LM138 and LM338 5-Amp Adjustable Regulators

1 Features
• Specified 7-A Peak Output Current
• Specified 5-A Output Current
• Adjustable Output Down to 1.2 V
• Specified Thermal Regulation
• Current Limit Constant With Temperature
• P+ Product Enhancement Tested
• Output is Short-Circuit Protected

2 Applications
• Adjustable Power Supplies
• Constant Current Regulators
• Battery Chargers

3 Description
The LM138 series of adjustable 3-terminal positive voltage regulators is capable of supplying in excess of 5 A over a 1.2-V to 32-V output range. They are exceptionally easy to use and require only 2 resistors to set the output voltage. Careful circuit design has resulted in outstanding load and line regulation, comparable to many commercial power supplies. The LM138 family is supplied in a standard 3-lead transistor package.

A unique feature of the LM138 family is time-dependent current limiting. The current limit circuitry allows peak currents of up to 12 A to be drawn from the regulator for short periods of time. This allows the LM138 to be used with heavy transient loads and speeds start-up under full-load conditions. Under sustained loading conditions, the current limit decreases to a safe value protecting the regulator. Also included on the chip are thermal overload protection and safe area protection for the power transistor. Overload protection remains functional even if the adjustment (ADJ) pin is accidentally disconnected.

Normally, no capacitors are needed unless the device is situated more than 6 inches from the input filter capacitors in which case an input bypass is needed. An output capacitor can be added to improve transient response, while bypassing the adjustment pin increases the ripple rejection of the regulator.

Besides replacing fixed regulators or discrete designs, the LM138 is useful in a wide variety of other applications. Because the regulator is floating and receives only the input-to-output differential voltage, supplies of several hundred volts can be regulated as long as the maximum input to output differential is not exceeded; do not short-circuit output to ground. The part numbers in the LM138 series which have a K suffix are packaged in a standard steel TO-CAN package, while those with a T suffix are packaged in a TO-220 plastic package. The LM138 is rated for TJ = –55°C to 150°C, and the LM338 is rated for TJ = 0°C to 125°C.

Device Information

<table>
<thead>
<tr>
<th>PART NUMBER</th>
<th>PACKAGE</th>
<th>BODY SIZE (NOM)</th>
</tr>
</thead>
<tbody>
<tr>
<td>LM138</td>
<td>TO-CAN (2)</td>
<td>25.40 mm × 38.94 mm</td>
</tr>
<tr>
<td>LM338</td>
<td>TO-220 (3)</td>
<td>10.16 mm × 14.986 mm</td>
</tr>
<tr>
<td></td>
<td>TO-CAN (2)</td>
<td>25.40 mm × 38.94 mm</td>
</tr>
</tbody>
</table>

(1) For all available packages, see the orderable addendum at the end of the data sheet.
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4 Revision History
NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision B (April 2013) to Revision C Page

• Added Device Information table, Pin Configuration and Functions section, Specifications section, Recommended Operating Conditions table, Thermal Information table, Detailed Description section, Application and Implementation section, Power Supply Recommendations section, Layout section, Device and Documentation Support section, and Mechanical, Packaging, and Orderable Information section .......................................................... 1
• Deleted RETS138K military specification reference from Absolute Maximum Ratings table .................................................................................... 3
• Changed Junction to Ambient, \( R_{\text{JA}} \), value in Thermal Information table From: 50°C/W To: 22.9°C/W (NDE) ................. 4
• Changed Junction to Case, \( R_{\text{JC(top)}} \), value in Thermal Information table From: 4°C/W To: 15.7°C/W (NDE) ....................... 4

Changes from Revision A (April 2013) to Revision B Page

• Changed layout of National Semiconductor Data Sheet to TI format .......................................................... 1
5 Pin Configuration and Functions

![Pin Configuration Diagram]

Pin Functions

<table>
<thead>
<tr>
<th>PIN</th>
<th>I/O</th>
<th>DESCRIPTION</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADJ</td>
<td>1</td>
<td>Output voltage adjustment pin. Connect to a resistor divider to set $V_O$</td>
</tr>
<tr>
<td>$V_{IN}$</td>
<td>3</td>
<td>Supply input pin</td>
</tr>
<tr>
<td>$V_{OUT}$</td>
<td>Case</td>
<td>Voltage output pin</td>
</tr>
</tbody>
</table>

