











TSV911, TSV912, TSV914

SBOS878D-JULY 2017-REVISED OCTOBER 2019

# TSV91x Rail-to-Rail Input/Output, 8-MHz Operational Amplifiers

#### **Features**

Rail-to-rail input and output

Low noise: 18 nV/√Hz at 1 kHz

Low power consumption: 550 µA (typical)

High-gain bandwidth: 8 MHz

Operating supply voltage from 2.5 V to 5.5 V

Low input bias current: 1 pA (typical)

Low input offset voltage: 1.5 mV (maximum)

Low offset voltage drift: ±0.5 µV/°C (typical)

ESD internal protection: ±4-kV human-body model

Extended temperature range: -40°C to 125°C

# **Applications**

- Battery-powered applications
- Motor control
- Power modules
- HVAC: heating, ventilating, and air conditioning
- Washing machines
- Refrigerators
- Medical instrumentation
- Active filters
- Sensor signal conditioning
- Audio receiver
- Automotive infotainment

# 3 Description

The TSV91x family, which includes single-, dual-, and quad-channel operational amplifiers (op amps), is for specifically designed general-purpose applications. Featuring rail-to-rail input and output (RRIO) swings, wide bandwidth (8 MHz), and low offset voltage (0.3 mV, typical), this family is designed for a variety of applications that require a good balance between speed and power consumption. The op amps are unity-gain stable and feature an ultralow input bias current, which enables the family to be used in applications with high-source impedances. The low input bias current allows the devices to be used for sensor interfaces, battery-supplied and portable applications, and active filtering.

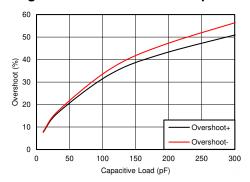
The robust design of the TSV91x provides ease-ofuse to the circuit designer. Features include a unitygain stable, integrated RFI-EMI rejection filter, no phase reversal in overdrive condition, and high electrostatic discharge (ESD) protection (4-kV HBV).

## Device Information<sup>(1)</sup>

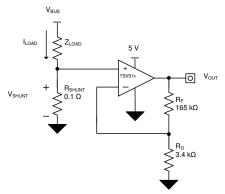
PART NUMBER	PACKAGE	BODY SIZE (NOM)		
TSV911	SOT-23 (5)	1.60 mm × 2.90 mm		
	SC70 (5)	1.25 mm × 2.00 mm		
	SOIC (8)	3.91 mm × 4.90 mm		
TSV912	WSON (8)	2.00 mm × 2.00 mm		
	SOT-23 (8)	1.60 mm × 2.90 mm		
TC\/04.4	SOIC (14)	8.65 mm × 3.91 mm		
TSV914	TSSOP (14)	4.40 mm × 5.00 mm		

(1) For all available packages, see the orderable addendum at the end of the data sheet.

#### Small-Signal Overshoot vs Load Capacitance



# **Low-Side Motor Control**





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har	nges from Revision C (January 2019) to Revision D		Pa	ıge
Α	dded SOT-23 (8) (DDF) package information to data sheet			1
har	nges from Revision B (April 2018) to Revision C		Pa	ige
	nges from Revision B (April 2018) to Revision C Deleted preview notations for TSV911IDBV			ıge
D	Deleted preview notations for TSV911IDBV			ige . 1
D A	Deleted preview notations for TSV911IDBVdded SC70 package information to Device Information table	······		ige . 1 . 1
D A D	Deleted preview notations for TSV911IDBV	23) pacl	kage	. 1 . 1

# Changes from Revision A (October 2017) to Revision B

•	Deleted package preview note from 8-pin WSON package in Device Information table	. 1
•	Deleted package preview note from PW (TSSOP) package from Device Comparison table	. 4
•	Deleted package preview note from DSG (WSON) package from Device Comparison table	4
•	Deleted package preview note from TSV912 DSG package pinout drawing in Pin Configuration and Functions section	. 6
•	Added DGK (VSSOP) thermal information to Thermal Information: TSV912 table	9
•	Deleted package preview note to TSV914 PW (TSSOP) package Thermal Information table	9
•	Added PW (TSSOP) package information to Thermal Information: TSV914 table	9
•	Changed TSV914 PW (TSSOP) junction-to-ambient thermal resistance from 135.8°C/W to 205.8°C/W	. 9
•	Changed TSV914 PW (TSSOP) junction-to-case(top) thermal resistance from 64°C/W to 106.7°C/W	. 9
•	Changed TSV914 PW (TSSOP) junction-to-board thermal resistance from 79°C/W to 133.9°C/W	9

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•	Changed TSV914 PW (TSSOP) junction-to-top characterization parameter from 15.7°C/W to 34.4°C/W	9
•	Changed TSV914 PW (TSSOP) junction-to-board characterization parameter from 78.4°C/W to 132.6°C/W	9
Cł	hanges from Original (July 2017) to Revision A	Page
•	Changed TSV914 14-pin SOIC package from preview to production data in <i>Device Information</i> table	1
•	Deleted TSV911 SC70, SOT-553 and SOIC packages from Device Information table	1
•	Deleted TSV912 VSSOP packages from Device Information table	1
•	Deleted TSV911 SC70 and SOIC packages from pinout drawings and Pin Functions table	5
•	Deleted TSV912 DGK and DGS packages from pinout images Pin Functions table	6
•	Deleted package preview note from TSV914 pinout drawing and Pin Functions table	7
•	Added TSV914 Thermal Information table	9
•	Added 2017 copyright notice to Figure 35	20



# 5 Device Comparison Table

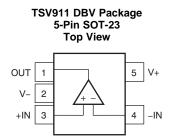
DEVICE	NO. OF	PACKAGE LEADS						
DEVICE	CHANNELS	DBV	DCK	D	DSG	PW	DDF	
TSV911	1	5	5	_	_	_	_	
TSV912	2	_	_	8	8	_	8	
TS\/01/	1	_		1/		1/		

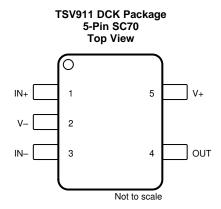
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# 6 Pin Configuration and Functions



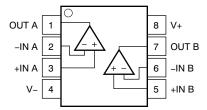


# Pin Functions: TSV911

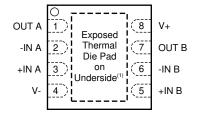
PIN						
NAME	NO.		1/0	DESCRIPTION		
NAIVIE	DBV (SOT-23)	DCK (SC70)				
-IN	4	3	ı	Inverting input		
+IN	3	1	ı	Noninverting input		
OUT	1	4	0	Output		
V-	2	2	_	Negative (lowest) supply or ground (for single-supply operation)		
V+	5	5	_	Positive (highest) supply		



#### TSV912 D, DGK, DDF Packages 8-Pin SOIC, VSSOP Top View



# TSV912 DSG Package (1) 8-Pin WSON With Exposed Thermal Pad Top View



 Connect exposed thermal pad to V-. See Packages with an Exposed Thermal Pad section for more information.

