TEXAS INSTRUMENTS
CC2652P7
JAJSNG1A – MAY 2021 – REVISED NOVEMBER 2021

CC2652P7 SimpleLink™ パワー・アンプ内蔵マルチプロトコル 2.4GHz ワイヤレス MCU

1 特長

ワイヤレス・マイコン

- 強力な 48MHz Arm® Cortex®-M4F プロセッサ
- 704KB フラッシュ・プログラム・メモリ
- プロトコルおよびライブラリ機能用の 256KB ROM
- 8KB のキャッシュ SRAM
- 高信頼性動作に適した 144KB の超低リーク SRAM (パリティ付き)
- ダイナミック・マルチプロトコル・マネージャ (DMM) ドライバ
- プログラマブルな無線機能には、2-(G)FSK、4-(G)FSK、MSK、*Bluetooth*® 5.2 Low Energy、IEEE 802.15.4 PHY、MAC のサポートが含まれます。
- OTA (Over-The-Air) アップグレードに対応

超低消費電力センサ・コントローラ

- 4KB の SRAM を備えた自律型 MCU
- センサ・データのサンプリング、保存、処理
- 高速ウェイクアップにより低消費電力動作を実現
- ソフトウェア定義ペリフェラル、静電容量式タッチ、流量計、LCD

低い消費電力

- MCU の消費電流:
  - 3.10mA (アクティブ・モード、CoreMark)
  - 65µA/MHz (CoreMark 実行中)
  - 0.9µA (スタンバイ・モード、RTC、144KB RAM)
  - 0.1µA (シャットダウン・モード、ウェイクアップ・オン・ピン)
- 超低消費電力センサ・コントローラの消費電流:
  - 29.2µA (2MHz モード)
  - 799µA (24MHz モード)
- 無線の消費電流:
  - 6.4mA (RX)
  - 21mA (TX、+10dBm)
  - 101mA (TX、+20dBm)

無線プロトコルのサポート

- Thread、Zigbee®、Matter
- *Bluetooth*® 5.2 Low Energy
- SimpleLink™ TI 15.4 スタック
- 6LoWPAN
- 独自システム

高性能の無線

- -104dBm (*Bluetooth*® Low Energy、125kbps)
- 温度補償付きで最大 +20dBm の出力電力

法規制の順守

- 以下の規格への準拠を目的としたシステムに最適:
  - ETSI EN 300 328、EN 300 440 Cat.2 および 3
  - FCC CFR47 Part 15
  - ARIB STD-T66

マイコンのペリフェラル

- デジタル・ペリフェラルを任意の GPIO に接続可能
- 4 つの 32 ビットまたは 8 つの 16 ビット汎用タイマ
- 12 ビット ADC、200k サンプル/秒、8 チャネル
- 8 ビット DAC
- 2 つのコンパレータ
- プログラマブルな電流ソース
- 2 つの UART、2 つの SSI、I^2C、I^2S
- リアルタイム・クロック (RTC)
- 温度およびバッテリ・モニタを内蔵

セキュリティ・イネーブラ (実現機能)

- AES 128/256 ビット暗号化アクセラレータ
- ECC および RSA 公開鍵ハードウェア・アクセラレータ
- SHA2 アクセラレータ (SHA-512 までのフル・スイート)
- TRNG (True Random Number Generator)

開発ツールとソフトウェア

- LP-CC1352P7 開発キット
- SimpleLink™ CC13xx/CC26xx ソフトウェア開発キット (SDK)
- SmartRF™ Studio による容易な無線構成
- Sensor Controller Studio により低消費電力のセンシング・アプリケーションを構築
- SysConfig システム・コンフィギュレーション・ツール

動作範囲

- オンチップの降圧型 DC/DC コンバータ
- 1.8V～3.8V のシングル電源電圧
- -40～+105°C

パッケージ

- 7mm × 7mm RGZ VQFN48 (26GPIO)
- RoHS 準拠のパッケージ

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2 アプリケーション

- 2400～2500MHz の ISM および SRD システム [1] (最小 4kHz の受信帯域幅)
- ビル・オートメーション
  - ビル・セキュリティ・システム – モーション検出器、電子スマート・ロック、ドアおよび窓センサ、ガレージ・ドア・システム、ゲートウェイ
  - HVAC – サーモスタット、ワイヤレス環境センサ、HVAC システム・コントローラ、ゲートウェイ
  - 防火システム – 煙および熱感知器、火災警報制御パネル (FACP)
  - ビデオ監視 – IP ネットワーク・カメラ
  - エレベータとエスカレータ – エレベータおよびエスカレータのためのエレベータ・メイン制御パネル
- 産業用輸送 – アセット・トラッキング
- ファクトリ・オートメーション / 制御
- 医療用
- 電子 POS (EPOS) – 電子棚札 (ESL)
- 通信機器
  - 有線ネットワーク – 無線 LAN または Wi-Fi アクセス・ポイント、エッジ・ルータ 、小規模企業向けルータ
- パーソナル・エレクトロニクス
  - ホーム・シアター / エンターテインメント – スマート・スピーカ、スマート・ディスプレイ、セット・トップ・ボックス
  - ウェアラブル (医療以外) – スマート・トラッカー、スマート衣料

3 概要

SimpleLink™ CC2652P7 デバイスはマルチプロトコル 2.4GHz ワイヤレス・マイクロコントローラ (MCU) です。本デバイスは Thread、Zigbee®、Matter、*Bluetooth*® 5.2 Low Energy、IEEE 802.15.4g、IPv6 対応スマート・オブジェクト (6LoWPAN)、TI 15.4 スタック (2.4GHz)、DMM (ダイナミック・マルチプロトコル・マネージャ) ドライバを使った同時マルチプロトコルをサポートしています。CC2652P7 は、Arm® Cortex® M4F メイン・プロセッサを基に設計されており、グリッド・インフラストラクチャ、ビル・オートメーション、リテール・オートメーション、パーソナル・エレクトロニクス、医療用アプリケーションの低消費電力の無線通信と先進センシングに最適化されています。

CC2652P7 は、Arm® Cortex® M0 で実行するソフトウェア無線を内蔵しているため、各種の物理層と RF 規格をサポートできます。本デバイスは、2360～2500MHz の周波数帯域での動作をサポートしています。PHY と周波数の切り換えは、DMM (ダイナミック・マルチプロトコル・マネージャ) ドライバを使って実行時に行うことができます。CC2652P7 は、2.4GHz 帯で +10dBm TX (21mA)、+20dBm TX (101mA) をサポートする高効率 PA を内蔵しています。 CC2652P7 は -104dBm (125kbps の Bluetooth® Low Energy Coded PHY) の受信感度を実現しています。

CC2652P7 は 0.9µA の低スリープ電流 (RTC 動作、144KB RAM を保持) を実現しています。メインの Cortex® M4F プロセッサに加えて、高速ウェイクアップ機能を備えた自律型超低消費電力センサ・コントローラ CPU も内蔵しています。たとえば、このセンサ・コントローラは、1µA のシステム電流で 1Hz の ADC サンプリングが可能です。

CC2652P7 は、動作寿命を伸ばすために低 SER (ソフト・エラー・レート) FIT (Failure-in-time) を実現しています。常時オンの SRAM パリティにより、潜在的な放射線イベントによる破損のリスクを最小限に抑えています。多くのお客様の 10～15 年またはそれ以上の長いライフ・サイクル要件に沿って、TI は、製品寿命と製品供給の継続性を約束する製品ライフ・サイクル・ポリシーを制定しています。

CC2652P7 デバイスは SimpleLink™ MCU プラットフォームの一部です。このプラットフォームは Wi-Fi®、*Bluetooth*® Low Energy、Thread、Zigbee、Wi-SUN®、Amazon Sidewalk、mioty®、Sub-1GHz MCU、ホスト MCU で構成されます。 CC2652P7 は、32KB～704KB のフラッシュ・サイズに対応し、ピン互換パッケージを選択可能なスケーラブルなポートフォリオの一部です。共通の SimpleLink™ CC13xx および CC26xx ソフトウェア開発キット (SDK) と SysConfig システム・コンフィギュレーション・ツールはポートフォリオ内のデバイス間の移行を支援します。多数のソフトウェア・スタック、アプリケーション例、SimpleLink™ Academy トレーニング・セッションが本 SDK に含まれます。詳細については、ワイヤレス・コネクティビティ製品をご覧ください。

製品情報

部品番号(1)	パッケージ	本体サイズ (公称)
CC2652P74T0RGZR	VQFN (48)	7.00mm × 7.00mm

(1) 提供中の全デバイスに関する最新の製品、パッケージ、および注文情報についてはセクション 11 のパッケージ・オプションについての付録、または TI Web サイトを参照してください。

[1] サポートしているプロトコル規格、変調フォーマット、データ・レートの詳細については、*RF コア*を参照してください。

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3.1 機能ブロック図

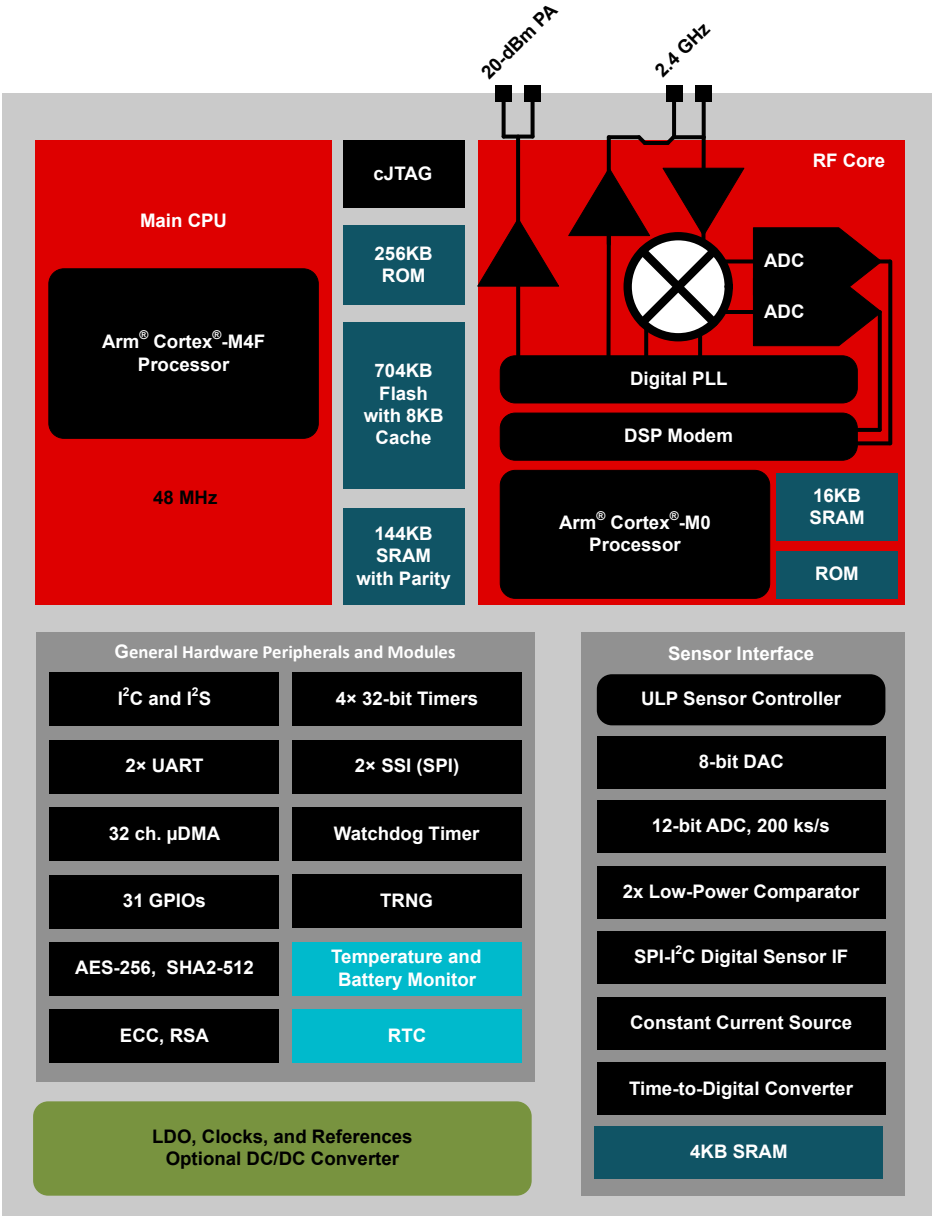


図 3-1. CC2652P7 ブロック図

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4 Revision History

資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

DATE	REVISION	NOTES
November 2021	*	Initial Release

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5 Device Comparison

表 5-1. Device Family Overview

DEVICE	RADIO SUPPORT	FLASH (KB)	RAM (KB)	GPIO	PACKAGE SIZE
CC1310	Sub-1 GHz Wireless M-Bus	32-128	16-20	10-30	RGZ (7-mm × 7-mm VQFN48) RHB (5 mm × 5 mm VQFN32) RSM (4 mm × 4 mm VQFN32)
CC1312R	Sub-1 GHz Wi-SUN® Amazon Sidewalk Wireless M-Bus	352-704	80-144	30	RGZ (7-mm × 7-mm VQFN48)
CC1352P	Multiprotocol Sub-1 GHz Wi-SUN® Amazon Sidewalk Wireless M-Bus Bluetooth 5.2 Low Energy Zigbee Thread 2.4 GHz proprietary FSK-based formats +20-dBm high-power amplifier	352-704	80-144	26	RGZ (7-mm × 7-mm VQFN48)
CC1352R	Multiprotocol Sub-1 GHz Wi-SUN® Wireless M-Bus Bluetooth 5.2 Low Energy Zigbee Thread 2.4 GHz proprietary FSK-based formats	352	80	28	RGZ (7-mm × 7-mm VQFN48)
CC2642R	Bluetooth 5.2 Low Energy 2.4 GHz proprietary FSK-based formats	352	80	31	RGZ (7-mm × 7-mm VQFN48)
CC2642R-Q1	Bluetooth 5.2 Low Energy	352	80	31	RTC (7-mm × 7-mm VQFN48)
CC2652R	Multiprotocol Bluetooth 5.2 Low Energy Zigbee Thread 2.4 GHz proprietary FSK-based formats	352-704	80-144	31	RGZ (7-mm × 7-mm VQFN48)
CC2652RB	Multiprotocol Bluetooth 5.2 Low Energy Zigbee Thread	352	80	31	RGZ (7-mm × 7-mm VQFN48)
CC2652P	Multiprotocol Bluetooth 5.2 Low Energy Zigbee Thread 2.4 GHz proprietary FSK-based formats +19.5-dBm high-power amplifier	352-704	80-144	26	RGZ (7-mm × 7-mm VQFN48)

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6 Terminal Configuration and Functions

6.1 Pin Diagram – RGZ Package (Top View)

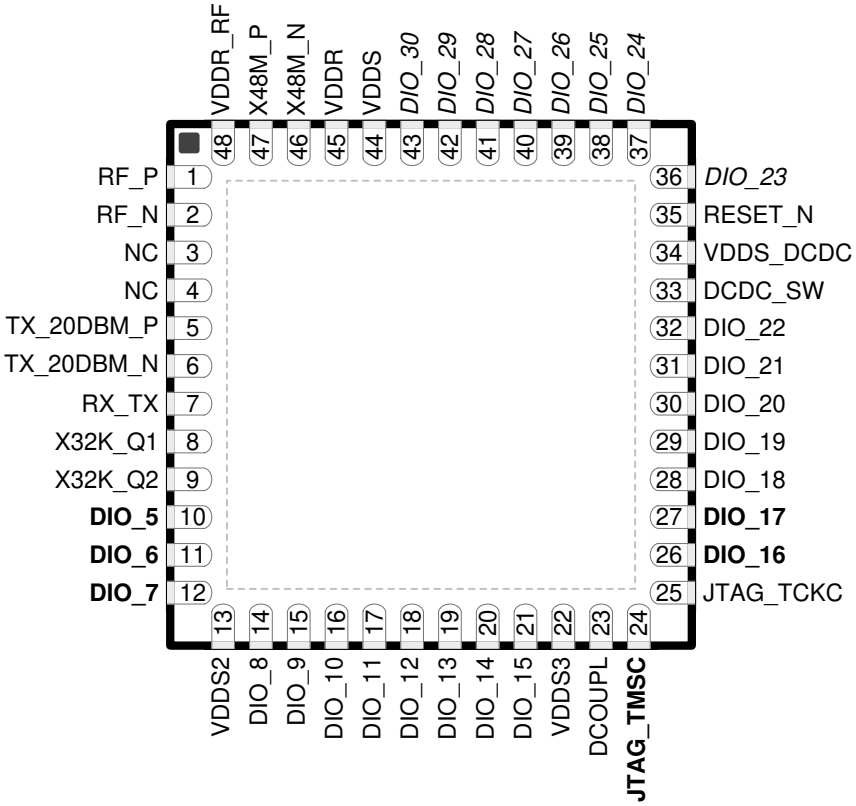


図 6-1. RGZ (7-mm × 7-mm) Pinout, 0.5-mm Pitch (Top View)

The following I/O pins marked in 図 6-1 in **bold** have high-drive capabilities:

- Pin 10, DIO_5
- Pin 11, DIO_6
- Pin 12, DIO_7
- Pin 24, JTAG_TMSC
- Pin 26, DIO_16
- Pin 27, DIO_17

The following I/O pins marked in 図 6-1 in *italics* have analog capabilities:

- Pin 36, DIO_23
- Pin 37, DIO_24
- Pin 38, DIO_25
- Pin 39, DIO_26
- Pin 40, DIO_27
- Pin 41, DIO_28
- Pin 42, DIO_29
- Pin 43, DIO_30

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6.2 Signal Descriptions – RGZ Package

表 6-1. Signal Descriptions – RGZ Package

PIN		I/O	TYPE	DESCRIPTION
NAME	NO.			
DCDC_SW	33	—	Power	Output from internal DC/DC converter(1)
DCOUPL	23	—	Power	For decoupling of internal 1.27 V regulated digital-supply (2)
DIO_5	10	I/O	Digital	GPIO, high-drive capability
DIO_6	11	I/O	Digital	GPIO, high-drive capability
DIO_7	12	I/O	Digital	GPIO, high-drive capability
DIO_8	14	I/O	Digital	GPIO
DIO_9	15	I/O	Digital	GPIO
DIO_10	16	I/O	Digital	GPIO
DIO_11	17	I/O	Digital	GPIO
DIO_12	18	I/O	Digital	GPIO
DIO_13	19	I/O	Digital	GPIO
DIO_14	20	I/O	Digital	GPIO
DIO_15	21	I/O	Digital	GPIO
DIO_16	26	I/O	Digital	GPIO, JTAG_TDO, high-drive capability
DIO_17	27	I/O	Digital	GPIO, JTAG_TDI, high-drive capability
DIO_18	28	I/O	Digital	GPIO
DIO_19	29	I/O	Digital	GPIO
DIO_20	30	I/O	Digital	GPIO
DIO_21	31	I/O	Digital	GPIO
DIO_22	32	I/O	Digital	GPIO
DIO_23	36	I/O	Digital or Analog	GPIO, analog capability
DIO_24	37	I/O	Digital or Analog	GPIO, analog capability
DIO_25	38	I/O	Digital or Analog	GPIO, analog capability
DIO_26	39	I/O	Digital or Analog	GPIO, analog capability
DIO_27	40	I/O	Digital or Analog	GPIO, analog capability
DIO_28	41	I/O	Digital or Analog	GPIO, analog capability
DIO_29	42	I/O	Digital or Analog	GPIO, analog capability
DIO_30	43	I/O	Digital or Analog	GPIO, analog capability
EGP	—	—	GND	Ground – exposed ground pad(3)
JTAG_TMSC	24	I/O	Digital	JTAG TMSC, high-drive capability
JTAG_TCKC	25	I	Digital	JTAG TCKC
RESET_N	35	I	Digital	Reset, active low. No internal pullup resistor
RF_P	1	—	RF	Positive RF input signal to LNA during RX Positive RF output signal from PA during TX
RF_N	2	—	RF	Negative RF input signal to LNA during RX Negative RF output signal from PA during TX
RX_TX	7	—	RF	Optional bias pin for the RF LNA
TX_20DBM_P	5	—	RF	Positive high-power TX signal
TX_20DBM_N	6	—	RF	Negative high-power TX signal
VDDR	45	—	Power	Internal supply, must be powered from the internal DC/DC converter or the internal LDO(2) (4) (6)
VDDR_RF	48	—	Power	Internal supply, must be powered from the internal DC/DC converter or the internal LDO(2) (5) (6)
VDDS	44	—	Power	1.8-V to 3.8-V main chip supply(1)

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表 6-1. Signal Descriptions – RGZ Package (continued)

PIN		I/O	TYPE	DESCRIPTION
NAME	NO.			
VDDS2	13	—	Power	1.8-V to 3.8-V DIO supply(1)
VDDS3	22	—	Power	1.8-V to 3.8-V DIO supply(1)
VDDS_DCDC	34	—	Power	1.8-V to 3.8-V DC/DC converter supply
X48M_N	46	—	Analog	48-MHz crystal oscillator pin 1
X48M_P	47	—	Analog	48-MHz crystal oscillator pin 2
X32K_Q1	8	—	Analog	32-kHz crystal oscillator pin 1
X32K_Q2	9	—	Analog	32-kHz crystal oscillator pin 2

(1) For more details, see technical reference manual listed in セクション 10.3.
(2) Do not supply external circuitry from this pin.
(3) EGP is the only ground connection for the device. Good electrical connection to device ground on printed circuit board (PCB) is imperative for proper device operation.
(4) If internal DC/DC converter is not used, this pin is supplied internally from the main LDO.
(5) If internal DC/DC converter is not used, this pin must be connected to VDDR for supply from the main LDO.
(6) Output from internal DC/DC and LDO is trimmed to 1.68 V.