6 Specifications

6.1 Absolute Maximum Ratings

<table>
<thead>
<tr>
<th>MIN</th>
<th>MAX</th>
<th>UNIT</th>
</tr>
</thead>
<tbody>
<tr>
<td>$-0.3$</td>
<td>40</td>
<td>V</td>
</tr>
<tr>
<td>Internally limited</td>
<td></td>
<td></td>
</tr>
<tr>
<td>$300$</td>
<td></td>
<td>°C</td>
</tr>
<tr>
<td>$260$</td>
<td></td>
<td>°C</td>
</tr>
<tr>
<td>$-55$</td>
<td>150</td>
<td>°C</td>
</tr>
<tr>
<td>$0$</td>
<td>125</td>
<td>°C</td>
</tr>
<tr>
<td>$-65$</td>
<td>150</td>
<td>°C</td>
</tr>
</tbody>
</table>

(1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.2 Recommended Operating Conditions

<table>
<thead>
<tr>
<th>MIN</th>
<th>MAX</th>
<th>UNIT</th>
</tr>
</thead>
<tbody>
<tr>
<td>3</td>
<td>40</td>
<td>V</td>
</tr>
<tr>
<td>5</td>
<td></td>
<td>A</td>
</tr>
</tbody>
</table>

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6.3 Thermal Information

<table>
<thead>
<tr>
<th>THERMAL METRIC(1)</th>
<th>LM338</th>
<th>LM338</th>
<th>UNIT</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>NDS (TO-CAN)</td>
<td>NDE (TO-220)</td>
<td>NDS (TO-CAN)</td>
</tr>
<tr>
<td></td>
<td>2 PINS</td>
<td>3 PINS</td>
<td>2 PINS</td>
</tr>
<tr>
<td>$R_{JJA}$ Junction-to-ambient thermal resistance</td>
<td>35</td>
<td>22.9</td>
<td>35</td>
</tr>
<tr>
<td>$R_{JJC(top)}$ Junction-to-case (top) thermal resistance</td>
<td>1</td>
<td>15.7</td>
<td>1</td>
</tr>
<tr>
<td>$R_{JJB}$ Junction-to-board thermal resistance</td>
<td>—</td>
<td>4.1</td>
<td>—</td>
</tr>
<tr>
<td>$\psi_{JT}$ Junction-to-top characterization parameter</td>
<td>—</td>
<td>2.1</td>
<td>—</td>
</tr>
<tr>
<td>$\psi_{JB}$ Junction-to-board characterization parameter</td>
<td>—</td>
<td>4.1</td>
<td>—</td>
</tr>
<tr>
<td>$R_{JJC(bot)}$ Junction-to-case (bottom) thermal resistance</td>
<td>—</td>
<td>0.7</td>
<td>—</td>
</tr>
</tbody>
</table>

(1) For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

6.4 Electrical Characteristics: LM138

Values apply for $T_J = 25^\circ C$; $V_{IN} - V_{OUT} = 5$ V; and $I_{OUT} = 10$ mA (unless otherwise noted).(1)

