

LogiCORE™ FIFO Generator v3.2

User Guide

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Revision History

The following table shows the revision history for this document.

Date	Version	Revision
04/28/05	1.1	Initial Xilinx release.
8/31/05	2.0	Updated guide for release v2.2, added SP3 to ISEv7.1i, incorporated edits from engineering specific for this release, including FWFT, and Built-in FIFO flags, etc.
1/11/06	3.0	Updated for v2.3 release, ISE v8.1i.
7/13/06	4.0	Added Virtex-5 support, reorganized Chapter 5, added ISE v8.2i, version to 3.1
9/21/06	5.0	Core version updated to v3.2; support added for Spartan-3A.

Table of Contents

Table of Contents 3

Preface: About This Guide

Guide Contents	11
Additional Resources	12
Conventions	12
Typographical	12
Online Document	13

Chapter 1: Introduction

About the Core	15
Recommended Design Experience	15
Technical Support	15
Feedback	16
FIFO Generator	16
Document	16

Chapter 2: Core Overview

System Overview	17
Clock Implementation Operation	17
Built-in FIFO Support in Virtex-5	17
Built-in FIFO Support in Virtex-4	17
First-Word Fall-Through	17
Memory Types	18
Non-Symmetric Aspect Ratio	18
Core Configuration and Implementation	18
Independent Clocks: Block RAM and Distributed RAM	19
Independent Clocks: Virtex-5 and Virtex-4 Built-in FIFO	19
Common Clock: Block RAM, Distributed RAM, Shift Register	19
Common Clock: Virtex-5 and Virtex-4 Built-in FIFO	19
FIFO Generator Features	20
FIFO Interfaces	21
Interface Signals: FIFOs With Independent Clocks	21
Interface Signals: FIFOs with Common Clock	25

Chapter 3: Generating the Core

CORE Generator Graphical User Interface	29
FIFO Implementation	30
Component Name	30
FIFO Implementation	30
Common Clock (CLK), Block RAM	30
Common Clock (CLK), Distributed RAM	30

Common Clock (CLK), Shift Register	30
Common Clock (CLK), Built-in FIFO	31
Independent Clocks (RD_CLK, WR_CLK), Block RAM	31
Independent Clocks (RD_CLK, WR_CLK), Distributed RAM	31
Independent Clocks (RD_CLK, WR_CLK), Built-in FIFO	31
Performance Options and Data Port Parameters	31
Performance Options	32
Standard FIFO	32
First-word Fall-through FIFO	32
Data Port Parameters	32
Input Data Width	32
Input Depth	32
Output Data Width	32
Output Depth	32
Built-in FIFO Options	32
Optional Flags, Handshaking, and Initialization	33
Optional Flags	33
Almost Full Flag	33
Almost Empty Flag	33
Write Port Handshaking	33
Write Acknowledge	33
Overflow (Write Error)	33
Read Port Handshaking	34
Valid (Read Acknowledge)	34
Underflow (Read Error)	34
Initialization	34
Reset Pin	34
Programmable Flags	34
Programmable Flags	35
Programmable Full Type	35
Programmable Empty Type	35
Data Count and Reset	36
Data Count and Reset Options	36
Data Count	36
Resets	37
Summary	37

Chapter 4: Designing with the Core

General Design Guidelines	39
Know the Degree of Difficulty	39
Understand Signal Pipelining and Synchronization	39
Synchronization Considerations	39
Initializing the FIFO Generator	40
FIFO Implementations	41
Independent Clocks: Block RAM and Distributed RAM	41
Independent Clocks: Built-in FIFO	42
Common Clock: Built-in FIFO	43
Common Clock FIFO: Block RAM and Distributed RAM	44
Common Clock FIFO: Shift Registers	44
FIFO Usage and Control	45
Write Operation	45

ALMOST_FULL and FULL Flags	45
Example Operation	46
Read Operation	46
ALMOST_EMPTY and EMPTY Flags	46
Modes of Read Operation	47
Handshaking Flags	49
Write Acknowledge	49
Valid	49
Example Operation	50
Underflow	51
Overflow	51
Example Operation	51
Programmable Flags	52
Programmable Full	52
Programmable Empty	54
Data Counts	55
Read Data Count	56
Write Data Count	56
First-Word Fall-Through Data Count	57
Example Operation	57
Non-symmetric Aspect Ratios	58
Reset Behavior	60
Asynchronous Reset	60
Synchronous Reset	62

Chapter 5: Special Design Considerations

Resetting the FIFO	65
Continuous Clocks	65
Pessimistic Full and Empty	65
Programmable Full and Empty	66
Write Data Count and Read Data Count	66
Setup and Hold Time Violations	66

Chapter 6: Simulating Your Design

Simulation Models	67
-----------------------------	----

Appendix A: Performance Information

Resource Utilization and Performance	69
--	----

Appendix B: Core Parameters

FIFO Parameters	73
---------------------------	----

Schedule of Figures

Schedule of Figures	7
----------------------------------	---

Preface: About This Guide

Chapter 1: Introduction

Chapter 2: Core Overview

<i>Figure 2-1: FIFO with Independent Clocks: Interface Signals</i>	21
--	----

Chapter 3: Generating the Core

<i>Figure 3-1: Main FIFO Generator Screen</i>	30
<i>Figure 3-2: Performance Options and Data Port Parameters Screen</i>	31
<i>Figure 3-3: Optional Flags, Handshaking, and Initialization Options Screen</i>	33
<i>Figure 3-4: Programmable Flags Screen</i>	34
<i>Figure 3-5: Data Count and Reset Screen</i>	36
<i>Figure 3-6: Summary Screen</i>	37

Chapter 4: Designing with the Core

<i>Figure 4-1: FIFO with Independent Clocks: Write and Read Clock Domains</i>	40
<i>Figure 4-2: Functional Implementation of a FIFO with Independent Clock Domains</i> .	41
<i>Figure 4-3: Functional Implementation of Built-in FIFO</i>	43
<i>Figure 4-4: Functional Implementation of a Common Clock FIFO using Block RAM or Distributed RAM</i>	44
<i>Figure 4-5: Functional Implementation of a Common Clock FIFO using Shift Registers</i>	45
<i>Figure 4-6: Write Operation for a FIFO with Independent Clocks</i>	46
<i>Figure 4-7: Standard Read Operation for a FIFO with Independent Clocks</i>	48
<i>Figure 4-8: FWFT Read Operation for a FIFO with Independent Clocks</i>	48
<i>Figure 4-9: Write and Read Operation for a FIFO with Common Clocks</i>	49
<i>Figure 4-10: Handshaking Signals for a FIFO with Independent Clocks</i>	50
<i>Figure 4-11: Handshaking Signals for a FIFO with Common Clocks</i>	51
<i>Figure 4-12: Programmable Full Single Threshold: Threshold Set to 7</i>	53
<i>Figure 4-13: Programmable Full with Assert and Negate Thresholds: Assert Set to 10 and Negate Set to 7</i>	53
<i>Figure 4-14: Programmable Empty with Single Threshold: Threshold Set to 4</i>	54
<i>Figure 4-15: Programmable Empty with Assert and Negate Thresholds: Assert Set to 7 and Negate Set to 10</i>	55
<i>Figure 4-16: Write and Read Data Counts for FIFO with Independent Clocks</i>	57
<i>Figure 4-17: 1:4 Aspect Ratio: Data Ordering</i>	58

<i>Figure 4-18: 1:4 Aspect Ratio: Status Flag Behavior</i>	59
<i>Figure 4-19: 4:1 Aspect Ratio: Data Ordering</i>	59
<i>Figure 4-20: 4:1 Aspect Ratio: Status Flag Behavior</i>	60
<i>Figure 4-21: Synchronous Reset: FIFO with Independent Clock</i>	61
<i>Figure 4-22: Asynchronous Reset: FIFO with Common Clock</i>	61
<i>Figure 4-23: Synchronous Reset: FIFO with a Common Clock</i>	62

Schedule of Tables

Chapter 2: Core Overview

<i>Table 2-1: Memory Configuration Benefits</i>	18
<i>Table 2-2: FIFO Configurations</i>	18
<i>Table 2-3: FIFO Configurations Summary</i>	20
<i>Table 2-4: Reset Signal for FIFOs with Independent Clocks</i>	21
<i>Table 2-5: Write Interface Signals for FIFOs with Independent Clocks</i>	22
<i>Table 2-6: Read Interface Signals for FIFOs with Independent Clocks</i>	23
<i>Table 2-7: Interface Signals for FIFOs with a Common Clock</i>	25

Chapter 3: Generating the Core

Chapter 4: Designing with the Core

<i>Table 4-1: Interface Signals and Corresponding Clock Domains</i>	42
<i>Table 4-2: Interface Signals and Corresponding Clock Domains</i>	43
<i>Table 4-3: Implementation-Specific Support for First-Word Fall-Through</i>	47
<i>Table 4-4: Implementation-specific Support for Data Counts</i>	56
<i>Table 4-5: Implementation-specific Support for Non-symmetric Aspect Ratios</i>	58
<i>Table 4-6: FIFO Reset Values</i>	61
<i>Table 4-7: FIFO Reset and Power-up Values</i>	63

Appendix A: Performance Information

<i>Table A-1: Benchmarks: FIFO Configured without Optional Features</i>	69
<i>Table A-2: Benchmarks: FIFO Configured with Multiple Programmable Thresholds</i> ..	70
<i>Table A-3: Benchmarks: FIFO Configured with Virtex-5 FIFO36 Resources</i>	71
<i>Table A-4: Benchmarks: FIFO Configured with Virtex-4 FIFO16 Patch</i>	72

Appendix B: Core Parameters

<i>Table B-1: FIFO Parameter Table</i>	73
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About This Guide

The *LogicCORE™ FIFO Generator User Guide v3.2* describes the function and operation of the FIFO Generator, and contains information about designing, customizing, and implementing the core.

Guide Contents

The following chapters are included:

- [“Preface, About this Guide”](#) describes how the user guide is organized, the conventions used in the guide, and provides information about additional resources.
- [Chapter 1, “Introduction,”](#) describes the core and related information, including recommended design experience, additional resources, technical support, and submitting feedback to Xilinx.
- [Chapter 2, “Core Overview,”](#) describes the core configuration options and their interfaces.
- [Chapter 3, “Generating the Core,”](#) describes how to generate the core using the Xilinx CORE Generator GUI.
- [Chapter 4, “Designing with the Core,”](#) discusses how to use the core in a user application.
- [Chapter 5, “Special Design Considerations,”](#) discusses specific design features that must be considered when using with the core.
- [Chapter 6, “Simulating Your Design,”](#) provides instructions for simulating the design with either behavioral or structural simulation models.
- [Appendix A, “Performance Information,”](#) provides a summary of the core’s performance data.
- [Appendix B, “Core Parameters,”](#) provides a comprehensive list of the parameters set by in the CORE Generator GUI for the FIFO Generator.

Additional Resources

For additional information, go to www.xilinx.com/support. The following table lists some of the resources you can access from this website or by using the provided URLs.

Resource	Description/URL
Tutorials	Tutorials covering Xilinx design flows, from design entry to verification and debugging www.xilinx.com/support/techsup/tutorials/index.htm
Answer Browser	Database of Xilinx solution records www.xilinx.com/xlnx/xil_ans_browser.jsp
Data Sheets	Device-specific information on Xilinx device characteristics, including readback, boundary scan, configuration, length count, and debugging www.xilinx.com/xlnx/xweb/xil_publications_index.jsp
Problem Solvers	Interactive tools that allow you to troubleshoot your design issues www.xilinx.com/support/troubleshoot/psolvers.htm
Tech Tips	Latest news, design tips, and patch information for the Xilinx design environment www.xilinx.com/xlnx/xil_tt_home.jsp

Conventions

This document uses the following conventions. An example illustrates each convention.

Typographical

The following typographical conventions are used in this document:

Convention	Meaning or Use	Example
Courier font	Messages, prompts, and program files that the system displays and signal names	speed grade: - 100
Courier bold	Literal commands you enter in a syntactical statement	ngdbuild design_name
<i>Italic font</i>	Variables in a syntax statement for which you must supply values	See the <i>Development System Reference Guide</i> for more information.
	References to other manuals	See the <i>User Guide</i> for details.
	Emphasis in text	If a wire is drawn so that it overlaps the pin of a symbol, the two nets are <i>not</i> connected.
Dark Shading	Items that are not supported or reserved	This feature is not supported

Convention	Meaning or Use	Example
Square brackets []	An optional entry or parameter. However, in bus specifications, such as bus [7:0] , they are required.	ngdbuild [option_name] design_name
Braces { }	A list of items from which you must choose one or more	lowpwr = {on off}
Vertical bar	Separates items in a list of choices	lowpwr = {on off}
Vertical ellipsis . . .	Repetitive material that has been omitted	IOB #1: Name = QOUT' IOB #2: Name = CLKIN' . . .
Horizontal ellipsis ...	Omitted repetitive material	allow block block_name loc1 loc2 ... locn;
Notations	The prefix '0x' or the suffix 'h' indicate hexadecimal notation	A read of address 0x00112975 returned 45524943h.
	An '_n' means the signal is active low	usr_teof_n is active low.

Online Document

The following linking conventions are used in this document:

Convention	Meaning or Use	Example
Blue text	Cross-reference link to a location in the current document	See the section " Additional Resources " for details. See " Title Formats " in Chapter 1 for details.
Blue, underlined text	Hyperlink to a website (URL)	Go to http://www.xilinx.com for the latest speed files.

Introduction

The FIFO Generator core, a fully verified first-in first-out memory queue for use in any application requiring in-order storage and retrieval, enables high-performance and area-optimized designs. This core can be customized using the Xilinx CORE Generator™ system as a complete solution with control logic already implemented, including management of the read and write pointers and the generation of status flags.

This chapter introduces the FIFO Generator and provides related information, including recommended design experience, additional resources, technical support, and submitting feedback to Xilinx.

About the Core

The FIFO Generator is a Xilinx CORE Generator IP core, included in the latest IP Update on the Xilinx IP Center. The core is free of charge and no license is required. For detailed information about the core, see

www.xilinx.com/xlnx/xebiz/designResources/ip_product_details.jsp?key=FIFO_Generator.

Recommended Design Experience

The FIFO Generator is a fully verified solution, and can be used by all levels of design engineers.

Important! When implementing a FIFO with independent write and read clocks, special care must be taken to ensure the FIFO Generator is correctly used. [Chapter 4, “Designing with the Core,”](#) in this guide—specifically the section “[Synchronization Considerations](#),” [page 39](#)—has important information to help you ensure correct design configuration.

Similarly, asynchronous designs should also be aware that the Behavioral models are not cycle accurate across clock domains. See [Chapter 6, “Simulating Your Design,”](#) for details.

Technical Support

For technical support, visit www.support.xilinx.com/. Questions are routed to a team of engineers with FIFO Generator expertise.

Xilinx will provide technical support for use of this product as described in the *LogiCORE FIFO Generator User Guide*. Xilinx cannot guarantee timing, functionality, or support of this product for designs that do not follow these guidelines.

Feedback

Xilinx welcomes comments and suggestions about the FIFO Generator and the documentation supplied with the core.

FIFO Generator

For comments or suggestions about the FIFO Generator, please submit a WebCase from www.support.xilinx.com/. Be sure to include the following information:

- Product name
- Core version number
- Explanation of your comments

Document

For comments or suggestions about this document, please submit a WebCase from www.support.xilinx.com/. Be sure to include the following information:

- Document title
- Document number
- Page number(s) to which your comments refer
- Explanation of your comments

Core Overview

This chapter provides an overview of the FIFO Generator configuration options and their interfaces.

System Overview

Clock Implementation Operation

The FIFO Generator enables FIFOs to be configured with either independent or common clock domains for write and read operations. The independent clock configuration of the FIFO Generator enables the user to implement unique clock domains on the write and read ports. The FIFO Generator handles the synchronization between clock domains, placing no requirements on phase and frequency relationships between clocks. A common clock domain implementation optimizes the core for data buffering within a single clock domain.

Built-in FIFO Support in Virtex-5

The FIFO Generator supports the Virtex™-5 built-in FIFO modules, enabling large FIFOs to be created by cascading the built-in FIFOs in both width and depth. The core expands the capabilities of the built-in FIFOs by utilizing the FPGA fabric to create optional status flags not implemented in the built-in FIFO macro.

Built-in FIFO Support in Virtex-4

The FIFO Generator supports a single instantiation of the Virtex-4 built-in FIFO module. The core adds the required patch based upon estimated clock frequencies. This patch is implemented in fabric. See [Appendix A, “Performance Information”](#) for resource utilization estimates.

First-Word Fall-Through

The first-word fall-through (FWFT) feature provides the ability to look ahead to the next word available from the FIFO without having to issue a read operation. The FIFO accomplishes this by using output registers which are automatically loaded with data, when data appears in the FIFO. This causes the first word written to the FIFO to automatically appear on the data out bus (DOUT). Subsequent user read operations cause the output data to update with the next word, as long as data is available in the FIFO. The use of registers on the FIFO DOUT bus improves clock-to-output timing, and the FWFT functionality provides low-latency access to data. This is ideal for applications that require throttling, based on the contents of the data that are read.

See [Table 2-2](#) for FWFT availability. The use of this feature impacts the behavior of many other features, such as:

- Read operations (see [“First-Word Fall-Through FIFO Read Operation,”](#) page 48)
- Programmable empty (see [“Programmable Empty for First-Word Fall-Through,”](#) page 55)
- Data counts (see [“First-Word Fall-Through Data Count,”](#) page 57)

Memory Types

The FIFO Generator implements FIFOs built from block RAM, distributed RAM, shift registers, or the Virtex-4 and Virtex-5 built-in FIFOs. The core combines memory primitives in an optimal configuration based on the selected width and depth of the FIFO. [Table 2-1](#) provides best-use recommendations for specific design requirements.

Table 2-1: Memory Configuration Benefits

	Independent Clocks	Common Clock	Small Buffering	Medium-Large Buffering	High Performance	Minimal Resources
Built-in FIFO	✓	✓		✓	✓	✓
Block RAM	✓	✓		✓	✓	✓
Shift Register		✓	✓		✓	
Distributed RAM	✓	✓	✓		✓	

Non-Symmetric Aspect Ratio

The core supports generating FIFOs whose write and read ports have different widths, enabling automatic width conversion of the data width. Non-symmetric aspect ratios ranging from 1:8 to 8:1 are supported for the write and read port widths. This feature is available for FIFOs implemented with block RAM that are configured to have independent write and read clocks.

