TMS320C6412 Fixed-Point Digital Signal Processor

Data Manual

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Revision History

This data manual revision history highlights the technical changes made to the SPRS219I device-specific data sheet to make it an SPRS219J revision.

PAGE(s) NO.	ADDS/CHANGES/DELETES
101	Added note for V_{OH} and V_{OL}.

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1   Features

- **High-Performance Fixed-Point Digital Signal Processor (TMS320C6412)**
  - 2-, 1.67-, 1.39-ns Instruction Cycle Time
  - 500-, 600-, 720-MHz Clock Rate
  - Eight 32-Bit Instructions/Cycle
  - 4000, 4800, 5760 MIPS
  - Fully Software-Compatible With C64x™

- **VelociTI.2™ Extensions to VelociTI™ Advanced Very-Long-Instruction-Word (VLIW) TMS320C64x™ DSP Core**
  - Eight Highly Independent Functional Units With VelociTI.2™ Extensions:
    - Six ALUs (32-/40-Bit), Each Supports Single 32-Bit, Dual 16-Bit, or Quad 8-Bit Arithmetic per Clock Cycle
    - Two Multipliers Support Four 16 x 16-Bit Multiplies (32-Bit Results) per Clock Cycle or Eight 8 x 8-Bit Multiplies (16-Bit Results) per Clock Cycle
  - Load-Store Architecture With Non-Aligned Support
  - 64 32-Bit General-Purpose Registers
  - Instruction Packing Reduces Code Size
  - All Instructions Conditional

- **Instruction Set Features**
  - Byte-Addressable (8-/16-/32-/64-Bit Data)
  - 8-Bit Overflow Protection
  - Bit-Field Extract, Set, Clear
  - Normalization, Saturation, Bit-Counting
  - VelociTI.2™ Increased Orthogonality

- **L1/L2 Memory Architecture**
  - 128K-Bit (16K-Byte) L1P Program Cache (Direct Mapped)
  - 128K-Bit (16K-Byte) L1D Data Cache (2-Way Set-Associative)
  - 2M-Bit (256K-Byte) L2 Unified Mapped RAM/Cache (Flexible RAM/Cache Allocation)

- **Endianess: Little Endian, Big Endian**

- **64-Bit External Memory Interface (EMIF)**
  - Glueless Interface to Asynchronous Memories (SRAM and EPROM) and Synchronous Memories (SDRAM, SBSRAM, ZBT SRAM, and FIFO)
  - 1024M-Byte Total Addressable External Memory Space

- **Enhanced Direct-Memory-Access (EDMA) Controller (64 Independent Channels)**

- **10/100 Mb/s Ethernet MAC (EMAC)**
  - IEEE 802.3 Compliant
  - Media Independent Interface (MII)
  - 8 Independent Transmit (TX) Channels and 1 Receive (RX) Channel

- **Management Data Input/Output (MDIO)**

- **Host-Port Interface (HPI) [32-/16-Bit]**

- **32-Bit/66-MHz, 3.3-V Peripheral Component Interconnect (PCI) Master/Slave Interface Conforms to PCI Specification 2.2**

- **Inter-Integrated Circuit (I²C) Bus**

- **Two Multichannel Buffered Serial Ports**

- **Three 32-Bit General-Purpose Timers**

- **Sixteen General-Purpose I/O (GPIO) Pins**

- **Flexible PLL Clock Generator**

- **IEEE-1149.1 (JTAG†) Boundary-Scan-Compatible**

- **548-Pin Ball Grid Array (BGA) Package (GDK and ZDK Suffixes), 0.8-mm Ball Pitch**

- **548-Pin Ball Grid Array (BGA) Package (GNZ and ZNZ Suffixes), 1.0-mm Ball Pitch**

- **0.13-µm/6-Level Cu Metal Process (CMOS)**

- **3.3-V I/Os, 1.2-V Internal (-500)**

- **3.3-V I/Os, 1.4-V Internal (A-500, A–600, -600, -720)**

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1.1 GDK and ZDK BGA Packages (Bottom View)

**GDK and ZDK 548-PIN BALL GRID ARRAY (BGA) PACKAGES
(BOTTOM VIEW)**
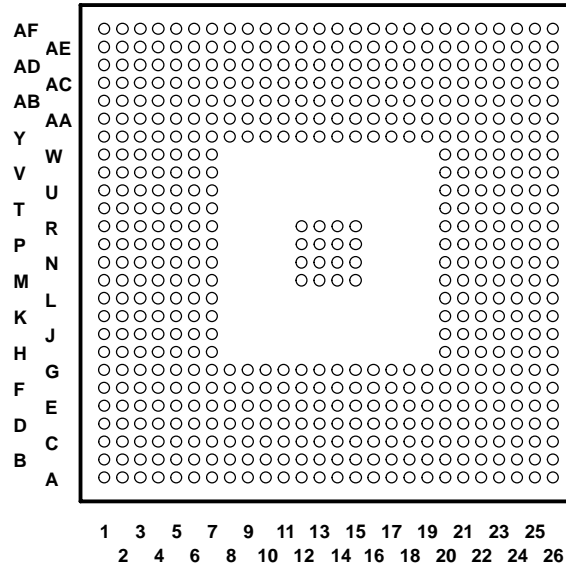
Figure 1–1. GDK and ZDK BGA Packages (Bottom View)

1.2 GNZ and ZNZ BGA Packages (Bottom View)

**GNZ and ZNZ 548-PIN BALL GRID ARRAY (BGA) PACKAGEs
(BOTTOM VIEW)**
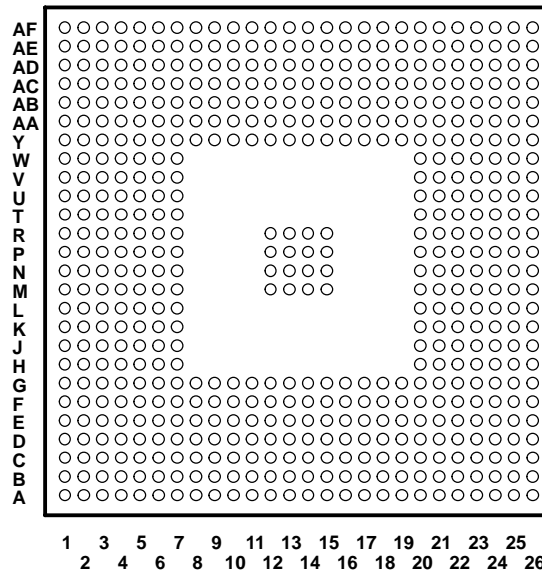
Figure 1–2. GNZ and ZNZ BGA Packages (Bottom View)

1.3   Description

The TMS320C64x™ DSPs (including the TMS320C6412 device) are the highest-performance fixed-point DSP generation in the TMS320C6000™ DSP platform. The TMS320C6412 (C6412) device is based on the second-generation high-performance, advanced VelociTI™ very-long-instruction-word (VLIW) architecture (VelociTI.2™) developed by Texas Instruments (TI), making these DSPs an excellent choice for multichannel and multifunction applications. The C64x™ is a code-compatible member of the C6000™ DSP platform.

With performance of up to 5760 million instructions per second (MIPS) at a clock rate of 720 MHz, the C6412 device offers cost-effective solutions to high-performance DSP programming challenges. The C6412 DSP possesses the operational flexibility of high-speed controllers and the numerical capability of array processors. The C64x™ DSP core processor has 64 general-purpose registers of 32-bit word length and eight highly independent functional units—two multipliers for a 32-bit result and six arithmetic logic units (ALUs)—with VelociTI.2™ extensions. The VelociTI.2™ extensions in the eight functional units include new instructions to accelerate the performance in applications and extend the parallelism of the VelociTI™ architecture. The C6412 can produce four 16-bit multiply-accumulates (MACs) per cycle for a total of 2400 million MACs per second (MMACS), or eight 8-bit MACs per cycle for a total of 4800 MMACS. The C6412 DSP also has application-specific hardware logic, on-chip memory, and additional on-chip peripherals similar to the other C6000™ DSP platform devices.

The C6412 uses a two-level cache-based architecture and has a powerful and diverse set of peripherals. The Level 1 program cache (L1P) is a 128-Kbit direct mapped cache and the Level 1 data cache (L1D) is a 128-Kbit 2-way set-associative cache. The Level 2 memory/cache (L2) consists of an 2-Mbit memory space that is shared between program and data space. L2 memory can be configured as mapped memory, cache, or combinations of the two. The peripheral set includes: a 10/100 Mb/s Ethernet MAC (EMAC); a management data input/output (MDIO) module; an inter-integrated circuit (I2C) Bus module; two multichannel buffered serial ports (McBSPs); three 32-bit general-purpose timers; a user-configurable 16-bit or 32-bit host-port interface (HPI16/HPI32); a peripheral component interconnect (PCI); a 16-pin general-purpose input/output port (GP0) with programmable interrupt/event generation modes; and a 64-bit glueless external memory interface (EMIFA), which is capable of interfacing to synchronous and asynchronous memories and peripherals.

The ethernet media access controller (EMAC) provides an efficient interface between the C6412 DSP core processor and the network. The C6412 EMAC support both 10Base-T and 100Base-TX, or 10 Mbits/second (Mbps) and 100 Mbps in either half- or full-duplex, with hardware flow control and quality of service (QOS) support. The C6412 EMAC makes use of a custom interface to the DSP core that allows efficient data transmission and reception. For more details on the EMAC, see the *TMS320C6000 DSP Ethernet Media Access Controller (EMAC) / Management Data Input/Output (MDIO) Module Reference Guide* (literature number SPRU628).

The management data input/output (MDIO) module continuously polls all 32 MDIO addresses in order to enumerate all PHY devices in the system. Once a PHY candidate has been selected by the DSP, the MDIO module transparently monitors its link state by reading the PHY status register. Link change events are stored in the MDIO module and can optionally interrupt the DSP, allowing the DSP to poll the link status of the device without continuously performing costly MDIO accesses. For more details on the MDIO port, see the *TMS320C6000 DSP Ethernet Media Access Controller (EMAC) / Management Data Input/Output (MDIO) Module Reference Guide* (literature number SPRU628).

The I2C0 port on the TMS320C6412 allows the DSP to easily control peripheral devices and communicate with a host processor. In addition, the standard multichannel buffered serial port (McBSP) may be used to communicate with serial peripheral interface (SPI) mode peripheral devices.

The C6412 has a complete set of development tools which includes: a new C compiler, an assembly optimizer to simplify programming and scheduling, and a Windows™ debugger interface for visibility into source code execution.

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1.4 Device Characteristics

Table 1–1 provides an overview of the C6412 DSP. The table shows significant features of the C6412 device, including the capacity of on-chip RAM, the peripherals, the CPU frequency, and the package type with pin count.

Table 1–1. Characteristics of the C6412 Processor

HARDWARE FEATURES		C6412
Peripherals Not all peripherals pins are available at the same time (For more detail, see the Device Configuration section).	EMIFA (64-bit bus width) (clock source = AECLKIN)	1
	EDMA (64 independent channels)	1
	I2C0 (uses Peripheral Clock)	1
	HPI (32- or 16-bit user selectable)	1 (HPI16 or HPI32)
	PCI (32-bit), 66-MHz/33-MHz [DeviceID Register value 0x9065]	1
	McBSPs (internal clock source = CPU/4 clock frequency)	2
	10/100 Ethernet MAC (EMAC)	1
	Management Data Input/Output (MDIO)	1
	32-Bit Timers (internal clock source = CPU/8 clock frequency)	3
	General-Purpose Input/Output Port (GP0)	16
On-Chip Memory	Size (Bytes)	288K
	Organization	16K-Byte (16KB) L1 Program (L1P) Cache 16KB L1 Data (L1D) Cache 256KB Unified Mapped RAM/Cache (L2)
CPU ID + CPU Rev ID	Control Status Register (CSR.[31:16])	0x0C01
JTAG BSDL_ID	JTAGID register (address location: 0x01B3F008)	0x0007902F
Frequency	MHz	500, 600, 720
Cycle Time	ns	2 ns (C6412-500) and (C6412**A**-500) [500 MHz CPU, 100 MHz EMIF[†], 33 MHz PCI port] 1.67 ns (C6412-600) and (C6412**A**–600) [600 MHz CPU, 133 MHz EMIF[†], 66 MHz PCI port] 1.39 ns (C6412-720) [720 MHz CPU, 133 MHz EMIF[†], 66 MHz PCI port]
Voltage	Core (V)	1.2 V (-500) 1.4 V (**A**-500, **A**–600, -600, -720)
	I/O (V)	3.3 V
PLL Options	CLKIN frequency multiplier	Bypass (x1), x6, x12
BGA Package[‡]	23 x 23 mm	548-Pin BGA (GDK and ZDK)
	27 x 27 mm	548-Pin BGA (GNZ and ZNZ)
Process Technology	μm	0.13 μm
Product Status[§]	Product Preview (PP), Advance Information (AI), or Production Data (PD)	PD

[†] On this C64x™ device, the rated EMIF speed affects only the SDRAM interface on the EMIF. For more detailed information, see the EMIF device speed portion of this data sheet.

[‡] For the exact markings of pin A1, see the *TMS320C6412 Digital Signal Processor Silicon Errata* (Literature Number: SPRZ199).

[§] PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

1.4.1 *Device Compatibility*

The C6412 device is a code-compatible member of the C6000™ DSP platform.

The C64x™ DSP generation of devices has a diverse and powerful set of peripherals.

For more detailed information on the device compatibility and similarities/differences among the DM642, C6412, and other C64x™ devices, see the *TMS320DM642 Technical Overview* (literature number SPRU615).

1.4.2 **Functional Block Diagram**

Figure 1−3 shows the functional block diagram of the C6412 device.
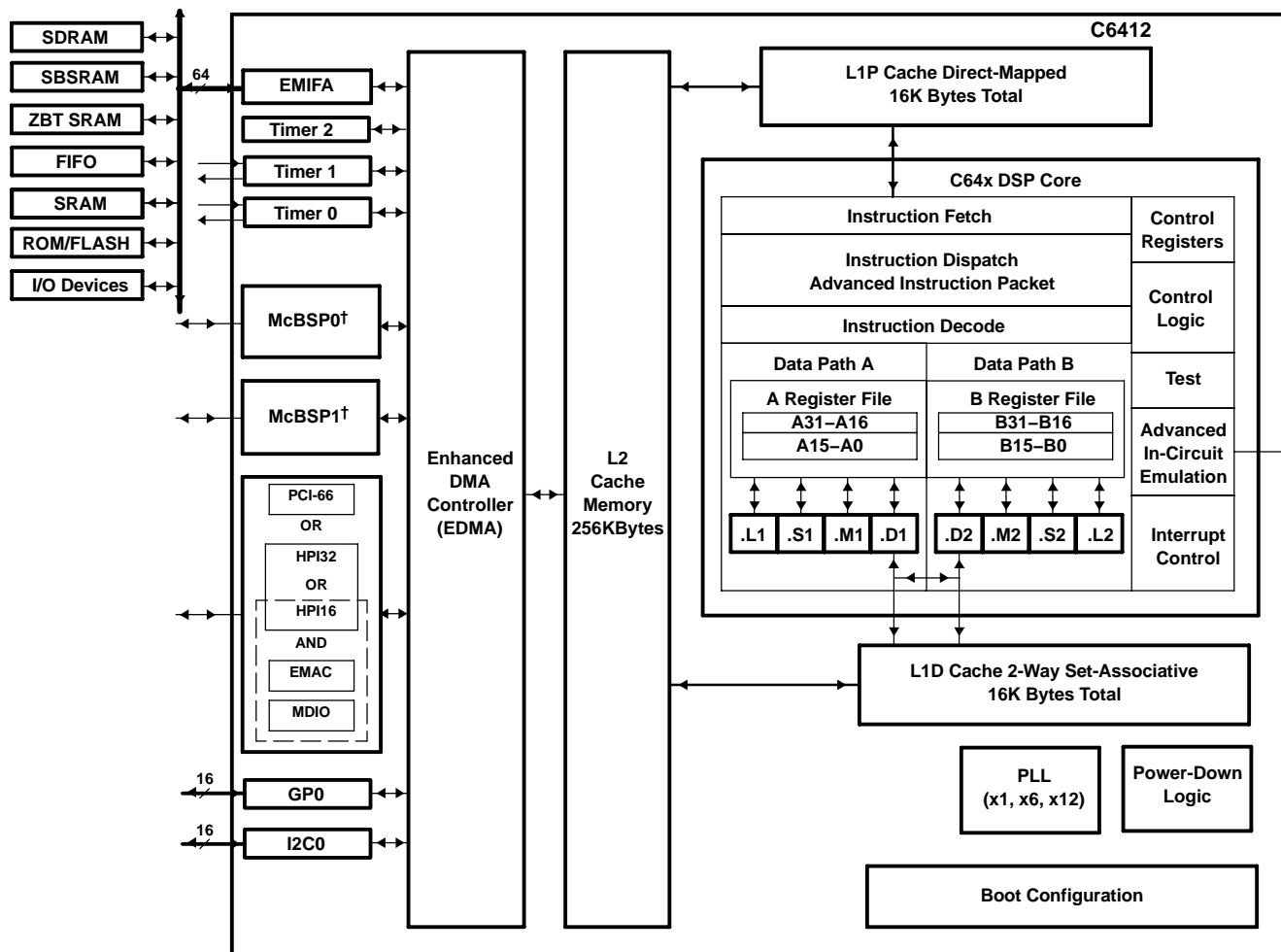
† McBSPs: Framing Chips – H.100, MVIP, SCSA, T1, E1; AC97 Devices; SPI Devices; Codecs

Figure 1−3. Functional Block Diagram

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1.5   CPU (DSP Core) Description

The CPU fetches VelociTI™ advanced very-long instruction words (VLIWs) (256 bits wide) to supply up to eight 32-bit instructions to the eight functional units during every clock cycle. The VelociTI™ VLIW architecture features controls by which all eight units do not have to be supplied with instructions if they are not ready to execute. The first bit of every 32-bit instruction determines if the next instruction belongs to the same execute packet as the previous instruction, or whether it should be executed in the following clock as a part of the next execute packet. Fetch packets are always 256 bits wide; however, the execute packets can vary in size. The variable-length execute packets are a key memory-saving feature, distinguishing the C64x CPUs from other VLIW architectures. The C64x™ VelociTI.2™ extensions add enhancements to the TMS320C62x™ DSP VelociTI™ architecture. These enhancements include:

- Register file enhancements
- Data path extensions
- Quad 8-bit and dual 16-bit extensions with data flow enhancements
- Additional functional unit hardware
- Increased orthogonality of the instruction set
- Additional instructions that reduce code size and increase register flexibility

The CPU features two sets of functional units. Each set contains four units and a register file. One set contains functional units .L1, .S1, .M1, and .D1; the other set contains units .D2, .M2, .S2, and .L2. The two register files each contain 32 32-bit registers for a total of 64 general-purpose registers. In addition to supporting the packed 16-bit and 32-/40-bit fixed-point data types found in the C62x™ VelociTI™ VLIW architecture, the C64x™ register files also support packed 8-bit data and 64-bit fixed-point data types. The two sets of functional units, along with two register files, compose sides A and B of the CPU [see the functional block and CPU (DSP core) diagram, and Figure 1–4]. The four functional units on each side of the CPU can freely share the 32 registers belonging to that side. Additionally, each side features a "data cross path"—a single data bus connected to all the registers on the other side, by which the two sets of functional units can access data from the register files on the opposite side. The C64x CPU pipelines data-cross-path accesses over multiple clock cycles. This allows the same register to be used as a data-cross-path operand by multiple functional units in the same execute packet. All functional units in the C64x CPU can access operands via the data cross path. Register access by functional units on the same side of the CPU as the register file can service all the units in a single clock cycle. On the C64x CPU, a delay clock is introduced whenever an instruction attempts to read a register via a data cross path if that register was updated in the previous clock cycle.

In addition to the C62x™ DSP fixed-point instructions, the C64x™ DSP includes a comprehensive collection of quad 8-bit and dual 16-bit instruction set extensions. These VelociTI.2™ extensions allow the C64x CPU to operate directly on packed data to streamline data flow and increase instruction set efficiency. This is a key factor for video and imaging applications.

Another key feature of the C64x CPU is the load/store architecture, where all instructions operate on registers (as opposed to data in memory). Two sets of data-addressing units (.D1 and .D2) are responsible for all data transfers between the register files and the memory. The data address driven by the .D units allows data addresses generated from one register file to be used to load or store data to or from the other register file. The C64x .D units can load and store bytes (8 bits), half-words (16 bits), and words (32 bits) with a single instruction. And with the new data path extensions, the C64x .D unit can load and store doublewords (64 bits) with a single instruction. Furthermore, the non-aligned load and store instructions allow the .D units to access words and doublewords on any byte boundary. The C64x CPU supports a variety of indirect addressing modes using either linear- or circular-addressing with 5- or 15-bit offsets. All instructions are conditional, and most can access any one of the 64 registers. Some registers, however, are singled out to support specific addressing modes or to hold the condition for conditional instructions (if the condition is not automatically "true").

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The two .M functional units perform all multiplication operations. Each of the C64x .M units can perform two 16 × 16-bit multiplies or four 8 × 8-bit multiplies per clock cycle. The .M unit can also perform 16 × 32-bit multiply operations, dual 16 × 16-bit multiplies with add/subtract operations, and quad 8 × 8-bit multiplies with add operations. In addition to standard multiplies, the C64x .M units include bit-count, rotate, Galois field multiplies, and bidirectional variable shift hardware.

The two .S and .L functional units perform a general set of arithmetic, logical, and branch functions with results available every clock cycle. The arithmetic and logical functions on the C64x CPU include single 32-bit, dual 16-bit, and quad 8-bit operations.

The processing flow begins when a 256-bit-wide instruction fetch packet is fetched from a program memory. The 32-bit instructions destined for the individual functional units are "linked" together by "1" bits in the least significant bit (LSB) position of the instructions. The instructions that are "chained" together for simultaneous execution (up to eight in total) compose an execute packet. A "0" in the LSB of an instruction breaks the chain, effectively placing the instructions that follow it in the next execute packet. A C64x™ DSP device enhancement now allows execute packets to cross fetch-packet boundaries. In the TMS320C62x™/TMS320C67x™ DSP devices, if an execute packet crosses the fetch-packet boundary (256 bits wide), the assembler places it in the next fetch packet, while the remainder of the current fetch packet is padded with NOP instructions. In the C64x™ DSP device, the execute boundary restrictions have been removed, thereby, eliminating all of the NOPs added to pad the fetch packet, and thus, decreasing the overall code size. The number of execute packets within a fetch packet can vary from one to eight. Execute packets are dispatched to their respective functional units at the rate of one per clock cycle and the next 256-bit fetch packet is not fetched until all the execute packets from the current fetch packet have been dispatched. After decoding, the instructions simultaneously drive all active functional units for a maximum execution rate of eight instructions every clock cycle. While most results are stored in 32-bit registers, they can be subsequently moved to memory as bytes, half-words, or doublewords. All load and store instructions are byte-, half-word-, word-, or doubleword-addressable.

For more details on the C64x CPU functional units enhancements, see the following documents:

- *TMS320C6000 CPU and Instruction Set Reference Guide* (literature number SPRU189)
- *TMS320C64x Technical Overview* (literature number SPRU395)

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NOTE A: For the .M functional units, the long dst is 32 MSBs and the dst is 32 LSBs.
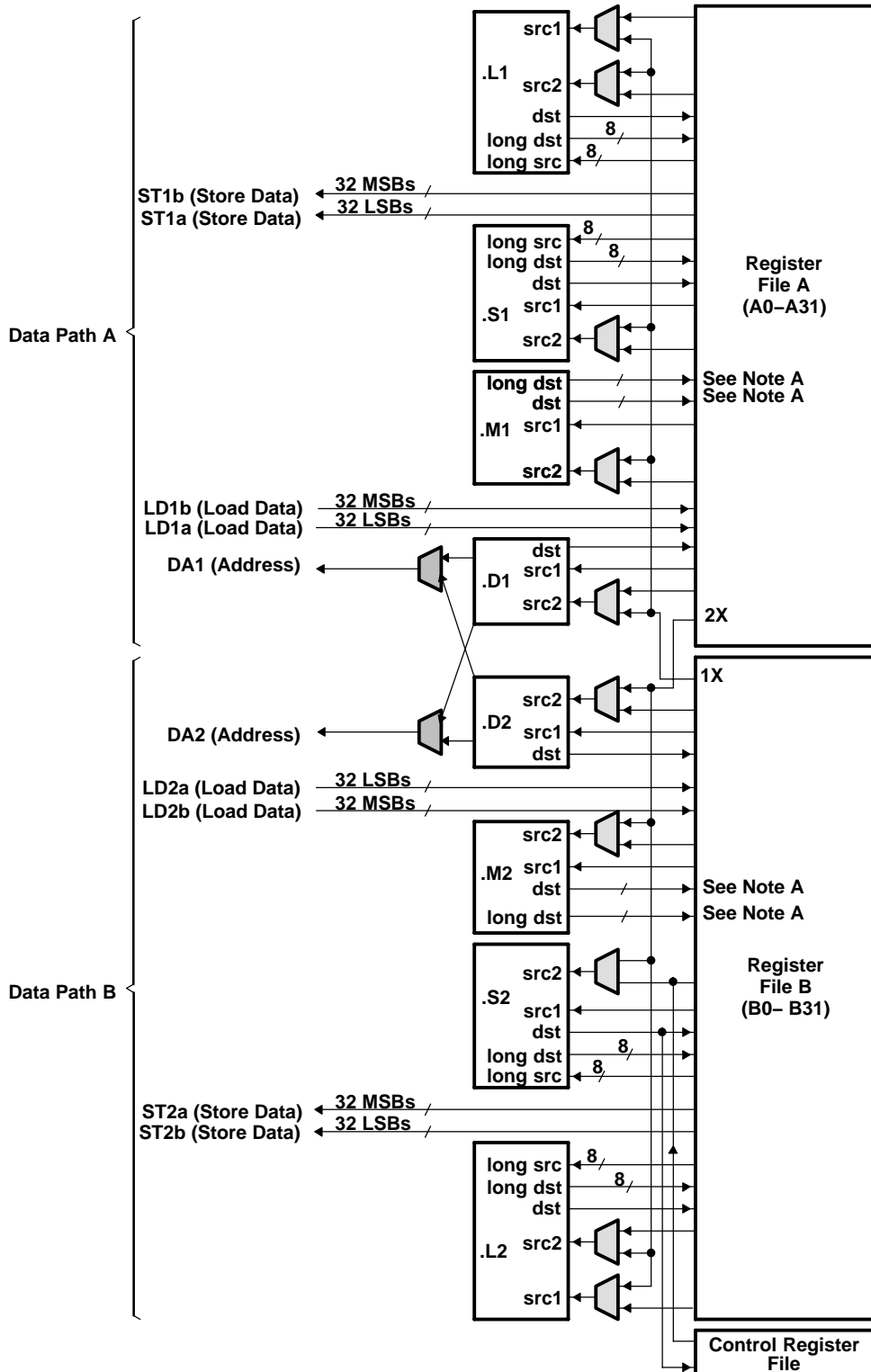
Figure 1−4. TMS320C64x™ CPU (DSP Core) Data Paths

1.6 Memory Map Summary

Table 1–2 shows the memory map address ranges of the C6412 device. Internal memory is always located at address 0 and can be used as both program and data memory. The external memory address ranges in the C6412 device begin at the hex address location 0x8000 0000 for EMIFA.

Table 1–2. TMS320C6412 Memory Map Summary

MEMORY BLOCK DESCRIPTION	BLOCK SIZE (BYTES)	HEX ADDRESS RANGE
Internal RAM (L2)	256K	0000 0000 – 0003 FFFF
Reserved	768K	0004 0000 – 000F FFFF
Reserved	23M	0010 0000 – 017F FFFF
External Memory Interface A (EMIFA) Registers	256K	0180 0000 – 0183 FFFF
L2 Registers	256K	0184 0000 – 0187 FFFF
HPI Registers	256K	0188 0000 – 018B FFFF
McBSP 0 Registers	256K	018C 0000 – 018F FFFF
McBSP 1 Registers	256K	0190 0000 – 0193 FFFF
Timer 0 Registers	256K	0194 0000 – 0197 FFFF
Timer 1 Registers	256K	0198 0000 – 019B FFFF
Interrupt Selector Registers	256K	019C 0000 – 019F FFFF
EDMA RAM and EDMA Registers	256K	01A0 0000 – 01A3 FFFF
Reserved	512K	01A4 0000 – 01AB FFFF
Timer 2 Registers	256K	01AC 0000 – 01AF FFFF
GP0 Registers	256K – 4K	01B0 0000 – 01B3 EFFF
Device Configuration Registers	4K	01B3 F000 – 01B3 FFFF
I2C0 Data and Control Registers	16K	01B4 0000 – 01B4 3FFF
Reserved	496K	01B4 4000 – 01BB BFFF
Emulation	256K	01BC 0000 – 01BF FFFF
PCI Registers	256K	01C0 0000 – 01C3 FFFF
Reserved	256K	01C4 0000 – 01C7 FFFF
EMAC Control	4K	01C8 0000 – 01C8 0FFF
EMAC Wrapper	8K	01C8 1000 – 01C8 2FFF
EWRAP Registers	2K	01C8 3000 – 01C8 37FF
MDIO Control Registers	2K	01C8 3800 – 01C8 3FFF
Reserved	3.5M	01C8 4000 – 01FF FFFF
QDMA Registers	52	0200 0000 – 0200 0033
Reserved	736M – 52	0200 0034 – 2FFF FFFF
McBSP 0 Data	64M	3000 0000 – 33FF FFFF
McBSP 1 Data	64M	3400 0000 – 37FF FFFF
Reserved	64M	3800 0000 – 3BFF FFFF
Reserved	1M	3C00 0000 – 3C0F FFFF
Reserved	64M – 1M	3C10 0000 – 3FFF FFFF
Reserved	832M	4000 0000 – 73FF FFFF
Reserved	192M	7400 0000 – 75FF FFFF
Reserved	192M	7600 0000 – 77FF FFFF

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Table 1–2. TMS320C6412 Memory Map Summary (Continued)

MEMORY BLOCK DESCRIPTION	BLOCK SIZE (BYTES)	HEX ADDRESS RANGE
Reserved	192M	7800 0000 – 79FF FFFF
Reserved	192M	7A00 0000 – 7BFF FFFF
Reserved	192M	7C00 0000 – 7DFF FFFF
Reserved	192M	7E00 0000 – 7FFF FFFF
EMIFA CE0	256M	8000 0000 – 8FFF FFFF
EMIFA CE1	256M	9000 0000 – 9FFF FFFF
EMIFA CE2	256M	A000 0000 – AFFF FFFF
EMIFA CE3	256M	B000 0000 – BFFF FFFF
Reserved	1G	C000 0000 – FFFF FFFF

1.6.1   L2 Architecture Expanded

Figure 1−5 shows the detail of the L2 architecture on the TMS320C6412 device. For more information on the L2MODE bits, see the cache configuration (CCFG) register bit field descriptions in the *TMS320C64x Two-Level Internal Memory Reference Guide* (literature number SPRU610).
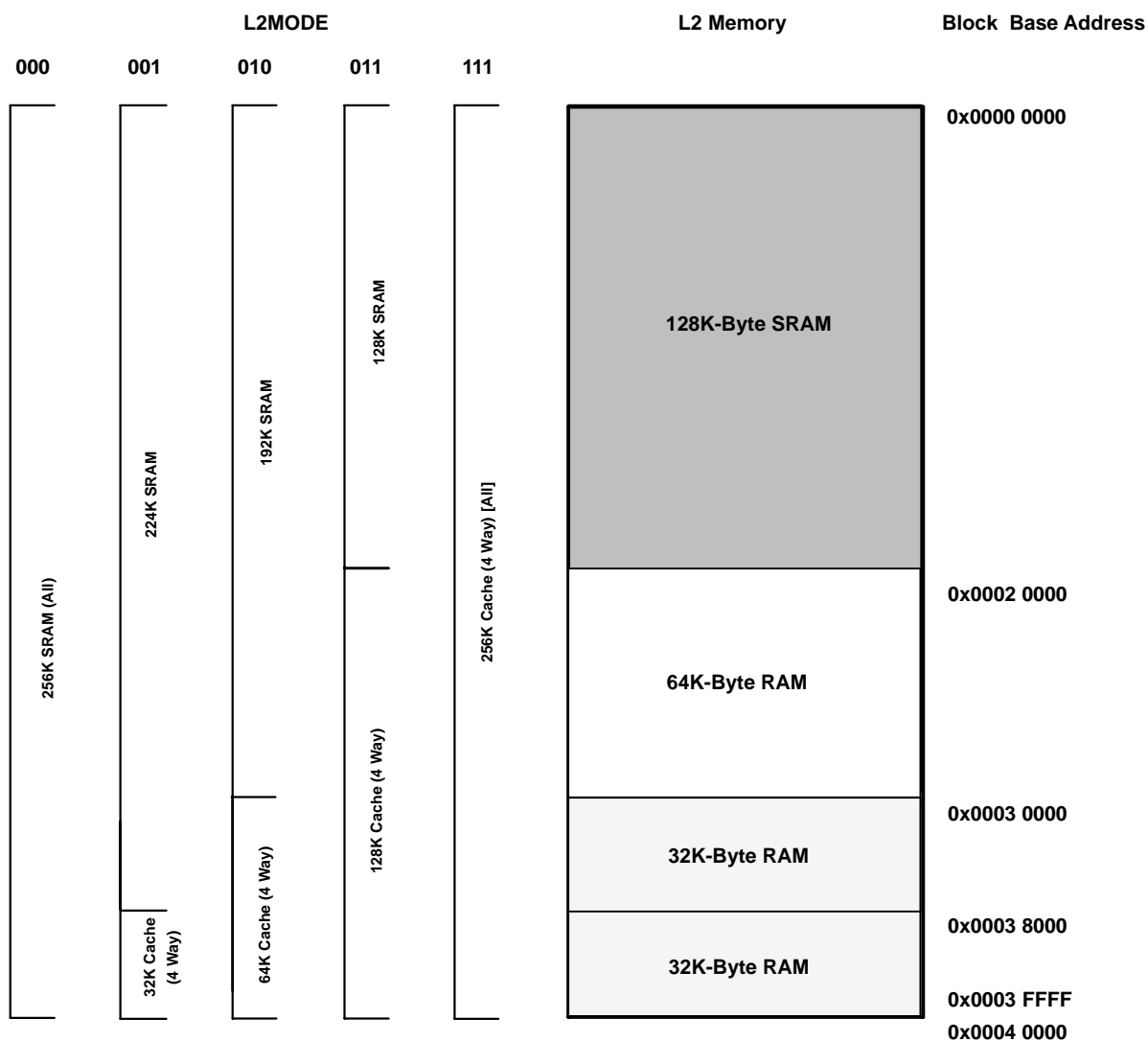
Figure 1−5. TMS320C6412 L2 Architecture Memory Configuration

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1.7 Peripheral Register Descriptions

Table 1−3 through Table 1−23 identify the peripheral registers for the C6412 device by their register names, acronyms, and hex address or hex address range. For more detailed information on the register contents, bit names and their descriptions, see the specific peripheral reference guide listed in the *TMS320C6000 DSP Peripherals Overview Reference Guide* (literature number SPRU190).