6.3 Connections for Unused Pins and Modules

表 6-2. Connections for Unused Pins

FUNCTION	SIGNAL NAME	PIN NUMBER	ACCEPTABLE PRACTICE(1)	PREFERRED PRACTICE(1)
GPIO	DIO_n	10–12 14–21 26–32 36–43	NC or GND	NC
32.768-kHz crystal	X32K_Q1	8	NC or GND	NC
	X32K_Q2	9		
No Connects	NC	3–4	NC	NC
DC/DC converter(2)	DCDC_SW	33	NC	NC
	VDDS_DCDC	34	VDDS	VDDS

(1) NC = No connect
(2) When the DC/DC converter is not used, the inductor between DCDC_SW and VDDR can be removed. VDDR and VDDR_RF must still be connected and the 22 uF DCDC capacitor must be kept on the VDDR net.

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7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1) (2)

			MIN	MAX	UNIT
VDDS(3)	Supply voltage		–0.3	4.1	V
	Voltage on any digital pin(4)		–0.3	VDDS + 0.3, max 4.1	V
	Voltage on crystal oscillator pins, X32K_Q1, X32K_Q2, X48M_N and X48M_P		–0.3	VDDR + 0.3, max 2.25	V
V_{in}	Voltage on ADC input	Voltage scaling enabled	–0.3	VDDS	V
		Voltage scaling disabled, internal reference	–0.3	1.49	
		Voltage scaling disabled, VDDS as reference	–0.3	VDDS / 2.9	
	Input level, RF pins (RF_P and RF_N)			5	dBm
T_{stg}	Storage temperature		–40	150	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) All voltage values are with respect to ground, unless otherwise noted.
(3) VDDS_DCDC, VDDS2 and VDDS3 must be at the same potential as VDDS.
(4) Including analog capable DIOs.

7.2 ESD Ratings

				VALUE	UNIT
V_{ESD}	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS001(1)	All pins	±2000	V
		Charged device model (CDM), per JESD22-C101(2)	All pins	±500	V

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	MAX	UNIT
Operating ambient temperature(1) (3)	–40	105	°C
Operating junction temperature(1) (3)	–40	115	°C
Operating supply voltage (VDDS)	1.8	3.8	V
Rising supply voltage slew rate	0	100	mV/µs
Falling supply voltage slew rate(2)	0	20	mV/µs

(1) Operation at or near maximum operating temperature for extended durations will result in a reduction in lifetime.
(2) For small coin-cell batteries, with high worst-case end-of-life equivalent source resistance, a 22-µF VDDS input capacitor must be used to ensure compliance with this slew rate.
(3) For thermal resistance characteristics refer to セクション 7.8.

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7.4 Power Supply and Modules

over operating free-air temperature range (unless otherwise noted)

PARAMETER		MIN	TYP	MAX	UNIT
VDDS Power-on-Reset (POR) threshold			1.1 - 1.55		V
VDDS Brown-out Detector (BOD) (1)	Rising threshold		1.77		V
VDDS Brown-out Detector (BOD), before initial boot (2)	Rising threshold		1.70		V
VDDS Brown-out Detector (BOD) (1)	Falling threshold		1.75		V

(1) For boost mode (VDDR =1.95 V), TI drivers software initialization will trim VDDS BOD limits to maximum (approximately 2.0 V)
(2) Brown-out Detector is trimmed at initial boot, value is kept until device is reset by a POR reset or the RESET_N pin

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7.5 Power Consumption - Power Modes

When measured on the CC1352-P7EM-XD7793-XD24-PA24 reference design with T_c = 25 °C, V_{DDS} = 3.0 V with DC/DC enabled unless otherwise noted.

PARAMETER		TEST CONDITIONS	TYP	UNIT
Core Current Consumption				
I_{core}	Reset and Shutdown	Reset. RESET_N pin asserted or VDDS below power-on-reset threshold	100	nA
		Shutdown. No clocks running, no retention	100	
	Standby without cache retention	RTC running, CPU, 144KB RAM and (partial) register retention. RCOSC_LF	0.9	µA
		RTC running, CPU, 64KB RAM and (partial) register retention. RCOSC_LF	0.8	µA
		RTC running, CPU, 144KB RAM and (partial) register retention XOSC_LF	1.0	µA
	Standby with cache retention	RTC running, CPU, 144KB RAM and (partial) register retention. RCOSC_LF	2.3	µA
		RTC running, CPU, 144KB RAM and (partial) register retention. XOSC_LF	2.4	µA
	Idle	Supply Systems and RAM powered RCOSC_HF	669	µA
I_{core}	Active	MCU running CoreMark at 48 MHz RCOSC_HF	3.1	mA
Peripheral Current Consumption				
I_{peri}	Peripheral power domain	Delta current with domain enabled	49	µA
	Serial power domain	Delta current with domain enabled	3.6	
	RF Core	Delta current with power domain enabled, clock enabled, RF core idle	109	
	µDMA	Delta current with clock enabled, module is idle	69	
	Timers	Delta current with clock enabled, module is idle(3)	115	
	I2C	Delta current with clock enabled, module is idle	11.6	
	I2S	Delta current with clock enabled, module is idle	27.6	
	SSI	Delta current with clock enabled, module is idle	61	
	UART	Delta current with clock enabled, module is idle(1)	131	
	CRYPTO (AES)	Delta current with clock enabled, module is idle(2)	19.5	
	PKA	Delta current with clock enabled, module is idle	70	
	TRNG	Delta current with clock enabled, module is idle	25	
Sensor Controller Engine Consumption				
I_{SCE}	Active mode	24 MHz, infinite loop	799	µA
	Low-power mode	2 MHz, infinite loop	29.2	

(1) Only one UART running
(2) Only one SSI running
(3) Only one GPTimer running

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7.6 Power Consumption - Radio Modes

When measured on the CC1352-P7EM-XD7793-XD24-PA24 reference design with T_c = 25 °C, V_{DDS} = 3.0 V with DC/DC enabled unless otherwise noted.
High-power PA connected to V_{DDS} unless otherwise noted.

PARAMETER		TEST CONDITIONS	TYP	UNIT
	Radio receive current	2440 MHz	6.4	mA
	Radio transmit current 2.4 GHz PA (Bluetooth Low Energy)	0 dBm output power setting 2440 MHz	7.3	mA
		+5 dBm output power setting 2440 MHz	9.7	mA
	Radio transmit current High-power PA	+20 dBm output power setting 2440 MHz. VDDS = 3 V	101	mA
	Radio transmit current High-power PA, 10 dBm configuration(1)	+10 dBm output power setting 2440 MHz VDDR = 1.67 V	21	mA

(1) Measured on the CC1352-P7EM-XD7793-XD24-PA24_10dBm reference design.

7.7 Nonvolatile (Flash) Memory Characteristics

Over operating free-air temperature range and V_{DDS} = 3.0 V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Flash sector size			8		KB
Supported flash erase cycles before failure, single-bank(1) (5)		30			k Cycles
Supported flash erase cycles before failure, single sector(2)		60			k Cycles
Maximum number of write operations per row before sector erase(3)				83	Write Operations
Flash retention	105 °C	11.4			Years at 105 °C
Flash sector erase current	Average delta current		9.5		mA
Flash sector erase time(4)	Zero cycles		10		ms
	30k cycles			4000	ms
Flash write current	Average delta current, 4 bytes at a time		5.2		mA
Flash write time(4)	4 bytes at a time		21.6		µs

(1) A full bank erase is counted as a single erase cycle on each sector. If both flash banks are always cycled simultaneously they can be cycled 30K times each. Alternatively, the banks can be cycled a total of 30K times, e.g. the main bank X times and the second bank Y times (X+Y=30K)
(2) Up to 4 customer-designated sectors can be individually erased an additional 30k times beyond the baseline bank limitation of 30k cycles
(3) Each wordline is 2048 bits (or 256 bytes) wide. This limitation corresponds to sequential memory writes of 4 (3.1) bytes minimum per write over a whole wordline. If additional writes to the same wordline are required, a sector erase is required once the maximum number of write operations per row is reached.
(4) This number is dependent on Flash aging and increases over time and erase cycles
(5) Aborting flash during erase or program modes is not a safe operation.

7.8 Thermal Resistance Characteristics

THERMAL METRIC(1)		PACKAGE RGZ (VQFN) 48 PINS	UNIT
$R_{\theta JA}$	Junction-to-ambient thermal resistance	23.7	°C/W(2)
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	13.0	°C/W(2)
$R_{\theta JB}$	Junction-to-board thermal resistance	7.7	°C/W(2)
ψ_{JT}	Junction-to-top characterization parameter	0.1	°C/W(2)
ψ_{JB}	Junction-to-board characterization parameter	7.6	°C/W(2)

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7.8 Thermal Resistance Characteristics (continued)

THERMAL METRIC[1]		PACKAGE	UNIT
		RGZ (VQFN)	
		48 PINS	
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	1.9	°C/W[2]

(1) For more information about traditional and new thermal metrics, see Semiconductor and IC Package Thermal Metrics.
(2) °C/W = degrees Celsius per watt.

7.9 RF Frequency Bands

Over operating free-air temperature range (unless otherwise noted).

PARAMETER	MIN	TYP	MAX	UNIT
Frequency bands	2360		2500	MHz

7.10 Bluetooth Low Energy - Receive (RX)

Measured on the CC1352-P7EM-XD7793-XD24-PA24 reference design with T_c = 25 °C, V_{DDS} = 3.0 V, f_{RF}= 2440 MHz with DC/DC enabled and high power PA connected to V_{DDS} unless otherwise noted.
All measurements are performed at the antenna input with a combined RX and TX path, except for high power PA which is measured at a dedicated antenna connection. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
125 kbps (LE Coded)					
Receiver sensitivity	Differential mode. BER = 10^{-3}		–104		dBm
Receiver saturation	Differential mode. BER = 10^{-3}		>5		dBm
Frequency error tolerance	Difference between the incoming carrier frequency and the internally generated carrier frequency		> (–300 / 300)		kHz
Data rate error tolerance	Difference between incoming data rate and the internally generated data rate (37-byte packets)		> (–320 / 240)		ppm
Data rate error tolerance	Difference between incoming data rate and the internally generated data rate (255-byte packets)		> (–125 / 100)		ppm
Co-channel rejection[1]	Wanted signal at –79 dBm, modulated interferer in channel, BER = 10^{-3}		–1.5		dB
Selectivity, ±1 MHz[1]	Wanted signal at –79 dBm, modulated interferer at ±1 MHz, BER = 10^{-3}		8 / 4.5[2]		dB
Selectivity, ±2 MHz[1]	Wanted signal at –79 dBm, modulated interferer at ±2 MHz, BER = 10^{-3}		44 / 37[2]		dB
Selectivity, ±3 MHz[1]	Wanted signal at –79 dBm, modulated interferer at ±3 MHz, BER = 10^{-3}		46 / 44[2]		dB
Selectivity, ±4 MHz[1]	Wanted signal at –79 dBm, modulated interferer at ±4 MHz, BER = 10^{-3}		44 / 46[2]		dB
Selectivity, ±6 MHz[1]	Wanted signal at –79 dBm, modulated interferer at ≥ ±6 MHz, BER = 10^{-3}		48 / 44[2]		dB
Selectivity, ±7 MHz	Wanted signal at –79 dBm, modulated interferer at ≥ ±7 MHz, BER = 10^{-3}		51 / 45[2]		dB
Selectivity, Image frequency[1]	Wanted signal at –79 dBm, modulated interferer at image frequency, BER = 10^{-3}		37		dB
Selectivity, Image frequency ±1 MHz[1]	Note that Image frequency + 1 MHz is the Co- channel – 1 MHz. Wanted signal at –79 dBm, modulated interferer at ±1 MHz from image frequency, BER = 10^{-3}		4.5 / 44 [2]		dB
RSSI Range			89		dB
RSSI Accuracy (+/-)			±4		dB
500 kbps (LE Coded)					
Receiver sensitivity	Differential mode. BER = 10^{-3}		–100		dBm
Receiver saturation	Differential mode. BER = 10^{-3}		> 5		dBm
Frequency error tolerance	Difference between the incoming carrier frequency and the internally generated carrier frequency		> (–300 / 300)		kHz

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7.10 Bluetooth Low Energy - Receive (RX) (continued)

Measured on the CC1352-P7EM-XD7793-XD24-PA24 reference design with T_c = 25 °C, V_{DDS} = 3.0 V, f_{RF}= 2440 MHz with DC/DC enabled and high power PA connected to V_{DDS} unless otherwise noted.
All measurements are performed at the antenna input with a combined RX and TX path, except for high power PA which is measured at a dedicated antenna connection. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Data rate error tolerance	Difference between incoming data rate and the internally generated data rate (37-byte packets)		> (–450 / 450)		ppm
Data rate error tolerance	Difference between incoming data rate and the internally generated data rate (255-byte packets)		> (–150 / 175)		ppm
Co-channel rejection(1)	Wanted signal at –72 dBm, modulated interferer in channel, BER = 10^{-3}		–3.5		dB
Selectivity, ±1 MHz(1)	Wanted signal at –72 dBm, modulated interferer at ±1 MHz, BER = 10^{-3}		8 / 4(2)		dB
Selectivity, ±2 MHz(1)	Wanted signal at –72 dBm, modulated interferer at ±2 MHz, BER = 10^{-3}		43 / 35(2)		dB
Selectivity, ±3 MHz(1)	Wanted signal at –72 dBm, modulated interferer at ±3 MHz, BER = 10^{-3}		46 / 46(2)		dB
Selectivity, ±4 MHz(1)	Wanted signal at –72 dBm, modulated interferer at ±4 MHz, BER = 10^{-3}		45 / 47(2)		dB
Selectivity, ±6 MHz(1)	Wanted signal at –72 dBm, modulated interferer at ≥ ±6 MHz, BER = 10^{-3}		46 / 45(2)		dB
Selectivity, ±7 MHz	Wanted signal at –72 dBm, modulated interferer at ≥ ±7 MHz, BER = 10^{-3}		49 / 45(2)		dB
Selectivity, Image frequency(1)	Wanted signal at –72 dBm, modulated interferer at image frequency, BER = 10^{-3}		35		dB
Selectivity, Image frequency ±1 MHz(1)	Note that Image frequency + 1 MHz is the Co- channel – 1 MHz. Wanted signal at –72 dBm, modulated interferer at ±1 MHz from image frequency, BER = 10^{-3}		4 / 46(2)		dB
RSSI Range			90		dB
RSSI Accuracy (+/-)			±4		dB
1 Mbps (LE 1M)					
Receiver sensitivity	Differential mode. BER = 10^{-3}		–97		dBm
Receiver saturation	Differential mode. BER = 10^{-3}		> 5		dBm
Frequency error tolerance	Difference between the incoming carrier frequency and the internally generated carrier frequency		> (–350 / 350)		kHz
Data rate error tolerance	Difference between incoming data rate and the internally generated data rate (37-byte packets)		> (–650 / 750)		ppm
Co-channel rejection(1)	Wanted signal at –67 dBm, modulated interferer in channel, BER = 10^{-3}		–6		dB
Selectivity, ±1 MHz(1)	Wanted signal at –67 dBm, modulated interferer at ±1 MHz, BER = 10^{-3}		7 / 4(2)		dB
Selectivity, ±2 MHz(1)	Wanted signal at –67 dBm, modulated interferer at ±2 MHz,BER = 10^{-3}		39 / 33(2)		dB
Selectivity, ±3 MHz(1)	Wanted signal at –67 dBm, modulated interferer at ±3 MHz, BER = 10^{-3}		36 / 40(2)		dB
Selectivity, ±4 MHz(1)	Wanted signal at –67 dBm, modulated interferer at ±4 MHz, BER = 10^{-3}		36 / 45(2)		dB
Selectivity, ±5 MHz or more(1)	Wanted signal at –67 dBm, modulated interferer at ≥ ±5 MHz, BER = 10^{-3}		40		dB
Selectivity, image frequency(1)	Wanted signal at –67 dBm, modulated interferer at image frequency, BER = 10^{-3}		33		dB
Selectivity, image frequency ±1 MHz(1)	Note that Image frequency + 1 MHz is the Co- channel – 1 MHz. Wanted signal at –67 dBm, modulated interferer at ±1 MHz from image frequency, BER = 10^{-3}		4 / 41(2)		dB
Out-of-band blocking(3)	30 MHz to 2000 MHz		–10		dBm
Out-of-band blocking	2003 MHz to 2399 MHz		–18		dBm
Out-of-band blocking	2484 MHz to 2997 MHz		–12		dBm
Out-of-band blocking	3000 MHz to 12.75 GHz		–2		dBm

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7.10 Bluetooth Low Energy - Receive (RX) (continued)

Measured on the CC1352-P7EM-XD7793-XD24-PA24 reference design with T_c = 25 °C, V_{DDS} = 3.0 V, f_{RF}= 2440 MHz with DC/DC enabled and high power PA connected to V_{DDS} unless otherwise noted.
All measurements are performed at the antenna input with a combined RX and TX path, except for high power PA which is measured at a dedicated antenna connection. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Intermodulation	Wanted signal at 2402 MHz, –64 dBm. Two interferers at 2405 and 2408 MHz respectively, at the given power level		–42		dBm
Spurious emissions, 30 to 1000 MHz	Measurement in a 50-Ω single-ended load.		< –59		dBm
Spurious emissions, 1 to 12.75 GHz	Measurement in a 50-Ω single-ended load.		< –47		dBm
RSSI dynamic range			70		dB
RSSI accuracy			±4		dB
2 Mbps (LE 2M)					
Receiver sensitivity	Differential mode. Measured at SMA connector, BER = 10^{-3}		–91		dBm
Receiver saturation	Differential mode. Measured at SMA connector, BER = 10^{-3}		> 5		dBm
Frequency error tolerance	Difference between the incoming carrier frequency and the internally generated carrier frequency		> (–500 / 500)		kHz
Data rate error tolerance	Difference between incoming data rate and the internally generated data rate (37-byte packets)		> (–700 / 750)		ppm
Co-channel rejection(1)	Wanted signal at –67 dBm, modulated interferer in channel,BER = 10^{-3}		–7		dB
Selectivity, ±2 MHz(1)	Wanted signal at –67 dBm, modulated interferer at ±2 MHz, Image frequency is at –2 MHz, BER = 10^{-3}		8 / 4(2)		dB
Selectivity, ±4 MHz(1)	Wanted signal at –67 dBm, modulated interferer at ±4 MHz, BER = 10^{-3}		36 / 34(2)		dB
Selectivity, ±6 MHz(1)	Wanted signal at –67 dBm, modulated interferer at ±6 MHz, BER = 10^{-3}		37 / 36(2)		dB
Selectivity, image frequency(1)	Wanted signal at –67 dBm, modulated interferer at image frequency, BER = 10^{-3}		4		dB
Selectivity, image frequency ±2 MHz(1)	Note that Image frequency + 2 MHz is the Co-channel. Wanted signal at –67 dBm, modulated interferer at ±2 MHz from image frequency, BER = 10^{-3}		–7 / 36(2)		dB
Out-of-band blocking(3)	30 MHz to 2000 MHz		–16		dBm
Out-of-band blocking	2003 MHz to 2399 MHz		–21		dBm
Out-of-band blocking	2484 MHz to 2997 MHz		–15		dBm
Out-of-band blocking	3000 MHz to 12.75 GHz		–12		dBm
Intermodulation	Wanted signal at 2402 MHz, –64 dBm. Two interferers at 2408 and 2414 MHz respectively, at the given power level		–38		dBm
RSSI Range			60		dB
RSSI Accuracy (+/-)			±4		dB

(1) Numbers given as I/C dB
(2) X / Y, where X is +N MHz and Y is –N MHz
(3) Excluding one exception at F_{wanted} / 2, per Bluetooth Specification

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7.11 Bluetooth Low Energy - Transmit (TX)

Measured on the CC1352-P7EM-XD7793-XD24-PA24 reference design with T_c = 25 °C, V_{DDS} = 3.0 V, f_{RF}= 2440 MHz with DC/DC enabled and high power PA connected to V_{DDS} unless otherwise noted.
All measurements are performed at the antenna input with a combined RX and TX path, except for high power PA which is measured at a dedicated antenna connection. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
General Parameters						
Max output power, high power PA	Differential mode, delivered to a single-ended 50 Ω load through a balun			20		dBm
Output power programmable range high power PA	Differential mode, delivered to a single-ended 50 Ω load through a balun			6		dB
Max output power, high power PA, 10 dBm configuration(3)	Differential mode, delivered to a single-ended 50 Ω load through a balun			10.5		dBm
Output power programmable range high power PA, 10 dBm configuration(3)	Differential mode, delivered to a single-ended 50 Ω load through a balun			5		dB
Max output power, regular PA	Differential mode, delivered to a single-ended 50 Ω load through a balun			5		dBm
Output power programmable range, regular PA	Differential mode, delivered to a single-ended 50 Ω load through a balun			26		dB
Spurious emissions and harmonics						
Spurious emissions, high-power PA(1)	f < 1 GHz, outside restricted bands	+20 dBm setting		< -36		dBm
	f < 1 GHz, restricted bands FCC			< -55		dBm
	f > 1 GHz, including harmonics			-37		dBm
Harmonics, high-power PA(2)	Second harmonic			-35		dBm
	Third harmonic			-42		dBm
Spurious emissions, high-power PA, 10 dBm configuration(1) (3)	f < 1 GHz, outside restricted bands	+10 dBm setting(3)		< -36		dBm
	f < 1 GHz, restricted bands ETSI			< -54		dBm
	f < 1 GHz, restricted bands FCC,			< -55		dBm
	f > 1 GHz, including harmonics			-41		dBm
Harmonics, high-power PA, 10 dBm configuration(3)	Second harmonic			< -42		dBm
	Third harmonic			< -42		dBm
Spurious emissions, regular PA	f < 1 GHz, outside restricted bands	+5 dBm setting		< –36		dBm
	f < 1 GHz, restricted bands ETSI			< –54		dBm
	f < 1 GHz, restricted bands FCC			< –55		dBm
	f > 1 GHz, including harmonics			< –42		dBm
Harmonics, regular PA	Second harmonic			< –42		dBm
	Third harmonic			< –42		dBm

(1) To ensure margins for passing FCC band edge requirements at 2483.5 MHz, a lower than maximum output-power setting may be required when operating at the upper Bluetooth Low Energy channel(s).
(2) To ensure margins for passing FCC requirements for harmonic emission, a reduction of maximum output-power may be required.
(3) Measured on the CC1352-P7EM-XD7793-XD24-PA24_10dbm reference design.

