

INAx133 High-Speed, Precision Difference Amplifiers

1 Features

- Single, dual versions
- Low offset voltage: $\pm 450\mu\text{V}$ (maximum)
- Low offset voltage drift: $\pm 5\mu\text{V}/^\circ\text{C}$ (maximum)
- Low gain error: 0.05% (maximum)
- Wide bandwidth: 1.5MHz
- High slew rate: 5V/ μs
- Fast settling time: 5.5 μs to 0.01%
- Low quiescent current: 950 μA
- Wide supply range: $\pm 2.25\text{V}$ to $\pm 18\text{V}$

2 Applications

- [Battery cell formation & test equipment](#)
- [Sensor tag & data logger](#)
- [Servo drive position feedback](#)
- [Level transmitter](#)
- [String inverter](#)

3 Description

The INA133 and INA2133 are high slew rate, unity gain difference amplifiers made up of a precision operational amplifier with a precision resistor network. The on-chip resistors are laser trimmed for accurate gain and high common-mode rejection. Excellent tracking of resistors (TCR) maintains gain accuracy and common-mode rejection over temperature. The INAx133 operates over a wide supply range, $\pm 2.25\text{V}$ to $\pm 18\text{V}$ ($+4.5\text{V}$ to $+36\text{V}$ single supply), and that input common-mode voltage range extends beyond the positive and negative supply rails.

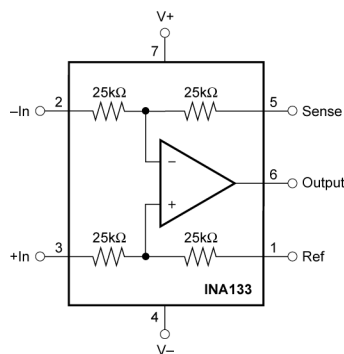
The differential amplifier is the foundation of many commonly used circuits. The INAx133 provides this precision circuit function for cost-optimized designs.

The single INA133 version is available in the SOIC-8 surface mount package. The dual INA2133 version is available in SOIC-14 package. Both are specified for operation over the industrial temperature range -40°C to $+85^\circ\text{C}$.

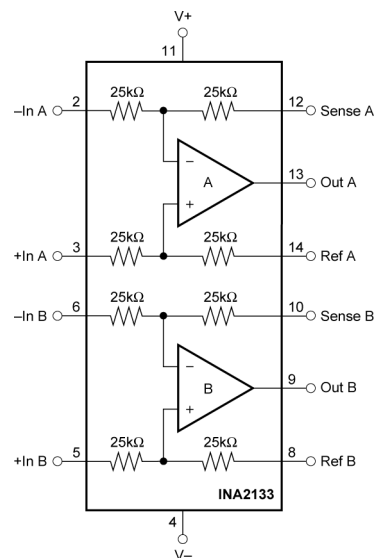
Package Information

PART NUMBER	PACKAGE ⁽¹⁾	PACKAGE SIZE ⁽²⁾
INA133	D (SOIC, 8)	4.90mm × 6.00mm
INA2133	D (SOIC, 14)	8.65mm × 6.00mm

- (1) For all available packages, see the orderable addendum in [Section 10](#).
- (2) The package size (length × width) is a nominal value and includes pins, where applicable.



INA133 Simplified Internal Schematic



INA2133 Simplified Internal Schematic



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4 Pin Configuration and Functions

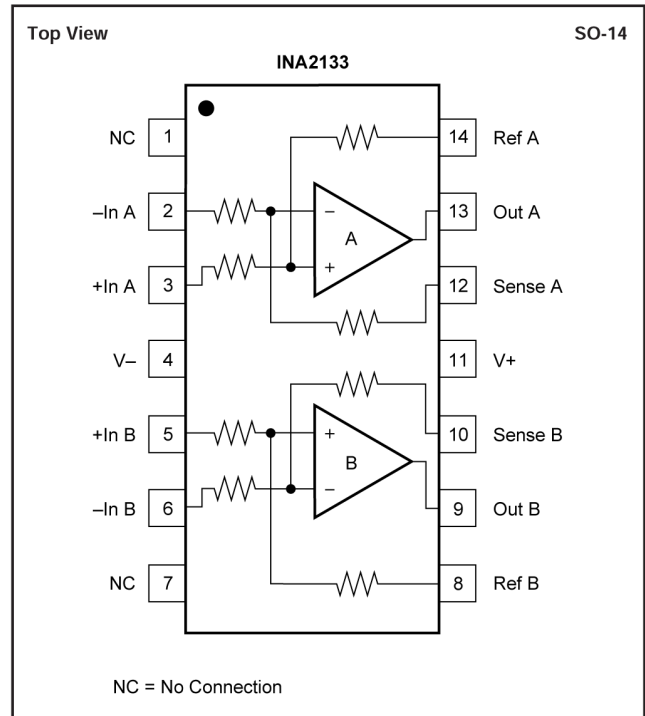
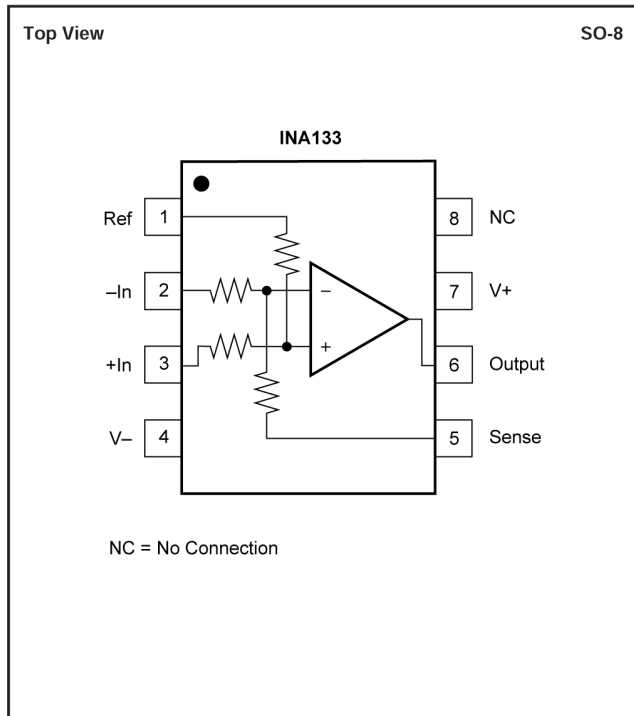


Table 4-1. Pin Functions: INA133

NAME	NO.	TYPE	DESCRIPTION
+In	3	Input	Positive (noninverting) input 25kΩ resistor to noninverting terminal of op amp
-In	2	Input	Negative (inverting) input 25kΩ resistor to inverting terminal of op amp
Output	6	Output	Output
Ref	1	Input	Reference input 25kΩ resistor to noninverting terminal of op amp
V+	7	–	Positive (highest) power supply
V-	4	–	Negative (lowest) power supply
Sense	5	Input	Sense input 25kΩ resistor to inverting terminal of op amp
NC	8	–	No internal connection (can be left floating)

Table 4-2. Pin Functions: INA2133

NAME	NO.	TYPE	DESCRIPTION
+In A	3	Input	Positive (noninverting) input, Channel A 25kΩ resistor to noninverting terminal of op amp
-In A	2	Input	Negative (inverting) input, Channel A 25kΩ resistor to inverting terminal of op amp
+In B	5	Input	Positive (noninverting) input, Channel B 25kΩ resistor to non-inverting terminal of op amp
-In B	6	Input	Negative (inverting) input, Channel B 25kΩ resistor to inverting terminal of op amp
Out A	13	Output	Output, Channel A
Out B	9	Output	Output, Channel B
Ref A	14	Input	Reference input, Channel A 25kΩ resistor to noninverting terminal of op amp
Ref B	8	Input	Reference input, Channel B 25kΩ resistor to noninverting terminal of op amp
Sense A	12	Input	Sense input, Channel A 25kΩ resistor to inverting terminal of op amp
Sense B	10	Input	Sense input, Channel B 25kΩ resistor to inverting terminal of op amp
V+	11	–	Positive (highest) power supply

Table 4-2. Pin Functions: INA2133 (continued)

NAME	NO.	TYPE	DESCRIPTION
V-	4	-	Negative (lowest) power supply

5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
Supply voltage, $V_S = (V+) - (V-)$		0	36	V
Signal input pins	Input pins	0	$2 \times V_S$	V
	Sense, and REF pins	0	V_S	
Output short-circuit ⁽²⁾		Continuous		
Temperature	Operating, T_A	-55	125	°C
	Junction, T_J		150	
	Storage, T_{stg}	-55	125	

- (1) Operation outside the *Absolute Maximum Ratings* may cause permanent device damage. *Absolute Maximum Ratings* do not imply functional operation of the device at these or any other conditions beyond those listed under *Recommended Operating Conditions*. If used outside the *Recommended Operating Conditions* but within the *Absolute Maximum Ratings*, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.
- (2) Short-circuit to $V_S/2$, one channel per package.

