

LMP2232 Dual Micropower, 1.6V, Precision, Operational Amplifier with CMOS Input

Check for Samples: [LMP2232](#)

FEATURES

(For $V_S = 5V$, Typical Unless Otherwise Noted)

- Supply Current at 1.8V 16 μA
- Operating Voltage Range 1.6V to 5.5V
- Low $TCV_{OS} \pm 0.5 \mu V/^{\circ}C$ (max)
- $V_{OS} \pm 150 \mu V$ (max)
- Input Bias Current 20 fA
- PSRR 120 dB
- CMRR 97 dB
- Open Loop Gain 120 dB
- Gain Bandwidth Product 130 kHz
- Slew Rate 58 V/ms
- Input Voltage Noise, $f = 1 \text{ kHz}$ 60 nV/ \sqrt{Hz}
- Temperature Range $-40^{\circ}C$ to $125^{\circ}C$

APPLICATIONS

- Precision Instrumentation Amplifiers
- Battery Powered Medical Instrumentation
- High Impedance Sensors
- Strain Gauge Bridge Amplifier
- Thermocouple Amplifiers

DESCRIPTION

The LMP2232 is a dual micropower precision amplifier designed for battery powered applications. The 1.6V to 5.5V operating supply voltage range and quiescent power consumption of only 26 μW extend the battery life in portable systems. The LMP2232 is part of the LMP™ precision amplifier family. The high impedance CMOS input makes it ideal for instrumentation and other sensor interface applications.

The LMP2232 has a maximum offset voltage of 150 μV and maximum offset voltage drift of only $0.5 \mu V/^{\circ}C$ along with low bias current of only ± 20 fA. These precise specifications make the LMP2232 a great choice for maintaining system accuracy and long term stability.

The LMP2232 has a rail-to-rail output that swings 15 mV from the supply voltage, which increases system dynamic range. The common mode input voltage range extends 200 mV below the negative supply, thus the LMP2232 is ideal for ground sensing in single supply applications.

The LMP2232 is offered in 8-pin SOIC and VSSOP packages.

The LMP2231 is the single version of this product and the LMP2234 is the quad version of this product. Both of these products are available on Texas Instruments' website.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

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Typical Application

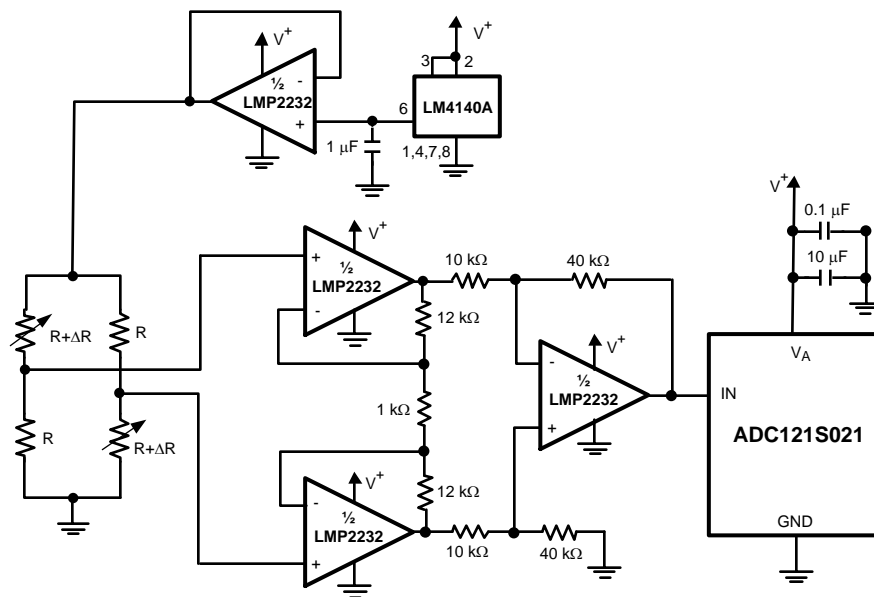


Figure 1. Strain Gauge Bridge Amplifier



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

Absolute Maximum Ratings ⁽¹⁾⁽²⁾

ESD Tolerance ⁽³⁾	Human Body Model	2000V
	Machine Model	100V
Differential Input Voltage		±300 mV
Supply Voltage ($V_S = V^+ - V^-$)		6V
Voltage on Input/Output Pins		$V^+ + 0.3V, V^- - 0.3V$
Storage Temperature Range		-65°C to 150°C
Junction Temperature ⁽⁴⁾		150°C
Mounting Temperature Infrared or Convection (20 sec.)		+235°C
Wave Soldering Lead Temperature (10 sec.)		+260°C

- (1) Absolute Maximum Ratings indicate limits beyond which damage may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not ensured. For ensured specifications and test conditions, see the Electrical Characteristics.
- (2) If Military/Aerospace specified devices are required, please contact the TI Sales Office/ Distributors for availability and specifications.
- (3) Human Body Model, applicable std. MIL-STD-883, Method 3015.7. Machine Model, applicable std. JESD22-A115-A (ESD MM std. of JEDEC) Field-Induced Charge-Device Model, applicable std. JESD22-C101-C (ESD FICDM std. of JEDEC).
- (4) The maximum power dissipation is a function of $T_{J(MAX)}$, θ_{JA} , and T_A . The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} - T_A) / \theta_{JA}$. All numbers apply for packages soldered directly onto a PC board.

Operating Ratings ⁽¹⁾

Operating Temperature Range ⁽²⁾		–40°C to 125°C
Supply Voltage ($V_S = V^+ - V^-$)		1.6V to 5.5V
Package Thermal Resistance (θ_{JA}) ⁽²⁾	8-Pin SOIC	111.2 °C/W
	8-Pin VSSOP	147.4 °C/W

- (1) Absolute Maximum Ratings indicate limits beyond which damage may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not ensured. For ensured specifications and test conditions, see the Electrical Characteristics.
- (2) The maximum power dissipation is a function of $T_{J(MAX)}$, θ_{JA} , and T_A . The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} - T_A) / \theta_{JA}$. All numbers apply for packages soldered directly onto a PC board.

5V DC Electrical Characteristics ⁽¹⁾

Unless otherwise specified, all limits ensured for $T_A = 25^\circ\text{C}$, $V^+ = 5\text{V}$, $V^- = 0\text{V}$, $V_{CM} = V_O = V^+/2$, and $R_L > 1\text{ M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min ⁽²⁾	Typ ⁽³⁾	Max ⁽²⁾	Units
V_{OS}	Input Offset Voltage			± 10	± 150 ± 230	μV
TCV_{OS}	Input Offset Voltage Drift	LMP2232A		± 0.3	± 0.5	$\mu\text{V}/^\circ\text{C}$
		LMP2232B		± 0.3	± 2.5	
I_{BIAS}	Input Bias Current			0.02	± 3 ± 125	pA
I_{OS}	Input Offset Current			5		fA
CMRR	Common Mode Rejection Ratio	$0\text{V} \leq V_{CM} \leq 4\text{V}$	81 80	97		dB
PSRR	Power Supply Rejection Ratio	$1.6\text{V} \leq V^+ \leq 5.5\text{V}$ $V^- = 0\text{V}$, $V_{CM} = 0\text{V}$	83 83	120		dB
CMVR	Common Mode Voltage Range	CMRR $\geq 80\text{ dB}$ CMRR $\geq 79\text{ dB}$	–0.2 –0.2		4.2 4.2	V
A_{VOL}	Large Signal Voltage Gain	$V_O = 0.3\text{V}$ to 4.7V $R_L = 10\text{ k}\Omega$ to $V^+/2$	110 108	120		dB
V_O	Output Swing High	$R_L = 10\text{ k}\Omega$ to $V^+/2$ $V_{IN}(\text{diff}) = 100\text{ mV}$		17	50 50	mV from either rail
	Output Swing Low	$R_L = 10\text{ k}\Omega$ to $V^+/2$ $V_{IN}(\text{diff}) = -100\text{ mV}$		17	50 50	
I_O	Output Current ⁽⁴⁾	Sourcing, V_O to V^- $V_{IN}(\text{diff}) = 100\text{ mV}$	27 19	30		mA
		Sinking, V_O to V^+ $V_{IN}(\text{diff}) = -100\text{ mV}$	17 12	22		
I_S	Supply Current			19	27 28	μA

- (1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that $T_J = T_A$. No specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where $T_J > T_A$. Absolute Maximum Ratings indicate junction temperature limits beyond which the device may be permanently degraded, either mechanically or electrically.
- (2) All limits are specified by testing, statistical analysis or design.
- (3) Typical values represent the most likely parametric norm at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not ensured on shipped production material.
- (4) The short circuit test is a momentary open loop test.

