CC1354R10SimpleLink™ 高性能マルチバンド ワイヤレス マイコン

1 特長

ワイヤレス マイクロコントローラ

- 強力な 48MHz Arm® Cortex®-M33 プロセッサと TrustZone® の組み合わせ
- FPU と DSP の拡張機能
- 1024kB フラッシュ プログラム メモリ
- 8kB のキャッシュ SRAM
- 高信頼性動作に適した 256kB の超低リーク SRAM (パリティ付き)
  - パリティが無効になっている場合は 32kB の SRAM を追加可能
- デュアル バンド Sub-1GHz および 2.4GHz 動作
- ダイナミック マルチプロトコル マネージャ (DMM) ドライバ
- プログラマブルな無線機能には、2-(G)FSK、4-(G)FSK、MSK、OOK、IEEE 802.15.4 PHY、MAC のサポートが含まれます
- OTA (Over-the-air) 更新をサポート

超低消費電力センサ コントローラ

- 4kB の SRAM を備えた自律型 MCU
- センサ データのサンプリング、保存、処理
- 高速ウェークアップによる低消費電力動作
- ソフトウェア定義ペリフェラル、静電容量式タッチ、流量計、LCD

低消費電力

- MCU の消費電流:
  - 3.4mA のアクティブ モード電流、CoreMark®
  - 71µA/MHz (CoreMark® 実行中)
  - 0.98µA (スタンバイ モード、RTC、256kB SRAM)
  - 0.17µA (シャットダウン モード、ウェークアップ オンピン)
- 超低消費電力センサ コントローラの消費電流
  - 2MHz モードで 32µA
  - 24MHz モードで 849µA
- 無線の消費電流:
  - 868MHz で 5.8mA RX
  - 2.4GHz で 6.9mA RX
  - 868MHz、+14dBm で 25.8mA TX
  - 2.4GHz、0dBm で 7.1mA TX

無線プロトコルのサポート

- Thread、Zigbee®、Matter
- Bluetooth® 5.3 Low Energy
- Wi-SUN®
- mioty®
- Amazon Sidewalk
- ワイヤレス M-Bus
- SimpleLink™ TI 15.4 スタック (Sub-1GHz)
- 6LoWPAN
- 独自システム

高性能の無線

- - 121dBm (2.5kbps、長距離モード)
- - 110dBm (50kbps、802.15.4、868MHz)
- –104dBm (Bluetooth® Low Energy、125kbps)
- –105dBm (IEEE 802.15.4-2006 2.4GHz OQPSK、コヒーレント モデム)

法規制の順守

- 以下の規格への準拠を目的としたシステムに最適:
  - ETSI EN 300 220 Receiver Cat.1.5 および 2、EN 300 328、EN 303 131、EN 303 204、EN 300 440 Cat.2 および 3
  - FCC CFR47 Part 15
  - ARIB STD-T66、STD-T67、STD-T108

MCU 周辺機器

- ほとんどのデジタル ペリフェラルを任意の GPIO に配線可能
- 4 つの 32 ビットまたは 8 つの 16 ビット汎用タイマ
- 12 ビット SAR ADC、200ksps、8 チャネル
- 8 ビット DAC
- 2 つのコンパレータ
- プログラマブル電流ソース
- 4 つの UART、4 つの SPI、2 つの I^2C、1 つの I^2S
- リアルタイム クロック (RTC)
- 温度およびバッテリ モニタを内蔵

セキュリティ イネーブラ (実現機能)

- セキュア ブートのサポート
- セキュア キー ストレージとデバイス ID をサポート
- 信頼できる実行環境に対応する Arm® TrustZone®
- AES 128 および 256 ビット暗号化アクセラレータ
- 公開鍵アクセラレータ
- SHA2 アクセラレータ (SHA-512 までのフル スイート)
- 真性乱数生成器 (TRNG)
- セキュア デバッグ ロック
- ソフトウェア アンチロールバック保護

開発ツールとソフトウェア

- 868MHz または 915MHz でのデュアルバンドおよび +20dBm 出力電力用の LP-EM-CC1354P10-1
- 2.4GHz でのデュアルバンドおよび +10dBm 出力電力用の LP-EM-CC1354P10-6

このリソースの元の言語は英語です。翻訳は概要を便宜的に提供するもので、自動化ツール (機械翻訳) を使用していることがあり、TI では翻訳の正確性および妥当性につきましては一切保証いたしません。実際の設計などの前には、ti.com で必ず最新の英語版をご参照くださいますようお願いいたします。

- LP-XDS110、LP-XDS110ET または TMDSEMU110-U (TMDSEMU110-ETH アドオン使用) デバッグ プローブ
- SimpleLink™ LOWPOWER F2 ソフトウェア開発キット (SDK)
- SmartRF™ Studio による簡素な無線構成
- Sensor Controller Studio により低消費電力のセンシング アプリケーションを構築
- SysConfig システム コンフィギュレーション ツール

動作範囲

- オンチップの降圧型 DC/DC コンバータ
- 1.8V～3.8V のシングル電源電圧
- -40°C～+105°C

パッケージ

- 7mm × 7mm RGZ VQFN48 (GPIO 28 本)
- 8mm × 8mm RSK VQFN64 (GPIO 42 本)
- RoHS 準拠のパッケージ

2 アプリケーション

- 315MHz、433MHz、470MHz から 510MHz、868MHz、902MHz から 928MHz、2400 から 2480MHzMHz の ISM および SRD システム[1]
  (最小 4kHz の受信帯域幅)
- ビル オートメーション
  - ビルディングのセキュリティ システム – モーション検出器、電子スマート ロック、ドアおよび窓センサ、ガレージ ドア システム、ゲートウェイ
  - HVAC – サーモスタット、ワイヤレス環境センサ、HVAC システム コントローラ、ゲートウェイ
  - 防火システム - 煙および熱感知器、火災警報制御パネル (FACP)
  - ビデオ監視 – IP ネットワーク カメラ
  - エレベータとエスカレータ – エレベータとエスカレータのエレベータ メイン制御パネル
- グリッド インフラ
  - スマート メータ – 水道メータ、ガス メータ、電気メータ、ヒート コスト アロケータ
  - グリッド通信 – 無線通信と長距離センサ アプリケーション
  - その他の代替エネルギー - 環境発電
- 産業用輸送 - アセット トラッキング
- ファクトリ オートメーション / 制御
- 医療用
- 通信機器
  - 有線ネットワーク – ワイヤレス LAN または Wi-Fi アクセス ポイント、エッジ ルータ
- パーソナル エレクトロニクス
  - ホーム シアターおよびエンターテインメント – スマート スピーカ、スマート ディスプレイ、セット トップ ボックス
  - ウェアラブル (医療以外) – スマート トラッカー、スマート衣料

3 概要

SimpleLink™ CC1354R10 デバイスはマルチプロトコル、マルチバンドの Sub-1GHz および 2.4GHz ワイヤレス マイコン (MCU) です。本デバイスは Thread、Zigbee、Bluetooth 5.3 Low Energy、IEEE 802.15.4g、IPv6 対応スマート オブジェクト (6LoWPAN)、mioty、Wi-SUN、Amazon Sidewalk、TI 15.4 スタック (Sub-1GHz および 2.4GHz) を含む独自システム、DMM (ダイナミック マルチプロトコル マネージャ) ドライバを使った同時マルチプロトコルをサポートしています。本デバイスは低消費電力のワイヤレス通信向けに最適化されており、高度なセキュリティ機能とオンチップのワイヤレス (OAD) 更新機能を備えています。本デバイスは、ビルディングのセキュリティ システム、HVAC、スマート メータ、医療、有線ネットワーク、携帯型電子機器、ホーム シアター & エンターテインメント、ネットワーク接続周辺機器市場向けの長距離かつ信頼性の高い通信を可能にします。このデバイスの主な特長を以下に示します。

- Arm® TrustZone® ベースのセキュア キー ストレージ、デバイス ID、トラステッド機能をサポート。
- DMM ドライバを使用して、Sub-1GHz と 2.4GHz の両方の同時マルチプロトコルをサポートするマルチバンド デバイス。
- SimpleLink LOWPOWER F2 ソフトウェア開発キット (SDK) で幅広いプロトコル スタックを柔軟にサポート。
- 0.98 µA の小さいスタンバイ電流 (全 SRAM 保持) により、ワイヤレス アプリケーションのバッテリ駆動時間を延長。
- 産業用温度に対応し、85℃で 5µA の最小スタンバイ電流。
- 高速ウェークアップ機能を備えたプログラマブルな自律型超低消費電力センサ コントローラ CPU による先進センシング。たとえば、このセンサ コントローラは、1µA のシステム電流で 1Hz の ADC サンプリングが可能です。

[1] サポートしているプロトコル規格、変調フォーマット、データ レートの詳細については、RF コアを参照してください。

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- 潜在的な放射線イベントによるデータ破損を防止する常時オン SRAM パリティを備え、低い SER (ソフト エラー レート) FIT (Failure-In-Time、故障率) により、産業用市場向けに中断のない長い動作寿命を実現。
- 柔軟性の高い低消費電力 RF トランシーバ機能を備えた専用のソフトウェア制御無線コントローラ (Arm® Cortex®-M0) により、複数の物理層および RF 規格をサポート。
- SimpleLink™ 長距離モードに対応する優れた無線感度 (-121dBm) と堅牢 (選択度、ブロッキング) 性能。

この CC1354R10 デバイスは、SimpleLink™ MCU プラットフォームの一部です。本プラットフォームは、単一のコア SDK (ソフトウェア開発キット) と豊富なツール セットを備えた使いやすい共通の開発環境を共有する Wi-Fi®、Bluetooth® Low Energy、Thread、Zigbee®、Sub-1GHz MCU、およびホスト MCU で構成されています。SimpleLink™ プラットフォームの統合を 1 回行うだけで、どのようなデバイスの組み合わせでも、ポートフォリオから自分の設計に追加でき、設計の要件が変化した場合でも、コードを 100% 再利用することができます。詳細については、SimpleLink マイコン プラットフォームをご覧ください。

ソフトウェアの互換性に加えて、マルチバンド ワイヤレス MCU は 7mm × 7mm の QFN パッケージに封止済みで、352kB のフラッシュから 1MB のフラッシュまでのピン互換性があるので、設計のスケーラビリティを最大限に高めることができます。テキサス・インスツルメンツの Sub-1GHz デバイスの詳細については、www.ti.com/sub1ghz をご覧ください 。

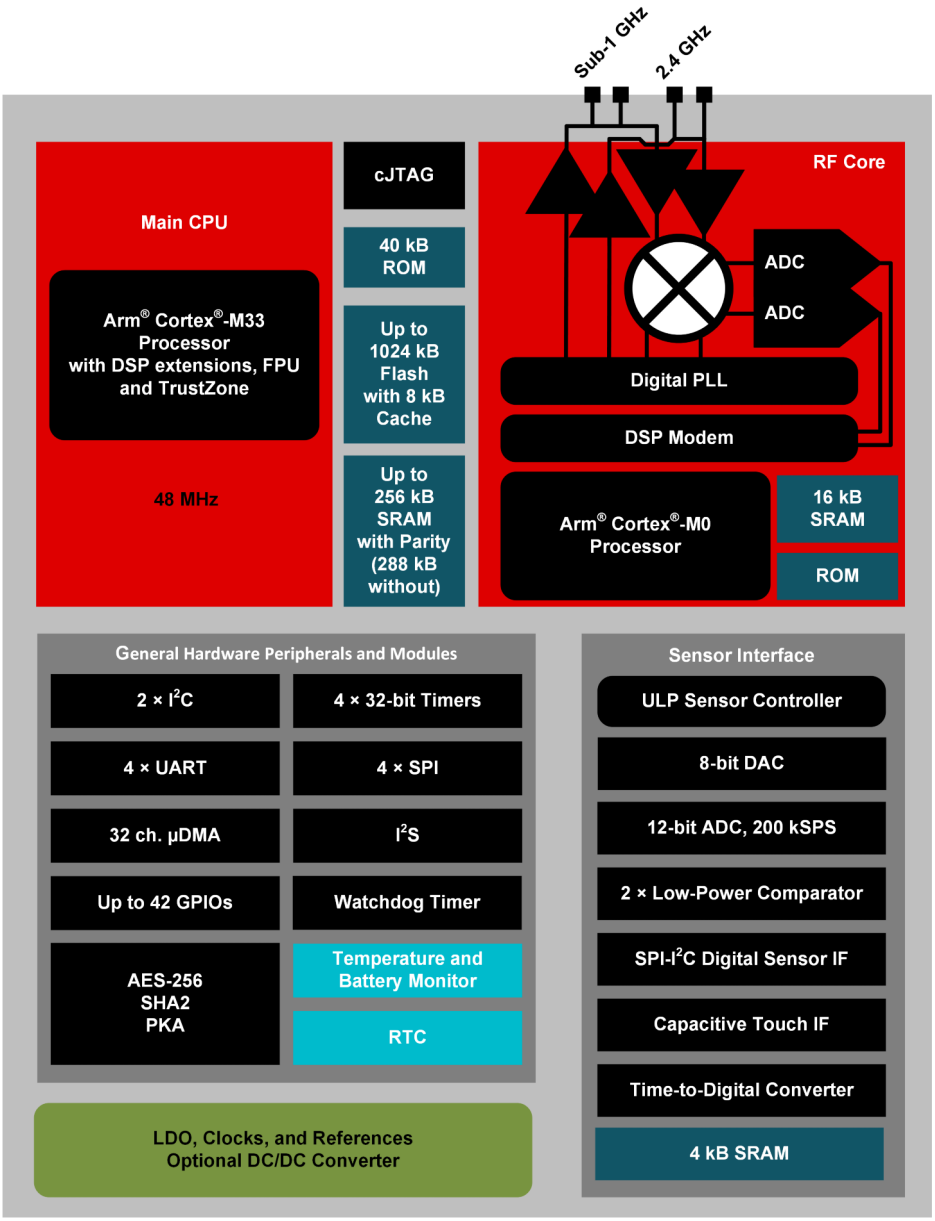
製品情報

部品番号	パッケージ(1)	パッケージ サイズ
CC1354R106T0RGZ	VQFN (48)	7.00mm × 7.00mm
CC1354R106T0RSK	VQFN (64)	8.00mm × 8.00mm

(1) 詳細については、「メカニカル、パッケージ、および注文情報」を参照してください。

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4 機能ブロック図



CC1354R10 ブロック図

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5 Device Comparison

Device	RADIO SUPPORT											FLASH (kB)	RAM + Cache (kB)	GPIO	PACKAGE SIZE					
	Sub-1GHz Prop.	2.4GHz Prop.	Wireless M-Bus	mioty	Wi-SUN®	Sidewalk	*Bluetooth*® LE	Zigbee	Thread	Multiprotocol	+20dBm PA				4 × 4 mm VQFN (24)	4 × 4 mm VQFN (32)	5 × 5 mm VQFN (32)	5 × 5 mm VQFN (40)	7 × 7 mm VQFN (48)	8 × 8 mm VQFN (64)
CC1310	√		√	√								32-128	16-20 + 8	10-30		√	√		√	
CC1311R3	√		√	√								352	32 + 8	22-30				√	√	
CC1311P3	√		√	√							√	352	32 + 8	26					√	
CC1312R	√		√	√	√							352	80 + 8	30					√	
CC1312R7	√		√	√	√	√				√		704	144 + 8	30					√	
CC1314R10	√		√	√	√	√				√		1024	256 + 8	30-46					√	√
CC1352R	√	√	√	√	√		√	√	√	√		352	80 + 8	28					√	
CC1354R10	√	√	√	√	√		√	√	√	√		1024	256 + 8	28-42					√	√
CC1352P	√	√	√	√	√		√	√	√	√	√	352	80 + 8	26					√	
CC1352P7	√	√	√	√	√	√	√	√	√	√	√	704	144 + 8	26					√	
CC1354P10	√	√	√	√	√	√	√	√	√	√	√	1024	256 + 8	26-42					√	√
CC2340R2 (1)		√					√	√				256	28	12	√					
CC2340R5 (2)		√					√	√	√			512	36	12-26	√			√		
CC2340R5-Q1							√					512	36	19			√			
CC2640R2F							√					128	20 + 8	10-31		√	√		√	
CC2642R							√					352	80 + 8	31					√	
CC2642R-Q1							√					352	80 + 8	31					√	
CC2651R3		√					√	√				352	32 + 8	23-31				√	√	
CC2651P3		√					√	√			√	352	32 + 8	22-26				√	√	
CC2652R		√					√	√	√	√		352	80 + 8	31					√	
CC2652RB		√					√	√	√	√		352	80 + 8	31					√	
CC2652R7		√					√	√	√	√		704	144 + 8	31					√	
CC2652P		√					√	√	√	√	√	352	80 + 8	26					√	
CC2652P7		√					√	√	√	√	√	704	144 + 8	26					√	
CC2674R10		√					√	√	√	√		1024	256 + 8	31-45					√	√
CC2674P10		√					√	√	√	√	√	1024	256 + 8	26-45					√	√

(1) Zigbee and Proprietary RF support enabled by a future software update
(2) Zigbee and Thread support enabled by a future software update

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6 Pin Configuration and Functions

6.1 Pin Diagram—RGZ Package (Top View)

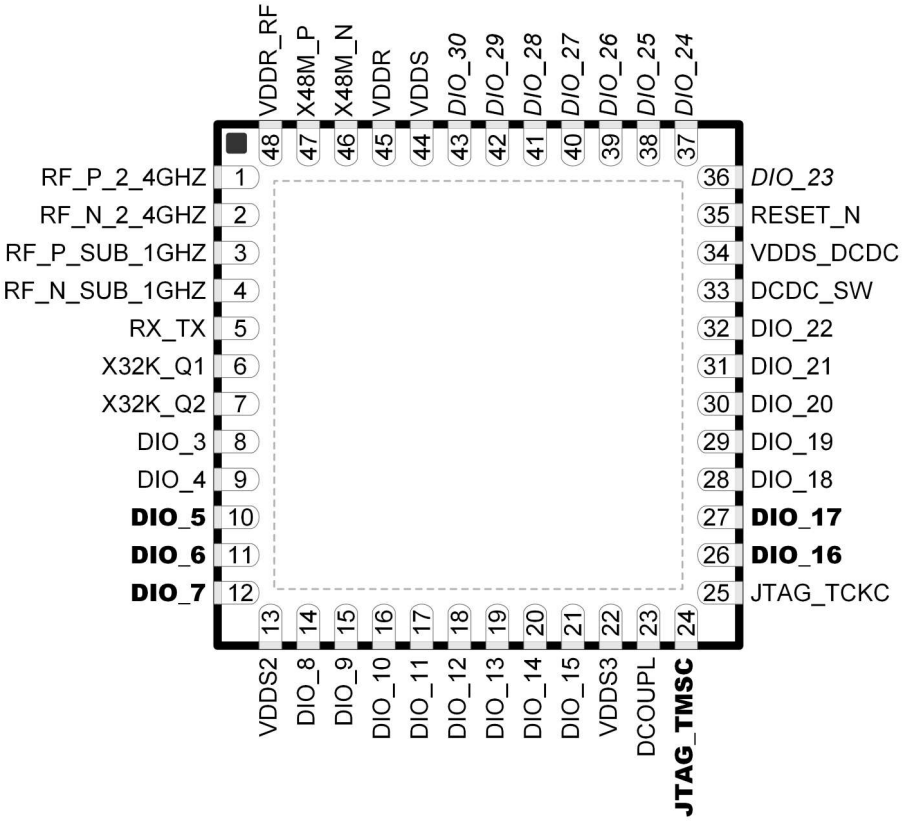


図 6-1. RGZ (7mm × 7mm) Pinout, 0.5mm Pitch (Top View)

The following I/O pins marked in 図 6-1 in **bold** have high-drive capabilities:

- Pin 10, DIO_5
- Pin 11, DIO_6
- Pin 12, DIO_7
- Pin 24, JTAG_TMSC
- Pin 26, DIO_16
- Pin 27, DIO_17

The following I/O pins marked in 図 6-1 in *italics* have analog capabilities:

- Pin 36, DIO_23
- Pin 37, DIO_24
- Pin 38, DIO_25
- Pin 39, DIO_26
- Pin 40, DIO_27
- Pin 41, DIO_28
- Pin 42, DIO_29
- Pin 43, DIO_30

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6.2 Signal Descriptions—RGZ Package

表 6-1. Signal Descriptions—RGZ Package

PIN		I/O	TYPE	DESCRIPTION
NAME	NO.			
DCDC_SW	33	—	Power	Output from internal DC/DC converter(1)
DCOUPL	23	—	Power	For decoupling of internal 1.27V regulated digital-supply (2)
DIO_3	8	I/O	Digital	GPIO
DIO_4	9	I/O	Digital	GPIO
DIO_5	10	I/O	Digital	GPIO, high-drive capability
DIO_6	11	I/O	Digital	GPIO, high-drive capability
DIO_7	12	I/O	Digital	GPIO, high-drive capability
DIO_8	14	I/O	Digital	GPIO
DIO_9	15	I/O	Digital	GPIO
DIO_10	16	I/O	Digital	GPIO
DIO_11	17	I/O	Digital	GPIO
DIO_12	18	I/O	Digital	GPIO
DIO_13	19	I/O	Digital	GPIO
DIO_14	20	I/O	Digital	GPIO
DIO_15	21	I/O	Digital	GPIO
DIO_16	26	I/O	Digital	GPIO, JTAG_TDO, high-drive capability
DIO_17	27	I/O	Digital	GPIO, JTAG_TDI, high-drive capability
DIO_18	28	I/O	Digital	GPIO
DIO_19	29	I/O	Digital	GPIO
DIO_20	30	I/O	Digital	GPIO
DIO_21	31	I/O	Digital	GPIO
DIO_22	32	I/O	Digital	GPIO
DIO_23	36	I/O	Digital or Analog	GPIO, analog capability
DIO_24	37	I/O	Digital or Analog	GPIO, analog capability
DIO_25	38	I/O	Digital or Analog	GPIO, analog capability
DIO_26	39	I/O	Digital or Analog	GPIO, analog capability
DIO_27	40	I/O	Digital or Analog	GPIO, analog capability
DIO_28	41	I/O	Digital or Analog	GPIO, analog capability
DIO_29	42	I/O	Digital or Analog	GPIO, analog capability
DIO_30	43	I/O	Digital or Analog	GPIO, analog capability
EGP	—	—	GND	Ground—exposed ground pad(3)
JTAG_TMSC	24	I/O	Digital	JTAG TMSC, high-drive capability
JTAG_TCKC	25	I	Digital	JTAG TCKC
RESET_N	35	I	Digital	Reset, active low. No internal pullup resistor
RF_P_SUB_1GHZ	3	—	RF	Positive Sub-1GHz RF input signal to LNA during RX Positive Sub-1GHz RF output signal from PA during TX
RF_N_SUB_1GHZ	4	—	RF	Negative Sub-1GHz RF input signal to LNA during RX Negative Sub-1GHz RF output signal from PA during TX
RX_TX	5	—	RF	Optional bias pin for the RF LNA
VDDR	45	—	Power	Internal supply, must be powered from the internal DC/DC converter or the internal LDO(2) (4) (6)
VDDR_RF	48	—	Power	Internal supply, must be powered from the internal DC/DC converter or the internal LDO(2) (5) (6)

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表 6-1. Signal Descriptions—RGZ Package (続き)

PIN		I/O	TYPE	DESCRIPTION
NAME	NO.			
VDDS	44	—	Power	1.8V to 3.8V main chip supply(1)
VDDS2	13	—	Power	1.8V to 3.8V DIO supply(1)
VDDS3	22	—	Power	1.8V to 3.8V DIO supply(1)
VDDS_DCDC	34	—	Power	1.8V to 3.8V DC/DC converter supply
X48M_N	46	—	Analog	48MHz crystal oscillator pin N
X48M_P	47	—	Analog	48MHz crystal oscillator pin P
X32K_Q1	6	—	Analog	32kHz crystal oscillator pin 1
X32K_Q2	7	—	Analog	32kHz crystal oscillator pin 2

(1) For more details, see the technical reference manual listed in セクション 10.2.
(2) Do not supply external circuitry from this pin.
(3) EGP is the only ground connection for the device. A good electrical connection to the device ground on a printed circuit board (PCB) is imperative for proper device operation.
(4) If an internal DC/DC converter is not used, this pin is supplied internally from the main LDO.
(5) If an internal DC/DC converter is not used, this pin must be connected to VDDR for supply from the main LDO.
(6) Output from internal DC/DC and LDO is trimmed to 1.68V.

6.3 Connections for Unused Pins and Modules—RGZ Package

表 6-2. Connections for Unused Pins—RGZ Package

FUNCTION	SIGNAL NAME	PIN NUMBER	ACCEPTABLE PRACTICE(1)	PREFERRED PRACTICE(1)
GPIO	DIO_n	8–12 14–21 26–32 36–43	NC or GND	NC
32.768 kHz crystal	X32K_Q1	6 8	NC or GND	NC
	X32K_Q2	7		
DC/DC converter(2)	DCDC_SW	33	NC	NC
	VDDS_DCDC	34	VDDS	VDDS

(1) NC = No connect
(2) When the DC/DC converter is not used, the inductor between DCDC_SW and VDDR can be removed. VDDR and VDDR_RF must still be connected and the 22µF DCDC capacitor must be kept on the VDDR net.

