# Design Guide: TIDA-010030 Accurate gauging and 50-µA standby current, 13S, 48-V Li-Ion battery pack reference design

# Texas Instruments

### Description

This reference design is a low standby and ship mode current consumption and high SOC gauging accuracy 13S, 48-V Li-ion battery pack design. It monitors each cell voltage, pack current and temperature with high accuracy and protects the Li-ion battery pack against overvoltage, undervoltage, over temperature and over current situations. The SOC gauging, based on bq34z100-g1, takes advantage of an impedance tracking algorithm and achieves as high as 2% accuracy at room temperature. Through a welldesigned auxiliary power supply strategy and high efficiency, low quiescent current DC-DC converter LM5164, this design achieves 50-µA stand-by and 5µA ship mode consumption, saving more energy and allowing longer shipping time and idle time. This reference design was implemented for a 2-layer PCB. Also, this design supports a well running firmware, which helps to decrease product research time.

#### Resources

TIDA-010030	Design Folder
BQ34Z100-G1, BQ76940, BQ76200, BQ7718	Product Folder
MSP430FR2155	Product Folder
LM5164	Product Folder
SN6501	Product Folder
THVD1500	Product Folder
ISO1042	Product Folder

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# Features

- 2% battery pack SOC accuracy at room temperature.
- 50-µA current consumption when in standby mode
- 5-µA current consumption when in shipping mode
- Robust and programmable protection, including: cell over voltage, cell under voltage, overcurrent discharge, short circuit, overtemperature and undertemperature
- Support 100-mA cell balancing
- High-side charging and discharging MOSFETs and support pre-discharge function

#### Applications

- e-bike battery pack
- e-scooter battery pack
- Mowing robot battery pack
- UPS Li-Ion battery pack
- Other over 10S battery pack applications



PACK 

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#### System Description



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#### 1 System Description

The e-bike market is growing rapidly. Because of the weight limited and longer endurance request, the battery cell chemistries of the battery pack is shifting to Li-ion. Li-polymer, or Li-iron phosphate types. These chemistries are good in both volumetric and gravimetric energy density. While these chemistries provide high energy density and thereby lower volume and weight as an advantage, they are associated with safety concerns and need more complicated monitoring and accuracy capacity gauging. Those concerns are undervoltage (UV) and overvoltage (OV), over temperature (OT), and overcurrent (OC), over discharge and charge current, all which contribute to the accelerating cell degradation and may lead to thermal runaway and explosion. Therefore the pack current, temperature and each cell voltages should be timely monitored in case of some unusual situations. And pack must be protected against all these situations. The capacity gauging is able to calculate how much capacity is still stored in the pack and predict the run-time and charge time, avoiding suddenly shutdown when ridding the ebike. A good accuracy gauging design can also extend run-time because it allows users using all available battery capacity without damaging the battery cells. Another important feature is the current consumption when the e-bike is in ship mode or idle.

This design is mainly focus on e-bike battery pack applications and also suitable for other high cell applications, such as e-scooter battery pack, mowing robot battery pack, etc. It contains both primary and secondary protection to ensure safety use of the battery pack. The primary protection will protect the battery pack against all unusual situations, including: over voltage, under voltage, over discharge current, short circuit, over temperature and under temperature. The secondary protection which uses hardware over voltage protection is an add on board with is easily remove and added based on actual demands. This design utilizes Impedance Track™ gauging algorithm which helps to achieve very high SOC gauging accuracy even for old battery cells or at low or high temperature. Meanwhile, this design carefully designs the auxiliary power and firmware, which achieves quite low ship mode and standby mode current consumption (5–µA for ship mode and 50-µA for standby mode). Also, this design supports a well running firmware based on a low power consumption MSP430<sup>™</sup> MCU MSP430FR2155. This will definitely give a lot help for customers' system design.



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# 1.1 Key System Specifications

The most important there features are  $50-\mu$ A standby current consumption, smooth SOC gauging at low temperature and  $5-\mu$ A ship mode current consumption. Table 1 lists other features.

PARAMETER	Conditions		Spec Range		Unit
		Minimum	Typical	Maximum	
Cell architecture	Li-ion Battery		13S		
Stack Voltage	Li-ion Battery	36	48	55	V
Cell Balancing Current	Li-ion Battery			100	mA
SOC accuracy of fuel gauging	25°C		2		%
Charge Current			10		А
Discharge Current			20	30	А
Pre-discharge Current			0.2		А
Primary OV protection	Threshold		4200		mV
	Delay		2		S
Secondary OV protection	Threshold		4300		mV
	Delay		4		S
UV protection			2900		mV
OCD protection			36		А
SCD protection			56		А
Over Temperature Protection	Charge		55		°C
	Discharge		60		°C
Under Temperature Protection	Charge		0		°C
	Discharge		-20		°C
Communication speed	RS-485		9600	19200	kbps
Standby mode current consumption			50		μΑ
Shipping mode current consumption			5		μA

### **Table 1. Key System Specifications**



System Overview

#### 2 System Overview

#### 2.1 Block Diagram

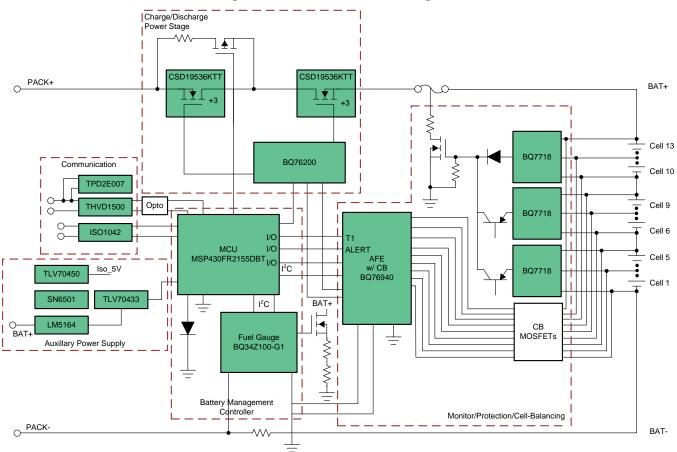


Figure 1. TIDA-010030 Block Diagram

Figure 1 shows the system diagram containing several sub systems: monitor/protection/cell-balancing, battery management controller, auxiliary power supply, charge/discharge power stage, and communication. The battery cell monitoring and balancing uses TI 9-15S AFE bq76940 to monitoring each cell voltage and pack current and temperature, it also integrates over voltage and under voltage hardware protection, cell balancing and charge and discharge FET drivers. The battery gauging utilizes Impedance Track<sup>™</sup> gauging algorithm based device bg34z100-g1 which can handle large cell packs. There is a lower power MSP430<sup>™</sup> MCU MSP430FR2155 which will configure AFE, read data from AFE and gauge and upload all the requested information to system side. The design firmware is open framed which will give some help for customers' system design. The secondary protection is a add-on function, which is easily removed and added based customers' request.

#### 2.2 Highlighted Products

#### 2.2.1 **BQ76940**

The BQ76940 of robust AFE device serves as part of a complete pack monitoring and protection solution for next-generation high-power systems, such as light electric vehicles, power tools, and uninterruptible power supplies. The BQ76940 is designed with low power in mind: Sub-blocks within the IC may be enabled or disabled to control the overall chip current consumption, and a SHIP mode provides a simple way to put the pack into an ultra-low power state. The BQ76940 supports 9 cells to 15 cells providing measurement of individual cell voltages. This AFE can measure a variety of battery chemistries, including Li-ion, Li-iron phosphate, and more. A coulomb counter is provided for current measurement. Three thermistors are provided for temperature measurement. Hardware protection features are configured by



registers set by the system controller and automatically switch off charge and discharge. Through I<sup>2</sup>C, a host controller can use the BQ76940 to implement many battery pack management functions such as monitoring (cell voltages, pack current, pack temperatures), protection (controlling charge or discharge FETs), and balancing. Integrated ADCs enable a purely digital readout of critical system parameters with calibration handled in TI's manufacturing process.