5V AC Electrical Characteristics⁽¹⁾

Unless otherwise specified, all limits ensured for $T_A = 25^\circ\text{C}$, $V^+ = 5\text{V}$, $V^- = 0\text{V}$, $V_{\text{CM}} = V_O = V^+/2$, and $R_L > 1\text{ M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min ⁽²⁾	Typ ⁽³⁾	Max ⁽²⁾	Units
GBW	Gain-Bandwidth Product	$C_L = 20\text{ pF}$, $R_L = 10\text{ k}\Omega$		130		kHz
SR	Slew Rate	$A_V = +1$	Falling Edge	33 32	58	V/ms
			Rising Edge	33 32	48	
θ_m	Phase Margin	$C_L = 20\text{ pF}$, $R_L = 10\text{ k}\Omega$		68		deg
G_m	Gain Margin	$C_L = 20\text{ pF}$, $R_L = 10\text{ k}\Omega$		27		dB
e_n	Input-Referred Voltage Noise Density	$f = 1\text{ kHz}$		60		$\text{nV}/\sqrt{\text{Hz}}$
	Input Referred Voltage Noise	0.1 Hz to 10 Hz		2.3		μV_{PP}
i_n	Input-Referred Current Noise	$f = 1\text{ kHz}$		10		$\text{fA}/\sqrt{\text{Hz}}$
THD+N	Total Harmonic Distortion + Noise	$f = 100\text{ Hz}$, $R_L = 10\text{ k}\Omega$		0.002		%

- (1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that $T_J = T_A$. No specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where $T_J > T_A$. Absolute Maximum Ratings indicate junction temperature limits beyond which the device may be permanently degraded, either mechanically or electrically.
- (2) All limits are specified by testing, statistical analysis or design.
- (3) Typical values represent the most likely parametric norm at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not ensured on shipped production material.

3.3V DC Electrical Characteristics⁽¹⁾

Unless otherwise specified, all limits ensured for $T_A = 25^\circ\text{C}$, $V^+ = 3.3\text{V}$, $V^- = 0\text{V}$, $V_{\text{CM}} = V_O = V^+/2$, and $R_L > 1\text{ M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min ⁽²⁾	Typ ⁽³⁾	Max ⁽²⁾	Units
V_{OS}	Input Offset Voltage			± 10	± 160 ± 250	μV
TCV_{OS}	Input Offset Voltage Drift	LMP2232A		± 0.3	± 0.5	$\mu\text{V}/^\circ\text{C}$
		LMP2232B		± 0.3	± 2.5	
I_{BIAS}	Input Bias Current			0.02	± 3 ± 125	pA
I_{OS}	Input Offset Current			5		fA
CMRR	Common Mode Rejection Ratio	$0\text{V} \leq V_{\text{CM}} \leq 2.3\text{V}$	79 77	92		dB
PSRR	Power Supply Rejection Ratio	$1.6\text{V} \leq V^+ \leq 5.5\text{V}$ $V^- = 0\text{V}$, $V_{\text{CM}} = 0\text{V}$	83 83	120		dB
CMVR	Common Mode Voltage Range	CMRR $\geq 78\text{ dB}$ CMRR $\geq 77\text{ dB}$	-0.2 -0.2		2.5 2.5	V
A_{VOL}	Large Signal Voltage Gain	$V_O = 0.3\text{V}$ to 3V $R_L = 10\text{ k}\Omega$ to $V^+/2$	108 107	120		dB
V_O	Output Swing High	$R_L = 10\text{ k}\Omega$ to $V^+/2$ $V_{\text{IN}}(\text{diff}) = 100\text{ mV}$		14	50 50	mV from either rail
	Output Swing Low	$R_L = 10\text{ k}\Omega$ to $V^+/2$ $V_{\text{IN}}(\text{diff}) = -100\text{ mV}$		14	50 50	

- (1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that $T_J = T_A$. No specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where $T_J > T_A$. Absolute Maximum Ratings indicate junction temperature limits beyond which the device may be permanently degraded, either mechanically or electrically.
- (2) All limits are specified by testing, statistical analysis or design.
- (3) Typical values represent the most likely parametric norm at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not ensured on shipped production material.

3.3V DC Electrical Characteristics⁽¹⁾ (continued)

Unless otherwise specified, all limits ensured for $T_A = 25^\circ\text{C}$, $V^+ = 3.3\text{V}$, $V^- = 0\text{V}$, $V_{\text{CM}} = V_O = V^+/2$, and $R_L > 1\text{M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min ⁽²⁾	Typ ⁽³⁾	Max ⁽²⁾	Units
I_O	Output Current ⁽⁴⁾	Sourcing, V_O to V^- $V_{\text{IN(diff)}} = 100\text{ mV}$	11 8	14		mA
		Sinking, V_O to V^+ $V_{\text{IN(diff)}} = -100\text{ mV}$	8 5	11		
I_S	Supply Current			17	25 26	μA

(4) The short circuit test is a momentary open loop test.

3.3V AC Electrical Characteristics⁽¹⁾

Unless otherwise is specified, all limits ensured for $T_A = 25^\circ\text{C}$, $V^+ = 3.3\text{V}$, $V^- = 0\text{V}$, $V_{\text{CM}} = V_O = V^+/2$, and $R_L > 1\text{M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min ⁽²⁾	Typ ⁽³⁾	Max ⁽²⁾	Units
GBW	Gain-Bandwidth Product	$C_L = 20\text{ pF}$, $R_L = 10\text{ k}\Omega$		128		kHz
SR	Slew Rate	$A_V = +1$, $C_L = 20\text{ pF}$ $R_L = 10\text{ k}\Omega$	Falling Edge	58		V/ms
			Rising Edge	48		
θ_m	Phase Margin	$C_L = 20\text{ pF}$, $R_L = 10\text{ k}\Omega$		66		deg
G_m	Gain Margin	$C_L = 20\text{ pF}$, $R_L = 10\text{ k}\Omega$		26		dB
e_n	Input-Referred Voltage Noise Density	$f = 1\text{ kHz}$		60		$\text{nV}/\sqrt{\text{Hz}}$
	Input-Referred Voltage Noise	0.1 Hz to 10 Hz		2.4		μV_{PP}
i_n	Input-Referred Current Noise	$f = 1\text{ kHz}$		10		$\text{fA}/\sqrt{\text{Hz}}$
THD+N	Total Harmonic Distortion + Noise	$f = 100\text{ Hz}$, $R_L = 10\text{ k}\Omega$		0.003		%

- (1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that $T_J = T_A$. No specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where $T_J > T_A$. Absolute Maximum Ratings indicate junction temperature limits beyond which the device may be permanently degraded, either mechanically or electrically.
- (2) All limits are specified by testing, statistical analysis or design.
- (3) Typical values represent the most likely parametric norm at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not ensured on shipped production material.

2.5V DC Electrical Characteristics⁽¹⁾

Unless otherwise specified, all limits ensured for $T_A = 25^\circ\text{C}$, $V^+ = 2.5\text{V}$, $V^- = 0\text{V}$, $V_{\text{CM}} = V_O = V^+/2$, and $R_L > 1\text{M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min ⁽²⁾	Typ ⁽³⁾	Max ⁽²⁾	Units
V_{OS}	Input Offset Voltage			± 10	± 190 ± 275	μV
TCV_{OS}	Input Offset Voltage Drift	LMP2232A		± 0.3	± 0.5	$\mu\text{V}/^\circ\text{C}$
		LMP2232B		± 0.3	± 2.5	
I_{Bias}	Input Bias Current			0.02	± 3 ± 125	pA
I_{OS}	Input Offset Current			5		fA
CMRR	Common Mode Rejection Ratio	$0\text{V} \leq V_{\text{CM}} \leq 1.5\text{V}$	77 76	91		dB

- (1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that $T_J = T_A$. No specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where $T_J > T_A$. Absolute Maximum Ratings indicate junction temperature limits beyond which the device may be permanently degraded, either mechanically or electrically.
- (2) All limits are specified by testing, statistical analysis or design.
- (3) Typical values represent the most likely parametric norm at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not ensured on shipped production material.

2.5V DC Electrical Characteristics⁽¹⁾ (continued)

Unless otherwise specified, all limits ensured for $T_A = 25^\circ\text{C}$, $V^+ = 2.5\text{V}$, $V^- = 0\text{V}$, $V_{CM} = V_O = V^+/2$, and $R_L > 1\text{M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min ⁽²⁾	Typ ⁽³⁾	Max ⁽²⁾	Units
PSRR	Power Supply Rejection Ratio	$1.6\text{V} \leq V^+ \leq 5.5\text{V}$ $V^- = 0\text{V}$, $V_{CM} = 0\text{V}$	83 83	120		dB
CMVR	Common Mode Voltage Range	$\text{CMRR} \geq 77\text{ dB}$ $\text{CMRR} \geq 76\text{ dB}$	-0.2 -0.2		1.7 1.7	V
A_{VOL}	Large Signal Voltage Gain	$V_O = 0.3\text{V}$ to 2.2V $R_L = 10\text{ k}\Omega$ to $V^+/2$	104 104	120		dB
V_O	Output Swing High	$R_L = 10\text{ k}\Omega$ to $V^+/2$ $V_{IN}(\text{diff}) = 100\text{ mV}$		12	50 50	mV from either rail
	Output Swing Low	$R_L = 10\text{ k}\Omega$ to $V^+/2$ $V_{IN}(\text{diff}) = -100\text{ mV}$		13	50 50	
I_O	Output Current ⁽⁴⁾	Sourcing, V_O to V^- $V_{IN}(\text{diff}) = 100\text{ mV}$	5 4	8		mA
		Sinking, V_O to V^+ $V_{IN}(\text{diff}) = -100\text{ mV}$	3.5 2.5	7		
I_S	Supply Current			16	24 25	μA

(4) The short circuit test is a momentary open loop test.

