

CDCE72010 Phase Noise Performance and Jitter Cleaning Ability

Madhu Balasubramanian

Serial Link Products

ABSTRACT

This application report presents phase noise data taken on the <u>CDCE72010</u> jitter cleaner and synchronizer PLL device from Texas Instruments. The phase noise performance of the CDCE72010 depends on the phase noise of the reference clock, VCXO clock, and the CDCE72010 itself. This application report shows the phase noise performance at several of the most popular CDMA frequencies. This data helps the user to choose the right clocking solution for specific applications. These test results confirm that the CDCE72010 can provide clocks better than -145dBc/Hz phase noise at 1-MHz offset from the carrier frequency. This type of low phase noise is required for wireless applications as well as many other high-performance sampling systems.

Contents

1	Introduction	2
2	Test Equipment and Setup	4
3	Total Phase Noise Measurements	6
4	Total Phase Noise Measurements	14
5	Additive Phase Noise Measurements	26

List of Figures

Passive Loop Filter Circuit	4
Phase Noise Measurement Test Setup	4
CDCE72010 Device Configuration	5
491.52-MHz HS-LVPECL Output Phase Noise	7
491.52-MHz HS-LVDS Output Phase Noise	7
245.76-MHz LVCMOS Output Phase Noise	8
245.76-MHz HS-LVPECL Output Phase Noise	8
245.76-MHz HS-LVDS Output Phase Noise	9
163.84-MHz LVCMOS Output Phase Noise	9
163.84-MHz HS-LVPECL Output Phase Noise	10
163.84-MHz HS-LVDS Output Phase Noise	10
122.88-MHz LVCMOS Output Phase Noise	11
122.88-MHz HS-LVPECL Output Phase Noise	11
122.88-MHz HS-LVDS Output Phase Noise	12
98.304-MHz LVCMOS Output Phase Noise	12
98.304-MHz HS-LVPECL Output Phase Noise	13
98.304-MHz HS-LVDS Output Phase Noise	13
122.88-MHz LVCMOS Output Jitter Cleaning Ability	15
122.88-MHz HS-LVPECL Output Jitter Cleaning Ability	15
122.88-MHz HS-LVDS Output Jitter Cleaning Ability	15
30.72-MHz LVCMOS Input Phase Noise Profile (1 ps, RMS)	16
122.88-MHz LVCMOS Output Phase Noise Profile (1 ps, RMS Input)	16
122.88-MHz HS-LVPECL Output Phase Noise Profile (1 ps, RMS Input)	17
122.88-MHz HS-LVDS Output Phase Noise Profile (1 ps, RMS Input)	17
	Passive Loop Filter Circuit Phase Noise Measurement Test Setup CDCE72010 Device Configuration

All trademarks are the property of their respective owners.

1



25	30.72-MHz LVCMOS Input Phase Noise Profile (5 ps, RMS)	18
26	122.88-MHz LVCMOS Output Phase Noise Profile (5 ps, RMS Input)	18
27	122.88-MHz HS-LVPECL Output Phase Noise Profile (5 ps, RMS Input)	19
28	122.88-MHz HS-LVDS Output Phase Noise Profile (5 ps, RMS Input)	19
29	30.72-MHz LVCMOS Input Phase Noise Profile (10 ps, RMS)	20
30	122.88-MHz LVCMOS Output Phase Noise Profile (10 ps, RMS Input)	20
31	122.88-MHz HS-LVPECL Output Phase Noise Profile (10 ps, RMS Input)	21
32	122.88-MHz HS-LVDS Output Phase Noise Profile (10 ps, RMS Input)	21
33	30.72-MHz LVCMOS Input Phase Noise Profile (25 ps, RMS)	22
34	122.88-MHz LVCMOS Output Phase Noise Profile (25 ps, RMS Input)	22
35	122.88-MHz HS-LVPECL Output Phase Noise Profile (25 ps, RMS Input)	23
36	122.88-MHz HS-LVDS Output Phase Noise Profile (25 ps, RMS Input)	23
37	30.72-MHz LVCMOS Input Phase Noise Profile (50 ps, RMS)	24
38	122.88-MHz LVCMOS Output Phase Noise Profile (50 ps, RMS Input)	24
39	122.88-MHz HS-LVPECL Output Phase Noise Profile (50 ps, RMS Input)	25
40	122.88-MHz HS-LVDS Output Phase Noise Profile (50 ps, RMS Input)	25
41	491.52-MHz LVPECL VCXO (TCO-2111) Phase Noise Profile	27

1 Introduction

The CDCE72010 is a low phase noise/low jitter clock synthesizer and jitter cleaner with programmable outputs and inputs. An external low-pass loop filter in addition to an external VCXO or VCO is required to complete the phase-locked loop (PLL). Proper selection of the VCXO and loop bandwidth is critical to achieve the best performance from the CDCE72010.

This report includes phase noise plots of the most common frequencies used in wireless basestation applications. In addition, the phase noise of the clock source that feeds the CDCE72010 is included for completeness. Phase noise measurements of the 30.72-MHz reference, the 491.52-MHz Epson-Toyocom VCXO (TCO-2111), and output phase noise of the CDCE72010 at various frequencies are included. The root-mean-square (RMS) jitter was calculated from the various phase noise plots over the following ranges with the Agilent E5052A phase noise analyzer:

- 1 kHz to 40 MHz for frequencies greater than 100 MHz
- 1 kHz to 20 MHz for frequencies above 40 MHz
- 1 kHz to 5 MHz for frequencies above 10 MHz

In addition, additive phase noise jitter of the CDCE72010 is included in these integrated bandwidths. However, these data do not include jitter contributions from the external clock reference and the VCXO.