Table 1−3. EMIFA Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
0180 0000	GBLCTL	EMIFA global control	
0180 0004	CECTL1	EMIFA CE1 space control	
0180 0008	CECTL0	EMIFA CE0 space control	
0180 000C	–	Reserved	
0180 0010	CECTL2	EMIFA CE2 space control	
0180 0014	CECTL3	EMIFA CE3 space control	
0180 0018	SDCTL	EMIFA SDRAM control	
0180 001C	SDTIM	EMIFA SDRAM refresh control	
0180 0020	SDEXT	EMIFA SDRAM extension	
0180 0024 – 0180 003C	–	Reserved	
0180 0040	PDTCTL	Peripheral device transfer (PDT) control	
0180 0044	CESEC1	EMIFA CE1 space secondary control	
0180 0048	CESEC0	EMIFA CE0 space secondary control	
0180 004C	–	Reserved	
0180 0050	CESEC2	EMIFA CE2 space secondary control	
0180 0054	CESEC3	EMIFA CE3 space secondary control	
0180 0058 – 0183 FFFF	–	Reserved	

Table 1−4. L2 Cache Registers (C64x)

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
0184 0000	CCFG	Cache configuration register	
0184 0004 – 0184 0FFC	–	Reserved	
0184 1000	EDMAWEIGHT	L2 EDMA access control register	
0184 1004 – 0184 1FFC	–	Reserved	
0184 2000	L2ALLOC0	L2 allocation register 0	
0184 2004	L2ALLOC1	L2 allocation register 1	
0184 2008	L2ALLOC2	L2 allocation register 2	
0184 200C	L2ALLOC3	L2 allocation register 3	
0184 2010 – 0184 3FFC	–	Reserved	
0184 4000	L2WBAR	L2 writeback base address register	
0184 4004	L2WWC	L2 writeback word count register	
0184 4010	L2WIBAR	L2 writeback invalidate base address register	
0184 4014	L2WIWC	L2 writeback invalidate word count register	
0184 4018	L2IBAR	L2 invalidate base address register	
0184 401C	L2IWC	L2 invalidate word count register	
0184 4020	L1PIBAR	L1P invalidate base address register	
0184 4024	L1PIWC	L1P invalidate word count register	
0184 4030	L1DWIBAR	L1D writeback invalidate base address register	

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Table 1–4. L2 Cache Registers (C64x) (Continued)

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
0184 4034	L1DWIWC	L1D writeback invalidate word count register	
0184 4038 – 0184 4044	–	Reserved	
0184 4048	L1DIBAR	L1D invalidate base address register	
0184 404C	L1DIWC	L1D invalidate word count register	
0184 4050 – 0184 4FFC	–	Reserved	
0184 5000	L2WB	L2 writeback all register	
0184 5004	L2WBINV	L2 writeback invalidate all register	
0184 5008 – 0184 7FFC	–	Reserved	
0184 8000 –0184 81FC	MAR0 to MAR127	Reserved	
0184 8200	MAR128	Controls EMIFA CE0 range 8000 0000 – 80FF FFFF	
0184 8204	MAR129	Controls EMIFA CE0 range 8100 0000 – 81FF FFFF	
0184 8208	MAR130	Controls EMIFA CE0 range 8200 0000 – 82FF FFFF	
0184 820C	MAR131	Controls EMIFA CE0 range 8300 0000 – 83FF FFFF	
0184 8210	MAR132	Controls EMIFA CE0 range 8400 0000 – 84FF FFFF	
0184 8214	MAR133	Controls EMIFA CE0 range 8500 0000 – 85FF FFFF	
0184 8218	MAR134	Controls EMIFA CE0 range 8600 0000 – 86FF FFFF	
0184 821C	MAR135	Controls EMIFA CE0 range 8700 0000 – 87FF FFFF	
0184 8220	MAR136	Controls EMIFA CE0 range 8800 0000 – 88FF FFFF	
0184 8224	MAR137	Controls EMIFA CE0 range 8900 0000 – 89FF FFFF	
0184 8228	MAR138	Controls EMIFA CE0 range 8A00 0000 – 8AFF FFFF	
0184 822C	MAR139	Controls EMIFA CE0 range 8B00 0000 – 8BFF FFFF	
0184 8230	MAR140	Controls EMIFA CE0 range 8C00 0000 – 8CFF FFFF	
0184 8234	MAR141	Controls EMIFA CE0 range 8D00 0000 – 8DFF FFFF	
0184 8238	MAR142	Controls EMIFA CE0 range 8E00 0000 – 8EFF FFFF	
0184 823C	MAR143	Controls EMIFA CE0 range 8F00 0000 – 8FFF FFFF	
0184 8240	MAR144	Controls EMIFA CE1 range 9000 0000 – 90FF FFFF	
0184 8244	MAR145	Controls EMIFA CE1 range 9100 0000 – 91FF FFFF	
0184 8248	MAR146	Controls EMIFA CE1 range 9200 0000 – 92FF FFFF	
0184 824C	MAR147	Controls EMIFA CE1 range 9300 0000 – 93FF FFFF	
0184 8250	MAR148	Controls EMIFA CE1 range 9400 0000 – 94FF FFFF	
0184 8254	MAR149	Controls EMIFA CE1 range 9500 0000 – 95FF FFFF	
0184 8258	MAR150	Controls EMIFA CE1 range 9600 0000 – 96FF FFFF	
0184 825C	MAR151	Controls EMIFA CE1 range 9700 0000 – 97FF FFFF	
0184 8260	MAR152	Controls EMIFA CE1 range 9800 0000 – 98FF FFFF	
0184 8264	MAR153	Controls EMIFA CE1 range 9900 0000 – 99FF FFFF	
0184 8268	MAR154	Controls EMIFA CE1 range 9A00 0000 – 9AFF FFFF	
0184 826C	MAR155	Controls EMIFA CE1 range 9B00 0000 – 9BFF FFFF	
0184 8270	MAR156	Controls EMIFA CE1 range 9C00 0000 – 9CFF FFFF	
0184 8274	MAR157	Controls EMIFA CE1 range 9D00 0000 – 9DFF FFFF	
0184 8278	MAR158	Controls EMIFA CE1 range 9E00 0000 – 9EFF FFFF	
0184 827C	MAR159	Controls EMIFA CE1 range 9F00 0000 – 9FFF FFFF	

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Table 1–4. L2 Cache Registers (C64x) (Continued)

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
0184 8280	MAR160	Controls EMIFA CE2 range A000 0000 – A0FF FFFF	
0184 8284	MAR161	Controls EMIFA CE2 range A100 0000 – A1FF FFFF	
0184 8288	MAR162	Controls EMIFA CE2 range A200 0000 – A2FF FFFF	
0184 828C	MAR163	Controls EMIFA CE2 range A300 0000 – A3FF FFFF	
0184 8290	MAR164	Controls EMIFA CE2 range A400 0000 – A4FF FFFF	
0184 8294	MAR165	Controls EMIFA CE2 range A500 0000 – A5FF FFFF	
0184 8298	MAR166	Controls EMIFA CE2 range A600 0000 – A6FF FFFF	
0184 829C	MAR167	Controls EMIFA CE2 range A700 0000 – A7FF FFFF	
0184 82A0	MAR168	Controls EMIFA CE2 range A800 0000 – A8FF FFFF	
0184 82A4	MAR169	Controls EMIFA CE2 range A900 0000 – A9FF FFFF	
0184 82A8	MAR170	Controls EMIFA CE2 range AA00 0000 – AAFF FFFF	
0184 82AC	MAR171	Controls EMIFA CE2 range AB00 0000 – ABFF FFFF	
0184 82B0	MAR172	Controls EMIFA CE2 range AC00 0000 – ACFF FFFF	
0184 82B4	MAR173	Controls EMIFA CE2 range AD00 0000 – ADFF FFFF	
0184 82B8	MAR174	Controls EMIFA CE2 range AE00 0000 – AEFF FFFF	
0184 82BC	MAR175	Controls EMIFA CE2 range AF00 0000 – AFFF FFFF	
0184 82C0	MAR176	Controls EMIFA CE3 range B000 0000 – B0FF FFFF	
0184 82C4	MAR177	Controls EMIFA CE3 range B100 0000 – B1FF FFFF	
0184 82C8	MAR178	Controls EMIFA CE3 range B200 0000 – B2FF FFFF	
0184 82CC	MAR179	Controls EMIFA CE3 range B300 0000 – B3FF FFFF	
0184 82D0	MAR180	Controls EMIFA CE3 range B400 0000 – B4FF FFFF	
0184 82D4	MAR181	Controls EMIFA CE3 range B500 0000 – B5FF FFFF	
0184 82D8	MAR182	Controls EMIFA CE3 range B600 0000 – B6FF FFFF	
0184 82DC	MAR183	Controls EMIFA CE3 range B700 0000 – B7FF FFFF	
0184 82E0	MAR184	Controls EMIFA CE3 range B800 0000 – B8FF FFFF	
0184 82E4	MAR185	Controls EMIFA CE3 range B900 0000 – B9FF FFFF	
0184 82E8	MAR186	Controls EMIFA CE3 range BA00 0000 – BAFF FFFF	
0184 82EC	MAR187	Controls EMIFA CE3 range BB00 0000 – BBFF FFFF	
0184 82F0	MAR188	Controls EMIFA CE3 range BC00 0000 – BCFF FFFF	
0184 82F4	MAR189	Controls EMIFA CE3 range BD00 0000 – BDFF FFFF	
0184 82F8	MAR190	Controls EMIFA CE3 range BE00 0000 – BEFF FFFF	
0184 82FC	MAR191	Controls EMIFA CE3 range BF00 0000 – BFFF FFFF	
0184 8300 –0184 83FC	MAR192 to MAR255	Reserved	
0184 8400 –0187 FFFF	–	Reserved	

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Table 1–5. Quick DMA (QDMA) and Pseudo Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME
0200 0000	QOPT	QDMA options parameter register
0200 0004	QSRC	QDMA source address register
0200 0008	QCNT	QDMA frame count register
0200 000C	QDST	QDMA destination address register
0200 0010	QIDX	QDMA index register
0200 0014 – 0200 001C		Reserved
0200 0020	QSOPT	QDMA pseudo options register
0200 0024	QSSRC	QDMA psuedo source address register
0200 0028	QSCNT	QDMA psuedo frame count register
0200 002C	QSDST	QDMA destination address register
0200 0030	QSIDX	QDMA psuedo index register

Table 1–6. EDMA Registers (C64x)

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME
01A0 0800 – 01A0 FF98	–	Reserved
01A0 FF9C	EPRH	Event polarity high register
01A0 FFA4	CIPRH	Channel interrupt pending high register
01A0 FFA8	CIERH	Channel interrupt enable high register
01A0 FFAC	CCERH	Channel chain enable high register
01A0 FFB0	ERH	Event high register
01A0 FFB4	EERH	Event enable high register
01A0 FFB8	ECRH	Event clear high register
01A0 FFBC	ESRH	Event set high register
01A0 FFC0	PQAR0	Priority queue allocation register 0
01A0 FFC4	PQAR1	Priority queue allocation register 1
01A0 FFC8	PQAR2	Priority queue allocation register 2
01A0 FFCC	PQAR3	Priority queue allocation register 3
01A0 FFDC	EPRL	Event polarity low register
01A0 FFE0	PQSR	Priority queue status register
01A0 FFE4	CIPRL	Channel interrupt pending low register
01A0 FFE8	CIERL	Channel interrupt enable low register
01A0 FFEC	CCERL	Channel chain enable low register
01A0 FFF0	ERL	Event low register
01A0 FFF4	EERL	Event enable low register
01A0 FFF8	ECRL	Event clear low register
01A0 FFFC	ESRL	Event set low register
01A1 0000 – 01A3 FFFF	–	Reserved

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Table 1–7. EDMA Parameter RAM (C64x)[†]

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
01A0 0000 – 01A0 0017	–	Parameters for Event 0 (6 words)	Parameters for Event 0 (6 words) or Reload/Link Parameters for other Event
01A0 0018 – 01A0 002F	–	Parameters for Event 1 (6 words)	
01A0 0030 – 01A0 0047	–	Parameters for Event 2 (6 words)	
01A0 0048 – 01A0 005F	–	Parameters for Event 3 (6 words)	
01A0 0060 – 01A0 0077	–	Parameters for Event 4 (6 words)	
01A0 0078 – 01A0 008F	–	Parameters for Event 5 (6 words)	
01A0 0090 – 01A0 00A7	–	Parameters for Event 6 (6 words)	
01A0 00A8 – 01A0 00BF	–	Parameters for Event 7 (6 words)	
01A0 00C0 – 01A0 00D7	–	Parameters for Event 8 (6 words)	
01A0 00D8 – 01A0 00EF	–	Parameters for Event 9 (6 words)	
01A0 00F0 – 01A0 00107	–	Parameters for Event 10 (6 words)	
01A0 0108 – 01A0 011F	–	Parameters for Event 11 (6 words)	
01A0 0120 – 01A0 0137	–	Parameters for Event 12 (6 words)	
01A0 0138 – 01A0 014F	–	Parameters for Event 13 (6 words)	
01A0 0150 – 01A0 0167	–	Parameters for Event 14 (6 words)	
01A0 0168 – 01A0 017F	–	Parameters for Event 15 (6 words)	
01A0 0180 – 01A0 0197	–	Parameters for Event 16 (6 words)	
01A0 0198 – 01A0 01AF	–	Parameters for Event 17 (6 words)	
...		...	
01A0 05D0 – 01A0 05E7	–	Parameters for Event 62 (6 words)	
01A0 05E8 – 01A0 05FF	–	Parameters for Event 63 (6 words)	
01A0 0600 – 01A0 0617	–	Reload/link parameters for Event 0 (6 words)	Reload/Link Parameters for other Event 0–15
01A0 0618 – 01A0 062F	–	Reload/link parameters for Event 1 (6 words)	
...		...	
01A0 07E0 – 01A0 07F7	–	Reload/link parameters for Event 20 (6 words)	
01A0 07F8 – 01A0 080F	–	Reload/link parameters for Event 21 (6 words)	
01A0 0810 – 01A0 0827	–	Reload/link parameters for Event 22 (6 words)	
...		...	
01A0 13C8 – 01A0 13DF	–	Reload/link parameters for Event 147 (6 words)	
01A0 13E0 – 01A0 13F7	–	Reload/link parameters for Event 148 (6 words)	
01A0 13F8 – 01A0 13FF	–	Scratch pad area (2 words)	
01A0 1400 – 01A3 FFFF	–	Reserved	

[†] The C6412 device has 213 EDMA parameters total: 64-Event/Reload channels and 149-Reload only parameter sets [six (6) words each] that can be used to reload/link EDMA transfers.

Table 1–8. Interrupt Selector Registers (C64x)

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
019C 0000	MUXH	Interrupt multiplexer high	Selects which interrupts drive CPU interrupts 10–15 (INT10–INT15)
019C 0004	MUXL	Interrupt multiplexer low	Selects which interrupts drive CPU interrupts 4–9 (INT04–INT09)
019C 0008	EXTPOL	External interrupt polarity	Sets the polarity of the external interrupts (EXT_INT4–EXT_INT7)
019C 000C – 019F FFFF	–	Reserved	

Table 1–9. Ethernet MAC (EMAC) Control Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME
01C8 0000	TXIDVER	Transmit Identification and Version Register
01C8 0004	TXCONTROL	Transmit Control Register
01C8 0008	TXTEARDOWN	Transmit Teardown Register
01C8 000C	–	Reserved
01C8 0010	RXIDVER	Receive Identification and Version Register
01C8 0014	RXCONTROL	Receive Control Register
01C8 0018	RXTEARDOWN	Receive Teardown Register (RXTDNCH field only supports writes of 0.)
01C8 001C – 01C8 00FF	–	Reserved
01C8 0100	RXMBPENABLE	Receive Multicast/Broadcast/Promiscuous Channel Enable Register (The RXQOSEN field is reserved and only supports writes of 0. The PROMCH, BROADCH, and MUCTCH bit fields only support writes of 0.)
01C8 0104	RXUNICASTSET	Receive Unicast Set Register (Bits 7–1 are reserved and only support writes of 0.)
01C8 0108	RXUNICASTCLEAR	Receive Unicast Clear Register (Bits 7–1 are reserved and only support writes of 0.)
01C8 010C	RXMAXLEN	Receive Maximum Length Register
01C8 0110	RXBUFFEROFFSET	Receive Buffer Offset Register
01C8 0114	RXFILTERLOWTHRESH	Receive Filter Low Priority Packets Threshold Register
01C8 0118 – 01C8 011F	–	Reserved
01C8 0120	RX0FLOWTHRESH	Receive Channel 0 Flow Control Threshold Register
01C8 0124	RX1FLOWTHRESH	Reserved. Do not write.
01C8 0128	RX2FLOWTHRESH	
01C8 012C	RX3FLOWTHRESH	
01C8 0130	RX4FLOWTHRESH	
01C8 0134	RX5FLOWTHRESH	
01C8 0138	RX6FLOWTHRESH	
01C8 013C	RX7FLOWTHRESH	
01C8 0140	RX0FREEBUFFER	Receive Channel 0 Free Buffer Count Register
01C8 0144	RX1FREEBUFFER	Reserved. Do not write.
01C8 0148	RX2FREEBUFFER	
01C8 014C	RX3FREEBUFFER	
01C8 0150	RX4FREEBUFFER	
01C8 0154	RX5FREEBUFFER	
01C8 0158	RX6FREEBUFFER	
01C8 015C	RX7FREEBUFFER	

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Table 1−9. Ethernet MAC (EMAC) Control Registers (Continued)

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME
01C8 0160	MACCONTROL	MAC Control Register
01C8 0164	MACSTATUS	MAC Status Register (RXQOSACT field is reserved.)
01C8 0168 – 01C8 016C	–	Reserved
01C8 0170	TXINTSTATRAW	Transmit Interrupt Status (Unmasked) Register
01C8 0174	TXINTSTATMASKED	Transmit Interrupt Status (Masked) Register
01C8 0178	TXINTMASKSET	Transmit Interrupt Mask Set Register
01C8 017C	TXINTMASKCLEAR	Transmit Interrupt Mask Clear Register
01C8 0180	MACINVECTOR	MAC Input Vector Register
01C8 0184 – 01C8 018F	–	Reserved
01C8 0190	RXINTSTATRAW	Receive Interrupt Status (Unmasked) Register (Bits 7−1 are reserved.)
01C8 0194	RXINTSTATMASKED	Receive Interrupt Status (Masked) Register (Bits 7−1 are reserved.)
01C8 0198	RXINTMASKSET	Receive Interrupt Mask Set Register (Bits 7−1 are reserved and only support writes of 0.)
01C8 019C	RXINTMASKCLEAR	Receive Interrupt Mask Clear Register (Bits 7−1 are reserved and only support writes of 0.)
01C8 01A0	MACINTSTATRAW	MAC Interrupt Status (Unmasked) Register
01C8 01A4	MACINTSTATMASKED	MAC Interrupt Status (Masked) Register
01C8 01A8	MACINTMASKSET	MAC Interrupt Mask Set Register
01C8 01AC	MACINTMASKCLEAR	MAC Interrupt Mask Clear Register
01C8 01B0	MACADDRL0	MAC Address Channel 0 Lower Byte Register
01C8 01B4	MACADDRL1	
01C8 01B8	MACADDRL2	
01C8 01BC	MACADDRL3	
01C8 01C0	MACADDRL4	Reserved. Do not write.
01C8 01C4	MACADDRL5	
01C8 01C8	MACADDRL6	
01C8 01CC	MACADDRL7	
01C8 01D0	MACADDRM	MAC Address Middle Byte Register
01C8 01D4	MACADDRH	MAC Address High Bytes Register
01C8 01D8	MACHASH1	MAC Address Hash 1 Register
01C8 01DC	MACHASH2	MAC Address Hash 2 Register
01C8 01E0	BOFFTEST	Backoff Test Register
01C8 01E4	TPACETEST	Transmit Pacing Test Register
01C8 01E8	RXPAUSE	Receive Pause Timer Register
01C8 01EC	TXPAUSE	Transmit Pause Timer Register
01C8 01F0 – 01C8 01FF	–	Reserved
01C8 0200 – 01C8 05FF	(see Table 1−10)	EMAC Statistics Registers
01C8 0600	TX0HDP	Transmit Channel 0 DMA Head Descriptor Pointer Register
01C8 0604	TX1HDP	Transmit Channel 1 DMA Head Descriptor Pointer Register
01C8 0608	TX2HDP	Transmit Channel 2 DMA Head Descriptor Pointer Register
01C8 060C	TX3HDP	Transmit Channel 3 DMA Head Descriptor Pointer Register
01C8 0610	TX4HDP	Transmit Channel 4 DMA Head Descriptor Pointer Register

Table 1–9. Ethernet MAC (EMAC) Control Registers (Continued)

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME
01C8 0614	TX5HDP	Transmit Channel 5 DMA Head Descriptor Pointer Register
01C8 0618	TX6HDP	Transmit Channel 6 DMA Head Descriptor Pointer Register
01C8 061C	TX7HDP	Transmit Channel 7 DMA Head Descriptor Pointer Register
01C8 0620	RX0HDP	Receive Channel 0 DMA Head Descriptor Pointer Register
01C8 0624	RX1HDP	Reserved. Do not write.
01C8 0628	RX2HDP	
01C8 062C	RX3HDP	
01C8 0630	RX4HDP	
01C8 0634	RX5HDP	
01C8 0638	RX6HDP	
01C8 063C	RX7HDP	
01C8 0640	TX0INTACK	Transmit Channel 0 Interrupt Acknowledge Register
01C8 0644	TX1INTACK	Transmit Channel 1 Interrupt Acknowledge Register
01C8 0648	TX2INTACK	Transmit Channel 2 Interrupt Acknowledge Register
01C8 064C	TX3INTACK	Transmit Channel 3 Interrupt Acknowledge Register
01C8 0650	TX4INTACK	Transmit Channel 4 Interrupt Acknowledge Register
01C8 0654	TX5INTACK	Transmit Channel 5 Interrupt Acknowledge Register
01C8 0658	TX6INTACK	Transmit Channel 6 Interrupt Acknowledge Register
01C8 065C	TX7INTACK	Transmit Channel 7 Interrupt Acknowledge Register
01C8 0660	RX0INTACK	Receive Channel 0 Interrupt Acknowledge Register
01C8 0664	RX1INTACK	Reserved. Do not write.
01C8 0668	RX2INTACK	
01C8 066C	RX3INTACK	
01C8 0670	RX4INTACK	
01C8 0674	RX5INTACK	
01C8 0678	RX6INTACK	
01C8 067C	RX7INTACK	
01C8 0680 – 01C8 0FFF	–	Reserved

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Table 1–10. EMAC Statistics Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	
01C8 0200	RXGOODFRAMES	Good Receive Frames Register	
01C8 0204	RXBCASTFRAMES	Broadcast Receive Frames Register	
01C8 0208	RXMCASTFRAMES	Multicast Receive Frames Register	
01C8 020C	RXPAUSEFRAMES	Pause Receive Frames Register	
01C8 0210	RXCRCERRORS	Receive CRC Errors Register	
01C8 0214	RXALIGNCODEERRORS	Receive Alignment/Code Errors Register	
01C8 0218	RXOVERSIZED	Receive Oversized Frames Register	
01C8 021C	RXJABBER	Receive Jabber Frames Register	
01C8 0220	RXUNDERSIZED	Receive Undersized Frames Register	
01C8 0224	RXFRAGMENTS	Receive Frame Fragments Register	
01C8 0228	RXFILTERED	Filtered Receive Frames Register	
01C8 022C	RXQOSFILTERED	Reserved	
01C8 0230	RXOCTETS	Receive Octet Frames Register	
01C8 0234	TXGOODFRAMES	Good Transmit Frames Register	
01C8 0238	TXBCASTFRAMES	Broadcast Transmit Frames Register	
01C8 023C	TXMCASTFRAMES	Multicast Transmit Frames Register	
01C8 0240	TXPAUSEFRAMES	Pause Transmit Frames Register	
01C8 0244	TXDEFERRED	Deferred Transmit Frames Register	
01C8 0248	TXCOLLISION	Collision Register	
01C8 024C	TXSINGLECOLL	Single Collision Transmit Frames Register	
01C8 0250	TXMULTICOLL	Multiple Collision Transmit Frames Register	
01C8 0254	TXEXCESSIVECOLL	Excessive Collisions Register	
01C8 0258	TXLATECOLL	Late Collisions Register	
01C8 025C	TXUNDERRUN	Transmit Underrun Register	
01C8 0260	TXCARRIERSLOSS	Transmit Carrier Sense Errors Register	
01C8 0264	TXOCTETS	Transmit Octet Frames Register	
01C8 0268	FRAME64	Transmit and Receive 64 Octet Frames Register	
01C8 026C	FRAME65T127	Transmit and Receive 65 to 127 Octet Frames Register	
01C8 0270	FRAME128T255	Transmit and Receive 128 to 255 Octet Frames Register	
01C8 0274	FRAME256T511	Transmit and Receive 256 to 511 Octet Frames Register	
01C8 0278	FRAME512T1023	Transmit and Receive 512 to 1023 Octet Frames Register	
01C8 027C	FRAME1024TUP	Transmit and Receive 1024 or Above Octet Frames Register	
01C8 0280	NETOCTETS	Network Octet Frames Register	
01C8 0284	RXSOFOVERRUNS	Receive Start of Frame Overruns Register	
01C8 0288	RXMOFOVERRUNS	Receive Middle of Frame Overruns Register	
01C8 028C	RXDMAOVERRUNS	Receive DMA Overruns Register	
01C8 0290 – 01C8 05FF	–	Reserved	

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Table 1−11. EMAC Wrapper

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME
01C8 1000 – 01C8 1FFF		EMAC Control Module Descriptor Memory
01C8 2000 – 01C8 2FFF	–	Reserved

Table 1−12. EWRAP Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME
01C8 3000	EWTRCTRL	TR control
01C8 3004	EWCTL	Interrupt control register
01C8 3008	EWINTTCNT	Interrupt timer count
01C8 300C – 01C8 37FF	–	Reserved

Table 1−13. Device Configuration Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
01B3 F000	PERCFG	Peripheral Configuration Register	Enables or disables specific peripherals. This register is also used for power-down of disabled peripherals.
01B3 F004	DEVSTAT	Device Status Register	Read-only. Provides status of the User's device configuration on reset.
01B3 F008	JTAGID	JTAG Identification Register	Read-only. Provides 32-bit JTAG ID of the device.
01B3 F00C – 01B3 F014	–	Reserved	
01B3 F018	PCFGLOCK	Peripheral Configuration Lock Register	
01B3 F01C – 01B3 FFFF	–	Reserved	

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Table 1–14. McBSP 0 Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
018C 0000	DRR0	McBSP0 data receive register via Configuration Bus	The CPU and EDMA controller can only read this register; they cannot write to it.
0x3000 0000 – 0x33FF FFFF	DRR0	McBSP0 data receive register via Peripheral Bus	
018C 0004	DXR0	McBSP0 data transmit register via Configuration Bus	
0x3000 0000 – 0x33FF FFFF	DXR0	McBSP0 data transmit register via Peripheral Bus	
018C 0008	SPCR0	McBSP0 serial port control register	
018C 000C	RCR0	McBSP0 receive control register	
018C 0010	XCR0	McBSP0 transmit control register	
018C 0014	SRGR0	McBSP0 sample rate generator register	
018C 0018	MCR0	McBSP0 multichannel control register	
018C 001C	RCERE00	McBSP0 enhanced receive channel enable register 0	
018C 0020	XCERE00	McBSP0 enhanced transmit channel enable register 0	
018C 0024	PCR0	McBSP0 pin control register	
018C 0028	RCERE10	McBSP0 enhanced receive channel enable register 1	
018C 002C	XCERE10	McBSP0 enhanced transmit channel enable register 1	
018C 0030	RCERE20	McBSP0 enhanced receive channel enable register 2	
018C 0034	XCERE20	McBSP0 enhanced transmit channel enable register 2	
018C 0038	RCERE30	McBSP0 enhanced receive channel enable register 3	
018C 003C	XCERE30	McBSP0 enhanced transmit channel enable register 3	
018C 0040 – 018F FFFF	–	Reserved	

Table 1–15. McBSP 1 Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
0190 0000	DRR1	McBSP1 data receive register via Configuration Bus	The CPU and EDMA controller can only read this register; they cannot write to it.
0x3400 0000 – 0x37FF FFFF	DRR1	McBSP1 data receive register via peripheral bus	
0190 0004	DXR1	McBSP1 data transmit register via configuration bus	
0x3400 0000 – 0x37FF FFFF	DXR1	McBSP1 data transmit register via peripheral bus	
0190 0008	SPCR1	McBSP1 serial port control register	
0190 000C	RCR1	McBSP1 receive control register	
0190 0010	XCR1	McBSP1 transmit control register	
0190 0014	SRGR1	McBSP1 sample rate generator register	
0190 0018	MCR1	McBSP1 multichannel control register	
0190 001C	RCERE01	McBSP1 enhanced receive channel enable register 0	
0190 0020	XCERE01	McBSP1 enhanced transmit channel enable register 0	
0190 0024	PCR1	McBSP1 pin control register	
0190 0028	RCERE11	McBSP1 enhanced receive channel enable register 1	
0190 002C	XCERE11	McBSP1 enhanced transmit channel enable register 1	
0190 0030	RCERE21	McBSP1 enhanced receive channel enable register 2	
0190 0034	XCERE21	McBSP1 enhanced transmit channel enable register 2	
0190 0038	RCERE31	McBSP1 enhanced receive channel enable register 3	
0190 003C	XCERE31	McBSP1 enhanced transmit channel enable register 3	
0190 0040 – 0193 FFFF	–	Reserved	

Table 1–16. Timer 0 Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
0194 0000	CTL0	Timer 0 control register	Determines the operating mode of the timer, monitors the timer status, and controls the function of the TOUT pin.
0194 0004	PRD0	Timer 0 period register	Contains the number of timer input clock cycles to count. This number controls the TSTAT signal frequency.
0194 0008	CNT0	Timer 0 counter register	Contains the current value of the incrementing counter.
0194 000C – 0197 FFFF	–	Reserved	

Table 1–17. Timer 1 Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
0198 0000	CTL1	Timer 1 control register	Determines the operating mode of the timer, monitors the timer status, and controls the function of the TOUT pin.
0198 0004	PRD1	Timer 1 period register	Contains the number of timer input clock cycles to count. This number controls the TSTAT signal frequency.
0198 0008	CNT1	Timer 1 counter register	Contains the current value of the incrementing counter.
0198 000C – 019B FFFF	–	Reserved	

Table 1–18. Timer 2 Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
01AC 0000	CTL2	Timer 2 control register	Determines the operating mode of the timer, monitors the timer status.
01AC 0004	PRD2	Timer 2 period register	Contains the number of timer input clock cycles to count. This number controls the TSTAT signal frequency.
01AC 0008	CNT2	Timer 2 counter register	Contains the current value of the incrementing counter.
01AC 000C – 01AF FFFF	–	Reserved	

Table 1–19. HPI Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
–	HPID	HPI data register	Host read/write access only
0188 0000	HPIC	HPI control register	HPIC has both Host/CPU read/write access
0188 0004	HPIA (HPIAW)[†]	HPI address register (Write)	HPIA has both Host/CPU read/write access
0188 0008	HPIA (HPIAR)[†]	HPI address register (Read)	
0188 000C – 0189 FFFF	–	Reserved	
018A 0000	HPI_TRCTL	HPI transfer request control register	
018A 0004 – 018B FFFF	–	Reserved	

[†] Host access to the HPIA register updates both the HPIAW and HPIAR registers. The CPU can access HPIAW and HPIAR independently.