2.5V AC Electrical Characteristics⁽¹⁾

Unless otherwise specified, all limits specified for $T_A = 25^\circ\text{C}$, $V^+ = 2.5\text{V}$, $V^- = 0\text{V}$, $V_{CM} = V_O = V^+/2$, and $R_L > 1\text{M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min ⁽²⁾	Typ ⁽³⁾	Max ⁽²⁾	Units
GBW	Gain-Bandwidth Product	$C_L = 20\text{ pF}$, $R_L = 10\text{ k}\Omega$		128		kHz
SR	Slew Rate	$A_V = +1$, $C_L = 20\text{ pF}$ $R_L = 10\text{ k}\Omega$	Falling Edge	58		V/ms
			Rising Edge	48		
θ_m	Phase Margin	$C_L = 20\text{ pF}$, $R_L = 10\text{ k}\Omega$		64		deg
G_m	Gain Margin	$C_L = 20\text{ pF}$, $R_L = 10\text{ k}\Omega$		26		dB
e_n	Input-Referred Voltage Noise Density	$f = 1\text{ kHz}$		60		$\text{nV}/\sqrt{\text{Hz}}$
	Input-Referred Voltage Noise	0.1 Hz to 10 Hz		2.5		μV_{PP}
i_n	Input-Referred Current Noise	$f = 1\text{ kHz}$		10		$\text{fA}/\sqrt{\text{Hz}}$
THD+N	Total Harmonic Distortion + Noise	$f = 100\text{ Hz}$, $R_L = 10\text{ k}\Omega$		0.005		%

- (1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that $T_J = T_A$. No specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where $T_J > T_A$. Absolute Maximum Ratings indicate junction temperature limits beyond which the device may be permanently degraded, either mechanically or electrically.
- (2) All limits are specified by testing, statistical analysis or design.
- (3) Typical values represent the most likely parametric norm at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not ensured on shipped production material.

1.8V DC Electrical Characteristics ⁽¹⁾

Unless otherwise specified, all limits ensured for $T_A = 25^\circ\text{C}$, $V^+ = 1.8\text{V}$, $V^- = 0\text{V}$, $V_{\text{CM}} = V_O = V^+/2$, and $R_L > 1\text{ M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min ⁽²⁾	Typ ⁽³⁾	Max ⁽²⁾	Units
V_{OS}	Input Offset Voltage			± 10	± 230 ± 325	μV
TCV_{OS}	Input Offset Voltage Drift	LMP2232A		± 0.3	± 0.5	$\mu\text{V}/^\circ\text{C}$
		LMP2232B		± 0.3	± 2.5	
I_{BIAS}	Input Bias Current			0.02	± 3 ± 125	pA
I_{OS}	Input Offset Current			5		fA
CMRR	Common Mode Rejection Ratio	$0\text{V} \leq V_{\text{CM}} \leq 0.8\text{V}$	76 75	92		dB
PSRR	Power Supply Rejection Ratio	$1.6\text{V} \leq V^+ \leq 5.5\text{V}$ $V^- = 0\text{V}$, $V_{\text{CM}} = 0\text{V}$	83 83	120		dB
CMVR	Common Mode Voltage Range	CMRR $\geq 76\text{ dB}$ CMRR $\geq 75\text{ dB}$	-0.2 0		1.0 1.0	V
A_{VOL}	Large Signal Voltage Gain	$V_O = 0.3\text{V}$ to 1.5V $R_L = 10\text{ k}\Omega$ to $V^+/2$	103 103	120		dB
V_O	Output Swing High	$R_L = 10\text{ k}\Omega$ to $V^+/2$ $V_{\text{IN}}(\text{diff}) = 100\text{ mV}$		12	50 50	mV from either rail
	Output Swing Low	$R_L = 10\text{ k}\Omega$ to $V^+/2$ $V_{\text{IN}}(\text{diff}) = -100\text{ mV}$		13	50 50	
I_O	Output Current ⁽⁴⁾	Sourcing, V_O to V^- $V_{\text{IN}}(\text{diff}) = 100\text{ mV}$	2.5 2	5		mA
		Sinking, V_O to V^+ $V_{\text{IN}}(\text{diff}) = -100\text{ mV}$	2 1.5	5		
I_S	Supply Current			16	24 25	μA

- (1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that $T_J = T_A$. No specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where $T_J > T_A$. Absolute Maximum Ratings indicate junction temperature limits beyond which the device may be permanently degraded, either mechanically or electrically.
- (2) All limits are specified by testing, statistical analysis or design.
- (3) Typical values represent the most likely parametric norm at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not ensured on shipped production material.
- (4) The short circuit test is a momentary open loop test.

1.8V AC Electrical Characteristics ⁽¹⁾

Unless otherwise is specified, all limits ensured for $T_A = 25^\circ\text{C}$, $V^+ = 1.8\text{V}$, $V^- = 0\text{V}$, $V_{\text{CM}} = V_O = V^+/2$, and $R_L > 1\text{ M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min ⁽²⁾	Typ ⁽³⁾	Max ⁽²⁾	Units
GBW	Gain-Bandwidth Product	$C_L = 20\text{ pF}$, $R_L = 10\text{ k}\Omega$		127		kHz
SR	Slew Rate	$A_V = +1$, $C_L = 20\text{ pF}$ $R_L = 10\text{ k}\Omega$	Falling Edge	58		V/ms
			Rising Edge	48		
θ_m	Phase Margin	$C_L = 20\text{ pF}$, $R_L = 10\text{ k}\Omega$		60		deg
G_m	Gain Margin	$C_L = 20\text{ pF}$, $R_L = 10\text{ k}\Omega$		25		dB
e_n	Input-Referred Voltage Noise Density	$f = 1\text{ kHz}$		60		$\text{nV}/\sqrt{\text{Hz}}$
	Input-Referred Voltage Noise	0.1 Hz to 10 Hz		2.4		μV_{PP}

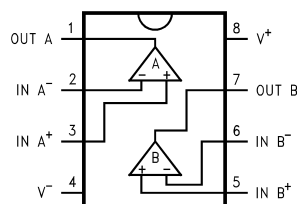
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- (2) All limits are specified by testing, statistical analysis or design.
- (3) Typical values represent the most likely parametric norm at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not ensured on shipped production material.

1.8V AC Electrical Characteristics ⁽¹⁾ (continued)

Unless otherwise is specified, all limits ensured for $T_A = 25^\circ\text{C}$, $V^+ = 1.8\text{V}$, $V^- = 0\text{V}$, $V_{\text{CM}} = V_O = V^+/2$, and $R_L > 1\text{ M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min ⁽²⁾	Typ ⁽³⁾	Max ⁽²⁾	Units
i_n	Input-Referred Current Noise	$f = 1\text{ kHz}$		10		$\text{fA}/\sqrt{\text{Hz}}$
THD+N	Total Harmonic Distortion + Noise	$f = 100\text{ Hz}$, $R_L = 10\text{ k}\Omega$		0.005		%

Connection Diagram



**Figure 2. 8-Pin VSSOP/SOIC (Top View)
Package Numbers DGK0008A and D0008A**

Typical Performance Characteristics

Unless otherwise Specified: $T_A = 25^\circ\text{C}$, $V_S = 5\text{V}$, $V_{CM} = V_S/2$, where $V_S = V^+ - V^-$

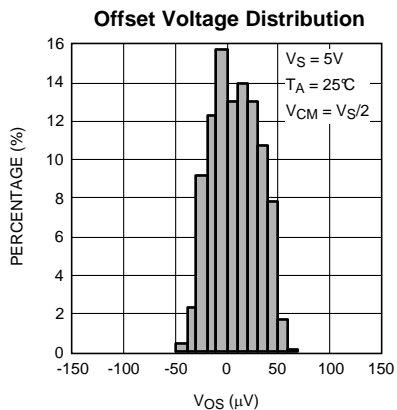


Figure 3.

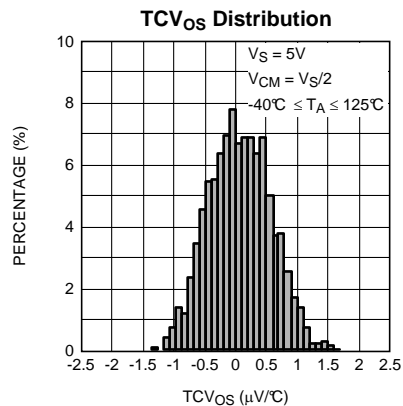


Figure 4.

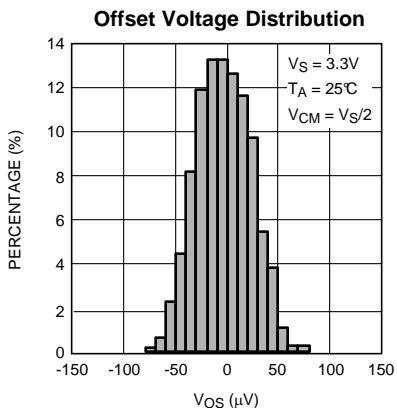


Figure 5.

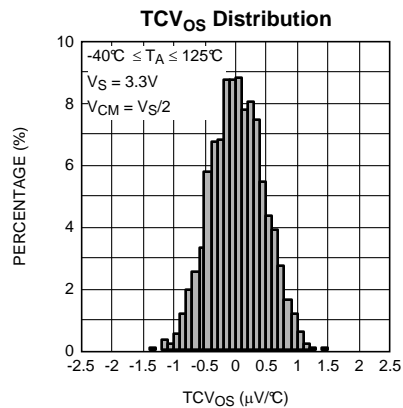


Figure 6.