# **Pin Functions: TSV912**

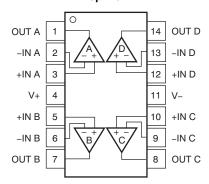
PIN		1/0	DESCRIPTION	
NAME	NO.	1/0	DESCRIPTION	
–IN A	2	I	Inverting input, channel A	
+IN A	3	I	Noninverting input, channel A	
–IN B	6	I	Inverting input, channel B	
+IN B	5	I	Noninverting input, channel B	
OUT A	1	0	Output, channel A	
OUT B	7	0	Output, channel B	
V-	4	_	Negative (lowest) supply or ground (for single-supply operation)	
V+	8	_	Positive (highest) supply	

Product Folder Links: TSV911 TSV912 TSV914

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# TSV914 D, PW Packages 14-Pin SOIC, TSSOP Top View



#### Pin Functions: TSV914

PIN		1/0	DESCRIPTION	
NAME	NO.	1/0	DESCRIPTION	
–IN A	2	I	Inverting input, channel A	
+IN A	3	1	Noninverting input, channel A	
–IN B	6	1	Inverting input, channel B	
+IN B	5	I	Noninverting input, channel B	
–IN C	9	I	Inverting input, channel C	
+IN C	10	1	Noninverting input, channel C	
–IN D	13	1	Inverting input, channel D	
+IN D	12	1	Noninverting input, channel D	
OUT A	1	0	Output, channel A	
OUT B	7	0	Output, channel B	
OUT C	8	0	Output, channel C	
OUT D	14	0	Output, channel D	
V-	11	_	Negative (lowest) supply or ground (for single-supply operation)	
V+	4	_	Positive (highest) supply	



# 7 Specifications

## 7.1 Absolute Maximum Ratings

over operating free-air temperature (unless otherwise noted)(1)

			MIN	MAX	UNIT	
Supply voltage				6	V	
Signal input pins	Voltage <sup>(2)</sup>	Common-mode	(V-) - 0.5	(V+) + 0.5		
	voitage (=)	Differential		(V+) - (V-) + 0.2	V	
	Current <sup>(2)</sup>		-10	10	mA	
Output short-circuit	(3)	Continuous		mA		
Specified, T <sub>A</sub>			-40	125	°C	
Junction, T <sub>J</sub>				150	°C	
Storage, T <sub>stg</sub>			-65	150	°C	

<sup>(1)</sup> Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

# 7.2 ESD Ratings

over operating free-air temperature range (unless otherwise noted)

			VALUE	UNIT
V	Clastrostatio discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±4000	\/
V <sub>(ESD)</sub>	Electrostatic discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 (2)	±1500	V

<sup>1)</sup> JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

# 7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	MAX	UNIT
V <sub>S</sub> Supply voltage	2.5	5.5	V
Specified temperature	-40	125	°C

#### 7.4 Thermal Information: TSV911

		TSV		
	THERMAL METRIC <sup>(1)</sup>	DBV (SOT-23)	DCK (SC70)	UNIT
		5 PINS	5 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	221.7	263.3	°C/W
$R_{\theta JC(top)}$	Junction-to-case(top) thermal resistance	144.7	75.5	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	49.7	51.0	°C/W
ΨЈТ	Junction-to-top characterization parameter	26.1	1.0	°C/W
ΨЈВ	Junction-to-board characterization parameter	49.0	50.3	°C/W

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

<sup>(2)</sup> Input pins are diode-clamped to the power-supply rails. Current limit input signals that can swing more than 0.5 V beyond the supply rails to 10 mA or less.

<sup>(3)</sup> Short-circuit to ground, one amplifier per package.

<sup>(2)</sup> JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



## 7.5 Thermal Information: TSV912

		TSV912							
	THERMAL METRIC <sup>(1)</sup>	D (SOIC)	DGK (VSSOP)	DSG (WSON)	DDF (SOT-23)	UNIT			
		8 PINS	8 PINS	8 PINS	8 PINS				
$R_{\theta JA}$	Junction-to-ambient thermal resistance	157.6	201.2	94.4	184.4	°C/W			
$R_{\theta JC(top)}$	Junction-to-case(top) thermal resistance	104.6	85.7	116.5	112.8	°C/W			
$R_{\theta JB}$	Junction-to-board thermal resistance	99.7	122.9	61.3	99.9	°C/W			
ΨЈТ	Junction-to-top characterization parameter	55.6	21.2	13	18.7	°C/W			
ΨЈВ	Junction-to-board characterization parameter	99.2	121.4	61.7	99.3	°C/W			
R <sub>θJC(bot)</sub>	Junction-to-case(bottom) thermal resistance	N/A	N/A	34.4	N/A	°C/W			

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

# 7.6 Thermal Information: TSV914

			TSV914				
	THERMAL METRIC <sup>(1)</sup>	D (SOIC)	PW (TSSOP)	UNIT			
		14 PINS	14 PINS				
$R_{\theta JA}$	Junction-to-ambient thermal resistance	106.9	205.8	°C/W			
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	69	106.7	°C/W			
$R_{\theta JB}$	Junction-to-board thermal resistance	63	133.9	°C/W			
ΨЈТ	Junction-to-top characterization parameter	25.9	34.4	°C/W			
ΨЈВ	Junction-to-board characterization parameter	62.7	132.6	°C/W			

<sup>(1)</sup> For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.