### 2.2.2 BQ34Z100-G1

The bq34z100-G1 device is an Impedance Track<sup>™</sup> fuel gauge for Li-Ion, PbA, NiMH, and NiCd batteries, and works independently of battery series-cell configurations. Batteries from 3 V to 65 V can be easily supported through an external voltage translation circuit that is controlled automatically to reduce system power consumption. The bq34z100-G1 device provides several interface options, including an I<sup>2</sup>C slave, an HDQ slave, one or four direct LEDs, and an ALERT output pin. Additionally, the bq34z100-G1 provides support for an external port expander for more than four LEDs.

### 2.2.3 BQ76200

The bq76200 device is a low-power, high-side, N-Channel system. A high-side protection avoids ground disconnection in the system and also allows continuous communication between the battery pack and host system. The device has additional P-Channel FET control to allow low-current pre-charge to a deeply depleted battery or pre-discharge to a capacitive load, and a PACK+ voltage monitor control for the host to sense the PACK+ voltage. The independent enable inputs allow CHG and DSG FETs to be turned on and off separately, offering great implementation flexibility in battery systems. The bq76200 device can be used with a companion Analog Front End device such as the bq76920/30/40 family, a 3-series to 15-series Cell Analog Front End Monitoring, and a host microcontroller or dedicated state-of-charge (SOC) tracking gas gauge device.

### 2.2.4 MSP430FR2155

MSP430FR215x micro-controller (MCUs) is part of the MSP430<sup>™</sup> MCU value line portfolio of ultra-lowpower low-cost devices for sensing and measurement applications. The device includes a 12-channels 12bit ADC and two comparators. The MSP430FR215x supports an extended temperature range from -40° up to 105°C, so higher temperature industrial applications can benefit from the FRAM data-logging capabilities of the device. The extended temperature range allows developers to meet requirements of applications such as smoke detectors, sensor transmitters, and circuit breakers. The MSP430FR215x features a powerful 16-bit RISC CPU, 16-bit registers, and a constant generator that contribute to maximum code efficiency. The digitally controlled oscillator (DCO) allows the device to wake up from lowpower modes to active mode typically in less than 10 µs. The MSP430 ultra-low-power (ULP) FRAM microcontroller platform combines uniquely embedded FRAM and a holistic ultra-low-power system architecture, allowing system designers to increase performance while lowering energy consumption. FRAM technology combines the low-energy fast writes, flexibility, and endurance of RAM with the nonvolatile behavior of flash. MSP430FR215x is supported by an extensive hardware and software ecosystem with reference designs and code examples to get your design started quickly.

#### 2.2.5 LM5164

The LM5164 synchronous buck converter is designed to regulate over a wide input voltage range, minimizing the need for external surge suppression components. A minimum controllable on-time of 50 ns facilitates large step-down conversion ratios, enabling the direct step-down from a 48 V nominal input to low-voltage rails for reduced system complexity and solution cost. The LM5164 operates during input voltage dips as low as 6 V, at nearly 100% duty cycle if needed, making it an excellent choice for wide input supply range industrial and high cell count battery pack applications. With integrated high-side and low-side power MOSFETs, the LM5164 delivers up to 1 A of output current. A constant on-time (COT) control architecture provides nearly constant switching frequency with excellent load and line transient response. Additional features of the LM5164 include ultra-low IQ and diode emulation mode operation for high light-load efficiency, innovative peak and valley overcurrent protection, integrated VCC bias supply and bootstrap diode, precision enable and input UVLO, and thermal shutdown protection with automatic recovery. An open-drain PGOOD indicator provides sequencing, fault reporting, and output voltage monitoring. The LM5164 is available in a thermally-enhanced, 8- pin SO PowerPAD<sup>™</sup> package. Its 1.27mm pin pitch provides adequate spacing for high-voltage applications.

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#### **THVD1500** 2.2.6

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THVD1500 is a robust half-duplex RS-485 transceiver for industrial applications. The bus pins are immune to high levels of IEC Contact Discharge ESD events eliminating need of additional system level protection components. The device operates from a single 5-V supply. The wide common-mode voltage range and low input leakage on bus pins make THVD1500 suitable for multi-point applications over long cable runs. THVD1500 is available in industry standard 8-pin SOIC package for drop-in compatibility. The device is characterized from -40°C to 125°C.

#### 2.2.7 ISO1042

The ISO1042 device is a galvanically-isolated controller area network (CAN) transceiver that meets the specifications of the ISO11898-2 (2016) standard. The ISO1042 device offers ±70-V DC bus fault protection and ±30-V common-mode voltage range. The device supports up to 5-Mbps data rate in CAN FD mode allowing much faster transfer of payload compared to classic CAN. This device uses a silicon dioxide (SiO2) insulation barrier with a withstand voltage of 5000 VRMS and a working voltage of 1060 VRMS. Electromagnetic compatibility has been significantly enhanced to enable system-level ESD, EFT, surge, and emissions compliance. Used in conjunction with isolated power supplies, the device protects against high voltage, and prevents noise currents from the bus from entering the local ground. The ISO1042 device is available for both basic and reinforced isolation (see Reinforced and Basic Isolation Options). The ISO1042 device supports a wide ambient temperature range of -40°C to +125°C. The device is available in the SOIC-16 (DW) package and a smaller SOIC-8 (DWV) package.

#### **TLV704** 2.2.8

The TLV704 series of low-dropout (LDO) regulators are ultralow guiescent current devices designed for extremely power-sensitive applications. Quiescent current is virtually constant over the complete load current and ambient temperature range. These devices are an ideal power-management attachment to low-power microcontrollers, such as the MSP430. The TLV704 operates over a wide operating input voltage of 2.5 V to 24 V. Thus, the device is an excellent choice for both battery-powered systems as well as industrial applications that undergo large line transients. The TLV704 is available in a 3-mm x 3-mm SOT23-5 package, which is ideal for cost-effective board manufacturing.

#### 2.2.9 SN6501

The SN6501 is a monolithic oscillator/power-driver, specifically designed for small form factor, isolated power supplies in isolated interface applications. The device drives a low-profile, center-tapped transformer primary from a 3.3-V or 5-V DC power supply. The secondary can be wound to provide any isolated voltage based on transformer turns ratio. The SN6501 consists of an oscillator followed by a gate drive circuit that provides the complementary output signals to drive the ground referenced N-channel power switches. The internal logic ensures break-before-make action between the two switches. The SN6501 is available in a small SOT-23 (5) package, and is specified for operation at temperatures from -40°C to 125°C.

#### **TPD2E007** 2.2.10

This device is a transient voltage suppressor (TVS) based electrostatic discharge (ESD) protection device designed to offer system level ESD solutions for wide range of portable and industrial applications. The back-to-back diode array allows AC-coupled or negative-going data transmission (audio interface, LVDS, RS-485, RS-232, and so forth) without compromising signal integrity. This device exceeds the IEC 61000-4-2 (Level 4) ESD protection and is ideal for providing system level ESD protection for the internal ICs when placed near the connector. The TPD2E007 is offered in a 4-bump PicoStar and 3-pin SOT (DGK) packages. The PicoStar package (YFM), with only 0.15 mm (Max) package height, is recommended for ultra space saving application where the package height is a key concern. The PicoStar package can be used in either embedded PCB board applications or in surface mount applications. The industry standard SOT package offers straightforward board layout option in legacy designs.



