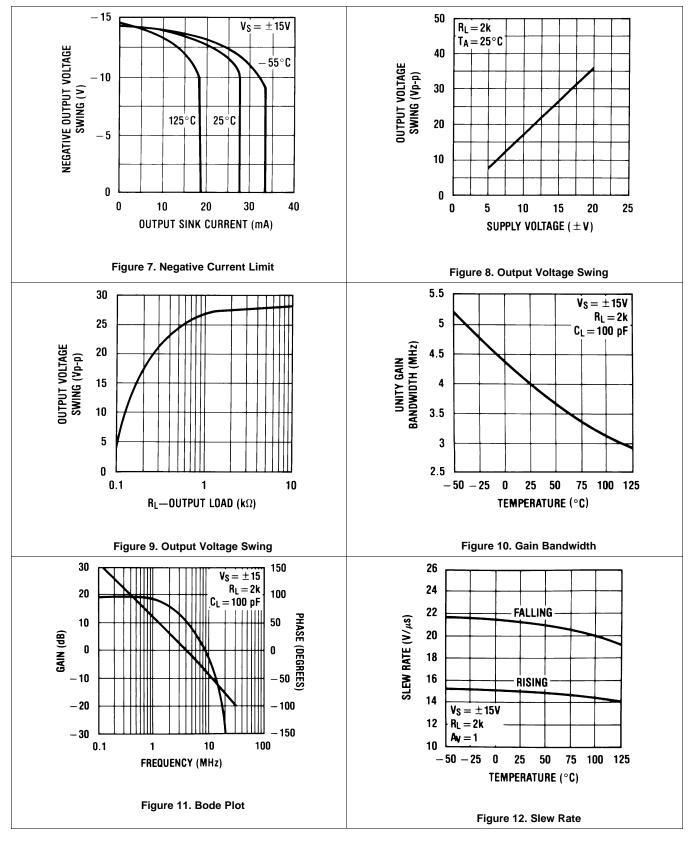
over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	LF412A ⁽¹⁾			LF412 ⁽¹⁾			LINUT
		TEST CONDITIONS	MIN	MIN TYP MAX		MIN TYP		MAX	UNIT
SR	Slew Rate	$V_S=\pm 15V, T_A=25^{\circ}C$	10	15		8	15		V/µs
GBW	Gain-Bandwidth Product	V _S =±15V, T _A =25°C	3	4		2.7	4		MHz
THD	Total Harmonic Dist	A _V =+10, R _L =10k, V _O =20 Vp-p, BW=20 Hz-20 kHz		≤0.02%		≤0.02%			
e _n	Equivalent Input Noise Voltage	$T_A=25^{\circ}C, R_S=100\Omega, f=1 \text{ kHz}$	25 25			nV / √Hz			
i _n	Equivalent Input Noise Current	T _A =25°C, f=1 kHz	0.01			0.01		pA / √Hz	

