LMT87-Q1 2.7-V、SC70、Class-AB 出力のアナログ温度センサ

1 特長

- LMT87-Q1-Q1 は車載用アプリケーション向けに AEC-Q100 認定済み:
  - デバイス温度グレード 0:–40℃～+150℃
  - デバイス HBM ESD 分類レベル 2
  - デバイス CDM ESD 分類レベル C6
- 機能安全対応
  - 機能安全システムの設計に役立つ資料を利用可能
- 非常に高精度:±0.4°C (標準値)
- 低電圧 2.7V での動作
- 平均センサ・ゲイン:–13.6mV/℃
- 低い静止電流:5.4μA
- 広い温度範囲:-50℃～150℃
- 出力の短絡保護
- 駆動能力 ±50μA のプッシュプル出力
- 業界標準の LM20/19 および LM35 温度センサとフットプリント互換
- コスト効率に優れたサーミスタの代替

2 アプリケーション

- 車載用
- インフォテインメントおよびクラスタ
- パワートレイン・システム
- 煙および熱検出器
- ドローン
- 家電製品

3 概要

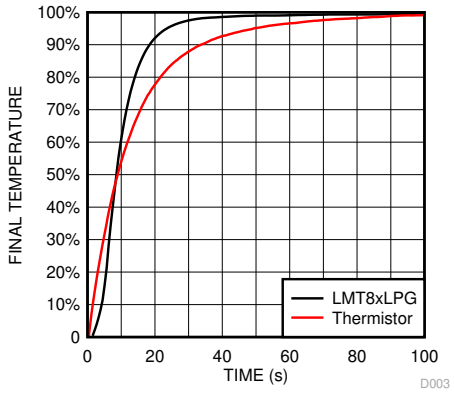
LMT87-Q1 デバイスは高精度の CMOS 温度センサであり、標準精度が ±0.4℃ (最大±2.7℃) で、リニアなアナログ出力電圧を備え、この電圧は温度に反比例します。2.7V の電源電圧で動作し、静止電流が 5.4μA、パワーオン時間が 0.7ms で、効率的なパワーサイクリング・アーキテクチャを実現し、ドローンやセンサ・ノードなどバッテリ駆動のアプリケーションで消費電力を最小化できます。LMT87-Q1-Q1 デバイスは、AEC-Q100 グレード 0 認定済みで、較正なしで全動作温度範囲にわたって ±2.7°C の最大精度を維持します。このため、LMT87-Q1-Q1 はインフォテインメント、クラスタ、パワートレイン・システムなどの車載アプリケーションに適しています。広い動作範囲での精度や、その他の特長から、LMT87-Q1 はサーミスタの優れた代替となります。

異なる平均センサ・ゲインおよび類似の精度を持つデバイスについては、「*類似の代替デバイス*」で、LMT8x ファミリの他のデバイスを参照してください。

デバイス情報(1)

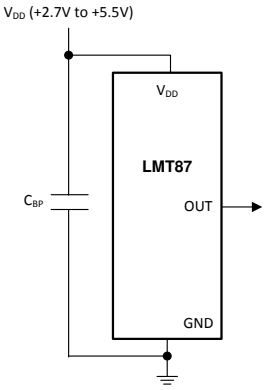
部品番号	パッケージ	本体サイズ (公称)
LMT87-Q1	SOT (5)	2.00mm×1.25mm

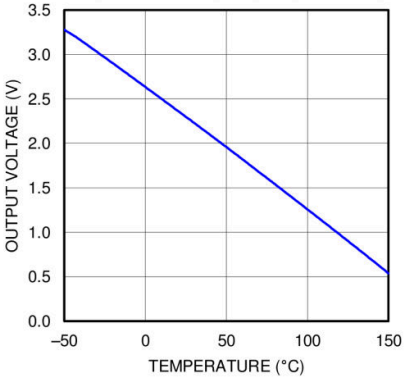
(1) 利用可能なパッケージについては、このデータシートの末尾にある注文情報を参照してください。



* 高速な熱応答 NTC

熱時定数





出力電圧と温度との関係

Table of Contents

4 Revision History

資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

Changes from Revision * (October 2017) to Revision A (June 2022)	Page
• 文書全体にわたって表、図、相互参照の採番方法を更新	1
• 「*特長*」セクションに機能安全の箇条書き項目を追加	1

5 Device Comparison

表 5-1. Available Device Packages

ORDER NUMBER(1)	PACKAGE	PIN	BODY SIZE (NOM)	MOUNTING TYPE
LMT87DCK	SOT (AKA(2): SC70, DCK)	5	2.00 mm × 1.25 mm	Surface Mount
LMT87LP	TO-92 (AKA(2): LP)	3	4.30 mm × 3.50 mm	Through-hole; straight leads
LMT87LPG	TO-92S (AKA(2): LPG)	3	4.00 mm × 3.15 mm	Through-hole; straight leads
LMT87LPM	TO-92 (AKA(2): LPM)	3	4.30 mm × 3.50 mm	Through-hole; formed leads
LMT87DCK-Q1	SOT (AKA(2): SC70, DCK)	5	2.00 mm × 1.25 mm	Surface Mount

(1) For all available packages and complete order numbers, see the Package Option addendum at the end of the data sheet.
(2) AKA = Also Known As

表 5-2. Comparable Alternative Devices

DEVICE NAME	AVERAGE OUTPUT SENSOR GAIN	POWER SUPPLY RANGE
LMT84-Q1	–5.5 mV/°C	1.5 V to 5.5 V
LMT85-Q1	–8.2 mV/°C	1.8 V to 5.5 V
LMT86-Q1	–10.9 mV/°C	2.2 V to 5.5 V
LMT87-Q1	–13.6 mV/°C	2.7 V to 5.5 V

6 Pin Configuration and Functions

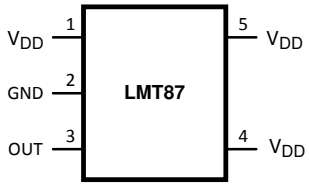
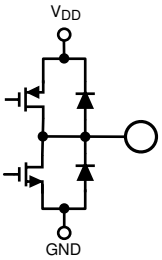


図 6-1. DCK Package 5-Pin SOT (SC70) Top View

表 6-1. Pin Functions

PIN		TYPE	DESCRIPTION	
NAME	SOT (SC70)		EQUIVALENT CIRCUIT	FUNCTION
GND	2(1)	Ground	N/A	Power Supply Ground
OUT	3	Analog Output		Outputs a voltage that is inversely proportional to temperature
V_{DD}	1, 4, 5	Power	N/A	Positive Supply Voltage

(1) Direct connection to the back side of the die

7 Specifications

7.1 Absolute Maximum Ratings

See (1) (3)

	MIN	MAX	UNIT
Supply voltage	–0.3	6	V
Voltage at output pin	–0.3	(V_{DD} + 0.5)	V
Output current	–7	7	mA
Input current at any pin(2)	–5	5	mA
Maximum junction temperature (T_{JMAX})		150	°C
Storage temperature T_{stg}	–65	150	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) When the input voltage (V_I) at any pin exceeds power supplies (V_I < GND or V_I > V), the current at that pin should be limited to 5 mA.
(3) *Soldering process must comply with Reflow Temperature Profile specifications. Refer to* www.ti.com/packaging.