### 2.3 System Design Theory

#### 2.3.1 Auxiliary Power Supply Design

This auxiliary power circuit consists of a DC/DC BUCK converter, Isolated power supplies and several LDOs. The DC/DC BUCK converter is controlled by LM5164, a constant on-time (COT) control architecture device. The on-time is calculated with Equation 1. In CCM, the operating frequency is programmed by the  $R_{RON}$  resister and can be calculated using Equation 2. Therefor the  $R_{RON}$  can be determined using Equation 3.

$$t_{on}(\mu s) = \frac{R_{RON}(k\Omega)}{V_{IN}(V) \times 2.5}$$

$$F_{sw}(kHz) = \frac{V_{OUT}(V) \times 2500}{R_{ROW}(k\Omega)}$$
(1)

$$R_{RON}(k\Omega) = \frac{V_{OUT}(V) \times 2500}{F_{SW}(kHz)}$$
(2)  
(2)  
(3)

To decrease the standby power consumption, the light load efficiency should be carefully designed. When at quite light load, the switching loss is the mainly loss and to decrease the switching loss, the switching frequency is designed at a low value. In this design, the CCM frequency is designed as 50 kHz. Therefore the  $R_{RON}$  is 250 k $\Omega$ .

The BUCK inductor is calculated using Equation 4. where  $\Delta I$  is the current decrease value when the main FET is off. In this design, the critical mode load is designed at 600 mA, therefore L is 68  $\mu$ H.

$$L(H) = \frac{V_{OUT}(V) \times (1 - ton(ms) \times F_{sw}(kHz))}{F_{sw}(kHz)}$$
(4)

#### 2.3.2 Pack Current Sensing Design

The pack current sensing is using a high accuracy resister. To ensure accurate current measurement, the input voltage generated across the current sense resistor should not exceed  $\pm 125$  mV. For applications with a very high dynamic range, it is allowable to extend this range to absolute maximum of  $\pm 300$  mV for overload conditions where a protector device will be taking independent protective action. In such an overloaded state, current reporting and gauging accuracy will not function correctly. The normal current is designed as 30 A and the OCD threshold is design as 60 A. Therefore the current sensing resister is designed less than 2.08 m $\Omega$ . A 2 m $\Omega$  with 50-ppm temperature coefficient and power rating of 5 W is used in this design.

#### 2.3.3 Wake-up Circuit Design

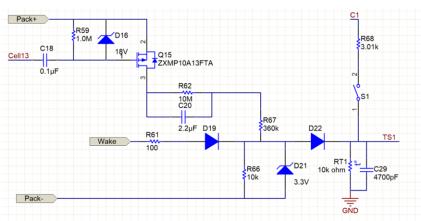
When the battery pack is first time powered on or with quite low battery voltage, system will enter ship mode and the whole system except for BQ76940 is powered off to save power. In this mode, the BQ76940 device is monitoring the wake up command. This design gives two wake up method, the first method is through a button which will generate a wake up signal on TS1 pin. The second method is detecting if charger attached with circuit shown as Figure 2. When a right charge is attached, Q15 will be turned on and pull TS1 high, and because of C20 and C18, Q15 will be turned off, generating a wake up signal on TS1 pin. *Wake* signal is from the MCU and can wake up the BQ76940 from shutdown mode, also.

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Figure 2. Wake Up Circuit





# 3 Hardware, Software, Testing Requirements, and Test Results

# 3.1 Required Hardware and Software

# 3.1.1 Hardware

The board is configured during assembly with a selection of components to be installed and there are no configuration jumpers or options prior to connection. Figure 3 shows the top side of the board.



Figure 3. Board Top Side

Table 2 describes the connections for the board.

#### **Table 2. Board Connections**

CONNECTOR AND PIN ASSIGNMENTS	FUNCTION OR SCHEMATIC NET	NOTES
J1	BAT-	Cell stack negative; this provides a reference for the electronics and the high current path from the cells
J2	BAT+	Cell stack positive; this provides power for the high-side driver and the high current path from the cells
J3-1	RS-485-B	RS485 BUS port, B
J3-2	RS-485-A	RS485 BUS port, A
J4-1	CAN-H	High-level CAN BUS
J4-2	CAN-L	Low-level CAN BUS

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CONNECTOR AND PIN ASSIGNMENTS	FUNCTION OR SCHEMATIC NET	NOTES
J5-1	Commu_EN	Communication control from system side
J5-2	SGND	Isolated ground
J6	PACK+	Charger positive or load positive
J7	PACK-	Charger negative or load negative
P1-1	SDA	I2C communication SDA from bq34z100- g1 and bq76940
P1-2	SDA_MCU	I2C communication SDA from MCU
P1-3	SCL	I2C communication SCL from bq34z100- g1 and bq76940
P1-4	SCL_MCU	I2C communication SCL from MCU
P2-1	3V3	SBW program connector 3.3 V
P2-2	SBCLK	SBW program connector SBCLK
P2-3	SBIO	SBW program connector SBIO
P2-4	GND	SBW program connector ground
P3-1	Fuse_G	Secondary protection output signal
P3-2	Fuse_G	Secondary protection output signal
P3-3	GND	Secondary protection output reference
P3-4	GND	Secondary protection output reference
P4, P5	Cn	(n = 0 to 13) Cell monitor, balance, and electronics power connections. P4 is the connector with battery cells. P3 and P5 are the connectors with Secondary protection board
P6-1	PRES	Detect system load attached
P6-2	GND	Ground

#### Table 2 Board Connections (continued)

#### 3.1.2 Software

#### 3.1.2.1 Using the bg769X0 Eval Software

Download the bg769X0 Eval Software from ti.com. It can act as a host controller to read voltage, current and temperature data from bq76940. See the bq76930 and bq76940 Evaluation Module user's guide to learn how to use this tool.

#### 3.1.2.2 Using bqStudio

The bqStudio software can be started from the desktop icon or the Start  $\rightarrow$  Texas Instruments  $\rightarrow$  Battery Management Studio menu or equivalent for supported operating systems. When bqStudio is started while connected to a working gauge, the software detects the gauge and open its display similar to Figure 4. The dashboard section on the left side of the display shows the connected interface and part and the voltage and current of the part. The dashboard refreshes periodically but can be turned off if desired. The registers tab is displayed in the middle pane, which shows a snapshot of when the tool started or from the last refresh. The tab can be refreshed by clicking the Refresh button in the upper-right corner of the tab or by selecting the scan button for periodic update. The Start Log button starts a log of the data, which is useful for recording the gauge data over time. Across the top of the bgStudio window and below the File menu are other tools which typically open in the center tab area of the display. On the right side of the display is a Command window which the user can scrolled through to display various commands that can be selected as a button. Below the Commands window is the Log Panel, which shows the command and its result. Status messages display in the bottom border of the bostudio window.



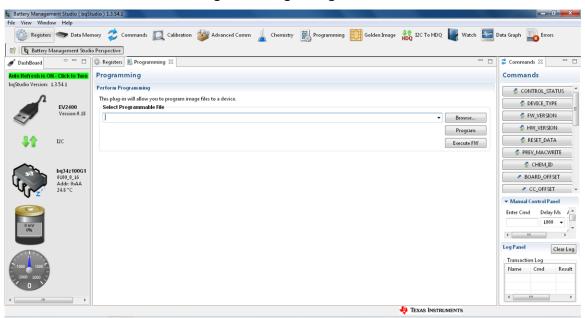
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1000 📩 1000 了		Flags B (high)	0x2400	SOH		FE	FIRSTDOD	RSVD	RSVD	DODEOC	DTRC		SVD	Name	Cmd	R
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Figure 4. bqStudio Register Display

#### 3.1.2.2.1 Programming

Before configure bq34z100-g1, it is recommend to program the device. Select the Programing Update window, the Program screen is shown as Figure 5. This window allows the user to program the device to a new version of firmware.