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6.4 Pin Diagram—RSK Package (Top View)

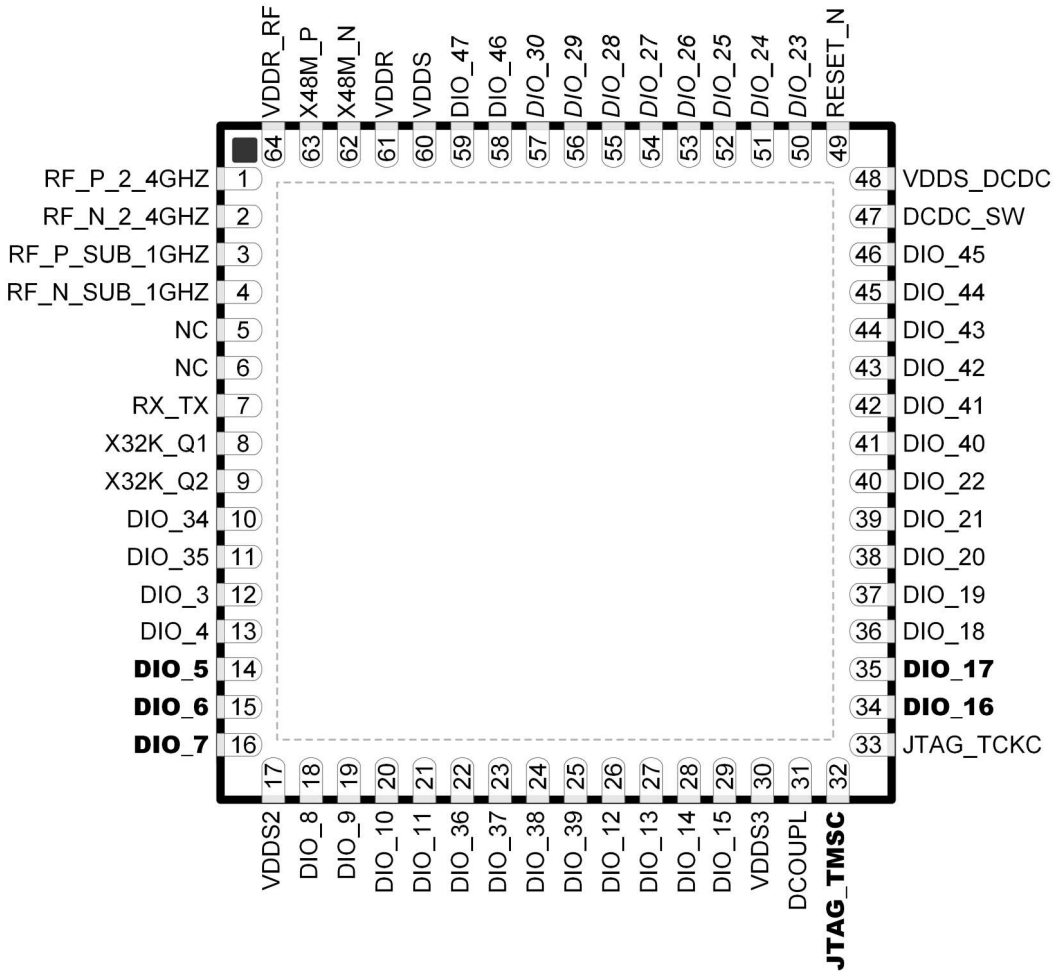


図 6-2. RSK (8mm × 8mm) Pinout, 0.4mm Pitch (Top View)

The following I/O pins marked in 図 6-2 in **bold** have high-drive capabilities:

- Pin 14, DIO_5
- Pin 15, DIO_6
- Pin 16, DIO_7
- Pin 32, JTAG_TMSC
- Pin 34, DIO_16
- Pin 35, DIO_17

The following I/O pins marked in 図 6-2 in *italics* have analog capabilities:

- Pin 50, DIO_23
- Pin 51, DIO_24
- Pin 52, DIO_25
- Pin 53, DIO_26
- Pin 54, DIO_27
- Pin 55, DIO_28
- Pin 56, DIO_29
- Pin 57, DIO_30

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6.5 Signal Descriptions—RSK Package

表 6-3. Signal Descriptions—RSK Package

PIN		I/O	TYPE	DESCRIPTION
NAME	NO.			
DCDC_SW	47	—	Power	Output from internal DC/DC converter(1)
DCOUPL	31	—	Power	For decoupling of internal 1.27V regulated digital-supply (2)
DIO_3	12	I/O	Digital	GPIO
DIO_4	13	I/O	Digital	GPIO
DIO_5	14	I/O	Digital	GPIO, high-drive capability
DIO_6	15	I/O	Digital	GPIO, high-drive capability
DIO_7	16	I/O	Digital	GPIO, high-drive capability
DIO_8	18	I/O	Digital	GPIO
DIO_9	19	I/O	Digital	GPIO
DIO_10	20	I/O	Digital	GPIO
DIO_11	21	I/O	Digital	GPIO
DIO_12	26	I/O	Digital	GPIO
DIO_13	27	I/O	Digital	GPIO
DIO_14	28	I/O	Digital	GPIO
DIO_15	29	I/O	Digital	GPIO
DIO_16	34	I/O	Digital	GPIO, JTAG_TDO, high-drive capability
DIO_17	35	I/O	Digital	GPIO, JTAG_TDI, high-drive capability
DIO_18	36	I/O	Digital	GPIO
DIO_19	37	I/O	Digital	GPIO
DIO_20	38	I/O	Digital	GPIO
DIO_21	39	I/O	Digital	GPIO
DIO_22	40	I/O	Digital	GPIO
DIO_23	50	I/O	Digital or Analog	GPIO, analog capability
DIO_24	51	I/O	Digital or Analog	GPIO, analog capability
DIO_25	52	I/O	Digital or Analog	GPIO, analog capability
DIO_26	53	I/O	Digital or Analog	GPIO, analog capability
DIO_27	54	I/O	Digital or Analog	GPIO, analog capability
DIO_28	55	I/O	Digital or Analog	GPIO, analog capability
DIO_29	56	I/O	Digital or Analog	GPIO, analog capability
DIO_30	57	I/O	Digital	GPIO, analog capability
DIO_34	10	I/O	Digital	GPIO
DIO_35	11	I/O	Digital	GPIO
DIO_36	22	I/O	Digital	GPIO
DIO_37	23	I/O	Digital	GPIO
DIO_38	24	I/O	Digital	GPIO
DIO_39	25	I/O	Digital	GPIO
DIO_40	41	I/O	Digital	GPIO
DIO_41	42	I/O	Digital	GPIO
DIO_42	43	I/O	Digital	GPIO
DIO_43	44	I/O	Digital	GPIO
DIO_44	45	I/O	Digital	GPIO
DIO_45	46	I/O	Digital	GPIO

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表 6-3. Signal Descriptions—RSK Package (続き)

PIN		I/O	TYPE	DESCRIPTION
NAME	NO.			
DIO_46	58	I/O	Digital	GPIO
DIO_47	59	I/O	Digital	GPIO
EGP	—	—	GND	Ground—exposed ground pad(3)
JTAG_TMSC	32	I/O	Digital	JTAG TMSC, high-drive capability
JTAG_TCKC	33	I	Digital	JTAG TCKC
RESET_N	49	I	Digital	Reset, active low. No internal pullup resistor
RF_P_2_4GHZ	1	—	RF	Positive 2.4GHz RF input signal to LNA during RX Positive 2.4GHz RF output signal from PA during TX
RF_N_2_4GHZ	2	—	RF	Negative 2.4GHz RF input signal to LNA during RX Negative 2.4GHz RF output signal from PA during TX
RF_P_SUB_1GHZ	3	—	RF	Positive Sub-1GHz RF input signal to LNA during RX Positive Sub-1GHz RF output signal from PA during TX
RF_N_SUB_1GHZ	4	—	RF	Negative Sub-1GHz RF input signal to LNA during RX Negative Sub-1GHz RF output signal from PA during TX
RX_TX	7	—	RF	Optional bias pin for the RF LNA
VDDR	61	—	Power	Internal supply, must be powered from the internal DC/DC converter or the internal LDO(2) (4) (6)
VDDR_RF	64	—	Power	Internal supply, must be powered from the internal DC/DC converter or the internal LDO(2) (5) (6)
VDDS	60	—	Power	1.8V to 3.8V main chip supply(1)
VDDS2	17	—	Power	1.8V to 3.8V DIO supply(1)
VDDS3	30	—	Power	1.8V to 3.8V DIO supply(1)
VDDS_DCDC	48	—	Power	1.8V to 3.8V DC/DC converter supply
X48M_N	62	—	Analog	48MHz crystal oscillator pin N
X48M_P	63	—	Analog	48MHz crystal oscillator pin P
NC	5	—	—	No connect
NC	6	—	—	No connect
X32K_Q1	8	—	Analog	32kHz crystal oscillator pin 1

(1) For more details, see technical reference manual listed in the documentation support section.
(2) Do not supply external circuitry from this pin.
(3) EGP is the only ground connection for the device. Good electrical connection to device ground on printed circuit board (PCB) is imperative for proper device operation.
(4) If internal DC/DC converter is not used, this pin is supplied internally from the main LDO.
(5) If internal DC/DC converter is not used, this pin must be connected to VDDR for supply from the main LDO.
(6) Output from internal DC/DC and LDO is trimmed to 1.68V.

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6.6 Connection of Unused Pins and Module—RSK Package

表 6-4. Connections for Unused Pins—RSK Package

FUNCTION	SIGNAL NAME	PIN NUMBER	ACCEPTABLE PRACTICE(1)	PREFERRED PRACTICE(1)
GPIO	DIO_n	10–16 14–21 26–32 36–43	NC or GND	NC
32.768kHz crystal	X32K_Q1	8	NC or GND	NC
	X32K_Q2	9		
No Connects	NC	5–8	NC	NC
DC/DC converter(2)	DCDC_SW	47	NC	NC
	VDDS_DCDC	48	VDDS	VDDS

(1) NC = No connect
(2) When the DC/DC converter is not used, the inductor between DCDC_SW and VDDR can be removed. VDDR and VDDR_RF must still be connected and the 22µF DCDC capacitor must be kept on the VDDR net.

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7 Specifications

7.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)(1) (2)

			MIN	MAX	UNIT
VDDS(3)	Supply voltage		–0.3	4.1	V
	Voltage on any digital pin(4) (5)		–0.3	VDDS + 0.3, max 4.1	V
	Voltage on crystal oscillator pins, X32K_Q1, X32K_Q2, X48M_N and X48M_P		–0.3	VDDR + 0.3, max 2.25	V
V_{in}	Voltage on ADC input	Voltage scaling enabled	–0.3	VDDS	V
		Voltage scaling disabled, internal reference	–0.3	1.49	
		Voltage scaling disabled, VDDS as reference	–0.3	VDDS / 2.9	
	Input level, Sub-1GHz RF pins			10	dBm
	Input level, 2.4GHz RF pins			5	dBm
T_{stg}	Storage temperature		–40	150	°C

(1) Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.
(2) All voltage values are with respect to ground, unless otherwise noted.
(3) VDDS_DCDC, VDDS2, and VDDS3 must be at the same potential as VDDS.
(4) Including analog capable DIOs.
(5) Injection current is not supported on any GPIO pin.

7.2 ESD Ratings

				VALUE	UNIT
V_{ESD}	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001(1)	All pins	±2000	V
		Charged device model (CDM), per ANSI/ESDA/JEDEC JS-002(2)	All pins	±500	V

(1) JEDEC document JEP155 states that 500V HBM allows safe manufacturing with a standard ESD control process
(2) JEDEC document JEP157 states that 250V CDM allows safe manufacturing with a standard ESD control process

7.3 Recommended Operating Conditions

Over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
Operating ambient temperature(1) (3)		–40	105	°C
Operating supply voltage (VDDS)		1.8	3.8	V
Operating supply voltage (VDDS), boost mode	VDDR = 1.95V +14dBm RF output power	2.1	3.8	V
Rising supply voltage slew rate		0	100	mV/µs
Falling supply voltage slew rate(2)		0	20	mV/µs

(1) Operation at or near maximum operating temperature for extended durations will result in lifetime reduction.
(2) For small coin-cell batteries, with high worst-case end-of-life equivalent source resistance, a 22 µF VDDS input capacitor must be used to ensure compliance with this slew rate.
(3) For thermal resistance characteristics refer to Thermal Resistance Characteristics. For application considerations, refer to SPRA953

7.4 Power Supply and Modules

Over operating free-air temperature range (unless otherwise noted).

PARAMETER		MIN	TYP	MAX	UNIT
VDDS Power-on-Reset (POR) threshold			1.1 - 1.55		V
VDDS Brown-out Detector (BOD) (1)	Rising threshold		1.77		V
VDDS Brown-out Detector (BOD), before initial boot (2)	Rising threshold		1.70		V

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7.4 Power Supply and Modules (続き)

Over operating free-air temperature range (unless otherwise noted).

PARAMETER		MIN	TYP	MAX	UNIT
VDDS Brown-out Detector (BOD) (1)	Falling threshold		1.75		V

(1) For boost mode (VDDR =1.95 V), TI drivers software initialization will trim VDDS BOD limits to maximum (approximately 2.0 V).
(2) Brown-out Detector is trimmed at initial boot, value is kept until device is reset by a POR reset or the RESET_N pin.

7.5 Power Consumption—Power Modes

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.6V with DC/DC enabled unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
Core Current Consumption						
I_{core}	Reset and Shutdown	Reset. RESET_N pin asserted or VDDS below power-on-reset threshold		150		nA
		Shutdown. No clocks running, no retention		171		
	Standby without cache retention	RTC running, CPU, 256kB RAM and (partial) register retention. RCOSC_LF		0.98		µA
		RTC running, CPU, 128kB RAM and (partial) register retention. RCOSC_LF		0.88		µA
		RTC running, CPU, 256kB RAM and (partial) register retention XOSC_LF		1.08		µA
		RTC running, CPU, 128kB RAM and (partial) register retention XOSC_LF		0.99		µA
	Standby with cache retention	RTC running, CPU, 256kB RAM and (partial) register retention. RCOSC_LF		2.24		µA
		RTC running, CPU, 128kB RAM and (partial) register retention. RCOSC_LF		2.16		µA
		RTC running, CPU, 256kB RAM and (partial) register retention. XOSC_LF		2.34		µA
		RTC running, CPU, 128kB RAM and (partial) register retention. XOSC_LF		2.25		µA
	Idle	Supply Systems and RAM powered RCOSC_HF		635		µA
	Active	MCU running CoreMark at 48MHz with parity enabled RCOSC_HF		3.5		mA
		MCU running CoreMark at 48MHz with parity disabled RCOSC_HF		3.4		mA
Peripheral Current Consumption						
I_{peri}	Peripheral power domain	Delta current with domain enabled		62.4		µA
	Serial power domain	Delta current with domain enabled		5.83		
	RF Core	Delta current with power domain enabled, clock enabled, RF core idle		102.0		
	µDMA	Delta current with clock enabled, module is idle		58.0		
	Timers	Delta current with clock enabled, module is idle(1)		97.2		
	I2C	Delta current with clock enabled, module is idle		9.8		
	I2S	Delta current with clock enabled, module is idle		22.2		
	SPI	Delta current with clock enabled, module is idle(2)		55.8		
	UART	Delta current with clock enabled, module is idle(3)		114.2		
	CRYPTO (AES)	Delta current with clock enabled, module is idle		15.5		
	PKA	Delta current with clock enabled, module is idle		66.6		
	TRNG	Delta current with clock enabled, module is idle		21.0		
Sensor Controller Engine Consumption						

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7.5 Power Consumption—Power Modes (続き)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.6V with DC/DC enabled unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I_{SCE}	Active mode	24MHz, infinite loop, V_{DDS} = 3.0V		849		µA
	Low-power mode	2MHz, infinite loop, V_{DDS} = 3.0V		32		

(1) Only one GPTimer running
(2) Only one SPI running
(3) Only one UART running

7.6 Power Consumption—Radio Modes

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.6V with DC/DC enabled unless otherwise noted.
Using boost mode (increasing VDDR up to 1.95 V), will increase system current by 15% (does not apply to TX +14dBm setting where this current is already included).
Relevant I_{core} and I_{peri} currents are included in below numbers.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I_{radio}	Radio receive current, 868MHz			5.8		mA
I_{radio}	Radio receive current, 2.44GHz (BLE)	V_{DDS} = 3.0V		6.9		mA
I_{radio}	Radio transmit current Sub-1GHz PA	0dBm output power setting 868MHz		9.5		mA
	Radio transmit current Sub-1GHz PA	+10dBm output power setting 868MHz		14.1		mA
	Radio transmit current Sub-1GHz PA	+14dBm output power setting 868MHz		25.8		mA
I_{radio}	Radio transmit current 2.4GHz PA (BLE)	0dBm output power setting, V_{DDS} = 3.0V		7.1		mA
	Radio transmit current 2.4GHz PA (BLE)	+5dBm output power setting 2440MHz, V_{DDS} = 3.0V		9.6		mA

7.7 Nonvolatile (Flash) Memory Characteristics

Over operating free-air temperature range and V_{DDS} = 3.0V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Flash sector size			2		kB
Supported flash erase cycles before failure, full bank(1) (2)		30			k Cycles
Supported flash erase cycles before failure, single sector(3)		60			k Cycles
Maximum number of write operations per row before sector erase(4)				83	Write Operations
Flash retention	105 °C Tj	11.4			Years
Flash sector erase current	Average delta current		1.0		mA
Flash sector erase time(5)	Zero cycles		10		ms
	30k cycles			4000	ms

(1) A full bank erase is counted as a single erase cycle on each sector.
(2) Aborting flash during erase or program modes is not a safe operation.
(3) Up to 4 customer-designated sectors can be individually erased an additional 30k times beyond the baseline bank limitation of 30k cycles.
(4) Each wordline is 2048 bits (or 256 bytes) wide. This limitation corresponds to sequential memory writes of 4 (3.1) bytes minimum per write over a whole wordline. If additional writes to the same wordline are required, a sector erase is required once the maximum number of write operations per row is reached.
(5) This number is dependent on Flash aging and increases over time and erase cycles.

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7.8 Thermal Resistance Characteristics

THERMAL METRIC(1)		PACKAGE		UNIT
		RGZ (VQFN)	RSK (VQFN)	
		48 PINS	64 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	23.4	25.1	°C/W(2)
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	13.3	11.5	°C/W(2)
$R_{\theta JB}$	Junction-to-board thermal resistance	8.0	8.9	°C/W(2)
ψ_{JT}	Junction-to-top characterization parameter	0.1	0.1	°C/W(2)
ψ_{JB}	Junction-to-board characterization parameter	7.9	8.8	°C/W(2)
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	1.7	1.2	°C/W(2)

(1) For more information about traditional and new thermal metrics, see *Semiconductor and IC Package Thermal Metrics*.
(2) °C/W = degrees Celsius per watt.

7.9 RF Frequency Bands

Over operating free-air temperature range (unless otherwise noted).

PARAMETER	MIN	TYP	MAX	UNIT
Frequency bands	2360		2500	MHz
	1076		1315	
	861		1054	
	431		527	
	359		439	
	287		351	

7.10 861MHz to 1054MHz—Receive (RX)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
General Parameters					
Digital channel filter programmable receive bandwidth		4		4000	kHz
Data rate step size			1.5		bps
Spurious emissions 25MHz to 1GHz	868MHz Conducted emissions measured according to ETSI EN 300 220		< –57		dBm
Spurious emissions 1GHz to 13GHz			< –47		dBm
Wi-SUN, 50kbps, ±12.5kHz deviation, 2-GFSK, 78 kHz RX BW, #1a					
Sensitivity	MRFSK, 866.6MHz, 10% PER, 250 byte payload		–106		dBm
Saturation limit	10% PER, 250 byte payload, 866.6MHz		10		dBm
Selectivity, +100kHz	10% PER, 250 byte payload, 866.6MHz. Wanted signal 3dB above sensitivity level.		33		dB
Selectivity, -100kHz			31		dB
Selectivity, +200kHz			38		dB
Selectivity, -200kHz			37		dB
RSSI dynamic range	Starting from the sensitivity limit		93		dB
RSSI accuracy	Starting from the sensitivity limit across the given dynamic range		±3		dB
Wi-SUN, 50kbps, ±25kHz deviation, 2-GFSK, 100kHz RX BW, #1b					
Sensitivity	MRFSK, 918.2MHz, 10% PER, 250 byte payload		–106		dBm
Saturation limit	10% PER, 250 byte payload, 918.2MHz		10		dBm
Selectivity, +200kHz	10% PER, 250 byte payload, 918.2MHz. Wanted signal 3dB above sensitivity level.		37		dB
Selectivity, -200kHz			35		dB
Selectivity, +400kHz			42		dB
Selectivity, -400kHz			41		dB

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7.10 861MHz to 1054MHz—Receive (RX) (続き)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
RSSI dynamic range	Starting from the sensitivity limit		95		dB
RSSI accuracy	Starting from the sensitivity limit across the given dynamic range		±3		dB
Wi-SUN, 100kbps, ±25kHz deviation, 2-GFSK, 135kHz RX BW, #2a					
Sensitivity	MRFSK, 866.6MHz, 10% PER, 250 byte payload		–103		dBm
Saturation limit	10% PER, 250 byte payload, 866.6MHz		10		dBm
Selectivity, +200kHz	10% PER, 250 byte payload, 866.6MHz. Wanted signal 3dB above sensitivity level.		40		dB
Selectivity, -200kHz			38		dB
Selectivity, +400kHz			46		dB
Selectivity, -400kHz			44		dB
RSSI dynamic range	Starting from the sensitivity limit		95		dB
RSSI accuracy	Starting from the sensitivity limit across the given dynamic range		±3		dB
Wi-SUN, 100kbps, ±50kHz deviation, 2-GFSK, 208 kHz RX BW, #2b					
Sensitivity	MRFSK, 920.9MHz, 10% PER, 250 byte payload		–102		dBm
Saturation limit	10% PER, 250 byte payload, 920.9MHz		10		dBm
Selectivity, +400kHz	10% PER, 250 byte payload, 920.9MHz. Wanted signal 3dB above sensitivity level, modulated blocker.		42		dB
Selectivity, -400kHz			39		dB
Selectivity, +800kHz			52		dB
Selectivity, -800kHz			46		dB
RSSI dynamic range	Starting from the sensitivity limit		91		dB
RSSI accuracy	Starting from the sensitivity limit across the given dynamic range		±3		dB
Wi-SUN, 150kbps, ±37.5kHz deviation, 2-GFSK, 273 kHz RX BW, #3					
Sensitivity	MRFSK, 918.4MHz, 10% PER, 250 byte payload		–99		dBm
Saturation limit	10% PER, 250 byte payload, 918.4MHz		10		dBm
Selectivity, +400kHz	10% PER, 250 byte payload, 918.4MHz. Wanted signal 3dB above sensitivity level.		41		dB
Selectivity, -400kHz			39		dB
Selectivity, +800kHz			50		dB
Selectivity, -800kHz			46		dB
RSSI dynamic range	Starting from the sensitivity limit		86		dB
RSSI accuracy	Starting from the sensitivity limit across the given dynamic range		±3		dB
Wi-SUN, 200kbps, ±50kHz deviation, 2-GFSK, 335kHz RX BW, #4a					
Sensitivity	MRFSK, 918.4MHz, 10% PER, 250 byte payload		–99		dBm
Saturation limit	10% PER, 250 byte payload, 918.4MHz		10		dBm
Selectivity, +400kHz	10% PER, 250 byte payload, 918.4MHz. Wanted signal 3dB above sensitivity level.		42		dB
Selectivity, -400kHz			40		dB
Selectivity, +800kHz			51		dB
Selectivity, -800kHz			47		dB
RSSI dynamic range	Starting from the sensitivity limit		91		dB
RSSI accuracy	Starting from the sensitivity limit across the given dynamic range		±3		dB
Wi-SUN, 200kbps, ±100kHz deviation, 2-GFSK, 416kHz RX BW, #4b					
Sensitivity	MRFSK, 920.8MHz, 10% PER, 250 byte payload		–98		dBm
Saturation limit	10% PER, 250 byte payload, 920.8MHz		10		dBm
Selectivity, +600kHz	10% PER, 250 byte payload, 920.8MHz. Wanted signal 3dB above sensitivity level, modulated blocker.		46		dB
Selectivity, -600kHz			43		dB
Selectivity, +1200kHz			54		dB
Selectivity, -1200kHz			51		dB
RSSI dynamic range	Starting from the sensitivity limit		86		dB

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7.10 861MHz to 1054MHz—Receive (RX) (続き)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
RSSI accuracy	Starting from the sensitivity limit across the given dynamic range		±3		dB
Wi-SUN, 300kbps, ±75kHz deviation, 2-GFSK, 496kHz RX BW, #5					
Sensitivity	MRFSK, 917.6MHz, 10% PER, 250 byte payload		–97		dBm
Saturation limit	10% PER, 250 byte payload, 917.6MHz		10		dBm
Selectivity, +600kHz	10% PER, 250 byte payload, 917.6MHz. Wanted signal 3dB above sensitivity level.		42		dB
Selectivity, -600kHz			37		dB
Selectivity, +1200kHz			51		dB
Selectivity, -1200kHz			40		dB
RSSI dynamic range	Starting from the sensitivity limit		86		dB
RSSI accuracy	Starting from the sensitivity limit across the given dynamic range		±3		dB
802.15.4-2020, 10kbps, ±5kHz deviation, 2-FSK, 26kHz RX BW, Mode #1a					
Sensitivity	FSK, 915.0MHz, 20 byte PSDU < 10% PER		–113		dBm
Sensitivity	FSK, 868.3MHz, 20 byte PSDU < 10% PER		–113		dBm
Saturation limit	PSDU length 20 octets; PER < 10%, 868.3MHz		10		dBm
Selectivity, +50kHz	PSDU length 20 octets; PER < 10%, 868.3MHz		36		dB
Selectivity, -50kHz			36		dB
Selectivity, +100kHz			40		dB
Selectivity, -100kHz			39		dB
Selectivity, +200kHz			44		dB
Selectivity, -200kHz			37		dB
Blocking, +1MHz			60		dB
Blocking, -1MHz			59		dB
Blocking, +2MHz			64		dB
Blocking, -2MHz			64		dB
Blocking, +5MHz			75		dB
Blocking, -5MHz			74		dB
Blocking, +10MHz			79		dB
Blocking, -10MHz			79		dB
Blocking + 5% Fc. (45.75MHz)	10% PER, 20 byte payload, 866.6MHz 802.15.4g mandatory mode, wanted signal -94dBm. 3dB above usable sensitivity limit according to ETSI EN 300 220 V3.1.1 (usable sensitivity -97dBm). Limit is Cat 1.5 requirement.		–15		dBm
Blocking - 5% Fc. (-45.75MHz)			–15		dBm
Image rejection (image compensation enabled)	20 byte PSDU < 10% PER, 866.6MHz. Wanted signal 3dB above sensitivity limit.		39		dB
Image rejection (image compensation enabled)	20 byte PSDU < 10% PER, 866.6MHz(1)		39		dB
RSSI dynamic range	Starting from the sensitivity limit		100		dB
RSSI accuracy	Starting from the sensitivity limit across the given dynamic range		±3		dB
Frequency error tolerance (ppm)	10% PER, 20 byte payload, measured at 10dB above sensitivity level. Negative offset		–12		ppm
Frequency error tolerance (ppm)	10% PER, 20 byte payload, measured at 10dB above sensitivity level. Positive offset		12		ppm
Symbol rate error tolerance (ppm)	10% PER, 20 byte payload, measured at 10dB above sensitivity level. Negative offset		–1000		ppm
Symbol rate error tolerance (ppm)	10% PER, 20 byte payload, measured at 10dB above sensitivity level. Positive offset		1000		ppm
802.15.4-2020, 20kbps, ±10kHz deviation, 2-FSK, 52 kHz RX BW, Mode #1b					

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7.10 861MHz to 1054MHz—Receive (RX) (続き)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Sensitivity	FSK, 20kbps, ±10kHz deviation, 2-GFSK, 915.0MHz, 52 kHz RX BW, 20 byte PSDU < 10% PER		–110		dBm
Sensitivity	FSK, 20kbps, ±10kHz deviation, 2-GFSK, 868.3MHz, 52 kHz RX BW, 20 byte PSDU < 10% PER		–110		dBm
Saturation limit	20 byte PSDU < 10% PER, 868.3MHz		10		dBm
Selectivity, +100kHz	20-byte PSDU < 10% PER, 868.3MHz		38		dB
Selectivity, -100kHz			36		dB
Selectivity, +200kHz			44		dB
Selectivity, -200kHz			42		dB
Selectivity, +400kHz			49		dB
Selectivity, -400kHz			44		dB
Blocking, +1MHz			58		dB
Blocking, -1MHz			54		dB
Blocking, -2MHz			61		dB
Blocking, +2MHz			61		dB
Blocking, -5MHz			70		dB
Blocking, +5MHz			70		dB
Blocking, -10MHz			75		dB
Blocking, +10MHz			76		dB
Blocking + 5% Fc. (45.75MHz)	20 byte PSDU < 10% PER, 866.6MHz, wanted signal -94dBm. 3dB above usable sensitivity limit according to ETSI EN 300 220 V3.1.1 (usable sensitivity -97dBm). Limit is Cat 1.5 requirement.		–13		dBm
Blocking - 5% Fc. (-45.75MHz)			–13		dBm
Image rejection (image compensation enabled)	20 byte PSDU < 10% PER, 866.6MHz. Wanted signal 3dB above sensitivity limit.		39		dB
Image rejection (image compensation enabled)	20 byte PSDU < 10% PER, 866.6MHz(1)		39		dB
RSSI dynamic range	Starting from the sensitivity limit		100		dB
RSSI accuracy	Starting from the sensitivity limit across the given dynamic range		±3		dB
Frequency error tolerance (ppm)	10% PER, 20 byte payload, measured at 10dB above sensitivity level. Negative offset		24		ppm
Frequency error tolerance (ppm)	10% PER, 20 byte payload, measured at 10dB above sensitivity level. Positive offset		24		ppm
Symbol rate error tolerance (ppm)	10% PER, 20 byte payload, measured at 10dB above sensitivity level. Negative offset		–1000		ppm
Symbol rate error tolerance (ppm)	10% PER, 20 byte payload, measured at 10dB above sensitivity level. Positive offset		1000		ppm
802.15.4, 200kbps, ±50kHz deviation, 2-GFSK, 311kHz RX BW					
Sensitivity	1% BER, 868MHz		–103		dBm
Sensitivity	1% BER, 915MHz		–103		dBm
Selectivity, +400kHz	1% BER, 915MHz. Wanted signal 3dB above sensitivity limit.		45		dB
Selectivity, -400kHz	1% BER, 915MHz. Wanted signal 3dB above sensitivity limit.		45		dB
Selectivity, +800kHz	1% BER, 915MHz. Wanted signal 3dB above sensitivity limit.		52		dB
Selectivity, –800kHz	1% BER, 915MHz. Wanted signal 3dB above sensitivity limit.		47		dB
Blocking, +2MHz	1% BER, 915MHz. Wanted signal 3dB above sensitivity limit.		59		dB
Blocking, –2MHz	1% BER, 915MHz. Wanted signal 3dB above sensitivity limit.		56		dB
Blocking, +10MHz	1% BER, 915MHz. Wanted signal 3dB above sensitivity limit.		71		dB
Blocking, –10MHz	1% BER, 915MHz. Wanted signal 3dB above sensitivity limit.		70		dB