5.2 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
Supply voltage	Single supply	4.5	30	36	V
	Dual supply	± 2.25	± 15	± 18	
Specified temperature		-40		85	°C

5.3 Thermal Information

THERMAL METRIC ⁽¹⁾		INA133	INA2133	UNIT
		8 PINS	14 PINS	
		D (SOIC)	D (SOIC)	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	108.9	71.4	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	45.9	33.1	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	56.6	31.2	°C/W
Ψ_{JT}	Junction-to-top characterization parameter	4.8	3.4	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	55.7	30.9	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	N/A	N/A	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

5.4 Electrical Characteristics

at $T_A = 25^\circ\text{C}$, $V_S = \pm 15\text{V}$, $R_L = 10\text{k}\Omega$ connected to ground, $V_{REF} = 0\text{V}$, and $G = 1$, applies to all packages and part variants (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
INPUT							
V_{OS}	Offset voltage	RTO ^{(1) (2)} , $V_{CM} = 0\text{V}$	U variant		± 150	± 450	μV
			UA variant		± 150	± 900	μV
		RTO ^{(1) (2)} , $V_{CM} = 0\text{V}$, $V_S = \pm 5\text{V}$	U variant		± 300	± 750	μV
			UA variant		± 300	± 1500	μV
	Offset voltage drift	RTO ^{(1) (2)} , $T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$			± 2	± 5	$\mu\text{V}/^\circ\text{C}$
			$V_S = \pm 5\text{V}$		± 2		$\mu\text{V}/^\circ\text{C}$
PSRR	Power-supply rejection ratio	RTO ^{(1) (2)} , $V_S = \pm 2.25\text{V}$ to $\pm 18\text{V}$	U variant		± 10	± 30	$\mu\text{V}/\text{V}$
			UA variant	850	900	950	$\mu\text{V}/\text{V}$
	Long-term stability	RTO ^{(1) (2)}		0.3			$\mu\text{V}/\sqrt{\text{mO}}$
Z_{IN-DM}	Differential impedance ⁽³⁾				50		$\text{k}\Omega$
Z_{IN-CM}	Common-mode impedance ⁽³⁾	$V_{CM} = 0\text{V}$			25		$\text{k}\Omega$
V_{CM}	Common-mode voltage range ⁽⁴⁾	$V_O = 0\text{V}$, $V_S = \pm 15\text{V}$ and $V_S = \pm 5\text{V}$	Positive	$2 \times (V+) - 3$	$2 \times (V+) - 2$		V
			Negative	$2 \times (V-) + 3$	$2 \times (V-) + 2$		V
CMRR	Common-mode rejection ratio	$V_S = \pm 15\text{V}$ and $V_S = \pm 5\text{V}$, $V_{CM} = 2 \times (V-) + 3$ to $2 \times (V+) - 3$, $R_S = 0\Omega$	U variant	80	90		dB
			UA variant	74	90		dB
NOISE VOLTAGE							
e_N	Voltage noise	RTO ^{(2) (5)}	$f = 0.01\text{Hz}$ to 10Hz		2		μV_{PP}
			$f = 10\text{Hz}$		80		$\text{nV}/\sqrt{\text{Hz}}$
			$f = 100\text{Hz}$		60		
			$f = 1\text{kHz}$		57		
GAIN							
G	Initial gain	$V_S = \pm 15\text{V}$ and $V_S = \pm 5\text{V}$			1		V/V
GE	Gain error	$V_S = \pm 15\text{V}$ and $V_S = \pm 5\text{V}$, $(V-) + 1\text{V} \leq V_O \leq (V+) - 1.5\text{V}$	U variant		± 0.02	± 0.05	%
			UA variant		± 0.02	± 0.1	
	Gain drift	$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$			± 1	± 10	$\text{ppm}/^\circ\text{C}$
	Gain nonlinearity	$V_S = \pm 15\text{V}$ and $V_S = \pm 5\text{V}$, $(V-) + 1\text{V} \leq V_O \leq (V+) - 1.5\text{V}$	U variant		± 0.0001	± 0.001	% of FSR
			UA variant		± 0.0001	± 0.002	% of FSR
OUTPUT							
V_O	Output voltage, gain error < 0.1%	Positive, $V_S = \pm 15\text{V}$ and $V_S = \pm 5\text{V}$		$(V+) - 1.5$	$(V+) - 1.3$		V
			$R_L = 100\text{k}\Omega$		$(V+) - 0.8$		
		Negative, $V_S = \pm 15\text{V}$ and $V_S = \pm 5\text{V}$		$(V-) + 1$	$(V+) - 0.8$		
			$R_L = 100\text{k}\Omega$		$(V-) + 0.3$		
	Capacitive load (stable operation)				1000		pF
I_{SC}	Short-circuit current	Continuous to $V_S / 2$			$+32/ -25$		mA

at $T_A = 25^\circ\text{C}$, $V_S = \pm 15\text{V}$, $R_L = 10\text{k}\Omega$ connected to ground, $V_{REF} = 0\text{V}$, and $G = 1$, applies to all packages and part variants (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
FREQUENCY RESPONSE						
BW	Bandwidth, -3dB			1.5		MHz
SR	Slew rate			5		V/ μs
t_s	Settling time	0.1%, $V_{STEP} = 10\text{V}$, $C_L = 100\text{pF}$		4		μs
		0.01%, $V_{STEP} = 10\text{V}$, $C_L = 1000\text{pF}$		5.5		μs
	Overload recovery time	50% Overdrive		4		μs
POWER SUPPLY						
I_Q	Quiescent current (per amplifier)	$I_O = 0\text{V}$		± 0.95	± 1.2	mA
		$I_O = 0\text{V}$, $V_S = \pm 5\text{V}$		± 0.92	± 1.2	mA

- (1) Referred to output in unity-gain difference configuration.
- (2) Includes effects of input bias and offset currents of the amplifier.
- (3) 25k Ω resistors are ratio matched but have $\pm 20\%$ absolute value.
- (4) Maximum input voltage without protection is 10V more than either $\pm 15\text{V}$ supply ($\pm 25\text{V}$). Limit I_{IN} to 1mA.
- (5) Includes effects of input current noise of the amplifier and thermal noise contribution of resistor network.

5.5 Typical Characteristics

At $T_A = +25^\circ\text{C}$, $V_S = \pm 15\text{V}$, $R_L = 10\text{k}\Omega$ connected to ground, and $V_{REF} = 0\text{V}$, unless otherwise noted.

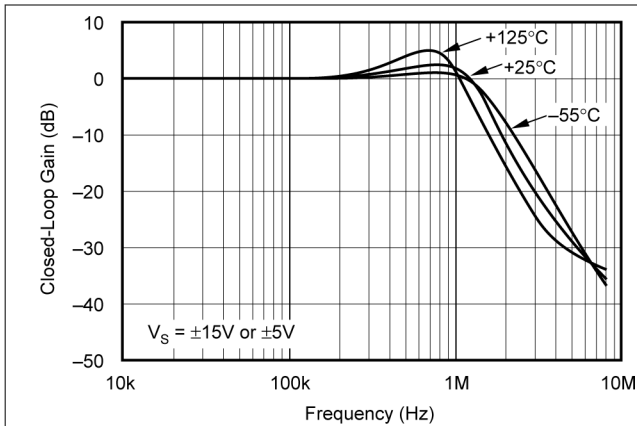


Figure 5-1. Gain vs Frequency

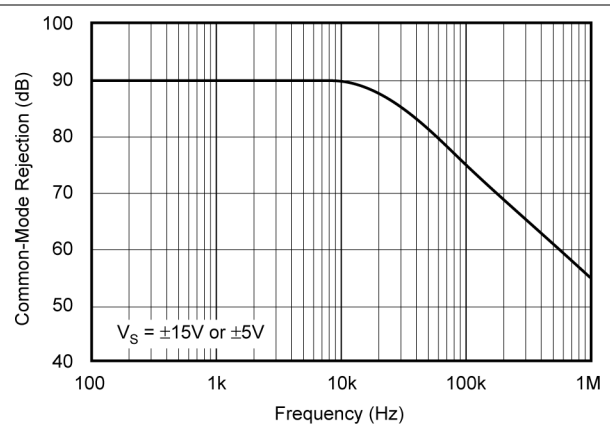


Figure 5-2. Common-Mode Rejection vs Frequency

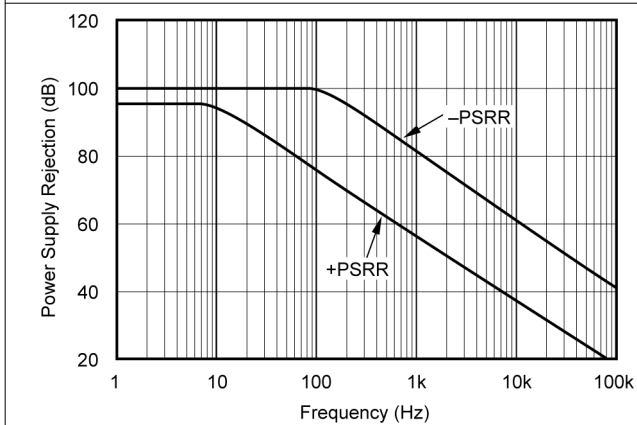


Figure 5-3. Power Supply Rejection vs Frequency

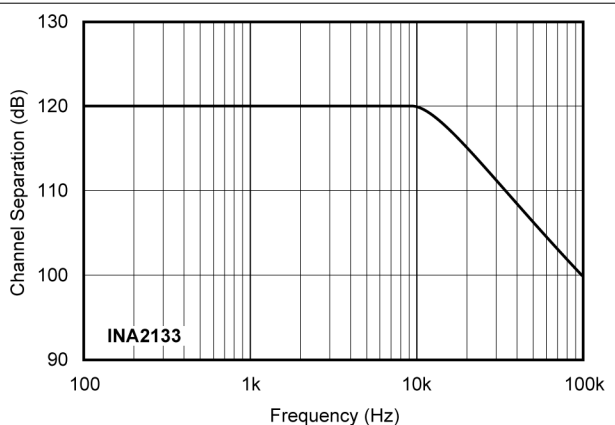


Figure 5-4. Channel Separation vs Frequency

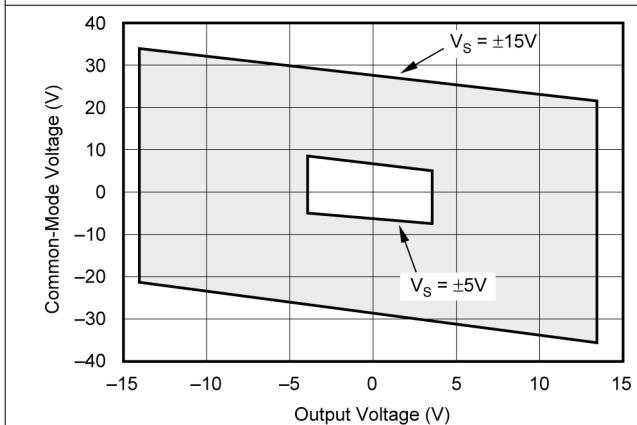


Figure 5-5. Input Common-Mode Voltage vs Output Voltage

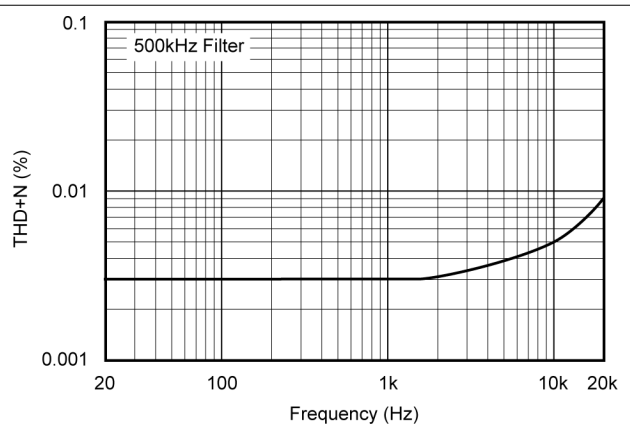


Figure 5-6. Total Harmonic Distortion + Noise vs Frequency

5.5 Typical Characteristics (continued)

At $T_A = +25^\circ\text{C}$, $V_S = \pm 15\text{V}$, $R_L = 10\text{k}\Omega$ connected to ground, and $V_{REF} = 0\text{V}$, unless otherwise noted.