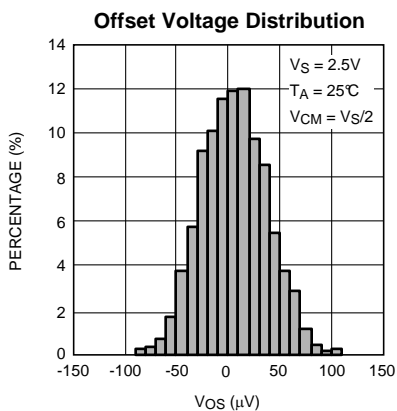


Figure 7.

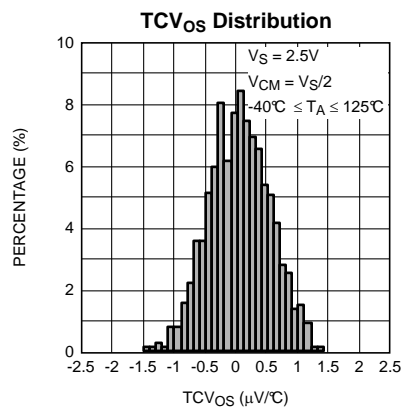


Figure 8.

Typical Performance Characteristics (continued)

Unless otherwise Specified: $T_A = 25^\circ\text{C}$, $V_S = 5\text{V}$, $V_{CM} = V_S/2$, where $V_S = V^+ - V^-$

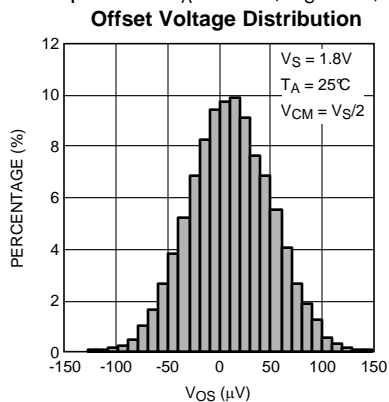


Figure 9.

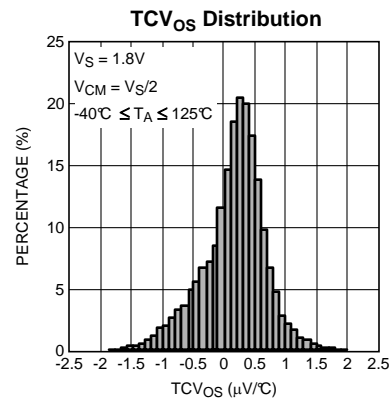


Figure 10.

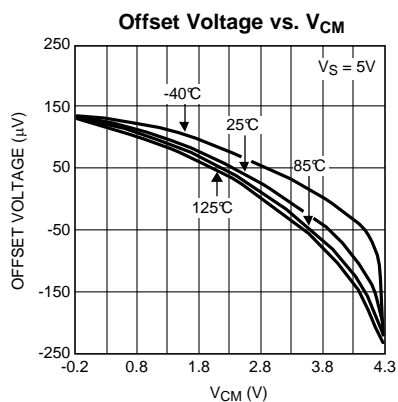


Figure 11.

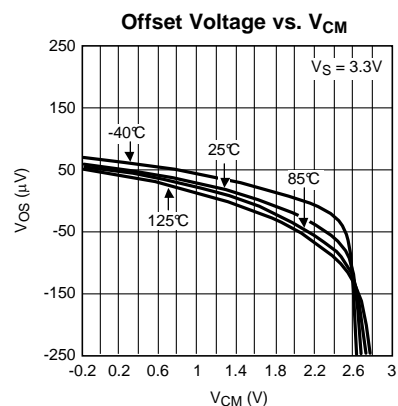


Figure 12.

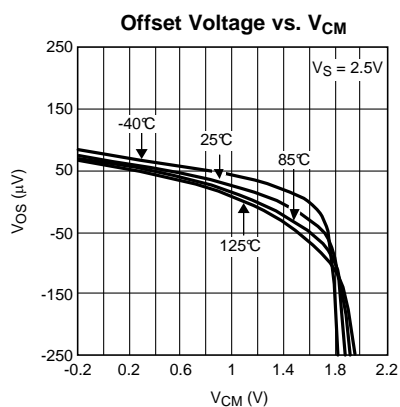


Figure 13.

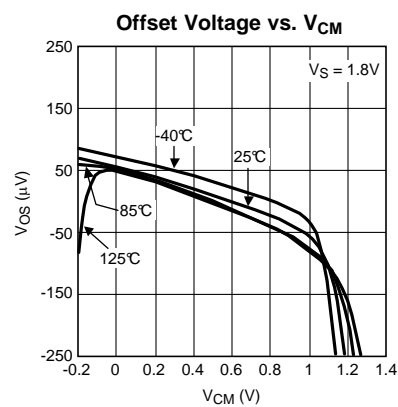


Figure 14.

Typical Performance Characteristics (continued)

Unless otherwise Specified: $T_A = 25^\circ\text{C}$, $V_S = 5\text{V}$, $V_{CM} = V_S/2$, where $V_S = V^+ - V^-$

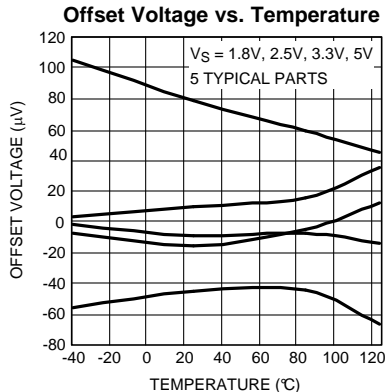


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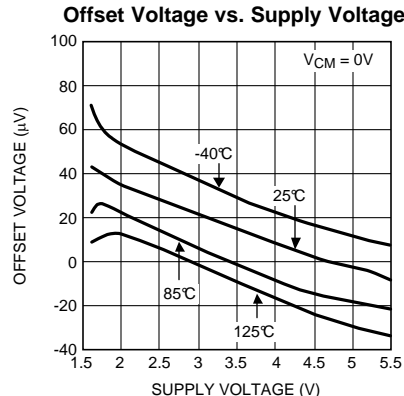


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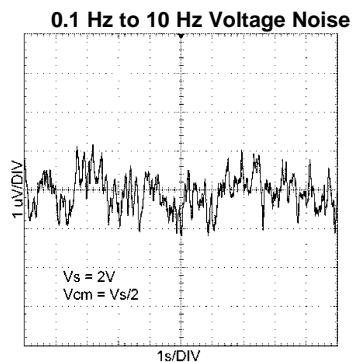


Figure 17.

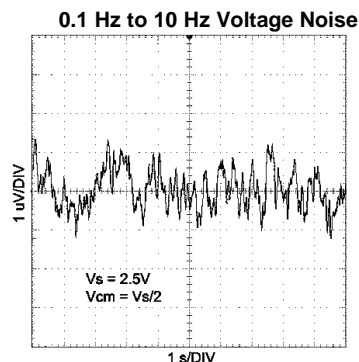


Figure 18.

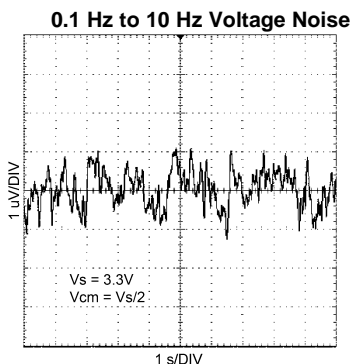


Figure 19.

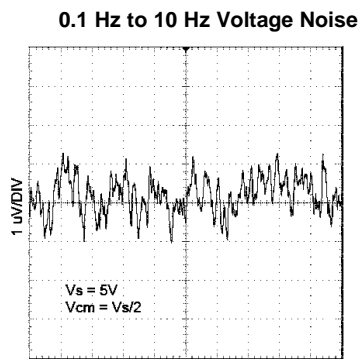


Figure 20.

Typical Performance Characteristics (continued)

Unless otherwise Specified: $T_A = 25^\circ\text{C}$, $V_S = 5\text{V}$, $V_{CM} = V_S/2$, where $V_S = V^+ - V^-$

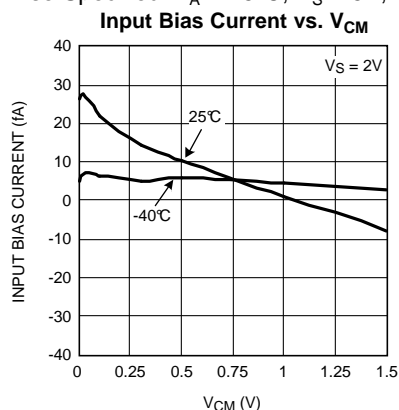


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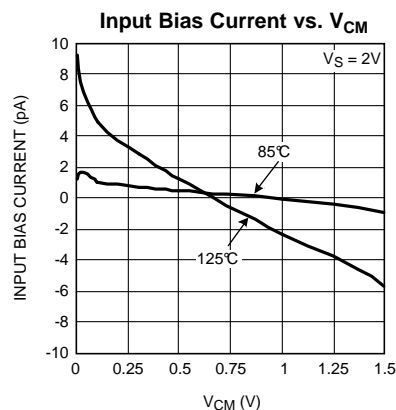


Figure 22.