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7.12 Zigbee and Thread - IEEE 802.15.4-2006 2.4 GHz (OQPSK DSSS1:8, 250 kbps) - RX

Measured on the CC1352-P7EM-XD7793-XD24-PA24 reference design with T_c = 25 °C, V_{DDS} = 3.0 V, f_{RF}= 2440 MHz with DC/DC enabled and high power PA connected to V_{DDS} unless otherwise noted.
All measurements are performed at the antenna input with a combined RX and TX path, except for high power PA which is measured at a dedicated antenna connection. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
General Parameters					
Receiver sensitivity	PER = 1%		–99		dBm
Receiver saturation	PER = 1%		> 5		dBm
Adjacent channel rejection	Wanted signal at –82 dBm, modulated interferer at ±5 MHz, PER = 1%		36		dB
Alternate channel rejection	Wanted signal at –82 dBm, modulated interferer at ±10 MHz, PER = 1%		57		dB
Channel rejection, ±15 MHz or more	Wanted signal at –82 dBm, undesired signal is IEEE 802.15.4 modulated channel, stepped through all channels 2405 to 2480 MHz, PER = 1%		59		dB
Blocking and desensitization, 5 MHz from upper band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		57		dB
Blocking and desensitization, 10 MHz from upper band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		62		dB
Blocking and desensitization, 20 MHz from upper band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		62		dB
Blocking and desensitization, 50 MHz from upper band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		65		dB
Blocking and desensitization, –5 MHz from lower band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		59		dB
Blocking and desensitization, –10 MHz from lower band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		59		dB
Blocking and desensitization, –20 MHz from lower band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		63		dB
Blocking and desensitization, –50 MHz from lower band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		65		dB
Spurious emissions, 30 MHz to 1000 MHz	Measurement in a 50-Ω single-ended load		–66		dBm
Spurious emissions, 1 GHz to 12.75 GHz	Measurement in a 50-Ω single-ended load		–53		dBm
Frequency error tolerance	Difference between the incoming carrier frequency and the internally generated carrier frequency		> 350		ppm
Symbol rate error tolerance	Difference between incoming symbol rate and the internally generated symbol rate		> 1000		ppm
RSSI dynamic range			95		dB
RSSI accuracy			±4		dB

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7.13 Zigbee and Thread - IEEE 802.15.4-2006 2.4 GHz (OQPSK DSSS1:8, 250 kbps) - TX

Measured on the CC1352-P7EM-XD7793-XD24-PA24 reference design with T_c = 25 °C, V_{DDS} = 3.0 V, f_{RF}= 2440 MHz with DC/DC enabled and high power PA connected to V_{DDS} unless otherwise noted.
All measurements are performed at the antenna input with a combined RX and TX path, except for high power PA which is measured at a dedicated antenna connection. All measurements are performed conducted.

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
General Parameters						
Max output power, high power PA	Differential mode, delivered to a single-ended 50-Ω load through a balun			20		dBm
Output power programmable range, high power PA	Differential mode, delivered to a single-ended 50-Ω load through a balun			6		dB
Max output power, high power PA, 10 dBm configuration(4)	Differential mode, delivered to a single-ended 50-Ω load through a balun			10.5		dBm
Output power programmable range, high power PA, 10 dBm configuration(4)	Differential mode, delivered to a single-ended 50-Ω load through a balun			5		dB
Max output power, regular PA	Differential mode, delivered to a single-ended 50-Ω load through a balun			5		dBm
Output power programmable range, regular PA	Differential mode, delivered to a single-ended 50-Ω load through a balun			26		dB
Spurious emissions and harmonics						
Spurious emissions, high-power PA(2)	f < 1 GHz, outside restricted bands	+20 dBm setting		< -39		dBm
	f < 1 GHz, restricted bands FCC			< -49		dBm
	f > 1 GHz, including harmonics			-40		dBm
Harmonics, high-power PA(3)	Second harmonic			-35		dBm
	Third harmonic			-42		dBm
Spurious emissions, high-power PA, 10 dBm configuration(2) (4)	f < 1 GHz, outside restricted bands	+10 dBm setting(4)		< -36		dBm
	f < 1 GHz, restricted bands ETSI			< -47		dBm
	f < 1 GHz, restricted bands FCC			< -55		dBm
	f > 1 GHz, including harmonics			-42		dBm
Harmonics, high-power PA, 10 dBm configuration(4)	Second harmonic			< -42		dBm
	Third harmonic			< -42		dBm
Spurious emissions, regular PA (1)	f < 1 GHz, outside restricted bands	+5 dBm setting		< -36		dBm
	f < 1 GHz, restricted bands ETSI			< -47		dBm
	f < 1 GHz, restricted bands FCC			< -55		dBm
	f > 1 GHz, including harmonics			< –42		dBm
Harmonics, regular PA	Second harmonic			< -42		dBm
	Third harmonic			< -42		dBm
IEEE 802.15.4-2006 2.4 GHz (OQPSK DSSS1:8, 250 kbps)						
Error vector magnitude, high power PA	+20 dBm setting			2		%
Error vector magnitude, high power PA, 10 dBm configuration(4)	+10 dBm setting			2		%
Error vector magnitude Regular PA	+5 dBm setting			2		%

(1) To ensure margins for passing FCC band edge requirements at 2483.5 MHz, a lower than maximum output-power setting or less than 100% duty cycle may be used when operating at 2480 MHz.
(2) To ensure margins for passing FCC band edge requirements at 2483.5 MHz, a lower than maximum output-power setting or less than 100% duty cycle may be used when operating at the upper 802.15.4 channel(s).

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(3) To ensure margins for passing FCC requirements for harmonic emission, duty cycling may be required.
(4) Measured on the CC1352-P7EM-XD7793-XD24-PA24_10dbm reference design.

7.14 Timing and Switching Characteristics

7.14.1 Reset Timing

PARAMETER	MIN	TYP	MAX	UNIT
RESET_N low duration	1			µs

7.14.2 Wakeup Timing

Measured over operating free-air temperature with V_{DDS} = 3.0 V (unless otherwise noted). The times listed here do not include software overhead.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
MCU, Reset to Active(1)			850 - 4000		µs
MCU, Shutdown to Active(1)			850 - 4000		µs
MCU, Standby to Active			165		µs
MCU, Active to Standby			39		µs
MCU, Idle to Active			15		µs

(1) The wakeup time is dependent on remaining charge on VDDR capacitor when starting the device, and thus how long the device has been in Reset or Shutdown before starting up again. The wake up time increases with a higher capacitor value.

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7.14.3 Clock Specifications

7.14.3.1 48 MHz Crystal Oscillator (XOSC_HF)

Measured on a Texas Instruments reference design with T_c = 25 °C, V_{DDS} = 3.0 V, unless otherwise noted.(1)

	PARAMETER	MIN	TYP	MAX	UNIT
	Crystal frequency		48		MHz
ESR	Equivalent series resistance 6 pF < C_L ≤ 9 pF		20	60	Ω
ESR	Equivalent series resistance 5 pF < C_L ≤ 6 pF			80	Ω
L_M	Motional inductance, relates to the load capacitance that is used for the crystal (C_L in Farads)(5)		< 3 × 10^{-25} / C_L ^{2}		H
C_L	Crystal load capacitance(4)	5	7(3)	9	pF
	Start-up time(2)		200		µs

(1) Probing or otherwise stopping the crystal while the DC/DC converter is enabled may cause permanent damage to the device.
(2) Start-up time using the TI-provided power driver. Start-up time may increase if driver is not used.
(3) On-chip default connected capacitance including reference design parasitic capacitance. Connected internal capacitance is changed through software in the Customer Configuration section (CCFG).
(4) Adjustable load capacitance is integrated into the device. External load capacitors are required for systems targeting compliance with certain regulations. See the device errata for further details.
(5) The crystal manufacturer's specification must satisfy this requirement for proper operation.

7.14.3.2 48 MHz RC Oscillator (RCOSC_HF)

Measured on a Texas Instruments reference design with T_c = 25 °C, V_{DDS} = 3.0 V, unless otherwise noted.

	MIN	TYP	MAX	UNIT
Frequency		48		MHz
Uncalibrated frequency accuracy		±1		%
Calibrated frequency accuracy(1)		±0.25		%
Start-up time		5		µs

(1) Accuracy relative to the calibration source (XOSC_HF)

7.14.3.3 2 MHz RC Oscillator (RCOSC_MF)

Measured on a Texas Instruments reference design with T_c = 25 °C, V_{DDS} = 3.0 V, unless otherwise noted.

	MIN	TYP	MAX	UNIT
Calibrated frequency		2		MHz
Start-up time		5		µs

7.14.3.4 32.768 kHz Crystal Oscillator (XOSC_LF)

Measured on a Texas Instruments reference design with T_c = 25 °C, V_{DDS} = 3.0 V, unless otherwise noted.

		MIN	TYP	MAX	UNIT
	Crystal frequency		32.768		kHz
ESR	Equivalent series resistance		30	100	kΩ
C_L	Crystal load capacitance	6	7(1)	12	pF

(1) Default load capacitance using TI reference designs including parasitic capacitance. Crystals with different load capacitance may be used.

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7.14.3.5 32 kHz RC Oscillator (RCOSC_LF)

Measured on a Texas Instruments reference design with T_c = 25 °C, V_{DDS} = 3.0 V, unless otherwise noted.

		MIN	TYP	MAX	UNIT
Frequency			32.8		kHz
Calibrated RTC variation(1)	Calibrated periodically against XOSC_HF(2)		±600(3)		ppm
Temperature coefficient			50		ppm/°C

(1) When using RCOSC_LF as source for the low frequency system clock (SCLK_LF), the accuracy of the SCLK_LF-derived Real Time Clock (RTC) can be improved by measuring RCOSC_LF relative to XOSC_HF and compensating for the RTC tick speed. This functionality is available through the TI-provided Power driver.
(2) TI driver software calibrates the RTC every time XOSC_HF is enabled.
(3) Some device's variation can exceed 1000 ppm. Further calibration will not improve variation.

7.14.4 Synchronous Serial Interface (SSI) Characteristics

7.14.4.1 Synchronous Serial Interface (SSI) Characteristics

over operating free-air temperature range (unless otherwise noted)

PARAMETER NO.	PARAMETER		MIN	TYP	MAX	UNIT
S1	t_{clk_per}	SSIClk cycle time	12		65024	System Clocks (2)
S2(1)	t_{clk_high}	SSIClk high time		0.5		t_{clk_per}
S3(1)	t_{clk_low}	SSIClk low time		0.5		t_{clk_per}

(1) Refer to SSI timing diagrams 図 7-1, 図 7-2 and 図 7-3.
(2) When using the TI-provided Power driver, the SSI system clock is always 48 MHz.

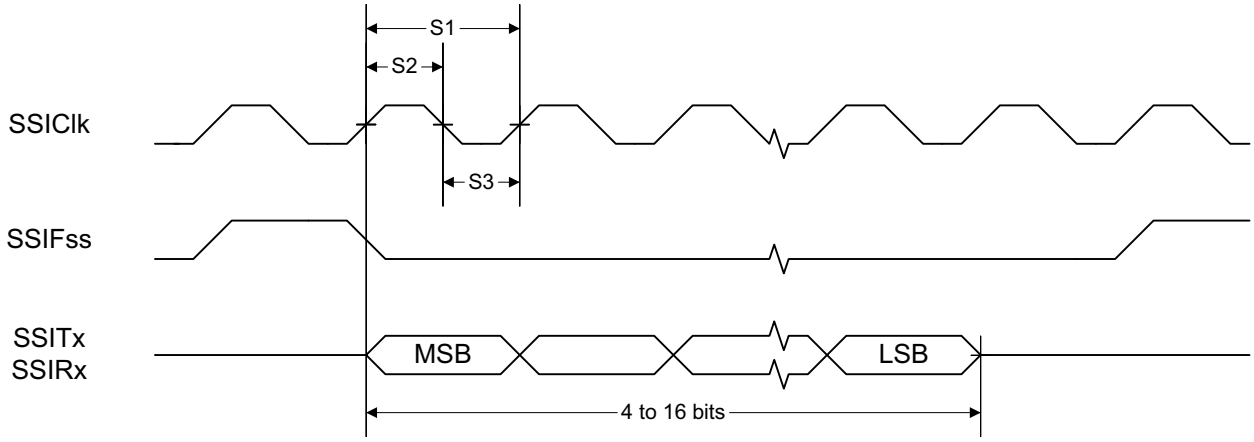


図 7-1. SSI Timing for TI Frame Format (FRF = 01), Single Transfer Timing Measurement

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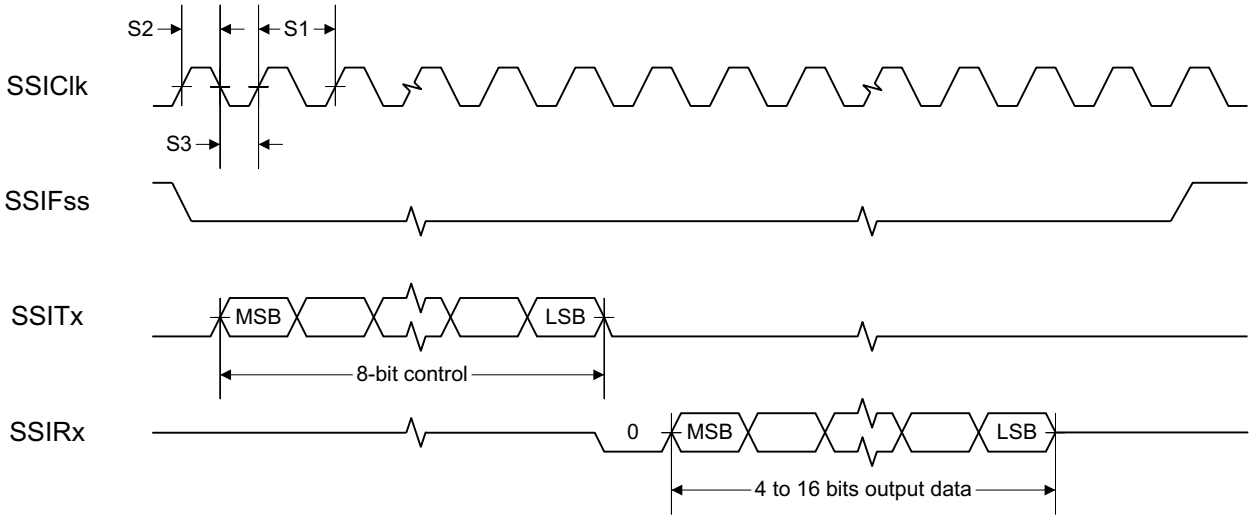


図 7-2. SSI Timing for MICROWIRE Frame Format (FRF = 10), Single Transfer

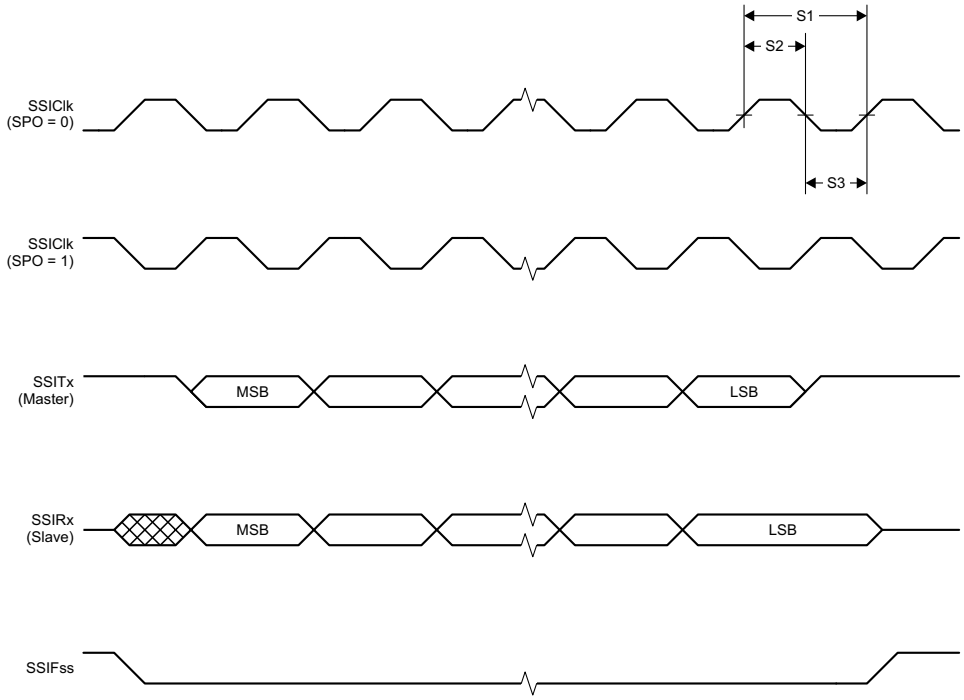


図 7-3. SSI Timing for SPI Frame Format (FRF = 00), With SPH = 1

7.14.5 UART

7.14.5.1 UART Characteristics

over operating free-air temperature range (unless otherwise noted)

PARAMETER	MIN	TYP	MAX	UNIT
UART rate			2.89	MBaud

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7.15 Peripheral Characteristics

7.15.1 ADC

7.15.1.1 Analog-to-Digital Converter (ADC) Characteristics

T_c = 25 °C, V_{DDS} = 3.0 V and voltage scaling enabled, unless otherwise noted.(1)
Performance numbers require use of offset and gain adjustments in software by TI-provided ADC drivers.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Input voltage range		0		VDDS	V
	Resolution			12		Bits
	Sample Rate				200	ksps
	Offset	Internal 4.3 V equivalent reference(2)		±2		LSB
	Gain error	Internal 4.3 V equivalent reference(2)		±7		LSB
DNL(4)	Differential nonlinearity			>–1		LSB
INL	Integral nonlinearity			±4		LSB
ENOB	Effective number of bits	Internal 4.3 V equivalent reference(2), 200 kSamples/s, 9.6 kHz input tone		9.8		Bits
		Internal 4.3 V equivalent reference(2), 200 kSamples/s, 9.6 kHz input tone, DC/DC enabled		9.8		
		VDDS as reference, 200 kSamples/s, 9.6 kHz input tone		10.1		
		Internal reference, voltage scaling disabled, 32 samples average, 200 kSamples/s, 300 Hz input tone		11.1		
		Internal reference, voltage scaling disabled, 14-bit mode, 200 kSamples/s, 600 Hz input tone (5)		11.3		
		Internal reference, voltage scaling disabled, 15-bit mode, 200 kSamples/s, 150 Hz input tone (5)		11.6		
THD	Total harmonic distortion	Internal 4.3 V equivalent reference(2), 200 kSamples/s, 9.6 kHz input tone		–65		dB
		VDDS as reference, 200 kSamples/s, 9.6 kHz input tone		–70		
		Internal reference, voltage scaling disabled, 32 samples average, 200 kSamples/s, 300 Hz input tone		–72		
SINAD, SNDR	Signal-to-noise and distortion ratio	Internal 4.3 V equivalent reference(2), 200 kSamples/s, 9.6 kHz input tone		60		dB
		VDDS as reference, 200 kSamples/s, 9.6 kHz input tone		63		
		Internal reference, voltage scaling disabled, 32 samples average, 200 kSamples/s, 300 Hz input tone		68		
SFDR	Spurious-free dynamic range	Internal 4.3 V equivalent reference(2), 200 kSamples/s, 9.6 kHz input tone		70		dB
		VDDS as reference, 200 kSamples/s, 9.6 kHz input tone		73		
		Internal reference, voltage scaling disabled, 32 samples average, 200 kSamples/s, 300 Hz input tone		75		
	Conversion time	Serial conversion, time-to-output, 24 MHz clock		50		Clock Cycles
	Current consumption	Internal 4.3 V equivalent reference(2)		0.40		mA
	Current consumption	VDDS as reference		0.57		mA
	Reference voltage	Equivalent fixed internal reference (input voltage scaling enabled). For best accuracy, the ADC conversion should be initiated through the TI-RTOS API in order to include the gain/offset compensation factors stored in FCFG1		4.3(2) (3)		V
	Reference voltage	Fixed internal reference (input voltage scaling disabled). For best accuracy, the ADC conversion should be initiated through the TI-RTOS API in order to include the gain/offset compensation factors stored in FCFG1. This value is derived from the scaled value (4.3 V) as follows: V_{ref} = 4.3 V × 1408 / 4095		1.48		V
	Reference voltage	VDDS as reference, input voltage scaling enabled		VDDS		V
	Reference voltage	VDDS as reference, input voltage scaling disabled		VDDS / 2.82(3)		V

