TEXAS INSTRUMENTS
TLV701
JAJSPM6A – NOVEMBER 2011 – REVISED APRIL 2023

TLV701 24V、150mA、3.2µA 静止電流の低ドロップアウト・リニア・レギュレータ

1 特長

- 入力電圧範囲:
  - 2.5V～24V (新しいチップのみ最大 30V)
- 選択可能な出力電圧:
  - 固定:3V および 3.3V
- 出力電流:最大 150mA
- 超低 I_Q:100mA の負荷電流で 3.4µA
- 0.47µF 以上の出力コンデンサで安定動作
- 過電流保護
- パッケージ:5 ピン SOT-23 (DBV)
- 動作時接合部温度:-40℃～+125℃

2 アプリケーション

- ホーム / ビルディング・オートメーション
- リテール・オートメーションおよびペイメント
- グリッド・インフラ
- 医療用アプリケーション
- 照明用途

3 説明

TLV701 低ドロップアウト (LDO) リニア電圧レギュレータは低静止時電流のデバイスで、幅広い入力電圧範囲と低消費電力動作の利点を小型パッケージで提供します。このため、TLV701 はバッテリ駆動アプリケーションや、低消費電力マイコンの電力管理外付け機能用に設計されています。

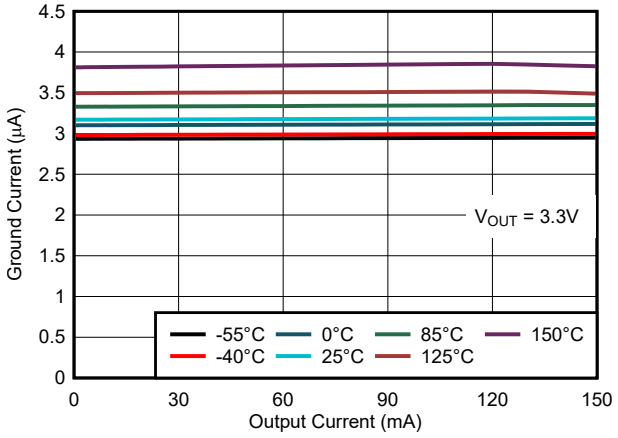
TLV701 LDO は、100mA の負荷電流で標準 850mV の低ドロップアウトをサポートしています。静止時電流が低く (標準値 3.4µA)、出力負荷電流の全範囲 (0mA～150mA) にわたって安定しています。また、TLV701 には内部ソフトスタートが搭載されており、突入電流を低減できます。過電流制限保護機能が組み込まれているため、負荷の短絡やフォルトが発生してもレギュレータが保護されます。

TLV701 は 2.90mm x 1.60mm の SOT23-5 パッケージで供給され、コスト効率の優れた基板製造に役立ちます。

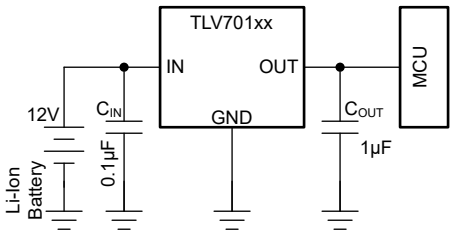
パッケージ情報

部品番号	パッケージ(1)	本体サイズ (公称)
TLV701	DBV (SOT-23、5)	2.90mm x 1.60mm

(1) 利用可能なパッケージについては、このデータシートの末尾にある注文情報を参照してください。



TLV701 の静止時電流と負荷電流との関係 (新しいチップのみ)



代表的なアプリケーション

英語版の TI 製品についての情報を翻訳したこの資料は、製品の概要を確認する目的で便宜的に提供しているものです。該当する正式な英語版の最新情報は、www.ti.com で閲覧でき、その内容が常に優先されます。TI では翻訳の正確性および妥当性につきましては一切保証いたしません。実際の設計などの前には、必ず最新版の英語版をご参照くださいますようお願いいたします。

Table of Contents

1 特長 ... 1
2 アプリケーション ... 1
3 説明 ... 1
4 Revision History ... 2
5 Pin Configuration and Functions ... 3
6 Specifications ... 4
6.1 Absolute Maximum Ratings ... 4
6.2 ESD Ratings ... 4
6.3 Recommended Operating Conditions ... 4
6.4 Thermal Information ... 5
6.5 Electrical Characteristics ... 5
6.6 Typical Characteristics ... 6
7 Detailed Description ... 11
7.1 Overview ... 11
7.2 Functional Block Diagram ... 11
7.3 Feature Description ... 11
7.4 Device Functional Modes ... 13
8 Application and Implementation ... 14
8.1 Application Information ... 14
8.2 Typical Application ... 14
8.3 Best Design Practices ... 18
8.4 Power Supply Recommendations ... 18
8.5 Layout ... 18
9 Device and Documentation Support ... 20
9.1 Device Support ... 20
9.2 Documentation Support ... 20
9.3 ドキュメントの更新通知を受け取る方法 ... 20
9.4 サポート・リソース ... 20
9.5 Trademarks ... 20
9.6 静電気放電に関する注意事項 ... 20
9.7 用語集 ... 20
10 Mechanical, Packaging, and Orderable Information ... 21

4 Revision History

資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

Changes from Revision * (November 2011) to Revision A (April 2023)	Page
• ドキュメントのステータスを製品プレビューから量産データへ変更	1

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5 Pin Configuration and Functions

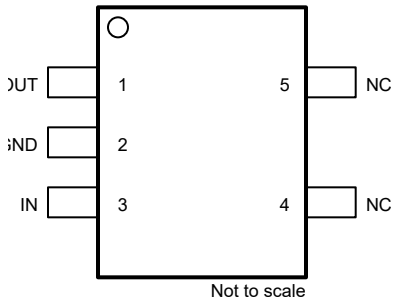


図 5-1. DBV Package, 5-Pin SOT-23 (Top View)

表 5-1. Pin Functions

PIN		TYPE	DESCRIPTION
NAME	DBV		
OUT	1	O	Output of the regulator. A capacitor with a value of 1 µF or larger is required from this pin to ground.(1) See the *Input and Output Capacitor Requirements* section for more information.
GND	2	—	Ground pin.
IN	3	I	Input supply pin. A capacitor with a value of 0.1 µF or larger is recommended from this pin to ground. See the *Input and Output Capacitor Requirements* section for more information.
NC	4, 5	—	Not internally connected. This pin can be left open or tied to ground for improved thermal performance.

(1) The nominal output capacitance must be greater than 0.47 µF. Throughout this document, the nominal derating on these capacitors is 50%. Make sure that the effective capacitance at the pin is greater than 0.47 µF.

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6 Specifications

6.1 Absolute Maximum Ratings

over operating temperature range (unless otherwise noted)(1) (2)

		MIN	MAX	UNIT
Voltage	V_{IN} (for legacy chip only)	–0.3	24	V
	V_{IN} (for new chip only)	–0.3	30	
Voltage	V_{OUT} (for legacy chip only)	–0.3	5.0	V
Voltage	V_{OUT} (for fixed output new chip only)	–0.3	2 × $V_{OUT(typ)}$ or V_{IN} + 0.3 or 5.5 (whichever is lower)	V
Current	Peak output current	Internally limited		
Temperature	Junction, T_J	–40	150	°C
	Storage, T_{stg}	–65	150	

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) All voltage values are with respect to network ground terminal.

6.2 ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins(1)	±2000	V
		Charged device model (CDM), per JEDEC specification JESD22-C101, all pins(2)	±500	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)(1)

		MIN	NOM	MAX	UNIT
V_{IN}	Input supply voltage	2.5		24	V
V_{OUT}	Output voltage	1.205		5	V
I_{OUT}	Output current	0		150	mA
C_{IN}	Input capacitor(2)	0	0.047		μF
C_{OUT}	Output capacitor (for legacy chip only)	0.47	1		
	Output capacitor (for new chip only) (3)	1			
T_J	Operating junction temperature	-40		125	°C

(1) All voltages are with respect to GND.
(2) An input capacitor is not required for LDO stability. However, an input capacitor with an effective value of 0.047 μF is recommended to counteract the effect of source resistance and inductance, which may in some cases cause symptoms of system level instability such as ringing or oscillation, especially in the presence of load transients.
(3) All capacitor values are assumed to derate to 50% of the nominal capacitor value. Maintain an effective output capacitance of 0.47 μF minimum for the stability.

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6.4 Thermal Information

THERMAL METRIC[1]		Legacy Chip DBV (SOT-23) 5 PINS	New Chip DBV (SOT-23) 5 PINS	UNIT
$R_{\theta JA}$	Junction-to-ambient thermal resistance	213.1	170.8	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	110.9	68.7	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	97.4	76.7	°C/W
ψ_{JT}	Junction-to-top characterization parameter	22.0	10.3	°C/W
ψ_{JB}	Junction-to-board characterization parameter	78.4	76.3	°C/W

(1) For more information about traditional and new thermal metrics, see the Semiconductor and IC package thermal metrics application report.

