

## 用于电源门控的 TPL5110 毫微功耗系统计时器

### 1 特性

- 电源电压范围为 1.8V 至 5.5V
- 电压为 2.5V 时，电流消耗为 35nA（典型值）
- 可选计时间隔：100ms 至 7200s
- 计时器精度：1%（典型值）
- 可通过电阻选择时间间隔
- 手动为 MOSFET 上电
- 单次触发功能

### 2 应用

- 电池供电系统
- 物联网 (IoT)
- 出入探测
- 篡改检测
- 家庭自动化传感器
- 温度调节装置
- 消费类电子产品
- 远程传感器
- 白色家电

### 3 说明

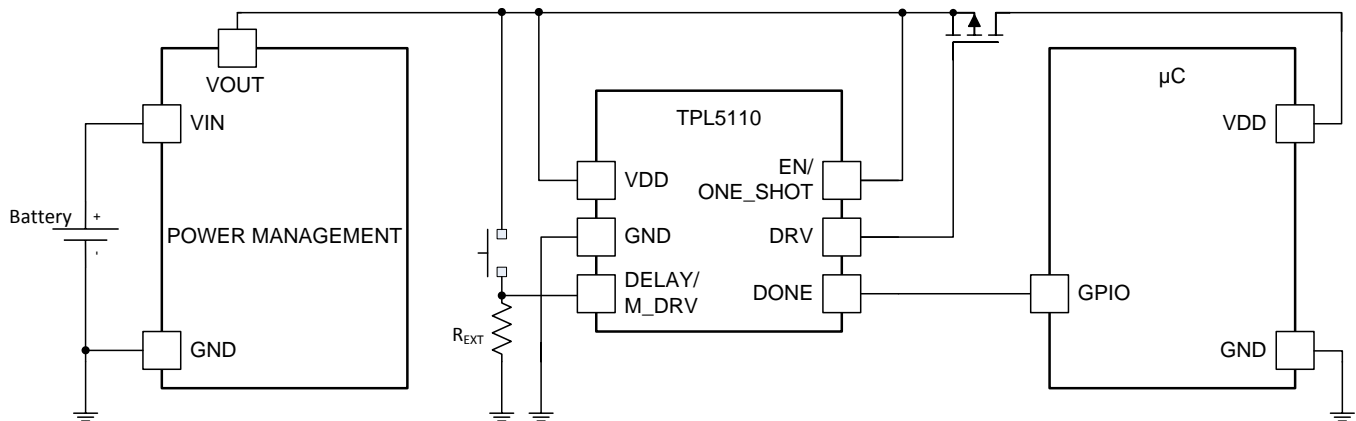
TPL5110 是一款低功耗计时器，具有集成 MOSFET 驱动器，专为占空比或电池供电型应用中的电源门控而设计。TPL5110 的电流消耗仅为 35nA，可用于支持电源线路，并可大幅降低睡眠期间的总系统待机电流。这一节能特性可以大幅减小能量采集或无线传感器应用中所使用的电池尺寸。TPL5110 提供 100ms 至 7200s 的可选计时间隔，适合用于电源门控应用。此外，TPL5110 还具有独特的单次触发功能，可使计时器仅为 MOSFET 供电一个周期。TPL5110 采用 6 引脚 SOT23 封装。

器件信息<sup>(1)</sup>

器件编号	封装	封装尺寸（标称值）
TPL5110	SOT23 (6)	3.00mm x 3.00mm

(1) 如需了解所有可用封装，请见数据表末尾的可订购产品附录。

简化应用电路原理图



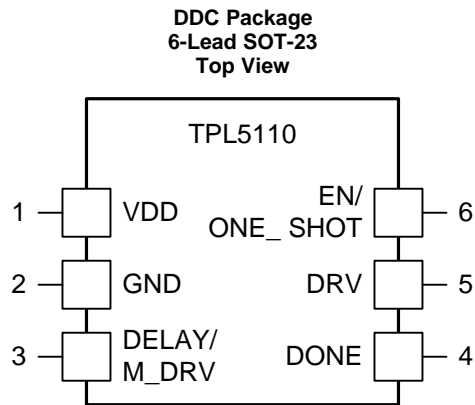
## 目录

1	特性 .....	1	7.4	Device Functional Modes.....	9
2	应用 .....	1	7.5	Programming .....	10
3	说明 .....	1	8	<b>Application and Implementation .....</b>	<b>16</b>
4	修订历史记录 .....	2	8.1	Application Information.....	16
5	<b>Pin Configuration and Functions .....</b>	<b>3</b>	8.2	Typical Application .....	16
6	<b>Specifications.....</b>	<b>4</b>	9	<b>Power Supply Recommendations.....</b>	<b>17</b>
6.1	Absolute Maximum Ratings .....	4	10	<b>Layout.....</b>	<b>17</b>
6.2	ESD Ratings .....	4	10.1	Layout Guidelines .....	17
6.3	Recommended Operating Ratings .....	4	10.2	Layout Example .....	18
6.4	Thermal Information .....	4	11	<b>器件和文档支持 .....</b>	<b>19</b>
6.5	Electrical Characteristics.....	5	11.1	接收文档更新通知 .....	19
6.6	Timing Requirements .....	6	11.2	社区资源.....	19
6.7	Typical Characteristics.....	7	11.3	商标.....	19
7	<b>Detailed Description .....</b>	<b>8</b>	11.4	静电放电警告.....	19
7.1	Overview .....	8	11.5	术语表 .....	19
7.2	Functional Block Diagram .....	8	12	<b>机械、封装和可订购信息.....</b>	<b>19</b>
7.3	Feature Description.....	8			

## 4 修订历史记录

Changes from Original (January 2015) to Revision A	Page
• 已更改 说明文本.....	1
• Changed max IDD .....	5
• Changed <i>TPL5110 Timing</i> labels .....	6
• Changed <i>Circuitry</i> text.....	12
• Changed $T_{ADC}$ and $R_D$ equations in the <i>Quantization Error</i> section .....	14
• 添加了接收文档更新通知 部分 .....	19

## 5 Pin Configuration and Functions



### Pin Functions

PIN		TYPE <sup>(1)</sup>	DESCRIPTION	APPLICATION INFORMATION
NO.	NAME			
1	VDD	P	Supply voltage	
2	GND	G	Ground	
3	DELAY/ M_DRV	I	Time interval set and manual MOSFET Power ON	Resistance between this pin and GND is used to select the time interval. The manual MOSFET power ON switch is also connected to this pin.
4	DONE	I	Logic Input for watchdog functionality	Digital signal driven by the $\mu$ C to indicate successful processing.
5	DRV	O	Power Gating output signal generated every $t_{IP}$	The Gate of the MOSFET is connected to this pin. When DRV = LOW, the MOSFET is ON.
6	EN/ ONE_SHOT	I	Selector of mode of operation	When EN/ONE_SHOT = HIGH, the TPL5110 works as a TIMER. When EN/ONE_SHOT = LOW, the TPL5110 turns on the MOSFET one time for the programmed time interval. The next power on of the MOSFET is enabled by the manual power ON.

(1) G= Ground, P= Power, O= Output, I= Input.

## 6 Specifications

### 6.1 Absolute Maximum Ratings<sup>(1)</sup>

	MIN	MAX	UNIT
Supply Voltage (VDD-GND)	−0.3	6	V
Input Voltage at any pin <sup>(2)</sup>	−0.3	VDD + 0.3	V
Input Current on any pin	−5	+5	mA
Junction Temperature, T <sub>J</sub> <sup>(3)</sup>		150	°C
Storage Temperature, T <sub>stg</sub>	−65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) The voltage between any two pins should not exceed 6 V.
- (3) The maximum power dissipation is a function of T<sub>J</sub>(MAX), θ<sub>JA</sub>, and the ambient temperature, T<sub>A</sub>. The maximum allowable power dissipation at any ambient temperature is P<sub>DMAX</sub> = (T<sub>J</sub>(MAX) - T<sub>A</sub>) / θ<sub>JA</sub>. All numbers apply for packages soldered directly onto a printed-circuit board (PCB).

