Evaluation Board for the TLC320AD50C DSP Analog Interface Circuit

User’s Guide
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About This Manual

This user’s guide discusses the design, use, and performance of the TLC320AD50 Evaluation Module.

How to Use This Manual

This document contains the following chapters:

- Chapter 1 – Introduction
- Chapter 2 – AD50-EVM Design and Construction
- Chapter 3 – Setting Up the AD50-EVM
- Chapter 4 – Results Obtained With AD50 EVM
- Appendix A – Installing the AD50-EVM
- Appendix B – Programmable Logic
- Appendix C – Converting DSK+ Software for the AD50-EVM

Notational Conventions

This document uses the following conventions.

- Program listings, program examples, and interactive displays are shown in a special typeface similar to a typewriter’s. Examples use a bold version of the special typeface for emphasis; interactive displays use a bold version of the special typeface to distinguish commands that you enter from items that the system displays (such as prompts, command output, error messages, etc.).

Here is a sample program listing:

```
0011  0005  0001  .field  1, 2
0012  0005  0003  .field  3, 4
0013  0005  0006  .field  6, 3
0014  0006  .even
```
Here is an example of a system prompt and a command that you might enter:

```bash
C: csr -a /user/ti/simuboard/utilities
```

- In syntax descriptions, the instruction, command, or directive is in a **bold typeface** font and parameters are in an *italic typeface*. Portions of a syntax that are in **bold** should be entered as shown; portions of a syntax that are in *italics* describe the type of information that should be entered. Here is an example of a directive syntax:

```plaintext
.asect "section name", address
```

.asect is the directive. This directive has two parameters, indicated by *section name* and *address*. When you use .asect, the first parameter must be an actual section name, enclosed in double quotes; the second parameter must be an address.

- Square brackets ([ and ]) identify an optional parameter. If you use an optional parameter, you specify the information within the brackets; you don’t enter the brackets themselves. Here’s an example of an instruction that has an optional parameter:

```plaintext
LALK 16-bit constant [, shift]
```

The LALK instruction has two parameters. The first parameter, *16-bit constant*, is required. The second parameter, *shift*, is optional. As this syntax shows, if you use the optional second parameter, you must precede it with a comma.

Square brackets are also used as part of the pathname specification for VMS pathnames; in this case, the brackets are actually part of the pathname (they are not optional).

- Braces ({ and }) indicate a list. The symbol | (read as or) separates items within the list. Here’s an example of a list:

```
{ * | *+ | *– }
```

This provides three choices: *, *+, or *–.

Unless the list is enclosed in square brackets, you must choose one item from the list.

- Braces ({ and }) indicate a list. The symbol | (read as or) separates items within the list. Here’s an example of a list:

```
{ * | *+ | *– }
```

This provides three choices: *, *+, or *–.

Unless the list is enclosed in square brackets, you must choose one item from the list.

- Braces ({ and }) indicate a list. The symbol | (read as or) separates items within the list. Here’s an example of a list:

```
{ * | *+ | *– }
```

This provides three choices: *, *+, or *–.

Unless the list is enclosed in square brackets, you must choose one item from the list.

- Some directives can have a varying number of parameters. For example, the .byte directive can have up to 100 parameters. The syntax for this directive is:

```plaintext
.byte value1 [, ... , value100]
```

This syntax shows that .byte must have at least one value parameter, but you have the option of supplying additional value parameters, separated by commas.
Related Documentation From Texas Instruments

TLC320AD50C Data Manual, Literature number SLAS131
Data Acquisition Data Book, Literature number SLAD001
Data Converter Selection Guide, Literature number SLABE05
Operational Amplifiers Data Book Volume A, Literature number SLYD011
Operational Amplifiers Data Book Volume B, Literature number SLYD012
Rail-to-Rail Operational Amplifier Selection Guide,
Literature number SLOBE02
Single Supply Operational Amplifier Selection Guide,
Literature number SLOBE03
Mixed Signal Analog CD-ROM, Literature number SLYC005
TMS320C54x CPU and Peripherals, Literature number SPRU131
TMS320C54x Algebraic Instruction Set, Literature number SPRU179
TMS320C54x DSKplus User’s Guide, Literature number SPRU191

Much useful software is available from the TI Internet site. The main TI Web site is at

http://www.ti.com/

Information on the DSK+ is at

http://www.ti.com/sc/docs/dsps/tools/c54x/c54xdskp.htm

DSK+ software can be downloaded from

http://www.ti.com/sc/docs/dsps/tools/c54x/softsupp.htm
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This user's guide discusses the design of the AD50-EVM evaluation board and its use to demonstrate the performance of the TLC320AD50C (AD50) analog interface circuit.

The TLC320AD50C provides high-resolution low-speed signal conversion from digital-to-analog (D/A) and from analog-to-digital (A/D) using oversampling sigma-delta technology. This device consists of two serial synchronous conversion paths (one for each data direction) and includes an interpolation filter before the DAC and a decimation filter after the ADC. Other overhead functions provide on-chip timing and control. The sigma-delta architecture produces high-resolution analog-to-digital and digital-to-analog conversion at low system speeds and low cost.

The options and the circuit configurations of this device can be programmed through the serial interface. The options include reset, power down, communications protocol, serial clock rate, signal sampling rate, gain control, and test mode.

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The TLC320AD50 is an analog interface circuit (AIC) with many features that make it suitable for DSP-based applications:

- Single 5-V supply or dual (5-V analog and 3.3-V digital) supplies
- 16-bit resolution sigma-delta ADC and DAC
- 85 dB (min) signal to noise
- Inherent antialiasing filtering and \( \sin(x)/x \) compensation
- High input impedance
- Supports up to 4 devices on one serial interface
- Low operating power (175 mW max)
- Power down mode (20 mW max)
- Small package size

1.1 Scope of User’s Guide

This application note discusses the design of the AD50-EVM evaluation board and its use to demonstrate the performance of the TLC320AD50C (AD50) analog interface circuit. The results come from measurements on a small number of samples. For specifications refer to the datasheet.

The AD50-EVM has two AD50 devices for stereo operation. Two AD50-EVMs can be configured as a four-channel system using a single serial interface. The AD50-EVM can be interfaced directly to the DSK+ DSP starter kit or other systems which have a compatible synchronous serial interface.

The objective was to design a development board (the AD50-EVM) which would allow prospective users of the AD50 to determine its capabilities with a minimum effort. The board can be directly connected to the low cost TMS320C54x DSP starter kit (DSK+), or to any other system with a compatible synchronous serial interface. Directly compatible DSP devices include TMS320C2x, C2xx, C3x, C5x, C54x and C6xxx.

An example program for the DSK+ development system allows the board to be used as a sine-wave generator, or to output on the DAC samples read in from the ADC. In this echo mode signal processing functions such as filtering can easily be included. The AD50-EVM board was also interfaced to a TMS320C25 development board, which was used to transfer analog data to a personal computer running real-time FFT spectrum analysis software. This system was used to prepare the ADC and DAC FFT spectrograms shown in this user’s guide.
1.2 Block Diagram of AD50-EVM

Figure 1–1 shows the block diagram of the AD50-EVM evaluation board.

Figure 1–1. AD50-EVM Block Diagram

The input signal is first buffered and optionally amplified by the preamp stage before being level-shifted and converted to a differential pair of signals. The external antialiasing filter is a simple continuous time filter to remove RF noise. The output from the codec is converted to a single ended signal and is filtered to remove high frequency noise.

A negative power supply is generated on the AD50-EVM to allow for dc coupled input and output signals. The clock signal for the codec can be generated by an oscillator on the board or from an external clock. The sampling frequency is set using the programmable clock divider within the AD50. The AD50 FC, RESET, and POWERDOWN signals can be controlled via a software programmable register.
1.3 Special Considerations When Using Sigma-Delta Converters

Sigma-delta analog-to-digital converters typically consist of an analog modulator (fourth order in the case of the AD50) followed by a digital filter section. The modulator contains a 1-bit ADC (a comparator) which produces a 1-bit wide data stream, which is applied to the input of the digital filter. It is also applied to the input of a 1-bit DAC, the output of which is fed back to the input of the modulator. The 1-bit ADC is clocked much faster than the desired output sampling frequency ($F_S$), ($64 \times F_S$ for the AD50) and the large amount of quantization noise generated is uniformly spread over a wide range of frequencies. A noise-shaping filter within the analog modulator reduces the noise in the pass-band, increasing it elsewhere. A low-pass digital filter then removes the unwanted high frequency quantization noise and the signal is resampled at the desired output frequency.

The sigma-delta DAC in the AD50 consists of a digital interpolating filter operating at $256 \times F_S$ followed by a 1-bit DAC and a second-order modulator. The digital filter removes most of the image frequencies, that would otherwise be present at the output of a conventional DAC. This avoids the need for a high-order analog low-pass reconstruction filter. Separate $\sin(x)/x$ compensation is not needed, as this is inherent in the DAC architecture.

Sigma-delta converters have several advantages and a few disadvantages relative to other types of ADC which are summarized below.

1.3.1 Advantages

- Sigma-delta conversion is inherently linear because there is no reference resistor chain as in flash or successive approximation converters. This results in extremely low distortion.
- Inherent monotonicity
- No missing codes
- Antialias filtering is inherent within sigma-delta converters. This greatly simplifies their application, because only a simple external RC filter is required at the input of the ADC to achieve the necessary alias rejection.
- The digital filters which form an integral part of the ADC and DAC are usually of the finite impulse response type, which gives a linear phase characteristic with high stability.
- The cutoff frequency of the digital filters automatically tracks the sampling rate.
- DAC anti-image filtering is greatly simplified because an internal digital filter attenuates image frequencies. A simple analog filter is sufficient to remove HF noise.
- No need for $\sin(x)/x$ compensation.
- The sigma-delta architecture is compatible with dense, low-cost, digital IC processes.
1.3.2 Disadvantages

- The digital filters in sigma-delta converters introduce into the conversion process a time delay that makes them unsuitable for some control applications. The AD50 ADC has a delay of 17 samples and the DAC has an 18 sample delay.