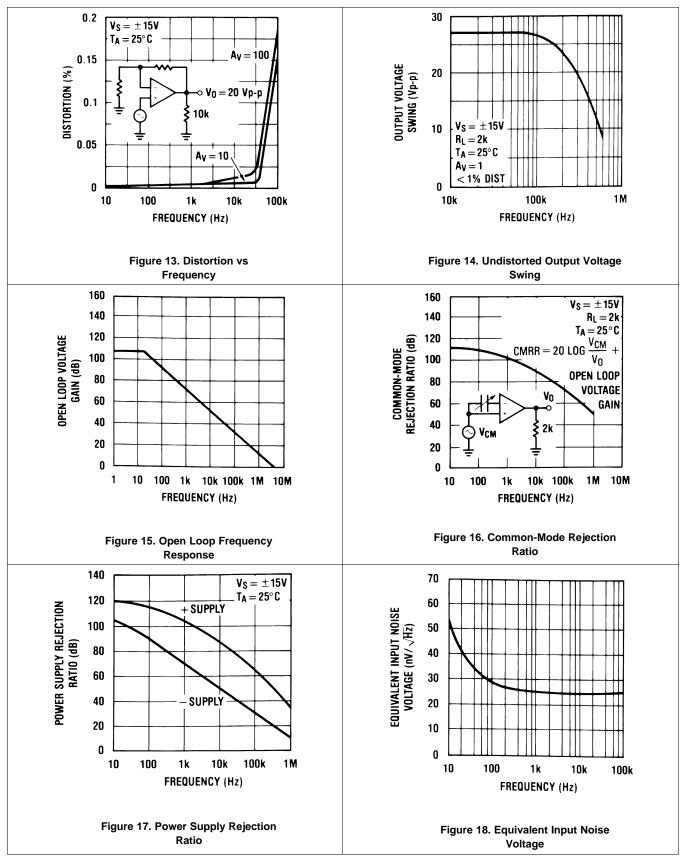


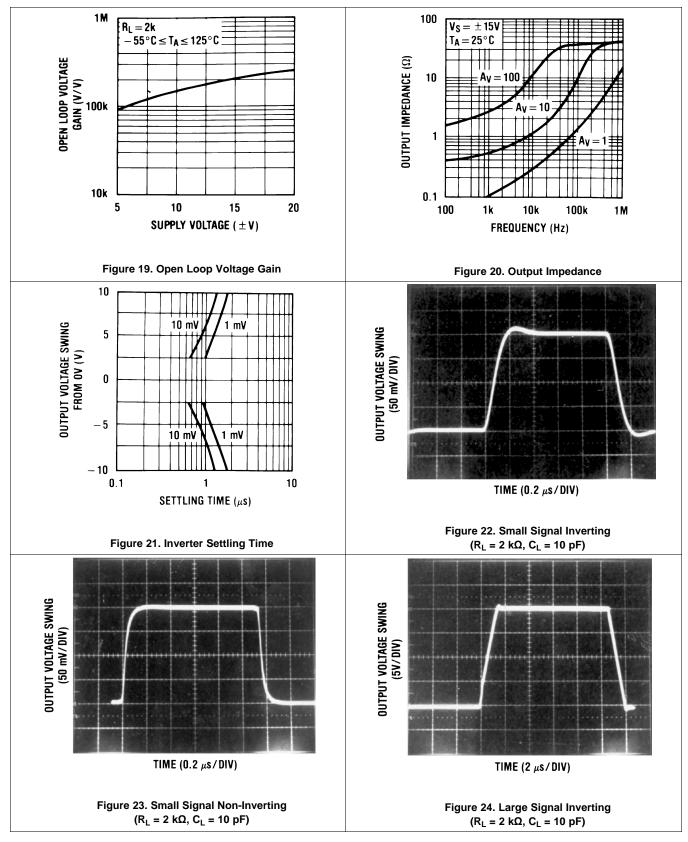
6.7 Typical Characteristics



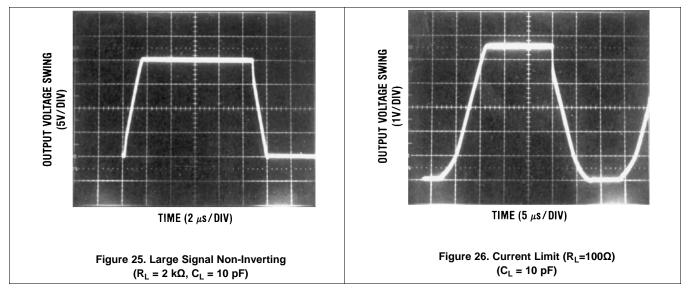












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7 Detailed Description

7.1 Overview

The LF412 devices are low cost, high speed, JFET input operational amplifiers with very low input offset voltage and input offset voltage drift. They require low supply current yet maintain a large gain bandwidth product and fast slew rate. In addition, well matched high voltage JFET input devices provide very low input bias and offset currents. The LF412-N dual is pin compatible with the LM1558, allowing designers to immediately upgrade the overall performance of existing designs.

These amplifiers may be used in applications such as high speed integrators, fast D/A converters, sample and hold circuits and many other circuits requiring low input offset voltage and drift, low input bias current, high input impedance, high slew rate and wide bandwidth.

7.2 Functional Block Diagram

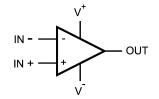


Figure 27. Each Amplifier

7.3 Feature Description

The amplifier's differential inputs consist of a non-inverting input (+IN) and an inverting input (-IN). The amplifier amplifies only the difference in voltage between the two inputs, which is called the differential input voltage. The output voltage of the op-amp V_{OUT} is given by the equation $V_{OUT} = A_{OL}(IN + -IN)$.



7.4 Device Functional Modes

7.4.1 Input and Output Stage

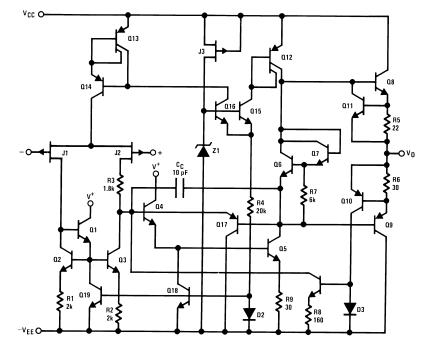


Figure 28. 1/2 Dual LF412

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8 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

The LF412-N series of JFET input dual op amps are internally trimmed (BI-FET II[™]) providing very low input offset voltages and input offset voltage drift. These JFETs have large reverse breakdown voltages from gate to source and drain eliminating the need for clamps across the inputs. Therefore, large differential input voltages can easily be accommodated without a large increase in input current. The maximum differential input voltage is independent of the supply voltages. However, neither of the input voltages should be allowed to exceed the negative supply as this will cause large currents to flow which can result in a destroyed unit.

8.2 Typical Application

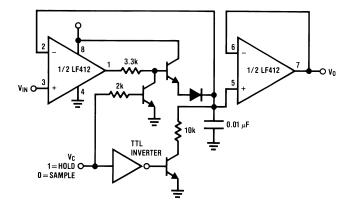


Figure 29. Single Supply Sample and Hold

8.2.1 Design Requirements

Single supply.

8.2.2 Detailed Design Procedure

Exceeding the negative common-mode limit on either input will cause a reversal of the phase to the output and force the amplifier output to the corresponding high or low state.