Core Configuration and Implementation

[Table 2-2](#) provides a summary of the supported memory and clock configurations.

Table 2-2: FIFO Configurations

Clock Domain	Memory Type	Supported Configuration	Non-symmetric Aspect Ratios	First-Word Fall-Through
Common	Block RAM	✓		
Common	Distributed RAM	✓		
Common	Shift Register	✓		
Common	Built-in FIFO ¹	✓		✓ ²
Independent	Block RAM	✓	4	✓

Table 2-2: FIFO Configurations

Clock Domain	Memory Type	Supported Configuration	Non-symmetric Aspect Ratios	First-Word Fall-Through
Independent	Distributed RAM	✓		✓
Independent	Built-in FIFO ¹	✓		✓ ²

1. The built-in FIFO primitive is only available in Virtex-5 and Virtex-4 architectures.

2. Only valid in Virtex-5 built-in FIFO primitives.

Independent Clocks: Block RAM and Distributed RAM

This implementation category allows the user to select block RAM or distributed RAM and supports independent clock domains for write and read data accesses. Operations in the read domain are synchronous to the read clock and operations in the write domain are synchronous to the write clock.

The feature set supported for this type of FIFO includes non-symmetric aspect ratios (different write and read port widths), status flags (full, almost full, empty, and almost empty), as well as programmable full and empty flags generated with user-defined thresholds. Optional read data count and write data count indicators provide the number of words in the FIFO relative to their respective clock domains. In addition, optional handshaking and error flags are available (write acknowledge, overflow, valid, and underflow).

Independent Clocks: Virtex-5 and Virtex-4 Built-in FIFO

This implementation category allows you to select the built-in FIFO that is available in the Virtex-5 and Virtex-4 architectures. Operations in the read domain are synchronous to the read clock and operations in the write domain are synchronous to the write clock.

The feature set supported for this configuration includes status flags (full and empty) and programmable full and empty flags generated with user-defined thresholds. In addition, optional handshaking and error flags are available (write acknowledge, overflow, valid, and underflow).

Common Clock: Block RAM, Distributed RAM, Shift Register

This implementation category allows the user to select block RAM, distributed RAM, or shift register and supports a common clock for write and read data accesses.

The feature set supported for this configuration includes status flags (full, almost full, empty, and almost empty) and programmable empty and full flags generated with user-defined thresholds. In addition, optional handshaking and error flags are supported (write acknowledge, overflow, valid, and underflow), and an optional data count provides the number of words in the FIFO. For block RAM and distributed RAM implementations, optional synchronous or asynchronous reset pin is available.

Common Clock: Virtex-5 and Virtex-4 Built-in FIFO

This implementation category allows you to select the built-in FIFO that is available in the Virtex-5 and Virtex-4 architectures, and supports a common clock for write and read data accesses.

The feature set supported for this configuration includes status flags (full and empty) and optional programmable full and empty flags with user-defined thresholds. In addition, optional handshaking and error flags are available (write acknowledge, overflow, valid, and underflow).

FIFO Generator Features

Table 2-3 summarizes the FIFO Generator features supported for each clock configuration and memory type.

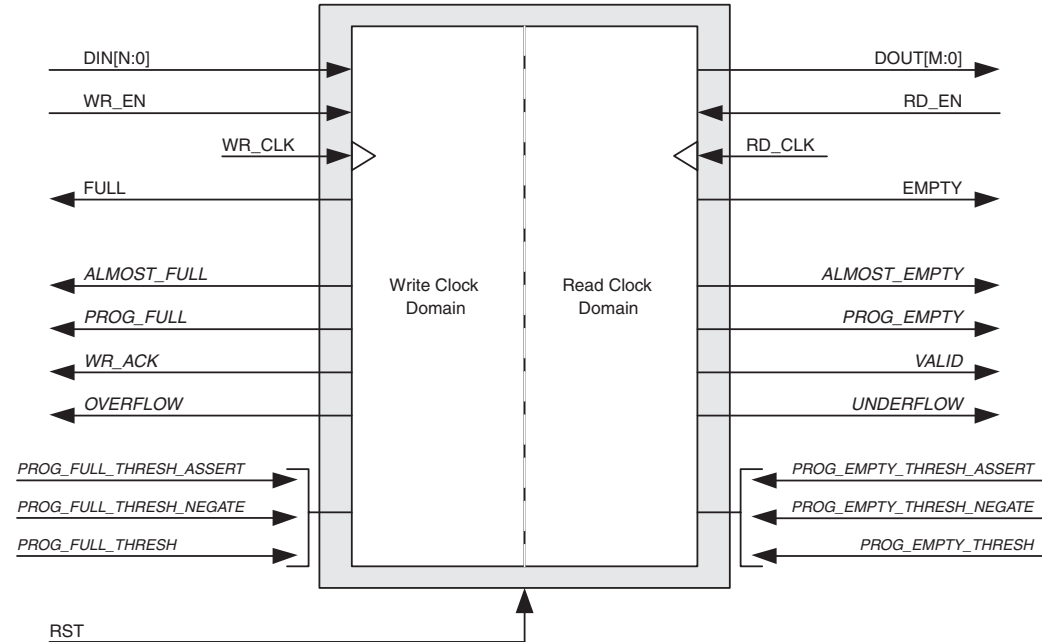
Table 2-3: FIFO Configurations Summary

FIFO Feature	Independent Clocks			Common Clock		
	Block RAM	Distributed RAM	Built-in FIFO	Block RAM	Distributed RAM, Shift Register	Built-in FIFO
Non-symmetric Aspect Ratios ¹	✓					
Symmetric Aspect Ratios	✓	✓	✓	✓	✓	✓
Almost Full	✓	✓		✓	✓	
Almost Empty	✓	✓		✓	✓	
Handshaking	✓	✓	✓	✓	✓	✓
Data Count	✓	✓		✓	✓	
Programmable Empty/Full Thresholds	✓	✓	✓	✓	✓	✓
First-Word Fall-Through	✓	✓	✓ ²			✓ ²
Synchronous Reset				✓	✓ ³	
Asynchronous Reset	✓ ⁴	✓ ⁴	✓	✓ ⁴	✓ ⁴	✓
DOUT Reset Value	✓ ⁵	✓		✓ ⁵	✓	

1. For applications with a single clock that require non-symmetric ports, use the independent clock configuration and connect the write and read clocks to the same source. A dedicated solution for common clocks will be available in a future release. Contact your Xilinx representative for more details.
2. Only supported for Virtex-5 built-in FIFOs.
3. Synchronous Reset available for common clock configurations using distributed and block RAM only.
4. Asynchronous reset is optional for all FIFOs built using distributed and block RAM.
5. All architectures except for Virtex, Virtex-E, Spartan-II, and Spartan-IIE.

FIFO Interfaces

The following two sections provide definitions for the FIFO interface signals. [Figure 2-1](#) illustrates these signals (both the standard and optional ports) for a FIFO core that supports independent write and read clocks.



Note: Optional ports represented in *italics*

Figure 2-1: FIFO with Independent Clocks: Interface Signals

Interface Signals: FIFOs With Independent Clocks

The signal (RST) causes a reset of the entire core logic (both write and read clock domains) and is defined in [Table 2-4](#). It is an asynchronous input which is synchronized internally in the core before being used. The initial hardware reset should be generated by the user. When the core is configured to have independent clocks, the reset signal should be High for at least three read clock and write clock cycles to ensure all internal states are reset to the correct values.

Table 2-4: Reset Signal for FIFOs with Independent Clocks

Name	Direction	Description
RST	Input	Reset: This signal is an asynchronous reset that initializes all internal pointers and output registers.

Table 2-5 defines the signals for the write interface for FIFOs with independent clocks. The write interface signals are divided into required and optional signals and all signals are synchronous to the write clock (WR_CLK).

Table 2-5: Write Interface Signals for FIFOs with Independent Clocks

Name	Direction	Description
<i>Required</i>		
WR_CLK	Input	Write Clock: All signals on the write domain are synchronous to this clock.
DIN[N:0]	Input	Data Input: The input data bus used when writing the FIFO.
WR_EN	Input	Write Enable: If the FIFO is not full, asserting this signal causes data (on DIN) to be written to the FIFO.
FULL	Output	Full Flag: When asserted, this signal indicates that the FIFO is full. Write requests are ignored when the FIFO is full, initiating a write when the FIFO is full is non-destructive to the contents of the FIFO.
<i>Optional</i>		
ALMOST_FULL	Output	Almost Full: When asserted, this signal indicates that only one more write can be performed before the FIFO is full.
PROG_FULL	Output	Programmable Full: This signal is asserted when the number of words in the FIFO is greater than or equal to the assert threshold. It is deasserted when the number of words in the FIFO is less than the negate threshold.
WR_DATA_COUNT [D:0]	Output	Write Data Count: This bus indicates the number of words stored in the FIFO. The count is guaranteed to never under-report the number of words in the FIFO, to ensure the user never overflows the FIFO. If D is less than $\log_2(\text{FIFO depth})-1$, the bus is truncated by removing the least-significant bits.
WR_ACK	Output	Write Acknowledge: This signal indicates that a write request (WR_EN) during the prior clock cycle succeeded.
OVERFLOW	Output	Overflow: This signal indicates that a write request (WR_EN) during the prior clock cycle was rejected, because the FIFO is full. Overflowing the FIFO is non-destructive to the contents of the FIFO.

Table 2-5: Write Interface Signals for FIFOs with Independent Clocks (*Continued*)

Name	Direction	Description
PROG_FULL_THRESH	Input	Programmable Full Threshold: This signal is used to input the threshold value for the assertion and deassertion of the programmable full (PROG_FULL) flag. The threshold can be dynamically set in-circuit during reset. The user can either choose to set the assert and negate threshold to the same value (using PROG_FULL_THRESH), or the user can control these values independently (using PROG_FULL_THRESH_ASSERT and PROG_FULL_THRESH_NEGATE).
PROG_FULL_THRESH_ASSERT	Input	Programmable Full Threshold Assert: This signal is used to set the upper threshold value for the programmable full flag, which defines when the signal is asserted. The threshold can be dynamically set in-circuit during reset.
PROG_FULL_THRESH_NEGATE	Input	Programmable Full Threshold Negate: This signal is used to set the lower threshold value for the programmable full flag, which defines when the signal is deasserted. The threshold can be dynamically set in-circuit during reset.

Table 2-6 defines the signals on the read interface of a FIFO with independent clocks. The read interface signals are divided into required signals and optional signals, and all signals are synchronous to the read clock (RD_CLK).

Table 2-6: Read Interface Signals for FIFOs with Independent Clocks

Name	Direction	Description
<i>Required</i>		
RD_CLK	Input	Read Clock: All signals on the read domain are synchronous to this clock.
DOUT[M:0]	Output	Data Output: The output data bus is driven when reading the FIFO.
RD_EN	Input	Read Enable: If the FIFO is not empty, asserting this signal causes data to be read from the FIFO (output on DOUT).
EMPTY	Output	Empty Flag: When asserted, this signal indicates that the FIFO is empty. Read requests are ignored when the FIFO is empty, initiating a read while empty is non-destructive to the FIFO.
<i>Optional</i>		
ALMOST_EMPTY	Output	Almost Empty Flag: When asserted, this signal indicates that the FIFO is almost empty and one word remains in the FIFO.

Table 2-6: Read Interface Signals for FIFOs with Independent Clocks (Continued)

Name	Direction	Description
PROG_EMPTY	Output	Programmable Empty: This signal is asserted when the number of words in the FIFO is less than or equal to the programmable threshold. It is deasserted when the number of words in the FIFO exceeds the programmable threshold.
RD_DATA_COUNT [C:0]	Output	Read Data Count: This bus indicates the number of words available for reading in the FIFO. The count is guaranteed to never over-report the number of words available for reading, to ensure that the user does not underflow the FIFO. If C is less than $\log_2(\text{FIFO depth})-1$, the bus is truncated by removing the least-significant bits.
VALID	Output	Valid: This signal indicates that valid data is available on the output bus (DOUT).
UNDERFLOW	Output	Underflow: Indicates that the read request (RD_EN) during the previous clock cycle was rejected because the FIFO is empty. Underflowing the FIFO is not destructive to the FIFO.
PROG_EMPTY_THRESH	Input	Programmable Empty Threshold: This signal is used to input the threshold value for the assertion and deassertion of the programmable empty (PROG_EMPTY) flag. The threshold can be dynamically set in-circuit during reset. The user can either choose to set the assert and negate threshold to the same value (using PROG_EMPTY_THRESH), or the user can control these values independently (using PROG_EMPTY_THRESH_ASSERT and PROG_EMPTY_THRESH_NEGATE).
PROG_EMPTY_THRESH_ASSERT	Input	Programmable Empty Threshold Assert: This signal is used to set the lower threshold value for the programmable empty flag, which defines when the signal is asserted. The threshold can be dynamically set in-circuit during reset.
PROG_EMPTY_THRESH_NEGATE	Input	Programmable Empty Threshold Negate: This signal is used to set the upper threshold value for the programmable empty flag, which defines when the signal is deasserted. The threshold can be dynamically set in-circuit during reset.

Interface Signals: FIFOs with Common Clock

Table 2-7 defines the interface signals of a FIFO with a common write and read clock. The table is divided into standard and optional interface signals, and all signals (except reset) are synchronous to the common clock (CLK). Users have the option to select synchronous or asynchronous reset for the distributed or block RAM FIFO implementation.

Table 2-7: Interface Signals for FIFOs with a Common Clock

Name	Direction	Description
<i>Required</i>		
RST	Input	Reset: An asynchronous reset that initializes all internal pointers and output registers.
SRST	Input	Synchronous Reset: A synchronous reset that initializes all internal pointers and output registers.
CLK	Input	Clock: All signals on the write and read domains are synchronous to this clock.
DIN[N:0]	Input	Data Input: The input data bus used when writing the FIFO.
WR_EN	Input	Write Enable: If the FIFO is not full, asserting this signal causes data (on DIN) to be written to the FIFO.
FULL	Output	Full Flag: When asserted, this signal indicates that the FIFO is full. Write requests are ignored when the FIFO is full, initiating a write when the FIFO is full is non-destructive to the contents of the FIFO.
DOUT[M:0]	Output	Data Output: The output data bus driven when reading the FIFO.
RD_EN	Input	Read Enable: If the FIFO is not empty, asserting this signal causes data to be read from the FIFO (output on DOUT).
EMPTY	Output	Empty Flag: When asserted, this signal indicates that the FIFO is empty. Read requests are ignored when the FIFO is empty, initiating a read while empty is non-destructive to the FIFO.
<i>Optional</i>		
DATA_COUNT [C:0]	Output	Data Count: This bus indicates the number of words stored in the FIFO. If C is less than $\log_2(\text{FIFO depth}) - 1$, the bus is truncated by removing the least-significant bits.
ALMOST_FULL	Output	Almost Full: When asserted, this signal indicates that only one more write can be performed before the FIFO is full.

Table 2-7: Interface Signals for FIFOs with a Common Clock (Continued)

Name	Direction	Description
PROG_FULL	Output	Programmable Full: This signal is asserted when the number of words in the FIFO is greater than or equal to the assert threshold. It is deasserted when the number of words in the FIFO is less than the negate threshold.
WR_ACK	Output	Write Acknowledge: This signal indicates that a write request (WR_EN) during the prior clock cycle succeeded.
OVERFLOW	Output	Overflow: This signal indicates that a write request (WR_EN) during the prior clock cycle was rejected, because the FIFO is full. Overflowing the FIFO is non-destructive to the contents of the FIFO.
PROG_FULL_THRESH	Input	<p>Programmable Full Threshold: This signal is used to set the threshold value for the assertion and deassertion of the programmable full flag (PROG_FULL). The threshold can be dynamically set in-circuit during reset.</p> <p>The user can either choose to set the assert and negate threshold to the same value (using PROG_FULL_THRESH), or the user can control these values independently (using PROG_FULL_THRESH_ASSERT and PROG_FULL_THRESH_NEGATE).</p>
PROG_FULL_THRESH_ASSERT	Input	Programmable Full Threshold Assert: This signal is used to set the upper threshold value for the programmable full flag, which defines when the signal is asserted. The threshold can be dynamically set in-circuit during reset.
PROG_FULL_THRESH_NEGATE	Input	Programmable Full Threshold Negate: This signal is used to set the lower threshold value for the programmable full flag, which defines when the signal is deasserted. The threshold can be dynamically set in-circuit during reset.
ALMOST_EMPTY	Output	Almost Empty Flag: When asserted, this signal indicates that the FIFO is almost empty and one word remains in the FIFO.
PROG_EMPTY	Output	Programmable Empty: This signal is asserted after the number of words in the FIFO is less than or equal to the programmable threshold. It is deasserted when the number of words in the FIFO exceeds the programmable threshold.

Table 2-7: Interface Signals for FIFOs with a Common Clock (Continued)

Name	Direction	Description
VALID	Output	Valid: This signal indicates that valid data is available on the output bus (DOUT).
UNDERFLOW	Output	Underflow: Indicates that read request (RD_EN) during the previous clock cycle was rejected because the FIFO is empty. Underflowing the FIFO is not destructive to the FIFO.
PROG_EMPTY_THRESH	Input	<p>Programmable Empty Threshold: This signal is used to set the threshold value for the assertion and deassertion of the programmable empty flag (PROG_EMPTY). The threshold can be dynamically set in-circuit during reset.</p> <p>The user can either choose to set the assert and negate threshold to the same value (using PROG_EMPTY_THRESH), or the user can control these values independently (using PROG_EMPTY_THRESH_ASSERT and PROG_EMPTY_THRESH_NEGATE).</p>
PROG_EMPTY_THRESH_ASSERT	Input	Programmable Empty Threshold Assert: This signal is used to set the lower threshold value for the programmable empty flag, which defines when the signal is asserted. The threshold can be dynamically set in-circuit during reset.
PROG_EMPTY_THRESH_NEGATE	Input	Programmable Empty Threshold Negate: This signal is used to set the upper threshold value for the programmable empty flag, which defines when the signal is deasserted. The threshold can be dynamically set in-circuit during reset.

Generating the Core

This chapter contains information and instructions for using the Xilinx CORE Generator system to customize the FIFO Generator.

CORE Generator Graphical User Interface

The FIFO Generator graphical user interface (GUI) contains six configuration screens.