- It is not practical to multiplex several inputs to a single sigma-delta ADC except at very low rates, because each channel would be corrupted by the earlier samples from other channels still propagating through the digital filter. Although the AD50 has an input multiplexer, this is only for selecting one or another input, not for interleaving two input channels onto one data stream.

- Audio band converters are optimized for ac signals and a small dc offset may be present.

- Spurious low level tones can sometimes occur at very low input signal levels, especially if clock signals at Fs/2 are allowed to couple into the reference voltage pins. Such tones can be identified because their frequency is affected by small changes in dc offset.
This chapter discusses the printed-circuit board design considerations for the AD50-EVM.

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2.1 PCB Construction

The AD50-EVM printed-circuit board is constructed of 4 layers, with ground and power planes sandwiched between the top and bottom signal-carrying layers. This minimizes the coupling of RF noise into the system by providing very low impedance to power and ground, and by shielding signal tracks. The AD50 is available in two surface-mount packages, the DW and PT. The PT has a smaller footprint size and is much thinner (only 1.6 mm high). It has 48 pins at a pitch of 0.5 mm, making it very suitable for PCMCIA and other miniature applications. However it is difficult to hand solder, and for prototype evaluation the use of the DW package is recommended. This package has 28 pins at a pitch of 1.27 mm (0.05 inch) and so is much easier to handle. Surface-mounted components were used except for the connectors, which through-hole connectors offering robustness.

The printed-circuit board is the same size as the DSK+ board. The AD50-EVM and the DSK+ can be stacked one above the other using inter-board links. Either board can be on top, but placing the AD50-EVM on top gives better access to the EVM test points.

A separate connector has been provided for interfacing to systems other than the DSK+. This brings out the SCLK, FS, DIN, DOUT, and RESET signals. Each signal is interleaved with a ground conductor in the ribbon cable, allowing a longer cable length without crosstalk. Nevertheless, this cable should be as short as reasonably possible, since ringing in the unterminated cable may become excessive for lengths greater than 3 feet. To minimize RF emissions, the cable should be shielded, with the shield connected to the ground plane at the ground point provided adjacent to pins 1 and 2 of the connector. This connector must not be used simultaneously with the connections around the edge of the AD50-EVM board that are specifically intended for the DSK+.

For the best signal-to-noise ratio, it is often recommended that the ground plane be split into separate analog and digital sections, joined together in only one place with the split passing under the AD50 package between its analog and digital ground pins. The split prevents digital noise currents from inducing noise voltages in the analog ground of the device. The DSK+ has ground and power pins along both edges of the board, making the use of split ground planes ineffective. Instead short breaks were inserted on the ground plane to separate the analog and digital grounds and reduce the common impedance.
2.2 Power Supply

The AD50-EVM board requires only +5 Vdc; a –5-Vdc supply for some of the op-amps is generated on the board using a CMOS 7660 charge pump phase-locked to the frame sync output of the AD50. Phase-locking minimizes the risk of audible beats between the sampling clock and the 7660 clock. The OSC pin of the 7660 is coupled to the master FSD with a 22 pF capacitor. This method allows the 7660 oscillator to free-run when there is no frame sync, ensuring that the negative power is always maintained, even when the AD50 is held in reset. The 7660 divides the signal on the OSC pin by 2 internally, ensuring that the charge pump operates at a 50% duty cycle, even with a grossly asymmetric input such as that provided by frame sync. Not all negative supply generators contain a divider. Check before using any other type.

The digital part of the AD50 can be operated at 3.3 V. Users wishing to investigate this should remove R16 and R36 and connect 3.3 V power to L_VD and R_VD.

2.3 System Clock

When the AD50-EVM is used in standalone mode, in conjunction with a separate DSP system, a crystal oscillator must be fitted to the socket provided. A frequency of 10.24 MHz is suggested, as this will allow standard sampling frequencies such as 10, 16, and 20 kHz to be achieved by programming the AD50 clock divider and phase-locked loop. Table 2–1 shows the sampling frequencies that can be selected for a number of master clock frequencies. Frequencies shown in brackets are above the maximum sampling frequency specified for the AD50.

In standalone mode the AD50-EVM can be clocked at frequencies up to 22.579 MHz. However, when used in conjunction with the DSK+ the upper frequency is limited by the maximum clock rate of the TMS320C542 DSP device. The 10.24 MHz oscillator supplied with the AD50-EVM will drive the TMS320C542 on the DSK+ at 40.96 MHz, which is only slightly higher than its maximum operating frequency of 40.00 MHz. However, at room temperature this should not cause any problems, because the TMS320C542 is tested by TI at 40.00 MHz over the temperature range 0-70 °C. The authors have successfully operated a DSK+ in conjunction with an AD50-EVM at MCLK frequencies from 4 MHz to 13 MHz at room temperature, but this may not work with all DSK+ units. Alternatively, the DSK+ and AD50-EVM clocks can be separated by removing resistor R80 from the AD50-EVM board and a suitable oscillator installed on each board. (If this modification is made, 3 and 4 channel modes will not work correctly because all the AD50s must have an identical MCLK. Removing R80 prevents the clock from being transmitted between the boards.)

It is important to ensure that only one oscillator is used in the system (except as described above), otherwise results will be unpredictable. No damage will be caused, however, as current limiting resistors are provided on the output of each oscillator. Electrically it makes no difference whether the oscillator is fitted to the DSK+, or to the first or second AD50-EVM board.
### Table 2–1. Sampling Frequencies for Various MCLK and Register 4 Values

<table>
<thead>
<tr>
<th>Reg 4, Bit 7</th>
<th>Reg 4, Bits 6–4</th>
<th>8.192 MHz</th>
<th>10.000 MHz</th>
<th>10.240 MHz</th>
<th>11.2896 MHz</th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7 = 0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>PLL on</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>0 (default)</td>
<td>8 kHz</td>
<td>9.765 kHz</td>
<td>10 kHz</td>
<td>11.025 kHz</td>
<td></td>
</tr>
<tr>
<td>1</td>
<td>(64 kHz)</td>
<td>(78.125 kHz)</td>
<td>(80 kHz)</td>
<td>(88.2 kHz)</td>
<td></td>
</tr>
<tr>
<td>2</td>
<td>(32 kHz)</td>
<td>(39.063 kHz)</td>
<td>(40 kHz)</td>
<td>(44.1 kHz)</td>
<td></td>
</tr>
<tr>
<td>3</td>
<td>21.333 kHz</td>
<td>(26.042 kHz)</td>
<td>(26.666 kHz)</td>
<td>(29.4 kHz)</td>
<td></td>
</tr>
<tr>
<td>4</td>
<td>16 kHz</td>
<td>19.531 kHz</td>
<td>20 kHz</td>
<td>22.05 kHz</td>
<td></td>
</tr>
<tr>
<td>5</td>
<td>12.8 kHz</td>
<td>15.625 kHz</td>
<td>16 kHz</td>
<td>17.64 kHz</td>
<td></td>
</tr>
<tr>
<td>6</td>
<td>10.666 kHz</td>
<td>13.021 kHz</td>
<td>13.333 kHz</td>
<td>14.7 kHz</td>
<td></td>
</tr>
<tr>
<td>7</td>
<td>9.1432 kHz</td>
<td>11.161 kHz</td>
<td>11.429 kHz</td>
<td>12.601 kHz</td>
<td></td>
</tr>
<tr>
<td>bit 7 = 1</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>PLL off</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>0</td>
<td>2 kHz</td>
<td>2.441 kHz</td>
<td>2.500 kHz</td>
<td>2.756 kHz</td>
<td></td>
</tr>
<tr>
<td>1</td>
<td>16 kHz</td>
<td>19.531 kHz</td>
<td>20 kHz</td>
<td>22.05 kHz</td>
<td></td>
</tr>
<tr>
<td>2</td>
<td>8 kHz</td>
<td>9.766 kHz</td>
<td>10 kHz</td>
<td>11.025 kHz</td>
<td></td>
</tr>
<tr>
<td>3</td>
<td>5.333 kHz</td>
<td>6.510 kHz</td>
<td>6.666 kHz</td>
<td>7.35 kHz</td>
<td></td>
</tr>
<tr>
<td>4</td>
<td>4 kHz</td>
<td>4.883 kHz</td>
<td>5 kHz</td>
<td>5.513 kHz</td>
<td></td>
</tr>
<tr>
<td>5</td>
<td>3.2 kHz</td>
<td>3.906 kHz</td>
<td>4 kHz</td>
<td>4.41 kHz</td>
<td></td>
</tr>
<tr>
<td>6</td>
<td>2.666 kHz</td>
<td>3.255 kHz</td>
<td>3.333 kHz</td>
<td>3.675 kHz</td>
<td></td>
</tr>
<tr>
<td>7</td>
<td>0.798 kHz</td>
<td>2.79 kHz</td>
<td>2.857 kHz</td>
<td>3.15 kHz</td>
<td></td>
</tr>
</tbody>
</table>

#### 2.4 Reset

In all operating modes the AD50-EVM is reset when the power is switched on. In DSK+ mode it is also reset when the DSK+ itself is reset. The programmable logic on the AD50-EVM is configured so that in the reset state all the AD50s are powered down and in slave mode and that the AC01 on the DSK+ is powered up as normal. This means that even with one or two AD50-EVMs attached to a DSK+ system, all the DSK+ demonstration software works normally with the AC01 AIC.