# 7.7 Electrical Characteristics: $V_s$ (Total Supply Voltage) = (V+) - (V-) = 2.5 V to 5.5 V

at  $T_A = 25^{\circ}$ C,  $R_L = 10 \text{ k}\Omega$  connected to  $V_S / 2$ ,  $V_{CM} = V_S / 2$ , and  $V_{OUT} = V_S / 2$  (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
OFFSET	VOLTAGE					
		V <sub>S</sub> = 5 V		±0.3	±1.5	
V <sub>OS</sub>	Input offset voltage	V <sub>S</sub> = 5 V T <sub>A</sub> = -40°C to 125°C			±3	mV
dV <sub>OS</sub> /dT	Drift	$V_S = 5 V$ $T_A = -40$ °C to 125°C		±0.5		μV/°C
PSRR	Power-supply rejection ratio	$V_S = 2.5 \text{ V} - 5.5 \text{ V}, V_{CM} = (V-)$		±7		μV/V
	Channel separation, DC	At DC		100		dB
NPUT V	OLTAGE RANGE				·	
V <sub>CM</sub>	Common-mode voltage range	V <sub>S</sub> = 2.5 V to 5.5 V	(V-) - 0.1		(V+) + 0.1	V
		$V_S = 5.5 \text{ V}$ $(V-) - 0.1 \text{ V} < V_{CM} < (V+) - 1.4 \text{ V}$ $T_A = -40^{\circ}\text{C}$ to 125°C	80	103		
CMRR	Common-mode rejection ratio	$V_S = 5.5 \text{ V}, V_{CM} = -0.1 \text{ V to } 5.6 \text{ V}$ $T_A = -40^{\circ}\text{C to } 125^{\circ}\text{C}$	57	87		dB
		$V_S = 2.5 \text{ V}, (V-) - 0.1 \text{ V} < V_{CM} < (V+) - 1.4 \text{ V}$ $T_A = -40^{\circ}\text{C}$ to 125°C		88		
		V <sub>S</sub> = 2.5 V, V <sub>CM</sub> = -0.1 V to 1.9 V T <sub>A</sub> = -40°C to 125°C	81			
NPUT B	IAS CURRENT		•		+	
I <sub>B</sub>	Input bias current			±1		pA
los	Input offset current			±0.05		pА
NOISE			"			
E <sub>n</sub>	Input voltage noise (peak-to-peak)	$V_S = 5 \text{ V}, f = 0.1 \text{ Hz to } 10 \text{ Hz}$		4.77		μV <sub>PP</sub>
		$V_S = 5 \text{ V, f} = 10 \text{ kHz}$		12		
e <sub>n</sub>	Input voltage noise density	$V_S = 5 \text{ V, f} = 1 \text{ kHz}$		18		nV/√Hz
i <sub>n</sub>	Input current noise density	f = 1 kHz		10		fA/√Hz
	APACITANCE					
C <sub>ID</sub>	Differential			2		pF
C <sub>IC</sub>	Common-mode			4		pF
	OOP GAIN			•		Ρ.
SI EN EGGI GAIN		$V_S = 2.5 \text{ V}, (V-) + 0.04 \text{ V} < V_O < (V+) - 0.04 \text{ V}$ $R_L = 10 \text{ k}\Omega$		100		
		$V_S = 5.5 \text{ V}, (V-) + 0.05 \text{ V} < V_O < (V+) - 0.05 \text{ V}$ $R_L = 10 \text{ k}\Omega$	104	130		
A <sub>OL</sub>	Open-loop voltage gain	loop voltage gain $ \begin{array}{c} V_S = 2.5 \text{ V, } (V-) + 0.06 \text{ V} < V_O < (V+) - 0.06 \text{ V} \\ R_L = 2  k\Omega \end{array} $			dB	
		$V_S = 5.5 \text{ V}, (V-) + 0.15 \text{ V} < V_O < (V+) - 0.15 \text{ V}$ $R_L = 2 \text{ k}\Omega$				
FREQUE	NCY RESPONSE					
GBP	Gain bandwidth product	V <sub>S</sub> = 5 V, G = 1		8		MHz
Pm	Phase margin	V <sub>S</sub> = 5 V, G = 1		55		0
SR	Slew rate	$\begin{array}{l} V_S=5~V,~G=1\\ R_L=2~k\Omega\\ C_L=100~pF \end{array}$		4.5		V/µs
	Cottling time	To 0.1%, $V_S = 5 \text{ V}$ , 2-V step , $G = 1$ $C_L = 100 \text{ pF}$		0.5		
s	Settling time	To 0.01%, $V_S = 5 \text{ V}$ , 2-V step , $G = 1$ $C_L = 100 \text{ pF}$		1		μs
OR	Overload recovery time	$V_S = 5 \text{ V}, V_{IN} \times \text{gain} > V_S$		0.2		μs
THD + N	Total harmonic distortion + noise <sup>(1)</sup>	V <sub>S</sub> = 5 V, V <sub>O</sub> = 1 V <sub>RMS</sub> , G = 1, f = 1 kHz		0.0008%		
OUTPUT					"	
,	Voltage output swing from supply	$V_{S} = 5.5 \text{ V}, R_{L} = 10 \text{ k}\Omega$			15	
/ <sub>0</sub>	rails	$V_S = 5.5 \text{ V}, R_L = 2 \text{ k}\Omega$		50	mV	

Product Folder Links: TSV911 TSV912 TSV914

Third-order filter; bandwidth = 80 kHz at −3 dB. (1)

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# Electrical Characteristics: $V_S$ (Total Supply Voltage) = (V+) - (V-) = 2.5 V to 5.5 V (continued)

