# Power-Dissipation Calculations for TI FIFO Products 

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#### Abstract

Power consumption has become a major consideration in today's circuit design. Low power consumption is one of the major advantages of Texas Instruments (TI) FIFO products. Power calculations are required to meet the design requirements relating to the chip temperature and system power. The only way that a designer can minimize the power requirements of a board or a system is to understand and control the causes. This application report assists the component and system design engineer to evaluate the power consumption of TI ACT and ABT FIFO products.


## Introduction

A simple method of calculating power dissipation for FIFO products under varying conditions is presented in this application report. Power calculations include the power dissipation when a FIFO is transferring data as well as when only clocks are running and no data is being transferred. Power consumption in a FIFO product is dependent on clock switching frequency, data-input switching frequency, data-output switching frequency, and data-output capacitive loading.
In this report, a simplified introduction to the physics of CMOS devices regarding power consumption is presented. In the second part of this application report, power-calculation examples for FIFO products are presented in two subsections. The first subsection includes power-dissipation calculations for advanced CMOS (ACT) FIFO products, including an example using the SN74ACT3632. The second subsection includes power-dissipation calculations for advanced BiCMOS (ABT) FIFO products, including an example using the SN74ABT3614. Appendix A includes several graphs of $\mathrm{I}_{\mathrm{CC}}$ versus frequency to assist the design engineer with the information required for calculating power consumption. In addition, the graphs of $\mathrm{I}_{\mathrm{CC}}$ versus frequency assist the designer in selecting a device with the lowest power consumption. Appendix B presents reliability information and a table of maximum power versus ambient temperature for different package types. Finally, the goals, achievements, and results of this application report are included in the summary.

## CMOS and BiCMOS Power Basics

Power dissipation is dependent on supply voltage $\left(\mathrm{V}_{\mathrm{CC}}\right)$ and supply current $\left(\mathrm{I}_{\mathrm{CC}}\right)$. It is calculated using the formula:

$$
\mathrm{P}=\mathrm{V}_{\mathrm{CC}} \times \mathrm{I}_{\mathrm{CC}}
$$

Any CMOS function can be broken down to a gate-level model. The simplest CMOS circuit is an inverter as shown in Figure 1. When the input voltage is at ground or $\mathrm{V}_{\mathrm{CC}}$ level, one transistor is fully on and the other transistor is fully off. This results in a negligible $\mathrm{I}_{\mathrm{CC}}$ that is simply the reverse-leakage current flowing through the nonconducting transistor. Quiescent power is due to this current (referred to as idle $\mathrm{I}_{\mathrm{CC}}$ in this application report).


Figure 1. Gate Model of a CMOS Inverter

When the input switches from $V_{\mathrm{CC}}$ to ground or vice versa during the transition period, both transistors are on, resulting in current flow from $\mathrm{V}_{\mathrm{CC}}$ to ground. This current is called through current (referred to as active current in this application report). For many applications using CMOS and BiCMOS devices, switching power accounts for most of the power consumption.

The through-current waveform supplied by $\mathrm{V}_{\mathrm{CC}}$ to a CMOS gate is shown in Figure 2. As the switching frequency increases, the number of current spikes also increases. For instance, if the switching frequency is doubled, the number of current spikes double. Figure 3 shows the current spikes for the same device at twice the frequency of the signal shown in Figure 2. Since power is directly proportional to the RMS current, the increase in frequency results in increased power dissipation. Power consumption due to the load should also be considered. For a CMOS device with an entirely capacitive load, the transient power due to the load is calculated using the formula:

$$
\mathrm{P}=\mathrm{C}_{\mathrm{L}} \times \mathrm{V}_{\mathrm{CC}}{ }^{2} \times \mathrm{f}_{\mathrm{o}}
$$

Where:

$$
\begin{aligned}
\mathrm{C}_{\mathrm{L}} & =\text { load capacitance } \\
\mathrm{V}_{\mathrm{CC}} & =\text { supply voltage } \\
\mathrm{f}_{\mathrm{O}} & =\text { output switching frequency }
\end{aligned}
$$

Power calculations are presented in more detail in the following sections of this application report.


Figure 2. Current Waveform Supplied by $\mathrm{V}_{\mathrm{Cc}}$ to a CMOS Gate


Figure 3. Current Waveform When Switching Frequency Is Doubled

## Power Calculation

The total power consumption in the ACT and ABT FIFOs is the sum of the quiescent or nonswitching power $\left(\mathrm{P}_{\mathrm{q}}\right)$ and dynamic or switching power $\left(\mathrm{P}_{\mathrm{d}}\right)$ :

$$
\mathrm{P}_{\mathrm{T}}=\mathrm{P}_{\mathrm{d}}\left(\mathrm{DC}_{\mathrm{d}}\right)+\mathrm{P}_{\mathrm{q}}\left(1-\mathrm{DC}_{\mathrm{d}}\right)
$$

Where:

| $\mathrm{P}_{\mathrm{T}}$ | $=$ total power |
| :--- | :--- |
| $\mathrm{P}_{\mathrm{d}}$ | $=$ dynamic or switching power |
| $\mathrm{P}_{\mathrm{q}}$ | $=$ quiescent or nonswitching power |
| DC |  |
| $1-\mathrm{DC}_{\mathrm{d}}$ | $=\%$ time FIFO is switching |
|  | $\%$ time FIFO is not switching |