6.5 Electrical Characteristics

over operating junction temperature range (T_J = –40°C to 125°C), V_{IN} = $V_{OUT(nom)}$ + 1 V, I_{OUT} = 1 mA, and C_{OUT} = 1 µF (unless otherwise noted); typical values are at T_J = 25°C

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{IN}	Input voltage range ((1))	T_J = 25°C			24	V
V_{OUT}	Output voltage range (1)	T_J = 25°C	1.2		5	V
V_{OUT}	DC output accuracy((1))	T_J = 25°C	–2		2	%
I_{GND}	Ground pin current (legacy chip) ((3))	I_{OUT} = 0 mA, T_J = 25°C		3.2	4.5	µA
		I_{OUT} = 100 mA, T_J = 25°C		3.2	5.5	
	Ground pin current (new chip) ((3))	I_{OUT} = 0 mA, T_J = 25°C		3.2	4.1	
		I_{OUT} = 100 mA, T_J = 25°C		3.4	4.5	
$\Delta V_{OUT\ (\Delta IOUT)}$	Load regulation	1 mA < I_{OUT} < 10 mA		6		mV
		1 mA < I_{OUT} < 50 mA		19		
		1 mA < I_{OUT} < 100 mA		29	50	
$\Delta V_{OUT\ (\Delta VIN)}$	Line regulation (1)	$V_{OUT(NOM)}$ + 1 V ≤ V_{IN} ≤ 24 V , T_J = 25°C		20	50	mV
I_{CL}	Output current limit (legacy chip)	V_{OUT} = 0 V , T_J = 25°C	160		1000	mA
	Output current limit (new chip)		160		500	
PSRR	Power-supply ripple rejection	f = 100 kHz, C_{OUT} = 10 µF		60		dB
V_{DO}	Dropout voltage	V_{IN} = $V_{OUT(nom)}$ – 0.1 V, I_{OUT} = 10 mA		75		mV
		V_{IN} = $V_{OUT(nom)}$ – 0.1 V, I_{OUT} = 50 mA		400		

(1) Minimum V_{IN} = V_{OUT} + V_{DO} or the value shown for *Input voltage* in this table, whichever is greater.
(2) This device employs a leakage null control circuit. This circuit is active only if output current is less than pass transistor leakage current. The circuit is typically active when output load is less than 5 µA, V_{IN} is greater than 18 V, and die temperature is greater than 100°C.

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6.6 Typical Characteristics

at operating temperature T_J = 25°C, V_{IN} = $V_{OUT(NOM)}$ + 1.0 V or 2.5 V (whichever is greater), I_{OUT} = 1 mA, C_{IN} = 1 µF, and C_{OUT} = 1 µF (unless otherwise noted)

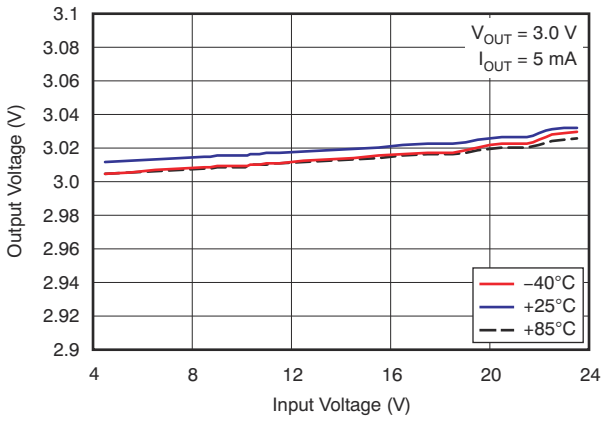


図 6-1. Line Regulation (V_{OUT} = 3.0 V) for Legacy Chip

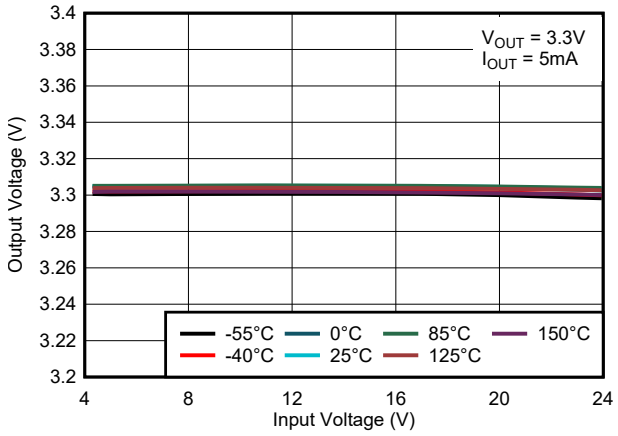


図 6-2. Line Regulation (V_{OUT} = 3.3 V) for New Chip

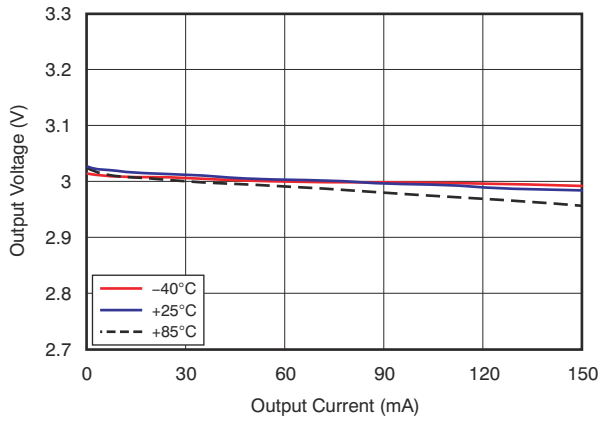


図 6-3. Load Regulation (V_{OUT} = 3.0 V) for Legacy Chip

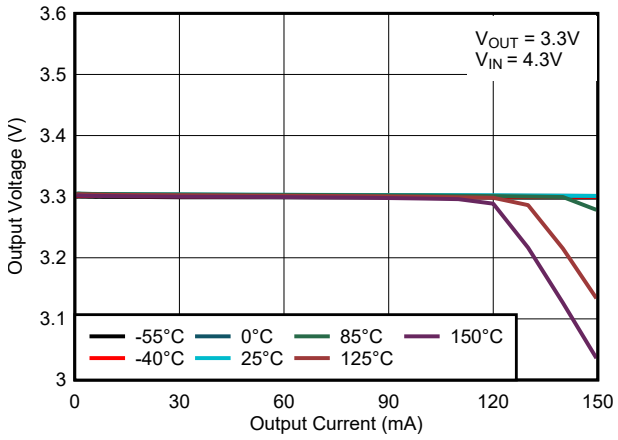


図 6-4. Load Regulation (V_{OUT} = 3.3 V) for New Chip

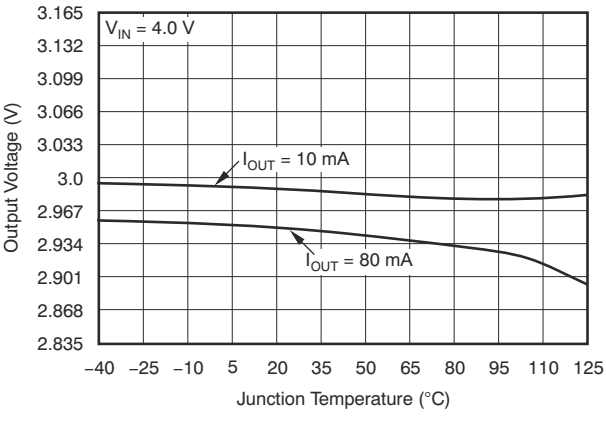


図 6-5. Output Voltage vs Junction Temperature for Legacy Chip

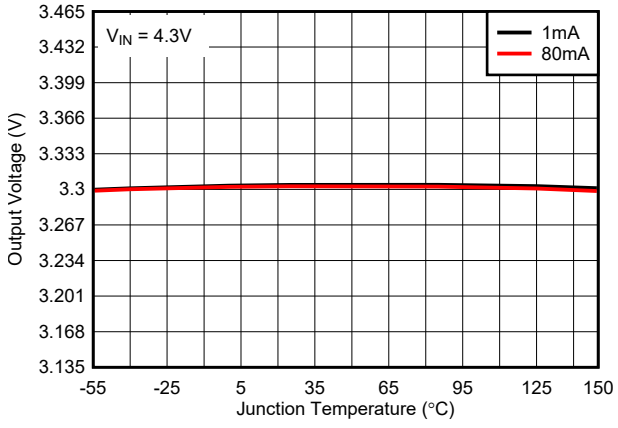


図 6-6. Output Voltage vs Junction Temperature for New Chip

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6.6 Typical Characteristics (continued)

at operating temperature T_J = 25°C, V_{IN} = $V_{OUT(NOM)}$ + 1.0 V or 2.5 V (whichever is greater), I_{OUT} = 1 mA, C_{IN} = 1 µF, and C_{OUT} = 1 µF (unless otherwise noted)