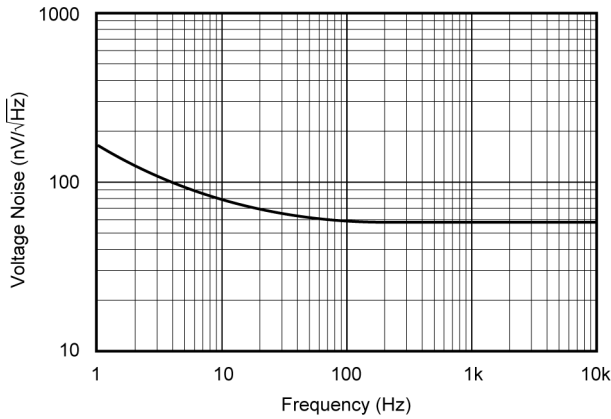


Figure 5-7. Voltage Noise Density vs Frequency

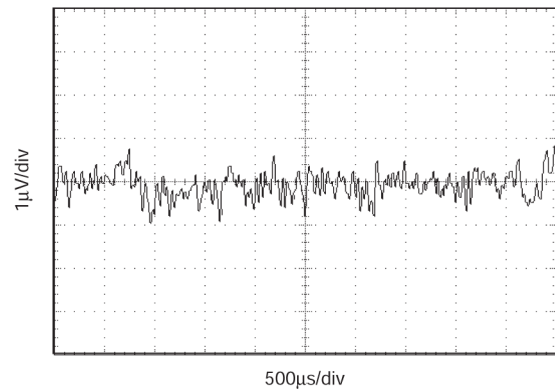


Figure 5-8. 0.1Hz to 10Hz Peak-to-Peak Voltage Noise

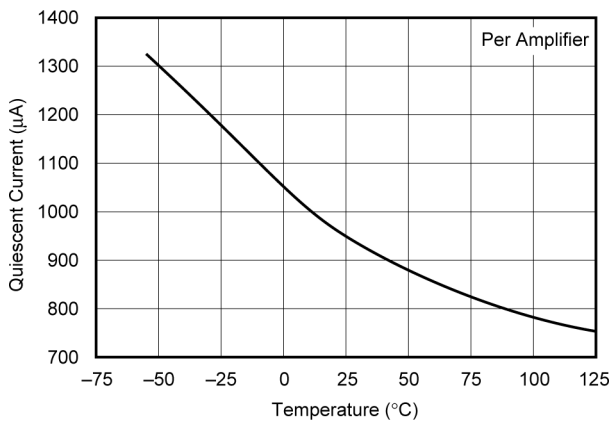


Figure 5-9. Quiescent Current vs Temperature

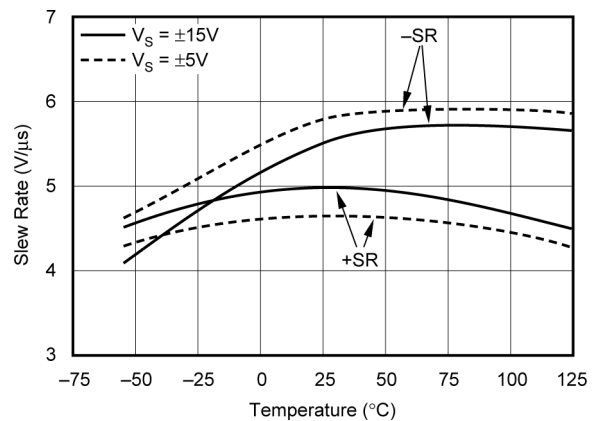


Figure 5-10. Slew Rate vs Temperature

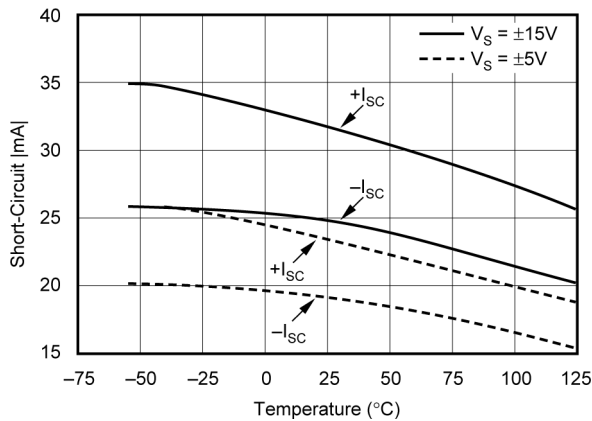


Figure 5-11. Short-Circuit Current vs Temperature

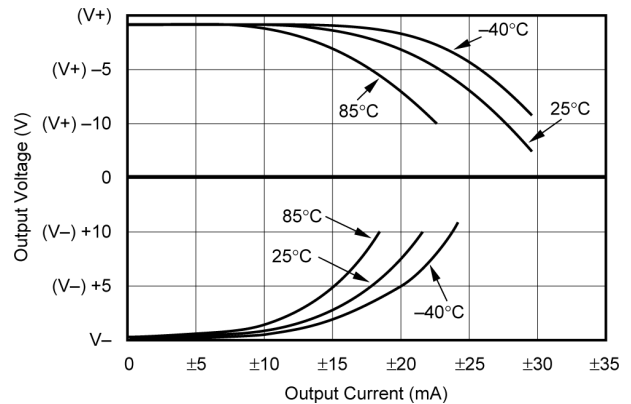


Figure 5-12. Output Voltage Swing vs Output Current

5.5 Typical Characteristics (continued)

At $T_A = +25^\circ\text{C}$, $V_S = \pm 15\text{V}$, $R_L = 10\text{k}\Omega$ connected to ground, and $V_{REF} = 0\text{V}$, unless otherwise noted.