To use one or more AD50s it is necessary to program bits 0-3 of I/O address 0. Bit 0 controls the reset state of the AD50s. Bit 0 = 0 resets all AD50s and bit 0 = 1 allows them to run. Bits 1, 2, and 3 are encoded to control the number of AD50s that are powered up (see Table B–1 for details). Note that no provision has been made to operate the AD50s and the AC01 together. This is because their control registers are incompatible and it would be difficult to prevent bus conflicts on the serial interface.

The reset pins are connected to both the SERIAL and CONTROL cable headers for use in standalone mode (see Table A–2 for pinouts).
2.5 Serial Port Interfacing

The AD50-EVM has been designed to interface to the same synchronous serial port that is used by the AC01 on the DSK+ system. This is achieved by sharing the serial clock, frame sync and data signals, and by powering down the AC01 when AD50s are in use.

When two AD50-EVMs are connected together the serial interface frame sync signals need to be configured differently on the primary and secondary boards. This is achieved using the 22V10 programmable logic device as shown in Figure 2–1. The logic equations are listed in Appendix B. The serial clock, serial data in, and serial data out signals are connected to all devices. Only one device at a time drives these lines.

When two AD50-EVMs are used in standalone mode together with a separate DSP system, the ribbon cable must be attached to the primary board so that the master device frame sync signal is transmitted to the external system. The two AD50-EVM boards have all the necessary signals linked via the four connectors around the periphery.

Figure 2–1. Configuration of Serial Interface on Primary and Secondary AD50-EVMs
2.6 Analog Input

The AD50 can be operated from a single 5 V supply or from a 5 V analog supply and a 3.3 V digital supply. The AD50-EVM uses 5 V for both analog and digital supplies to be compatible with the DSK+, and generates an additional –5 V supply on the board. This negative supply allows the input and output circuitry to be dc coupled. In an ac coupled system alternative configurations may be used that do not require a negative supply.

The input circuitry consists of a switchable gain preamp, a single-ended to differential converter with level shifting, and an antialiasing filter.

2.6.1 Preamp Design

This amplifier stage, illustrated in Figure 2–2 allows the use of either a line level input (approximately 1 Vrms full scale) or a low level input (approximately 100mVrms full scale). With the link labeled GAIN removed the preamp has unity gain from dc to 20 kHz. With this link inserted the preamp gain is increased by 20 dB.

The signal can be ac coupled, by removing the link labeled DC. The use of two tantalum capacitors back-to-back allows ±10 V dc voltage bias at the input without damaging the capacitors.

Figure 2–2. AD50-EVM Input Preamp

This preamp uses the op amp in the inverting configuration to ensure that the input common mode range of the op amp cannot be exceeded, since both inputs are always at a voltage close to ground. In the non-inverting configuration the op amp inputs can approach the supply rails, causing distortion.

2.6.2 Input Stage Design

The AD50 analog to digital section uses differential inputs biased at 2.5 V. Maximum code is generated with 3 V PEAK-PEAK on both differential inputs. The input stage provides the necessary level shifting and single-ended to differential conversion, prior to the inputs of the AD50.
The first op amp inverts the incoming signal, to provide the differential pair while the second two op amps do the necessary level shifting. The design of this stage ensures that any noise on FILT appears equally, and in phase, on both differential inputs. The first op amp needs ±5-V supplies, while the second two are powered from just 5-V and ground to ensure that the inputs to the AD50 never go below ground.

It is recognized that this device is likely to be used in predominantly digital designs, powered from a single 5 V supply, where the need for a negative supply could be perceived as a problem. However, since the current drawn from this negative supply will be small, it is possible to use an inverter chip to provide a local negative rail for the op amps. The 7660 device requires no external inductors and just 2 external capacitors. Its internal clock can be synchronized to the codec’s sampling clock, ensuring that any interference generated by the 7660 appears on the codec’s output as a dc offset or at the Nyquist frequency rather than as noise in the pass-band.

### 2.6.3 Op Amp Selection

The design of the input stage uses op amps in the inverting configuration. This has the advantage of keeping both inputs at the reference voltage, preventing problems with input common mode range.
For a full-scale signal the outputs of the op amps that drive the codec inputs will reach 4 V. Many op amps will not behave well at this output level with a 5-V supply (in fact the system was tested at 4.75 V to allow for the minimum rated power supply voltage of the AD50).

Miniature systems often require that analog and digital circuits be in close proximity to each other. This can lead to problems where radio frequency interference from clock signals or DSP bus lines is demodulated by op amps, leading to an increased DC offset. If the RF is modulated, then that modulation may appear at the op amp output. CMOS and BiFET op amps can be more resistant to demodulating RF than bipolar devices. Take particular care to keep digital signals away from analog ones, and be generous with power supply decoupling and filtering. Power planes help a great deal in reducing system noise.

The authors chose the TI device TLC2272 which is a low noise CMOS dual op amp designed for single supply operation. It has full rail to rail output swing and low distortion. However the distortion increases when driving loads of about 2 kΩ or less.

### 2.6.4 Anti-aliasing Filter

Sigma-delta converters have the advantage of providing anti-aliasing filtering as an integral part of their operation. However this filtering has holes in it, at multiples of the oversampling frequency. The AD50 is a 64 times oversampling converter so for a sampling rate of 20 kHz the first hole will be centered at 1.28 MHz and will be 20 kHz wide. It is important to ensure that no energy at this frequency is present across the inputs to the codec. A simple single pole RC filter is sufficient. If ceramic capacitors are used for this filter they should be of the COG or NPO dielectric type. Significant distortion can be introduced by the voltage dependent capacitance of other types of dielectric. Two filter configurations are possible, as shown in Figure 2–4.

**Figure 2–4. AD50-EVM Anti-aliasing Filter Options**

![Figure 2–4a](image) ![Figure 2–4b](image)

Figure 2–4a gives good rejection of differential noise signals but common mode noise is not removed.

Figure 2–4b gives good rejection of common mode noise signals and, if the capacitors and resistors are well matched, good rejection of differential noise.
Since most noise sources are likely to be coupled equally into both inputs the filter in Figure 2–4b is usually the best choice (the evaluation board has footprints to allow for either or both configurations, with the capacitor in Figure 2–4a not connected).

The AD50 allows higher source impedances at its inputs than many sigma-delta converters, easing the drive requirements.

The aux inputs also have antialiasing filters connected, allowing the use of these inputs from external sources without any special filtering requirements.

The antialiasing filter components are placed as close as possible to the inputs to the AD50 to reduce the possibility of noise pickup between the filter and the ADC. The use of 0603 footprint surface-mount components makes this possible.

2.7 Analog Output

The AD50 uses a pair of differential voltage outputs. The output stage needs to convert the differential signals to a single ended output, and to attenuate noise outside the pass-band. This filter is not a conventional reconstruction filter, since the AD50 has an integral low-pass reconstruction filter. The filter is a 12-kHz second order low-pass filter designed for use at 20 ksps and will be less effective at lower sampling rates. Since this filter only removes out of band noise it will be unnecessary in many applications.

*Figure 2–5. AD50-EVM Differential to Single-Ended Output Converter*
Setting up the AD50-EVM normally involves coupling it to the synchronous serial port of a digital signal processing system. Two configurations are supported. The first is to couple the AD50-EVM to a DSK+ development system by means of four connectors directly linking the circuit boards together. The DSK+ consists of a TMS320C542 16-bit fixed point digital signal processor combined with a 14-bit analog interface circuit and PC printer port compatible data interface. This allows the downloading and debugging of C or assembler programs from a DOS or Windows environment. Many example programs are supplied with the DSK+, which can readily be adapted to interface to the AD50-EVM. (See Appendix C for detailed information on adapting example programs.) The second configuration is standalone mode where a ribbon cable up to 1 m long connects the AD50-EVM to a separate DSP system. In standalone mode some options such as the number of active AD50s must be set using jumpers or by control signals from the remote processor. In DSK+ mode, almost everything is set from software.

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</table>
3.1 Setting Up the AD50-EVM With the DSK+

The first step is to couple the two circuit boards together using four inter-board connectors. See Appendix A for full details. The AD50-EVM takes its power directly from the DSK+. Inputs and outputs are provided on stereo 3.5 mm jacks. Note that in single channel mode, the active device connects to the ring contact on the jack. This is conventionally the left channel on stereo headphones. Attaching the AD50-EVM to the DSK+ does not affect the operation of the AC01 on the DSK+ because the reset default is for both AD50 devices to be powered down and in slave mode. The master clock source can be the 10.00 MHz oscillator supplied with the DSK+, the 10.24 MHz oscillator provided with the AD50-EVM or any other oscillator module within the capabilities of the DSK+ (see section 2.3 for details). It does not matter which socket is used, so long as only one oscillator module is present in the system. Check that the operation of the SELFTEST and example programs has not been affected by adding the AD50-EVM or the oscillator selection.

Either type in and assemble a program such as the sine-wave generator listed below, or modify an appropriate example program such as OSCOPE supplied with the DSK+. Appendix C shows how OSCOPE was modified to work with the AD50-EVM.

3.2 Setting Up the AD50-EVM in Stand-Alone Mode

In stand-alone mode the AD50-EVM needs to be connected to a system with a suitable DSP serial interface. This should include a means of controlling the RESET signal. The connector marked SERIAL has all the necessary signals, with interleaved grounds. The POWERDOWN and other control signals are available on the CONTROL connector. In stand-alone mode all the devices are powered up by default.