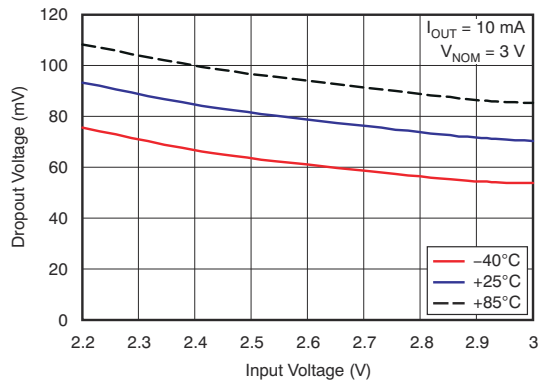


図 6-7. Dropout Voltage vs Input Voltage (V_{OUT} = 3.0 V) for Legacy Chip

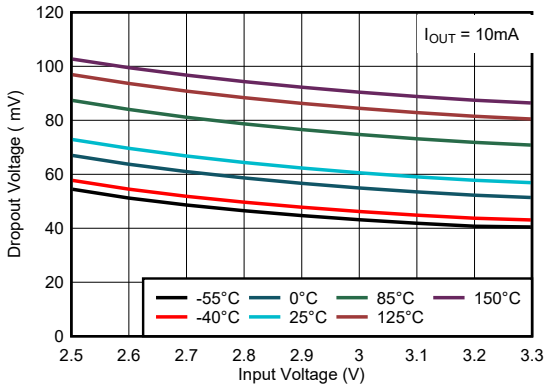


図 6-8. Dropout Voltage vs Input Voltage (V_{OUT} = 3.3 V) for New Chip

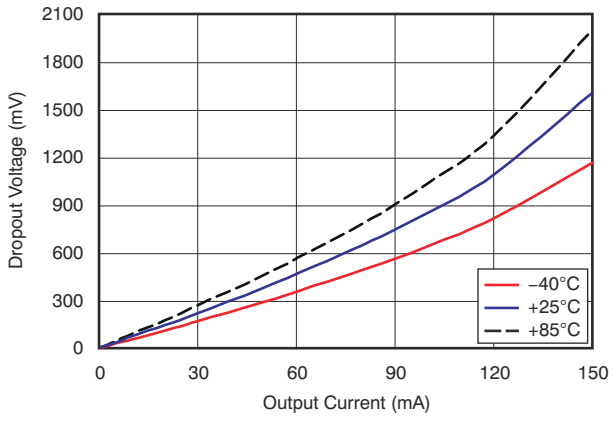


図 6-9. Dropout Voltage vs Output Current for Legacy Chip

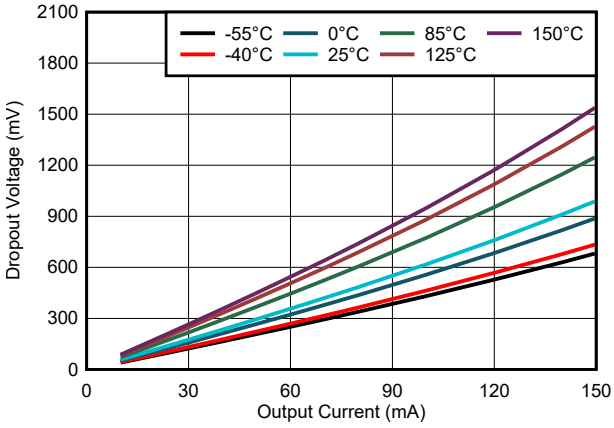


図 6-10. Dropout Voltage vs Output Current for New Chip

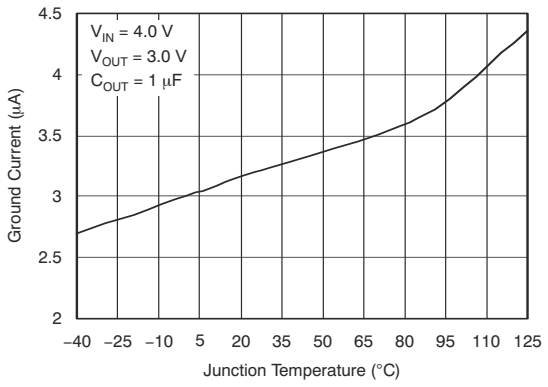


図 6-11. Ground Current vs Junction Temperature for Legacy Chip

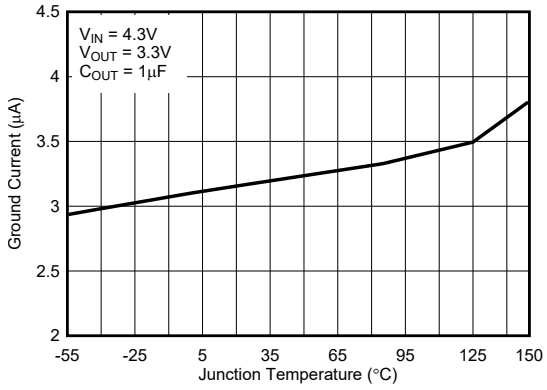


図 6-12. Ground Current vs Junction Temperature for New Chip

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6.6 Typical Characteristics (continued)

at operating temperature T_J = 25°C, V_{IN} = $V_{OUT(NOM)}$ + 1.0 V or 2.5 V (whichever is greater), I_{OUT} = 1 mA, C_{IN} = 1 µF, and C_{OUT} = 1 µF (unless otherwise noted)