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7.10 861MHz to 1054MHz—Receive (RX) (続き)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
802.15.4, 500kbps, ±190kHz deviation, 2-GFSK, 622 kHz RX BW					
Sensitivity 500kbps	915MHz, 1% PER, 127 byte payload		–95		dBm
Selectivity, ±1MHz	915MHz, 1% PER, 127 byte payload. Wanted signal at -88dBm		34		dB
Selectivity, ±2MHz	915MHz, 1% PER, 127 byte payload. Wanted signal at -88dBm		46		dB
Co-channel rejection	915MHz, 1% PER, 127 byte payload. Wanted signal at -71dBm		–8		dB
SimpleLink™ Long Range 2.5/5kbps (20ksps), ±5kHz Deviation, 2-GFSK, 34kHz RX Bandwidth, FEC = 1:2, DSSS = 1:4/1:2					
Sensitivity	2.5kbps, 1% BER, 868MHz		–121		dBm
Sensitivity	2.5kbps, 1% BER, 915MHz		–121		dBm
Sensitivity	5kbps, 1% BER, 868MHz		–119		dBm
Sensitivity	5kbps, 1% BER, 915MHz		–119		dBm
Saturation limit	2.5kbps, 1% BER, 868MHz		10		dBm
Selectivity, +100kHz	2.5kbps, 1% BER, 868MHz(1)		49		dB
Selectivity, -100kHz	2.5kbps, 1% BER, 868MHz(1)		49		dB
Selectivity, +200kHz	2.5kbps, 1% BER, 868MHz(1)		52		dB
Selectivity, -200kHz	2.5kbps, 1% BER, 868MHz(1)		48		dB
Selectivity, +300kHz	2.5kbps, 1% BER, 868MHz(1)		54		dB
Selectivity, -300kHz	2.5kbps, 1% BER, 868MHz(1)		48		dB
Blocking, +1MHz	2.5kbps, 1% BER, 868MHz(1)		65		dB
Blocking, -1MHz	2.5kbps, 1% BER, 868MHz(1)		60		dB
Blocking, +2MHz	2.5kbps, 1% BER, 868MHz(1)		70		dB
Blocking, -2MHz	2.5kbps, 1% BER, 868MHz(1)		68		dB
Blocking, +5MHz	2.5kbps, 1% BER, 868MHz(1)		78		dB
Blocking, -5MHz	2.5kbps, 1% BER, 868MHz(1)		77		dB
Blocking, +10MHz	2.5kbps, 1% BER, 868MHz(1)		87		dB
Blocking, -10MHz	2.5kbps, 1% BER, 868MHz(1)		92		dB
Image rejection (image compensation enabled)	2.5kbps, 1% BER, 868MHz(1)		47		dB
RSSI accuracy	Starting from the sensitivity limit across the given dynamic range		±3		dB
Frequency error tolerance (ppm)	2.5kbps, measured at –110dBm.		–24/26		ppm
Symbolrate error tolerance (ppm)	2.5kbps, measured at –110dBm.		–90/70		ppm
Narrowband, 9.6kbps ±2.4kHz Deviation, 2-GFSK, 868MHz, 17.1kHz RX BW					
Sensitivity	1% BER		-117		dBm
Adjacent Channel Rejection	1% BER. Wanted signal 3dB above usable sensitivity limit (usable sensitivity –104.6dBm). Interferer ±20kHz		42		dB
Alternate Channel Rejection	1% BER. Wanted signal 3dB above usable sensitivity limit (usable sensitivity –104.6dBm). Interferer ±40kHz		42		dB
Blocking, ±1MHz	1% BER. Wanted signal 3dB above usable sensitivity limit (usable sensitivity –104.6dBm)		66		dB
Blocking, ±2MHz	1% BER. Wanted signal 3dB above usable sensitivity limit (usable sensitivity –104.6dBm)		71		dB
Blocking, ±10MHz	1% BER. Wanted signal 3dB above usable sensitivity limit (usable sensitivity –104.6dBm)		85		dB
802.15.4, 50kbps, ±25kHz Deviation, 2-GFSK, 100kHz RX BW (Legacy)					
Sensitivity	1% BER, 868MHz		–110		dBm
Sensitivity	1% BER, 915MHz		–110		dBm
Saturation limit	1% BER, 868MHz		10		dBm

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7.10 861MHz to 1054MHz—Receive (RX) (続き)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Selectivity, +200kHz	1% BER, 868MHz(1)		44		dB
Selectivity, -200kHz			44		dB
Selectivity, +400kHz			54		dB
Selectivity, -400kHz			44		dB
Blocking, +1MHz			57		dB
Blocking, -1MHz			57		dB
Blocking, +2MHz			61		dB
Blocking, -2MHz			61		dB
Blocking, +5MHz			67		dB
Blocking, -5MHz			67		dB
Blocking, +10MHz			76		dB
Blocking, -10MHz			76		dB
Blocking + 5% Fc. (43.42MHz)	1% BER, 868MHz 802.15.4g mandatory mode, wanted signal –94dBm. 3dB above usable sensitivity limit according to ETSI EN 300 220 V3.1.1 (usable sensitivity -97dBm). Limit is Cat 1.5 requirement.		–15		dBm
Blocking - 5% Fc. (-43.42MHz)			–15		dBm
Image rejection (image compensation enabled)	1% BER, 868MHz. Wanted signal 3dB above sensitivity limit		39		dB
RSSI dynamic range	Starting from the sensitivity limit		95		dB
RSSI accuracy	Starting from the sensitivity limit across the given dynamic range		±3		dB
Frequency error tolerance (ppm)	1% BER, measured at –100dBm (10dB above sensitivity level). Negative offset		–30		ppm
Frequency error tolerance (ppm)	1% BER, measured at –100dBm (10dB above sensitivity level). Positive offset		25		ppm
Symbol rate error tolerance (ppm)	1% BER, measured at –100dBm (10dB above sensitivity level). Negative offset		–2000		ppm
Symbol rate error tolerance (ppm)	1% BER, measured at –100dBm (10dB above sensitivity level) Positive offset		2000		ppm
802.15.4, 100kbps, ±25kHz Deviation, 2-GFSK, 137kHz RX BW					
Sensitivity 100kbps	868MHz, 1% PER, 127 byte payload		–103		dBm
Selectivity, ±200kHz	868MHz, 1% PER, 127 byte payload. Wanted signal at –96dBm		38		dB
Selectivity, ±400kHz			44		dB
Co-channel rejection	868MHz, 1% PER, 127 byte payload. Wanted signal at –79dBm		–9		dB
Generic OOK (16.384kbps, OOK w / Manchester encoding, 100kHz RX BW)					
Sensitivity	OOK, 915.0MHz, 1% BER		–114		dBm
Sensitivity	OOK, 868.8MHz, 1% BER		–113		dBm
Saturation limit	868.3MHz		0		dBm
Selectivity, +200kHz	868.3MHz. Wanted signal 3dB above sensitivity level.		52		dB
Selectivity, –200kHz			47		dB
Selectivity, +400kHz			42		dB
Selectivity, –400kHz			42		dB
Blocking, +1MHz			68		dB
Blocking, –1MHz			64		dB
Blocking, +2MHz			68		dB
Blocking, –2MHz			64		dB
Blocking, +5MHz			74		dB
Blocking, –5MHz			73		dB
Blocking, +10MHz			68		dB
Blocking, –10MHz			64		dB

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7.10 861MHz to 1054MHz—Receive (RX) (続き)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
RSSI dynamic range	Starting from the sensitivity limit		95		dB
RSSI accuracy	Starting from the sensitivity limit across the given dynamic range		±3		dB
Frequency error tolerance (ppm)	Measured at 10dB above sensitivity level. Negative offset		-40		ppm
Frequency error tolerance (ppm)	Measured at 10dB above sensitivity level. Positive offset		40		ppm
Symbol rate error tolerance (ppm)	Measured at 10dB above sensitivity level. Negative offset		-2000		ppm
Symbol rate error tolerance (ppm)	Measured at 10dB above sensitivity level Positive offset		2000		ppm
WB-DSSS, 240/120/60/30kbps (480ksym/s, 2-GFSK, ±195kHz Deviation, FEC (Half Rate), DSSS = 1/2/4/8, 622 kHz RX BW)					
Sensitivity	240kbps, DSSS = 1, 1% BER, 915.0MHz		–105		dBm
Sensitivity	120kbps, DSSS = 2, 1% BER, 915.0MHz		–106		dBm
Sensitivity	60kbps, DSSS = 4, 1% BER, 915.0MHz		–108		dBm
Sensitivity	30kbps, DSSS = 8, 1% BER, 915.0MHz		–109		dBm
Saturation limit	915.0MHz		0		dBm
Blocking +1MHz	240kbps, DSSS = 1		54		dB
	120kbps, DSSS = 2		57		dB
	60kbps, DSSS = 4		57		dB
	30kbps, DSSS = 8		57		dB
Blocking -1MHz	240kbps, DSSS = 1		49		dB
	120kbps, DSSS = 2		50		dB
	60kbps, DSSS = 4		52		dB
	30kbps, DSSS = 8		53		dB
Blocking +2MHz	240kbps, DSSS = 1		54		dB
	120kbps, DSSS = 2		55		dB
	60kbps, DSSS = 4		57		dB
	30kbps, DSSS = 8		58		dB
Blocking -2MHz	240kbps, DSSS = 1		53		dB
	120kbps, DSSS = 2		54		dB
	60kbps, DSSS = 4		56		dB
	30kbps, DSSS = 8		56		dB
Blocking +5MHz	240kbps, DSSS = 1		55		dB
	120kbps, DSSS = 2		56		dB
	60kbps, DSSS = 4		58		dB
	30kbps, DSSS = 8		59		dB
Blocking -5MHz	240kbps, DSSS = 1		54		dB
	120kbps, DSSS = 2		55		dB
	60kbps, DSSS = 4		57		dB
	30kbps, DSSS = 8		58		dB
Blocking +10MHz	240kbps, DSSS = 1		69		dB
	120kbps, DSSS = 2		70		dB
	60kbps, DSSS = 4		72		dB
	30kbps, DSSS = 8		73		dB
Blocking -10MHz	240kbps, DSSS = 1		65		dB
	120kbps, DSSS = 2		67		dB
	60kbps, DSSS = 4		69		dB
	30kbps, DSSS = 8		70		dB
RSSI dynamic range	Starting from the sensitivity limit		85		dB

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7.10 861MHz to 1054MHz—Receive (RX) (続き)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
RSSI accuracy	Starting from the sensitivity limit across the given dynamic range		±3		dB

(1) Wanted signal 3dB above usable sensitivity limit according to ETSI EN 300 220 v. 3.1.1

7.11 861MHz to 1054MHz—Transmit (TX)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted. (1)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
General parameters						
Max output power, boost mode Sub-1GHz PA (2)		VDDR = 1.95V Minimum supply voltage (VDDS) for boost mode is 2.1V 868MHz and 915MHz		14		dBm
Max output power, Sub-1GHz PA (2)		868MHz and 915MHz		12		dBm
Output power programmable range Sub-1GHz PA		868MHz and 915MHz, 1dB step size.		34		dB
Output power variation over temperature Sub-1GHz PA		+10dBm setting Over recommended temperature operating range		±2		dB
Output power variation over temperature Boost mode, Sub-1GHz PA		+14dBm setting Over recommended temperature operating range		±1.5		dB
Spurious emissions and harmonics						
Spurious emissions (excluding harmonics) Sub-1GHz PA, 868MHz (3)	30MHz to 1GHz	+14dBm setting ETSI restricted bands		< –54		dBm
		+14dBm setting ETSI outside restricted bands		< –36		dBm
	1GHz to 12.75GHz (outside ETSI restricted bands)	+14dBm setting measured in 1MHz bandwidth (ETSI)		< –30		dBm
Spurious emissions out-of-band Sub-1GHz PA, 915MHz (3)	30MHz to 88MHz (within FCC restricted bands)	+14dBm setting		< –56		dBm
	88MHz to 216MHz (within FCC restricted bands)	+14dBm setting		< –52		dBm
	216MHz to 960MHz (within FCC restricted bands)	+14dBm setting		< –50		dBm
	960MHz to 2390MHz and above 2483.5MHz (within FCC restricted band)	+14dBm setting		< –42		dBm
	1GHz to 12.75GHz (outside FCC restricted bands)	+14dBm setting		< –40		dBm
Spurious emissions out-of-band Sub-1GHz PA, 920.6/928MHz (3)	Below 710MHz (ARIB T-108)	+14dBm setting		< –36		dBm
	710MHz to 900MHz (ARIB T-108)	+14dBm setting		< –55		dBm
	900MHz to 915MHz (ARIB T-108)	+14dBm setting		< –55		dBm
	930MHz to 1000MHz (ARIB T-108)	+14dBm setting		< –55		dBm
	1000MHz to 1215MHz (ARIB T-108)	+14dBm setting		< –45		dBm
	Above 1215 MHz (ARIB T-108)	+14dBm setting		< –30		dBm

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7.11 861MHz to 1054MHz—Transmit (TX) (続き)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted. (1)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
Harmonics Sub-1GHz PA	Second harmonic	+14dBm setting, 868MHz		< –30		dBm
		+14dBm setting, 915MHz		< –30		
	Third harmonic	+14dBm setting, 868MHz		< –30		dBm
		+14dBm setting, 915MHz		< –42		
	Fourth harmonic	+14dBm setting, 868MHz		< –30		dBm
		+14dBm setting, 915MHz		< –42		
	Fifth harmonic	+14dBm setting, 868MHz		< –30		dBm
		+14dBm setting, 915MHz		< –42		

(1) Some combinations of frequency, data rate and modulation format requires use of external crystal load capacitors for regulatory compliance. More details can be found in the device errata.
(2) Output power is dependent on RF match. For dual-band devices in the CC13X4 platform, output power might be slightly reduced depending on RF layout trade-offs.
(3) Suitable for systems targeting compliance with EN 300 220, EN 303 131, EN 303 204, FCC CFR47 Part 15, ARIB STD-T108.

7.12 861MHz to 1054MHz - PLL Phase Noise Wideband Mode

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Phase noise in the 868- and 915-MHz bands 20kHz PLL loop bandwidth	±10kHz offset		–74		dBc/Hz
	±100kHz offset		–97		dBc/Hz
	±200kHz offset		–107		dBc/Hz
	±400kHz offset		–113		dBc/Hz
	±1000kHz offset		–120		dBc/Hz
	±2000kHz offset		–127		dBc/Hz
	±10000kHz offset		–141		dBc/Hz

7.13 861MHz to 1054MHz - PLL Phase Noise Narrowband Mode

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Phase noise in the 868- and 915-MHz bands 150kHz PLL loop bandwith	±10kHz offset		–96		dBc/Hz
	±100kHz offset		–95		dBc/Hz
	±200kHz offset		–94		dBc/Hz
	±400kHz offset		–104		dBc/Hz
	±1000kHz offset		–121		dBc/Hz
	±2000kHz offset		–130		dBc/Hz
	±10000kHz offset		–140		dBc/Hz

7.14 Bluetooth Low Energy—Receive (RX)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V, f_{RF}= 2440MHz with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
125kbps (LE Coded)					
Receiver sensitivity	Differential mode. BER = 10^{-3}		–105		dBm
Receiver saturation	Differential mode. BER = 10^{-3}		>5		dBm

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7.14 Bluetooth Low Energy—Receive (RX) (続き)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V, f_{RF}= 2440MHz with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Frequency error tolerance	Difference between the incoming carrier frequency and the internally generated carrier frequency		> (–300 / 300)		kHz
Data rate error tolerance	Difference between incoming data rate and the internally generated data rate (37-byte packets)		> (–320 / 240)		ppm
Data rate error tolerance	Difference between incoming data rate and the internally generated data rate (255-byte packets)		> (–100 / 125)		ppm
Co-channel rejection(1)	Wanted signal at –79dBm, modulated interferer in channel, BER = 10^{-3}		–1.5		dB
Selectivity, ±1MHz(1)	Wanted signal at –79dBm, modulated interferer at ±1MHz, BER = 10^{-3}		8 / 4.5(2)		dB
Selectivity, ±2MHz(1)	Wanted signal at –79dBm, modulated interferer at ±2MHz, BER = 10^{-3}		44 / 39(2)		dB
Selectivity, ±3MHz(1)	Wanted signal at –79dBm, modulated interferer at ±3MHz, BER = 10^{-3}		43 / 43(2)		dB
Selectivity, ±4MHz(1)	Wanted signal at –79dBm, modulated interferer at ±4MHz, BER = 10^{-3}		44 / 43(2)		dB
Selectivity, ±6MHz(1)	Wanted signal at –79dBm, modulated interferer at ≥ ±6MHz, BER = 10^{-3}		48 / 43(2)		dB
Selectivity, ±7MHz	Wanted signal at –79dBm, modulated interferer at ≥ ±7MHz, BER = 10^{-3}		51 / 45(2)		dB
Selectivity, Image frequency(1)	Wanted signal at –79dBm, modulated interferer at image frequency, BER = 10^{-3}		39		dB
Selectivity, Image frequency ±1MHz(1)	Note that Image frequency + 1MHz is the Co- channel – 1MHz. Wanted signal at –79dBm, modulated interferer at ±1MHz from image frequency, BER = 10^{-3}		4.5 / 44 (2)		dB
RSSI Range			89		dB
RSSI Accuracy (+/-)			±4		dB
500kbps (LE Coded)					
Receiver sensitivity	Differential mode. BER = 10^{-3}		–100		dBm
Receiver saturation	Differential mode. BER = 10^{-3}		> 5		dBm
Frequency error tolerance	Difference between the incoming carrier frequency and the internally generated carrier frequency		> (–300 / 300)		kHz
Data rate error tolerance	Difference between incoming data rate and the internally generated data rate (37-byte packets)		> (–450 / 450)		ppm
Data rate error tolerance	Difference between incoming data rate and the internally generated data rate (255-byte packets)		> (–175 / 175)		ppm
Co-channel rejection(1)	Wanted signal at –72dBm, modulated interferer in channel, BER = 10^{-3}		–3.5		dB
Selectivity, ±1MHz(1)	Wanted signal at –72dBm, modulated interferer at ±1MHz, BER = 10^{-3}		8 / 4(2)		dB
Selectivity, ±2MHz(1)	Wanted signal at –72dBm, modulated interferer at ±2MHz, BER = 10^{-3}		41 / 37(2)		dB
Selectivity, ±3MHz(1)	Wanted signal at –72dBm, modulated interferer at ±3MHz, BER = 10^{-3}		44 / 41(2)		dB
Selectivity, ±4MHz(1)	Wanted signal at –72dBm, modulated interferer at ±4MHz, BER = 10^{-3}		44 / 43(2)		dB
Selectivity, ±6MHz(1)	Wanted signal at –72dBm, modulated interferer at ≥ ±6MHz, BER = 10^{-3}		46 / 43(2)		dB
Selectivity, ±7MHz	Wanted signal at –72dBm, modulated interferer at ≥ ±7MHz, BER = 10^{-3}		49 / 45(2)		dB
Selectivity, Image frequency(1)	Wanted signal at –72dBm, modulated interferer at image frequency, BER = 10^{-3}		37		dB

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7.14 Bluetooth Low Energy—Receive (RX) (続き)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V, f_{RF}= 2440MHz with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Selectivity, Image frequency ±1MHz(1)	Note that Image frequency + 1MHz is the Co- channel – 1MHz. Wanted signal at –72dBm, modulated interferer at ±1MHz from image frequency, BER = 10^{-3}		4 / 46(2)		dB
RSSI Range			85		dB
RSSI Accuracy (+/-)			±4		dB
1 Mbps (LE 1M)					
Receiver sensitivity	Differential mode. BER = 10^{-3}		–97		dBm
Receiver saturation	Differential mode. BER = 10^{-3}		> 5		dBm
Frequency error tolerance	Difference between the incoming carrier frequency and the internally generated carrier frequency		> (–350 / 350)		kHz
Data rate error tolerance	Difference between incoming data rate and the internally generated data rate (37-byte packets)		> (–750 / 750)		ppm
Co-channel rejection(1)	Wanted signal at –67dBm, modulated interferer in channel, BER = 10^{-3}		–6		dB
Selectivity, ±1MHz(1)	Wanted signal at –67dBm, modulated interferer at ±1MHz, BER = 10^{-3}		7 / 4(2)		dB
Selectivity, ±2MHz(1)	Wanted signal at –67dBm, modulated interferer at ±2MHz,BER = 10^{-3}		40 / 33(2)		dB
Selectivity, ±3MHz(1)	Wanted signal at –67dBm, modulated interferer at ±3MHz, BER = 10^{-3}		36 / 41(2)		dB
Selectivity, ±4MHz(1)	Wanted signal at –67dBm, modulated interferer at ±4MHz, BER = 10^{-3}		36 / 45(2)		dB
Selectivity, ±5MHz or more(1)	Wanted signal at –67dBm, modulated interferer at ≥ ±5MHz, BER = 10^{-3}		40		dB
Selectivity, image frequency(1)	Wanted signal at –67dBm, modulated interferer at image frequency, BER = 10^{-3}		33		dB
Selectivity, image frequency ±1MHz(1)	Note that Image frequency + 1MHz is the Co- channel – 1MHz. Wanted signal at –67dBm, modulated interferer at ±1MHz from image frequency, BER = 10^{-3}		4 / 41(2)		dB
Out-of-band blocking(3)	30MHz to 2000MHz		–10		dBm
Out-of-band blocking	2003MHz to 2399MHz		–18		dBm
Out-of-band blocking	2484MHz to 2997MHz		–12		dBm
Out-of-band blocking	3000MHz to 12.75GHz		–2		dBm
Intermodulation	Wanted signal at 2402MHz, –64dBm. Two interferers at 2405 and 2408MHz respectively, at the given power level		–42		dBm
Spurious emissions, 30 to 1000MHz(4)	Measurement in a 50 Ω single-ended load.		< –59		dBm
Spurious emissions, 1 to 12.75GHz(4)	Measurement in a 50 Ω single-ended load.		< –47		dBm
RSSI dynamic range			70		dB
RSSI accuracy			±4		dB
2 Mbps (LE 2M)					
Receiver sensitivity	Differential mode. Measured at SMA connector, BER = 10^{-3}		–92		dBm
Receiver saturation	Differential mode. Measured at SMA connector, BER = 10^{-3}		> 5		dBm
Frequency error tolerance	Difference between the incoming carrier frequency and the internally generated carrier frequency		> (–500 / 500)		kHz
Data rate error tolerance	Difference between incoming data rate and the internally generated data rate (37-byte packets)		> (–700 / 750)		ppm
Co-channel rejection(1)	Wanted signal at –67dBm, modulated interferer in channel,BER = 10^{-3}		–7		dB

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7.14 Bluetooth Low Energy—Receive (RX) (続き)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V, f_{RF}= 2440MHz with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Selectivity, ±2MHz(1)	Wanted signal at –67dBm, modulated interferer at ±2MHz, Image frequency is at –2MHz, BER = 10^{-3}		8 / 4(2)		dB
Selectivity, ±4MHz(1)	Wanted signal at –67dBm, modulated interferer at ±4MHz, BER = 10^{-3}		35 / 32(2)		dB
Selectivity, ±6MHz(1)	Wanted signal at –67dBm, modulated interferer at ±6MHz, BER = 10^{-3}		37 / 34(2)		dB
Selectivity, image frequency(1)	Wanted signal at –67dBm, modulated interferer at image frequency, BER = 10^{-3}		4		dB
Selectivity, image frequency ±2MHz(1)	Note that Image frequency + 2MHz is the Co-channel. Wanted signal at –67dBm, modulated interferer at ±2MHz from image frequency, BER = 10^{-3}		–7 / 36(2)		dB
Out-of-band blocking(3)	30MHz to 2000MHz		–16		dBm
Out-of-band blocking	2003MHz to 2399MHz		–21		dBm
Out-of-band blocking	2484MHz to 2997MHz		–15		dBm
Out-of-band blocking	3000MHz to 12.75GHz		–20		dBm
Intermodulation	Wanted signal at 2402MHz, –64dBm. Two interferers at 2408 and 2414MHz respectively, at the given power level		–37		dBm
RSSI Range			64		dB
RSSI Accuracy (+/-)			±4		dB

(1) Numbers given as I/C dB.
(2) X / Y, where X is +N MHz and Y is –N MHz.
(3) Excluding one exception at F_{wanted} / 2, per Bluetooth Specification.
(4) Suitable for systems targeting compliance with worldwide radio-frequency regulations ETSI EN 300 328 and EN 300 440 Class 2 (Europe), FCC CFR47 Part 15 (US), and ARIB STD-T66 (Japan).

7.15 Bluetooth Low Energy—Transmit (TX)

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V, f_{RF}= 2440MHz with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
General Parameters						
Max output power, 2.4GHz PA	Differential mode, delivered to a single-ended 50Ω load through a balun			5		dBm
Output power programmable range, 2.4GHz PA	Differential mode, delivered to a single-ended 50Ω load through a balun			26		dB
Spurious emissions and harmonics						
Spurious emissions, 2.4GHz PA	f < 1GHz, outside restricted bands	+5dBm setting		< –36		dBm
	f < 1GHz, restricted bands ETSI			< –54		dBm
	f < 1GHz, restricted bands FCC			< –55		dBm
	f > 1GHz, including harmonics			< –42		dBm
Harmonics, 2.4GHz PA	Second harmonic			< –42		dBm
	Third harmonic			< –42		dBm

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7.16 Zigbee and Thread - IEEE 802.15.4-2006 2.4GHz (OQPSK DSSS1:8, 250kbps) - RX

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V, f_{RF}= 2440MHz with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
General Parameters					
Receiver sensitivity	Coherent mode PER = 1%		–105		dBm
Receiver saturation	PER = 1%		> –10		dBm
Adjacent channel rejection	Wanted signal at –82dBm, modulated interferer at ±5MHz, PER = 1%		36		dB
Alternate channel rejection	Wanted signal at –82dBm, modulated interferer at ±10MHz, PER = 1%		55		dB
Channel rejection, ±15MHz or more	Wanted signal at –82dBm, undesired signal is IEEE 802.15.4 modulated channel, stepped through all channels 2405 to 2480MHz, PER = 1%		59		dB
Blocking and desensitization, 5MHz from upper band edge	Wanted signal at 3 dB above the sensitivity level, CW jammer, PER = 1%		57		dB
Blocking and desensitization, 10MHz from upper band edge	Wanted signal at 3 dB above the sensitivity level, CW jammer, PER = 1%		62		dB
Blocking and desensitization, 20MHz from upper band edge	Wanted signal at 3 dB above the sensitivity level, CW jammer, PER = 1%		62		dB
Blocking and desensitization, 50MHz from upper band edge	Wanted signal at 3 dB above the sensitivity level, CW jammer, PER = 1%		65		dB
Blocking and desensitization, –5MHz from lower band edge	Wanted signal at 3 dB above the sensitivity level, CW jammer, PER = 1%		60		dB
Blocking and desensitization, –10MHz from lower band edge	Wanted signal at 3 dB above the sensitivity level, CW jammer, PER = 1%		60		dB
Blocking and desensitization, –20MHz from lower band edge	Wanted signal at 3 dB above the sensitivity level, CW jammer, PER = 1%		60		dB
Blocking and desensitization, –50MHz from lower band edge	Wanted signal at 3 dB above the sensitivity level, CW jammer, PER = 1%		62		dB
Spurious emissions, 30MHz to 1000MHz	Measurement in a 50 Ω single-ended load		–66		dBm
Spurious emissions, 1GHz to 12.75GHz	Measurement in a 50 Ω single-ended load		–53		dBm
Frequency error tolerance	Difference between the incoming carrier frequency and the internally generated carrier frequency		> 100		ppm
Symbol rate error tolerance	Difference between incoming symbol rate and the internally generated symbol rate		> 800		ppm
RSSI dynamic range			95		dB
RSSI accuracy			±4		dB

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7.17 Zigbee and Thread - IEEE 802.15.4-2006 2.4GHz (OQPSK DSSS1:8, 250kbps) - TX

When measured on the LP-EM-CC1354P10-1 reference design with T_c = 25°C, V_{DDS} = 3.0V, f_{RF}= 2440MHz with DC/DC enabled unless otherwise noted. All measurements are performed at the antenna input with a combined RX and TX path. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
General Parameters						
Max output power, 2.4GHz PA	Differential mode, delivered to a single-ended 50 Ω load through a balun			5		dBm
Output power programmable range, 2.4GHz PA	Differential mode, delivered to a single-ended 50 Ω load through a balun			26		dB
Spurious emissions and harmonics						
Spurious emissions, 2.4GHz PA	f < 1GHz, outside restricted bands	+5dBm setting		< -36		dBm
	f < 1GHz, restricted bands ETSI			< -47		dBm
	f < 1GHz, restricted bands FCC			< -55		dBm
	f > 1GHz, including harmonics			< –42		dBm
Harmonics, 2.4GHz PA	Second harmonic			< -42		dBm
	Third harmonic			< -42		dBm
IEEE 802.15.4-2006 2.4GHz (OQPSK DSSS1:8, 250kbps)						
Error vector magnitude, 2.4-GHz PA	+5dBm setting			2		%

7.18 Timing and Switching Characteristics

7.18.1 Reset Timing

PARAMETER	MIN	TYP	MAX	UNIT
RESET_N low duration	1			µs

7.18.2 Wakeup Timing

Measured over operating free-air temperature with V_{DDS} = 3.0V (unless otherwise noted). The times listed here do not include software overhead.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
MCU, Reset to Active(1)			850 - 4000		µs
MCU, Shutdown to Active(1)			850 - 4000		µs
MCU, Standby to Active			160		µs
MCU, Active to Standby			39		µs
MCU, Idle to Active			15		µs

(1) The wakeup time is dependent on remaining charge on VDDR capacitor when starting the device, and thus how long the device has been in Reset or Shutdown before starting up again. The wake up time increases with a higher capacitor value.