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Table 1–20. GP0 Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME
01B0 0000	GPEN	GP0 enable register
01B0 0004	GPDIR	GP0 direction register
01B0 0008	GPVAL	GP0 value register
01B0 000C	–	Reserved
01B0 0010	GPDH	GP0 delta high register
01B0 0014	GPHM	GP0 high mask register
01B0 0018	GPDL	GP0 delta low register
01B0 001C	GPLM	GP0 low mask register
01B0 0020	GPGC	GP0 global control register
01B0 0024	GPPOL	GP0 interrupt polarity register
01B0 0028 – 01B3 EFFF	–	Reserved

Table 1–21. PCI Peripheral Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME
01C0 0000	RSTSRC	DSP Reset source/status register
01C0 0004	–	Reserved
01C0 0008	PCIIS	PCI interrupt source register
01C0 000C	PCIIEN	PCI interrupt enable register
01C0 0010	DSPMA	DSP master address register
01C0 0014	PCIMA	PCI master address register
01C0 0018	PCIMC	PCI master control register
01C0 001C	CDSPA	Current DSP address register
01C0 0020	CPCIA	Current PCI address register
01C0 0024	CCNT	Current byte count register
01C0 0028	–	Reserved
01C0 002C – 01C1 FFEF	–	Reserved
0x01C1 FFF0	HSR	Host status register
0x01C1 FFF4	HDCR	Host-to-DSP control register
0x01C1 FFF8	DSPP	DSP page register
0x01C1 FFFC	–	Reserved
01C2 0000	EEADD	EEPROM address register
01C2 0004	EEDAT	EEPROM data register
01C2 0008	EECTL	EEPROM control register
01C2 000C – 01C2 FFFF	–	Reserved
01C3 0000	PCI_TRCTL	PCI transfer request control register
01C3 0004 – 01C3 FFFF	–	Reserved

Table 1–22. MDIO Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME
01C8 3800	VERSION	MDIO Version Register
01C8 3804	CONTROL	MDIO Control Register
01C8 3808	ALIVE	MDIO PHY Alive Indication Register
01C8 380C	LINK	MDIO PHY Link Status Register
01C8 3810	LINKINTRAW	MDIO Link Status Change Interrupt Register (MAC1 field is reserved and only supports writes of 0.)
01C8 3814	LINKINTMASKED	MDIO Link Status Change Interrupt (Masked) Register (MAC1 field is reserved and only supports writes of 0.)
01C8 3818 – 01C8 381F	–	Reserved
01C8 3820	USERINTRAW	MDIO User Command Complete Interrupt Register (MAC1 field is reserved and only supports writes of 0.)
01C8 3824	USERINTMASKED	MDIO User Command Complete Interrupt (Masked) Register (MAC1 field is reserved and only supports writes of 0.)
01C8 3828	USERINTMASKSET	MDIO User Command Complete Interrupt Mask Set Register (MAC1 field is reserved and only supports writes of 0.)
01C8 382C	USERINTMASKCLEAR	MDIO User Command Complete Interrupt Mask Clear Register (MAC1 field is reserved and only supports writes of 0.)
01C8 3830 – 01C8 387F	–	Reserved
01C8 3880	USERACCESS0	MDIO User Access Register 0
01C8 3884	USERPHYSEL0	MDIO User PHY Select Register 0
01C8 3888	USERACCESS1	Reserved. Do not write.
01C8 388C	USERPHYSEL1	Reserved. Do not write.
01C8 3890 – 01C8 3FFF	–	Reserved

Table 1–23. I2C0 Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME
01B4 0000	I2COAR0	I2C0 own address register
01B4 0004	I2CIER0	I2C0 interrupt enable register
01B4 0008	I2CSTR0	I2C0 interrupt status register
01B4 000C	I2CCLKL0	I2C0 clock low-time divider register
01B4 0010	I2CCLKH0	I2C0 clock high-time divider register
01B4 0014	I2CCNT0	I2C0 data count register
01B4 0018	I2CDRR0	I2C0 data receive register
01B4 001C	I2CSAR0	I2C0 slave address register
01B4 0020	I2CDXR0	I2C0 data transmit register
01B4 0024	I2CMDR0	I2C0 mode register
01B4 0028	I2CISRC0	I2C0 interrupt source register
01B4 002C	–	Reserved
01B4 0030	I2CPSC0	I2C0 prescaler register
01B4 0034	I2CPID10	I2C0 Peripheral Identification register 1 [Value: 0x0000 0101]
01B4 0038	I2CPID20	I2C0 Peripheral Identification register 2 [Value: 0x0000 0005]
01B4 003C – 01B4 3FFF	–	Reserved

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1.8 EDMA Channel Synchronization Events

The C64x EDMA supports up to 64 EDMA channels which service peripheral devices and external memory. Table 1–24 lists the source of C64x EDMA synchronization events associated with each of the programmable EDMA channels. For the C6412 device, the association of an event to a channel is fixed; each of the EDMA channels has one specific event associated with it. These specific events are captured in the EDMA event registers (ERL, ERH) even if the events are disabled by the EDMA event enable registers (EERL, EERH). The priority of each event can be specified independently in the transfer parameters stored in the EDMA parameter RAM. For more detailed information on the EDMA module and how EDMA events are enabled, captured, processed, linked, chained, and cleared, etc., see the *TMS320C6000 DSP Enhanced Direct Memory Access (EDMA) Controller Reference Guide* (literature number SPRU234).

Table 1–24. TMS320C6412 EDMA Channel Synchronization Events[†]

EDMA CHANNEL	EVENT NAME	EVENT DESCRIPTION
0	DSP_INT	HPI/PCI-to-DSP interrupt
1	TINT0	Timer 0 interrupt
2	TINT1	Timer 1 interrupt
3	SD_INTA	EMIFA SDRAM timer interrupt
4	GPINT4/EXT_INT4	GP0 event 4/External interrupt pin 4
5	GPINT5/EXT_INT5	GP0 event 5/External interrupt pin 5
6	GPINT6/EXT_INT6	GP0 event 6/External interrupt pin 6
7	GPINT7/EXT_INT7	GP0 event 7/External interrupt pin 7
8	GPINT0	GP0 event 0
9	GPINT1	GP0 event 1
10	GPINT2	GP0 event 2
11	GPINT3	GP0 event 3
12	XEVT0	McBSP0 transmit event
13	REVT0	McBSP0 receive event
14	XEVT1	McBSP1 transmit event
15	REVT1	McBSP1 receive event
16–18	–	None
19	TINT2	Timer 2 interrupt
20–43	–	None
44	ICREVT0	I2C0 receive event
45	ICXEVT0	I2C0 transmit event
46–47	–	None
48	GPINT8	GP0 event 8
49	GPINT9	GP0 event 9
50	GPINT10	GP0 event 10
51	GPINT11	GP0 event 11
52	GPINT12	GP0 event 12

[†] In addition to the events shown in this table, each of the 64 channels can also be synchronized with the transfer completion or alternate transfer completion events. For more detailed information on EDMA event-transfer chaining, see the *TMS320C6000 DSP Enhanced Direct Memory Access (EDMA) Controller Reference Guide* (literature number SPRU234).

Table 1−24. TMS320C6412 EDMA Channel Synchronization Events† (Continued)

EDMA CHANNEL	EVENT NAME	EVENT DESCRIPTION
53	GPINT13	GP0 event 13
54	GPINT14	GP0 event 14
55	GPINT15	GP0 event 15
56−63	−	None

† In addition to the events shown in this table, each of the 64 channels can also be synchronized with the transfer completion or alternate transfer completion events. For more detailed information on EDMA event-transfer chaining, see the *TMS320C6000 DSP Enhanced Direct Memory Access (EDMA) Controller Reference Guide* (literature number SPRU234).

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1.9   Interrupt Sources and Interrupt Selector

The C64x DSP core supports 16 prioritized interrupts, which are listed in Table 1−25. The highest-priority interrupt is INT_00 (dedicated to RESET) while the lowest-priority interrupt is INT_15. The first four interrupts (INT_00−INT_03) are non-maskable and fixed. The remaining interrupts (INT_04−INT_15) are maskable and default to the interrupt source specified in Table 1−25. The interrupt source for interrupts 4−15 can be programmed by modifying the selector value (binary value) in the corresponding fields of the Interrupt Selector Control registers: MUXH (address 0x019C0000) and MUXL (address 0x019C0004).

Table 1−25. C6412 DSP Interrupts

CPU INTERRUPT NUMBER	INTERRUPT SELECTOR CONTROL REGISTER	SELECTOR VALUE (BINARY)	INTERRUPT EVENT	INTERRUPT SOURCE
INT_00[†]	–	–	RESET	
INT_01[†]	–	–	NMI	
INT_02[†]	–	–	Reserved	Reserved. Do not use.
INT_03[†]	–	–	Reserved	Reserved. Do not use.
INT_04[‡]	MUXL[4:0]	00100	GPINT4/EXT_INT4	GP0 interrupt 4/External interrupt pin 4
INT_05[‡]	MUXL[9:5]	00101	GPINT5/EXT_INT5	GP0 interrupt 5/External interrupt pin 5
INT_06[‡]	MUXL[14:10]	00110	GPINT6/EXT_INT6	GP0 interrupt 6/External interrupt pin 6
INT_07[‡]	MUXL[20:16]	00111	GPINT7/EXT_INT7	GP0 interrupt 7/External interrupt pin 7
INT_08[‡]	MUXL[25:21]	01000	EDMA_INT	EDMA channel (0 through 63) interrupt
INT_09[‡]	MUXL[30:26]	01001	EMU_DTDMA	EMU DTDMA
INT_10[‡]	MUXH[4:0]	00011	SD_INTA	EMIFA SDRAM timer interrupt
INT_11[‡]	MUXH[9:5]	01010	EMU_RTDXRX	EMU real-time data exchange (RTDX) receive
INT_12[‡]	MUXH[14:10]	01011	EMU_RTDXTX	EMU RTDX transmit
INT_13[‡]	MUXH[20:16]	00000	DSP_INT	HPI/PCI-to-DSP interrupt
INT_14[‡]	MUXH[25:21]	00001	TINT0	Timer 0 interrupt
INT_15[‡]	MUXH[30:26]	00010	TINT1	Timer 1 interrupt
–	–	01100	XINT0	McBSP0 transmit interrupt
–	–	01101	RINT0	McBSP0 receive interrupt
–	–	01110	XINT1	McBSP1 transmit interrupt
–	–	01111	RINT1	McBSP1 receive interrupt
–	–	10000	GPINT0	GP0 interrupt 0
–	–	10001	Reserved	Reserved. Do not use.
–	–	10010	Reserved	Reserved. Do not use.
–	–	10011	TINT2	Timer 2 interrupt
–	–	10100	Reserved	Reserved. Do not use.
–	–	10101	Reserved	Reserved. Do not use.
–	–	10110	ICINT0	I2C0 interrupt
–	–	10111	Reserved	Reserved. Do not use.
–	–	11000	EMAC_MDIO_INT	EMAC/MDIO interrupt
–	–	11001 – 11111	Reserved	Reserved. Do not use.

[†] Interrupts INT_00 through INT_03 are non-maskable and fixed.

[‡] Interrupts INT_04 through INT_15 are programmable by modifying the binary selector values in the Interrupt Selector Control registers fields. Table 1−25 shows the default interrupt sources for Interrupts INT_04 through INT_15. For more detailed information on interrupt sources and selection, see the *TMS320C6000 DSP Interrupt Selector Reference Guide* (literature number SPRU646).

1.10   Signal Groups Description

Figure 1–6. CPU and Peripheral Signals
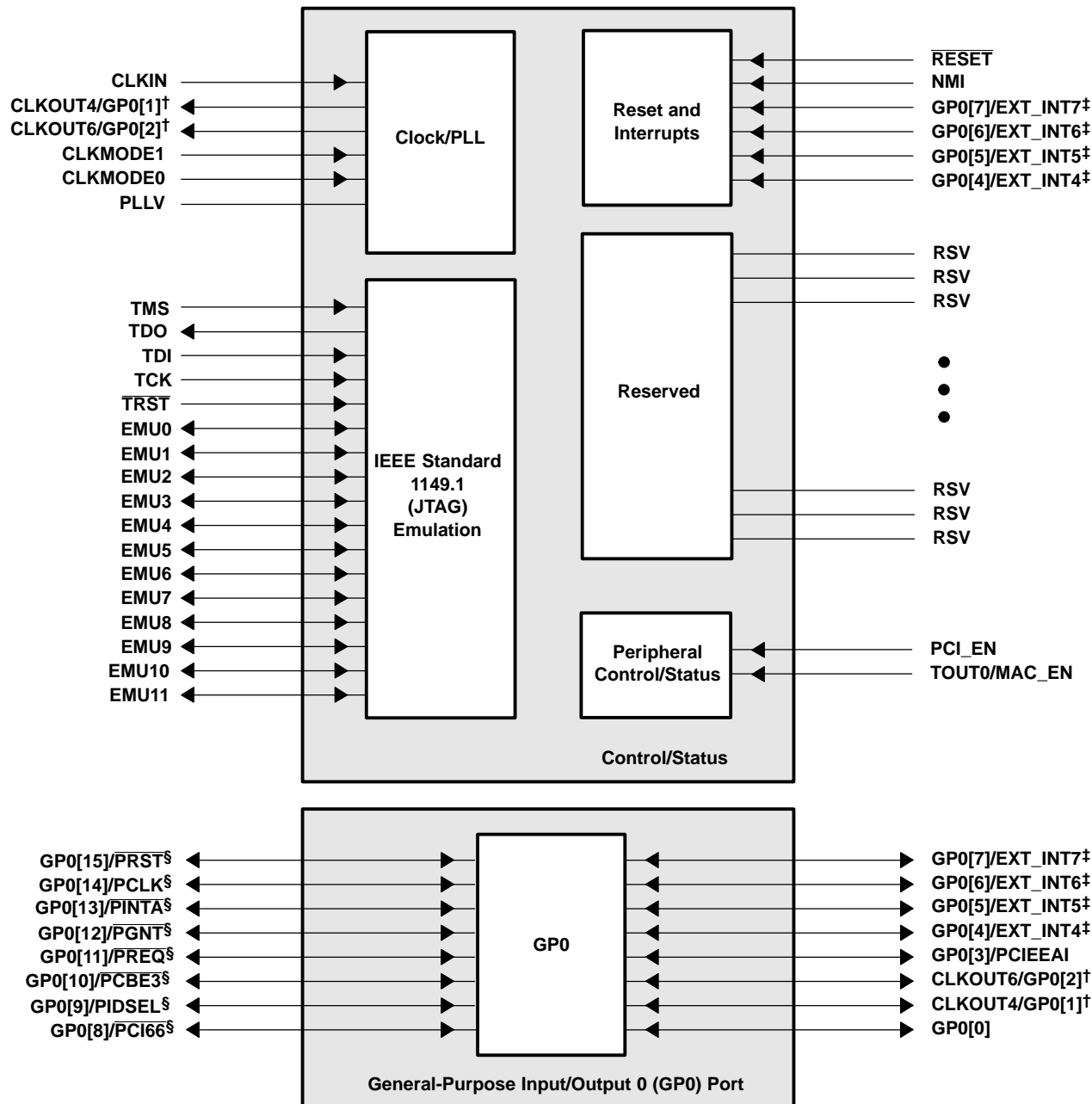
† These pins are muxed with the GP0 pins and by default these signals function as clocks (CLKOUT4 or CLKOUT6). To use these muxed pins as GPIO signals, the appropriate GPIO register bits (GPxEN and GPxDIR) must be properly enabled and configured. For more details, see the Device Configurations section of this data sheet.

‡ These pins are GP0 pins that can also function as external interrupt sources (EXT_INT[7:4]). Default after reset is EXT_INTx or GPIO as input-only.

§ These GP0 pins are muxed with the PCI peripheral pins and by default these signals are set up to no function with both the GPIO and PCI pin functions *disabled*. For more details on these muxed pins, see the Device Configurations section of this data sheet.

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Figure 1–7. Peripheral Signals

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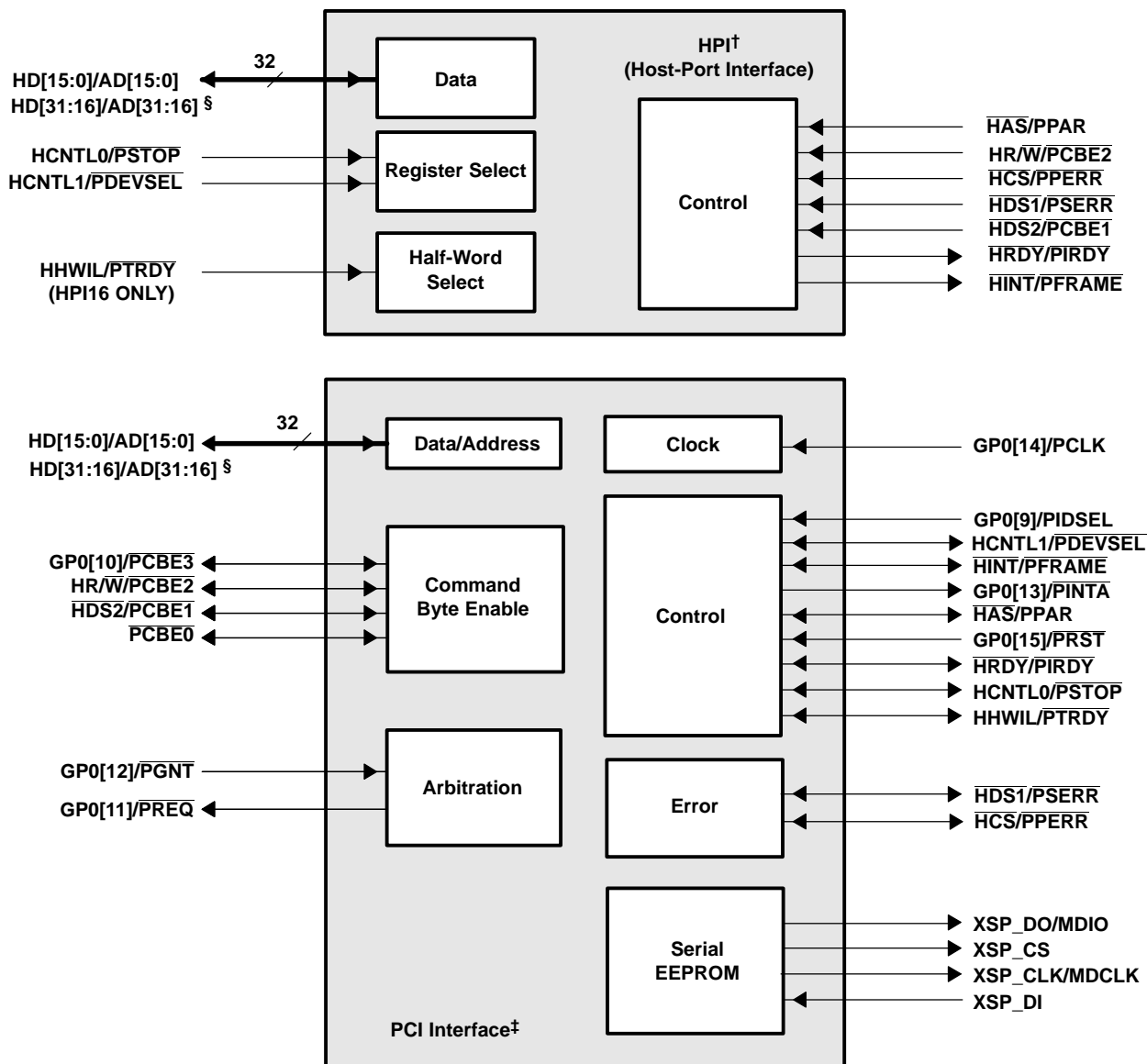
† These HPI pins are muxed with the PCI peripheral. By default, these signals function as HPI. For more details on these muxed pins, see the Device Configurations section of this data sheet.

‡ These PCI pins (excluding PCBE0 and XSP_CS) are muxed with the HPI or MDIO or GP0 peripherals. By default, these signals function as HPI and no function, respectively. For more details on these muxed pins, see the Device Configurations section of this data sheet.

§ These HPI/PCI data pins (HD[31:16/AD[31:16]) are muxed with the EMAC peripheral. By default, these pins function as HPI. For more details on the EMAC pin functions, see the Ethernet MAC (EMAC) peripheral signals section and the terminal functions table portions of this data sheet.

Figure 1−7. Peripheral Signals (Continued)

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McBSP1

CLKX1

FSX1

DX1

Transmit

CLKR1

FSR1

DR1

Receive

CLKS1

Clock

McBSP0

CLKX0

FSX0

DX0

Transmit

CLKR0

FSR0

DR0

Receive

CLKS0

Clock

**McBSPs
(Multichannel Buffered
Serial Ports)**

TOUT1/LENDIAN

TINP1

Timer 1

Timer 0

TOUT0/MACEN

TINP0

Timer 2

Timers

I2C0

SCL0

SDA0

I2C0

Figure 1–7. Peripheral Signals (Continued)

TEXAS
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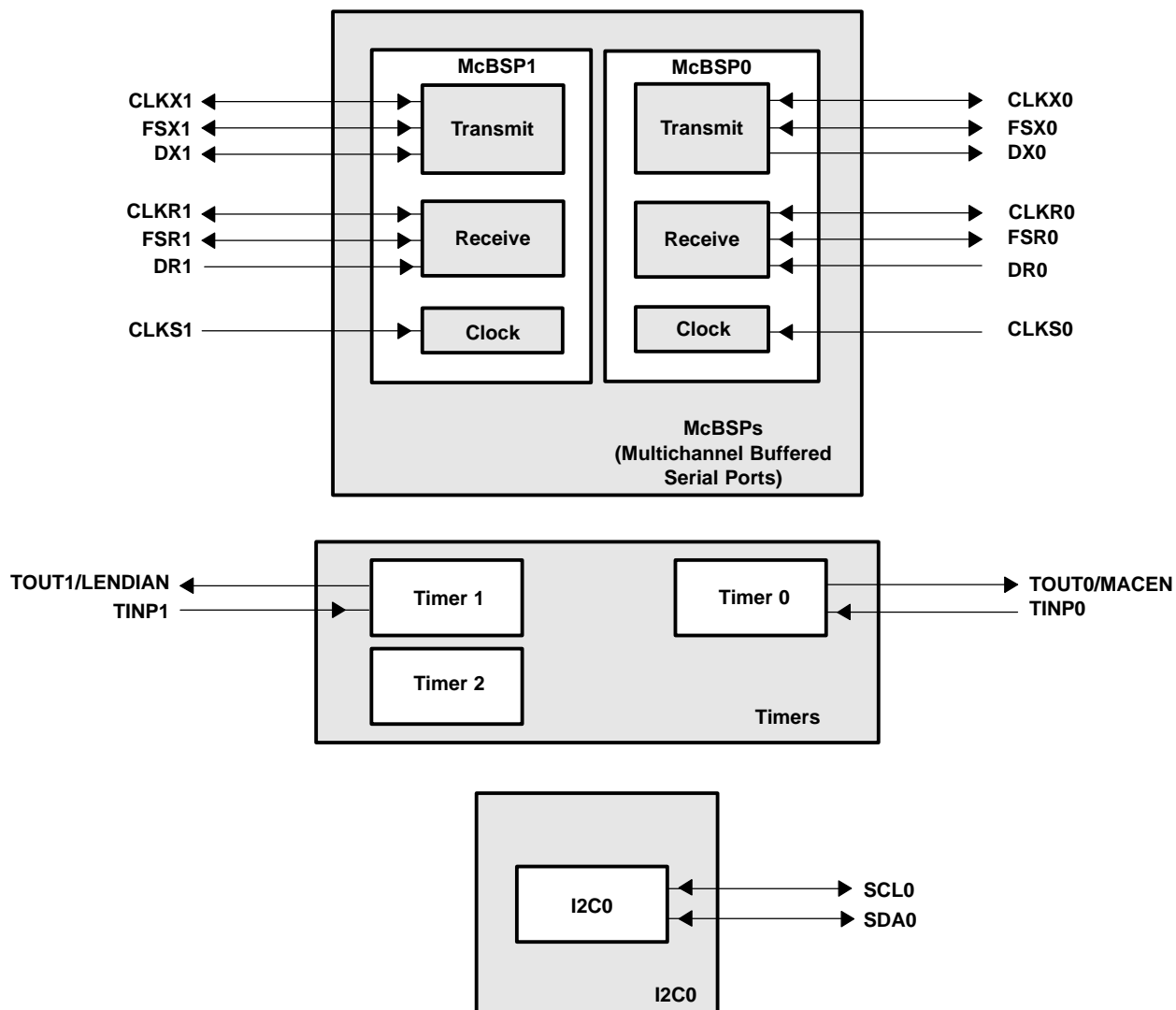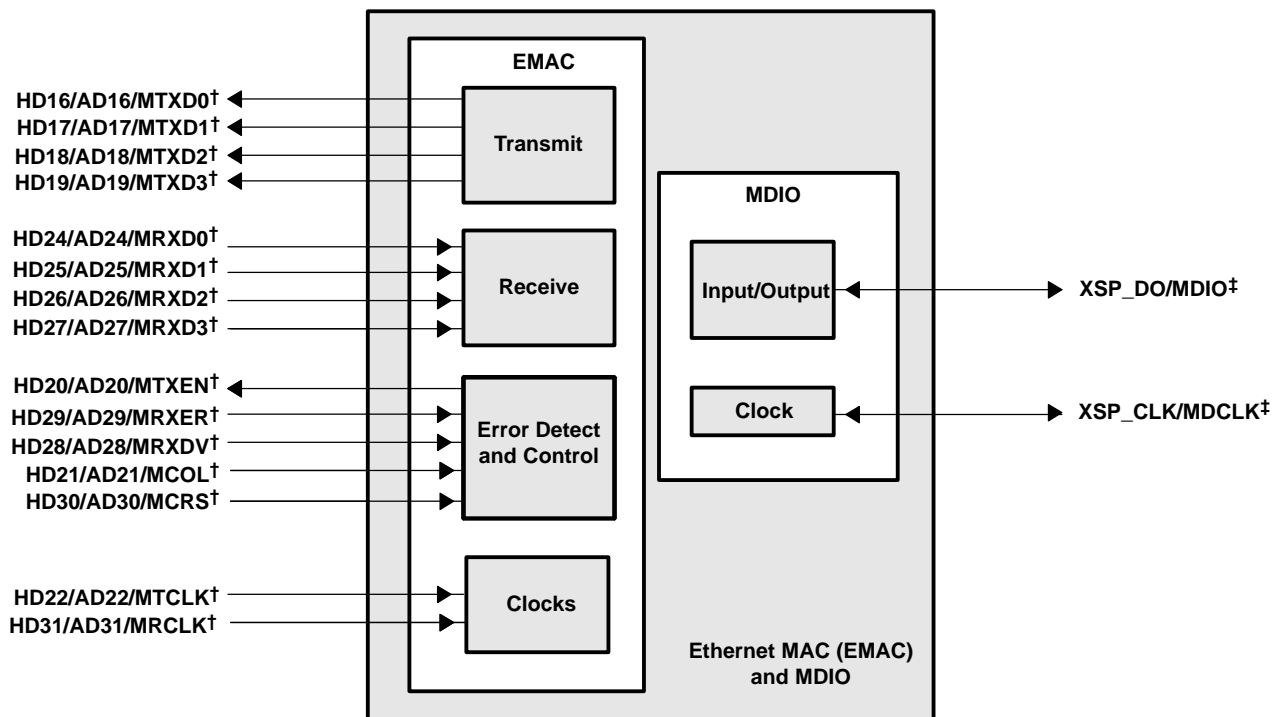
† These EMAC pins are muxed with the upper data pins of the HPI or PCI peripherals. By default, these signals function as HPI. For more details on these muxed pins, see the Device Configurations section of this data sheet.
‡ These MDIO pins are muxed with the PCI peripherals. By default, these signals function as PCI. For more details on these muxed pins, see the Device Configurations section of this data sheet.

Figure 1−7. Peripheral Signals (Continued)

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2 Device Configurations

On the C6412 device, bootmode and certain device configurations/peripheral selections are determined at device reset, while other device configurations/peripheral selections are software-configurable via the peripheral configurations register (PERCFG) [address location 0x01B3F000] after device reset.

2.1 Peripheral Selection at Device Reset

Some C6412 peripherals share the same pins (internally muxed) and are mutually exclusive (i.e., HPI, general-purpose input/output pins GP0[15:9], PCI and its internal EEPROM, EMAC, and MDIO). Other C6412 peripherals (i.e., the Timers, I2C0, and the GP0[7:0] pins), are always available.

- HPI, GP0[15:9], PCI, EEPROM (internal to PCI), and EMAC peripherals

  The PCI_EN and MAC_EN pins are latched at reset. They determine specific peripheral selection, summarized in Table 2–1.

Table 2–1. PCI_EN, HD5, and MAC_EN Peripheral Selection (HPI, GP0[15:9], PCI, EMAC, and MDIO)

PERIPHERAL SELECTION				PERIPHERALS SELECTED					
PCI_EN Pin [E2]	PCI_EEAI Pin [L5]	HD5 Pin [Y1]	MAC_EN Pin [C5]	HPI Data Lower	HPI Data Upper	32-Bit PCI	EEPROM (Auto-Init)	EMAC and MDIO	GP0[15:9]
0	0	0	0	√	Hi-Z	Disabled	N/A	Disabled	√
0	0	0	1	√	Hi-Z	Disabled	N/A	√	√
0	0	1	0	√	√	Disabled	N/A	Disabled	√
0	0	1	1	Disabled		Disabled	N/A	√	√
1	1	X	X	Disabled		√	Enabled (via External EEPROM)	Disabled	Disabled
1	0	X	X	Disabled		√	Disabled (default values)	Disabled	Disabled

- If the PCI is disabled (PCI_EN = 0), the HPI peripheral is enabled and based on the HD5 and MAC_EN pin configuration at reset, HPI16 mode or EMAC and MDIO can be selected. When the PCI is disabled (PCI_EN = 0), the GP0[15:9] pins can also be programmed as GPIO, provided the GPxEN and GPxDIR bits are properly configured.

  This means all multiplexed HPI/PCI pins function as HPI and all standalone PCI pins ($\overline{PCBE0}$ and XSP_CS) are tied-off (Hi-Z). Also, the multiplexed GP0/PCI pins can be used as GPIO with the proper software configuration of the GPIO enable and direction registers (for more details, see Table 2–9).

- If the PCI is enabled (PCI_EN = 1), the HPI peripheral is disabled.

  This means all multiplexed HPI/PCI pins function as PCI. Also, the multiplexed GP0/PCI pins function as PCI pins (for more details, see Table 2–9).

- The MAC_EN pin, in combination with the PCI_EN and HD5 pins, controls the selection of the EMAC and MDIO peripherals (for more details, see Table 2–2).

- The PCI_EN pin (= 1) and the PCI_EEAI pin control the whether the PCI initializes its internal registers via external EEPROM (PCI_EEAI = 1) or if the internal default values are used instead (PCI_EEAI = 0).

Table 2–2. HPI vs. EMAC Peripheral Pin Selection

CONFIGURATION SELECTION			PERIPHERALS SELECTED	
GP0[0] Pin [M5][†]	HD5 Pin [Y1]	MAC_EN Pin [C5]	HD[15:0]	HD[31:16]
0	0	0	HPI16	Hi-Z
0	0	1	HPI16	used for EMAC
0	1	0	HPI32 (HD[31:0])	
0	1	1	Hi-Z	used for EMAC
1	X	X	Invalid configuration: The GP0[0] pin *must* remain low during device reset.	

2.2 Device Configuration at Device Reset

Table 2–3 describes the C6412 device configuration pins, which are set up via external pullup/pulldown resistors through the specified EMIFA address bus pins (AEA[22:19]), and the TOUT1/LENDIAN, GP0[3]/PCIEEAI, and the HD5 pins (all of which are latched during device reset).