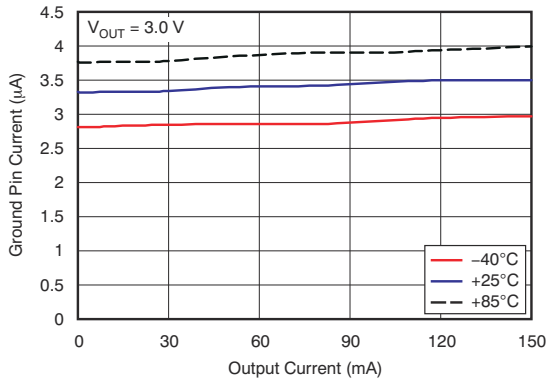


図 6-13. Ground Pin Current vs Load Current for Legacy Chip

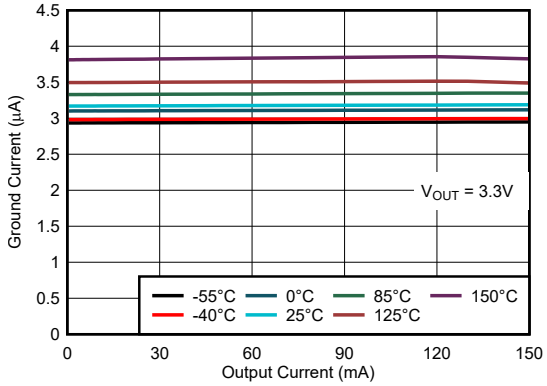


図 6-14. Ground Pin Current vs Load Current for New Chip

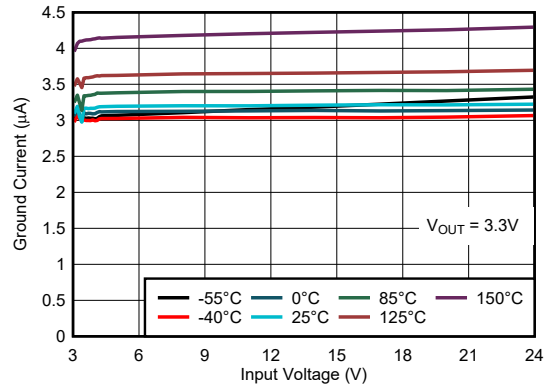


図 6-15. Ground Pin Current vs Input Voltage for New Chip

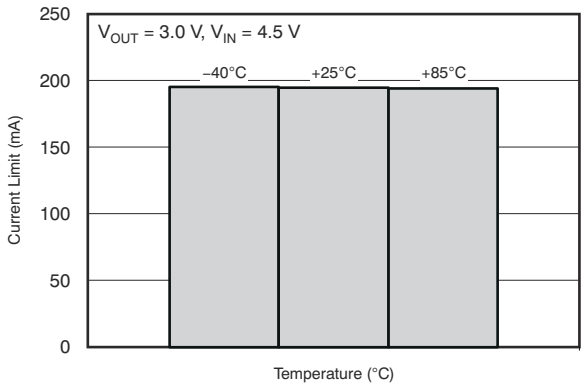


図 6-16. Current Limit vs Junction Temperature for Legacy Chip

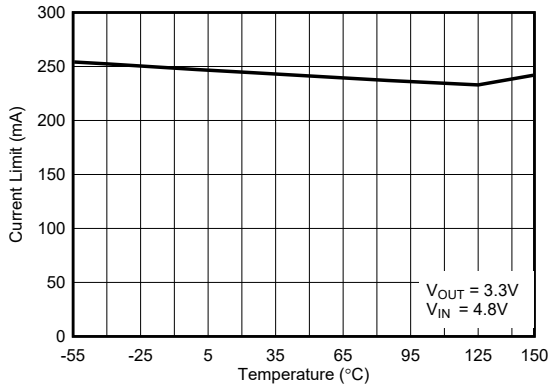


図 6-17. Current Limit vs Junction Temperature for New Chip

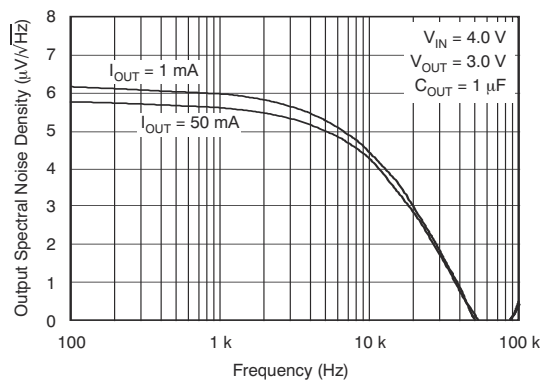


図 6-18. Output Spectral Noise Density vs Frequency for Legacy Chip

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6.6 Typical Characteristics (continued)

at operating temperature T_J = 25°C, V_{IN} = $V_{OUT(NOM)}$ + 1.0 V or 2.5 V (whichever is greater), I_{OUT} = 1 mA, C_{IN} = 1 µF, and C_{OUT} = 1 µF (unless otherwise noted)

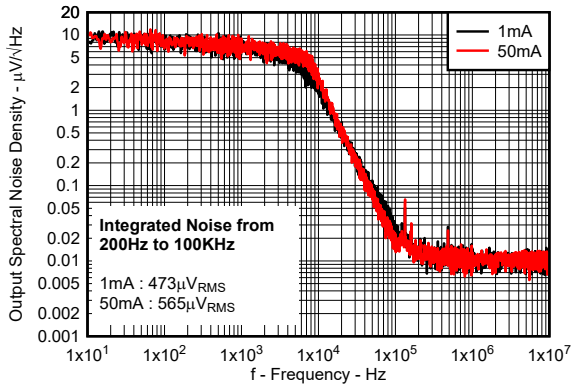


図 6-19. Output Spectral Noise Density vs Frequency for New Chip

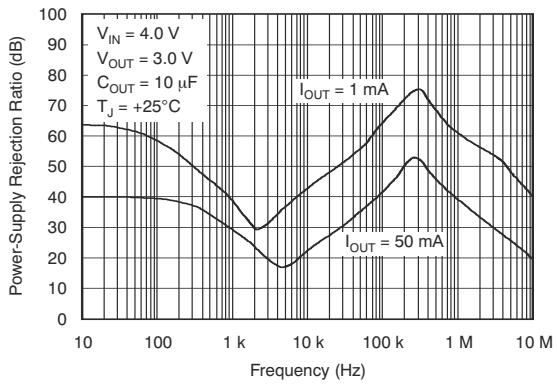


図 6-20. Power-Supply Ripple Rejection vs Frequency for Legacy Chip

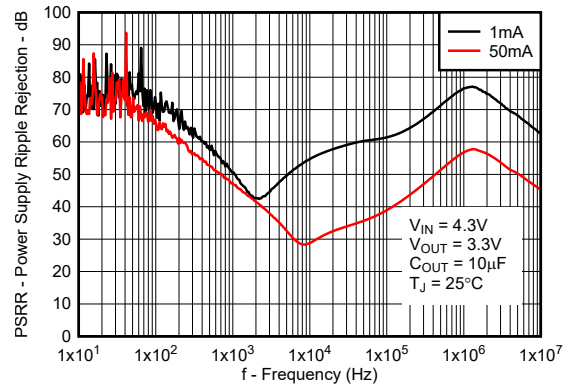


図 6-21. Power-Supply Ripple Rejection vs Frequency for New Chip

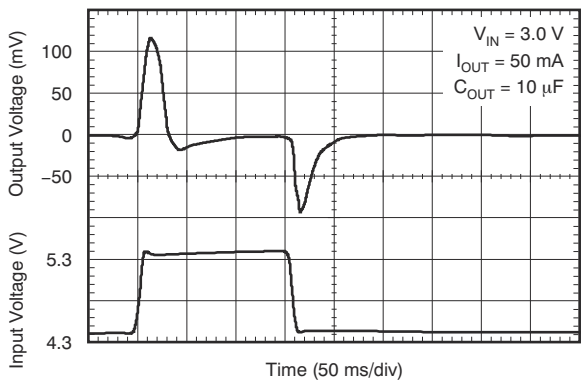


図 6-22. Line Transient Response for Legacy Chip

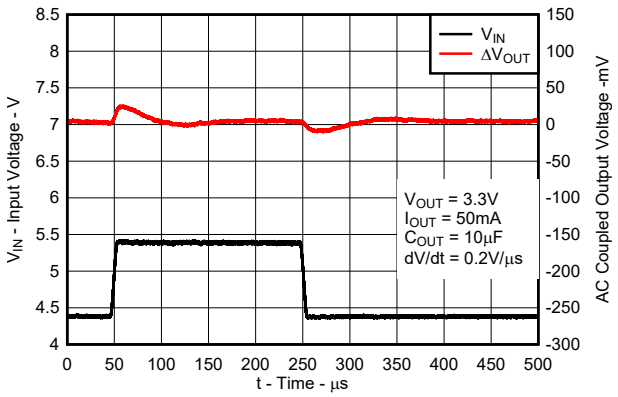


図 6-23. Line Transient Response for New Chip

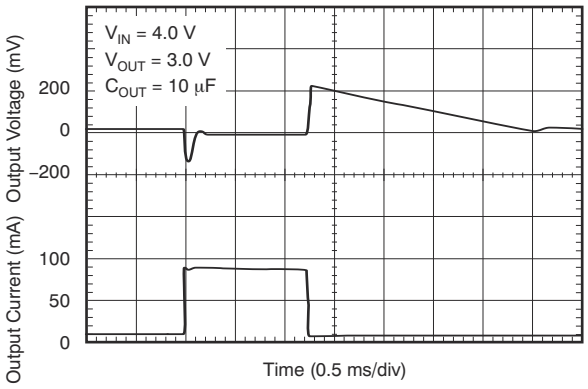


図 6-24. Load Transient Response for Legacy Chip

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6.6 Typical Characteristics (continued)

at operating temperature T_J = 25°C, V_{IN} = $V_{OUT(NOM)}$ + 1.0 V or 2.5 V (whichever is greater), I_{OUT} = 1 mA, C_{IN} = 1 µF, and C_{OUT} = 1 µF (unless otherwise noted)

