

## LM9044 Lambda Sensor Interface Amplifier

 Check for Samples: [LM9044](#)

### FEATURES

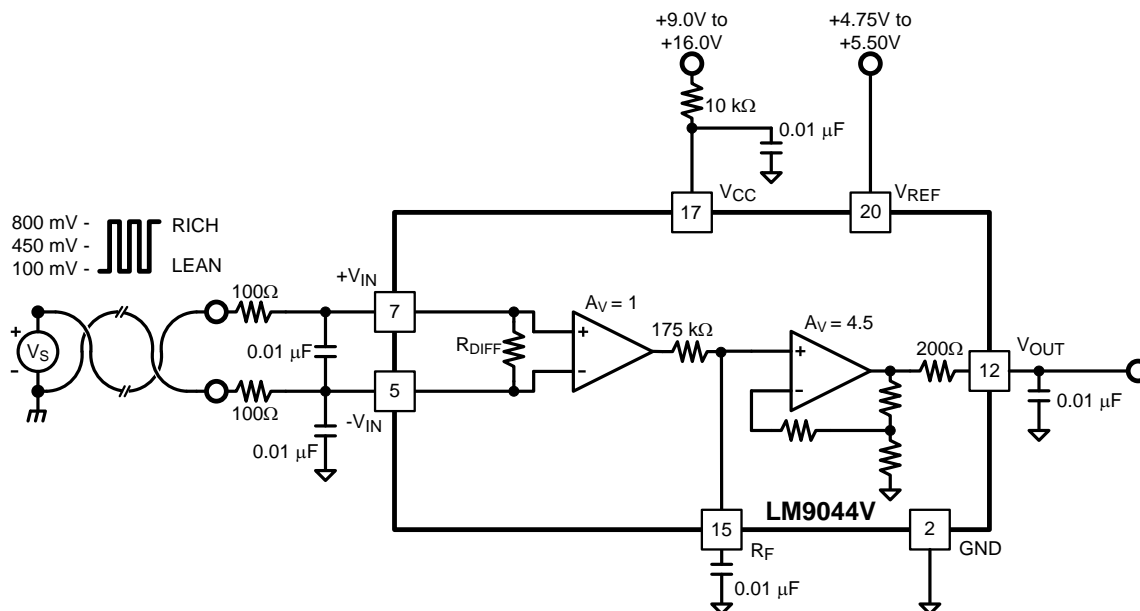
- Normal Circuit Operation Specified with Inputs up to 3V Below Ground on a Single Supply.
- Gain Factory Trimmed and Specified over Temperature ( $\pm 3\%$  of Full-scale from  $-40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ )
- Low Power Consumption (Typically 1 mA)
- Fully Protected Inputs
- Input Open Circuit Detection
- Operation Specified over the Entire Automotive Temperature Range ( $-40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ )

### DESCRIPTION

The LM9044 is a precision differential amplifier specifically designed for operation in the automotive environment. Gain accuracy is specified over the entire automotive temperature range ( $-40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ ) and is factory trimmed after package assembly. The input circuitry has been specifically designed to reject common-mode signals as much as 3V below ground without the need for a negative voltage supply. This facilitates the use of sensors which are grounded at the engine block while the LM9044 itself is grounded at chassis potential. An external capacitor on the  $R_F$  pin sets the maximum operating frequency of the amplifier, thereby filtering high frequency transients. Both inputs are protected against accidental shorting to the battery and against load dump transients. The input impedance is typically 1.2 M $\Omega$ .

The output op amp is capable of driving capacitive loads and is fully protected. Also, internal circuitry has been provided to detect open circuit conditions on either or both inputs and force the output to a "home" position (a ratio of the external reference voltage).

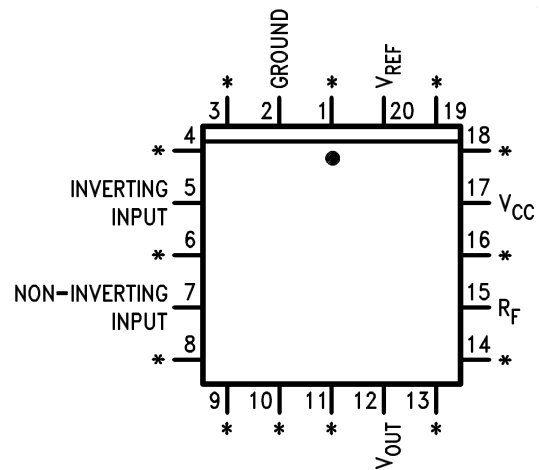
### Typical Application



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

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## Connection Diagram



\*Pins 1, 3, 4, 6, 8, 9, 10, 11, 13, 14, 16, 18, 19 are trim pins and should be left floating.