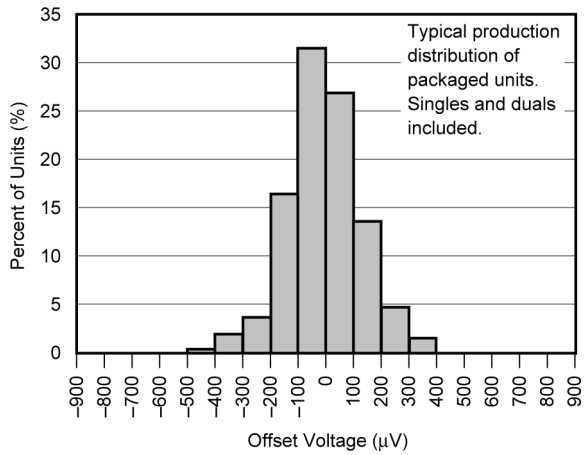


Figure 5-13. Offset Voltage Production Distribution $V_S = \pm 15\text{V}$

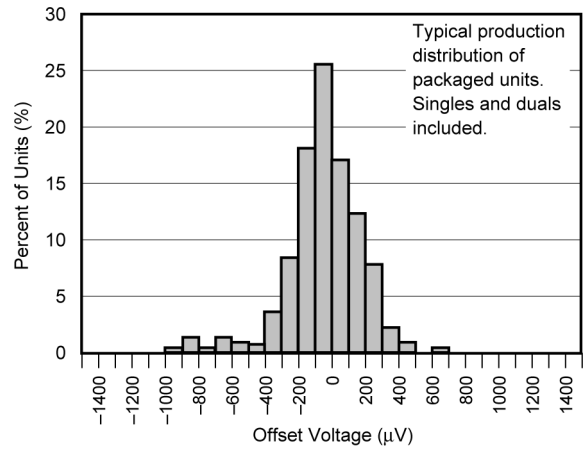


Figure 5-14. Offset Voltage Production Distribution $V_S = \pm 5\text{V}$

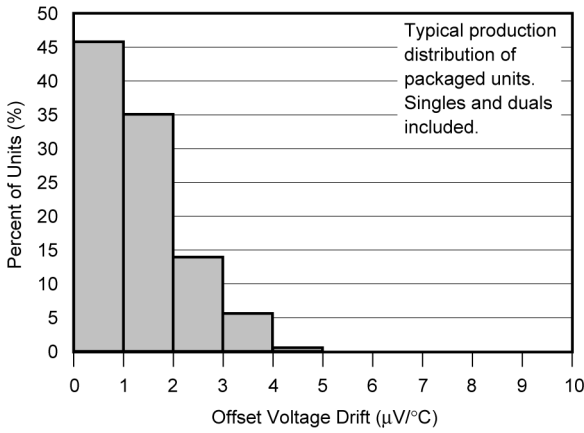


Figure 5-15. Offset Voltage Drift Production Distribution $V_S = \pm 15\text{V}$

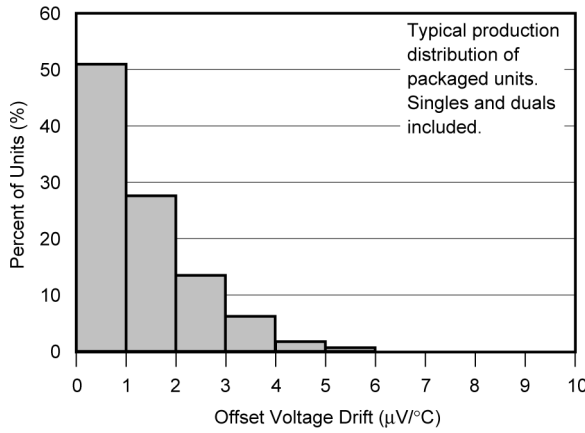


Figure 5-16. Offset Voltage Drift Production Distribution $V_S = \pm 5\text{V}$

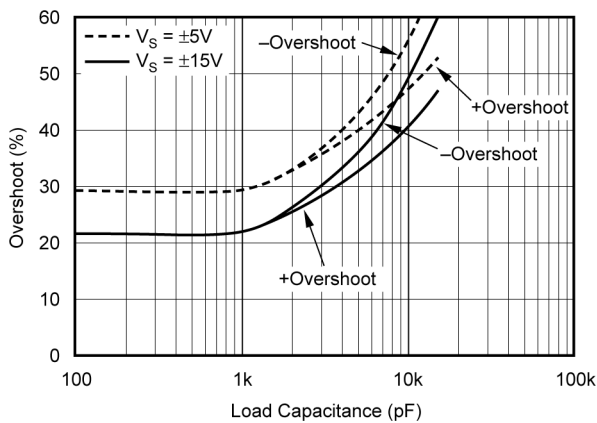


Figure 5-17. Small-Signal Overshoot vs Load Capacitance

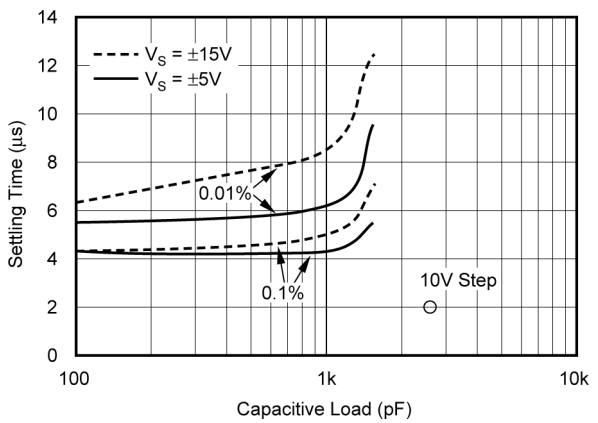


Figure 5-18. Settling Time vs Load Capacitance

5.5 Typical Characteristics (continued)

At $T_A = +25^\circ\text{C}$, $V_S = \pm 15\text{V}$, $R_L = 10\text{k}\Omega$ connected to ground, and $V_{REF} = 0\text{V}$, unless otherwise noted.

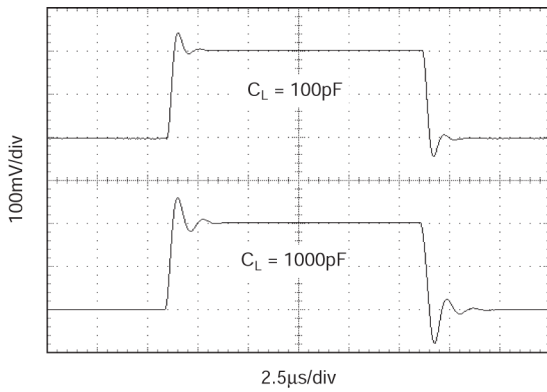


Figure 5-19. Small-Signal Step Response

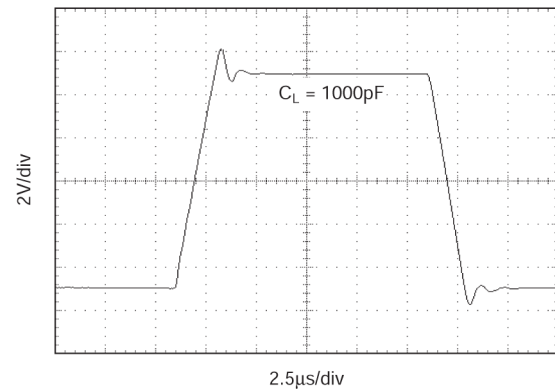


Figure 5-20. Large-Signal Step Response

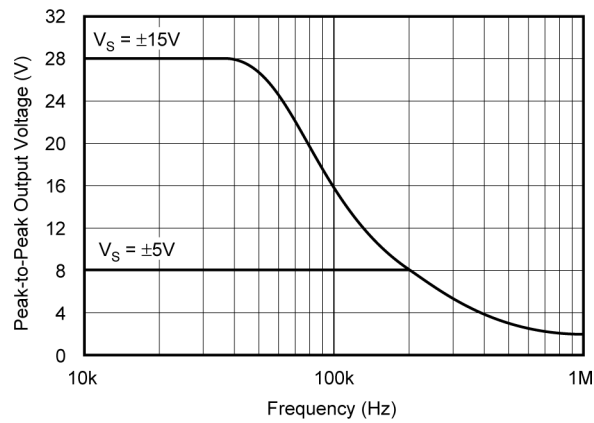


Figure 5-21. Maximum Output Voltage vs Frequency

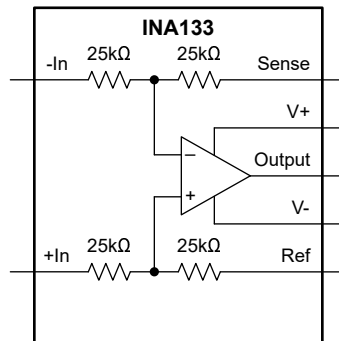
6 Detailed Description

6.1 Overview

The INAx133 has a high-precision operational amplifier and four trimmed, on-chip resistors. The device can be configured to make a wide variety of amplifier configurations, including difference, noninverting, and inverting configurations. The integrated, matched resistors provide an advantage over discrete implementation.

Much of the DC performance of op amp circuits depends on the accuracy of the surrounding resistors. The resistors on the INAx133 are laid out to be tightly matched. The resistors of each part are matched on-chip and tested for matching accuracy. As a result, the INAx133 provides high accuracy for specifications such as gain drift, common-mode rejection ratio, and gain error.

6.2 Functional Block Diagram



6.3 Feature Description

6.3.1 Gain Error and Drift

Gain error in the INAx133 is limited by the mismatch of the integrated precision resistors. Gain drift is limited by the slight mismatch of the temperature coefficient of integrated resistors. The integrated resistors are precision-matched with low temperature coefficient resistors to improve overall gain drift compared to the discrete implementation of differences amplifiers build when using external resistors.

6.3.2 Input Voltage Range

The INAx133 difference amplifier is able to achieve a wide input common-mode voltage range by dividing down the input signal with a high-precision resistor divider. The internal resistors divide down the voltage before the voltage reaches the internal op amp and provide protection to the op amp inputs. Figure 6-1 shows an example of how the voltage division works in a difference-amplifier configuration.

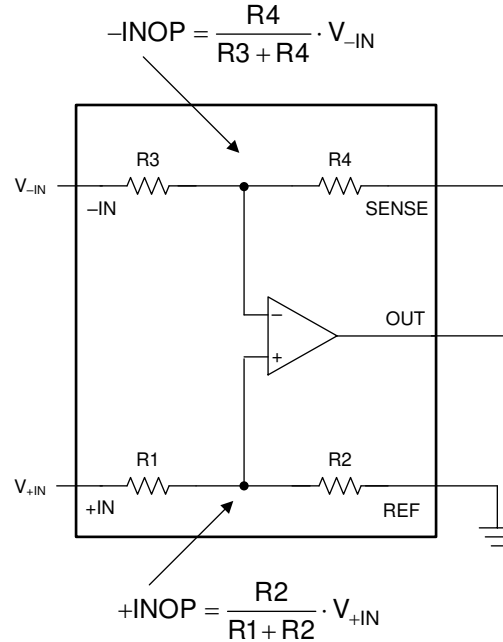


Figure 6-1. Voltage Division in the Difference Amplifier Configuration

6.4 Device Functional Modes

The INAx133 has one functional mode. The device is specified on a power supply of $\pm 15V$ or $\pm 5V$ and can operate on a power supply from $\pm 2.25V$ to $\pm 18V$ with derated performance. See [Typical Characteristics](#).

7 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

7.1 Application Information

The INA133 and INA2133 are high-speed difference amplifiers are designed for a wide range of general-purpose applications. [Figure 7-2](#) shows the basic connections required for operation of the INA133. Place decoupling capacitors close to the device pins as shown in [Figure 7-2](#) in applications with noisy or high impedance power supplies. All circuitry is completely independent in the dual version to provide lowest crosstalk and normal behavior when one amplifier is overdriven or short-circuited.

As shown in [Figure 7-2](#), the differential input signal is connected to pins 2 and 3. The source impedances connected to the inputs must be nearly equal to maintain good common-mode rejection. A 5Ω mismatch in source impedance degrades the common-mode rejection of a typical device to approximately 80dB (a 10Ω mismatch degrades CMR to 74dB). If the source has a known impedance mismatch, an additional resistor in series with the opposite input can be used to preserve good common-mode rejection.

The internal resistors of the INA133 are accurately ratio trimmed to match. That is, R_1 is trimmed to match R_2 and R_3 is trimmed to match R_4 . However, the absolute values may not be equal ($R_1 + R_2$ may be slightly different than $R_3 + R_4$). Thus, large series resistors on the input (greater than 250Ω), even if well matched, can degrade common-mode rejection.

Circuit board layout constraints can suggest possible variations in connections of the internal resistors, like that pins 1 and 3 can be interchanged. However, because of the ratio trimming technique used, CMRR can degrade. If pins 1 and 3 are interchanged, pins 2 and 5 must also be interchanged to maintain proper ratio matching.

7.1.1 Operating Voltage

The INA133 and INA2133 operate from single (+4.5V to +36V) or dual ($\pm 2.25V$ to $\pm 18V$) supplies with excellent performance. Specifications are production tested with $\pm 5V$ and $\pm 15V$ supplies. Most behavior remains unchanged throughout the full operating voltage range. Parameters which vary significantly with operating voltage are shown in the [Typical Characteristics](#).

7.1.2 Input Voltage

The INA133 and INA2133 can accurately measure differential signals that are above and below the supply rails. Linear common-mode range extends from $2 \times (V+) - 3V$ to $2 \times (V-) + 3V$ (nearly twice the supplies). See [Figure 5-5](#).

7.1.3 Offset Voltage Trim

The INA133 and INA2133 are laser trimmed for low offset voltage and drift. Most applications require no external offset adjustment. [Figure 7-1](#) shows an optional circuit for trimming the output offset voltage. The output is referred to the output reference terminal (pin 1), which is normally grounded. A voltage applied to the Ref terminal is summed with the output signal. This can be used to null offset voltage as shown in [Figure 7-1](#). To maintain good common-mode rejection, keep the source impedance of a signal applied to the Ref terminal less than 10Ω.