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7.18.3 Clock Specifications

7.18.3.1 48MHz Clock Input (TCXO)

Measured on a Texas Instruments reference design with T_c = 25°C, V_{DDS} = 3.0V, unless otherwise noted.(1)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Clock frequency			48		MHz
TCXO clipped sine output, peak-to-peak	TCXO clipped sine output connected to pin X48M_P through series capacitor	0.8		1.7	V
TCXO with CMOS output, High input voltage	TCXO with CMOS output directly coupled to pin X48M_P	1.3		VDDR	V
TCXO with CMOS output, Low input voltage		0		0.3	V

(1) Probing or otherwise stopping the TCXO while the DC/DC converter is enabled may cause permanent damage to the device.

7.18.3.2 48MHz Crystal Oscillator (XOSC_HF)

Measured on a Texas Instruments reference design with T_c = 25°C, V_{DDS} = 3.0V, unless otherwise noted.(1)

	PARAMETER	MIN	TYP	MAX	UNIT
F	Crystal frequency		48		MHz
ESR	Equivalent series resistance 6 pF < C_L ≤ 9 pF		20	60	Ω
ESR	Equivalent series resistance 5 pF < C_L ≤ 6 pF			80	Ω
L_M	Motional inductance, relates to the load capacitance that is used for the crystal (C_L in Farads)(2)		< 3 × 10^{-25} / C_L ^{2}		H
C_L	Crystal load capacitance(3)	5	7(4)	9	pF
t	Start-up time(5)		200		µs

(1) Probing or otherwise stopping the crystal while the DC/DC converter is enabled may cause permanent damage to the device.
(2) The crystal manufacturer's specification must satisfy this requirement for proper operation.
(3) Adjustable load capacitance is integrated into the device. External load capacitors are required for systems targeting compliance with certain regulations. See the device errata for further details.
(4) On-chip default connected capacitance including reference design parasitic capacitance. Connected internal capacitance is changed through software in the Customer Configuration section (CCFG).
(5) Start-up time using the TI-provided power driver. Start-up time may increase if driver is not used.

7.18.3.3 48MHz RC Oscillator (RCOSC_HF)

Measured on a Texas Instruments reference design with T_c = 25°C, V_{DDS} = 3.0V, unless otherwise noted.

	MIN	TYP	MAX	UNIT
Frequency		48		MHz
Uncalibrated frequency accuracy		±1		%
Calibrated frequency accuracy(1)		±0.25		%
Start-up time		5		µs

(1) Accuracy relative to the calibration source (XOSC_HF).

7.18.3.4 2MHz RC Oscillator (RCOSC_MF)

Measured on a Texas Instruments reference design with T_c = 25°C, V_{DDS} = 3.0V, unless otherwise noted.

	MIN	TYP	MAX	UNIT
Calibrated frequency		2		MHz
Start-up time		5		µs

7.18.3.5 32.768 kHz Crystal Oscillator (XOSC_LF)

Measured on a Texas Instruments reference design with T_c = 25°C, V_{DDS} = 3.0V, unless otherwise noted.

		MIN	TYP	MAX	UNIT
	Crystal frequency		32.768		kHz
ESR	Equivalent series resistance		30	100	kΩ

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7.18.3.5 32.768 kHz Crystal Oscillator (XOSC_LF) (続き)

Measured on a Texas Instruments reference design with T_c = 25°C, V_{DDS} = 3.0V, unless otherwise noted.

		MIN	TYP	MAX	UNIT
C_L	Crystal load capacitance	6	7(1)	12	pF

(1) Default load capacitance using TI reference designs including parasitic capacitance. Crystals with different load capacitance may be used.

7.18.3.6 32 kHz RC Oscillator (RCOSC_LF)

Measured on a Texas Instruments reference design with T_c = 25°C, V_{DDS} = 3.0V, unless otherwise noted.

	MIN	TYP	MAX	UNIT
Calibrated frequency		32.8 (1)		kHz
Temperature coefficient.		50		ppm/°C

(1) When using RCOSC_LF as source for the low frequency system clock (SCLK_LF), the accuracy of the SCLK_LF-derived Real Time Clock (RTC) can be improved by measuring RCOSC_LF relative to XOSC_HF and compensating for the RTC tick speed. This functionality is available through the TI-provided Power driver.

7.18.4 Serial Peripheral Interface (SPI) Characteristics

7.18.4.1 SPI Characteristics

over operating free-air temperature range (unless otherwise noted).

PARAMETERS		TEST CONDITIONS	MIN	TYP	MAX	UNIT
f_{SCLK} $1/t_{sclk}$	SPI clock frequency	Master Mode 1.8 < VDDS < 3.8			12	MHz
		Slave Mode 2.7 < VDDS < 3.8			8	
		Slave Mode VDDS < 2.7			7	
DC_{SCK}	SCK Duty Cycle		45	50	55	%

7.18.4.2 SPI Master Mode

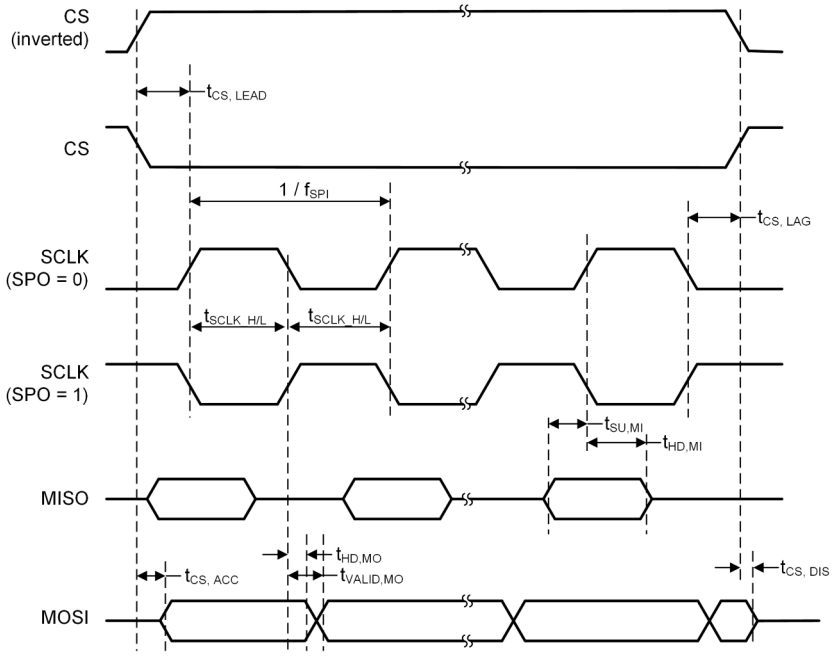
over operating free-air temperature range (unless otherwise noted).

PARAMETERS		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{SCLK_H/L}$	SCLK High or Low time		$(t_{SPI}/2)$ - 1	t_{SPI} / 2	$(t_{SPI}/2)$ + 1	ns
$t_{CS.LEAD}$	CS lead-time, CS active to clock		1			SCLK
$t_{CS.LAG}$	CS lag time, Last clock to CS inactive		1			SCLK
$t_{CS.ACC}$	CS access time, CS active to MOSI data out				1	SCLK
$t_{CS.DIS}$	CS disable time, CS inactive to MOSI high inpedance				1	SCLK
$t_{SU.MI}$	MISO input data setup time(1)	VDDS = 3.3V	12.5			ns
$t_{SU.MI}$	MISO input data setup time	VDDS = 1.8V	23.5			ns
$t_{HD.MI}$	MISO input data hold time		0			ns
$t_{VALID.MO}$	MOSI output data valid time(2)	SCLK edge to MOSI valid,CL = 20 pF (4)			13	ns
$t_{HD.MO}$	MOSI output data hold time(3)	CL = 20 pF	0			ns

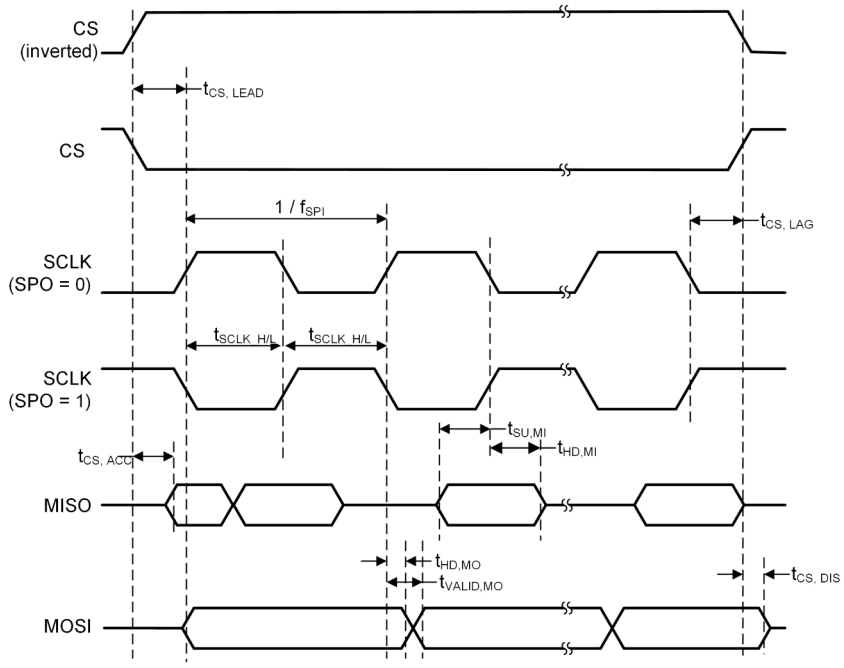
(1) The MISO input data setup time can be fully compensated when delayed sampling feature is enabled.
(2) Specifies the time to drive the next valid data to the output after the output changing SCLK clock edge.
(3) Specifies how long data on the output is valid after the output changing SCLK clock edge.

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7.18.4.3 SPI Master Mode Timing Diagrams



Master Mode, SPH = 0



Master Mode, SPH = 1

図 7-1. SPI Master Mode Timing

7.18.4.4 SPI Slave Mode

over operating free-air temperature range (unless otherwise noted).

PARAMETERS		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{CS.LEAD}$	CS lead-time, CS active to clock		1			SCLK
$t_{CS.LAG}$	CS lag time, Last clock to CS inactive		1			SCLK
$t_{CS.ACC}$	CS access time, CS active to MISO data out	VDDS = 3.3V			56	ns
$t_{CS.ACC}$	CS access time, CS active to MISO data out	VDDS = 1.8V			70	ns
$t_{CS.DIS}$	CS disable time, CS inactive to MISO high inpedance	VDDS = 3.3V			56	ns
$t_{CS.DIS}$	CS disable time, CS inactive to MISO high inpedance	VDDS = 1.8V			70	ns
$t_{SU.SI}$	MOSI input data setup time		30			ns
$t_{HD.SI}$	MOSI input data hold time		0			ns
$t_{VALID.SO}$	MISO output data valid time(1)	SCLK edge to MISO valid,C_L = 20 pF, 3.3V (4)			50	ns
$t_{VALID.SO}$	MISO output data valid time(1)	SCLK edge to MISO valid,C_L = 20 pF, 1.8V (4)			65	ns
$t_{HD.SO}$	MISO output data hold time(2)	C_L = 20 pF	0			ns

(1) Specifies the time to drive the next valid data to the output after the output changing SCLK clock edge.
(2) Specifies how long data on the output is valid after the output changing SCLK clock edge.

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7.18.4.5 SPI Slave Mode Timing Diagrams

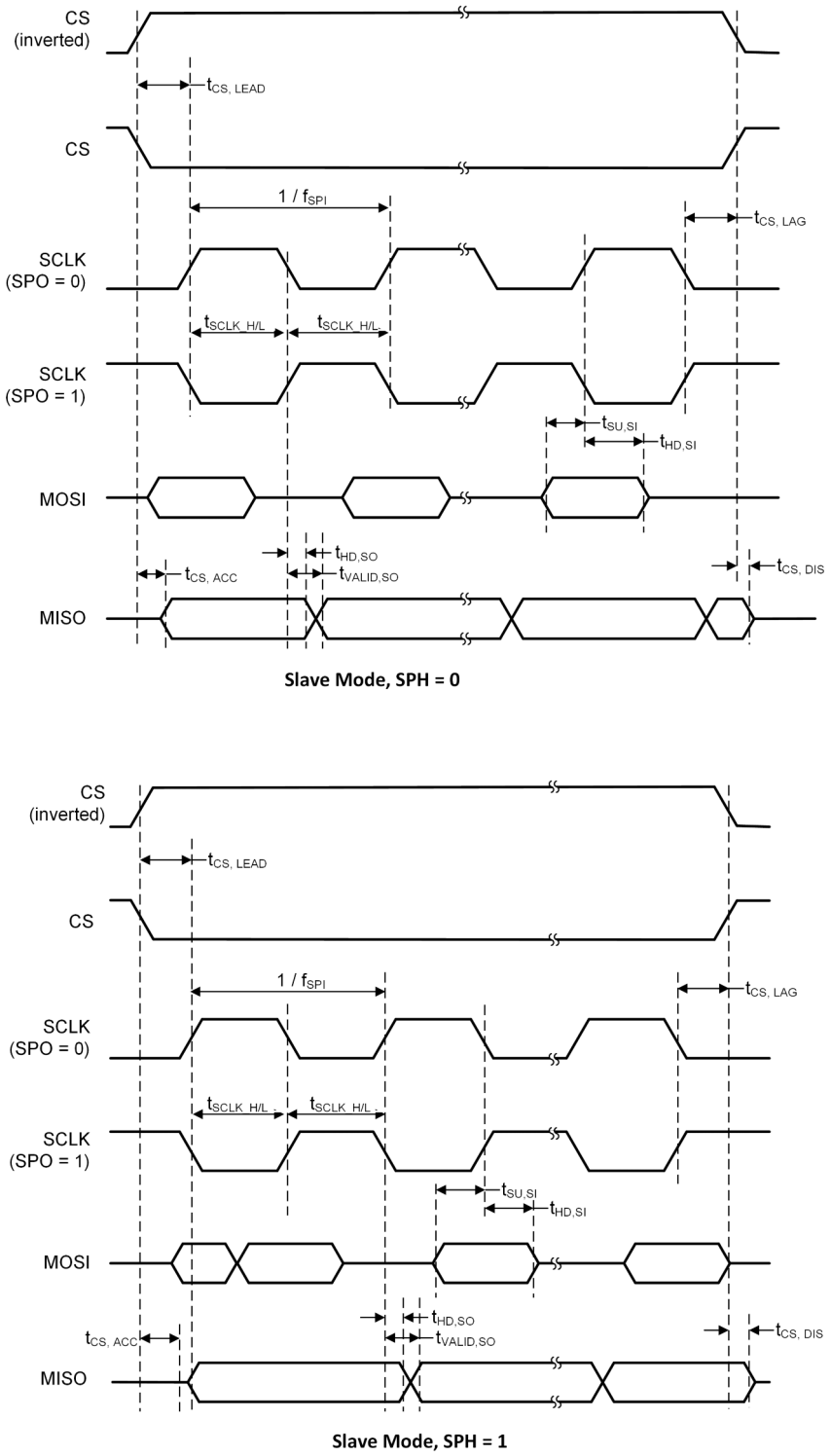


図 7-2. SPI Slave Mode Timing

7.18.5 UART

7.18.5.1 UART Characteristics

over operating free-air temperature range (unless otherwise noted).

PARAMETER	MIN	TYP	MAX	UNIT
UART rate			3	MBaud

7.19 Peripheral Characteristics

7.19.1 ADC

7.19.1.1 Analog-to-Digital Converter (ADC) Characteristics

T_c = 25°C, V_{DDS} = 3.0V and voltage scaling enabled, unless otherwise noted.(1)
Performance numbers require use of offset and gain adjustements in software by TI-provided ADC drivers.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Input voltage range		0		VDDS	V
	Resolution			12		Bits
	Sample Rate				200	ksps
	Offset	Internal 4.3V equivalent reference(2)		–0.24		LSB
	Gain error	Internal 4.3V equivalent reference(2)		7.14		LSB
DNL(3)	Differential nonlinearity			>–1		LSB
INL	Integral nonlinearity			±4		LSB
ENOB	Effective number of bits	Internal 4.3V equivalent reference(2), 200 kSamples/s, 9.6kHz input tone		9.8		Bits
		Internal 4.3V equivalent reference(2), 200 kSamples/s, 9.6kHz input tone, DC/DC enabled		9.8		
		VDDS as reference, 200 kSamples/s, 9.6kHz input tone		10.1		
		Internal reference, voltage scaling disabled, 32 samples average, 200 kSamples/s, 300Hz input tone		11.1		
		Internal reference, voltage scaling disabled, 14-bit mode, 200 kSamples/s, 300Hz input tone (4)		11.3		
		Internal reference, voltage scaling disabled, 15-bit mode, 200 kSamples/s, 300Hz input tone (4)		11.6		
THD	Total harmonic distortion	Internal 4.3V equivalent reference(2), 200 kSamples/s, 9.6kHz input tone		–65		dB
		VDDS as reference, 200 kSamples/s, 9.6kHz input tone		–70		
		Internal reference, voltage scaling disabled, 32 samples average, 200 kSamples/s, 300Hz input tone		–72		
SINAD, SNDR	Signal-to-noise and distortion ratio	Internal 4.3V equivalent reference(2), 200 kSamples/s, 9.6kHz input tone		60		dB
		VDDS as reference, 200 kSamples/s, 9.6kHz input tone		63		
		Internal reference, voltage scaling disabled, 32 samples average, 200 kSamples/s, 300Hz input tone		68		
SFDR	Spurious-free dynamic range	Internal 4.3V equivalent reference(2), 200 kSamples/s, 9.6kHz input tone		70		dB
		VDDS as reference, 200 kSamples/s, 9.6kHz input tone		73		
		Internal reference, voltage scaling disabled, 32 samples average, 200 kSamples/s, 300Hz input tone		75		
	Conversion time	Serial conversion, time-to-output, 24MHz clock		50		Clock Cycles
	Current consumption	Internal 4.3V equivalent reference(2)		0.42		mA
	Current consumption	VDDS as reference		0.6		mA
	Reference voltage	Equivalent fixed internal reference (input voltage scaling enabled). For best accuracy, the ADC conversion should be initiated through the TI-RTOS API in order to include the gain/ offset compensation factors stored in FCFG1		4.3(2) (5)		V

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7.19.1.1 Analog-to-Digital Converter (ADC) Characteristics (続き)

T_c = 25°C, V_{DDS} = 3.0V and voltage scaling enabled, unless otherwise noted.(1)
Performance numbers require use of offset and gain adjustements in software by TI-provided ADC drivers.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Reference voltage	Fixed internal reference (input voltage scaling disabled). For best accuracy, the ADC conversion should be initiated through the TI-RTOS API in order to include the gain/offset compensation factors stored in FCFG1. This value is derived from the scaled value (4.3V) as follows: V_{ref} = 4.3V × 1408 / 4095		1.48		V
	Reference voltage	VDDS as reference, input voltage scaling enabled		VDDS		V
	Reference voltage	VDDS as reference, input voltage scaling disabled		VDDS / 2.82(5)		V
	Input impedance	200 kSamples/s, voltage scaling enabled. Capacitive input, Input impedance depends on sampling frequency and sampling time		>1		MΩ

(1) Using IEEE Std 1241-2010 for terminology and test methods.
(2) Input signal scaled down internally before conversion, as if voltage range was 0V to 4.3V.
(3) No missing codes.
(4) ADC_output = Σ(4^n samples) >> n, n = desired extra bits.
(5) Applied voltage must be within Absolute Maximum Ratings at all times.

7.19.2 DAC

7.19.2.1 Digital-to-Analog Converter (DAC) Characteristics

T_c = 25°C, V_{DDS} = 3.0V, unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
General Parameters						
	Resolution			8		Bits
V_{DDS}	Supply voltage	Any load, any V_{REF}, pre-charge OFF, DAC charge-pump ON	1.8		3.8	V
		External Load(1), any V_{REF}, pre-charge OFF, DAC charge-pump OFF	2.0		3.8	
		Any load, V_{REF} = DCOUPL, pre-charge ON	2.6		3.8	
F_{DAC}	Clock frequency	Buffer ON (recommended for external load)	16		250	kHz
		Buffer OFF (internal load)	16		1000	
	Voltage output settling time	V_{REF} = VDDS, buffer OFF, internal load		13		1 / F_{DAC}
		V_{REF} = VDDS, buffer ON, external capacitive load = 20pF(2)		13.8		
	External capacitive load			20	200	pF
	External resistive load		10			MΩ
	Short circuit current				400	µA
Z_{MAX}	Max output impedance Vref = VDDS, buffer ON, CLK 250kHz	VDDS = 3.8V, DAC charge-pump OFF		50.8		kΩ
		VDDS = 3.0V, DAC charge-pump ON		51.7		
		VDDS = 3.0V, DAC charge-pump OFF		53.2		
		VDDS = 2.0 V, DAC charge-pump ON		48.7		
		VDDS = 2.0 V, DAC charge-pump OFF		70.2		
		VDDS = 1.8V, DAC charge-pump ON		46.3		
		VDDS = 1.8V, DAC charge-pump OFF		88.9		
Internal Load - Continuous Time Comparator / Low Power Clocked Comparator						
DNL	Differential nonlinearity	V_{REF} = VDDS, load = Continuous Time Comparator or Low Power Clocked Comparator F_{DAC} = 250kHz		±1		LSB(3)
	Differential nonlinearity	V_{REF} = VDDS, load = Continuous Time Comparator or Low Power Clocked Comparator F_{DAC} = 16 kHz		±1.2		
	Offset error(4) Load = Continuous Time Comparator	V_{REF} = VDDS = 3.8V		±0.64		LSB(3)
		V_{REF} = VDDS = 3.0V		±0.81		
		V_{REF} = VDDS = 1.8V		±1.27		
		V_{REF} = DCOUPL, pre-charge ON		±3.43		
		V_{REF} = DCOUPL, pre-charge OFF		±2.88		
		V_{REF} = ADCREF		±2.37		
	Offset error(4) Load = Low Power Clocked Comparator	V_{REF} = VDDS = 3.8V		±0.78		LSB(3)
		V_{REF} = VDDS = 3.0V		±0.77		
		V_{REF} = VDDS = 1.8V		±3.46		
		V_{REF} = DCOUPL, pre-charge ON		±3.44		
		V_{REF} = DCOUPL, pre-charge OFF		±4.70		
		V_{REF} = ADCREF		±4.11		
	Max code output voltage variation(4) Load = Continuous Time Comparator	V_{REF} = VDDS = 3.8V		±1.53		LSB(3)
		V_{REF} = VDDS = 3.0V		±1.71		
		V_{REF} = VDDS = 1.8V		±2.10		
		V_{REF} = DCOUPL, pre-charge ON		±6.00		
		V_{REF} = DCOUPL, pre-charge OFF		±3.85		
		V_{REF} = ADCREF		±5.84		

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7.19.2.1 Digital-to-Analog Converter (DAC) Characteristics (続き)

T_c = 25°C, V_{DDS} = 3.0V, unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Max code output voltage variation[4] Load = Low Power Clocked Comparator	V_{REF} = VDDS = 3.8V		±2.92		LSB[3]
		V_{REF} =VDDS = 3.0V		±3.06		
		V_{REF} = VDDS = 1.8V		±3.91		
		V_{REF} = DCOUPL, pre-charge ON		±7.84		
		V_{REF} = DCOUPL, pre-charge OFF		±4.06		
		V_{REF} = ADCREF		±6.94		
	Output voltage range[4] Load = Continuous Time Comparator	V_{REF} = VDDS = 3.8V, code 1		0.03		V
		V_{REF} = VDDS = 3.8V, code 255		3.62		
		V_{REF} = VDDS = 3.0V, code 1		0.02		
		V_{REF} = VDDS = 3.0V, code 255		2.86		
		V_{REF} = VDDS = 1.8V, code 1		0.01		
		V_{REF} = VDDS = 1.8V, code 255		1.71		
		V_{REF} = DCOUPL, pre-charge OFF, code 1		0.01		
		V_{REF} = DCOUPL, pre-charge OFF, code 255		1.21		
		V_{REF} = DCOUPL, pre-charge ON, code 1		1.27		
		V_{REF} = DCOUPL, pre-charge ON, code 255		2.46		
		V_{REF} = ADCREF, code 1		0.01		
		V_{REF} = ADCREF, code 255		1.41		
	Output voltage range[4] Load = Low Power Clocked Comparator	V_{REF} = VDDS = 3.8V, code 1		0.03		V
		V_{REF} = VDDS = 3.8V, code 255		3.61		
		V_{REF} = VDDS = 3.0V, code 1		0.02		
		V_{REF} = VDDS = 3.0V, code 255		2.85		
		V_{REF} = VDDS = 1.8V, code 1		0.01		
		V_{REF} = VDDS = 1.8V, code 255		1.71		
		V_{REF} = DCOUPL, pre-charge OFF, code 1		0.01		
		V_{REF} = DCOUPL, pre-charge OFF, code 255		1.21		
		V_{REF} = DCOUPL, pre-charge ON, code 1		1.27		
		V_{REF} = DCOUPL, pre-charge ON, code 255		2.46		
		V_{REF} = ADCREF, code 1		0.01		
		V_{REF} = ADCREF, code 255		1.41		
External Load (Keysight 34401A Multimeter)						
INL	Integral nonlinearity	V_{REF} = VDDS, F_{DAC} = 250kHz		±1		LSB[3]
		V_{REF} = DCOUPL, F_{DAC} = 250kHz		±1		
		V_{REF} = ADCREF, F_{DAC} = 250kHz		±1		
DNL	Differential nonlinearity	V_{REF} = VDDS, F_{DAC} = 250kHz		±1		LSB[3]
	Offset error	V_{REF} = VDDS = 3.8V		±0.20		LSB[3]
		V_{REF} = VDDS = 3.0V		±0.25		
		V_{REF} = VDDS = 1.8V		±0.45		
		V_{REF} = DCOUPL, pre-charge ON		±1.55		
		V_{REF} = DCOUPL, pre-charge OFF		±1.30		
		V_{REF} = ADCREF		±1.10		
	Max code output voltage variation	V_{REF} = VDDS = 3.8V		±0.60		LSB[3]
		V_{REF} = VDDS = 3.0V		±0.55		
		V_{REF} = VDDS = 1.8V		±0.60		
		V_{REF} = DCOUPL, pre-charge ON		±3.45		
		V_{REF} = DCOUPL, pre-charge OFF		±2.10		
		V_{REF} = ADCREF		±1.90		

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7.19.2.1 Digital-to-Analog Converter (DAC) Characteristics (続き)

T_c = 25°C, V_{DDS} = 3.0V, unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Output voltage range Load = Low Power Clocked Comparator	V_{REF} = VDDS = 3.8V, code 1		0.03		V
		V_{REF} = VDDS = 3.8V, code 255		3.61		
		V_{REF} = VDDS = 3.0V, code 1		0.02		
		V_{REF} = VDDS = 3.0V, code 255		2.85		
		V_{REF} = VDDS = 1.8V, code 1		0.02		
		V_{REF} = VDDS = 1.8V, code 255		1.71		
		V_{REF} = DCOUPL, pre-charge OFF, code 1		0.02		
		V_{REF} = DCOUPL, pre-charge OFF, code 255		1.20		
		V_{REF} = DCOUPL, pre-charge ON, code 1		1.27		
		V_{REF} = DCOUPL, pre-charge ON, code 255		2.46		
		V_{REF} = ADCREF, code 1		0.02		
		V_{REF} = ADCREF, code 255		1.42		

(1) Keysight 34401A Multimeter.
(2) A load > 20pF increincreasesettling time.
(3) 1 LSB (V_{REF} 3.8V/3.0V/1.8V/DCOUPL/ADCREF) = 14.10mV/11.13mV/6.68mV/4.67mV/5.48mV.
(4) Includes comparator offset.

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7.19.3 Temperature and Battery Monitor

7.19.3.1 Temperature Sensor

Measured on a Texas Instruments reference design with T_c = 25°C, V_{DDS} = 3.0V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Resolution			2		°C
Accuracy	-40°C to 0°C		±5.0		°C
Accuracy	0°C to 105 °C		±3.5		°C
Supply voltage coefficient(1)			3.6		°C/V

(1) The temperature sensor is automatically compensated for VDDS variation when using the TI-provided driver.