Figure 5. Programing Screen



- Search for the bq34z100\_G1\_v0\_16\_build\_17.srec file using the Browse button.
- · Press the Program button and wait for the download to complete.
- Press the Execute FW button after the download has completed.
- Select File | Restart to initialize bqStudio to the new firmware.



#### 3.1.2.2.2 Chemistry

The chemistry file contains parameters that the simulations use to model the cell and its operating profile. It is critical to program a Chemistry ID that matches the cell into the device. Some of these parameters can be viewed in the Data Flash section of the Battery Management Studio. Press the Chemistry button to select the Chemistry window, see Figure 6.

Battery Management Studio ( bq	Studio ) 1.3.54.1				_ 0 ×
File View Window Help					
🛞 Registers 🖘 Data Me	emory 🥏 Commands 🔟 C	alibration 🛯 🍟 Advanced Comm 📗	Chemistry	Programming 🛄 Golden Image ដ I2C To HDQ 📗 Watch 🔛	Data Graph 🙀 Errors
📑 🔓 Battery Management Stud	lio Perspective				
🖋 DashBoard 🗢 🗖 🗖	🚳 Registers 🛓 Chemistry 🛙	🔣 Programming			💈 Commands 🛛 🗖 🗖
Auto Refresh is ON - Click to Turn	Chemistry Programmin	g			Commands
bqStudio Version: 1.3.54.1	Program Battery Chemistry				
0	Mart Lisian calls use LiCaO2 ent	hode and graphitized carbon anode which i	e composited by the	default firmware in the Impedance track fuel gauges.	CONTROL_STATUS
EV2400		to be set up for various alternate battery cher		deradic filmware in the impedance track rule gauges.	DEVICE_TYPE
Version:0.18				their cells use a different chemistry than LiCoO2 cathode and graphite anode.	FW_VERSION
~	Note : Right Click on the select	ted chemistry to apply it to individual cells	. The menu appea	rs only if the f/w supports individual cell chemistries.	
	Manufacturer	Model	Chemistry ID	Description	HW_VERSION
<b>1</b>	Panasonic	UF18650RX (2050mAh)	2057	NiCoMn/carbon	🔮 RESET_DATA
<b>V</b>	🛃 Moli	INR 18650A (2500mAh)	2058	NiCoMn/carbon, -30C	
	LGC	INR18650MU1 (3500mAh)	2059	NiCoMn/carbon, -20C	PREV_MACWRITE
	SAFT SAFT	MP176065xc (6400mAh)	2060	NiCoMn/carbon	CHEM_ID
bq34z100G1 0100 0 16	Ranasonic Ranasonic	NCA653864 (2000mAh)	2061	NiCoMn/carbon	BOARD_OFFSET
Addr: 0xAA	Panasonic	NCA793540 (1570mAh)	2062	NiCoMn/carbon	BOARD_OFFSET
24.6 °C	🔜 Sanyo	UR18650FM (2600mAh)	2063	NiCoMn/carbon	CC_OFFSET +
0 2	R LGC	INR18650-MH1 (3200mAh)	2064	NiCoMn/carbon	▼ Manual Control Panel
	Sony	US18650V3 (2250mAh)	2065	NiCoMn/carbon	<ul> <li>Manual Control Panel</li> </ul>
	Sony Sony	US496764E15 (2420mAh)	2066	NiCoMn/carbon	Enter Cmd Delay Ms
0 mV	Panasonic	UR18650NSX (2600mAh)	2067	NiCoMn/carbon	1000 -
0%	🔀 Samsung	INR18650-25R (2500mAh)	2068	NiCoMn/carbon	
	🔀 Shenzhen	WD3288101P (3800mAh)	2069	NiCoMn/carbon,-10C *	* III +
	•			•	Log Panel Clear Log
AN TON					
S.1000 1000					Transaction Log
		Lindate Chemister from Dat	abaco Undato	Chemistry from External File	Name Cmd Result
-2000 2000		Opuate Chemistry from Date	opuate	Chemistry from External File	
0					
<	Chemistry Version : 472				4 III >
				💠 Texas Instruments	

#### Figure 6. Chemistry Screen

- Select the ChemID that matches your cell from the table. ٠
- Press the Update chemistry in the data flash button to update the chemistry in the device. ٠

#### 3.1.2.2.3 Configuration

The bq34z100-g1 gauge is configured primarily using the Data Memory tab. Different sections are displayed by selecting the button for the section on the left side of the pane. The selection is written by clicking the Write To Data Memory button at the bottom of the popup. The popup must be closed with the X button in the top-left corner. All the registers configurations below Configuration of this design is shown as Figure 7, Figure 8, and Figure 9.



### Figure 7. Configuration-1

ata Memory		Filter/Search	Auto Export Export Import Virite_All Rea
ad/Write Data	Memory Contents		
Configuration	Name	Value	Unit
System Data	~ Safety		
	OT Chg	55.0	1degC
Gas Gauging	OT Chg Time	2	Seconds
Ra Tables	OT Chg Recovery	50.0	1degC
Calibration	OT Dsg	60.0	1degC
	OT Dsg Time	2	Seconds
Security	OT Dsg Recovery	55.0	1degC
	~ Charge Inhibit Cfg		
	Chg Inhibit Temp Low	0	1degC
	Chg Inhibit Temp High	45.0	1degC
	Temp Hys	5.0	1degC
	~ Charge		
	Suspend Low Temp	-5.0	1degC
	Suspend High Temp	55.0	1degC
	Pb EFF Efficiency	100	%
	Pb Temp Comp	24.960	%
	Pb Drop Off Percent	96	%
	Pb Reduction Rate	10.000	%
	~ Charge Termination		
	Taper Current	300	mAmp
	Min Taper Capacity	25	mAmpHr
	Cell Taper Voltage	100	mVolt
	Current Taper Window	40	Seconds
	TCA Set %	99	Percent
	TCA Clear %	95	Percent
	FC Set %	100	Percent
	FC Clear %	98	Percent
	DODatEOC Delta T	10.0	1degC
	NiMH Delta Temp	3.0	1degC
	NiMH Delta Temp Time	180	Seconds
	NiMH Hold Off Time	100	Seconds
	NiMH Hold Off Current	240	mAmp
	NiMH Hold Off Temp	25.0	1degC
	NiMH Cell Negative Delta Volt	17	m√olt
	NiMH Cell Negative Delta Time	16	Seconds
	NiMH Cell Neg Delta Qual Volt	4200	mVolt