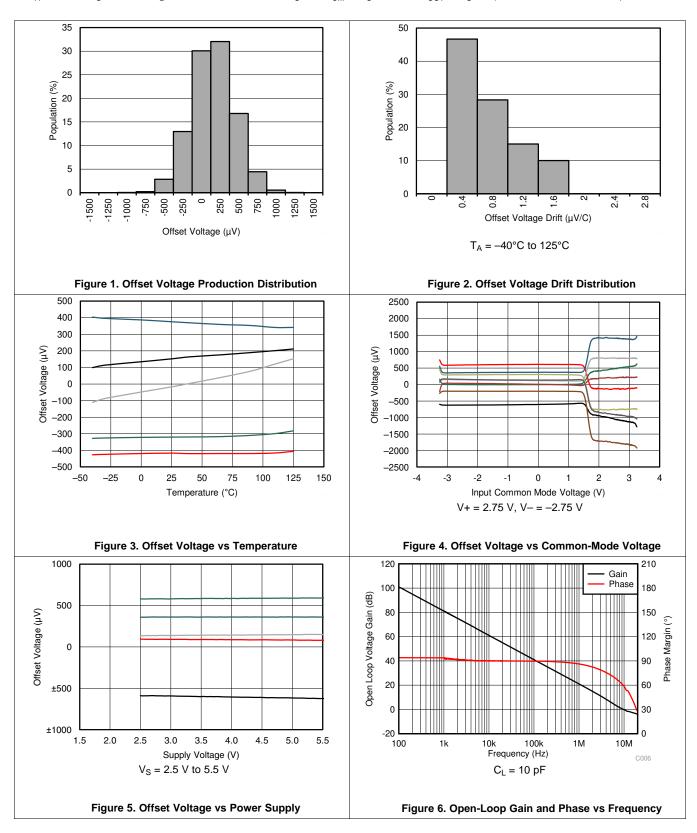
at  $T_A = 25$ °C,  $R_L = 10$  k $\Omega$  connected to  $V_S$  / 2,  $V_{CM} = V_S$  / 2, and  $V_{OUT} = V_S$  / 2 (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT				
I <sub>SC</sub>	Short-circuit current	V <sub>S</sub> = 5 V		±50		mA				
Z <sub>O</sub>	Open-loop output impedance	V <sub>S</sub> = 5 V, f = 10 MHz		100		Ω				
POWE	POWER SUPPLY									
		V <sub>S</sub> = 5.5 V, I <sub>O</sub> = 0 mA		550	750					
IQ	Quiescent current per amplifier	$V_S = 5.5 \text{ V}, I_O = 0 \text{ mA T}_A = -40^{\circ}\text{C to } 125^{\circ}\text{C}$				μA				



# 7.8 Typical Characteristics

at  $T_A = 25$  °C,  $V_S = 5.5$  V,  $R_L = 10$  k $\Omega$  connected to  $V_S$  / 2,  $V_{CM} = V_S$  / 2, and  $V_{OUT} = V_S$  / 2 (unless otherwise noted)

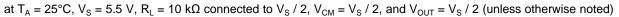


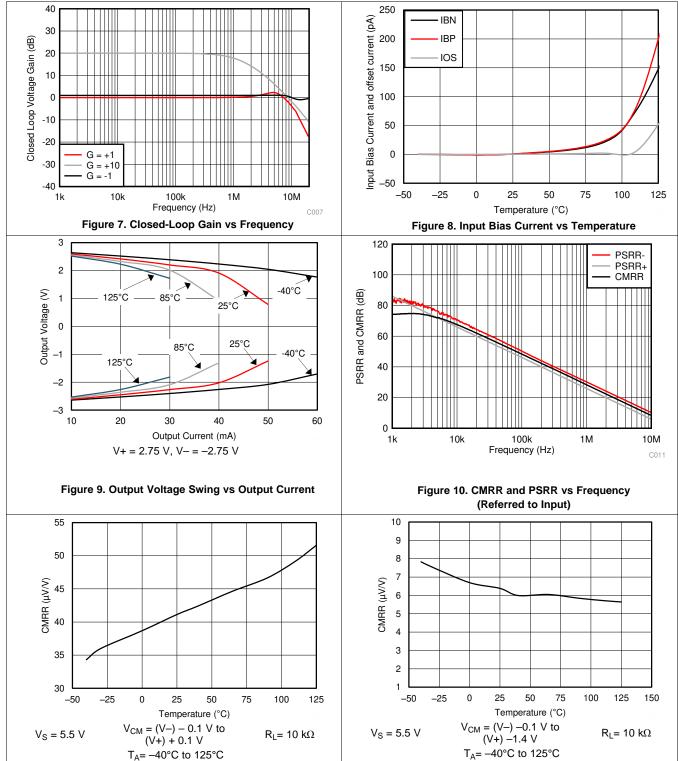
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# **Typical Characteristics (continued)**





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Figure 11. CMRR vs Temperature

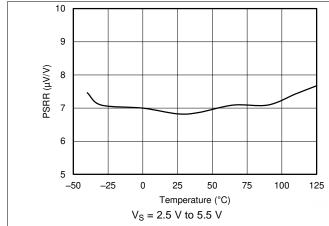
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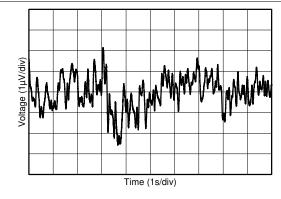
Figure 12. CMRR vs Temperature

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# **Typical Characteristics (continued)**

at  $T_A = 25$  °C,  $V_S = 5.5$  V,  $R_L = 10$  k $\Omega$  connected to  $V_S$  / 2,  $V_{CM} = V_S$  / 2, and  $V_{OUT} = V_S$  / 2 (unless otherwise noted)

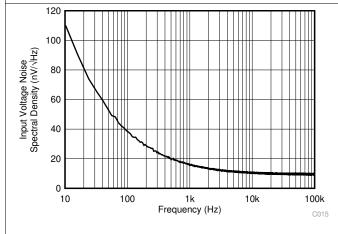




 $V_S = 2.5 \text{ V to } 5.5 \text{ V}$ 

Figure 13. PSRR vs Temperature

Figure 14. 0.1-Hz to 10-Hz Input Voltage Noise



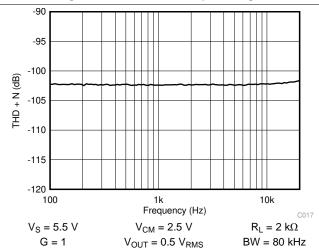
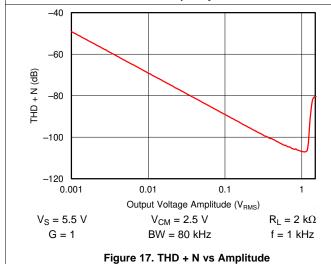


Figure 15. Input Voltage Noise Spectral Density vs Frequency

Figure 16. THD + N vs Frequency



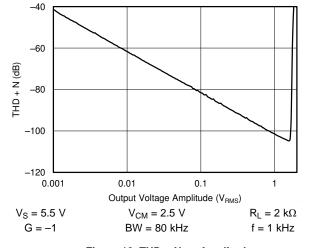
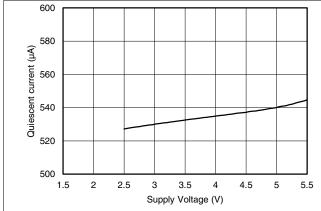


