



LM2991 可変型低ドロップアウト負電圧レギュレータ

1 特長

- 出力電圧は-3V～-24V(通常-2V～-25V)の範囲で可変
- 1Aを超える出力電流
- ドロップアウト電圧: 0.6V (1A負荷時の標準値)
- 低い静止電流
- 内部的な短絡電流制限
- 内部的なサーマル・シャットダウン、ヒステリシスあり
- TTL、CMOS互換の $\overline{\text{ON/OFF}}$ スイッチ
- LM2941シリーズを機能的に補完

2 アプリケーション

- スイッチャ後のレギュレータ
- ローカル、オンカード・レギュレーション
- バッテリー駆動の機器
- 産業用
- 計装機器

3 概要

LM2991は低ドロップアウトの可変負電圧レギュレータで、出力電圧範囲は-3V～-24Vです。LM2991は最大1Aの負荷電流を供給し、 $\overline{\text{ON/OFF}}$ ピンを搭載しているためリモートでシャットダウン可能です。

LM2991は新しい回路設計手法により、低いドロップアウト電圧、低い静止電流、低い温度ドリフト係数の高精度基準電圧を実現しています。1A負荷電流でのドロップアウト電圧は通常0.6V (標準値)で、動作温度範囲の全体にわたってワーストケースの1V (最大値)が保証されています。静止電流は負荷電流1A、入出力の電圧差分が3Vを超えると、1mA (標準値)です。内部バイアス電源の独自の回路設計により、レギュレータがドロップアウト・モード($V_{\text{OUT}} - V_{\text{IN}} \leq -3\text{V}$)のときは静止電流がわずか9mA (標準値)に制限されます。

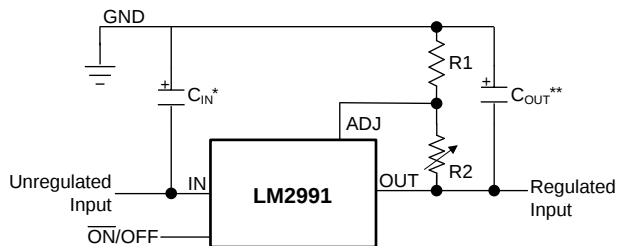
LM2991は短絡防止が保証され、サーマル・シャットダウンにヒステリシスが組み込まれていることで、意図せずに長期間の過負荷が発生した場合にもデバイスの信頼性が強化されています。LM2991は5リードのTO-220およびDDPAK/TO-263パッケージで供給され、車載用温度範囲の-40℃～+125℃での動作が規定されています。Mil-Aeroバージョンも利用できます。

製品情報⁽¹⁾

型番	パッケージ	本体サイズ(公称)
LM2991	DDPAK/TO-263 (5)	10.20mm×9.00mm
	TO-220 (5)	14.99mm×10.16mm

(1) 提供されているすべてのパッケージについては、巻末の注文情報を参照してください。

代表的なアプリケーション



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$$V_{\text{OUT}} = V_{\text{REF}} (1 + R2/R1)$$

* レギュレータが電源フィルタ・コンデンサから6インチ以上離れている場合に必要です。1μFのソリッド・タンタルまたは10μFのアルミ電解コンデンサをお勧めします。

** 安定性のため必要です。安定性を維持するため、最低10μFのアルミ電解、または1μFの固形タンタルが必要となります。過渡事象中にレギュレーションを維持するために、制限なく増やすことができます。コンデンサは、レギュレータのできるだけ近くに配置します。等価直列抵抗(ESR)は非常に重要で、レギュレータと同じ動作温度範囲にわたって10Ω未満を維持する必要があります。



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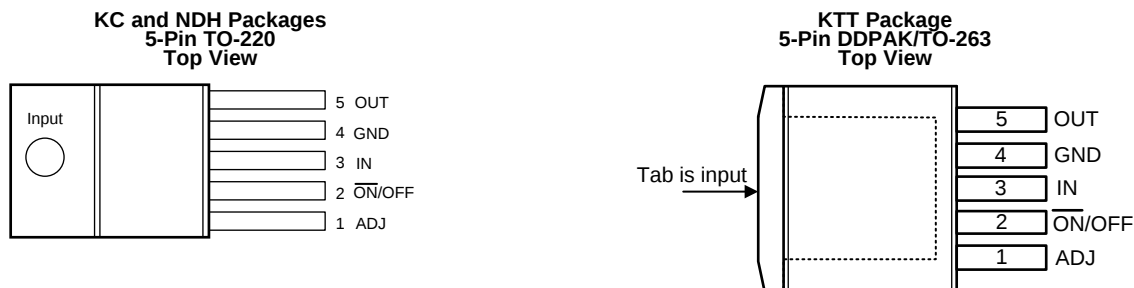
4 改訂履歴

資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

Revision H (June 2013) から Revision I に変更	Page
一部の曲線を「アプリケーション曲線」セクションへ移動し、「製品情報」および「ESD定格」の表、「機能概要」セクション、「デバイスの機能モード」セクション、「アプリケーションと実装」セクション、「電源に関する推奨事項」セクション、「レイアウト」セクション、「デバイスおよびドキュメントのサポート」セクション、「メカニカル、パッケージ、および注文情報」セクションを追加	1
Changed footnote 4 of Abs Max table and footnote 1 to Typical Characteristics to eliminate obsolete thermal values for thetaJA; updated values are in Thermal Information	4
Changed Figure 14 as previous thermal values have been updated	8

Revision G (April 2013) から Revision H に変更	Page
ナショナル・セミコンダクターのデータシート・レイアウトからTIフォーマットへ 変更	1

5 Pin Configuration and Functions



Pin Functions

PIN		I/O	DESCRIPTION
NO.	NAME		
1	ADJ	I	Feedback pin to the control loop for programming the output voltage.
2	$\overline{\text{ON/OFF}}$	I	Logic high enable input
3	IN	I	Negative Input voltage. Internally connected directly to the thermal tab.
4	GND	—	Ground
5	OUT	O	Regulated output voltage
—	TAB	I	Negative Input voltage. Internally connected directly to the device pin 3. The thermal tab must be connected to a copper area on the PCB at the same potential as device pin 3 (IN) to assure thermal performance, or leave the thermal tab floating. Do NOT connect the thermal tab to any potential other than the same potential at device pin 3. Do NOT connect the thermal tab to ground.

6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾⁽²⁾

	MIN	MAX	UNIT
Input voltage	–26	0.3	V
Power dissipation ⁽³⁾	Internally limited		
Storage temperature, T _{stg}	–65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) If Military/Aerospace specified devices are required, contact the Texas Instruments Sales Office/Distributors for availability and specifications.
- (3) The maximum power dissipation is a function of T_{J(MAX)}, R_{θJA}, and T_A. The maximum allowable power dissipation at any ambient temperature is P_D = (T_{J(MAX)} – T_A)/R_{θJA}. If this dissipation is exceeded, the die temperature will rise above 125°C, and the LM2991 will eventually go into thermal shutdown at a T_J of approximately 160°C. Refer to [Thermal Shutdown](#) for more details.