7.2 ESD Ratings

			VALUE	UNIT
LMT87DCK-Q1 in SC70 package				
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per AEC Q100-002(1)	±2500	V
		Charged-device model (CDM), per AEC Q100-011	±1000	

(1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

7.3 Recommended Operating Conditions

	MIN	MAX	UNIT
Specified temperature	$T_{MIN} \le T_A \le T_{MAX}$		°C
	$-50 \le T_A \le 150$		°C
Supply voltage (V_{DD})	2.7	5.5	V

7.4 Thermal Information

THERMAL METRIC(1) (2)		LMT87-Q1 DCK (SOT/SC70) 5 PINS	UNIT
$R_{\theta JA}$	Junction-to-ambient thermal resistance (3) (4)	275	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	84	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	56	°C/W
ψ_{JT}	Junction-to-top characterization parameter	1.2	°C/W
ψ_{JB}	Junction-to-board characterization parameter	55	°C/W

(1) For information on self-heating and thermal response time see section *Mounting and Thermal Conductivity*.
(2) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report.
(3) The junction to ambient thermal resistance ($R_{\theta JA}$) under natural convection is obtained in a simulation on a JEDEC-standard, High-K board as specified in JESD51-7, in an environment described in JESD51-2. Exposed pad packages assume that thermal vias are included in the PCB, per JESD 51-5.
(4) Changes in output due to self-heating can be computed by multiplying the internal dissipation by the thermal resistance.

7.5 Accuracy Characteristics

These limits do not include DC load regulation. These stated accuracy limits are with reference to the values in 表 8-1.

PARAMETER	CONDITIONS	MIN(1)	TYP	MAX(1)	UNIT
Temperature accuracy(2)	70°C to 150°C; V_{DD} = 3.0 V to 5.5 V	–2.7	±0.4	2.7	°C
	20°C to 40°C; V_{DD} = 2.7 V to 5.5 V		±0.6		°C
	20°C to 40°C; V_{DD} = 3.4 V to 5.5 V		±0.3		°C
	0°C; V_{DD} = 3.0 V to 5.5 V	–2.7	±0.6	2.7	°C
	0°C; V_{DD} = 3.6 V to 5.5 V		±0.3		°C
	–50°C; V_{DD} = 3.6 V to 5.5 V	–2.7	±0.6	2.7	°C
	–50°C; V_{DD} = 4.2 V to 5.5 V		±0.3		°C

(1) Limits are specific to TI's AOQL (Average Outgoing Quality Level).
(2) Accuracy is defined as the error between the measured and reference output voltages, tabulated in the Transfer Table at the specified conditions of supply gain setting, voltage, and temperature (expressed in °C). Accuracy limits include line regulation within the specified conditions. Accuracy limits do not include load regulation; they assume no DC load.

7.6 Electrical Characteristics

Unless otherwise noted, these specifications apply for +V_{DD} = 2.7 V to 5.5 V. MIN and MAX limits apply for T_A = T_J = T_{MIN} to T_{MAX} ; typical limits apply for T_A = T_J = 25°C.

PARAMETER		TEST CONDITIONS	MIN(2)	TYP (1)	MAX (2)	UNIT
	Sensor gain (output transfer function slope)			–13.6		mV/°C
	Load regulation(3)	Source ≤ 50 µA, (V_{DD} – V_{OUT}) ≥ 200 mV	–1	–0.22		mV
		Sink ≤ 50 µA, V_{OUT} ≥ 200 mV		0.26	1	mV
	Line regulation(4)			200		µV/V
I_S	Supply current	T_A = 30°C to 150°C, (V_{DD} – V_{OUT}) ≥ 100 mV		5.4	8.1	µA
		T_A = –50°C to 150°C, (V_{DD} – V_{OUT}) ≥ 100 mV		5.4	9	µA
C_L	Output load capacitance			1100		pF
	Power-on time(5)	C_L= 0 pF to 1100 pF		0.7	1.9	ms
	Output drive	T_A = T_J = 25°C	–50		50	µA

(1) Typicals are at T_J = T_A = 25°C and represent most likely parametric norm.
(2) Limits are specific to TI's AOQL (Average Outgoing Quality Level).
(3) Source currents are flowing out of the LMT87-Q1. Sink currents are flowing into the LMT87-Q1.
(4) Line regulation (DC) is calculated by subtracting the output voltage at the highest supply voltage from the output voltage at the lowest supply voltage. The typical DC line regulation specification does not include the output voltage shift discussed in *Output Voltage Shift*.
(5) Specified by design and characterization.