## Quiescent or Nonswitching Power Dissipation in ACT FIFO Devices (CMOS FIFO products)

The quiescent power consumed by a CMOS device is given by the formula:

$$
\mathrm{P}_{\mathrm{q}}=\mathrm{V}_{\mathrm{CC}} \times \mathrm{I}_{\mathrm{CC}}(\text { total })
$$

Where $\mathrm{V}_{\mathrm{CC}}$ is the supply voltage and $\mathrm{I}_{\mathrm{CC}}$ (total) includes the increase in $\mathrm{I}_{\mathrm{CC}}$ due to inputs being driven by TTL devices. This is calculated as:

$$
\mathrm{I}_{\mathrm{CC}}(\text { total })=\left[\mathrm{N}_{\mathrm{TTL}} \times \Delta \mathrm{I}_{\mathrm{CC}} \times \mathrm{DC}_{\mathrm{VIH}}\right]+\mathrm{I}_{\mathrm{CCI}}
$$

Where:

| $\mathrm{I}_{\mathrm{CCI}}$ | $=\mathrm{f}_{\text {clock }} \times \mathrm{pF}_{\text {(clock) }}=$ supply current when FIFO is idle (the clocks are running |
| ---: | :--- |
|  | but no data is written to or read from the FIFO) |
| $\mathrm{f}_{\text {clock }}$ | $=$ clock switching frequency |
| $\mathrm{pF}_{(\text {clock }}$ | $=$ clock switching power factor, the slope of $\mathrm{I}_{\mathrm{CC}}$ versus $\mathrm{f}_{\text {clock }}$ curve (see data sheet) |
| $\mathrm{N}_{\mathrm{TTL}}$ | $=$ number of inputs driven by TTL levels |
| $\Delta \mathrm{I}_{\mathrm{CC}}$ | $=$ increase in supply current for each input at a TTL high level (see data sheet) |
| $\mathrm{DC}_{\mathrm{VIH}}$ | $=\%$ of TTL signals at a high level of 3.4 V |

Therefore, the quiescent power is calculated by:

$$
\mathrm{P}_{\mathrm{q}}=\mathrm{V}_{\mathrm{CC}} \times\left[\mathrm{I}_{\mathrm{CCI}}+\left(\mathrm{N}_{\mathrm{TTL}} \times \Delta \mathrm{I}_{\mathrm{CC}} \times \mathrm{DC}_{\mathrm{VIH}}\right)\right]
$$

## Dynamic or Switching Power Dissipation in ACT FIFO Devices (CMOS FIFO products)

For most applications, dynamic power accounts for most of the total power dissipation of a CMOS device. Dynamic power is dependent on the load capacitance, output switching frequency, input switching frequency, and the power-dissipation capacitance of the device. The following equation is typically used to calculate power consumption in a CMOS device (refer to Texas Instruments Advanced CMOS Logic Designer's Handbook, literature number SCAA001A).

$$
\mathrm{P}_{\mathrm{d}}=\left(\mathrm{C}_{\mathrm{pd}} \times \mathrm{V}_{\mathrm{CC}}^{2} \times \mathrm{f}_{\mathrm{i}}\right)+\sum\left(\mathrm{C}_{\mathrm{L}} \times \mathrm{V}_{\mathrm{CC}}^{2} \times \mathrm{f}_{\mathrm{o}}\right)
$$

The value $\mathrm{C}_{\mathrm{pd}}$ is not provided for most FIFO devices. It is more accurate to calculate power of a FIFO device by obtaining active $\mathrm{I}_{\mathrm{CC}}$ versus frequency curves, $\mathrm{I}_{\mathrm{CCf}}$ (supply current when the FIFO is transferring data) and the slope of the $\mathrm{I}_{\mathrm{CC}}$ versus frequency curve (which essentially display the same information). Consequently, dynamic power includes the power dissipation due to active $\mathrm{I}_{\mathrm{CC}}$ without the output load and power dissipation due to the output load current.

$$
\mathrm{P}_{\mathrm{d}}=\mathrm{V}_{\mathrm{CC}} \times \mathrm{I}_{\mathrm{CC}}(\text { total })+\sum\left(\mathrm{C}_{\mathrm{L}} \times \mathrm{V}_{\mathrm{CC}}^{2} \times \mathrm{f}_{\mathrm{o}}\right)
$$

Where:

$$
\begin{array}{ll}
\mathrm{V}_{\mathrm{CC}} \times \mathrm{I}_{\mathrm{CC}} \text { (total) } & =\text { device switching power without load } \\
\mathrm{I}_{\mathrm{CC}} \text { (total) } & =\left[\mathrm{I}_{\mathrm{CCf}}+\left(\mathrm{N}_{\mathrm{TTL}} \times \Delta \mathrm{I}_{\mathrm{CC}} \times \mathrm{DC}_{\mathrm{VIH}}\right)\right]
\end{array}
$$

$\Sigma\left(\mathrm{C}_{\mathrm{L}} \times \mathrm{V}_{\mathrm{CC}}{ }^{2} \times \mathrm{f}_{\mathrm{o}}\right)=$ power due to output switching frequency and load capacitance

Therefore, dynamic power consumption is calculated by:

$$
\mathrm{P}_{\mathrm{d}}=\mathrm{V}_{\mathrm{CC}} \times\left[\mathrm{I}_{\mathrm{CCf}}+\left(\mathrm{N}_{\mathrm{TTL}} \times \Delta \mathrm{I}_{\mathrm{CC}} \times \mathrm{DC}_{\mathrm{VIH}}\right)\right]+\sum\left(\mathrm{C}_{\mathrm{L}} \times \mathrm{V}_{\mathrm{CC}}^{2} \times \mathrm{f}_{\mathrm{o}}\right)
$$