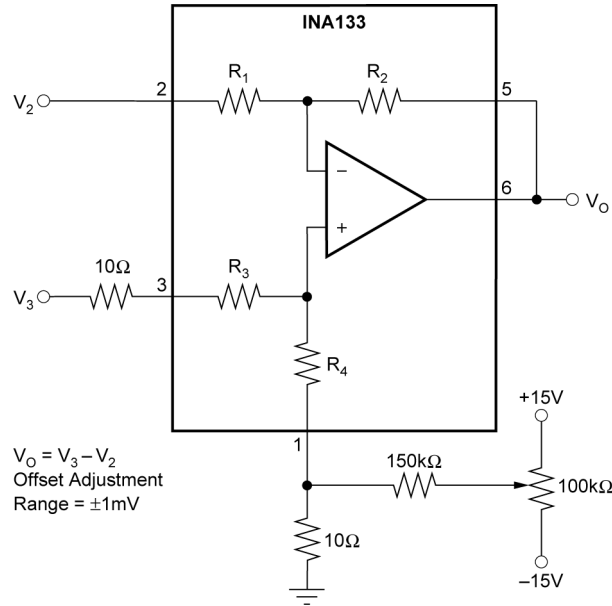


Figure 7-1. Offset Adjustment

7.2 Typical Application

The INAx133 can be used in a variety of applications. [Figure 7-2](#) shows one example.

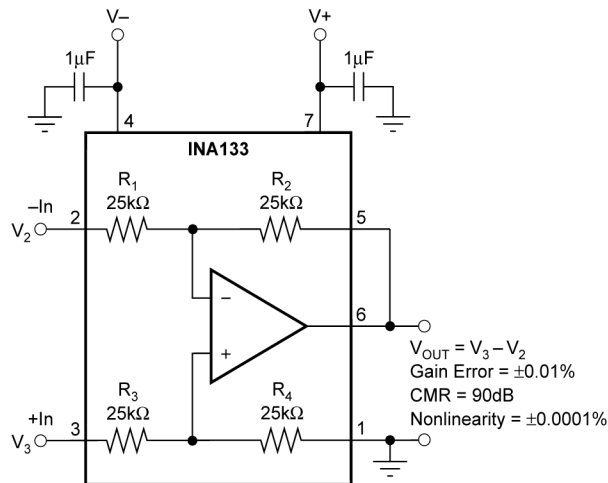


Figure 7-2. Precision Difference Amplifier (Basic Power Supply and Signal Connections)

7.3 Additional Applications

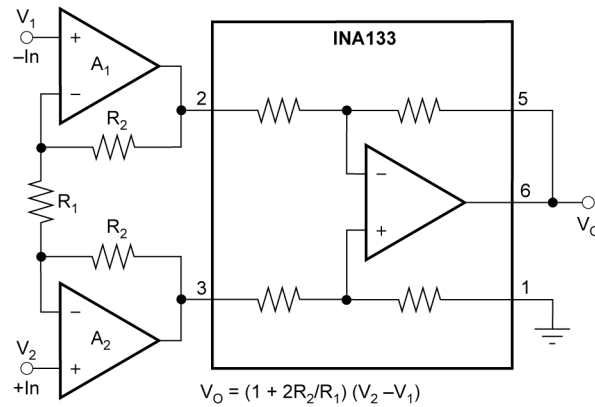
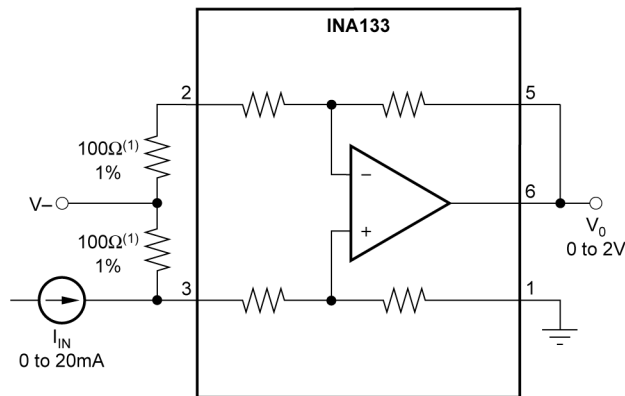


Figure 7-3. Precision Instrumentation Amplifier



NOTE: (1) Input series resistors should be less than 250Ω (1% max mismatch) to maintain excellent CMR. With 100Ω resistors, gain error is increased to 0.5%.