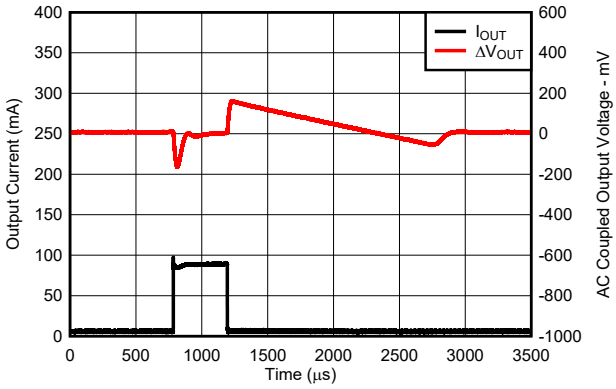


図 6-25. Load Transient Response for New Chip

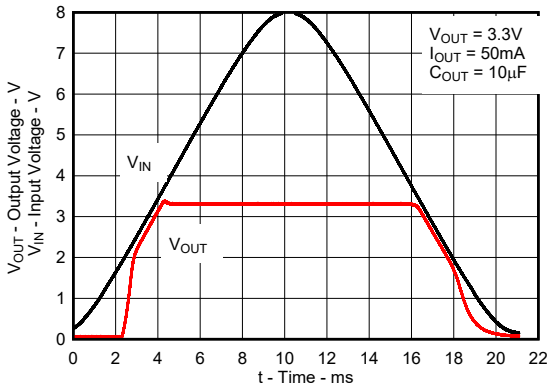


図 6-26. Power-Up, Power-Down for New Chip

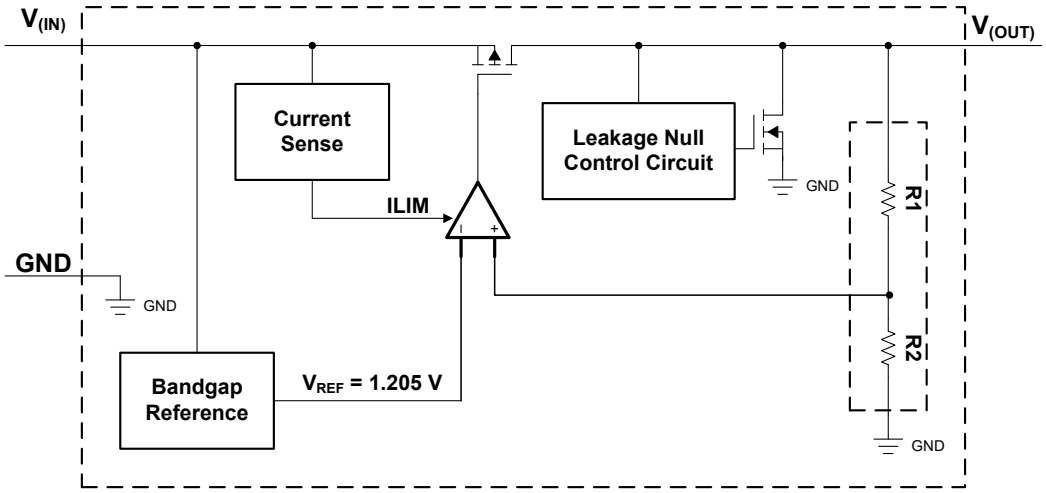
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7 Detailed Description

7.1 Overview

The TLV701 low-dropout regulator (LDO) consumes only 3.4 µA of quiescent current across the entire output current range, and offers a wide input voltage range and low-dropout voltage in a small package. The device, which operates over an input range of 2.5 V to 24 V, is stable with any output capacitor greater than or equal to 0.47 µF. The low quiescent current across the complete load current range makes the TLV701 designed for powering battery-operated applications. The TLV701 has internal soft-start to control inrush current into the output capacitor. This LDO also has overcurrent protection during a load-short or fault condition on the output.

7.2 Functional Block Diagram



7.3 Feature Description

7.3.1 Wide Supply Range

This device has an operational input supply range of 2.5 V to 24 V, allowing for a wide range of applications. This wide supply range is designed for applications that have either large transients or high DC voltage supplies.

7.3.2 Low Quiescent Current

This device only requires 3.4 µA (typical) of quiescent current across the complete load current range (0 mA to 150 mA) and has a maximum current consumption of 4.5 µA (for new device only) at –40°C to +125°C.

7.3.3 Dropout Voltage (V_{DO})

Dropout voltage (V_{DO}) is defined as the input voltage minus the output voltage ($V_{IN} – V_{OUT}$) at the rated output current (I_{RATED}), where the pass transistor is fully on. I_{RATED} is the maximum I_{OUT} listed in the *Recommended Operating Conditions* table. In dropout operation, the pass transistor is in the ohmic or triode region of operation, and acts as a switch. The dropout voltage indirectly specifies a minimum input voltage greater than the nominal programmed output voltage at which the output voltage is expected to stay in regulation. If the input voltage falls to less than the value required to maintain output regulation, the output voltage falls as well.

For a CMOS regulator, the dropout voltage is determined by the drain-source, on-state resistance ($R_{DS(ON)}$) of the pass transistor. Therefore, if the linear regulator operates at less than the rated current, the dropout voltage for that current scales accordingly. Use 式 1 to calculate the $R_{DS(ON)}$ of the device.

$$R_{DS(ON)} = \frac{V_{DO}}{I_{RATED}} \quad (1)$$

7.3.4 Current Limit

The device has an internal current limit circuit that protects the regulator during transient high-load current faults or shorting events. The current limit is a brick-wall scheme. In a high-load current fault, the brick-wall scheme limits the output current to the current limit (I_{CL}). I_{CL} is listed in the *Electrical Characteristics* table.

The output voltage is not regulated when the device is in current limit. When a current limit event occurs, the device begins to heat up because of the increase in power dissipation. When the device is in brick-wall current limit, the pass transistor dissipates power [(V_{IN} – V_{OUT}) × I_{CL}]. For more information on current limits, see the *Know Your Limits* application note.

図 7-1 shows a diagram of the current limit.

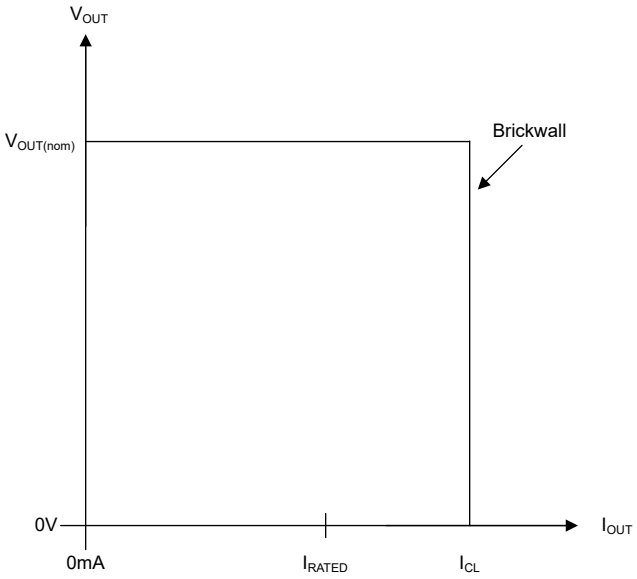


図 7-1. Current Limit

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7.4 Device Functional Modes

表 7-1 provides a quick comparison between the normal and dropout modes of operation.

表 7-1. Device Functional Mode Comparison

OPERATING MODE	PARAMETER	
	V_{IN}	I_{OUT}
Normal	$V_{IN} > V_{OUT(nom)} + V_{DO}$	$I_{OUT} < I_{CL}$
Dropout	$V_{IN} < V_{OUT(nom)} + V_{DO}$	$I_{OUT} < I_{CL}$

7.4.1 Normal Operation

The device regulates to the nominal output voltage under the following conditions:

- The input voltage is greater than the nominal output voltage plus the dropout voltage ($V_{OUT(nom)} + V_{DO}$)
- The output current is less than the current limit ($I_{OUT} < I_{CL}$)
- The device junction temperature is greater than –40°C and less than +125°C

7.4.2 Dropout Operation

If the input voltage is lower than the nominal output voltage plus the specified dropout voltage, but all other conditions are met for normal operation, the device operates in dropout mode. In this mode, the output voltage tracks the input voltage. During this mode, the transient performance of the device becomes significantly degraded because the pass transistor is in the ohmic or triode region, and acts as a switch. Line or load transients in dropout can result in large output voltage deviations.

When the device is in a steady dropout state (defined as when the device is in dropout, $V_{IN} < V_{OUT(NOM)} + V_{DO}$, directly after being in a normal regulation state, but *not* during start up), the pass transistor is driven into the ohmic or triode region. When the input voltage returns to a value greater than or equal to the nominal output voltage plus the dropout voltage ($V_{OUT(NOM)} + V_{DO}$), the output voltage can overshoot for a short period of time while the device pulls the pass transistor back into the linear region.

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8 Application and Implementation

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8.1 Application Information

The TLV701 LDO regulator is designed for battery-powered applications and is a good attachment to low-power microcontrollers (such as the MSP430) because of the device low I_Q performance across the entire load current range. The ultra-low supply current of the TLV701 maximizes efficiency at light loads, and the high input voltage range and flexibility of output voltage selection in fixed output levels makes the device applicable for supplies such as unconditioned solar panels.

8.2 Typical Application

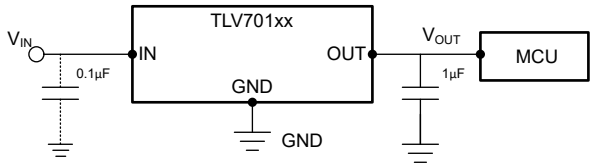


図 8-1. Typical Application

8.2.1 Design Requirements

Select the desired device based on the output voltage.

Provide an input supply with adequate headroom to account for dropout and output current to account for the GND pin current, and power the load.

8.2.2 Detailed Design Procedure

8.2.2.1 External Capacitor Requirements

The device is designed to be stable using low equivalent series resistance (ESR) ceramic capacitors at the input and output. Multilayer ceramic capacitors have become the industry standard for these types of applications and are recommended, but must be used with good judgment. Ceramic capacitors that employ X7R-, X5R-, and C0G-rated dielectric materials provide relatively good capacitive stability across temperature, whereas the use of Y5V-rated capacitors is discouraged because of large variations in capacitance.

Regardless of the ceramic capacitor type selected, the effective capacitance varies with operating voltage and temperature. Generally, expect the effective capacitance to decrease by as much as 50%. The input and output capacitors recommended in the *Recommended Operating Conditions* table account for an effective capacitance of approximately 50% of the nominal value.