Where:

| $\mathrm{V}_{\mathrm{CC}}$ | $=$ supply voltage |
| :--- | :--- |
| $\mathrm{I}_{\mathrm{CCf}}$ | $=$ supply current when the FIFO is transferring data (see active $\mathrm{I}_{\mathrm{CC}}$ versus frequency plot |
|  | in data sheet) |
| $\mathrm{N}_{\mathrm{TTL}}$ | $=$ number of inputs driven by TTL levels |
| $\Delta \mathrm{I}_{\mathrm{CC}}$ | $=$ increase in supply current for each input at a TTL high level (see data sheet) |
| $\mathrm{DC}_{\mathrm{VIH}}=$ | $\%$ of inputs at a TTL high level |
| $\mathrm{C}_{\mathrm{L}}$ | $=$ load capacitance |
| $\mathrm{f}_{\mathrm{O}}$ | $=$ output switching frequency |

## Example 1

This example shows how to calculate the power dissipation for an SN74ACT3632 bidirectional FIFO used in a system under the following conditions:

- Data input lines (A0-A35) are driven by a TTL device.
- Control signals (CLKA, CLKB, $\overline{\mathrm{CSA}}, \overline{\mathrm{CSB}}, \mathrm{W} / \overline{\mathrm{R}} \mathrm{A}, \overline{\mathrm{W}} / \mathrm{RB}, \overline{\mathrm{RST}}, \overline{\mathrm{RST}}, \mathrm{ENA}, \mathrm{ENB}, \mathrm{MBA}, \mathrm{MBB}, \mathrm{FSO}, \mathrm{FS} 1)$ are driven by a CMOS device.
- The output of the FIFO is fed to a memory device.
- Only $3 / 4$ of the inputs (or outputs) are switching at a given time.
- The port-B clock rate is 33.3 MHz and the port-A clock rate is 40 MHz .
- The SN74ACT3632 is only used about $1 / 3$ of the time by the system.
- The load capacitance on each output is about 30 pF , and the supply voltage is set at 5 V .
- The SN74ACT3632 is used equally in both directions.

The following parametric values are needed to calculate power dissipation:
Where:

```
\(\mathrm{V}_{\mathrm{CC}} \quad=\) supply voltage \(=5 \mathrm{~V}\)
\(\mathrm{DC}_{\mathrm{d}}=\%\) time FIFO is switching \(=1 / 3\)
\(1-\mathrm{DC}_{\mathrm{d}}=\%\) time FIFO is not switching \(=2 / 3\)
\(\mathrm{f}_{\text {clockA }}=\) clock switching frequency of port \(\mathrm{A}=40 \mathrm{MHz}\)
\(\mathrm{f}_{\text {clock }}=\) clock switching frequency of port \(\mathrm{B}=33.3 \mathrm{MHz}\)
\(\mathrm{pF} \quad=\) clock switching power factor, the slope of \(\mathrm{I}_{\mathrm{CC}}\) versus \(\mathrm{f}_{\text {clock }}\) curve (see data sheet)
    \(=0.184 \mathrm{~mA} / \mathrm{MHz}\)
\(\mathrm{I}_{\mathrm{CCI}}=\mathrm{I}_{\mathrm{CLKA}}+\mathrm{I}_{\mathrm{CLKB}}=\left(\mathrm{f}_{\text {clockA }} \times \mathrm{pF}\right)+\left(\mathrm{f}_{\text {clockB }} \times \mathrm{pF}\right)\)
    \(=(40 \mathrm{MHz}+33.3 \mathrm{MHz}) \times 0.184 \mathrm{~mA} / \mathrm{MHz}=13.4872 \mathrm{~mA}\)
\(\mathrm{N}_{\text {TTL }}=\) number of inputs driven by TTL levels \(=36\)
\(\mathrm{I}_{\mathrm{CCf}}=\) active supply current when FIFO is transferring data from the curve if
    \(\mathrm{I}_{\mathrm{CC}}\) versus frequency \(=115 \mathrm{~mA} @ 33.3 \mathrm{MHz}\)
\(\Delta \mathrm{I}_{\mathrm{CC}} \quad=\) increase in supply current for each input at a TTL high level (see data sheet)
    \(=0 \mathrm{~mA}\) if \(\overline{\mathrm{CSA}}=\mathrm{V}_{\mathrm{IH}}\) or \(\overline{\mathrm{CSB}}=\mathrm{V}_{\mathrm{IH}}\)
    \(=1 \mathrm{~mA}\) if \(\overline{\mathrm{CSA}}=\mathrm{V}_{\mathrm{IL}}\) or \(\overline{\mathrm{CSB}}=\mathrm{V}_{\mathrm{IL}}\)
    \(=1 \mathrm{~mA}\) for all other inputs (see data sheet)
```

In this example ( $\Delta \mathrm{I}_{\mathrm{CC}}=1 \mathrm{~mA}$ is assumed ):

```
\(\mathrm{DC}_{\text {VIH }}=3 / 4\)
\(\mathrm{C}_{\mathrm{L}}=\) load capacitance \(=30 \mathrm{pF}\)
\(\mathrm{f}_{\mathrm{O}} \quad=\) output switching frequency \(=1 / 2\) (since maximum data rate is \(1 / 2\) clock frequency)
    \(\times 3 / 4\) (since \(3 / 4\) of the outputs are switching at a given time)
    \(\times 33.3\) (slowest of the two clock frequencies, \(\mathrm{f}_{\text {clockA }}\) or \(\left.\mathrm{f}_{\text {clockB }}\right)=12.4875 \mathrm{MHz}\)
```