7.7 Typical Characteristics

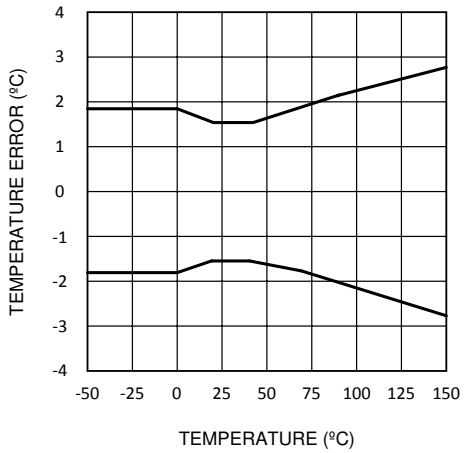


図 7-1. Temperature Error vs Temperature

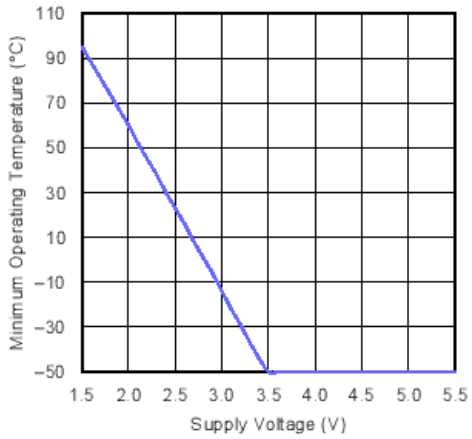


図 7-2. Minimum Operating Temperature vs Supply Voltage

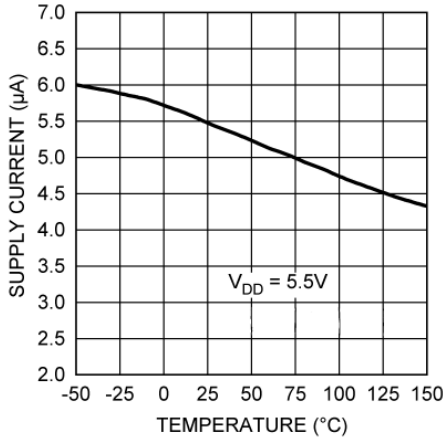


図 7-3. Supply Current vs Temperature

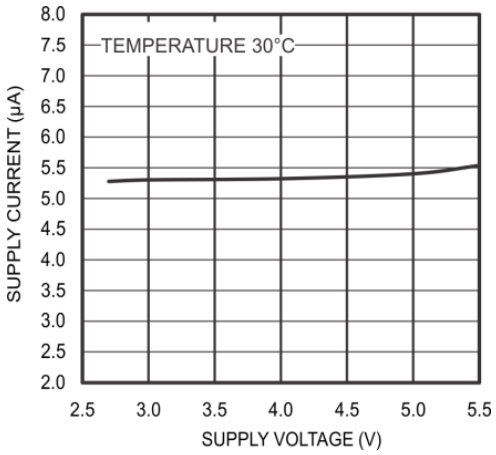


図 7-4. Supply Current vs Supply Voltage

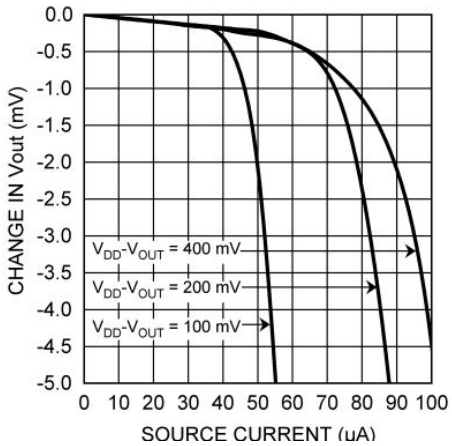


図 7-5. Load Regulation, Sourcing Current

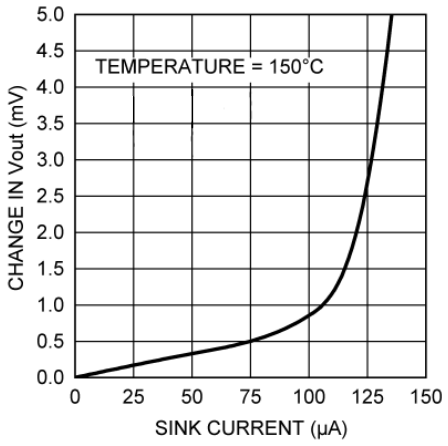


図 7-6. Load Regulation, Sinking Current

7.7 Typical Characteristics (continued)

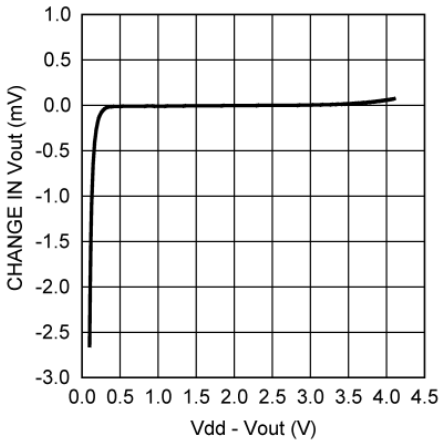


図 7-7. Change in V_{OUT} vs Overhead Voltage

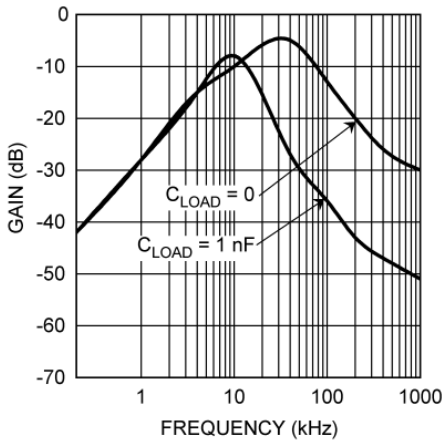


図 7-8. Supply-Noise Gain vs Frequency

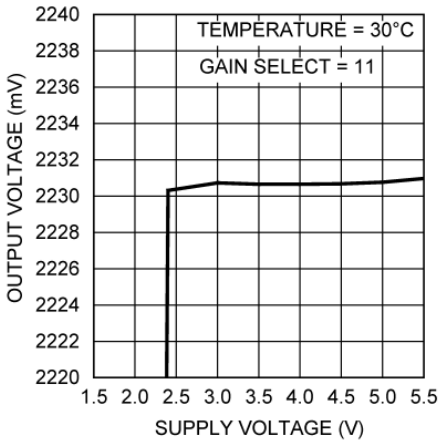


図 7-9. Output Voltage vs Supply Voltage

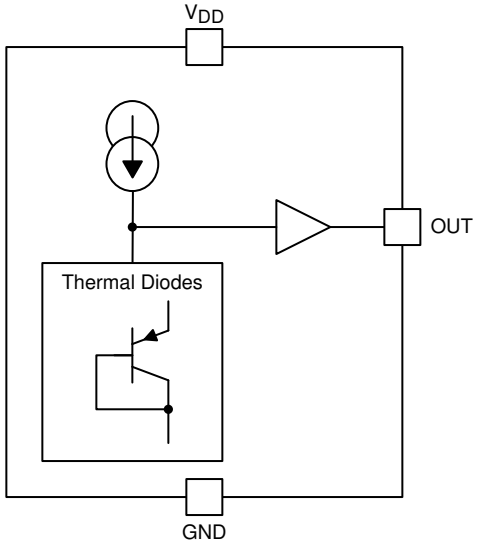
8 Detailed Description

8.1 Overview

The LMT87-Q1 is an analog output temperature sensor. The temperature-sensing element is comprised of a simple base emitter junction that is forward biased by a current source. The temperature-sensing element is then buffered by an amplifier and provided to the OUT pin. The amplifier has a simple push-pull output stage thus providing a low impedance output source.