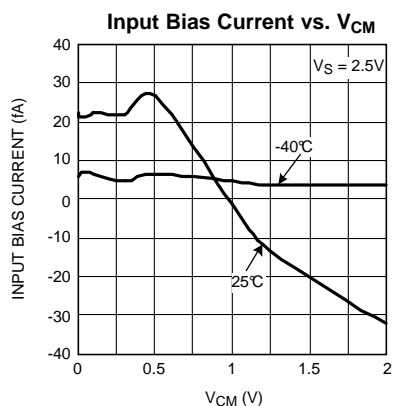


Figure 23.

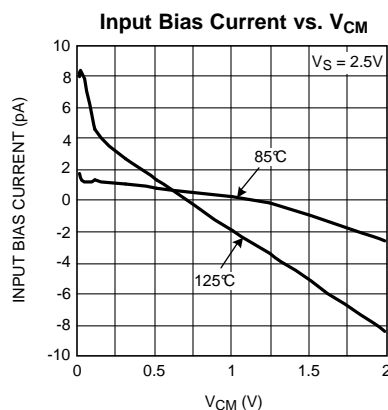


Figure 24.

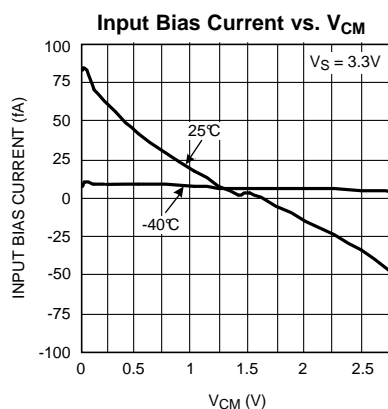


Figure 25.

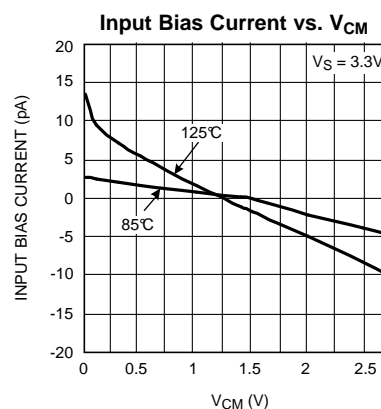


Figure 26.

Typical Performance Characteristics (continued)

Unless otherwise Specified: $T_A = 25^\circ\text{C}$, $V_S = 5\text{V}$, $V_{CM} = V_S/2$, where $V_S = V^+ - V^-$

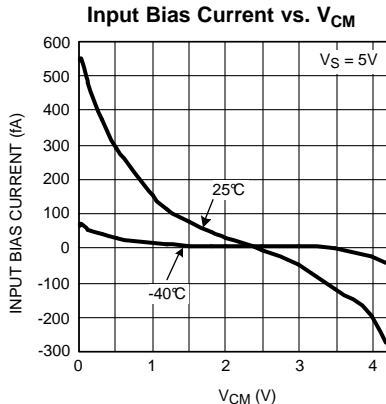


Figure 27.

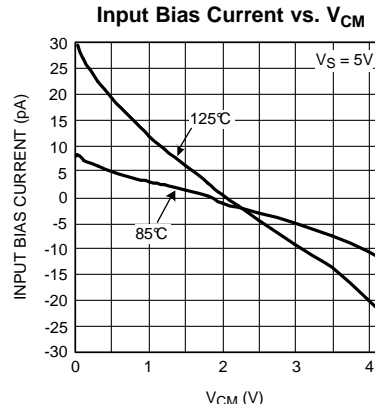


Figure 28.

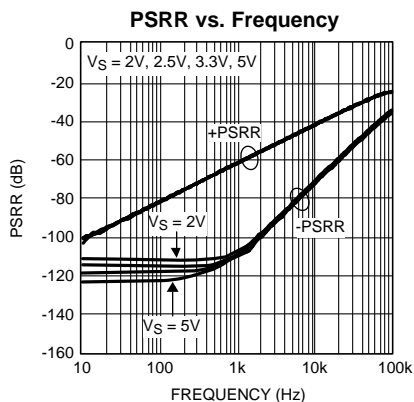


Figure 29.

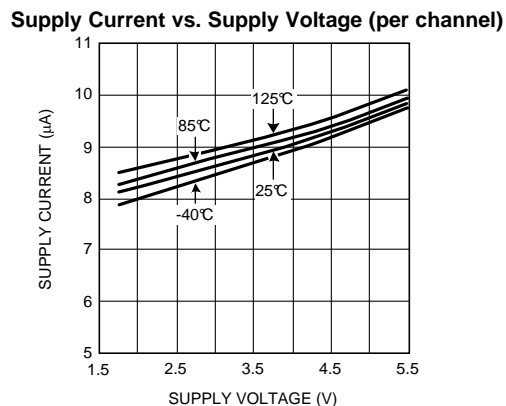


Figure 30.

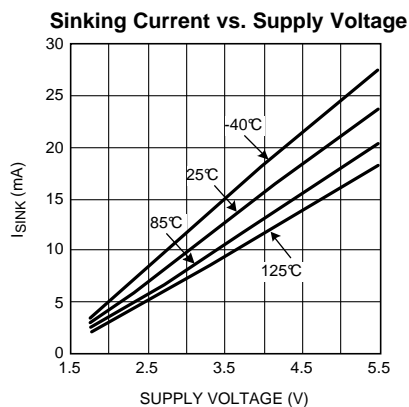


Figure 31.

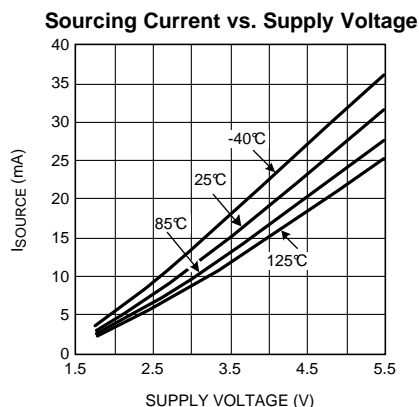


Figure 32.

Typical Performance Characteristics (continued)

Unless otherwise Specified: $T_A = 25^\circ\text{C}$, $V_S = 5\text{V}$, $V_{CM} = V_S/2$, where $V_S = V^+ - V^-$

Output Swing High vs. Supply Voltage

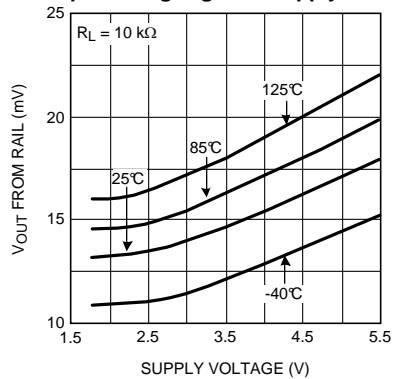


Figure 33.

Output Swing Low vs. Supply Voltage

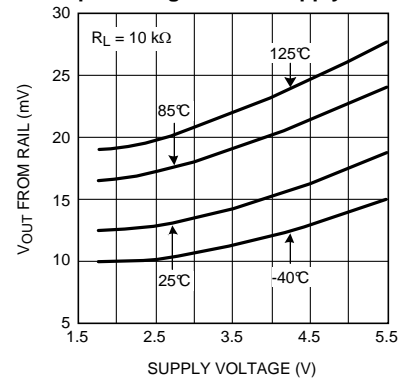


Figure 34.

Open Loop Frequency Response

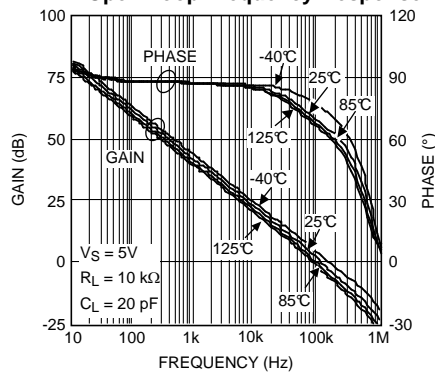


Figure 35.

Open Loop Frequency Response

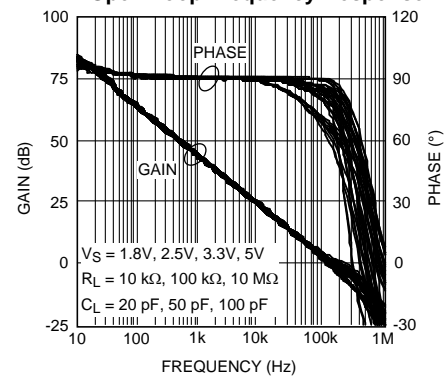


Figure 36.

Phase Margin vs. Capacitive Load

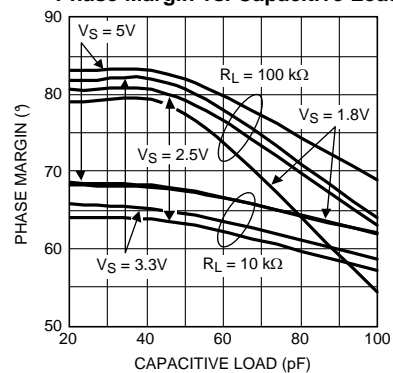


Figure 37.

Slew Rate vs. Supply Voltage

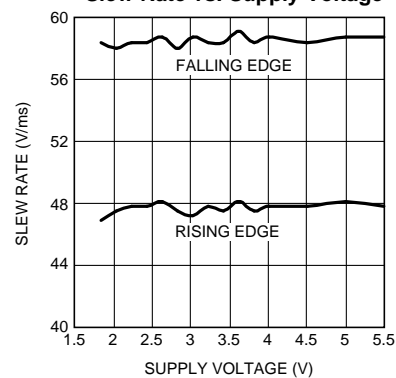


Figure 38.