$\mathrm{I}_{\mathrm{CCf}}$ and $\mathrm{I}_{\mathrm{CCI}}$ are taken from graphs of $\mathrm{I}_{\mathrm{CC}}$ versus clock frequency. In the case of SN74ACT3632, $\mathrm{I}_{\mathrm{CCI}}$ is taken for either of the two clocks while only one clock is switching. All other inputs are tied to 0 or to $\mathrm{V}_{\mathrm{CC}}-0.2 \mathrm{~V}$ and all the outputs are disconnected. Later, $\mathrm{I}_{\mathrm{CCf}}$ is measured while simultaneously reading and writing a FIFO with both CLKA and CLKB set to $\mathrm{f}_{\text {clock }}$.


Figure 4. SN74ACT3632 Active ICC Versus Frequency

## Solution

$$
\begin{aligned}
\mathrm{P}_{\mathrm{T}} & =\mathrm{P}_{\mathrm{T}}(\text { from } \mathrm{A} \text { to } \mathrm{B})+\mathrm{P}_{\mathrm{T}}(\text { from } \mathrm{B} \text { to } \mathrm{A}) \\
& =\mathrm{P}_{\mathrm{q}}(\text { from } \mathrm{A} \text { to } \mathrm{B})+\mathrm{P}_{\mathrm{q}}(\text { from } \mathrm{B} \text { to } \mathrm{A})+\mathrm{P}_{\mathrm{d}}(\text { from } \mathrm{A} \text { to } \mathrm{B})+\mathrm{P}_{\mathrm{d}}(\text { from } \mathrm{B} \text { to } \mathrm{A}) \\
& =\mathrm{P}_{\mathrm{q}}+\mathrm{P}_{\mathrm{d}}
\end{aligned}
$$

Where $\mathrm{P}_{\mathrm{q}}$ and $\mathrm{P}_{\mathrm{d}}$ include power from A to B and from B to A directions.

## Quiescent Power

$$
\begin{aligned}
\mathrm{P}_{\mathrm{q}} & =\mathrm{V}_{\mathrm{CC}} \times\left[\mathrm{I}_{\mathrm{CCI}}+\left(\mathrm{N}_{\mathrm{TTL}} \times \Delta \mathrm{I}_{\mathrm{CC}} \times \mathrm{DC}_{\mathrm{VIH}}\right)\right] \\
& =5 \times[13.4872 \mathrm{~mA}+(36 \times 1 \mathrm{~mA} \times 3 / 4)]=202.436 \mathrm{~mW}
\end{aligned}
$$

## Dynamic Power

$$
\begin{aligned}
\mathrm{P}_{\mathrm{d}} & =\mathrm{V}_{\mathrm{CC}} \times\left[\mathrm{I}_{\mathrm{CCf}}+\left(\mathrm{N}_{\mathrm{TTL}} \times \Delta \mathrm{I}_{\mathrm{CC}} \times \mathrm{DC}_{\mathrm{VIH}}\right)\right]+\sum\left(\mathrm{C}_{\mathrm{L}} \times \mathrm{V}_{\mathrm{CC}}^{2} \times \mathrm{f}_{\mathrm{O}}\right) \\
& =5 \times[115 \mathrm{~mA}+(36 \times 1 \mathrm{~mA} \times 3 / 4)]+\left[36 \times 30 \mathrm{pF} \times(5 \mathrm{~V})^{2} \times 12.4875 \mathrm{MHz}\right] \\
& =710 \mathrm{~mW}+337.1625 \mathrm{~mW}=1047.1625 \mathrm{~mW}
\end{aligned}
$$

## Total Power

$$
\begin{aligned}
\mathrm{P}_{\mathrm{T}} & =\mathrm{P}_{\mathrm{d}}\left(\mathrm{DC}_{\mathrm{d}}\right)+\mathrm{P}_{\mathrm{q}}\left(1-\mathrm{DC}_{\mathrm{d}}\right) \\
& =1047.1625 \mathrm{~mW} \times 1 / 3+202.436 \mathrm{~mW} \times 2 / 3=484.0115 \mathrm{~mW}
\end{aligned}
$$

Therefore, total power is approximately:

$$
\mathrm{P}_{\mathrm{T}}=484 \mathrm{~mW}
$$

The SN74ACT3632 is available in 120-pin TQFP and 132-pin PQFP packages (refer to Appendix B for the maximum power curve calculated for reliability purposes). At maximum ambient temperature $\left(70^{\circ} \mathrm{C}\right)$ and no air flow for 132 -pin PQFP and 120-pin TQFP packages, the maximum power that the packages can dissipate to free air is 1610 mW and 1606 mW , respectively. The SN74ACT3632, in this example, meets the reliability requirement since 484 mW is much less than 1610 mW or 1606 mW .