A power supply of 5 V at approximately 150 mA needs to be supplied to the positive (+) and 0 pins of the PSU connector. The negative (–) pin is an output from the negative voltage converter.

In stand-alone mode an oscillator of up to 22.579 MHz needs to be inserted in the socket on the AD50-EVM. A 10.24 MHz oscillator is supplied.

Configuration of the registers in the AD50s is software controlled via the serial interface, as described for the DSK+ mode. However the RESET, POWERDOWN and FC signals need to be set via the CONTROL connector.

3.3 Configuring the Serial Interface

First, the four least significant bits of I/O address 0 must be configured with the value 2. This selects one AD50 and disables the AC01, holding the AD50 in reset. Note that the AC01 on the DSK+ is reset by manipulating the XF output of the C542, whereas the AD50-EVM implements the software reset control as bit 0 of I/O address 0. This frees the XF bit for other uses, such as driving
Programming the AD50 Registers

3-3 Setting Up the AD50-EVM

an oscilloscope for debugging purposes. Then the ’C542 serial port is initialized to 16-bit mode with external frame sync pulses. The AD50 is then taken out of reset by writing 3 to I/O address 0. The AD50 starts sending frame sync pulses and a serial clock signal to the ’C542, which responds by transmitting data from the serial port. This data consists of a list of alternating primary and secondary transmissions. The primary transmissions are dummy words with bit 0 set to 1 to request the interleaved secondary transmissions. Each secondary transmission is a 16-bit word containing commands to initialize the four control registers of the AD50.

For each secondary transmission, bits 12-8 encode the binary address of the register to be accessed, bit 13 specifies a read or write access and bits 7-0 contain the data to be written for write operations.

3.4 Programming the AD50 Registers

Register 0 is a dummy register, equivalent in principle to a NOP instruction. Writing to it does nothing. It is needed so that when several AD50s are cascaded it is possible to modify the control registers of one device without disturbing the others.

Register 1 controls reset and power-down status, input selection, monitor amplifier signal source and gain, digital loop back test mode and 15 or 16 bit DAC mode. Once 16 bit mode has been selected, it is not possible to request further secondary communications by setting bit 0 of a serial port data word. However, the AD50-EVM provides a means for driving the FC pin of the AD50 chip by writing to bit 0 of I/O address 1. The value written is applied to the FC pin, and a 1 initiates a hardware secondary communication request. FC is sampled by the AD50 on the rising edge of frame sync.

Register 2 allows control of the flag bit, reading of the decimator overflow flag, selection of 15 or 16 bit ADC mode and control of the analog loop back test mode.

Register 3 is used to select the number of cascaded AD50 devices (up to a maximum of 1 master and 3 slaves) and to control the time between communications from each device. It is important that register 3 is only programmed after the sampling frequency has been set. It is convenient to program everything common to all the devices prior to setting register 3. Thereafter, when secondary communications are requested, each device takes part in turn and must either be programmed individually or a NOP instruction sent to register 0.

Register 4 controls the analog input and output gains and the sampling frequency.

It is only necessary to reprogram those registers that have unsuitable default values. If none of the AD50 registers are programmed, the system will default to a sampling frequency of 10 kHz for a MCLK frequency of 10.24 MHz. If I/O address 0 is not programmed as described above, the system will default to using the AC01 ADC and DAC as normal.
3.5 Sine-Wave Generator and Loopback Program

The following code is for a very simple sine-wave generator and loop back program that demonstrates how the DSK+ can be programmed to drive the AD50-EVM. The sine-wave is generated by means of a lookup table which is cyclically indexed. This program was used to generate the test waveforms for the DAC distortion measurements in section 4 below. The sine table delivers an output waveform at -3dB relative to clipping, to allow direct comparison with the levels used in the AD50 data book.

This program (AD50SIN.ASM) generates a sine-wave when the assembler variable makesine is set to 1. Otherwise it reads data from the ADC and copies it to the DAC. Filtering code can be inserted if desired. The variable mode16bit specifies whether the least significant bit of data written to the DAC is masked out.

The program should be edited using a text editor (such as MSDOS EDIT) and assembled as follows:

```
DSKPLASM AD50SIN
```