7.19.3.2 Battery Monitor

Measured on a Texas Instruments reference design with T_c = 25°C, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Resolution			25		mV
Range		1.8		3.8	V
Integral nonlinearity (max)			23		mV
Accuracy	VDDS = 3.0V		22.5		mV
Offset error			-32		mV
Gain error			-1		%

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7.19.4 Comparators

7.19.4.1 Low-Power Clocked Comparator

T_c = 25°C, V_{DDS} = 3.0V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Input voltage range		0		V_{DDS}	V
Clock frequency			SCLK_LF		
Internal reference voltage(1)	Using internal DAC with VDDS as reference voltage, DAC code = 0 - 255		0.024 - 2.865		V
Offset	Measured at V_{DDS} / 2, includes error from internal DAC		±5		mV
Decision time	Step from –50 mV to 50 mV		1		Clock Cycle

(1) The comparator can use an internal 8 bits DAC as its reference. The DAC output voltage range depends on the reference voltage selected. See DAC Characteristics.

7.19.4.2 Continuous Time Comparator

T_c = 25°C, V_{DDS} = 3.0V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Input voltage range(1)		0		V_{DDS}	V
Offset	Measured at V_{DDS} / 2		±5		mV
Decision time	Step from –10mV to 10mV		0.78		µs
Current consumption	Internal reference		8.6		µA

(1) The input voltages can be generated externally and connected throughout I/Os or an internal reference voltage can be generated using the DAC.

7.19.5 Current Source

7.19.5.1 Programmable Current Source

T_c = 25°C, V_{DDS} = 3.0V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Current source programmable output range (logarithmic range)			0.25 - 20		µA
Resolution			0.25		µA

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7.19.6 GPIO

7.19.6.1 GPIO DC Characteristics

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
T_A = 25°C, V_{DDS} = 1.8V					
GPIO VOH at 8mA load	IOCURR = 2, high-drive GPIOs only		1.56		V
GPIO VOL at 8mA load	IOCURR = 2, high-drive GPIOs only		0.24		V
GPIO VOH at 4mA load	IOCURR = 1		1.59		V
GPIO VOL at 4mA load	IOCURR = 1		0.21		V
GPIO pullup current	Input mode, pullup enabled, Vpad = 0 V		73		µA
GPIO pulldown current	Input mode, pulldown enabled, Vpad = VDDS		19		µA
GPIO low-to-high input transition, with hysteresis	IH = 1, transition voltage for input read as 0 → 1		1.08		V
GPIO high-to-low input transition, with hysteresis	IH = 1, transition voltage for input read as 1 → 0		0.73		V
GPIO input hysteresis	IH = 1, difference between 0 → 1 and 1 → 0 points		0.35		V
T_A = 25°C, V_{DDS} = 3.0V					
GPIO VOH at 8mA load	IOCURR = 2, high-drive GPIOs only		2.59		V
GPIO VOL at 8mA load	IOCURR = 2, high-drive GPIOs only		0.42		V
GPIO VOH at 4mA load	IOCURR = 1		2.63		V
GPIO VOL at 4mA load	IOCURR = 1		0.40		V
T_A = 25°C, V_{DDS} = 3.8V					
GPIO pullup current	Input mode, pullup enabled, Vpad = 0 V		282		µA
GPIO pulldown current	Input mode, pulldown enabled, Vpad = VDDS		110		µA
GPIO low-to-high input transition, with hysteresis	IH = 1, transition voltage for input read as 0 → 1		1.97		V
GPIO high-to-low input transition, with hysteresis	IH = 1, transition voltage for input read as 1 → 0		1.55		V
GPIO input hysteresis	IH = 1, difference between 0 → 1 and 1 → 0 points		0.42		V
T_A = 25°C					
VIH	Lowest GPIO input voltage reliably interpreted as a *High*	$0.8*V_{DDS}$			V
VIL	Highest GPIO input voltage reliably interpreted as a *Low*			$0.2*V_{DDS}$	V

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7.20 Typical Characteristics

All measurements in this section are done with T_c = 25°C and V_{DDS} = 3.0V, unless otherwise noted. See *Recommended Operating Conditions*, セクション 7.3, for device limits. Values exceeding these limits are for reference only.

7.20.1 MCU Current

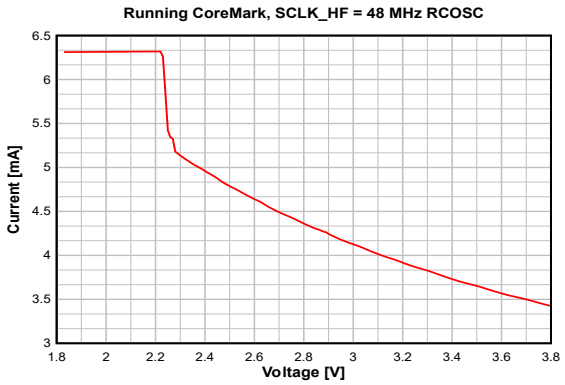


図 7-3. Active Mode (MCU) Current vs. Supply Voltage (VDDS)

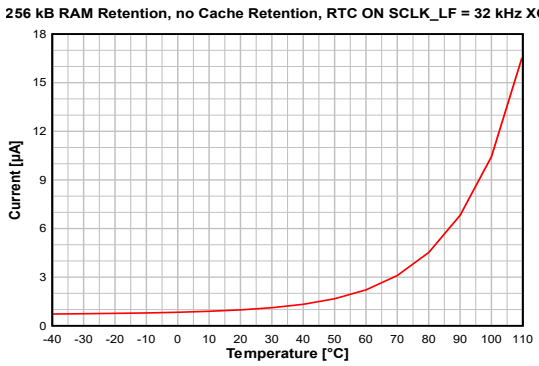


図 7-4. Standby Mode (MCU) Current vs. Temperature

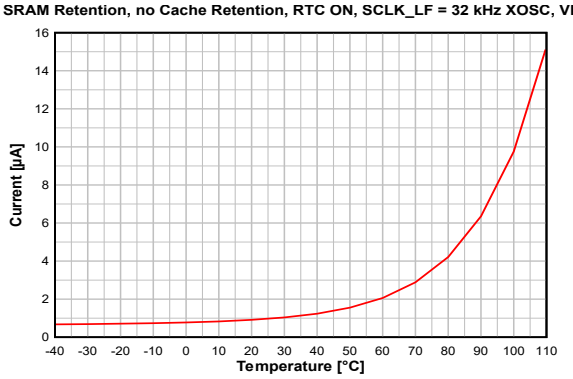


図 7-5. Standby Mode (MCU) Current vs. Temperature (VDDS = 3.6V)

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7.20.2 RX Current

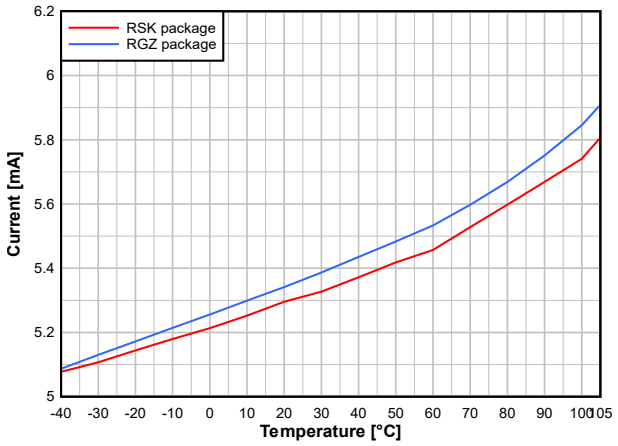


図 7-6. RX Current vs. Temperature (50kbps, 868.3MHz, VDDS = 3.6V)

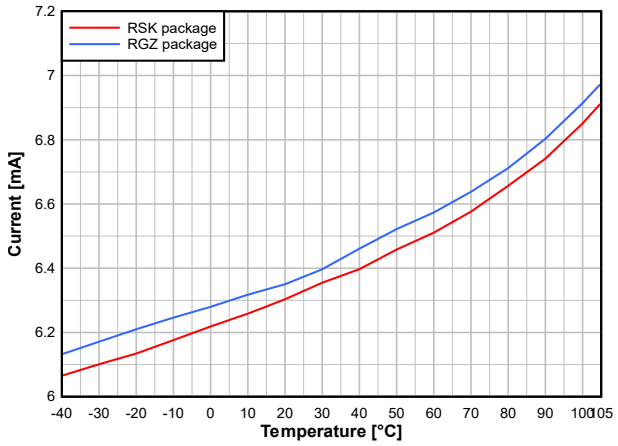


図 7-7. RX Current vs. Temperature (BLE 1Mbps, 2.44GHz)

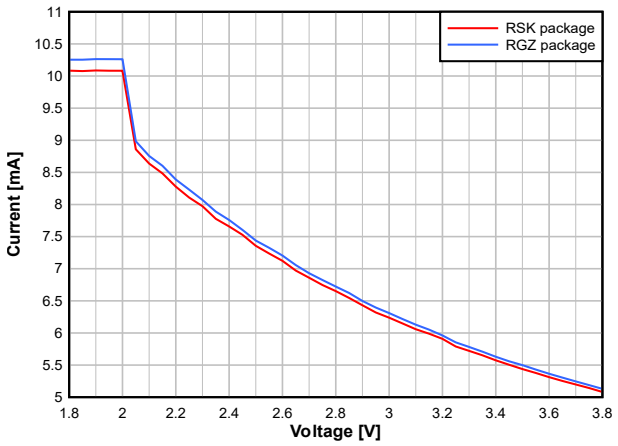


図 7-8. RX Current vs. Supply Voltage (VDDS) (50kbps, 868.3MHz)

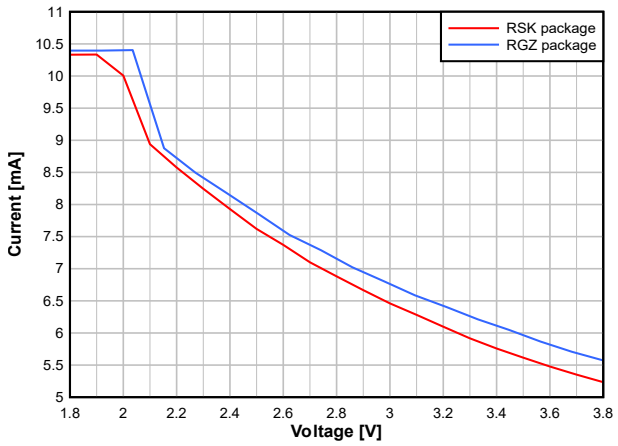


図 7-9. RX Current vs. Supply Voltage (VDDS) (BLE 1Mbps, 2.44GHz)

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7.20.3 TX Current

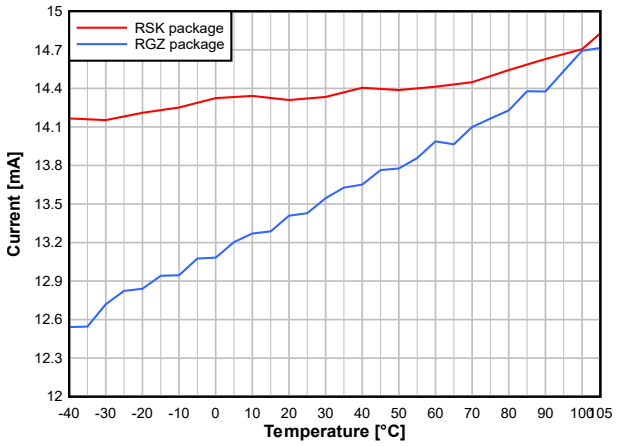


図 7-10. TX Current vs. Temperature (50kbps, 868.3MHz, +10dBm, VDDS = 3.6V)

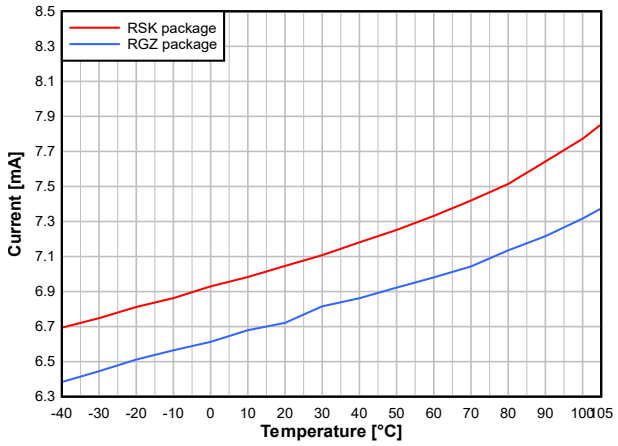


図 7-11. TX Current vs. Temperature (BLE 1Mbps, 2.44GHz, 0dBm)

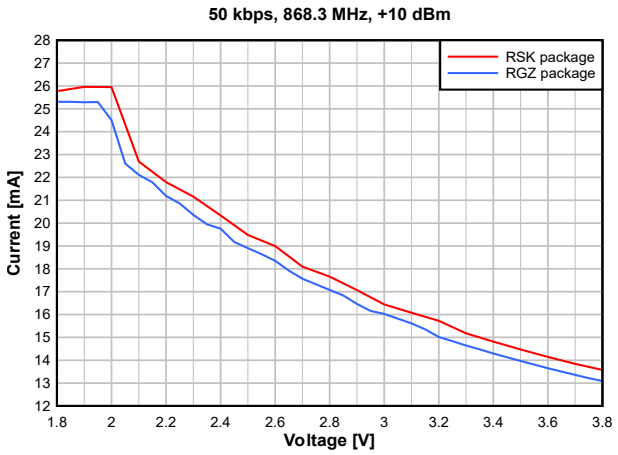


図 7-12. TX Current vs. Supply Voltage (VDDS) (50kbps, 868.3MHz)

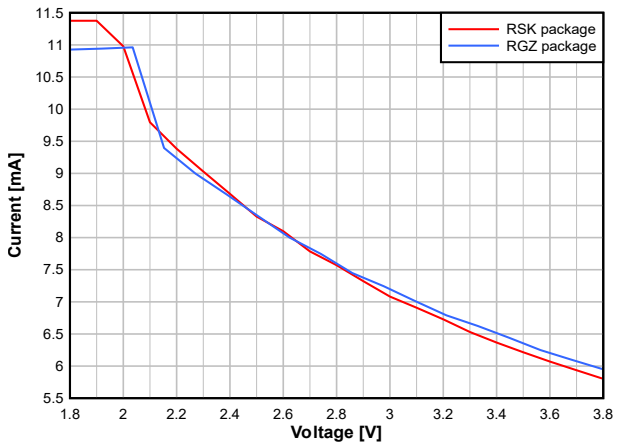


図 7-13. TX Current vs. Supply Voltage (VDDS) (BLE 1 Mbps, 2.44GHz, 0dBm)

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表 7-1 and 表 7-2 for RGZ (7 × 7) package and 表 7-3 and 表 7-4 for RSK (8 × 8) package show typical TX current and output power for different output power settings.

表 7-1. Typical TX Current and Output Power (868MHz, VDDS = 3.6V)

CC1354R10 RGZ at 868MHz, VDDS = 3.6V (Measured on CC1354R10EM-7X7-XD7793-XD24)			
txPower	**TX Power Setting (SmartRF Studio)**	**Typical Output Power [dBm]**	**Typical Current Consumption [mA]**
0x13F	14	13.5	24.3
0xB02B	12	12.0	17.2
0x785E	11	11.0	15.1
0x2E04	10	10.1	13.5
0x70A4	9	9.0	12.6
0x16A66	8	8.0	12.2
0x14E5D	7	7.0	11.1
0x14057	6	6.1	10.4
0x5CE5	5	5.0	9.6
0x158B3	4	4.0	9.9
0x14EA9	3	3.0	9.1
0x11C4A	2	2.1	8.2
0x1329C	1	1.0	7.9
0x12A98	0	0.1	7.5
0x138F4	-1	-1.0	8.2
0x132EB	-2	-2.0	7.6
0x12AE4	-3	-3.0	7.1
0x21CA8	-4	-4.0	7.2
0x11CDA	-5	-5.0	6.3
0x116D6	-6	-6.1	6.0
0x20EF7	-7	-7.0	7.5
0x20EEF	-8	-8.0	7.0
0x20EE8	-9	-9.0	6.5
0x300AB	-10	-9.9	6.8
0x208DE	-11	-10.9	5.9
0x208DA	-12	-12.0	5.7
0x208D7	-13	-12.8	5.5
0x308F0	-14	-14.0	6.8
0x300EA	-15	-14.9	6.4
0x300E4	-16	-16.1	6.1
0x308E0	-17	-17.0	5.8
0x300DC	-18	-18.0	5.6
0x300D9	-19	-19.0	5.4
0x300D6	-20	-20.0	5.2

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表 7-2. Typical TX Current and Output Power (2.4GHz, VDDS = 3.0V)

CC1354R10 RGZ at 2.4GHz, VDDS = 3.0V (Measured on CC1354R10EM-7X7-XD7793-XD24)			
txPower	**TX Power Setting (SmartRF Studio)**	**Typical Output Power [dBm]**	**Typical Current Consumption [mA]**
0x802B	5	4.3	9.2
0x661E	4	3.1	8.4
0x4E16	3	1.8	7.8
0x4011	2	0.5	7.2
0x2E5C	1	-1.5	6.6
0x209C	0	-3.6	6.1
0x2050	-3	-6.7	5.4
0x184C	-5	-9.7	5.0
0x184B	-6	-10.1	4.9
0xE8B	-9	-14.0	4.6
0x8CF	-10	-15.1	4.5
0x8CC	-12	-17.9	4.3
0x4C9	-15	-21.3	4.2
0x884	-18	-24.0	4.0
0x883	-20	-26.3	4.0

表 7-3. Typical TX Current and Output Power (868MHz, VDDS = 3.6V)

CC1354R10 RSK at 868MHz, VDDS = 3.6V (Measured on LP-EM-CC1354P10-1)			
txPower	**TX Power Setting (SmartRF Studio)**	**Typical Output Power [dBm]**	**Typical Current Consumption [mA]**
0x13F	14	13.5	25.2
0xBE33	12	12.1	18.3
0x945F	11	11.1	16.0
0x3404	10	10.1	14.2
0x78A3	9	9.2	13.5
0x5499	8	8.2	12.3
0x4693	7	7.3	11.4
0x78EC	6	6.2	11.3
0x62E3	5	5.2	10.3
0x24E77	4	4.0	11.1
0x24E6D	3	3.0	10.2
0x24665	2	2.0	9.4
0x2385F	1	1.0	8.8
0x2325A	0	-0.0	8.2
0x22A56	-1	-1.1	7.7
0x22AB6	-2	-2.0	8.7
0x22AAD	-3	-3.0	8.0
0x224A6	-4	-4.0	7.5
0x21CA1	-5	-4.9	7.1
0x21C9C	-6	-6.0	6.7
0x21CF6	-7	-6.9	7.9
0x30EB8	-8	-8.0	8.0
0x216E7	-9	-8.9	6.8
0x216E1	-10	-10.0	6.4
0x20EDD	-11	-10.8	6.1

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表 7-3. Typical TX Current and Output Power (868MHz, VDDS = 3.6V) (続き)

CC1354R10 RSK at 868MHz, VDDS = 3.6V (Measured on LP-EM-CC1354P10-1)			
txPower	**TX Power Setting (SmartRF Studio)**	**Typical Output Power [dBm]**	**Typical Current Consumption [mA]**
0x20ED9	-12	-11.9	5.8
0x20ED6	-13	-12.8	5.6
0x308EF	-14	-13.9	7.1
0x208D1	-15	-14.8	5.2
0x208FC	-16	-15.8	5.0
0x308DF	-17	-16.8	6.0
0x308DB	-18	-17.9	5.7
0x300D8	-19	-18.8	5.6
0x300D5	-20	19.9	5.3

表 7-4. Typical TX Current and Output Power (2.4GHz, VDDS = 3.0V)

CC1354R10 RSK at 2.4GHz, VDDS = 3.0V (Measured on LP-EM-CC1354P10-1)			
txPower	**TX Power Setting (SmartRF Studio)**	**Typical Output Power [dBm]**	**Typical Current Consumption [mA]**
0x003F	5	4.7	9.4
0x8029	4	3.9	8.7
0x5C1D	3	3.0	8.1
0x4616	2	2.1	7.6
0x3263	1	1.1	7.2
0x2A5E	0	0.2	6.9
0x1CE6	-3	-2.8	6.1
0x1695	-5	-4.6	5.6
0x1693	-6	-5.6	5.4
0x0E8E	-9	-8.6	5.0
0x00D2	-10	-9.9	4.9
0x088A	-12	-12.0	4.6
0x08CC	-15	-14.6	4.4
0x00C9	-18	-17.6	4.3
0x00C7	-20	-20.2	4.1

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7.20.4 RX Performance

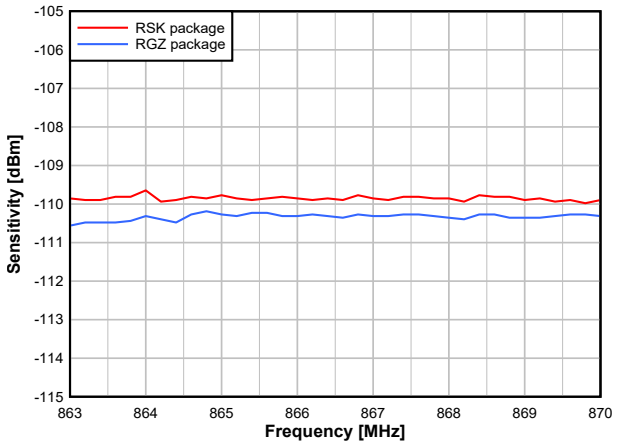


図 7-14. Sensitivity vs. Frequency (50kbps, 868MHz)

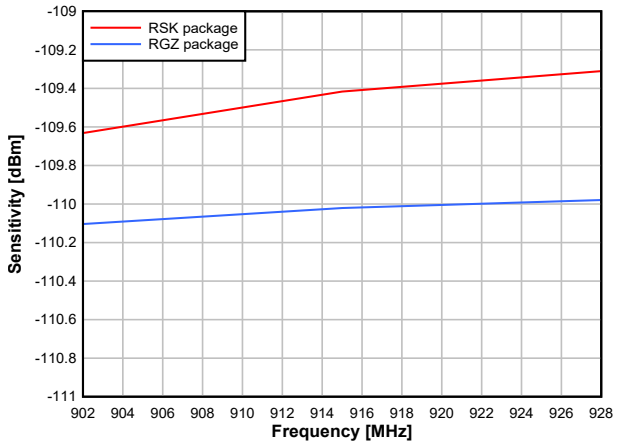


図 7-15. Sensitivity vs. Frequency (50kbps, 915MHz)

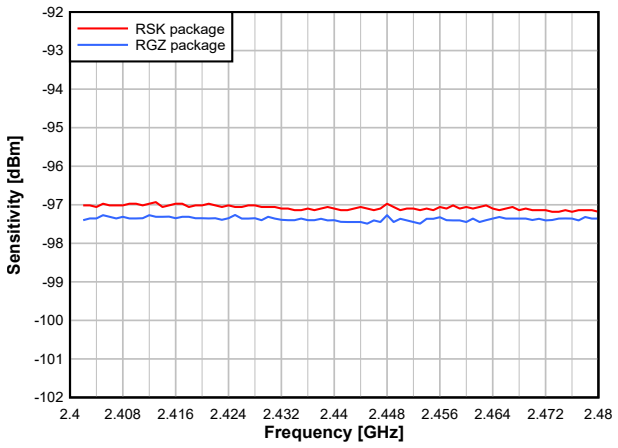


図 7-16. Sensitivity vs. Frequency (BLE 1Mbps, 2.44GHz)

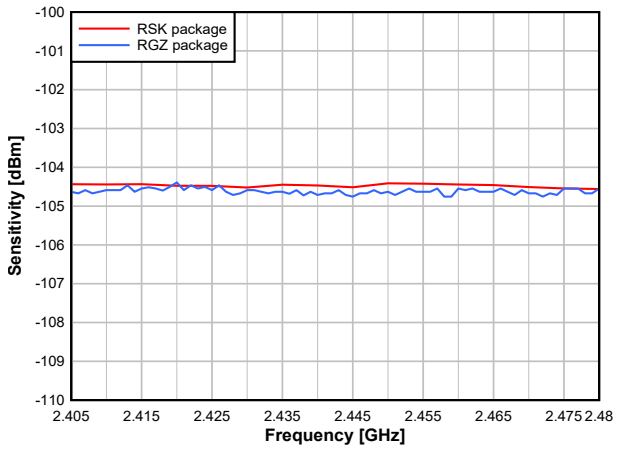


図 7-17. Sensitivity vs. Frequency (IEEE 802.15.4 OQPSK DSSS1:8, 250kbps, 2.44GHz)

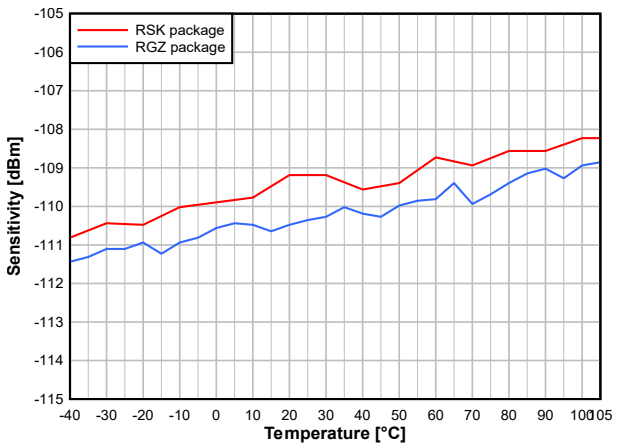


図 7-18. Sensitivity vs. Temperature (50kbps, 868.3MHz)

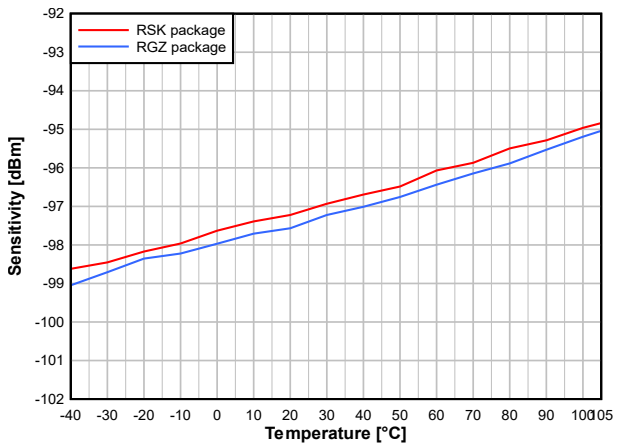


図 7-19. Sensitivity vs. Temperature (BLE 1 Mbps, 2.44GHz)

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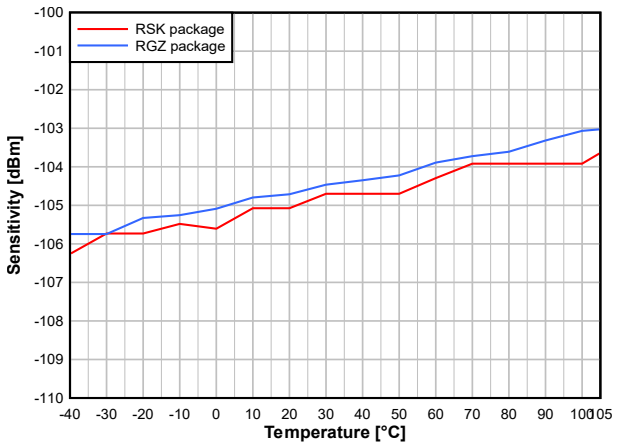


図 7-20. Sensitivity vs. Temperature (250kbps, 2.44GHz)

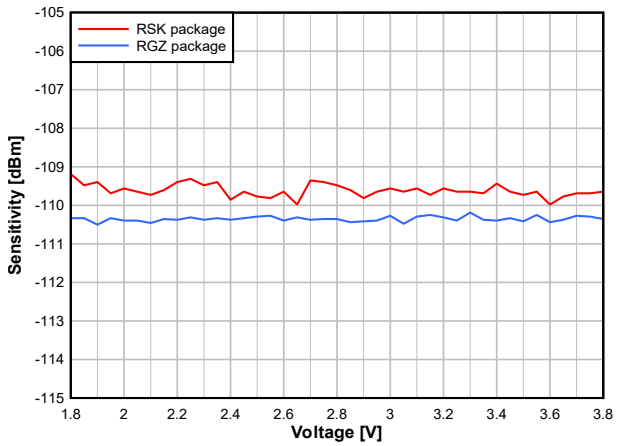


図 7-21. Sensitivity vs. Supply Voltage (VDDS) (50kbps, 868.3MHz)