# Figure 8. Configuration-2

ata Memory		Filter/Searc	Auto Export Export Import Virite_All Re
nd/Write Data	Memory Contents		nuo capore capore impore - mitegai n
Configuration	Name	Value	Unit
	~ Data		
System Data	Manufacture Date	1980-1-1	Day + Mo*32 + (Yr - 1980)*256
Gas Gauging	Serial Number	0001	hex
Ra Tables	Cycle Count	2	Count
Calibration	CC Threshold	5760	mAmpHr
	Max Error Limit	100	%
Security	Design Capacity	6400	MilliAmpHour
	Design Energy	23488	MilliWattHour
	SOH Load I	-1920	MilliAmp
	Cell Charge Voltage T1-T2	4200	mV
	Cell Charge Voltage T2-T3	4200	m∨
	Cell Charge Voltage T3-T4	4200	mV
	Charge Current T1-T2	10	Percent
	Charge Current T2-T3	50	Percent
	Charge Current T3-T4	30	Percent
	JEITA T1	0	degC
	JEITA T2	10	degC
	JEITA T3	45	degC
	JEITA T4	55	degC
	Design Energy Scale	13	Number
	Device Name	bq34z100-G1	
	Manufacturer Name	Texas Inst.	
	Device Chemistry	LION	
	~ Discharge		
	SOC1 Set Threshold	512	mAh
	SOC1 Clear Threshold	640	mAh
	SOCF Set Threshold	192	mAh
	SOCF Clear Threshold	320	mAh
	Cell BL Set Volt Threshold	2850	mVolt
	Cell BL Set Volt Time	2	Seconds
	Cell BL Clear Volt Threshold	2950	mVolt
	Cell BH Set Volt Threshold	4250	mVolt
	Cell BH Volt Time	2	Seconds
	Cell BH Clear Volt Threshold	4200	mVolt
	Cycle Delta	0.05	%
	~ Manufacturer Data		
	Pack Lot Code	0000	hex
	PCB Lot Code	0000	hex
	Firmware Version	0000	hex
	Hardware Revision	0000	hex
	Cell Revision	0000	hex
	DF Config Version	0000	hex
	~ Lifetime Data		



egisters ⇒ Data I ata Memory		Filter/Search	
ata memory		At	ito Export Export Import * Write_All Re-
d/Write Data	Memory Contents		
onfiguration	Name	Value	Unit
ystem Data	Manufacturer Name	Texas Inst.	
	Device Chemistry	LION	
is Gauging	~ Discharge		
Ra Tables	SOC1 Set Threshold	512	mAh
alibration	SOC1 Clear Threshold	640	mAh
	SOCF Set Threshold	192	mAh
Security	SOCF Clear Threshold	320	mAh
	Cell BL Set Volt Threshold	2850	mVolt
	Cell BL Set Volt Time	2	Seconds
	Cell BL Clear Volt Threshold	2950	mVolt
	Cell BH Set Volt Threshold	4250	mVolt
	Cell BH Volt Time	2	Seconds
	Cell BH Clear Volt Threshold	4200	mVolt
	Cycle Delta	0.05	%
	~ Manufacturer Data		
	Pack Lot Code	0000	hex
	PCB Lot Code	0000	hex
	Firmware Version	0000	hex
	Hardware Revision	0000	hex
	Cell Revision	0000	hex
	DF Config Version	0000	hex
	~ Lifetime Data		
	Lifetime Max Temp	30.0	1degC
	Lifetime Min Temp	20.0	1degC
	Lifetime Max Chg Current	0	mAmp
	Lifetime Max Dsg Current	0	mA
	Lifetime Max Pack Voltage	160	20mV
	Lifetime Min Pack Voltage	175	20mV
	<ul> <li>Lifetime Temp Samples</li> </ul>		
	LT Flash Cnt	0	Count
	~ Registers		
	Pack Configuration	09d9	flags
	Pack Configuration B	af	flags
	Pack Configuration C	33	flags
	LED_Comm Configuration	00	flags
	Alert Configuration	c000	flags
	Number of series cell	13	num
	Lifetime Resolution	.,	
	LT Temp Res	1.0	1degC
	LT Cur Res	100	mA
	LT V Res	1	20mV
	LT Update Time	60	Seconds
	<ul> <li>LED Display</li> </ul>		Seconds
	LED Hold Time	4	Num
	<ul> <li>Power</li> </ul>	4	Pulli
	Flash Update OK Cell Volt	2895	mVolt
	Sleep Current	2095	mAmp
	FS Wait	0	Seconds

# Figure 9. Configuration-3

All the registers configurations below Gas Gauging of this design is shown as Figure 10.

### Figure 10. Gas Gauging Screen

ta Memory		Filter/Search	Auto Export Export Import Virite_All Read
d/Write Data	Memory Contents		ato sport sport import i magin head
onfiguration	Name	Value	Unit
ystem Data	~ IT Cfg		
	Load Select	1	Number
as Gauging	Load Mode	0	Number
Ra Tables	Res Current	100	mAmp
alibration	Max Res Factor	50	num
	Min Res Factor	1	num
Security	Ra Filter	500	num
	Min PassedChg NiMH-LA 1st Qmax	50	%
	Maximum Qmax Change	100	%
	Cell Terminate Voltage	3000	mVolt
	Cell Term V Delta	200	mVolt
	ResRelax Time	500	Seconds
	User Rate-mA	0	MilliAmp
	User Rate-Pwr	0	mW/cW
	Reserve Cap-mAh	0	MilliAmpHour
	Reserve Energy	0	mWh/cWh
	Max Scale Back Grid	4	num
	Cell Min DeltaV	0	mVolt
	Ra Max Delta	15	%
	Design Resistance	42	mOhms
	Reference Grid	4	
	Qmax Max Delta %	10	mAmpHour
	Max Res Scale	32000	Num
	Min Res Scale	1	Num
	Fast Scale Start SOC	10	%
	Charge Hys V Shift	40	mVolt
	Smooth Relax Time	1000	s
	~ Current Thresholds		
	Dsg Current Threshold	75	mAmp
	Chg Current Threshold	100	mAmp
	Quit Current	50	mAmp
	Dsg Relax Time	60	Seconds
	Chg Relax Time	60	Seconds
	Cell Max IR Correct	400	mV
	~ State		
	Qmax Cell 0	6400	mAmpHr
	Cycle Count	0	num
	Update Status	00	num
	Cell V at Chg Term	4200	mVolt
	Avg I Last Run	-299	mAmp
	Avg P Last Run	-1131	MilliWattHour
	Cell Delta Voltage	2	mVolt
	T Rise	20	Num
	T Time Constant	1000	Num

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#### 3.1.2.2.4 Calibration

The Calibration of this design is shown as Figure 11. The voltage and current must be calibrated before normal operation.

#### Figure 11. Calibration Screen

Registers ~ Data	Memory =		° 0
Data Memory	/	Filter/Search Au	to Export Export Import Virite_All Read All
Read/Write Data	Memory Contents		
Configuration	Name	Value	Unit
System Data	~ Data		
	CC Gain	1.051	mohm
Gas Gauging	CC Delta	1.049	mohm
Ra Tables	CC Offset	-1372	num
Calibration	Board Offset	-15	num
	Int Temp Offset	0	degC
Security	Ext Temp Offset	0	degC
	Voltage Divider	62862	mVolt
	~ Current		
	Deadband	15	mAmp

#### Voltage Calibration

- Measure the voltage from BAT+ to BAT- and enter this value in the Applied Voltage field and select the Calibrate Voltage box.
- Press the Calibrate Gas Gauge button to calibrate the voltage measurement system.
- Deselect the Calibrate Voltage boxes after voltage calibration has completed.

If this doesn't work, users can also manually input the Voltage Divider value to do the voltage calibration.

#### **Temperature Calibration**

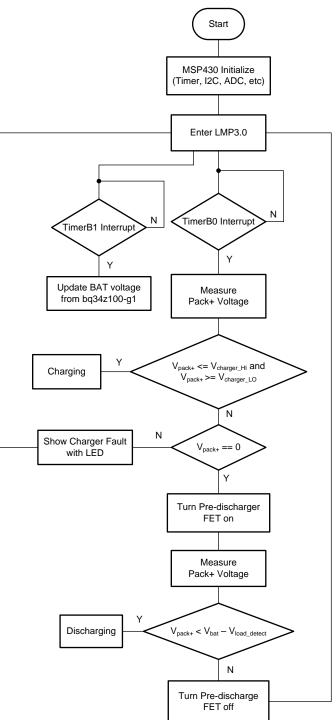
- Enter the room temperature in the Applied Temperature field and select the Calibrate Temperature box and select the thermistor to be calibrated. The temperature value must be entered in degrees Celsius.
- Press the Calibrate Gas Gauge button to calibrate the temperature measurement system.
- Deselect the Calibrate Temperature box after temperature calibration has completed.