1.1 Definitions

- **Crosstalk** This characteristic is used to measure parasitic coupling between signals, and is the effect of capacitive coupling that causes a logic transition. Capacitive coupling is the transfer of energy between nearby switching integrated circuits. The coupling depends on factors such as the distance between the traces, the signal swing, the operating frequency, and the permissiveness of the silicon dioxide. Coupling can be improved by physically increasing the distance between traces. Power and ground planes also act as shields to minimize crosstalk.
- **Cycle-to-cycle period jitter**—Also known as adjacent cycle jitter; the variation in cycle time of a signal between consecutive cycles over a random sample of successive cycle pairs. Cycle-to-cycle jitter is also a good value to calculate the setup and hold time budgets because it defines the minimum and maximum variations of the timing variation from ideal for the next clock edge.
- **Jitter** Any edge deviation from the ideal occurrence. The causes of jitter include: power-supply noise, thermal and mechanical noise from the input signal and other external sources, reflection, electromagnetic interference (EMI), and other random noise. Suggestions to reduce jitter include: power-supply bypassing (10 μ F to 47 μ F) to prevent voltage droop and ripple because of current surges; filtering each VCC pin (with 0.1- μ F, low effective series resistance [ESR] capacitor); using proper termination to remove reflections; using differential signaling as opposed to single-ended signaling; and minimizing noise coupling by isolating other high-frequency signals from the clock driver.
- **Peak-to-peak period jitter**—The total jitter range from minimum to maximum values of a clock signal. Peak-to-peak jitter increases indefinitely with recording time. Thus, peak-to-peak jitter values are only meaningful if either the recording length or the relative bit error rate is known.
- **Period jitter**—The deviation in cycle time of a signal with respect to an ideal period over a random sample of cycles. Period jitter is important because it includes the maximum and minimum frequencies, and it specifies the shortest clock period. It is important for the setup and hold time budgets. Calculations with period jitter are sufficient for subsystems that use clock and data signals derived from the same clock source. Period jitter can be measured with any oscilloscope.
- **Phase jitter** Phase jitter, or accumulated jitter, is the absolute deviation of a clock edge from its ideal position in timing. While period jitter only accounts for the variation between clock periods, phase jitter accumulates the error of each period and is therefore always larger. The wider the recording time window, the more frequency bandwidth becomes integrated into the total phase jitter. Phase jitter can also be measured by integrating phase noise over the frequency band of interest. Either way, the system designer must specify the minimum and maximum frequency for the integration.

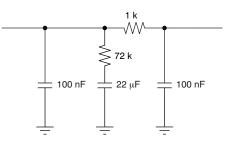
For setup and hold time budget calculations, the peak-to-peak (PP) value of the phase jitter is important. Note that only the added phase noise by the clock driver is of interest to find the worst edge position between the master clock in the system and the subsystem. The absolute phase jitter of the master clock itself adds to all clock signals in the system, thus canceling its effect.

- **Phase noise**—The short-term instability caused by frequency variation (phase) of a signal referenced to the carrier level and a function of the carrier offset (that is, relative noise level within a 1-Hz bandwidth). Integration of PN over a given frequency band yields phase jitter RMS.
- **RMS period jitter**—One standard deviation (1 σ) of the peak-to-peak jitter of a clock signal. RMS jitter is only valid for Gaussian (that is, normal) distribution. RMS jitter is independent of the sampling window, and therefore more suitable for comparing the performance of two or more devices where the sampling time window differs or is unknown.
- **Timing budget**—Defined by dynamic (jitter) and static errors (skew). Depending on the system architecture, only a subset of parameters from the datasheet affect the timing budget. Jitter is a timing distribution of the clock signal that expresses the edge deviation from the ideal occurrence. Jitter is composed of both deterministic and random (Gaussian) content.

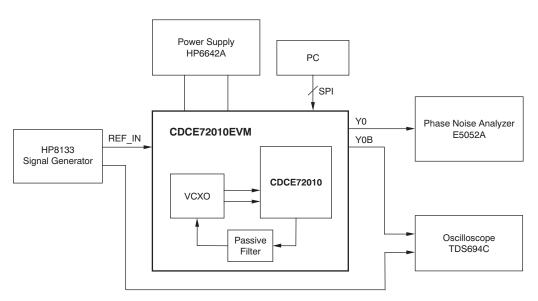


2 Test Equipment and Setup

All measurements discussed in this application report were taken under nominal conditions: a 3.3-V power supply, at room temperature, and in a PLL lock condition except for additive jitter, which was performed for a VCXO held at a control voltage of 1.65 V. The CDCE72010EVM evaluation module is used with a 491.52-MHz Epson-Toyocom (TCO-2111) VCXO. Figure 3 shows the CDCE72010 device configuration. The power supply is provided by the HP6624A; a reference input of 30.72 MHz LVCMOS is provided by an HP8133. Phase noise is measured using the Agilent E5052A. Figure 2 shows the test setup that was used for all phase noise testing. Output dividers were set to /1, /2, /3, /4, and /5 for overall and additive phase noise measurements. An output divider of /4 was used for jitter cleaner tests where the 30.72 MHz LVCMOS input was fed into a NoiseCom noise generator box to increase the noise floor, thus raising the input jitter. The RMS jitter was calculated from the phase noise plots over a 1 kHz to 40 MHz range for frequencies greater than 100 MHz, from 1 kHz to 20 MHz for frequencies above 40 MHz, and from 1 kHz to 5 MHz for frequencies above 10 MHz. A 60-Hz loop filter bandwidth is used; Figure 1 illustrates the filter topology for this loop filter. All HS-LVPECL, LVCMOS, and HS-LVDS outputs were properly terminated and tested for jitter.











Test Equipment and Setup

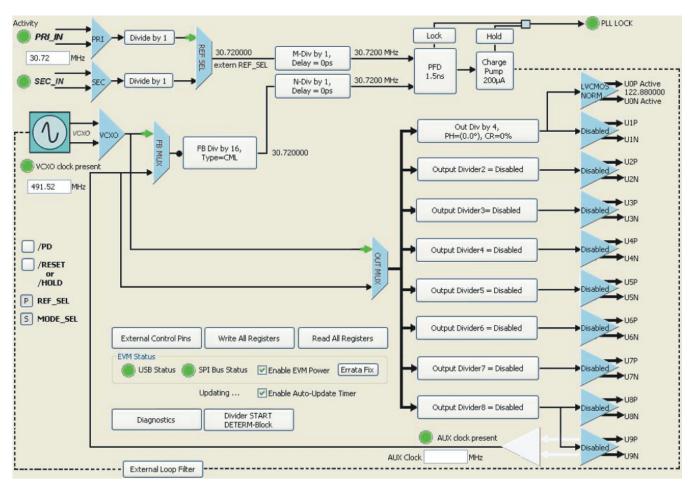


Figure 3. CDCE72010 Device Configuration



3 Total Phase Noise Measurements

Section 3.1 summarizes the total phase noise/jitter measurements made on the CDCE72010 with a 491.52-MHz VCXO and for /1, /2, /3, /4, and /5 output divider configurations for HS-LVDS, HS-LVPECL and LVCMOS output types. Section 3.2 shows the measurement results at different output frequencies and for different output types.