Figure 7-4. Current Receiver With Compliance to Rails

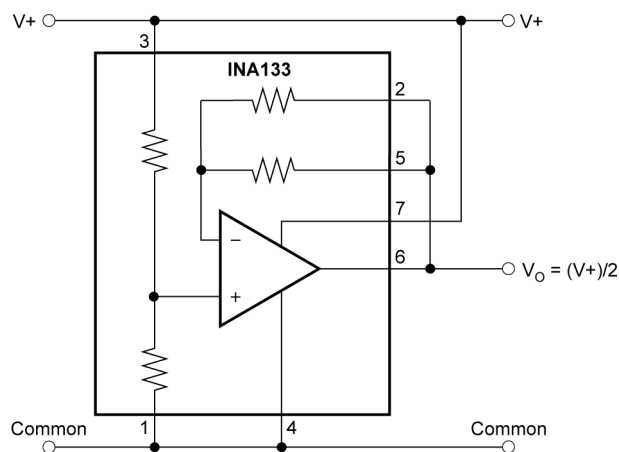


Figure 7-5. Pseudoground Generator

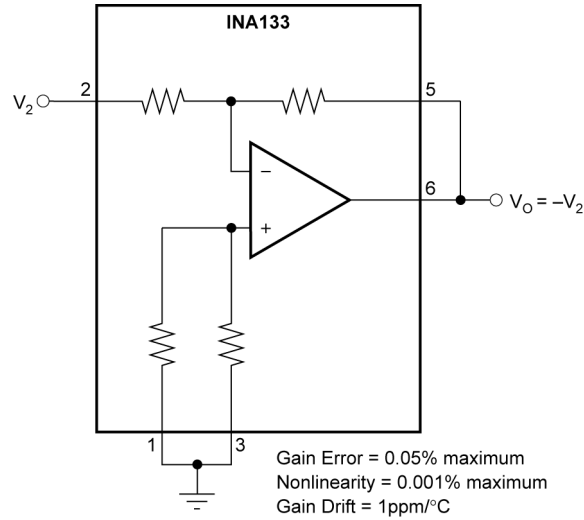


Figure 7-6. Precision Unity-Gain Inverting Amplifier

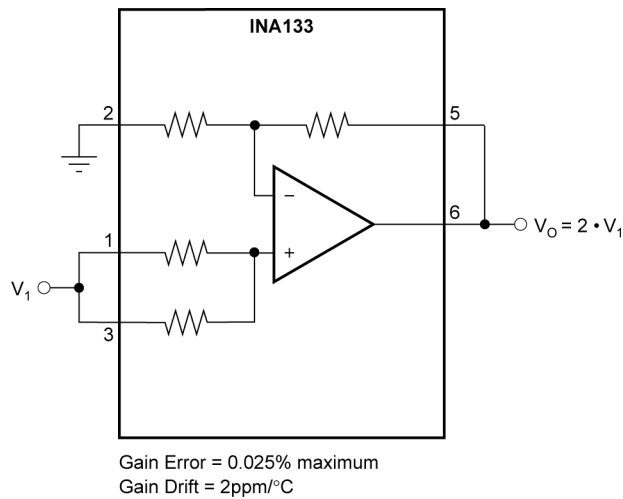


Figure 7-7. Precision Gain = 2 Amplifier

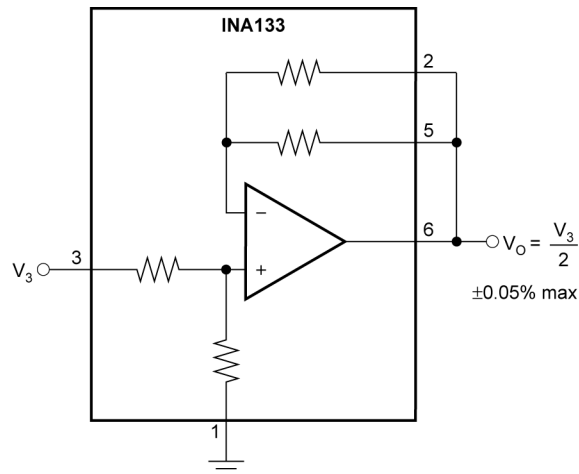


Figure 7-8. Precision Gain = 1/2 Amplifier

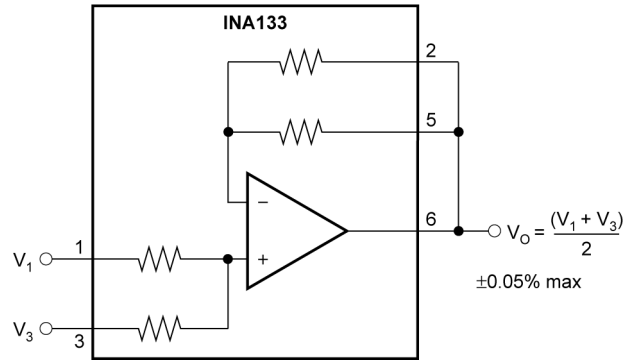


Figure 7-9. Precision Average Value Amplifier

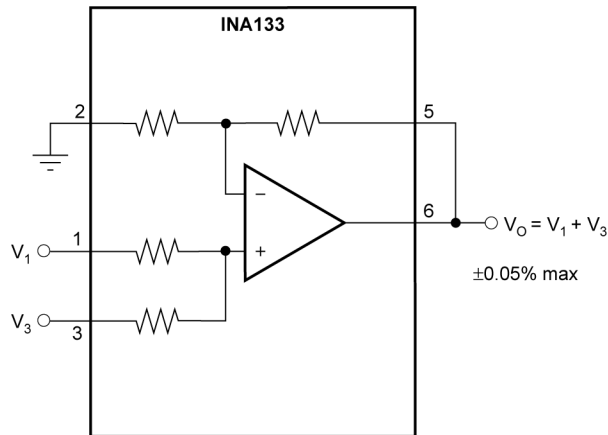


Figure 7-10. Precision Summing Amplifier

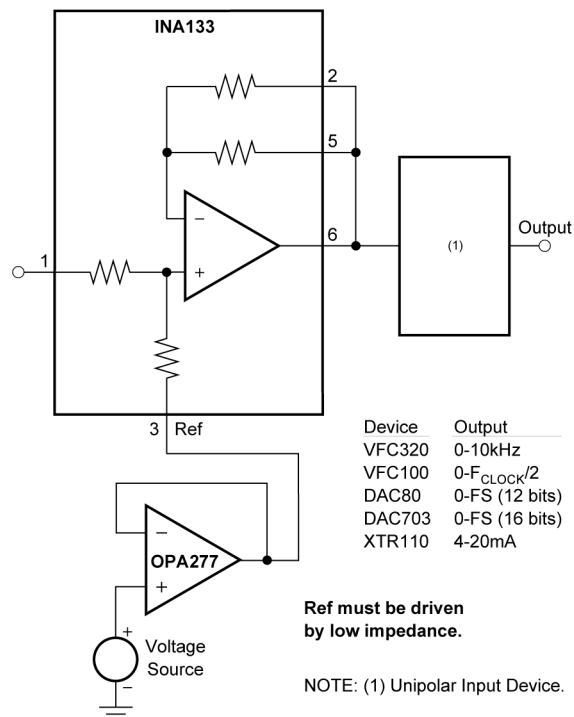


Figure 7-11. Precision Bipolar Offsetting

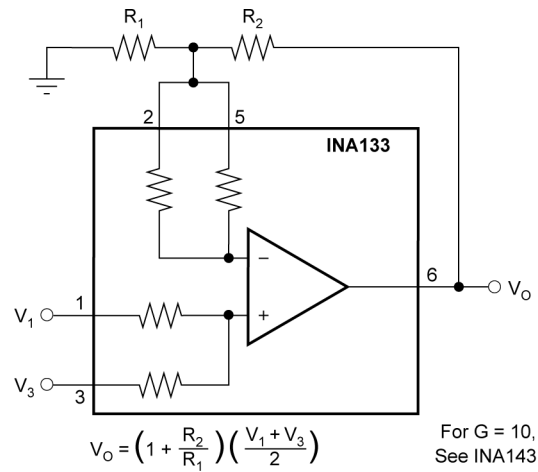


Figure 7-12. Precision Summing Amplifier With Gain