Figure 18. THD + N vs Amplitude



# **Typical Characteristics (continued)**

at  $T_A = 25$  °C,  $V_S = 5.5$  V,  $R_L = 10$  k $\Omega$  connected to  $V_S$  / 2,  $V_{CM} = V_S$  / 2, and  $V_{OUT} = V_S$  / 2 (unless otherwise noted)



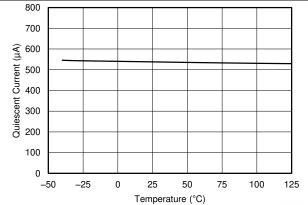
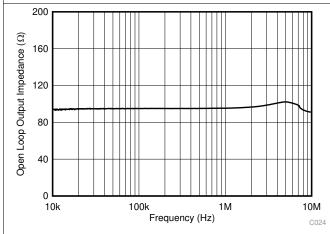


Figure 19. Quiescent Current vs Supply Voltage

Figure 20. Quiescent Current vs Temperature



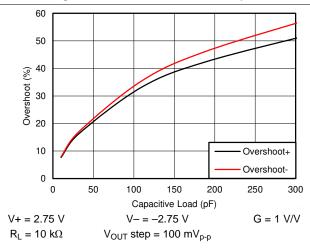
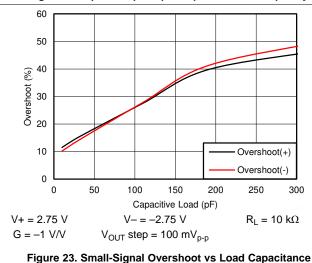


Figure 21. Open-Loop Output Impedance vs Frequency

Figure 22. Small-Signal Overshoot vs Load Capacitance



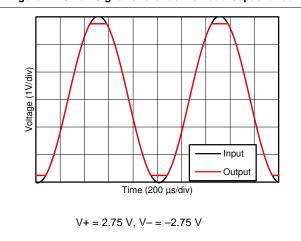
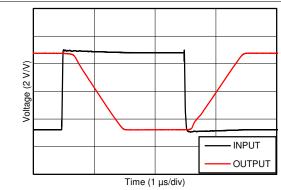


Figure 24. No Phase Reversal

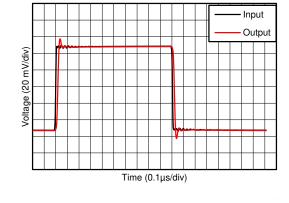
# TEXAS INSTRUMENTS

# **Typical Characteristics (continued)**

at  $T_A = 25$ °C,  $V_S = 5.5$  V,  $R_L = 10$  k $\Omega$  connected to  $V_S$  / 2,  $V_{CM} = V_S$  / 2, and  $V_{OUT} = V_S$  / 2 (unless otherwise noted)



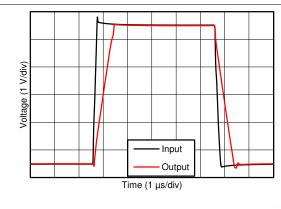
$$V+ = 2.75 V, V- = -2.75 V, G = -10 V/V$$



$$V+ = 2.75 \text{ V}, V- = -2.75 \text{ V}, G = 1 \text{ V/V}$$

Figure 25. Overload Recovery







$$V- = -2.75 V$$

$$C_L = 100 pF$$

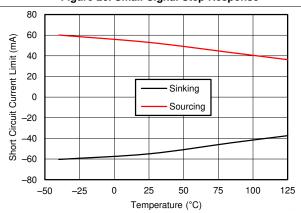


Figure 27. Large-Signal Step Response

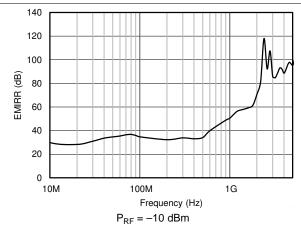
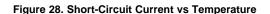


Figure 29. Electromagnetic Interference Rejection Ratio Referred to Noninverting Input (EMIRR+) vs Frequency



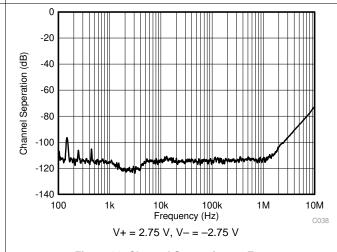
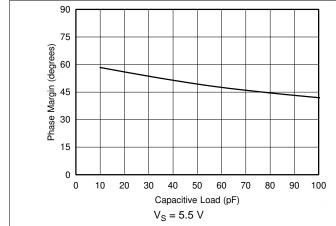


Figure 30. Channel Separation vs Frequency



# **Typical Characteristics (continued)**

at  $T_A = 25$ °C,  $V_S = 5.5$  V,  $R_L = 10$  k $\Omega$  connected to  $V_S$  / 2,  $V_{CM} = V_S$  / 2, and  $V_{OUT} = V_S$  / 2 (unless otherwise noted)



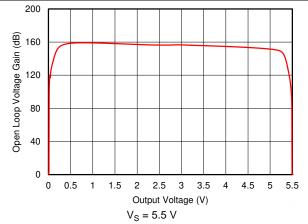
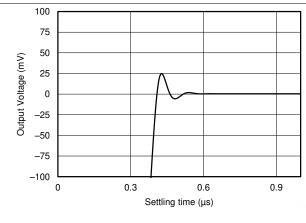


Figure 31. Phase Margin vs Capacitive Load

Figure 32. Open Loop Voltage Gain vs Output Voltage



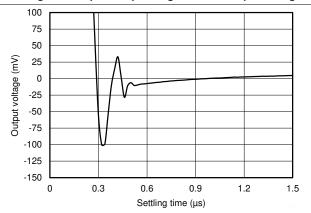


Figure 33. Large Signal Settling Time (Positive)

Figure 34. Large Signal Settling Time (Negative)

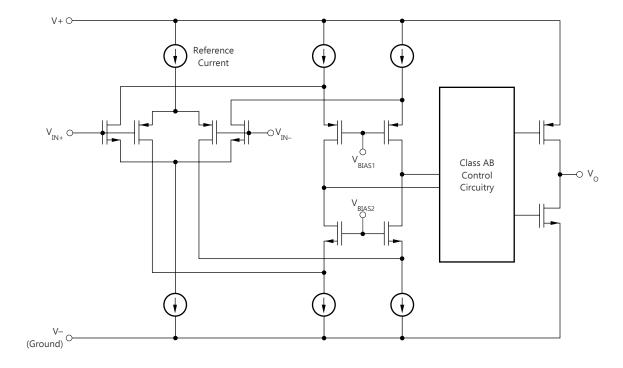


# 8 Detailed Description

#### 8.1 Overview

The TSV91x series is a family of low-power, rail-to-rail input and output op amps. These devices operate from 2.5 V to 5.5 V, are unity-gain stable, and are designed for a wide range of general-purpose applications. The input common-mode voltage range includes both rails and allows the TSV91x series to be used in virtually any single-supply application. Rail-to-rail input and output swing significantly increases dynamic range, especially in low-supply applications and are designed for driving sampling analog-to-digital converters (ADCs).