Table 2–3. C6412 Device Configuration Pins (TOUT1/**LENDIAN, AEA[22:19],** GP0[3]/**PCIEEAI,**
GP0[8]/$\overline{\text{PCI66}}$**, HD5**/AD5**, PCI_EN**, and TOUT0/**MAC_EN)**

CONFIGURATION PIN	NO.	FUNCTIONAL DESCRIPTION
TOUT1/**LENDIAN**	B5	Device Endian mode (LEND) 0 – System operates in Big Endian mode 1 – System operates in Little Endian mode (default)
AEA[22:21]	[U23, V24]	Bootmode [1:0] 00 – No boot (default mode) 01 – HPI/PCI boot (based on PCI_EN pin) 10 – Reserved 11 – EMIFA 8–bit ROM boot
AEA[20:19]	[V25, V26]	EMIFA input clock select Clock mode select for EMIFA (AECLKIN_SEL[1:0]) 00 – AECLKIN (default mode) 01 – CPU/4 Clock Rate 10 – CPU/6 Clock Rate 11 – Reserved
GP0[3]/**PCIEEAI**	L5	PCI EEPROM Auto-Initialization (PCIEEAI) PCI auto-initialization via external EEPROM 0 – PCI auto-initialization through EEPROM is disabled; the PCI peripheral uses the specified PCI default values (default). 1 – PCI auto-initialization through EEPROM is enabled; the PCI peripheral is configured through EEPROM provided the PCI peripheral pin is enabled (PCI_EN = 1). **Note**: If the PCI peripheral is disabled (PCI_EN pin = 0), this pin *must not* be pulled up. For more information on the PCI EEPROM default values, see the *TMS320C6000 DSP Peripheral Component Interconnect (PCI) Reference Guide* (literature number SPRU581).
GP0[8]/$\overline{\text{PCI66}}$	AD1	PCI frequency selection ($\overline{\text{PCI66}}$) [PCI peripheral *needs* be enabled (PCI_EN = 1) to use this function] Selects the PCI operating frequency of 66 MHz or 33 MHz PCI operating frequency is selected at reset via the pullup/pulldown resistor on the $\overline{\text{PCI66}}$ pin: 0 – PCI operates at 66 MHz (default). 1 – PCI operates at 33 MHz. The -500 speed device supports PCI at 33 MHz only. For proper -500 device operation when the PCI is enabled (PCI_EN = 1), this pin *must* be pulled up with a 1-kΩ resistor at device reset. **Note**: If the PCI peripheral is disabled (PCI_EN pin = 0), this pin *must not* be pulled up.
HD5/AD5	Y1	HPI peripheral bus width (HPI_WIDTH) 0 – HPI operates as an HPI16. (HPI bus is 16 bits wide. HD[15:0] pins are used and the remaining HD[31:16] pins are reserved pins in the Hi-Z state.) 1 – HPI operates as an HPI32. (HPI bus is 32 bits wide. All HD[31:0] pins are used for host-port operations.) (Also see the PCI_EN; TOUT0/MAC_EN functional description in this table)
PCI_EN; TOUT0/**MAC_EN**	[E2; C5]	Peripheral Selection 00 – HPI (default mode) [HPI32, if HD5 = 1; HPI16 if HD5 = 0 01 – EMAC and MDIO; HPI16, if HD5 = 0; HPI disabled, if HD5 = 1 10 – PCI 11 – Reserved

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2.3 Peripheral Selection After Device Reset

McBSP1, McBSP0, and I2C0

The C6412 device has designated registers for peripheral configuration (PERCFG), device status (DEVSTAT), and JTAG identification (JTAGID). These registers are part of the Device Configuration module and are mapped to a 4K block memory starting at 0x01B3F000. The CPU accesses these registers via the CFGBUS.

The peripheral configuration register (PERCFG), allows the user to control the peripheral selection of the McBSP0, McBSP1, and I2C0 peripherals. For more detailed information on the PERCFG register control bits, see Figure 2–1 and Table 2–4.

31					24
Reserved					
R-0					

23					16
Reserved					
R-0					

15					8
Reserved					
R-0					

7	4	3	2	1	0
Reserved		I2C0EN	MCBSP1EN	MCBSP0EN	Reserved
R-0		R/W-0	**R/W-1**	**R/W-1**	R-0
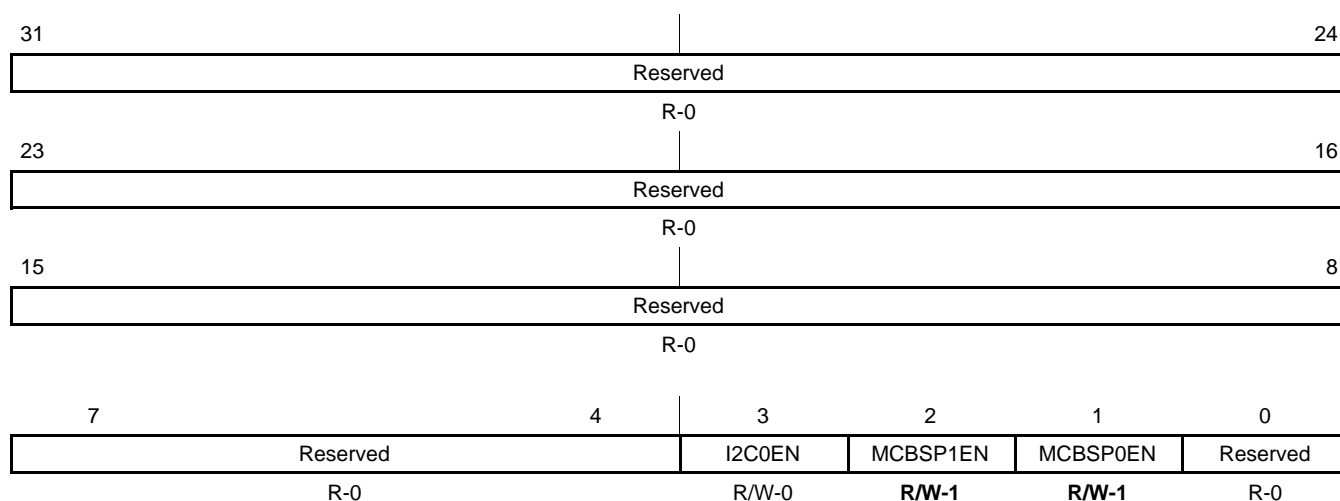
Legend: R = Read only; R/W = Read/Write; -n = value after reset

Figure 2–1. Peripheral Configuration Register (PERCFG) [Address Location: 0x01B3F000 – 0x01B3F003]

Table 2–4. Peripheral Configuration (PERCFG) Register Selection Bit Descriptions

BIT	NAME	DESCRIPTION
31:4	Reserved	Reserved. Read-only, writes have no effect.
3	I2C0EN	Inter-integrated circuit 0 (I2C0) enable bit. Selects whether I2C0 peripheral is enabled or disabled (default). 0 = I2C0 is disabled, and the module is powered down (default). 1 = I2C0 is enabled.
2	MCBSP1EN	McBSP1 enable bit. 0 = Reserved. Do not use. 1 = McBSP1 is enabled (default).
1	MCBSP0EN	McBSP0 enable bit. 0 = Reserved. Do not use. 1 = McBSP0 is enabled (default).
0	Reserved	Reserved. Read-only, writes have no effect.

2.4 Peripheral Configuration Lock

By default, the I2C peripheral is disabled on power up. In order to use this peripheral on the C6412 device, the peripheral must first be enabled in the Peripheral Configuration register (PERCFG). **Software muxed pins should not be programmed to switch functionalities during run-time. Care should also be taken to ensure that no accesses are being performed before disabling the peripherals.** To help minimize power consumption in the C6412 device, unused peripherals may be disabled.

Figure 2–2 shows the flow needed to enable (or disable) a given peripheral on the C6412 device.

**Unlock the PERCFG Register
Using the PCFGLOCK Register**

**Write to
PERCFG Register
to Enable/Disable Peripherals**

**Read from
PERCFG Register**

**Wait 128 CPU Cycles Before
Accessing Enabled Peripherals**
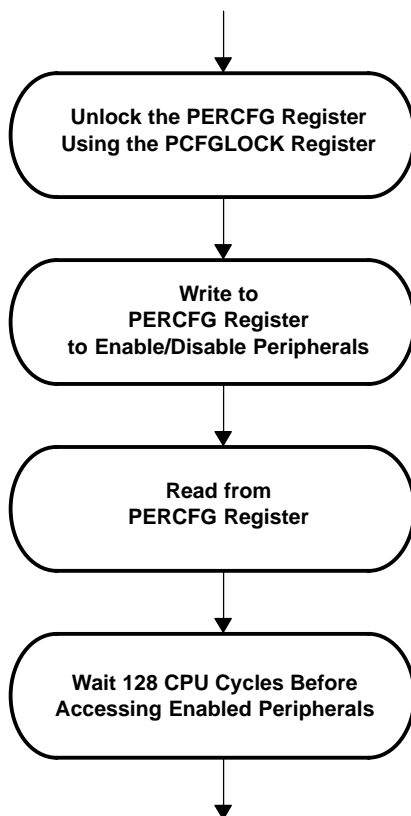
Figure 2–2. Peripheral Enable/Disable Flow Diagram

A 32-bit key (value = 0x10C0010C) must be written to the Peripheral Configuration Lock register (PCFGLOCK) in order to unlock access to the PERCFG register. Reading the PCFGLOCK register determines whether the PERCFG register is currently locked (LOCKSTAT bit = 1) or unlocked (LOCKSTAT bit = 0), see Figure 2–3. A peripheral can only be enabled when the PERCFG register is "unlocked" (LOCKSTAT bit = 0).

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Read Accesses

31		1	0
Reserved			LOCKSTAT
R-0			R-1

Write Accesses

31	0
LOCK	
W-0	
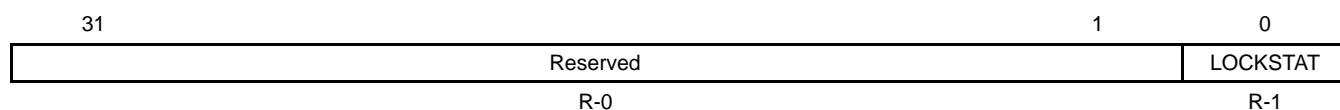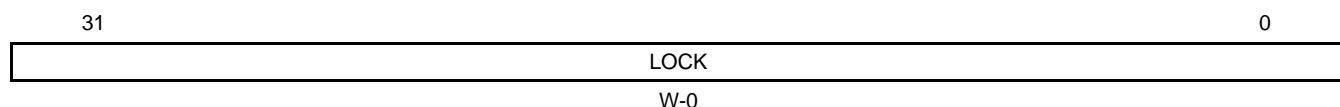
Legend: R = Read only; R/W = Read/Write; -n = value after reset

Figure 2–3. PCFGLOCK Register Diagram [Address Location: 0x01B3 F018] – Read/Write Accesses

Table 2–5. PCFGLOCK Register Selection Bit Descriptions – Read Accesses

BIT	NAME	DESCRIPTION
31:1	Reserved	Reserved. Read-only, writes have no effect.

0	LOCKSTAT	Lock status bit.
		Determines whether the PERCFG register is locked or unlocked.
		    0 = Unlocked, read accesses to the PERCFG register allowed.
		    1 = Locked, write accesses to the PERCFG register do **not** modify the register state [default].
		Reads are unaffected by Lock Status.

Table 2–6. PCFGLOCK Register Selection Bit Descriptions – Write Accesses

BIT	NAME	DESCRIPTION
31:0	LOCK	Lock bits. 0x10C0010C = Unlocks PERCFG register accesses.

Any write to the PERCFG register will automatically relock the register. In order to avoid the unnecessary overhead of multiple unlock/enable sequences, all peripherals should be enabled with a single write to the PERCFG register with the necessary enable bits set.

Prior to waiting 128 CPU cycles, the PERCFG register should be read. There is no direct correlation between the CPU issuing a write to the PERCFG register and the write actually occurring. Reading the PERCFG register after the write is issued forces the CPU to wait for the write to the PERCFG register to occur.

Once a peripheral is enabled, the DSP (or other peripherals such as the HPI) must wait a minimum of 128 CPU cycles before accessing the enabled peripheral. The user *must* ensure that no accesses are performed to a peripheral while it is disabled.

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2.5 Device Status Register Description

The device status register depicts the status of the device peripheral selection. For the actual register bit names and their associated bit field descriptions, see Figure 2–4 and Table 2–7.

31							24
Reserved							
R-0							

23							16
Reserved							
R-0							

15	14	13	12	11	10	9	8
Reserved				MAC_EN	HPI_WIDTH	PCI_EEAI	PCI_EN
R-0				R-x	R-x	R-x	R-x

7	6	5	4	3	2	1	0
Reserved	CLKMODE1	CLKMODE0	LENDIAN	BOOTMODE1	BOOTMODE0	AECLKINSEL1	AECLKINSEL0
R-x	R-x	R-x	R-x	R-x	R-x	R-x	R-x
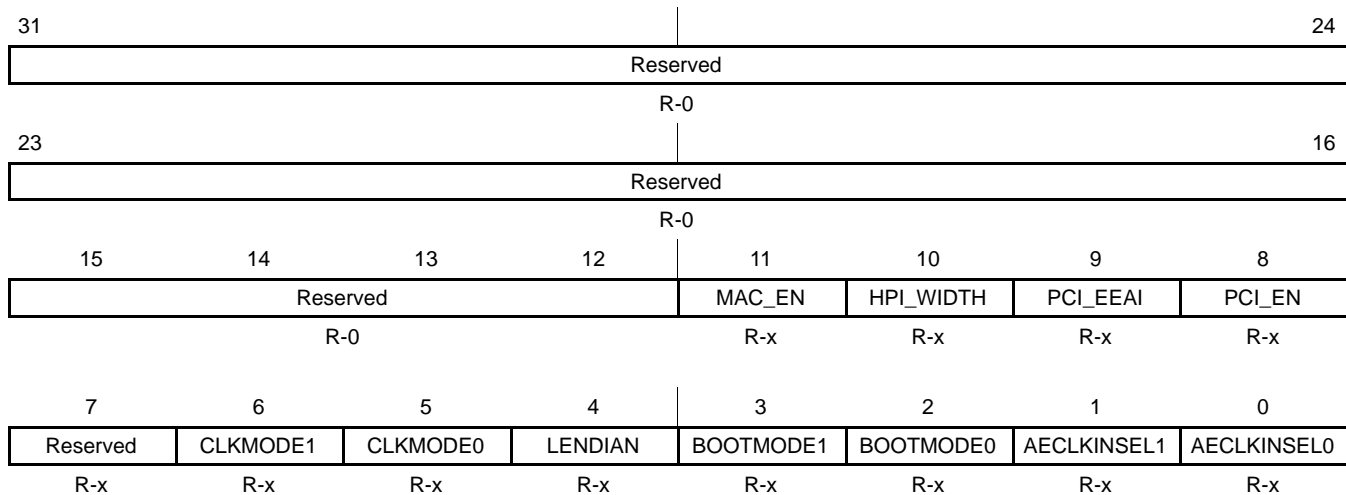
Legend: R = Read only; R/W = Read/Write; -n = value after reset

Figure 2–4. Device Status Register (DEVSTAT) Description – 0x01B3 F004

Table 2–7. Device Status (DEVSTAT) Register Selection Bit Descriptions

BIT	NAME	DESCRIPTION
31:12	Reserved	Reserved. Read-only, writes have no effect.

11	MAC_EN	EMAC enable bit.
		Shows the status of whether EMAC peripheral is enabled or disabled (default).
		   0 = EMAC is disabled, and the module is powered down (default).
		   1 = EMAC is enabled.
		This bit has no effect if the PCI peripheral is enabled (PCI_EN = 1).

10	HPI_WIDTH	HPI bus width control bit.
		Shows the status of whether the HPI bus operates in 32-bit mode or in 16-bit mode (default).
		   0 = HPI operates in 16-bit mode. (default).
		   1 = HPI operates in 32-bit mode.

9	PCI_EEAI	PCI EEPROM auto-initialization bit (PCI auto-initialization via external EEPROM).
		Shows the status of whether the PCI module initializes internal registers via external EEPROM or if the internal PCI default values are used instead (default).
		
		   0 = PCI auto-initialization through EEPROM is disabled; the PCI peripheral uses the specified PCI default values (default).
		   1 = PCI auto-initialization through EEPROM is enabled; the PCI peripheral is configured through EEPROM provided the PCI peripheral pin is enabled (PCI_EN = 1).

8	PCI_EN	PCI enable bit.
		Shows the status of whether the PCI peripheral is enabled or disabled (default).
		   0 = PCI disabled. (default).
		   1 = PCI enabled.
		
		Global select for the PCI vs. HPI/EMAC/MDIO/GPIO peripherals.

7	Reserved	Reserved. Read-only, writes have no effect.

Table 2–7. Device Status (DEVSTAT) Register Selection Bit Descriptions (Continued)

BIT	NAME	DESCRIPTION

6	CLKMODE1	Clock mode select bits
		Shows the status of whether the CPU clock frequency equals the input clock frequency X1 (Bypass), x6, or x12.
		    Clock mode select for CPU clock frequency (CLKMODE[1:0])
		    00 – Bypass (x1) (default mode)

5	CLKMODE0	01 – x6 10 – x12 11 – Reserved For more details on the CLKMODE pins and the PLL multiply factors, see the Clock PLL section of this data sheet.

4	LENDIAN	Device Endian mode (LEND)
		Shows the status of whether the system is operating in Big Endian mode or Little Endian mode (default).
		    0 – System is operating in Big Endian mode
		    1 – System is operating in Little Endian mode (default)

3	BOOTMODE1	Bootmode configuration bits Shows the status of what device bootmode configuration is operational. Bootmode [1:0] 00 – No boot (default mode)
2	BOOTMODE0	01 – HPI/PCI boot (based on PCI_EN pin) 10 – Reserved 11 – EMIFA 8–bit ROM boot

1	AECLKINSEL1	EMIFA input clock select
		Shows the status of what clock mode is enabled or disabled for the EMIF.
		    Clock mode select for EMIFA (AECLKIN_SEL[1:0])
		    00 – AECLKIN (default mode)

0	AECLKINSEL0	01 – CPU/4 Clock Rate 10 – CPU/6 Clock Rate 11 – Reserved

2.6 JTAG ID Register Description

The JTAG ID register is a read-only register that identifies to the customer the JTAG/Device ID. For the C6412 device, the JTAG ID register resides at address location 0x01B3 F008. The register hex value for the C6412 device is: 0x0007 902F. For the actual register bit names and their associated bit field descriptions, see Figure 2–5 and Table 2–8.

31–28	27–12	11–1	0
VARIANT (4-Bit)	PART NUMBER (16-Bit)	MANUFACTURER (11-Bit)	LSB
R-0000	R-0000 0000 0111 1001	R-0000 0010 111	R-1

Legend: R = Read only; -n = value after reset

Figure 2–5. JTAG ID Register Description – TMS320C6412 Register Value – 0x0007 902F

Table 2–8. JTAG ID Register Selection Bit Descriptions

BIT	NAME	DESCRIPTION
31:28	VARIANT	Variant (4-Bit) value. C6412 value: 0000.
27:12	PART NUMBER	Part Number (16-Bit) value. C6412 value: 0000 0000 0111 1001.
11–1	MANUFACTURER	Manufacturer (11-Bit) value. C6412 value: 0000 0010 111.
0	LSB	LSB. This bit is read as a "1" for C6412.

2.7   Multiplexed Pins

Multiplexed pins are pins that are shared by more than one peripheral and are internally multiplexed. Some of these pins are configured by software, and the others are configured by external pullup/pulldown resistors only at reset. Those muxed pins that are configured by software should *not* be programmed to switch functionalities during run-time. Those muxed pins that are configured by external pullup/pulldown resistors are mutually exclusive; only one peripheral has primary control of the function of these pins after reset. Table 2–9 identifies the multiplexed pins on the C6412 device; shows the default (primary) function and the default settings after reset; and describes the pins, registers, etc. necessary to configure specific multiplexed functions.

2.8   Debugging Considerations

It is recommended that external connections be provided to device configuration pins, including TOUT1/LENDIAN, AEA[22:19], GP0[3]/PCIEEAI, GP0[8]/$\overline{\text{PCI66}}$, HD5/AD5, PCI_EN, and TOUT0/MAC_EN. Although internal pullup/pulldown resistors exist on these pins, providing external connectivity adds convenience to the user in debugging and flexibility in switching operating modes.

Internal pullup/pulldown resistors also exist on the non-configuration pins on the AEA bus (AEA[18:0]). Do *not* oppose the internal pullup/pulldown resistors on these non-configuration pins with external pullup/pulldown resistors. If an external controller provides signals to these non-configuration pins, these signals must be driven to the default state of the pins at reset, or not be driven at all.

For the internal pullup/pulldown resistors for all device pins, see the terminal functions table.

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Table 2–9. C6412 Device Multiplexed Pins[†]

MULTIPLEXED PINS		DEFAULT FUNCTION	DEFAULT SETTING	DESCRIPTION
NAME	NO.			
CLKOUT4/GP0[1]	D6	CLKOUT4	GP1EN = 0 (disabled)	These pins are software-configurable. To use these pins as GPIO pins, the GPxEN bits in the GPIO Enable Register and the GPxDIR bits in the GPIO Direction Register must be properly configured.
CLKOUT6/GP0[2]	C6	CLKOUT6	GP2EN = 0 (disabled)	GPxEN = 1: GPx pin enabled GPxDIR = 0: GPx pin is an input GPxDIR = 1: GPx pin is an output
GP0[3]/PCIEEAI	L5	PCIEEAI	GP3EN = 0 (disabled)	To use the PCI auto–initialization EEPROM (PCIEEAI) the PCI needs to be enabled (PCI_EN = 1): 0 – PCI auto-init through EEPROM disabled (default). 1 – PCI auto-init through EEPROM is enabled. To use GP0[3] as a GPIO pin, the PCI needs to be disabled (PCI_EN = 0), the GP3EN bits in the GPIO Enable Register and the GP3DIR bits in the GPIO Direction Register must be properly configured. GP3EN = 1: GP3 pin enabled GP3DIR = 0: GP3 pin is an input GP3DIR = 1: GP3 pin is an output
GP0[8]/$\overline{PCI66}$	AD1	$\overline{PCI66}$	GP8EN = 0 (disabled) MAC_EN = 0 (disabled)	To use GP0[8] as a GPIO pin, the PCI needs to be disabled (PCI_EN = 0), the GPxEN bits in the GPIO Enable Register and the GPxDIR bits in the GPIO Direction Register must be properly configured. GP8EN = 1: GP8 pin enabled GP8DIR = 0: GP8 pin is an input GP8DIR = 1: GP8 pin is an output To use the $\overline{PCI66}$ pin function, which changes the PCI operating frequency selection, the PCI needs to be enabled (PCI_EN = 1): 0 – PCI operates at 66 MHz (default). 1 – PCI operates at 33 MHz. The -500 device supports PCI at 33 MHz only. For proper -500 device operation when the PCI peripheral is enabled (PCI_EN = 1), this pin **must** be pulled up with a 1-kΩ resistor at device reset. If the PCI peripheral is disabled (PCI_EN pin = 0), this pin must **not** be pulled up.
GP0[9]/PIDSEL	K3	None	GPxEN = 0 (disabled) PCI_EN = 0 (disabled)[†]	To use GP0[15:9] as GPIO pins, the PCI needs to be disabled (PCI_EN = 0), the GPxEN bits in the GPIO Enable Register and the GPxDIR bits in the GPIO Direction Register must be properly configured. GPxEN = 1: GPx pin enabled GPxDIR = 0: GPx pin is an input GPxDIR = 1: GPx pin is an output
GP0[10]/$\overline{PCBE3}$	J2			
GP0[11]/$\overline{PREQ}$	F1			
GP0[12]/$\overline{PGNT}$	H4			
GP0[13]/$\overline{PINTA}$	G4			
GP0[14]/PCLK	C1			
GP0[15]/$\overline{PRST}$	G3			

[†] All other standalone PCI pins are tied-off internally (pins in Hi-Z) when the peripheral is disabled [PCI_EN = 0].

[‡] For the HD[31:0]/AD[31:0] multiplexed pins pin numbers, see the *Terminal Functions* table.

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Table 2–9. C6412 Device Multiplexed Pins† (Continued)

MULTIPLEXED PINS NAME	NO.	DEFAULT FUNCTION	DEFAULT SETTING	DESCRIPTION
XSP_CLK/MDCLK	R5	None	PCI_EN = 0 (disabled)† MAC_EN = 0 (disabled)†	By default, no functions enabled upon reset (PCI is disabled). To enable the PCI peripheral, an external pullup resistor (1 kΩ) must be provided on the PCI_EN pin (setting PCI_EN = 1 at reset) To enable the MDIO peripheral (which also enables the EMAC peripheral), an external pullup resistor (1 kΩ) must be provided on the MAC_EN pin (setting MAC_EN = 1 at reset)
XSP_DO/MDIO	P5			
HAS/PPAR	P3	HAS	PCI_EN = 0 (disabled)†	By default, HPI is enabled upon reset (PCI is disabled). To enable the PCI peripheral, an external pullup resistor (1 kΩ) must be provided on the PCI_EN pin (setting PCI_EN = 1 at reset).
HCNTL1/PDEVSEL	P1	HCNTL1		
HCNTL0/PSTOP	R3	HCNTL0		
HDS1/PSERR	R2	HDS1		
HDS2/PCBE1	T2	HDS2		
HR/W/PCBE2	M1	HR/W		
HHWIL/PTRDY	N3	HHWIL (HPI16 only)		
HINT/PFRAME	N4	HINT		
HCS/PPERR	R1	HCS		
HRDY/PIRDY	N1	HRDY		
HD[23,15:0]/AD[23,15:0]	‡	HD[23, 15:0]	PCI_EN = 0 (disabled)†	By default, HPI is enabled upon reset (PCI is disabled). To enable the PCI peripheral, an external pullup resistor (1 kΩ) must be provided on the PCI_EN pin (setting PCI_EN = 1 at reset).
HD31/AD31/MRCLK	G1	HD31	PCI_EN = 0 (disabled)† MAC_EN = 0 (disabled)†	By default, HPI is enabled upon reset (PCI is disabled). To enable the PCI peripheral, an external pullup resistor (1 kΩ) must be provided on the PCI_EN pin (setting PCI_EN = 1 at reset). To enable the EMAC peripheral, an external pullup resistor (1 kΩ) must be provided on the MAC_EN pin (setting MAC_EN = 1 at reset).
HD30/AD30/MCRS	H3	HD30		
HD29/AD29/MRXER	G2	HD29		
HD28/AD28/MRXDV	J4	HD28		
HD27/AD27/MRXD3	H2	HD27		
HD26/AD26/MRXD2	J3	HD26		
HD25/AD25/MRXD1	J1	HD25		
HD24/AD24/MRXD0	K4	HD24		
HD22/AD22/MTCLK	L4	HD22		
HD21/AD21/MCOL	K2	HD21		
HD20/AD20/MTXEN	L3	HD20		
HD19/AD19/MTXD3	L2	HD19		
HD18/AD18/MTXD2	M4	HD18		
HD17/AD17/MTXD1	M2	HD17		
HD16/AD16/MTXD0	M3	HD16		

† All other standalone PCI pins are tied-off internally (pins in Hi-Z) when the peripheral is disabled [PCI_EN = 0].
‡ For the HD[31:0]/AD[31:0] multiplexed pins pin numbers, see the *Terminal Functions* table.

2.9 Configuration Examples

Figure 2–6 illustrates an example of peripheral selections that are configurable on the C6412 device.

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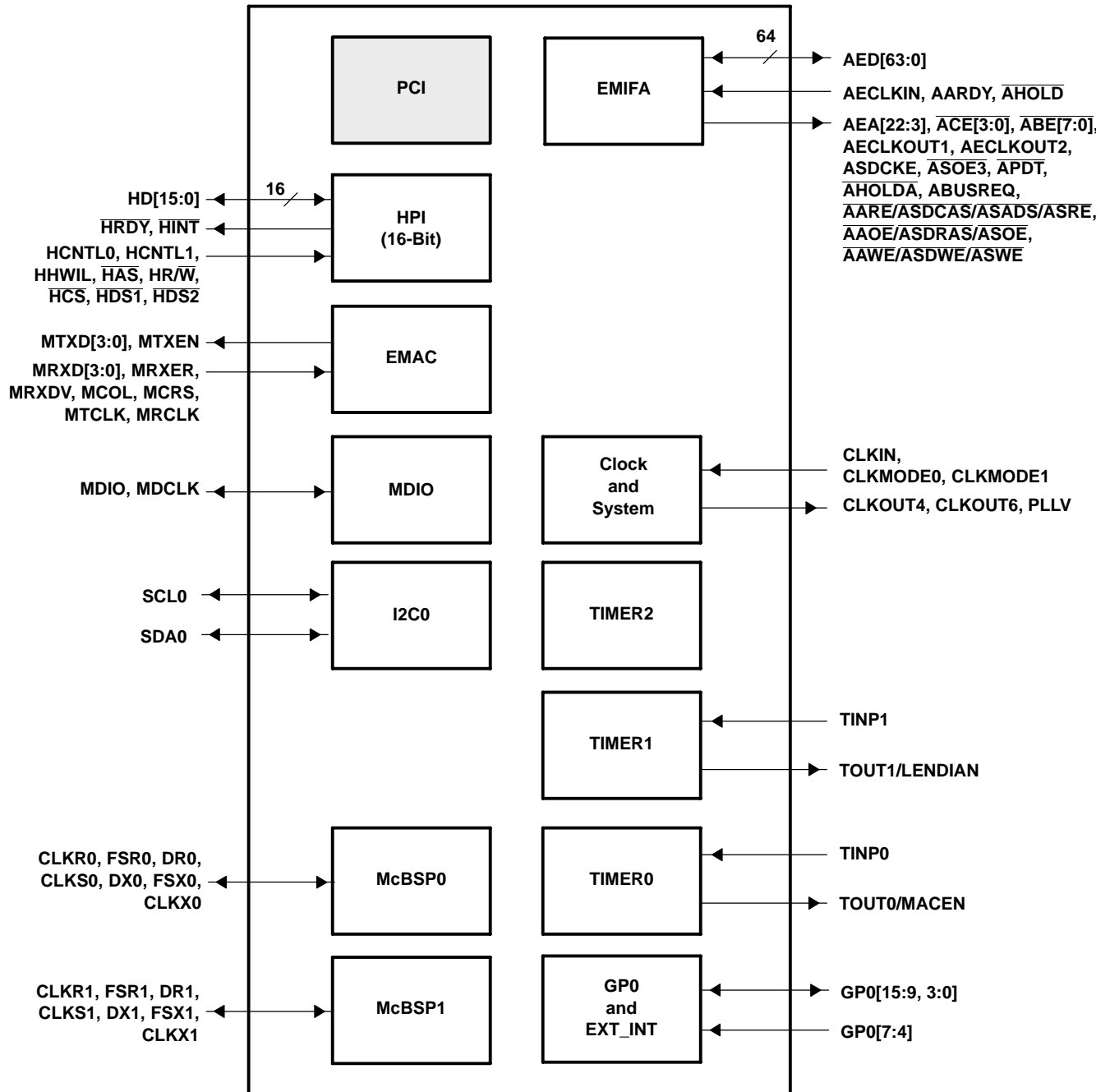
□ Shading denotes a peripheral module not available for this configuration.

PERCFG Register Value: 0x0000 000D
Extenal Pins: PCI_EN = 0   GP0[3]/PCIEEAI = 0   HD5 = 0   TOUT0/MAC_EN = 1

**Figure 2−6. Configuration Example
(2 McBSPs + EMAC + MDIO + I2C0 + EMIF + HPI + 3 Timers)**

2.10 Terminal Functions

The terminal functions table (Table 2−10) identifies the external signal names, the associated pin (ball) numbers along with the mechanical package designator, the pin type (I, O/Z, or I/O/Z), whether the pin has any internal pullup/pulldown resistors and a functional pin description. For more detailed information on device configuration, peripheral selection, multiplexed/shared pins, and debugging considerations, see the Device Configurations section of this data sheet.

Table 2–10. Terminal Functions

SIGNAL NAME	GDK/ GNZ	TYPE[†]	IPD/ IPU[‡]	DESCRIPTION
CLOCK/PLL CONFIGURATION				
CLKIN	AC2	I		Clock Input. This clock is the input to the on-chip PLL.
CLKOUT4/GP0[1][§]	D6	I/O/Z	IPU	Clock output at 1/4 of the device speed (**O/Z**) [default] or this pin can be programmed as a GP0 1 pin (**I/O/Z**).
CLKOUT6/GP0[2][§]	C6	I/O/Z	IPU	Clock output at 1/6 of the device speed (**O/Z**) [default] or this pin can be programmed as a GP0 2 pin (**I/O/Z**).
CLKMODE1	AE4	I	IPD	Clock mode select • Selects whether the CPU clock frequency = input clock frequency x1 (Bypass), x6, or x12.
CLKMODE0	AA2	I	IPD	For more details on the CLKMODE pins and the PLL multiply factors, see the Clock PLL section of this data sheet.
PLLV[¶]	V6	A[#]		PLL voltage supply
JTAG EMULATION				
TMS	E15	I	IPU	JTAG test-port mode select
TDO	B18	O/Z	IPU	JTAG test-port data out
TDI	A18	I	IPU	JTAG test-port data in
TCK	A16	I	IPU	JTAG test-port clock
$\overline{\text{TRST}}$	D14	I	IPD	JTAG test-port reset. For IEEE 1149.1 JTAG compatibility, see the IEEE 1149.1 JTAG compatibility statement portion of this data sheet.
EMU11	D17	I/O/Z	IPU	Emulation clock 1. Reserved for future use, leave unconnected.
EMU10	C17	I/O/Z	IPU	Emulation clock 0. Reserved for future use, leave unconnected.
EMU9	B17	I/O/Z	IPU	Emulation pin 9. Reserved for future use, leave unconnected.
EMU8	D16	I/O/Z	IPU	Emulation pin 8. Reserved for future use, leave unconnected.
EMU7	A17	I/O/Z	IPU	Emulation pin 7. Reserved for future use, leave unconnected.
EMU6	C16	I/O/Z	IPU	Emulation pin 6. Reserved for future use, leave unconnected.
EMU5	B16	I/O/Z	IPU	Emulation pin 5. Reserved for future use, leave unconnected.
EMU4	D15	I/O/Z	IPU	Emulation pin 4. Reserved for future use, leave unconnected.
EMU3	C15	I/O/Z	IPU	Emulation pin 3. Reserved for future use, leave unconnected.
EMU2	B15	I/O/Z	IPU	Emulation pin 2. Reserved for future use, leave unconnected.
EMU1	C14	I/O/Z	IPU	Emulation pin 1[‖]
EMU0	A15	I/O/Z	IPU	Emulation pin 0[‖]

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground

‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)

§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.

¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.

A = Analog signal (PLL Filter)

‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
RESETS, INTERRUPTS, AND GENERAL-PURPOSE INPUT/OUTPUTS				
$\overline{\text{RESET}}$	P4	I		Device reset
NMI	B4	I	IPD	Nonmaskable interrupt, edge-driven (rising edge) **Note:** Any noise on the NMI pin may trigger an NMI interrupt; therefore, if the NMI pin is not used, it is recommended that the NMI pin be grounded versus relying on the IPD.
GP0[7]/EXT_INT7	E1	I/O/Z	IPU	General-purpose input/output (GPIO) pins (**I/O/Z**) or external interrupts (**input only**). The default after reset setting is GPIO enabled as input-only.
GP0[6]/EXT_INT6	F2	I/O/Z	IPU	• When these pins function as External Interrupts [by selecting the corresponding interrupt enable register bit (IER.[7:4])], they are edge-driven and the polarity can be independently selected via the External Interrupt Polarity Register bits (EXTPOL.[3:0]).
GP0[5]/EXT_INT5	F3	I/O/Z	IPU	
GP0[4]/EXT_INT4	F4	I/O/Z	IPU	
GP0[15]/$\overline{\text{PRST}}$§	G3			General-purpose input/output (GP0) 15 pin (**I/O/Z**) or PCI reset (**I**). No function at default.
GP0[14]/PCLK§	C1			GP0 14 pin (**I/O/Z**) or PCI clock (**I**). No function at default.
GP0[13]/$\overline{\text{PINTA}}$§	G4			GP0 13 pin (**I/O/Z**) or PCI interrupt A (**O/Z**). No function at default.
GP0[12]/$\overline{\text{PGNT}}$§	H4	I/O/Z		GP0 12 pin (**I/O/Z**) or PCI bus grant (**I**). No function at default.
GP0[11]/$\overline{\text{PREQ}}$§	F1			GP0 11 pin (**I/O/Z**) or PCI bus request (**O/Z**). No function at default.
GP0[10]/$\overline{\text{PCBE3}}$§	J2			GP0 10 pin (**I/O/Z**) or PCI command/byte enable 3 (**I/O/Z**). No function at default.
GP0[9]/PIDSEL§	K3			GP0 9 pin (**I/O/Z**) or PCI initialization device select (**I**). No function at default.
GP0[3]/PCIEEAI	L5		IPD	GP0 3 pin (**I/O/Z**) and PCI EEPROM Auto-Initialization (EEAI). If the PCI peripheral is disabled (PCI_EN pin = 0), this pin must *not* be pulled up. 0 – PCI auto-initialization through EEPROM is disabled (default). 1 – PCI auto-initialization through EEPROM is enabled.
GP0[0]	M5	I/O/Z	IPD	GP0 0 pin (**I/O/Z**) [default]. The general-purpose 0 pin (GP0[0]) (**I/O/Z**) can be programmed as GPIO 0 (**input only**) [default] or as GP0[0] (**output only**) pin or output as a general-purpose interrupt (GP0INT) signal (**output only**) **Note:** This pin *must* remain low during device reset. For more details, see the Device Configurations section of this data manual.
GP0[8]/$\overline{\text{PCI66}}$§	AD1	I/O/Z	IPD	This pin can be programmed as a GP0 8 pin (**I/O/Z**) or PCI frequency selection ($\overline{\text{PCI66}}$). If the PCI peripheral is enabled (PCI_EN pin = 1), then: 0 – PCI operates at 66 MHz (default). 1 – PCI operates at 33 MHz. The -500 device supports PCI at 33 MHz only. For proper -500 device operation when the PCI peripheral is enabled (PCI_EN = 1), this pin *must* be pulled up with a 1-kΩ resistor at device reset. If the PCI peripheral is disabled (PCI_EN pin = 0), this pin must *not* be pulled up.
CLKOUT6/GP0[2]§	C6	I/O/Z	IPU	Clock output at 1/6 of the device speed (**O/Z**) [default] or this pin can be programmed as a GP0 2 pin (**I/O/Z**).
CLKOUT4/GP0[1]§	D6	I/O/Z	IPU	Clock output at 1/4 of the device speed (**O/Z**) [default] or this pin can be programmed as a GP0 1 pin (**I/O/Z**).

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground

‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)

§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.

¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.

A = Analog signal (PLL Filter)

‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

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Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE[†]	IPD/ IPU[‡]	DESCRIPTION
HOST-PORT INTERFACE (HPI) OR PERIPHERAL COMPONENT INTERCONNECT (PCI) OR EMAC				
PCI_EN	E2	I	IPD	PCI enable pin. This pin and the MAC_EN pin control the selection (enable/disable) of the HPI, EMAC, MDIO, and GP0[15:8], or PCI peripherals. The pins work in conjunction to enable/disable these peripherals (for more details, see the Device Configurations section of this data sheet).
$\overline{\text{HINT}}$/PFRAME[§]	N4	I/O/Z		Host interrupt from DSP to host (**O**) [default] or PCI frame (**I/O/Z**)
HCNTL1/ $\overline{\text{PDEVSEL}}$[§]	P1	I/O/Z		Host control – selects between control, address, or data registers (**I**) [default] or PCI device select (**I/O/Z**).
HCNTL0/ $\overline{\text{PSTOP}}$[§]	R3	I/O/Z		Host control – selects between control, address, or data registers (**I**) [default] or PCI stop (**I/O/Z**)
HHWIL/$\overline{\text{PTRDY}}$[§]	N3	I/O/Z		Host half-word select – first or second half-word (not necessarily high or low order) [For HPI16 bus width selection only] (**I**) [default] or PCI target ready (**I/O/Z**)
H$\overline{\text{R}}$/$\overline{\text{W}}$/PCBE2[§]	M1	I/O/Z		Host read or write select (**I**) [default] or PCI command/byte enable 2 (**I/O/Z**)
$\overline{\text{HAS}}$/PPAR[§]	P3	I/O/Z		Host address strobe (**I**) [default] or PCI parity (**I/O/Z**)
$\overline{\text{HCS}}$/$\overline{\text{PPERR}}$[§]	R1	I/O/Z		Host chip select (**I**) [default] or PCI parity error (**I/O/Z**)
$\overline{\text{HDS1}}$/$\overline{\text{PSERR}}$[§]	R2	I/O/Z		Host data strobe 1 (**I**) [default] or PCI system error (**I/O/Z**)
$\overline{\text{HDS2}}$/PCBE1[§]	T2	I/O/Z		Host data strobe 2 (**I**) [default] or PCI command/byte enable 1 (**I/O/Z**)
$\overline{\text{HRDY}}$/$\overline{\text{PIRDY}}$[§]	N1	I/O/Z		Host ready from DSP to host (**O**) [default] or PCI initiator ready (**I/O/Z**).

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground
‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)
§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.
¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.
# A = Analog signal (PLL Filter)
‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
HOST-PORT INTERFACE (HPI) OR PERIPHERAL COMPONENT INTERCONNECT (PCI) OR EMAC (CONTINUED)
HD31/AD31/MRCLK§	G1			
HD30/AD30/MCRS§	H3			
HD29/AD29/MRXER§	G2			
HD28/AD28/MRXDV§	J4			
HD27/AD27/MRXD3§	H2			
HD26/AD26/MRXD2§	J3			
HD25/AD25/MRXD1§	J1			
HD24/AD24/MRXD0§	K4			Host-port data (**I/O/Z**) [default] or PCI data-address bus (**I/O/Z**) or EMAC transmit/receive or control pins
HD23/AD23§	K1			
HD22/AD22/MTCLK§	L4			As HPI data bus (PCI_EN pin = 0)
HD21/AD21/MCOL§	K2			• Used for transfer of data, address, and control
HD20/AD20/MTXEN§	L3			• Host-Port bus width user-configurable at device reset via a 10-kΩ resistor pullup/ pulldown resistor on the HD5 pin:
HD19/AD19/MTXD3§	L2			
HD18/AD18/MTXD2§	M4			HD5 pin = 0: HPI operates as an HPI16.
HD17/AD17/MTXD1§	M2			(HPI bus is 16 bits wide. HD[15:0] pins are used and the remaining HD[31:16] pins are reserved pins in the high-impedance state.)
HD16/AD16/MTXD0§	M3	I/O/Z		
HD15/AD15§	T3			HD5 pin = 1: HPI operates as an HPI32.
HD14/AD14§	U1			(HPI bus is 32 bits wide. All HD[31:0] pins are used for host-port operations.)
HD13/AD13§	U3			
HD12/AD12§	U2			As PCI data-address bus (PCI_EN pin = 1)
HD11/AD11§	U4			• Used for transfer of data and address
HD10/AD10§	V1			
HD9/AD9§	V3			For superset devices like C6412, the HD31/AD31 through HD16/AD16 pins can also function as EMAC transmit/receive or control pins (when PCI_EN pin = 0; MAC_EN pin = 1). For more details on the EMAC pin functions, see the *Ethernet MAC (EMAC) peripheral* section of this table and for more details on how to configure the EMAC pin, see the device configuration section of this data sheet.
HD8/AD8§	V2			
HD7/AD7§	W2			
HD6/AD6§	W4			
HD5/AD5§	Y1			
HD4/AD4§	W3			
HD3/AD3§	Y2			
HD2/AD2§	Y4			
HD1/AD1§	AA1			
HD0/AD0§	Y3			

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground

‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)

§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.

¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.

A = Analog signal (PLL Filter)

‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

TEXAS INSTRUMENTS

Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
HOST-PORT INTERFACE (HPI) OR PERIPHERAL COMPONENT INTERCONNECT (PCI) OR EMAC (CONTINUED)				
$\overline{\text{PCBE0}}$	V4	I/O/Z		PCI command/byte enable 0 (**I/O/Z**). When PCI is disabled (PCI_EN = 0), this pin is tied-off.
GP0[15]/$\overline{\text{PRST}}$§	G3	I/O/Z		General-purpose input/output (GP0) 15 pin (**I/O/Z**) or PCI reset (**I**). No function at default.
XSP_CS	T4	O	IPD	PCI serial interface chip select (**O**). When PCI is disabled (PCI_EN = 0), this pin is tied-off.
XSP_CLK/MDCLK§	R5	I/O/Z	IPD	PCI serial interface clock (**O**) [default] or MDIO serial clock input/output (**I/O/Z**).
XSP_DI	R4	I	IPU	PCI serial interface data in (**I**) [default]. In PCI mode, this pin is connected to the output data pin of the serial PROM.
XSP_DO/MDIO§	P5	I/O/Z	IPU	PCI serial interface data out (**O**) [default] or MDIO serial data input/output (**I/O/Z**). In PCI mode, this pin is connected to the input data pin of the serial PROM.
GP0[14]/PCLK§	C1	I/O/Z		GP0 14 pin (**I/O/Z**) or PCI clock (**I**). No function at default.
GP0[13]/$\overline{\text{PINTA}}$§	G4			GP0 13 pin (**I/O/Z**) or PCI interrupt A (**O/Z**). No function at default.
GP0[12]/$\overline{\text{PGNT}}$§	H4			GP0 12 pin (**I/O/Z**) or PCI bus grant (**I**). No function at default.
GP0[11]/$\overline{\text{PREQ}}$§	F1			GP0 11 pin (**I/O/Z**) or PCI bus request (**O/Z**). No function at default.
GP0[10]/$\overline{\text{PCBE3}}$§	J2			GP0 10 pin (**I/O/Z**) or PCI command/byte enable 3 (**I/O/Z**). No function at default.
GP0[9]/PIDSEL§	K3			GP0 9 pin (**I/O/Z**) or PCI initialization device select (**I**). No function at default.
GP0[3]/PCIEEAI	L5	I/O/Z	IPD	GP0 3 pin (**I/O/Z**) and PCI EEPROM Auto-Initialization (EEAI). If the PCI peripheral is disabled (PCI_EN pin = 0), this pin must *not* be pulled up. 0 – PCI auto-initialization through EEPROM is disabled (default). 1 – PCI auto-initialization through EEPROM is enabled.
GP0[8]/$\overline{\text{PCI66}}$§	AD1	I/O/Z	IPD	This pin can be programmed as a GP0 8 pin (**I/O/Z**) or PCI frequency selection (PCI66). If the PCI peripheral is enabled (PCI_EN pin = 1), then: 0 – PCI operates at 66 MHz (default). 1 – PCI operates at 33 MHz. The -500 device supports PCI at 33 MHz only. For proper -500 device operation when the PCI peripheral is enabled (PCI_EN = 1), this pin *must* be pulled up with a 1-kΩ resistor at device reset. If the PCI peripheral is disabled (PCI_EN pin = 0), this pin must *not* be pulled up.
EMIFA (64-BIT) – CONTROL SIGNALS COMMON TO ALL TYPES OF MEMORY				
$\overline{\text{ACE3}}$	L26	O/Z	IPU	EMIFA memory space enables • Enabled by bits 28 through 31 of the word address • Only one pin is asserted during any external data access
$\overline{\text{ACE2}}$	K23	O/Z	IPU	
$\overline{\text{ACE1}}$	K24	O/Z	IPU	
$\overline{\text{ACE0}}$	K25	O/Z	IPU	

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground
‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)
§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.
¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.
# A = Analog signal (PLL Filter)
‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
EMIFA (64-BIT) – CONTROL SIGNALS COMMON TO ALL TYPES OF MEMORY (CONTINUED)
$\overline{ABE7}$	T22	O/Z	IPU	EMIFA byte-enable control • Decoded from the low-order address bits. The number of address bits or byte enables used depends on the width of external memory. • Byte-write enables for most types of memory • Can be directly connected to SDRAM read and write mask signal (SDQM)
$\overline{ABE6}$	T23	O/Z	IPU	
$\overline{ABE5}$	R25	O/Z	IPU	
$\overline{ABE4}$	R26	O/Z	IPU	
$\overline{ABE3}$	M25	O/Z	IPU	
$\overline{ABE2}$	M26	O/Z	IPU	
$\overline{ABE1}$	L23	O/Z	IPU	
$\overline{ABE0}$	L24	O/Z	IPU	
$\overline{APDT}$	M22	O/Z	IPU	EMIFA peripheral data transfer, allows direct transfer between external peripherals
EMIFA (64-BIT) – BUS ARBITRATION☆
$\overline{AHOLDA}$	N22	O	IPU	EMIFA hold-request-acknowledge to the host
$\overline{AHOLD}$	W24	I	IPU	EMIFA hold request from the host
ABUSREQ	P22	O	IPU	EMIFA bus request output
EMIFA (64-BIT) – ASYNCHRONOUS/SYNCHRONOUS MEMORY CONTROL
AECLKIN	H25	I	IPD	EMIFA external input clock. The EMIFA input clock (AECLKIN, CPU/4 clock, or CPU/6 clock) is selected at reset via the pullup/pulldown resistors on the AEA[20:19] pins. AECLKIN is the default for the EMIFA input clock.
AECLKOUT2	J23	O/Z	IPD	EMIFA output clock 2. Programmable to be EMIFA input clock (AECLKIN, CPU/4 clock, or CPU/6 clock) frequency divided-by-1, -2, or -4.
AECLKOUT1	J26	O/Z	IPD	EMIFA output clock 1 [at EMIFA input clock (AECLKIN, CPU/4 clock, or CPU/6 clock) frequency].

$\overline{AARE}$/ $\overline{ASDCAS}$/ $\overline{ASADS}$/$\overline{ASRE}$	J25	O/Z	IPU	EMIFA asynchronous memory read-enable/SDRAM column-address strobe/programmable synchronous interface-address strobe or read-enable
				• For programmable synchronous interface, the RENEN field in the CE Space Secondary
				  Control Register (CExSEC) selects between $\overline{ASADS}$ and $\overline{ASRE}$:
				  If RENEN = 0, then the $\overline{ASADS}$/$\overline{ASRE}$ signal functions as the $\overline{ASADS}$ signal.
				  If RENEN = 1, then the $\overline{ASADS}$/$\overline{ASRE}$ signal functions as the $\overline{ASRE}$ signal.

$\overline{AAOE}$/ $\overline{ASDRAS}$/ $\overline{ASOE}$	J24	O/Z	IPU	EMIFA asynchronous memory output-enable/SDRAM row-address strobe/programmable synchronous interface output-enable
$\overline{AAWE}$/ $\overline{ASDWE}$/ $\overline{ASWE}$	K26	O/Z	IPU	EMIFA asynchronous memory write-enable/SDRAM write-enable/programmable synchronous interface write-enable
ASDCKE	L25	O/Z	IPU	EMIFA SDRAM clock-enable (used for self-refresh mode). [EMIFA module only.] • If SDRAM is not in system, ASDCKE can be used as a general-purpose output.
$\overline{ASOE3}$	R22	O/Z	IPU	EMIFA synchronous memory output-enable for $\overline{ACE3}$ (for glueless FIFO interface)
AARDY	L22	I	IPU	Asynchronous memory ready input

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground

‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)

§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.

¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.

A = Analog signal (PLL Filter)

‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

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Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
EMIFA (64-BIT) – ADDRESS				

AEA22	U23	O/Z	IPD	EMIFA external address (doubleword address)
				Note: EMIFA address numbering for the C6412 device starts with AEA3 to maintain signal name compatibility with other C64x™ devices (e.g., C6414, C6415, and C6416) [see the 64-bit EMIF addressing scheme in the *TMS320C6000 DSP External Memory Interface (EMIF) Reference Guide* (literature number SPRU266)].
				
				• Also controls initialization of DSP modes at reset (**I**) via pullup/pulldown resistors
				  – Boot mode (AEA[22:21]):
				    00 – No boot (default mode)
				    01 – HPI/PCI boot (based on PCI_EN pin)
				    10 – Reserved
				    11 – EMIFA 8–bit ROM boot
				
				  – EMIF clock select
				
				  – AEA[20:19]: Clock mode select for EMIFA (AECLKIN_SEL[1:0])
				    00 – AECLKIN (default mode)
				    01 – CPU/4 Clock Rate
				    10 – CPU/6 Clock Rate
				    11 – Reserved
				
				For more details, see the Device Configurations section of this data sheet.

AEA21	V24			
AEA20	V25			
AEA19	V26			
AEA18	V23			
AEA17	U24			
AEA16	U25			
AEA15	U26			
AEA14	T24			
AEA13	T25			
AEA12	R23			
AEA11	R24			
AEA10	P23			
AEA9	P24			
AEA8	P26			
AEA7	N23			
AEA6	N24			
AEA5	N26			
AEA4	M23			
AEA3	M24			
EMIFA (64-BIT) – DATA				
AED63	AF24	I/O/Z	IPU	EMIFA external data
AED62	AF23			
AED61	AE23			
AED60	AD23			
AED59	AD22			
AED58	AE22			
AED57	AD21			
AED56	AE21			
AED55	AC21			
AED54	AF21			
AED53	AD20			
AED52	AE20			
AED51	AC20			

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‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)

§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.

¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.

A = Analog signal (PLL Filter)

‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
EMIFA (64-BIT) – DATA (CONTINUED)				
AED50	AF20			
AED49	AC19			
AED48	AD19			
AED47	W23			
AED46	Y26			
AED45	Y23			
AED44	Y25			
AED43	Y24			
AED42	AA26			
AED41	AA23			
AED40	AA25			
AED39	AA24			
AED38	AB23			
AED37	AB25			
AED36	AB24			
AED35	AC26			
AED34	AC25	I/O/Z	IPU	EMIFA external data
AED33	AD25			
AED32	AD26			
AED31	C26			
AED30	C25			
AED29	D26			
AED28	D25			
AED27	E24			
AED26	E25			
AED25	F24			
AED24	F25			
AED23	F23			
AED22	F26			
AED21	G24			
AED20	G25			
AED19	G23			
AED18	G26			
AED17	H23			

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‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)

§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.

¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.

A = Analog signal (PLL Filter)

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TEXAS INSTRUMENTS

Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
EMIFA (64-BIT) – DATA (CONTINUED)				
AED16	H24			
AED15	C19			
AED14	D19			
AED13	A20			
AED12	D20			
AED11	B20			
AED10	C20			
AED9	A21			
AED8	D21	I/O/Z	IPU	EMIFA external data
AED7	B21			
AED6	C21			
AED5	A23			
AED4	C22			
AED3	B22			
AED2	B23			
AED1	A24			
AED0	B24			
MANAGEMENT DATA INPUT/OUTPUT (MDIO)				
XSP_CLK/MDCLK§	R5	I/O/Z	IPD	PCI serial interface clock (**O**) [default] or MDIO serial clock input/output (**I/O/Z**).
XSP_DO/MDIO§	P5	I/O/Z	IPU	PCI serial interface data out (**O**) [default] or MDIO serial data input/output (**I/O/Z**). In PCI mode, this pin is connected to the input data pin of the serial PROM.
TIMER 2				
	–			No external pins. The timer 2 peripheral pins are ***not*** pinned out as external pins.
TIMER 1				
TOUT1/LENDIAN	B5	O/Z	IPU	Timer 1 output (**O/Z**) or device endian mode (**I**). Also controls initialization of DSP modes at reset via pullup/pulldown resistors – Device Endian mode 0 – Big Endian 1 – Little Endian (default) For more details on LENDIAN, see the Device Configurations section of this data sheet.
TINP1	A5	I	IPD	Timer 1 or general-purpose input

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§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.

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Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
				TIMER 0
TOUT0/MAC_EN	C5	O/Z	IPD	Timer 0 output (**O/Z**) or MAC enable select bit (**I**) MAC enable pin. This pin and the MAC_EN pin control the selection (enable/disable) of the HPI, EMAC, MDIO, and GP0[15:9], or PCI peripherals. The pins work in conjunction to enable/disable these peripherals (for more details, see the Device Configurations section of this data sheet). For more details, see the Device Configurations section of this data sheet.
TINP0	A4	I	IPD	Timer 0 or general-purpose input
				INTER-INTEGRATED CIRCUIT 0 (I2C0)
SCL0	E4	I/O/Z	—	I2C0 clock.
SDA0	D3	I/O/Z	—	I2C0 data.
				MULTICHANNEL BUFFERED SERIAL PORT 1 (McBSP1)
CLKR1	AD8	I/O/Z	IPD	McBSP1 receive clock (**I/O/Z**)
FSR1	AC7	I/O/Z	IPD	McBSP1 receive frame sync (**I/O/Z**)
DR1	AD7	I	IPD	McBSP1 receive data (**I**)
CLKS1	AE7	I	IPD	McBSP1 external clock source (**I**) (as opposed to internal)
DX1	AC6	I/O/Z	IPD	McBSP1 transmit data (**O/Z**)
FSX1	AD6	I/O/Z	IPD	McBSP1 transmit frame sync (**I/O/Z**)
CLKX1	AE6	I/O/Z	IPD	McBSP1 transmit clock (**I/O/Z**)
				MULTICHANNEL BUFFERED SERIAL PORT 0 (McBSP0)
CLKR0	AE15	I/O/Z	IPD	McBSP0 receive clock (**I/O/Z**)
FSR0	AB16	I/O/Z	IPD	McBSP0 receive frame sync (**I/O/Z**)
DR0	AC16	I	IPD	McBSP0 receive data (**I**)
CLKS0	AD16	I	IPD	McBSP0 external clock source (**I**) (as opposed to internal)
DX0	AE16	O/Z	IPD	McBSP0 transmit data (**O/Z**)
FSX0	AF16	I/O/Z	IPD	McBSP0 transmit frame sync (**I/O/Z**)
CLKX0	AF17	I/O/Z	IPD	McBSP0 transmit clock (**I/O/Z**)

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§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.
¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.
# A = Analog signal (PLL Filter)
‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

TEXAS INSTRUMENTS

Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
ETHERNET MAC (EMAC)				
HD31/AD31/MRCLK§	G1	I		EMAC Media Independent I/F (MII) data, clocks, and control pins for Transmit/Receive. MII transmit clock (MTCLK), Transmit clock source from the attached PHY. MII transmit data (MTXD[3:0]), Transmit data nibble synchronous with transmit clock (MTCLK). MII transmit enable (MTXEN), This signal indicates a valid transmit data on the transmit data pins (MTDX[3:0]). MII collision sense (MCOL) Assertion of this signal during half-duplex operation indicates network collision. During full-duplex operation, transmission of new frames will not begin if this pin is asserted. MII carrier sense (MCRS) Indicates a frame carrier signal is being received. MII receive data (MRXD[3:0]), Receive data nibble synchronous with receive clock (MRCLK). MII receive clock (MRCLK), Receive clock source from the attached PHY. MII receive data valid (MRXDV), This signal indicates a valid data nibble on the receive data pins (MRDX[3:0]). and MII receive error (MRXER), Indicates reception of a coding error on the receive data.
HD30/AD30/MCRS§	H3	I		
HD29/AD29/MRXER§	G2	I		
HD28/AD28/MRXDV§	J4	I		
HD27/AD27/MRXD3§	H2	I		
HD26/AD26/MRXD2§	J3	I		
HD25/AD25/MRXD1§	J1	I		
HD24/AD24/MRXD0§	K4	I		
HD22/AD22/MTCLK§	L4	I		
HD21/AD21/MCOL§	K2	I		
HD20/AD20/MTXEN§	L3	O/Z		
HD19/AD19/MTXD3§	L2	O/Z		
HD18/AD18/MTXD2§	M4	O/Z		
HD17/AD17/MTXD1§	M2	O/Z		
HD16/AD16/MTXD0§	M3	O/Z		
RESERVED FOR TEST				
RSV07	H7	A	—	Reserved. This pin must be connected directly to CV$_{DD}$ for proper device operation.
RSV08	R6	A	—	Reserved. This pin must be connected directly to DV$_{DD}$ for proper device operation.
RSV46	A7			Reserved. This pin must be connected directly to CV$_{DD}$ for proper device operation.
RSV47	A13			
RSV48	B8			
RSV53	A9			
RSV57	A10			
RSV61	A11			

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‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)

§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.

¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.

A = Analog signal (PLL Filter)

‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
RESERVED FOR TEST (CONTINUED)				
RSV00	AA3	A	—	
RSV01	AB3	I	—	
RSV02	AC4	O/Z	—	
RSV05	E14	I	IPD	
RSV06	W7	A	—	
RSV09	AC1			
RSV19	AB15			
RSV22	AB14			
RSV25	AB13			
RSV35	AC8			
RSV44	AB11			
RSV45	AB12			
RSV49	D7			
RSV50	C7			
RSV51	C8			
RSV52	D8			Reserved (leave unconnected, *do not* connect to power or ground)
RSV54	B9			
RSV55	C9			
RSV56	D9			
RSV58	B10			
RSV59	C10			
RSV60	D10			
RSV62	B11			
RSV63	C11			
RSV64	D11			
RSV65	E11			
RSV66	B12			
RSV67	C12			
RSV68	D12			
RSV69	E12			
RSV70	E13			

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‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)

§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.

¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.

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Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
RESERVED FOR TEST (CONTINUED)				
RSV38	AC9			
RSV40	AC10			
RSV43	AC11			
RSV27	AC12			
RSV24	AC13			
RSV21	AC14			
RSV18	AC15			
RSV13	AC17			
RSV03	AD3	O/Z	—	
RSV32	AD5			
RSV37	AD9			
RSV39	AD10			
RSV42	AD11			
RSV26	AD12			
RSV23	AD13			
RSV20	AD14			
RSV17	AD15			Reserved (leave unconnected, ***do not*** connect to power or ground)
RSV14	AD17			
RSV31	AE5			
RSV36	AE9			
RSV41	AE11			
RSV12	AE17			
RSV16	AE18			
RSV04	AF3	O	IPU	
RSV30	AF4			
RSV33	AF5			
RSV34	AF6			
RSV28	AF8			
RSV29	AF10			
RSV11	AF12			
RSV10	AF14			
RSV15	AF18			

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground
‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)
§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.
¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.
# A = Analog signal (PLL Filter)
‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
SUPPLY VOLTAGE PINS				
DV$_{DD}$	A2	S		3.3-V supply voltage (see the Power-Supply Decoupling section of this data manual)
	A25			
	B1			
	B2			
	B14			
	B25			
	B26			
	C3			
	C24			
	D4			
	D23			
	E5			
	E7			
	E8			
	E10			
	E17			
	E19			
	E20			
	E22			
	F9			
	F12			
	F15			
	F18			
	G5			
	G22			

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground
‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)
§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.
¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.
# A = Analog signal (PLL Filter)
‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

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Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
SUPPLY VOLTAGE PINS (CONTINUED)				
DV$_{DD}$	H5	S		3.3-V supply voltage (see the Power-Supply Decoupling section of this data manual)
	H22			
	J6			
	J21			
	K5			
	K22			
	M6			
	M21			
	N2			
	P25			
	R21			
	U5			
	U22			
	V21			
	W5			
	W22			
	W25			
	Y5			
	Y22			
	AA9			
	AA12			
	AA15			
	AA18			
	AB5			
	AB7			
	AB8			
	AB10			
	AB17			
	AB19			
	AB20			
	AB22			
	AC23			
	AD24			
	AE1			

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground

‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)

§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.

¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.

A = Analog signal (PLL Filter)

‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE[†]	IPD/ IPU[‡]	DESCRIPTION
SUPPLY VOLTAGE PINS (CONTINUED)				
DV$_{DD}$	AE2	S		3.3-V supply voltage (see the Power-Supply Decoupling section of this data manual)
	AE13			
	AE25			
	AE26			
	AF2			
	AF25			
CV$_{DD}$	F6	S		1.2-V supply voltage (-500 device) 1.4 V supply voltage (A-500, A–600, -600, -720 devices) (see the Power-Supply Decoupling section of this data manual)
	F7			
	F20			
	F21			
	G6			
	G7			
	G8			
	G10			
	G11			
	G13			
	G14			
	G16			
	G17			
	G19			
	G20			
	G21			
	H20			
	K7			
	K20			
	L7			
	L20			
	M12			
	M14			

[†] I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground

[‡] IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)

[§] These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.

[¶] PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.

[#] A = Analog signal (PLL Filter)

[||] The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

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Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
				SUPPLY VOLTAGE PINS (CONTINUED)
CV_{DD}	N7			
	N13			
	N15			
	N20			
	P7			
	P12			
	P14			
	P20			
	R13			
	R15			
	T7			
	T20			
	U7			
	U20			
	W20			
	Y6	S		1.2-V supply voltage (-500 device)
	Y7			1.4 V supply voltage (A-500, A–600, -600, -720 devices)
	Y8			(see the Power-Supply Decoupling section of this data manual)
	Y10			
	Y11			
	Y13			
	Y14			
	Y16			
	Y17			
	Y19			
	Y20			
	Y21			
	AA6			
	AA7			
	AA20			
	AA21			

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground

‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)

§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data manual.

¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.

A = Analog signal (PLL Filter)

‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
				GROUND PINS
V$_{SS}$	A1	GND		Ground pins
	A3			
	A6			
	A8			
	A12			
	A14			
	A19			
	A22			
	A26			
	B3			
	B6			
	B7			
	B13			
	B19			
	C2			
	C4			
	C13			
	C18			
	C23			
	D1			
	D2			
	D5			
	D13			
	D18			
	D22			
	D24			
	E3			
	E6			
	E9			
	E16			
	E18			
	E21			
	E23			

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground
‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)
§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data sheet.
¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.
# A = Analog signal (PLL Filter)
‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

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Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
		GROUND PINS (CONTINUED)		
V$_{SS}$	E26	GND		Ground pins
	F5			
	F8			
	F10			
	F11			
	F13			
	F14			
	F16			
	F17			
	F19			
	F22			
	G9			
	G12			
	G15			
	G18			
	H1			
	H6			
	H21			
	H26			
	J5			
	J7			
	J20			
	J22			
	K6			
	K21			
	L1			
	L6			
	L21			
	M7			
	M13			
	M15			
	M20			
	N5			
	N6			

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground

‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)

§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data sheet.

¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.