**Figure 1. Top View  
PLCC Package  
See Package Number FN0020A**



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

## ABSOLUTE MAXIMUM RATINGS<sup>(1)(2)</sup>

V <sub>CC</sub> Supply Voltage (R <sub>VCC</sub> = 15 kΩ)	±60V
V <sub>REF</sub> Supply Voltage	-0.3V to +6V
DC Input Voltage (Either input) <sup>(3)</sup>	-3V to +16V
Input Transients <sup>(4)</sup>	±60V
Power Dissipation see <sup>(5)</sup>	1350 mW
Output Short Circuit Duration	Indefinite
Operating Temperature Range	-40°C to +125°C
Storage Temperature Range	-65°C to +150°C
Soldering Information	
PLCC Package	
Vapor Phase (60 seconds)	215°C
Infrared (15 seconds)	220°C

See <http://www.ti.com> for other methods of soldering surface mount devices.

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur.
- (2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/Distributors for availability and specifications.
- (3) With a 100Ω series resistor on each input pin.
- (4) This test is performed with a 1000Ω source impedance.
- (5) For operation in ambient temperatures above 25°C the device must be derated based on a maximum junction temperature of 150°C and a thermal resistance of 93°C/W junction to ambient.

## ELECTRICAL CHARACTERISTICS

V<sub>CC</sub> = 12V, V<sub>REF</sub> = 5V, -40°C ≤ T<sub>A</sub> ≤ 125°C unless otherwise noted

Parameter	Conditions	(1)			(2)			Units
		Min	Typ	Max	Min	Typ	Max	
Differential Voltage Gain	V <sub>DIFF</sub> = 0.5, -1V ≤ V <sub>CM</sub> ≤ +1V	4.41	4.50	4.59	-	-	-	V/V
	V <sub>DIFF</sub> = 0.5, -3V ≤ V <sub>CM</sub> ≤ +1V	-	-	-	4.36	4.50	4.64	V/V
Gain Error <sup>(3)</sup>	0 ≤ V <sub>DIFF</sub> ≤ 1V, -1V ≤ V <sub>CM</sub> ≤ +1V	-2	0	2	-	-	-	%/FS
	0 ≤ V <sub>DIFF</sub> ≤ 1V, -3 ≤ V <sub>CM</sub> ≤ +1V	-	-	-	-3	0	3	%/FS
Differential Input Resistance	0 ≤ V <sub>DIFF</sub> ≤ 1V, -1V ≤ V <sub>CM</sub> ≤ +1V	0.95	1.20	3.00	-	-	-	MΩ
	0 ≤ V <sub>DIFF</sub> ≤ 1V, -3 ≤ V <sub>CM</sub> ≤ +1V	-	-	-	0.70	1.20	4.00	MΩ
Non-Inverting Input Bias Current	0 ≤ V <sub>DIFF</sub> ≤ 1V, -1 ≤ V <sub>CM</sub> ≤ +1V	-	±0.38	±0.65	-	-	-	μA
	0 ≤ V <sub>DIFF</sub> ≤ 1V, -3 ≤ V <sub>CM</sub> ≤ +1V	-	-	-	-	±0.38	±1.5	μA
Inverting Input Bias Current	0 ≤ V <sub>DIFF</sub> ≤ 1V, -1 ≤ V <sub>CM</sub> ≤ +1V	-25	-65	-100	-	-	-	μA
	0 ≤ V <sub>DIFF</sub> ≤ 1V, -3 ≤ V <sub>CM</sub> ≤ +1V	-	-	-	-	-45	-150	μA
V <sub>CC</sub> Supply Current	V <sub>CC</sub> = 12V, R <sub>VCC</sub> = 15k	-	300	500	-	-	-	μA
V <sub>REF</sub> Supply Current	4.75V ≤ V <sub>REF</sub> ≤ 5.5V	-	0.5	1.0	-	-	-	mA
Common-Mode Voltage Range <sup>(4)</sup>		-1	-	1	-3	-	1	V
DC Common-Mode Rejection Ratio	Input Referred							
	-1V ≤ V <sub>CM</sub> ≤ +1V, V <sub>DIFF</sub> = 0.5V	50	60	-	-	-	-	dB
Open Circuit Output Voltage	One or Both Inputs Open							
	-1V ≤ V <sub>CM</sub> ≤ +1V	0.371	0.397	0.423	-	-	-	xV <sub>REF</sub>
	-3V ≤ V <sub>CM</sub> ≤ +1V	-	-	-	0.365	0.397	0.439	xV <sub>REF</sub>
Short Circuit Output Current	Output Grounded	1.0	2.7	5.0	-	-	-	mA

- (1) These parameters are specified and 100% production tested.
- (2) These parameters will be specified but not 100% production tested.
- (3) Gain error is given as a percent of full-scale. Full-scale is defined as 1V at the input and 4.5V at the output.
- (4) The LM9044 has been designed to common-mode to -3V, but production testing is only performed at ±1V.