<table>
<thead>
<tr>
<th>PARAMETER</th>
<th>TEST CONDITIONS</th>
<th>MIN</th>
<th>TYP</th>
<th>MAX</th>
<th>UNIT</th>
</tr>
</thead>
<tbody>
<tr>
<td>$V_{REF}$ Reference voltage</td>
<td>$3 , V \leq (V_{IN} - V_{OUT}) \leq 35 , V$, $10 , mA \leq I_{OUT} \leq 5 , A$, $P \leq 50 , W$, $T_J = -55^\circ C$ to $150^\circ C$</td>
<td>1.19</td>
<td>1.24</td>
<td>1.29</td>
<td>V</td>
</tr>
<tr>
<td>$V_{RLINE}$ Line regulation</td>
<td>$3 , V \leq (V_{IN} - V_{OUT}) \leq 35 , V$</td>
<td>$T_J = 25^\circ C$</td>
<td>0.005%</td>
<td>0.01%</td>
<td>V</td>
</tr>
<tr>
<td>$V_{RLOAD}$ Load regulation</td>
<td>$10 , mA \leq I_{OUT} \leq 5 , A$</td>
<td>$T_J = 25^\circ C$</td>
<td>0.02%</td>
<td>0.04%</td>
<td>V</td>
</tr>
<tr>
<td>Thermal regulation</td>
<td>20 ms pulse</td>
<td>0.002%</td>
<td>0.01%</td>
<td>W</td>
<td></td>
</tr>
<tr>
<td>$I_{ADJ}$ Adjustment pin current</td>
<td>$T_J = -55^\circ C$ to $150^\circ C$</td>
<td>45</td>
<td>100</td>
<td>µA</td>
<td></td>
</tr>
<tr>
<td>$\Delta I_{ADJ}$ Adjustment pin current change</td>
<td>$10 , mA \leq I_{OUT} \leq 5 , A$, $3 , V \leq (V_{IN} - V_{OUT}) \leq 35 , V$, $T_J = -55^\circ C$ to $150^\circ C$</td>
<td>0.2</td>
<td>5</td>
<td>µA</td>
<td></td>
</tr>
<tr>
<td>$\Delta V_{R/T}$ Temperature stability</td>
<td>$T_J = -55^\circ C$ to $150^\circ C$</td>
<td>1%</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>$I_{LOAD(MIN)}$ Minimum load current</td>
<td>$V_{IN} - V_{OUT} = 35$ V, $T_J = -55^\circ C$ to $150^\circ C$</td>
<td>3.5</td>
<td>5</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td>$I_{CL}$ Current limit</td>
<td>$V_{IN} - V_{OUT} \leq 10$ V</td>
<td>DC, $T_J = -55^\circ C$ to $150^\circ C$</td>
<td>5</td>
<td>8</td>
<td>A</td>
</tr>
<tr>
<td>     </td>
<td>0.5-ms peak, $T_J = -55^\circ C$ to $150^\circ C$</td>
<td>7</td>
<td>12</td>
<td>A</td>
<td></td>
</tr>
<tr>
<td>$V_{N}$ RMS output noise (percent of $V_{OUT}$)</td>
<td>10 Hz $\leq f \leq 10$ kHz</td>
<td>0.003%</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>$\Delta V_{R}/\Delta V_{IN}$ Ripple rejection ratio</td>
<td>$V_{OUT} = 10$ V, $f = 120$ Hz, $C_{ADJ} = 0$ µF, $T_J = -55^\circ C$ to $150^\circ C$</td>
<td>60</td>
<td>dB</td>
<td></td>
<td></td>
</tr>
<tr>
<td>   </td>
<td>$V_{OUT} = 10$ V, $f = 120$ Hz, $C_{ADJ} = 10$ µF, $T_J = -55^\circ C$ to $150^\circ C$</td>
<td>60</td>
<td>75</td>
<td>dB</td>
<td></td>
</tr>
<tr>
<td>Long-term stability</td>
<td>$T_J = 125^\circ C$, 1000 Hrs</td>
<td>0.3%</td>
<td>1%</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

(1) These specifications are applicable for power dissipations up to 50 W for the TO-3 (NDS) package and 25 W for the TO-220 (NDE) package. Power dissipation is specified at these values up to 15-V input-output differential. Above 15-V differential, power dissipation is limited by internal protection circuitry. All limits (that is, the numbers in the minimum and maximum columns) are specified to TI’s AOQL (Average Outgoing Quality Level).

(2) Regulation is measured at a constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered under the specifications for thermal regulation.
### 6.5 Electrical Characteristics: LM338

Values apply for $T_J = 25^\circ\text{C}$; $V_{IN} - V_{OUT} = 5 \text{ V}$; and $I_{OUT} = 10 \text{ mA}$ (unless otherwise noted).  