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7.15.1.1 Analog-to-Digital Converter (ADC) Characteristics (continued)

T_c = 25 °C, V_{DDS} = 3.0 V and voltage scaling enabled, unless otherwise noted.(1)
Performance numbers require use of offset and gain adjustments in software by TI-provided ADC drivers.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Input impedance	200 kSamples/s, voltage scaling enabled. Capacitive input, Input impedance depends on sampling frequency and sampling time		>1		MΩ

(1) Using IEEE Std 1241-2010 for terminology and test methods
(2) Input signal scaled down internally before conversion, as if voltage range was 0 to 4.3 V
(3) Applied voltage must be within Absolute Maximum Ratings at all times
(4) No missing codes
(5) ADC_output = Σ(4^n samples) >> n, n = desired extra bits

7.15.2 DAC

7.15.2.1 Digital-to-Analog Converter (DAC) Characteristics

T_c = 25 °C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
General Parameters						
	Resolution			8		Bits
V_{DDS}	Supply voltage	Any load, any V_{REF}, pre-charge OFF, DAC charge-pump ON	1.8		3.8	V
		External Load(4), any V_{REF}, pre-charge OFF, DAC charge-pump OFF	2.0		3.8	
		Any load, V_{REF} = DCOUPL, pre-charge ON	2.6		3.8	
F_{DAC}	Clock frequency	Buffer ON (recommended for external load)	16		250	kHz
		Buffer OFF (internal load)	16		1000	
	Voltage output settling time	V_{REF} = VDDS, buffer OFF, internal load		13		1 / F_{DAC}
		V_{REF} = VDDS, buffer ON, external capacitive load = 20 pF(3)		13.8		
	External capacitive load			20	200	pF
	External resistive load		10			MΩ
	Short circuit current				400	µA
Z_{MAX}	Max output impedance Vref = VDDS, buffer ON, CLK 250 kHz	VDDS = 3.8 V, DAC charge-pump OFF		50.8		kΩ
		VDDS = 3.0 V, DAC charge-pump ON		51.7		
		VDDS = 3.0 V, DAC charge-pump OFF		53.2		
		VDDS = 2.0 V, DAC charge-pump ON		48.7		
		VDDS = 2.0 V, DAC charge-pump OFF		70.2		
		VDDS = 1.8 V, DAC charge-pump ON		46.3		
		VDDS = 1.8 V, DAC charge-pump OFF		88.9		
Internal Load - Continuous Time Comparator / Low Power Clocked Comparator						
DNL	Differential nonlinearity	V_{REF} = VDDS, load = Continuous Time Comparator or Low Power Clocked Comparator F_{DAC} = 250 kHz		±1		LSB(1)
	Differential nonlinearity	V_{REF} = VDDS, load = Continuous Time Comparator or Low Power Clocked Comparator F_{DAC} = 16 kHz		±1.2		
	Offset error(2) Load = Continuous Time Comparator	V_{REF} = VDDS = 3.8 V		±0.64		LSB(1)
		V_{REF} = VDDS= 3.0 V		±0.81		
		V_{REF} = VDDS = 1.8 V		±1.27		
		V_{REF} = DCOUPL, pre-charge ON		±3.43		
		V_{REF} = DCOUPL, pre-charge OFF		±2.88		
		V_{REF} = ADCREF		±2.37		

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7.15.2.1 Digital-to-Analog Converter (DAC) Characteristics (continued)

T_c = 25 °C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Offset error[2] Load = Low Power Clocked Comparator	V_{REF} = VDDS= 3.8 V		±0.78		LSB[1]
		V_{REF} = VDDS = 3.0 V		±0.77		
		V_{REF} = VDDS= 1.8 V		±3.46		
		V_{REF} = DCOUPL, pre-charge ON		±3.44		
		V_{REF} = DCOUPL, pre-charge OFF		±4.70		
		V_{REF} = ADCREF		±4.11		
	Max code output voltage variation[2] Load = Continuous Time Comparator	V_{REF} = VDDS = 3.8 V		±1.53		LSB[1]
		V_{REF} = VDDS = 3.0 V		±1.71		
		V_{REF} = VDDS= 1.8 V		±2.10		
		V_{REF} = DCOUPL, pre-charge ON		±6.00		
		V_{REF} = DCOUPL, pre-charge OFF		±3.85		
		V_{REF} = ADCREF		±5.84		
	Max code output voltage variation[2] Load = Low Power Clocked Comparator	V_{REF} = VDDS= 3.8 V		±2.92		LSB[1]
		V_{REF} =VDDS= 3.0 V		±3.06		
		V_{REF} = VDDS= 1.8 V		±3.91		
		V_{REF} = DCOUPL, pre-charge ON		±7.84		
		V_{REF} = DCOUPL, pre-charge OFF		±4.06		
		V_{REF} = ADCREF		±6.94		
	Output voltage range[2] Load = Continuous Time Comparator	V_{REF} = VDDS = 3.8 V, code 1		0.03		V
		V_{REF} = VDDS = 3.8 V, code 255		3.62		
		V_{REF} = VDDS= 3.0 V, code 1		0.02		
		V_{REF} = VDDS= 3.0 V, code 255		2.86		
		V_{REF} = VDDS= 1.8 V, code 1		0.01		
		V_{REF} = VDDS = 1.8 V, code 255		1.71		
		V_{REF} = DCOUPL, pre-charge OFF, code 1		0.01		
		V_{REF} = DCOUPL, pre-charge OFF, code 255		1.21		
		V_{REF} = DCOUPL, pre-charge ON, code 1		1.27		
		V_{REF} = DCOUPL, pre-charge ON, code 255		2.46		
		V_{REF} = ADCREF, code 1		0.01		
		V_{REF} = ADCREF, code 255		1.41		
	Output voltage range[2] Load = Low Power Clocked Comparator	V_{REF} = VDDS = 3.8 V, code 1		0.03		V
		V_{REF} = VDDS= 3.8 V, code 255		3.61		
		V_{REF} = VDDS= 3.0 V, code 1		0.02		
		V_{REF} = VDDS= 3.0 V, code 255		2.85		
		V_{REF} = VDDS = 1.8 V, code 1		0.01		
		V_{REF} = VDDS = 1.8 V, code 255		1.71		
		V_{REF} = DCOUPL, pre-charge OFF, code 1		0.01		
		V_{REF} = DCOUPL, pre-charge OFF, code 255		1.21		
		V_{REF} = DCOUPL, pre-charge ON, code 1		1.27		
		V_{REF} = DCOUPL, pre-charge ON, code 255		2.46		
		V_{REF} = ADCREF, code 1		0.01		
		V_{REF} = ADCREF, code 255		1.41		
External Load						
INL	Integral nonlinearity	V_{REF} = VDDS, F_{DAC} = 250 kHz		±1		LSB[1]
		V_{REF} = DCOUPL, F_{DAC} = 250 kHz		±2		
		V_{REF} = ADCREF, F_{DAC} = 250 kHz		±1		
DNL	Differential nonlinearity	V_{REF} = VDDS, F_{DAC} = 250 kHz		±1		LSB[1]

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7.15.2.1 Digital-to-Analog Converter (DAC) Characteristics (continued)

T_c = 25 °C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Offset error	V_{REF} = VDDS= 3.8 V		±0.40		LSB(1)
		V_{REF} = VDDS= 3.0 V		±0.50		
		V_{REF} = VDDS = 1.8 V		±0.75		
		V_{REF} = DCOUPL, pre-charge ON		±1.55		
		V_{REF} = DCOUPL, pre-charge OFF		±1.30		
		V_{REF} = ADCREF		±1.10		
	Max code output voltage variation	V_{REF} = VDDS= 3.8 V		±1.00		LSB(1)
		V_{REF} = VDDS= 3.0 V		±1.00		
		V_{REF} = VDDS= 1.8 V		±1.00		
		V_{REF} = DCOUPL, pre-charge ON		±3.45		
		V_{REF} = DCOUPL, pre-charge OFF		±2.10		
		V_{REF} = ADCREF		±1.90		
	Output voltage range Load = Low Power Clocked Comparator	V_{REF} = VDDS = 3.8 V, code 1		0.03		V
		V_{REF} = VDDS = 3.8 V, code 255		3.61		
		V_{REF} = VDDS = 3.0 V, code 1		0.02		
		V_{REF} = VDDS= 3.0 V, code 255		2.85		
		V_{REF} = VDDS= 1.8 V, code 1		0.02		
		V_{REF} = VDDS = 1.8 V, code 255		1.71		
		V_{REF} = DCOUPL, pre-charge OFF, code 1		0.02		
		V_{REF} = DCOUPL, pre-charge OFF, code 255		1.20		
		V_{REF} = DCOUPL, pre-charge ON, code 1		1.27		
		V_{REF} = DCOUPL, pre-charge ON, code 255		2.46		
		V_{REF} = ADCREF, code 1		0.02		
		V_{REF} = ADCREF, code 255		1.42		

(1) 1 LSB (V_{REF} 3.8 V/3.0 V/1.8 V/DCOUPL/ADCREF) = 14.10 mV/11.13 mV/6.68 mV/4.67 mV/5.48 mV
(2) Includes comparator offset
(3) A load > 20 pF will increases the settling time
(4) Keysight 34401A Multimeter

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7.15.3 Temperature and Battery Monitor

7.15.3.1 Temperature Sensor

Measured on a Texas Instruments reference design with T_c = 25 °C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Resolution			2		°C
Accuracy	-40 °C to 0 °C		±4.0		°C
Accuracy	0 °C to 105 °C		±2.5		°C
Supply voltage coefficient(1)			3.6		°C/V

(1) The temperature sensor is automatically compensated for VDDS variation when using the TI-provided temperature driver.

7.15.3.2 Battery Monitor

Measured on a Texas Instruments reference design with T_c = 25 °C, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Resolution			25		mV
Range		1.8		3.8	V
Integral nonlinearity (max)			23		mV
Accuracy	VDDS = 3.0 V		22.5		mV
Offset error			-32		mV
Gain error			-1		%

7.15.4 Comparators

7.15.4.1 Low-Power Clocked Comparator

T_c = 25 °C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Input voltage range		0		V_{DDS}	V
Clock frequency			SCLK_LF		
Internal reference voltage(1)	Using internal DAC with VDDS as reference voltage, DAC code = 0 - 255		0.024 - 2.865		V
Offset	Measured at V_{DDS} / 2, includes error from internal DAC		±5		mV
Decision time	Step from –50 mV to 50 mV		1		Clock Cycle

(1) The comparator can use an internal 8 bits DAC as its reference. The DAC output voltage range depends on the reference voltage selected. See #none#

7.15.4.2 Continuous Time Comparator

T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Input voltage range(1)		0		V_{DDS}	V
Offset	Measured at V_{DDS} / 2		±5		mV
Decision time	Step from –10 mV to 10 mV		0.70		µs
Current consumption	Internal reference		8.0		µA

(1) The input voltages can be generated externally and connected throughout I/Os or an internal reference voltage can be generated using the DAC

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7.15.5 Current Source

7.15.5.1 Programmable Current Source

T_c = 25 °C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Current source programmable output range (logarithmic range)			0.25 - 20		µA
Resolution			0.25		µA

7.15.6 GPIO

7.15.6.1 GPIO DC Characteristics

Measurements CBSed to PG2.1:

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
T_A = 25 °C, V_{DDS} = 1.8 V					
GPIO VOH at 8 mA load	IOCURR = 2, high-drive GPIOs only		1.56		V
GPIO VOL at 8 mA load	IOCURR = 2, high-drive GPIOs only		0.24		V
GPIO VOH at 4 mA load	IOCURR = 1		1.59		V
GPIO VOL at 4 mA load	IOCURR = 1		0.21		V
GPIO pullup current	Input mode, pullup enabled, Vpad = 0 V		73		µA
GPIO pulldown current	Input mode, pulldown enabled, Vpad = VDDS		19		µA
GPIO low-to-high input transition, with hysteresis	IH = 1, transition voltage for input read as 0 → 1		1.08		V
GPIO high-to-low input transition, with hysteresis	IH = 1, transition voltage for input read as 1 → 0		0.73		V
GPIO input hysteresis	IH = 1, difference between 0 → 1 and 1 → 0 points		0.35		V
T_A = 25 °C, V_{DDS} = 3.0 V					
GPIO VOH at 8 mA load	IOCURR = 2, high-drive GPIOs only		2.59		V
GPIO VOL at 8 mA load	IOCURR = 2, high-drive GPIOs only		0.42		V
GPIO VOH at 4 mA load	IOCURR = 1		2.63		V
GPIO VOL at 4 mA load	IOCURR = 1		0.40		V
T_A = 25 °C, V_{DDS} = 3.8 V					
GPIO pullup current	Input mode, pullup enabled, Vpad = 0 V		282		µA
GPIO pulldown current	Input mode, pulldown enabled, Vpad = VDDS		110		µA
GPIO low-to-high input transition, with hysteresis	IH = 1, transition voltage for input read as 0 → 1		1.97		V
GPIO high-to-low input transition, with hysteresis	IH = 1, transition voltage for input read as 1 → 0		1.55		V
GPIO input hysteresis	IH = 1, difference between 0 → 1 and 1 → 0 points		0.42		V
T_A = 25 °C					
VIH	Lowest GPIO input voltage reliably interpreted as a *High*	0.8*V_{DDS}			V
VIL	Highest GPIO input voltage reliably interpreted as a *Low*			0.2*V_{DDS}	V

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7.16 Typical Characteristics

All measurements in this section are done with T_c = 25 °C and V_{DDS} = 3.0 V, unless otherwise noted. See セクション 7.3 for device limits. Values exceeding these limits are for reference only.

7.16.1 MCU Current

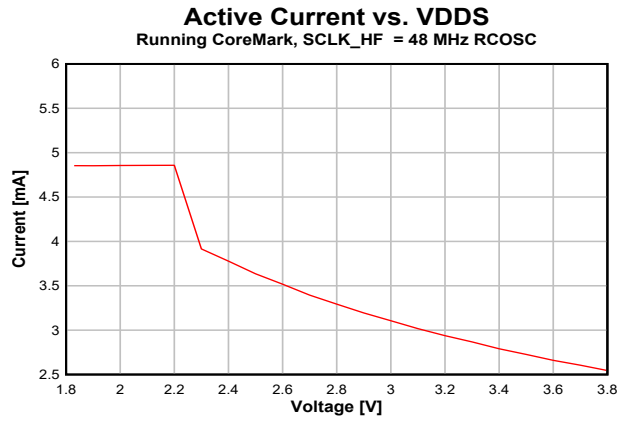


図 7-4. Active Mode (MCU) Current vs. Supply Voltage (VDDS)

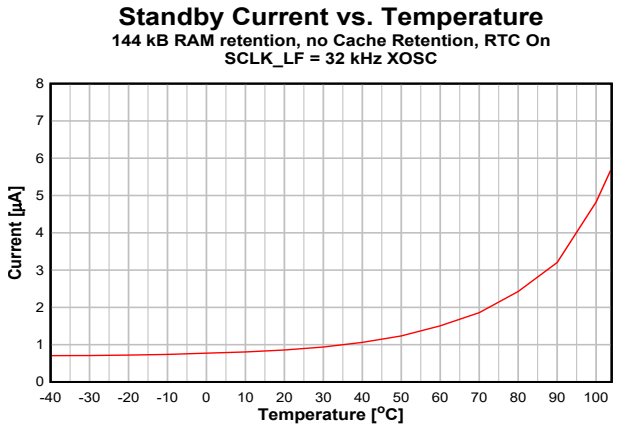


図 7-5. Standby Mode (MCU) Current vs. Temperature

7.16.2 RX Current

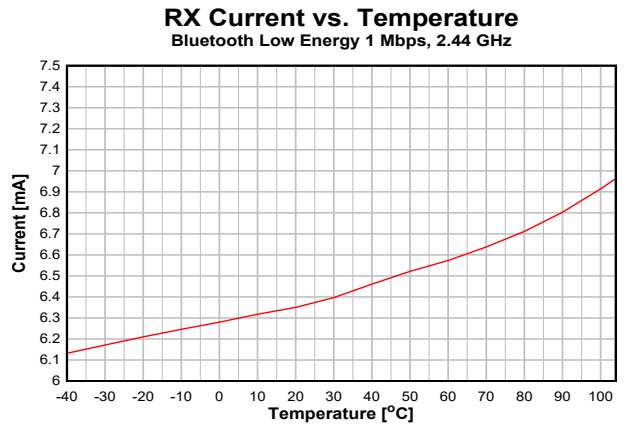


図 7-6. RX Current vs. Temperature (Bluetooth Low Energy 1 Mbps, 2.44 GHz)

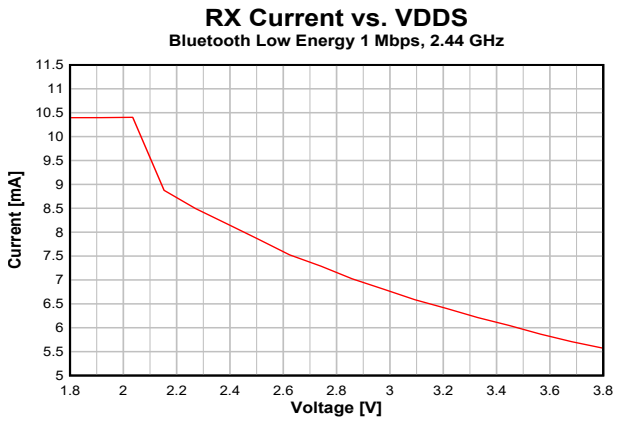


図 7-7. RX Current vs. Supply Voltage (VDDS) (Bluetooth Low Energy 1 Mbps, 2.44 GHz)

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7.16.3 TX Current

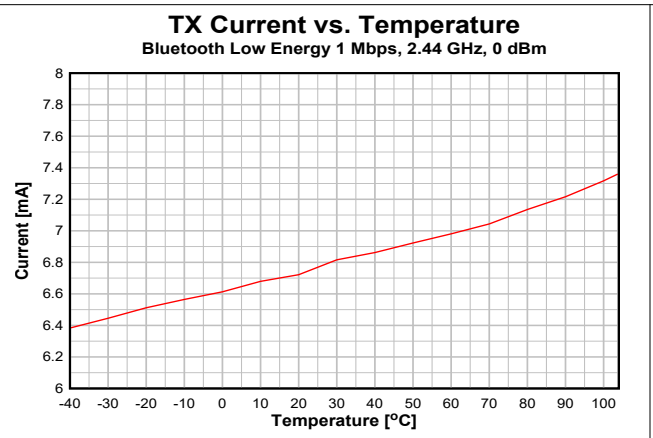


図 7-8. TX Current vs. Temperature (Bluetooth Low Energy 1 Mbps, 2.44 GHz)

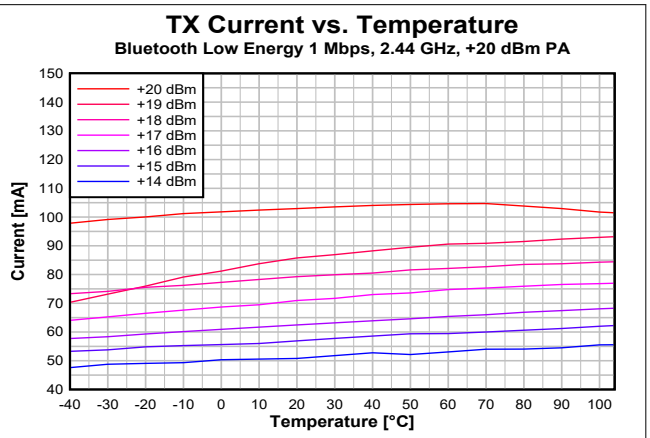


図 7-9. TX Current vs. Temperature (Bluetooth Low Energy 1 Mbps, 2.44 GHz, VDDS = 3.3 V)

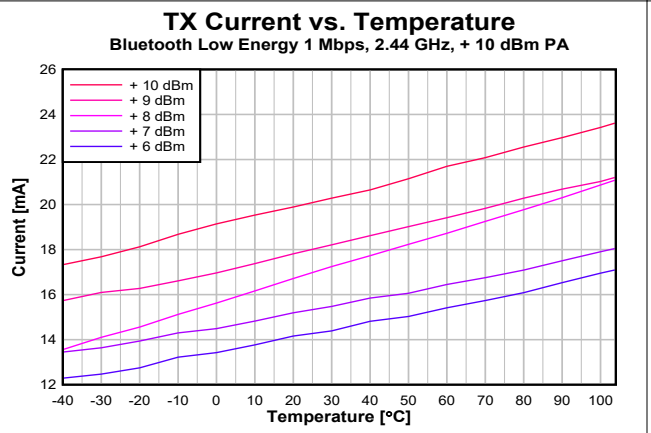


図 7-10. TX Current vs. Temperature (250 kbps, 2.44 GHz, +10 dBm PA)