6.2 ESD Ratings

	VALUE	UNIT
V _(ESD) Electrostatic discharge Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±2000	V

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

	MIN	NOM	MAX	UNIT
Junction temperature, T _J	–40		125	°C
ON/OFF pin	0		5	V
Maximum input voltage (operational)	–26			V

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾	LM2991			UNIT
	TO-263 (KTT)	TO-220 (NDH) ⁽²⁾	TO-220 (KC) ⁽²⁾	
	5 PINS	5 PINS	5 PINS	
R _{θJA} ⁽³⁾ Junction-to-ambient thermal resistance, High-K	27.8	54.4	56.4	°C/W
R _{θJC(top)} Junction-to-case (top) thermal resistance	41.4	30.1	40.0	°C/W
R _{θJB} Junction-to-board thermal resistance	10.9	33.2	38.6	°C/W
ψ _{JT} Junction-to-top characterization parameter	6.0	11.6	12.8	°C/W
ψ _{JB} Junction-to-board characterization parameter	10.6	36.2	35.3	°C/W
R _{θJC(bot)} Junction-to-case (bottom) thermal resistance	0.7	0.5	0.6	°C/W

- (1) For more information about traditional and new thermal metrics, see [Semiconductor and IC Package Thermal Metrics](#).
- (2) The TO-220 package is vertically mounted in center of a JEDEC High-K test board (JESD 51-7) with no additional heat sink attached. This is a through-hole package; this is NOT a surface-mount package.
- (3) Thermal resistance value R_{θJA} is based on the EIA/JEDEC High-K printed circuit board defined by JESD51-7 - *High Effective Thermal Conductivity Test Board for Leadless Surface Mount Packages*.

6.5 Electrical Characteristics

 $V_{IN} = -10\text{ V}$, $V_{OUT} = -3\text{ V}$, $I_{OUT} = 1\text{ A}$, $C_{OUT} = 47\text{ }\mu\text{F}$, $R1 = 2.7\text{ k}\Omega$, $T_J = 25^\circ\text{C}$, unless otherwise specified.

PARAMETER	TEST CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT
Reference voltage	$5\text{ mA} \leq I_{OUT} \leq 1\text{ A}$	-1.234	-1.210	-1.186	V
	$5\text{ mA} \leq I_{OUT} \leq 1\text{ A}$, $V_{OUT} - 1\text{ V} \geq V_{IN} > -26\text{ V}$ $-40^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$	-1.27		-1.15	V
Output voltage (V_{OUT})			-2	-3	V
	$V_{IN} = -26\text{ V}$	-24	-25		V
Line regulation	$I_{OUT} = 5\text{ mA}$, $V_{OUT} - 1\text{ V} > V_{IN} > -26\text{ V}$		0.004	0.04	%/V
Load regulation	$50\text{ mA} \leq I_{OUT} \leq 1\text{ A}$		0.04%	0.4%	
Dropout voltage	$I_{OUT} = 0.1\text{ A}$, $\Delta V_{OUT} \leq 100\text{ mV}$		0.1	0.2	V
	$I_{OUT} = 0.1\text{ A}$, $\Delta V_{OUT} \leq 100\text{ mV}$ $-40^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$			0.3	
	$I_{OUT} = 1\text{ A}$, $\Delta V_{OUT} \leq 100\text{ mV}$		0.6	0.8	V
	$I_{OUT} = 1\text{ A}$, $\Delta V_{OUT} \leq 100\text{ mV}$ $-40^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$			1	
Quiescent current	$I_{OUT} \leq 1\text{ A}$		0.7		mA
	$I_{OUT} = 1\text{ A}$, $-40^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$			5	
Dropout quiescent current	$V_{IN} = V_{OUT}$, $I_{OUT} \leq 1\text{ A}$		16	50	mA
Ripple rejection	$V_{\text{ripple}} = 1\text{ V}_{\text{RMS}}$, $f_{\text{ripple}} = 1\text{ kHz}$, $I_{OUT} = 5\text{ mA}$	50	60		dB
Output noise	10 Hz to 100 kHz, $I_{OUT} = 5\text{ mA}$		200	450	μV
$\overline{\text{ON}}/\text{OFF}$ input voltage	V_{OUT} : ON		1.2		V
	V_{OUT} : ON $-40^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$			0.8	
	V_{OUT} : OFF		1.3		V
	V_{OUT} : OFF, $-40^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$	2.4			
$\overline{\text{ON}}/\text{OFF}$ input current	$V_{\overline{\text{ON}}/\text{OFF}} = 0.8\text{ V}$, V_{OUT} : ON		0.1	10	μA
	$V_{\overline{\text{ON}}/\text{OFF}} = 2.4\text{ V}$, V_{OUT} : OFF		40	100	
Output leakage current	$V_{IN} = -26\text{ V}$, $V_{\overline{\text{ON}}/\text{OFF}} = 2.4\text{ V}$, $V_{OUT} = 0\text{ V}$		60	150	μA
Current limit	$V_{OUT} = 0\text{ V}$	1.5	2		A

(1) Typicals are at $T_J = 25^\circ\text{C}$ and represent the most likely parametric norm.