## Quiescent or Nonswitching Power Dissipation in ABT FIFO Devices (BiCMOS FIFO products)

Unlike CMOS devices that have a single value for $\mathrm{I}_{\mathrm{CC}}$, BiCMOS devices have varying static current levels depending on the state of the output ( $\mathrm{I}_{\mathrm{CCL}}, \mathrm{I}_{\mathrm{CCH}}, \mathrm{I}_{\mathrm{CCZ}}$ ). Quiescent power includes the power consumed while outputs are active, the power consumed when outputs are disabled, and the power consumed by the switching clocks. The design of the BiCMOS inputs is such that when a TTL high level is applied at the input, it does not increase the current; therefore, the $\Delta \mathrm{I}_{\mathrm{CC}}$ term $\left(\mathrm{N}_{\mathrm{TTL}} \times \Delta \mathrm{I}_{\mathrm{CC}} \times \mathrm{DC}_{\mathrm{d}}\right)$ is excluded from the following equation (from Texas Instruments ABT Advanced BiCMOS Technology Data Book, 1993, literature \# SCBD002A) in calculating power for ABT FIFO products.

$$
\mathrm{P}_{\mathrm{q}}=\mathrm{V}_{\mathrm{CC}} \times\left[\mathrm{DC}_{\mathrm{EN}} \times\left(\mathrm{N}_{\mathrm{H}} \times \mathrm{I}_{\mathrm{CCH}} / \mathrm{N}_{\mathrm{T}}+\mathrm{N}_{\mathrm{L}} \times \mathrm{I}_{\mathrm{CCL}} / \mathrm{N}_{\mathrm{T}}\right)+\left(1-\mathrm{DC}_{\mathrm{EN}}\right) \mathrm{I}_{\mathrm{CCZ}}+\mathrm{I}_{\mathrm{CCI}}\right]
$$

Where:

$$
\left.\begin{array}{rl}
\mathrm{V}_{\mathrm{CC}} \times\left[\mathrm{DC}_{\mathrm{EN}} \times\left(\mathrm{N}_{\mathrm{H}} \times \mathrm{I}_{\mathrm{CCH}} / \mathrm{N}_{\mathrm{T}}+\mathrm{N}_{\mathrm{L}} \times \mathrm{I}_{\mathrm{CCL}} / \mathrm{N}_{\mathrm{T}}\right)\right]=\text { power consumed while outputs are active } \\
\mathrm{V}_{\mathrm{CC}} \times(1-\mathrm{DC} \\
\left.\mathrm{V}_{\mathrm{EN}}\right) \mathrm{I}_{\mathrm{CCZ}}=\text { power consumed when outputs are disabled }
\end{array}\right)
$$

## Dynamic or Switching Power Dissipation in ABT FIFO Devices (BiCMOS FIFO products)

For most applications, dynamic power accounts for most of the total power consumption of a BiCMOS device. Dynamic power consumption includes the device switching power consumed without the load, as well as the power consumed due to the capacitive load.

$$
\mathrm{P}_{\mathrm{d}}=\mathrm{V}_{\mathrm{CC}} \times \mathrm{I}_{\mathrm{CCf}}+\sum\left[\mathrm{v}_{\mathrm{CC}} \times \mathrm{C}_{\mathrm{L}} \times\left(\mathrm{V}_{\mathrm{OH}}-\mathrm{V}_{\mathrm{OL}}\right) \times \mathrm{f}_{\mathrm{o}}\right]
$$

Where:
$\mathrm{V}_{\mathrm{CC}} \times \mathrm{I}_{\mathrm{CCf}}=$ device switching power without the load
$\Sigma\left[\mathrm{V}_{\mathrm{CC}} \times \mathrm{C}_{\mathrm{L}} \times\left(\mathrm{V}_{\mathrm{OH}}-\mathrm{V}_{\mathrm{OL}}\right) \times \mathrm{f}_{\mathrm{o}}\right]=$ power consumed due to the output switching frequency and the load capacitance
$\mathrm{V}_{\mathrm{CC}} \quad=$ supply voltage
$\mathrm{I}_{\mathrm{CCf}} \quad=$ supply current when FIFO is transferring data, active current (see active $\mathrm{I}_{\mathrm{CC}}$ versus frequency plot in data sheet)
$\mathrm{f}_{\mathrm{o}} \quad=$ output switching frequency
$\mathrm{V}_{\mathrm{OH}} \quad=$ output voltage in high state
$\mathrm{V}_{\mathrm{OL}} \quad=$ output voltage in low state
$\mathrm{C}_{\mathrm{L}} \quad=$ load capacitance

## Example 2

This example shows how to calculate the power dissipation for an SN74ABT3614 bidirectional FIFO used in a system under the following conditions:

- Data input lines (A0-A35) and the control signals (CLKA, CLKB, $\overline{\mathrm{CSA}}, \overline{\mathrm{CSB}}, \mathrm{W} / \overline{\mathrm{R}} A, \mathrm{~W} / \overline{\mathrm{R}} \mathrm{B}, \mathrm{ENA}, \mathrm{ENB}, \mathrm{MBA}$, $\overline{\mathrm{BE}}, \overline{\mathrm{RST}}, \mathrm{SIZ0}, \mathrm{SIZ1}, \mathrm{ODD} / \overline{\mathrm{EVEN}}, \mathrm{SW0}, \mathrm{SW} 1$, PGA, PGB) are driven by a CMOS device.
- Only $2 / 3$ of the inputs (or outputs) are switching at a given time.
- The output of the FIFO is fed to a memory device.
- The port-B clock rate is 33.3 MHz and the port-A clock rate is 40 MHz .
- The SN74ABT3614 is only used $60 \%$ of the time by the system.
- The load capacitance of each output is about 50 pF , and the supply voltage is set to 5 V .
- During the FIFO active period, the bus is enabled $75 \%$ of the time.
- When the bus is enabled, the output is in the high state $80 \%$ of the time.
- The SN74ABT3614 is used equally in both directions.