**ELECTRICAL CHARACTERISTICS (continued)**
 $V_{CC} = 12V$ ,  $V_{REF} = 5V$ ,  $-40^{\circ}C \leq T_A \leq 125^{\circ}C$  unless otherwise noted

Parameter	Conditions	(1)			(2)			Units
		Min	Typ	Max	Min	Typ	Max	
$V_{CC}$ Power Supply Rejection Ratio	$V_{CC} = 12V$ , $R_{V_{CC}} = 15k$ $V_{DIFF} = 0.5V$	50	65	-	-	-	-	dB
$V_{REF}$ Power Supply Rejection Ratio	$V_{REF} = 5 V_{DC}$ $V_{DIFF} = 0.5V$	60	74	-	-	-	-	dB

TYPICAL PERFORMANCE CHARACTERISTICS

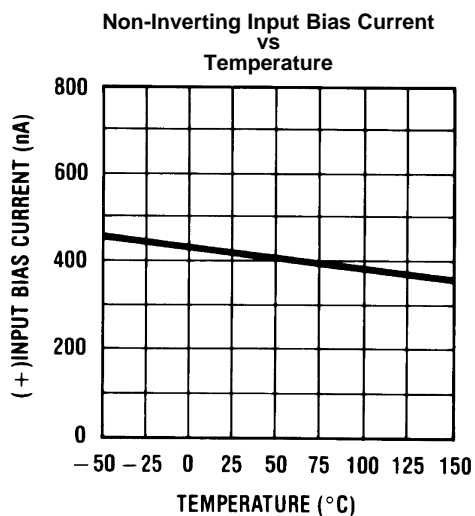


Figure 2.

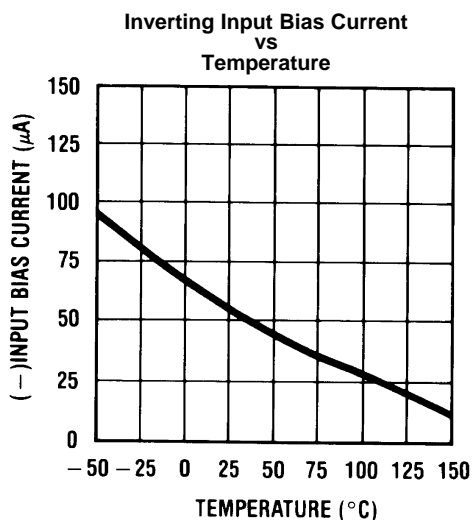


Figure 3.

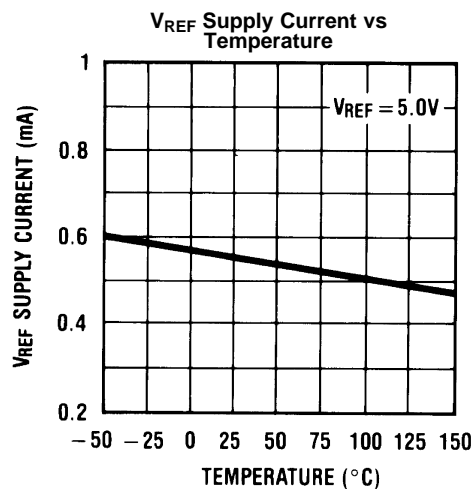


Figure 4.

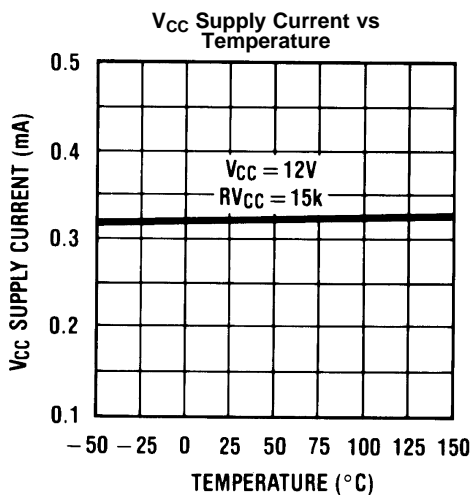


Figure 5.

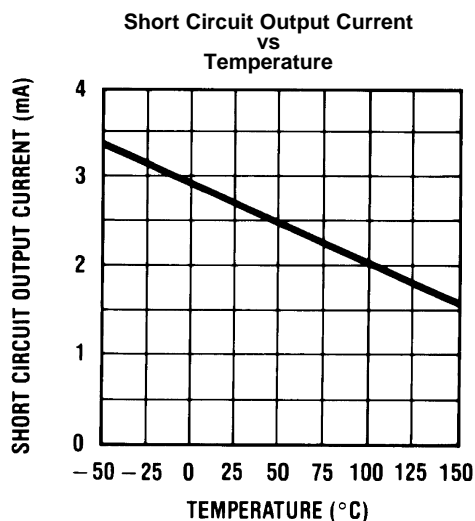


Figure 6.