#### **Current Calibration**

- Select the Calibrate CC Offset and Calibrate Board Offset boxes and insure that there is no current flow.
- Press the Calibrate Gas Gauge button to calibrate.
- Deselect the Calibrate CC Offset and Calibrate Board Offset boxes after current calibration has completed.
- Connect and measure a 2-A load from PACK- and BAT- to calibrate the current gain.
- Enter –2000 in the Applied Current field and select the Calibrate Current box.
- Press the Calibrate Gas Gauge button to calibrate.
- Deselect the Calibrate Current box after current calibration has completed.

#### 3.1.2.3 MSP430 MCU<sup>™</sup> Firmware

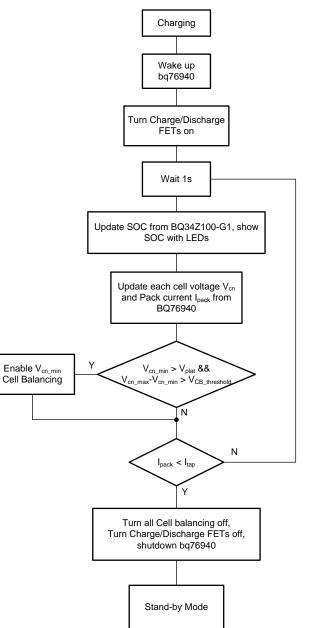
To achieve quite low standby current consumption, the MSP430 MCU<sup>™</sup> firmware has to be carefully designed. This design gives an example of the firmware. The program flow chart is shown in Figure 12, Figure 13, and Figure 14. Figure 15 shows the faults handling.



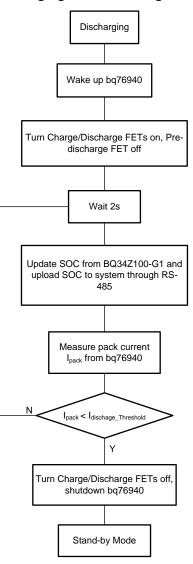
# Figure 12. MSP430 MCU<sup>™</sup> Firmware Program Flow Chart







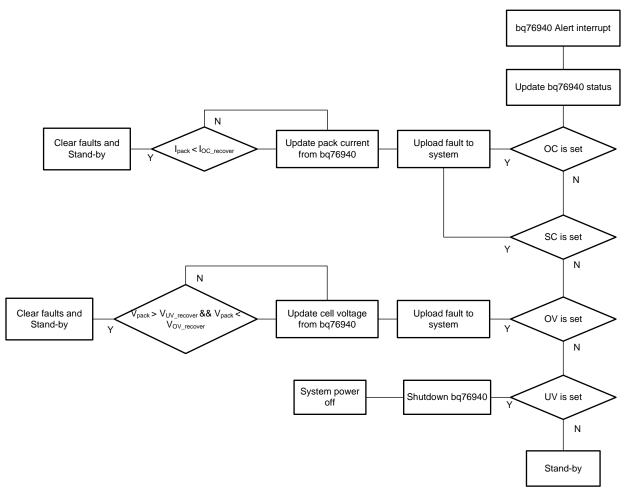
# Figure 13. Charging Firmware Program Flow Chart



#### Figure 14. Discharging Firmware Program Flow Chart



Figure 15. Faults Handling





#### 3.2 Testing and Results

#### 3.2.1 **Test Setup**

The following procedures should be followed before running this design board. The design was constructed with 13S pack configurations and tested using both equipment and battery cells. Board calibration was performed using simulated cells powered by a DC power supply. Learning cycle and normal testing was performed using a 13S2P cell assembly. Table 3 summarizes the equipment used for testing.

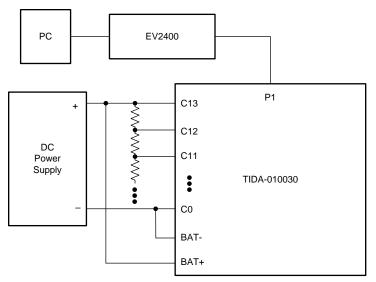
EQUIPMENT	MODEL OR DESCRIPTION
Oscilloscope	Tektronix DPO 3054
DC Power supply	Chroma 62050P-100-100
Electronic load	Chroma 63106
Multimeter	Agilent 34401A
Communication adapter	Texas Instruments EV2300 or EV2400
MSP430 programmer	MSP430 LaunchPad™
Battery cell	INR18650 MH1 3200 mAh

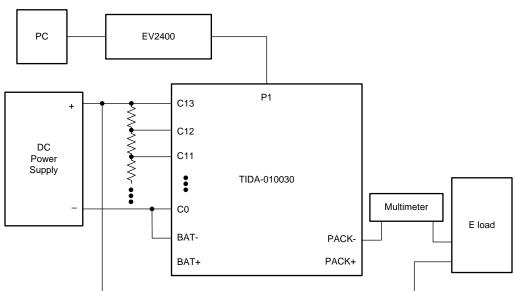
**Table 3. Test Equipment Summary** 

#### 3.2.1.1 **Calibration Setup**

The board voltage and current calibration was performed using DC power supply. Figure 16 shows the voltage calibration setup example, Figure 17 shows the current calibration setup example. After setup the board, press S1 button to wake up the system, then open bqStudio software and follow Section 3.1.2.2.4 to finish voltage and current calibration.





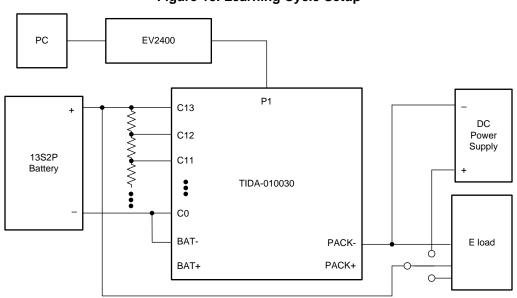


### Figure 17. Current Calibration Setup

### 3.2.1.2 Learning Cycle

After the board calibration and configuration, a learning cycle should be finished with battery cells. The matched Chemistry ID much be programmed to the device before doing learning cycle (See Section 3.1.2.2.2). Figure 18 shows the example setup for learning cycle. Follow below process to finish a learning cycle.

- · Discharge the battery until the voltage reaches the term voltage
- Relax for at least 5 hours
- Use IT\_ENABLE command to set the QEN in Control Status register
- Charge the battery to full with 0.2C
- Relax at least 2 hours until VOK is clear
- Discharge the battery until the voltage reaches the Term Voltage with 0.2C
- Relax at least 5 hours



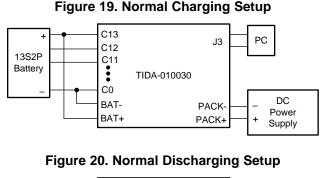
### Figure 18. Learning Cycle Setup

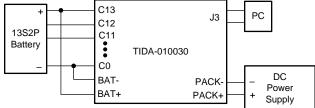


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After a successful learning cycle, the board is ready for testing. Figure 19 shows the normal charging setup. Figure 20 shows the normal discharging setup.