<table>
<thead>
<tr>
<th>PARAMETER</th>
<th>TEST CONDITIONS</th>
<th>MIN</th>
<th>TYP</th>
<th>MAX</th>
</tr>
</thead>
<tbody>
<tr>
<td>$V_{REF}$ Reference voltage</td>
<td>$3 \text{ V} \leq (V_{IN} - V_{OUT}) \leq 35 \text{ V}$, 10 mA $\leq I_{OUT} \leq 5 \text{ A}$, $P \leq 50 \text{ W}$, $T_J = 0^\circ\text{C}$ to $125^\circ\text{C}$</td>
<td>1.19</td>
<td>1.24</td>
<td>1.29</td>
</tr>
<tr>
<td>$V_{R\text{LINE}}$ Line regulation</td>
<td>$3 \text{ V} \leq (V_{IN} - V_{OUT}) \leq 35 \text{ V}$</td>
<td>$T_J = 25^\circ\text{C}$</td>
<td>0.005%</td>
<td>0.03%</td>
</tr>
<tr>
<td></td>
<td>$T_J = 0^\circ\text{C}$ to $125^\circ\text{C}$</td>
<td>0.02%</td>
<td>0.06%</td>
<td>V</td>
</tr>
<tr>
<td>$V_{R\text{LOAD}}$ Load regulation</td>
<td>$10 \text{ mA} \leq I_{OUT} \leq 5 \text{ A}$</td>
<td>$T_J = 25^\circ\text{C}$</td>
<td>0.1</td>
<td>0.5</td>
</tr>
<tr>
<td></td>
<td>$T_J = 0^\circ\text{C}$ to $125^\circ\text{C}$</td>
<td>0.3</td>
<td>1</td>
<td></td>
</tr>
<tr>
<td>Thermal regulation</td>
<td>20-ms pulse</td>
<td>0.002%</td>
<td>0.02%</td>
<td>W</td>
</tr>
<tr>
<td>$I_{ADJ}$ Adjustment pin current</td>
<td>$T_J = 0^\circ\text{C}$ to $125^\circ\text{C}$</td>
<td>45</td>
<td>100</td>
<td>μA</td>
</tr>
<tr>
<td>$\Delta I_{ADJ}$ Adjustment pin current change</td>
<td>$10 \text{ mA} \leq I_{OUT} \leq 5 \text{ A}$, $3 \text{ V} \leq (V_{IN} - V_{OUT}) \leq 35 \text{ V}$, $T_J = 0^\circ\text{C}$ to $125^\circ\text{C}$</td>
<td>0.2</td>
<td>5</td>
<td>μA</td>
</tr>
<tr>
<td>$\Delta V_{R/T}$ Temperature stability</td>
<td>$T_J = 0^\circ\text{C}$ to $125^\circ\text{C}$</td>
<td>1</td>
<td></td>
<td></td>
</tr>
<tr>
<td>$I_{LOAD(MIN)}$ Minimum load current</td>
<td>$V_{IN} - V_{OUT} = 35 \text{ V}$, $T_J = 0^\circ\text{C}$ to $125^\circ\text{C}$</td>
<td>3.5</td>
<td>10</td>
<td>mA</td>
</tr>
<tr>
<td>$I_{CL}$ Current limit</td>
<td>$V_{IN} - V_{OUT} = 10 \text{ V}$</td>
<td>DC, $T_J = 0^\circ\text{C}$ to $125^\circ\text{C}$</td>
<td>5</td>
<td>8</td>
</tr>
<tr>
<td></td>
<td>$V_{IN} - V_{OUT} = 30 \text{ V}$</td>
<td>0.5-ms peak, $T_J = 0^\circ\text{C}$ to $125^\circ\text{C}$</td>
<td>7</td>
<td>12</td>
</tr>
<tr>
<td>$V_N$ RMS output noise (percent of $V_{OUT}$)</td>
<td>$10 \text{ Hz} \leq f \leq 10 \text{ kHz}$</td>
<td>0.003%</td>
<td></td>
<td></td>
</tr>
<tr>
<td>$\Delta V_{P/\Delta V_{IN}}$ Ripple rejection ratio</td>
<td>$V_{OUT} = 10 \text{ V}$, $f = 120 \text{ Hz}$, $C_{ADJ} = 0 \text{ µF}$, $T_J = 0^\circ\text{C}$ to $125^\circ\text{C}$</td>
<td>60</td>
<td></td>
<td>dB</td>
</tr>
<tr>
<td></td>
<td>$V_{OUT} = 10 \text{ V}$, $f = 120 \text{ Hz}$, $C_{ADJ} = 10 \text{ µF}$, $T_J = 0^\circ\text{C}$ to $125^\circ\text{C}$</td>
<td>60</td>
<td>75</td>
<td>dB</td>
</tr>
<tr>
<td>Long-term stability</td>
<td>$T_J = 125^\circ\text{C}$, 1000 Hrs</td>
<td>0.3%</td>
<td>1%</td>
<td></td>
</tr>
</tbody>
</table>

(1) These specifications are applicable for power dissipations up to 50 W for the TO-3 (NDS) package and 25 W for the TO-220 (NDE) package. Power dissipation is specified at these values up to 15-V input-output differential. Above 15-V differential, power dissipation is limited by internal protection circuitry. All limits (that is, the numbers in the minimum and maximum columns) are specified to TI’s AOQL (Average Outgoing Quality Level).

