

Control-mode quick reference guide

Overview

TI is active in the development of leading-edge control circuits to help engineers address specific design challenges. Since no control mode is optimal for every application, various control modes for non-isolated step-down controllers and converters are referenced with their advantages and how to learn more about each mode. The TI portfolio contains 15 types of control architectures for non-isolated TPS- and LM-series switching DC/DC converters and controllers.

Voltage mode	Internally-compensated advanced current mode (ACM)	Direct connection to the output capacitor (D-CAP™)
Voltage mode with voltage feed-forward	Hysteretic control mode	D-CAP+™ control mode
Peak current mode	Constant on-time	D-CAP2™ control mode
Average current mode	Constant on-time with emulated ripple mode	D-CAP3™ control mode
Emulated current mode	DCS-Control™; Direct control with seamless transition into power-save mode	D-CAP4™ control mode

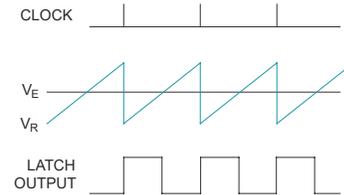
Voltage mode

Pulse-width modulation (latch output) is accomplished by comparing a voltage error signal (V_E) from the output voltage and reference voltage to a constant saw-tooth-ramp waveform. The ramp is initiated by a clock signal from an oscillator. Good noise-margin performance is attained with a fixed ramp amplitude (V_R). Voltage regulation is independent of the output current. Voltage mode uses type-3 compensation addressing a double-pole power stage to support a wide range of output filter combinations for externally compensated devices.

When to use: When a fixed, predictable switching frequency is desired. Also useful when wide output-load variations are possible.

Popular devices: [TPS54610](#), [TPS40040](#), [LM22670](#)

Learn more: [Switching Power Supply Topology Voltage Mode vs. Current Mode](#)



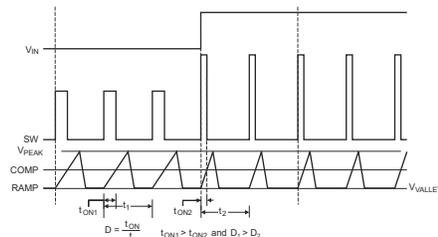
Voltage mode with voltage feed-forward

Similar to voltage mode, but ramp generator varies the PWM ramp slope with the input voltage at a constant ramp magnitude and delivers an instantaneous response to input voltage variations. The PWM does not have to wait for loop delays to change the duty cycle.

When to use: When a fixed, predictable switching frequency is desired. Also useful when wide variations of input voltage and output load are possible.

Popular devices: [TPS40057](#), [TPS40170](#), [TPS56121](#)

Learn more: [Effect of Programmable UVLO on Maximum Duty Cycle Achievable With the TPS4005x and TPS4006x Family of Synchronous Buck Controllers](#)



Peak current mode

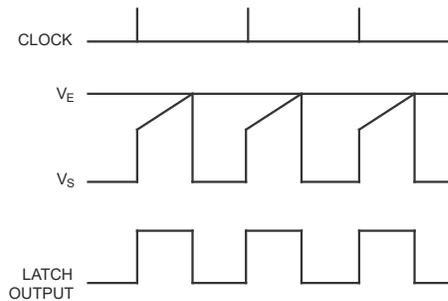
Pulse-width modulation (latch output) is accomplished by comparing a voltage error signal (V_E) and a ramp waveform (V_S) derived from the output current. The ramp is initiated by the clock signal. This mode offers fast

response to output current changes. However, it can be susceptible to noise sensitivity at low duty cycles due to leading-edge current spike. It uses type-2 compensation addressing a single-pole power stage for externally compensated devices.

When to use: When a fixed, predictable switching frequency is needed with a lower parts count than the externally-compensated, double-pole voltage mode. Peak current mode uses a single zero compensator, which is easier to design than voltage mode’s double-zero compensator.