Typical Performance Characteristics (continued)

Unless otherwise Specified: $T_A = 25^\circ\text{C}$, $V_S = 5\text{V}$, $V_{CM} = V_S/2$, where $V_S = V^+ - V^-$

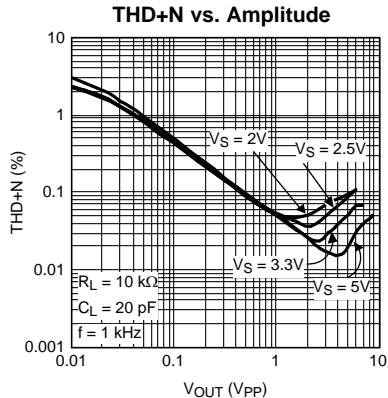


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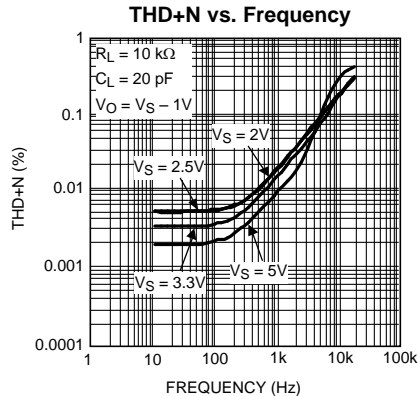


Figure 40.

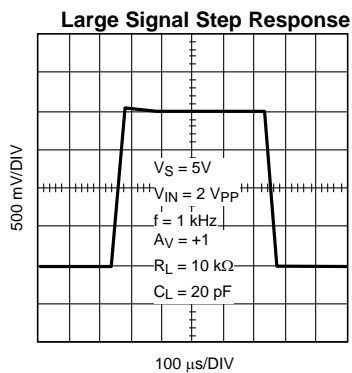


Figure 41.

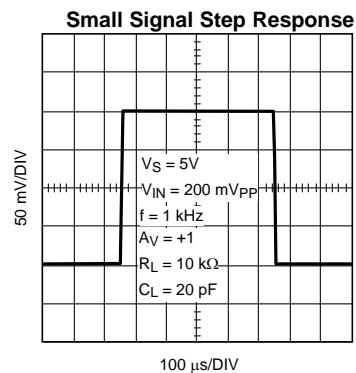


Figure 42.

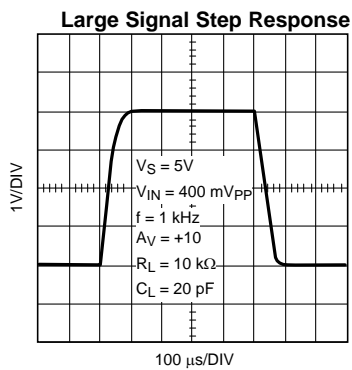


Figure 43.

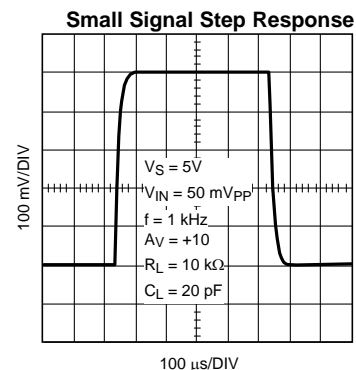


Figure 44.

Typical Performance Characteristics (continued)

Unless otherwise Specified: $T_A = 25^\circ\text{C}$, $V_S = 5\text{V}$, $V_{CM} = V_S/2$, where $V_S = V^+ - V^-$

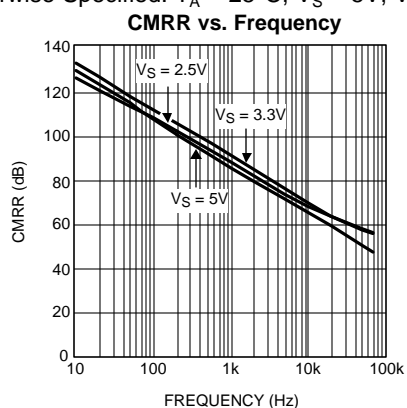


Figure 45.

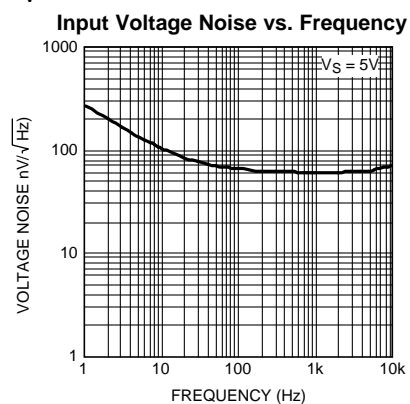


Figure 46.

APPLICATION INFORMATION

LMP2232

The LMP2232 is a quad CMOS precision amplifier that offers low offset voltage, low offset voltage drift, and high gain while consuming less than 10 μA of supply current per channel.

The LMP2232 is a micropower op amp, consuming only 36 μA of current. Micropower op amps extend the run time of battery powered systems and reduce energy consumption in energy limited systems. The ensured supply voltage range of 1.8V to 5.0V along with the ultra-low supply current extend the battery run time in two ways. The extended ensured power supply voltage range of 1.8V to 5.0V enables the op amp to function when the battery voltage has depleted from its nominal value down to 1.8V. In addition, the lower power consumption increases the life of the battery.

The LMP2232 has input referred offset voltage of only $\pm 150 \mu\text{V}$ maximum at room temperature. This offset is ensured to be less than $\pm 230 \mu\text{V}$ over temperature. This minimal offset voltage along with very low TCV_{OS} of only $0.3 \mu\text{V}/^\circ\text{C}$ typical allows more accurate signal detection and amplification in precision applications.

The low input bias current of only $\pm 20 \text{ fA}$ gives the LMP2232 superiority for use in high impedance sensor applications. Bias current of an amplifier flows through source resistance of the sensor and the voltage resulting from this current flow appears as a noise voltage on the input of the amplifier. The low input bias current enables the LMP2232 to interface with high impedance sensors while generating negligible voltage noise. Thus the LMP2232 provides better signal fidelity and a higher signal-to-noise ratio when interfacing with high impedance sensors.

Texas Instruments is heavily committed to precision amplifiers and the market segments they serve. Technical support and extensive characterization data is available for sensitive applications or applications with a constrained error budget.

The operating voltage range of 1.6V to 5.5V over the extensive temperature range of -40°C to 125°C makes the LMP2232 an excellent choice for low voltage precision applications with extensive temperature requirements.

The LMP2232 is offered in the 8-pin VSSOP and 8-pin SOIC packages. These small packages are ideal solutions for area constrained PC boards and portable electronics.

TOTAL NOISE CONTRIBUTION

The LMP2232 has very low input bias current, very low input current noise, and low input voltage noise for micropower amplifiers. As a result, these amplifiers make great choices for circuits with high impedance sensor applications.

[Figure 47](#) shows the typical input noise of the LMP2232 as a function of source resistance where:

e_n denotes the input referred voltage noise

e_i is the voltage drop across source resistance due to input referred current noise or $e_i = R_S * i_n$

e_t shows the thermal noise of the source resistance

e_{ni} shows the total noise on the input.

Where:

$$e_{ni} = \sqrt{e_n^2 + e_i^2 + e_t^2}$$

The input current noise of the LMP2232 is so low that it will not become the dominant factor in the total noise unless source resistance exceeds 300 $\text{M}\Omega$, which is an unrealistically high value. As is evident in [Figure 47](#), at lower R_S values, total noise is dominated by the amplifier's input voltage noise. Once R_S is larger than a 100 $\text{k}\Omega$, then the dominant noise factor becomes the thermal noise of R_S . As mentioned before, the current noise will not be the dominant noise factor for any practical application.

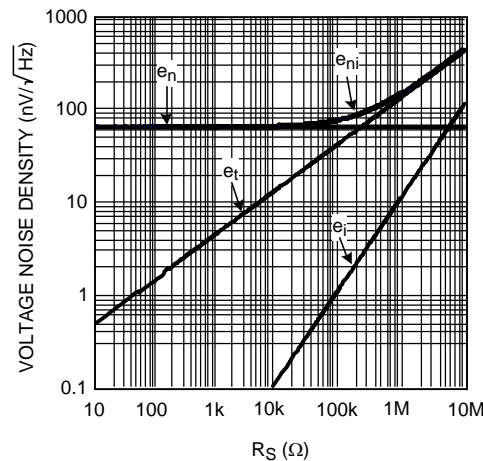


Figure 47. Total Input Noise

VOLTAGE NOISE REDUCTION

The LMP2232 has an input voltage noise of $60\text{nV}/\sqrt{\text{Hz}}$. While this value is very low for micropower amplifiers, this input voltage noise can be further reduced by placing N amplifiers in parallel as shown in Figure 48. The total voltage noise on the output of this circuit is divided by the square root of the number of amplifiers used in this parallel combination. This is because each individual amplifier acts as an independent noise source, and the average noise of independent sources is the quadrature sum of the independent sources divided by the number of sources. For N identical amplifiers, this means:

$$\begin{aligned}
 \text{REDUCED INPUT VOLTAGE NOISE} &= \frac{1}{N} \sqrt{e_{n1}^2 + e_{n2}^2 + \dots + e_{nN}^2} \\
 &= \frac{1}{N} \sqrt{N e_n^2} = \frac{\sqrt{N}}{N} e_n \\
 &= \frac{1}{\sqrt{N}} e_n
 \end{aligned}$$

Figure 48 shows a schematic of this input voltage noise reduction circuit. Typical resistor values are: $R_G = 10\Omega$, $R_F = 1\text{ k}\Omega$, and $R_O = 1\text{ k}\Omega$.