- FIFO Implementation
- Performance Options and Data Port Parameters
- Optional Flags and Handshaking Options
- Programmable Flags
- Data Count and Reset
- Summary

FIFO Implementation

The main FIFO Generator screen is used to define the component name and provides configuration options for the core.

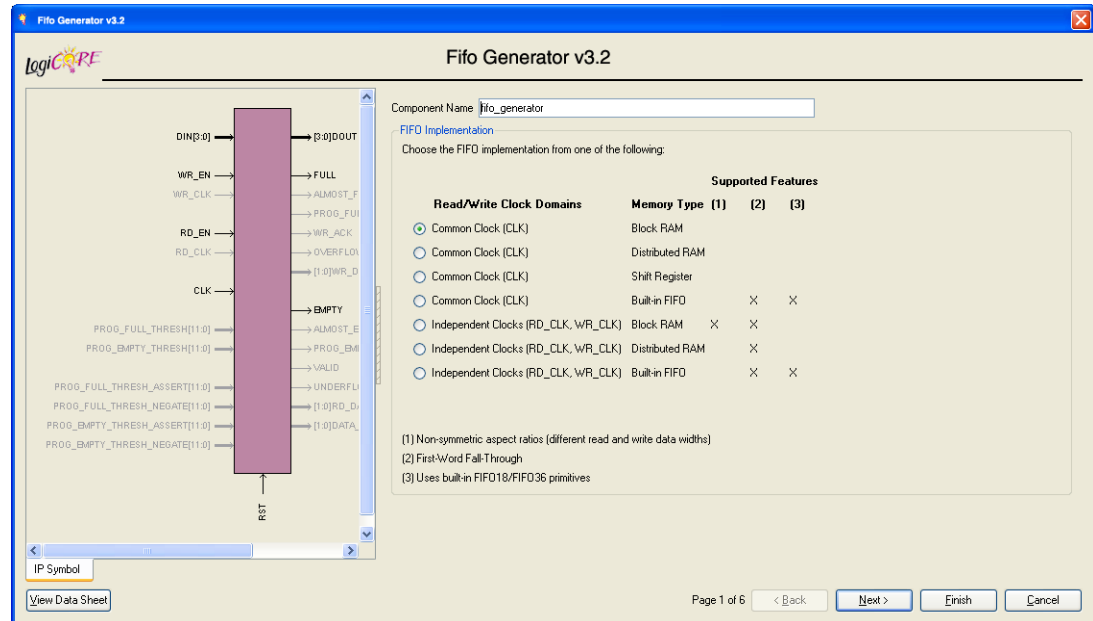


Figure 3-1: Main FIFO Generator Screen

Component Name

Base name of the output files generated for this core. The name must begin with a letter and be composed of the following characters: a to z, 0 to 9, and “_”.

FIFO Implementation

This section of the GUI allows the user to select from a set of available FIFO implementations and supported features. The key supported features that are only available for certain implementations are highlighted by checks in the right-margin. The available options are listed below, with cross-references to additional information.

Common Clock (CLK), Block RAM

For details, see [“Common Clock FIFO: Block RAM and Distributed RAM,” page 44.](#)

Common Clock (CLK), Distributed RAM

For details, see [“Common Clock FIFO: Block RAM and Distributed RAM,” page 44.](#)

Common Clock (CLK), Shift Register

For details, see [“Common Clock FIFO: Shift Registers,” page 44.](#) This implementation is only available in Virtex-II and newer architectures.

Common Clock (CLK), Built-in FIFO

For details, see [“Common Clock: Built-in FIFO,” page 43](#). This implementation is only available when using the Virtex-5 or Virtex-4 architectures. This implementation optionally supports first-word fall-through (selectable in the second GUI screen, shown in [Figure 3-2](#)).

Independent Clocks (RD_CLK, WR_CLK), Block RAM

For details, see [“Independent Clocks: Block RAM and Distributed RAM,” page 41](#). This implementation optionally supports asymmetric read/write ports and first-word fall-through (selectable in the second GUI screen, shown in [Figure 3-2](#)).

Independent Clocks (RD_CLK, WR_CLK), Distributed RAM

For more information, see [“Independent Clocks: Block RAM and Distributed RAM,” page 41](#). This implementation optionally supports first-word fall-through (selectable in the second GUI screen, shown in [Figure 3-2](#)).

Independent Clocks (RD_CLK, WR_CLK), Built-in FIFO

For more information, see [“Independent Clocks: Built-in FIFO,” page 42](#). This implementation is only available when using Virtex-5 or Virtex-4 architectures. This implementation optionally supports first-word fall-through (selectable in the second GUI screen, shown in [Figure 3-2](#)).

Performance Options and Data Port Parameters

This screen provides performance options and data port parameters for the core.

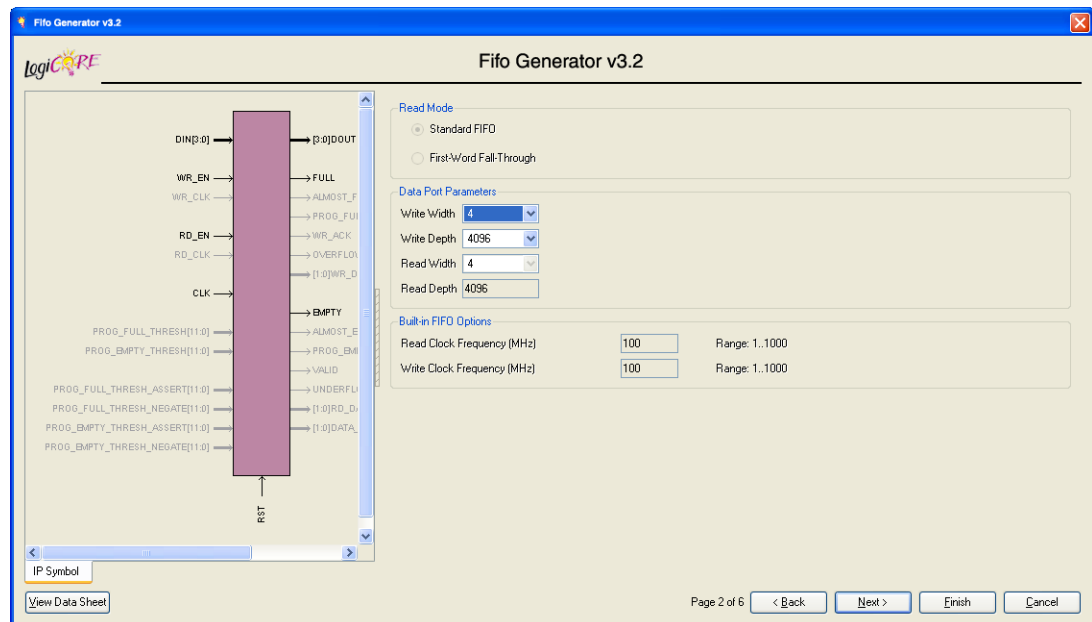


Figure 3-2: Performance Options and Data Port Parameters Screen

Performance Options

Only available when Virtex-5 built-in FIFO or independent clock FIFO with block RAM or distributed RAM FIFOs is selected. For more information, see [“Read Operation,” page 46](#).

Standard FIFO

Implements a FIFO with standard latencies, and without using output registers.

First-word Fall-through FIFO

Implements a FIFO with registered outputs. For more information about FWFT functionality, see [“First-Word Fall-Through FIFO Read Operation,” page 48](#).

Data Port Parameters

Input Data Width

Valid range is 1 to 256.

Input Depth

Valid range is 16 to 4194394. Only depths with powers of 2 are allowed.

Output Data Width

Available if independent clocks configuration with block RAM is selected. Valid range must comply with asymmetric port rules. See [“Non-symmetric Aspect Ratios,” page 58](#).

Output Depth

Automatically calculated based on Input Data Width, Input Depth, and Output Data Width.

Built-in FIFO Options

The Read Clock Frequency and Write Clock Frequency fields can be any integer from 1 to 999. They are used to determine the optimal implementation of the domain-crossing logic in the core. This option is only available for built-in FIFOs with independent clocks. If the desired frequency is not within the allowable range, scale the read and write clock frequencies so that they fit within the valid range, while maintaining their ratio relationship.

Optional Flags, Handshaking, and Initialization

This screen allows you to select the optional status flags and set the handshaking options.

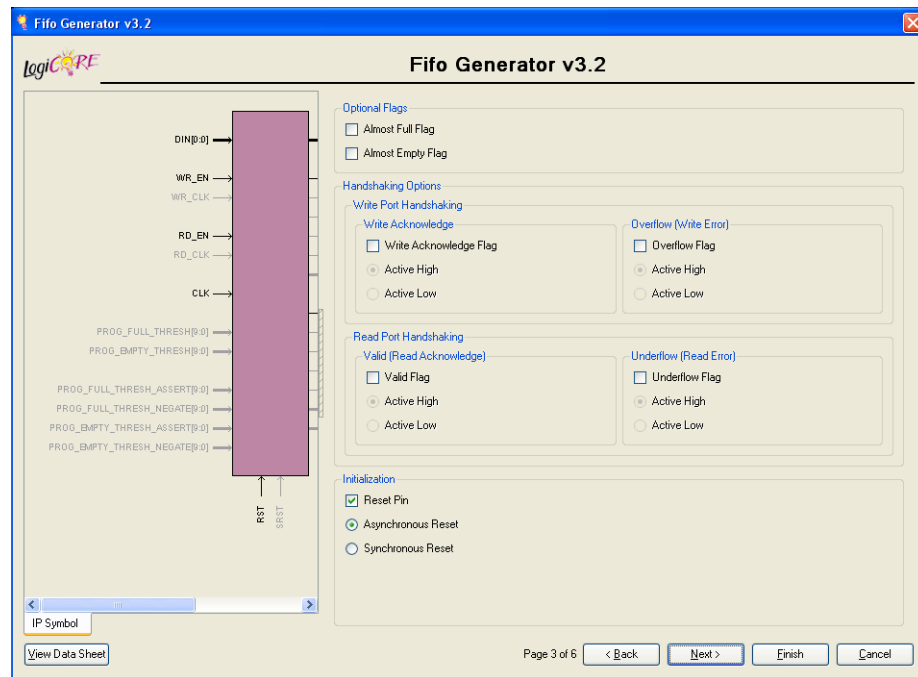


Figure 3-3: Optional Flags, Handshaking, and Initialization Options Screen

Optional Flags

Almost Full Flag

Available in all FIFO implementations except those using Virtex-5 or Virtex-4 built-in FIFOs. Generates an output port that indicates the FIFO is almost full (only one more word can be written).

Almost Empty Flag

Available in all FIFO implementations except in those using Virtex-5 or Virtex-4 built-in FIFOs. Generates an output port that indicates the FIFO is almost empty (only one more word can be read).

Write Port Handshaking

Write Acknowledge

Generates write acknowledge flag which reports the status of a write operation. This signal can be configured to be active high or low (default active high).

Overflow (Write Error)

Generates overflow flag which indicates when the previous write operation was not successful. This signal can be configured to be active high or low (default active high).

Read Port Handshaking

Valid (Read Acknowledge)

Generates valid flag which indicates when the data on the output bus is valid. This signal can be configured to be active high or low (default active high).

Underflow (Read Error)

Generates underflow flag to indicate that the previous read request was not successful. This signal can be configured to be active high or low (default active high).

Initialization

Reset Pin

For FIFOs implemented with block RAM or distributed RAM, a reset pin is not required, and the input pin is optional. For a common clock FIFO implemented using distributed or block RAM, the user has the option to generate a synchronous or asynchronous reset.

Programmable Flags

Use this screen to select the programmable flag type when generating a specific FIFO Generator configuration.

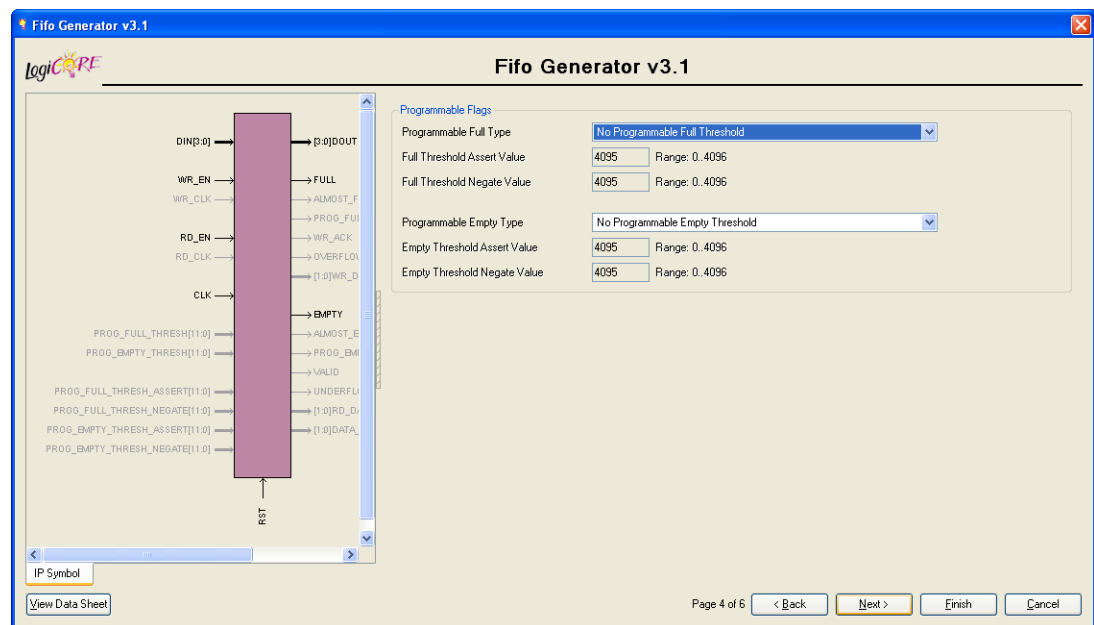


Figure 3-4: Programmable Flags Screen

Programmable Flags

Programmable Full Type

Select a programmable full threshold type from the drop-down menu. The valid range for each threshold is displayed, and will vary, depending on options selected elsewhere in the GUI.

Full Threshold Assert Value

Available when Programmable Full with Single or Multiple Threshold Constants is selected. Enter a user-defined value, or select a preset value from the drop-down menu. The valid range for this threshold is provided in the GUI. When using a single threshold constant, only the assert threshold value is used.

Full Threshold Negate Value

Available when Programmable Full with Multiple Threshold Constants is selected. Enter a user-defined value, or select a preset value from the drop-down menu. The valid range for this threshold is provided in the GUI.

Programmable Empty Type

Select a programmable empty threshold type from the drop-down menu. The valid range for each threshold is displayed, and will vary depending on options selected elsewhere in the GUI.

Empty Threshold Assert Value

Available when Programmable Empty with Single or Multiple Threshold Constants is selected. Enter a user-defined value, or select a preset value from the drop-down menu. The valid range for this threshold is provided in the GUI. When using a single threshold constant, only the assert value is used.

Empty Threshold Negate Value

Available when Programmable Empty with Multiple Threshold Constants is selected. Enter a user-defined value, or select a preset value from the drop-down menu. The valid range for this threshold is provided in the GUI.

Data Count and Reset

Use this screen to set data count and reset parameters.

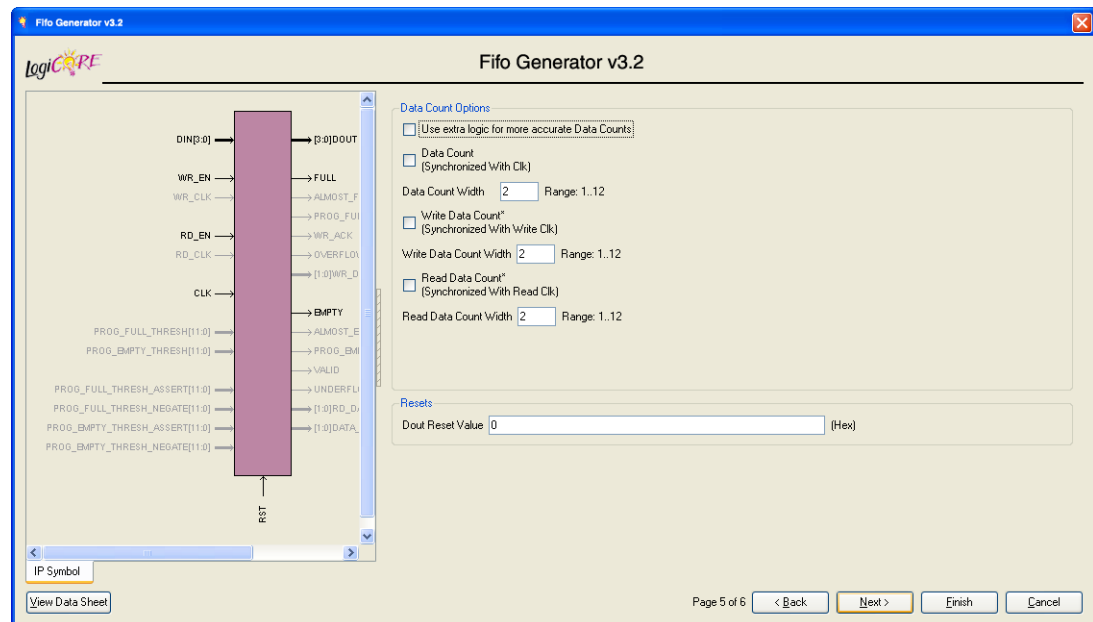


Figure 3-5: Data Count and Reset Screen

Data Count and Reset Options

Data Count

Use Extra Logic For More Accurate Data Counts

Only available for independent clocks FIFO with block RAM or distributed RAM, and when using first-word fall-through. This option uses additional external logic to generate a more accurate data count. See [“First-Word Fall-Through Data Count,” page 57](#) for details.

Write Data Count

Available when an independent clocks FIFO with block RAM or distributed RAM is selected. Valid range is from 1 to \log_2 (write depth).

Read Data Count

Available when an independent clocks FIFO with block RAM or distributed RAM is selected. Valid range is from 1 to \log_2 (read depth).

Data Count

Available when a common clock FIFO with block RAM, distributed RAM or shift registers is selected. Valid range is from 1 to \log_2 of input depth.

Resets

Dout Reset Value

Available in Virtex-II and newer architectures for all implementations using block RAM, distributed RAM, or shift register memory. This text box indicates the hexadecimal value which is asserted on the output of the FIFO when RST is asserted.

Summary

This screen summarizes the FIFO type, dimensions, and any additional features selected. In the Additional Features section, most features display either *Not Selected* (if the feature is not used), or *Selected* (if the feature is used).