# 8.2 Functional Block Diagram



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#### 8.3 Feature Description

# 8.3.1 Rail-to-Rail Input

The input common-mode voltage range of the TSV91x family extends 100 mV beyond the supply rails for the full supply voltage range of 2.5 V to 5.5 V. This performance is achieved with a complementary input stage: an N-channel input differential pair in parallel with a P-channel differential pair, as shown in the *Functional Block Diagram*. The N-channel pair is active for input voltages close to the positive rail, typically (V+) - 1.4 V to 100 mV above the positive supply, whereas the P-channel pair is active for inputs from 100 mV below the negative supply to approximately (V+) - 1.4 V. There is a small transition region, typically (V+) - 1.2 V to (V+) - 1 V, in which both pairs are on. This 200-mV transition region can vary up to 200 mV with process variation. Thus, the transition region (with both stages on) can range from (V+) - 1.4 V to (V+) - 1.2 V on the low end, and up to (V+) - 1 V to (V+) - 0.8 V on the high end. Within this transition region, PSRR, CMRR, offset voltage, offset drift, and THD can degrade compared to device operation outside this region.

#### 8.3.2 Rail-to-Rail Output

Designed as a low-power, low-voltage operational amplifier, the TSV91x series delivers a robust output drive capability. A class AB output stage with common-source transistors achieves full rail-to-rail output swing capability. For resistive loads of 10 k $\Omega$ , the output swings to within 15 mV of either supply rail, regardless of the applied power-supply voltage. Different load conditions change the ability of the amplifier to swing close to the rails.

#### 8.3.3 Packages with an Exposed Thermal Pad

The TSV91x family is available in packages such as the WSON-8 (DSG) which feature an exposed thermal pad. Inside the package, the die is attached to this thermal pad using an electrically conductive compound. For this reason, when using a package with an exposed thermal pad, the thermal pad must either be connected to V- or left floating. Attaching the thermal pad to a potential other then V- is not allowed, and the performance of the device is not assured when doing so.

#### 8.3.4 Overload Recovery

Overload recovery is defined as the time required for the operational amplifier output to recover from a saturated state to a linear state. The output devices of the operational amplifier enter a saturation region when the output voltage exceeds the rated operating voltage, because of the high input voltage or the high gain. After the device enters the saturation region, the charge carriers in the output devices require time to return to the linear state. After the charge carriers return to the linear state, the device begins to slew at the specified slew rate. Therefore, the propagation delay (in case of an overload condition) is the sum of the overload recovery time and the slew time. The overload recovery time for the TSV91x series is approximately 200 ns.

#### 8.4 Device Functional Modes

The TSV91x family has a single functional mode. These devices are powered on as long as the power-supply voltage is between 2.5 V ( $\pm$ 1.25 V) and 5.5 V ( $\pm$ 2.75 V).



# 9 Application and Implementation

#### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

# 9.1 Application Information

The TSV91x series features 8-MHz bandwidth and 4.5-V/ $\mu$ s slew rate with only 550  $\mu$ A of supply current per channel, providing good AC performance at low power consumption. DC applications are well served with a low input noise voltage of 18 nV /  $\sqrt{\text{Hz}}$  at 1 kHz, low input bias current, and a typical input offset voltage of 0.3 mV.

# 9.2 Typical Application

Figure 35 shows the TSV91x configured in a low-side, motor-control application.

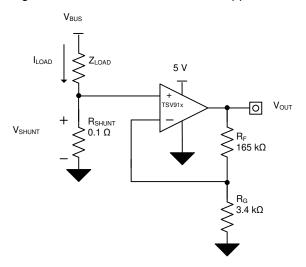


Figure 35. TSV91x in a Low-Side, Motor-Control Application

# 9.2.1 Design Requirements

The design requirements for this design are:

Load current: 0 A to 1 AOutput voltage: 4.95 V

Maximum shunt voltage: 100 mV

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## **Typical Application (continued)**

#### 9.2.2 Detailed Design Procedure

The transfer function of the circuit in Figure 35 is shown in Equation 1.

$$V_{OUT} = I_{LOAD} \times R_{SHUNT} \times Gain$$
 (1)

The load current ( $I_{LOAD}$ ) produces a voltage drop across the shunt resistor ( $R_{SHUNT}$ ). The load current is set from 0 A to 1 A. To keep the shunt voltage below 100 mV at maximum load current, the largest shunt resistor is defined using Equation 2.

$$R_{SHUNT} = \frac{V_{SHUNT\_MAX}}{I_{LOAD\_MAX}} = \frac{100mV}{1A} = 100m\Omega$$
(2)

Using Equation 2,  $R_{SHUNT}$  is 100 m $\Omega$ . The voltage drop produced by  $I_{LOAD}$  and  $R_{SHUNT}$  is amplified by the TSV91x to produce an output voltage of approximately 0 V to 4.95 V. The gain required by the TSV91x to produce the necessary output voltage is calculated using Equation 3:

$$Gain = \frac{\left(V_{OUT\_MAX} - V_{OUT\_MIN}\right)}{\left(V_{IN\_MAX} - V_{IN\_MIN}\right)}$$
(3)

Using Equation 3, the required gain is calculated to be 49.5 V/V, which is set with resistors  $R_F$  and  $R_G$ . Equation 4 is used to size the resistors,  $R_F$  and  $R_G$ , to set the gain of the TSV91x to 49.5 V/V.

$$Gain = 1 + \frac{(R_F)}{(R_G)}$$
(4)

Selecting  $R_F$  as 165  $k\Omega$  and  $R_G$  as 3.4  $k\Omega$  provides a combination that equals roughly 49.5 V/V. Figure 36 shows the measured transfer function of the circuit shown in Figure 35.