6.6 Typical Characteristics

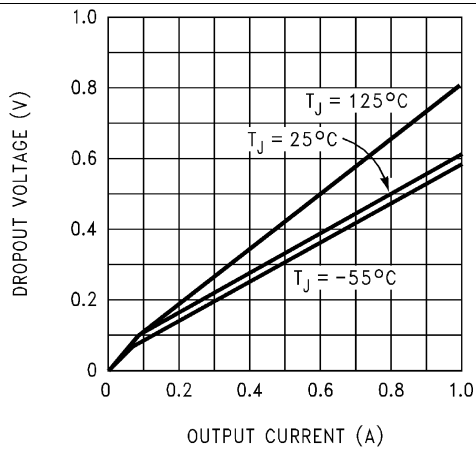


Figure 1. Dropout Voltage

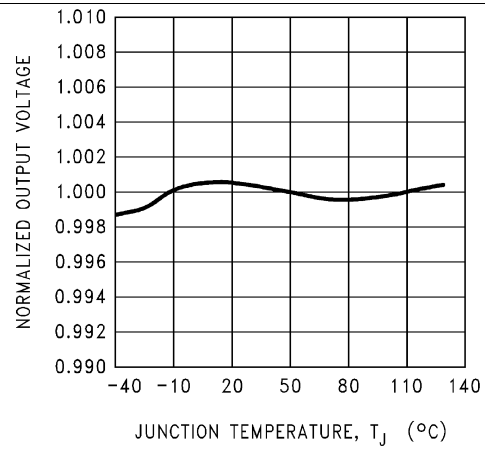


Figure 2. Normalized Output Voltage

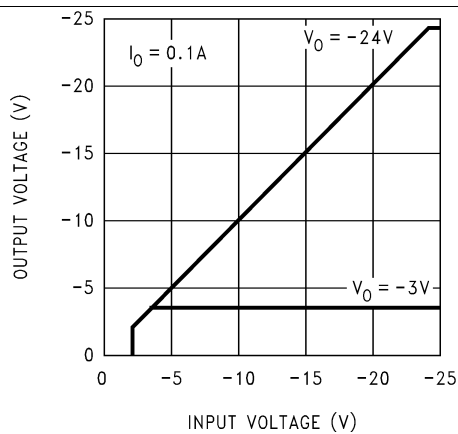


Figure 3. Output Voltage

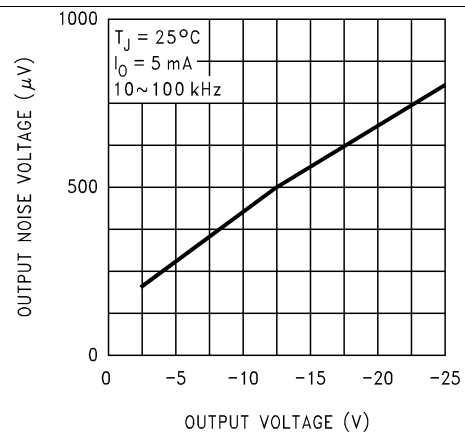


Figure 4. Output Noise Voltage

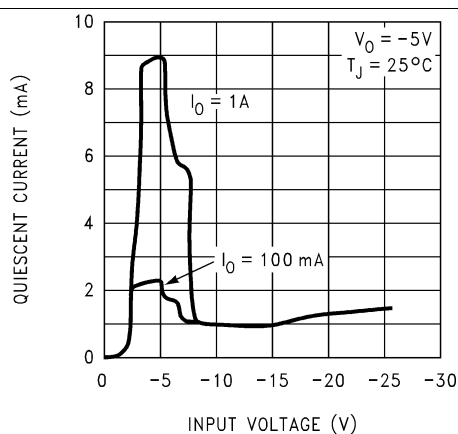


Figure 5. Quiescent Current

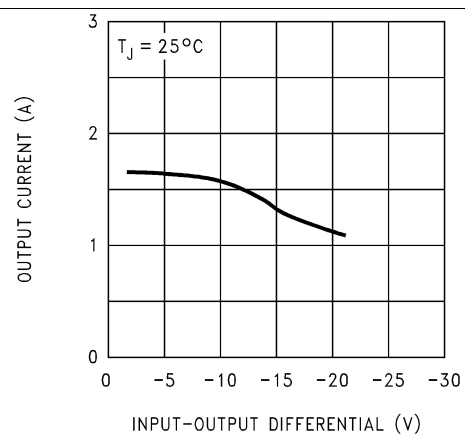


Figure 6. Maximum Output Current

Typical Characteristics (continued)

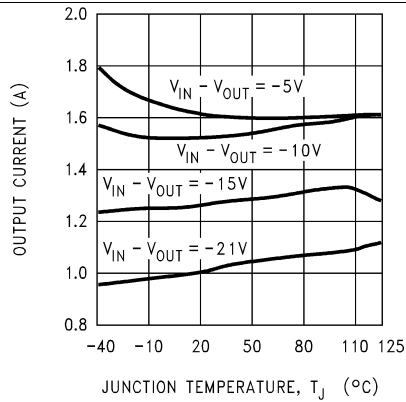


Figure 7. Maximum Output Current

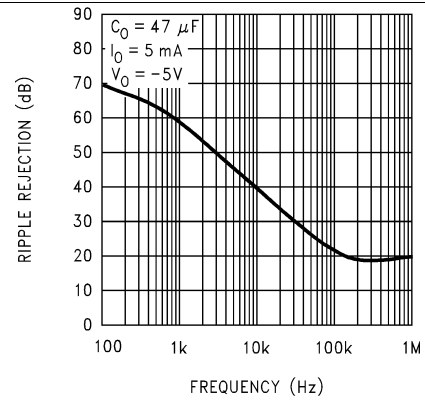


Figure 8. Ripple Rejection

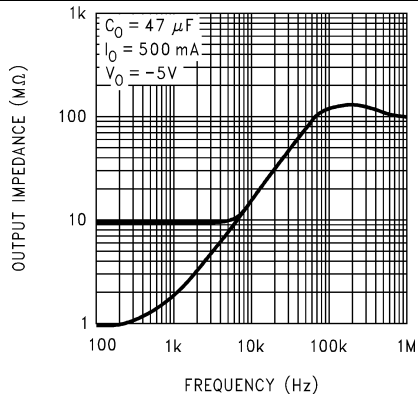


Figure 9. Output Impedance

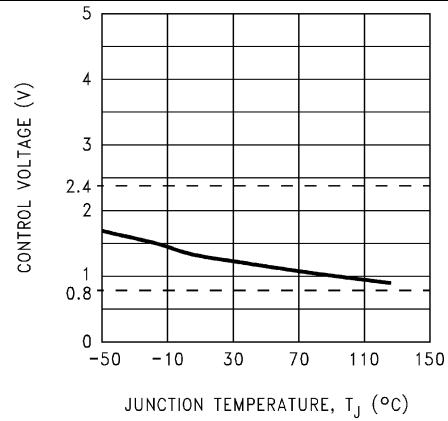


Figure 10. $\overline{\text{ON/OFF}}$ Control Voltage

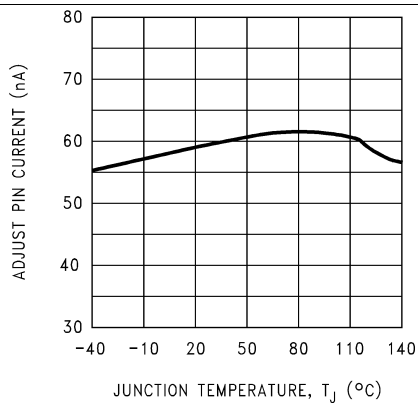


Figure 11. Adjust Pin Current

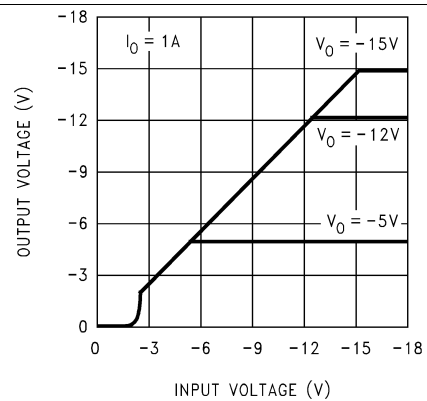


Figure 12. Low Voltage Behavior

Typical Characteristics (continued)

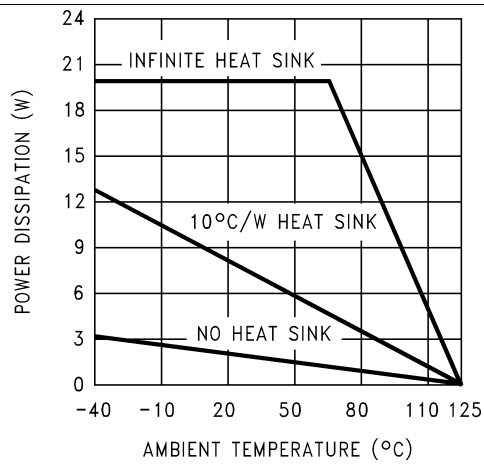


Figure 13. Maximum Power Dissipation (TO-220)