(2) Regulation is measured at a constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered under the specifications for thermal regulation.
### 6.6 Typical Characteristics

#### Figure 1. Current Limit

- **Preload Current = 0**
- **Tcase = 25°C**
- **Vin - Vout = 10V**
- **Vin - Vout = 15V**
- **Vin - Vout = 20V**
- **Vin - Vout = 30V**

#### Figure 2. Current Limit

- **Peak Current Limit**
- **DC Current Limit**
- **Tcase = 25°C**
- **Preload = 0**
- **Preload = 3A**
- **Preload = 5A**
- **Preload = 1A**

#### Figure 3. Current Limit

- **Preload = 0**
- **Preload = 1A**
- **Preload = 5A**
- **Preload = 3A**
- **Vin = 10V**
- **Vout = 5V**
- **Tcase = 25°C**

#### Figure 4. Load Regulation

- **Vout = 3A**
- **Vout = 5A**
- **Vin = 15V**
- **Vout = 10V**
- **Preload = 50 mA**

#### Figure 5. Dropout Voltage

- **ΔVout = 100 mV**
- **Vout = 5A**
- **Vout = 3A**
- **Vout = 1A**

#### Figure 6. Adjustment Current

- **Temperature (°C)**
- **Adjustment Current (µA)**
- **75 50 25 150**
Typical Characteristics (continued)

Figure 7. Temperature Stability

Figure 8. Output Impedance

Figure 9. Minimum Operating Current

Figure 10. Ripple Rejection

Figure 11. Ripple Rejection

Figure 12. Ripple Rejection
Typical Characteristics (continued)

Figure 13. Line Transient Response

Figure 14. Load Transient Response
7 Detailed Description

7.1 Overview
The LM138 and LM338 devices are adjustable, three-terminal, positive-voltage regulators capable of supplying more than 5 A over an output-voltage range of 1.2 V to 32 V. It requires only two external resistors to set the output voltage. These devices feature a typical line regulation of 0.005% and typical load regulation of 0.1%. It includes time-dependent current limiting, thermal overload protection, and safe operating area protection. Overload protection remains functional even if the ADJUST terminal is disconnected.

The LM138 and LM338 devices are versatile in their applications, including uses in programmable output regulation and local on-card regulation. Or, by connecting a fixed resistor between the ADJUST and OUTPUT terminals, the LM138 and LM338 devices can function as a precision current regulators. An optional output capacitor can be added to improve transient response. The ADJUST terminal can be bypassed to achieve very high ripple-rejection ratios, which are difficult to achieve with standard three-terminal regulators.

7.2 Functional Block Diagram

![Functional Block Diagram](image)

7.3 Feature Description

7.3.1 NPN Darlington Output Drive
NPN Darlington output topology provides naturally low output impedance and an output capacitor is optional. To support maximum current and lowest temperature, 3-V headroom is recommended (\(V_I - V_O\)).

7.3.2 Overload Block
Overcurrent and overtemperature shutdown protects the device against overload or damage from operating in excessive heat.

7.3.3 Programmable Feedback
Op amp with 1.25-V offset input at the ADJUST terminal provides easy output voltage or current (not both) programming. For current regulation applications, a single resistor whose resistance value is 1.25 V/IO and power rating is greater than 1.25 V^2/R must be used. For voltage regulation applications, two resistors set the output voltage.
7.4 Device Functional Modes

7.4.1 Normal Operation
The device OUTPUT pin sources current necessary to make OUTPUT pin 1.25 V greater than ADJUST terminal to provide output regulation.

7.4.2 Operation With Low Input Voltage
The device requires up to 3-V headroom (V_i – V_O) to operate in regulation. With less headroom, the device may drop out and OUTPUT voltage is INPUT voltage minus drop out voltage.

7.4.3 Operation at Light Loads
The device passes its bias current to the OUTPUT pin. The load or feedback must consume this minimum current for regulation or the output may be too high. A 250-Ω feedback resistor between OUTPUT and ADJUST consumes the worst case minimum load current of 5 mA.