This generates a binary object file called AD50SIN.OBJ which is loaded into the DSK+ as follows:

```
LOADAPP -A AD50SIN.OBJ
```

```
.title "Sine-wave generator for AD50-EVM and C54x DSK+"
.width 80
.length 55
; Adapted from DSK+ example programs by John Walliker and Julian Daley
; June 1997

.mmregs
.setsect ".text",0x1800,0 ; these assembler directives specify
.setsect ".data",0x0200,1 ; the absolute addresses of different
.setsect "vectors",0x0180,0 ; sections of the assembly code

makesine .set 1
model6bit .set 0
sinestepsize .set 1
sinetablesize .set 20 ; for 1kHz at 20 ksps

; The vectors in this table can be configured for processing external and
; internal software interrupts. The DSKplus debugger uses four interrupt
; vectors. These are RESET, TRAP2, INT2, and HPIINT.
; * DO NOT MODIFY THESE FOUR VECTORS IF YOU PLAN TO USE THE DEBUGGER *

; All other vector locations are free to use. When programming always be sure
; the HPIINT bit is unmasked (IMR=200h) to allow the communications kernel and
; host PC interact. INT2 should normally be masked (IMR(bit 2) = 0) so that the
; DSP will not interrupt itself during a HINT. HINT is tied to INT2 externally.

.reset goto #80h ;00; RESET * DO NOT MODIFY IF USING DEBUGGER *
.nop
.nop
nmi return_enable ;04; non-maskable external interrupt
    nop
    nop
    nop

trap2 goto #88h ;08; trap2 * DO NOT MODIFY IF USING DEBUGGER *
    nop
    nop
    .space 52*16 ;0C-3F: vectors for software interrupts 18-30
int0 return_enable ;40; external interrupt int0
    nop
    nop
    nop

int1 return_enable ;44; external interrupt int1
    nop
    nop
    nop

int2 return_enable ;48; external interrupt int2
    nop
    nop
    nop

tint return_enable ;4C; internal timer interrupt
    nop
    nop
    nop

brint return_enable ;50; BSP receive interrupt
    nop
    nop
    nop

bxint return_enable ;54; BSP transmit interrupt
    nop
    nop
    nop

trint dgoto receive ;58; TDM receive interrupt
    nop
    nop

txint return_enable ;5C; TDM transmit interrupt
    nop
    nop

int3 return_enable ;60; external interrupt int3
    nop
    nop
    nop

hpiint goto #0e4h ;64; HPIint * DO NOT MODIFY IF USING DEBUGGER *
    nop
    nop
    .space 24*16 ;68-7F; reserved area
    .data
sineindex .word 0 ; index to sine-wave table
sinevalue .word 0 ; workspace for sine-wave generation
Sine-Wave Generator and Loopback Program

initwork .word 0 ; workspace for initialization

.text
start: intm = 1 ; disable all interrupts

pmst = #01a0h ; Interrupt pointer maps vectors to address 180h
sp = #0ffah ; stack pointer located in Communications Kernel
imr = #240h ; unmask TDM RINT and HPIINT(host port interface)
OVM = 0 ; OVM = 0 (do not clip overflow)
ASM = #0 ; ASM = 0 (shift=0)
DP = #0

AC01INIT:

; initialize serial port and ad50-evm on-board logic
arl = #initwork ; workspace for storing value to be output to ad50-evm
*arl = #0010b ; switch on one AD50 and hold it in reset
port(0) = *arl ; write the data to i/o port 0
repeat(#23) ; hold reset low for at least 6 mclk cycles
nop ; (4 dsp clock cycles == 1 mclk cycle)
tcr = #10h ; stop timer
imr = #240h ; unmask RXINT and HPIINT
tspc = #0008h ; stop TDM serial port
tdxr = #0h ; send 0 as first xmit word
tspc = #00c8h ; reset and start TDM serial port
*arl = #0011b ; bring ad50 out of reset with one channel powered up
port(0) = *arl

; initialize ad50 registers
a = #0000010010010000b ; register 4
call ad50init ; change fsamp to 20kHz
a = #0000001011000000b ; register 4
call ad50init ; light led
; a = #0000000100000001b ; select 16 bit dac mode
; call ad50init ; this must be the last command

intm = 0 ; enable all interrupts

WAIT idle(1) ; idle waiting for interrupts
nop
goto WAIT

;-------------------------------------------------------------------------------
; ad50 init routines here

ad50init
ifr = #080h ; clear flag from IFR
tdxr = #01h ; request secondary when AD50 starts
call waitfortdx ; wait for primary to xmit
tdxr = a ; send register value to serial port
ifr = #080h ; clear flag from IFR
call waitfortdx ; wait for secondary to xmit
tdxr = #0h ; send neutral state in case last init
ifr = #080h ; clear flag from IFR
call waitfortdx ; wait for neutral state to xmit
Setting Up the AD50-EVM

Sine-Wave Generator and Loopback Program

:return: return from subroutine

:waitfortdx a robust way of waiting for a tdm port

:b = mmr(ifr) transmission to complete

:b &= #10000000b

:nop

:nop

:if(BEQ) goto waitfortdx

:return

;--------------------------------------------------------------------------------

:receive

; save context if necessary

:B = mmr(TRCV) always read serial port to avoid buffer overflow

; which would stop interrupts from being generated

:.if makesine = 0 don’t make a sinewave – just echo data from adc to dac

; put code for filtering etc here

:.else make a sinewave on LEFT channel

:AR1 = #sineindex

:AR2 = #sinevalue

:A = *AR1

:A += #sinestepsize

:B = A

:A -= #sinetablesize

:nop ; two nops to allow for pipeline delay

:nop

:if (ALT) execute(1) ; check whether index points outside table

:A = B

:*AR1 = A

:A += #sinetable20_3dB

:*AR2 = prog(A) ; get sinewave value from table in program memory

:B = *AR2 ; (could use sine table in rom for this)

:.endif

:.if mode16bit = 0

:B &= #1111111111111110b ; mask least significant bit if in 15 bit mode to

:endif ; prevent inadvertent secondary communication requests

:mmr(TDXR) = B

; restore context if necessary

:return_enable ; return to waiting loop

;--------------------------------------------------------------------------------

:transmit: ; Transmit and receive interrupts are inherently

; synchronized so there is no point in using both

:return_enable

;--------------------------------------------------------------------------------

:sinetable20_3dB ; -3dB re clipping, 1kHz @ 20 ksps, 20 samples/cycle

:.word 0, 7160, 13619, 18745, 22036, 23170, 22036, 18745, 13619, 7160

:.word 0, -7160, -13619, -18745, -22036, -23170, -22036, -18745, -13619, -7160

:sinetable8_3dB ; -3dB re clipping, 1kHz @ 8 ksps, 8 samples/cycle

:.word 0, 16384, 23170, 16384, 0, -16384, -23170, -16384

:end
Chapter 4

Results Obtained With AD50 EVM

This chapter describes the results possible with the AD50 EVM. Actual measurements are presented.

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4.1 ADC Results

To evaluate the performance of the ADC channel of the AD50-EVM, FFT measurements were performed.

Figure 4–1. AD50-EVM ADC Distortion Measurement at 8 ksps

Figure 4–1 shows an FFT plot obtained from the ADC of the AD50-EVM. The input signal was a 1.2-kHz sine wave at -3 dB relative to maximum input. A digital oscillator using a 20 bit DAC was used to produce the test signal. By adding up the energy in each frequency bin within (a) the signal, (b) the harmonics and (c) the rest of the noise floor, figures for SNR, SDR and SINAD can be calculated. The figures were 82.9 dB SNR, 79.4 dB SDR and 77.8 dB SINAD. These figures have not been compensated for noise and distortion components in the oscillator output.
Figure 4–2. AD50-EVM ADC Distortion Measurement at 20 ksp/s

Figure 4–2 shows an FFT plot obtained from the ADC while sampling at 20 ksp/s. The performance figures were 82.7 dB SNR, 81.0 dB SDR and 78.7 dB SINAD.
4.2 DAC Results

The DAC was characterized using a sine-wave table lookup program running on a TMS320C5x DSK+ coupled to an AD50-EVM.

Figure 4–3. AD50-EVM DAC Distortion Measurement at 8 ksps

Figure 4–3 shows an FFT plot of the output from the AD50-EVM. This was measured using a Texas Instruments AD55-EVM acquisition system running at 40 ksps coupled to a DSP development system in a PC.

The signal to noise plus distortion ratio (SINAD) measured in the pass-band (excluding dc) was 80.5 dB. The SNR was 83.6 dB and the SDR was 83.4 dB. These measurements have not been adjusted for the noise and distortion of the acquisition system.

The rise in the noise floor above 4 kHz is due to the digital noise-shaping filter in the AD50 DAC, which decreases pass-band noise at the expense of increased high frequency noise. This is partially attenuated by the output filter on the AD50-EVM but continues to rise to a plateau at approximately –100 dB. This filter is optimized for use at 20 kbps and therefore the out of band noise is worse than at higher sampling frequencies.
Similar measurements were made at 20 ksps. The results were 78.6 dB SNR, 84.8 dB SDR and 77.6 dB SINAD. Again this has not been compensated for the characteristics of the acquisition system.

The spike at 10 kHz is caused by breakthrough from the negative supply generator. The larger peaks above 16 kHz are images that have not been completely removed by the reconstruction filters. These are well outside the passband of the DAC.

All the above measurements were made using a well-regulated low noise 5 V power supply because the power supply shipped with the DSK+ was found to introduce extra noise.
Installing the AD50-EVM

This chapter discusses the configuration jumper settings for the AD50 EVM, along with guidelines for connecting it to the DSK+. The parts list, schematics, gerber plots, and board outline drawings are included.