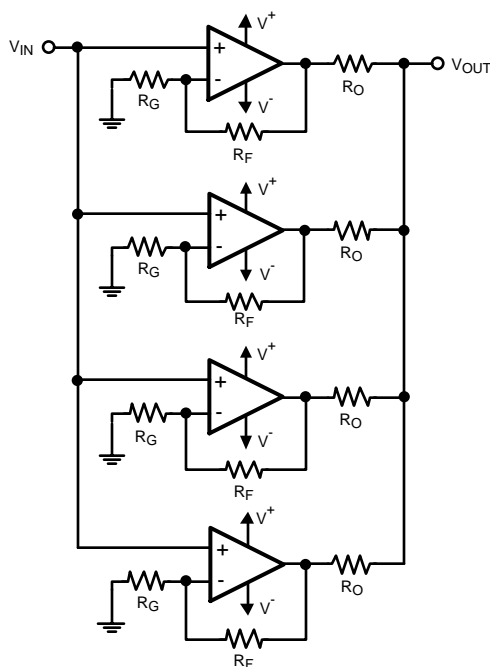


Figure 48. Noise Reduction Circuit

PRECISION INSTRUMENTATION AMPLIFIER

Measurement of very small signals with an amplifier requires close attention to the input impedance of the amplifier, gain of the signal on the inputs, and the gain on each input of the amplifier. This is because the difference of the input signal on the two inputs is of the interest and the common signal is considered noise. A classic circuit implementation is an instrumentation amplifier. Instrumentation amplifiers have a finite, accurate, and stable gain. They also have extremely high input impedances and very low output impedances. Finally they have an extremely high CMRR so that the amplifier can only respond to the differential signal. A typical instrumentation amplifier is shown in Figure 49.

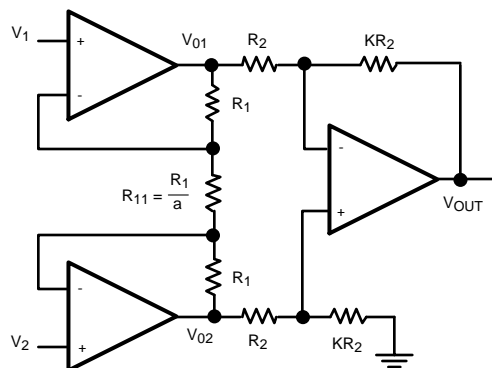


Figure 49. Instrumentation Amplifier

There are two stages in this amplifier. The last stage, output stage, is a differential amplifier. In an ideal case the two amplifiers of the first stage, the input stage, would be set up as buffers to isolate the inputs. However they cannot be connected as followers because of mismatch of amplifiers. That is why there is a balancing resistor between the two. The product of the two stages of gain will give the gain of the instrumentation amplifier. Ideally, the CMRR should be infinite. However the output stage has a small non-zero common mode gain which results from resistor mismatch.

In the input stage of the circuit, current is the same across all resistors. This is due to the high input impedance and low input bias current of the LMP2232.

$$\text{GIVEN: } I_{R_1} = I_{R_{11}} \quad (1)$$

By Ohm's Law:

$$\begin{aligned} V_{O1} - V_{O2} &= (2R_1 + R_{11}) I_{R_{11}} \\ &= (2a + 1) R_{11} \cdot I_{R_{11}} \\ &= (2a + 1) V_{R_{11}} \end{aligned} \quad (2)$$

However:

$$V_{R_{11}} = V_1 - V_2 \quad (3)$$

So we have:

$$V_{O1} - V_{O2} = (2a + 1)(V_1 - V_2) \quad (4)$$

Now looking at the output of the instrumentation amplifier:

$$\begin{aligned} V_O &= \frac{KR_2}{R_2} (V_{O2} - V_{O1}) \\ &= -K (V_{O1} - V_{O2}) \end{aligned} \quad (5)$$

Substituting from [Equation 4](#):

$$V_O = -K (2a + 1) (V_1 - V_2) \quad (6)$$

This shows the gain of the instrumentation amplifier to be:

$$-K(2a + 1) \quad (7)$$

Typical values for this circuit can be obtained by setting: $a = 12$ and $K = 4$. This results in an overall gain of -100 .

SINGLE SUPPLY STRAIN GAGE BRIDGE AMPLIFIER

Strain gauges are popular electrical elements used to measure force or pressure. Strain gauges are subjected to an unknown force which is measured as the deflection on a previously calibrated scale. Pressure is often measured using the same technique; however this pressure needs to be converted into force using an appropriate transducer. Strain gauges are often resistors which are sensitive to pressure or to flexing. Sense resistor values range from tens of ohms to several hundred kilo-ohms. The resistance change which is a result of applied force across the strain gauge might be 1% of its total value. An accurate and reliable system is needed to measure this small resistance change. Bridge configurations offer a reliable method for this measurement.

Bridge sensors are formed of four resistors, connected as a quadrilateral. A voltage source or a current source is used across one of the diagonals to excite the bridge while a voltage detector across the other diagonal measures the output voltage.

Bridges are mainly used as null circuits or to measure differential voltages. Bridges will have no output voltage if the ratios of two adjacent resistor values are equal. This fact is used in null circuit measurements. These are particularly used in feedback systems which involve electrochemical elements or human interfaces. Null systems force an active resistor, such as a strain gauge, to balance the bridge by influencing the measured parameter.

Often in sensor applications at least one of the resistors is a variable resistor, or a sensor. The deviation of this active element from its initial value is measured as an indication of change in the measured quantity. A change in output voltage represents the sensor value change. Since the sensor value change is often very small, the resulting output voltage is very small in magnitude as well. This requires an extensive and very precise amplification circuitry so that signal fidelity does not change after amplification.

Sensitivity of a bridge is the ratio of its maximum expected output change to the excitation voltage change.

Figure 50(a) shows a typical bridge sensor and Figure 50(b) shows the bridge with four sensors. R in Figure 50(b) is the nominal value of the sense resistor and the deviations from R are proportional to the quantity being measured.

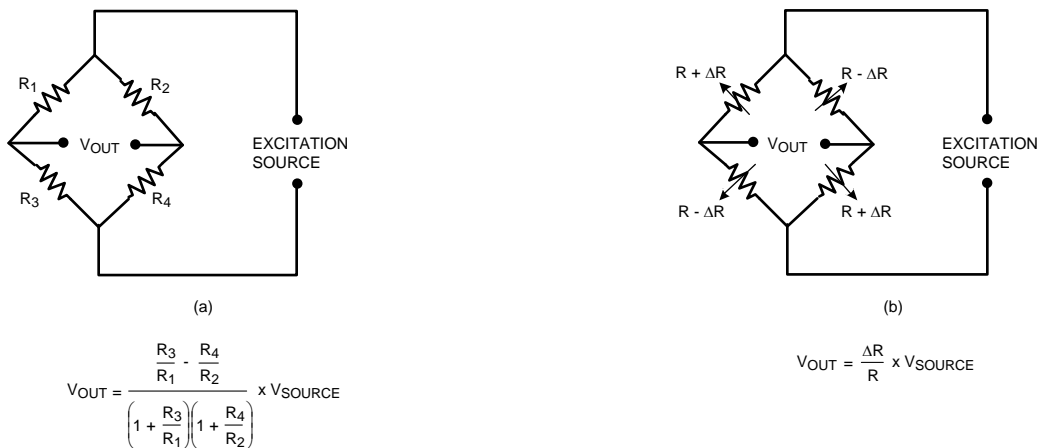


Figure 50. Bridge Sensor

Instrumentation amplifiers are great for interfacing with bridge sensors. Bridge sensors often sense a very small differential signal in the presence of a larger common mode voltage. Instrumentation amplifiers reject this common mode signal.

Figure 51 shows a strain gauge bridge amplifier. In this application one of the LMP2232 amplifiers is used to buffer the LM4140A's precision output voltage. The LM4140A is a precision voltage reference. The other three amplifiers in the LMP2232 are used to form an instrumentation amplifier. This instrumentation amplifier uses the LMP2232's high CMRR and low V_{OS} and TCV_{OS} to accurately amplify the small differential signal generated by the output of the bridge sensor. This amplified signal is then fed into the ADC121S021 which is a 12-bit analog to digital converter. This circuit works on a single supply voltage of 5V.