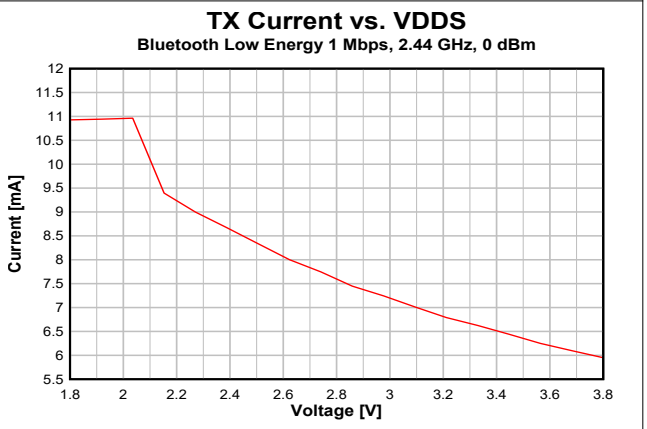


図 7-11. TX Current vs. Supply Voltage (VDDS) (Bluetooth Low Energy 1 Mbps, 2.44 GHz)

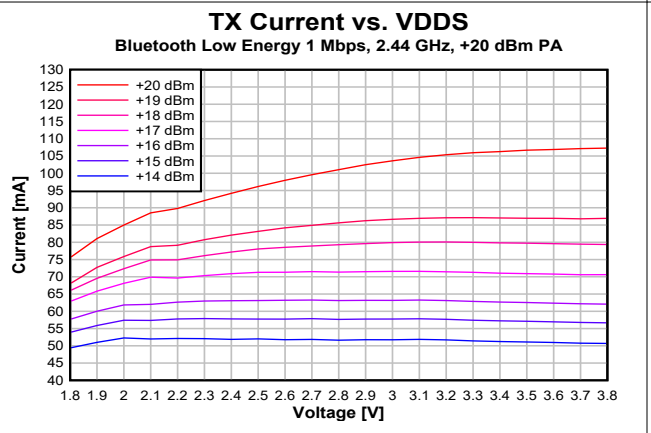


図 7-12. TX Current vs. Supply Voltage (VDDS) (Bluetooth Low Energy 1 Mbps, 2.44 GHz, +20 dBm PA)

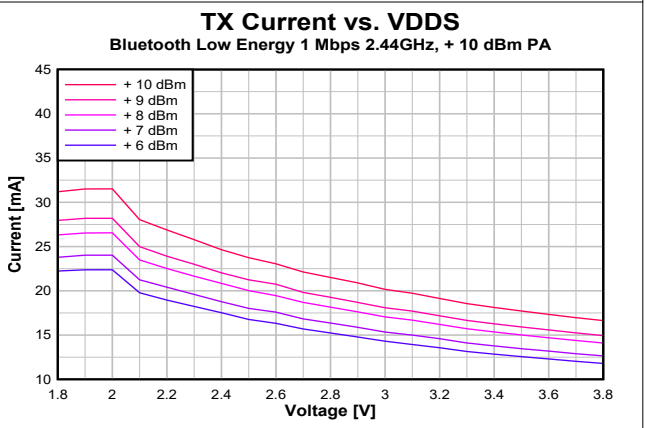


図 7-13. TX Current vs. Supply Voltage (VDDS) (250 kbps, 2.44 GHz, +10 dBm PA)

表 7-1. Typical TX Current and Output Power

CC2652P7 at 2.4 GHz, VDDS = 3.3 V (1) (Measured on CC1352-7PEM-XD7793-XD24-PA24)			
txPower	TX Power Setting (SmartRF Studio)	Typical Output Power [dBm]	Typical Current Consumption [mA]
0x3F75F5	20	19.6	102
0x3F61E2	19	18.3	86

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7.16.3 TX Current (continued)

表 7-1. Typical TX Current and Output Power (continued)

CC2652P7 at 2.4 GHz, VDDS = 3.3 V (1) (Measured on CC1352-7PEM-XD7793-XD24-PA24)			
txPower	**TX Power Setting (SmartRF Studio)**	**Typical Output Power [dBm]**	**Typical Current Consumption [mA]**
0x3047E0	18	17.4	79
0x1B4FE5	17	16.3	71
0x1B39DE	16	15.2	63
0x1B2FDA	15	14.3	58
0x1B27D6	14	13.2	52

(1) VDDS powers the PA, therefore the output power is affected by variation in VDDS voltage.

表 7-2. Typical TX Current and Output Power

CC2652P7 at 2.4 GHz, VDDS = 3.0 V (1) (Measured on CC1352-7PEM-XD7793-XD24-PA24_10dBm)			
txPower	**TX Power Setting (SmartRF Studio)**	**Typical Output Power [dBm]**	**Typical Current Consumption [mA]**
0x103F5F	10	10.7	21
0x10335A	9	9.6	19
0x143661	8	8.5	19
0x144F2A	7	7.6	17
0x144F26	6	6.6	16
0x144722	5	5.4	15

(1) Internal regulated voltage powers the PA, therefore the output power is not affected by variation in VDDS voltage.

表 7-3. Typical TX Current and Output Power

CC2652P7 at 2.4 GHz, VDDS = 3.0 V (1) (Measured on CC1352-7PEM-XD7793-XD24-PA24)			
txPower	**TX Power Setting (SmartRF Studio)**	**Typical Output Power [dBm]**	**Typical Current Consumption [mA]**
0x762E	5	4.7	10
0x8220	4	3.7	9
0x5617	3	2.8	8
0x3E66	2	1.9	8
0x3261	1	0.9	8
0x2C5D	0	0.0	7
0x1899	-3	-3.1	6
0x1695	-5	-4.9	6
0x1693	-6	-6.0	6
0x0CD4	-9	-9.1	5
0x0AD3	-10	-9.8	5
0x0AD0	-12	-11.9	5
0x06CD	-15	-14.6	5
0x04CA	-18	-17.8	5
0x04C8	-20	-20.4	4

(1) Internal regulated voltage powers the PA, therefore the output power is not affected by variation in VDDS voltage.

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7.16.4 RX Performance

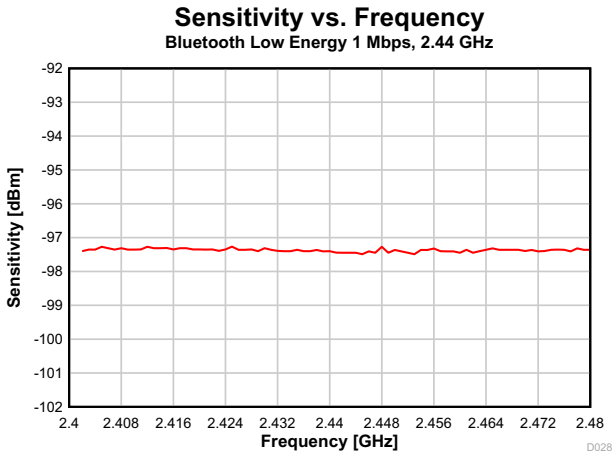


図 7-14. Sensitivity vs. Frequency (Bluetooth Low Energy 1 Mbps, 2.44 GHz)

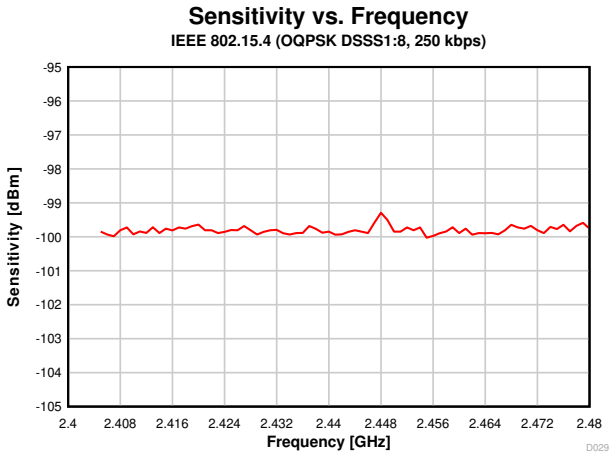


図 7-15. Sensitivity vs. Frequency (250 kbps, 2.44 GHz)

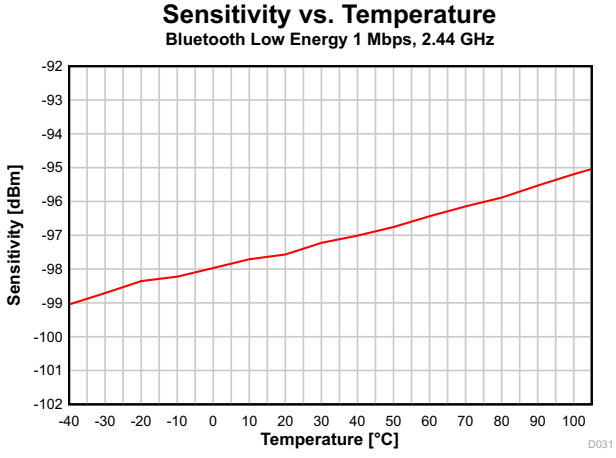


図 7-16. Sensitivity vs. Temperature (Bluetooth Low Energy 1 Mbps, 2.44 GHz)

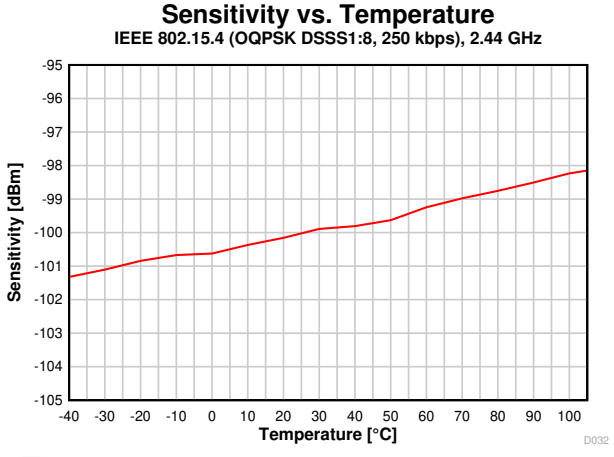


図 7-17. Sensitivity vs. Temperature (250 kbps, 2.44 GHz)

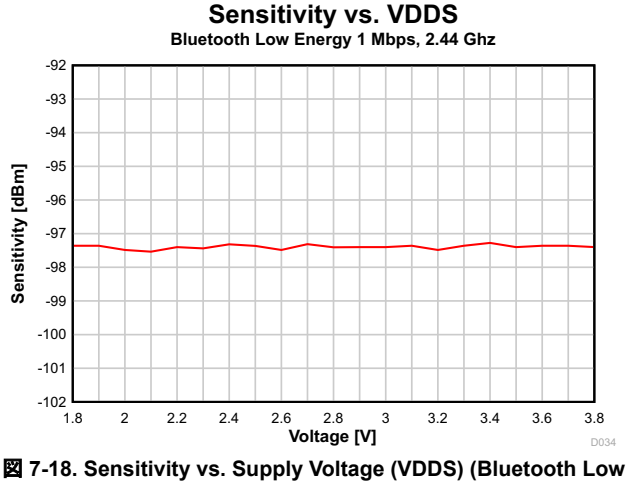


図 7-18. Sensitivity vs. Supply Voltage (VDDS) (Bluetooth Low Energy 1 Mbps, 2.44 GHz)

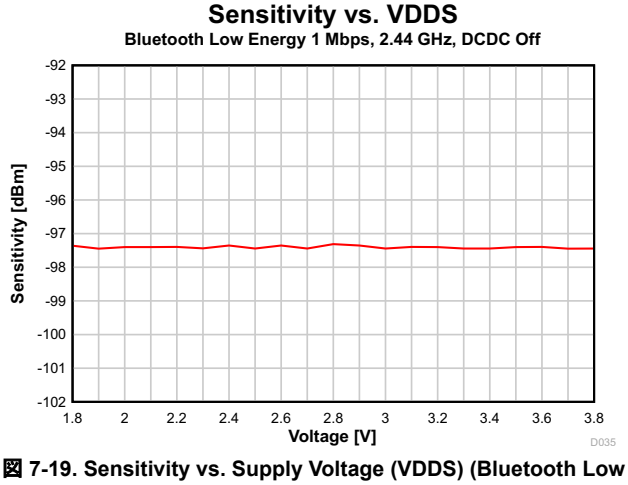


図 7-19. Sensitivity vs. Supply Voltage (VDDS) (Bluetooth Low Energy 1 Mbps, 2.44 GHz, DCDC Off)

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7.16.4 RX Performance (continued)

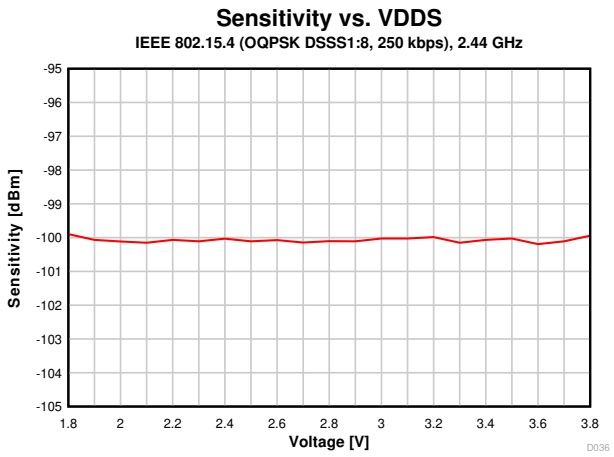


図 7-20. Sensitivity vs. Supply Voltage (VDDS) (250 kbps, 2.44 GHz)

7.16.5 TX Performance

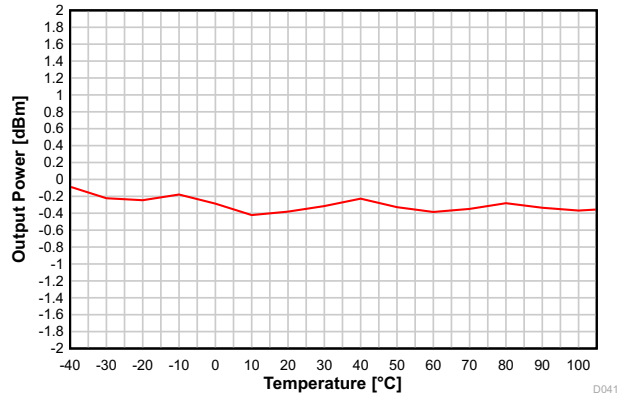


図 7-21. Output Power vs. Temperature (Bluetooth Low Energy 1 Mbps, 2.44 GHz)

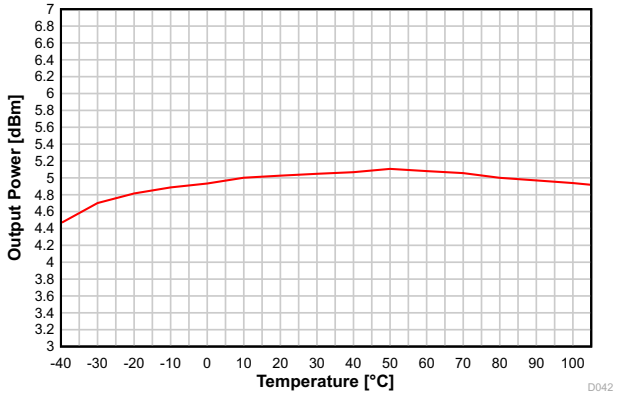


図 7-22. Output Power vs. Temperature (Bluetooth Low Energy 1 Mbps, 2.44 GHz, +5 dBm)

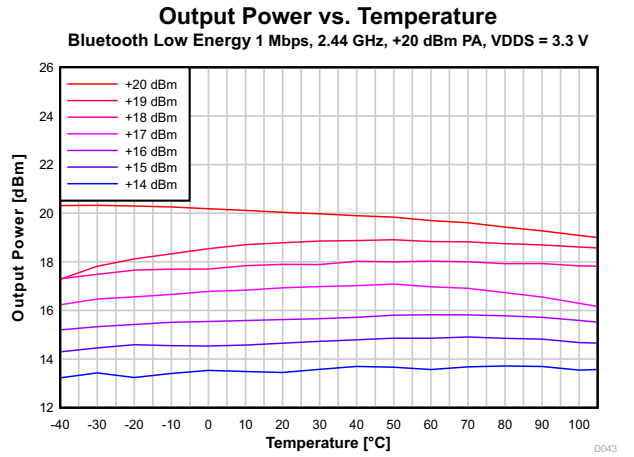


図 7-23. Output Power vs. Temperature (Bluetooth Low Energy 1 Mbps, 2.44 GHz, +20 dBm PA)

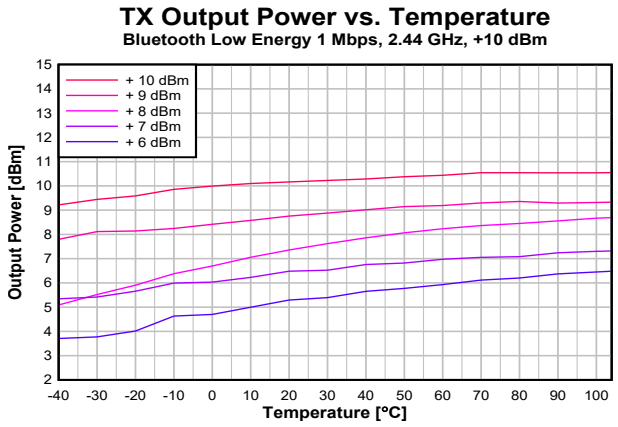


図 7-24. Output Power vs. Temperature (2.44 GHz, +10 dBm PA)

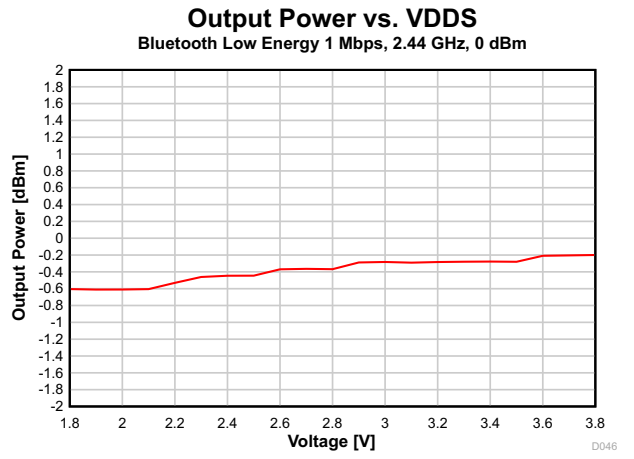


図 7-25. Output Power vs. Supply Voltage (VDDS) (Bluetooth Low Energy 1 Mbps, 2.44 GHz)

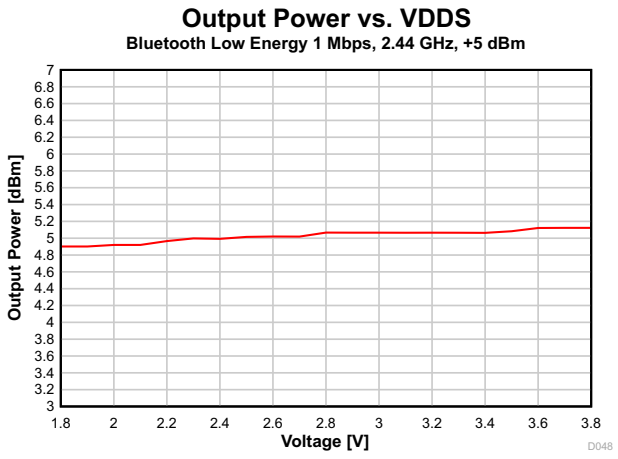


図 7-26. Output Power vs. Supply Voltage (VDDS) (Bluetooth Low Energy 1 Mbps, 2.44 GHz, +5 dBm)

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7.16.5 TX Performance (continued)

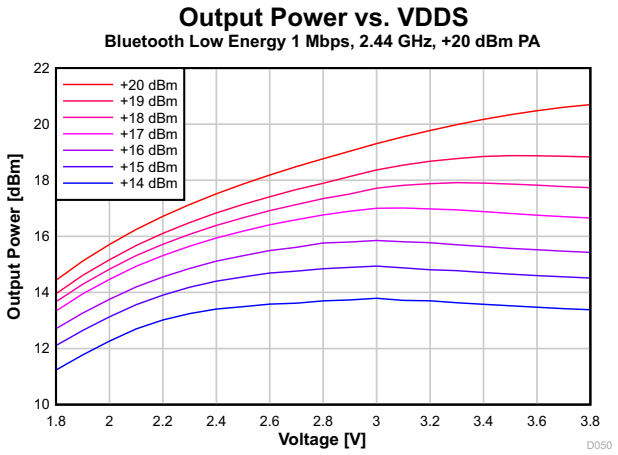


図 7-27. Output Power vs. Supply Voltage (VDDS) (Bluetooth Low Energy 1 Mbps, 2.44 GHz, +20 dBm PA)

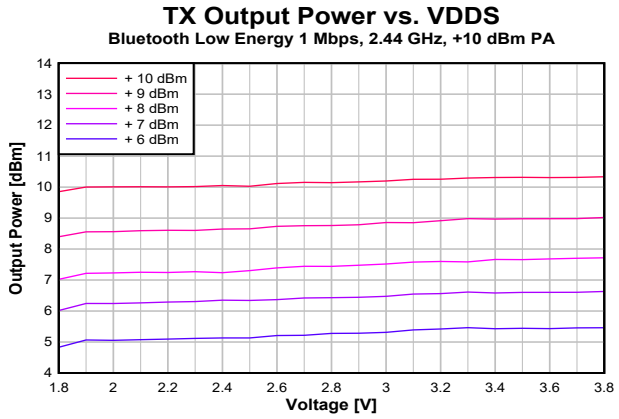


図 7-28. Output Power vs. Supply Voltage (VDDS) (2.44 GHz, +10 dBm PA)