A = Analog signal (PLL Filter)

‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
GROUND PINS (CONTINUED)				
V$_{SS}$	N12	GND		Ground pins
	N14			
	N21			
	N25			
	P2			
	P6			
	P13			
	P15			
	P21			
	R7			
	R12			
	R14			
	R20			
	T1			
	T5			
	T6			
	T21			
	T26			
	U6			
	U21			
	V5			
	V7			
	V20			
	V22			
	W1			
	W6			
	W21			
	W26			
	Y9			
	Y12			
	Y15			
	Y18			

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground
‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)
§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data sheet.
¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.
# A = Analog signal (PLL Filter)
‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
	GROUND PINS (CONTINUED)			
V_{SS}	AA4	GND		Ground pins
	AA5			
	AA8			
	AA10			
	AA11			
	AA13			
	AA14			
	AA16			
	AA17			
	AA19			
	AA22			
	AB1			
	AB2			
	AB4			
	AB6			
	AB9			
	AB18			
	AB21			
	AB26			
	AC3			
	AC5			
	AC18			
	AC22			
	AC24			
	AD2			
	AD4			
	AD18			
	AE3			
	AE8			
	AE10			
	AE12			

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground

‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)

§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data sheet.

¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.

A = Analog signal (PLL Filter)

|| The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

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Table 2–10. Terminal Functions (Continued)

SIGNAL NAME	GDK/ GNZ	TYPE†	IPD/ IPU‡	DESCRIPTION
			GROUND PINS (CONTINUED)	
V_{SS}	AE14	GND		Ground pins
	AE19			
	AE24			
	AF1			
	AF7			
	AF9			
	AF11			
	AF13			
	AF15			
	AF19			
	AF22			
	AF26			

† I = Input, O = Output, Z = High impedance, S = Supply voltage, GND = Ground

‡ IPD = Internal pulldown, IPU = Internal pullup. (These IPD/IPU signal pins feature a 30-kΩ IPD or IPU resistor. To pull up a signal to the opposite supply rail, a 1-kΩ resistor should be used.)

§ These pins are multiplexed pins. For more details, see the Device Configurations section of this data sheet.

¶ PLLV is not part of external voltage supply. See the Clock PLL section for information on how to connect this pin.

A = Analog signal (PLL Filter)

‖ The EMU0 and EMU1 pins are internally pulled up with 30-kΩ resistors; therefore, for emulation and normal operation, no external pullup/pulldown resistors are necessary. However, for boundary scan operation, pull down the EMU1 and EMU0 pins with a dedicated 1-kΩ resistor.

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2.11 Development Support

TI offers an extensive line of development tools for the TMS320C6000™ DSP platform, including tools to evaluate the performance of the processors, generate code, develop algorithm implementations, and fully integrate and debug software and hardware modules.

The following products support development of C6000™ DSP-based applications:

Software Development Tools:
Code Composer Studio™ Integrated Development Environment (IDE): including Editor
C/C++/Assembly Code Generation, and Debug plus additional development tools
Scalable, Real-Time Foundation Software (DSP/BIOS™), which provides the basic run-time target software needed to support any DSP application.

Hardware Development Tools:
Extended Development System (XDS™) Emulator (supports C6000™ DSP multiprocessor system debug)
EVM (Evaluation Module)

For a complete listing of development-support tools for the TMS320C6000™ DSP platform, visit the Texas Instruments web site on the Worldwide Web at http://www.ti.com uniform resource locator (URL). For information on pricing and availability, contact the nearest TI field sales office or authorized distributor.

Code Composer Studio, DSP/BIOS, XDS, and TMS320 are trademarks of Texas Instruments.

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2.12 Device Support

2.12.1 Device and Development-Support Tool Nomenclature

To designate the stages in the product development cycle, TI assigns prefixes to the part numbers of all DSP devices and support tools. Each DSP commercial family member has one of three prefixes: TMX, TMP, or TMS (e.g., **TMS**320C6412AGDK6). Texas Instruments recommends two of three possible prefix designators for its support tools: TMDX and TMDS. These prefixes represent evolutionary stages of product development from engineering prototypes (TMX/TMDX) through fully qualified production devices/tools (TMS/TMDS).

Device development evolutionary flow:

TMX   Experimental device that is not necessarily representative of the final device's electrical specifications

TMP   Final silicon die that conforms to the device's electrical specifications but has not completed quality and reliability verification

TMS   Fully qualified production device

Support tool development evolutionary flow:

TMDX   Development-support product that has not yet completed Texas Instruments internal qualification testing.

TMDS   Fully qualified development-support product

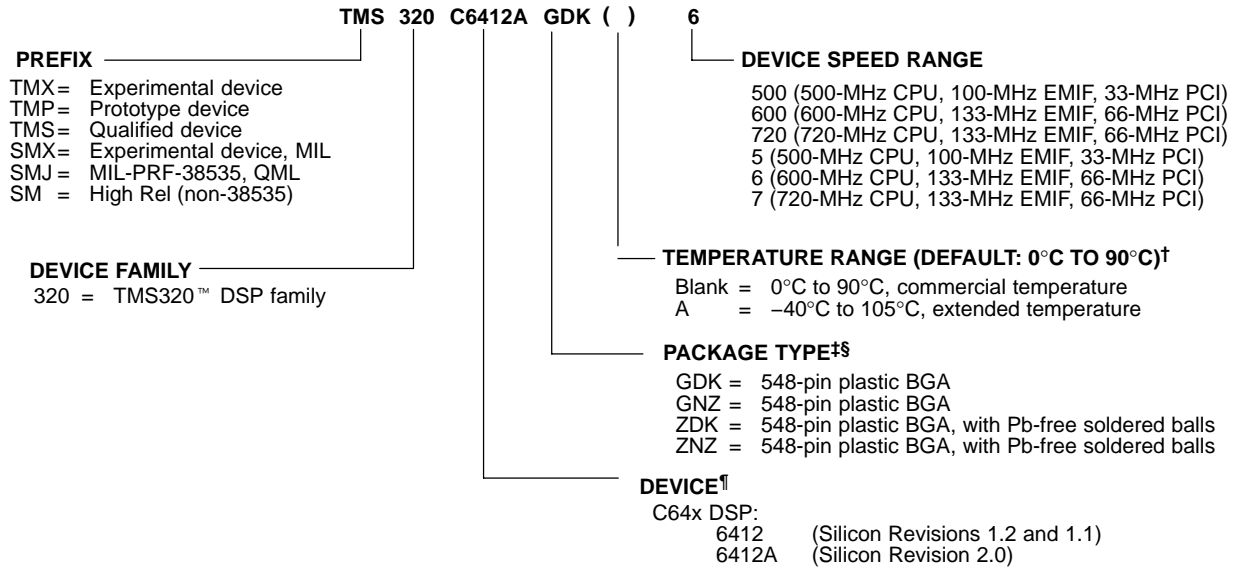TMX and TMP devices and TMDX development-support tools are shipped against the following disclaimer:

"Developmental product is intended for internal evaluation purposes."

TMS devices and TMDS development-support tools have been characterized fully, and the quality and reliability of the device have been demonstrated fully. TI's standard warranty applies.

Predictions show that prototype devices (TMX or TMP) have a greater failure rate than the standard production devices. Texas Instruments recommends that these devices not be used in any production system because their expected end-use failure rate still is undefined. Only qualified production devices are to be used.

TI device nomenclature also includes a suffix with the device family name. This suffix indicates the package type (for example, GDK), the temperature range (for example, blank is the default commercial temperature range), and the device speed range in megahertz (for example, -6 is 600 MHz). Figure 2–7 provides a legend for reading the complete device name for any DSP platform member.

The ZDK package, like the GDK package, is a 548-ball plastic BGA *only* with Pb-free balls. The ZNZ package is the Pb–free version of the GNZ package. For device part numbers and further ordering information for TMS320C6412 in the GDK, GNZ, ZDK and ZNZ package types, see the TI website (http://www.ti.com) or contact your TI sales representative.

Texas Instruments

TMS 320 C6412A GDK ( )   6

PREFIX
TMX = Experimental device
TMP = Prototype device
TMS = Qualified device
SMX = Experimental device, MIL
SMJ = MIL-PRF-38535, QML
SM = High Rel (non-38535)

DEVICE FAMILY
320 = TMS320™ DSP family

DEVICE SPEED RANGE

500 (500-MHz CPU, 100-MHz EMIF, 33-MHz PCI)
600 (600-MHz CPU, 133-MHz EMIF, 66-MHz PCI)
720 (720-MHz CPU, 133-MHz EMIF, 66-MHz PCI)
5 (500-MHz CPU, 100-MHz EMIF, 33-MHz PCI)
6 (600-MHz CPU, 133-MHz EMIF, 66-MHz PCI)
7 (720-MHz CPU, 133-MHz EMIF, 66-MHz PCI)

TEMPERATURE RANGE (DEFAULT: 0°C TO 90°C)[†]
Blank = 0°C to 90°C, commercial temperature
A   = −40°C to 105°C, extended temperature

PACKAGE TYPE[‡][§]
GDK = 548-pin plastic BGA
GNZ = 548-pin plastic BGA
ZDK = 548-pin plastic BGA, with Pb-free soldered balls
ZNZ = 548-pin plastic BGA, with Pb-free soldered balls

DEVICE[¶]
C64x DSP:
6412   (Silicon Revisions 1.2 and 1.1)
6412A   (Silicon Revision 2.0)

[†] The extended temperature "A version" devices may have different operating conditions than the commercial temperature devices. For more details, see the recommended operating conditions portion of this data sheet.
[‡] BGA = Ball Grid Array
[§] The ZDK and ZNZ mechanical package designators represent the version of the GDK and GNZ packages with Pb-free balls. For more detailed information, see the *Mechanical Data* section of this document.
[¶] For actual device part numbers (P/Ns) and ordering information, see the TI website (www.ti.com).

Figure 2–7. TMS320C64x™ DSP Device Nomenclature (Including the TMS320C6412 Device)

For additional information, see the *TMS320C6412 Digital Signal Processor Silicon Errata (*literature number SPRZ199).

2.12.2 *Documentation Support*

Extensive documentation supports all TMS320™ DSP family generations of devices from product announcement through applications development. The types of documentation available include: data sheets, such as this document, with design specifications; complete user's reference guides for all devices and tools; technical briefs; development-support tools; on-line help; and hardware and software applications. The following is a brief, descriptive list of support documentation specific to the C6000™ DSP devices:

The *TMS320C6000 CPU and Instruction Set Reference Guide* (literature number SPRU189) describes the C6000™ DSP CPU (core) architecture, instruction set, pipeline, and associated interrupts.

The *TMS320C6000 DSP Peripherals Overview Reference Guide* (literature number SPRU190) provides an overview and briefly describes the functionality of the peripherals available on the C6000™ DSP platform of devices. This document also includes a table listing the peripherals available on the C6000 devices along with literature numbers and hyperlinks to the associated peripheral documents.

The *TMS320C64x Technical Overview* (literature number SPRU395) gives an introduction to the C64x™ digital signal processor, and discusses the application areas that are enhanced by the C64x™ DSP VelociTI.2™ VLIW architecture.

TMS320C6000 DSP Inter-Integrated Circuit (I2C) Module Peripheral Reference Guide (literature number SPRU175) describes the functionality of the I^{2}C peripheral.

TMS320C6000 DSP Ethernet Media Access Controller (EMAC)/ Management Data Input/Output (MDIO) Module Reference Guide (literature number SPRU628) describes the functionality of the EMAC and MDIO peripherals.

The *TMS320C6412 Digital Signal Processor Silicon Errata* (literature number SPRZ199) describes the known exceptions to the functional specifications for particular silicon revisions of the TMS320C6412 device.

The *TMS320C6412 Power Consumption Summary* application report (literature number SPRA967) discusses the power consumption for user applications with the TMS320C6412 DSP device.

The *TMS320C6412 Hardware Designer's Resource Guide* (literature number SPRAA34) is organized by development flow and functional areas to make design efforts as seamless as possible. This document includes getting started, board design, system testing, and checklists to aid in initial designs and debug efforts. Each section of this document includes pointers to valuable information including: technical documentation, models, symbols, and reference designs for use in each phase of design. Particular attention is given to peripheral interfacing and system-level design concerns.

The *Using IBIS Models for Timing Analysis* application report (literature number SPRA839) describes how to properly use IBIS models to attain accurate timing analysis for a given system.

The tools support documentation is electronically available within the Code Composer Studio™ Integrated Development Environment (IDE). For a complete listing of C6000™ DSP latest documentation, visit the Texas Instruments web site on the Worldwide Web at http://www.ti.com uniform resource locator (URL).

2.12.2.1 *Device Silicon Revision*

The device silicon revision can be determined by the "Die PG code" marked on the top of the package. For more detailed information on the DM642 silicon revision, package markings, and the known exceptions to the functional specifications as well as any usage notes, refer to the device-specific silicon errata: *TMS320C6412 Digital Signal Processor Silicon Errata* (literature number SPRZ199).
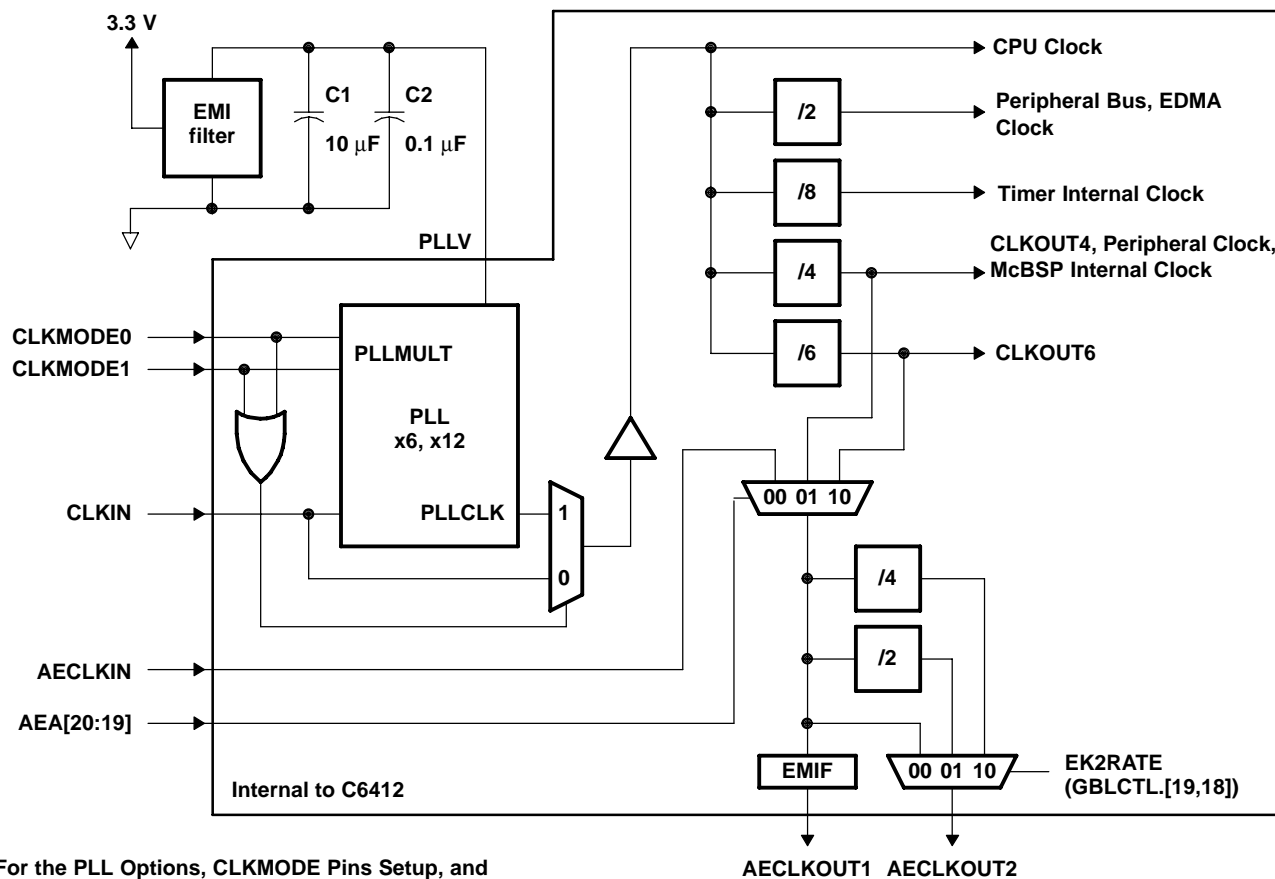
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2.13 Clock PLL

Most of the internal C64x™ DSP clocks are generated from a single source through the CLKIN pin. This source clock either drives the PLL, which multiplies the source clock frequency to generate the internal CPU clock, or bypasses the PLL to become the internal CPU clock.

To use the PLL to generate the CPU clock, the external PLL filter circuit must be properly designed. Figure 2–8 shows the external PLL circuitry for either x1 (PLL bypass) or other PLL multiply modes.

To minimize the clock jitter, a single clean power supply should power both the C64x™ DSP device and the external clock oscillator circuit. The minimum CLKIN rise and fall times should also be observed. For the input clock timing requirements, see the *input and output clocks* electricals section.

Rise/fall times, duty cycles (high/low pulse durations), and the load capacitance of the external clock source must meet the DSP requirements in this data sheet (see the *electrical characteristics over recommended ranges of supply voltage and operating case temperature* table and the *input and output clocks* electricals section).

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(For the PLL Options, CLKMODE Pins Setup, and
PLL Clock Frequency Ranges, see Table 9.)

NOTES: A. Place all PLL external components (C1, C2, and the EMI Filter) as close to the C6000™ DSP device as possible. For the best
performance, TI recommends that all the PLL external components be on a single side of the board without jumpers, switches, or
components other than the ones shown.
  B. For reduced PLL jitter, maximize the spacing between switching signals and the PLL external components (C1, C2, and the EMI
Filter).
  C. The 3.3-V supply for the EMI filter must be from the same 3.3-V power plane supplying the I/O voltage, DV_{DD}.
  D. EMI filter manufacturer TDK part number ACF451832-333, -223, -153, -103. Panasonic part number EXCCET103U.

Figure 2–8. External PLL Circuitry for Either PLL Multiply Modes or x1 (Bypass) Mode

Table 2–11. TMS320C6412 PLL Multiply Factor Options, Clock Frequency Ranges, and Typical Lock Time†‡

GDK and ZDK PACKAGES – 23 x 23 mm BGA, GNZ and ZNZ PACKAGES – 27 x 27 mm BGA							
CLKMODE1	CLKMODE0	CLKMODE (PLL MULTIPLY FACTORS)	CLKIN RANGE (MHz)	CPU CLOCK FREQUENCY RANGE (MHz)	CLKOUT4 RANGE (MHz)	CLKOUT6 RANGE (MHz)	TYPICAL LOCK TIME (μs)§
0	0	Bypass (x1)	30–75	30–75	7.5–18.8	5–12.5	N/A
0	1	x6	30–75	180–450	45–112.5	30–75	75
1	0	x12	30–60	360–720	90–180	60–120	
1	1	Reserved	–	–	–	–	–

† These clock frequency range values are applicable to a C6412–720 speed device. For –500 and -600 device speed values, see the CLKIN timing requirements table for the specific device speed.

‡ Use external pullup resistors on the CLKMODE pins (CLKMODE1 and CLKMODE0) to set the C6412 device to one of the valid PLL multiply clock modes (x6 or x12). With internal pulldown resistors on the CLKMODE pins (CLKMODE1, CLKMODE0), the default clock mode is x1 (bypass).

§ Under some operating conditions, the maximum PLL lock time may vary by as much as 150% from the specified typical value. For example, if the typical lock time is specified as 100 μs, the maximum value may be as long as 250 μs.

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2.14 I2C

The I2C module on the TMS320C6412 may be used by the DSP to control local peripherals ICs (DACs, ADCs, etc.) while the other may be used to communicate with other controllers in a system or to implement a user interface.

The I2C port supports:

- Compatible with Philips I2C Specification Revision 2.1 (January 2000)
- Fast Mode up to 400 Kbps (no fail-safe I/O buffers)
- Noise Filter to Remove Noise 50 ns or less
- Seven- and Ten-Bit Device Addressing Modes
- Master (Transmit/Receive) and Slave (Transmit/Receive) Functionality
- Events: DMA, Interrupt, or Polling
- Slew-Rate Limited Open-Drain Output Buffers

Figure 2–9 is a block diagram of the I2C0 module.
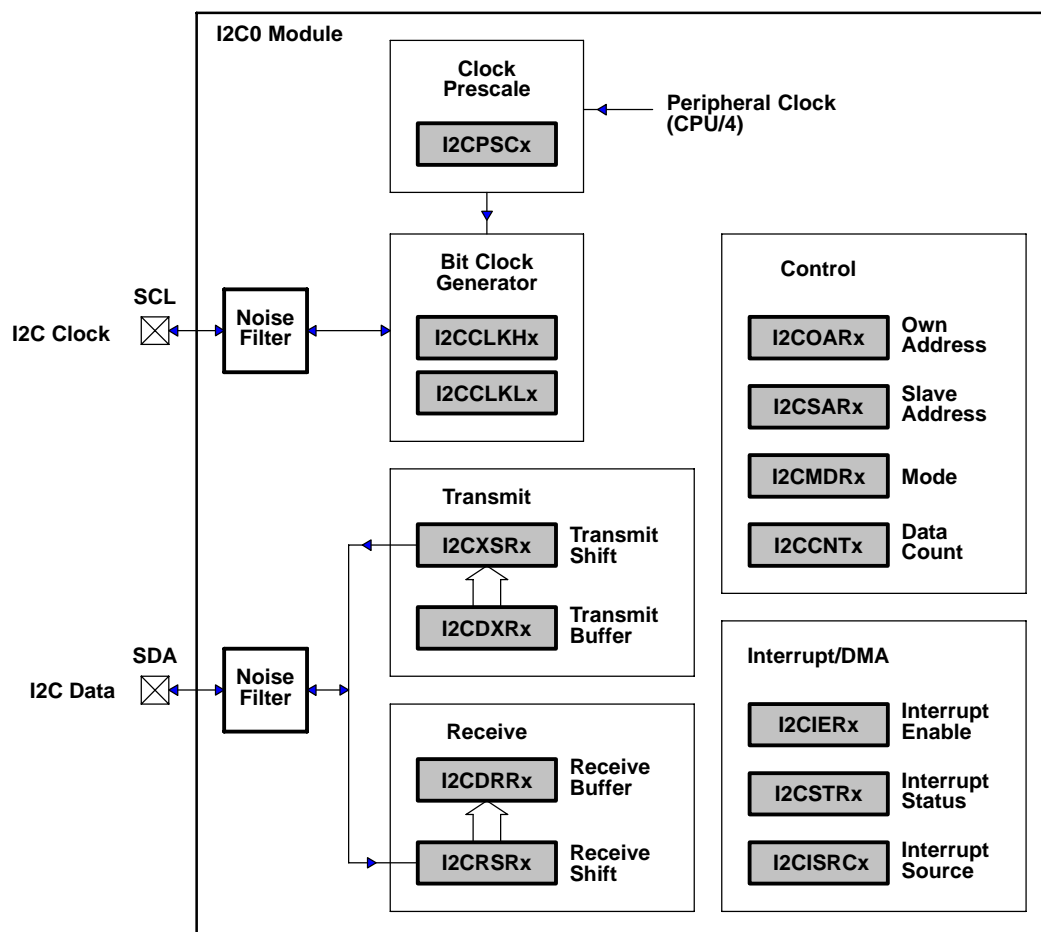
NOTE A: Shading denotes control/status registers.

Figure 2–9. I2C0 Module Block Diagram

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2.15 PCI

On the C6412 device, the PCI interface is multiplexed with the 32-bit Host Port Interface (HPI), or with a combination of 16-bit HPI and EMAC/MDIO. This provides the following flexibility options to the user:

- 32-bit 66 MHz PCI bus
- 32-bit HPI
- Combination of 16-bit HPI and EMAC/MDIO

The PCI port for the TMS320C6000 supports connection of the DSP to a PCI host via the integrated PCI master/slave bus interface. For the C64x devices, like the C6412, the PCI port interfaces to the DSP via the EDMA internal address generation hardware. This architecture allows for both PCI Master and Slave transactions, while keeping the EDMA channel resources available for other applications.

For more details on the PCI port peripheral module, see the *TMS320C6000 DSP Peripheral Component Interconnect (PCI) Reference Guide* (literature number SPRU581).

2.16 EMAC

The ethernet media access controller (EMAC) provides an efficient interface between the C6412 DSP core processor and the network. The C6412 EMAC support both 10Base-T and 100Base-TX, or 10 Mbits/second (Mbps) and 100 Mbps in either half- or full-duplex, with hardware flow control and quality of service (QOS) support. The C6412 EMAC makes use of a custom interface to the DSP core that allows efficient data transmission and reception.

The EMAC controls the flow of packet data from the DSP to the PHY. The MDIO module controls PHY configuration and status monitoring.

Both the EMAC and the MDIO modules interface to the DSP through a custom interface that allows efficient data transmission and reception. This custom interface is referred to as the EMAC control module, and is considered integral to the EMAC/MDIO peripheral. The control module is also used to control device reset, interrupts, and system priority.

The *TMS320C6000 DSP Ethernet Media Access Controller (EMAC) / Management Data Input/Output (MDIO) Module Reference Guide* (literature number SPRU628) describes the C6412 EMAC peripheral in detail. Some of the features documented in this peripheral reference guide are not supported on the C6412 at this time. The C6412 supports one receive channel and does not support receive quality of service (QOS). For a list of supported registers and register fields, see Table 1–9 [Ethernet MAC (EMAC) Control Registers] and Table 1–10 [EMAC Statistics Registers] in this data manual.

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2.17 MDIO

The management data input/output (MDIO) module continuously polls all 32 MDIO addresses in order to enumerate all PHY devices in the system.

The management data input/output (MDIO) module implements the 802.3 serial management interface to interrogate and control Ethernet PHY(s) using a shared two-wire bus. Host software uses the MDIO module to configure the auto-negotiation parameters of each PHY attached to the EMAC, retrieve the negotiation results, and configure required parameters in the EMAC module for correct operation. The module is designed to allow almost transparent operation of the MDIO interface, with very little maintenance from the core processor.

The *TMS320C6000 DSP Ethernet Media Access Controller (EMAC) / Management Data Input/Output (MDIO) Module Reference Guide* (literature number SPRU628) describes the C6412 MDIO peripheral in detail. Some of the features documented in this peripheral reference guide are not supported on the C6412 at this time. The C6412 only supports one EMAC module. For a list of supported registers and register fields, see Table 1–22 [MDIO Registers] in this data manual.

2.18 General-Purpose Input/Output (GPIO)

To use the GP[15:0] software-configurable GPIO pins, the GPxEN bits in the GP Enable (GPEN) Register and the GPxDIR bits in the GP Direction (GPDIR) Register must be properly configured.
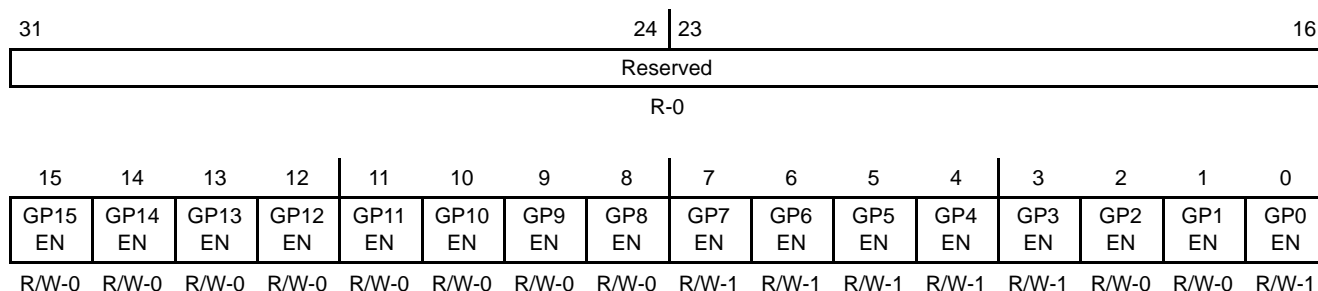
GPxEN = 1   GP[x] pin is enabled

GPxDIR = 0   GP[x] pin is an input

GPxDIR = 1   GP[x] pin is an output

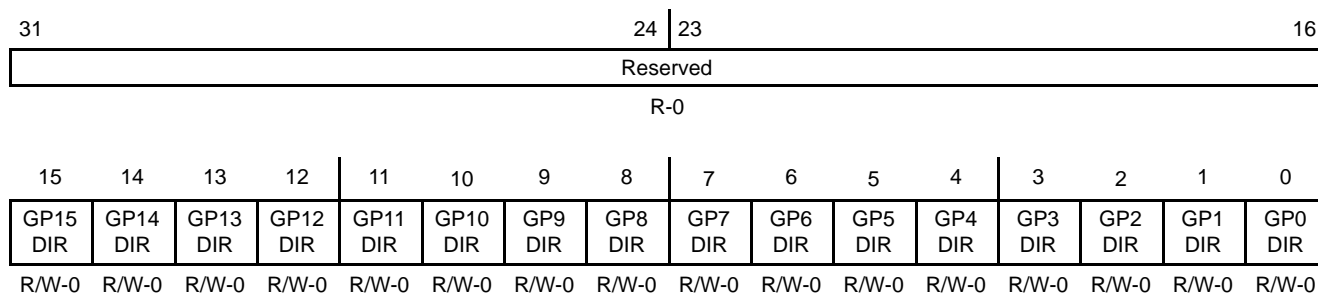where "x" represents one of the 15 through 0 GPIO pins

Figure 2–10 shows the GPIO enable bits in the GPEN register for the C6412 device. To use any of the GPx pins as general-purpose input/output functions, the corresponding GPxEN bit must be set to "1" (enabled). Default values are device-specific, so refer to Figure 2–10 for the C6412 default configuration.

31															16
Reserved															
R-0															

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
GP15 EN	GP14 EN	GP13 EN	GP12 EN	GP11 EN	GP10 EN	GP9 EN	GP8 EN	GP7 EN	GP6 EN	GP5 EN	GP4 EN	GP3 EN	GP2 EN	GP1 EN	GP0 EN
R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-0	R/W-0	R/W-1

Legend: R/W = Readable/Writeable; -*n* = value after reset, -x = undefined value after reset

Figure 2–10. GPIO Enable Register (GPEN) [Hex Address: 01B0 0000]

Figure 2–11 shows the GPIO direction bits in the GPDIR register. This register determines if a given GPIO pin is an input or an output providing the corresponding GPxEN bit is enabled (set to "1") in the GPEN register. By default, all the GPIO pins are configured as input pins.

31															16
Reserved															
R-0															

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
GP15 DIR	GP14 DIR	GP13 DIR	GP12 DIR	GP11 DIR	GP10 DIR	GP9 DIR	GP8 DIR	GP7 DIR	GP6 DIR	GP5 DIR	GP4 DIR	GP3 DIR	GP2 DIR	GP1 DIR	GP0 DIR
R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0

Legend: R/W = Readable/Writeable; -*n* = value after reset, -x = undefined value after reset

Figure 2–11. GPIO Direction Register (GPDIR) [Hex Address: 01B0 0004]

For more detailed information on general-purpose inputs/outputs (GPIOs), see the *TMS320C6000 DSP General-Purpose Input/Output (GPIO) Reference Guide* (literature number SPRU584).