### 6.2 ESD Ratings

		VALUE	UNIT
V <sub>(ESD)</sub> Electrostatic discharge	Human Body Model, per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±1000	V
	Charged-device model (CDM), per JEDEC specification JESD22-101 <sup>(2)</sup>	±250	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### 6.3 Recommended Operating Ratings

	MIN	MAX	UNIT
Supply Voltage (VDD-GND)	1.8	5.5	V
Temperature	−40	105	°C

### 6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		TPL5110	UNIT
		DDC (SOT-23)	
		6 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	163	°C/W
R <sub>θJC(top)</sub>	Junction-to-case (top) thermal resistance	26	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	57	°C/W
ψ <sub>JT</sub>	Junction-to-top characterization parameter	7.5	°C/W
ψ <sub>JB</sub>	Junction-to-board characterization parameter	57	°C/W
R <sub>θJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	N/A	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report (SPRA953).

## 6.5 Electrical Characteristics<sup>(1)</sup>

Specifications are for  $T_A = 25^\circ\text{C}$ ,  $V_{DD-GND} = 2.5\text{ V}$ , unless otherwise stated.

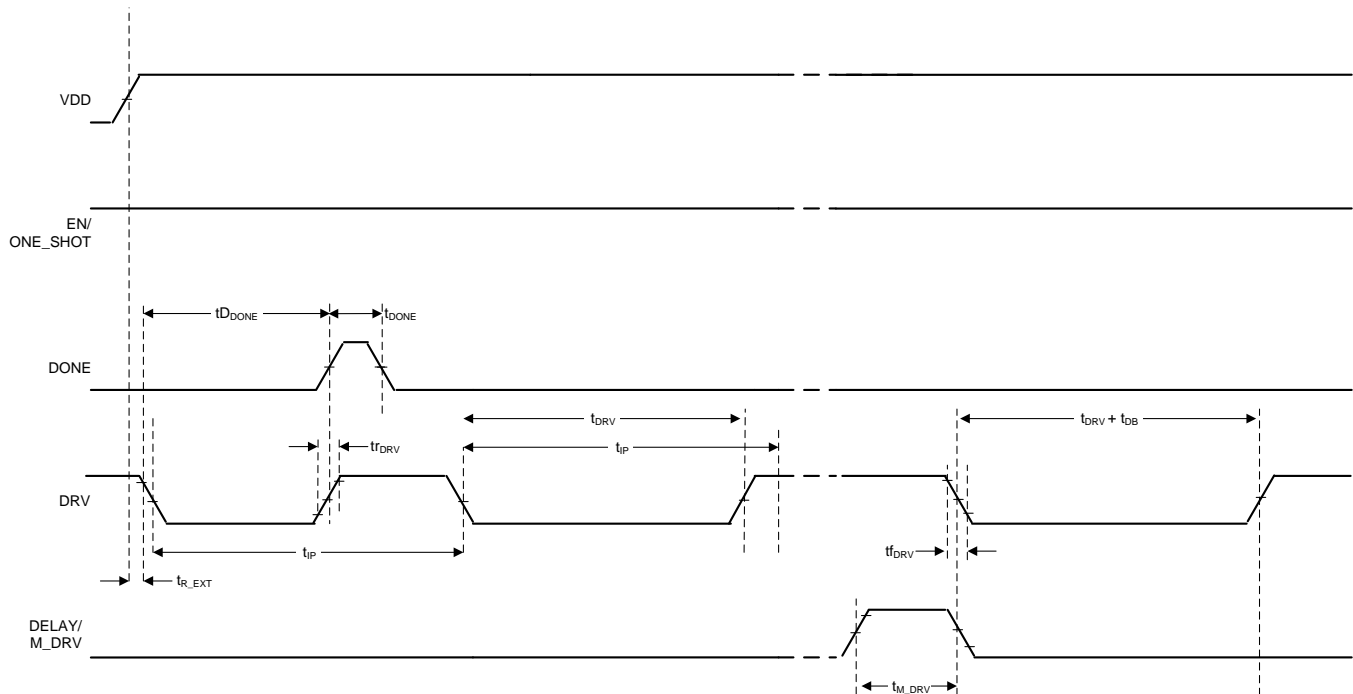
PARAMETER		TEST CONDITIONS		MIN <sup>(2)</sup>	TYP <sup>(3)</sup>	MAX <sup>(2)</sup>	UNIT
POWER SUPPLY							
IDD	Supply current <sup>(4)</sup>	Operation mode		35	50		nA
		Digital conversion of external resistance (Rext)		200	400		μA
TIMER							
t <sub>IP</sub>	Time interval Period	1650 selectable Time intervals	Minimum time interval	100			ms
			Maximum time interval	7200			s
	Time interval Setting Accuracy <sup>(5)</sup>	Excluding the precision of Rext	±0.6%				
	Time interval Setting Accuracy over supply voltage	1.8 V ≤ VDD ≤ 5.5 V	±25			ppm/V	
t <sub>OSC</sub>	Oscillator Accuracy			−0.5%	0.5%		
	Oscillator Accuracy over temperature <sup>(6)</sup>	−40°C ≤ T <sub>A</sub> ≤ 105°C		±100	±400		ppm/°C
	Oscillator Accuracy over supply voltage	1.8 V ≤ VDD ≤ 5.5 V		±0.4			%/V
	Oscillator Accuracy over life time <sup>(7)</sup>			±0.24%			
t <sub>DONE</sub>	DONE Pulse width <sup>(6)</sup>			100			ns
t <sub>DRV</sub>	DRV Pulse width	DONE signal not received		t <sub>IP</sub> -50 ms			
t <sub>Rext</sub>	Time to convert Rext			100	120		ms
DIGITAL LOGIC LEVELS							
VIH	Logic High Threshold DONE pin			0.7 × VDD			V
VIL	Logic Low Threshold DONE pin					0.3 × VDD	V
VOH	Logic output High Level DRV pin	Iout = 100 μA		VDD − 0.3			V
		Iout = 1 mA		VDD − 0.7			V
VOL	Logic output Low Level DRV pin	Iout = -100 μA				0.3	V
		Iout = −1 mA				0.7	V
VIH <sub>M_DRV</sub>	Logic High Threshold DELAY/M_DRV pin			1.5			V

- (1) *Electrical Characteristics* values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that  $T_J = T_A$ . No specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where  $T_J > T_A$ . *Absolute Maximum Ratings* indicate junction temperature limits beyond which the device may be permanently degraded, either mechanically or electrically.
- (2) Limits are specified by testing, design, or statistical analysis at 25°C. Limits over the operating temperature range are specified through correlations using statistical quality control (SQC) method.
- (3) Typical values represent the most likely parametric norm as determined at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not specified on shipped production material.
- (4) The supply current excludes load and pullup resistor current. Input pins are at GND or VDD.
- (5) The accuracy for time interval settings below 1 second is ±100 ms.
- (6) This parameter is specified by design and/or characterization and is not tested in production.
- (7) Operational life time test procedure equivalent to 10 years.

## 6.6 Timing Requirements

			MIN <sup>(1)</sup>	NOM <sup>(2)</sup>	MAX <sup>(1)</sup>	UNIT
$t_{r_{DRV}}$	Rise Time DRV <sup>(3)</sup>	Capacitive load 50 pF		50		ns
$t_{f_{DRV}}$	Fall Time DRV <sup>(3)</sup>	Capacitive load 50 pF		50		ns
$t_{D_{DONE}}$	DONE to DRV delay	Minimum delay <sup>(4)</sup>		100		ns
		Maximum delay <sup>(4)</sup>		$t_{DRV}$		
$t_{M\_DRV}$	Valid manual MOSFET Power ON	Observation time 30 ms	20			ms
$t_{DB}$	De-bounce manual MOSFET Power ON			20		ms

- (1) Limits are specified by testing, design, or statistical analysis at 25°C. Limits over the operating temperature range are specified through correlations using statistical quality control (SQC) method.
- (2) Typical values represent the most likely parametric norm as determined at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not specified on shipped production material.
- (3) This parameter is specified by design and/or characterization and is not tested in production.
- (4) From DRV falling edge.