8.2 Functional Block Diagram

Full-Range Celsius Temperature Sensor (−50°C to +150°C)



8.3 Feature Description

8.3.1 LMT87-Q1 Transfer Function

表 8-1 shows the output voltage of the LMT87-Q1 across the complete operating temperature range. This table is the reference from which the LMT87-Q1 accuracy specifications (listed in the *Accuracy Characteristics* table) are determined. This table can be used, for example, in a host processor look-up table. A file containing this data is available for download at the LMT87-Q1 product folder under *Tools and Software Models*.

表 8-1. LMT87-Q1 Transfer Table

TEMP (°C)	V_{OUT} (mV)	TEMP (°C)	V_{OUT} (mV)	TEMP (°C)	V_{OUT} (mV)	TEMP (°C)	V_{OUT} (mV)	TEMP (°C)	V_{OUT} (mV)
–50	3277	–10	2767	30	2231	70	1679	110	1115
–49	3266	–9	2754	31	2217	71	1665	111	1101
–48	3254	–8	2740	32	2204	72	1651	112	1087
–47	3243	–7	2727	33	2190	73	1637	113	1073
–46	3232	–6	2714	34	2176	74	1623	114	1058
–45	3221	–5	2700	35	2163	75	1609	115	1044
–44	3210	–4	2687	36	2149	76	1595	116	1030
–43	3199	–3	2674	37	2136	77	1581	117	1015
–42	3186	–2	2660	38	2122	78	1567	118	1001
–41	3173	–1	2647	39	2108	79	1553	119	987
–40	3160	0	2633	40	2095	80	1539	120	973
–39	3147	1	2620	41	2081	81	1525	121	958
–38	3134	2	2607	42	2067	82	1511	122	944
–37	3121	3	2593	43	2054	83	1497	123	929
–36	3108	4	2580	44	2040	84	1483	124	915
–35	3095	5	2567	45	2026	85	1469	125	901

表 8-1. LMT87-Q1 Transfer Table (continued)

TEMP (°C)	V_{OUT} (mV)	TEMP (°C)	V_{OUT} (mV)	TEMP (°C)	V_{OUT} (mV)	TEMP (°C)	V_{OUT} (mV)	TEMP (°C)	V_{OUT} (mV)
–34	3082	6	2553	46	2012	86	1455	126	886
–33	3069	7	2540	47	1999	87	1441	127	872
–32	3056	8	2527	48	1985	88	1427	128	858
–31	3043	9	2513	49	1971	89	1413	129	843
–30	3030	10	2500	50	1958	90	1399	130	829
–29	3017	11	2486	51	1944	91	1385	131	814
–28	3004	12	2473	52	1930	92	1371	132	800
–27	2991	13	2459	53	1916	93	1356	133	786
–26	2978	14	2446	54	1902	94	1342	134	771
–25	2965	15	2433	55	1888	95	1328	135	757
–24	2952	16	2419	56	1875	96	1314	136	742
–23	2938	17	2406	57	1861	97	1300	137	728
–22	2925	18	2392	58	1847	98	1286	138	713
–21	2912	19	2379	59	1833	99	1272	139	699
–20	2899	20	2365	60	1819	100	1257	140	684
–19	2886	21	2352	61	1805	101	1243	141	670
–18	2873	22	2338	62	1791	102	1229	142	655
–17	2859	23	2325	63	1777	103	1215	143	640
–16	2846	24	2311	64	1763	104	1201	144	626
–15	2833	25	2298	65	1749	105	1186	145	611
–14	2820	26	2285	66	1735	106	1172	146	597
–13	2807	27	2271	67	1721	107	1158	147	582
–12	2793	28	2258	68	1707	108	1144	148	568
–11	2780	29	2244	69	1693	109	1130	149	553
								150	538

Although the LMT87-Q1 is very linear, the response does have a slight umbrella parabolic shape. 表 8-1 very accurately reflects this shape. The transfer table can be calculated by using the parabolic equation (式 1).

$$V_{TEMP}(mV) = 2230.8mV - \left[13.582\frac{mV}{°C}(T - 30°C)\right] - \left[0.00433\frac{mV}{°C^2}(T - 30°C)^2\right] \quad (1)$$

The parabolic equation is an approximation of the transfer table and the accuracy of the equation degrades slightly at the temperature range extremes. 式 1 can be solved for T resulting in:

$$T = \frac{13.582 - \sqrt{(-13.582)^2 + 4 \times 0.00433 \times (2230.8 - V_{TEMP}(mV))}}{2 \times (-0.00433)} + 30 \quad (2)$$

For an even less accurate linear transfer function approximation, a line can easily be calculated over the desired temperature range from 表 8-1 using the two-point equation (式 3):

$$V - V_1 = \left(\frac{V_2 - V_1}{T_2 - T_1}\right) \times (T - T_1) \quad (3)$$

where

- V is in mV,
- T is in °C,
- T_1 and V_1 are the coordinates of the lowest temperature,
- and T_2 and V_2 are the coordinates of the highest temperature.

For example, if the user wanted to resolve this equation, over a temperature range of 20°C to 50°C, they would proceed as follows:

$$V - 2365\ mV = \left(\frac{1958\ mV - 2365\ mV}{50^{\circ}C - 20^{\circ}C}\right) \times (T - 20^{\circ}C) \tag{4}$$

$$V - 2365\ mV = (-13.6\ mV / {}^{\circ}C) \times (T - 20^{\circ}C) \tag{5}$$

$$V = (-13.6\ mV / {}^{\circ}C) \times T + 2637\ mV \tag{6}$$

Using this method of linear approximation, the transfer function can be approximated for one or more temperature ranges of interest.

8.4 Device Functional Modes

8.4.1 Mounting and Thermal Conductivity

The LMT87-Q1 can be applied easily in the same way as other integrated-circuit temperature sensors. It can be glued or cemented to a surface.

To ensure good thermal conductivity, the backside of the LMT87-Q1 die is directly attached to the GND pin. The temperatures of the lands and traces to the other leads of the LMT87-Q1 will also affect the temperature reading.