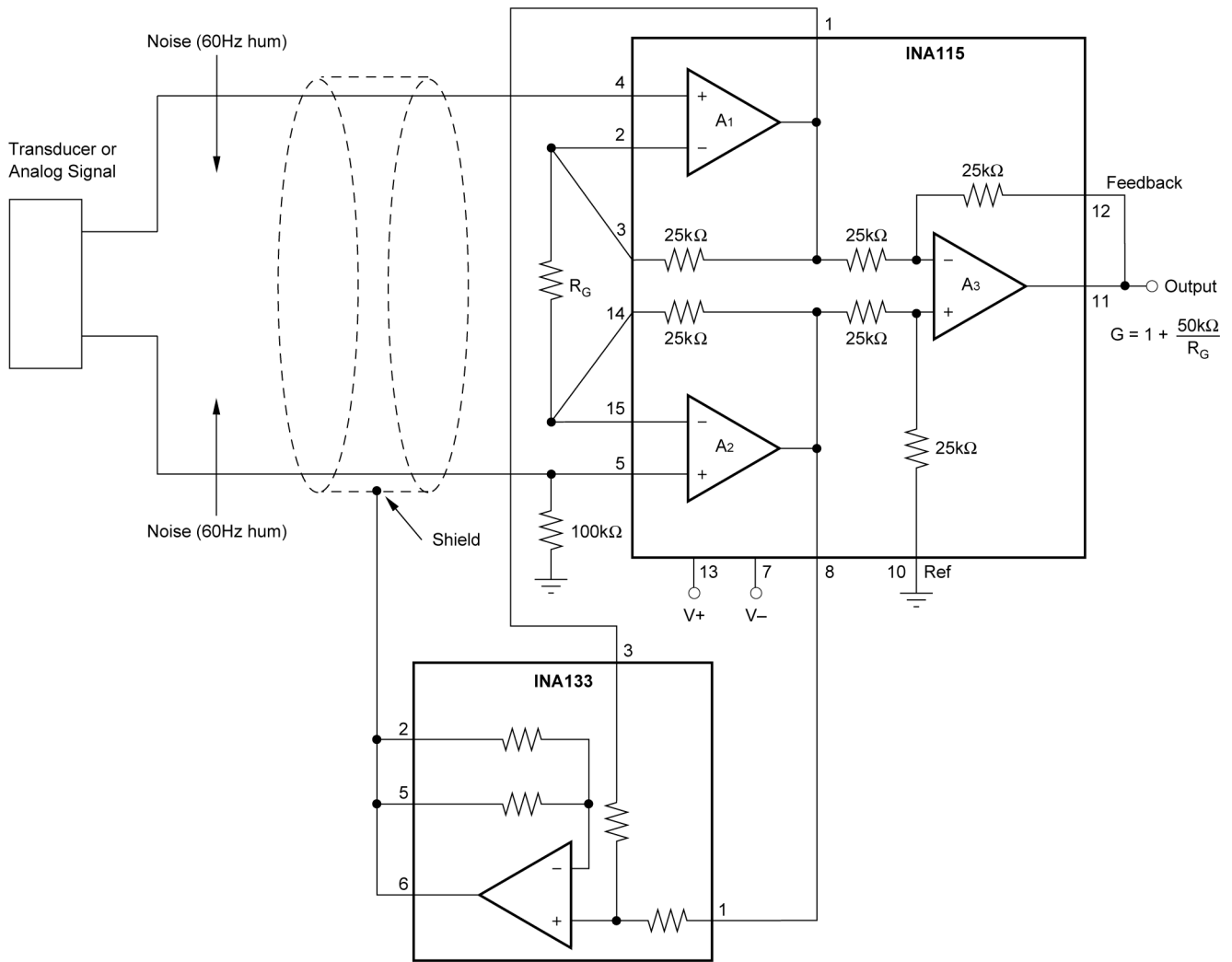


Figure 7-13. Instrumentation Amplifier Guard Drive Generator

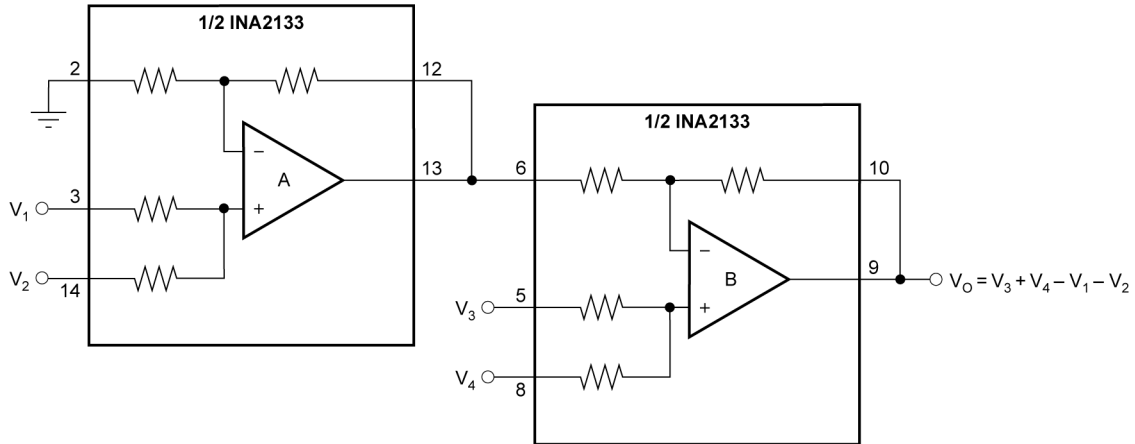


Figure 7-14. Precision Summing Instrumentation Amplifier

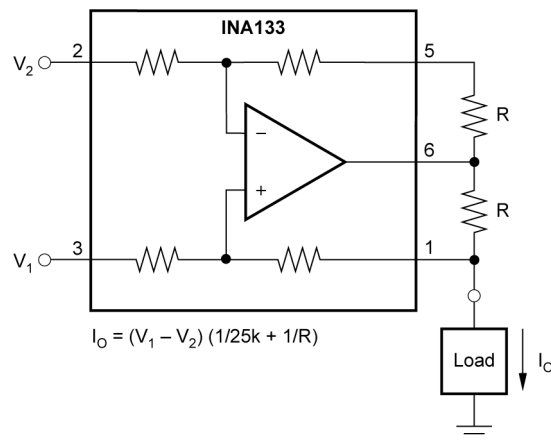


Figure 7-15. Precision Voltage-to-Current Converter With Differential Inputs

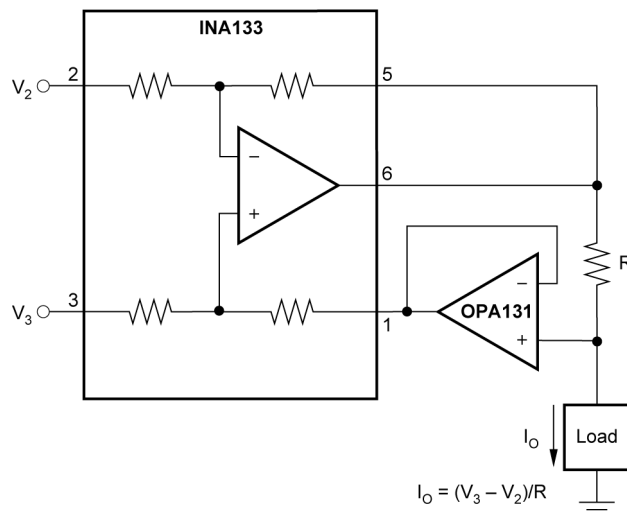


Figure 7-16. Differential Input Voltage-to-Current Converter for Low I_{OUT}

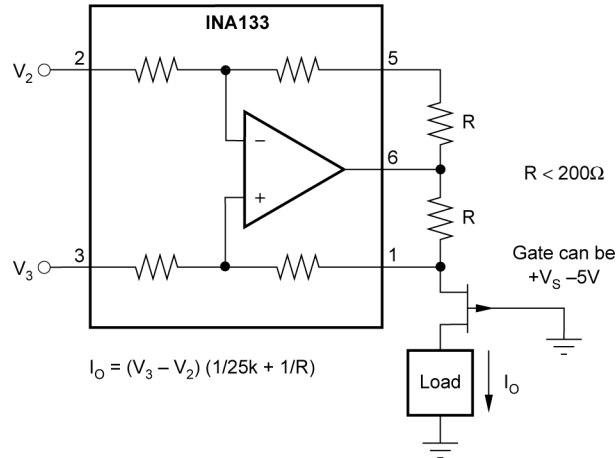


Figure 7-17. Isolating Current Source

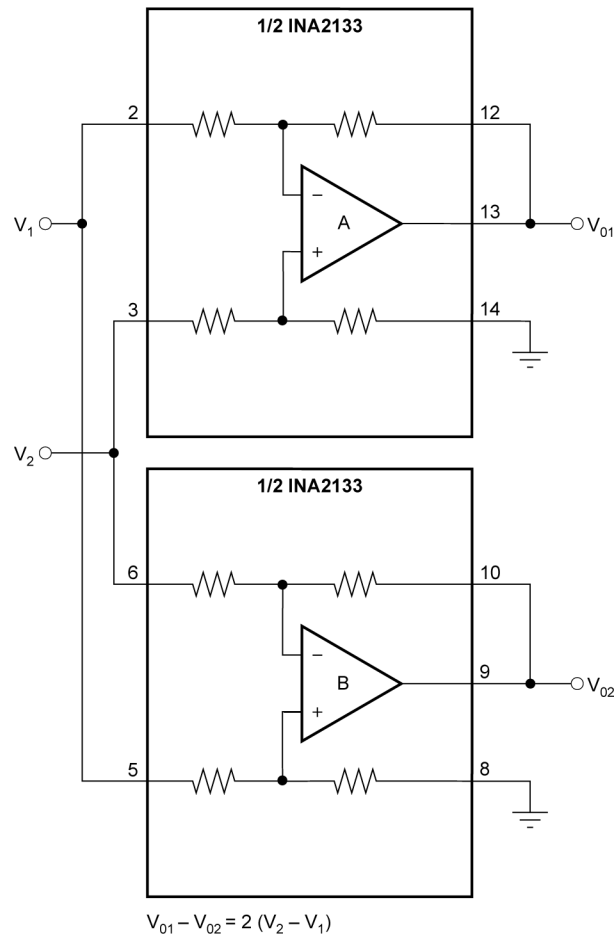


Figure 7-18. Differential Output Difference Amplifier

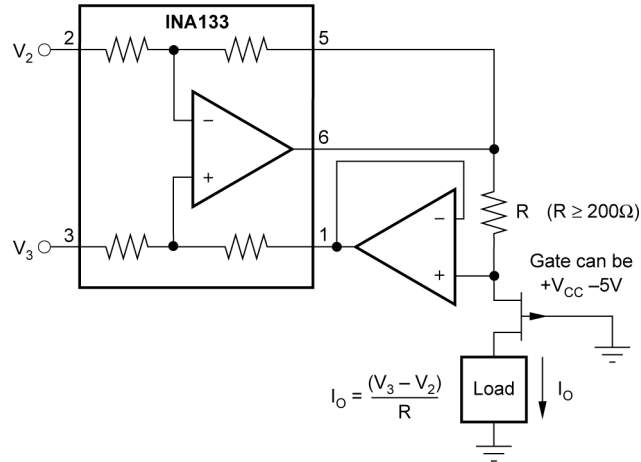


Figure 7-19. Isolating Current Source With Buffering Amplifier for Greater Accuracy