**Figure 1. TPL5110 Timing**

## 6.7 Typical Characteristics

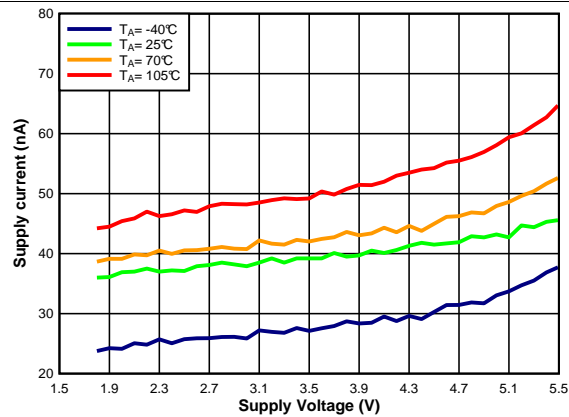


Figure 2.  $I_{DD}$  vs.  $V_{DD}$

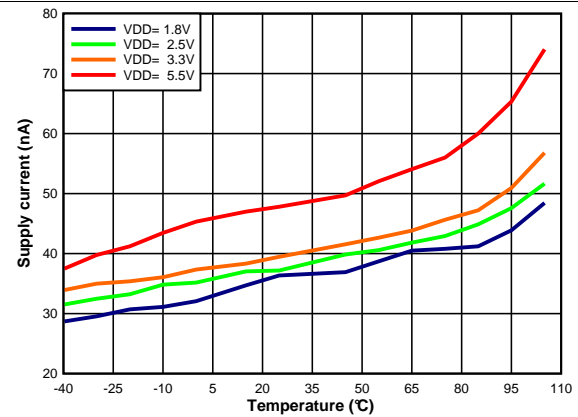


Figure 3.  $I_{DD}$  vs. Temperature

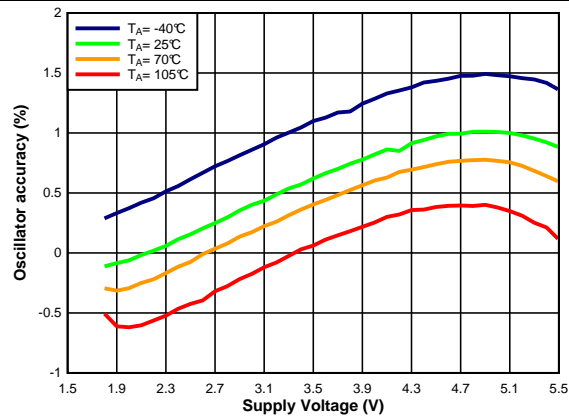


Figure 4. Oscillator Accuracy vs.  $V_{DD}$

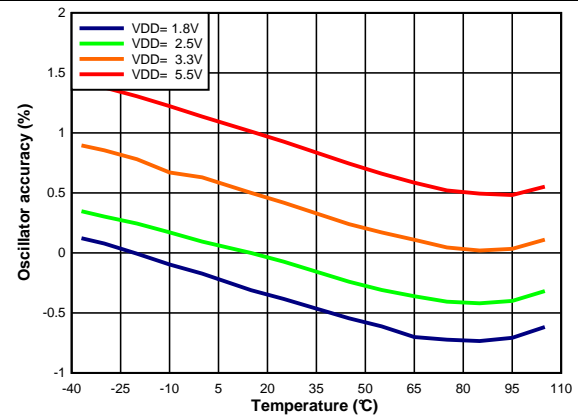


Figure 5. Oscillator Accuracy vs. Temperature

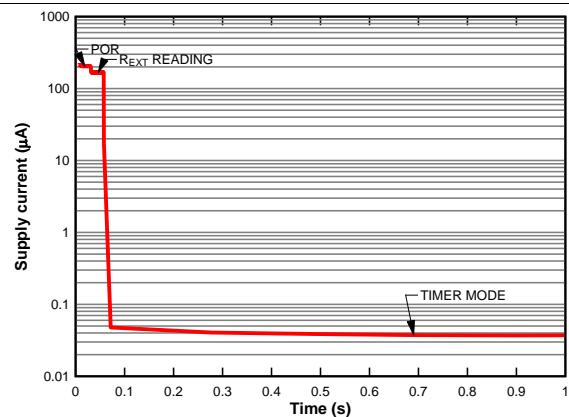


Figure 6.  $I_{DD}$  vs. Time

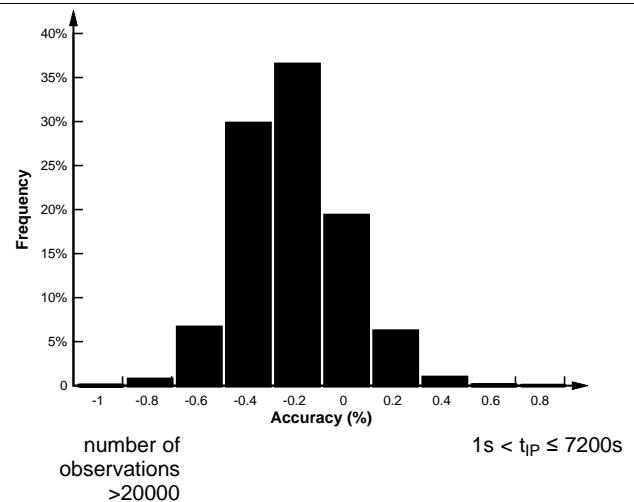


Figure 7. Time interval Setting Accuracy

## 7 Detailed Description

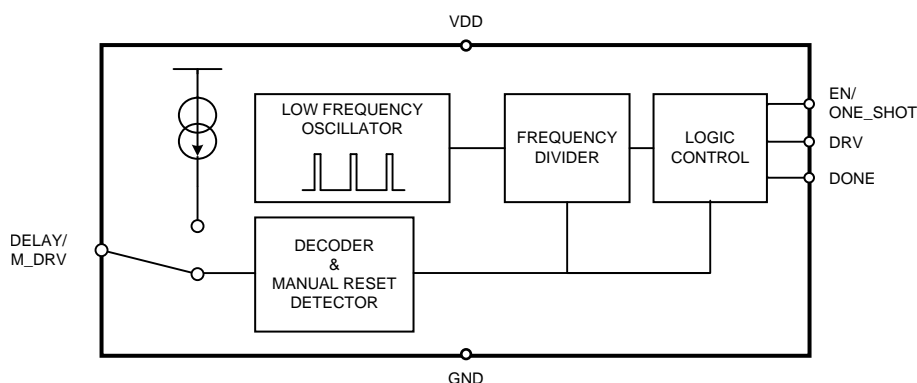
### 7.1 Overview

The TPL5110 is a timer with power gating feature. It is ideal for use in power-cycled applications and provides selectable timing from 100 ms to 7200 s.

Once configured in timer mode (EN/ONE\_SHOT= HIGH) the TPL5110 periodically sends out a DRV signal to a MOSFET to turn on the  $\mu\text{C}$ . If the  $\mu\text{C}$  replies with a DONE signal within the programmed time interval ( $t_{\text{DRV}}$ ) the TPL5110 turns off the  $\mu\text{C}$ , otherwise the TPL5110 keeps the  $\mu\text{C}$  in the on state for a time equal to  $t_{\text{DRV}}$ .

The TPL5110 can work also in a one-shot mode (EN/ONE\_SHOT= LOW). In this mode the DRV signal is sent out just one time at the power on of the TPL5110 to turn on the  $\mu\text{C}$ . If the  $\mu\text{C}$  replies with a DONE signal within the programmed time interval ( $t_{\text{DRV}}$ ) the TPL5110 turns off the  $\mu\text{C}$ , otherwise the TPL5110 keeps the  $\mu\text{C}$  in the on state for a time equal to  $t_{\text{DRV}}$ .

### 7.2 Functional Block Diagram



### 7.3 Feature Description

The TPL5110 implements a periodical power gating feature or one shot power gating according to the EN/ONE\_SHOT voltage. A manual MOSFET Power ON function is realized by momentarily pulling the DELAY/M\_DRV pin to VDD.

#### 7.3.1 DRV

The gate of the MOSFET is connected to the DRV pin. When DRV=LOW, the MOSFET is turned ON. The pulse generated at DRV is equal to the selected time interval period, minus 50 ms. It is shorter in the case of a DONE signal received from the  $\mu\text{C}$ . If the DONE signal is not received within the programmed time interval (minus 50 ms), the DRV signal will be high for the last 50 ms of the time interval to turn off the MOSFET before the next cycle starts.

The default value (after resistance reading) is HIGH. The signal is sent out from the TPL5110 when the programmed time interval starts. When the DRV is LOW, the manual power ON signal is ignored.

#### 7.3.2 DONE

The DONE pin is driven by a  $\mu\text{C}$  to signal that the  $\mu\text{C}$  is working properly. The TPL5110 recognizes a valid DONE signal as a low to high transition. If two or more DONE signals are received within the time interval, only the first DONE signal is processed. The minimum DONE signal pulse length is 100 ns. When the TPL5110 receives the DONE signal it asserts DRV logic HIGH.