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A.1 Board Outline Drawing With Jumper Locations on the AD50 EVM

Figure A–1 shows a diagram of the AD50-EVM board showing locations of connectors and jumpers etc.

Figure A–1. Diagram of the AD50-EVM

Table A–1. Jumper Positions

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<th>Jumper</th>
<th>OPEN</th>
<th>CLOSED</th>
<th>DEFAULT</th>
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<tr>
<td>L_DC</td>
<td>Left channel ac input coupling</td>
<td>Left channel dc input coupling</td>
<td>OPEN</td>
</tr>
<tr>
<td>R_DC</td>
<td>Right channel ac input coupling</td>
<td>Right channel dc input coupling</td>
<td>OPEN</td>
</tr>
<tr>
<td>L_GAIN</td>
<td>Left channel 0 dB input gain</td>
<td>Left channel 20 dB input gain</td>
<td>OPEN</td>
</tr>
<tr>
<td>R_GAIN</td>
<td>Right channel 0 dB input gain</td>
<td>Right channel 20 dB input gain</td>
<td>OPEN</td>
</tr>
<tr>
<td>PRI/SEC</td>
<td>Primary board</td>
<td>Secondary board</td>
<td>OPEN</td>
</tr>
</tbody>
</table>

The connectors marked SERIAL and CONTROL are for use in interfacing the AD50-EVM to systems other than the DSK+. The pinouts are shown in Table A–2.

If two AD50-EVMs are used the serial connection should be made to the primary board.
Table A–2. Control and Serial Connections

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<tr>
<th>Pin</th>
<th>Control</th>
<th>Serial</th>
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<td>1</td>
<td>GROUND</td>
<td>GROUND</td>
</tr>
<tr>
<td>2</td>
<td>PRI/SEC</td>
<td>SERIAL DATA IN</td>
</tr>
<tr>
<td>3</td>
<td>GROUND</td>
<td>GROUND</td>
</tr>
<tr>
<td>4</td>
<td>RIGHT ALTDATA</td>
<td>SERIAL DATA OUT</td>
</tr>
<tr>
<td>5</td>
<td>GROUND</td>
<td>GROUND</td>
</tr>
<tr>
<td>6</td>
<td>RIGHT PWRDN</td>
<td>LEFT FRAME SYNC</td>
</tr>
<tr>
<td>7</td>
<td>RESET</td>
<td>GROUND</td>
</tr>
<tr>
<td>8</td>
<td>LEFT PWRDN</td>
<td>SERIAL CLOCK</td>
</tr>
<tr>
<td>9</td>
<td>GROUND</td>
<td>GROUND</td>
</tr>
<tr>
<td>10</td>
<td>FC</td>
<td>RESET</td>
</tr>
<tr>
<td>11</td>
<td>GROUND</td>
<td></td>
</tr>
<tr>
<td>12</td>
<td>RIGHT FLAG</td>
<td></td>
</tr>
<tr>
<td>13</td>
<td>GROUND</td>
<td></td>
</tr>
<tr>
<td>14</td>
<td>LEFT FLAG</td>
<td></td>
</tr>
<tr>
<td>15</td>
<td>GROUND</td>
<td></td>
</tr>
<tr>
<td>16</td>
<td>LEFT ALTDATA</td>
<td></td>
</tr>
</tbody>
</table>

There are 4 sets of connectors that provide direct access to the analog I/O pins of the AD50. The pinouts for these are shown in Table A–3.

Table A–3. Analog Input/Output Connectors

<table>
<thead>
<tr>
<th>Label</th>
<th>Aux ip</th>
<th>Diff-op</th>
<th>Mon out</th>
</tr>
</thead>
<tbody>
<tr>
<td>Left</td>
<td>L_AUXIN</td>
<td>L_OP</td>
<td>L_MON</td>
</tr>
<tr>
<td>Right</td>
<td>R_AUXIN</td>
<td>R_OP</td>
<td>R_MON</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Pin</th>
<th>Aux ip</th>
<th>Diff-op</th>
<th>Mon out</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>AUXP</td>
<td>OUTP</td>
<td>MONOUT</td>
</tr>
<tr>
<td>2</td>
<td>1.25 V</td>
<td>GROUND</td>
<td>GROUND</td>
</tr>
<tr>
<td>3</td>
<td>AUXM</td>
<td>OUTM</td>
<td></td>
</tr>
<tr>
<td>4</td>
<td>GROUND</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

There is a set of pins labeled PSU provided for connecting power supplies. The locally generated –5 V supply can be accessed via this connector.
A.2 Connecting the AD50-EVM to the DSK+

The two boards are stacked together using plug and socket connectors. Four pairs of connectors are needed, in JP2, JP3, JP4, and JP5.

The boards can be stacked either way round, although a slightly wider spacing is needed if the AD50-EVM is on top of the DSK+, due to the height of the audio jack connectors used on the DSK+. Suitable connectors that will allow the boards to be stacked either way with a spacing of approximately 13 mm, can be obtained from the following supplier.

<table>
<thead>
<tr>
<th>Manufacturer</th>
<th>Plug</th>
<th>Socket</th>
</tr>
</thead>
<tbody>
<tr>
<td>Samtec</td>
<td>TSW-112-14-T-T</td>
<td>SSW-112-21-G-T</td>
</tr>
<tr>
<td>Tel: 1-800-SAMTEC-9 (USA) +44 1236 739292 (UK)</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

In order to ensure that the connectors are properly aligned it is suggested that the plugs and sockets be mated before being soldered to the boards. To allow development work with the AD50-EVM it is suggested that the sockets be soldered to the bottom of the AD50-EVM and the pins to the top of the DSK+. This puts the AD50-EVM on top of the DSK+ and provides good access to the signals on the AD50-EVM.

If two AD50-EVM boards are being used in conjunction with a DSK+, then a different socket, with pass through pins, should be used on the middle board. The Samtec part number is SSQ-112-24-G-T.

A.3 Parts List

<table>
<thead>
<tr>
<th>Silk Screen Name</th>
<th>Component</th>
<th>Silk Screen Name</th>
<th>Component</th>
</tr>
</thead>
<tbody>
<tr>
<td>LEFT</td>
<td>TLC320AD50CDW</td>
<td>OA6</td>
<td>TLC2272ACD</td>
</tr>
<tr>
<td>RIGHT</td>
<td>TLC320AD50CDW</td>
<td>Q1</td>
<td>BC848 SOT23</td>
</tr>
<tr>
<td>GAL1</td>
<td>CMOS 22V10 25 ns, PLCC</td>
<td>Q2</td>
<td>BC848 SOT23</td>
</tr>
<tr>
<td>INV</td>
<td>ICL7660 8 pin SO</td>
<td>10–24 MHz</td>
<td>10.24 MHz 8-pin format TTL or CMOS oscillator module</td>
</tr>
<tr>
<td>ON</td>
<td>SM LED 2 mm × 1.25 mm green</td>
<td>R1</td>
<td>22K (all resistors 1% 0603)</td>
</tr>
<tr>
<td>L</td>
<td>SM LED 2 mm × 1.25 mm red</td>
<td>R2</td>
<td>(2.2 k) 2K2†</td>
</tr>
<tr>
<td>R</td>
<td>SM LED 2 mm × 1.25 mm red</td>
<td>R3</td>
<td>22K</td>
</tr>
<tr>
<td>OA1</td>
<td>TLC2272ACD</td>
<td>R4</td>
<td>10K</td>
</tr>
<tr>
<td>OA2</td>
<td>TLC2272ACD</td>
<td>R5</td>
<td>10K</td>
</tr>
<tr>
<td>OA3</td>
<td>TLC2272ACD</td>
<td>R6</td>
<td>10K</td>
</tr>
<tr>
<td>OA4</td>
<td>TLC2272ACD</td>
<td>R7</td>
<td>10K</td>
</tr>
<tr>
<td>OA5</td>
<td>TLC2272ACD</td>
<td>R8</td>
<td>10K</td>
</tr>
</tbody>
</table>

† British notation used in schematics A–4.
<table>
<thead>
<tr>
<th>Silk Screen Name</th>
<th>Component</th>
<th>Silk Screen Name</th>
<th>Component</th>
</tr>
</thead>
<tbody>
<tr>
<td>R9</td>
<td>10K</td>
<td>R58</td>
<td>(47) 47R†</td>
</tr>
<tr>
<td>R10</td>
<td>(4.7 k) 4K7†</td>
<td>R61</td>
<td>10K</td>
</tr>
<tr>
<td>R11</td>
<td>(4.7 k) 4K7†</td>
<td>R62</td>
<td>10K</td>
</tr>
<tr>
<td>R12</td>
<td>(4.7 k) 4K7†</td>
<td>R63</td>
<td>10K</td>
</tr>
<tr>
<td>R13</td>
<td>(4.7 k) 4K7†</td>
<td>R64</td>
<td>10K</td>
</tr>
<tr>
<td>R14</td>
<td>56K</td>
<td>R65</td>
<td>12K</td>
</tr>
<tr>
<td>R15</td>
<td>39K</td>
<td>R66</td>
<td>18K</td>
</tr>
<tr>
<td>R16</td>
<td>0 Ω link</td>
<td>R67</td>
<td>12K</td>
</tr>
<tr>
<td>R21</td>
<td>22K</td>
<td>R68</td>
<td>47R</td>
</tr>
<tr>
<td>R22</td>
<td>(2.2 k)2K2†</td>
<td>R71</td>
<td>(1.8 k) 1K8†</td>
</tr>
<tr>
<td>R23</td>
<td>22K</td>
<td>R77</td>
<td>10k</td>
</tr>
<tr>
<td>R24</td>
<td>10K</td>
<td>R78</td>
<td>(4.7 k) 4K7†</td>
</tr>
<tr>
<td>R25</td>
<td>10K</td>
<td>R79</td>
<td>(4.7 k) 4K7†</td>
</tr>
<tr>
<td>R26</td>
<td>10K</td>
<td>R80</td>
<td>(4.7 k) 4K7†</td>
</tr>
<tr>
<td>R27</td>
<td>10K</td>
<td>R81</td>
<td>10k</td>
</tr>
<tr>
<td>R28</td>
<td>10K</td>
<td>R82</td>
<td>10k</td>
</tr>
<tr>
<td>R29</td>
<td>10K</td>
<td>R83</td>
<td>(470) 470R†</td>
</tr>
<tr>
<td>R30</td>
<td>(4.7 k) 4K7†</td>
<td>R84</td>
<td>(470) 470R†</td>
</tr>
<tr>
<td>R31</td>
<td>(4.7 k) 4K7†</td>
<td>R85</td>
<td>10k</td>
</tr>
<tr>
<td>R32</td>
<td>(4.7 k) 4K7†</td>
<td>R86</td>
<td>10k</td>
</tr>
<tr>
<td>R33</td>
<td>(4.7 k) 4K7†</td>
<td>R87</td>
<td>10k</td>
</tr>
<tr>
<td>R36</td>
<td>0 Ω link</td>
<td>R88</td>
<td>10k</td>
</tr>
<tr>
<td>R51</td>
<td>10K</td>
<td>R89</td>
<td>10k</td>
</tr>
<tr>
<td>R52</td>
<td>10K</td>
<td>R90</td>
<td>10k</td>
</tr>
<tr>
<td>R53</td>
<td>10K</td>
<td>R91</td>
<td>10k</td>
</tr>
<tr>
<td>R54</td>
<td>10K</td>
<td>R92</td>
<td>10k</td>
</tr>
<tr>
<td>R55</td>
<td>12K</td>
<td>R93</td>
<td>10k</td>
</tr>
<tr>
<td>R56</td>
<td>18K</td>
<td>R94</td>
<td>10k</td>
</tr>
<tr>
<td>R57</td>
<td>12K</td>
<td>R95</td>
<td>10k</td>
</tr>
<tr>
<td>R96</td>
<td>10k</td>
<td>C30</td>
<td>100 nF X7R 0805</td>
</tr>
<tr>
<td>R97</td>
<td>10k</td>
<td>C31</td>
<td>100 nF X7R 0805</td>
</tr>
<tr>
<td>R98</td>
<td>10k</td>
<td>C33</td>
<td>100 nF X7R 0805</td>
</tr>
<tr>
<td>R99</td>
<td>10k</td>
<td>C34</td>
<td>100 nF X7R 0805</td>
</tr>
</tbody>
</table>

† British notation used in schematics A–4.
<table>
<thead>
<tr>
<th>Silk Screen Name</th>
<th>Component</th>
<th>Silk Screen Name</th>
<th>Component</th>
</tr>
</thead>
<tbody>
<tr>
<td>C1</td>
<td>4.7 μF 10 V size A tantalum</td>
<td>C35</td>
<td>100 nF X7R 0805</td>
</tr>
<tr>
<td>C2</td>
<td>4.7 μF 10 V size A tantalum</td>
<td>C36</td>
<td>100 nF X7R 0805</td>
</tr>
<tr>
<td>C3</td>
<td>470 pF COG/NPO 0603</td>
<td>C37</td>
<td>100 nF X7R 0805</td>
</tr>
<tr>
<td>C4</td>
<td>470 pF COG/NPO 0603</td>
<td>C51</td>
<td>2.2 nF COG/NPO 0805</td>
</tr>
<tr>
<td>C5</td>
<td>470 pF COG/NPO 0603</td>