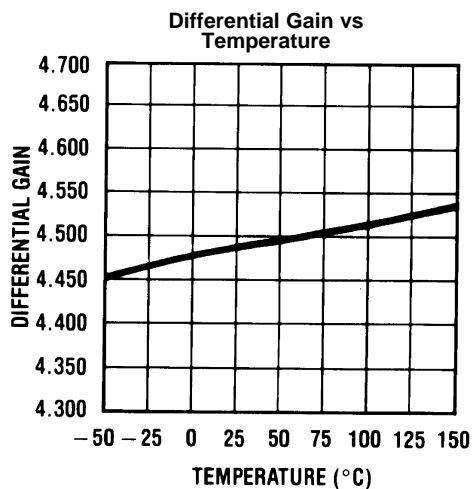


Figure 7.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

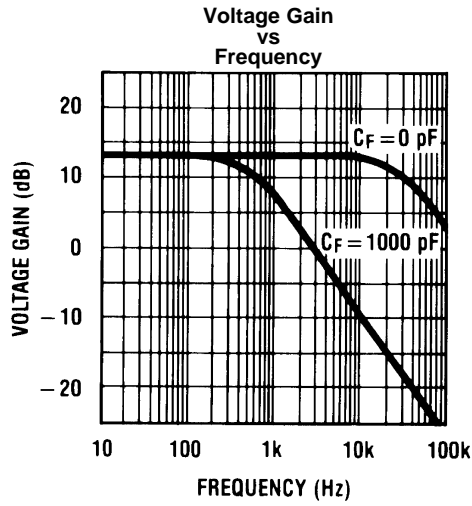


Figure 8.

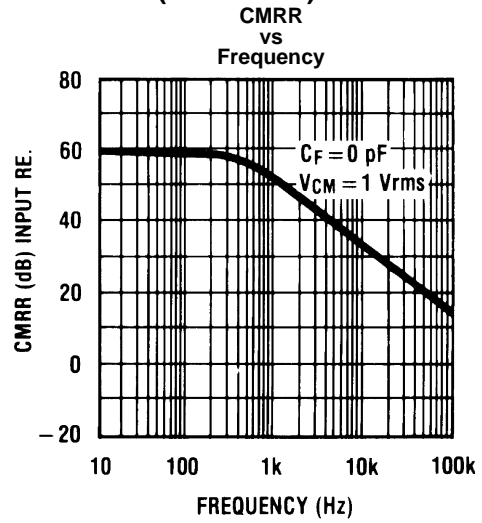


Figure 9.

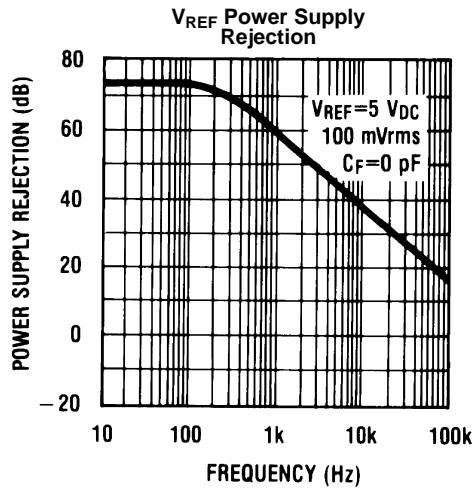


Figure 10.

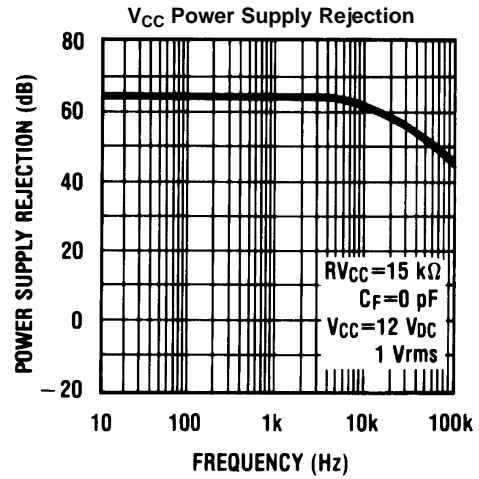
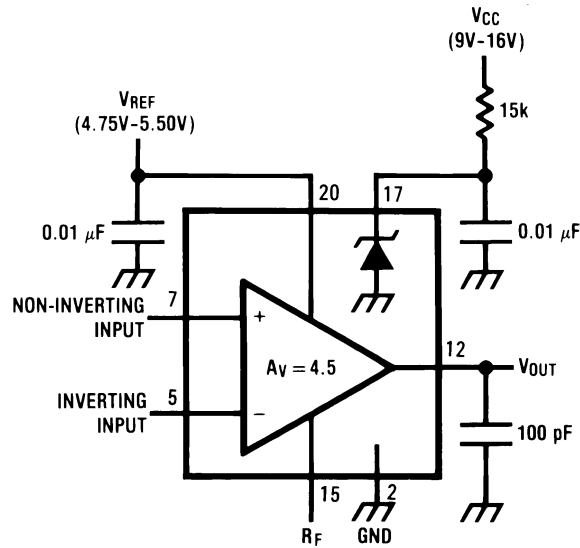
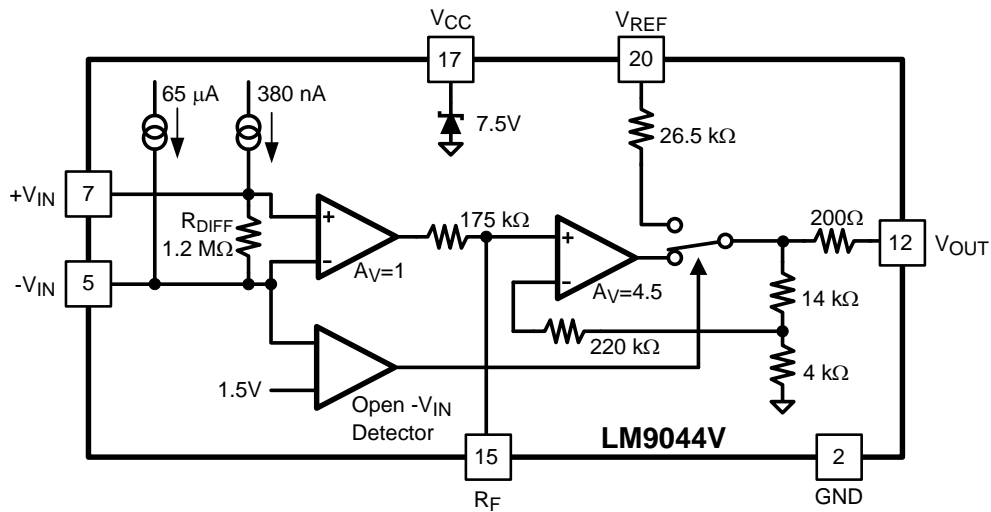