## 7.4 Device Functional Modes

### 7.4.1 Start-Up

During start-up, after POR, the TPL5110 executes a one-time measurement of the resistance attached to the DELAY/M\_DRV pin in order to determine the desired time interval for DRV. This measurement interval is  $t_{R\_EXT}$ . During this measurement a constant current is temporarily flowing into  $R_{EXT}$ .

Once the reading of the external resistance is complete, the TPL5110 enters automatically in one of the two modes according to the EN/ONE\_SHOT value. The EN/ONE\_SHOT pin needs to be hard wired to GND or VDD according to the required mode of operation.

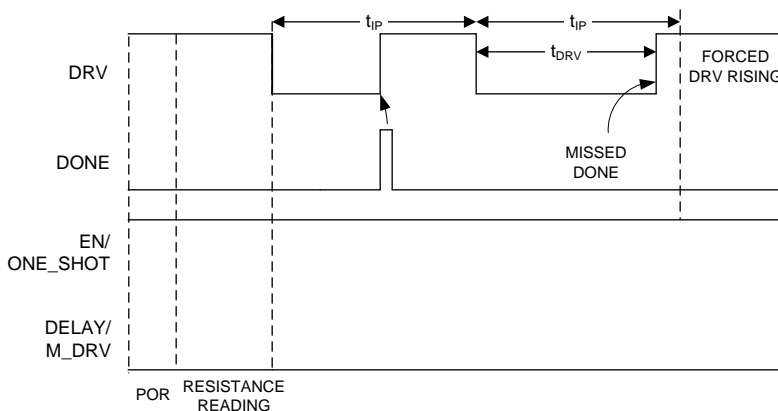


Figure 8. Start-Up - Timer Mode

### 7.4.2 Timer Mode

During timer mode (EN/ONE\_SHOT = HIGH), the TPL5110 asserts periodic DRV pulses according to the programmed time interval. The length of the DRV pulses is set by the receiving of a DONE pulse from the  $\mu C$ . See [Figure 8](#).

### 7.4.3 One-Shot Mode

During one-shot mode (EN/ONE\_SHOT = LOW), the TPL5110 generates just one pulse at the DRV pin which lasts according to the programmed time interval. In one-shot mode, other DRV pulses can be triggered using the DELAY/M\_DRV pin. If a valid manual power ON occurs when EN/ONE\_SHOT is LOW, the TPL5110 generates just one pulse at the DRV pin. The duration of the pulse is set by the programmed time interval. Also in this case, if a DONE signal is received within the programmed time interval (minus 50 ms), the MOSFET connected to the DRV pin is turned off. See [Figure 9](#) and [Figure 10](#).

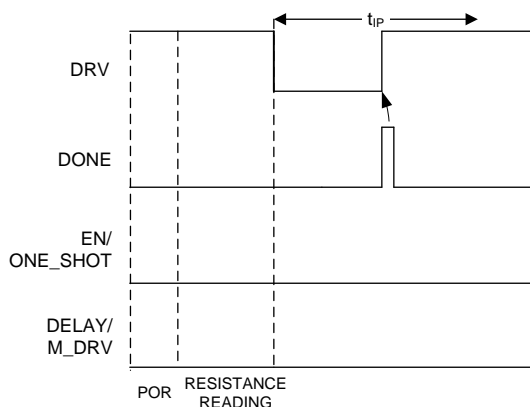
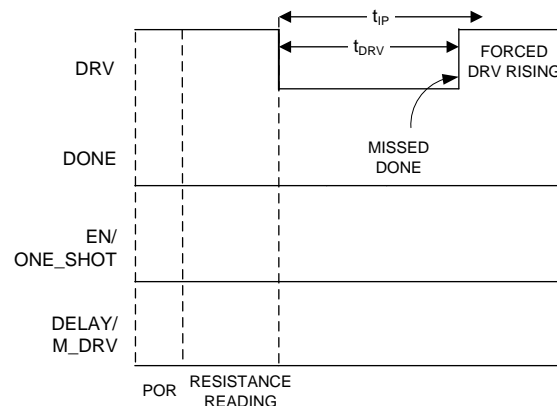


Figure 9. Start-Up One-Shot Mode (DONE Received Within  $t_{IP}$ )

## Device Functional Modes (continued)



**Figure 10. Start-Up One-Shot Mode (No DONE Received Within  $t_{IP}$ )**

## 7.5 Programming

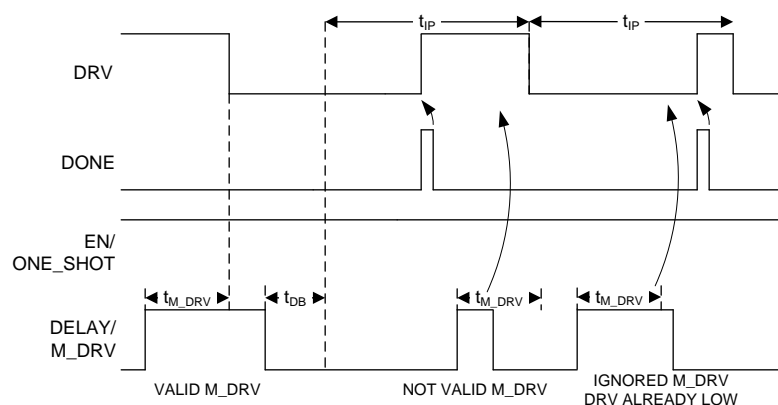
### 7.5.1 Configuring the Time Interval With the DELAY/M\_DRV Pin

The time interval between two adjacent DRV pulses (falling edges, in timer mode) is selectable through an external resistance ( $R_{EXT}$ ) between the DELAY/M\_DRV pin and ground. The resistance ( $R_{EXT}$ ) must be in the range between 500  $\Omega$  and 170 k $\Omega$ . At least a 1% precision resistance is recommended. See section [Selection of the External Resistance](#) on how to set the time interval using  $R_{EXT}$ .

### 7.5.2 Manual MOSFET Power ON Applied to the DELAY/M\_DRV Pin

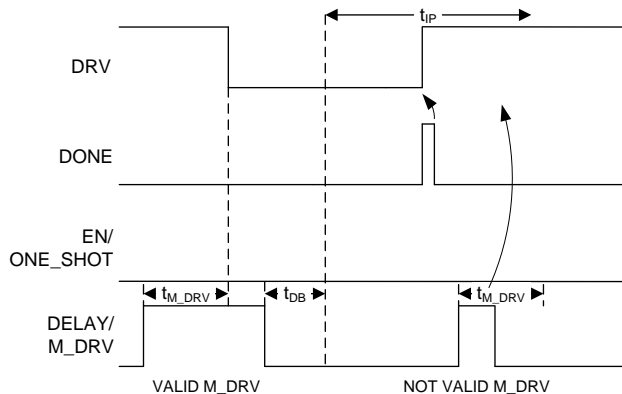
If VDD is connected to the DELAY/M\_DRV pin, the TPL5110 recognizes this as a manual MOSFET Power ON condition. In this case the time interval is not set. If the manual MOSFET Power ON is asserted during the POR or during the reading procedure, the reading procedure is aborted and is restarted as soon as the manual MOSFET Power ON switch is released. A pulse on the DELAY/M\_DRV pin is recognized as a valid manual MOSFET Power ON only if it lasts at least 20 ms (observation time is 30 ms). The manual MOSFET Power ON may be implemented using a switch (momentary mechanical action).

If the DRV is already LOW (MOSFET ON) the manual MOSFET Power ON is ignored.



**Figure 11. Manual MOSFET Power ON in Timer Mode**

## Programming (continued)



**Figure 12. Manual MOSFET Power ON in One-Shot Mode**

### 7.5.2.1 DELAY/M\_DRV

A resistance in the range between 500  $\Omega$  and 170 k $\Omega$  must be connected to the DELAY/M\_DRV pin to select a valid time interval. At the POR and during the reading of the resistance, the DELAY/M\_DRV is connected to an analog signal chain through a mux. After the reading of the resistance, the analog circuit is switched off and the DELAY/M\_DRV is connected to a digital circuit.

In this state, a logic HIGH applied to the DELAY/M\_DRV pin is interpreted by the TPL5110 as a manual power ON. The manual power ON detection is provided with a de-bounce feature (on both edges) which makes the TPL5110 insensitive to the glitches on the DELAY/M\_DRV.