Figures 5 through 7 and included information are needed to calculate power dissipation:


Figure 5. SN74ABT3614 Active ICc With CLKA and CLKB Switching, Simultaneous Read/Write and CLKB as Data Output


Figure 6. SN74ABT3614 Idle Icc With CLKA Switching, Other Inputs at 0 or $\mathrm{V}_{\mathrm{CC}}-0.2 \mathrm{~V}$ and Outputs Disconnected


Figure 7. SN74ABT3614 Idle Icc With CLKB Switching, Other Inputs at 0 or $\mathrm{V}_{\mathrm{CC}}-\mathbf{0 . 2} \mathrm{V}$ and Outputs Disconnected

The following parametric values are needed to calculate power dissipation:

```
\(\mathrm{V}_{\mathrm{CC}} \quad=\) supply voltage \(=5 \mathrm{~V}\)
\(\mathrm{V}_{\mathrm{OH}}=\mathrm{V}_{\mathrm{CC}}-1.3 \mathrm{~V}\)
\(\mathrm{V}_{\mathrm{OL}}=0.3 \mathrm{~V}\)
\(\mathrm{DC}_{\mathrm{d}}=\%\) time FIFO is switching \(=0.6\)
\(1-\mathrm{DC}_{\mathrm{d}}=\%\) time FIFO is not switching \(=0.4\)
\(\mathrm{f}_{\text {clockA }}=\) clock switching frequency of port \(\mathrm{A}=40 \mathrm{MHz}\)
\(\mathrm{f}_{\text {clockB }}=\) clock switching frequency of port \(\mathrm{B}=33.3 \mathrm{MHz}\)
\(\mathrm{pF}(\mathrm{A})=\) clock-A switching power factor, the slope of \(\mathrm{I}_{\mathrm{CC}}\) versus \(\mathrm{f}_{\text {clock }}\) curve (see data sheet) \(=0.25\)
\(\mathrm{pF}(\mathrm{B})=\) clock-B switching power factor, the slope of \(\mathrm{I}_{\mathrm{CC}}\) versus \(\mathrm{f}_{\text {clock }}\) curve (see data sheet) \(=0.28\)
\(\mathrm{DC}_{\mathrm{EN}}=\%\) duty cycle enabled \(=0.75\)
\(\mathrm{I}_{\mathrm{CCI}}=\left[\mathrm{f}_{\text {clockA }} \times \mathrm{pF}(\mathrm{A}]+\left[\mathrm{f}_{\text {clockB }} \times \mathrm{pF}(\mathrm{B})=(40 \times 0.25)+(33.3 \times 0.28)=19.32 \mathrm{~mA}\right.\right.\)
\(\mathrm{I}_{\mathrm{CCf}}=\) idle supply current when FIFO is transferring data \(=136.26 \mathrm{~mA}\)
\(\mathrm{I}_{\mathrm{CCH}}=\) active supply current when outputs are in high state (see data sheet) \(=30 \mathrm{~mA}\)
\(\mathrm{I}_{\mathrm{CCL}}=\) power supply current when outputs are in low state (see data sheet) \(=130 \mathrm{~mA}\)
\(\mathrm{I}_{\mathrm{CCZ}}=\) power supply current when outputs are in high-impedance state (see data sheet) \(=30 \mathrm{~mA}\)
\(\mathrm{N}_{\mathrm{L}} / \mathrm{N}_{\mathrm{T}}=\) ratio of number of outputs in low state to total number of outputs \(=0.2\)
\(\mathrm{N}_{\mathrm{H}} / \mathrm{N}_{\mathrm{T}}=\) ratio of number of outputs in high state to total number of outputs \(=0.8\)
\(\mathrm{C}_{\mathrm{L}} \quad=\) load capacitance \(=50 \mathrm{pF}\)
\(\mathrm{f}_{\mathrm{O}} \quad=1 / 2\) (since maximum data rate is \(1 / 2\) clock frequency \() \times 2 / 3\)
(since \(2 / 3\) of the outputs are switching at a given time) \(\times 33.3 \mathrm{MHz}\)
(slowest of the the two clock frequencies, \(\mathrm{f}_{\text {clockA }}\) or \(\left.\mathrm{f}_{\text {clockB }}\right)=11.1 \mathrm{MHz}\)
```


## Solution

$$
\begin{aligned}
\mathrm{P}_{\mathrm{T}} & =\mathrm{P}_{\mathrm{T}}(\text { from } \mathrm{A} \text { to } \mathrm{B})+\mathrm{P}_{\mathrm{T}}(\text { from } \mathrm{B} \text { to } \mathrm{A}) \\
& =\mathrm{P}_{\mathrm{q}}(\text { from } \mathrm{A} \text { to } \mathrm{B})+\mathrm{P}_{\mathrm{q}}(\text { from } \mathrm{B} \text { to } \mathrm{A})+\mathrm{P}_{\mathrm{d}}(\text { from } \mathrm{A} \text { to } \mathrm{B})+\mathrm{P}_{\mathrm{d}}(\text { from } \mathrm{B} \text { to } \mathrm{A}) \\
& =\mathrm{P}_{\mathrm{q}}+\mathrm{P}_{\mathrm{d}}
\end{aligned}
$$

Where $\mathrm{P}_{\mathrm{q}}$ and $\mathrm{P}_{\mathrm{d}}$ include power from A to B and from B to A directions.