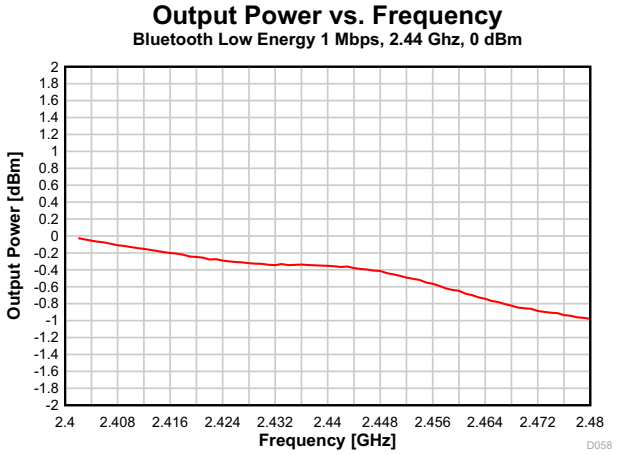


図 7-29. Output Power vs. Frequency (Bluetooth Low Energy 1 Mbps, 2.44 GHz)

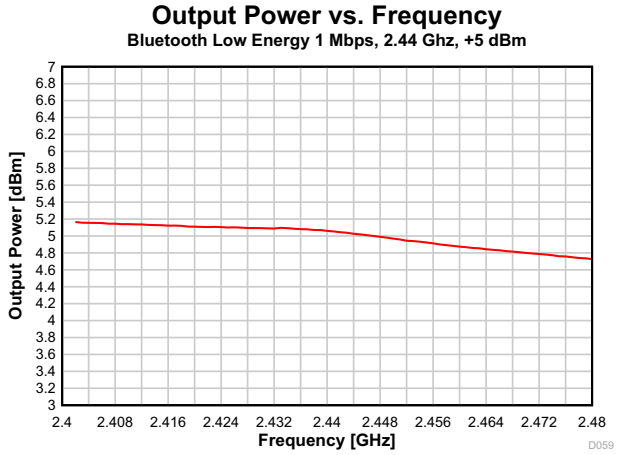


図 7-30. Output Power vs. Frequency (Bluetooth Low Energy 1 Mbps, 2.44 GHz, +5 dBm)

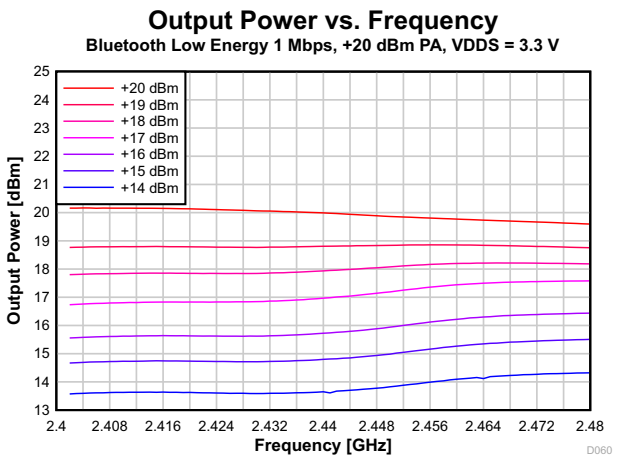


図 7-31. Output Power vs. Frequency (Bluetooth Low Energy 1 Mbps, 2.44 GHz, +20 dBm PA)

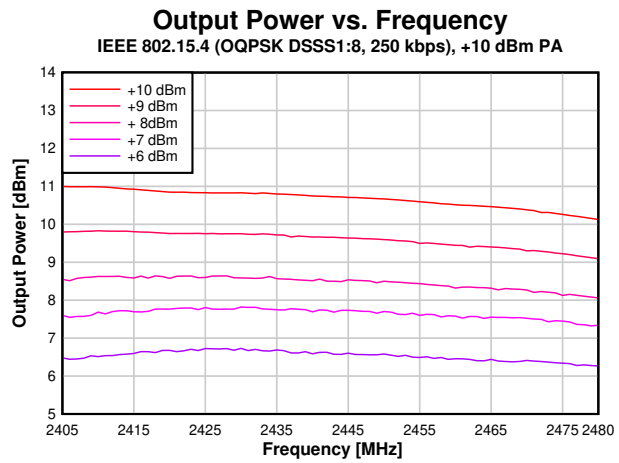


図 7-32. Output Power vs. Frequency (250 kbps, +10 dBm PA)

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7.16.6 ADC Performance

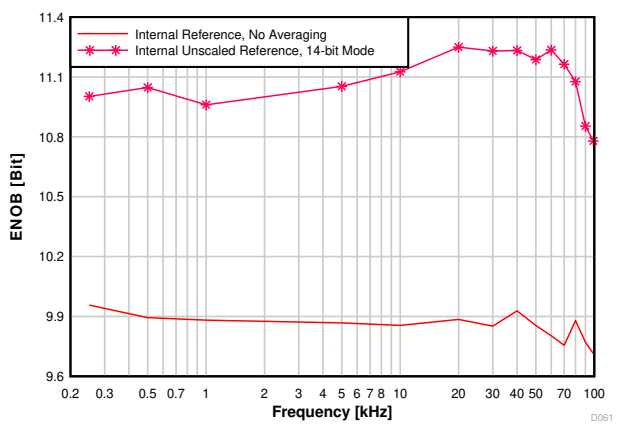


図 7-33. ENOB vs. Input Frequency

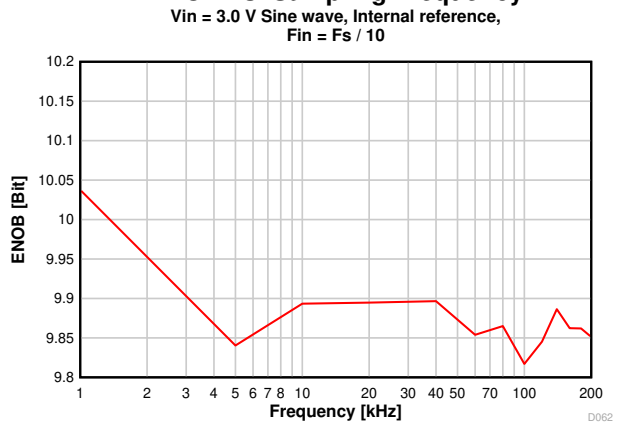


図 7-34. ENOB vs. Sampling Frequency

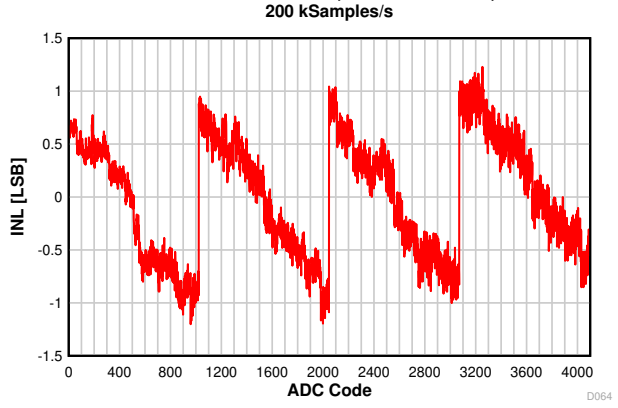


図 7-35. INL vs. ADC Code

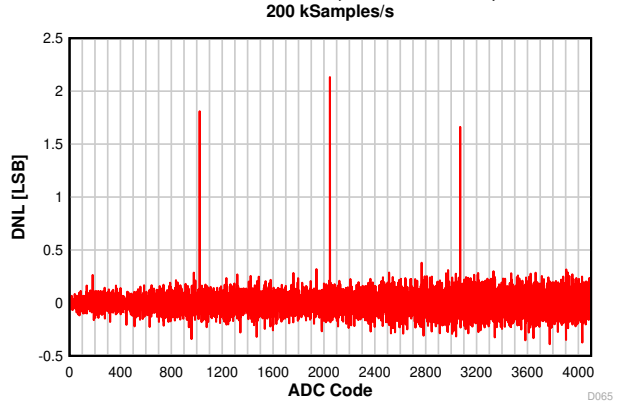


図 7-36. DNL vs. ADC Code

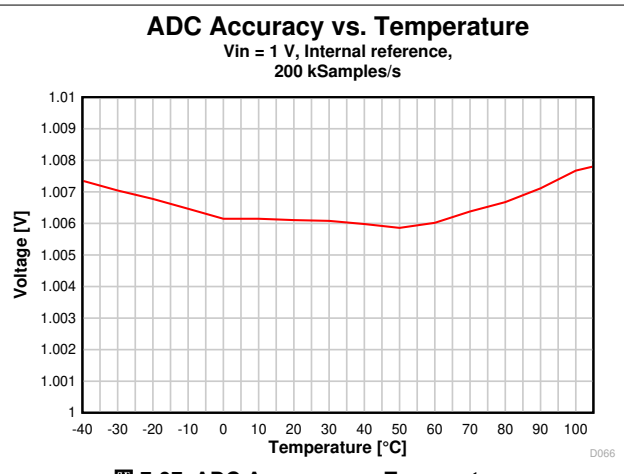


図 7-37. ADC Accuracy vs. Temperature

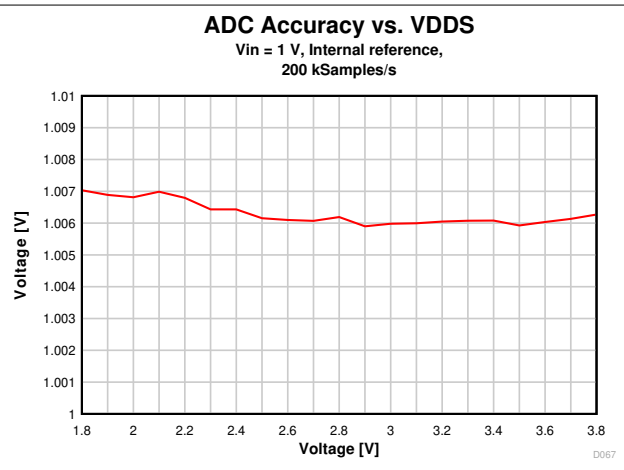


図 7-38. ADC Accuracy vs. Supply Voltage (VDDS)

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8 Detailed Description

8.1 Overview

セクション 3.1 shows the core modules of the CC2652P7 device.

8.2 System CPU

The CC2652P7 SimpleLink™ Wireless MCU contains an Arm® Cortex®-M4F system CPU, which runs the application and the higher layers of radio protocol stacks.

The system CPU is the foundation of a high-performance, low-cost platform that meets the system requirements of minimal memory implementation, and low-power consumption, while delivering outstanding computational performance and exceptional system response to interrupts.

Its features include the following:

- ARMv7-M architecture optimized for small-footprint embedded applications
- Arm Thumb®-2 mixed 16- and 32-bit instruction set delivers the high performance expected of a 32-bit Arm core in a compact memory size
- Fast code execution permits increased sleep mode time
- Deterministic, high-performance interrupt handling for time-critical applications
- Single-cycle multiply instruction and hardware divide
- Hardware division and fast digital-signal-processing oriented multiply accumulate
- Saturating arithmetic for signal processing
- IEEE 754-compliant single-precision Floating Point Unit (FPU)
- Memory Protection Unit (MPU) for safety-critical applications
- Full debug with data matching for watchpoint generation
  - Data Watchpoint and Trace Unit (DWT)
  - JTAG Debug Access Port (DAP)
  - Flash Patch and Breakpoint Unit (FPB)
- Trace support reduces the number of pins required for debugging and tracing
  - Instrumentation Trace Macrocell Unit (ITM)
  - Trace Port Interface Unit (TPIU) with asynchronous serial wire output (SWO)
- Optimized for single-cycle flash memory access
- Tightly connected to 8-KB 4-way random replacement cache for minimal active power consumption and wait states
- Ultra-low-power consumption with integrated sleep modes
- 48 MHz operation
- 1.25 DMIPS per MHz

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8.3 Radio (RF Core)

The RF Core is a highly flexible and future proof radio module which contains an Arm Cortex-M0 processor that interfaces the analog RF and base-band circuitry, handles data to and from the system CPU side, and assembles the information bits in a given packet structure. The RF core offers a high level, command-based API to the main CPU that configurations and data are passed through. The Arm Cortex-M0 processor is not programmable by customers and is interfaced through the TI-provided RF driver that is included with the SimpleLink Software Development Kit (SDK).

The RF core can autonomously handle the time-critical aspects of the radio protocols, thus offloading the main CPU, which reduces power and leaves more resources for the user application. Several signals are also available to control external circuitry such as RF switches or range extenders autonomously.

Multiprotocol solutions are enabled through time-sliced access of the radio, handled transparently for the application through the TI-provided RF driver and dual-mode manager.

The various physical layer radio formats are partly built as a software defined radio where the radio behavior is either defined by radio ROM contents or by non-ROM radio formats delivered in form of firmware patches with the SimpleLink SDKs. This allows the radio platform to be updated for support of future versions of standards even with over-the-air (OTA) updates while still using the same silicon.

8.3.1 Bluetooth 5.2 Low Energy

The RF Core offers full support for Bluetooth 5.2 Low Energy, including the high-speed 2-Mbps physical layer and the 500-kbps and 125-kbps long range PHYs (Coded PHY) through the TI provided Bluetooth 5.2 stack or through a high-level Bluetooth API. The Bluetooth 5.2 PHY and part of the controller are in radio and system ROM, providing significant savings in memory usage and more space available for applications.

The new high-speed mode allows data transfers up to 2 Mbps, twice the speed of Bluetooth 4.2 and five times the speed of Bluetooth 4.0, without increasing power consumption. In addition to faster speeds, this mode offers significant improvements for energy efficiency and wireless coexistence with reduced radio communication time.

Bluetooth 5.2 also enables unparalleled flexibility for adjustment of speed and range based on application needs, which capitalizes on the high-speed or long-range modes respectively. Data transfers are now possible at 2 Mbps, enabling development of applications using voice, audio, imaging, and data logging that were not previously an option using Bluetooth Low Energy. With high-speed mode, existing applications deliver faster responses, richer engagement, and longer battery life. Bluetooth 5.2 enables fast, reliable firmware updates.

8.3.2 802.15.4 (Thread, Zigbee, 6LoWPAN)

Through a dedicated IEEE radio API, the RF Core supports the 2.4-GHz IEEE 802.15.4-2011 physical layer (2 Mchips per second Offset-QPSK with DSSS 1:8), used in Thread, Zigbee, and 6LoWPAN protocols. The 802.15.4 PHY and MAC are in radio and system ROM. TI also provides royalty-free protocol stacks for Thread and Zigbee as part of the SimpleLink SDK, enabling a robust end-to-end solution.

8.4 Memory

The up to 704KB nonvolatile (flash) memory provides storage for code and data. The flash memory is in-system programmable and erasable. The last flash memory sector must contain a Customer Configuration section (CCFG) that is used by boot ROM and TI provided drivers to configure the device. This configuration is done through the ccfg.c source file that is included in all TI provided examples.

The ultra-low leakage system static RAM (SRAM) is split into four 32KB and one 16KB blocks and can be used for both storage of data and execution of code. Retention of SRAM contents in Standby power mode is enabled by default and included in Standby mode power consumption numbers. Parity checking for detection of bit errors in memory is built-in, which reduces chip-level soft errors and thereby increases reliability. System SRAM is always initialized to zeroes upon code execution from boot.

To improve code execution speed and lower power when executing code from nonvolatile memory, a 4-way nonassociative 8-KB cache is enabled by default to cache and prefetch instructions read by the system CPU. The cache can be used as a general-purpose RAM by enabling this feature in the Customer Configuration Area (CCFG).

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There is a 4KB ultra-low leakage SRAM available for use with the Sensor Controller Engine which is typically used for storing Sensor Controller programs, data and configuration parameters. This RAM is also accessible by the system CPU. The Sensor Controller RAM is not cleared to zeroes between system resets.

The ROM includes a TI-RTOS kernel and low-level drivers, as well as significant parts of selected radio stacks, which frees up flash memory for the application. The ROM also contains a serial (SPI and UART) bootloader that can be used for initial programming of the device.

8.5 Sensor Controller

The Sensor Controller contains circuitry that can be selectively enabled in both Standby and Active power modes. The peripherals in this domain can be controlled by the Sensor Controller Engine, which is a proprietary power-optimized CPU. This CPU can read and monitor sensors or perform other tasks autonomously; thereby significantly reducing power consumption and offloading the system CPU.

The Sensor Controller Engine is user programmable with a simple programming language that has syntax similar to C. This programmability allows for sensor polling and other tasks to be specified as sequential algorithms rather than static configuration of complex peripheral modules, timers, DMA, register programmable state machines, or event routing.

The main advantages are:

- Flexibility - data can be read and processed in unlimited manners while still ensuring ultra-low power
- 2 MHz low-power mode enables lowest possible handling of digital sensors
- Dynamic reuse of hardware resources
- 40-bit accumulator supporting multiplication, addition and shift
- Observability and debugging options

Sensor Controller Studio is used to write, test, and debug code for the Sensor Controller. The tool produces C driver source code, which the System CPU application uses to control and exchange data with the Sensor Controller. Typical use cases may be (but are not limited to) the following:

- Read analog sensors using integrated ADC or comparators
- Interface digital sensors using GPIOs, SPI, UART, or I^2C (I^2C and UART are bit-banged)
- Capacitive sensing
- Waveform generation
- Very low-power pulse counting (flow metering)
- Key scan

The peripherals in the Sensor Controller include the following:

- The low-power clocked comparator can be used to wake the system CPU from any state in which the comparator is active. A configurable internal reference DAC can be used in conjunction with the comparator. The output of the comparator can also be used to trigger an interrupt or the ADC.
- Capacitive sensing functionality is implemented through the use of a constant current source, a time-to-digital converter, and a comparator. The continuous time comparator in this block can also be used as a higher-accuracy alternative to the low-power clocked comparator. The Sensor Controller takes care of baseline tracking, hysteresis, filtering, and other related functions when these modules are used for capacitive sensing.
- The ADC is a 12-bit, 200-ksamples/s ADC with eight inputs and a built-in voltage reference. The ADC can be triggered by many different sources including timers, I/O pins, software, and comparators.
- The analog modules can connect to up to eight different GPIOs
- Dedicated SPI master with up to 6 MHz clock speed

The peripherals in the Sensor Controller can also be controlled from the main application processor.

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8.6 Cryptography

The CC2652P7 device comes with a wide set of modern cryptography-related hardware accelerators, drastically reducing code footprint and execution time for cryptographic operations. It also has the benefit of being lower power and improves availability and responsiveness of the system because the cryptography operations runs in a background hardware thread.
Together with a large selection of open-source cryptography libraries provided with the Software Development Kit (SDK), this allows for secure and future proof IoT applications to be easily built on top of the platform. The hardware accelerator modules are:

- **True Random Number Generator (TRNG)** module provides a true, nondeterministic noise source for the purpose of generating keys, initialization vectors (IVs), and other random number requirements. The TRNG is built on 24 ring oscillators that create unpredictable output to feed a complex nonlinear-combinatorial circuit.
- **Secure Hash Algorithm 2 (SHA-2)** with support for SHA224, SHA256, SHA384, and SHA512
- **Advanced Encryption Standard (AES)** with 128 and 256 bit key lengths
- **Public Key Accelerator** - Hardware accelerator supporting mathematical operations needed for elliptic curves up to 512 bits and RSA key pair generation up to 1024 bits.

Through use of these modules and the TI provided cryptography drivers, the following capabilities are available for an application or stack:

- **Key Agreement Schemes**
  - Elliptic curve Diffie–Hellman with static or ephemeral keys (ECDH and ECDHE)
  - Elliptic curve Password Authenticated Key Exchange by Juggling (ECJ-PAKE)
- **Signature Generation**
  - Elliptic curve Diffie-Hellman Digital Signature Algorithm (ECDSA)
- **Curve Support**
  - Short Weierstrass form (full hardware support), such as:
    - NIST-P224, NIST-P256, NIST-P384, NIST-P521
    - Brainpool-256R1, Brainpool-384R1, Brainpool-512R1
    - secp256r1
  - Montgomery form (hardware support for multiplication), such as:
    - Curve25519
- **SHA2 based MACs**
  - HMAC with SHA224, SHA256, SHA384, or SHA512
- Block cipher mode of operation
  - AESCCM
  - AESGCM
  - AESECB
  - AESCBC
  - AESCBC-MAC
- **True random number generation**

Other capabilities, such as RSA encryption and signatures as well as Edwards type of elliptic curves such as Curve1174 or Ed25519, can also be implemented using the provided hardware accelerators but are not part of the TI SimpleLink SDK for the CC2652P7 device.

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8.7 Timers

A large selection of timers are available as part of the CC2652P7 device. These timers are:

- **Real-Time Clock (RTC)**

  A 70-bit 3-channel timer running on the 32 kHz low frequency system clock (SCLK_LF). This timer is available in all power modes except Shutdown. The timer can be calibrated to compensate for frequency drift when using the RCOSC_LF as the low frequency system clock. If an external LF clock with frequency different from 32.768 kHz is used, the RTC tick speed can be adjusted to compensate for this. When using TI-RTOS, the RTC is used as the base timer in the operating system and should thus only be accessed through the kernel APIs such as the Clock module. The real time clock can also be read by the Sensor Controller Engine to timestamp sensor data and also has dedicated capture channels. By default, the RTC halts when a debugger halts the device.