Note: Write depth and read depth provide the actual FIFO depths for the selected configuration. These depths may differ slightly from the depth selected on page 2 of the FIFO GUI.

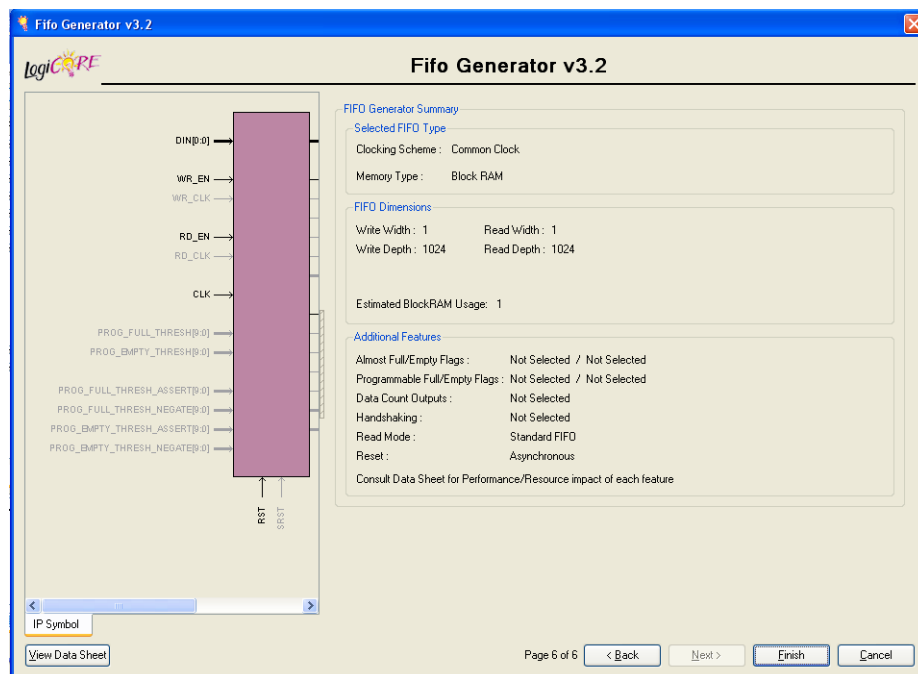


Figure 3-6: Summary Screen

Designing with the Core

This chapter describes the steps required to turn a FIFO Generator core into a fully functioning design integrated with the user application logic. It is important to note that depending on the configuration of the FIFO core, only a subset of the implementation details provided are applicable. For successful use of a FIFO core, the design guidelines discussed in this chapter must be observed.

General Design Guidelines

Know the Degree of Difficulty

A fully-compliant and feature-rich FIFO design is challenging to implement in any technology. For this reason, it is important to understand that the degree of difficulty can be significantly influenced by

- Maximum system clock frequency
- Targeted device architecture
- Specific user application

Ensure that design techniques are used to facilitate implementation, including pipelining and use of constraints (timing constraints, and placement and/or area constraints).

Understand Signal Pipelining and Synchronization

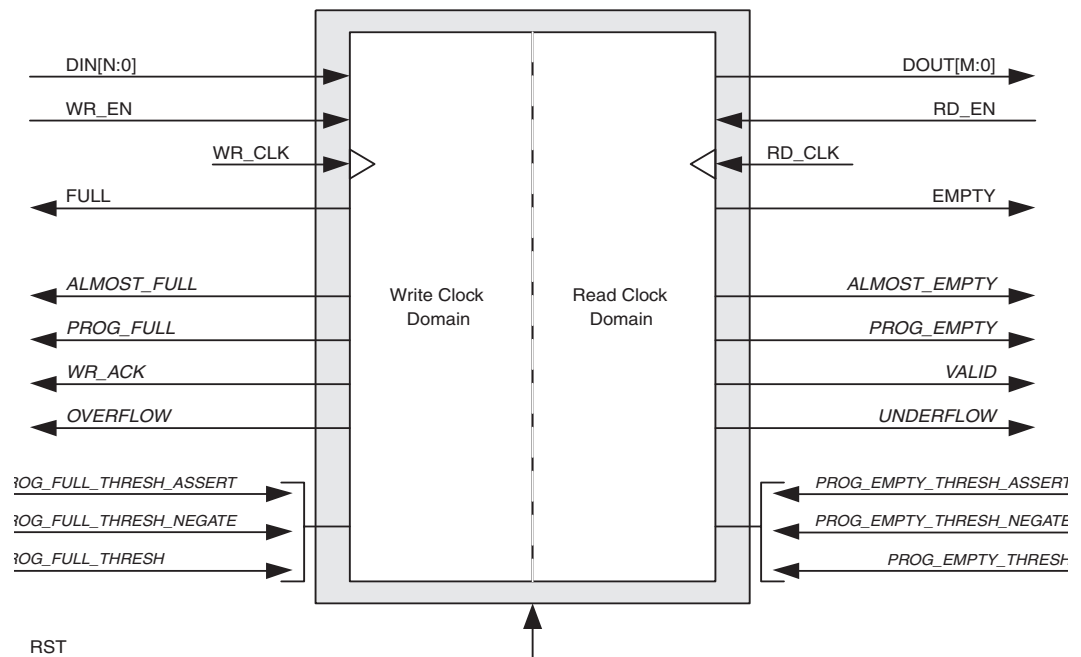
To understand the nature of FIFO designs, it is important to understand how pipelining is used to maximize performance and implement synchronization logic for clock-domain crossing. Data written into the write interface may take multiple clock cycles before it can be accessed on the read interface.

Synchronization Considerations

FIFOs with independent write and read clocks require that interface signals be used only in their respective clock domains. The independent clocks FIFO handles all synchronization requirements, enabling the user to cross between two clock domains that have no relationship in frequency or phase.

Important! FIFO Full and Empty flags must be used to guarantee proper behavior.

Figure 4-1 shows the signals with respect to their clock domains. All signals are synchronous to a specific clock, with the exception of RST, which performs an asynchronous reset of the entire FIFO.



Note: Optional ports represented in *italics*

Figure 4-1: FIFO with Independent Clocks: Write and Read Clock Domains

For write operations, the write enable signal (WR_EN) and data input (DIN) are synchronous to WR_CLK. For read operations, the read enable (RD_EN) and data output (DOUT) are synchronous to RD_CLK. All status outputs are synchronous to their respective clock domains and can only be used in that clock domain. The performance of the FIFO can be measured by independently constraining the clock period for the WR_CLK and RD_CLK input signals.

The interface signals are evaluated on their rising clock edge (WR_CLK and RD_CLK). They can be made falling-edge active (relative to the clock source) by inserting an inverter between the clock source and the FIFO clock inputs. This inverter is absorbed into the internal FIFO control logic and does not cause a decrease in performance or increase in logic utilization.

Initializing the FIFO Generator

When designing with the built-in FIFO or common clock shift register FIFO, the FIFO must be reset after the FPGA is configured and before operation begins. An asynchronous reset pin (RST) is provided, which is an asynchronous reset that clears the internal counters and output registers.

For FIFOs implemented with block RAM or distributed RAM, a reset is not required, and the input pin is optional. For these FIFO configurations, users have the option of asynchronous or synchronous reset (only in common clock implementations). When asynchronous reset is implemented, it is synchronized to the clock domain in which it is used to ensure that the FIFO initializes to a known state. This synchronization logic allows for proper reset timing of the core logic, avoiding glitches and metastable behavior. For independent clock implementations, the synchronization process mandates a 3-cycle delay post-reset prior to writing to the FIFO. For common clock implementations, a 2-cycle delay post-reset prior to writing to the FIFO is mandated. For synchronous reset, no additional

synchronization is required and the FIFO is available for writing in the next cycle after reset.

FIFO Implementations

Each FIFO configuration has a set of allowable features, as defined in [Table 2-3, page 20](#).

Independent Clocks: Block RAM and Distributed RAM

[Figure 4-2](#) illustrates the functional implementation of a FIFO configured with independent clocks. This implementation uses block RAM or distributed RAM for memory, counters for write and read pointers, conversions between binary and Gray code for synchronization across clock domains, and logic for calculating the status flags.

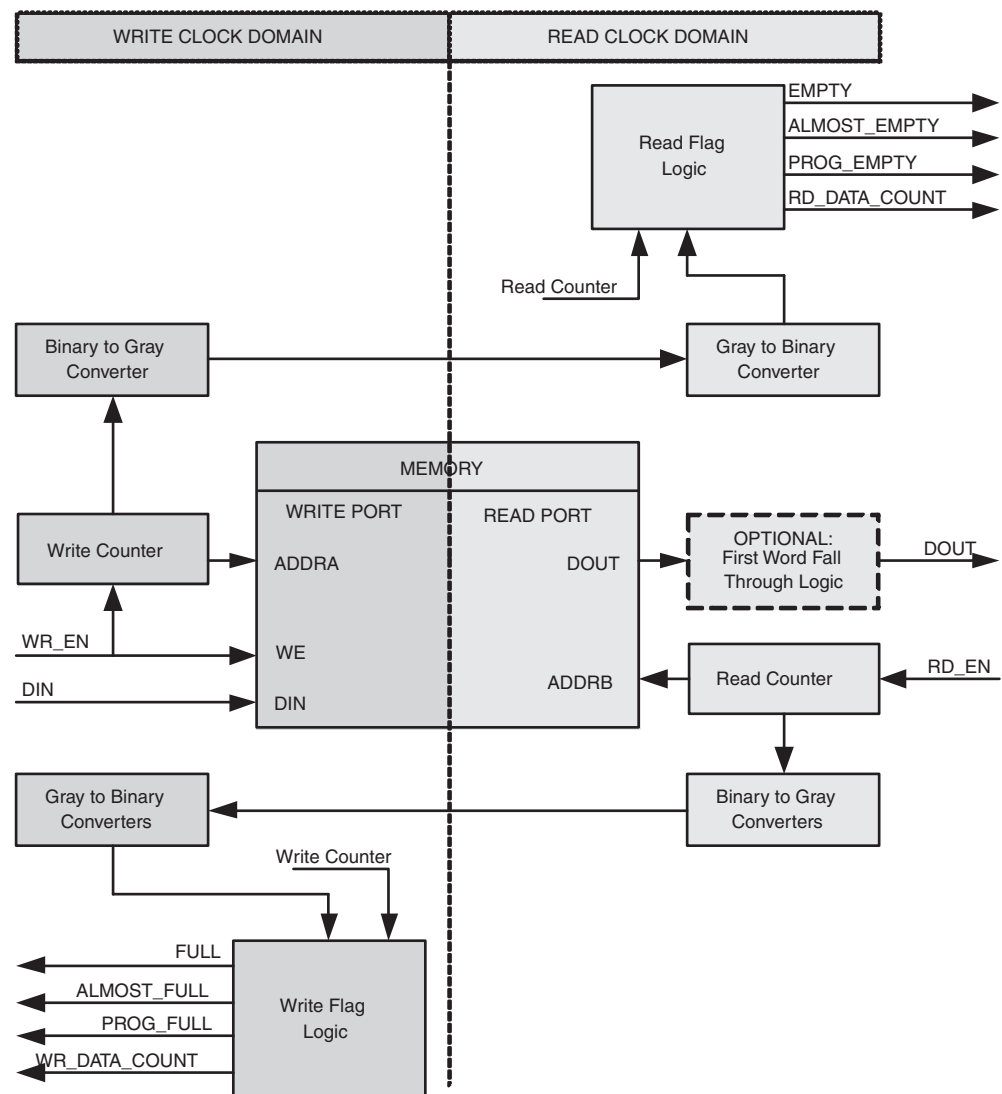


Figure 4-2: Functional Implementation of a FIFO with Independent Clock Domains

This FIFO is designed to support an independent read clock (RD_CLK) and write clock (WR_CLK); in other words, there is no required relationship between RD_CLK and WR_CLK with regard to frequency or phase. The FIFO interface signals are only valid in their respective clock domains and are summarize in [Table 4-1](#).

Table 4-1: Interface Signals and Corresponding Clock Domains

WR_CLK	RD_CLK
DIN	DOUT
WR_EN	RD_EN
FULL	EMPTY
ALMOST_FULL	ALMOST_EMPTY
PROG_FULL	PROG_EMPTY
WR_ACK	VALID
OVERFLOW	UNDERFLOW
WR_DATA_COUNT	RD_DATA_COUNT

For FIFO cores using independent clocks, the timing relationship between the write and read operations and the status flags is affected by the relationship of the two clocks. For example, the timing between writing to an empty FIFO and the deassertion of EMPTY is determined by the phase and frequency relationship between the write and read clocks. For additional information refer to the [“Synchronization Considerations,”](#) page 39.

Independent Clocks: Built-in FIFO

[Figure 4-3](#) illustrates the functional implementation of FIFO configured with independent clocks using the Virtex-5 built-in FIFO primitive. This design implementation consists of cascaded built-in FIFO primitives and handshaking logic. The number of built-in primitives depends on the FIFO width and depth requested.

The Virtex-4 built-in FIFO implementation allows generation of a single primitive. The generated core includes the FIFO patch as defined in [Answer Record 22462](#).

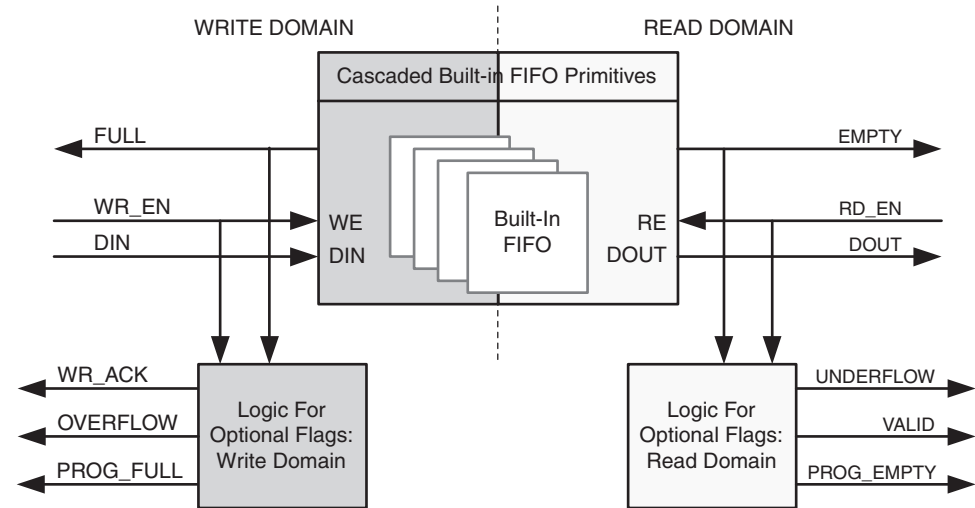


Figure 4-3: Functional Implementation of Built-in FIFO

This FIFO is designed to support an independent read clock (RD_CLK) and write clock (WR_CLK); in other words, there is no required relationship between RD_CLK and WR_CLK with regard to frequency or phase. The FIFO interface signals are only valid in their respective clock domains, and are summarized in [Table 4-2](#).

Table 4-2: Interface Signals and Corresponding Clock Domains

WR_CLK	RD_CLK
DIN	DOUT
WR_EN	RD_EN
FULL	EMPTY
PROG_FULL	PROG_EMPTY
WR_ACK	VALID
OVERFLOW	UNDERFLOW

For FIFO cores using independent clocks, the timing relationship between the write and read operations and the status flags is affected by the relationship of the two clocks. For example, the timing between writing to an empty FIFO and the deassertion of EMPTY is determined by the phase and frequency relationship between the write and read clocks. For additional information refer to the [“Synchronization Considerations,”](#) page 39.

Common Clock: Built-in FIFO

The FIFO Generator supports FIFO cores using the built-in FIFO primitive with a common clock. This provides users the ability to use the built-in FIFO, while requiring only a single clock interface. The behavior of the common clock configuration with built-in FIFO is identical to the independent clock configuration with built-in FIFO, except all operations are in relation to the common clock (CLK). See [“Independent Clocks: Built-in FIFO,”](#) page 42, for more information.

Common Clock FIFO: Block RAM and Distributed RAM

Figure 4-4 illustrates the functional implementation of a FIFO configured with a common clock using block RAM or distributed RAM for memory. All signals are synchronous to a single clock input (CLK). This design implements counters for write and read pointers and logic for calculating the status flags. An optional synchronous (SRST) or asynchronous (RST) reset signal is also available.

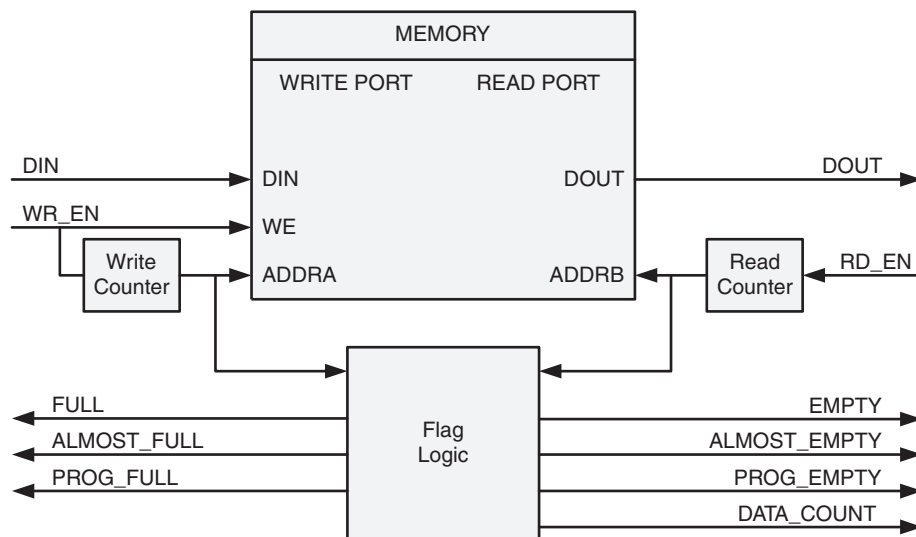


Figure 4-4: Functional Implementation of a Common Clock FIFO using Block RAM or Distributed RAM

Common Clock FIFO: Shift Registers

Figure 4-5 illustrates the functional implementation of a FIFO configured with a common clock using shift registers for memory. All operations are synchronous to the same clock

input (CLK). This design implements a single up/down counter for both the write and read pointers and logic for calculating the status flags.