The M\_DRV must stay high for at least 20 ms to be valid. Once a valid signal at DELAY/M\_DRV is understood as a manual power on, the DRV signal will be asserted in the next 10 ms. Its duration will be according to the programmed time interval (minus 50 ms), or less if the DONE is received.

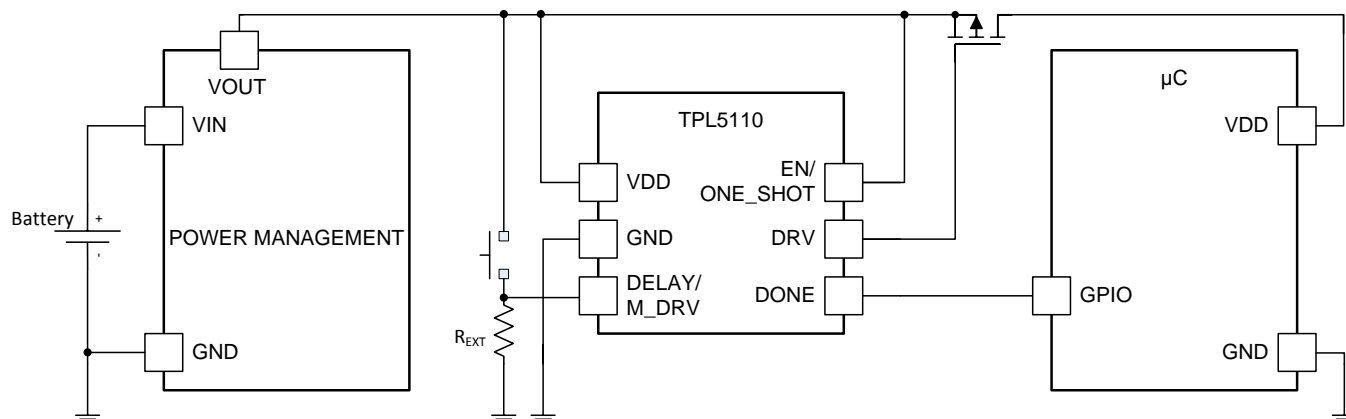
A manual power ON signal resets all the counters. The counters will restart as soon as a valid manual power ON signal is recognized and the signal at DELAY/M\_DRV pin is asserted LOW. Due to the asynchronous nature of the manual power ON signal and its arbitrary duration, the LOW status of the DRV signal may be affected by an uncertainty of about  $\pm 5$  ms.

An extended assertion of a logic HIGH at the DELAY/M\_DRV pin will turn on the MOSFET for a time longer than the programmed time interval. DONE signals received while the DELAY/M\_DRV is HIGH are ignored. If the DRV is already LOW (MOSFET ON) the manual power ON is ignored.

### 7.5.2.2 Circuitry

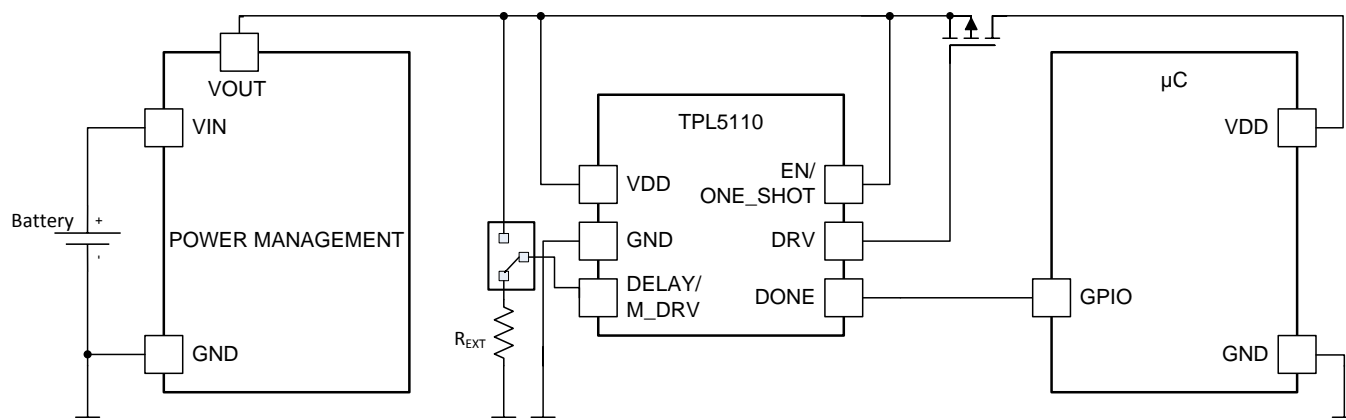
The manual Power ON may be implemented using a switch (momentary mechanical action). The TPL5110 offers two possible approaches according to the power consumption constraints of the application.

## Programming (continued)



**Figure 13. Manual MOSFET Power ON With SPST Switch**

For use cases that do not require the lowest power consumption, using a single-pole single-throw switch may offer a lower-cost solution. The DELAY/M\_DRV pin may be directly connected to VDD with  $R_{EXT}$  in the circuit. The current drawn from the supply voltage during the manual power ON is given by  $VDD/R_{EXT}$ .



**Figure 14. Manual MOSFET Power ON With SPDT Switch**

The manual MOSFET Power ON function may also be asserted by switching DELAY/M\_DRV from  $R_{EXT}$  to VDD using a single-pole double-throw switch, which will provide a lower power solution for the manual power ON, because no current flows.

### 7.5.3 Selection of the External Resistance

To set the time interval, the external resistance  $R_{EXT}$  is selected according to [Equation 1](#):

$$R_{EXT} = 100 \left( \frac{-b + \sqrt{b^2 - 4a(c - 100T)}}{2a} \right)$$

where

- T is the desired time interval in seconds.
- $R_{EXT}$  is the resistance value to use in  $\Omega$ .
- a, b, c are coefficients depending on the range of the time interval.

(1)

## Programming (continued)

**Table 1. Coefficients for Equation 1**

SET	Time Interval Range (s)	a	b	c
1	1 < T ≤ 5	0.2253	-20.7654	570.5679
2	5 < T ≤ 10	-0.1284	46.9861	-2651.8889
3	10 < T ≤ 100	0.1972	-19.3450	692.1201
4	100 < T ≤ 1000	0.2617	-56.2407	5957.7934
5	T > 1000	0.3177	-136.2571	34522.4680

### EXAMPLE

Required time interval: 8 s

The coefficient set to be selected is the number 2. The formula becomes Equation 2.

$$R_{EXT} = 100 \left( \frac{46.9861 - \sqrt{46.9861^2 + 4 \cdot 0.1284(-2651.8889 - 100 \cdot 8)}}{2 \cdot 0.1284} \right) \quad (2)$$

The resistance value is 10.18 kΩ.

Table 2 and Table 3 contain example values of t<sub>IP</sub> and their corresponding value of R<sub>EXT</sub>.

**Table 2. First 9 Time Intervals**

t <sub>IP</sub> (ms)	Resistance (Ω)	Closest real value (Ω)	Parallel of two 1% tolerance resistors, (kΩ)
100	500	500	1.0 // 1.0
200	1000	1000	-
300	1500	1500	2.43 // 3.92
400	2000	2000	-
500	2500	2500	4.42 // 5.76
600	3000	3000	5.36 // 6.81
700	3500	3500	4.75 // 13.5
800	4000	4000	6.19 // 11.3
900	4500	4501	6.19 // 16.5

**Table 3. Most Common Time Intervals Between 1s to 2h**

t <sub>IP</sub>	Calculated Resistance (kΩ)	Closest Real Value (kΩ)	Parallel of Two 1% Tolerance Resistors, (kΩ)
1s	5.20	5.202	7.15 // 19.1
2s	6.79	6.788	12.4 // 15.0
3s	7.64	7.628	12.7 // 19.1
4s	8.30	8.306	14.7 // 19.1
5s	8.85	8.852	16.5 // 19.1
6s	9.27	9.223	18.2 // 18.7
7s	9.71	9.673	19.1 // 19.6
8s	10.18	10.180	11.5 // 8.87
9s	10.68	10.68	17.8 // 26.7
10s	11.20	11.199	15.0 // 44.2
20s	14.41	14.405	16.9 // 97.6
30s	16.78	16.778	32.4 // 34.8
40s	18.75	18.748	22.6 // 110.0
50s	20.047	20.047	28.7 // 66.5