Popular devices: [TPS54620](#), [TPS62913](#), [LM5140-Q1](#)

Learn more: [Understanding and Applying Current-Mode Control Theory](#)



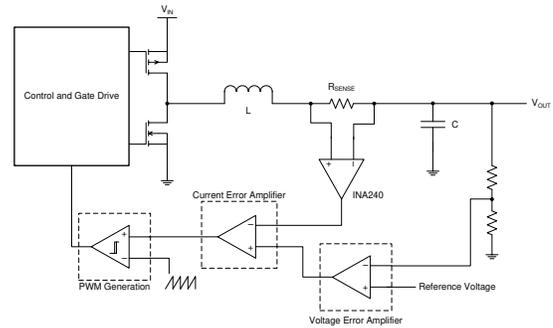
Average current mode

Average current mode addresses noise immunity issues, peak-to-average current errors, and slope compensation needs of peak current mode. Average current mode introduces a high gain integrating current error amplifier into the current loop. The voltage across a current sense resistor represents the actual inductor current. The difference, or current error, is amplified and compared to a large amplitude saw-tooth (oscillator ramp) at the PWM comparator inputs. The gain of the current loop effectively sets the slope compensation without restricting the minimum on-time or minimum-off time. Current sensing is usually inside the regulator, but can be external.

When to use: Effectively control currents other than inductor current, allowing a much broader range of topological application.

Popular devices: [TPS546D24S](#), [TPS546B24S](#)

Learn more: [Average Current Mode Control of Switching Power Supplies](#)



Emulated current mode

Similar to current mode, but employs a gated sample and hold circuit to capture current information emulated by measuring inductor voltage to estimate the ramp current. Eliminates the leading-edge spike issue of the traditional peak-current mode by allowing smaller duty cycles. Provides a clean current waveform when operating near the minimum on-time.

When to use: When low duty cycle is needed versus traditional current mode, without current noise susceptibility.

Popular devices: [LM5116](#), [LM5119](#)

Learn more: [Emulated Current Mode Control for Buck Regulators Using Sample and Hold Technique](#)

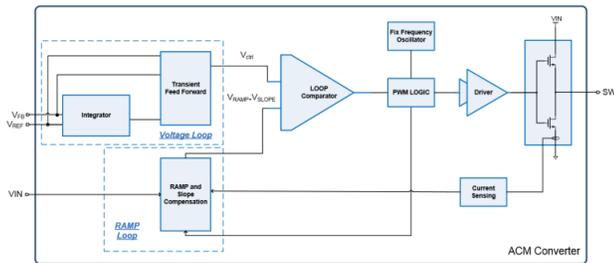
Internally-compensated advanced current mode (ACM)

Internally-compensated ACM is a ripple-based, peak-current-mode control scheme that uses an internally generated ramp to represent the inductor current. This control mode provides a balance between the fast transient response of non-linear control modes (D-CAP™, constant on-time, and so forth) and the broad capacitor stability of other externally-compensated, fixed-frequency control modes (voltage mode, current mode). Internally-compensated advanced current mode provides a fixed, predictable frequency and a simplified compensation selection to reduce external components.

When to use: When fixed frequency and/or stack ability is needed with good output capacitor tolerance and a simplified compensation selection.

Popular devices: [TPS543B22](#) , [TPS543C20A](#), [TPS543620](#)

Learn more: [Internally Compensated Advanced Current Mode \(ACM\)](#)



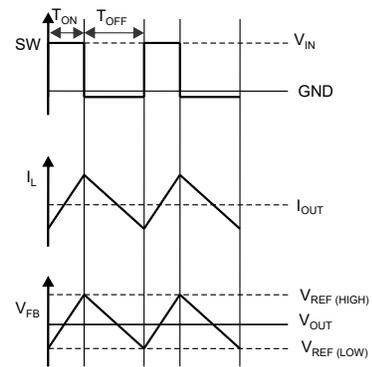
Hysteretic control mode

The simplest control scheme. The PWM (SW) on-time (T_{ON}) is terminated when the feedback voltage is greater than a reference-high threshold and the off-time (T_{OFF}) is terminated when the feedback voltage is less than a reference-low threshold. No compensation components are required. The PWM switching frequency is not controlled and varies with load current and delivers higher efficiency at lighter loads.

When to use: When fast transient response is required. There is no clock-signal time delay to initiate the ramp. A certain amount of ripple is required at the output from the output capacitor's ESR.

Popular devices: [LM3475](#), [LM3485](#)

Learn more: [LM3485 Hysteretic PFET Buck Controller Data Sheet](#)



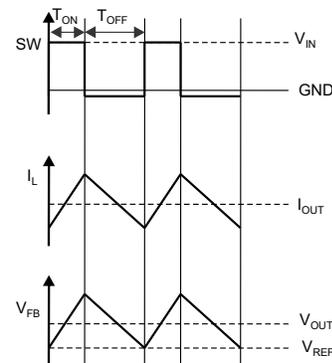
Constant on-time

A slight variation to hysteretic control minimizing frequency shift, but with a single voltage-threshold level, yet achieving fast transient response. The on-time is terminated by a one-shot on-timer and is proportional to the input voltage. The off-time is terminated when the feedback voltage falls below the reference-low threshold.