### 3.2.2 Test Results

#### 3.2.2.1 Boot

The 3.3-V MCU is powered by the DC/DC converter, LM5164, and the input of the LM5164 is controlled by two signals: REGOUT from the BQ76940 and SYS from the MCU. Either of them is high will enable the input of the LM5164 and therefore enable the MCU power supply. When the board is powered on for the first time, the board is in ship mode. The whole system, except for the BQ76940, is not powered on. Choose either of the following methods to wake the system up. The first method is through pressing button S1. Figure 21 shows BQ76940 TS1 PIN and REGOUT PIN voltage waveform. The REGOUT PIN is set high after 1-ms delay from when the S1 button is pressed. The second method to wake the board is to attach a charger. When the battery pack has low battery voltage, the board will also enter ship mode. When at this mode, wake the board up by attaching a charger. Figure 22 shows the test result. Both methods will generate a rising edge on TS1 PIN. Then set REGOUT high to enable the MCU power supply. When the MCU is powered on, it will set SYS high. The whole system has stable power supply regardless of if the BQ76940 is in shutdown or normal mode.



14: 24: 09

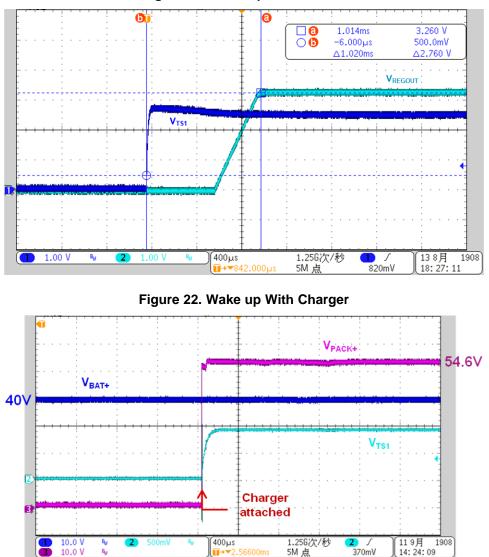


Figure 21. Wake up With Button

#### 3.2.2.2 Voltage & Current Accuracy

Before normal using, the board needs to be calibrated. After calibration, the voltage and current measurement should have good accuracy. Figure 23 shows the BQ76940 cell voltage measurement accuracy. The voltage error is typically below ±10 mV. Figure 24 shows BQ34z100-g1 voltage measurement accuracy. The voltage error is typically below ±20 mV. Figure 25 shows BQ34z100-g1 current measurement accuracy where the error is below ±10 mA. The BQ76940 measures the voltage across the current sensing resister and sends the voltage data out through I<sup>2</sup>C. The firmware should do the calculation and gets the current information. There is no BQ76940 current measurement accuracy test result. The information can be attained from the firmware.

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48000

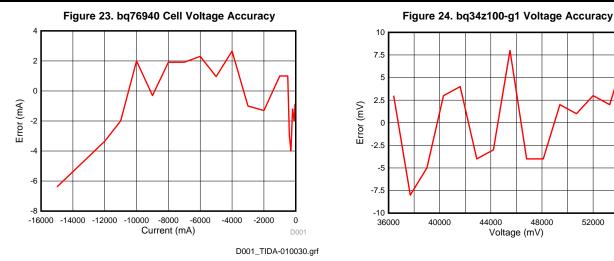
Voltage (mV)

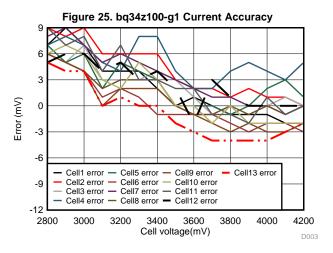
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### Hardware, Software, Testing Requirements, and Test Results







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# 3.2.2.3 DC/DC BUCK

This design is powered by a BUCK converter. To decrease light load consumption, LM5164 will enter DCM and decrease switching frequency. Figure 26 shows the BUCK converter start-up waveform.

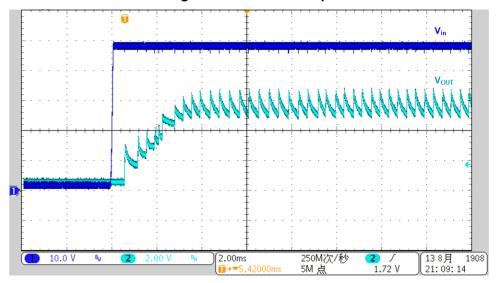
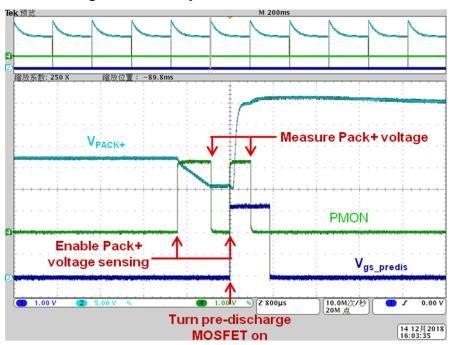


Figure 26. BUCK Start up

#### 3.2.2.4 Standby Mode

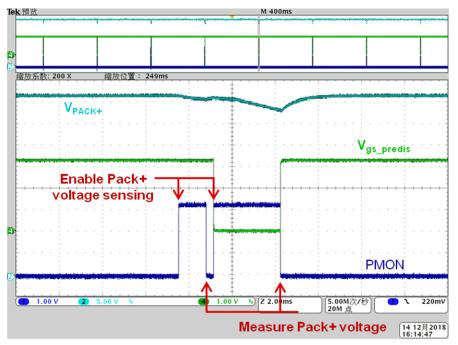
In this design, standby mode refers to two kind of scenarios: when the battery pack is outside of the e-bike and when the battery pack is put inside of the e-bike but not running. Because this design is for using the same charge and discharge port, the charger is only attached in the first scenario. The battery pack in the first scenario should be able to detect if the charger is attached or if the battery pack is put inside of the ebike. The battery pack in the second scenario should be able to detect if load is on or if the battery pack is plugged in outside of the e-bike. Figure 27 shows the standby test waveforms in the first scenario. The dark blue curve is the pre-discharge MOSFET drive voltage and the green curve is PMON of the BQ76200, which controls a switch to measure PACK side voltage. The light blue curve is the PACK side voltage. The first pulse of PMON is to measure PACK voltage with pre-discharge MOSFET off. If the charger is attached, the voltage should be equal to 54.6 V. If not, the pre-discharge will be turned on. After a short time delay, the PACK voltage will be measured again. If the battery pack is put in the e-bike, a large capacitor will be connected between PACK+ and PACK-, therefore, the measured voltage is lower than the battery voltage. Otherwise, the PACK voltage is equal to the battery voltage. The detection sequence happens every 200 ms, which will not cause obvious delay. In the second scenario, the predischarge MOSFET is mostly on. If load is on, there will be a voltage drop across the resister series with the pre-discharge MOSFET, therefore, the measured voltage is lower than the battery voltage. If not, the pre-discharge will turn off. After a some time delay, the PACK voltage is measured. If the battery pack is plugged outside of the e-bike, the output capacitor is much smaller, which will cause a larger voltage drop. The waveforms are shown in Figure 28.