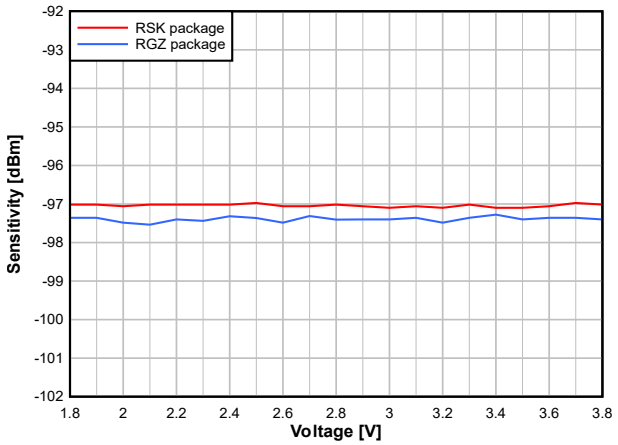


図 7-22. Sensitivity vs. Supply Voltage (VDDS) (BLE 1Mbps, 2.44GHz)

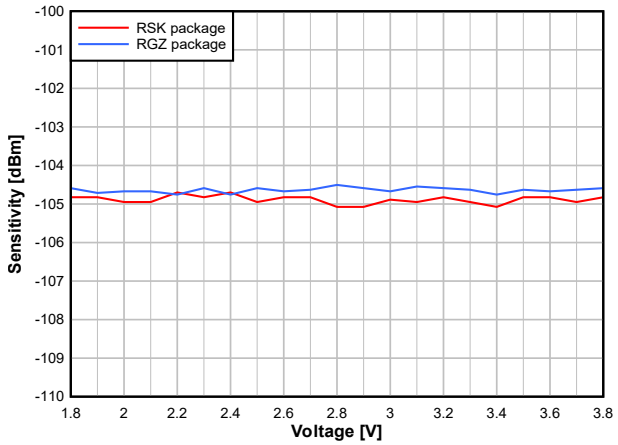


図 7-23. Sensitivity vs. Supply Voltage (VDDS) (250kbps, 2.44GHz)

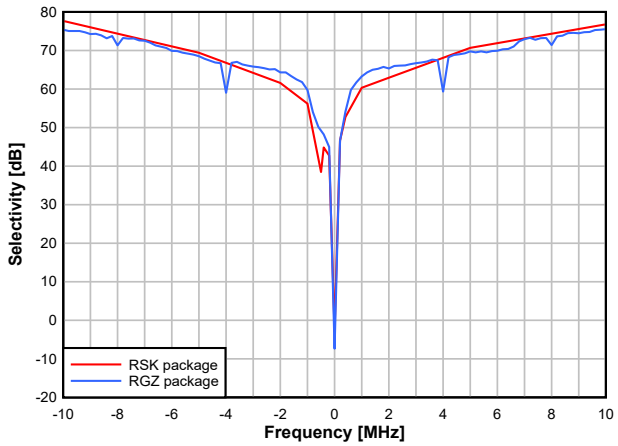


図 7-24. Selectivity vs. Frequency Offset (50kbps, 868.3MHz)

7.20.5 TX Performance

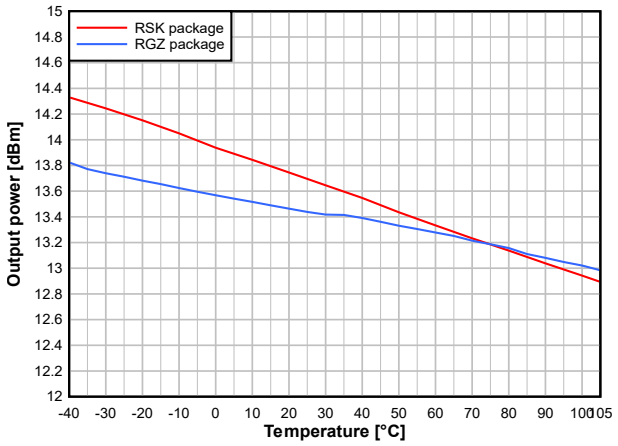


図 7-25. Output Power vs. Temperature (50kbps, 868.3MHz, +14dBm)

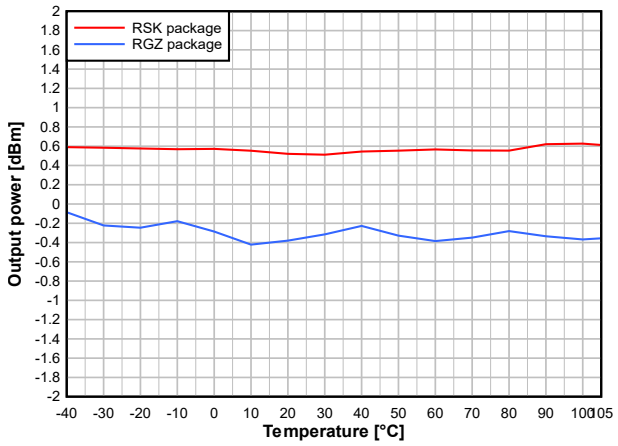


図 7-26. Output Power vs. Temperature (BLE 1Mbps, 2.44GHz, 0dBm)

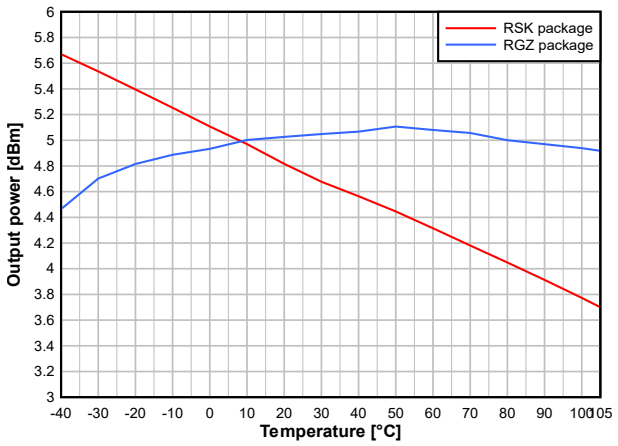


図 7-27. Output Power vs. Temperature (BLE 1Mbps, 2.44GHz, +5dBm)

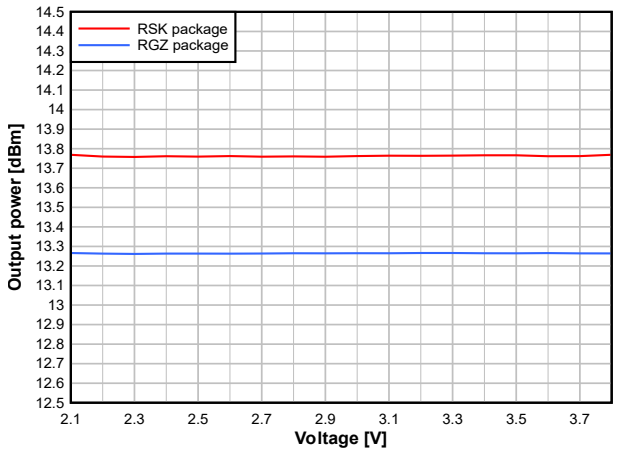


図 7-28. Output Power vs. Supply Voltage (VDDS) (50kbps, 868.3MHz, +14dBm)

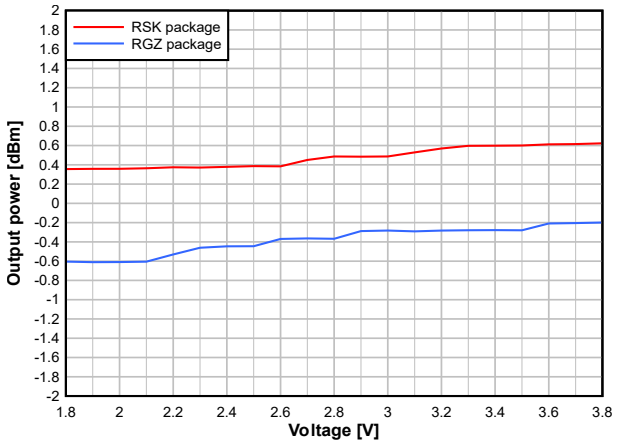


図 7-29. Output Power vs. Supply Voltage (VDDS) (BLE 1Mbps, 2.44GHz, 0dBm)

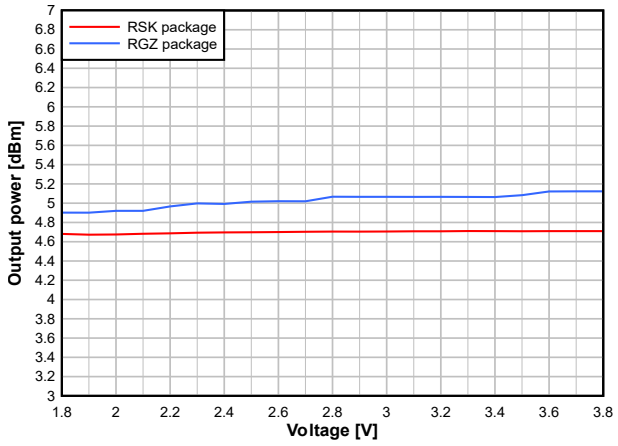


図 7-30. Output Power vs. Supply Voltage (VDDS) (BLE 1Mbps, 2.44GHz, +5dBm)

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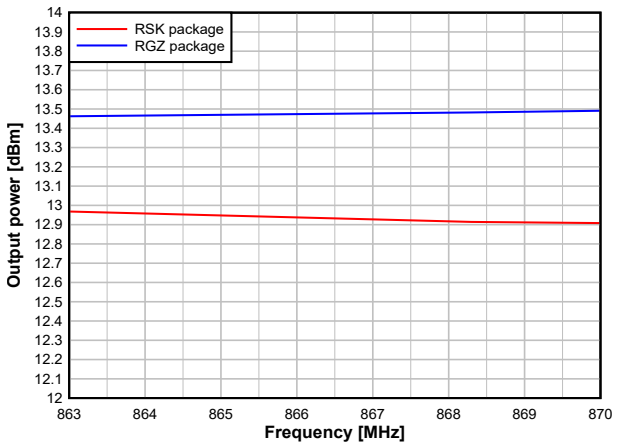


図 7-31. Output Power vs. Frequency (50kbps, 868MHz, +14dBm)

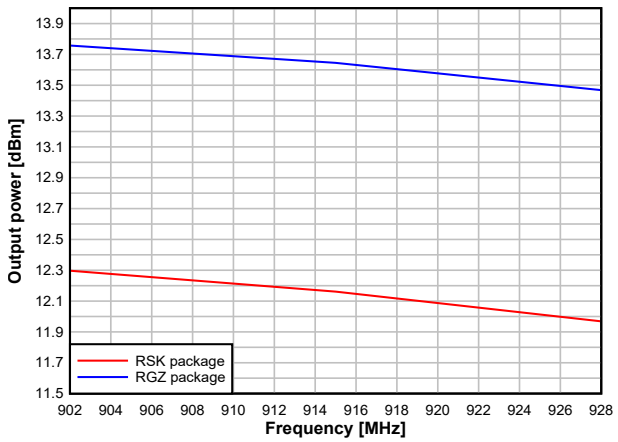


図 7-32. Output Power vs. Frequency (50kbps, 915MHz, +14dBm)

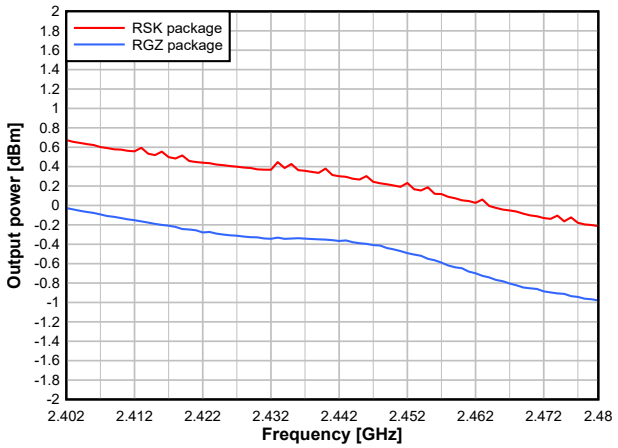


図 7-33. Output Power vs. Frequency (BLE 1Mbps, 2.44GHz, 0dBm)

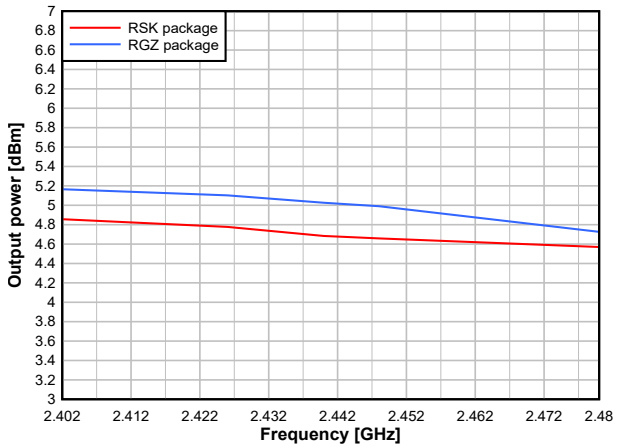


図 7-34. Output Power vs. Frequency (BLE 1Mbps, 2.44GHz, +5dBm)

7.20.6 ADC Performance

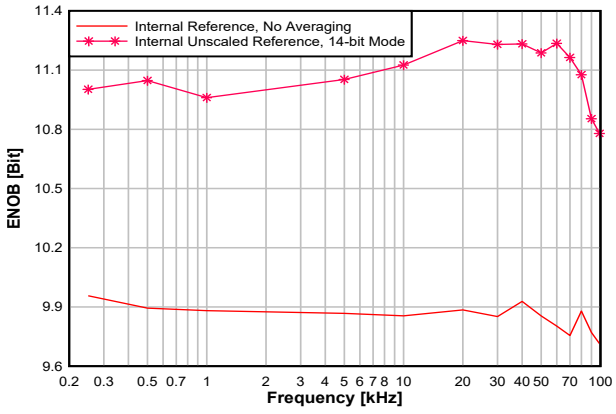


図 7-35. ENOB vs. Input Frequency

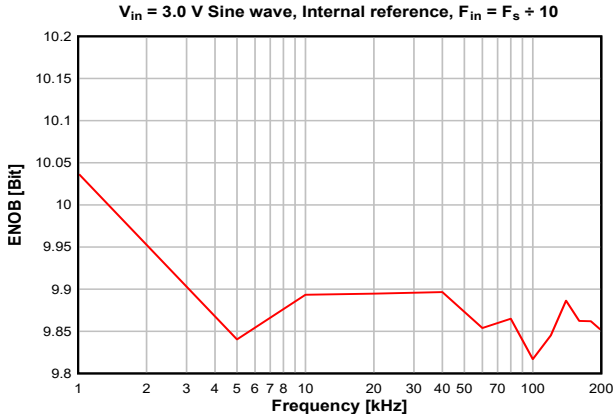


図 7-36. ENOB vs. Sampling Frequency

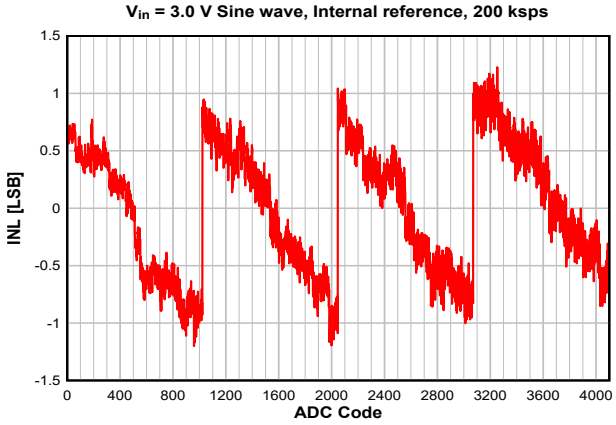


図 7-37. INL vs. ADC Code

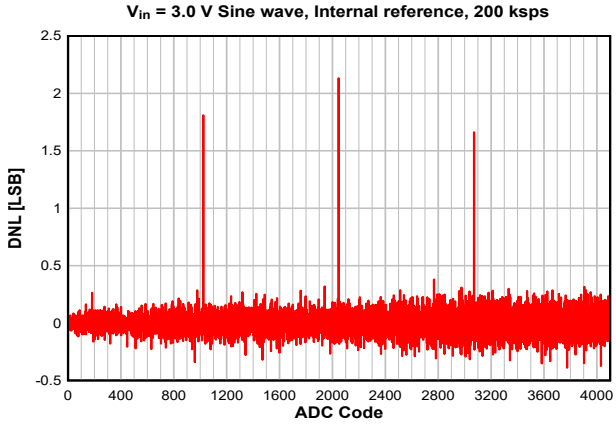


図 7-38. DNL vs. ADC Code

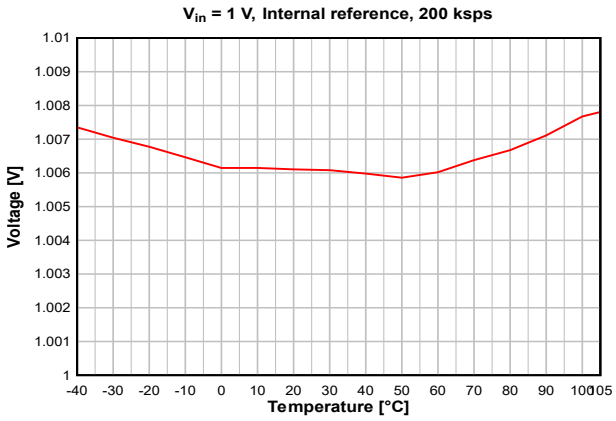


図 7-39. ADC Accuracy vs. Temperature

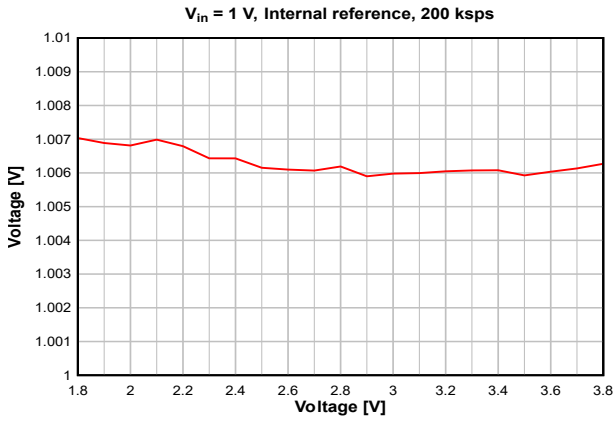


図 7-40. ADC Accuracy vs. Supply Voltage (VDDS)

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8 Detailed Description

8.1 Overview

CC1354R10 ブロック図 shows the core modules of the CC1354R10 device.

Throughout this section, see the Technical Reference Manual listed in Section 11.2 for more details.

8.2 System CPU

The CC1354R10 SimpleLink™ Wireless MCU contains an Arm® Cortex®-M33 system CPU with TrustZone®, which runs the application and the higher layers of radio protocol stacks.

The system CPU is the foundation of a high-performance, low-cost platform that meets the system requirements of minimal memory implementation, and low power consumption while delivering outstanding computational performance and exceptional system response to interrupts.

Its features include the following:

- ARMv8-M architecture with TrustZone® security extension optimized for small-footprint embedded applications
- Arm Thumb®-2 mixed 16- and 32-bit instruction set delivers the high performance expected of a 32-bit Arm core in a compact memory size
- 8 regions of non-secure memory-protected regions
- 8 regions of secure memory-protected regions
- 4 regions of Security Attribute Unit (SAU)
- Single-cycle multiply instruction and hardware divide
- Digital signal processing (DSP) extension
- IEEE 754-compliant single-precision Floating Point Unit (FPU)
- Fast code execution permits increased sleep mode time
- Deterministic, high-performance interrupt handling for time-critical applications
- Full debug with data matching for watchpoint generation
  - Data Watchpoint and Trace Unit (DWT)
  - JTAG Debug Access Port (DAP)
  - Flash Patch and Breakpoint Unit (FPB)
- Trace support reduces the number of pins required for debugging and tracing
  - Instrumentation Trace Macrocell Unit (ITM)
  - Trace Port Interface Unit (TPIU) with asynchronous serial wire output (SWO)
- Optimized for single-cycle flash memory access
- Tightly connected to 8kB 4-way random replacement cache for minimal active power consumption and wait states
- Ultra-low-power consumption with integrated sleep modes
- 48MHz operation

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8.3 Radio (RF Core)

The RF Core is a highly flexible and future-proof radio module that contains an Arm Cortex-M0 processor that interfaces the analog RF and base-band circuitry, handles data to and from the system CPU side, and assembles the information bits in a given packet structure. The RF core offers a high-level, command-based API to the main CPU that configurations and data are passed through. The Arm Cortex-M0 processor is not programmable by customers and is interfaced through the TI-provided RF driver that is included with the SimpleLink Software Development Kit (SDK).

The RF core can autonomously handle the time-critical aspects of the radio protocols, thus offloading the main CPU, which reduces power and leaves more resources for the user application. Several signals are also available to control external circuitry such as RF switches or range extenders autonomously.

Dual-band and multiprotocol solutions are enabled through time-sliced access to the radio, handled transparently for the application through the TI-provided RF driver and dual-mode manager.

The various physical layer radio formats are partly built as a software-defined radio where the radio behavior is either defined by radio ROM contents or by non-ROM radio formats delivered in the form of firmware patches with the SimpleLink SDKs. This allows the radio platform to be updated for support of future versions of standards even with over-the-air (OTA) updates while still using the same silicon.

注

Not all combinations of features, frequencies, data rates, and modulation formats described in this chapter are supported. Over time, TI can enable new physical radio formats (PHYs) for the device and provides performance numbers for selected PHYs in the data sheet. Supported radio formats for a specific device, including optimized settings to use with the TI RF driver, are included in the SmartRF Studio tool with performance numbers of selected formats found in セクション 7.

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8.3.1 Proprietary Radio Formats

The CC1354R10 radio can support a wide range of physical radio formats through a set of hardware peripherals combined with firmware available in the device ROM, covering various customer needs for optimizing parameters such as speed or sensitivity. This allows great flexibility in tuning the radio both to work with legacy protocols as well as customizing the behavior for specific application needs.

表 8-1 gives a simplified overview of features of the various radio formats available in ROM. Other radio formats may be available in the form of radio firmware patches or programs through the software development kit (SDK) and may combine features in a different manner, as well as add other features.

表 8-1. Feature Support

Feature	Main 2-(G)FSK Mode	High Data Rates	Low Data Rates	SimpleLink™ Long Range
Programmable preamble, sync word and CRC	Yes	Yes	Yes	No
Programmable receive bandwidth	Yes	Yes	Yes (down to 4 kHz)	Yes
Data / Symbol rate(3)	20 to 1000kbps	≤ 2 Msps	≤ 100 ksps	≤ 20 ksps
Modulation format	2-(G)FSK	2-(G)FSK 4-(G)FSK	2-(G)FSK 4-(G)FSK	2-(G)FSK
Dual Sync Word	Yes	Yes	No	No
Carrier Sense (1) (2)	Yes	No	No	No
Preamble Detection(2)	Yes	Yes	Yes	No
Data Whitening	Yes	Yes	Yes	Yes
Digital RSSI	Yes	Yes	Yes	Yes
CRC filtering	Yes	Yes	Yes	Yes
Direct-sequence spread spectrum (DSSS)	No	No	No	1:2 1:4 1:8
Forward error correction (FEC)	No	No	No	Yes
Link Quality Indicator (LQI)	Yes	Yes	Yes	Yes

(1) Carrier Sense can be used to implement HW-controlled listen-before-talk (LBT) and Clear Channel Assessment (CCA) for compliance with such requirements in regulatory standards. This is available through the CMD_PROP_CS radio API.
(2) Carrier Sense and Preamble Detection can be used to implement sniff modes where the radio is duty-cycled to save power.
(3) Data rates are only indicative. Data rates outside this range may also be supported. For some specific combinations of settings, a smaller range might be supported.

8.3.2 Bluetooth 5.3 Low Energy

The RF Core offers full support for Bluetooth 5.3 Low Energy, including the high-speed 2Mbps physical layer and the 500kbps and 125kbps long-range PHYs (coded PHY) through the TI-provided Bluetooth 5.3 stack or a high-level Bluetooth API. The Bluetooth 5.3 PHY and part of the controller are in radio and system ROM, providing significant savings in memory usage and more space available for applications.

The new high-speed mode allows data transfers up to 2Mbps, twice the speed of Bluetooth 4.2 and five times the speed of Bluetooth 4.0, without increasing power consumption. In addition to faster speeds, this mode offers significant improvements for energy efficiency and wireless coexistence with reduced radio communication time.

Bluetooth 5.3 also enables unparalleled flexibility for adjustment of speed and range based on application needs, which capitalizes on the high-speed or long-range modes respectively. Data transfers are now possible at 2Mbps, enabling the development of applications using voice, audio, imaging, and data logging that were not previously an option using Bluetooth low energy. With high-speed mode, existing applications deliver faster responses, richer engagement, and longer battery life. Bluetooth 5.3 enables fast, reliable firmware updates.

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8.3.3 802.15.4 Thread, Zigbee, and 6LoWPAN

Through a dedicated IEEE radio API, the RF Core supports the 2.4GHz IEEE 802.15.4-2011 physical layer (2 Mchips per second Offset-QPSK with DSSS 1:8), used in Thread, Zigbee, and 6LoWPAN protocols. The 802.15.4 PHY and MAC are in radio and system ROM. TI also provides royalty-free protocol stacks for Thread and Zigbee as part of the SimpleLink SDK, enabling a robust end-to-end solution.

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8.4 Memory

1024kB nonvolatile (Flash) memory provides storage for code and data in two banks. The flash memory is in-system programmable and erasable. The last flash memory sector must contain a Customer Configuration section (CCFG) that is used by boot ROM and TI-provided drivers to configure the device. This configuration is done through the ccfg.c source file that is included in all TI-provided examples.

The ultra-low leakage system static RAM (SRAM) is split into up to eight 32kB blocks and can be used for both storage of data and execution of code. Retention of SRAM contents in Standby power mode is enabled by default and included in Standby mode power consumption numbers. Parity checking for detection of bit errors in memory is built-in, which reduces chip-level soft errors and thereby increases reliability. Parity can be disabled for an additional 32kB that can be allocated for general-purpose SRAM. System SRAM is always initialized to zeroes upon code execution from boot.

To improve code execution speed and lower power when executing code from nonvolatile memory, a 4-way nonassociative 8kB cache is enabled by default to cache and prefetch instructions read by the system CPU. The cache can be used as a general-purpose RAM by enabling this feature in the Customer Configuration Area (CCFG).

There is a 4kB ultra-low leakage SRAM available for use with the Sensor Controller Engine which is typically used for storing Sensor Controller programs, data, and configuration parameters. This RAM is also accessible by the system CPU. The Sensor Controller RAM is not cleared to zeroes between system resets.

The ROM includes a TI-RTOS kernel and low-level drivers, as well as significant parts of selected radio stacks, which free up flash memory for the application. The ROM also contains a serial (SPI and UART) bootloader that can be used for the initial programming of the device.

8.5 Sensor Controller

The Sensor Controller contains circuitry that can be selectively enabled in both Standby and Active power modes. The peripherals in this domain can be controlled by the Sensor Controller Engine, which is a proprietary power-optimized CPU. This CPU can read and monitor sensors or perform other tasks autonomously; thereby significantly reducing power consumption and offloading the system CPU.

The Sensor Controller Engine is user-programmable with a simple programming language that has syntax similar to C. This programmability allows for sensor polling and other tasks to be specified as sequential algorithms rather than the static configuration of complex peripheral modules, timers, DMA, register programmable state machines, or event routing.

The main advantages are:

- Flexibility—Data can be read and processed in unlimited manners while still ensuring ultra-low power.
- 2MHz low-power mode enables the lowest possible handling of digital sensors
- Dynamic reuse of hardware resources
- 40-bit accumulator supporting multiplication, addition, and shift
- Observability and debugging options

Sensor Controller Studio is used to write, test, and debug code for the Sensor Controller. The tool produces C driver source code, which the System CPU application uses to control and exchange data with the Sensor Controller. Typical use cases may be (but are not limited to) the following:

- Read analog sensors using integrated ADC or comparators
- Interface digital sensors using GPIOs, SPI, UART, or I^2C (I^2C and UART are bit-banged)
- Capacitive sensing
- Waveform generation
- Very low-power pulse counting (flow metering)
- Key scan

The peripherals in the Sensor Controller include the following:

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- The low-power clocked comparator can be used to wake the system CPU from any state in which the comparator is active. A configurable internal reference DAC can be used in conjunction with the comparator. The output of the comparator can also be used to trigger an interrupt or the ADC.
- Capacitive sensing functionality is implemented through the use of a constant current source, a time-to-digital converter, and a comparator. The continuous time comparator in this block can also be used as a higher-accuracy alternative to the low-power clocked comparator. The Sensor Controller takes care of baseline tracking, hysteresis, filtering, and other related functions when these modules are used for capacitive sensing.
- The ADC is a 12-bit 200ksps ADC with eight inputs and a built-in voltage reference. The ADC can be triggered by many different sources including timers, I/O pins, software, and comparators.
- The analog modules can connect to up to eight different GPIOs.
- Dedicated SPI master with up to 6MHz clock speed.