3.1 Performance Summary

Table 1 lists the total jitter of the CDCE72010 with a 491.52-MHz VCXO.

Table 1. CDCE72010 and 491.52-MHz VCXO Total Jitter Summary			
Output Type	Phase Jitter (fs, RMS) 1 kHz to 40 MHz		
HS-LVPECL	154.35		
HS-LVDS	162.44		
LVCMOS	243.67		
HS-LVPECL	199.11		
HS-LVDS	216.09		
LVCMOS	246.74		
HS-LVPECL	215.02		
HS-LVDS	268.57		
LVCMOS	281.05		
HS-LVPECL	234.97		
HS-LVDS	321.86		
LVCMOS	228.91 (1 kHz to 20 MHz)		
HS-LVPECL	192.14 (1 kHz to 20 MHz)		
HS-LVDS	286.08 (1 kHz to 20 MHz)		
	Output TypeHS-LVPECLHS-LVDSLVCMOSHS-LVPECLHS-LVDSLVCMOSHS-LVDSLVCMOSHS-LVPECLHS-LVPECLHS-LVPECLHS-LVPECLHS-LVDSLVCMOSHS-LVDSLVCMOSHS-LVPECLHS-LVPECLHS-LVPECLHS-LVPECLHS-LVPECLLVCMOSHS-LVPECL		

Table 1. CDCE72010 and 491.52-MHz VCXO Total Jitter Summary



3.2 Measurement Results

Figure 4 through Figure 17 show the measured results for output phase noise over a range of frequencies.

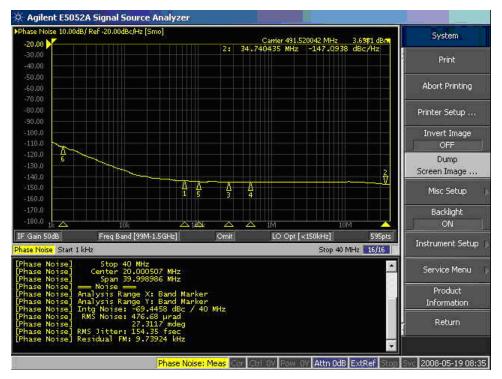


Figure 4. 491.52-MHz HS-LVPECL Output Phase Noise

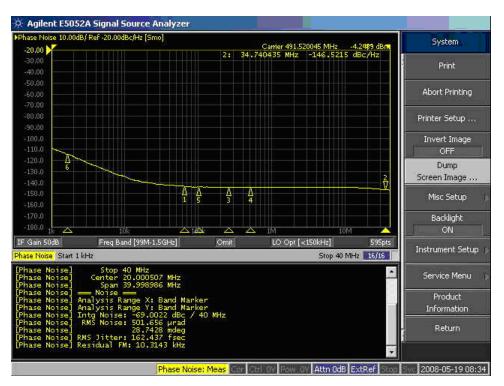


Figure 5. 491.52-MHz HS-LVDS Output Phase Noise



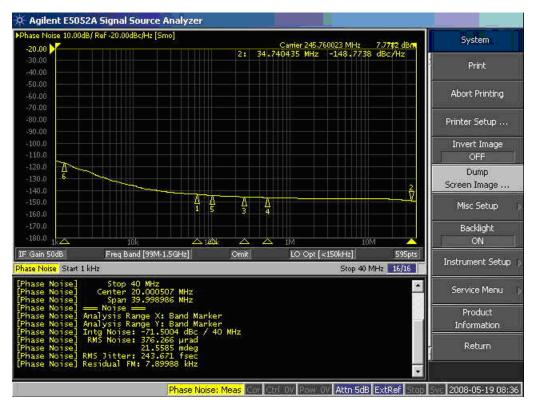


Figure 6. 245.76-MHz LVCMOS Output Phase Noise

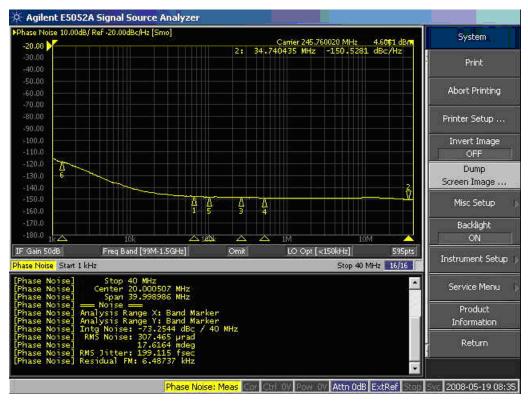
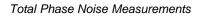