Figure 11.

TEST CIRCUIT



Block Diagram



## APPLICATION INFORMATION

### CIRCUIT DESCRIPTION

The LM9044 is a single channel device intended to act as a linear interface between a zirconium dioxide oxygen sensor and an A-to-D convertor. The LM9044 is fabricated in Bipolar technology and requires two supplies: a nominal 12V automotive supply (i.e.  $V_{BATTERY}$ ), and a well regulated 5V supply.

The IC consists of a single channel differential input amplifier with a nominal DC gain of 4.5 V/V. The differential inputs have a specified common mode voltage operating range of 1V above and below ground. The circuitry also contains provisions for default output voltage in the cases of cold sensor and open sensor wiring. Additional support circuitry includes one pin for an optional user programmed low pass filter.

### COLD SENSOR

Typically, a Lambda sensor will have an impedance of less than 10 k $\Omega$  when operating at temperatures between 300°C, and 500°C. When a Lambda sensor is not at operating temperature, its impedance can be more than 10 Meg $\Omega$ . Any voltage signal that may be developed is seriously attenuated. During this high impedance condition the LM9044 will provide a default output voltage.

While the Lambda sensor is high impedance the internal non-inverting input bias current (380 nA typical) will flow through the differential input resistance (1.2 M $\Omega$  typical) and out the inverting input pin to ground. This will cause a voltage to be developed across the differential inputs:

$$V_{IN(DIFF)} = 380 \text{ nA} \times 1.2 \text{ M}\Omega$$

$$V_{IN(DIFF)} = 456 \text{ mV}$$

The 456 mV across differential input resistance will be the dominant input signal, and the typical  $V_{OUT}$  will be:

$$V_{OUT} = V_{IN(DIFF)} \times 4.50$$

$$V_{OUT} = 456 \text{ mV} \times 4.50$$

$$V_{OUT} = 2.0\text{V}$$

As the Lambda sensor is heated, and the sensor impedance begins to drop, the voltage signal from the sensor will become the dominate signal.

The non-inverting input bias current is scaled to the  $V_{REF}$  voltage. As the  $V_{REF}$  voltage increases, or decreases, this bias current will change proportionally.

### OPEN INPUT PINS DEFAULTS

In any remote sensor application it is desirable to be able to deal with the possibility of open connections between the sensor and the control module. The LM9044 is capable of providing a default output voltage should either, or both, of the wires to the Lambda sensor open. The two inputs handle the open circuit condition differently.

For the case of an open connection at the non-inverting input, the device would react exactly the same as for the Cold Sensor condition. The internal non-inverting input bias current (380 nA typical) flowing through the differential input resistance (1.2 M $\Omega$  typical) would cause the typical output voltage to be at a value defined by:

$$V_{OUT} = ((380 \text{ nA} \times 1.2\text{M}\Omega) \times 4.50)$$

$$V_{OUT} = 2.0\text{V}$$

The inverting input would still be connected to the Lambda sensor ground, so common mode signals would still need to be considered in this condition.

For the case of an open connection of the inverting input, the device output stage switches from the amplifier output to a resistive voltage divider. The LM9044 has a comparator to monitor the voltage on the inverting input pin, and a 65  $\mu\text{A}$  (typical) current source that will force the pin high if the pin is open. When the voltage on the inverting pin goes above typically 1.5V, the comparator will switch the output pin from the amplifier output to the resistive voltage divider stage. In this case, the default  $V_{OUT}$  is not dependent on the gain stage, and any signal on the non-inverting input will be ignored.