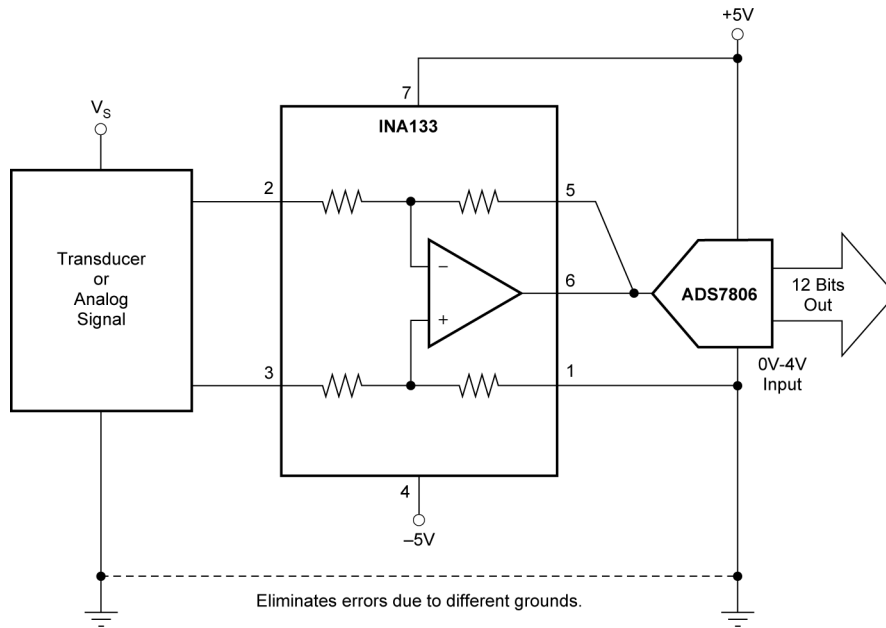


Figure 7-20. Differential Input Data Acquisition

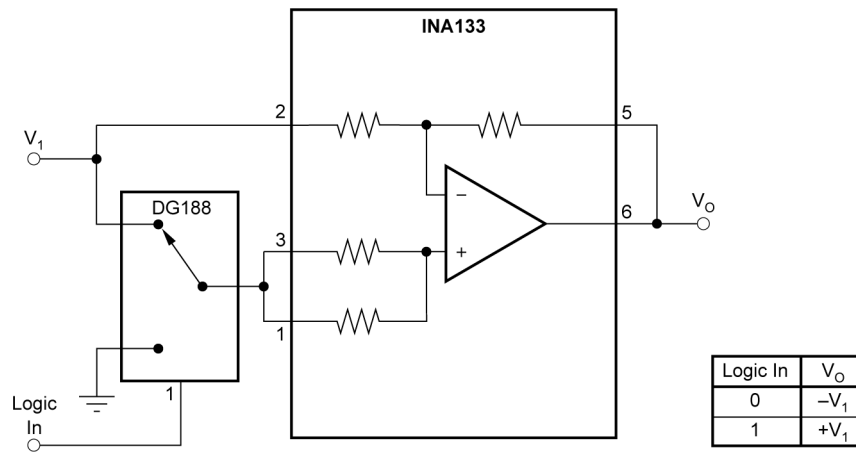


Figure 7-21. Digitally Controlled Gain of ± 1 Amplifier

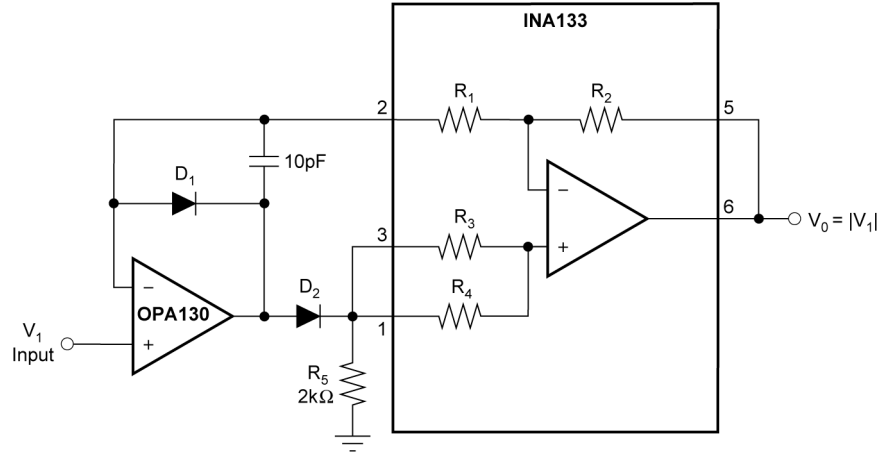


Figure 7-22. Precision Absolute Value Buffer

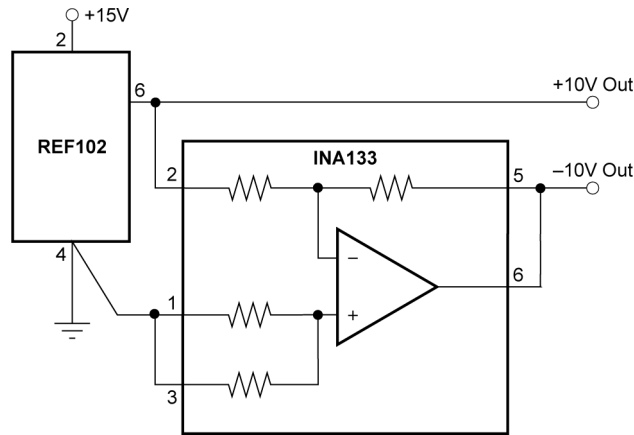


Figure 7-23. ±10V Precision Voltage Reference

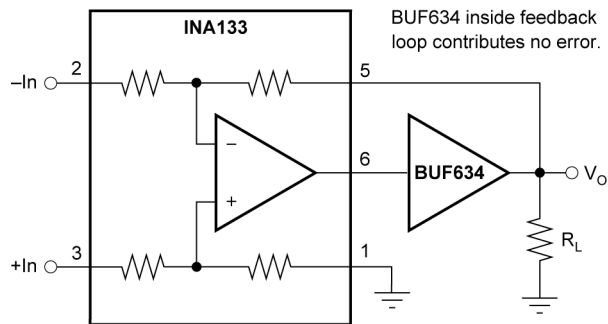


Figure 7-24. High Output Current Precision Difference Amplifier

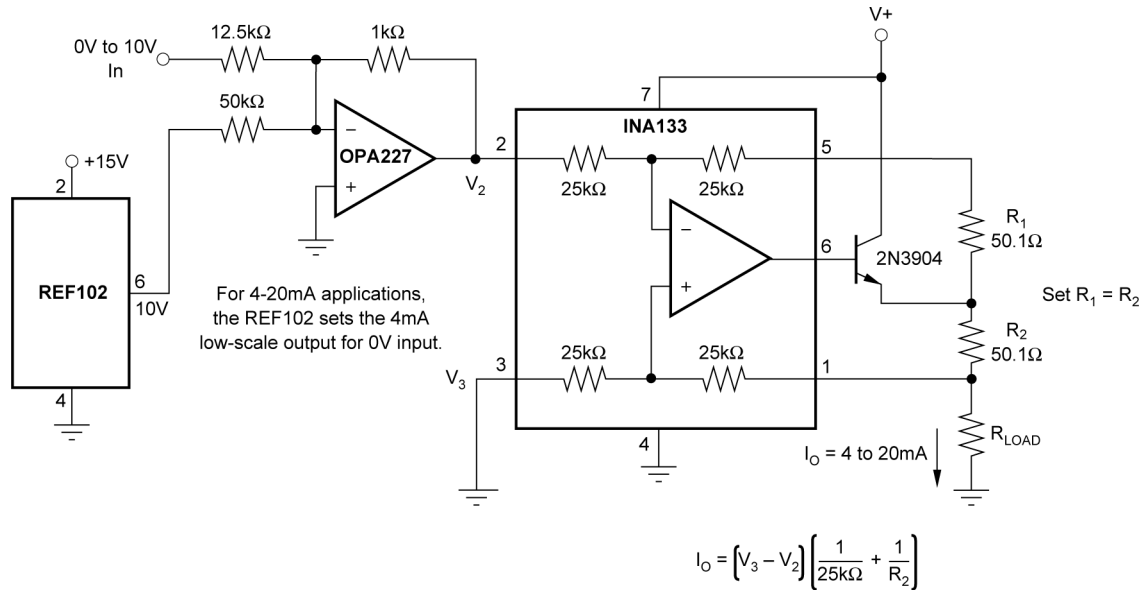


Figure 7-25. Precision Voltage-to-Current Conversion

7.4 Power Supply Recommendations

The nominal performance of the INA105 is specified with a supply voltage of $\pm 15V$ or $\pm 5V$. The device operates using power supplies from $\pm 2.25V$ to $\pm 18V$ with varying performance. Parameters varying across the operating voltage and reference voltage range can be referenced in the [Typical Characteristics](#).

TI highly recommends to add low-ESR ceramic bypass capacitors (C_{BYP}) between each supply pin and ground. Only one C_{BYP} is sufficient for single supply operation. Place the C_{BYP} as close to the device as possible to reduce coupling errors from noisy or high-impedance power supplies. Route the power supply trace through C_{BYP} before reaching the device power supply terminals. For more information, see [Layout Guidelines](#).

7.5 Layout

7.5.1 Layout Guidelines

Attention to good layout practices is always recommended. For best operational performance of the device, use good PCB layout practices, including:

- Make sure that both input paths are well-matched for source impedance and capacitance to avoid converting common-mode signals into differential signals.
- Noise propagates into analog circuitry through the power pins of the circuit as a whole and of the device. Bypass capacitors reduce the coupled noise by providing low-impedance power sources local to the analog circuitry.
 - Connect low-ESR, 0.1 μF ceramic bypass capacitors between each supply pin and ground, placed as close to the device as possible. A single bypass capacitor from V+ to ground is applicable for single-supply applications.
- To reduce parasitic coupling, route the input traces as far away from the supply or output traces as possible. If these traces cannot be kept separate, crossing the sensitive trace perpendicular is preferred over crossing in parallel with the noisy trace.
- Place the external components as close to the device as possible.
- Keep the traces as short as possible.

7.5.2 Layout Examples

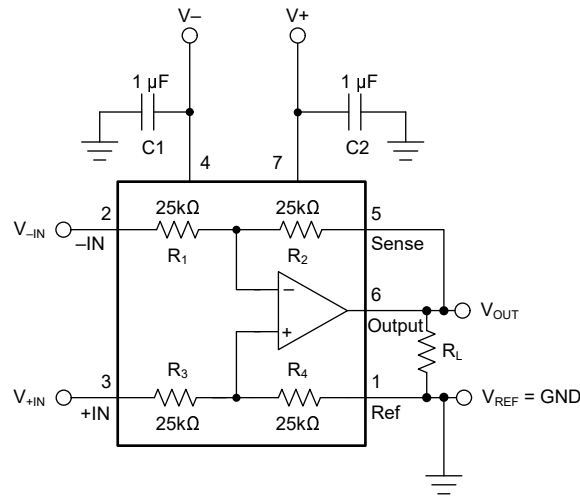


Figure 7-26. Example Schematic

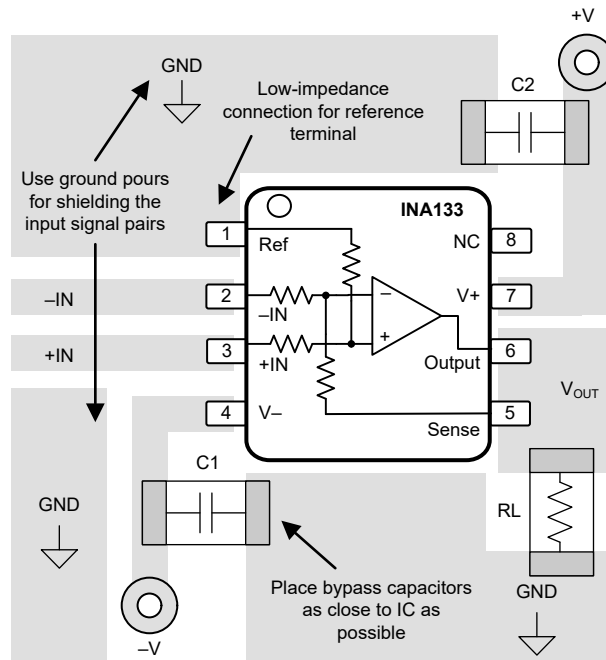


Figure 7-27. Associated PCB Layout for SOIC-8 Package

8 Device and Documentation Support

TI offers an extensive line of development tools. Tools and software to evaluate the performance of the device, generate code, and develop solutions are listed below.

8.1 Device Support

8.1.1 Development Support

For development support on this product, see the following:

8.1.1.1 PSpice® for TI

PSpice® for TI is a design and simulation environment that helps evaluate performance of analog circuits. Create subsystem designs and prototype solutions before committing to layout and fabrication, reducing development cost and time to market.

8.1.1.2 TINA-TI™ (Free Software Download)

TINA™ is a simple, powerful, and easy-to-use circuit simulation program based on a SPICE engine. TINA-TI is a free, fully-functional version of the TINA software, preloaded with a library of macro models in addition to a range of both passive and active models. TINA-TI provides all the conventional dc, transient, and frequency domain analysis of SPICE, as well as additional design capabilities.

Available as a [free download](#) from the Analog eLab Design Center, TINA-TI offers extensive post-processing capability that allows users to format results in a variety of ways. Virtual instruments offer the ability to select input waveforms and probe circuit nodes, voltages, and waveforms, creating a dynamic quick-start tool.

Note

These files require that either the TINA software (from DesignSoft™) or TINA-TI software be installed. Download the free TINA-TI software from the [TINA-TI folder](#).

8.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on [ti.com](#). Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

8.3 Support Resources

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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8.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

8.6 Glossary

[TI Glossary](#) This glossary lists and explains terms, acronyms, and definitions.

9 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision * (June 1999) to Revision A (March 2025)	Page
• Updated the numbering and format for tables, figures, and cross-references throughout the document.....	1
• Added the Pin Configuration and Functions, Specifications, Recommended Operating Conditions, Thermal Information, Detailed Description, Overview, Functional Block Diagram, Feature Description, Device Functional Modes, Application and Implementation, Power Supply Recommendations, Layout, Layout Guidelines, Layout Example, Device and Documentation Support, and Mechanical, Packaging, and Orderable Information sections.....	1
• Changed the Package Information table.....	1
• Added <i>Pin Functions</i> table for INA133 and INA2133.....	3
• Added test conditions to <i>Electrical Characteristics</i> table.....	6
• Combined $V_S = \pm 15V$ and $V_S = \pm 5V$ specification table in <i>Electrical Characteristics</i>	6
• Changed parameter name in <i>Electrical Characteristics</i> from <i>Offset Voltage Initial vs Temperature</i> to <i>Offset voltage drift</i>	6
• Changed parameter name in <i>Electrical Characteristics</i> from <i>Offset Voltage Initial vs Power Supply</i> to <i>Power supply rejection ratio</i>	6
• Changed parameter name in <i>Electrical Characteristics</i> from <i>Offset Voltage vs Time</i> to <i>Long-term stability</i>	6
• Changed parameter name in <i>Electrical Characteristics</i> from <i>Current Limit, Continuous-to-Common</i> to <i>Short-circuit current</i> and added test condition.....	6
• Moved the power supply and temperature ranges from the <i>Electrical Characteristics</i> table to the <i>Recommended Operating Conditions</i> and <i>Absolute Maximum Ratings</i> table.....	6
• Changed the Applications section.....	14

10 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
INA133U	ACTIVE	SOIC	D	8	75	RoHS & Green	Call TI	Level-3-260C-168 HR	-40 to 85	INA 133U	Samples
INA133U/2K5	ACTIVE	SOIC	D	8	2500	RoHS & Green	Call TI	Level-3-260C-168 HR		INA 133U	Samples
INA133UA	ACTIVE	SOIC	D	8	75	RoHS & Green	Call TI	Level-3-260C-168 HR		INA 133U A	Samples
INA133UA/2K5	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-3-260C-168 HR		INA 133U A	Samples
INA2133U	ACTIVE	SOIC	D	14	50	RoHS & Green	Call TI	Level-3-260C-168 HR	-55 to 125	INA2133U	Samples
INA2133UA	ACTIVE	SOIC	D	14	50	RoHS & Green	Call TI	Level-3-260C-168 HR	-55 to 125	INA2133U A	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSELETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
INA133U/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
INA133UA/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
INA133U/2K5	SOIC	D	8	2500	356.0	356.0	35.0
INA133UA/2K5	SOIC	D	8	2500	367.0	367.0	35.0

TUBE


*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
INA133U	D	SOIC	8	75	506.6	8	3940	4.32
INA133UA	D	SOIC	8	75	506.6	8	3940	4.32
INA2133U	D	SOIC	14	50	506.6	8	3940	4.32
INA2133UA	D	SOIC	14	50	506.6	8	3940	4.32

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