Figure 7. 245.76-MHz HS-LVPECL Output Phase Noise

8





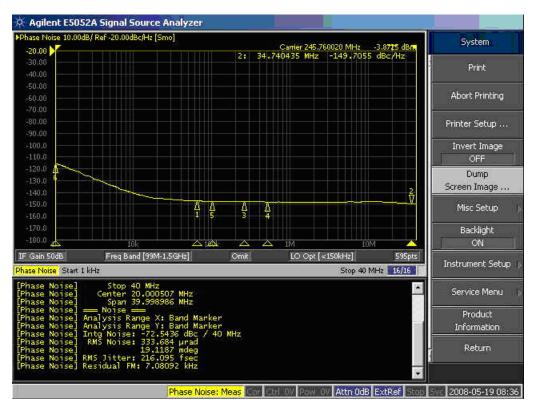


Figure 8. 245.76-MHz HS-LVDS Output Phase Noise

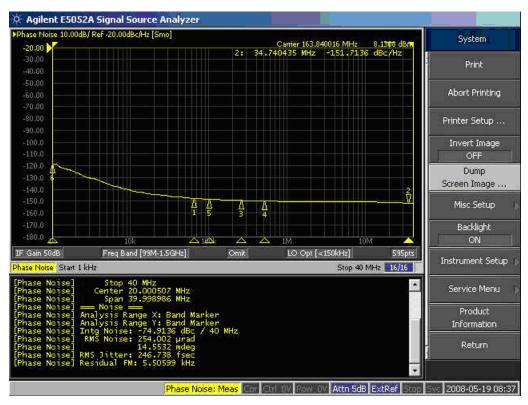


Figure 9. 163.84-MHz LVCMOS Output Phase Noise



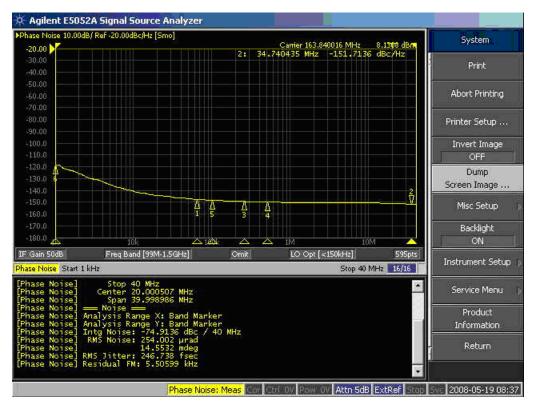


Figure 10. 163.84-MHz HS-LVPECL Output Phase Noise

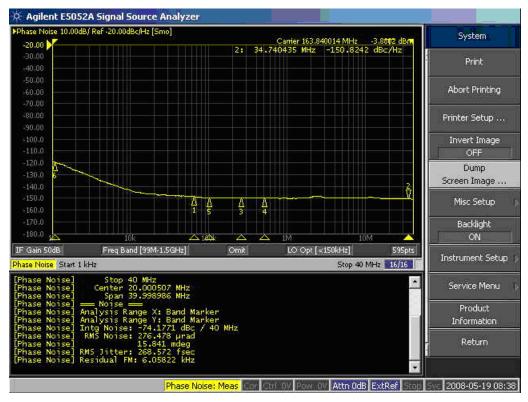
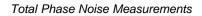


Figure 11. 163.84-MHz HS-LVDS Output Phase Noise





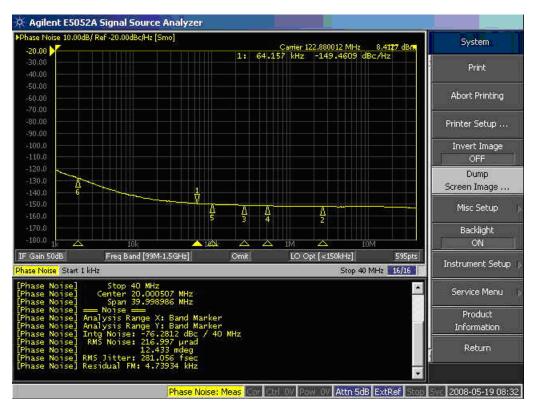


Figure 12. 122.88-MHz LVCMOS Output Phase Noise

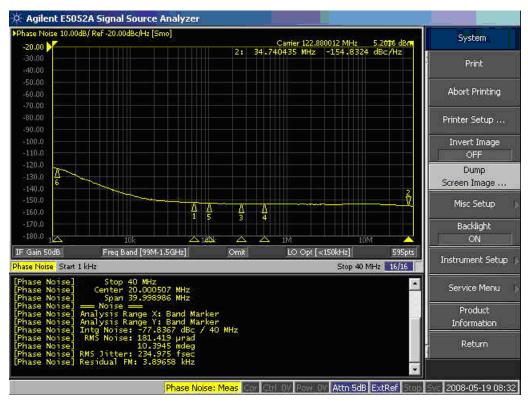


Figure 13. 122.88-MHz HS-LVPECL Output Phase Noise



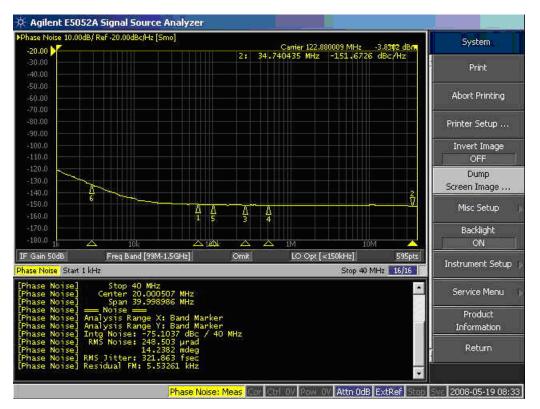


Figure 14. 122.88-MHz HS-LVDS Output Phase Noise

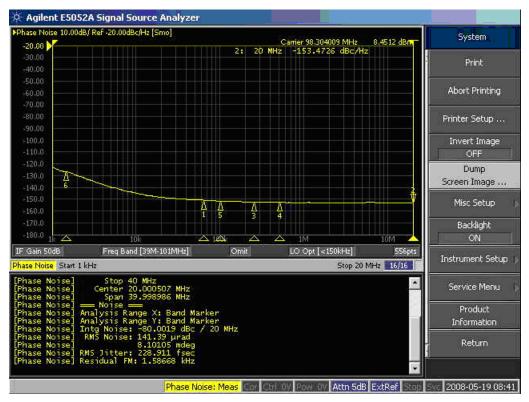


Figure 15. 98.304-MHz LVCMOS Output Phase Noise



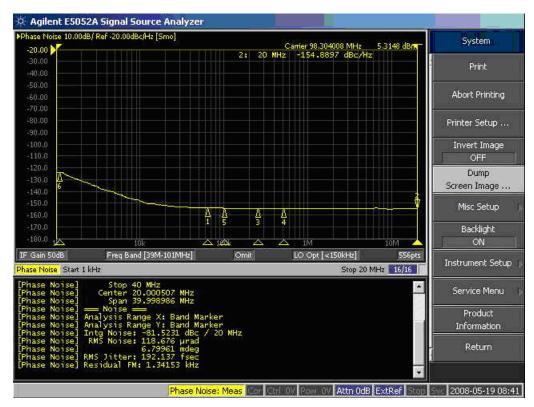


Figure 16. 98.304-MHz HS-LVPECL Output Phase Noise

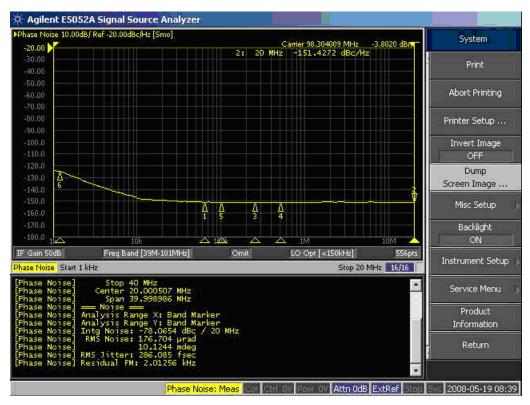


Figure 17. 98.304-MHz HS-LVDS Output Phase Noise



4 Total Phase Noise Measurements

Section 4.1 shows a summary of the CDCE72010 jitter cleaning ability with a 491.52-MHz VCXO and output divider of /4, for different output types. The input is a 30.72-MHz LVCMOS with varying phase jitter characteristics. Section 4.2 shows the measurement results.