In this condition  $V_{OUT}$  is:

$$V_{OUT} = V_{REF} \times ((14k + 4k) / (26.5k + 14k + 4k))$$

$$V_{OUT} = V_{REF} \times 0.4045$$

When  $V_{REF}$  is at 5.0V,  $V_{OUT}$  is defined as:

$$V_{OUT} = 5.0V \times 0.4045$$

$$V_{OUT} = 2.0V$$

In the cases where both the inverting and non-inverting pins are open, the open inverting pin condition (i.e.: a voltage divider across the output) will be the dominant condition.

Any common mode voltage transient on the inverting input pin which goes above the comparator threshold will immediately cause the output to switch to the resistive voltage divider mode. The output will return to normal operation when the voltage on the inverting input falls below the 1.5V threshold.

## OUTPUT RESISTANCE

Under normal operating conditions the output pin resistance is typically 200Ω.

If the LM9044 is operating in a default output mode due an open connection on the inverting input, the output resistance will typically appear to be close to 11 kΩ.

An external output filter capacitor value of no more than 0.01 μF is generally recommended. Since the output pin voltage drive is basically a simple NPN emitter follower, the output pin pull-down is done by the internal feedback resistor string. With larger value capacitors on the output pin the effect will be somewhat similar to a voltage peak detector where the output capacitor is charged through the 200Ω resistor, and discharged back through the 200Ω resistor and the 18 kΩ feedback resistor string to ground.

The output resistance provides current limiting for the output stage should it become shorted to Ground. Any DC loading of the output will cause an error in the output voltage.

## SUPPLY BYPASSING

For best performance the LM9044 requires a  $V_{REF}$  supply which is stable and noise free. The same 5V reference supply used for the A/D converter is the recommended LM9044  $V_{REF}$  supply.

The LM9044  $V_{CC}$  pin has an internal zener shunt voltage regulator, typically 7.5V, and requires a series resistor to limit the current. The  $V_{CC}$  pin should be bypassed with a minimum 0.01 μF capacitor to the Ground pin, and should be located as close to the device as possible. Some applications may require an additional bypass capacitance if the system voltage is unusually noisy.

## SETTING THE BANDWIDTH

The LM9044 bandwidth is limited by an external capacitor ( $C_F$ ) on the  $R_F$  pin.

This pin has an internal 175 kΩ resistor. The external capacitor and the internal resistor form a simple RC low-pass filter with a corner frequency ( $f_C$ ) defined as:

$$f_C = 1 / (2 \times \pi \times 175 \text{ k}\Omega \times C_F)$$

With a  $C_F$  capacitor value of 0.001 μF, the corner frequency is:

$$f_C = 1 / (2 \times \pi \times 175 \text{ k}\Omega \times 0.001 \text{ }\mu\text{F})$$

$$f_C = 909 \text{ Hz}$$

## INPUT FILTERING

Filtering at the differential inputs is strongly recommended. Both the differential voltage signal and the common mode voltage signal should have low pass filters.

Input filtering is accomplished with series resistors on the input pins, and appropriate bypass capacitors. Typical input pin series resistance values are in the 100Ω to 1kΩ range. Series resistance values larger than 1kΩ will generate offset voltages that affect the accuracy of the signal voltage seen at the differential input pins.



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**REVISION HISTORY**

<b>Changes from Revision C (March 2013) to Revision D</b>	<b>Page</b>
<hr/> <ul style="list-style-type: none"><li>• Changed layout of National Data Sheet to TI format .....</li></ul>	<hr/> <a href="#">10</a>

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LM9044V/NOPB	ACTIVE	PLCC	FN	20	40	Green (RoHS & no Sb/Br)	CU SN	Level-2A-250C-4 WEEK	-40 to 125	LM9044V	<a href="#">Samples</a>
LM9044VX/NOPB	ACTIVE	PLCC	FN	20	1000	Green (RoHS & no Sb/Br)	CU SN	Level-2A-250C-4 WEEK	-40 to 125	LM9044V	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

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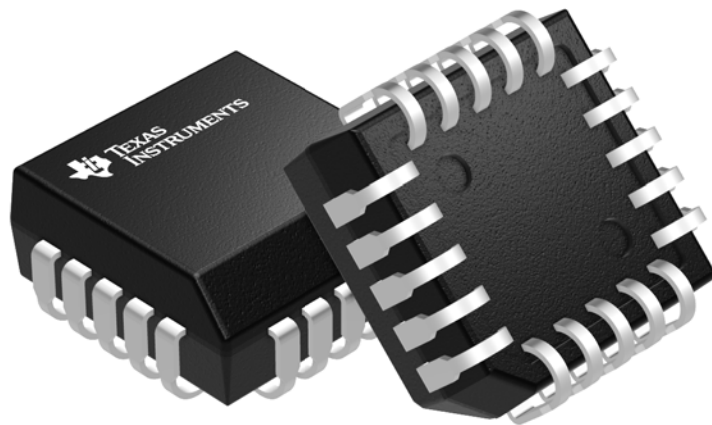
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**FN 20**

**GENERIC PACKAGE VIEW**

**PLCC - 4.57 mm max height**

PLASTIC CHIP CARRIER



Images above are just a representation of the package family, actual package may vary.  
Refer to the product data sheet for package details.