Exceeding the negative common-mode limit on both inputs will force the amplifier output to a high state. In neither case does a latch occur since raising the input back within the common-mode range again puts the input stage and thus the amplifier in a normal operating mode.

Exceeding the positive common-mode limit on a single input will not change the phase of the output, however, if both inputs exceed the limit, the output of the amplifier may be forced to a high state.

The amplifiers will operate with a common-mode input voltage equal to the positive supply; however, the gain bandwidth and slew rate may be decreased in this condition. When the negative common-mode voltage swings to within 3V of the negative supply, an increase in input offset voltage may occur.

Each amplifier is individually biased by a zener reference which allows normal circuit operation on $\pm 6.0V$ power supplies. Supply voltages less than these may result in lower gain bandwidth and slew rate.

The amplifiers will drive a 2 k Ω load resistance to ±10V over the full temperature range. If the amplifier is forced to drive heavier load currents, however, an increase in input offset voltage may occur on the negative voltage swing and finally reach an active current limit on both positive and negative swings.

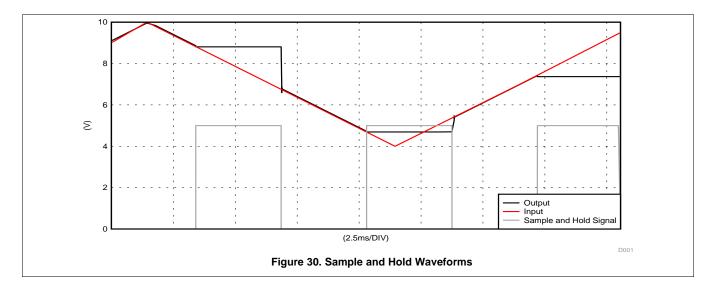


Typical Application (continued)

Precautions should be taken to ensure that the power supply for the integrated circuit never becomes reversed in polarity or that the unit is not inadvertently installed backwards in a socket as an unlimited current surge through the resulting forward diode within the IC could cause fusing of the internal conductors and result in a destroyed unit.

As with most amplifiers, care should be taken with lead dress, component placement and supply decoupling in order to ensure stability. For example, resistors from the output to an input should be placed with the body close to the input to minimize "pick-up" and maximize the frequency of the feedback pole by minimizing the capacitance from the input to ground.

A feedback pole is created when the feedback around any amplifier is resistive. The parallel resistance and capacitance from the input of the device (usually the inverting input) to AC ground set the frequency of the pole. In many instances the frequency of this pole is much greater than the expected 3 dB frequency of the closed loop gain and consequently there is negligible effect on stability margin. However, if the feedback pole is less than approximately 6 times the expected 3 dB frequency a lead capacitor should be placed from the output to the input of the op amp. The value of the added capacitor should be such that the RC time constant of this capacitor and the resistance it parallels is greater than or equal to the original feedback pole time constant.



8.2.3 Application Curves



9 Power Supply Recommendations

For proper operation, the power supplies must be properly decoupled. For decoupling the supply lines it is suggested that 0.1μ F capacitors be placed as close as possible to the op amp power supply pins. The minimum power supply voltage is ±5V.

10 Layout

10.1 Layout Guidelines

As with most amplifiers, care should be taken with lead dress, component placement and supply decoupling in order to ensure stability. For example, resistors from the output to an input should be placed with the body close to the input to minimize "pick-up" and maximize the frequency of the feedback pole by minimizing the capacitance from the input to ground.

10.2 Layout Example

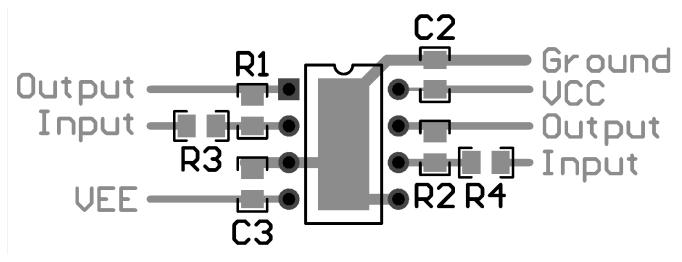


Figure 31. LF412 Layout



11 Device and Documentation Support

11.1 Trademarks

BI-FET II is a trademark of Texas Instruments. All other trademarks are the property of their respective owners.

11.2 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

11.3 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
							(6)				
LF412ACN/NOPB	ACTIVE	PDIP	Ρ	8	40	RoHS & Green	NIPDAU	Level-1-NA-UNLIM	0 to 70	LF 412ACN	Samples
LF412CN/NOPB	ACTIVE	PDIP	Р	8	40	RoHS & Green	NIPDAU	Level-1-NA-UNLIM	0 to 70	LF 412CN	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE OPTION ADDENDUM

25-Jun-2022



11-Feb-2022

TUBE



*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	Τ (μm)	B (mm)
LF412ACN/NOPB	Р	PDIP	8	40	502	14	11938	4.32
LF412CN/NOPB	Р	PDIP	8	40	502	14	11938	4.32

P(R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



- A. All linear dimensions are in inches (millimeters).B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-001 variation BA.



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