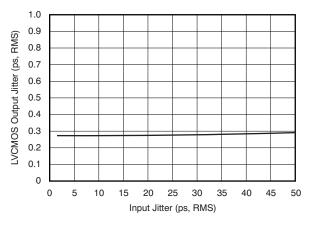
4.1 Performance Measurement Summary

Table 2 describes the CDCE72010 jitter cleaning ability for a range of output types.

In-Phase Jitter (ps, RMS) 1 kHz to 5 MHz	Output Type	Out Phase Jitter (fs, RMS) 1 kHz to 40 MHz
1	LVCMOS	275.63
1	HS-LVPECL	225.98
1	HS-LVDS	314.26
5	LVCMOS	274.84
5	HS-LVPECL	227.99
5	HS-LVDS	314.08
10	LVCMOS	275.82
10	HS-LVPECL	226.39
10	HS-LVDS	315.18
25	LVCMOS	279.79
25	HS-LVPECL	233.76
25	HS-LVDS	318.49
50	LVCMOS	290.41
50	HS-LVPECL	238.39
50	HS-LVDS	319.32

Table 2. CDCE72010 Jitter Cleaning Ability Summary





LVCMOS OUTPUT JITTER vs INPUT JITTER



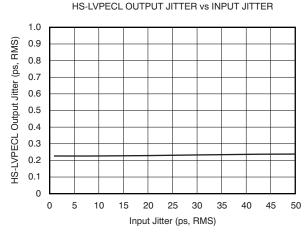


Figure 19. 122.88-MHz HS-LVPECL Output Jitter Cleaning Ability

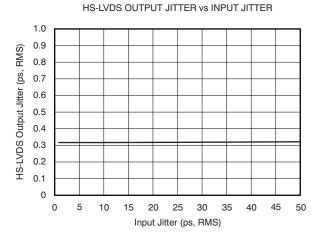


Figure 20. 122.88-MHz HS-LVDS Output Jitter Cleaning Ability

Total Phase Noise Measurements



www.ti.com

4.2 Measurement Results

Figure 21 through Figure 40 show the measured results for output phase noise over a range of frequencies.

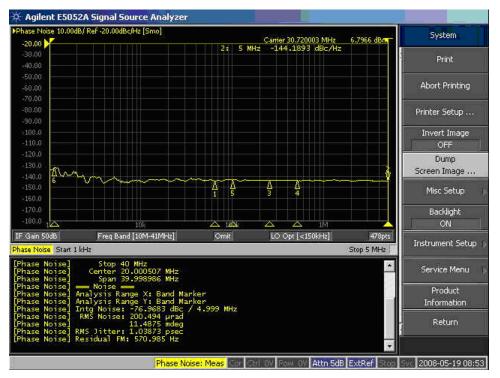


Figure 21. 30.72-MHz LVCMOS Input Phase Noise Profile (1 ps, RMS)

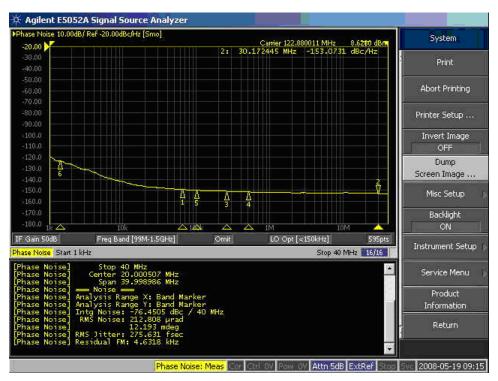


Figure 22. 122.88-MHz LVCMOS Output Phase Noise Profile (1 ps, RMS Input)



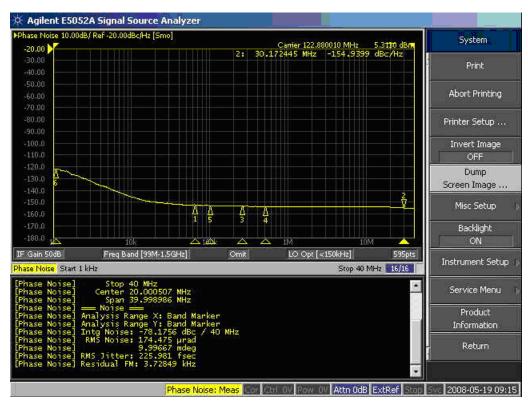


Figure 23. 122.88-MHz HS-LVPECL Output Phase Noise Profile (1 ps, RMS Input)

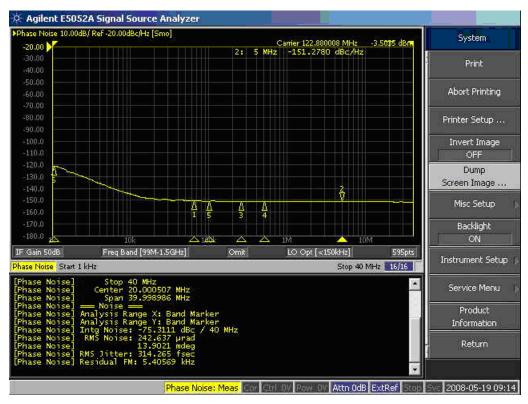


Figure 24. 122.88-MHz HS-LVDS Output Phase Noise Profile (1 ps, RMS Input)