#### 9.2.3 Application Curve

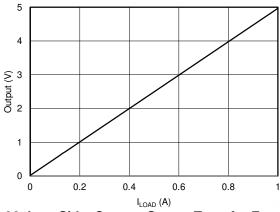


Figure 36. Low-Side, Current-Sense, Transfer Function



# 10 Power Supply Recommendations

The TSV91x series is specified for operation from 2.5 V to 5.5 V (±1.25 V to ±2.75 V); many specifications apply from -40°C to 125°C. The *Typical Characteristics* section presents parameters that can exhibit significant variance with regard to operating voltage or temperature.

#### **CAUTION**

Supply voltages larger than 6 V can permanently damage the device; see the *Absolute Maximum Ratings* table.

Place 0.1-µF bypass capacitors close to the power-supply pins to reduce errors coupling in from noisy or high-impedance power supplies. For more detailed information on bypass capacitor placement, see the *Layout Example* section.

# 10.1 Input and ESD Protection

The TSV91x series incorporates internal ESD protection circuits on all pins. For input and output pins, this protection consists of current-steering diodes connected between the input and power-supply pins. These ESD protection diodes provide in-circuit, input overdrive protection, as long as the current is limited to 10-mA, as stated in the *Absolute Maximum Ratings* table. Figure 37 shows how a series input resistor is added to the driven input to limit the input current. The added resistor contributes thermal noise at the amplifier input and the value must be kept to a minimum in noise-sensitive applications.

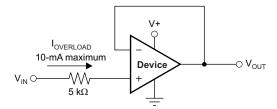


Figure 37. Input Current Protection

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# 11 Layout

# 11.1 Layout Guidelines

For best operational performance of the device, use good printed-circuit board (PCB) layout practices, including:

- Noise can propagate into analog circuitry through the power pins of the circuit as a whole and of op amp itself. Bypass capacitors are used to reduce the coupled noise by providing low-impedance power sources local to the analog circuitry.
  - Connect low-ESR, 0.1-µF ceramic bypass capacitors between each supply pin and ground, placed as close to the device as possible. A single bypass capacitor from V+ to ground is applicable for singlesupply applications.
- Separate grounding for analog and digital portions of circuitry is one of the simplest and most-effective
  methods of noise suppression. One or more layers on multilayer PCBs are usually devoted to ground
  planes. A ground plane helps distribute heat and reduces electromagnetic interference (EMI) noise
  pickup. Make sure to physically separate digital and analog grounds, paying attention to the flow of the
  ground current. For more detailed information, see Circuit Board Layout Techniques.
- To reduce parasitic coupling, run the input traces as far away from the supply or output traces as possible. If these traces cannot be kept separate, crossing the sensitive trace perpendicular is much better as opposed to in parallel with the noisy trace.
- Place the external components as close to the device as possible. As shown in Figure 39, keeping RF and RG close to the inverting input minimizes parasitic capacitance on the inverting input.
- Keep the length of input traces as short as possible. Always remember that the input traces are the most sensitive part of the circuit.
- Consider a driven, low-impedance guard ring around the critical traces. A guard ring can significantly reduce leakage currents from nearby traces that are at different potentials.
- Cleaning the PCB following board assembly is recommended for best performance.
- Any precision integrated circuit can experience performance shifts resulting from moisture ingress into the
  plastic package. Following any aqueous PCB cleaning process, baking the PCB assembly is
  recommended to remove moisture introduced into the device packaging during the cleaning process. A
  low-temperature, post-cleaning bake at 85°C for 30 minutes is sufficient for most circumstances.

## 11.2 Layout Example

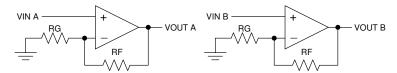


Figure 38. Schematic Representation for Figure 39

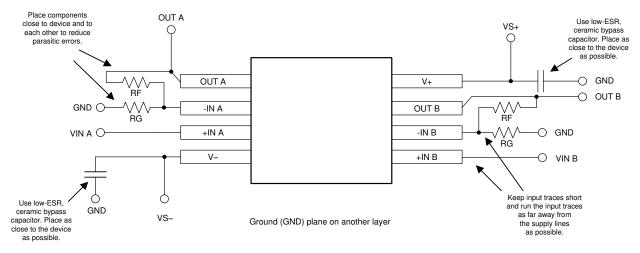


Figure 39. Layout Example

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# 12 Device and Documentation Support

## 12.1 Documentation Support

#### 12.1.1 Related Documentation

For related documentation see the following:

Texas Instruments, Circuit Board Layout Techniques, SLOA089

#### 12.2 Related Links

The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to order now.

Table 1. Related Links

PARTS	PRODUCT FOLDER	ORDER NOW	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY
TSV911	Click here	Click here	Click here	Click here	Click here
TSV912	Click here	Click here	Click here	Click here	Click here
TSV914	Click here	Click here	Click here	Click here	Click here

#### 12.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

# 12.4 Community Resources

TI E2E™ support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

#### 12.5 Trademarks

E2E is a trademark of Texas Instruments.

All other trademarks are the property of their respective owners.

#### 12.6 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

# 12.7 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

# 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



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#### **PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
TSV911AIDBVR	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU   SN	Level-1-260C-UNLIM	-40 to 125	1U2F	Samples
TSV911AIDCKR	ACTIVE	SC70	DCK	5	3000	RoHS & Green	SN	Level-2-260C-1 YEAR	-40 to 125	1EK	Samples
TSV912AIDDFR	ACTIVE	SOT-23-THIN	DDF	8	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	T12A	Samples
TSV912AIDGKR	ACTIVE	VSSOP	DGK	8	2500	RoHS & Green	NIPDAU   SN   NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	T912	Samples
TSV912AIDGKT	ACTIVE	VSSOP	DGK	8	250	RoHS & Green	NIPDAU   SN   NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	T912	Samples
TSV912AIDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU   SN	Level-2-260C-1 YEAR	-40 to 125	TSV912	Samples
TSV912AIDSGR	ACTIVE	WSON	DSG	8	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	T912	Samples
TSV912AIDSGT	ACTIVE	WSON	DSG	8	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	T912	Samples
TSV912AIPWR	ACTIVE	TSSOP	PW	8	2000	RoHS & Green	NIPDAU   SN	Level-2-260C-1 YEAR	-40 to 125	TSV912	Samples
TSV914AIDR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	TSV914AD	Samples
TSV914AIPWR	ACTIVE	TSSOP	PW	14	2000	RoHS & Green	SN	Level-2-260C-1 YEAR	-40 to 125	TSV914	Samples
TSV914AIPWT	ACTIVE	TSSOP	PW	14	250	RoHS & Green	SN	Level-2-260C-1 YEAR	-40 to 125	TSV914	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

**ACTIVE**: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

<sup>(2)</sup> **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".