## Quiescent Power

$$
\begin{aligned}
\mathrm{P}_{\mathrm{q}} & =\mathrm{V}_{\mathrm{CC}} \times\left[\mathrm{DC}_{\mathrm{EN}} \times\left(\mathrm{N}_{\mathrm{H}} \times \mathrm{I}_{\mathrm{CCH}} / \mathrm{N}_{\mathrm{T}}+\mathrm{N}_{\mathrm{L}} \times \mathrm{I}_{\mathrm{CCL}} / \mathrm{N}_{\mathrm{T}}\right)+\left(1-\mathrm{DC}_{\mathrm{EN}}\right) \mathrm{I}_{\mathrm{CCZ}}+\mathrm{I}_{\mathrm{CCI}}\right] \\
& =5 \mathrm{~V} \times[0.75 \times(0.8 \times 30 \mathrm{~mA}+0.2 \times 130 \mathrm{~mA})+(1-0.75) 30 \mathrm{~mA}+19.32 \mathrm{~mA}] \\
& =5 \mathrm{~V} \times[37.5 \mathrm{~mA}+7.5 \mathrm{~mA}+19.32 \mathrm{~mA}]=321.6 \mathrm{~mW}
\end{aligned}
$$

## Dynamic Power

$$
\begin{aligned}
\mathrm{P}_{\mathrm{d}} & =\mathrm{V}_{\mathrm{CC}} \times \mathrm{I}_{\mathrm{CCf}}+\sum\left[\mathrm{V}_{\mathrm{CC}} \times \mathrm{C}_{\mathrm{L}} \times\left(\mathrm{V}_{\mathrm{OH}}-\mathrm{V}_{\mathrm{OL}}\right) \times \mathrm{f}_{\mathrm{o}}\right] \\
& =5 \mathrm{~V} \times 136.25 \mathrm{~mA}+\sum[5 \times 50 \mathrm{pF} \times(5 \mathrm{~V}-1.3-0.3 \mathrm{~V}) \times(11.1 \mathrm{MHz})] \\
& =681.25 \mathrm{~mW}+(36 \times 9.44 \mathrm{~mW})=681.25 \mathrm{~mW}+339.66 \mathrm{~mW}=1020.91 \mathrm{~mW}
\end{aligned}
$$

## Total Power

$$
\begin{aligned}
\mathrm{P}_{\mathrm{T}} & =\mathrm{P}_{\mathrm{d}}\left(\mathrm{DC}_{\mathrm{d}}\right)+\mathrm{P}_{\mathrm{q}}\left(1-\mathrm{DC}_{\mathrm{d}}\right) \\
& =1020.91 \mathrm{~mW} \times 0.6+321.6 \times 0.4=741.19 \mathrm{~mW}
\end{aligned}
$$

Therefore, total power is approximately:

$$
\mathrm{P}_{\mathrm{T}}=741.19 \mathrm{~mW}
$$

The SN74ABT3614 is available in 120-pin TQFP and 132-pin PQFP packages (refer to Appendix B for the maximum power curve calculated for reliability purposes). At maximum ambient temperature $\left(70^{\circ} \mathrm{C}\right)$ and no air flow for the 132-pin PQFP and 120-pin TQFP packages, the maximum power that the package can dissipate to free air is 1610 mW and 1606 mW ,
respectively. The $\mathrm{SN} 74 \mathrm{ABT3614}$, in this example, meets the reliability requirement since 741.19 mW is much less than 1610 mW or 1606 mW .

## Summary

Power-dissipation calculations are essential to meet the design requirements related to the chip temperature and the system power. In this application report, a simple method of calculating power is provided to assist the design engineer with power-dissipation calculations for TI CMOS and BiCMOS FIFO products. Total power includes quiescent power and dynamic power. For most applications using CMOS and BiCMOS FIFOs, dynamic power accounts for most of the power requirement. Examples of power-dissipation calculations are provided to show the practical use of this application report. In each example, the reliability of the chip was tested against the absolute maximum power dissipation in free air. For example, the total calculated power consumption for the SN74ACT3632 and SN74ABT3614 examples resulted in 484 mW and 741 mW , respectively. These values are much less than the maximum power dissipation of the 120-pin TQFP $(1606 \mathrm{~mW})$ or 132-pin PQFP $(1610 \mathrm{~mW})$ packages in still air. $I_{C C}$ versus frequency curves are provided in Appendix A. These graphs assist the design engineer in the search for the FIFO device with the minimum power consumption. After total power is calculated for a system, the design engineer can ensure that this value does not exceed the maximum power capability of the package type. The table of maximum power versus ambient temperature for different package options are included in Appendix B.