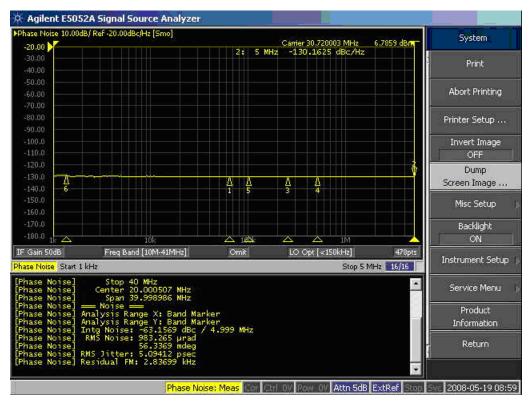


Figure 25. 30.72-MHz LVCMOS Input Phase Noise Profile (5 ps, RMS)

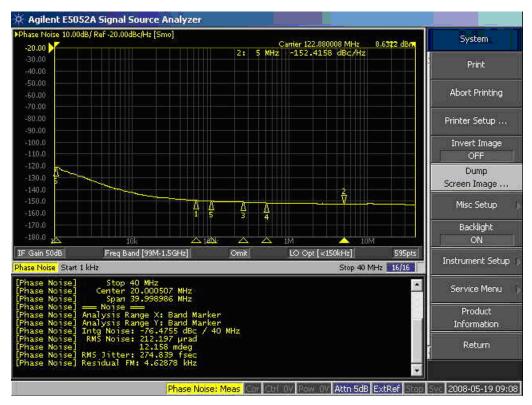


Figure 26. 122.88-MHz LVCMOS Output Phase Noise Profile (5 ps, RMS Input)



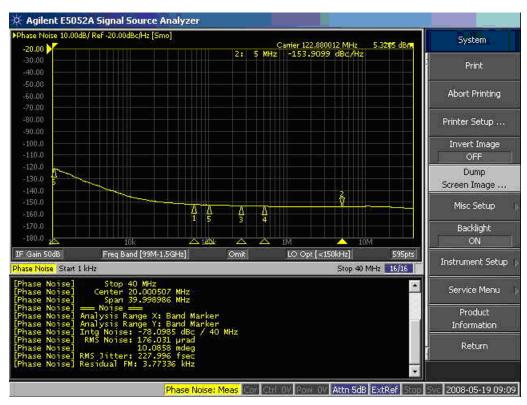


Figure 27. 122.88-MHz HS-LVPECL Output Phase Noise Profile (5 ps, RMS Input)

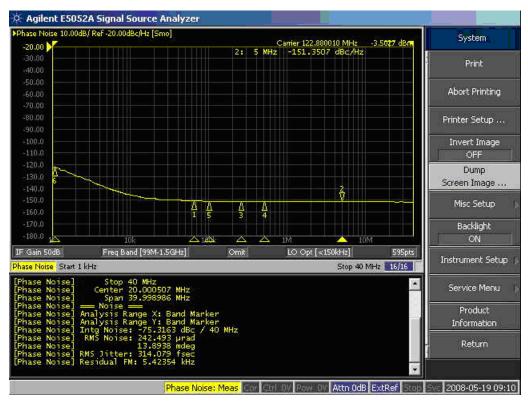


Figure 28. 122.88-MHz HS-LVDS Output Phase Noise Profile (5 ps, RMS Input)



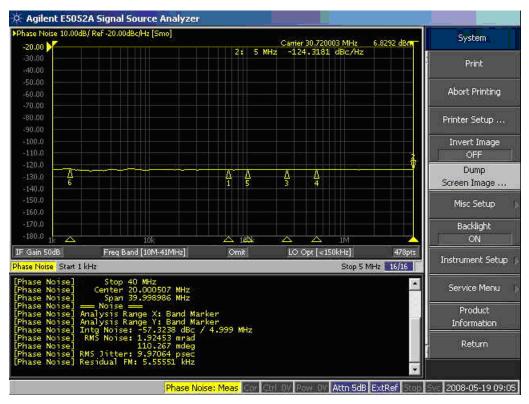


Figure 29. 30.72-MHz LVCMOS Input Phase Noise Profile (10 ps, RMS)

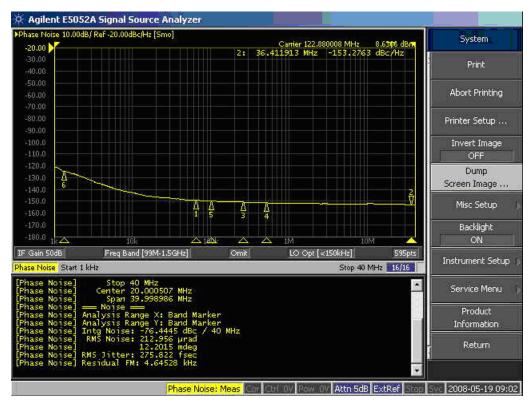


Figure 30. 122.88-MHz LVCMOS Output Phase Noise Profile (10 ps, RMS Input)



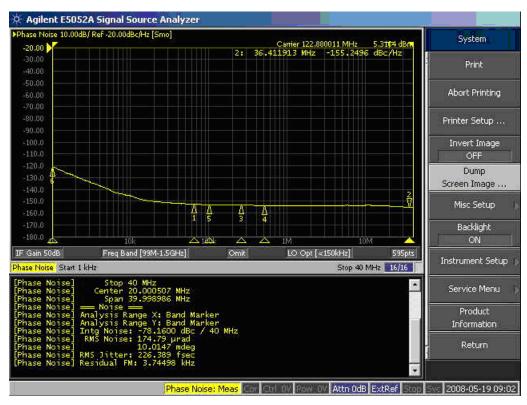


Figure 31. 122.88-MHz HS-LVPECL Output Phase Noise Profile (10 ps, RMS Input)

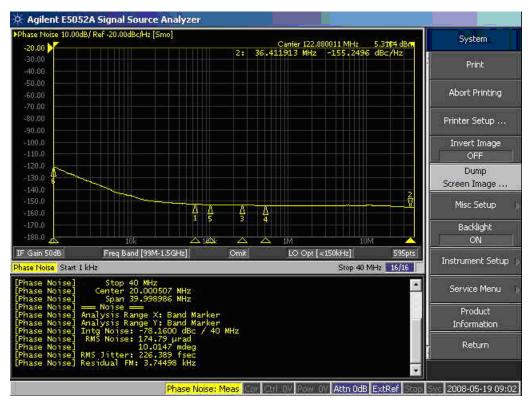


Figure 32. 122.88-MHz HS-LVDS Output Phase Noise Profile (10 ps, RMS Input)