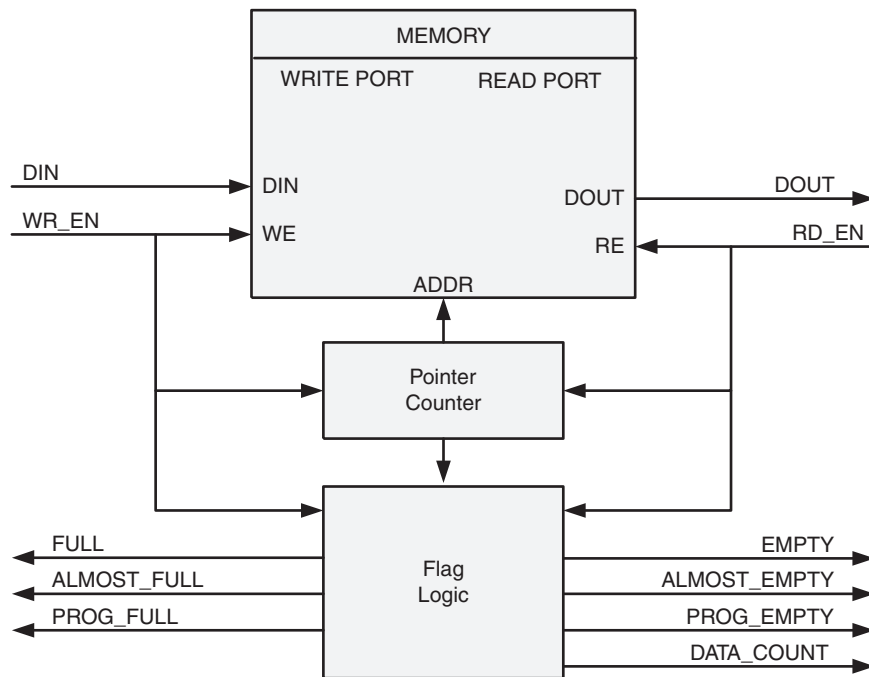


Figure 4-5: Functional Implementation of a Common Clock FIFO using Shift Registers

FIFO Usage and Control

Write Operation

This section describes the behavior of a FIFO write operation and the associated status flags. When write enable is asserted and the FIFO is not full, data is added to the FIFO from the input bus (DIN) and write acknowledge (WR_ACK) is asserted. If the FIFO is continuously written to without being read, it fills with data. Write operations are only successful when the FIFO is not full. When the FIFO is full and a write is initiated, the request is ignored, the overflow flag is asserted and there is no change in the state of the FIFO (overflowing the FIFO is non-destructive).

ALMOST_FULL and FULL Flags

Note: The Virtex-5 and Virtex-4 built-in FIFO does not support the ALMOST_FULL flag.

The almost full flag (ALMOST_FULL) indicates that only one more write can be performed before FULL is asserted. This flag is active high and synchronous to the write clock (WR_CLK).

The full flag (FULL) indicates that the FIFO is full and no more writes can be performed until data is read out. This flag is active high and synchronous to the write clock (WR_CLK). If a write is initiated when FULL is asserted, the write request is ignored and OVERFLOW is asserted.

Important! For the Virtex-4 built-in FIFO implementation, the Full signal has an extra cycle of latency. User Write Acknowledge to verify success or Programmable Full for an earlier indication.

Example Operation

Figure 4-6 shows a typical write operation. The user asserts `WR_EN`, causing a write operation to occur on the next rising edge of the `WR_CLK`. Since the FIFO is not full, `WR_ACK` is asserted, acknowledging a successful write operation. When only one additional word can be written into the FIFO, the FIFO asserts the `ALMOST_FULL` flag. When `ALMOST_FULL` is asserted, one additional write causes the FIFO to assert `FULL`. When a write occurs after `FULL` is asserted, `WR_ACK` is deasserted and `OVERFLOW` is asserted, indicating an overflow condition. Once the user performs one or more read operations, the FIFO deasserts `FULL`, and data can successfully be written to the FIFO, as is indicated by the assertion of `WR_ACK` and deassertion of `OVERFLOW`.

Note: The Virtex-4 built-in FIFO implementation will show an extra cycle of latency on the `FULL` flag.

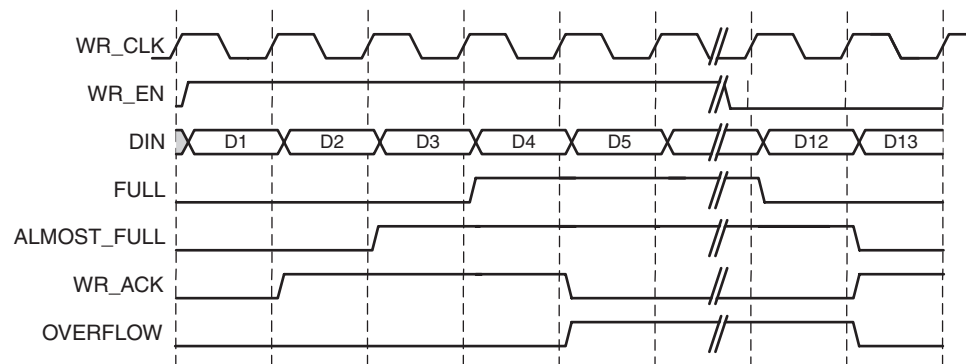


Figure 4-6: Write Operation for a FIFO with Independent Clocks

Read Operation

This section describes the behavior of a FIFO read operation and the associated status flags. When read enable is asserted and the FIFO is not empty, data is read from the FIFO on the output bus (`DOUT`), and the valid flag (`VALID`) is asserted. If the FIFO is continuously read without being written, the FIFO empties. Read operations are successful when the FIFO is not empty. When the FIFO is empty and a read is requested, the read operation is ignored, the underflow flag is asserted and there is no change in the state of the FIFO (underflowing the FIFO is non-destructive).

ALMOST_EMPTY and EMPTY Flags

Note: The Virtex-5 and Virtex-4 built-in FIFO does not support the `ALMOST_EMPTY` flag.

The almost empty flag (`ALMOST_EMPTY`) indicates that the FIFO will be empty after one more read operation. This flag is active high and synchronous to `RD_CLK`. This flag is asserted when the FIFO has one remaining word that can be read.

The empty flag (`EMPTY`) indicates that the FIFO is empty and no more reads can be performed until data is written into the FIFO. This flag is active high and synchronous to the read clock (`RD_CLK`). If a read is initiated when `EMPTY` is asserted, the request is ignored and `UNDERFLOW` is asserted.

Common Clock Note

When write and read operations occur simultaneously while `EMPTY` is asserted, the write operation is accepted and the read operation is ignored. On the next clock cycle, `EMPTY` is deasserted and `UNDERFLOW` is asserted.

Modes of Read Operation

The FIFO Generator supports two modes of read options, standard read operation and first-word fall-through (FWFT) read operation. The standard read operation provides the user data on the cycle after it was requested. The FWFT read operation provides the user data on the same cycle in which it is requested.

Table 4-3 details the supported implementations for FWFT.

Table 4-3: Implementation-Specific Support for First-Word Fall-Through

FIFO Implementation		FWFT Support
Independent Clocks	Block RAM	✓
	Distributed RAM	✓
	Built-in	✓ [∞]
Common Clock	Block RAM	
	Distributed RAM	
	Shift Register	
	Built-in	✓ ¹

1. Only supported in Virtex-5 built-in FIFO.

Standard FIFO Read Operation

For a standard FIFO read operation, after read enable is asserted and if the FIFO is not empty, the next data stored in the FIFO is driven on the output bus (DOUT) and the valid flag (VALID) is asserted.

Figure 4-7 shows a standard read access. Once the user writes at least one word into the FIFO, EMPTY is deasserted—indicating data is available to be read. The user asserts RD_EN, causing a read operation to occur on the next rising edge of RD_CLK. The FIFO outputs the next available word on DOUT and asserts VALID, indicating a successful read operation. When the last data word is read from the FIFO, the FIFO asserts EMPTY. If the user continues to assert RD_EN while EMPTY is asserted, the read request is ignored, VALID is deasserted, and UNDERFLOW is asserted. Once the user performs a write operation, the FIFO deasserts EMPTY, allowing the user to resume valid read operations, as indicated by the assertion of VALID and deassertion of UNDERFLOW.

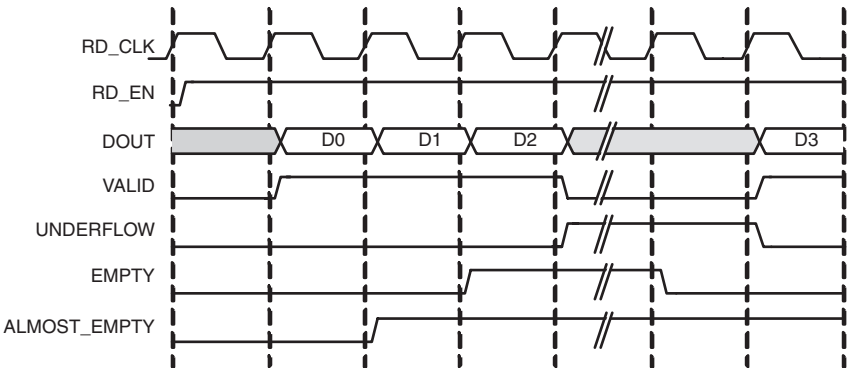


Figure 4-7: Standard Read Operation for a FIFO with Independent Clocks

First-Word Fall-Through FIFO Read Operation

The first-word fall-through (FWFT) feature provides the ability to look-ahead to the next word available from the FIFO without issuing a read operation. When data is available in the FIFO, the first word falls through the FIFO and appears automatically on the output bus (DOUT). Once the first word appears on DOUT, `EMPTY` is deasserted indicating one or more readable words in the FIFO, and `VALID` is asserted, indicating a valid word is present on DOUT.

Figure 4-8 shows a FWFT read access. Initially, the FIFO is not empty, the next available data word is placed on the output bus (DOUT), and `VALID` is asserted. When the user asserts `RD_EN`, the next rising clock edge of `RD_CLK` places the next data word onto DOUT. After the last data word has been placed on DOUT, an additional read request by the user causes the data on DOUT to become invalid, as indicated by the deassertion of `VALID` and the assertion of `EMPTY`. Any further attempts to read from the FIFO results in an underflow condition.

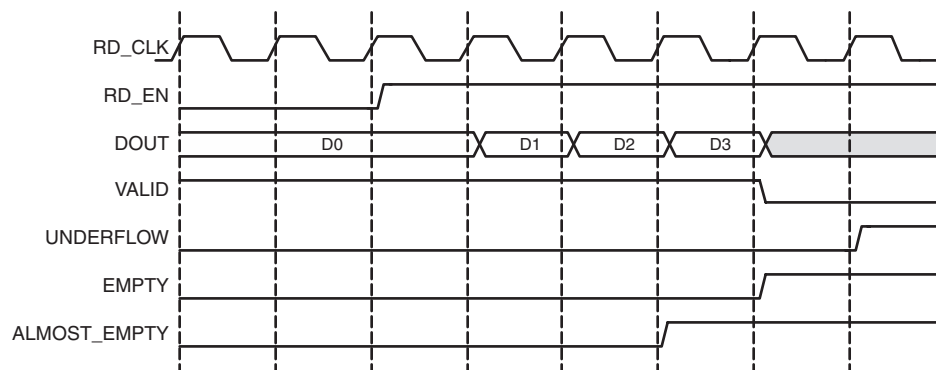


Figure 4-8: FWFT Read Operation for a FIFO with Independent Clocks

Common Clock FIFO, Simultaneous Read and Write Operation

Figure 4-9 shows atypical write and read operation. A write is issued to the FIFO, resulting in the deassertion of the `EMPTY` flag. A simultaneous write and read is then issued, resulting in no change in the status flags. Once two or more words are present in the FIFO, the `ALMOST_EMPTY` flag is deasserted. Write requests are then issued to the FIFO, resulting in the assertion of `ALMOST_FULL` when the FIFO can only accept one more write (without a read). A simultaneous write and read is then issued, resulting in no change in the status flags. Finally one additional write without a read results in the FIFO asserting `FULL`, indicating no further data can be written until a read request is issued.

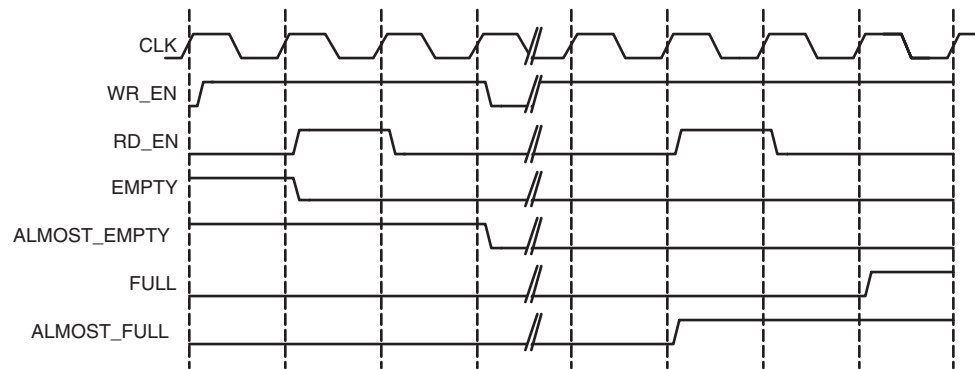


Figure 4-9: Write and Read Operation for a FIFO with Common Clocks

Handshaking Flags

Handshaking flags (valid, underflow, write acknowledge and overflow) are supported to provide additional information regarding the status of the write and read operations. The handshaking flags are optional, and can be configured as active high or active low through the CORE Generator GUI (see Handshaking Options in Chapter 4 for more information). These flags (configured as active high) are illustrated in [Figure 4-10](#).

Write Acknowledge

The write acknowledge flag (WR_ACK) is asserted at the completion of each successful write operation and indicates that the data on the DIN port has been stored in the FIFO. This flag is synchronous to the write clock (WR_CLK).

Valid

The operation of the valid flag (VALID) is dependent on the read mode of the FIFO. This flag is synchronous to the read clock (RD_CLK).

Standard FIFO Read Operation

For standard read operation, the VALID flag is asserted at the rising edge of RD_CLK for each successful read operation, and indicates that the data on the DOUT bus is valid. When a read request is unsuccessful (when the FIFO is empty), VALID is not asserted.

FWFT FIFO Read Operation

For FWFT read operation, the VALID flag indicates the data on the output bus (DOUT) is valid for the current cycle. A read request does not have to happen for data to be present and valid, as the first-word fall-through logic automatically places the next data to be read on the DOUT bus. VALID is asserted if there is one or more words in the FIFO. VALID is deasserted when there are no more words in the FIFO.

Example Operation

[Figure 4-10](#) illustrates the behavior of the FIFO flags. On the write interface, FULL is not asserted and writes to the FIFO are successful (as indicated by the assertion of WR_ACK). When a write occurs after FULL is asserted, WR_ACK is deasserted and OVERFLOW is asserted, indicating an overflow condition. On the read interface, once the FIFO is not EMPTY, the FIFO accepts read requests. In standard FIFO operation, VALID is asserted and

DOUT is updated on the clock cycle following the read request. In FWFT operation, VALID is asserted and DOUT is updated prior to a read request being issued. When a read request is issued while EMPTY is asserted, VALID is deasserted and UNDERFLOW is asserted, indicating an underflow condition.

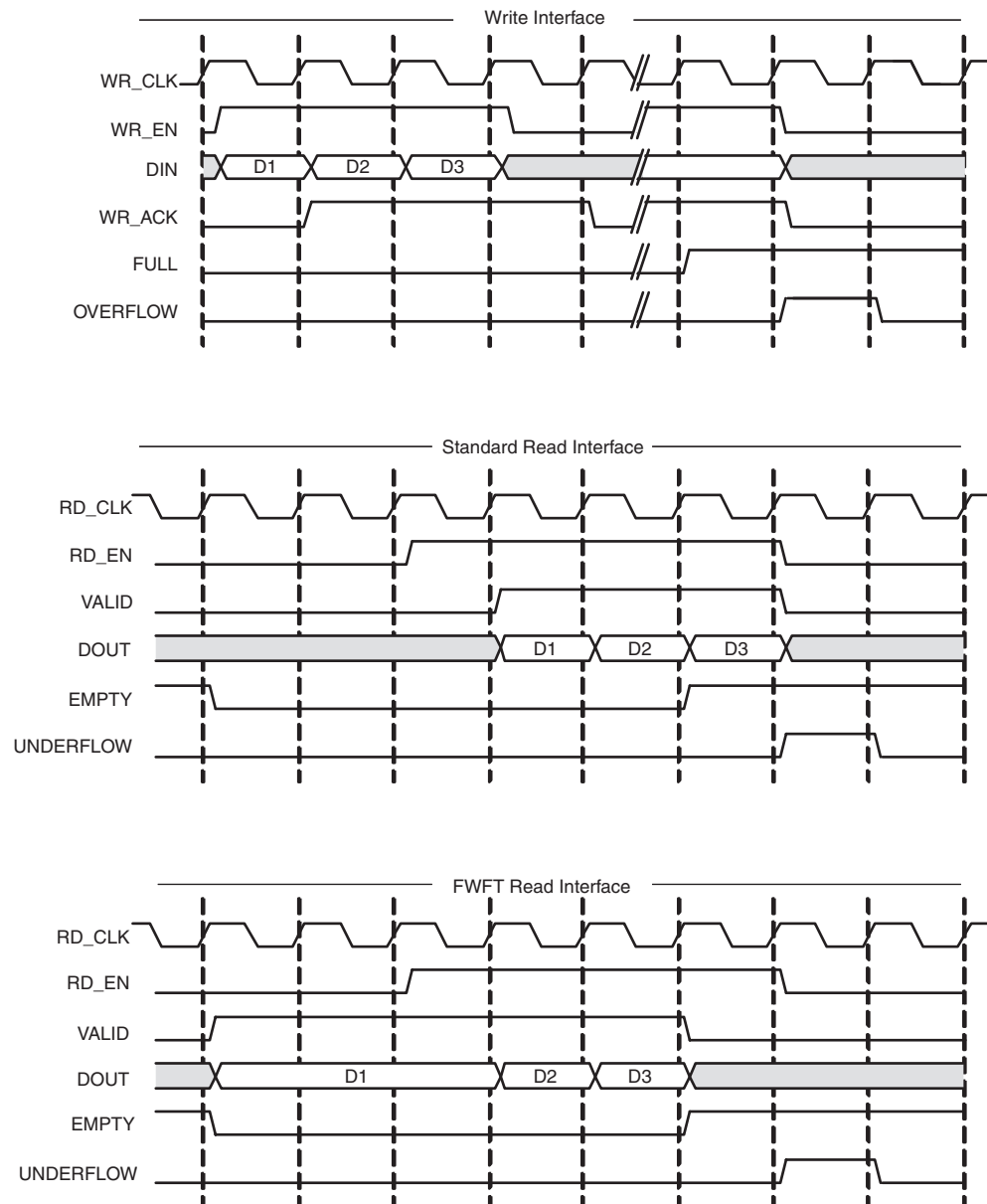


Figure 4-10: Handshaking Signals for a FIFO with Independent Clocks

Underflow

The underflow flag (UNDERFLOW) is used to indicate that a read operation is unsuccessful. This occurs when a read is initiated and the FIFO is empty. This flag is synchronous with the read clock (RD_CLK). Underflowing the FIFO does not change the state of the FIFO (it is non-destructive).