**Table 3. Most Common Time Intervals Between 1s to 2h (continued)**

<b>t<sub>IP</sub></b>	<b>Calculated Resistance (kΩ)</b>	<b>Closest Real Value (kΩ)</b>	<b>Parallel of Two 1% Tolerance Resistors, (kΩ)</b>
1min	22.02	22.021	40.2 // 48.7
2min	29.35	29.349	35.7 // 165.0
3min	34.73	34.729	63.4 // 76.8
4min	39.11	39.097	63.4 // 102.0
5min	42.90	42.887	54.9 // 196.0
6min	46.29	46.301	75.0 // 121.0
7min	49.38	49.392	97.6 // 100.0
8min	52.24	52.224	88.7 // 127.0
9min	54.92	54.902	86.6 // 150.0
10min	57.44	57.437	107.0 // 124.0
20min	77.57	77.579	140.0 // 174.0
30min	92.43	92.233	182.0 // 187.0
40min	104.67	104.625	130.0 // 536.00
50min	115.33	115.331	150.0 // 499.00
1h	124.91	124.856	221.0 // 287.00
1h30min	149.39	149.398	165.0 // 1580.0
2h	170.00	170.00	340.0 // 340.0

### 7.5.4 Quantization Error

The TPL5110 can generate 1650 discrete timer intervals in the range of 100 ms to 7200 s. The first 9 intervals are multiples of 100 ms. The remaining 1641 intervals cover the range between 1 s to 7200 s. Because they are discrete intervals, there is a quantization error associated with each value.

The quantization error can be evaluated according to [Equation 3](#):

$$Err = 100 \frac{(T_{DESIRED} - T_{ADC})}{T_{DESIRED}}$$

where

$$\begin{aligned} \bullet \quad T_{ADC} &= \text{INT} \left[ \frac{1}{100} (aR_D^2 + bR_D + c) \right] \\ \bullet \quad R_D &= \frac{R_{EXT}}{100} \end{aligned} \quad (3)$$

R<sub>EXT</sub> is the resistance calculated with [Equation 1](#) and a, b, c are the coefficients of the equation listed in [Table 1](#).

### 7.5.5 Error Due to Real External Resistance

R<sub>EXT</sub> is a theoretical value and may not be available in standard commercial resistor values. It is possible to closely approach the theoretical R<sub>EXT</sub> using two or more standard values in parallel. However, standard values are characterized by a certain tolerance. This tolerance will affect the accuracy of the time interval.

The accuracy can be evaluated using the following procedure:

1. Evaluate the min and max values of R<sub>EXT</sub> (R<sub>EXT\_MIN</sub>, R<sub>EXT\_MAX</sub> with [Equation 1](#) using the selected commercial resistance values and their tolerances.
2. Evaluate the time intervals (T<sub>ADC\_MIN</sub>[R<sub>EXT\_MIN</sub>], T<sub>ADC\_MAX</sub>[R<sub>EXT\_MAX</sub>]) with the T<sub>ADC</sub> equation mentioned in [Equation 3](#).
3. Find the errors using [Equation 3](#) with T<sub>ADC\_MIN</sub>, T<sub>ADC\_MAX</sub>.

The results of the formula indicate the accuracy of the time interval.

The example below illustrates the procedure.

- Desired time interval,  $T_{\text{desired}} = 600 \text{ s}$ ,
- Required  $R_{\text{EXT}}$  from Equation 1,  $R_{\text{EXT}} = 57.44 \text{ k}\Omega$ .

From Table 3,  $R_{\text{EXT}}$  can be built with a parallel combination of two commercial values with 1% tolerance:  $R_1 = 107 \text{ k}\Omega$ ,  $R_2 = 124 \text{ k}\Omega$ . The uncertainty of the equivalent parallel resistance can be found using Equation 4:

$$uR_{//} = R_{//} \sqrt{\left(\frac{u_{R1}}{R1}\right)^2 + \left(\frac{u_{R2}}{R2}\right)^2}$$

where

- $uR_n$  ( $n=1,2$ ) represent the uncertainty of a resistance (see Equation 5)

(4)

$$u_{R_n} = R_n \frac{\text{Tolerance}}{\sqrt{3}}$$

(5)

The uncertainty of the parallel resistance is 0.82%, which means the value of  $R_{\text{EXT}}$  may range between  $R_{\text{EXT\_MIN}} = 56.96 \text{ k}\Omega$  and  $R_{\text{EXT\_MAX}} = 57.90 \text{ k}\Omega$ .

Using these value of  $R_{\text{EXT}}$ , the digitized timer intervals calculated by  $T_{\text{ADC}}$  equation mentioned in Equation 3 are respectively  $T_{\text{ADC\_MIN}} = 586.85 \text{ s}$  and  $T_{\text{ADC\_MAX}} = 611.3 \text{ s}$ , giving an error range of  $-1.88\%$  /  $+2.19\%$ . The asymmetry of the error range is due to the quadratic transfer function of the resistance digitizer.

## 8 Application and Implementation

### NOTE

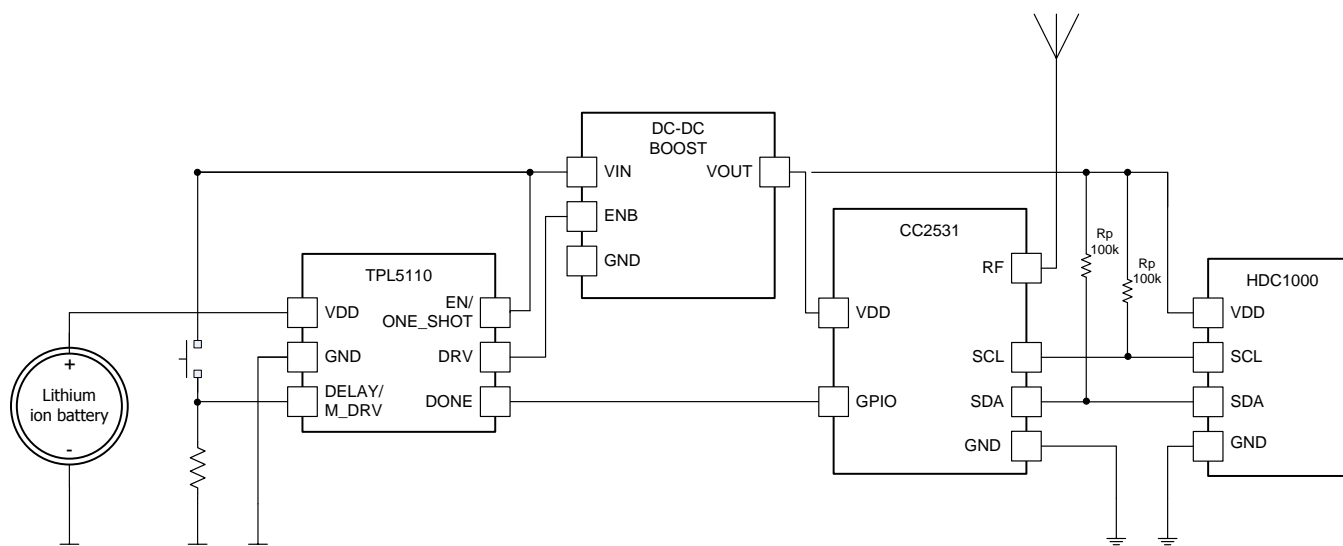
Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 8.1 Application Information

In battery-powered applications, one design constraint is the need for low current consumption. The TPL5110 is designed for applications where there is a need to monitor environmental conditions at a fixed time interval. Often in these applications a watchdog or other internal timer in a  $\mu\text{C}$  is used to implement a wake-up function. Typically, the power consumption of these functions is not optimized. Using the TPL5110 to implement a periodical power gating of the  $\mu\text{C}$  or of the entire system the current consumption will be only tens of nA.

### 8.2 Typical Application

The TPL5110 can be used in environment sensor nodes such as humidity and temperature sensor node. The sensor node has to measure the humidity and the temperature and transmit the data through a low power RF micro such as the CC2531. The temperature and the humidity in home application do not change so fast, so the measurement and the transmission of the data can be done at very low rate, such as every 30 seconds. The RF micro should spend most of the time in counting the elapsed time, but using the TPL5110 it is possible to complete turn off the RF micro and extend the battery life. The TPL5110 will turn on the RF micro when the programmed time interval elapses or for debug purpose with the manual MOSFET Power ON switch.