Alternatively, the LMT87-Q1 can be mounted inside a sealed-end metal tube, and can then be dipped into a bath or screwed into a threaded hole in a tank. As with any IC, the LMT87-Q1 and accompanying wiring and circuits must be kept insulated and dry, to avoid leakage and corrosion. This is especially true if the circuit may operate at cold temperatures where condensation can occur. If moisture creates a short circuit from the output to ground or V_{DD}, the output from the LMT87-Q1 will not be correct. Printed-circuit coatings are often used to ensure that moisture cannot corrode the leads or circuit traces.

The thermal resistance junction to ambient ($R_{\theta JA}$ or θ_{JA}) is the parameter used to calculate the rise of a device junction temperature due to its power dissipation. Use 式 7 to calculate the rise in the LMT87-Q1 die temperature:

$$T_J = T_A + \theta_{JA}\left[(V_{DD}I_S) + (V_{DD} - V_{OUT})\, I_L\right] \tag{7}$$

where

- T_A is the ambient temperature,
- I_S is the supply current,
- $I_L$is the load current on the output,
- and V_O is the output voltage.

For example, in an application where T_A = 30°C, V_{DD} = 5 V, I_S = 5.4 µA, V_{OUT} = 2231 mV, and I_L = 2 µA, the junction temperature would be 30.014°C, showing a self-heating error of only 0.014°C. Because the junction temperature of the LMT87-Q1 is the actual temperature being measured, take care to minimize the load current that the LMT87-Q1 is required to drive. The *Thermal Information* table shows the thermal resistance of the LMT87-Q1.

8.4.2 Output Noise Considerations

A push-pull output gives the LMT87-Q1 the ability to sink and source significant current. This is beneficial when, for example, driving dynamic loads like an input stage on an analog-to-digital converter (ADC). In these applications the source current is required to quickly charge the input capacitor of the ADC. The LMT87-Q1 is ideal for this and other applications which require strong source or sink current.

The LMT87-Q1 supply-noise gain (the ratio of the AC signal on V_{OUT} to the AC signal on V_{DD}) was measured during bench tests. 図 7-8 shows the typical attenuation found in the *Typical Characteristics* section. A load capacitor on the output can help to filter noise.

For operation in very noisy environments, some bypass capacitance should be present on the supply within approximately 5 centimeters of the LMT87-Q1.

8.4.3 Capacitive Loads

The LMT87-Q1 handles capacitive loading well. In an extremely noisy environment, or when driving a switched sampling input on an ADC, it may be necessary to add some filtering to minimize noise coupling. Without any precautions, 図 8-1 shows how the LMT87-Q1 can drive a capacitive load less than or equal to 1100 pF. For capacitive loads greater than 1100 pF, 図 8-2 shows how a series resistor may be required on the output.

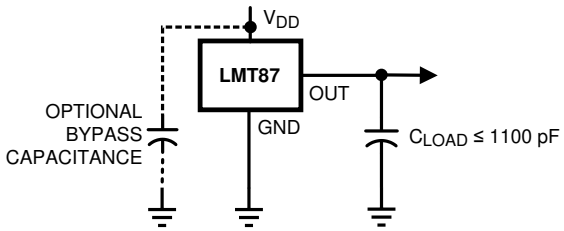


図 8-1. LMT87 No Decoupling Required for Capacitive Loads Less Than 1100 pF

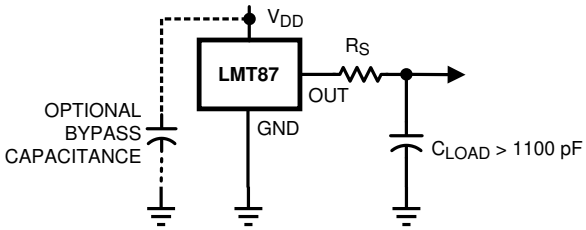


図 8-2. LMT87 with Series Resistor for Capacitive Loading Greater Than 1100 pF

表 8-2. Recommended Series Resistor Values

C_{LOAD}	MINIMUM R_S
1.1 nF to 99 nF	3 kΩ
100 nF to 999 nF	1.5 kΩ
1 µF	800 Ω

8.4.4 Output Voltage Shift

The LMT87-Q1 is very linear over temperature and supply voltage range. Due to the intrinsic behavior of an NMOS/PMOS rail-to-rail buffer, a slight shift in the output can occur when the supply voltage is ramped over the operating range of the device. The location of the shift is determined by the relative levels of V_{DD} and V_{OUT}. The shift typically occurs when V_{DD}- V_{OUT} = 1 V.

This slight shift (a few millivolts) takes place over a wide change (approximately 200 mV) in V_{DD} or V_{OUT}. Because the shift takes place over a wide temperature change of 5°C to 20°C, V_{OUT} is always monotonic. The accuracy specifications in the *Accuracy Characteristics* table already include this possible shift.

9 Application and Implementation

注

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9.1 Application Information

The LMT87-Q1 features make it suitable for many general temperature-sensing applications. It can operate down to 2.7-V supply with 5.4-µA power consumption.

9.2 Typical Applications

9.2.1 Connection to ADC

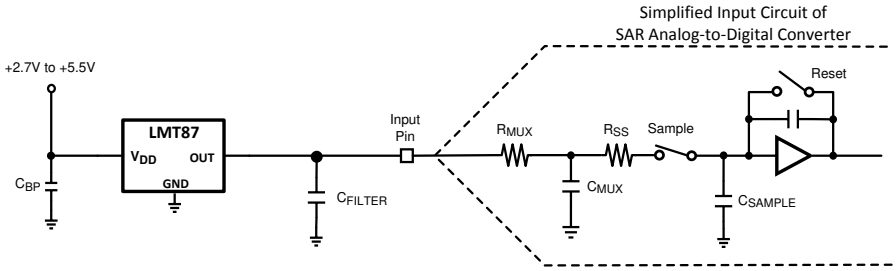


図 9-1. Suggested Connection to a Sampling Analog-to-Digital Converter Input Stage

9.2.1.1 Design Requirements

Most CMOS ADCs found in microcontrollers and ASICs have a sampled data comparator input structure. When the ADC charges the sampling cap, it requires instantaneous charge from the output of the analog source such as the LMT87-Q1 temperature sensor and many op amps. This requirement is easily accommodated by the addition of a capacitor (C_{FILTER}).