Overflow

The overflow flag (**OVERFLOW**) is used to indicate that a write operation is unsuccessful. This flag is asserted when a write is initiated to the FIFO while **FULL** is asserted. The overflow flag is synchronous to the write clock (**WR_CLK**). Overflowing the FIFO does not change the state of the FIFO (it is non-destructive).

Example Operation

Figure 4-11 illustrates the Handshaking flags. On the write interface, **FULL** is deasserted and therefore writes to the FIFO are successful (indicated by the assertion of **WR_ACK**). When a write occurs after **FULL** is asserted, **WR_ACK** is deasserted and **OVERFLOW** is asserted, indicating an overflow condition. On the read interface, once the FIFO is not **EMPTY**, the FIFO accepts read requests. Following a read request, **VALID** is asserted and **DOUT** is updated. When a read request is issued while **EMPTY** is asserted, **VALID** is deasserted and **UNDERFLOW** is asserted, indicating an underflow condition.

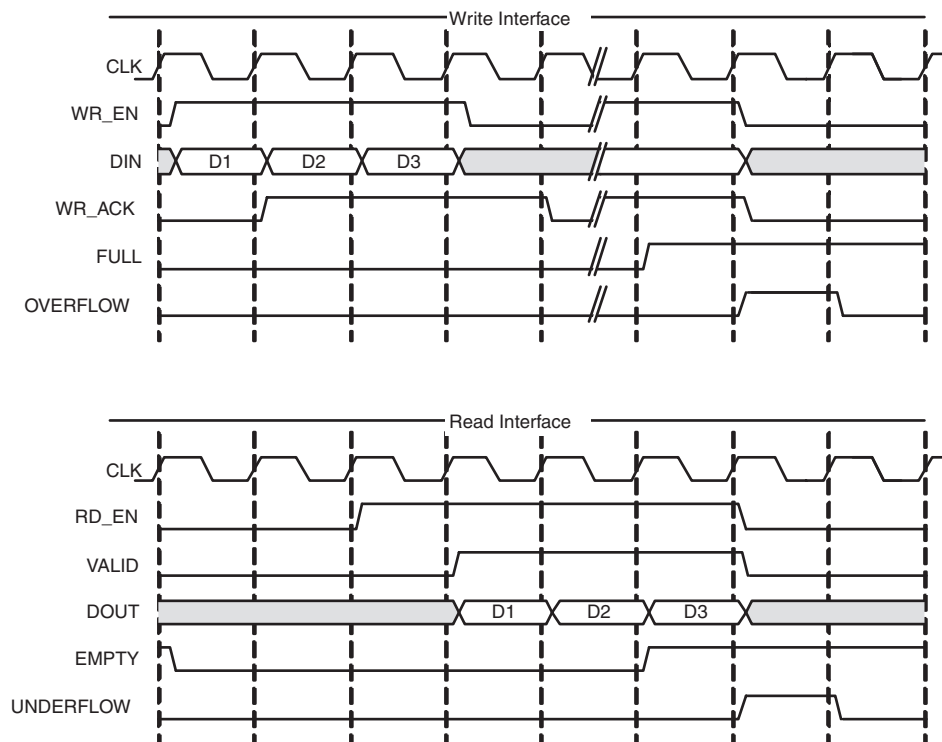


Figure 4-11: Handshaking Signals for a FIFO with Common Clocks

Programmable Flags

The FIFO supports programmable flags to indicate that the FIFO has reached a user-defined fill level.

- Programmable full (**PROG_FULL**) indicates that the FIFO has reached a user-defined full threshold.
- Programmable empty (**PROG_EMPTY**) indicates that the FIFO has reached a user-defined empty threshold.

For these thresholds, the user can set a constant value or choose to have dedicated input ports, enabling the thresholds to change dynamically in circuit. Hysteresis is also optionally supported, by providing unique assert and negate values for each flag. Detailed information about these options are provided below.

Programmable Full

The FIFO Generator supports four ways to define the programmable full threshold:

- Single threshold constant
- Single threshold with dedicated input port
- Assert and negate threshold constants (provides hysteresis)
- Assert and negate thresholds with dedicated input ports (provides hysteresis)

Note: The built-in FIFOs only support single-threshold constant programmable full.

These options are available in the CORE Generator GUI and accessed within the programmable flags window (Figure 3-4).

The programmable full flag (`PROG_FULL`) is asserted when the number of entries in the FIFO is greater than or equal to the user-defined assert threshold. When the programmable full flag is asserted, the FIFO can continue to be written to until the full flag (`FULL`) is asserted. If the number of words in the FIFO is less than the negate threshold, the flag is deasserted.

Note: If a write operation occurs on a rising clock edge that causes the number of words to meet or exceed the programmable full threshold, then the programmable full flag will assert on the next rising clock edge. The deassertion of the programmable full flag has a longer delay, and depends on the relationship between the write and read clocks.

Programmable Full: Single Threshold

This option enables the user to set a single threshold value for the assertion and deassertion of `PROG_FULL`. When the number of entries in the FIFO is greater than or equal to the threshold value, `PROG_FULL` is asserted. When the number of entries in the FIFO is less than the threshold value, `PROG_FULL` is deasserted.

There are two options for implementing this threshold:

- **Single threshold constant.** User specifies the threshold value through the CORE Generator GUI. Once the core is generated, this value can only be changed by re-generating the core. This option consumes fewer resources than the single threshold with dedicated input port.
- **Single threshold with dedicated input port.** User specifies the threshold value through an input port (`PROG_FULL_THRESH`) on the core. This input can be changed while the FIFO is in reset, providing the user the flexibility to change the programmable full threshold in-circuit without re-generating the core.

Note: Refer to the CORE Generator GUI for valid ranges for each threshold.

Figure 4-12 shows the programmable full flag with a single threshold. The user writes to the FIFO until there are seven words in the FIFO. Since the programmable full threshold is set to seven, the FIFO asserts `PROG_FULL` once seven words are written into the FIFO.

Note that both write data count (WR_DATA_COUNT) and PROG_FULL have one clock cycle of delay. Once the FIFO has six or fewer words in the FIFO, PROG_FULL is deasserted.

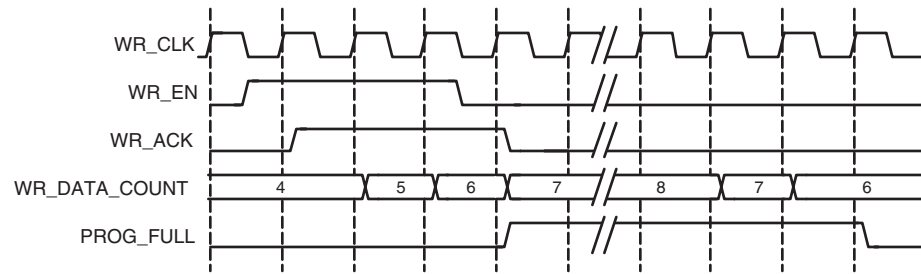


Figure 4-12: Programmable Full Single Threshold: Threshold Set to 7

Programmable Full: Assert and Negate Thresholds

This option enables the user to set separate values for the assertion and deassertion of PROG_FULL. When the number of entries in the FIFO is greater than or equal to the assert value, PROG_FULL is asserted. When the number of entries in the FIFO is less than the negate value, PROG_FULL is deasserted.

There are two options for implementing these thresholds:

- Assert and negate threshold constants: User specifies the threshold values through the CORE Generator GUI. Once the core is generated, these values can only be changed by re-generating the core. This option consumes fewer resources than the assert and negate thresholds with dedicated input ports.
- Assert and negate thresholds with dedicated input ports: User specifies the threshold values through input ports on the core. These input ports can be changed while the FIFO is in reset, providing the user the flexibility to change the values of the programmable full assert (PROG_FULL_THRESH_ASSERT) and negate (PROG_FULL_THRESH_NEGATE) thresholds in-circuit without re-generating the core.

Note: The full assert value must be larger than the full negate value. Refer to the CORE Generator GUI for valid ranges for each threshold.

Figure 4-13 shows the programmable full flag with assert and negate thresholds. The user writes to the FIFO until there are 10 words in the FIFO. Because the assert threshold is set to 10, the FIFO then asserts PROG_FULL. The negate threshold is set to seven, and the FIFO deasserts PROG_FULL once six words or fewer are in the FIFO. Both write data count (WR_DATA_COUNT) and PROG_FULL have one clock cycle of delay.

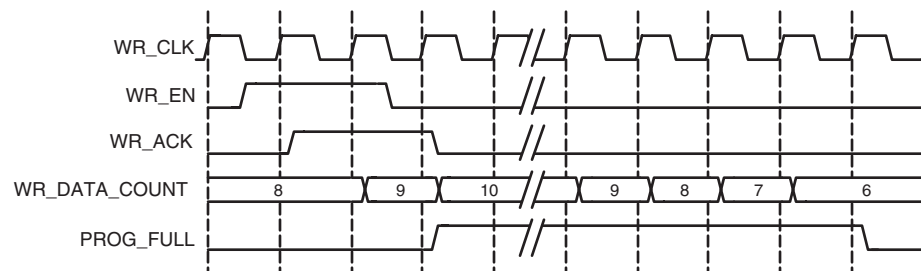


Figure 4-13: Programmable Full with Assert and Negate Thresholds: Assert Set to 10 and Negate Set to 7

Programmable Empty

The FIFO Generator supports four ways to define the programmable empty thresholds.

- Single threshold constant
- Single threshold with dedicated input port
- Assert and negate threshold constants (provides hysteresis)
- Assert and negate thresholds with dedicated input ports (provides hysteresis)

Note: The built-in FIFOs only support single-threshold constant programmable full.

These options are available in the CORE Generator GUI and accessed within the programmable flags window (Figure 3-4).

The programmable empty flag (PROG_EMPTY) is asserted when the number of entries in the FIFO is less than or equal to the user-defined assert threshold. If the number of words in the FIFO is greater than the negate threshold, the flag is deasserted.

Note: If a read operation occurs on a rising clock edge that causes the number of words in the FIFO to be equal to or less than the programmable empty threshold, then the programmable empty flag will assert on the next rising clock edge. The deassertion of the programmable empty flag has a longer delay, and depends on the read and write clocks.

Programmable Empty: Single Threshold

This option enables the user to set a single threshold value for the assertion and deassertion of PROG_EMPTY. When the number of entries in the FIFO is less than or equal to the threshold value, PROG_EMPTY is asserted. When the number of entries in the FIFO is greater than the threshold value, PROG_EMPTY is deasserted.

There are two options for implementing this threshold.

- **Single threshold constant:** User specifies the threshold value through the CORE Generator GUI. Once the core is generated, this value can only be changed by re-generating the core. This option consumes fewer resources than the single threshold with dedicated input port.
- **Single threshold with dedicated input port:** User specifies the threshold value through an input port (PROG_EMPTY_THRESH) on the core. This input can be changed while the FIFO is in reset, providing the user the flexibility to change the programmable empty threshold in-circuit without re-generating the core.

Note: Refer to the CORE Generator GUI for valid ranges for each threshold.

Figure 4-14 shows the programmable empty flag with a single threshold. The user writes to the FIFO until there are five words in the FIFO. Since the programmable empty threshold is set to four, PROG_EMPTY is asserted until more than four words are present in the FIFO. Once five words (or more) are present in the FIFO, PROG_EMPTY is deasserted. Both read data count (RD_DATA_COUNT) and PROG_EMPTY have one clock cycle of delay.

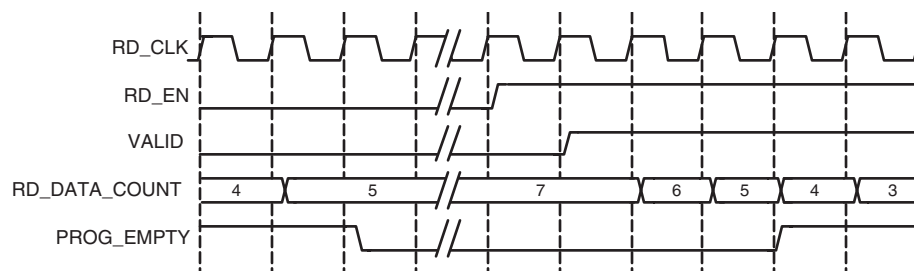


Figure 4-14: Programmable Empty with Single Threshold: Threshold Set to 4

Programmable Empty: Assert and Negate Thresholds

This option enables the user to set separate values for the assertion and deassertion of `PROG_EMPTY`. When the number of entries in the FIFO is less than or equal to the assert value, `PROG_EMPTY` is asserted. When the number of entries in the FIFO is greater than the negate value, `PROG_EMPTY` is deasserted.

There are two options for implementing the assert and negate thresholds.

- **Assert and negate threshold constants.** The threshold values are specified through the CORE Generator GUI. Once the core is generated, these values can only be changed by re-generating the core. This option consumes fewer resources than the assert and negate thresholds with dedicated input ports.
- **Assert and negate thresholds with dedicated input ports.** The threshold values are specified through input ports on the core. These input ports can be changed while the FIFO is in reset, providing the user the flexibility to change the values of the programmable empty assert (`PROG_EMPTY_THRESH_ASSERT`) and negate (`PROG_EMPTY_THRESH_NEGATE`) thresholds in-circuit without regenerating the core.

Note: The empty assert value must be less than the empty negate value. Refer to the CORE Generator GUI for valid ranges for each threshold.

Figure 4-15 shows the programmable empty flag with assert and negate thresholds. The user writes to the FIFO until there are eleven words in the FIFO. Since the programmable empty deassert value is set to ten, `PROG_EMPTY` is deasserted when there are more than ten words in the FIFO. Once the FIFO contains less than or equal to the programmable empty negate value (set to seven), `PROG_EMPTY` is asserted. Both read data count (`RD_DATA_COUNT`) and `PROG_EMPTY` have one clock cycle of delay.

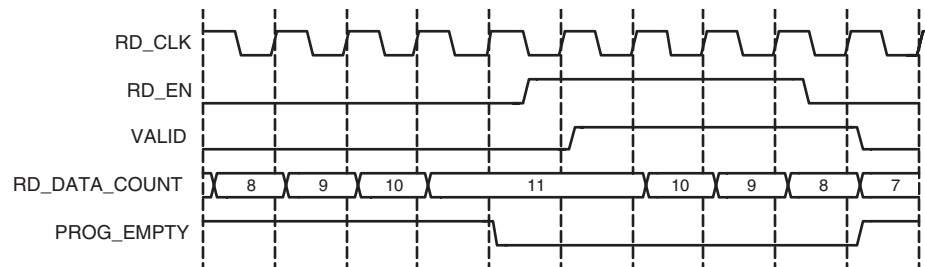


Figure 4-15: Programmable Empty with Assert and Negate Thresholds: Assert Set to 7 and Negate Set to 10

Programmable Empty for First-Word Fall-Through

For FWFT FIFOs, `PROG_EMPTY` is guaranteed to be asserted when the number of words in the FIFO is less than or equal to the programmable empty assert threshold. Under certain conditions, it is possible for `PROG_EMPTY` to violate this rule, but only while `EMPTY` is asserted. `PROG_EMPTY` should be assumed to be asserted whenever `EMPTY` is asserted.

Data Counts

`DATA_COUNT` tracks the number of words in the FIFO. You can specify the width of the data count bus with a maximum width of \log_2 (FIFO depth). If the width specified is smaller than the maximum allowable width, the bus is truncated by removing the lower

bits. These signals are optional outputs of the FIFO Generator, and are enabled through the CORE Generator GUI. Table 4-4 identifies FWFT support for each FIFO implementation.

Table 4-4: Implementation-specific Support for Data Counts

FIFO Implementation		FWFT Support
Independent Clocks	Block RAM	✓
	Distributed RAM	✓
	Built-in	
Common Clock	Block RAM	✓
	Distributed RAM	✓
	Shift Register	✓
	Built-in	

Read Data Count

Read data count (RD_DATA_COUNT) pessimistically reports the number of words available for reading. The count is guaranteed to never over-report the number of words available in the FIFO (although it may temporarily under-report the number of words available) to ensure that the user never underflows the FIFO. The user can specify the width of the read data count bus with a maximum width of log2 (read depth). If the width specified is smaller than the maximum allowable width, the bus is truncated with the lower bits removed.

For example, the user can specify to use two bits out of a maximum allowable three bits (provided a FIFO depth of eight). These two bits indicate the number of words in the FIFO, with a quarter resolution. This provides a status of the contents of the FIFO for the read clock domain.

Note: If a read operation occurs on a rising clock edge of RD_CLK, that read is reflected on the RD_DATA_COUNT signal following the next rising clock edge. A write operation on the WR_CLK clock domain may take a number of clock cycles before being reflected in the RD_DATA_COUNT.

Write Data Count

Write data count (WR_DATA_COUNT) pessimistically reports the number of words written into the FIFO. The count is guaranteed to never under-report the number of words in the FIFO (although it may temporarily over-report the number of words present) to ensure that the user never overflows the FIFO. The user can specify the width of the write data count bus with a maximum width of log2 (write depth). If the width specified is smaller than the maximum allowable width, the bus is truncated with the lower bits removed.

For example, you can only use two bits out of a maximum allowable three bits (provided a FIFO depth of eight). These two bits indicate the number of words in the FIFO, with a quarter resolution. This provides a status of the contents of the FIFO for the write clock domain.

Note: If a write operation occurs on a rising clock edge of WR_CLK, that write will be reflected on the WR_DATA_COUNT signal following the next rising clock edge. A read operation, which occurs on the RD_CLK clock domain, may take a number of clock cycles before being reflected in the WR_DATA_COUNT.