## Acknowledgements

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## Appendix A <br> Graphs of ICC Versus Frequency

The following information is provided to assist the designer with the power-consumption calculations. Graphs of $\mathrm{I}_{\mathrm{CC}}$ versus frequency are shown for the SN74ACT7803, SN74ACT7811, SN74ACT3641, SN74ACT7807, and SN74ABT7819. While the FIFOs were idle, data was taken on five units on an automatic test machine (HP 82000). Five readings were taken for each frequency and the average was used to plot the graphs. The tests were done by setting $\mathrm{V}_{\mathrm{IL}}$ and $\mathrm{V}_{\mathrm{IH}}$ as shown below:

$$
\begin{aligned}
\mathrm{V}_{\mathrm{IL}} & =0 \mathrm{~V} \\
\mathrm{~V}_{\mathrm{IH}} & =\mathrm{V}_{\mathrm{CC}}-0.2 \mathrm{~V}
\end{aligned}
$$

For each of the FIFOs, two graphs are provided for idle $\mathrm{I}_{\mathrm{CC}}$. One graph shows the $\mathrm{I}_{\mathrm{CC}}$ versus frequency when WRTCLK is running, whereas the other graph shows the $\mathrm{I}_{\mathrm{CC}}$ versus frequency when RDCLK is running. The slope of the 5-V supply voltage curve is calculated for both graphs and the largest of the two slopes is used as the power factor for power calculations.

The slopes of the $\mathrm{I}_{\mathrm{CCI}}$ versus frequency graphs in the tests performed were 0.09 (SN74ACT7807), 0.12 (SN74ACT7803, SN74ACT7811), 0.2 (SN74ACT3641), and 0.28 (SN74ABT3614); therefore, if the slope of the $\mathrm{I}_{\mathrm{CCI}}$ versus frequency plot is not readily available, it is appropriate to estimate the slope as 0.2 .


Figure A-1. SN74ACT7811 Idle ICc With RDCLK or WRTCLK Switching


Figure A-2. SN74ACT7803 Idle Icc With RDCLK or WRTCLK Switching


Figure A-3. SN74ACT3641 Idle ICC With CLKA Switching


Figure A-4. SN74ACT3641 Idle ICC With CLKB Switching


Figure A-5. SN74ACT3641 Active Icc With CLKA and CLKB Switching, Simultaneous Read/Write and CLKB as Data Output


Figure A-6. SN74ACT7807 Idle Icc With WRTCLK Switching, Other Inputs at 0 or $\mathrm{V}_{\mathrm{CC}}-0.2 \mathrm{~V}$ and Outputs Disconnected


Figure A-7. SN74ACT7807 Idle ICc With RDCLK Switching, Other Inputs at 0 or $\mathrm{V}_{\mathrm{CC}}-0.2 \mathrm{~V}$ and Outputs Disconnected


Figure A-8. SN74ACT7807 Active Icc With Simultaneous Read/Write


Figure A-9. SN74ABT7819 Active Icc Versus Frequency

## Appendix B Maximum Power Dissipation for Different Package Types

For reliability purposes, maximum power is calculated for each package option using the following equation: Chip temperature $=$ Power $\times \Theta_{\mathrm{JA}}+\mathrm{T}_{\mathrm{A}}$

Where:

| Chip temperature | $=150^{\circ} \mathrm{C}$ (absolute maximum chip temperature) |
| :--- | :--- |
| $\Theta_{\mathrm{JA}}$ | $=$ thermal characteristics of a package (known) |
| $\mathrm{T}_{\mathrm{A}}$ | $=$ ambient temperature (known) |

Table 1 lists maximum power dissipation by package type for ambient temperature from $25^{\circ} \mathrm{C}$ to $90^{\circ} \mathrm{C}$.
Table B-1. Maximum Power Dissipation (mW) for Packaged FIFOs

| AMBIENT <br> TEMPERATURE <br> $\left({ }^{\circ} \mathbf{C}\right)$ | $\mathbf{2 5}$ | $\mathbf{3 0}$ | $\mathbf{3 5}$ | $\mathbf{4 0}$ | $\mathbf{4 5}$ | $\mathbf{5 0}$ | $\mathbf{5 5}$ | $\mathbf{6 0}$ | $\mathbf{6 5}$ | $\mathbf{7 0}$ | $\mathbf{7 5}$ | $\mathbf{8 0}$ | $\mathbf{8 5}$ | $\mathbf{9 0}$ |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PQ132 PQFP | 2,515 | 2,414 | 2,314 | 2,213 | 2,113 | 2,012 | 1,911 | 1,811 | 1,710 | 1,509 | 1,509 | 1,408 | 1,308 | 1,207 |
| PCB120 TQFP | 2,510 | 2,410 | 2,309 | 2,209 | 2,108 | 2,008 | 1,908 | 1,807 | 1,707 | 1,505 | 1,505 | 1,406 | 1,305 | 1,205 |
| PN80 TQFP | 1,424 | 1,367 | 1,310 | 1,253 | 1,196 | 1,139 | 1,082 | 1,025 | 968 | 854 | 854 | 797 | 740 | 683 |
| PM64 TQFP | 1,351 | 1,297 | 1,243 | 1,189 | 1,135 | 1,081 | 1,027 | 973 | 919 | 811 | 811 | 757 | 703 | 649 |
| PH80 PQFP | 1,490 | 1,430 | 1,371 | 1,311 | 1,251 | 1,192 | 1,132 | 1,073 | 1,013 | 894 | 894 | 834 | 775 | 715 |
| DL56 SSOP | 1,330 | 1,277 | 1,223 | 1,170 | 1,117 | 1,064 | 1,011 | 957 | 904 | 798 | 798 | 745 | 691 | 638 |
| DW28 SOIC | 1,528 | 1,467 | 1,406 | 1,345 | 1,284 | 1,222 | 1,161 | 1,100 | 1,039 | 917 | 917 | 856 | 795 | 733 |
| DW24 SOIC | 1,420 | 1,364 | 1,307 | 1,250 | 1,193 | 1,136 | 1,080 | 1,023 | 966 | 852 | 852 | 795 | 739 | 682 |