8.2.2.2 Input and Output Capacitor Requirements

Although an input capacitor is not required for stability, good analog design practice is to connect a capacitor from IN to GND. This capacitor counteracts reactive input sources and improves transient response, input ripple, and PSRR. Use an input capacitor if the source impedance is more than 0.5 Ω. A higher value capacitor can be necessary if large, fast rise-time load or line transients are anticipated or if the device is located several inches from the input power source.

Dynamic performance of the device is improved by using a large output capacitor. Use an output capacitor within the range specified in the *Recommended Operating Conditions* table for stability.

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8.2.2.3 Reverse Current

Excessive reverse current can damage this device. Reverse current flows through the intrinsic body diode of the PMOS pass transistor instead of the normal conducting channel. At high magnitudes, this current flow degrades the long-term reliability of the device.

Conditions where reverse current can occur are outlined in this section, all of which can exceed the absolute maximum rating of $V_{OUT} \leq V_{IN} + 0.3$ V. These conditions are:

- If the device has a large C_{OUT} and the input supply collapses with little or no load current
- The output is biased when the input supply is not established
- The output is biased above the input supply

If reverse current flow is expected in the application, external protection is recommended to protect the device. Reverse current is not limited in the device, so external limiting is required if extended reverse voltage operation is anticipated. Limit reverse current to 5% or less of the rated output current of the device in the event this current cannot be avoided.

図 8-2 shows one approach for protecting the device.

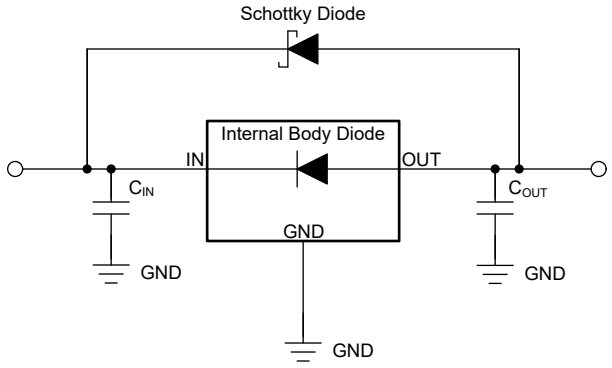


図 8-2. Example Circuit for Reverse Current Protection Using a Schottky Diode

8.2.2.4 Power Dissipation (P_D)

Circuit reliability requires consideration of the device power dissipation, location of the circuit on the printed circuit board (PCB), and correct sizing of the thermal plane. The PCB area around the regulator must have few or no other heat-generating devices that cause added thermal stress.

To first-order approximation, power dissipation in the regulator depends on the input-to-output voltage difference and load conditions. The following equation calculates power dissipation (P_D).

$$P_D = (V_{IN} - V_{OUT}) \times I_{OUT} \quad (2)$$

注

Power dissipation can be minimized, and therefore greater efficiency can be achieved, by correct selection of the system voltage rails. For the lowest power dissipation, use the minimum input voltage required for correct output regulation.

For devices with a thermal pad, the primary heat conduction path for the device package is through the thermal pad to the PCB. Solder the thermal pad to a copper pad area under the device. This pad area must contain an array of plated vias that conduct heat to additional copper planes for increased heat dissipation.

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The maximum power dissipation determines the maximum allowable ambient temperature (T_A) for the device. According to the following equation, power dissipation and junction temperature are most often related by the junction-to-ambient thermal resistance ($R_{\theta JA}$) of the combined PCB and device package and the temperature of the ambient air (T_A).

$$T_J = T_A + (R_{\theta JA} \times P_D) \quad (3)$$

Thermal resistance ($R_{\theta JA}$) is highly dependent on the heat-spreading capability built into the particular PCB design, and therefore varies according to the total copper area, copper weight, and location of the planes. The junction-to-ambient thermal resistance listed in the *Thermal Information* table is determined by the JEDEC standard PCB and copper-spreading area, and is used as a relative measure of package thermal performance.

8.2.2.5 Estimating Junction Temperature

The JEDEC standard now recommends the use of psi (Ψ) thermal metrics to estimate the junction temperatures of the linear regulator when in-circuit on a typical PCB board application. These metrics are not thermal resistance parameters and instead offer a practical and relative way to estimate junction temperature. These psi metrics are determined to be significantly independent of the copper area available for heat-spreading. The *Thermal Information* table lists the primary thermal metrics, which are the junction-to-top characterization parameter (ψ_{JT}) and junction-to-board characterization parameter (ψ_{JB}). These parameters provide two methods for calculating the junction temperature (T_J), as described in the following equations. Use the junction-to-top characterization parameter (ψ_{JT}) with the temperature at the center-top of device package (T_T) to calculate the junction temperature. Use the junction-to-board characterization parameter (ψ_{JB}) with the PCB surface temperature 1 mm from the device package (T_B) to calculate the junction temperature.

$$T_J = T_T + \psi_{JT} \times P_D \quad (4)$$

where:

- P_D is the dissipated power
- T_T is the temperature at the center-top of the device package

$$T_J = T_B + \psi_{JB} \times P_D \quad (5)$$

where:

- T_B is the PCB surface temperature measured 1 mm from the device package and centered on the package edge

For detailed information on the thermal metrics and how to use them, see the *Semiconductor and IC Package Thermal Metrics* application note.

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8.2.3 Application Curves

at operating temperature T_J = 25°C, V_{IN} = $V_{OUT(NOM)}$ + 1.0 V or 2.5 V (whichever is greater), I_{OUT} = 1 mA, C_{IN} = 1 µF, and C_{OUT} = 1 µF (unless otherwise noted)

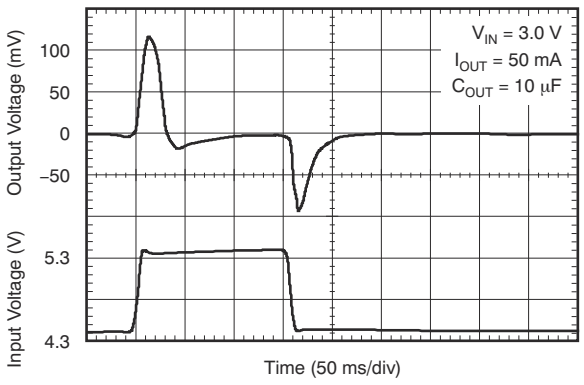


図 8-3. Line Transient Response for Legacy Chip

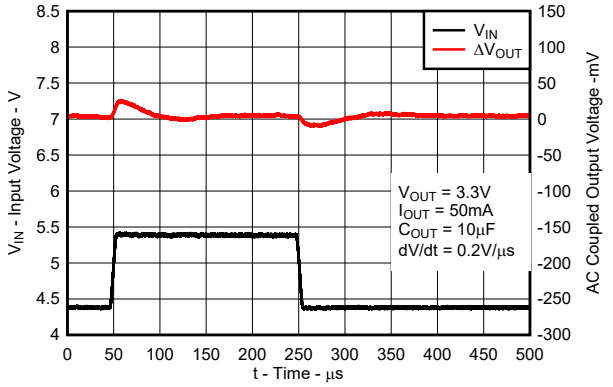


図 8-4. Line Transient Response for New Chip

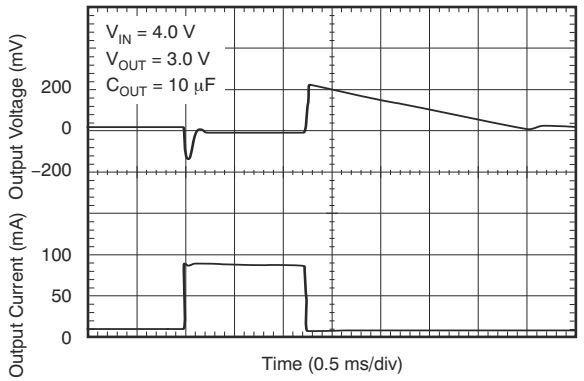


図 8-5. Load Transient Response for Legacy Chip

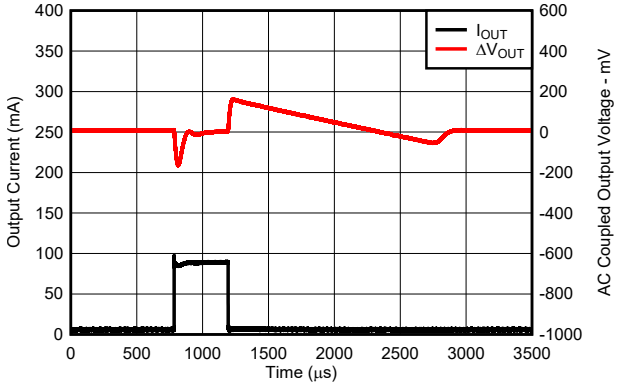


図 8-6. Load Transient Response for New Chip

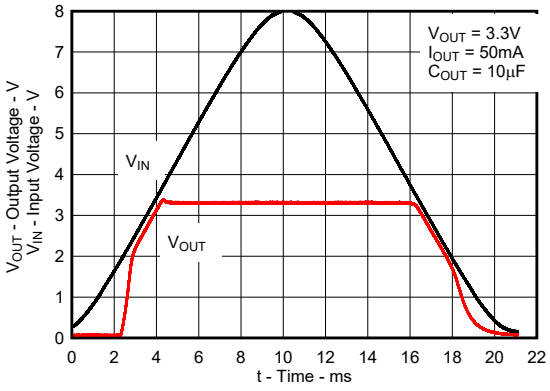


図 8-7. Power-Up, Power-Down for New Chip

8.3 Best Design Practices

Place at least one 0.47-µF capacitor as close as possible to the OUT and GND pins of the regulator.