<td>C52</td>
<td>220 pF COG/NPO 0603</td>
</tr>
<tr>
<td>C6</td>
<td>470 pF COG/NPO 0603</td>
<td>C53</td>
<td>100 nF X7R 0805</td>
</tr>
<tr>
<td>C7</td>
<td>470 pF COG/NPO 0603</td>
<td>C54</td>
<td>100 nF X7R 0805</td>
</tr>
<tr>
<td>C8</td>
<td>470 pF COG/NPO 0603</td>
<td>C61</td>
<td>2.2 nF COG/NPO 0805</td>
</tr>
<tr>
<td>C9</td>
<td>470 pF COG/NPO 0603</td>
<td>C62</td>
<td>220 pF COG/NPO 0603</td>
</tr>
<tr>
<td>C10</td>
<td>100 nF X7R 0805</td>
<td>C63</td>
<td>100 nF X7R 0805</td>
</tr>
<tr>
<td>C11</td>
<td>100 nF X7R 0805</td>
<td>C64</td>
<td>100 nF X7R 0805</td>
</tr>
<tr>
<td>C12</td>
<td>100 nF X7R 0805</td>
<td>C71</td>
<td>22 pF COG/NPO 0603</td>
</tr>
<tr>
<td>C13</td>
<td>100 nF X7R 0805</td>
<td>C72</td>
<td>100 μF 10 V size D tantalum</td>
</tr>
<tr>
<td>C14</td>
<td>100 nF X7R 0805</td>
<td>C73</td>
<td>100 μF 10 V size D tantalum</td>
</tr>
<tr>
<td>C15</td>
<td>100 nF X7R 0805</td>
<td>C74</td>
<td>100 nF X7R 0805</td>
</tr>
<tr>
<td>C16</td>
<td>100 nF X7R 0805</td>
<td>C75</td>
<td>100 nF X7R 0805</td>
</tr>
<tr>
<td>C17</td>
<td>100 nF X7R 0805</td>
<td>C76</td>
<td>100 μF 10 V size D tantalum</td>
</tr>
<tr>
<td>C21</td>
<td>4.7 μF 10 V size A tantalum</td>
<td>C81</td>
<td>100 nF X7R 0805</td>
</tr>
<tr>
<td>C22</td>
<td>4.7 μF 10 V size A tantalum</td>
<td>C82</td>
<td>100 nF X7R 0805</td>
</tr>
<tr>
<td>C23</td>
<td>470 pF COG/NPO 0603</td>
<td>INPUT</td>
<td>Marushin MJ156L 3.5 mm stereo pcb jack socket</td>
</tr>
<tr>
<td>C24</td>
<td>470 pF COG/NPO 0603</td>
<td>OUTPUT</td>
<td>Marushin MJ156L 3.5 mm stereo pcb jack socket</td>
</tr>
<tr>
<td>C25</td>
<td>470 pF COG/NPO 0603</td>
<td>PRI/SEC</td>
<td>Right angle header</td>
</tr>
<tr>
<td>C26</td>
<td>470 pF COG/NPO 0603</td>
<td>L_DC</td>
<td>Right angle header</td>
</tr>
<tr>
<td>C27</td>
<td>470 pF COG/NPO 0603</td>
<td>L_GAIN</td>
<td>Right angle header</td>
</tr>
<tr>
<td>C28</td>
<td>470 pF COG/NPO 0603</td>
<td>R_DC</td>
<td>Right angle header</td>
</tr>
<tr>
<td>C29</td>
<td>470 pF COG/NPO 0603</td>
<td>R_GAIN</td>
<td>Right angle header</td>
</tr>
</tbody>
</table>
Installing the AD50-EVM
C5/C25 adjacent to C8/C28
The components between C4 and C7 are R10, R11, C6, C9, R13, R12

Dimensions in thou
4.75
4.445
R83, R84, R71 adjacent to LEDs

AD50-EVM
JP3
JP5
JP4
JP2
JP1

R58 R68 C73
R2 R23 R21 DA3 DA6
R5 R9 R55 R52
R3 R1 R6 R7 R57 R56 R53 R54
C3 C1 C2 C11 C10 C12
R8 R9 R16 R94 R99 R92 R91
R80 R83 R81 R88 R85 R87 R89 R96

C51 C31 C32 C13 C14 C15
C18 C17 C16 L_VD
C8
GAL1

18-24 MHZ

Top Side Silkscreen
Ground plane
Installing the AD50-EVM

Power plane
Component positions
The 22V10 programmable logic device incorporates all the logic needed by the AD50-EVM when used in stand-alone mode or in conjunction with the DSK+. Note that in many applications little or no logic is required for interfacing the AD50 to a DSP system.

When used in stand-alone mode, most of the 22V10 outputs are high impedance to allow an external device complete freedom of control. This is controlled by the state of the DSK_PRESENT signal, which is pulled low when the AD50-EVM is connected to a DSK+. Otherwise a pull-up resistor makes it default to stand-alone mode.

One or two AD50-EVM boards can provide 1, 2, 3, or 4 simultaneously sampled channels. When two boards are coupled together the primary/secondary jumper controls the daisy-chaining of serial frame sync pulses and controls the way that data from the DSK+ is interpreted.

Designers familiar with the AC01 should note that the power-down behavior of the AD50 serial interface is different. The AC01 clock and frame sync outputs are always high impedance in power-down mode. The AD50 clock and frame sync outputs are driven high in power-down master mode, but are high impedance in power-down slave mode.

<table>
<thead>
<tr>
<th>Topic</th>
<th>Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>B.1 DSK+ Mode</td>
<td>B-2</td>
</tr>
<tr>
<td>B.2 Stand-Alone Mode</td>
<td>B-2</td>
</tr>
<tr>
<td>B.3 Logic Compiler Listing</td>
<td>B-2</td>
</tr>
</tbody>
</table>
B.1 DSK+ Mode

Address bits 0, 14, and 15 are decoded in conjunction with the I/O space select and the write enable from the C542 on the DSK+. Signals on the data bus are latched on the rising edge of the iostrobe signal if a valid write address has been decoded. Bits 14 and 15 are decoded so as to leave 48 k words of contiguous I/O address space free for other hardware that the user might wish to interface to the system. Bit 0 selects one of two control registers implemented with D latches within the 22V10 device. The first register is used to control the device reset and powerdown inputs for both codecs on each board and to control the powerdown signal to the TLC320AC01 codec on the DSK+ board. The second register just controls the FC signal. This is to simplify programming of interleaved secondary communications when the DAC is in 16 bit mode. When two AD50-EVM boards are used a jumper must be fitted to the primary/secondary header of one of the boards. The board with the jumper becomes the secondary one and has channels 3 and 4.

The 22V10 must have a propagation delay of 25 ns or less, assuming that one or more I/O wait states are programmed in the TMS320C542. Zero wait state operation would have negligible benefit and would need a much faster and more costly device.

The AD50s are reset either when the DSK+ is reset or when a 0 is programmed into bit 0 of address 0. Reset must be held low for at least 6 master clock cycles. The default configuration at reset is for all the AD50s to be powered down and for the AC01 on the DSK+ to be operating normally. This means that DSK+ will continue to operate normally until the 22V10 is programmed for AD50 operation.

B.2 Stand-Alone Mode

In stand-alone mode all the logic except that used to link the serial interface is 3-stated. This allows maximum flexibility to control the AD50s via the CONTROL and SERIAL cable headers.

Table B–1. AD50-EVM Registers

<table>
<thead>
<tr>
<th>I/O Address</th>
<th>Bits 15–4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>0</td>
<td>Hold AD50 in reset</td>
</tr>
<tr>
<td></td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>1</td>
<td>Release AD50 from reset</td>
</tr>
<tr>
<td></td>
<td>X</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>X</td>
<td>AC01 on DSK+ active</td>
</tr>
<tr>
<td></td>
<td>X</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>X</td>
<td>1 AD50 active</td>
</tr>
<tr>
<td></td>
<td>X</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>X</td>
<td>2 AD50s active</td>
</tr>
<tr>
<td></td>
<td>X</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>X</td>
<td>3 AD50s active</td>
</tr>
<tr>
<td></td>
<td>X</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>4 AD50s active</td>
</tr>
<tr>
<td>1</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>0</td>
<td>FC pin low</td>
</tr>
<tr>
<td></td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>1</td>
<td>FC pin high</td>
</tr>
</tbody>
</table>
B.3 Logic Compiler Listing

TITLE AD50 programmable address decoder & control port
AUTHOR John Walliker and Julian Daley
COMPANY For Texas Instruments
DATE 14 March 1997, 6 April 1997
OPTIONS
EXPAND = ON
INVERSION = OFF ; to ensure correct power-up reset state
MINIMIZATION = ON
CHIP AD50LOGIC EP22V10FN
;PIN  1 NC0
PIN  2 iostrb ; rising edge of DSK+ iostrobe clocks the D latches
PIN  3 DSK_reset ; global reset from DSK+
PIN  4 d0
PIN  5 d1 ; data bus
PIN  6 d2
PIN  7 d3
;PIN  8 NC1
PIN  9 a15
PIN  10 a14 ; address bus
PIN  11 a0
PIN  12 rw ; read/write control signal
PIN  13 primary ; pulled low with jumper when secondary board
;PIN  14 GRND
;PIN  15 NC2
PIN  16 DSK_present ; pulled low when DSK+ is connected
PIN  17 sec_sfs ; secondary serial frame sync
PIN  18 dsk_sfs ; serial frame sync to DSK+
PIN  19 fc ; control signal for initiating secondary comms
PIN  20 AC01_pwrdown ; power down AC01 on DSK+
PIN  21 L_ms ; left master/slave select
;PIN  22 NC3
PIN  23 L_pwrdown ; left power down
PIN  24 R_pwrdown ; right power down
PIN  25 AD50_rst ; reset all AD50s
PIN  26 L_sfs ; left serial frame sync
PIN  27 R_sfsD ; right serial frame sync delayed
;PIN  28 VCC
STRING select0 '!/(a0 * /a14 * /a15 * /rw)' ; address decode
STRING select1 '!/(a0 * /a14 * /a15 * /rw)' ;
EQUATIONS
AD50_rst := select0 * d0 ; primary and secondary boards are driven
+ /select0 * AD50_rst
AD50_rst.clkf = iostrb
AD50_rst.rstf = /DSK_reset
AD50_rst.trst = /DSK_present ; 3-state if in standalone mode
fc := select1 * d0
+ /select1 * fc
fc.clkf = iostrb
fc.rstf = /DSK_reset
fc.trst = /DSK_present

/AC01_pwrdsn := select0 * (d3 + d2 + d1) ; NB The logic assembler MUST NOT invert
+ /select0 * /AC01_pwrdsn ; equations, otherwise power-up reset state will be incorrect!

AC01_pwrdsn.clkf = iostrb
AC01_pwrdsn.rstf = /DSK_reset
L_pwrdsn := select0 * /d3 * /d2 * d1 * primary
+ select0 * /d3 * d2 * /d1 * primary
+ select0 * /d3 * d2 * d1
+ select0 * d3 * /d2 * /d1
+ /select0 * L_pwrdsn

L_pwrdsn.clkf = iostrb
L_pwrdsn.rstf = /DSK_reset
L_pwrdsn.trst = /DSK_present
L_ms = primary * L_pwrdsn * /DSK_present