4040005-2/C

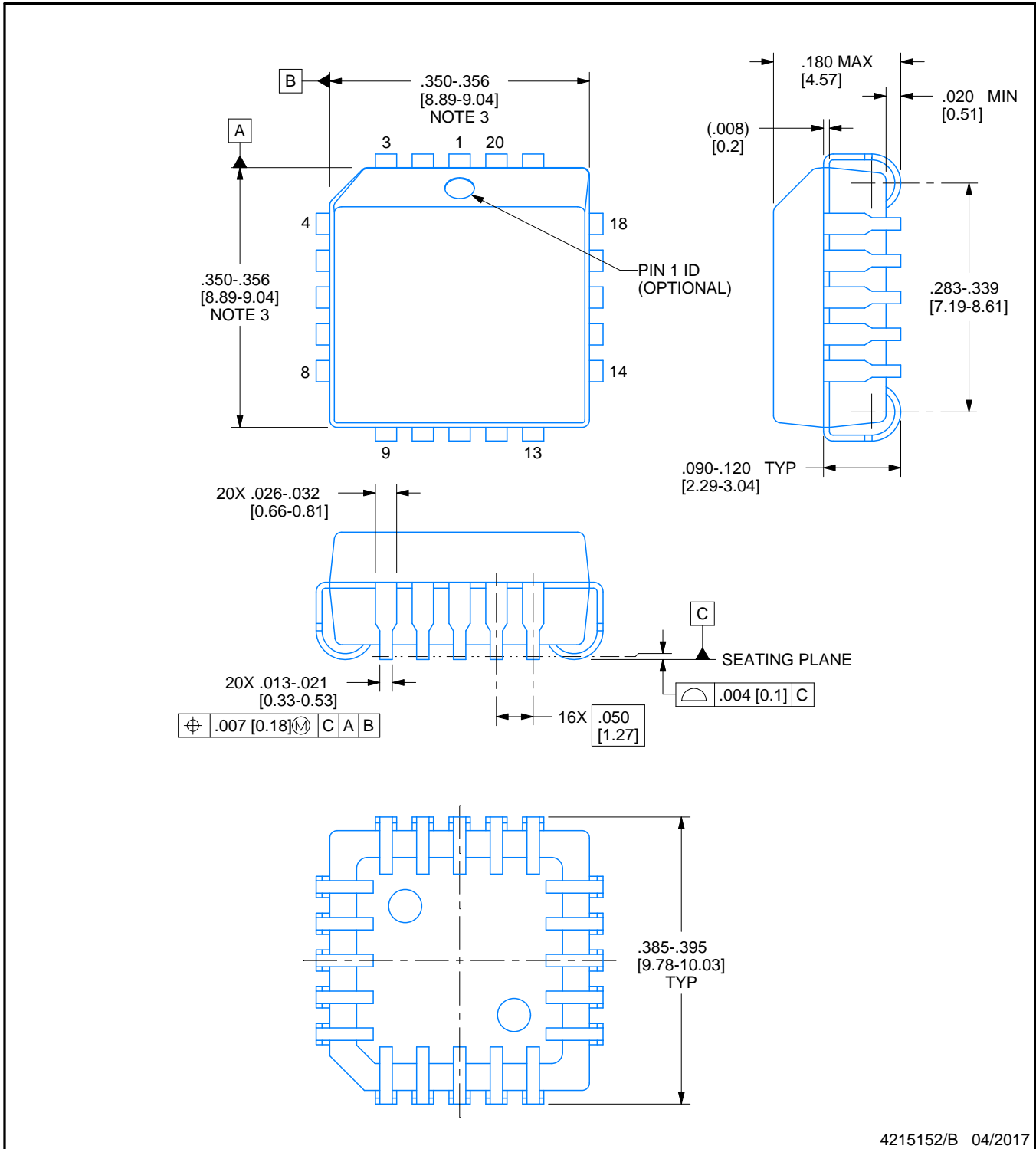


# FN0020A

# PACKAGE OUTLINE

## PLCC - 4.57 mm max height

PLASTIC CHIP CARRIER



4215152/B 04/2017

### NOTES:

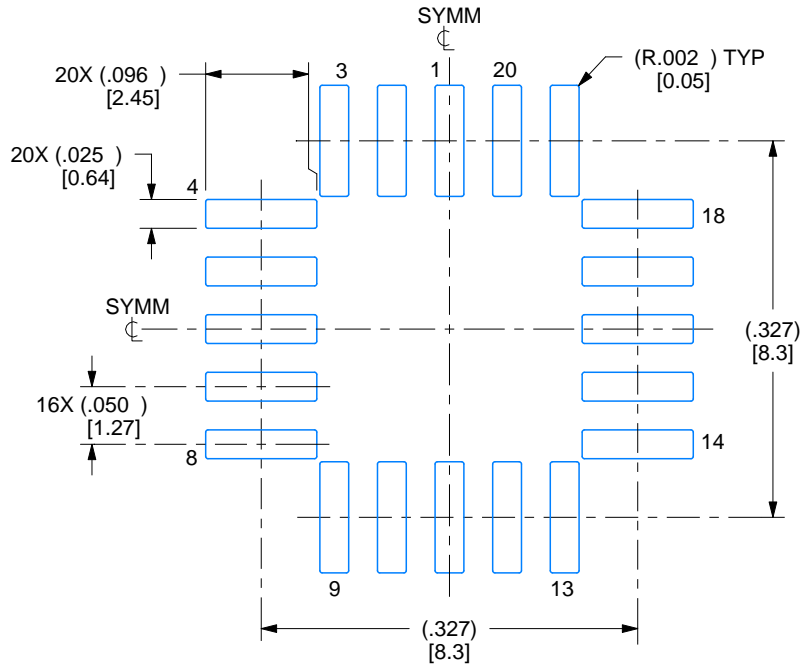
1. All linear dimensions are in inches. Any dimensions in brackets are in millimeters. Any dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. Dimension does not include mold protrusion. Maximum allowable mold protrusion .01 in [0.25 mm] per side.
4. Reference JEDEC registration MS-018.

# EXAMPLE BOARD LAYOUT

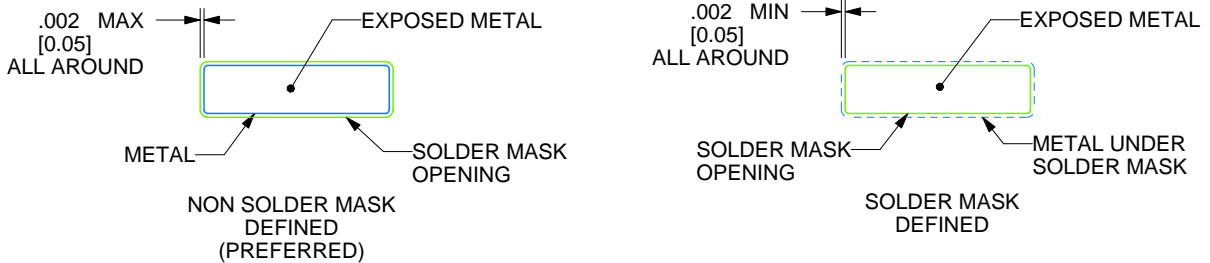
FN0020A

PLCC - 4.57 mm max height

PLASTIC CHIP CARRIER