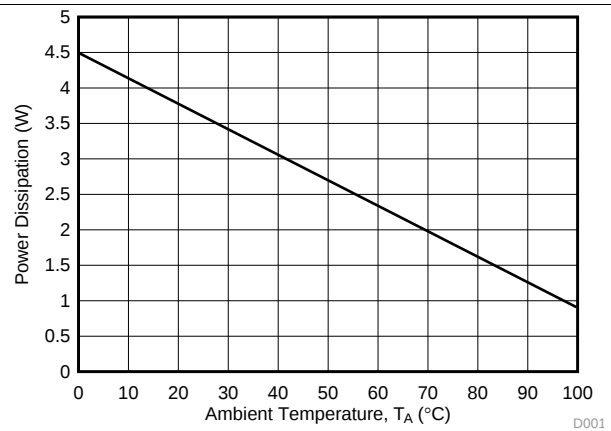


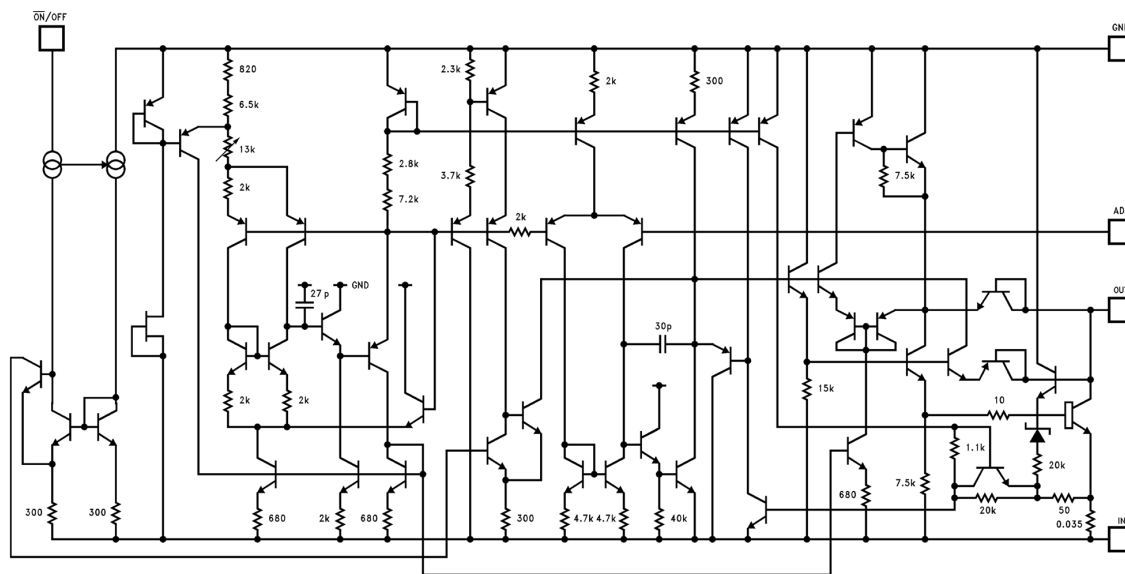
Figure 14. Maximum Power Dissipation (DDPAK/TO-263)

7 Detailed Description

7.1 Overview

The LM2991 is a five-pin, low-dropout, 1-A negative adjustable voltage regulator and negative power supply, ideally suited for a dual-supply system when using together with LM2941 series. The device may also be used as an adjustable current-sink load.

7.2 Functional Block Diagram



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7.3 Feature Description

7.3.1 $\overline{\text{ON/Off}}$ Pin

The LM2991 regulator can be turned off by applying a TTL or CMOS level high signal to the $\overline{\text{ON/Off}}$ pin. The impedance of the voltage source driving the $\overline{\text{ON/Off}}$ pin must be low enough to source the $\overline{\text{ON/Off}}$ pin input current to meet the OFF threshold voltage level, 100 μA maximum at 2.4 V.

If the $\overline{\text{ON/Off}}$ function is not needed, connect the pin to GND. The $\overline{\text{ON/Off}}$ pin should not be left floating, as this is not an ensured operating condition. See [Figure 15](#).

7.3.2 Forcing The Output Positive

Due to an internal clamp circuit, the LM2991 can withstand positive voltages on its output. If the voltage source pulling the output positive is DC, the current must be limited to 1.5 A. A current over 1.5 A fed back into the LM2991 could damage the device. The LM2991 output can also withstand fast positive voltage transients up to 26 V, without any current limiting of the source. However, if the transients have a duration of over 1 ms, the output should be clamped with a Schottky diode to ground.

7.3.3 Thermal Shutdown

The LM2991 has an internally set thermal shutdown point of typically 160°C, with approximately 10°C of hysteresis. This thermal shutdown temperature point is outside the specified [Recommended Operating Conditions](#) range, above the [Absolute Maximum Ratings](#), and is intended as a safety feature for momentary fault conditions only. Avoid continuous operation near the thermal shutdown temperature as it may have a negative affect on the life of the device.

7.4 Device Functional Modes

7.4.1 Operation with $V_{OUT(TARGET)} - 5\text{ V} \geq V_{IN} > -26\text{ V}$

The device operates if the input voltage is within $V_{OUT(TARGET)} - 5\text{ V}$ to -26 V range. At input voltages beyond the V_{IN} requirement, the devices do not operate correctly, and output voltage may not reach target value.

8 Application and Implementation

NOTE

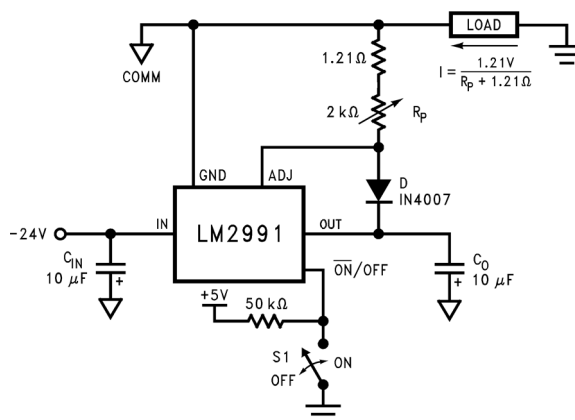
Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

The LM2991 is a 1-A negative adjustable voltage regulator with an operating V_{IN} range of -6 V to -26 V , and a regulated V_{OUT} having 5% accuracy with a maximum rated I_{OUT} current of 1 A. Efficiency is defined by the ratio of output voltage to input voltage because the LM2991 is a linear voltage regulator. To achieve high efficiency, the dropout voltage ($V_{IN} - V_{OUT}$) must be as small as possible, thus requiring a very low dropout LDO.