+ primary * DSK_present
R_pwrdsn := select0 * /d3 * d2 * /d1 * primary
+ select0 * /d3 * d2 * d1 * primary
+ select0 * d3 * /d2 * /d1
+ /select0 * R_pwrdsn

R_pwrdsn.clkf = iostrb
R_pwrdsn.rstf = /DSK_reset
R_pwrdsn.trst = /DSK_present

DSK_sfs := L_sfs
DSK_sfs.trst = primary * /AC01_pwrdsn
L_sfs := sec_sfs
L_sfs.trst = /primary
sec_sfs := R_sfsd
sec_sfs.trst = primary
Converting DSK+ Software for the AD50-EVM

The AD50-EVM has been designed to make conversion of DSK+ programs originally written for the AC01 analog interface circuit as easy as possible. The serial port used by the AC01 is shared with two or four AD50s fitted to one or two AD50-EVM boards, but logic ensures that only one type of AIC is used at a time.

For single channel operation the only essential changes to existing code relate to initialization. First, one AD50 must be switched on and held in reset for at least 6 master clock cycles. While it is held in reset the DSP serial port is initialized. The AD50 is then released from reset and the registers of the AD50 are programmed with the required clock division ratio, analog gains etc.

It is not essential to change the main program code for single channel operation, as the AD50 uses the same serial port as the AC01. However, the AC01 is a 14-bit device whereas the AD50 supports 15- or 16-bit operations. For best performance, therefore, masking operations that strip the two least significant bits from data sent to or from the AC01 should be modified to remove only bit 0 or removed completely, depending on whether 15- or 16-bit mode is to be used.

The following code shows how the oscilloscope program (OSCOPE) supplied with the DSK+ was modified to use one AD50 device instead of the AC01. OSCOPE provides an oscilloscope style display of the input waveform on the PC screen. This program is split up into several modules, with all the ADC initialization code in OSC_AC01.ASM.

The C54x algebraic assembly language is used. Code that has been added is shown in **bold**, while code that has been deleted is shown in *italics*. No changes were made to the C code running on the PC for transferring and displaying the acquired data, so this code is not listed below.
; ************************************************************************** *
; File: OSC_MAIN.ASM -> DSKplus Oscilloscope ASM file for the ’C54x DSKplus *
; Adapted to operate with AD50–EVM by John Walliker and Julian Daley *
; ************************************************************************** *
; .width 80
; .length 55
; .title "DSKplus Oscilloscope program"
; .mmregs
; .setsect ".text", 0x500
; .setsect "vectors",0x180
;
;Vectors
;
;***************************************************************************
; File: Osc_AC01.ASM -> AC01 Init Routine for the Oscilloscope application
; Adapted to operate with AD50–EVM by John Walliker and Julian Daley
;***************************************************************************
; .width 80
; .length 55

.start:
call AC01INIT
pmst = #01a0h ; set up iptr
sp = #0ffah ; init stack pointer.
ar2 = #1200h ; pointer to receive buffer.
*ar2+ = data(#0bh) ; store to rcv buffer
imr = #40h
intm = 0 ; ready to rcv int’s
.wait nop
.goto wait
;
; Receive Interrupt Routine
;
RINT:
b = trcv ; load acc b with input
b = #0FFFEh & b ; only strip out lsb for AD50
*ar2+ = data(#0bh) ; store to rcv buffer
tdxr = b ; transmit the data.
TC = (@ar2 == #01300h)
if (TC) goto restrt ; stop if rcv buffer is at 1300h

.return_enable
restrt
ar2 = #1200h ; set intm bit ...no int’s
hpic = #0ah ; flag host task completed
.return_enable
;
; Copy "osc_ac01.asm"
.end

;***************************************************************************
; File: Osc_AC01.ASM -> AC01 Init Routine for the Oscilloscope application
; Adapted to operate with AD50–EVM by John Walliker and Julian Daley
;***************************************************************************
; .width 80
; .length 55
* Certain AC01 registers can be initialized using a conditional assembly constant. By setting the constant REGISTER to the appropriate value, the assembler will either include initialization for certain registers or ignore register initialization.

* The constant REGISTER should be set to include the following AC01 register:

* REGISTER (binary) =

* 0000 0000 0000 0001 -> initialize Register 1 (A Register)
* 0000 0000 0000 0010 -> initialize Register 2 (B Register)
* 0000 0000 0000 0100 -> initialize Register 3 (A’ Register)
* 0000 0000 0000 1000 -> initialize Register 4 (Amplifier Gain-Select)
* 0000 0000 0001 0000 -> initialize Register 5 (Analog Configuration)
* 0000 0000 0010 0000 -> initialize Register 6 (Digital Configuration)
* 0000 0000 1000 0000 -> initialize Register 7 (Frame-Sync Delay)
* 0000 0000 1000 0000 -> initialize Register 8 (Frame-Sync number)

* For AD50-EVM

* 0000 0000 0000 0001 -> initialize Register 1
* 0000 0000 0000 0010 -> initialize Register 2
* 0000 0000 0000 0100 -> initialize Register 3
* 0000 0000 0000 1000 -> initialize Register 4

* Any combination of registers can be initialized by adding the binary number to the REGISTER constant. For example to initialize Registers 4 and 5, REGISTER = 18h. Upon assembly, only code for register 4 & 5 initialization is included in the AC01INIT module. When called the module will load the REG4 and REG5 values into internal AC01 registers.

* ;; REGISTER .set 09h ; Powerup default values:
* ;; REG1 .set 120h ;*A=32 112h
* ;; REG2 .set 212h ; 212h
* ;; REG3 .set 300h ; 300h
* ;; REG4 .set 409h ;* 405h
* ;; REG5 .set 501h ; 501h
* ;; REG6 .set 600h ; 600h
* ;; REG7 .set 700h ; 700h
* ;; REG8 .set 801h ; 801h

;For AD50-EVM:
REGISTER .set 1000b ; just program reg 4
REG1 .set 0000000100000000b ; default
REG2 .set 0000001000010000b ; select 16-bit ADC mode
REG3 .set 0000001100000000b ; default (no slaves)
REG4 .set 0000010010010000b ; 20ksa/s with 10.24MHz MCLK

; Note that once 16-bit DAC mode has been selected in reg 1, bit 0, other registers can only be configured by using FC to request secondary comms. On the
; AD50-EVM this can be achieved by writing 1 to i/o port 1, bit 0 before a primary
; communication and clearing port 1 afterwards.
; Once slave devices have been specified using reg 3, then a separate secondary
; communication (interleaved with normal communications) takes place for each
; device in turn.

AC01INIT:

ar2 = #1302h ; workspace for storing value to be output to AD50-EVM
*ar2 = #0010b ; switch on one AD50 and hold it in reset
port(0) = *ar2 ; write the data to i/o port 0
repeat(#50) ; hold reset low for at least 6 mclk cycles
nop ; ( 4 DSP clock cycles == 1 MCLK cycle )

;; xf = 0 ; reset ac01
intm = 1 ; disable all int service routines
tcr = #10h ; stop timer
tspc = #0008h ; stop TDM serial port
tdxr = #0h ; send 0 as first xmit word
tspc = #00c8h ; reset and start TDM serial port

;; xf = 1 ; release ac01 from reset

*ar2 = #0011b ; bring ad50 out of reset with 1 channel selected
port(0) = *ar2

; -- Register init’s ------------------------------------------
.eval REGISTER & 1h, SELECT ; if REG1 then include this source
.if SELECT = 1h
a = #REG1 ; load Acc A with REG1 value
call REQ2 ; Call REQ2 subroutine
.endif
.eval REGISTER & 2h, SELECT ; if REG2 then include this source
.if SELECT = 2h
a = #REG2
call REQ2
.endif
.eval REGISTER & 8h, SELECT ; if REG4 then include this source
.if SELECT = 8h
a = #REG4
call REQ2
.endif

; Note that AD50 register 3 must be programmed last!
.eval REGISTER & 4h, SELECT ; if REG3 then include this source
.if SELECT = 4h
a = #REG3
call REQ2
.endif

; Code for registers not used by AD50 deleted

return

REQ2
ifr = #080h ; clear flag from IFR
tdxr = #03h ; request secondary when AC01 starts
call waitfortdx ; wait for primary to xmit
tdxr = a ; send register value to serial port
ifr = #080h ; clear flag from IFR
call waitfortdx ; wait for secondary to xmit
tdxr = #0h ; send neutral state in case last init
ifr = #080h ; clear flag from IFR
call waitfortdx ; wait for neutral state to xmit
return ; return from subroutine

waitfortdx ; a robust way of waiting for a tdm port
b = mmr(ifr) ; transmission to complete
b &= #10000000b

nop
nop

if(BEQ) goto waitfortdx
return

;width   80
.length  55
.title "Vector Table"

reset  dgoto start ;00; RESET
IMR = #200h
nmi return_enable ;04; non-maskable external interrupt
nop
nop
nop
trap2  return_enable ;08; trap2
nop
nop
nop

.space 52*16 ;0C–3F: vectors for software interrupts 18–30
int0  return_enable ;40; external interrupt int0
nop
nop
nop
int1  return_enable ;44; external interrupt int1
nop
nop
nop
int2  return_enable ;48; external interrupt int2
nop
nop
nop
tint  return_enable ;4C; internal timer interrupt
nop
nop
nop
brint return_enable ;50; BSP receive interrupt
nop
nop
nop

bxint  return_enable ;54; BSP transmit interrupt	nop
nop
nop

trint  goto  RINT ;58; TDM receive interrupt	nop
nop

txint  return_enable ;5C; TDM transmit interrupt	nop
nop
nop

int3  return_enable ;60; external interrupt int3	nop
nop
nop

hpiint  return_enable ;64; HPIint	nop
nop
nop

.space  24*16 ;68-7F; reserved area