The peripherals in the Sensor Controller can also be controlled from the main application processor.

8.6 Cryptography

The CC1354R10 device comes with a wide set of modern cryptography-related hardware accelerators, drastically reducing code footprint and execution time for cryptographic operations. It also has the benefit of being lower power and improves availability and responsiveness of the system because the cryptography operations runs in a background hardware thread.
Together with a large selection of open-source cryptography libraries provided with the software development kit (SDK), this allows for secure and future proof IoT applications to be easily built on top of the platform. The hardware accelerator modules are:

- **True Random Number Generator (TRNG)** module provides a true, nondeterministic noise source for the purpose of generating keys, initialization vectors (IVs), and other random number requirements. The TRNG is built on 24 ring oscillators that create unpredictable output to feed a complex nonlinear-combinatorial circuit.
- **Secure Hash Algorithm 2 (SHA-2)** with support for SHA224, SHA256, SHA384, and SHA512.
- **Advanced Encryption Standard (AES)** with 128-bit, 192-bit, and 256-bit key lengths.
- **Public Key Accelerator**—Hardware accelerator supporting mathematical operations needed for elliptic curves up to 512 bits.

Through use of these modules and the TI provided cryptography drivers, the following capabilities are available for an application or stack:

- **Key Agreement Schemes**
  - Elliptic Curve Diffie–Hellman with static or ephemeral keys (ECDH and ECDHE)
  - Elliptic curve Password Authenticated Key Exchange by Juggling (ECJ-PAKE)
- **Signature Processing**
  - Elliptic curve Diffie-Hellman Digital Signature Algorithm (ECDSA)
  - Edwards-curve Digital Signature Algorithm (EdDSA)
- **Curve Support**
  - Short Weierstrass form, such as:
    - NIST-P224 (secp224r1), NIST-P256 (secp256r1), NIST-P384 (secp384r1), NIST-P521 (secp521r1)
    - Brainpool-256R1, Brainpool-384R1, Brainpool-512R1
  - Montgomery form, such as:
    - Curve25519
  - Twisted Edwards form, such as:
    - Ed25519
- **Message Authentication Codes**
  - AEC CBC-MAC
  - AES CMAC
  - HMAC with SHA224, SHA256, SHA384, and SHA512
- **Block cipher mode of operation**

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  - AES CCM and AES CCM-Star
  - AES GCM
  - AES ECB
  - AES CBC
  - AES CTR
- **Hash Algorithm**
  - SHA224
  - SHA256
  - SHA384
  - SHA512
- **True random number generation**

Other capabilities, such as RSA encryption and signatures (using keys as large as 2048 bits) as well as other ECC curves such as Curve1174, can be implemented using the provided public key accelerator but are not part of the TI SimpleLink SDK for the CC1354R10 device.

8.7 Timers

A large selection of timers are available as part of the CC1354R10 device. These timers are:

- **Real-Time Clock (RTC)**

  A 70-bit 3-channel timer running on the 32kHz low-frequency system clock (SCLK_LF). This timer is available in all power modes except Shutdown. The timer can be calibrated to compensate for frequency drift when using the LF RCOSC as the low-frequency system clock. If an external LF clock with a frequency different from 32.768kHz is used, the RTC tick speed can be adjusted to compensate for this. When using TI-RTOS, the RTC is used as the base timer in the operating system and should thus only be accessed through the kernel APIs such as the Clock module. The real-time clock can also be read by the Sensor Controller Engine to timestamp sensor data and also has dedicated capture channels. By default, the RTC halts when a debugger halts the device.

- **General Purpose Timers (GPTIMER)**

  The four flexible GPTIMERs can be used as either 4 × 32-bit timers or 8 × 16-bit timers, all running on up to 48MHz. Each of the 16- or 32-bit timers supports a wide range of features such as one-shot or periodic counting, pulse width modulation (PWM), time counting between edges, and edge counting. The inputs and outputs of the timer are connected to the device event fabric, which allows the timers to interact with signals such as GPIO inputs, other timers, DMA, and ADC. The GPTIMERs are available in Active and Idle power modes.

- **Sensor Controller Timers**

  The Sensor Controller contains three timers:

  The Sensor Controller contains three timers: AUX Timers 0 and 1 are 16-bit timers with a 2^N prescaler. Timers can either increment on a clock or each edge of a selected tick source. Both one-shot and periodical timer modes are available.

  AUX Timer 2 is a 16-bit timer that can operate at 24MHz, 2MHz, or 32kHz independent of the Sensor Controller functionality. There are four capture or compare channels, which can be operated in one-shot or periodical modes. The timer can be used to generate events for the Sensor Controller Engine or the ADC, as well as for PWM output or waveform generation.

- **Radio Timer**

  A multichannel 32-bit timer running at 4MHz is available as part of the device radio. The radio timer is typically used as the timing base in wireless network communication using the 32-bit timing word as the network time. The radio timer is synchronized with the RTC by using a dedicated radio API when the device radio is turned on or off. This ensures that for a network stack, the radio timer seems to always be running when the radio is enabled. The radio timer is in most cases used indirectly through the trigger time fields in

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the radio APIs and should only be used when running the accurate 48MHz high-frequency crystal is the source of SCLK_HF.

- **Watchdog Timer**

  The watchdog timer is used to regain control if the system operates incorrectly due to software errors. It is typically used to generate an interrupt and reset the device for the case where periodic monitoring of the system components and tasks fails to verify proper functionality. The watchdog timer runs on a 1.5MHz clock rate and cannot be stopped once enabled. The watchdog timer continues to run in Standby power mode but pauses when a debugger halts the device.

- **Always On Watchdog Timer (AON_WDT)**

  The Always On Watchdog Timer is used during standby to regain control when the system has failed due to a software error or failure of an external device to respond in the expected way. It generates a reset when its configured time-out counter reaches zero and cannot be stopped once started, unless by asserting a device reset. The Always-on watchdog timer runs in Standby power mode and may pause when a debugger halts the device.

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8.8 Serial Peripherals and I/O

The SPI interface provides a standardized synchronous serial interface to communicate with devices compatible with SPI (3 and 4 wire), MICROWIRE and TI Synchronous Serial Format. The SPIs support master/slave operation up to 12MHz, programmable clock bit rate with prescaler, as well as configurable phase and polarity.

The UART interface implements universal asynchronous receiver and transmitter functions. The UART supports flexible baud-rate generation up to a maximum of 3Mbps with FIFO, multiple data sizes, stop, and parity bits as well as hardware handshake.

The I^2S interface provides a standardized interface to exchange digital audio with devices compatible with this standard, including ADCs, DACs, and CODECs. The I^2S can also receive pulse-density modulation (PDM) data from devices such as digital microphones and perform conversion to PCM data.

The I^2C interface enables low-speed serial communications with devices compatible with the I^2C standard. The I^2C interface can handle both standard (100kHz) and fast (400kHz) speeds, as well as four modes of operation: master transmit/receive and slave transmit/receive.

The I/O controller (IOC) controls the digital I/O pins and contains multiplexer circuitry to allow a set of peripherals to be assigned to I/O pins in a flexible manner. All digital I/Os are interrupt and wake-up capable, have a programmable pullup and pulldown function, and can generate an interrupt on a negative or positive edge (configurable). When configured as an output, pins can function as either push-pull or open-drain. Five GPIOs have high-drive capabilities, which are marked in **bold** in セクション 6. All digital peripherals can be connected to any digital pin on the device.

8.9 Battery and Temperature Monitor

A combined temperature and battery voltage monitor is available in the CC1354R10 device. The battery and temperature monitor allows an application to continuously monitor on-chip temperature and supply voltage and respond to changes in environmental conditions as needed. The module contains window comparators to interrupt the system CPU when temperature or supply voltage goes outside defined windows. These events can also be used to wake up the device from Standby mode through the always-on (AON) event fabric.

8.10 µDMA

The device includes a direct memory access (µDMA) controller. The µDMA controller provides a way to offload data-transfer tasks from the system CPU, thus allowing for more efficient use of the processor and the available bus bandwidth. The µDMA controller can perform a transfer between memory and peripherals. The µDMA controller has dedicated channels for each supported on-chip module and can be programmed to automatically perform transfers between peripherals and memory when the peripheral is ready to transfer more data.

Some features of the µDMA controller include the following (this is not an exhaustive list):

- Highly flexible and configurable channel operation of up to 32 channels
- Transfer modes: memory-to-memory, memory-to-peripheral, peripheral-to-memory, and peripheral-to-peripheral
- Data sizes of 8 bits, 16 bits, and 32 bits
- Ping-pong mode for continuous streaming of data

8.11 Debug

The debug subsystem implements two IEEE standards for debug and test purposes:

IEEE 1149.7 Class 4: Reduced-pin and Enhanced-functionality Test Access Port and Boundary-scan Architecture. This is known by the acronym cJTAG (compact JTAG) and this device uses only two pins to communicate with the target: TMS (JTAG_TMSC) and TCK (JTAG_TCKC). This is the default mode of operation.

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IEEE standard 1149.1: Test Access Port and Boundary Scan Architecture Test Access Port (TAP). This standard is known by the acronym JTAG and this device uses four pins to communicate with the target: TMS (JTAG_TMSC), TCK (JTAG_TCKC), TDI (JTAG_TDI), and TDO (JTAG_TDO).

The debug subsystem also implements a user-configurable firewall to control unauthorized access to debug/test ports.

Also featured is **EnergyTrace**/**EnergyTrace++**. This technology implements an improved method for measuring MCU current consumption, which features a very high dynamic range (from sub-µA to hundreds of mA), high sample rate (up to 256 kSamples/s), and the ability to track the CPU and peripheral power states.

Two modes of operation can be configured. **EnergyTrace** measures the overall MCU current consumption and allows maximum accuracy and speed to track ultra-low-power states as well as the fast power transitions during radio transmission and reception. **EnergyTrace++** tracks the various power states of both the CPU and its Peripherals as well as the system clocks, allowing close monitoring of the overall device activity.

8.12 Power Management

To minimize power consumption, the CC1354R10 supports a number of power modes and power management features (see 表 8-2).

表 8-2. Power Modes

MODE	SOFTWARE CONFIGURABLE POWER MODES				RESET PIN HELD
	ACTIVE	IDLE	STANDBY	SHUTDOWN	
CPU	Active	Off	Off	Off	Off
Flash	On	Available	Off	Off	Off
SRAM	On	On	Retention	Off	Off
Supply System	On	On	Duty Cycled	Off	Off
Register and CPU retention	Full	Full	Partial	No	No
SRAM retention	Full	Full	Full	No	No
48MHz high-speed clock (SCLK_HF)	XOSC_HF or RCOSC_HF	XOSC_HF or RCOSC_HF	Off	Off	Off
2MHz medium-speed clock (SCLK_MF)	RCOSC_MF	RCOSC_MF	Available	Off	Off
32kHz low-speed clock (SCLK_LF)	XOSC_LF or RCOSC_LF	XOSC_LF or RCOSC_LF	XOSC_LF or RCOSC_LF	Off	Off
Peripherals	Available	Available	Off	Off	Off
Sensor Controller	Available	Available	Available	Off	Off
Wake-up on RTC	Available	Available	Available	Off	Off
Wake-up on pin edge	Available	Available	Available	Available	Off
Wake-up on reset pin	On	On	On	On	On
Brownout detector (BOD)	On	On	Duty Cycled	Off	Off
Power-on reset (POR)	On	On	On	Off	Off
Watchdog timer (WDT)	Available	Available	Paused	Off	Off
Always-on Watchdog timer (AON_WDT)	Available	Available	Available	Off	Off

In **Active** mode, the application system CPU is actively executing code. Active mode provides normal operation of the processor and all of the peripherals that are currently enabled. The system clock can be any available clock source (see 表 8-2).

In **Idle** mode, all active peripherals can be clocked, but the Application CPU core and memory are not clocked and no code is executed. Any interrupt event brings the processor back into active mode.

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In **Standby** mode, only the always-on (AON) domain is active. An external wake-up event, RTC event, or Sensor Controller event is required to bring the device back to active mode. MCU peripherals with retention do not need to be reconfigured when waking up again, and the CPU continues execution from where it went into standby mode. All GPIOs are latched in standby mode.

In **Shutdown** mode, the device is entirely turned off (including the AON domain and Sensor Controller), and the I/Os are latched with the value they had before entering shutdown mode. A change of state on any I/O pin defined as a *wake from shutdown pin* wakes up the device and functions as a reset trigger. The CPU can differentiate between reset in this way and reset-by-reset pin or power-on reset by reading the reset status register. The only state retained in this mode is the latched I/O state and the flash memory contents.

The Sensor Controller is an autonomous processor that can control the peripherals in the Sensor Controller independently of the system CPU. This means that the system CPU does not have to wake up, for example, to perform an ADC sampling or poll a digital sensor over SPI, thus saving both current and wake-up time that would otherwise be wasted. The Sensor Controller Studio tool enables the user to program the Sensor Controller, control its peripherals, and wake up the system CPU as needed. All Sensor Controller peripherals can also be controlled by the system CPU.

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The power, RF, and clock management for the CC1354R10 device require specific configuration and handling by software for optimized performance. This configuration and handling is implemented in the TI-provided drivers that are part of the CC1354R10 software development kit (SDK). Therefore, TI highly recommends using this software framework for all application development on the device. The complete SDK with TI-RTOS (optional), device drivers, and examples is offered free of charge in the source code.

8.13 Clock Systems

The CC1354R10 device has several internal system clocks.

The 48MHz SCLK_HF is used as the main system (MCU and peripherals) clock. This can be driven by the internal 48MHz RC Oscillator (RCOSC_HF) or an external 48MHz crystal (XOSC_HF). Radio operation requires an external 48MHz crystal.

SCLK_MF is an internal 2MHz clock that is used by the Sensor Controller in low-power mode and also for internal power management circuitry. The SCLK_MF clock is always driven by the internal 2MHz RC oscillator (RCOSC_MF).

SCLK_LF is the 32.768kHz internal low-frequency system clock. It can be used by the Sensor Controller for ultra-low-power operation and is also used for the RTC and to synchronize the radio timer before or after Standby power mode. SCLK_LF can be driven by the internal 32.8kHz RC Oscillator (RCOSC_LF), a 32.768kHz watch-type crystal, or a clock input on any digital IO.

When using a crystal or the internal RC oscillator, the device can output the 32kHz SCLK_LF signal to other devices, thereby reducing the overall system cost.

8.14 Network Processor

Depending on the product configuration, the CC1354R10 device can function as a wireless network processor (WNP), a device running the wireless protocol stack with the application running on a separate host MCU, or as a system-on-chip (SoC) with the application and protocol stack running on the system CPU inside the device.

In the first case, the external host MCU communicates with the device using SPI or UART. In the second case, the application must be written according to the application framework supplied with the wireless protocol stack.

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9 Application, Implementation, and Layout

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Information in the following Applications section is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

For general design guidelines and hardware configuration guidelines, refer to *CC13xx/CC26xx Hardware Configuration and PCB Design Considerations Application Report*.

9.1 Reference Designs

The following reference designs should be followed closely when implementing designs using the CC1354R10 device.

Special attention must be paid to RF component placement, decoupling capacitors, and DC/DC regulator components, as well as ground connections for all of these.

CC1352REM-XD7793-XD24 Design Files	The CC1352REM-XD7793-XD24 reference design provides schematic, layout, and production files for the characterization board used for deriving the performance number found in this document.
LP-EM-CC1354P10-1 Design Files	Detailed schematics and layouts for the multiband CC1354P10 LaunchPad evaluation board featuring 868/915MHz RF matching on the 20dBm PA output and up to 5dBm TX power at 2.4GHz.
LP-EM-CC1354P10-6 Design Files	Detailed schematics and layouts for the multiband CC1354P10 LaunchPad evaluation board featuring 2.4GHz RF matching optimized for 10dBm operation on the 20dBm PA output and up to 13dBm TX power at 433MHz.
Sub-1 GHz and 2.4 GHz Antenna Kit for LaunchPad™ Development Kit and SensorTag	The antenna kit allows real-life testing to identify the optimal antenna for your application. The antenna kit includes 16 antennas for frequencies from 169MHz to 2.4GHz, including: • PCB antennas • Helical antennas • Chip antennas • Dual-band antennas for 868MHz and 915MHz combined with 2.4GHz The antenna kit includes a JSC cable to connect to the Wireless MCU LaunchPad Development Kits and SensorTags.

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9.2 Junction Temperature Calculation

This section shows the different techniques for calculating the junction temperature under various operating conditions. For more details, see Semiconductor and IC Package Thermal Metrics.

There are three recommended ways to derive the junction temperature from other measured temperatures:

1. From package temperature:

$$T_J = \psi_{\mathrm{JT}} \times P + T_{\mathrm{case}} \quad (1)$$

2. From board temperature:

$$T_J = \psi_{\mathrm{JB}} \times P + T_{\mathrm{board}} \quad (2)$$

3. From ambient temperature:

$$T_J = R_{\theta\mathrm{JA}} \times P + T_A \quad (3)$$

P is the power dissipated from the device and can be calculated by multiplying current consumption with supply voltage. Thermal resistance coefficients are found in セクション 7.8.

$$T_J = 23.4^{\circ C}/_{W} \times 51.8mW + T_A = 1.2^{\circ}C + T_A \quad (4)$$

For various application use cases, current consumption for other modules may have to be added to calculate the appropriate power dissipation. For example, the MCU may be running simultaneously as the radio, peripheral modules may be enabled, and so on. Typically, the easiest way to find the peak current consumption, and thus the peak power dissipation in the device, is to measure as described in Measuring CC13xx and CC26xx current consumption.

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10 Device and Documentation Support

TI offers an extensive line of development tools. Tools and software to evaluate the performance of the device, generate code, and develop solutions are listed as follows.

10.1 Tools and Software

The CC1354R10 device is supported by a variety of software and hardware development tools.

Development Kit

CC1354P10-1 LaunchPad™ Development Kit	The CC1354P10-1 LaunchPad™ Development Kit enables the development of high-performance wireless applications in the 863MHz to 930MHz and 2.4GHz frequency bands that benefit from low-power operation. The kit features the CC1354P10 multiband and multiprotocol SimpleLink™ Wireless MCU with an integrated High-Power Amplifier. The kit works with the LaunchPad ecosystem, easily enabling additional functionality like sensors, displays, and more. The RF configuration of the LaunchPad enables up to +20dBm output power for 863MHz to 930MHz and +5dBm output power for 2.4GHz.
CC1354P10-6 LaunchPad™ Development Kit	The CC1354P10-6 LaunchPad™ Development Kit enables the development of high-performance wireless applications in the 863MHz to 930MHz and 2.4GHz frequency bands that benefit from low-power operation. The kit features the CC1354P10 multiband and multiprotocol SimpleLink™ Wireless MCU with an integrated High-Power Amplifier. The kit works with the LaunchPad ecosystem, easily enabling additional functionality like sensors, displays, and more. The built-in EnergyTrace™ software is an energy-based code analysis tool that measures and displays the application's energy profile and helps to optimize it for ultra-low power consumption. The RF configuration of the LaunchPad enables up to +14dBm output power for 863MHz to 930MHz and +20dBm output power for 2.4GHz.
LP-XDS110 LaunchPad™ Debug Probe	The LP-XDS110 LaunchPad™ Debug Probe enables the development of high-performance wireless applications in the entire family of LP-EM LaunchPad™ development boards. Featuring a seamless connection with the new 20-pin LP-EM Debug connector, it supports not only multiple standards such as JTAG/cJTAG/SWD but also a UART backchannel for maximum debugging flexibility. It also features an Arm® 10-pin Debug connector to perform debugging on any custom board.
LP-XDS110ET LaunchPad™ Debug Probe	The LP-XDS110ET LaunchPad™ Debug Probe enables the development of high-performance wireless applications in the entire family of LP-EM LaunchPad™ development boards. Featuring a seamless connection with the new 20-pin LP-EM Debug connector, it supports not only multiple standards such as JTAG/cJTAG/SWD but also a UART backchannel for maximum debugging flexibility. In addition, it also features an Arm® 10-pin Debug connector to perform debugging on any custom board. This Debug Probe also features the XDS110 EnergyTrace™ technology, which is a new method for measuring the current consumption that captures the complete operational profile of the wireless MCU.
TMDSEMU110-U Debug Probe	The TMDSEMU110-U Debug Probe enables the development of high-performance wireless applications in the entire family of SimpleLink™ LaunchPad™ development boards. Featuring a convenient enclosure, which grants the proper mechanical robustness for field and production environments, it supports not only multiple standards such as JTAG/cJTAG/SWD but also a UART backchannel and four GPIOs for maximum debugging flexibility. In addition, the expansion connector allows using the TMDSEMU110-ETH add-on (sold separately), which adds the full-featured XDS110 EnergyTrace™ technology with

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variable supply voltage from 1.8V to 3.6V and up to 800mA of supply current. The XDS110 EnergyTrace™ technology is a new method for measuring the current consumption that captures the complete operational profile of the wireless MCU.

Software

SimpleLink™ LOWPOWER F2 SDK

The SimpleLink™ LOWPOWER F2 Software Development Kit (SDK) provides a complete package for the development of wireless applications on the CC13XX / CC26XX family of devices. The SDK includes a comprehensive software package for the CC1354R10 device, including the following protocol stacks:

- Bluetooth Low Energy 4 and 5.3
- Thread (based on OpenThread)
- TI Z-Stack (Zigbee 3.0)
- TI 15.4-Stack—an IEEE 802.15.4-based star networking solution for Sub-1GHz and 2.4GHz
- EasyLink - a large set of building blocks for building proprietary RF software stacks
- Multiprotocol support—concurrent operation between stacks using the Dynamic Multiprotocol Manager (DMM)
- TI Wi-SUN FAN Stack
- Matter

The SimpleLink™ LOWPOWER F2 SDK is part of TI's SimpleLink™ MCU platform, offering a single development environment that delivers flexible hardware, software, and tool options for customers developing wired and wireless applications. For more information about the SimpleLink™ MCU Platform, visit ti.com/simplelink.

Development Tools

Code Composer Studio™ Integrated Development Environment (IDE)

Code Composer Studio is an integrated development environment (IDE) that supports TI's Microcontroller and Embedded Processors portfolio. Code Composer Studio comprises a suite of tools used to develop and debug embedded applications. It includes an optimizing C/C++ compiler, source code editor, project build environment, debugger, profiler, and many other features. The intuitive IDE provides a single user interface taking you through each step of the application development flow. Familiar tools and interfaces allow users to get started faster than ever before. Code Composer Studio combines the advantages of the Eclipse® software framework with advanced embedded debug capabilities from TI resulting in a compelling feature-rich development environment for embedded developers.

CCS has support for all SimpleLink™ Wireless MCUs and includes support for EnergyTrace™ software (application energy usage profiling). A real-time object viewer plugin is available for TI-RTOS, part of the SimpleLink™ SDK.

Code Composer Studio is provided free of charge when used in conjunction with the XDS debuggers included on a LaunchPad Development Kit.

Code Composer Studio™ Cloud IDE

Code Composer Studio (CCS) Cloud is a web-based IDE that allows you to create, edit, and build CCS and Energia™ projects. After you have successfully built your project, you can download and run on your connected LaunchPad. Basic debugging, including features like setting breakpoints and viewing variable values, is now supported with CCS Cloud.

IAR Embedded Workbench® for Arm®

IAR Embedded Workbench® is a set of development tools for building and debugging embedded system applications using Assembler, C, and C++. It provides a completely integrated development environment that includes a project manager, editor, and build tools. IAR has support for all SimpleLink™ Wireless MCUs. It offers broad debugger support,

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including XDS110, IAR I-jet™, and Segger J-Link™. A real-time object viewer plugin is available for TI-RTOS, part of the SimpleLink™ SDK. IAR is also supported out-of-the-box on most software examples provided as part of the SimpleLink™ SDK.

A 30-day evaluation or a 32kB size-limited version is available through iar.com.

SmartRF™ Studio 7

SmartRF™ Studio 7 is a Windows® application that can be used to evaluate and configure SimpleLink™ Wireless MCUs from Texas Instruments. The application will help designers of RF systems to easily evaluate the radio at an early stage in the design process. It is especially useful for the generation of configuration register values and for practical testing and debugging of the RF system. SmartRF Studio can be used either as a standalone application or together with applicable evaluation boards or debug probes for the RF device. Features of the SmartRF Studio include:

- Link tests—send and receive packets between nodes
- Antenna and radiation tests—set the radio in continuous wave TX and RX states
- Export radio configuration code for use with the TI SimpleLink™ SDK RF driver
- Custom GPIO configuration for signaling and control of external switches

Sensor Controller Studio

Sensor Controller Studio is used to write, test, and debug code for the Sensor Controller peripheral. The tool generates a Sensor Controller Interface driver, which is a set of C source files that are compiled into the System CPU application. These source files also contain the Sensor Controller binary image and allow the System CPU application to control and exchange data with the Sensor Controller. Features of the Sensor Controller Studio include:

- Ready-to-use examples for several common use cases
- Full toolchain with built-in compiler and assembler for programming in a C-like programming language
- Provides rapid development by using the integrated sensor controller task testing and debugging functionality, including visualization of sensor data and verification of algorithms

UniFlash

UniFlash is a standalone tool used to program on-chip flash memory on TI MCUs. UniFlash has a GUI, command line, and scripting interface. UniFlash is available free of charge.

10.1.1 SimpleLink™ Microcontroller Platform

The SimpleLink™ microcontroller platform sets a new standard for developers with the broadest portfolio of wired and wireless Arm® MCUs (System-on-Chip) in a single software development environment. Delivering flexible hardware, software and tool options for your IoT applications. Invest once in the SimpleLink™ software development kit and use throughout your entire portfolio. Learn more on ti.com/simplelink.

10.2 Documentation Support

To receive notification of documentation updates on data sheets, errata, application notes and similar, navigate to the device product folder on ti.com/product/CC1354R10 . In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

The current documentation that describes the MCU, related peripherals, and other technical collateral is listed as follows.

TI Resource Explorer

TI Resource Explorer Software examples, libraries, executables, and documentation are available for your device and development board.

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Errata

CC1354R10 Silicon Errata	The silicon errata describes the known exceptions to the functional specifications for each silicon revision of the device and description on how to recognize a device revision.

Application Reports

All application reports for the CC1354R10 device are found on the device product folder at: ti.com/product/CC1354R10/technicaldocuments .

Technical Reference Manual (TRM)

CC13x4, CC26x4 SimpleLink™ Wireless MCU Technical Reference Manual	The TRM provides detailed descriptions of all modules and peripherals available in the device family.