Successfully implementing an LDO in an application depends on the application requirements. If the requirements are simply input voltage and output voltage, compliance specifications (such as internal power dissipation or stability) must be verified to ensure a solid design. If timing, start-up, noise, PSRR, or any other transient specification is required, the design becomes more challenging.

8.2 Typical Application



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Figure 15. LM2991 Typical Application With Adjustable Current Sink

8.2.1 Design Requirements

For this design example, use the parameters listed in [Table 1](#) as the input parameters.

Table 1. Design Parameters

DESIGN PARAMETER	DESIGN REQUIREMENT
Input voltage	–26 V to –5 V
Output voltage	–2 V to –25 V (typical)
Output current	up to 1 A

8.2.2 Detailed Design Procedure

At 400-mA loading, the dropout of the LM2991 has 1-V maximum dropout over temperature, thus an –5 V headroom is sufficient for operation over both input and output voltage accuracy. The efficiency of the LM2991 in this configuration is $V_{OUT} / V_{IN} = 50\%$.

To achieve the smallest form factor, the TO-263 (KTT) package is selected. Select input and output capacitors in accordance with the [External Capacitors](#) section. Aluminum capacitances of 470 μF for the input and 50- μF capacitors for the output are selected. With an efficiency of 50% and a 400-mA maximum load, the internal power dissipation is 2000 mW, which corresponds to 82.5°C junction temperature rise for the TO-263 package. With an 25°C ambient temperature, the junction temperature is at 107.5°C.

8.2.2.1 External Capacitors

The LM2991 regulator requires an output capacitor to maintain stability. The capacitor must be at least 10- μF aluminum electrolytic or 1- μF solid tantalum. The equivalent series resistance (ESR) of the output capacitor must be less than 10 Ω , or the zero added to the regulator frequency response by the ESR could reduce the phase margin, creating oscillations. An input capacitor, of at least 1- μF solid tantalum or 10- μF aluminum electrolytic, is also needed if the regulator is situated more than 6 inches from the input power supply filter.

8.2.2.1.1 Input Capacitor

TI recommends a solid tantalum or ceramic capacitor whose value is at least 1 μF , but an aluminum electrolytic ($\geq 10 \mu\text{F}$) may be used. However, aluminum electrolytic types should not be used in applications where the ambient temperature can drop below 0°C because their internal impedance increases significantly at cold temperatures.

8.2.2.1.2 Output Capacitor

The output capacitor must meet the ESR limits shown in [Figure 16](#), which means it must have an ESR between about 25 m Ω and 10 Ω .

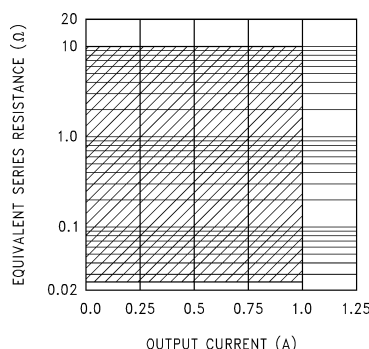


Figure 16. Output Capacitor ESR Range

A solid tantalum (value $\geq 1 \mu\text{F}$) is the best choice for the output capacitor. An aluminum electrolytic ($\geq 10 \mu\text{F}$) may be used if the ESR is in the stable range.

It should be noted that the ESR of a typical aluminum electrolytic will increase by as much as 50 $\times$ as the temperature is reduced from 25°C down to –40°C, while a tantalum exhibits an ESR increase of about 2 $\times$ over the same range. For this and other reasons, aluminum electrolytics should not be used in applications where low operating temperatures occur.

The lower stable ESR limit of 25 m Ω means that ceramic capacitors can not be used directly on the output of an LDO. A ceramic ($\geq 2.2 \mu\text{F}$) can be used on the output if some external resistance is placed in series with it (1 Ω recommended). Dielectric types X7R or X5R must be used if the temperature range of the application varies more than $\pm 25^\circ$ from ambient to assure the amount of capacitance is sufficient.

8.2.2.2 Ceramic Bypass Capacitors

Many designers place distributed ceramic capacitors whose value is in the range of 1000 pF to 0.1 μF at the power input pins of the IC's across a circuit board. These can cause reduced phase margin or oscillations in LDO regulators.

The advent of multi-layer boards with dedicated power and ground planes has removed the trace inductance that (previously) provided the necessary "de-coupling" to shield the output of the LDO from the effects of bypass capacitors.

Avoid these capacitors, if possible; if ceramic bypass capacitors are used, keep them as far away from the LDO output as is practical.

8.2.2.3 Minimum Load

A minimum load current of 500 μA is required for proper operation. The external resistor divider can provide the minimum load, with the resistor from the adjust pin to ground set to 2.4 kΩ.

8.2.2.4 Setting The Output Voltage

The output voltage of the LM2991 is set externally by a resistor divider using [Equation 1](#):

$$V_{OUT} = V_{REF} \times (1 + R_2/R_1) - (I_{ADJ} \times R_2)$$

where

- $V_{REF} = -1.21 \text{ V}$ (1)

The output voltage can be programmed within the range of –3 V to –24 V, typically an even greater range of –2 V to –25 V. The adjust pin current is about 60 nA, causing a slight error in the output voltage. However, using resistors lower than 100 kΩ makes the error due to the adjust pin current negligible. For example, neglecting the adjust pin current, and setting R2 to 100 kΩ and V_{OUT} to –5 V, results in an output voltage error of only 0.16%.

8.2.2.5 Power Dissipation

Knowing the device power dissipation and proper sizing of the thermal plane connected to the thermal tab is critical to ensuring reliable operation. Device power dissipation depends on input voltage, output voltage, and load conditions and can be calculated with [Equation 2](#).

$$P_{D(MAX)} = (V_{IN(MAX)} - V_{OUT}) \times I_{OUT} \quad (2)$$

Power dissipation can be minimized, and greater efficiency can be achieved, by using the lowest available voltage drop option that would still be greater than the dropout voltage (V_{DO}). However, keep in mind that higher voltage drops result in better dynamic (that is, PSRR and transient) performance.

Power dissipation and junction temperature are most often related by the junction-to-ambient thermal resistance (R_{θJA}) of the combined PCB and device package and the temperature of the ambient air (T_A), according to [Equation 3](#) or [Equation 4](#):

$$T_{J(MAX)} = T_{A(MAX)} + (R_{\theta JA} \times P_{D(MAX)}) \quad (3)$$

$$P_{D(MAX)} = (T_{J(MAX)} - T_{A(MAX)}) / R_{\theta JA} \quad (4)$$

Unfortunately, this R_{θJA} is highly dependent on the heat-spreading capability of the particular PCB design, and therefore varies according to the total copper area, copper weight, and location of the planes. The R_{θJA} recorded in [Thermal Information](#) is determined by the specific EIA/JEDEC JESD51-7 standard for PCB and copper-spreading area, and is to be used only as a relative measure of package thermal performance. For a well-designed thermal layout, R_{θJA} is actually the sum of the package junction-to-case (bottom) thermal resistance (R_{θJCbot}) plus the thermal resistance contribution by the PCB copper area acting as a heat sink.