- **General Purpose Timers (GPTIMER)**

  The four flexible GPTIMERs can be used as either 4× 32 bit timers or 8× 16 bit timers, all running on up to 48 MHz. Each of the 16- or 32-bit timers support a wide range of features such as one-shot or periodic counting, pulse width modulation (PWM), time counting between edges and edge counting. The inputs and outputs of the timer are connected to the device event fabric, which allows the timers to interact with signals such as GPIO inputs, other timers, DMA and ADC. The GPTIMERs are available in Active and Idle power modes.

- **Sensor Controller Timers**

  The Sensor Controller contains 3 timers:

  AUX Timer 0 and 1 are 16-bit timers with a 2^N prescaler. Timers can either increment on a clock or on each edge of a selected tick source. Both one-shot and periodical timer modes are available.

  AUX Timer 2 is a 16-bit timer that can operate at 24 MHz, 2 MHz or 32 kHz independent of the Sensor Controller functionality. There are 4 capture or compare channels, which can be operated in one-shot or periodical modes. The timer can be used to generate events for the Sensor Controller Engine or the ADC, as well as for PWM output or waveform generation.

- **Radio Timer**

  A multichannel 32-bit timer running at 4 MHz is available as part of the device radio. The radio timer is typically used as the timing base in wireless network communication using the 32-bit timing word as the network time. The radio timer is synchronized with the RTC by using a dedicated radio API when the device radio is turned on or off. This ensures that for a network stack, the radio timer seems to always be running when the radio is enabled. The radio timer is in most cases used indirectly through the trigger time fields in the radio APIs and should only be used when the accurate 48 MHz high frequency crystal is the source of SCLK_HF.

- **Watchdog timer**

  The watchdog timer is used to regain control if the system operates incorrectly due to software errors. It is typically used to generate an interrupt to and reset of the device for the case where periodic monitoring of the system components and tasks fails to verify proper functionality. The watchdog timer runs on a 1.5 MHz clock rate and cannot be stopped once enabled. The watchdog timer pauses to run in Standby power mode and when a debugger halts the device.

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8.8 Serial Peripherals and I/O

The SSIs are synchronous serial interfaces that are compatible with SPI, MICROWIRE, and TI's synchronous serial interfaces. The SSIs support both SPI master and slave up to 4 MHz. The SSI modules support configurable phase and polarity.

The UARTs implement universal asynchronous receiver and transmitter functions. They support flexible baud-rate generation up to a maximum of 3 Mbps.

The I^2S interface is used to handle digital audio and can also be used to interface pulse-density modulation microphones (PDM).

The I^2C interface is used to communicate with devices compatible with the I^2C standard. The I^2C interface can handle 100 kHz and 400 kHz operation, and can serve as both master and slave.

The I/O controller (IOC) controls the digital I/O pins and contains multiplexer circuitry to allow a set of peripherals to be assigned to I/O pins in a flexible manner. All digital I/Os are interrupt and wake-up capable, have a programmable pullup and pulldown function, and can generate an interrupt on a negative or positive edge (configurable). When configured as an output, pins can function as either push-pull or open-drain. Five GPIOs have high-drive capabilities, which are marked in **bold** in セクション 6. All digital peripherals can be connected to any digital pin on the device.

For more information, see the CC13x2x7, CC26x2x7 SimpleLink™ Wireless MCU Technical Reference Manual.

8.9 Battery and Temperature Monitor

A combined temperature and battery voltage monitor is available in the CC2652P7 device. The battery and temperature monitor allows an application to continuously monitor on-chip temperature and supply voltage and respond to changes in environmental conditions as needed. The module contains window comparators to interrupt the system CPU when temperature or supply voltage go outside defined windows. These events can also be used to wake up the device from Standby mode through the Always-On (AON) event fabric.

8.10 µDMA

The device includes a direct memory access (µDMA) controller. The µDMA controller provides a way to offload data-transfer tasks from the system CPU, thus allowing for more efficient use of the processor and the available bus bandwidth. The µDMA controller can perform a transfer between memory and peripherals. The µDMA controller has dedicated channels for each supported on-chip module and can be programmed to automatically perform transfers between peripherals and memory when the peripheral is ready to transfer more data.

Some features of the µDMA controller include the following (this is not an exhaustive list):

- Highly flexible and configurable channel operation of up to 32 channels
- Transfer modes: memory-to-memory, memory-to-peripheral, peripheral-to-memory, and peripheral-to-peripheral
- Data sizes of 8, 16, and 32 bits
- Ping-pong mode for continuous streaming of data

8.11 Debug

The on-chip debug support is done through a dedicated cJTAG (IEEE 1149.7) or JTAG (IEEE 1149.1) interface. The device boots by default into cJTAG mode and must be reconfigured to use 4-pin JTAG.

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8.12 Power Management

To minimize power consumption, the CC2652P7 supports a number of power modes and power management features (see 表 8-1).

表 8-1. Power Modes

MODE	SOFTWARE CONFIGURABLE POWER MODES				RESET PIN HELD
	ACTIVE	IDLE	STANDBY	SHUTDOWN	
CPU	Active	Off	Off	Off	Off
Flash	On	Available	Off	Off	Off
SRAM	On	On	Retention	Off	Off
Supply System	On	On	Duty Cycled	Off	Off
Register and CPU retention	Full	Full	Partial	No	No
SRAM retention	Full	Full	Full	No	No
48 MHz high-speed clock (SCLK_HF)	XOSC_HF or RCOSC_HF	XOSC_HF or RCOSC_HF	Off	Off	Off
2 MHz medium-speed clock (SCLK_MF)	RCOSC_MF	RCOSC_MF	Available	Off	Off
32 kHz low-speed clock (SCLK_LF)	XOSC_LF or RCOSC_LF	XOSC_LF or RCOSC_LF	XOSC_LF or RCOSC_LF	Off	Off
Peripherals	Available	Available	Off	Off	Off
Sensor Controller	Available	Available	Available	Off	Off
Wake-up on RTC	Available	Available	Available	Off	Off
Wake-up on pin edge	Available	Available	Available	Available	Off
Wake-up on reset pin	On	On	On	On	On
Brownout detector (BOD)	On	On	Duty Cycled	Off	Off
Power-on reset (POR)	On	On	On	Off	Off
Watchdog timer (WDT)	Available	Available	Paused	Off	Off

In **Active** mode, the application system CPU is actively executing code. Active mode provides normal operation of the processor and all of the peripherals that are currently enabled. The system clock can be any available clock source (see 表 8-1).

In **Idle** mode, all active peripherals can be clocked, but the Application CPU core and memory are not clocked and no code is executed. Any interrupt event brings the processor back into active mode.

In **Standby** mode, only the always-on (AON) domain is active. An external wake-up event, RTC event, or Sensor Controller event is required to bring the device back to active mode. MCU peripherals with retention do not need to be reconfigured when waking up again, and the CPU continues execution from where it went into standby mode. All GPIOs are latched in standby mode.

In **Shutdown** mode, the device is entirely turned off (including the AON domain and Sensor Controller), and the I/Os are latched with the value they had before entering shutdown mode. A change of state on any I/O pin defined as a *wake from shutdown pin* wakes up the device and functions as a reset trigger. The CPU can differentiate between reset in this way and reset-by-reset pin or power-on reset by reading the reset status register. The only state retained in this mode is the latched I/O state and the flash memory contents.

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The Sensor Controller is an autonomous processor that can control the peripherals in the Sensor Controller independently of the system CPU. This means that the system CPU does not have to wake up, for example to perform an ADC sampling or poll a digital sensor over SPI, thus saving both current and wake-up time that would otherwise be wasted. The Sensor Controller Studio tool enables the user to program the Sensor Controller, control its peripherals, and wake up the system CPU as needed. All Sensor Controller peripherals can also be controlled by the system CPU.

Note

The power, RF and clock management for the CC2652P7 device require specific configuration and handling by software for optimized performance. This configuration and handling is implemented in the TI-provided drivers that are part of the SimpleLink™ CC13xx and CC26xx software development kit (SDK). Therefore, TI highly recommends using this software framework for all application development on the device. The complete SDK with TI-RTOS (optional), device drivers, and examples are offered free of charge in source code.

8.13 Clock Systems

The CC2652P7 device has several internal system clocks.

The 48 MHz SCLK_HF is used as the main system (MCU and peripherals) clock. This can be driven by the internal 48 MHz RC Oscillator (RCOSC_HF) or an external 48 MHz crystal (XOSC_HF). Radio operation requires an external 48 MHz crystal.

SCLK_MF is an internal 2 MHz clock that is used by the Sensor Controller in low-power mode and also for internal power management circuitry. The SCLK_MF clock is always driven by the internal 2 MHz RC Oscillator (RCOSC_MF).

SCLK_LF is the 32.768 kHz internal low-frequency system clock. It can be used by the Sensor Controller for ultra-low-power operation and is also used for the RTC and to synchronize the radio timer before or after Standby power mode. SCLK_LF can be driven by the internal 32.8 kHz RC Oscillator (RCOSC_LF), a 32.768 kHz watch-type crystal, or a clock input on any digital IO.

When using a crystal or the internal RC oscillator, the device can output the 32 kHz SCLK_LF signal to other devices, thereby reducing the overall system cost.

8.14 Network Processor

Depending on the product configuration, the CC2652P7 device can function as a wireless network processor (WNP - a device running the wireless protocol stack with the application running on a separate host MCU), or as a system-on-chip (SoC) with the application and protocol stack running on the device's system CPU inside the device.

In the first case, the external host MCU communicates with the device using SPI or UART. In the second case, the application must be written according to the application framework supplied with the wireless protocol stack.

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9 Application, Implementation, and Layout

Note

Information in the following Applications section is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

For general design guidelines and hardware configuration guidelines, refer to CC13xx/CC26xx Hardware Configuration and PCB Design Considerations Application Report.

For optimum RF performance, especially when using the high-power PA, it is important to accurately follow the reference design with respect to component values and layout. Failure to do so may lead to reduced RF performance due to balun mismatch. For the high-power PA, the amplitude- and phase balance through the balun must be <1 dB and <6 degrees, respectively.

PCB stack-up is also critical for proper operation. The CC1352P7 EVMs and characterization boards are using a finished thickness between the top layer (RF signals) and layer 2 (ground plane) of 175 µm. It is very important to use the same substrate thickness, or slightly thicker, in an end product implementing the CC2652P7 device.

9.1 Reference Designs

The following reference designs should be followed closely when implementing designs using the CC2652P7 device.

Special attention must be paid to RF component placement, decoupling capacitors and DCDC regulator components, as well as ground connections for all of these.

All the CC1352P7 device reference designs are also applicable to the CC2652P7 device by simply disregarding the sub-1 GHz RF circuitry. For the CC2652P7 device, pins 3 and 4 must be left unconnected.

CC1352-P7EM-XD7793-XD24-PA24 Design Files	The differential CC1352-P7EM-XD7793-XD24-PA24 reference design provides schematic, layout and production files for the characterization board used for deriving the performance number found in this document. This board includes tuning for 20 dBm operation at 2.4 GHz on the high-power PA output. For the CC2652P7 device, the sub-1 GHz RF circuitry can be disregarded.
CC1352-P7EM-XD7793-XD24-PA24_10dBm Design Files	The differential CC1352-P7EM-XD7793-XD24-PA24_10dBm reference design provides schematic, layout and production files for the characterization board used for deriving the performance number found in this document. This board includes tuning for 10 dBm operation at 2.4 GHz on the high-power PA output. For the CC2652P7 device, the sub-1 GHz RF circuitry can be disregarded.
LP-CC1352P7-4 Design Files	Detailed schematics and layouts for the multi-band CC1352P7 LaunchPad evaluation board featuring 2.4 GHz RF matching optimized for 10 dBm operation on the 20 dBm PA output and up to 13 dBm TX power at 433 MHz. For evaluation of 20 dBm operation at 2.4 GHz the BOM can be modified as described in the schematics available in the Design Files. For CC2652P7, the sub-1 GHz RF circuitry can be disregarded.
Sub-1 GHz and 2.4 GHz Antenna Kit for LaunchPad™ Development Kit and SensorTag	The antenna kit allows real-life testing to identify the optimal antenna for your application. The antenna kit includes 16 antennas for frequencies from 169 MHz to 2.4 GHz, including: • PCB antennas • Helical antennas

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- Chip antennas
- Dual-band antennas for 868 MHz and 915 MHz combined with 2.4 GHz

The antenna kit includes a JSC cable to connect to the Wireless MCU LaunchPad Development Kits and SensorTags.

9.2 Junction Temperature Calculation

This section shows the different techniques for calculating the junction temperature under various operating conditions. For more details, see *Semiconductor and IC Package Thermal Metrics*.

There are three recommended ways to derive the junction temperature from other measured temperatures:

1. From package temperature:

$$T_J = \psi_{\mathrm{JT}} \times P + T_{\mathrm{case}} \quad (1)$$

2. From board temperature:

$$T_J = \psi_{\mathrm{JB}} \times P + T_{\mathrm{board}} \quad (2)$$

3. From ambient temperature:

$$T_J = R_{\theta\mathrm{JA}} \times P + T_A \quad (3)$$

P is the power dissipated from the device and can be calculated by multiplying current consumption with supply voltage. Thermal resistance coefficients are found in セクション 7.8.

Example:

Using 式 3, the temperature difference between ambient temperature and junction temperature is calculated. In this example, we assume a simple use case where the radio is transmitting continuously at 20dBm output power. Let us assume the ambient temperature is 105 °C and the supply voltage is 3.3 V. To calculate P, we need to look up the current consumption for Tx at 105 °C in 図 7-9. From the plot, we see that the current consumption is 101 mA. This means that P is 3.3 V × 101 mA = 303.0 mW.

The junction temperature is then calculated as:

$$T_J = 23.4^{\circ C}/_{W} \times 303.0mW + T_A = 7.1^{\circ}C + T_A \quad (4)$$

As can be seen from the example, the junction temperature is 7.1 °C higher than the ambient temperature when running continuous Tx at 105 °C and, thus, well within the recommended operating conditions of 115 °C.

For various application use cases current consumption for other modules may have to be added to calculate the appropriate power dissipation. For example, the MCU may be running simultaneously as the radio, peripheral modules may be enabled, etc. Typically, the easiest way to find the peak current consumption, and thus the peak power dissipation in the device, is to measure as described in *Measuring CC13xx and CC26xx Current Consumption*.

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10 Device and Documentation Support

TI offers an extensive line of development tools. Tools and software to evaluate the performance of the device, generate code, and develop solutions are listed as follows.

10.1 Device Nomenclature

To designate the stages in the product development cycle, TI assigns prefixes to all part numbers and/or date-code. Each device has one of three prefixes/identifications: X, P, or null (no prefix) (for example, XCC2652P7 is in preview; therefore, an X prefix/identification is assigned).

Device development evolutionary flow:

X Experimental device that is not necessarily representative of the final device's electrical specifications and may not use production assembly flow.

P Prototype device that is not necessarily the final silicon die and may not necessarily meet final electrical specifications.

null Production version of the silicon die that is fully qualified.

Production devices have been characterized fully, and the quality and reliability of the device have been demonstrated fully. TI's standard warranty applies.

Predictions show that prototype devices (X or P) have a greater failure rate than the standard production devices. Texas Instruments recommends that these devices not be used in any production system because their expected end-use failure rate still is undefined. Only qualified production devices are to be used.

TI device nomenclature also includes a suffix with the device family name. This suffix indicates the package type (for example, *RGZ*).

For orderable part numbers of *CC2652P7* devices in the RGZ (7-mm x 7-mm) package type, see the *Package Option Addendum* of this document, the Device Information in セクション 3, the TI website (www.ti.com), or contact your TI sales representative.

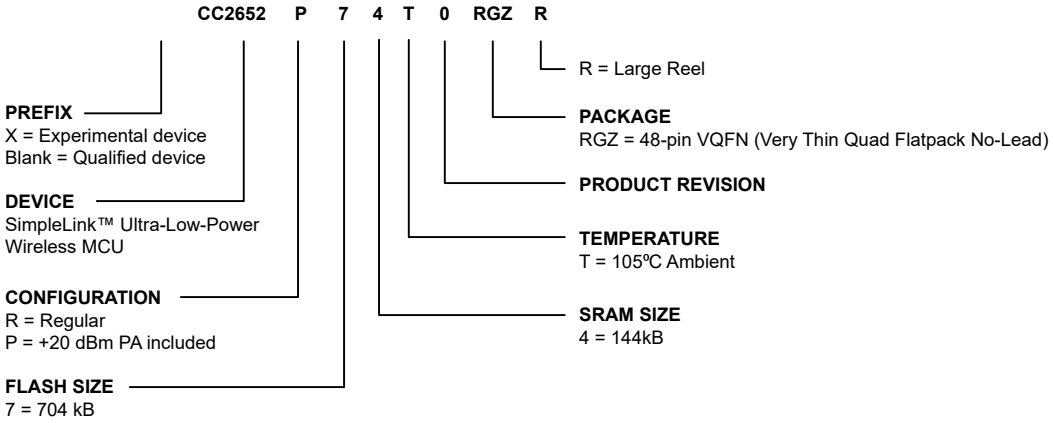


図 10-1. Device Nomenclature

10.2 Tools and Software

The CC2652P7 device is supported by a variety of software and hardware development tools.

Development Kit

CC1352P7-4 LaunchPad™ Development Kit

The CC1352P7-4 LaunchPad™ Development Kit enables development of high-performance wireless applications in the 433 MHz and 2.4 GHz frequency bands that benefit from low-power operation. The kit features the CC1352P7 dual-band and multiprotocol SimpleLink Wireless MCU with an integrated High Power Amplifier. The kit works with the LaunchPad

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ecosystem, easily enabling additional functionality like sensors, display, and more. The built-in EnergyTrace™ software is an energy-based code analysis tool that measures and displays the application's energy profile and helps to optimize it for ultra-low-power consumption.

The RF configuration of the LaunchPad enables up to +13 dBm output power for 433 MHz and +10 dBm output power for 2.4 GHz. The LaunchPad can also be used as a development kit when evaluating other device family devices such as CC1312R7 for use with 433 MHz frequency bands or CC2652P7 for +10 dBm in the 2.4 GHz band.

For evaluation of +20 dBm operation at 2.4 GHz the BOM can be modified as described in the schematics available in the Design Files.

Software

SimpleLink™ CC13XX-CC26XX SDK

The SimpleLink CC13xx and CC26xx Software Development Kit (SDK) provides a complete package for the development of wireless applications on the CC13XX / CC26XX family of devices. The SDK includes a comprehensive software package for the CC2652P7 device, including the following protocol stacks:

- Bluetooth Low Energy 4 and 5.2
- Thread (based on OpenThread)
- Zigbee 3.0
- Wi-SUN®
- TI 15.4-Stack - an IEEE 802.15.4-based star networking solution for Sub-1 GHz and 2.4 GHz
- Proprietary RF - a large set of building blocks for building proprietary RF software
- Multiprotocol support - concurrent operation between stacks using the Dynamic Multiprotocol Manager (DMM)

The SimpleLink CC13XX-CC26XX SDK is part of TI's SimpleLink MCU platform, offering a single development environment that delivers flexible hardware, software and tool options for customers developing wired and wireless applications. For more information about the SimpleLink MCU Platform, visit http://www.ti.com/simplelink.

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Development Tools

Code Composer Studio™ Integrated Development Environment (IDE)

Code Composer Studio is an integrated development environment (IDE) that supports TI's Microcontroller and Embedded Processors portfolio. Code Composer Studio comprises a suite of tools used to develop and debug embedded applications. It includes an optimizing C/C++ compiler, source code editor, project build environment, debugger, profiler, and many other features. The intuitive IDE provides a single user interface taking you through each step of the application development flow. Familiar tools and interfaces allow users to get started faster than ever before. Code Composer Studio combines the advantages of the Eclipse® software framework with advanced embedded debug capabilities from TI resulting in a compelling feature-rich development environment for embedded developers.

CCS has support for all SimpleLink Wireless MCUs and includes support for EnergyTrace™ software (application energy usage profiling). A real-time object viewer plugin is available for TI-RTOS, part of the SimpleLink SDK.

Code Composer Studio is provided free of charge when used in conjunction with the XDS debuggers included on a LaunchPad Development Kit.

Code Composer Studio™ Cloud IDE

Code Composer Studio (CCS) Cloud is a web-based IDE that allows you to create, edit and build CCS and Energia™ projects. After you have successfully built your project, you can download and run on your connected LaunchPad. Basic debugging, including features like setting breakpoints and viewing variable values is now supported with CCS Cloud.

IAR Embedded Workbench® for Arm®

IAR Embedded Workbench® is a set of development tools for building and debugging embedded system applications using assembler, C and C++. It provides a completely integrated development environment that includes a project manager, editor, and build tools. IAR has support for all SimpleLink Wireless MCUs. It offers broad debugger support, including XDS110, IAR I-jet™ and Segger J-Link™. A real-time object viewer plugin is available for TI-RTOS, part of the SimpleLink SDK. IAR is also supported out-of-the-box on most software examples provided as part of the SimpleLink SDK.

A 30-day evaluation or a 32 KB size-limited version is available through iar.com.