10.3 サポート・リソース

テキサス・インスツルメンツ E2E™ サポート・フォーラムは、エンジニアが検証済みの回答と設計に関するヒントをエキスパートから迅速かつ直接得ることができる場所です。既存の回答を検索したり、独自の質問をしたりすることで、設計で必要な支援を迅速に得ることができます。

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10.5 静電気放電に関する注意事項



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ESD による破損は、わずかな性能低下からデバイスの完全な故障まで多岐にわたります。精密な IC の場合、パラメータがわずかに変化するだけで公表されている仕様から外れる可能性があるため、破損が発生しやすくなっています。

10.6 用語集

テキサス・インスツルメンツ用語集　この用語集には、用語や略語の一覧および定義が記載されています。

11 Revision History

資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

Changes from June 20, 2023 to April 30, 2024 (from Revision A (June 2023) to Revision B (April 2024)) **Page**

- RSK パッケージの暫定情報の脚注を削除。 1
- Updated Device Comparison table 6
- Updated Sensor controller power consumption in セクション 7.5, *Power Consumption - Power Modes* 14
- Updated Flash specifications in セクション 7.7, *Nonvolatile (Flash) Memory Characteristics* 14
- Removed redundant blocking and selectivity rows in both *Generic OOK (16.384kbps, OOK w / Manchester encoding, 100kHz RX BW)* and *802.15.4, 50kbps, ±25kHz Deviation, 2-GFSK, 100kHz RX BW (Legacy)* of セクション 7.10, *861MHz to 1054MHz - Receive (RX)* 14
- Fixed typo in test conditions for Symbol Rate Tolerance in both *802.15.4-2020, 10kbps, 2-FSK, 26kHz RX BW, Mode #1a* and *802.15.4-2020, 20kbps, 2-FSK, 52 kHz RX BW, Mode #1b* of セクション 7.10, *861MHz to 1054MHz - Receive (RX)* 14
- Fixed typo in test conditions for Frequency Error Tolerance (ppm) in *802.15.4, 50kbps, ±25kHz Deviation, 2-GFSK, 100kHz RX BW (Legacy)* of セクション 7.10, *861MHz to 1054MHz - Receive (RX)* 14
- Merged parameter cells for Blocking -10MHz in *WB-DSSS, 240/120/60/30kbps (480 ksym/s, 2-GFSK, ±195kHz Deviation, FEC (Half Rate), DSSS = 1/2/4/8, 622 kHz RX BW)* of セクション 7.10, *861MHz to 1054MHz - Receive (RX)* 14
- Fixed incorrectly merged rows of test conditions for both Blocking +5% Fc. and Image rejection parameters in *802.15.4-2020, 10kbps, ±5kHz deviation, 2-FSK, 26kHz RX BW, Mode #1a* of セクション 7.10, *861MHz to 1054MHz - Receive (RX)* 14
- Updated test conditions for Frequency Error Tolerance in section *SimpleLink™ Long Range 2.5/5kbps (20 ksps), ±5kHz Deviation, 2-GFSK, 34 kHz RX Bandwidth, FEC = 1:2, DSSS = 1:4/1:2* of セクション 7.10, *861MHz to 1054MHz - Receive (RX)* 14
- Removed redundant Image Rejection row in *802.15.4, 50kbps, ±25kHz Deviation, 2-GFSK, 100kHz RX BW (Legacy)* of セクション 7.10, *861MHz to 1054MHz - Receive (RX)* 14
- Updated graphs and tables on Typical characteristics 44
- Added EnergyTrace information to セクション 8.11, *Debug* 63

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12 Mechanical, Packaging, and Orderable Information

12.1 Packaging Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package | Pins	Package qty | Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
CC1354R106T0RGZR	Active	Production	VQFN (RGZ) | 48	2500 | LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 105	CC1354 R106
CC1354R106T0RGZR.B	Active	Production	VQFN (RGZ) | 48	2500 | LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	See CC1354R106T0RGZR	CC1354 R106
CC1354R106T0RSKR	Active	Production	VQFN (RSK) | 64	2000 | LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 105	CC1354 R106
CC1354R106T0RSKR.B	Active	Production	VQFN (RSK) | 64	2000 | LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	See CC1354R106T0RSKR	CC1354 R106

(1) **Status:** For more details on status, see our product life cycle.

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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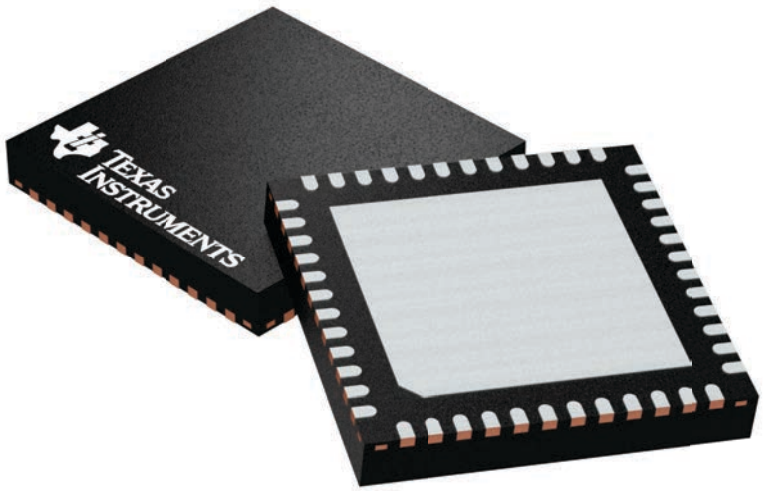
GENERIC PACKAGE VIEW

RGZ 48

VQFN - 1 mm max height

7 x 7, 0.5 mm pitch

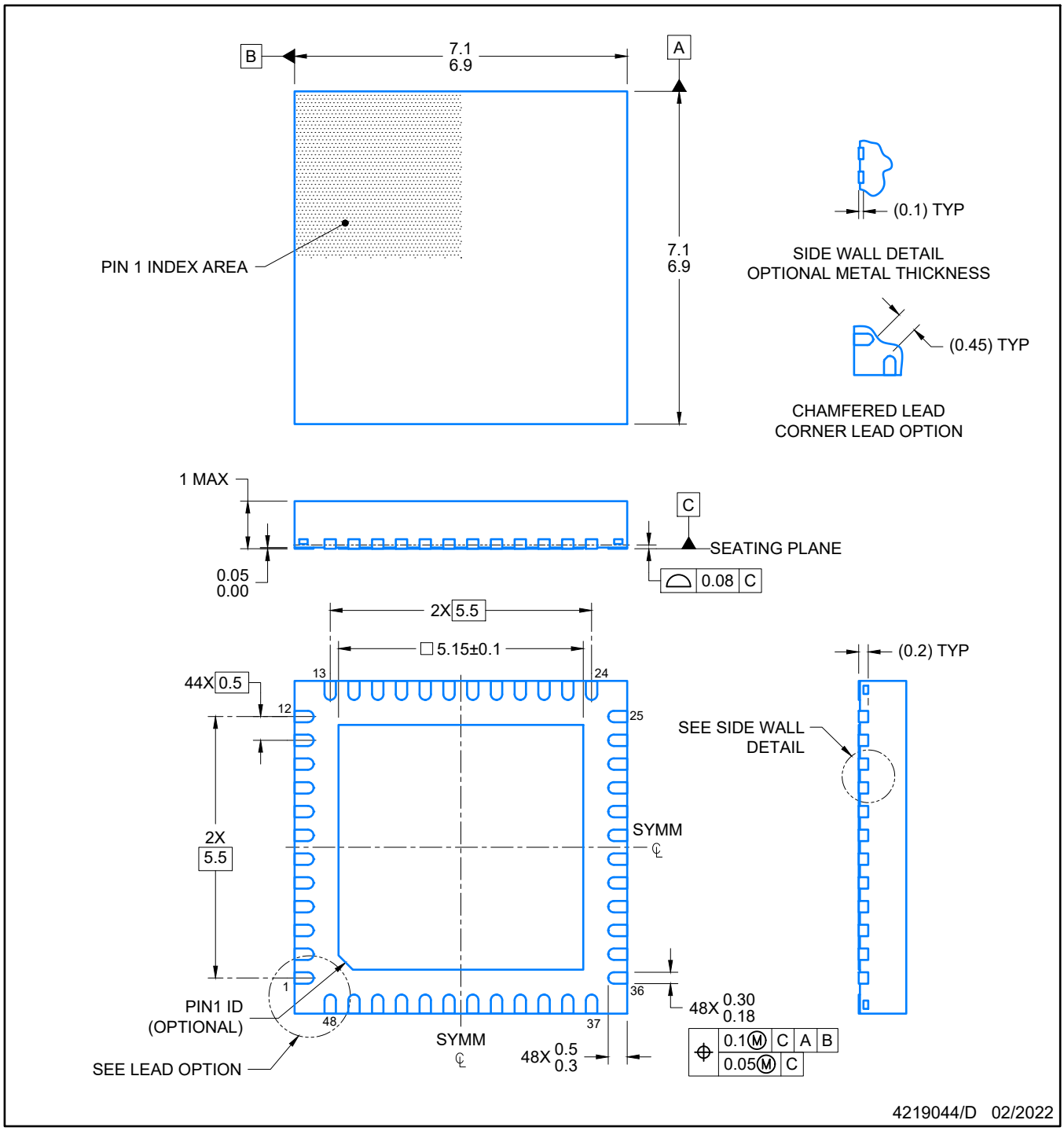
PLASTIC QUADFLAT PACK- NO LEAD



Images above are just a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.

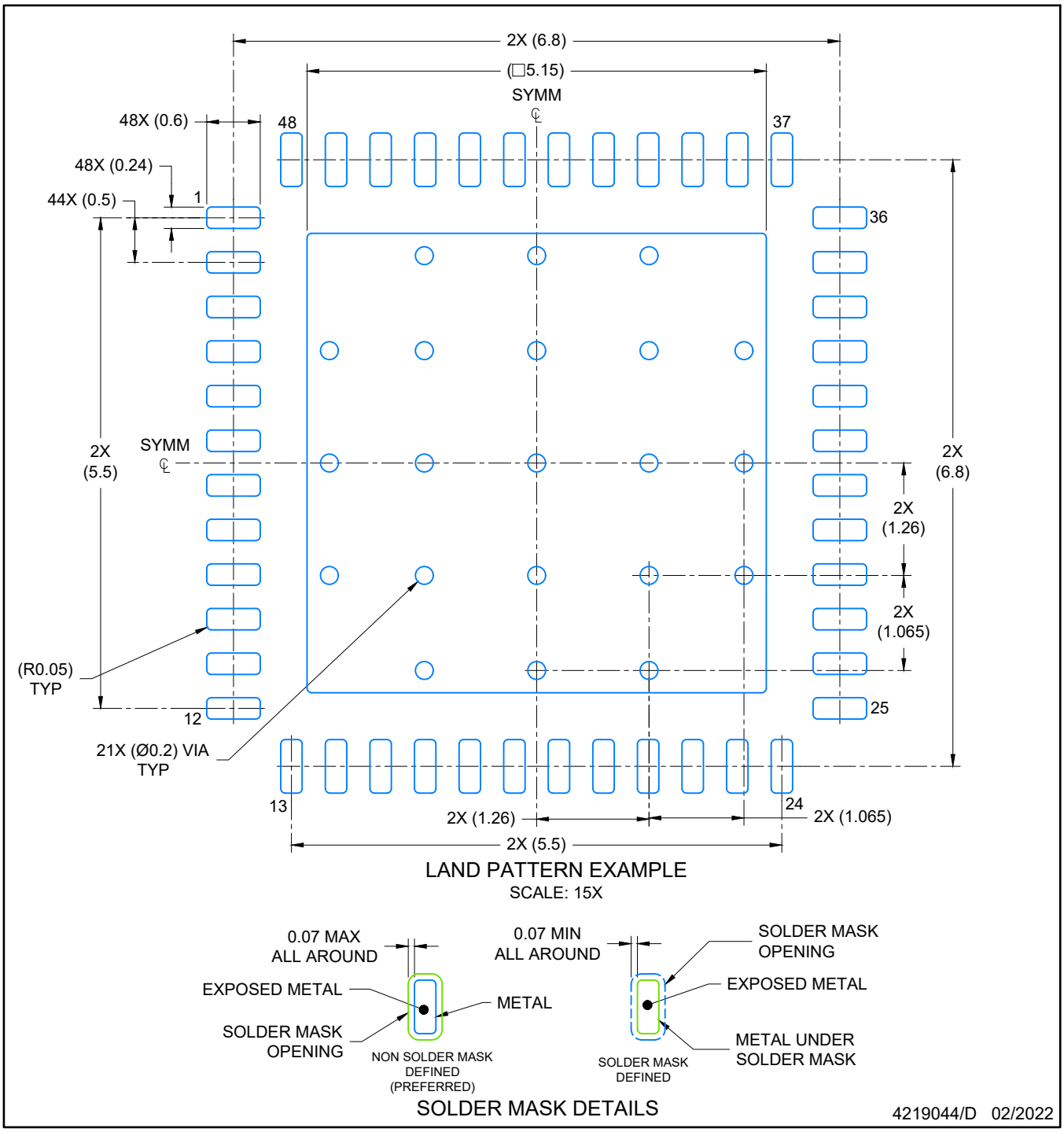
4224671/A

PLASTIC QUADFLAT PACK- NO LEAD



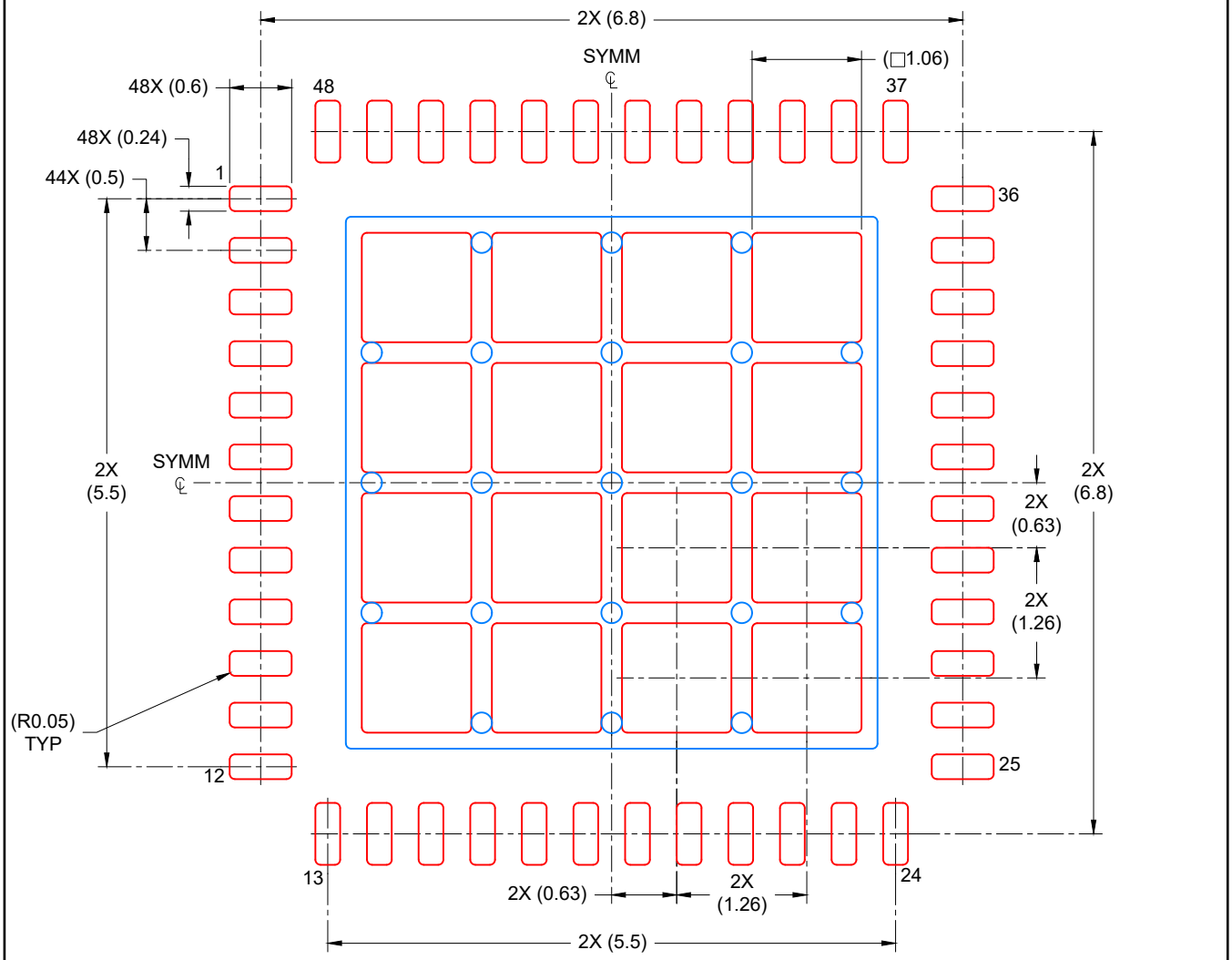
NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for optimal thermal and mechanical performance.



NOTES: (continued)

4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD
67% PRINTED COVERAGE BY AREA
SCALE: 15X

4219044/D 02/2022

NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

GENERIC PACKAGE VIEW

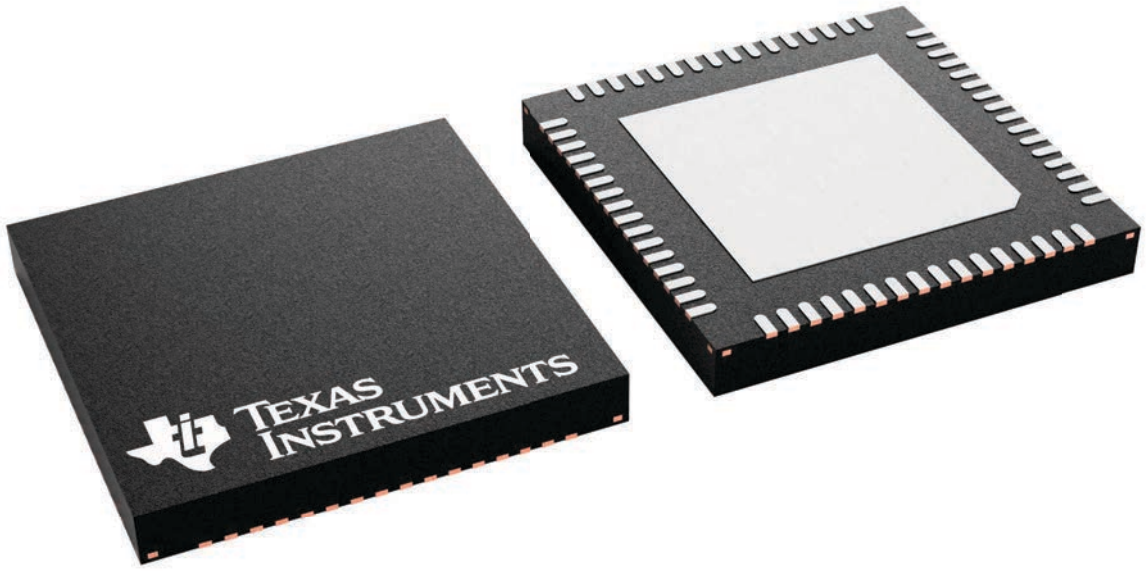
RSK 64

VQFN - 1 mm max height

8 x 8, 0.4 mm pitch

PLASTIC QUAD FLATPACK - NO LEAD

This image is a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.



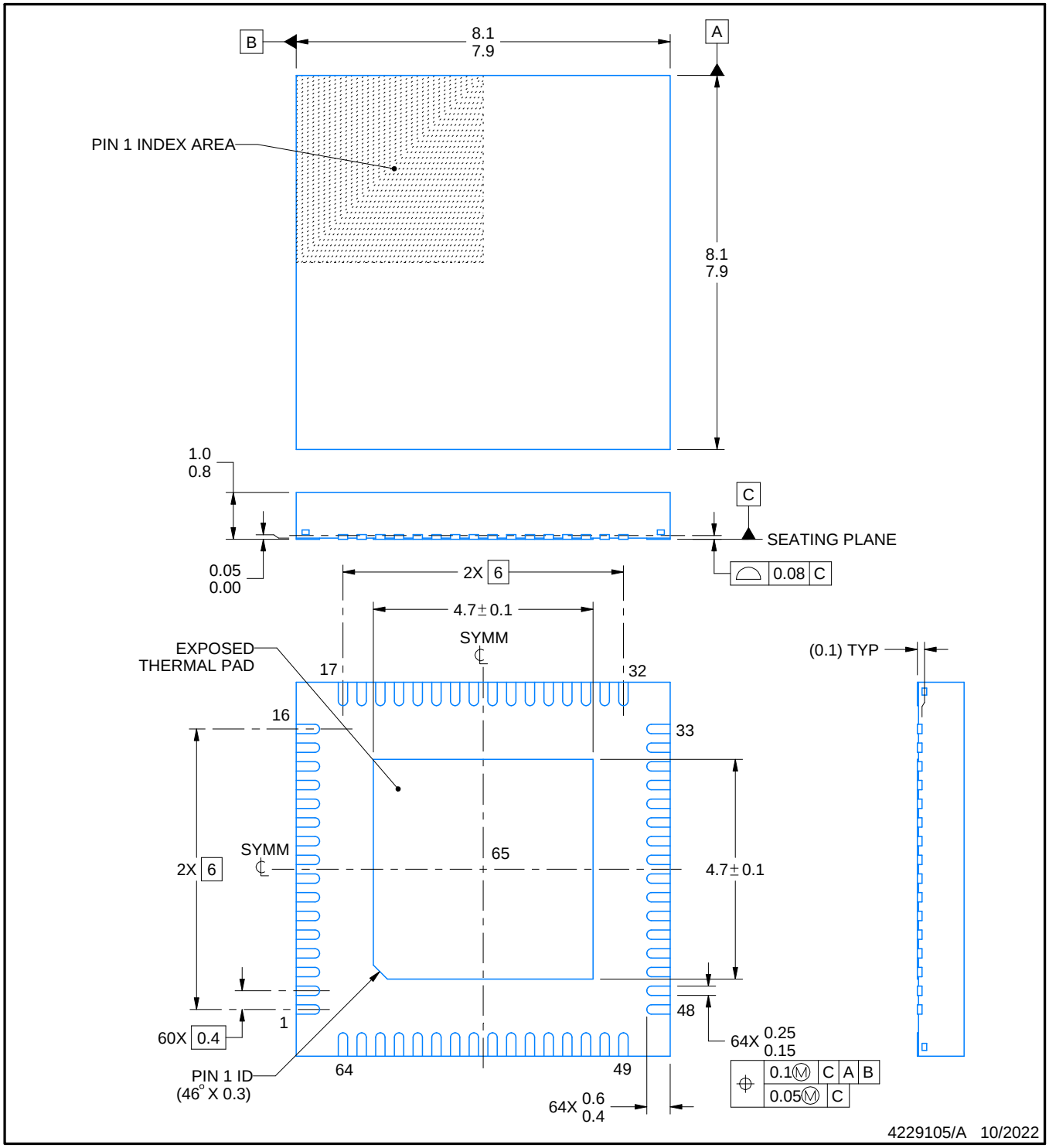
4231455/A

RSK0064D

PACKAGE OUTLINE

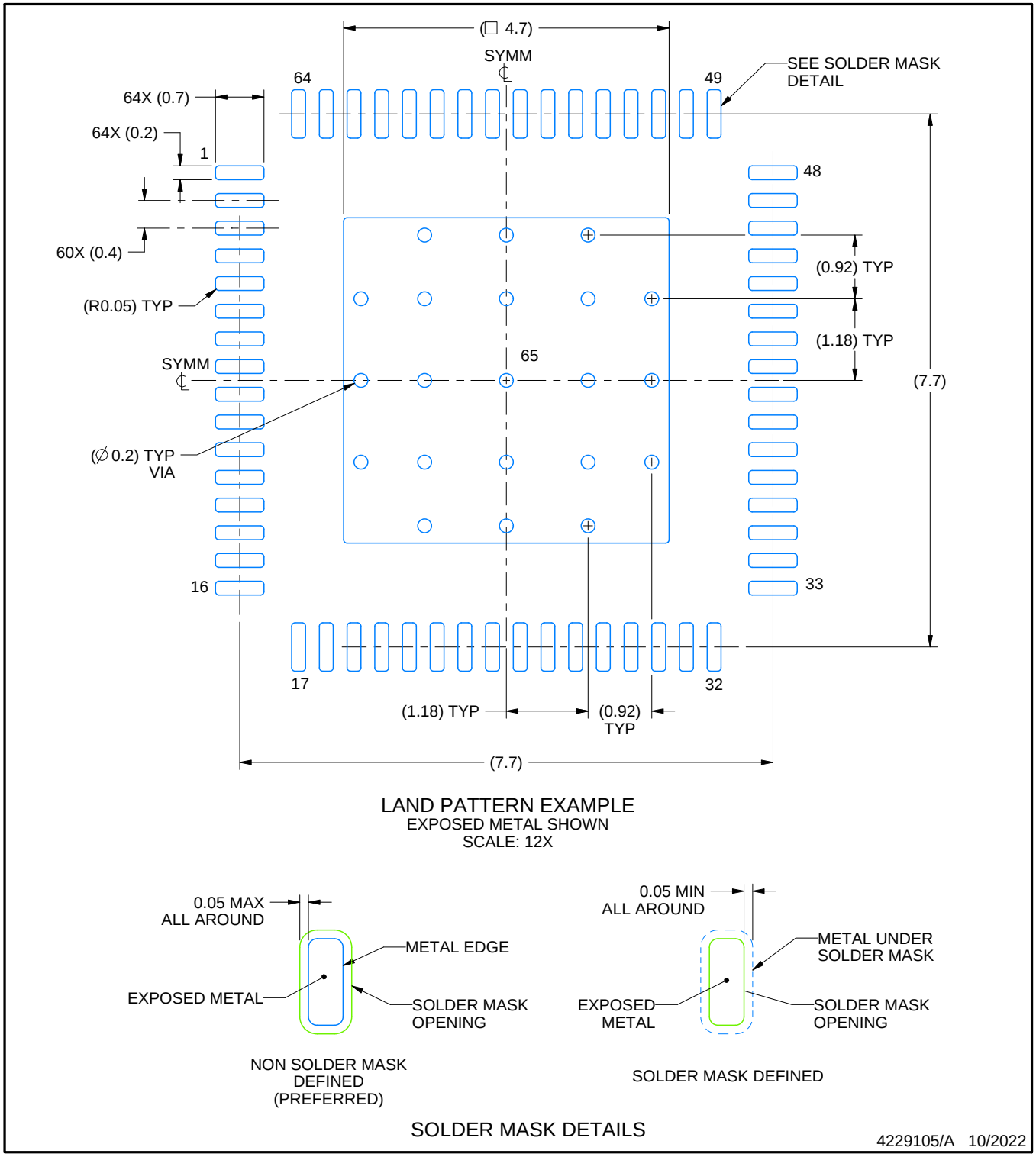
VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



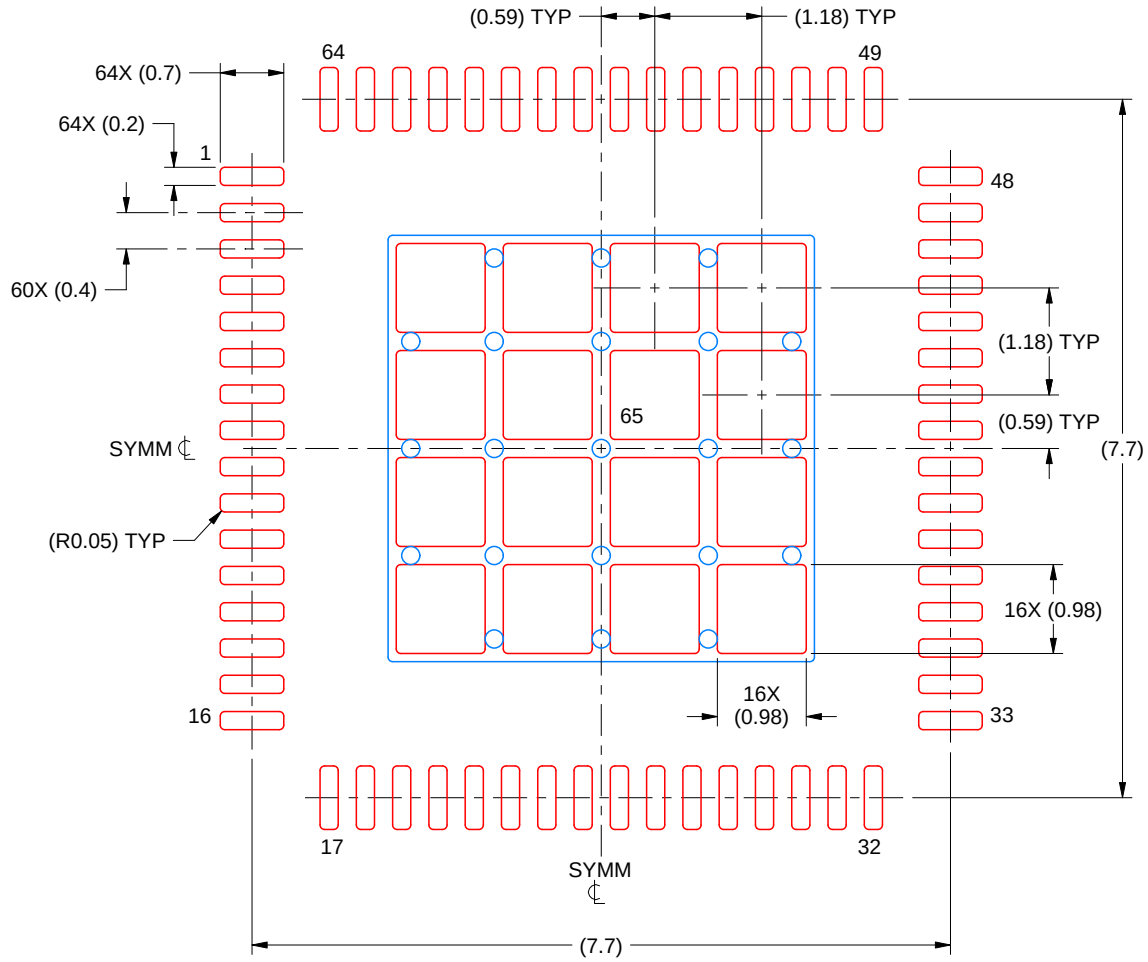
NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.



NOTES: (continued)

4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



SOLDER PASTE EXAMPLE
BASED ON 0.100 MM THICK STENCIL
SCALE: 12X

EXPOSED PAD 65
70% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE

4229105/A 10/2022

NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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