When to use: When fast transient response is required and a fixed or predictable switching frequency is not required. A certain amount of ripple is required at the output from the output capacitor ESR.

Popular devices: [LM5017](#), [LM2696](#), [TPS54A20](#)

Learn more: [Controlling Output Ripple and Achieving ESR Independence in Constant On-Time \(COT\) Regulator Designs](#)



Constant on-time with emulated ripple mode

A variation of the COT regulator that senses a portion of the low-side MOSFET's off-time current and injects it into the error comparator to emulate ripple. This control mode has the same fast transient response and fewer external component advantages of COT.

When to use: When employing low-ESR ceramic capacitors or when an external ripple injection circuit is undesirable.

Popular devices: LM3100, LM3150

Learn more: [Emulated ripple technique advances hysteretic switch-mode supplies](#)

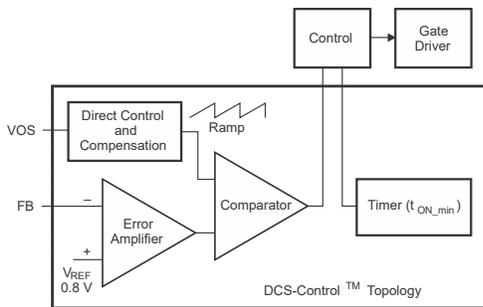
DCS-Control™: Direct control with seamless transition into power-save mode

Combines the advantages of hysteretic control for a fast transient response without compensation components, and the advantages of voltage-mode control for high DC accuracy with a seamless transition from PWM to power saving mode (PSM).

When to use: When light-load efficiency is needed with small, low-ESR ceramic capacitors.

Popular devices: TPS62872 , TPS628303, TPS62903, TPS82130

Learn more: [High-efficiency, low-ripple DCS-Control™ offers seamless PWM/power-save transitions](#)



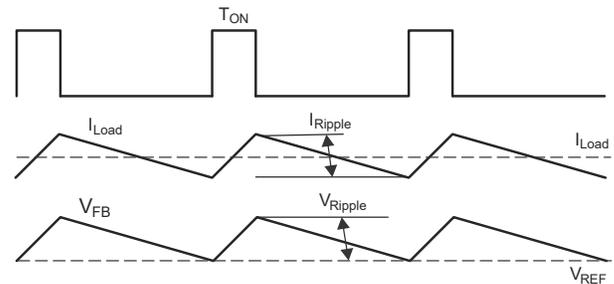
Direct connection to the output capacitor (D-CAP™)

Similar to COT control except a one-shot timer generates an on-time pulse that is proportional to the input voltage and the output voltage. When the falling feedback voltage equals the reference voltage, a new PWM on-pulse is generated. Fast response to load changes is achieved with a high-speed comparator in the control loop. D-CAP™ minimizes frequency shift compared to hysteretic control.

When to use: When a fast transient response is required and POSCAP or medium-ESR output capacitors are used. No loop-compensation calculation or components are needed.

Popular devices: TPS51116, TPS53219A, TPS53355

Learn more: [Adaptive Constant On-Time \(D-CAP™\) Control Study in Notebook Applications](#)



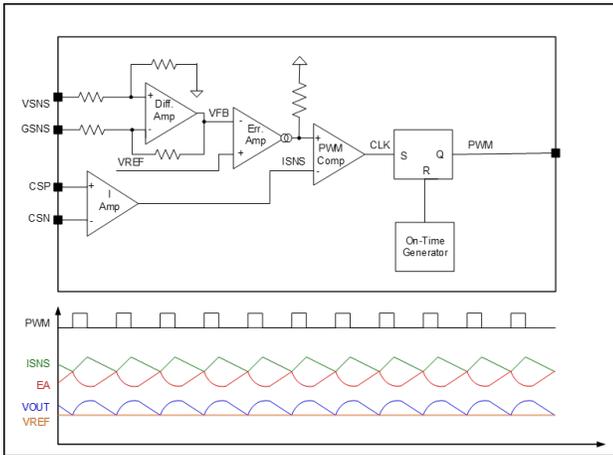
D-CAP+™

D-CAP+ adds an error amplifier to D-CAP that compares V_{FB} to V_{REF} for better output voltage accuracy and a current sense amplifier to sense the current directly, instead of relying on output ESR to act as the sense element. D-CAP+ is a true voltage-controlled current source without a clock limitation like most variants of current mode control. D-CAP+ is used where true current sensing is required, such as multi-phase and droop-compensation (load-line) applications with one output voltage. Current sensing may be accomplished either inside or outside of the power IC depending on the device.