SmartRF™ Studio

SmartRF™ Studio is a Windows® application that can be used to evaluate and configure SimpleLink Wireless MCUs from Texas Instruments. The application will help designers of RF systems to easily evaluate the radio at an early stage in the design process. It is especially useful for generation of configuration register values and for practical testing and debugging of the RF system. SmartRF Studio can be used either as a standalone application or together with applicable evaluation boards or debug probes for the RF device. Features of the SmartRF Studio include:

- Link tests - transmit and receive packets between nodes
- Antenna and radiation tests - set the radio in continuous wave TX and RX states
- Export radio configuration code for use with the TI SimpleLink SDK RF driver
- Custom GPIO configuration for signaling and control of external switches

Sensor Controller Studio

Sensor Controller Studio is used to write, test and debug code for the Sensor Controller peripheral. The tool generates a Sensor Controller Interface driver, which is a set of C source files that are compiled into the System CPU application. These source files also contain the Sensor Controller binary image and allow the System CPU application to control and exchange data with the Sensor Controller. Features of the Sensor Controller Studio include:

- Ready-to-use examples for several common use cases
- Full toolchain with built-in compiler and assembler for programming in a C-like programming language

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- Provides rapid development by using the integrated sensor controller task testing and debugging functionality, including visualization of sensor data and verification of algorithms

CCS UniFlash — CCS UniFlash is a standalone tool used to program on-chip flash memory on TI MCUs. UniFlash has a GUI, command line, and scripting interface. CCS UniFlash is available free of charge.

10.2.1 SimpleLink™ Microcontroller Platform

The SimpleLink microcontroller platform sets a new standard for developers with the broadest portfolio of wired and wireless Arm® MCUs (System-on-Chip) in a single software development environment. Delivering flexible hardware, software and tool options for your IoT applications. Invest once in the SimpleLink software development kit and use throughout your entire portfolio. Learn more on ti.com/simplelink.

10.3 Documentation Support

To receive notification of documentation updates on data sheets, errata, application notes and similar, navigate to the device product folder on ti.com/product/CC2652P7. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

The current documentation that describes the MCU, related peripherals, and other technical collateral is listed as follows.

TI Resource Explorer

TI Resource Explorer — Software examples, libraries, executables, and documentation are available for your device and development board.

Errata

CC2652P7 Silicon Errata — The silicon errata describes the known exceptions to the functional specifications for each silicon revision of the device and description on how to recognize a device revision.

Application Reports

All application reports for the CC2652P7 device are found on the device product folder at: ti.com/product/CC2652P7/#tech-docs.

Technical Reference Manual (TRM)

CC13x2x7, CC26x2x7 SimpleLink™ Wireless MCU TRM — The TRM provides a detailed description of all modules and peripherals available in the device family.

10.4 サポート・リソース

TI E2E™ サポート・フォーラムは、エンジニアが検証済みの回答と設計に関するヒントをエキスパートから迅速かつ直接得ることができる場所です。既存の回答を検索したり、独自の質問をしたりすることで、設計で必要な支援を迅速に得ることができます。

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10.6 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

10.7 Glossary

TI Glossary	This glossary lists and explains terms, acronyms, and definitions.

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11 Mechanical, Packaging, and Orderable Information

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PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package | Pins	Package qty | Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
CC2652P74T0RGZR	Active	Production	VQFN (RGZ) | 48	2500 | LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 105	CC2652 P74
CC2652P74T0RGZR.A	Active	Production	VQFN (RGZ) | 48	2500 | LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 105	CC2652 P74
CC2652P74T0RGZR.B	Active	Production	VQFN (RGZ) | 48	2500 | LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 105	CC2652 P74

(1) **Status:** For more details on status, see our product life cycle.

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

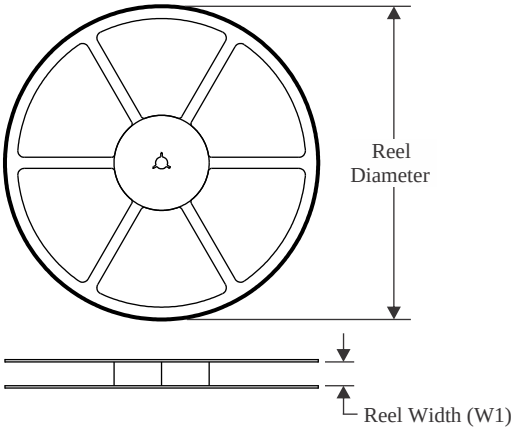
Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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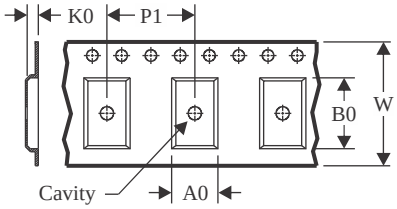
In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

TAPE AND REEL INFORMATION

REEL DIMENSIONS

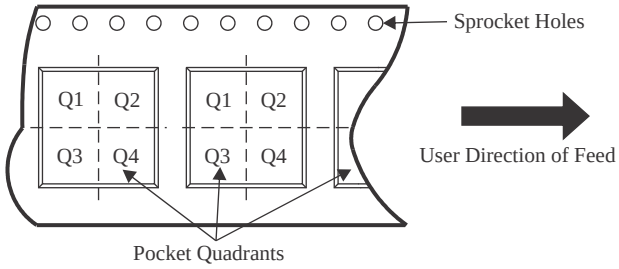


TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

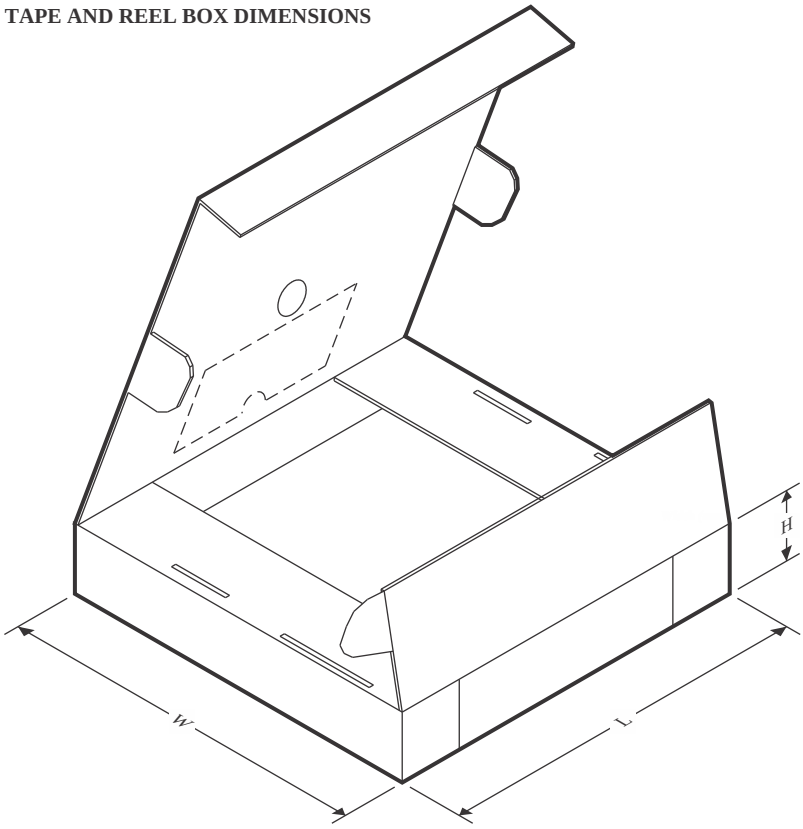
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
CC2652P74T0RGZR	VQFN	RGZ	48	2500	330.0	16.4	7.3	7.3	1.1	12.0	16.0	Q2
CC2652P74T0RGZR	VQFN	RGZ	48	2500	330.0	16.4	7.3	7.3	1.1	12.0	16.0	Q2
CC2652P74T0RGZR	VQFN	RGZ	48	2500	330.0	16.4	7.3	7.3	1.1	12.0	16.0	Q2
CC2652P74T0RGZR	VQFN	RGZ	48	2500	330.0	16.4	7.3	7.3	1.1	12.0	16.0	Q2

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
CC2652P74T0RGZR	VQFN	RGZ	48	2500	360.0	360.0	36.0
CC2652P74T0RGZR	VQFN	RGZ	48	2500	367.0	367.0	35.0
CC2652P74T0RGZR	VQFN	RGZ	48	2500	336.6	336.6	31.8
CC2652P74T0RGZR	VQFN	RGZ	48	2500	353.0	353.0	32.0

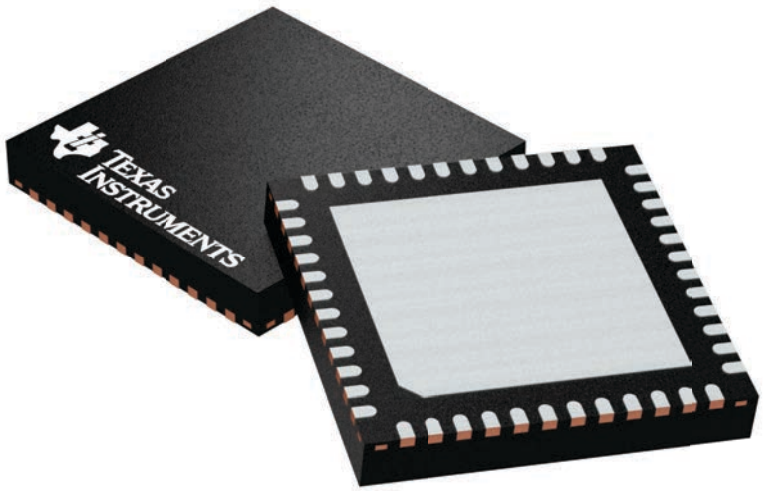
GENERIC PACKAGE VIEW

RGZ 48

VQFN - 1 mm max height

7 x 7, 0.5 mm pitch

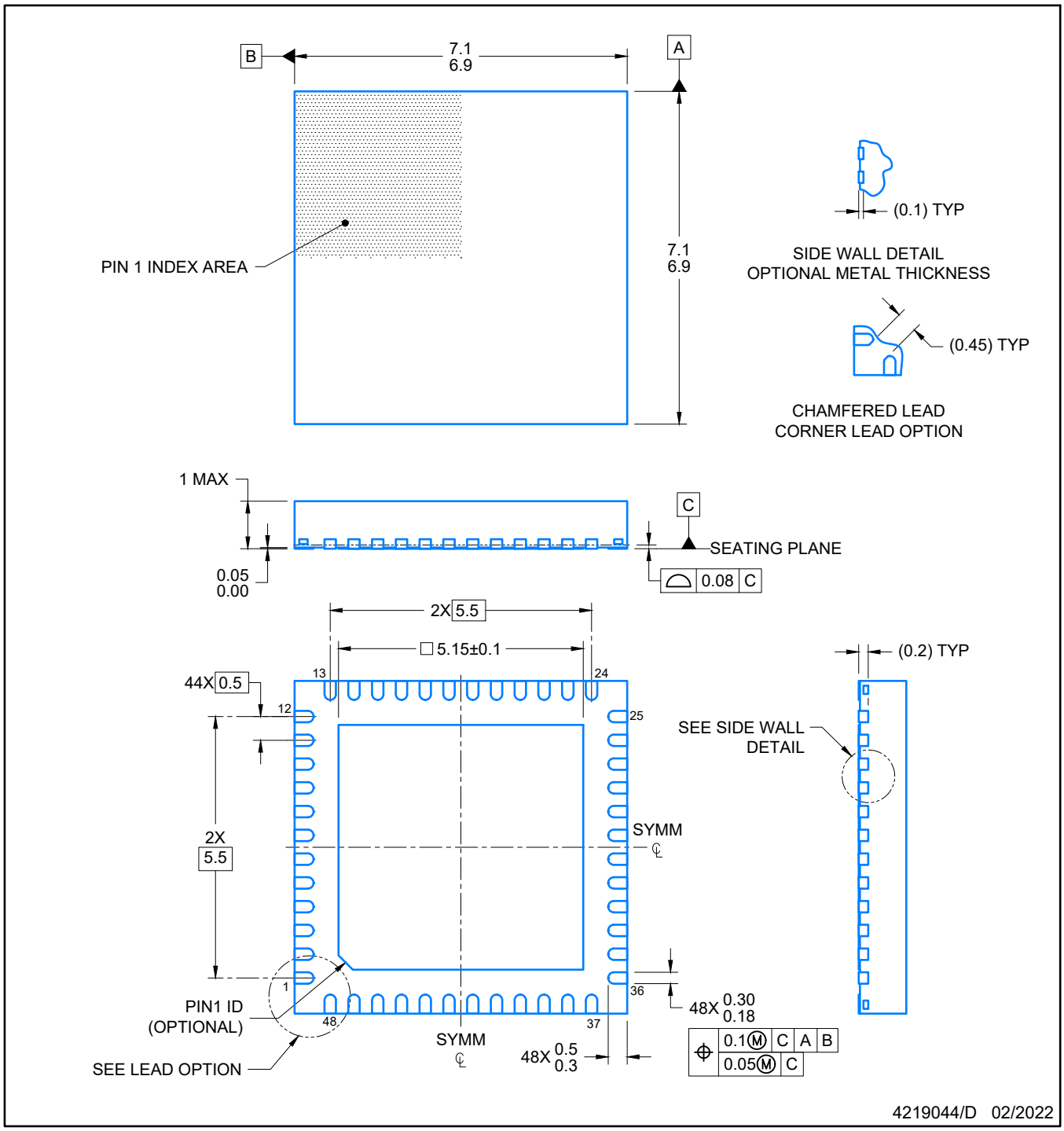
PLASTIC QUADFLAT PACK- NO LEAD



Images above are just a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.

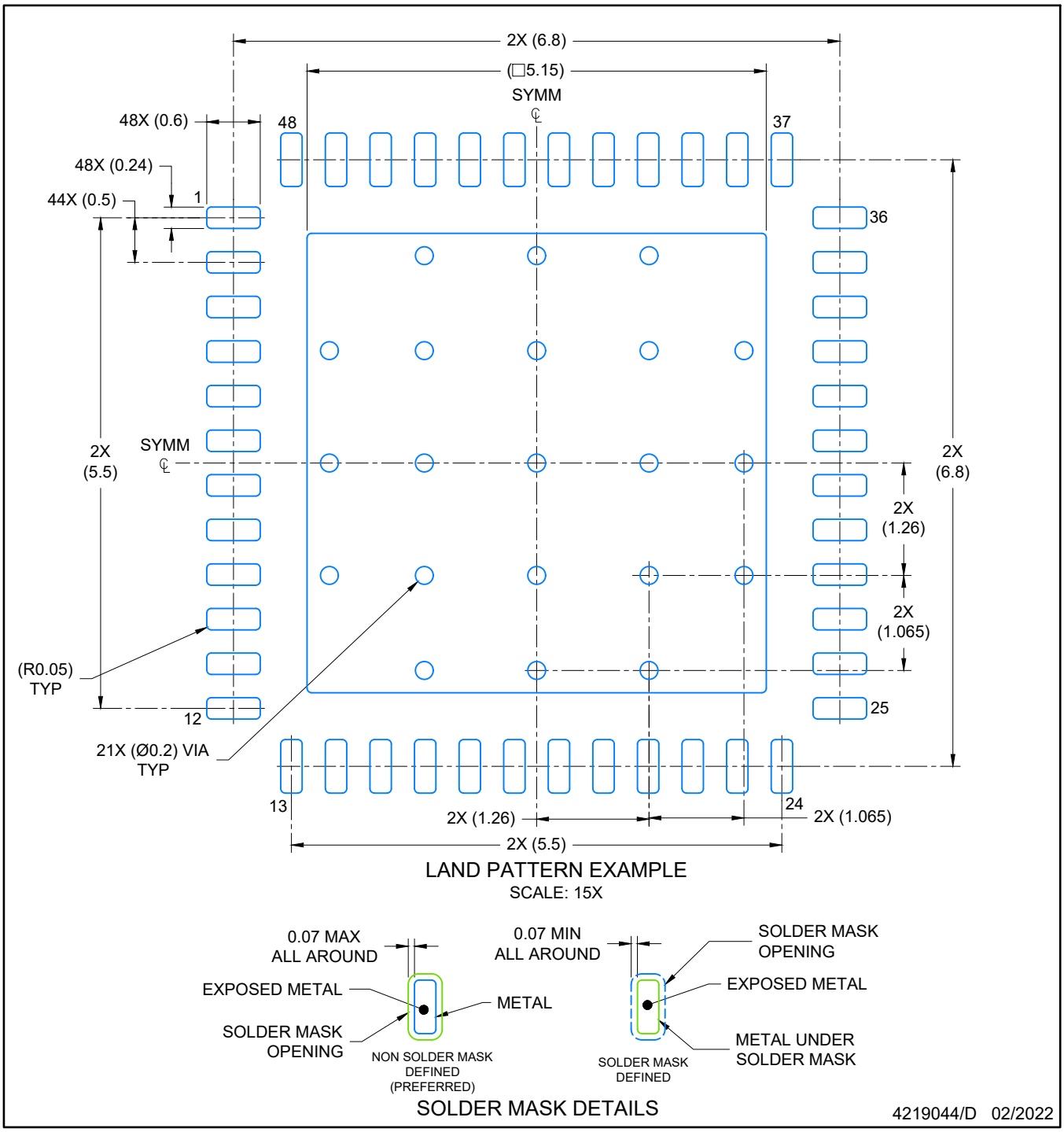
4224671/A

PLASTIC QUADFLAT PACK- NO LEAD



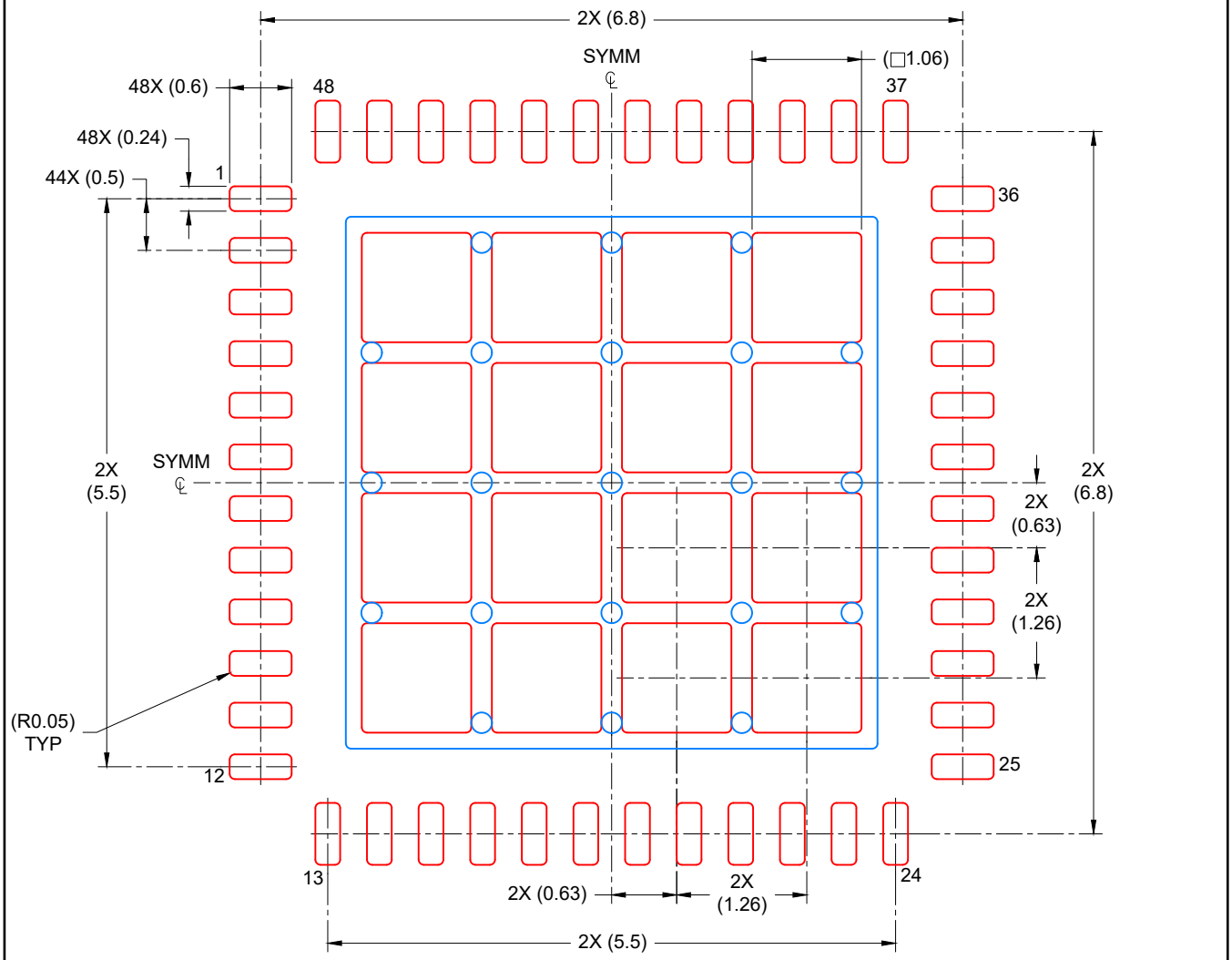
NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for optimal thermal and mechanical performance.



NOTES: (continued)

4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD
67% PRINTED COVERAGE BY AREA
SCALE: 15X

4219044/D 02/2022

NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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