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2.19 Power-Down Modes Logic

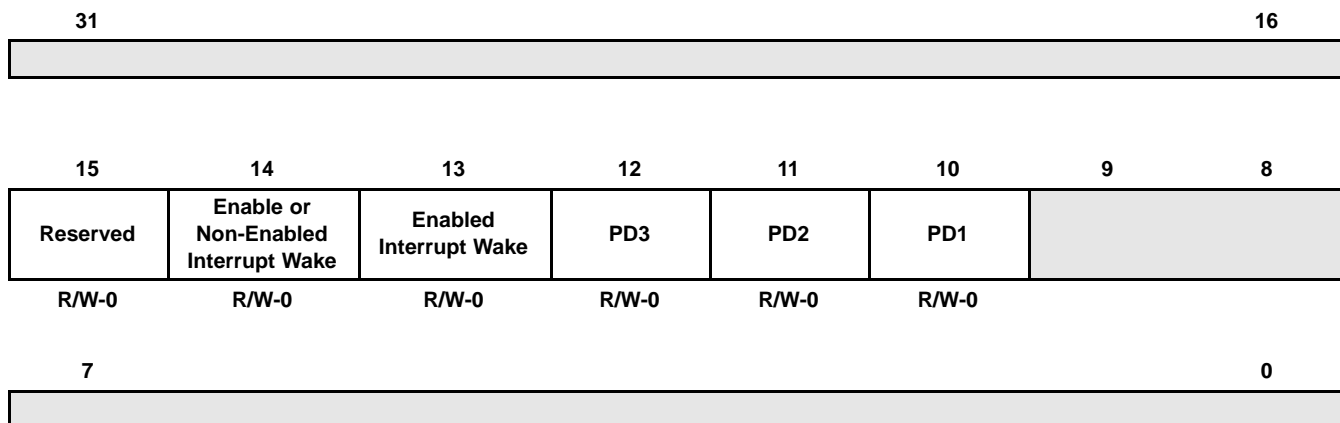Figure 2−12 shows the power-down mode logic on the C6412.
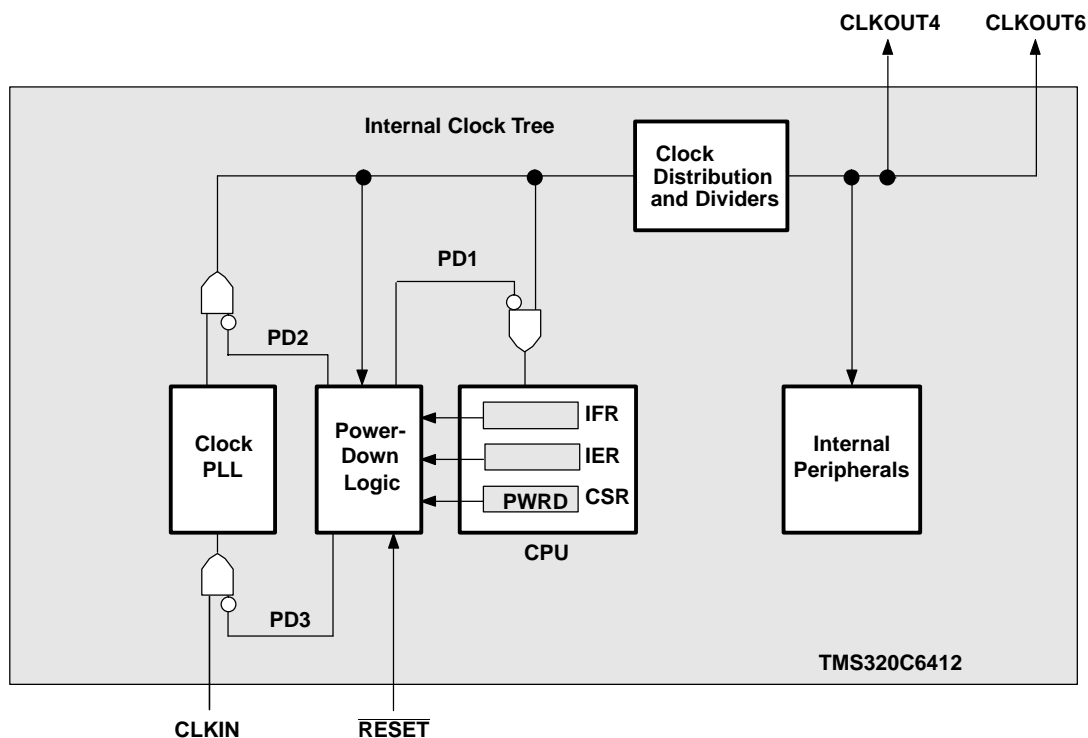
† External input clocks, with the exception of CLKIN, are *not* gated by the power-down mode logic.

Figure 2–12. Power-Down Mode Logic†

2.19.1 Triggering, Wake-up, and Effects

The power-down modes and their wake-up methods are programmed by setting the PWRD field (bits 15−10) of the control status register (CSR). The PWRD field of the CSR is shown in Figure 2−13 and described in Table 2−12. When writing to the CSR, all bits of the PWRD field should be set at the same time. Logic 0 should be used when writing to the reserved bit (bit 15) of the PWRD field. The CSR is discussed in detail in the *TMS320C6000 CPU and Instruction Set Reference Guide* (literature number SPRU189).

31							16

15	14	13	12	11	10	9	8
Reserved	Enable or Non-Enabled Interrupt Wake	Enabled Interrupt Wake	PD3	PD2	PD1		
R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0		

7							0

Legend: R/W–x = Read/write reset value

NOTE: The shadowed bits are not part of the power-down logic discussion and therefore are not covered here. For information on these other bit fields in the CSR register, see the *TMS320C6000 CPU and Instruction Set Reference Guide* (literature number SPRU189).

Figure 2–13. PWRD Field of the CSR Register

A delay of up to nine clock cycles may occur after the instruction that sets the PWRD bits in the CSR before the PD mode takes effect. As best practice, NOPs should be padded after the PWRD bits are set in the CSR to account for this delay.

If PD1 mode is terminated by a non-enabled interrupt, the program execution returns to the instruction where PD1 took effect. If PD1 mode is terminated by an enabled interrupt, the interrupt service routine will be executed first, then the program execution returns to the instruction where PD1 took effect. In the case with an enabled interrupt, the GIE bit in the CSR and the NMIE bit in the interrupt enable register (IER) must also be set in order for the interrupt service routine to execute; otherwise, execution returns to the instruction where PD1 took effect upon PD1 mode termination by an enabled interrupt.

PD2 and PD3 modes can only be aborted by device reset. Table 2–12 summarizes all the power-down modes.

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Table 2–12. Characteristics of the Power-Down Modes

PRWD FIELD (BITS 15–10)	POWER-DOWN MODE	WAKE-UP METHOD	EFFECT ON CHIP'S OPERATION
000000	No power-down	—	—
001001	PD1	Wake by an enabled interrupt	CPU halted (except for the interrupt logic) Power-down mode blocks the internal clock inputs at the boundary of the CPU, preventing most of the CPU's logic from switching. During PD1, EDMA transactions can proceed between peripherals and internal memory.
010001	PD1	Wake by an enabled or non-enabled interrupt	
011010	PD2[†]	Wake by a device reset	Output clock from PLL is halted, stopping the internal clock structure from switching and resulting in the entire chip being halted. All register and internal RAM contents are preserved. All functional I/O "freeze" in the last state when the PLL clock is turned off.
011100	PD3[†]	Wake by a device reset	Input clock to the PLL stops generating clocks. All register and internal RAM contents are preserved. All functional I/O "freeze" in the last state when the PLL clock is turned off. Following reset, the PLL needs time to re-lock, just as it does following power-up. Wake-up from PD3 takes longer than wake-up from PD2 because the PLL needs to be re-locked, just as it does following power-up.
All others	Reserved	—	—

[†] When entering PD2 and PD3, all functional I/O remains in the previous state. However, for peripherals which are asynchronous in nature or peripherals with an external clock source, output signals may transition in response to stimulus on the inputs. Under these conditions, peripherals will not operate according to specifications.

2.19.2 C64x Power-Down Mode with an Emulator

If user power-down modes are programmed, and an emulator is attached, the modes will be masked to allow the emulator access to the system. This condition prevails until the emulator is reset or the cable is removed from the header. If power measurements are to be performed when in a power-down mode, the emulator cable should be removed.

When the DSP is in power-down mode PD2 or PD3, emulation logic will force any emulation execution command (such as Step or Run) to spin in IDLE. For this reason, PC writes (such as loading code) will fail. A DSP reset will be required to get the DSP out of PD2/PD3.

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2.20 Power-Supply Sequencing

TI DSPs do not require specific power sequencing between the core supply and the I/O supply. However, systems should be designed to ensure that neither supply is powered up for extended periods of time (>1 second) if the other supply is below the proper operating voltage.

2.20.1 Power-Supply Design Considerations

A dual-power supply with simultaneous sequencing can be used to eliminate the delay between core and I/O power up. A Schottky diode can also be used to tie the core rail to the I/O rail (see Figure 2–14).
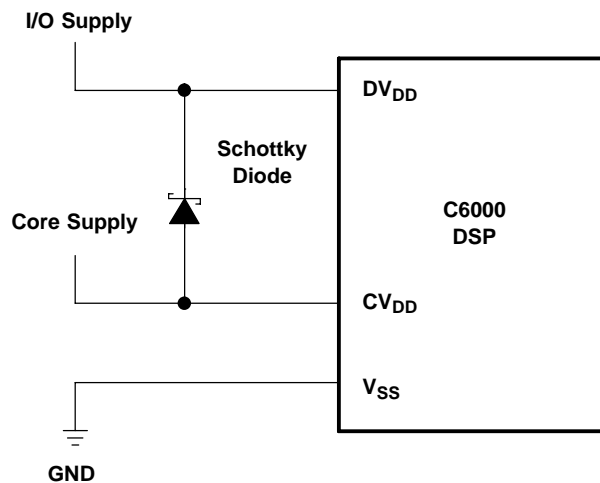
Figure 2–14. Schottky Diode Diagram

Core and I/O supply voltage regulators should be located close to the DSP (or DSP array) to minimize inductance and resistance in the power delivery path. Additionally, when designing for high-performance applications utilizing the C6000™ platform of DSPs, the PC board should include separate power planes for core, I/O, and ground, all bypassed with high-quality low-ESL/ESR capacitors.

TI DSPs do not require specific power sequencing between the core supply and the I/O supply. However, systems should be designed to ensure that neither supply is powered up for extended periods of time if the other supply is below the proper operating voltage.

2.20.2 Power-Supply Decoupling

In order to properly decouple the supply planes from system noise, place as many capacitors (caps) as possible close to the DSP. Assuming 0603 caps, the user should be able to fit a total of 60 caps, 30 for the core supply and 30 for the I/O supply. These caps need to be close to the DSP power pins, no more than 1.25 cm maximum distance to be effective. Physically smaller caps, such as 0402, are better because of their lower parasitic inductance. Proper capacitance values are also important. Small bypass caps (near 560 pF) should be closest to the power pins. Medium bypass caps (220 nF or as large as can be obtained in a small package) should be next closest. TI recommends no less than 8 small and 8 medium caps per supply (32 total) be placed immediately next to the BGA vias, using the "interior" BGA space and at least the corners of the "exterior".

Eight larger caps (4 for each supply) can be placed further away for bulk decoupling. Large bulk caps (on the order of 100 μF) should be furthest away (but still as close as possible). No less than 4 large caps per supply (8 total) should be placed outside of the BGA.

Any cap selection needs to be evaluated from a yield/manufacturing point-of-view. As with the selection of any component, verification of capacitor availability over the product's production lifetime should be considered.

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2.21 Power-Down Operation

The C6412 device can be powered down in three ways:

- Power-down due to pin configuration
- Power-down due to software configuration – relates to the default state of the peripheral configuration bits in the PERCFG register.
- Power-down during run-time via software configuration

On the C6412 device, the HPI, PCI, and EMAC and MDIO peripherals are controlled (selected) at the pin level during chip reset (e.g., PCI_EN, HD5, and MAC_EN pins).

The McBSP0, McBSP1, and I2C0 peripheral functions are selected via the peripheral configuration (PERCFG) register bits.

For more detailed information on the peripheral configuration pins and the PERCFG register bits, see the Device Configurations section of this document.

2.22 IEEE 1149.1 JTAG Compatibility Statement

The TMS320C6412 DSP requires that both $\overline{\text{TRST}}$ and $\overline{\text{RESET}}$ be asserted upon power up to be properly initialized. While $\overline{\text{RESET}}$ initializes the DSP core, $\overline{\text{TRST}}$ initializes the DSP's emulation logic. Both resets are required for proper operation.

Note: $\overline{\text{TRST}}$ is synchronous and **must** be clocked by TCLK; otherwise, BSCAN may not respond as expected after $\overline{\text{TRST}}$ is asserted.

While both $\overline{\text{TRST}}$ and $\overline{\text{RESET}}$ need to be asserted upon power up, only $\overline{\text{RESET}}$ needs to be released for the DSP to boot properly. $\overline{\text{TRST}}$ may be asserted indefinitely for normal operation, keeping the JTAG port interface and DSP's emulation logic in the reset state. $\overline{\text{TRST}}$ only needs to be released when it is necessary to use a JTAG controller to debug the DSP or exercise the DSP's boundary scan functionality. $\overline{\text{RESET}}$ must be released in order for boundary-scan JTAG to read the variant field of IDCODE correctly. Other boundary-scan instructions work correctly independant of current state of $\overline{\text{RESET}}$.

The TMS320C6412 DSP includes an internal pulldown (IPD) on the $\overline{\text{TRST}}$ pin to ensure that $\overline{\text{TRST}}$ will always be asserted upon power up and the DSP's internal emulation logic will always be properly initialized when this pin is not routed out. JTAG controllers from Texas Instruments actively drive $\overline{\text{TRST}}$ high. However, some third-party JTAG controllers may not drive $\overline{\text{TRST}}$ high but expect the use of an external pullup resistor on $\overline{\text{TRST}}$. When using this type of JTAG controller, assert $\overline{\text{TRST}}$ to initialize the DSP after powerup and externally drive $\overline{\text{TRST}}$ high before attempting any emulation or boundary scan operations.

Following the release of $\overline{\text{RESET}}$, the low-to-high transition of $\overline{\text{TRST}}$ must be "seen" to latch the state of EMU1 and EMU0. The EMU[1:0] pins configure the device for either Boundary Scan mode or Normal/Emulation mode. For more detailed information, see the terminal functions section of this data sheet.

Note: The DESIGN_WARNING section of the TMS320C6412 BSDL file contains information and constraints regarding proper device operation while in Boundary Scan Mode.

For more detailed information on the C6412 JTAG emulation, see the *TMS320C6000 DSP Designing for JTAG Emulation Reference Guide* (literature number SPRU641).

2.23 EMIF Device Speed

The rated EMIF speed of these devices only applies to the SDRAM interface when in a system that meets the following requirements:

- 1 chip-enable (CE) space (maximum of 2 chips) of SDRAM connected to EMIF
- up to 1 CE space of buffers connected to EMIF
- EMIF trace lengths between 1 and 3 inches
- 166-MHz SDRAM for 133-MHz operation
- 143-MHz SDRAM for 100-MHz operation

Other configurations may be possible, but timing analysis must be done to verify all AC timings are met. Verification of AC timings is mandatory when using configurations other than those specified above. TI recommends utilizing I/O buffer information specification (IBIS) to analyze all AC timings.

To properly use IBIS models to attain accurate timing analysis for a given system, see the *Using IBIS Models for Timing Analysis* application report (literature number SPRA839).

To maintain signal integrity, serial termination resistors should be inserted into all EMIF output signal lines (see the Terminal Functions table for the EMIF output signals).

For more detailed information on the C6412 EMIF peripheral, see the *TMS320C6000 DSP External Memory Interface (EMIF) Reference Guide* (literature number SPRU266).

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2.24 Bootmode

The C6412 device resets using the active-low signal $\overline{RESET}$. While $\overline{RESET}$ is low, the device is held in reset and is initialized to the prescribed reset state. Refer to reset timing for reset timing characteristics and states of device pins during reset. The release of $\overline{RESET}$ starts the processor running with the prescribed device configuration and boot mode.

The C6412 has three types of boot modes:

- Host boot

  If host boot is selected, upon release of $\overline{RESET}$, the CPU is internally "stalled" while the remainder of the device is released. During this period, an external host can initialize the CPU's memory space as necessary through the host interface, including internal configuration registers, such as those that control the EMIF or other peripherals. For the C6412 device, the HPI peripheral is used for host boot if PCI_EN = 0, and the PCI peripheral is used if PCI_EN = 1. Once the host is finished with all necessary initialization, it must set the DSPINT bit in the HPIC register to complete the boot process. This transition causes the boot configuration logic to bring the CPU out of the "stalled" state. The CPU then begins execution from address 0. The DSPINT condition is not latched by the CPU, because it occurs while the CPU is still internally "stalled". Also, DSPINT brings the CPU out of the "stalled" state only if the host boot process is selected. All memory may be written to and read by the host. This allows for the host to verify what it sends to the DSP if required. After the CPU is out of the "stalled" state, the CPU needs to clear the DSPINT, otherwise, no more DSPINTs can be received.

- EMIF boot (using default ROM timings)

  Upon the release of $\overline{RESET}$, the 1K-Byte ROM code located in the beginning of $\overline{ACE1}$ is copied to address 0 by the EDMA using the default ROM timings, while the CPU is internally "stalled". The data should be stored in the endian format that the system is using. In this case, the EMIF automatically assembles consecutive 8-bit bytes to form the 32-bit instruction words to be copied. The transfer is automatically done by the EDMA as a single-frame block transfer from the ROM to address 0. After completion of the block transfer, the CPU is released from the "stalled" state and starts running from address 0.

- No boot

  With no boot, the CPU begins direct execution from the memory located at address 0. Note: operation is undefined if invalid code is located at address 0.

2.25 Reset

A hardware reset ($\overline{RESET}$) is required to place the DSP into a known good state out of power-up. The $\overline{RESET}$ signal can be asserted (pulled low) prior to ramping the core and I/O voltages or after the core and I/O voltages have reached their proper operating conditions. As a best practice, reset should be held low during power-up. Prior to deasserting $\overline{RESET}$ (low-to-high transition), the core and I/O voltages should be at their proper operating conditions and CLKIN should also be running at the correct frequency.

3 Electrical Specifications

3.1 Absolute Maximum Ratings Over Operating Case Temperature Range (Unless Otherwise Noted)[†]

Supply voltage ranges:	CV_{DD} (see Note 1)	− 0.3 V to 1.8 V
	DV_{DD} (see Note 1)	−0.3 V to 4 V
Input voltage ranges:	(except PCI), V_I	−0.3 V to 4 V
	(PCI), V_{IP}	−0.5 V to DV_{DD} + 0.5 V
Output voltage ranges:	(except PCI), V_O	−0.3 V to 4 V
	(PCI), V_{OP}	−0.5 V to DV_{DD} + 0.5 V
Operating case temperature ranges, T_C:	(default)	0°C to 90°C
	(A version) [A−500 and A−600]	−40°C to 105°C
Storage temperature range, T_{stg}		−65°C to 150°C
Package Temperature Cycling:	Temperature Range	−40°C to 125°C
	Number of Cycles	500

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values are with respect to V_{SS}.

3.2 Recommended Operating Conditions

			MIN	NOM	MAX	UNIT
CV_{DD}	Supply voltage, Core (-500 device)[‡]		1.14	1.2	1.26	V
CV_{DD}	Supply voltage, Core (A-500, A−600, -600, -720 devices)[‡]		1.36	1.4	1.44	V
DV_{DD}	Supply voltage, I/O		3.14	3.3	3.46	V
V_{SS}	Supply ground		0	0	0	V
V_{IH}	High-level input voltage (except PCI)		2			V
V_{IL}	Low-level input voltage (except PCI)				0.8	V
V_{IP}	Input voltage (PCI)		−0.5		DV_{DD} + 0.5	V
V_{IHP}	High-level input voltage (PCI)		$0.5DV_{DD}$		DV_{DD} + 0.5	V
V_{ILP}	Low-level input voltage (PCI)		−0.5		$0.3DV_{DD}$	V
V_{OS}	Maximum voltage during overshoot				4.3[§]	V
V_{US}	Maximum voltage during undershoot		−1.0[§]			V
T_C	Operating case temperature	Default	0		90	°C
		A version [A−500 and A−600]	−40		105	°C

[‡] Future variants of the C64x DSPs may operate at voltages ranging from 0.9 V to 1.4 V to provide a range of system power/performance options. TI highly recommends that users design-in a supply that can handle multiple voltages within this range (i.e., 1.2 V, 1.25 V, 1.3 V, 1.35 V, 1.4 V with ± 3% tolerances) by implementing simple board changes such as reference resistor values or input pin configuration modifications. Examples of such supplies include the PT4660, PT5500, PT5520, PT6440, and PT6930 series from Power Trends, a subsidiary of Texas Instruments. Not incorporating a flexible supply may limit the system's ability to easily adapt to future versions of C64x devices.

[§] The absolute maximum ratings should *not* be exceeded for more than 30% of the cycle period.

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3.3 Electrical Characteristics Over Recommended Ranges of Supply Voltage and Operating Case Temperature (Unless Otherwise Noted)

PARAMETER		TEST CONDITIONS[†]	MIN	TYP	MAX	UNIT
V_{OH}	High-level output voltage (except PCI)	DV_{DD} = MIN, I_{OH} = MAX[☆]	2.4			V
V_{OHP}	High-level output voltage (PCI)	I_{OHP} = –0.5 mA, DV_{DD} = 3.3 V	$0.9DV_{DD}$[¶]			V
V_{OL}	Low-level output voltage (except PCI)	DV_{DD} = MIN, I_{OL} = MAX[☆]			0.4	V
V_{OLP}	Low-level output voltage (PCI)	I_{OLP} = 1.5 mA, DV_{DD} = 3.3 V			$0.1DV_{DD}$[¶]	V
I_I	Input current (except PCI)	V_I = V_{SS} to DV_{DD} no opposing internal resistor			±10	uA
		V_I = V_{SS} to DV_{DD} opposing internal pullup resistor[‡]	50	100	150	uA
		V_I = V_{SS} to DV_{DD} opposing internal pulldown resistor[‡]	–150	–100	–50	uA
I_{IP}	Input leakage current (PCI)[§]	0 < V_{IP} < DV_{DD} = 3.3 V			±10	uA
I_{OH}	High-level output current	EMIF, CLKOUT4, CLKOUT6, EMUx			–16	mA
		Timer, TDO, GPIO (Excluding GP0[15:9, 2, 1]), McBSP			–8	mA
		PCI/HPI			–0.5[¶]	mA
I_{OL}	Low-level output current	EMIF, CLKOUT4, CLKOUT6, EMUx			16	mA
		Timer, TDO, GPIO (Excluding GP0[15:9, 2, 1]), McBSP			8	mA
		SCL0 and SDA0			3	mA
		PCI/HPI			1.5[¶]	mA
I_{OZ}	Off-state output current	V_O = DV_{DD} or 0 V			±10	uA
I_{CDD}	Core supply current[#]	CV_{DD} = 1.4 V, CPU clock = 720 MHz		1090		mA
		CV_{DD} = 1.4 V, CPU clock = 600 MHz		890		mA
		CV_{DD} = 1.2 V, CPU clock = 500 MHz		620		mA
I_{DDD}	I/O supply current[#]	DV_{DD} = 3.3 V, CPU clock = 720 MHz		210		mA
		DV_{DD} = 3.3 V, CPU clock = 600 MHz		210		mA
		DV_{DD} = 3.3 V, CPU clock = 500 MHz		165		mA
C_i	Input capacitance				10	pF
C_o	Output capacitance				10	pF

[†] For test conditions shown as MIN, MAX, or NOM, use the appropriate value specified in the recommended operating conditions table.

[‡] Applies only to pins with an internal pullup (IPU) or pulldown (IPD) resistor.

[§] PCI input leakage currents include Hi-Z output leakage for all bidirectional buffers with 3-state outputs.

[¶] These rated numbers are from the PCI specification version 2.3. The DC specification and AC specification are defined in Tables 4-3 and 4-4, respectively.

[#] Measured with average activity (50% high/50% low power) at 25°C case temperature and 133-MHz EMIF for -600 and -720 speeds (100-MHz EMIF for -500 speed). This model represents a device performing high-DSP-activity operations 50% of the time, and the remainder performing low-DSP-activity operations. The high/low-DSP-activity models are defined as follows:

High-DSP-Activity Model:
  CPU: 8 instructions/cycle with 2 LDDW instructions [L1 Data Memory: 128 bits/cycle via LDDW instructions;
    L1 Program Memory: 256 bits/cycle; L2/EMIF EDMA: 50% writes, 50% reads to/from SDRAM (50% bit-switching)]
  McBSP: 2 channels at E1 rate
  Timers: 2 timers at maximum rate
Low-DSP-Activity Model:
  CPU: 2 instructions/cycle with 1 LDH instruction [L1 Data Memory: 16 bits/cycle; L1 Program Memory: 256 bits per 4 cycles;
    L2/EMIF EDMA: None]
  McBSP: 2 channels at E1 rate
 Timers: 2 timers at maximum rate
The actual current draw is highly application-dependent. For more details on core and I/O activity, refer to the *TMS320C6412 Power Consumption Summary* application report (literature number SPRA967).

[☆] Single pin driving I_{OH}/I_{OL} = MAX.

3.4 Recommended Clock and Control Signal Transition Behavior

All clocks and control signals ***must*** transition between V_{IH} and V_{IL} (or between V_{IL} and V_{IH}) in a monotonic manner.

4 Parameter Information
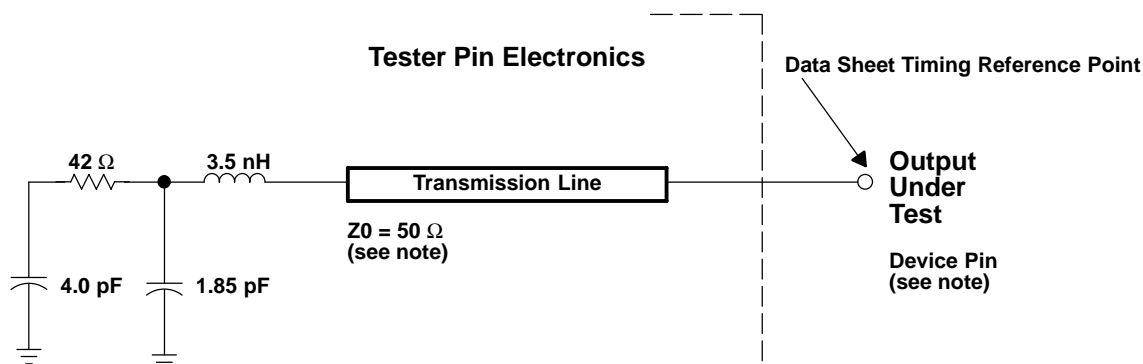
NOTE: The data sheet provides timing at the device pin. For output timing analysis, the tester pin electronics and its transmission line effects must be taken into account. A transmission line with a delay of 2 ns or longer can be used to produce the desired transmission line effect. The transmission line is intended as a load only. It is not necessary to add or subtract the transmission line delay (2 ns or longer) from the data sheet timings.

Input requirements in this data sheet are tested with an input slew rate of < 4 Volts per nanosecond (4 V/ns) at the device pin.

Figure 4–1. Test Load Circuit for AC Timing Measurements

The load capacitance value stated is only for characterization and measurement of AC timing signals. This load capacitance value does not indicate the maximum load the device is capable of driving.

4.1 Signal Transition Levels

All input and output timing parameters are referenced to 1.5 V for both "0" and "1" logic levels.
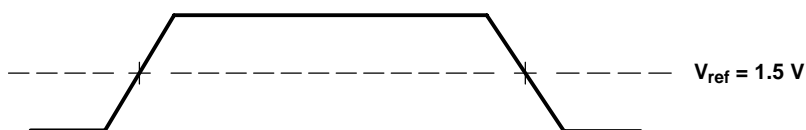

Figure 4–2. Input and Output Voltage Reference Levels for AC Timing Measurements

All rise and fall transition timing parameters are referenced to V_{IL} MAX and V_{IH} MIN for input clocks, V_{OL} MAX and V_{OH} MIN for output clocks, V_{ILP} MAX and V_{IHP} MIN for PCI input clocks, and V_{OLP} MAX and V_{OHP} MIN for PCI output clocks.
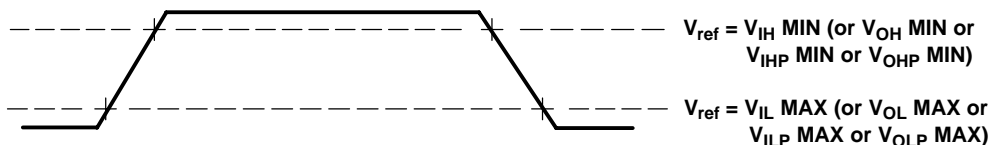

Figure 4–3. Rise and Fall Transition Time Voltage Reference Levels

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4.2 Signal Transition Rates

All timings are tested with an input edge rate of 4 Volts per nanosecond (4 V/ns).

4.2.1 AC Transient Rise/fall Time Specifications

Figure 4−4 and Figure 4−5 show the AC transient specifications for Rise and Fall Time. For device-specific information on these values, refer to the Recommended Operating Conditions section of this Data Sheet.
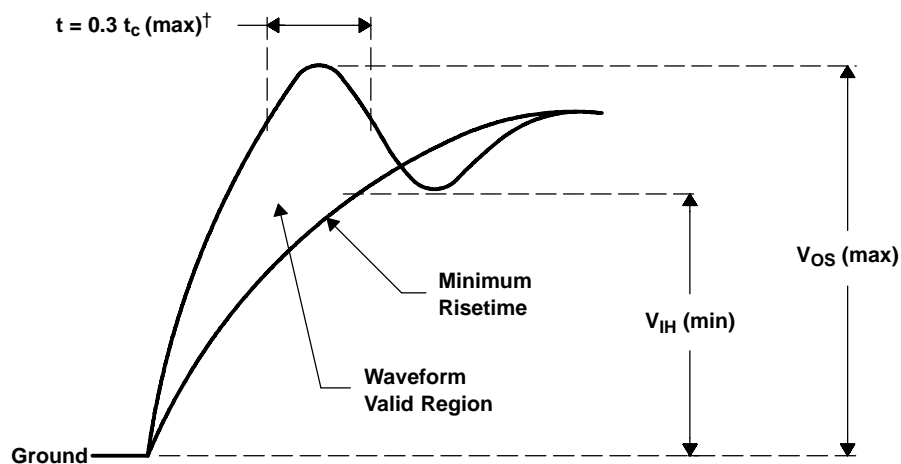
Figure 4−4. AC Transient Specification Rise Time
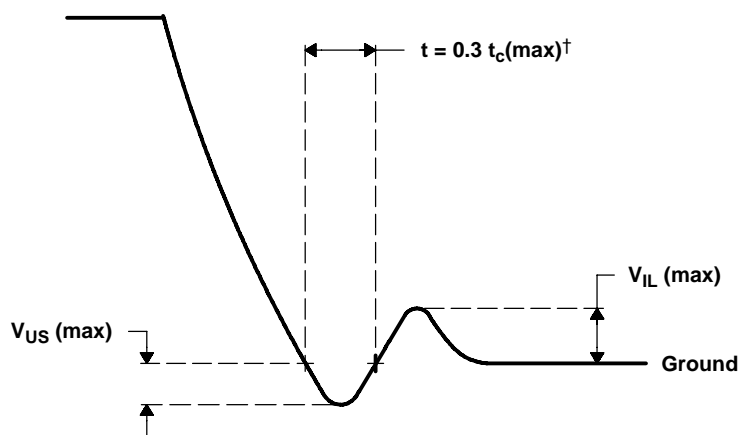
† t_c = the peripheral cycle time in nanoseconds (ns).

Figure 4−5. AC Transient Specification Fall Time

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† t_c = the peripheral cycle time in nanoseconds (ns).

## 4.3   Timing Parameters and Board Routing Analysis

The timing parameter values specified in this data sheet do *not* include delays by board routings. As a good board design practice, such delays must *always* be taken into account. Timing values may be adjusted by increasing/decreasing such delays. TI recommends utilizing the available I/O buffer information specification (IBIS) models to analyze the timing characteristics correctly. To properly use IBIS models to attain accurate timing analysis for a given system, see the *Using IBIS Models for Timing Analysis* application report (literature number SPRA839). If needed, external logic hardware such as buffers may be used to compensate for any timing differences.

For inputs, timing is most impacted by the round-trip propagation delay from the DSP to the external device and from the external device to the DSP. This round-trip delay tends to negatively impact the input setup time margin, but also tends to improve the input hold time margins (see Table 4–1 and Figure 4–6).

Figure 4–6 represents a general transfer between the DSP and an external device. The figure also represents board route delays and how they are perceived by the DSP and the external device.

Table 4–1. Board-Level Timing Example (see Figure 4–6)

NO.	DESCRIPTION
1	Clock route delay
2	Minimum DSP hold time
3	Minimum DSP setup time
4	External device hold time requirement
5	External device setup time requirement
6	Control signal route delay
7	External device hold time
8	External device access time
9	DSP hold time requirement
10	DSP setup time requirement
11	Data route delay
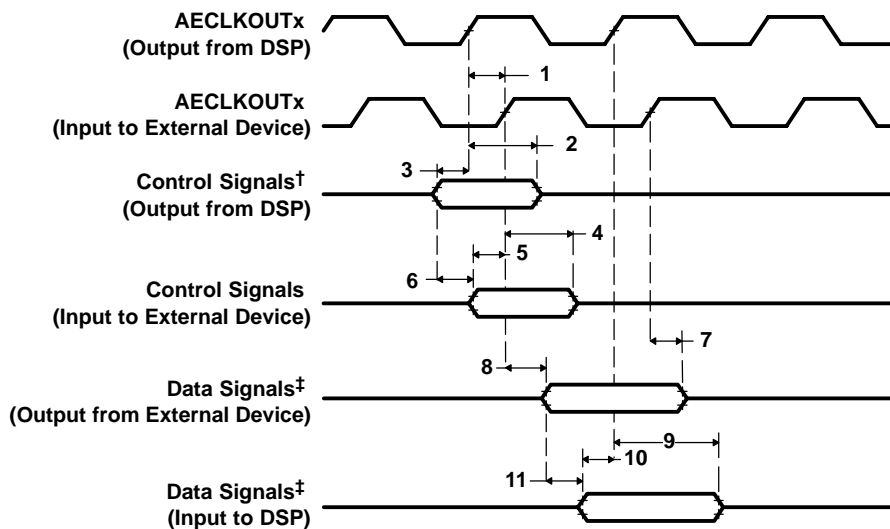
† Control signals include data for Writes.

‡ Data signals are generated during Reads from an external device.