# **PACKAGE OPTION ADDENDUM**

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- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.



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# TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TSV911AIDBVR	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TSV911AIDBVR	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TSV911AIDCKR	SC70	DCK	5	3000	180.0	8.4	2.3	2.5	1.2	4.0	8.0	Q3
TSV912AIDDFR	SOT-23- THIN	DDF	8	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TSV912AIDGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TSV912AIDGKT	VSSOP	DGK	8	250	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TSV912AIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TSV912AIDSGR	WSON	DSG	8	3000	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
TSV912AIDSGT	WSON	DSG	8	250	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
TSV912AIPWR	TSSOP	PW	8	2000	330.0	12.4	7.0	3.6	1.6	8.0	12.0	Q1
TSV914AIDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TSV914AIPWR	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TSV914AIPWT	TSSOP	PW	14	250	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1



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\*All dimensions are nominal

		,					
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TSV911AIDBVR	SOT-23	DBV	5	3000	210.0	185.0	35.0
TSV911AIDBVR	SOT-23	DBV	5	3000	210.0	185.0	35.0
TSV911AIDCKR	SC70	DCK	5	3000	210.0	185.0	35.0
TSV912AIDDFR	SOT-23-THIN	DDF	8	3000	210.0	185.0	35.0
TSV912AIDGKR	VSSOP	DGK	8	2500	356.0	356.0	35.0
TSV912AIDGKT	VSSOP	DGK	8	250	356.0	356.0	35.0
TSV912AIDR	SOIC	D	8	2500	356.0	356.0	35.0
TSV912AIDSGR	WSON	DSG	8	3000	210.0	185.0	35.0
TSV912AIDSGT	WSON	DSG	8	250	210.0	185.0	35.0
TSV912AIPWR	TSSOP	PW	8	2000	356.0	356.0	35.0
TSV914AIDR	SOIC	D	14	2500	356.0	356.0	35.0
TSV914AIPWR	TSSOP	PW	14	2000	366.0	364.0	50.0
TSV914AIPWT	TSSOP	PW	14	250	366.0	364.0	50.0



SMALL OUTLINE PACKAGE



#### NOTES:

PowerPAD is a trademark of Texas Instruments.

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-187.



SMALL OUTLINE PACKAGE



NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
- 8. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.
- 9. Size of metal pad may vary due to creepage requirement.



SMALL OUTLINE PACKAGE



NOTES: (continued)

- 11. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 12. Board assembly site may have different recommendations for stencil design.





PLASTIC SMALL OUTLINE



# NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.



PLASTIC SMALL OUTLINE



NOTES: (continued)

- 4. Publication IPC-7351 may have alternate designs.
- 5. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



PLASTIC SMALL OUTLINE



NOTES: (continued)

- 6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 7. Board assembly site may have different recommendations for stencil design.





SMALL OUTLINE TRANSISTOR



#### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
  2. This drawing is subject to change without notice.
  3. Reference JEDEC MO-178.

- 4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.
- 5. Support pin may differ or may not be present.



SMALL OUTLINE TRANSISTOR



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.





- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.







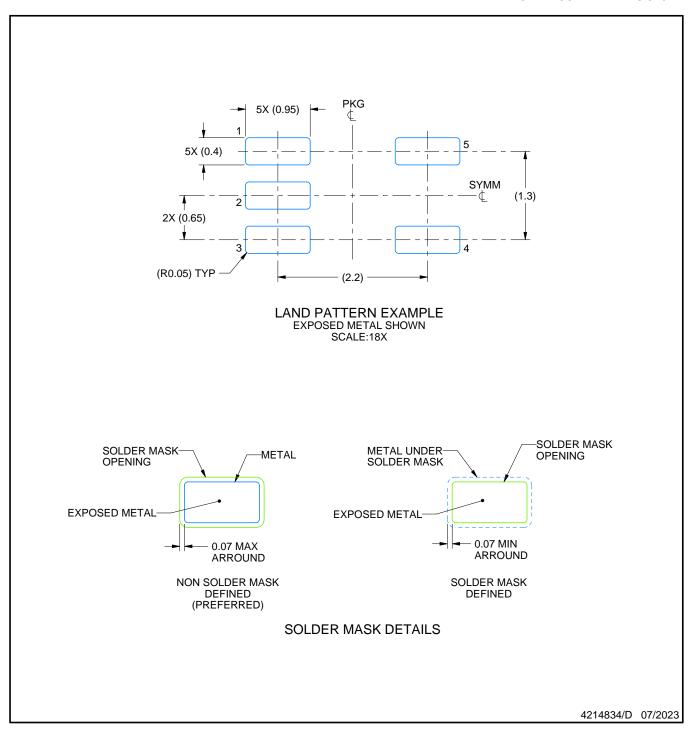
- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. Reference JEDEC MO-203.

- 4. Support pin may differ or may not be present.5. Lead width does not comply with JEDEC.





NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.





- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



# D (R-PDSO-G14)

## PLASTIC SMALL OUTLINE



- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AB.



PW (R-PDSO-G14)

# PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M—1994.
- B. This drawing is subject to change without notice.
  - Sody length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.
- E. Falls within JEDEC MO-153



# PW (R-PDSO-G14)

# PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.





SMALL OUTLINE INTEGRATED CIRCUIT



- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- 4. This dimension does not include interlead flash.
- 5. Reference JEDEC registration MS-012, variation AA.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE INTEGRATED CIRCUIT



- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



2 x 2, 0.5 mm pitch

PLASTIC SMALL OUTLINE - NO LEAD

This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.





PLASTIC SMALL OUTLINE - NO LEAD



- All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.



PLASTIC SMALL OUTLINE - NO LEAD



- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



PLASTIC SMALL OUTLINE - NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.





SMALL OUTLINE PACKAGE



- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-153, variation AA.



SMALL OUTLINE PACKAGE



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE PACKAGE



- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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