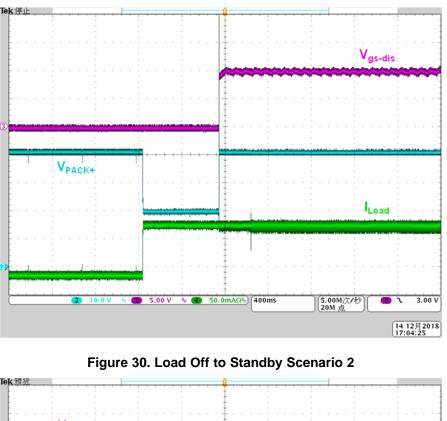
#### Figure 27. Standby Scenario 1 Test Waveforms



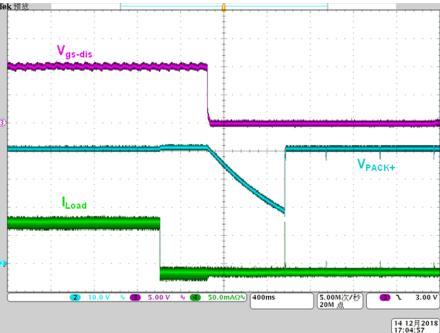


When in standby mode in the second scenario and when the load is turned on, the pre-discharge MOSFET will allow a small current to power the system. The charge and discharge MOSFETs are turned on after some delay. Figure 29 shows the test waveforms. The purple curve is the discharge MOSFET drive voltage, the blue curve is the PACK side voltage, and the green curve is the discharge load. Figure 30 shows the test waveforms when the load is off and when the system enters standby mode in the second scenario.

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When in standby mode in the first scenario and and the charger is attached, Figure 31 shows the resulting test waveforms. The purple curve is the charge MOSFET drive voltage, the light blue curve is the PACK side voltage, the dark blue curve is the battery voltage, and the green curve is the discharge load. When the charger MOSFET is turned on, the PACK side voltage is clamped to the battery voltage. Figure 32 shows the test waveforms when the charger is removed and the system enters standby mode in the first scenario.



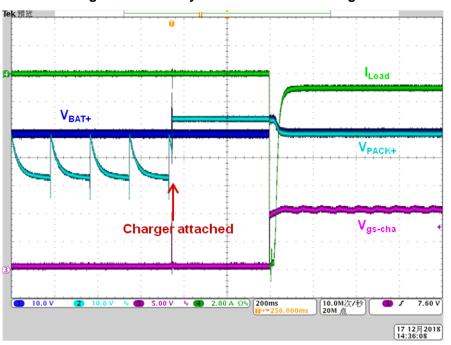
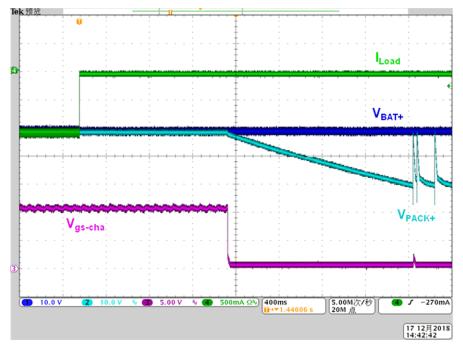


Figure 31. Standby Scenario 1 Detect Charger

Figure 32. Charger Removed to Standby Scenario 1



28 Accurate gauging and 50-μA standby current, 13S, 48-V Li-Ion battery pack TIDUEG7B-December 2018-Revised February 2019 reference design Submit Documentation Feedback Copyright © 2018–2019, Texas Instruments Incorporated



#### 3.2.2.5 Protections

The firmware configures the OCD threshold at 11 mV which turns out about 11-A discharge current with the delay is set at 320 ms. Figure 33 shows the waveform. The green curve is discharge current, the dark blue curve is the battery voltage, the light blue is the PACK side voltage, and the purple curve is the discharge MOSFET drive voltage. When the discharge current is switched from 10.1 A to 13.1 A after a 320-ms delay, the discharge MOSFET is turned off to stop the discharging.

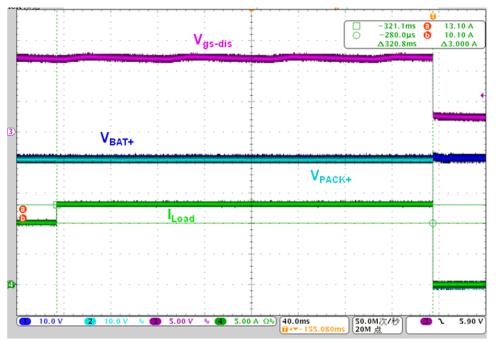
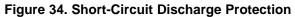
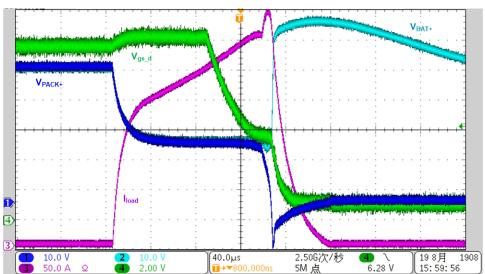


Figure 33. Overcurrent Discharge Protection

To test the design's short circuit protection feature, a switch is used to short PACK+ to PACK-. Figure 34 illustrates the SCD triggered. The purple curve is the discharge load, the dark blue curve is the PACK side voltage, the light blue curve is the battery voltage, and the green curve is the discharge MOSFET drive voltage. When the switch shorts PACK+ and PACK-, the current increases quickly to the SCD threshold, 22 A. Meanwhile, the PACK side voltage is pulled to a low value. After a delay of about 70 µs, the discharge MOSFET is turned off and stops the discharging.

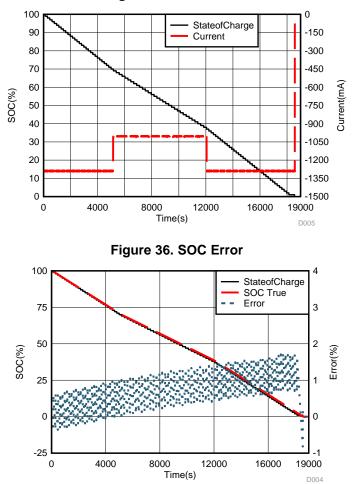


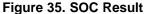




#### 3.2.2.6 SOC Gauging

The SOC gauging accuracy is tested at room temperature. The SOC result is shown in Figure 36. There are 3 stages of discharging: 1.28-A discharge, 1-A discharge, and 1.28-A discharge. Figure 35 shows the SOC error. The maximum error is less than 2%.







### 4 Design Files

### 4.1 Schematics

To download the schematics, see the design files at TIDA-010030.

### 4.2 Bill of Materials

To download the bill of materials (BOM), see the design files at TIDA-010030.

### 4.3 PCB Layout Recommendations

#### 4.3.1 Layout Prints

To download the layer plots, see the design files at TIDA-010030.

### 4.4 Altium Project

To download the Altium Designer® project files, see the design files at TIDA-010030.

### 4.5 Gerber Files

To download the Gerber files, see the design files at TIDA-010030.

### 4.6 Assembly Drawings

To download the assembly drawings, see the design files at TIDA-010030.

### 5 Software Files

To download the software files, see the design files at TIDA-010030.

### 6 Related Documentation

- 1. Texas Instruments, *bq34z100EVM* wide range impedance track<sup>™</sup> enabled battery fuel gauge solution user's guide
- 2. Texas Instruments, bq34z100-G1 wide range fuel gauge with impedance track<sup>™</sup> technology data sheet
- 3. Texas Instruments, bq76930 and bq76940 evaluation module user's guide
- 4. Texas Instruments, *bq769x0 3-series to 15-series cell battery monitor family for Li-lon and phosphate applications data sheet*
- 5. Texas Instruments, *LM5164 100-V input, 1-A synchronous buck DC/DC converter with ultra-low I*<sub>q</sub> data sheet

# 6.1 Trademarks

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Page

# **Revision History**

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from A Revision (February 2019) to B Revision		Page
•	Added This reference design was implemented for a 2-layer PCB.	1

#### Changes from Original (December 2018) to A Revision

•	Changed LM5164-Q1 to LM5164	. 1
	Changed 15-µA to 5-µA throughout document	
	Added information regarding Wake signal to Section 2.3.3	
	Added information to Section 3.2.2.	
•	Added content to Section 6	31
		-

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