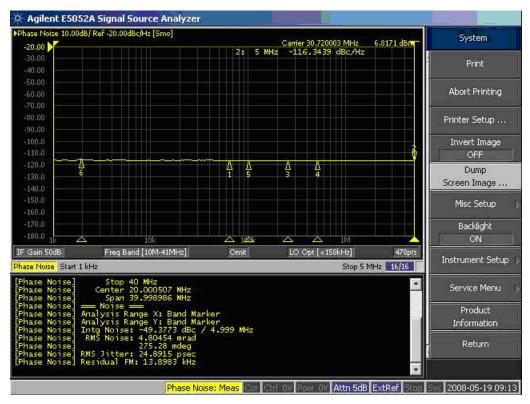


Figure 33. 30.72-MHz LVCMOS Input Phase Noise Profile (25 ps, RMS)

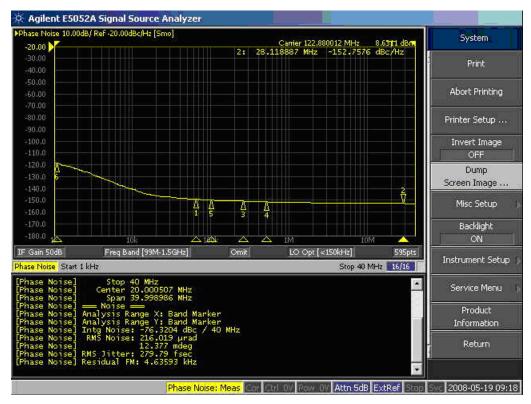


Figure 34. 122.88-MHz LVCMOS Output Phase Noise Profile (25 ps, RMS Input)



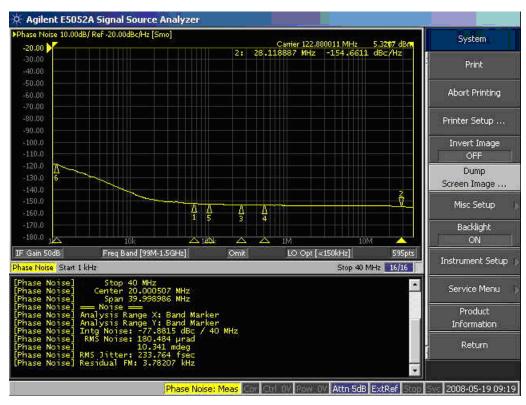


Figure 35. 122.88-MHz HS-LVPECL Output Phase Noise Profile (25 ps, RMS Input)

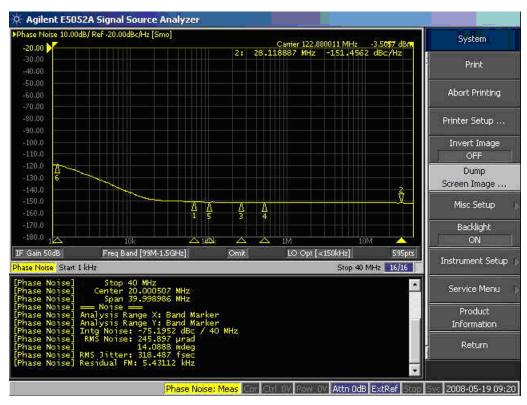


Figure 36. 122.88-MHz HS-LVDS Output Phase Noise Profile (25 ps, RMS Input)



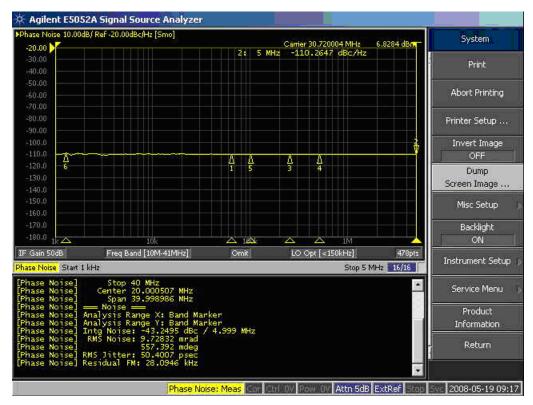


Figure 37. 30.72-MHz LVCMOS Input Phase Noise Profile (50 ps, RMS)

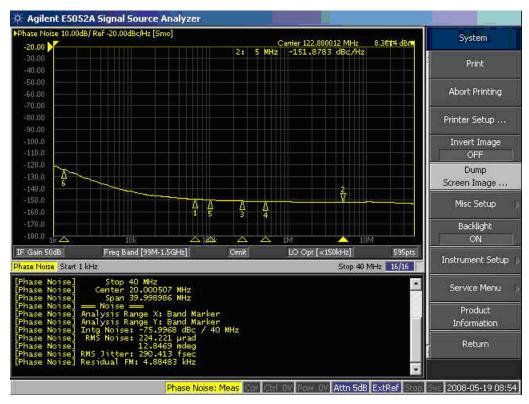


Figure 38. 122.88-MHz LVCMOS Output Phase Noise Profile (50 ps, RMS Input)



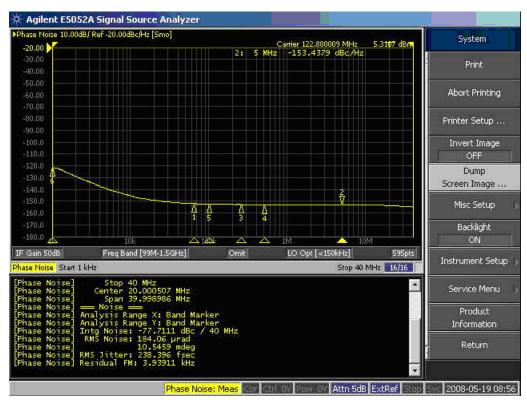


Figure 39. 122.88-MHz HS-LVPECL Output Phase Noise Profile (50 ps, RMS Input)