First-Word Fall-Through Data Count

By providing the capability to read the next data word before requesting it, FWFT implementations increase the depth of the FIFO. This depth increase causes the FWFT data counts to be estimates of the number of words in the FIFO, not exact representations of the number of words in the FIFO. By selecting “Use extra logic for more accurate Data Counts” on page 4 of the CORE Generator GUI, an extra bit is added to WR_DATA_COUNT and RD_DATA_COUNT to allow them to accurately express the full depth of the FIFO. When this option is selected, RD_DATA_COUNT and WR_DATA_COUNT behave as described in the sections above. When this option is not selected, the write and read data counts are only estimates of the contents of the FIFO.

For example, an independent clocks FIFO with a user-selected input and output depth of eight, has an actual depth (as reported on the summary page of the GUI) of seven. If FWFT is selected, the additional register stages increase the FIFO’s effective depth to nine. In this case, selecting “use extra logic for more accurate data counts” will increase the data count counter sizes to display the full range of the FIFO (zero to nine words).

Note: For FWFT implementations, WR_DATA_COUNT is guaranteed to be accurate when words are present in the FIFO, but may be incorrect by 2 at or near empty.

Example Operation

Figure 4-16 shows write and read data counts. When WR_EN is asserted and FULL is deasserted, WR_DATA_COUNT increments. Similarly, when RD_EN is asserted and EMPTY is deasserted, RD_DATA_COUNT decrements.

Note: In the first part of Figure 4-16, a successful write operation occurs on the third rising clock edge, and is not reflected on WR_DATA_COUNT until the next full clock cycle is complete. Similarly, RD_DATA_COUNT transitions one full clock cycle after a successful read operation.

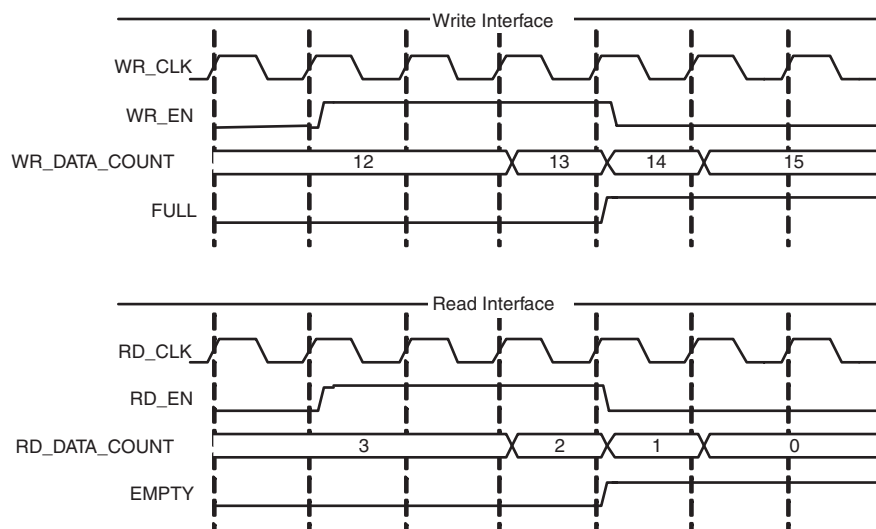


Figure 4-16: Write and Read Data Counts for FIFO with Independent Clocks

Non-symmetric Aspect Ratios

Table 4-5 identifies the FWFT support for non-symmetric aspect ratios.

Table 4-5: Implementation-specific Support for Non-symmetric Aspect Ratios

FIFO Implementation		FWFT Support
Independent Clocks	Block RAM	✓
	Distributed RAM	
	Built-in	
Common Clock	Block RAM	
	Distributed RAM	
	Shift Register	
	Built-in	

This feature is supported for FIFOs configured with independent clocks implemented with block RAM. Non-symmetric aspect ratios allow the input and output depths of the FIFO to be different. The following write-to-read aspect ratios are supported: 1:8, 1:4, 1:2, 1:1, 2:1, 4:1, 8:1. This feature is enabled by selecting unique write and read widths when customizing the FIFO using the CORE Generator. By default, the write and read widths are set to the same value (providing a 1:1 aspect ratio); but any ratio between 1:8 to 8:1 is supported, and the output depth of the FIFO is automatically calculated from the input depth and the write and read widths.

For non-symmetric aspect ratios, the full and empty flags are active only when one complete word can be written or read. The FIFO does not allow partial words to be accessed. For example, assuming a full FIFO, if the write width is 8 bits and read width is 2 bits, the user would have to complete four valid read operations before full deasserts and a write operation accepted. Write data count shows the number of FIFO words according to the write port ratio, and read data count shows the number of FIFO words according to the read port ratio.

Note: For non-symmetric aspect ratios where the write width is smaller than the read width (1:8, 1:4, 1:2), the most significant bits are read first (refer to Figure 4-17 and Figure 4-18).

Figure 4-17 is an example of a FIFO with a 1:4 aspect ratio (write width = 2, read width = 8). In this figure, four consecutive write operations are performed before a read operation can be performed. The first write operation is "10", followed by "11", "00", and finally "01." The memory is filling up from the right to the left (LSB to MSB). When a read operation is performed, the received data is "01_00_11_10."

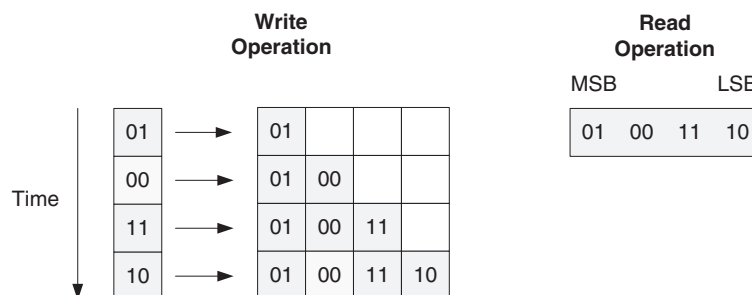


Figure 4-17: 1:4 Aspect Ratio: Data Ordering

Figure 4-18 shows DIN, DOUT and the handshaking signals for a FIFO with a 1:4 aspect ratio. After four words are written into the FIFO, EMPTY is deasserted. Then after a single read operation, EMPTY is asserted again.

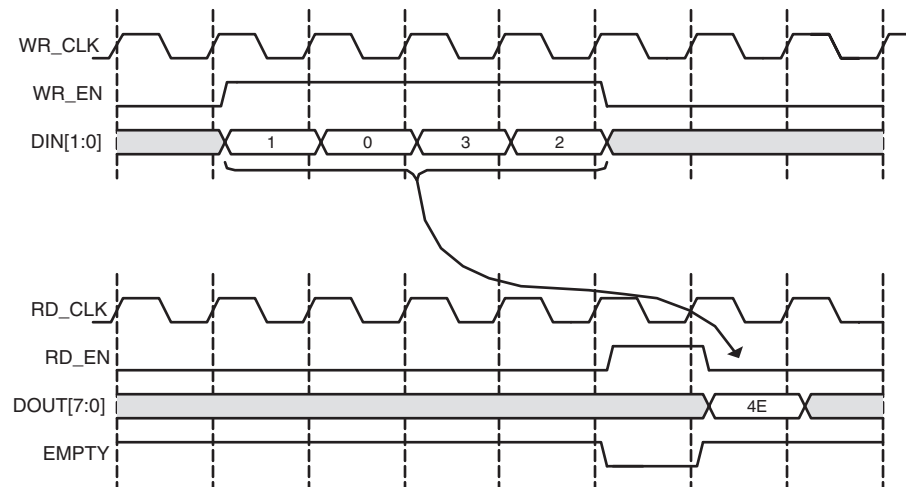


Figure 4-18: 1:4 Aspect Ratio: Status Flag Behavior

Figure 4-19 shows a FIFO with an aspect ratio of 4:1 (write width of 8, read width of 2). In this example, a single write operation is performed, after which four read operations are executed. The write operation is “11_00_01_11.” When a read operation is performed, the data is received left to right (MSB to LSB). As shown, the first read results in data of “11”, followed by “00”, “01”, and then “11.”

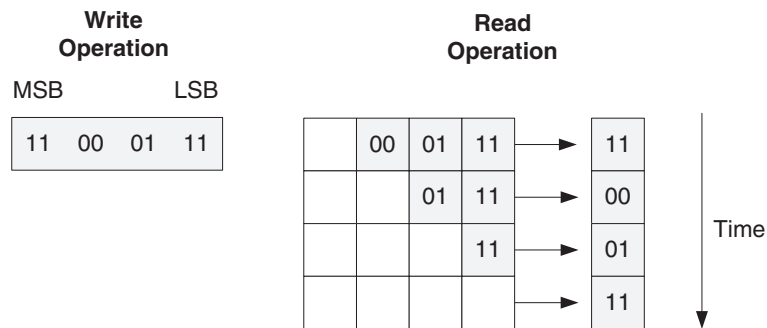


Figure 4-19: 4:1 Aspect Ratio: Data Ordering

Figure 4-20 shows DIN, DOUT, and the handshaking signals for a FIFO with an aspect ratio of 4:1. After a single write, the FIFO deasserts EMPTY. Since no other writes occur, the FIFO reasserts empty after four reads.

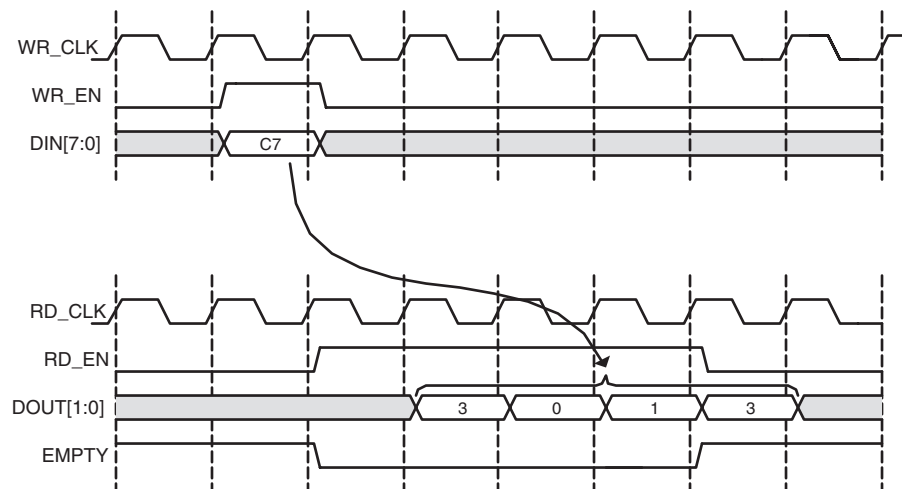


Figure 4-20: 4:1 Aspect Ratio: Status Flag Behavior

Reset Behavior

The FIFO Generator provides a single reset input that resets all counters, output registers, and memories when asserted. For block RAM or distributed RAM implementations, resetting the FIFO is not required, and the reset pin can be disabled in the FIFO. There are two reset options: asynchronous and synchronous.

Asynchronous Reset

The asynchronous reset (RST) input asynchronously resets all counters, output registers, and memories when asserted. When reset is implemented, it is synchronized internally to the core with each respective clock domain for setting the internal logic of the FIFO to a known state. This synchronization logic allows for proper timing of the reset logic within the core to avoid glitches and metastable behavior.

Due to the synchronization logic used, there is a latency in the deassertion of the FULL, ALMOST_FULL, and PROG_FULL signals. Figure 4-22 illustrates the reset behavior for a FIFO with an independent clock, and Figure 4-22 illustrates the reset behavior for a FIFO with a common clock.

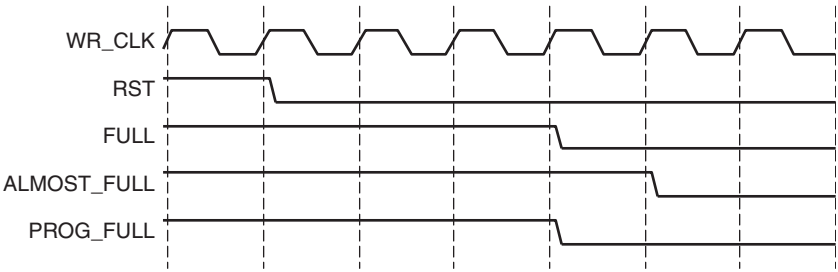


Figure 4-21: Synchronous Reset: FIFO with Independent Clock

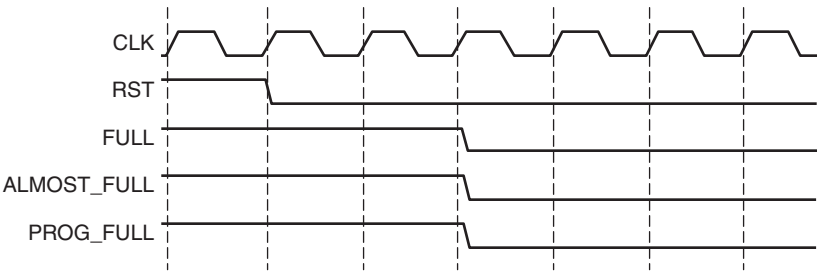


Figure 4-22: Asynchronous Reset: FIFO with Common Clock

Table 4-6 defines the values of the output ports during power-up and the reset state. DOUT reset value is supported for all architectures with the exception of Virtex and Spartan-II. If the user does not specify a DOUT reset value, it defaults to 0. The FIFO requires a reset pulse of only 1 clock cycle. FULL and ALMOST_FULL flags are asserted during reset to ensure that no write operations occur, and handshaking signals are deasserted during reset. Note that the power-up values for asynchronous reset are different than the reset state value.

The FIFO requires a reset pulse of at least three clock cycles in length. Independent clock FIFOs require four cycles post-reset prior to FIFO write availability—this is necessary for proper reset synchronization. Common clock FIFOs, however, require three cycles after reset prior to FIFO write availability. For the built-in FIFO, the RE_EN and WR_EN signal is required to be deasserted during reset.

Table 4-6: FIFO Reset Values

Signal	Block Memory Distributed Memory & Shift Register Values	Built-in FIFO Reset Values	Power-up Values
DOUT	DOUT Reset Value or 0 ¹	Content of memory at location 0	Same as reset values
FULL	1 ²	0	0
ALMOST FULL	1 ²	N/A	0

Table 4-6: FIFO Reset Values (Continued)

EMPTY	1	0	1
ALMOST EMPTY	1	N/A	1
VALID	0 (active high) or 1 (active low)	0 (active high) or 1 (active low)	0 (active high) or 1 (active low)
UNDERFLOW	0 (active high) or 1 (active low) ³	0 (active high) or 1 (active low)	0 (active high) or 1 (active low)
WR_ACK	0 (active high) or 1 (active low)	0 (active high) or 1 (active low)	0 (active high) or 1 (active low)
OVERFLOW	0 (active high) or 1 (active low) ⁴	0 (active high) or 1 (active low)	0 (active high) or 1 (active low)
PROG_FULL	1 ²	0	0
PROG_EMPTY	1	1	1
RD_DATA_COUNT	0	N/A	0
WR_DATA_COUNT	0	N/A	0

1. The ability to set DOUT to a user-defined value is not available for block RAM implementations in Virtex, Spartan-II and Spartan-IIE. DOUT resets to 0 when this feature is unavailable.
2. When reset is asserted, the FULL flags are asserted to prevent writes to the FIFO during reset.
3. The underflow signal is dependent on RD_EN. If RD_EN is asserted and the FIFO is empty, underflow is asserted.
4. The overflow signal is dependent on WR_EN. If WR_EN is asserted and the FIFO is full, overflow is asserted.

Synchronous Reset

The synchronous reset (SRST) input is only available for the Block and Distributed RAM implementation of the common clock FIFO. It synchronously resets all counters, output registers and memories when asserted. Because the reset pin is synchronous to the input clock and there is only one clock domain in the FIFO, no additional synchronization logic is necessary. Figure 4-23 illustrates the flags following the release of SRST.

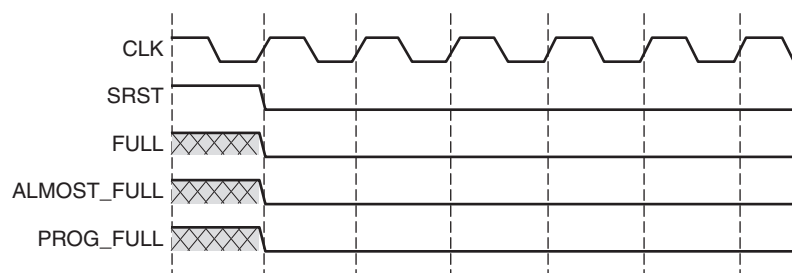


Figure 4-23: Synchronous Reset: FIFO with a Common Clock

Table 4-7 defines the values of the output ports during power-up and the reset state. DOUT reset value is supported for all architectures with the exception of Virtex and Spartan-II. If the user does not specify a DOUT reset value, it defaults to 0. The FIFO requires a reset pulse of only 1 clock cycle. The FIFOs are available for transaction on the clock cycle after

the reset is released. The power-up values for the synchronous reset are the same as the reset state.

Table 4-7: FIFO Reset and Power-up Values

Signal	Block Memory and Distributed Memory Values of Output Ports During Reset and Power-up
DOUT	DOUT Reset Value or 0 ¹
FULL	0
ALMOST FULL	0
EMPTY	1
ALMOST EMPTY	1
VALID	0 (active high) or 1 (active low)
UNDERFLOW	0 (active high) or 1 (active low) ²
WR_ACK	0 (active high) or 1 (active low)
OVERFLOW	0 (active high) or 1 (active low) ³
PROG_FULL	0
PROG_EMPTY	0
RD_DATA_COUNT	0
WR_DATA_COUNT	0

1. The ability to set DOUT to a user-defined value is not available for block RAM implementations in Virtex, Spartan-II, and Spartan-III. DOUT resets to 0 when this feature is unavailable.
2. The underflow signal is dependent on RD_EN. If RD_EN is asserted and the FIFO is empty, underflow is asserted.
3. The overflow signal is dependent on WR_EN. If WR_EN is asserted and the FIFO is full, overflow is asserted.

Special Design Considerations

This chapter provides additional design considerations for using the FIFO Generator core.

Resetting the FIFO

The FIFO Generator must be reset after the FPGA is configured and before operation begins. Two reset pins are available, asynchronous (RST) and synchronous (SRST), and both clear the internal counters and output registers.