**Figure 15. Sensor Node**

#### 8.2.1 Design Requirements

The design is driven by the low current consumption constraint. The data are usually acquired on a rate which is in the range between 30 s and 60 s. The highest necessity is the maximization of the battery life. The TPL5110 helps achieve this goal because it allows turning off the RF micro.

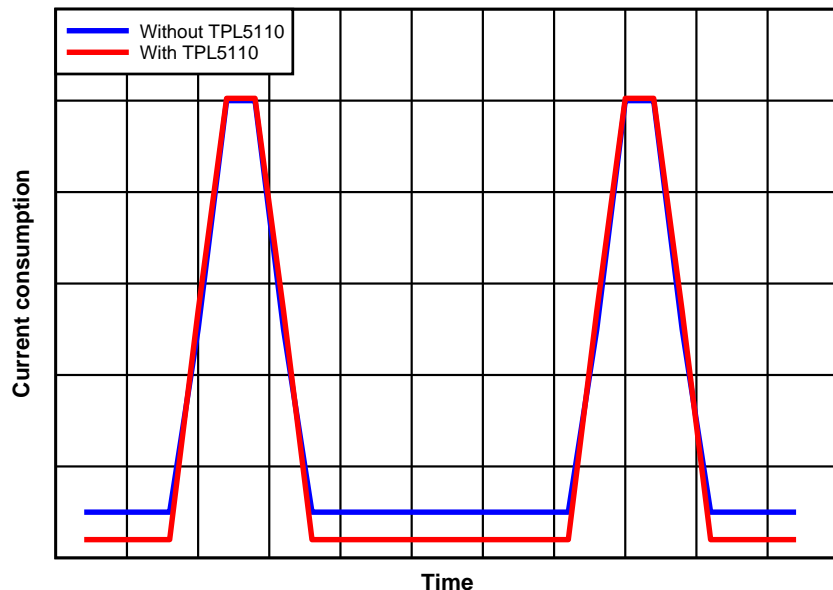
#### 8.2.2 Detailed Design Procedure

When the focal constraint is the battery life, the selection of a low power voltage regulator and low leakage MOSFET to power gate the  $\mu\text{C}$  is mandatory. The first step in the design is the calculation of the power consumption of each device in the different mode of operations. An example is the HDC1000, in measurement mode the RF micro is in normal operation and transmission. The different modes offer the possibility to select the appropriate time interval which respect the application constraint and maximize the life of the battery.



## Typical Application (continued)

### 8.2.3 Application Curve



**Figure 16. Effect of TPL5110 on Current Consumption**

## 9 Power Supply Recommendations

The TPL5110 requires a voltage supply within 1.8 V and 5.5 V. A multilayer ceramic bypass X7R capacitor of 0.1  $\mu$ F between VDD and GND pin is recommended.

## 10 Layout

### 10.1 Layout Guidelines

The DELAY/M\_DRV pin is sensitive to parasitic capacitance. TI suggests that the traces connecting the resistance on this pin to GROUND be kept as short as possible to minimize parasitic capacitance. This capacitance can affect the initial set up of the time interval. Signal integrity on the DRV pin is also improved by keeping the trace length between the TPL5110 and the gate of the MOSFET short to reduce the parasitic capacitance. The EN/ONE\_SHOT needs to be tied to GND or VDD with short traces.

## TPL5110

ZHCSEK5A – JANUARY 2015 – REVISED SEPTEMBER 2018

[www.ti.com.cn](http://www.ti.com.cn)

### 10.2 Layout Example

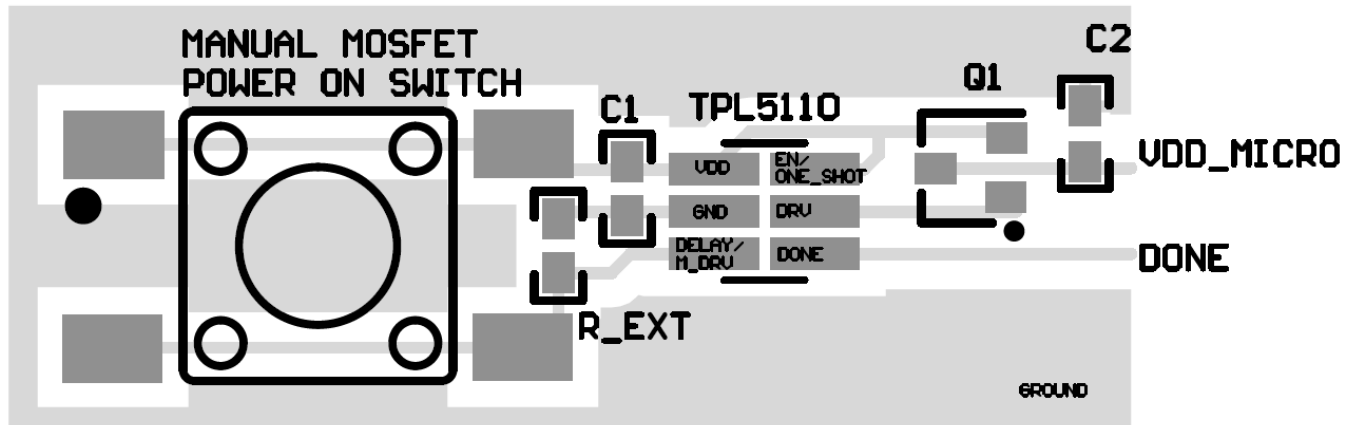


Figure 17. Layout

## 11 器件和文档支持

### 11.1 接收文档更新通知

要接收文档更新通知，请导航至 [TI.com.cn](http://TI.com.cn) 上的器件产品文件夹。单击右上角的 [通知我](#) 进行注册，即可每周接收产品信息更改摘要。有关更改的详细信息，请查看任何已修订文档中包含的修订历史记录。

### 11.2 社区资源

下列链接提供到 TI 社区资源的连接。链接的内容由各个分销商“按照原样”提供。这些内容并不构成 TI 技术规范，并且不一定反映 TI 的观点；请参阅 TI 的 [《使用条款》](#)。

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**设计支持** [TI 参考设计支持](#) 可帮助您快速查找有帮助的 E2E 论坛、设计支持工具以及技术支持的联系信息。

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ESD 的损坏小至导致微小的性能降级，大至整个器件故障。精密的集成电路可能更容易受到损坏，这是因为非常细微的参数更改都可能会导致器件与其发布的规格不相符。

### 11.5 术语表

[SLYZ022](#) — *TI 术语表*。

这份术语表列出并解释术语、缩写和定义。

## 12 机械、封装和可订购信息

以下页面包含机械、封装和可订购信息。这些信息是指定器件的最新可用数据。数据如有变更，恕不另行通知，且不会对此文档进行修订。如需获取此数据表的浏览器版本，请查阅左侧的导航栏。

## PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package   Pins	Package qty   Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
<a href="#">TPL5110DDCR</a>	Active	Production	SOT-23-THIN (DDC)   6	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 105	ZALX
TPL5110DDCR.A	Active	Production	SOT-23-THIN (DDC)   6	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 105	ZALX
TPL5110DDCR.B	Active	Production	SOT-23-THIN (DDC)   6	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 105	ZALX
<a href="#">TPL5110DDCT</a>	Active	Production	SOT-23-THIN (DDC)   6	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 105	ZALX
TPL5110DDCT.A	Active	Production	SOT-23-THIN (DDC)   6	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 105	ZALX
TPL5110DDCT.B	Active	Production	SOT-23-THIN (DDC)   6	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 105	ZALX
TPL5110DDCTG4	Active	Production	SOT-23-THIN (DDC)   6	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 105	ZALX
TPL5110DDCTG4.A	Active	Production	SOT-23-THIN (DDC)   6	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 105	ZALX
TPL5110DDCTG4.B	Active	Production	SOT-23-THIN (DDC)   6	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 105	ZALX

<sup>(1)</sup> **Status:** For more details on status, see our [product life cycle](#).