Do not connect the output capacitor to the regulator using a long, thin trace.

Connect an input capacitor as close as possible to the IN and GND pins of the regulator for best performance.

Do not exceed the absolute maximum ratings.

8.4 Power Supply Recommendations

The TLV701 is designed to operate from an input voltage supply range between 2.5 V and 24 V. The input voltage range must provide adequate headroom for the device to have a regulated output. Inductive impedances between the input supply and the IN pin can create significant voltage excursions at the IN pin during start-up or load transient events. If inductive impedances are unavoidable, use an input capacitor.

8.5 Layout

8.5.1 Layout Guidelines

For best overall performance, place all circuit components on the same side of the printed circuit board (PCB) and as near as practical to the respective LDO pin connections. Place ground return connections for the input and output capacitors as close to the GND pin as possible, using wide, component-side, copper planes. Do not use vias and long traces to create LDO circuit connections to the input capacitor, output capacitor, or the resistor divider because this practice negatively affects system performance. This grounding and layout scheme minimizes inductive parasitics, and thereby reduces load-current transients, minimizes noise, and increases circuit stability. A ground reference plane is also recommended and is either embedded in the PCB or located on the bottom side of the PCB opposite the components. This reference plane serves to assure accuracy of the output voltage and shield the LDO from noise.

8.5.1.1 Power Dissipation

To provide reliable operation, worst-case junction temperature must not exceed 125°C. This restriction limits the power dissipation the regulator can handle in any given application. To make sure the junction temperature is within acceptable limits, calculate the maximum allowable dissipation, $P_{D(max)}$, and the actual dissipation, P_D, which must be less than or equal to $P_{D(max)}$.

The maximum-power-dissipation limit is determined by:

$$P_{D(max)} = \frac{T_J max - T_A}{R_{\theta JA}} \tag{6}$$

where:

- $T_J$max is the maximum allowable junction temperature
- $R_{\theta JA}$ is the thermal resistance junction-to-ambient for the package (see the *Thermal Information* table)
- T_A is the ambient temperature

The regulator dissipation is calculated by:

$$P_D = (V_{IN} - V_{OUT}) \times I_{OUT} \tag{7}$$

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8.5.2 Layout Example

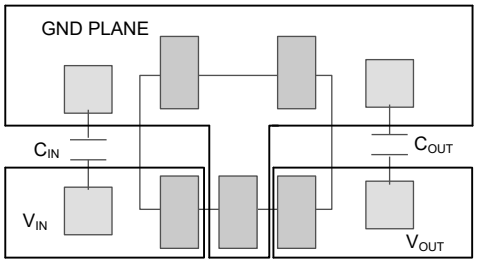


図 8-8. Layout Example for the DBV Package

9 Device and Documentation Support

9.1 Device Support

9.1.1 Development Support

9.1.1.1 Evaluation Module

An evaluation module (EVM) is available to assist in the initial circuit performance evaluation using the TLV701. The LP38693-ADJEV (and related user guide) can be requested at the Texas Instruments website through the product folders or purchased directly from the TI eStore.

9.1.2 Device Nomenclature

表 9-1. Available Options(1)

PRODUCT	V_{OUT}
TLV701**xxyyyz** Legacy chip	**xx** is the nominal output voltage (for example 33 = 3.3 V). **yyy** is the package designator. **z** is the package quantity.
TLV701**xxyyyzM3** New chip	**xx** is the nominal output voltage (for example 33 = 3.3 V). **yyy** is the package designator. **z** is the package quantity. **M3** is a suffix designator for newer chip redesigns, fabricated on the latest TI process technology.

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or visit the device product folder at www.ti.com.

9.2 Documentation Support

9.2.1 Related Documentation

For related documentation see the following:

- Texas Instruments, *TLV70433DBVEVM-712, TLV70433PKEVM-712 Evaluation Modules* user guide

9.3 ドキュメントの更新通知を受け取る方法

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10 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package | Pins	Package qty | Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
TLV70130DBVR	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	YBVA
TLV70130DBVR.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	YBVA
TLV70130DBVT	Active	Production	SOT-23 (DBV) | 5	250 | SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	YBVA
TLV70130DBVT.A	Active	Production	SOT-23 (DBV) | 5	250 | SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	YBVA
TLV70133DBVR	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	YBWA
TLV70133DBVR.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	YBWA
TLV70133DBVRG4	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	YBWA
TLV70133DBVRG4.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	YBWA
TLV70133DBVRM3	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	YBWA
TLV70133DBVRM3.A	Active	Production	SOT-23 (DBV) | 5	3000 | LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	YBWA
TLV70133DBVT	Obsolete	Production	SOT-23 (DBV) | 5	-	-	Call TI	Call TI	-40 to 125	YBWA

(1) **Status:** For more details on status, see our product life cycle.

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

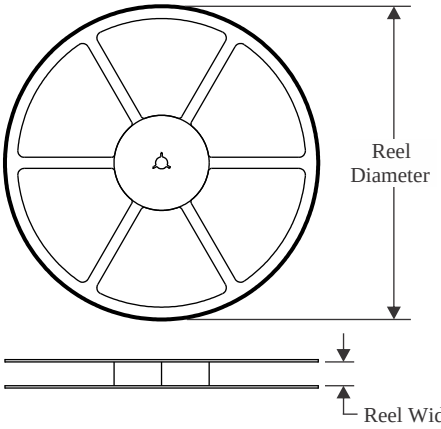
Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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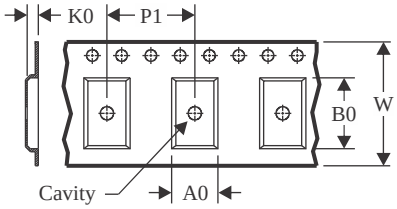
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TAPE AND REEL INFORMATION

REEL DIMENSIONS

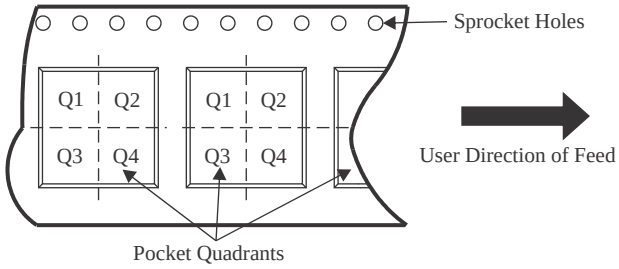


TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

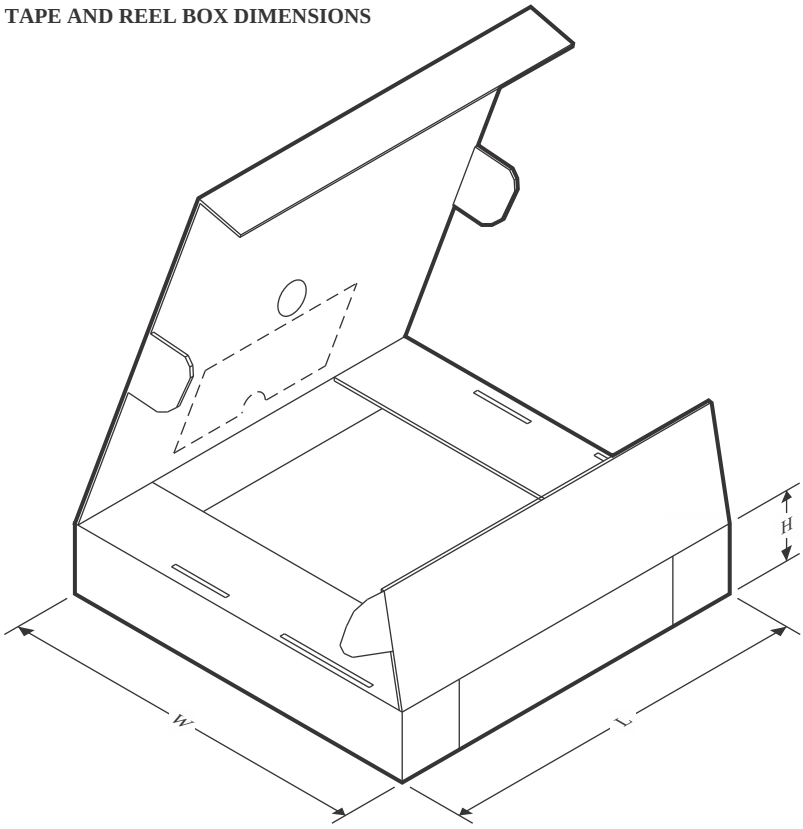
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLV70130DBVR	SOT-23	DBV	5	3000	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TLV70130DBVT	SOT-23	DBV	5	250	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TLV70133DBVR	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TLV70133DBVRG4	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TLV70133DBVRM3	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