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:6X



SOLDER MASK DETAILS

4215152/B 04/2017

NOTES: (continued)

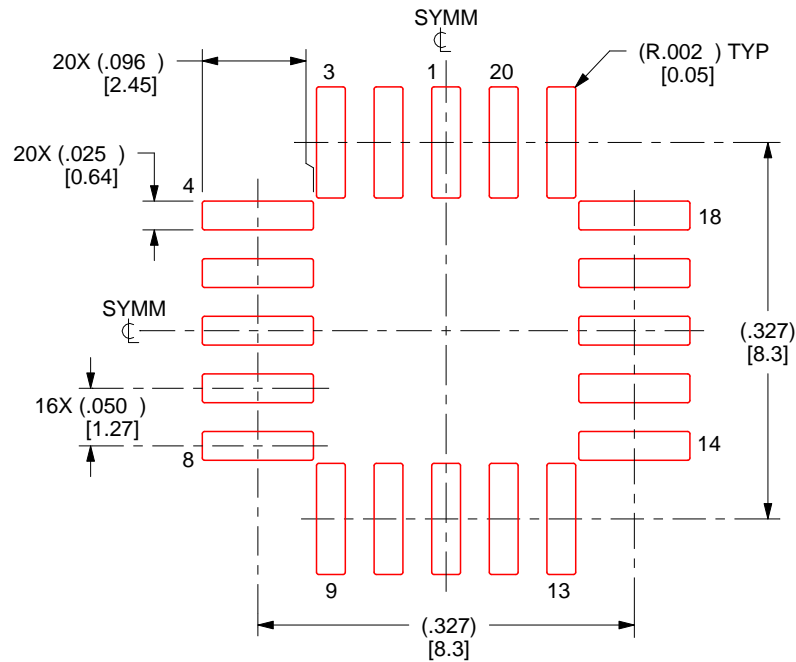
- Publication IPC-7351 may have alternate designs.
- Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

FN0020A

PLCC - 4.57 mm max height

PLASTIC CHIP CARRIER



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL  
SCALE:6X

4215152/B 04/2017

NOTES: (continued)

7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
8. Board assembly site may have different recommendations for stencil design.



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