8.2.2.6 Estimating Junction Temperature

The EIA/JEDEC standard recommends the use of ψ (Ψ) thermal characteristics to estimate the junction temperatures of surface mount devices on a typical PCB board application. These characteristics are not true thermal resistance values, but rather package specific thermal characteristics that offer practical and relative means of estimating junction temperatures. These ψ metrics are determined to be significantly independent of copper-spreading area. The key thermal characteristics (Ψ_{JT} and Ψ_{JB}) are given in [Thermal Information](#) and are used in accordance with [Equation 5](#) or [Equation 6](#).

$$T_{J(MAX)} = T_{TOP} + (\Psi_{JT} \times P_{D(MAX)})$$

where

- $P_{D(MAX)}$ is explained in [Equation 4](#)
- T_{TOP} is the temperature measured at the center-top of the device package. (5)

$$T_{J(MAX)} = T_{BOARD} + (\Psi_{JB} \times P_{D(MAX)})$$

where

- $P_{D(MAX)}$ is explained in [Equation 4](#)
- T_{BOARD} is the PCB surface temperature measured 1-mm from the device package and centered on the package edge. (6)

For more information about the thermal characteristics Ψ_{JT} and Ψ_{JB} , see [Semiconductor and IC Package Thermal Metrics](#); for more information about measuring T_{TOP} and T_{BOARD} , see [Using New Thermal Metrics](#); and for more information about the EIA/JEDEC JESD51 PCB used for validating $R_{\theta JA}$, see [Thermal Characteristics of Linear and Logic Packages Using JEDEC PCB Designs](#). These application notes are available at www.ti.com.

8.2.3 Application Curves

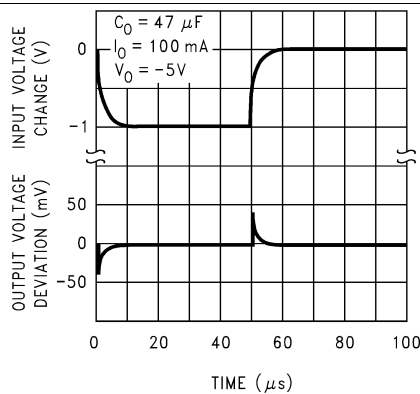


Figure 17. Line Transient Response

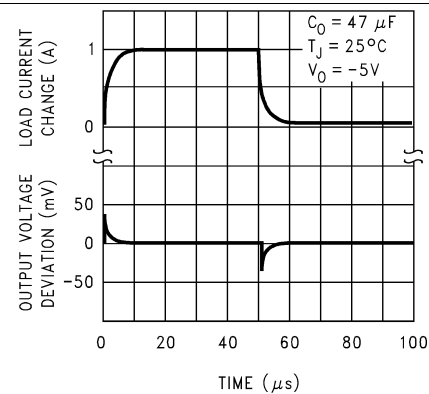


Figure 18. Load Transient Response

8.2.4 Additional Application Circuits

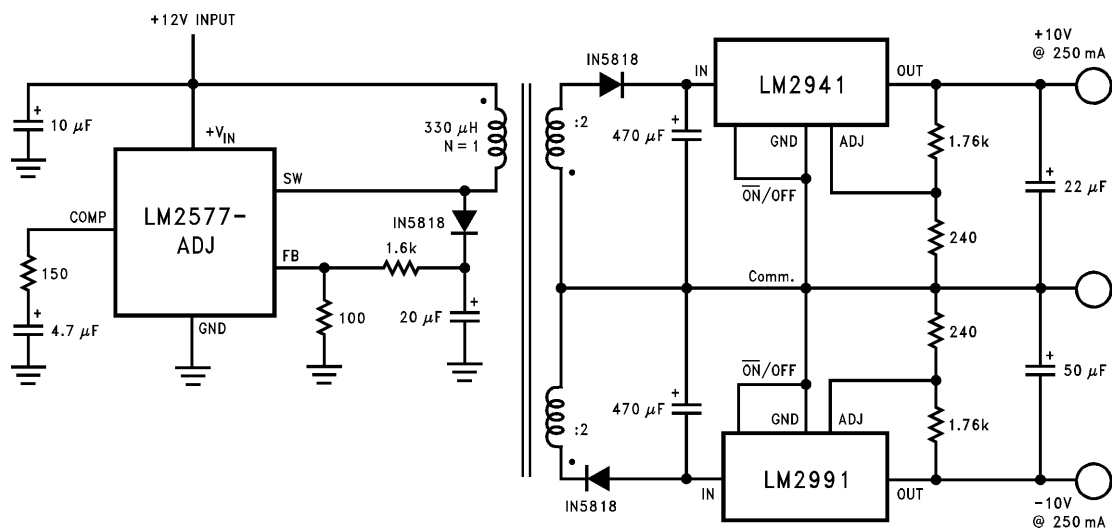


Figure 19. Fully Isolated Post-Switcher Regulator

9 Power Supply Recommendations

The LM2991 is designed to operate from an input voltage supply range between -6 V and -26 V . The input voltage range must provide adequate headroom in order for the device to have a regulated output. This input supply must be well regulated.

10 Layout

10.1 Layout Guidelines

The dynamic performance of the LM2991 is dependent on the layout of the PCB. PCB layout practices that are adequate for typical LDOs may degrade the PSRR, noise, or transient performance of the device. Best performance is achieved by placing C_{IN} and C_{OUT} on the same side of the PCB as the LM2991, and as close as is practical to the package. The ground connections for C_{IN} and C_{OUT} must be back to the LM2991 GND pin using as wide and short of a copper trace as is practical.

Good PC layout practices must be used or instability can be induced because of ground loops and voltage drops. The input and output capacitors must be directly connected to the IN, OUT, and GND pins of the LM2991 using traces which do not have other currents flowing in them (Kelvin connect). The best way to do this is to lay out C_{IN} and C_{OUT} near the device with short traces to the IN, OUT, and GND pins. The regulator ground pin must be connected to the external circuit ground so that the regulator and its capacitors have a single-point ground.

Stability problems have been seen in applications where vias to an internal ground plane were used at the ground points of the LM2991 device and the input and output capacitors. This was caused by varying ground potentials at these nodes resulting from current flowing through the ground plane. Using a single point ground technique for the regulator and its capacitors fixed the problem.

Because high current flows through the traces going into the IN pin and coming from the OUT pin, Kelvin connect the capacitor leads to these pins so there is no voltage drop in series with the input and output capacitors.