9.2.1.2 Detailed Design Procedure

The size of C_{FILTER} depends on the size of the sampling capacitor and the sampling frequency. Because not all ADCs have identical input stages, the charge requirements will vary. This general ADC application is shown as an example only.

9.2.1.3 Application Curve

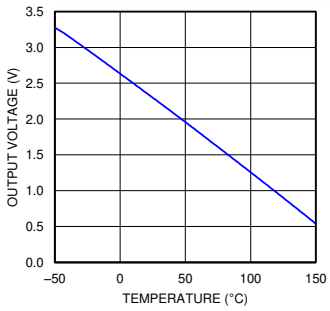


図 9-2. Analog Output Transfer Function

9.2.2 Conserving Power Dissipation With Shutdown

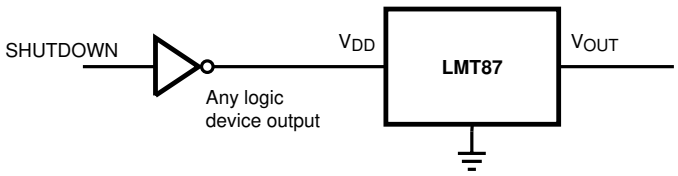


図 9-3. Simple Shutdown Connection of the LMT87-Q1

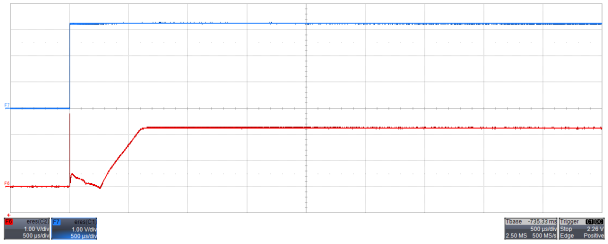
9.2.2.1 Design Requirements

Because the power consumption of the LMT87-Q1 is less than 9 µA, it can simply be powered directly from any logic gate output and therefore not require a specific shutdown pin. The device can even be powered directly from a microcontroller GPIO. In this way, it can easily be turned off for cases such as battery-powered systems where power savings are critical.

9.2.2.2 Detailed Design Procedure

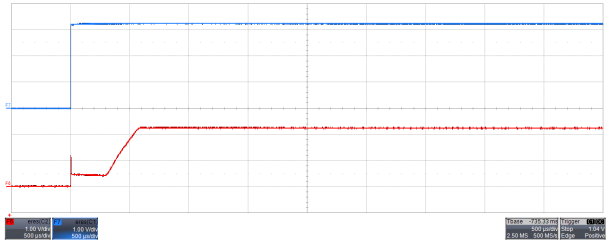
Simply connect the V_{DD} pin of the LMT87-Q1 directly to the logic shutdown signal from a microcontroller.

9.2.2.3 Application Curves



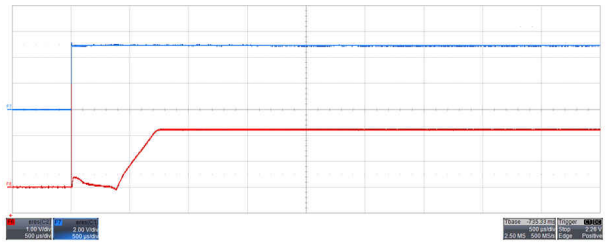
Time: 500 µs/div; Top trace: V_{DD} 1 V/div; Bottom trace: OUT 1 V/div

図 9-4. Output Turnon Response Time Without a Capacitive Load and V_{DD} = 3.3 V



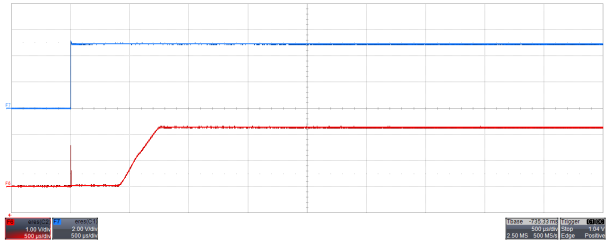
Time: 500 µs/div; Top trace: V_{DD} 1 V/div; Bottom trace: OUT 1 V/div

図 9-5. Output Turnon Response Time With a 1.1-nF Capacitive Load and V_{DD} = 3.3 V



Time: 500 µs/div; Top trace: V_{DD} 2 V/div; Bottom trace: OUT 1 V/div

図 9-6. Output Turnon Response Time Without a Capacitive Load and V_{DD} = 5 V



Time: 500 µs/div; Top trace: V_{DD} 2 V/div; Bottom trace: OUT 1 V/div

図 9-7. Output Turnon Response Time With a 1.1-nF Capacitive Load and V_{DD} = 5 V

10 Power Supply Recommendations

The low supply current and supply range (2.7 V to 5.5 V) of the LMT87-Q1 allow the device to easily be powered from many sources. Power supply bypassing is optional and is mainly dependent on the noise on the power supply used. In noisy systems it may be necessary to add bypass capacitors to lower the noise that is coupled to the output of the LMT87-Q1.

11 Layout

11.1 Layout Guidelines

The LMT87-Q1 is extremely simple to layout. If a power-supply bypass capacitor is used, the *Layout Example* shows how to connect the capacitor to the device.

11.2 Layout Example

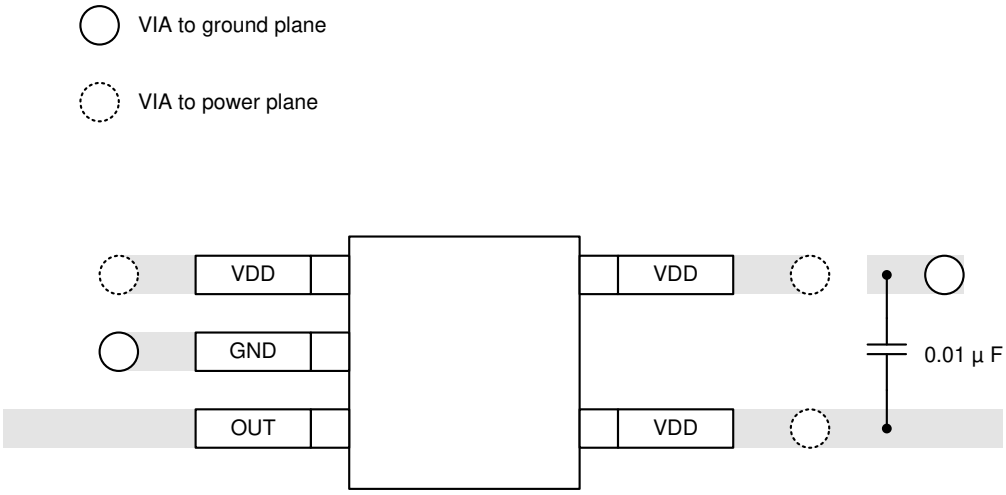


図 11-1. SC70 Package Recommended Layout

12 Device and Documentation Support

12.1 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Subscribe to updates* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