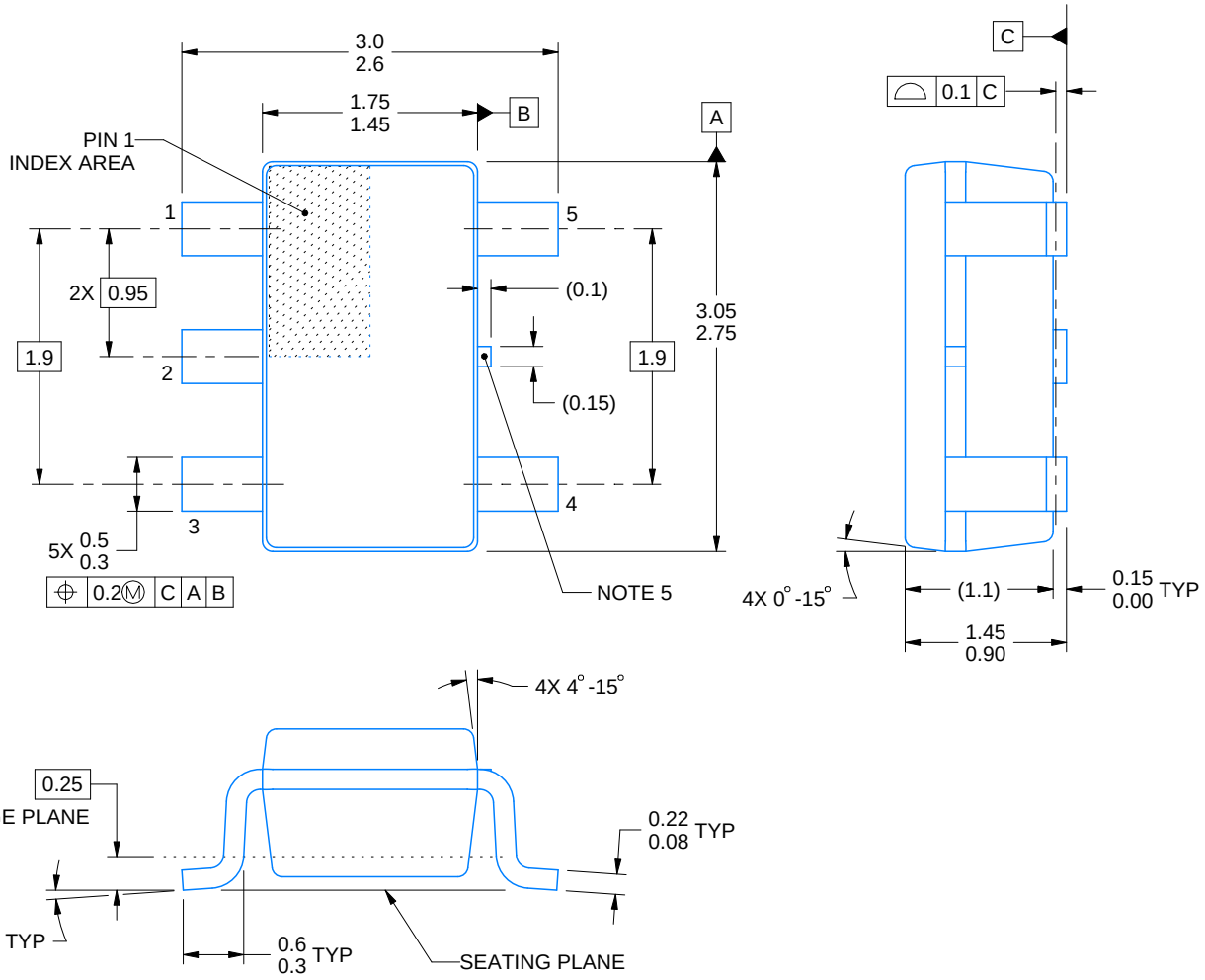
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLV70130DBVR	SOT-23	DBV	5	3000	180.0	180.0	18.0
TLV70130DBVT	SOT-23	DBV	5	250	180.0	180.0	18.0
TLV70133DBVR	SOT-23	DBV	5	3000	210.0	185.0	35.0
TLV70133DBVRG4	SOT-23	DBV	5	3000	210.0	185.0	35.0
TLV70133DBVRM3	SOT-23	DBV	5	3000	210.0	185.0	35.0

PACKAGE OUTLINE

DBV0005A

SOT-23 - 1.45 mm max height

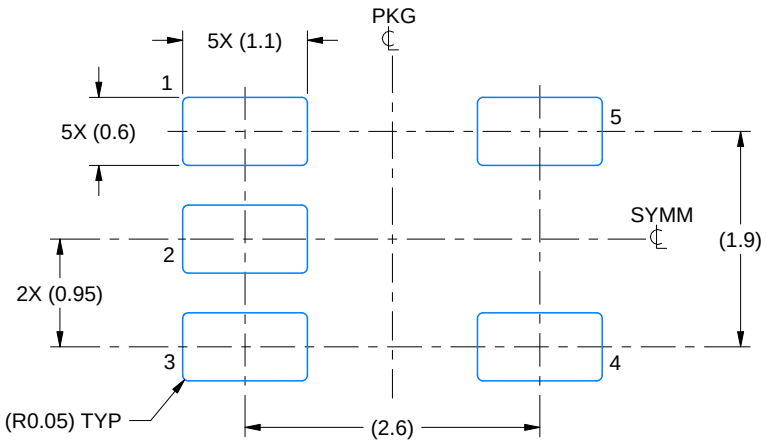
SMALL OUTLINE TRANSISTOR



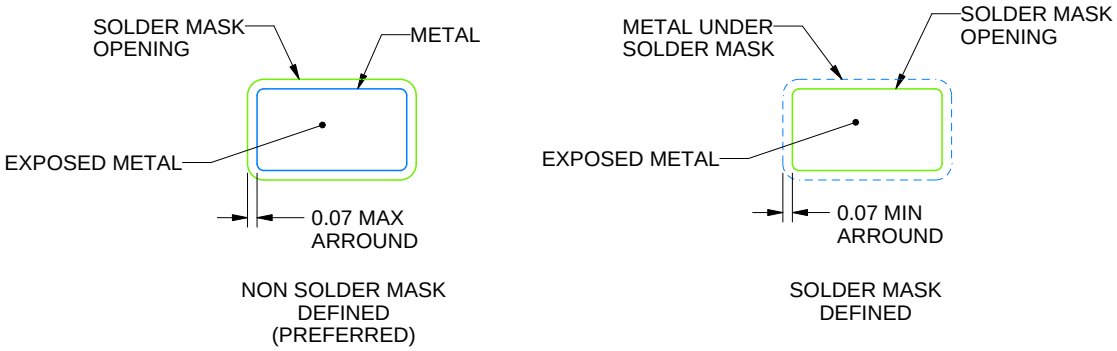
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NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. Refernce JEDEC MO-178.
4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.
5. Support pin may differ or may not be present.



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:15X

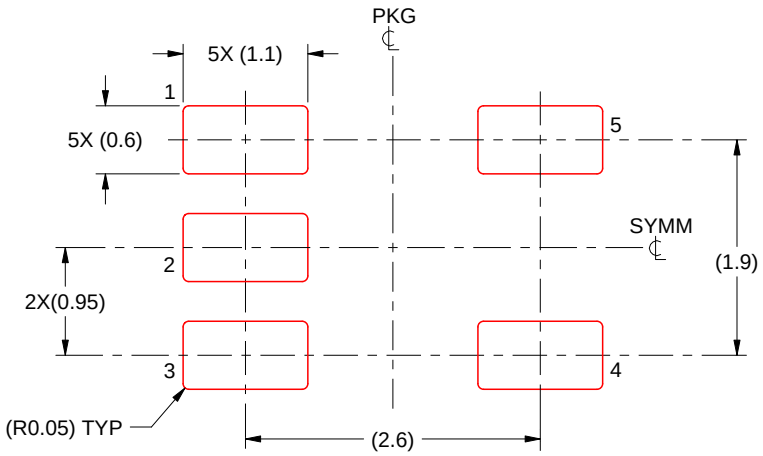


SOLDER MASK DETAILS

4214839/K 08/2024

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE:15X

4214839/K 08/2024

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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