10.2 Layout Example

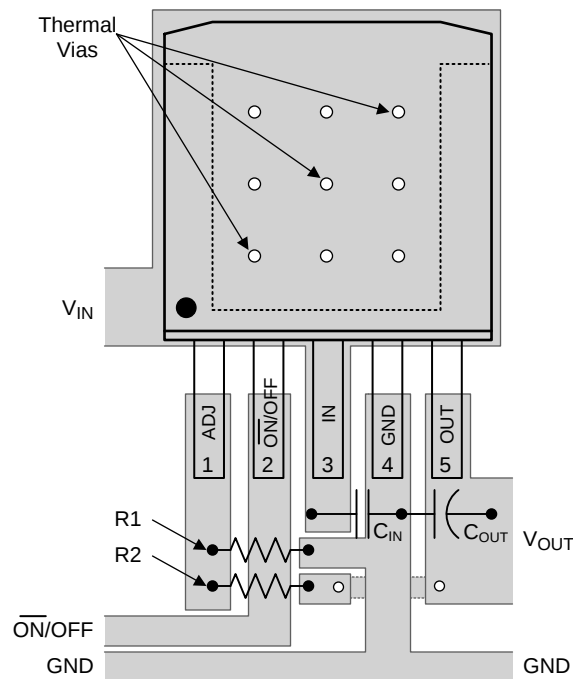


Figure 20. LM2991 TO-263 Board Layout

11 デバイスおよびドキュメントのサポート

11.1 デバイス・サポート

11.1.1 デバイスの関連用語

ドロップアウト電圧: 入出力電圧の差分がこの値に達すると、回路はそれ以上の入力電圧低減に対するレギュレートを中止します。ドロップアウト電圧は、出力電圧が($V_{OUT} + 5V$)入力で得られる公称値よりも100mV低下したときに測定され、負荷電流や接合部温度に依存します。

入力電圧: 入力端子に印加されるDC電圧で、グランドとの相対電圧。

入出力の差分: レギュレートされていない入力電圧と、レギュレータの動作対象であるレギュレートされた出力電圧との電位差。

ライン・レギュレーション: 入力電圧の変化に対する出力電圧の変化。この測定は、消費電力の低い状況、またはパルス技法を使用して、チップの平均温度が大きな影響を受けないように行われます。

負荷レギュレーション: 一定のチップ温度での、負荷電流の変化に対する出力電圧の変化。

長期的安定性: 加速寿命テスト状況において、最大定格の電圧と接合部温度で1000時間動作後の出力電圧の安定性。

出力ノイズ電圧: 出力でのrms AC電圧で、一定の負荷と入力リップルなしの状況において、指定の周波数範囲にわたって測定されます。

静止電流: 正の入力電流のうち、正の負荷電流に寄与しない部分。レギュレータのグランド・リード電流。

リップル除去: ピーク・ツー・ピークの入力リップル電圧と、ピーク・ツー・ピークの出力リップル電圧との比率。

V_{OUT} の温度安定性 室温から、上下いずれかの限界温度まで遷移したときの温度変化に対する出力電圧の変化割合。

11.2 関連資料

詳細情報については、以下を参照してください。

- 『[半導体およびICパッケージの熱指標](#)』
- 『[新しい温度指標の使用](#)』
- 『[JEDEC PCB設計を使用するリニアおよびロジック・パッケージの熱特性](#)』

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11.7 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

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PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
LM2991S/NOPB	Active	Production	DDPAK/ TO-263 (KTT) 5	45 TUBE	ROHS Exempt	SN	Level-3-245C-168 HR	-40 to 125	LM2991S P+
LM2991S/NOPB.B	Active	Production	DDPAK/ TO-263 (KTT) 5	45 TUBE	ROHS Exempt	SN	Level-3-245C-168 HR	-40 to 125	LM2991S P+
LM2991SX/NOPB	Active	Production	DDPAK/ TO-263 (KTT) 5	500 LARGE T&R	ROHS Exempt	SN	Level-3-245C-168 HR	-40 to 125	LM2991S P+
LM2991SX/NOPB.B	Active	Production	DDPAK/ TO-263 (KTT) 5	500 LARGE T&R	ROHS Exempt	SN	Level-3-245C-168 HR	-40 to 125	LM2991S P+
LM2991T/LB03	Obsolete	Production	TO-220 (NDH) 5	-	-	Call TI	Call TI	-	LM2991T P+
LM2991T/LF03	Active	Production	TO-220 (NDH) 5	45 TUBE	ROHS Exempt	SN	Level-1-NA-UNLIM	-	LM2991T P+
LM2991T/LF03.B	Active	Production	TO-220 (NDH) 5	45 TUBE	ROHS Exempt	SN	Level-1-NA-UNLIM	-40 to 125	LM2991T P+
LM2991T/NOPB	Active	Production	TO-220 (KC) 5	45 TUBE	ROHS Exempt	SN	Level-1-NA-UNLIM	-40 to 125	LM2991T P+
LM2991T/NOPB.B	Active	Production	TO-220 (KC) 5	45 TUBE	ROHS Exempt	SN	Level-1-NA-UNLIM	-40 to 125	LM2991T P+

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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TAPE AND REEL INFORMATION



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LM2991SX/NOPB	DDPAK/TO-263	KTT	5	500	330.0	24.4	10.75	14.85	5.0	16.0	24.0	Q2

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LM2991SX/NOPB	DDPAK/TO-263	KTT	5	500	356.0	356.0	45.0

TUBE



*All dimensions are nominal

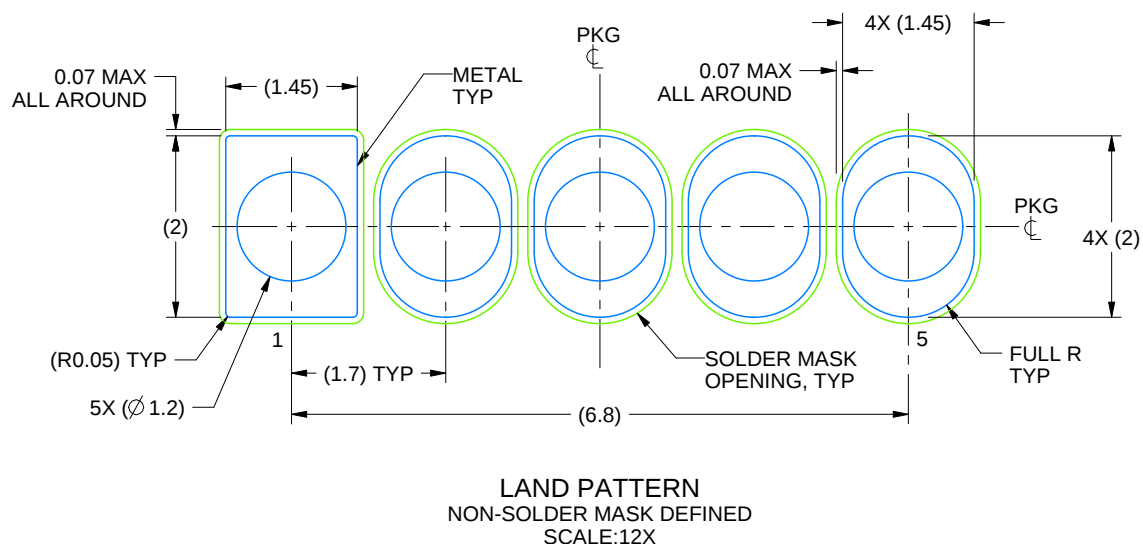
Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
LM2991S/NOPB	KTT	TO-263	5	45	502	25	8204.2	9.19
LM2991S/NOPB.B	KTT	TO-263	5	45	502	25	8204.2	9.19
LM2991T/LF03	NDH	TO-220	5	45	502	30	30048.2	10.74
LM2991T/LF03.B	NDH	TO-220	5	45	502	30	30048.2	10.74
LM2991T/NOPB	KC	TO-220	5	45	502	33	6985	4.06
LM2991T/NOPB.B	KC	TO-220	5	45	502	33	6985	4.06