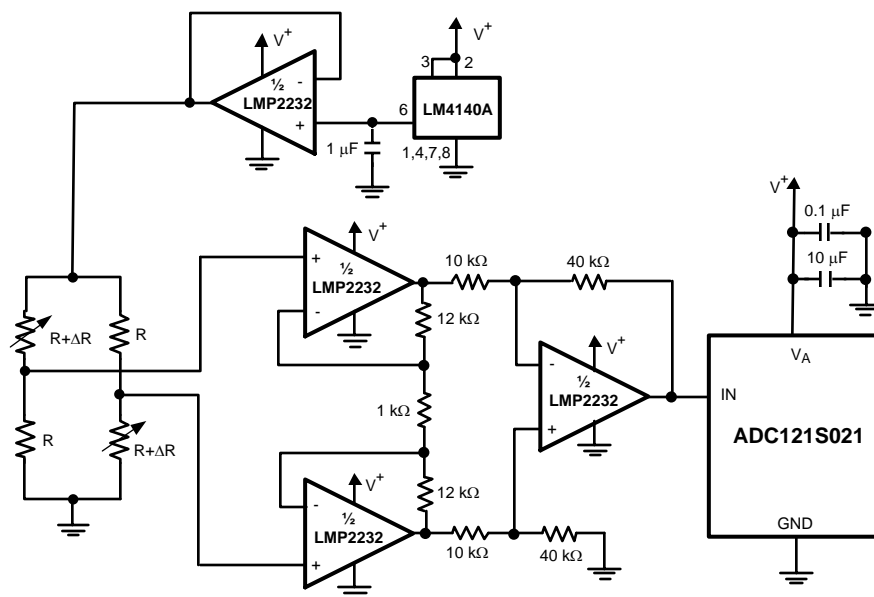


Figure 51. Strain Gauge Bridge Amplifier

PORTABLE GAS DETECTION SENSOR

Gas sensors are used in many different industrial and medical applications. They generate a current which is proportional to the percentage of a particular gas sensed in an air sample. This current goes through a load resistor and the resulting voltage drop is measured. Depending on the sensed gas and sensitivity of the sensor, the output current can be in the order of tens of microamperes to a few milliamperes. Gas sensor datasheets often specify a recommended load resistor value or they suggest a range of load resistors to choose from.

Oxygen sensors are used when air quality or oxygen delivered to a patient needs to be monitored. Fresh air contains 20.9% oxygen. Air samples containing less than 18% oxygen are considered dangerous. Oxygen sensors are also used in industrial applications where the environment must lack oxygen. An example is when food is vacuum packed. There are two main categories of oxygen sensors, those which sense oxygen when it is abundantly present (i.e. in air or near an oxygen tank) and those which detect very small traces of oxygen in ppm.

Figure 52 shows a typical circuit used to amplify the output signal of an oxygen detector. The LMP2232 makes an excellent choice for this application as it draws only 36 μA of current and operates on supply voltages down to 1.8V. This application detects oxygen in air. The oxygen sensor outputs a known current through the load resistor. This value changes with the amount of oxygen present in the air sample. Oxygen sensors usually recommend a particular load resistor value or specify a range of acceptable values for the load resistor. Oxygen sensors typically have a life of one to two years. The use of the micropower LMP2232 means minimal power usage by the op amp and it enhances the battery life. Depending on other components present in the circuit design, the battery could last for the entire life of the oxygen sensor. The precision specifications of the LMP2232, such as its very low offset voltage, low TCV_{OS} , low input bias current, low CMRR, and low PSRR are other factors which make the LMP2232 a great choice for this application..

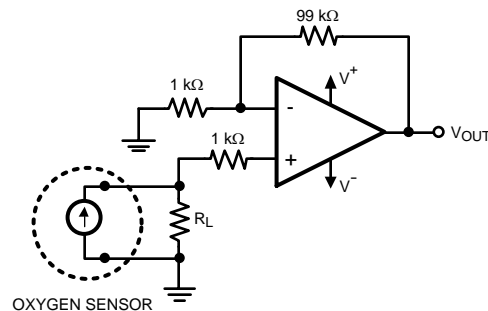


Figure 52. Precision Oxygen Sensor

REVISION HISTORY

Changes from Revision B (March 2013) to Revision C	Page
• Changed layout of National Data Sheet to TI format	22

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
LMP2232AMA/NOPB	Active	Production	SOIC (D) 8	95 TUBE	Yes	SN	Level-1-260C-UNLIM	-40 to 125	LMP22 32AMA
LMP2232AMA/NOPB.A	Active	Production	SOIC (D) 8	95 TUBE	Yes	SN	Level-1-260C-UNLIM	-40 to 125	LMP22 32AMA
LMP2232AMAE/NOPB	Active	Production	SOIC (D) 8	250 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	LMP22 32AMA
LMP2232AMAE/NOPB.A	Active	Production	SOIC (D) 8	250 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	LMP22 32AMA
LMP2232AMAX/NOPB	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	LMP22 32AMA
LMP2232AMAX/NOPB.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	LMP22 32AMA
LMP2232AMM/NOPB	Active	Production	VSSOP (DGK) 8	1000 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	AK5A
LMP2232AMM/NOPB.A	Active	Production	VSSOP (DGK) 8	1000 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	AK5A
LMP2232AMME/NOPB	Active	Production	VSSOP (DGK) 8	250 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	AK5A
LMP2232AMME/NOPB.A	Active	Production	VSSOP (DGK) 8	250 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	AK5A
LMP2232AMMX/NOPB	Active	Production	VSSOP (DGK) 8	3500 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	AK5A
LMP2232AMMX/NOPB.A	Active	Production	VSSOP (DGK) 8	3500 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	AK5A
LMP2232BMA/NOPB	Active	Production	SOIC (D) 8	95 TUBE	Yes	SN	Level-1-260C-UNLIM	-40 to 125	LMP22 32BMA
LMP2232BMA/NOPB.A	Active	Production	SOIC (D) 8	95 TUBE	Yes	SN	Level-1-260C-UNLIM	-40 to 125	LMP22 32BMA
LMP2232BMAE/NOPB	Active	Production	SOIC (D) 8	250 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	LMP22 32BMA
LMP2232BMAE/NOPB.A	Active	Production	SOIC (D) 8	250 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	LMP22 32BMA
LMP2232BMAX/NOPB	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	LMP22 32BMA
LMP2232BMAX/NOPB.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	LMP22 32BMA
LMP2232BMM/NOPB	Active	Production	VSSOP (DGK) 8	1000 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	AK5B
LMP2232BMM/NOPB.A	Active	Production	VSSOP (DGK) 8	1000 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	AK5B

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
LMP2232BMME/NOPB	Active	Production	VSSOP (DGK) 8	250 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	AK5B
LMP2232BMME/NOPB.A	Active	Production	VSSOP (DGK) 8	250 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	AK5B

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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TAPE AND REEL INFORMATION



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMP2232AMAE/NOPB	SOIC	D	8	250	177.8	12.4	6.5	5.4	2.0	8.0	12.0	Q1
LMP2232AMAX/NOPB	SOIC	D	8	2500	330.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1
LMP2232AMM/NOPB	VSSOP	DGK	8	1000	177.8	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LMP2232AMME/NOPB	VSSOP	DGK	8	250	177.8	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LMP2232AMMX/NOPB	VSSOP	DGK	8	3500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LMP2232BMAE/NOPB	SOIC	D	8	250	177.8	12.4	6.5	5.4	2.0	8.0	12.0	Q1
LMP2232BMAX/NOPB	SOIC	D	8	2500	330.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1
LMP2232BMM/NOPB	VSSOP	DGK	8	1000	177.8	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LMP2232BMME/NOPB	VSSOP	DGK	8	250	177.8	12.4	5.3	3.4	1.4	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMP2232AMAE/NOPB	SOIC	D	8	250	208.0	191.0	35.0
LMP2232AMAX/NOPB	SOIC	D	8	2500	356.0	356.0	35.0
LMP2232AMM/NOPB	VSSOP	DGK	8	1000	208.0	191.0	35.0
LMP2232AMME/NOPB	VSSOP	DGK	8	250	208.0	191.0	35.0
LMP2232AMMX/NOPB	VSSOP	DGK	8	3500	356.0	356.0	35.0
LMP2232BMAE/NOPB	SOIC	D	8	250	208.0	191.0	35.0
LMP2232BMAX/NOPB	SOIC	D	8	2500	356.0	356.0	35.0
LMP2232BMM/NOPB	VSSOP	DGK	8	1000	208.0	191.0	35.0
LMP2232BMME/NOPB	VSSOP	DGK	8	250	208.0	191.0	35.0

TUBE



*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
LMP2232AMA/NOPB	D	SOIC	8	95	495	8	4064	3.05
LMP2232AMA/NOPB.A	D	SOIC	8	95	495	8	4064	3.05
LMP2232BMA/NOPB	D	SOIC	8	95	495	8	4064	3.05
LMP2232BMA/NOPB.A	D	SOIC	8	95	495	8	4064	3.05



D0008A

PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES:

1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
4. This dimension does not include interlead flash.
5. Reference JEDEC registration MS-012, variation AA.

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:8X



SOLDER MASK DETAILS

4214825/C 02/2019

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE
BASED ON .005 INCH [0.125 MM] THICK STENCIL
SCALE:8X

4214825/C 02/2019

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

DGK0008A**PACKAGE OUTLINE****VSSOP - 1.1 mm max height**

SMALL OUTLINE PACKAGE



4214862/A 04/2023

NOTES:

PowerPAD is a trademark of Texas Instruments.

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-187.

EXAMPLE BOARD LAYOUT

DGK0008A

™ VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 15X



SOLDER MASK DETAILS

4214862/A 04/2023

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
8. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.
9. Size of metal pad may vary due to creepage requirement.

EXAMPLE STENCIL DESIGN

DGK0008A

TM VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE
SCALE: 15X

4214862/A 04/2023

NOTES: (continued)

11. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
12. Board assembly site may have different recommendations for stencil design.

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