12.2 サポート・リソース

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12.4 Electrostatic Discharge Caution

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

12.5 Glossary

TI Glossary	This glossary lists and explains terms, acronyms, and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LMT87QDCKRQ1	ACTIVE	SC70	DCK	5	3000	RoHS & Green	SN	Level-1-260C-UNLIM	-50 to 150	BVA	Samples

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".
RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.
Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

OTHER QUALIFIED VERSIONS OF LMT87-Q1 :

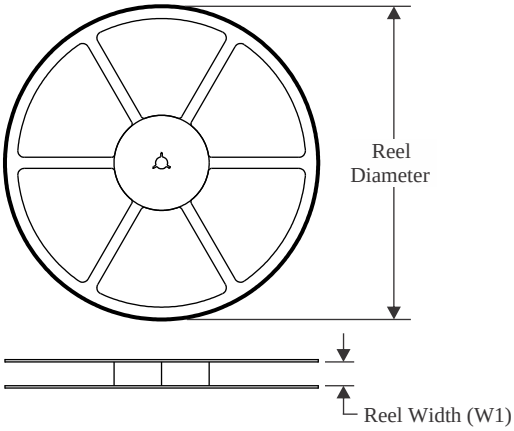
- Catalog : LMT87

NOTE: Qualified Version Definitions:

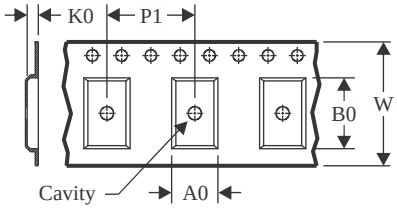
- Catalog - TI's standard catalog product

TAPE AND REEL INFORMATION

REEL DIMENSIONS

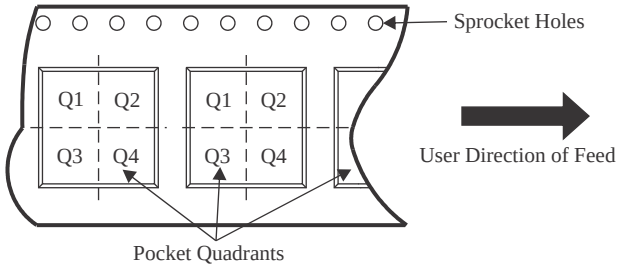


TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

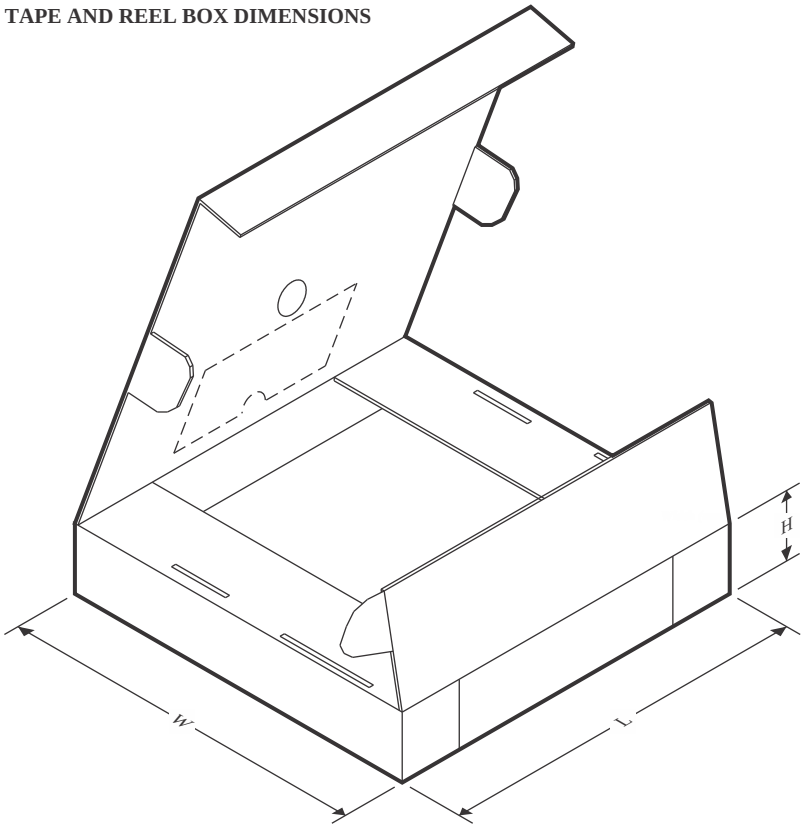
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMT87QDCKRQ1	SC70	DCK	5	3000	178.0	8.4	2.25	2.45	1.2	4.0	8.0	Q3