When to use: When high accurate current sensing is needed for load-line or multi-phase controller applications

Popular devices: TPS53661, TPS53667, TPS548C26

Update Learn more: [D-CAP+™ Control for Multi-phase, Step-Down Voltage Regulators for Powering Microprocessors](#)



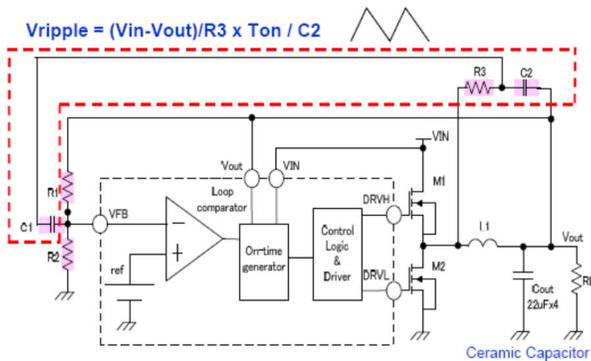
D-CAP2™

A slight variation of D-CAP with the same transient and external component advantages as D-CAP. This control mode supports ceramic output capacitance without external circuitry. A signal from an internal ripple-injection circuit is fed directly into the comparator, thus reducing the need for output voltage ripple from the capacitor’s ESR. The ramp is emulated by the output inductor.

When to use: When desiring fast transient response with low-ESR ceramic output capacitors.

Popular devices: [TPS563202](#), [TPS563210](#)

Learn more: [D-CAP2™ Frequency Response Model](#)



D-CAP3™

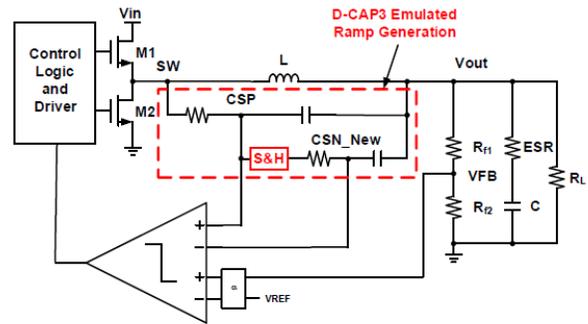
A variation of D-CAP2™ with the same transient and external component advantages. A sample-and-hold circuit is built-in to the converter to remove an offset voltage created by D-CAP2’s emulated ramp circuit,

improving the voltage reference accuracy. Well suited for powering low-core-voltage FPGAs, ASICs and DSPs.

When to use: When a tighter reference voltage accuracy and a fast transient response are desirable when using ceramic output capacitors.

Popular devices: [TPS565247](#) , [TPS56C231](#) , [TPS548B28](#) , [TPS563206](#)

Learn more: [Accuracy-Enhanced Ramp-Generation Design for D-CAP3 Modulation](#)

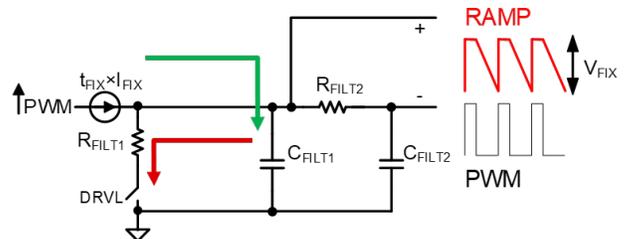


D-CAP4™

D-CAP4 includes the advantages as D-CAP3, but desensitizes the loop gain to the output voltage in order to improve the transient response at higher output voltages. The ramp injection principle is the same as D-CAP3, except the ramp common mode and amplitude are independent of the output voltage. The ramp common mode is inversely proportional to (1-D), keeping ramp amplitude constant, so there is less need to adjust the ramp for different output voltages.

When to use: When fast transient response time is needed with higher output voltages, like 3.3 V or 5 V.

Popular devices: [TPS54KB20](#)



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