<sup>(2)</sup> **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

<sup>(3)</sup> **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

<sup>(4)</sup> **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

<sup>(5)</sup> **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

**(6) Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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**OTHER QUALIFIED VERSIONS OF TPL5110 :**

- Automotive : [TPL5110-Q1](#)

NOTE: Qualified Version Definitions:

- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

## TAPE AND REEL INFORMATION



\*All dimensions are nominal

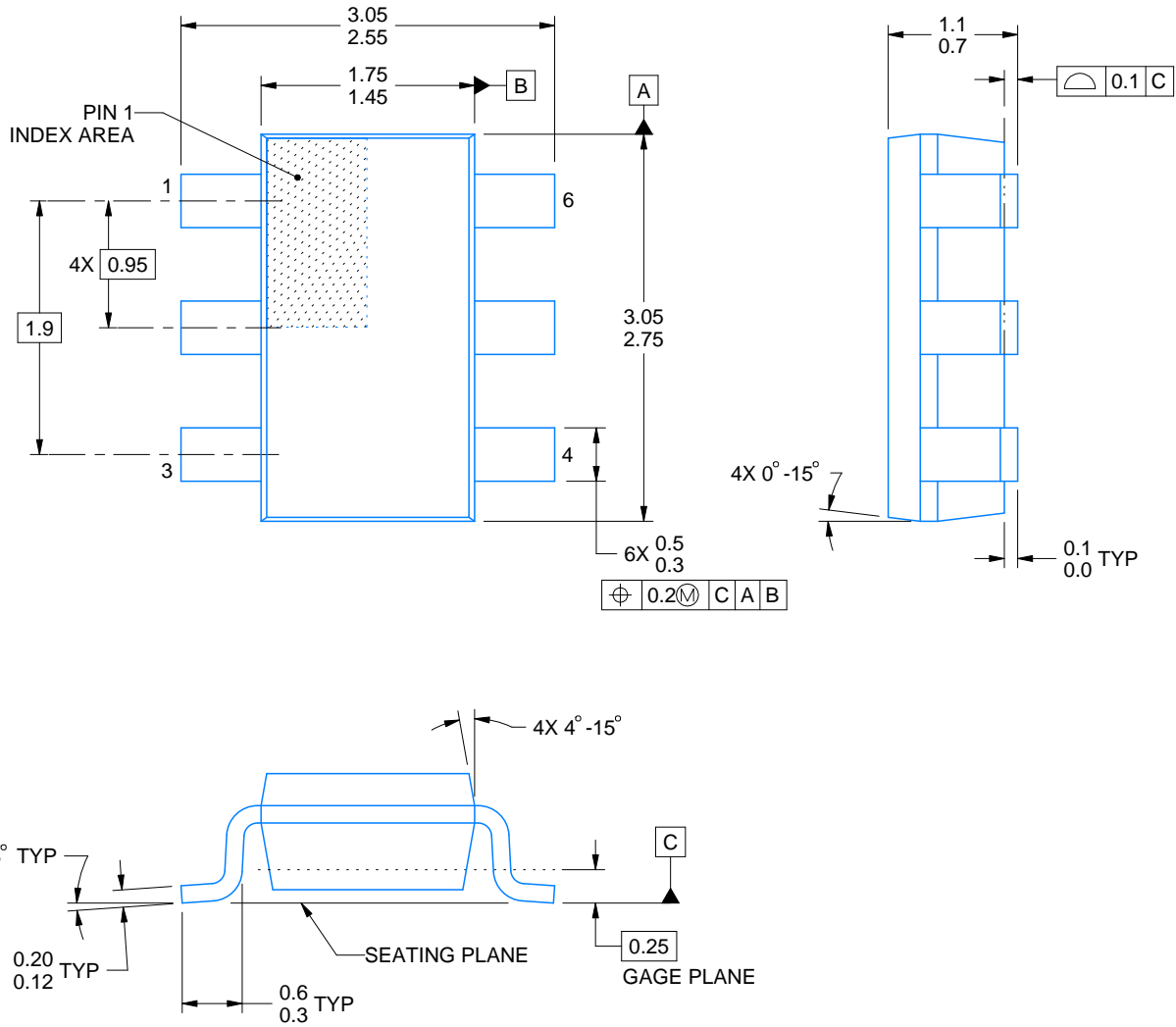
Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPL5110DDCR	SOT-23-THIN	DDC	6	3000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPL5110DDCT	SOT-23-THIN	DDC	6	250	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPL5110DDCTG4	SOT-23-THIN	DDC	6	250	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3

## TAPE AND REEL BOX DIMENSIONS



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPL5110DDCR	SOT-23-THIN	DDC	6	3000	208.0	191.0	35.0
TPL5110DDCT	SOT-23-THIN	DDC	6	250	208.0	191.0	35.0
TPL5110DDCTG4	SOT-23-THIN	DDC	6	250	208.0	191.0	35.0



4214841/E 08/2024

## NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. Reference JEDEC MO-193.

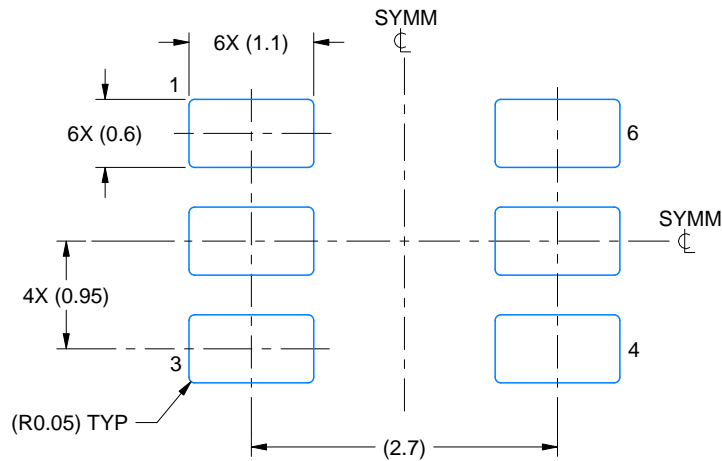


# EXAMPLE BOARD LAYOUT

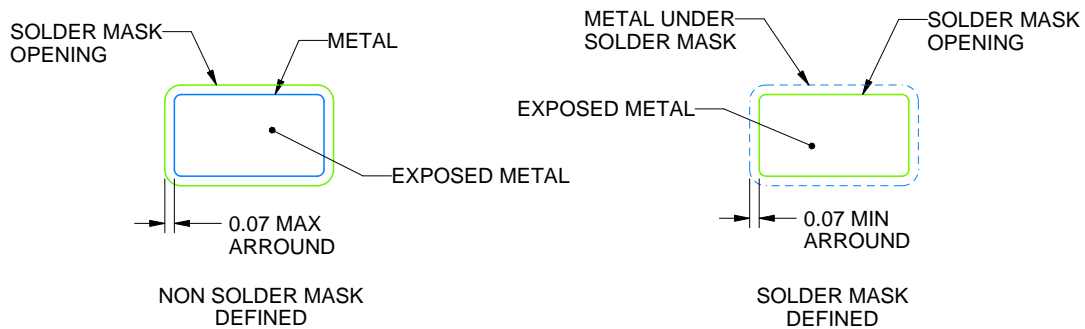
DDC0006A

SOT-23 - 1.1 max height

SMALL OUTLINE TRANSISTOR



LAND PATTERN EXAMPLE  
EXPLODED METAL SHOWN  
SCALE:15X



SOLDERMASK DETAILS

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NOTES: (continued)

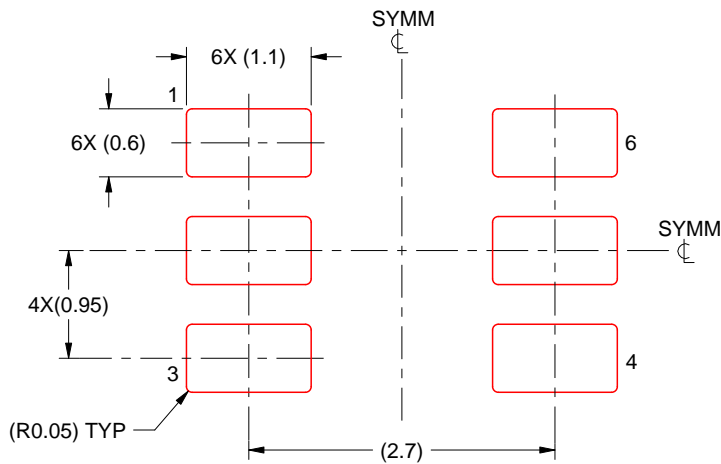
4. Publication IPC-7351 may have alternate designs.
5. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DDC0006A

SOT-23 - 1.1 max height

SMALL OUTLINE TRANSISTOR



SOLDER PASTE EXAMPLE  
BASED ON 0.125 THICK STENCIL  
SCALE:15X

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NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
7. Board assembly site may have different recommendations for stencil design.

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