- For asynchronous reset, internal to the core, RST is synchronized to the clock domain in which it is used, to ensure that the FIFO initializes to a known state. This synchronization logic allows for proper reset timing of the core logic, avoiding glitches and metastable behavior.
- For synchronous reset, because the reset pin is synchronous to the input clock and there is only one clock domain in the FIFO, no additional synchronization logic is needed.

The generated FIFO core will be initialized after reset to a known state. For details about reset values and behavior, see “Reset Behavior” in Chapter 4 of this guide.

Continuous Clocks

The FIFO Generator is designed to work only with free-running write and read clocks. Xilinx does not recommend controlling the core by manipulating RD_CLK and WR_CLK. If this functionality is required to gate FIFO operation, we recommend using the write enable (WR_EN) and read enable (RD_EN) signals.

Pessimistic Full and Empty

When independent clock domains are selected, the full flag (FULL, ALMOST_FULL) and empty flag (EMPTY, ALMOST_EMPTY) are pessimistic flags. FULL and ALMOST_FULL are synchronous to the write clock (WR_CLK) domain, while EMPTY and ALMOST_EMPTY are synchronous to the read clock (RD_CLK) domain.

The full flags are considered pessimistic flags because they assume that no read operations have taken place in the read clock domain. ALMOST_FULL is guaranteed to be asserted on the rising edge of WR_CLK when there is only one available location in the FIFO, and FULL is guaranteed to be asserted on the rising edge of WR_CLK when the FIFO is full. There may be a number of clock cycles between a read operation and the deassertion of FULL. The precise number of clock cycles for FULL to deassert is not predictable due to the crossing of clock domains and synchronization logic.

The EMPTY flags are considered pessimistic flags because they assume that no write operations have taken place in the write clock domain. ALMOST_EMPTY is guaranteed to be

asserted on the rising edge of RD_CLK when there is only one more word in the FIFO, and EMPTY is guaranteed to be asserted on the rising edge of RD_CLK when the FIFO is empty. There may be a number of clock cycles between a write operation and the deassertion of EMPTY. The precise number of clock cycles for EMPTY to deassert is not predictable due to the crossing of clock domains and synchronization logic.

See [Chapter 4, “Designing with the Core,”](#) for detailed information about the latency and behavior of the full and empty flags.

Programmable Full and Empty

The programmable full (PROG_FULL) and programmable empty (PROG_EMPTY) flags provide the user flexibility in specifying when the programmable flags assert and deassert. These flags can be set either by constant value(s) or by input port(s). These signals differ from the full and empty flags because they assert one (or more) clock cycle *after* the assert threshold has been reached. These signals are deasserted some time after the negate threshold has been passed. In this way, PROG_EMPTY and PROG_FULL are also considered pessimistic flags. See [“Programmable Flags” in Chapter 4](#) of this guide for more information about the latency and behavior of the programmable flags.

Write Data Count and Read Data Count

When independent clock domains are selected, write data count (WR_DATA_COUNT) and read data count (RD_DATA_COUNT) signals are provided as an indication of the number of words in the FIFO relative to the write or read clock domains, respectively.

Consider the following when using the WR_DATA_COUNT or RD_DATA_COUNT ports.

- The WR_DATA_COUNT and RD_DATA_COUNT outputs are not an instantaneous representation of the number of words in the FIFO, but can instantaneously provide an approximation of the number of words in the FIFO.
- WR_DATA_COUNT and RD_DATA_COUNT may skip values from clock cycle to clock cycle.
- Using non-symmetric aspect ratios, or running clocks which vary dramatically in frequency, will increase the disparity between the data count outputs and the actual number of words in the FIFO.

Note: The WR_DATA_COUNT and RD_DATA_COUNT outputs will always be correct after some period of time where RD_EN=0 and WR_EN=0 (generally, just a few clock cycles after read and write activity stops).

See [“Data Counts” in Chapter 4](#) of this guide for details about the latency and behavior of the data count flags.

Setup and Hold Time Violations

When generating a FIFO with independent clock domains, the core internally synchronizes the write and read clock domains. For this reason, setup and hold time violations are expected on certain registers within the core. In simulation, warning messages may be issued indicating these violations. If these warning messages are from the FIFO Generator core, they can be safely ignored. The core is designed to properly handle these conditions, regardless of the phase or frequency relationship between the write and read clocks.

Simulating Your Design

The FIFO Generator is provided as a Xilinx technology-specific netlist, and as a behavioral or structural simulation model. This chapter provides instructions for simulating the FIFO Generator in your design.

Simulation Models

The FIFO Generator supports two types of simulation models based on the Xilinx CORE Generator system project options. The models are available in both VHDL and Verilog. Both types of models are described in detail in this chapter.

To choose a model, do the following:

1. Open the CORE Generator.
2. Select Options from the Project drop-down list.
3. Click the Generation tab.
4. Choose to generate a behavioral model or a structural model.

Behavioral Models

Important! The behavioral models provided are designed to reproduce the behavior and functionality of the FIFO Generator, but are not cycle-accurate models (except for the common clock FIFO with block RAM, distributed RAM, or shift registers). If cycle accurate models are required (and common clock FIFO with block RAM, distributed RAM or shift registers is not selected), it is recommended to use the structural model.

The behavioral models are considered to be zero-delay models, as the modeled write-to-read latency is nearly zero. The behavioral models are functionally correct, and will represent the behavior of the configured FIFO, although the write-to-read latency and the behavior of the status flags will differ from the actual implementation of the FIFO design.

To generate behavioral models, select Behavioral and VHDL or Verilog in the Xilinx CORE Generator project options. Behavioral models are the default project options.

The following considerations apply to the behavioral models.

- Write operations always occur relative to the write clock (WR_CLK) or common clock (CLK) domain, as do the corresponding handshaking signals.
- Read operations always occur relative to the read clock (RD_CLK) or common clock (CLK) domain, as do the corresponding handshaking signals.
- The delay through the FIFO (write-to-read latency) will differ between the VHDL model, the Verilog model, and the core.
- The deassertion of the status flags (full, almost full, programmable full, empty, almost

empty, programmable empty) will vary between the VHDL model, the Verilog model, and the core.

Note: If independent clocks or common clocks with built-in FIFO is selected, it is strongly recommended to use the structural model, as the behavioral model does not correctly model the behavior of the status flags (full, programmable full, empty and programmable empty).

Structural Models

The structural models are designed to provide a more accurate model of FIFO behavior at the cost of simulation time. These models will provide a closer approximation of cycle accuracy across clock domains for asynchronous FIFOs. No asynchronous FIFO model can be 100% cycle accurate as physical relationships between the clock domains, including temperature, process, and frequency relationships, affect the domain crossing indeterminately.

To generate structural models, select Structural and VHDL or Verilog in the Xilinx CORE Generator project options.

Note: Simulation performance may be impacted when simulating the structural models compared to the behavioral models

Performance Information

Resource Utilization and Performance

Performance and resource utilization for a FIFO varies depending on the configuration and features selected when customizing the core. The tables below provide example FIFO configurations and the maximum performance and resources required.

[Table A-1](#) provides results for a FIFO configured without optional features. The benchmarks were performed using Virtex-II 2v3000, Virtex-4 4vlx15 -11, and Virtex-5 5vlx50-2 target devices.

Table A-1: Benchmarks: FIFO Configured without Optional Features

FIFO Type	Depth x Width	Family	Performance (MHz)	Resources				
				LUTs	FFs	Block RAM	Shift Register	Distributed RAM
Synchronous FIFO (Block RAM)	512 x 16	Virtex-5	350 MHz	62	45	1	0	0
		Virtex-4	200 MHz	64	43	1	0	0
		Virtex-II	175 MHz	64	43	1	0	0
Synchronous FIFO (Block RAM)	4096 x 16	Virtex-5	350 MHz	84	57	2	0	0
		Virtex-4	225 MHz	80	55	4	0	0
		Virtex-II	175 MHz	80	55	4	0	0
Synchronous FIFO (Distributed RAM)	64 x 16	Virtex-5	475 MHz	48	61	0	0	32
		Virtex-4	250 MHz	89	47	0	0	128
		Virtex-II	225 MHz	89	47	0	0	128
Synchronous FIFO (Distributed RAM)	512 x 16	Virtex-5	375 MHz	85	80	0	0	256
		Virtex-4	200 MHz	340	59	0	0	1024
		Virtex-II	150 MHz	340	59	0	0	1024
Independent Clocks FIFO (Block RAM)	512x 16	Virtex-5	350 MHz	105	124	1	0	0
		Virtex-4	250 MHz	103	105	1	0	0
		Virtex-II	200 MHz	103	105	1	0	0

Table A-1: Benchmarks: FIFO Configured without Optional Features (Continued)

FIFO Type	Depth x Width	Family	Performance (MHz)	Resources				
				LUTs	FFs	Block RAM	Shift Register	Distributed RAM
Independent Clocks FIFO (Block RAM)	4096 x 16	Virtex-5	350 MHz	147	163	2	0	0
		Virtex-4	250 MHz	134	138	4	0	0
		Virtex-II	175 MHz	134	138	4	0	0
Independent Clocks FIFO (Distributed RAM)	64 x 16	Virtex-5	475 MHz	76	104	0	0	32
		Virtex-4	325 MHz	112	100	0	0	128
		Virtex-II	225 MHz	112	100	0	0	128
Independent Clocks FIFO (Distributed RAM)	512 x 16	Virtex-5	350 MHz	126	148	0	0	256
		Virtex-4	225 MHz	382	139	0	0	1024
		Virtex-II	150 MHz	382	139	0	0	1024
Shift Register FIFO	64 x 16	Virtex-5	450 MHz	59	50	0	32	0
		Virtex-4	200 MHz	68	43	0	64	0
		Virtex-II	150 MHz	68	43	0	64	0
Shift Register FIFO	512 x 16	Virtex-5	275 MHz	159	67	0	256	0
		Virtex-4	200 MHz	312	82	0	512	0
		Virtex-II	150 MHz	312	82	0	512	0

Table A-2 provides results for FIFOs configured with multiple programmable thresholds. The benchmarks were performed using Virtex-II 2v3000, Virtex-4 4vlx15 -11, and Virtex-5 5vlx50-2 target devices.

Table A-2: Benchmarks: FIFO Configured with Multiple Programmable Thresholds

FIFO Type	Depth x Width	Family	Performance (MHz)	Resources				
				LUTs	FFs	Block RAM	Shift Register	Distributed RAM
Synchronous FIFO (Block RAM)	512 x 16	Virtex-5	350 MHz	91	70	1	0	0
		Virtex-4	200 MHz	109	68	1	0	0
		Virtex-II	175 MHz	109	68	1	0	0
Synchronous FIFO (Block RAM)	4096 x 16	Virtex-5	350 MHz	124	88	2	0	0
		Virtex-4	225 MHz	135	86	4	0	0
		Virtex-II	175 MHz	135	86	4	0	0
Synchronous FIFO (Distributed RAM)	64 x 16	Virtex-5	475 MHz	75	90	0	0	32
		Virtex-4	250 MHz	112	66	0	0	128
		Virtex-II	225 MHz	112	66	0	0	128

Table A-2: Benchmarks: FIFO Configured with Multiple Programmable Thresholds (Continued)

FIFO Type	Depth x Width	Family	Performance (MHz)	Resources				
				LUTs	FFs	Block RAM	Shift Register	Distributed RAM
Synchronous FIFO (Distributed RAM)	512 x 16	Virtex-5	375 MHz	115	105	0	0	256
		Virtex-4	200 MHz	370	84	0	0	1024
		Virtex-II	150 MHz	370	84	0	0	1024
Independent Clocks FIFO (Block RAM)	512x 16	Virtex-5	350 MHz	132	143	1	0	0
		Virtex-4	250 MHz	163	126	1	0	0
		Virtex-II	200 MHz	163	126	1	0	0
Independent Clocks FIFO (Block RAM)	4096 x 16	Virtex-5	350 MHz	185	186	2	0	0
		Virtex-4	250 MHz	214	165	4	0	0
		Virtex-II	175 MHz	214	165	4	0	0
Independent Clocks FIFO (Distributed RAM)	64 x 16	Virtex-5	475 MHz	96	123	0	0	32
		Virtex-4	325 MHz	134	115	0	0	128
		Virtex-II	225 MHz	134	115	0	0	128
Independent Clocks FIFO (Distributed RAM)	512 x 16	Virtex-5	350 MHz	154	168	0	0	256
		Virtex-4	225 MHz	412	160	0	0	1024
		Virtex-II	150 MHz	412	160	0	0	1024
Shift Register FIFO	64 x 16	Virtex-5	450 MHz	68	63	0	32	0
		Virtex-4	200 MHz	97	62	0	64	0
		Virtex-II	150 MHz	97	62	0	64	0
Shift Register FIFO	512 x 16	Virtex-5	275 MHz	174	91	0	256	0
		Virtex-4	200 MHz	355	106	0	512	0
		Virtex-II	150 MHz	355	106	0	512	0

Table A-3 provides results for FIFOs configured to use the Virtex-5 built-in FIFO. The benchmarks were performed using a Virtex-5 5vlx50-2 target device.

Table A-3: Benchmarks: FIFO Configured with Virtex-5 FIFO36 Resources

FIFO Type	Depth x Width	Read Mode	Performance (MHz)	Resources		
				LUTs	FFs	FIFO36s
Synchronous FIFO36 (basic)	512 x 72	Standard	500	0	2	1
		FWFT	400	2	4	1
	16k x 8	Standard	350	10	6	4
		FWFT	375	13	10	4

Table A-3: Benchmarks: FIFO Configured with Virtex-5 FIFO36 Resources

FIFO Type	Depth x Width	Read Mode	Performance (MHz)	Resources		
				LUTs	FFs	FIFO36s
Synchronous FIFO36 (with handshaking)	512 x 72	Standard	500	2	6	1
		FWFT	400	4	6	1
	16k x 8	Standard	350	12	12	4
		FWFT	375	16	13	4
Independent Clocks FIFO36 (basic)	512 x 72	Standard	500	0	2	1
		FWFT	500	0	2	1
	16k x 8	Standard	375	6	2	4
		FWFT	375	6	2	4
Independent Clocks FIFO36 (with handshaking)	512 x 72	Standard	500	2	6	1
		FWFT	500	2	4	1
	16k x 8	Standard	350	8	8	4
		FWFT	350	9	5	4

Table A-4 provides results for FIFOs configured to use the Virtex-4 built-in FIFO with patch. The benchmarks were performed using a Virtex-4 4vlx15 -11 target device

Table A-4: Benchmarks: FIFO Configured with Virtex-4 FIFO16 Patch

FIFO Type	Depth x Width	Clock Ratios	Performance (MHz)	Resources			
				LUTs	FFs	Distributed RAMs	FIFO16s
Built-in FIFO (basic)	512 x 36	WR_CLK ≥ RD_CLK	375	110	129	72	1
		RD_CLK > WR_CLK	400	92	115	72	1
Built-in FIFO (with handshaking)	512 x 36	WR_CLK ≥ RD_CLK	375	113	134	72	1
		RD_CLK > WR_CLK	400	95	120	72	1

Core Parameters

FIFO Parameters

Table B-1 describes the FIFO core parameters, including the XCO file value and the default settings.

Table B-1: FIFO Parameter Table

Parameter Name	XCO File Values	Default GUI Setting
Component_Name	instance_name ASCII text starting with a letter and using the following character set: a-z, 0-9, and _	fifo_generator_v3_1
FIFO Implementation	Common_Clock_Block_RAM Common_Clock_Distributed_RAM Common_Clock_Shift_Register Common_Clock_Builtin_FIFO Independent_Clocks_Block_RAM Independent_Clocks_Distributed_RAM Independent_Clocks_Builtin_FIFO	Common_Clock_Block_RAM
Input Data Width ¹	Integer in range 1 to 256	16
Output Data Width ¹	Integer in range 1 to 256	16
Input Depth ¹	2 ^N where N is an integer 4 to 24	1024
Output Depth ¹	2 ^M where M is an integer 4 to 24	1024
Data Count Width	Integer in range 1 to log ₂ (Output Depth)	2
Read Clock Frequency	Integer 1 to 999 (MHz)	1
Write Clock Frequency	Integer 1 to 999 (MHz)	1
Almost Full Flag	true, false	false
Almost Empty Flag	true, false	false
Programmable Full Type	No_Programmable_Full_Threshold Single_Programmable_Full_Threshold_Constant Multiple_Programmable_Full_Threshold_Constants Single_Programmable_Full_Threshold_Input_Port Multiple_Programmable_Full_Threshold_Input_Ports	No_Programmable_Full_Threshold
Full Threshold Assert Value	See range under Programmable Flags.	768
Full Threshold Negate Value	See range under “Programmable Flags,” page 52.	768

Table B-1: FIFO Parameter Table (Continued)

Parameter Name	XCO File Values	Default GUI Setting
Programmable Empty Type	No_Programmable_Empty_Threshold Sigle_Programmable_Empty_Threshold_Constant Multiple_Programmable_Empty_Threshold_Constants Single_Programmable_Empty_Threshold_Input_Port Multiple_Programmable_Empty_Threshold_Input_Ports	No_Programmable_Empty_Threshold
Empty Threshold Assert Value	See range under “Programmable Flags,” page 52.	256
Empty Threshold Negate Value	See range under “Programmable Flags,” page 52.	256
Write Acknowledge Flag	true, false	false
Write Acknowledge Sense	Active_High, Active_Low	Active_High
Overflow Flag	true, false	false
Overflow Sense	Active_High, Active_Low	Active_High
Valid Flag	true, false	false
Valid Sense	Active_High, Active_Low	Active_High
Underflow Flag	true, false	false
Underflow Sense	Active_High, Active_Low	Active_High
Dout Reset Value	Hex value in range of 0 to output data width - 1	0
Primitive Depth	512, 1024, 2048, 4096	1024
Read Data Count	true, false	false
Read Data Count Width	Integer in range 1 to $\log^2(\text{output depth})$	10
Write Data Count	true, false	false
Write Data Count Width	Integer in range 1 to $\log^2(\text{input depth})$	10
Data Count	true, false	false
Performance Options	First_Word_Fall_Through, Standard_Fifo	Standard_Fifo
Read Latency	integer range 0 to 1	1
Reset Pin	true, false	true

1. User customized core should not exceed the number of shift registers, built-in FIFOs, block RAM, or distributed RAM primitives available in the targeted architecture. It is the user’s responsibility to know the resource availability in the targeted device.