7.4.4 Operation in Self Protection
When an overload occurs, the device shuts down Darlington NPN output stage or reduces the output current to prevent device damage. The device automatically resets from the overload. The output may be reduced or alternate between on and off until the overload is removed.
8 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI’s customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

In operation, the LM138 develops a nominal 1.25-V reference voltage ($V_{REF}$) between the output and adjustment terminal. The reference voltage is impressed across program resistor $R_1$, and, since the voltage is constant, a constant current $I_1$ then flows through the output set resistor $R_2$, giving an output voltage calculated with Equation 1.

$$V_{OUT} = V_{REF}\left(1 + \frac{R_2}{R_2}\right) + I_{ADJ}R_2$$  \hspace{1cm} (1)

![Figure 15. Typical Application Circuit](image)

Because the 50-μA current from the adjustment terminal represents an error term, the LM138 was designed to minimize $I_{ADJ}$ and make it very constant with line and load changes. To do this, all quiescent operating current is returned to the output establishing a minimum load current requirement. If there is insufficient load on the output, the output rises.
8.2 Typical Applications

8.2.1 Constant 5-V Regulator

![Constant 5-V Regulator Diagram]

*Needed if device is more than 6 inches from filter capacitors.
†Optional—improves transient response

\[ V_{OUT} = 1.25 \left( 1 + \frac{R_2}{R_1} \right) f_{ADJ}(R_2) \]

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Figure 16. Constant 5-V Regulator

8.2.1.1 Design Requirements

R1: Because the LM138 produces a typical 1.24 V potential between the OUTPUT and ADJUST pins, placing a 270-Ω resistor between them causes 4.6 mA to flow through R1 and R2.

R2: To achieve a 5-V output, the sum of the voltages across R1 and R2 must equal 5 V. Therefore, \( V_{r2} \) must equal 3.76 V when 4.6 mA is flowing through it. \( R_2 = \frac{V_{r2}}{I} = \frac{3.76 \text{ V}}{4.6 \text{ mA}} = \approx 820 \Omega \).

\( C_{IN} \): 0.1 µF of input capacitance helps filter out unwanted noise, especially if the regulator is located far from the power supply filter capacitors.

\( C_{OUT} \): The regulator is stable without any output capacitance, but adding a 1-µF capacitor improves the transient response.

\( C_{ADJ} \): A 10-µF capacitor bypassing the ADJUST pin to ground improves the regulators ripple rejection.

D1: Protection diode D1 is recommended if \( C_{OUT} \) is used. The diode provides a low-impedance discharge path to prevent the capacitor from discharging into the output of the regulator (see Protection Diodes).

D2: Protection diode D2 is recommended if \( C_{ADJ} \) is used. The diode provides a low-impedance discharge path to prevent the capacitor from discharging into the output of the regulator (see Protection Diodes).

Table 1 lists the design parameters for this typical application.
Table 1. Design Parameters

<table>
<thead>
<tr>
<th>PARAMETER</th>
<th>VALUE</th>
</tr>
</thead>
<tbody>
<tr>
<td>Feedback resistor 1 (R1)</td>
<td>270 Ω</td>
</tr>
<tr>
<td>Feedback resistor 2 (R2)</td>
<td>820 Ω</td>
</tr>
<tr>
<td>Input capacitor (C_{IN})</td>
<td>0.1 µF</td>
</tr>
<tr>
<td>Output capacitor (C_{OUT})</td>
<td>1 µF</td>
</tr>
<tr>
<td>Adjust capacitor(C_{ADJ})</td>
<td>10 µF</td>
</tr>
</tbody>
</table>

8.2.1.2 Detailed Design Procedure

8.2.1.2.1 External Capacitors

An input bypass capacitor is recommended. A 0.1-µF disc or 1-µF solid tantalum on the input is suitable input bypassing for almost all applications. The device is more sensitive to the absence of input bypassing when adjustment or output capacitors are used but the above values eliminate the possibility of problems.

The adjustment terminal can be bypassed to ground on the LM138 to improve ripple rejection. This bypass capacitor prevents ripple from being amplified as the output voltage is increased. With a 10-µF bypass capacitor, 75-dB ripple rejection is obtainable at any output level. Increases over 20 µF do not appreciably improve the ripple rejection at frequencies above 120 Hz. If the bypass capacitor is used, it is sometimes necessary to include protection diodes to prevent the capacitor from discharging through internal low current paths and damaging the device.