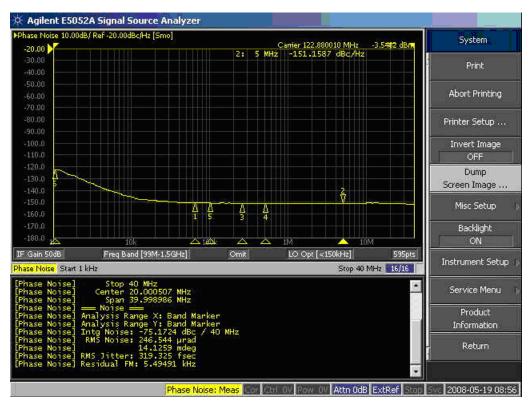


Figure 40. 122.88-MHz HS-LVDS Output Phase Noise Profile (50 ps, RMS Input)



5 Additive Phase Noise Measurements

Section 5.1 summarizes the additive phase noise/jitter measurements made on the CDCE72010 in an open loop configuration with a 491.52-MHz VCXO for /1, /2, /3, /4, and /5 output divider configurations for HS-LVDS, HS-LVPECL, and LVCMOS output types. Section 5.2 shows the 491.52-MHz LVPECL VCXO (TCO-2111) phase noise/jitter. The additive jitter can then be calculated in a specified integration band as shown in Equation 1.

 $t_{J,Additive} = \sqrt{t_{J,Output}^2 - t_{J,VCXO}^2}$

(1)

5.1 Performance Summary

Table 3 lists the additive jitter capability of the CDCE72010.

Frequency (MHz)	Output Type	Phase Jitter (fs, RMS) 1 kHz to 40 MHz
491.52	HS-LVPECL	57
491.52	HS-LVDS	76
245.76	LVCMOS	196
245.76	HS-LVPECL	138
245.76	HS-LVDS	161
163.84	LVCMOS	200
163.84	HS-LVPECL	160
163.84	HS-LVDS	226
122.88	LVCMOS	242
122.88	HS-LVPECL	186
122.88	HS-LVDS	287
98.304	LVCMOS	182 (1 kHz to 20 MHz)
98.304	HS-LVPECL	136 (1 kHz to 20 MHz)
98.304	HS-LVDS	252 (1 kHz to 20 MHz)



Additive Phase Noise Measurements

5.2 Measurement Result

Figure 41 shows the phase noise profile of the 491.52-MHz LVPECL VCXO.

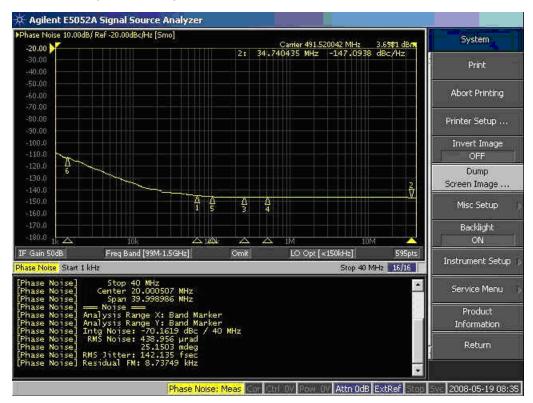


Figure 41. 491.52-MHz LVPECL VCXO (TCO-2111) Phase Noise Profile

IMPORTANT NOTICE

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, modifications, enhancements, improvements, and other changes to its products and services at any time and to discontinue any product or service without notice. Customers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All products are sold subject to TI's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its hardware products to the specifications applicable at the time of sale in accordance with TI's standard warranty. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by government requirements, testing of all parameters of each product is not necessarily performed.

TI assumes no liability for applications assistance or customer product design. Customers are responsible for their products and applications using TI components. To minimize the risks associated with customer products and applications, customers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any TI patent right, copyright, mask work right, or other TI intellectual property right relating to any combination, machine, or process in which TI products or services are used. Information published by TI regarding third-party products or services does not constitute a license from TI to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of TI information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. Reproduction of this information with alteration is an unfair and deceptive business practice. TI is not responsible or liable for such altered documentation. Information of third parties may be subject to additional restrictions.

Resale of TI products or services with statements different from or beyond the parameters stated by TI for that product or service voids all express and any implied warranties for the associated TI product or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

TI products are not authorized for use in safety-critical applications (such as life support) where a failure of the TI product would reasonably be expected to cause severe personal injury or death, unless officers of the parties have executed an agreement specifically governing such use. Buyers represent that they have all necessary expertise in the safety and regulatory ramifications of their applications, and acknowledge and agree that they are solely responsible for all legal, regulatory and safety-related requirements concerning their products and any use of TI products in such safety-critical applications, notwithstanding any applications-related information or support that may be provided by TI. Further, Buyers must fully indemnify TI and its representatives against any damages arising out of the use of TI products in such safety-critical applications.

TI products are neither designed nor intended for use in military/aerospace applications or environments unless the TI products are specifically designated by TI as military-grade or "enhanced plastic." Only products designated by TI as military-grade meet military specifications. Buyers acknowledge and agree that any such use of TI products which TI has not designated as military-grade is solely at the Buyer's risk, and that they are solely responsible for compliance with all legal and regulatory requirements in connection with such use.

TI products are neither designed nor intended for use in automotive applications or environments unless the specific TI products are designated by TI as compliant with ISO/TS 16949 requirements. Buyers acknowledge and agree that, if they use any non-designated products in automotive applications, TI will not be responsible for any failure to meet such requirements.

Following are URLs where you can obtain information on other Texas Instruments products and application solutions:

Products		Applications	
Amplifiers	amplifier.ti.com	Audio	www.ti.com/audio
Data Converters	dataconverter.ti.com	Automotive	www.ti.com/automotive
DSP	dsp.ti.com	Broadband	www.ti.com/broadband
Clocks and Timers	www.ti.com/clocks	Digital Control	www.ti.com/digitalcontrol
Interface	interface.ti.com	Medical	www.ti.com/medical
Logic	logic.ti.com	Military	www.ti.com/military
Power Mgmt	power.ti.com	Optical Networking	www.ti.com/opticalnetwork
Microcontrollers	microcontroller.ti.com	Security	www.ti.com/security
RFID	www.ti-rfid.com	Telephony	www.ti.com/telephony
RF/IF and ZigBee® Solutions	www.ti.com/lprf	Video & Imaging	www.ti.com/video
		Wireless	www.ti.com/wireless

Mailing Address: Texas Instruments, Post Office Box 655303, Dallas, Texas 75265 Copyright © 2008, Texas Instruments Incorporated