EXAMPLE BOARD LAYOUT

KC0005A

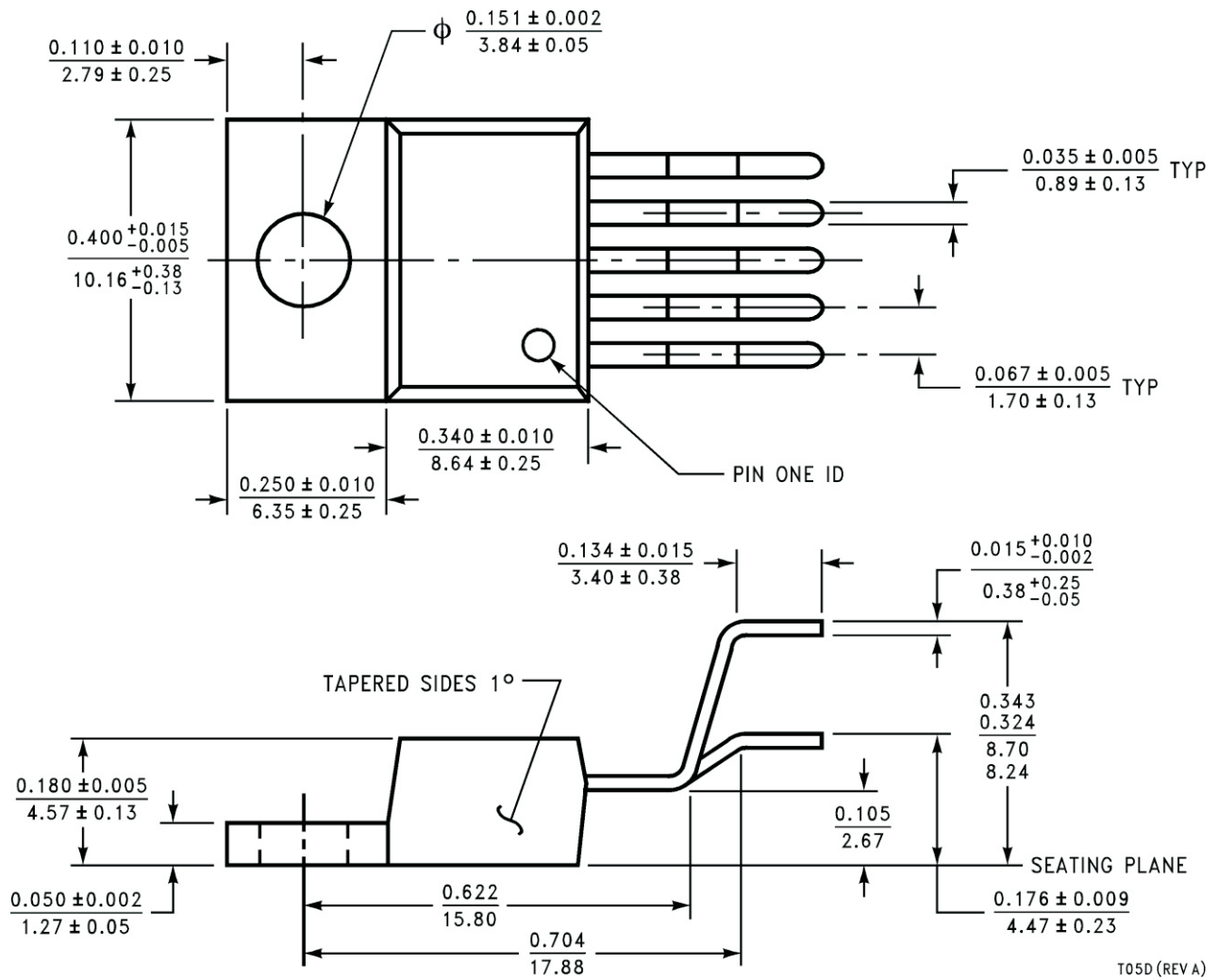
TO-220 - 16.51 mm max height

TO-220

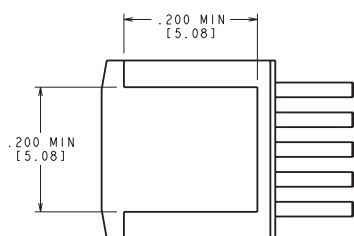
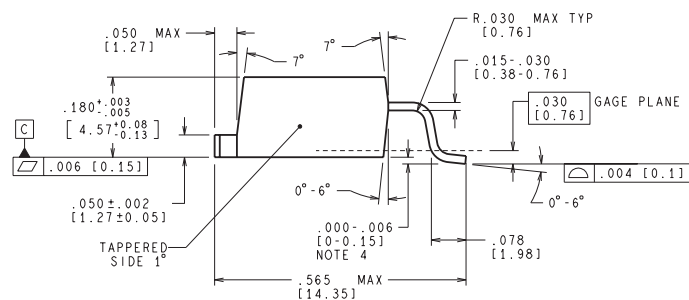
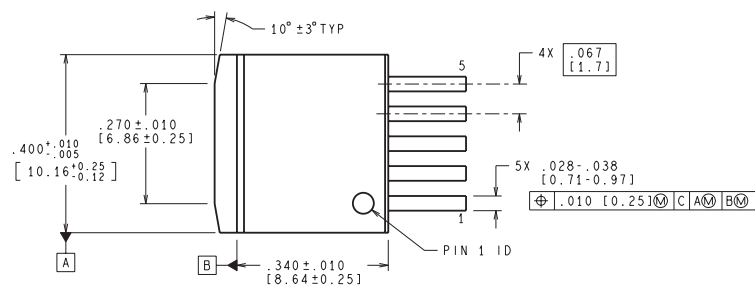


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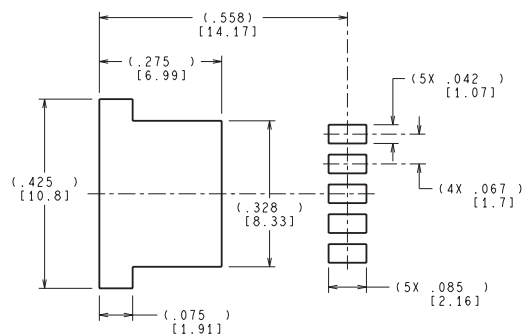
NDH0005D



KTT0005B



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