In general, the best type of capacitors to use are solid tantalum. Solid tantalum capacitors have low impedance even at high frequencies. Depending upon capacitor construction, it takes about 25 µF in aluminum electrolytic to equal 1-µF solid tantalum at high frequencies. Ceramic capacitors are also good at high frequencies; but some types have a large decrease in capacitance at frequencies around 0.5 MHz. For this reason, 0.01-µF disc may seem to work better than a 0.1-µF disc as a bypass.

Although the LM138 is stable with no output capacitors, like any feedback circuit, certain values of external capacitance can cause excessive ringing. This occurs with values between 500 pF and 5000 pF. A 1-µF solid tantalum (or 25-µF aluminum electrolytic) on the output swamps this effect and insures stability.

8.2.1.2.2 Load Regulation

The LM138 is capable of providing extremely good load regulation but a few precautions are needed to obtain maximum performance. The current set resistor connected between the adjustment terminal and the output terminal (usually 240 Ω) must be tied directly to the output of the regulator (case) rather than near the load. This eliminates line drops from appearing effectively in series with the reference and degrading regulation. For example, a 15-V regulator with 0.05-Ω resistance between the regulator and load has a load regulation due to line resistance of 0.05 Ω × I_L. If the set resistor is connected near the load, the effective line resistance is 0.05 Ω (1 + R2/R1) or in this case, 11.5 times worse.

Figure 17 shows the effect of resistance between the regulator and 240-Ω set resistor.
Figure 17. Regulator With Line Resistance in Output Lead

With the TO-3 package, it is easy to minimize the resistance from the case to the set resistor, by using 2 separate leads to the case. The ground of R2 can be returned near the ground of the load to provide remote ground sensing and improve load regulation.

8.2.1.2.3 Protection Diodes

When external capacitors are used with any IC regulator it is sometimes necessary to add protection diodes to prevent the capacitors from discharging through low current points into the regulator. Most 20-µF capacitors have low enough internal series resistance to deliver 20-A spikes when shorted. Although the surge is short, there is enough energy to damage parts of the IC.

When an output capacitor is connected to a regulator and the input is shorted, the output capacitor discharges into the output of the regulator. The discharge current depends on the value of the capacitor, the output voltage of the regulator, and the rate of decrease of $V_{IN}$. In the LM138 this discharge path is through a large junction that is able to sustain 25-A surge with no problem. This is not true of other types of positive regulators. For output capacitors of 100 µF or less at output of 15 V or less, there is no need to use diodes.

The bypass capacitor on the adjustment terminal can discharge through a low current junction. Discharge occurs when either the input or output is shorted. Internal to the LM138 is a 50-Ω resistor which limits the peak discharge current. No protection is needed for output voltages of 25-V or less and 10-µF capacitance. Figure 18 shows an LM138 with protection diodes included for use with outputs greater than 25 V and high values of output capacitance.
D1 protects against C1
D2 protects against C2

Figure 18. Regulator With Protection Diodes

8.2.1.3 Application Curves

Figure 19. Regulator Start-Up
Figure 20. Regulator Shutdown
Figure 21. Regulator Response to Load Step
8.3 System Examples

Full output current not available at high input-output voltages
†Optional—improves transient response. Output capacitors in the range of 1 µF to 1000 µF of aluminum or tantalum electrolytic are commonly used to provide improved output impedance and rejection of transients.
*Needed if device is more than 6 inches from filter capacitors.
**R1 = 240 Ω for LM138. R1, R2 as an assembly can be ordered from Bourns:
MIL part no. 7105A-AT2-502
COMM part no. 7105A-AT7-502

Figure 22. Regulator and Voltage Reference

Figure 23. 1.2-V to 25-V Adjustable Regulator

Figure 24. Temperature Controller

Figure 25. Precision Power Regulator With Low Temperature Coefficient
System Examples (continued)

Figure 26. Slow Turnon 15-V Regulator

Figure 27. Adjustable Regulator With Improved Ripple Rejection

Figure 28. High Stability 10-V Regulator

Figure 29. Digitally Selected Outputs
System Examples (continued)

* Minimum load—100 mA

** Minimum output = 1.2 V

Figure 30. 15-A Regulator

Figure 31. 5-V Logic Regulator With Electronic Shutdown

Figure 32. Light Controller

Figure 33. 0 to 22-V Regulator

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* R1 = 240 Ω, R2 = 5k for LM138
Full output current not available at high input-output voltages
System Examples (continued)