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

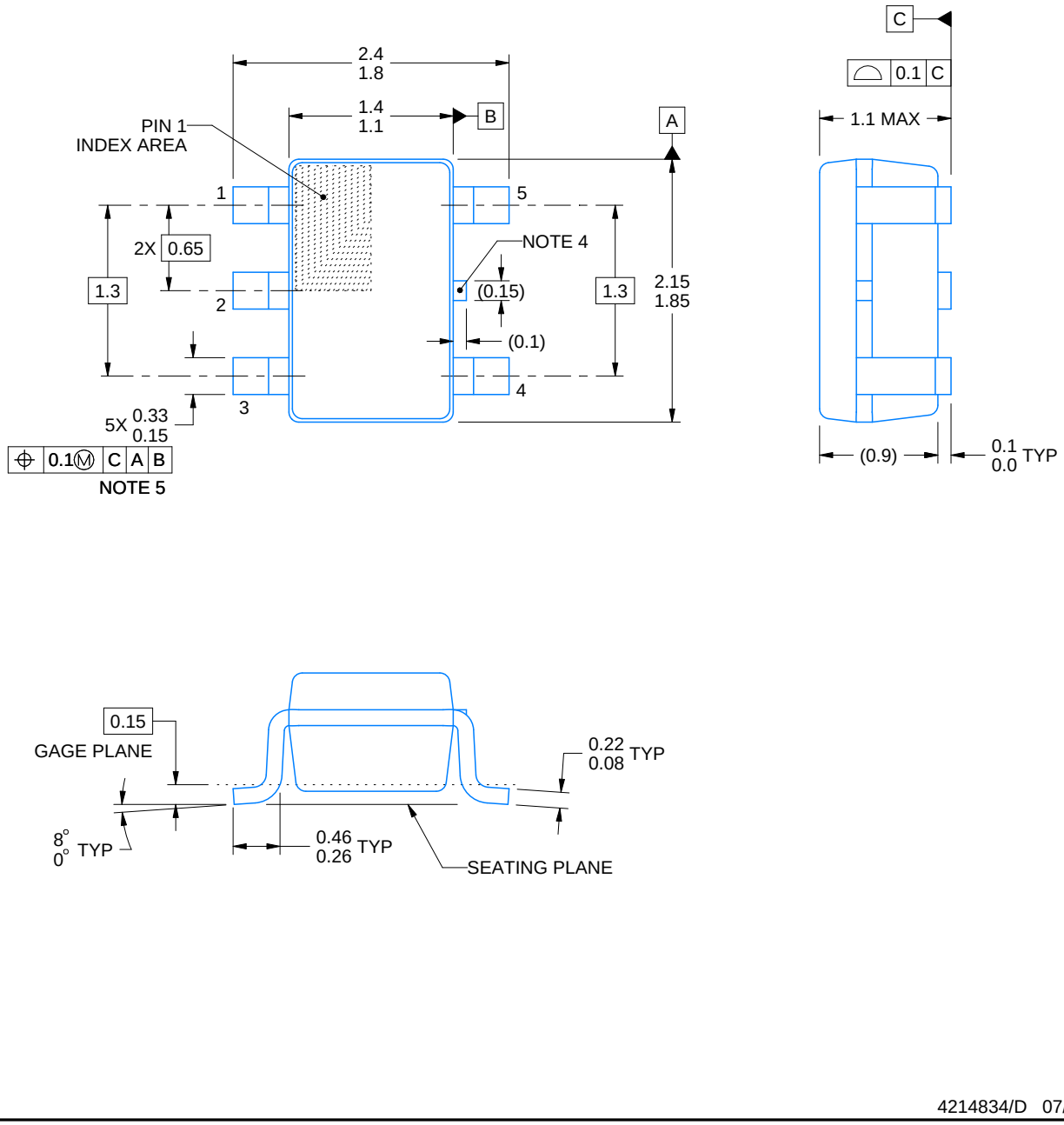
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMT87QDCKRQ1	SC70	DCK	5	3000	208.0	191.0	35.0

PACKAGE OUTLINE

DCK0005A

SOT - 1.1 max height

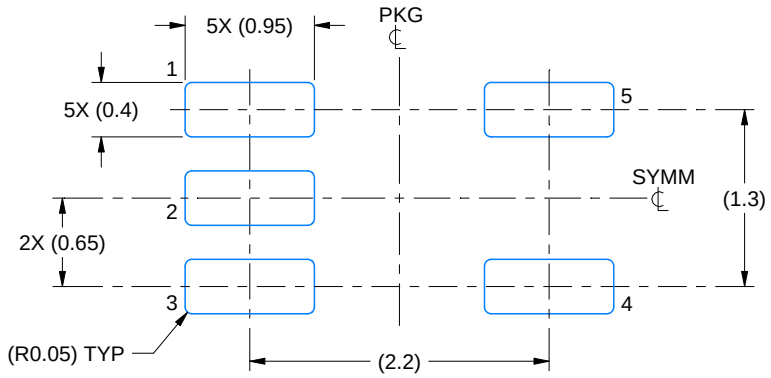
SMALL OUTLINE TRANSISTOR



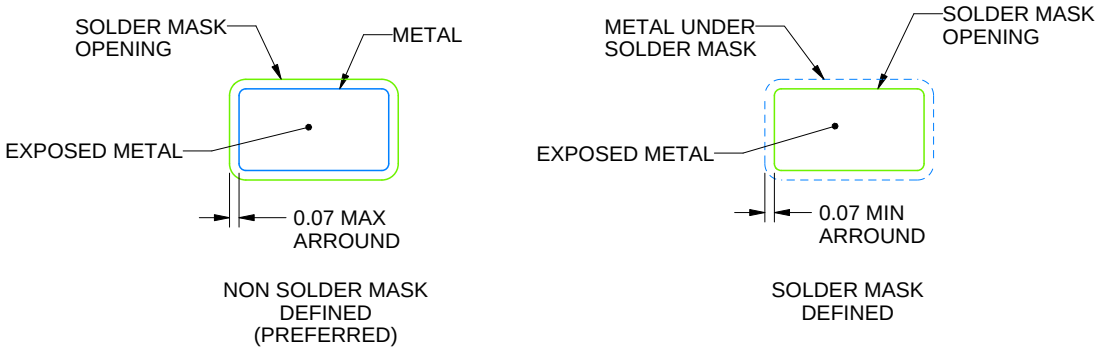
4214834/D 07/2023

NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. Refernce JEDEC MO-203.
4. Support pin may differ or may not be present.
5. Lead width does not comply with JEDEC.



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:18X

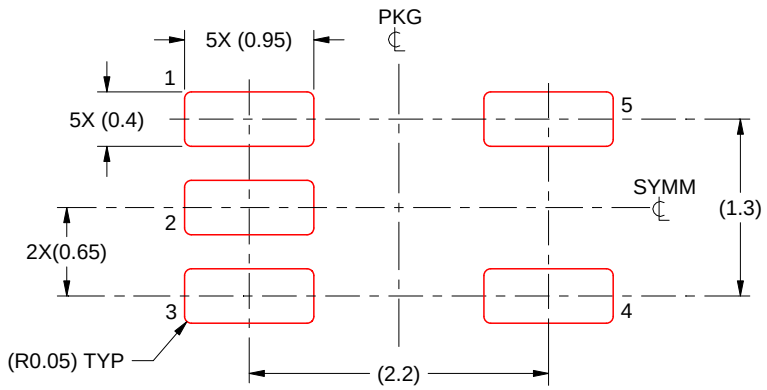


SOLDER MASK DETAILS

4214834/D 07/2023

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SOLDER PASTE EXAMPLE
BASED ON 0.125 THICK STENCIL
SCALE:18X

4214834/D 07/2023

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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