**Figure 34. 12-V Battery Charger**

**Figure 35. Adjustable Current Regulator**

**Figure 36. Precision Current Limiter**

**Figure 37. 5-A Current Regulator**
System Examples (continued)

Figure 38. Tracking Preregulator

Figure 39. Adjusting Multiple On-Card Regulators With Single Control*

Figure 40. Power Amplifier

Figure 41. Simple 12-V Battery Charger

$A_V = 1, R_F = 10k, C_F = 100 \text{ pF}$

$A_V = 10, R_F = 100k, C_F = 10 \text{ pF}$

Bandwidth $\geq 100 \text{ kHz}$

Distortion $\leq 0.1\%$

† Minimum load—10 mA

* All outputs within $\pm 100 \text{ mV}$

$\text{Use of } R_S \text{ allows low charging rates with fully charged battery.}$

**The 1000 $\mu$F is recommended to filter out input transients

$\text{Use of } R_S \text{ allows low charging rates with fully charged battery.}$

**The 1000 $\mu$F is recommended to filter out input transients

$R_S \text{—sets output impedance of charger } Z_{OUT} = R_S \left(1 + \frac{R_2}{R_1}\right)$

Submit Documentation Feedback

Product Folder Links: LM138, LM338
System Examples (continued)

* Set maximum charge current to 3 A
** The 1000 µF is recommended to filter out input transients.

Figure 42. Adjustable 15-A Regulator

Minimum load—100 mA

Figure 44. 10-A Regulator
9 Power Supply Recommendations

The input supply to LM138 and LM338 must be kept at a voltage level such that its maximum input to output differential voltage rating is not exceeded. The minimum dropout voltage must also be met with extra headroom when possible to keep the LM138 and LM338 in regulation. TI recommends a capacitor be placed at the input to bypass noise.

10 Layout

10.1 Layout Guidelines

Some layout guidelines must be followed to ensure proper regulation of the output voltage with minimum noise. Traces carrying the load current must be wide to reduce the amount of parasitic trace inductance and the feedback loop from $V_{OUT}$ to ADJ must be kept as short as possible. To improve PSRR, a bypass capacitor can be placed at the ADJ pin and must be placed as close as possible to the IC. In cases when $V_{IN}$ shorts to ground, an external diode must be placed from $V_{OUT}$ to $V_{IN}$ to divert the surge current from the output capacitor and protect the IC. Similarly, in cases when a large bypass capacitor is placed at the ADJ pin and VOUT shorts to ground, an external diode must be placed from ADJ to VOUT to provide a path for the bypass capacitor to discharge. These diodes must be placed close to the corresponding IC pins to increase their effectiveness.

10.2 Layout Example

![Figure 45. LMx38 Layout](image)
11 Device and Documentation Support

11.1 Receiving Notification of Documentation Updates
To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on Alert me to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

11.2 Related Links
The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

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<th>TECHNICAL DOCUMENTS</th>
<th>TOOLS &amp; SOFTWARE</th>
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<tr>
<td>LM138</td>
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<td>Click here</td>
<td>Click here</td>
<td>Click here</td>
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<tr>
<td>LM338</td>
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11.3 Community Resources
The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

**TI E2E™ Online Community** *TI's Engineer-to-Engineer (E2E) Community*. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

11.6 Glossary

SLYZ022 — *TI Glossary.*
This glossary lists and explains terms, acronyms, and definitions.

12 Mechanical, Packaging, and Orderable Information
The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.
## PACKAGING INFORMATION

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<th>Pins</th>
<th>Package Qty</th>
<th>Eco Plan (2)</th>
<th>Lead/Ball Finish (6)</th>
<th>MSL Peak Temp (3)</th>
<th>Op Temp (°C)</th>
<th>Device Marking (4/5)</th>
<th>Samples</th>
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<td>LM338T P+</td>
<td>Samples</td>
</tr>
</tbody>
</table>

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- **ACTIVE**: Product device recommended for new designs.
- **LIFEBUY**: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
- **NRND**: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
- **PREVIEW**: Device has been announced but is not in production. Samples may or may not be available.
- **OBSOLETE**: TI has discontinued the production of the device.

(2) **RoHS**: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".
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(3) **MSL, Peak Temp.** - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) **Lead/Ball Finish** - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.
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