Figure 4–6. Board-Level Input/Output Timings

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4.4 Input and Output Clocks

Table 4–2. Timing Requirements for CLKIN for –500 Devices[†‡§] (see Figure 4–7)

NO.			–500						UNIT
			PLL MODE x12		PLL MODE x6		x1 (BYPASS)		
			MIN	MAX	MIN	MAX	MIN	MAX	
1	$t_{c(CLKIN)}$	Cycle time, CLKIN	24	33.3	13.3	33.3	13.3	33.3	ns
2	$t_{w(CLKINH)}$	Pulse duration, CLKIN high	0.45C		0.45C		0.45C		ns
3	$t_{w(CLKINL)}$	Pulse duration, CLKIN low	0.45C		0.45C		0.45C		ns
4	$t_{t(CLKIN)}$	Transition time, CLKIN		5		5		1	ns
5	$t_{J(CLKIN)}$	Period jitter, CLKIN		0.02C		0.02C		0.02C	ns

[†] The reference points for the rise and fall transitions are measured at V_{IL} MAX and V_{IH} MIN.
[‡] For more details on the PLL multiplier factors (x6, x12), see the *Clock PLL* section of this data sheet.
[§] C = CLKIN cycle time in ns. For example, when CLKIN frequency is 50 MHz, use C = 20 ns.

Table 4–3. Timing Requirements for CLKIN for –600 Devices[†‡§] (see Figure 4–7)

NO.			–600						UNIT
			PLL MODE x12		PLL MODE x6		x1 (BYPASS)		
			MIN	MAX	MIN	MAX	MIN	MAX	
1	$t_{c(CLKIN)}$	Cycle time, CLKIN	20	33.3	13.3	33.3	13.3	33.3	ns
2	$t_{w(CLKINH)}$	Pulse duration, CLKIN high	0.45C		0.45C		0.45C		ns
3	$t_{w(CLKINL)}$	Pulse duration, CLKIN low	0.45C		0.45C		0.45C		ns
4	$t_{t(CLKIN)}$	Transition time, CLKIN		5		5		1	ns
5	$t_{J(CLKIN)}$	Period jitter, CLKIN		0.02C		0.02C		0.02C	ns

[†] The reference points for the rise and fall transitions are measured at V_{IL} MAX and V_{IH} MIN.
[‡] For more details on the PLL multiplier factors (x6, x12), see the *Clock PLL* section of this data sheet.
[§] C = CLKIN cycle time in ns. For example, when CLKIN frequency is 50 MHz, use C = 20 ns.
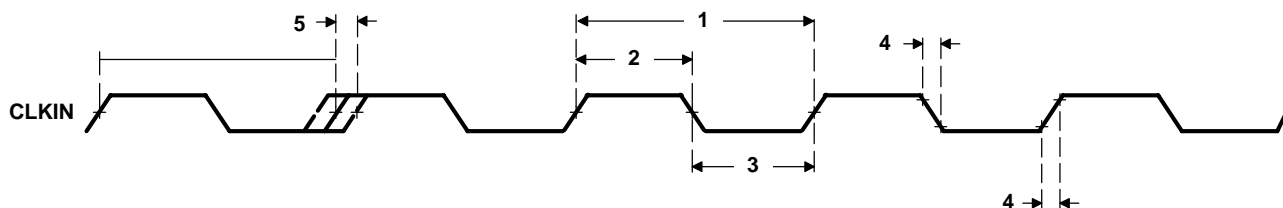
Table 4–4. Timing Requirements for CLKIN for –720 Devices[†‡§] (see Figure 4–7)

NO.			–720						UNIT
			PLL MODE x12		PLL MODE x6		x1 (BYPASS)		
			MIN	MAX	MIN	MAX	MIN	MAX	
1	$t_{c(CLKIN)}$	Cycle time, CLKIN	16.6	33.3	13.3	33.3	13.3	33.3	ns
2	$t_{w(CLKINH)}$	Pulse duration, CLKIN high	0.45C		0.45C		0.45C		ns
3	$t_{w(CLKINL)}$	Pulse duration, CLKIN low	0.45C		0.45C		0.45C		ns
4	$t_{t(CLKIN)}$	Transition time, CLKIN		5		5		1	ns
5	$t_{J(CLKIN)}$	Period jitter, CLKIN		0.02C		0.02C		0.02C	ns

[†] The reference points for the rise and fall transitions are measured at V_{IL} MAX and V_{IH} MIN.
[‡] For more details on the PLL multiplier factors (x6, x12), see the *Clock PLL* section of this data sheet.
[§] C = CLKIN cycle time in ns. For example, when CLKIN frequency is 50 MHz, use C = 20 ns.

Figure 4–7. CLKIN Timing

Table 4–5. Switching Characteristics Over Recommended Operating Conditions for CLKOUT4[†‡§] (see Figure 4–8)

NO.		PARAMETER	–500 –600 –720		UNIT
			CLKMODE = x1, x6, x12		
			MIN	MAX	
1	$t_{w(CKO4H)}$	Pulse duration, CLKOUT4 high	2P – 0.7	2P + 0.7	ns
2	$t_{w(CKO4L)}$	Pulse duration, CLKOUT4 low	2P – 0.7	2P + 0.7	ns
3	$t_{t(CKO4)}$	Transition time, CLKOUT4		1	ns

[†] The reference points for the rise and fall transitions are measured at V_{OL} MAX and V_{OH} MIN.
[‡] PH is the high period of CLKIN in ns and PL is the low period of CLKIN in ns.
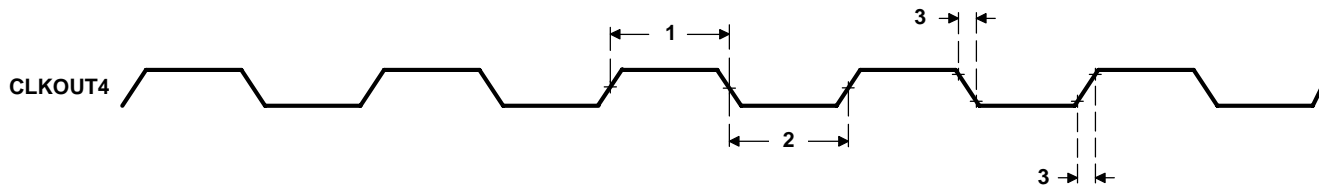[§] P = 1/CPU clock frequency in nanoseconds (ns)

Figure 4–8. CLKOUT4 Timing

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Table 4–6. Switching Characteristics Over Recommended Operating Conditions for CLKOUT6†‡§
(see Figure 4–9)

NO.	PARAMETER		−500 −600 −720 CLKMODE = x1, x6, x12		UNIT
			MIN	**MAX**	
1	$t_{w(CKO6H)}$	Pulse duration, CLKOUT6 high	3P − 0.7	3P + 0.7	ns
2	$t_{w(CKO6L)}$	Pulse duration, CLKOUT6 low	3P − 0.7	3P + 0.7	ns
3	$t_{t(CKO6)}$	Transition time, CLKOUT6		1	ns

† The reference points for the rise and fall transitions are measured at V_{OL} MAX and V_{OH} MIN.
‡ PH is the high period of CLKIN in ns and PL is the low period of CLKIN in ns.
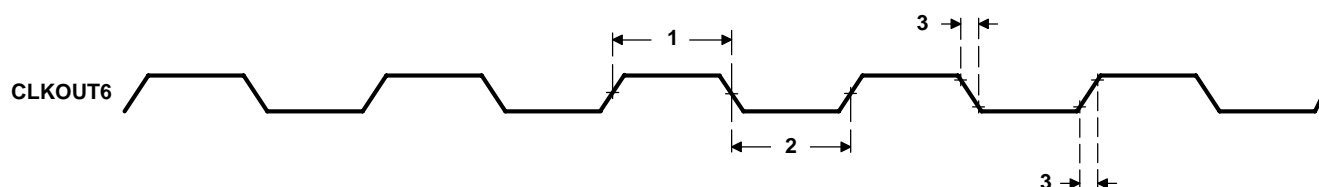§ P = 1/CPU clock frequency in nanoseconds (ns)

Figure 4–9. CLKOUT6 Timing

Table 4–7. Timing Requirements for AECLKIN for EMIFA[†‡§] (see Figure 4–10)

NO.			−500 −600 −720		UNIT
			MIN	MAX	
1	$t_{c(EKI)}$	Cycle time, AECLKIN	6[¶]	16P	ns
2	$t_{w(EKIH)}$	Pulse duration, AECLKIN high	2.7		ns
3	$t_{w(EKIL)}$	Pulse duration, AECLKIN low	2.7		ns
4	$t_{t(EKI)}$	Transition time, AECLKIN		3	ns
5	$t_{J(EKI)}$	Period jitter, AECLKIN		0.02E	ns

[†] P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
[‡] The reference points for the rise and fall transitions are measured at V_{IL} MAX and V_{IH} MIN.
[§] E = the EMIF input clock (AECLKIN, CPU/4 clock, or CPU/6 clock) period in ns for EMIFA.
[¶] Minimum AECLKIN cycle times *must* be met, even when AECLKIN is generated by an internal clock source. Minimum AECLKIN times are based on internal logic speed; the maximum useable speed of the EMIF may be lower due to AC timing requirements. On the 600 devices, 133-MHz operation is achievable if the requirements of the EMIF Device Speed section are met. On the 500 devices, 100-MHz operation is achievable if the requirements of the EMIF Device Speed section are met.
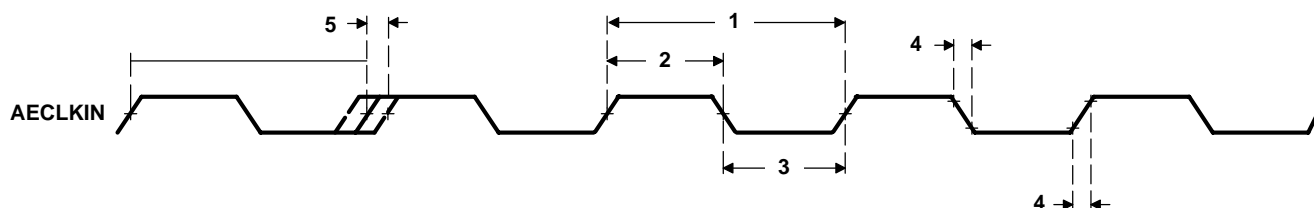

Figure 4–10. AECLKIN Timing for EMIFA

Table 4–8. Switching Characteristics Over Recommended Operating Conditions for AECLKOUT1 for the EMIFA Module[¶#‖] (see Figure 4–11)

NO.		PARAMETER	−500 −600 −720		UNIT
			MIN	MAX	
1	$t_{w(EKO1H)}$	Pulse duration, AECLKOUT1 high	EH − 0.7	EH + 0.7	ns
2	$t_{w(EKO1L)}$	Pulse duration, AECLKOUT1 low	EL − 0.7	EL + 0.7	ns
3	$t_{t(EKO1)}$	Transition time, AECLKOUT1		1	ns
4	$t_{d(EKIH-EKO1H)}$	Delay time, AECLKIN high to AECLKOUT1 high	1	8	ns
5	$t_{d(EKIL-EKO1L)}$	Delay time, AECLKIN low to AECLKOUT1 low	1	8	ns

[¶] The reference points for the rise and fall transitions are measured at V_{OL} MAX and V_{OH} MIN.
[#] E = the EMIF input clock (AECLKIN, CPU/4 clock, or CPU/6 clock) period in ns for EMIFA.
[‖] EH is the high period of E (EMIF input clock period) in ns and EL is the low period of E (EMIF input clock period) in ns for EMIFA.
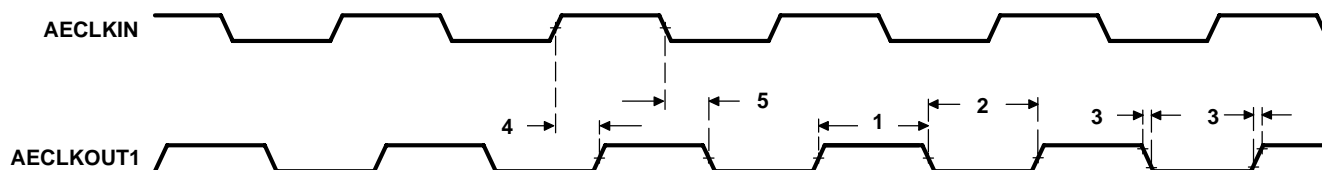[*] This cycle-to-cycle jitter specification was measured with CPU/4 or CPU/6 as the source of the EMIF input clock.

Figure 4–11. AECLKOUT1 Timing for EMIFA Module

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Table 4–9. Switching Characteristics Over Recommended Operating Conditions for AECLKOUT2 for the EMIFA Module†‡ (see Figure 4–12)

| NO.	PARAMETER		−500 −600 −720		UNIT
			MIN	MAX	
1	$t_{w(EKO2H)}$	Pulse duration, AECLKOUT2 high	0.5NE − 0.7	0.5NE + 0.7	ns
2	$t_{w(EKO2L)}$	Pulse duration, AECLKOUT2 low	0.5NE − 0.7	0.5NE + 0.7	ns
3	$t_{t(EKO2)}$	Transition time, AECLKOUT2		1	ns
4	$t_{d(EKIH-EKO2H)}$	Delay time, AECLKIN high to AECLKOUT2 high	1	8	ns
5	$t_{d(EKIL-EKO2L)}$	Delay time, AECLKIN low to AECLKOUT2 low	1	8	ns

† The reference points for the rise and fall transitions are measured at V_{OL} MAX and V_{OH} MIN.
‡ E = the EMIF input clock (AECLKIN, CPU/4 clock, or CPU/6 clock) period in ns for EMIFA.
  N = the EMIF input clock divider; N = 1, 2, or 4.
§ This cycle-to-cycle jitter specification was measured with CPU/4 or CPU/6 as the source of the EMIF input clock.
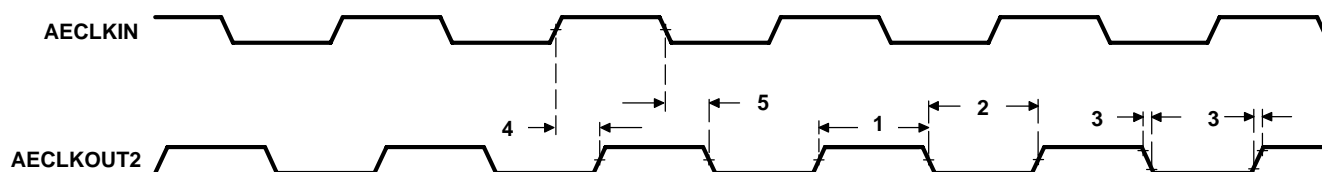
Figure 4–12. AECLKOUT2 Timing for the EMIFA Module

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5   Asynchronous Memory Timing

Table 5–1. Timing Requirements for Asynchronous Memory Cycles for EMIFA Module[†‡]
(see Figure 5–1 and Figure 5–2)

NO.		PARAMETER	–500 –600 –720		UNIT
			MIN	MAX	
3	$t_{su(EDV-AREH)}$	Setup time, AEDx valid before $\overline{AARE}$ high	6.5		ns
4	$t_{h(AREH-EDV)}$	Hold time, AEDx valid after $\overline{AARE}$ high	1		ns
6	$t_{su(ARDY-EKO1H)}$	Setup time, AARDY valid before AECLKOUTx high	3		ns
7	$t_{h(EKO1H-ARDY)}$	Hold time, AARDY valid after AECLKOUTx high	2.5		ns

[†] To ensure data setup time, simply program the strobe width wide enough. AARDY is internally synchronized. The AARDY signal is *only* recognized two cycles before the end of the programmed strobe time and while AARDY is low, the strobe time is extended cycle-by-cycle. When AARDY is recognized low, the end of the strobe time is two cycles after AARDY is recognized high. To use AARDY as an asynchronous input, the pulse width of the AARDY signal should be wide enough (e.g., pulse width = 2E) to ensure setup and hold time is met.

[‡] RS = Read setup, RST = Read strobe, RH = Read hold, WS = Write setup, WST = Write strobe, WH = Write hold. These parameters are programmed via the EMIF CE space control registers.

Table 5–2. Switching Characteristics Over Recommended Operating Conditions for Asynchronous
Memory Cycles for EMIFA Module[‡§¶] (see Figure 5–1 and Figure 5–2)

NO.		PARAMETER	–500 –600 –720		UNIT
			MIN	MAX	
1	$t_{osu(SELV-AREL)}$	Output setup time, select signals valid to $\overline{AARE}$ low	RS * E – 1.8		ns
2	$t_{oh(AREH-SELIV)}$	Output hold time, $\overline{AARE}$ high to select signals invalid	RH * E – 1.9		ns
5	$t_{d(EKO1H-AREV)}$	Delay time, AECLKOUTx high to $\overline{AARE}$ valid	1	7	ns
8	$t_{osu(SELV-AWEL)}$	Output setup time, select signals valid to $\overline{AAWE}$ low	WS * E – 2.0		ns
9	$t_{oh(AWEH-SELIV)}$	Output hold time, $\overline{AAWE}$ high to select signals invalid	WH * E – 2.5		ns
10	$t_{d(EKO1H-AWEV)}$	Delay time, AECLKOUTx high to $\overline{AAWE}$ valid	1.3	7.1	ns

[‡] RS = Read setup, RST = Read strobe, RH = Read hold, WS = Write setup, WST = Write strobe, WH = Write hold. These parameters are programmed via the EMIF CE space control registers.

[§] E = AECLKOUT1 period in ns for EMIFA

[¶] Select signals for EMIFA include: $\overline{ACEx}$, $\overline{ABE[7:0]}$, AEA[22:3], $\overline{AAOE}$; and for EMIFA writes, include AED[63:0].
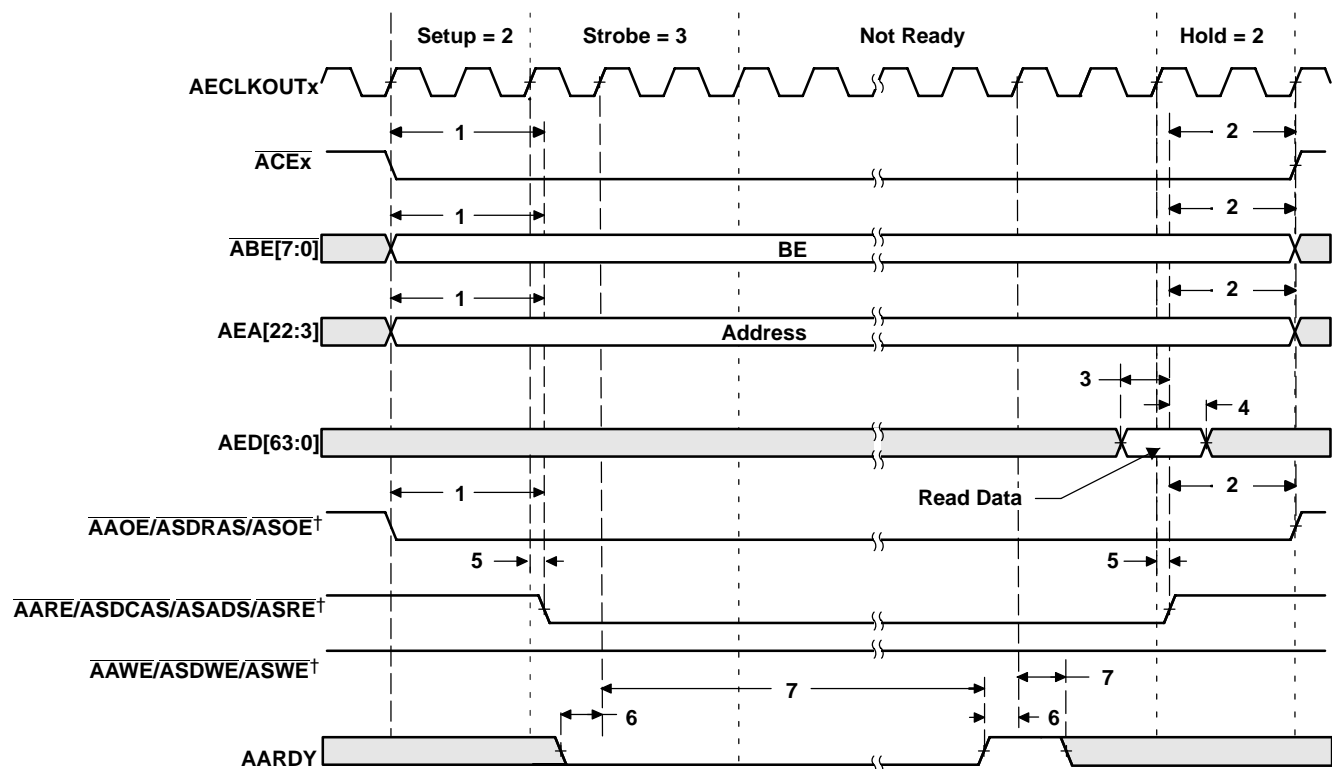
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† $\overline{\text{AAOE}}/\overline{\text{ASDRAS}}/\overline{\text{ASOE}}$, $\overline{\text{AARE}}/\overline{\text{ASDCAS}}/\overline{\text{ASADS}}/\overline{\text{ASRE}}$, and $\overline{\text{AAWE}}/\overline{\text{ASDWE}}/\overline{\text{ASWE}}$ operate as $\overline{\text{AAOE}}$ (identified under select signals), $\overline{\text{AARE}}$, and $\overline{\text{AAWE}}$, respectively, during asynchronous memory accesses.

Figure 5–1. Asynchronous Memory Read Timing for EMIFA
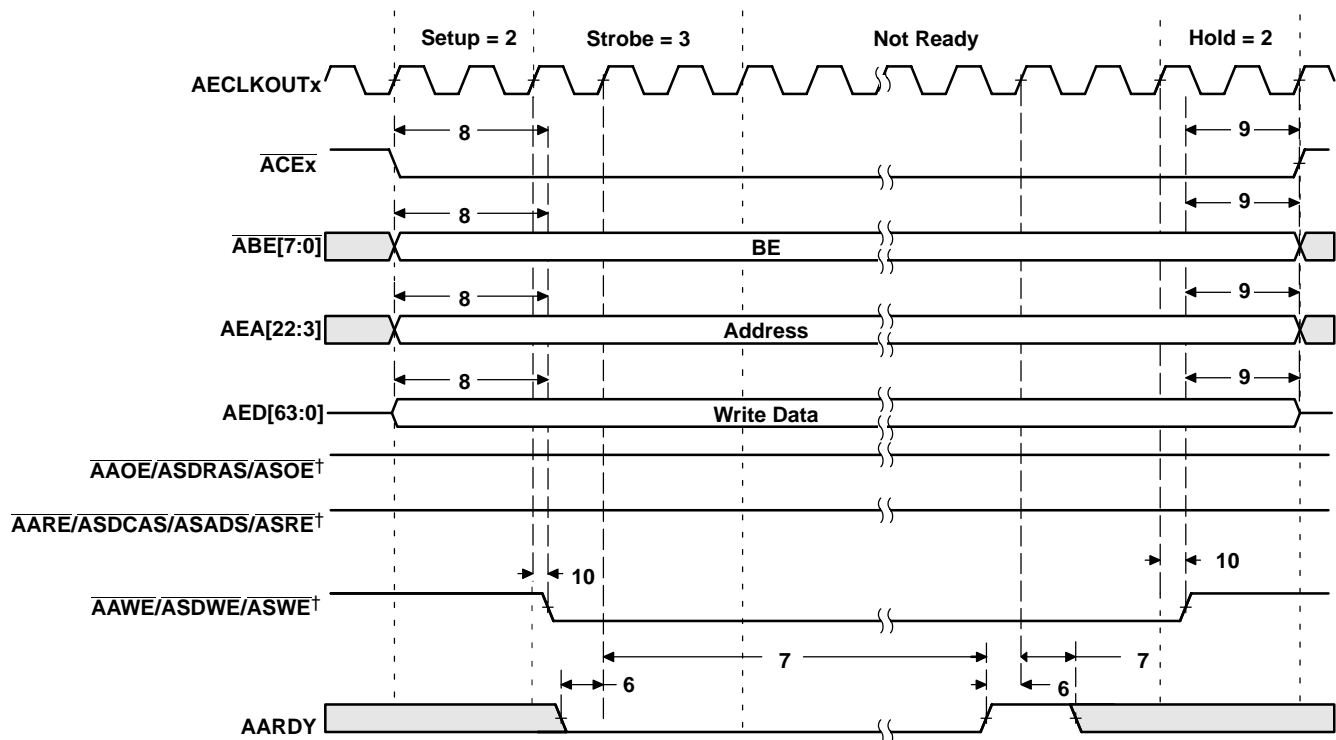
† AAOE/ASDRAS/ASOE, AARE/ASDCAS/ASADS/ASRE, and AAWE/ASDWE/ASWE operate as AAOE (identified under select signals), AARE, and AAWE, respectively, during asynchronous memory accesses.

Figure 5–2. Asynchronous Memory Write Timing for EMIFA

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6   Programmable Synchronous Interface Timing

**Table 6–1. Timing Requirements for Programmable Synchronous Interface Cycles for EMIFA Module[†]
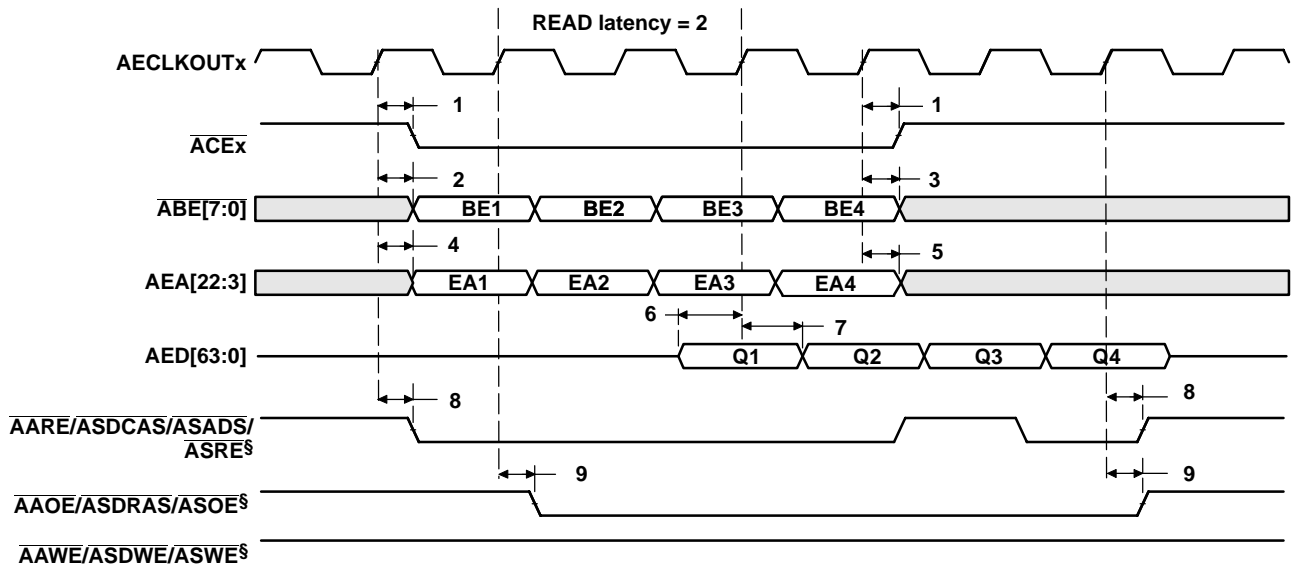(see Figure 6–1)**

NO.			−500		−600 −720		UNIT
			MIN	MAX	MIN	MAX	
6	$t_{su(EDV-EKOxH)}$	Setup time, read AEDx valid before AECLKOUTx high	3.1		2		ns
7	$t_{h(EKOxH-EDV)}$	Hold time, read AEDx valid after AECLKOUTx high	1.8		1.5		ns

**Table 6–2. Switching Characteristics Over Recommended Operating Conditions for Programmable
Synchronous Interface Cycles for EMIFA Module[†] (see Figure 6–1–Figure 6–3)**

NO.		PARAMETER	−500		−600 −720		UNIT
			MIN	MAX	MIN	MAX	
1	$t_{d(EKOxH-CEV)}$	Delay time, AECLKOUTx high to $\overline{ACEx}$ valid	1.1	6.4	1.1	4.9	ns
2	$t_{d(EKOxH-BEV)}$	Delay time, AECLKOUTx high to $\overline{ABEx}$ valid		6.4		4.9	ns
3	$t_{d(EKOxH-BEIV)}$	Delay time, AECLKOUTx high to $\overline{ABEx}$ invalid	1.1		1.1		ns
4	$t_{d(EKOxH-EAV)}$	Delay time, AECLKOUTx high to AEAx valid		6.4		4.9	ns
5	$t_{d(EKOxH-EAIV)}$	Delay time, AECLKOUTx high to AEAx invalid	1.1		1.1		ns
8	$t_{d(EKOxH-ADSV)}$	Delay time, AECLKOUTx high to $\overline{ASADS}/\overline{ASRE}$ valid	1.1	6.4	1.1	4.9	ns
9	$t_{d(EKOxH-OEV)}$	Delay time, AECLKOUTx high to $\overline{ASOE}$ valid	1.1	6.4	1.1	4.9	ns
10	$t_{d(EKOxH-EDV)}$	Delay time, AECLKOUTx high to AEDx valid		6.4		4.9	ns
11	$t_{d(EKOxH-EDIV)}$	Delay time, AECLKOUTx high to AEDx invalid	1.1		1.1		ns
12	$t_{d(EKOxH-WEV)}$	Delay time, AECLKOUTx high to $\overline{ASWE}$ valid	1.1	6.4	1.1	4.9	ns

[†] The following parameters are programmable via the EMIF CE Space Secondary Control register (CExSEC):
–   Read latency (SYNCRL): 0-, 1-, 2-, or 3-cycle read latency
–   Write latency (SYNCWL): 0-, 1-, 2-, or 3-cycle write latency
– $\overline{ACEx}$ assertion length (CEEXT): For standard SBSRAM or ZBT SRAM interface, $\overline{ACEx}$ goes inactive after the final command has been issued (CEEXT = 0). For synchronous FIFO interface with glue, $\overline{ACEx}$ is active when $\overline{ASOE}$ is active (CEEXT = 1).
–   Function of $\overline{ASADS}/\overline{ASRE}$ (RENEN): For standard SBSRAM or ZBT SRAM interface, $\overline{ASADS}/\overline{ASRE}$ acts as $\overline{ASADS}$ with deselect cycles (RENEN = 0). For FIFO interface, $\overline{ASADS}/\overline{ASRE}$ acts as $\overline{ASRE}$ with NO deselect cycles (RENEN = 1).
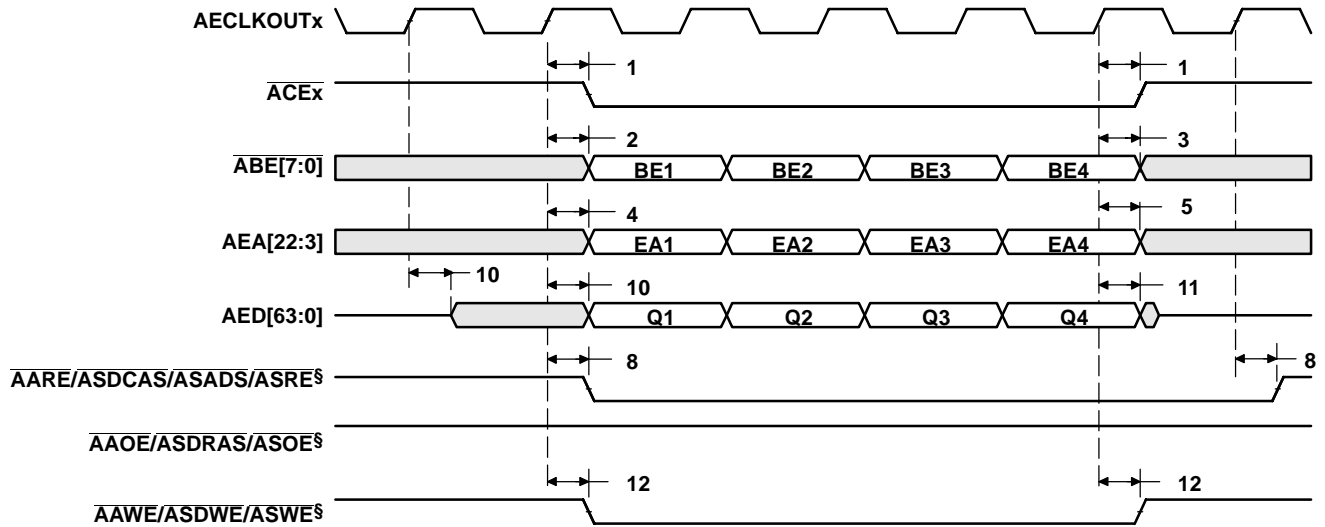–   Synchronization clock (SNCCLK): Synchronized to AECLKOUT1 or AECLKOUT2

† The read latency and the length of $\overline{\text{ACEx}}$ assertion are programmable via the SYNCRL and CEEXT fields, respectively, in the EMIFA CE Space Secondary Control register (CExSEC). In this figure, SYNCRL = 2 and CEEXT = 0.

‡ The following parameters are programmable via the EMIF CE Space Secondary Control register (CExSEC):
– Read latency (SYNCRL): 0-, 1-, 2-, or 3-cycle read latency
– Write latency (SYNCWL): 0-, 1-, 2-, or 3-cycle write latency
– $\overline{\text{ACEx}}$ assertion length (CEEXT): For standard SBSRAM or ZBT SRAM interface, $\overline{\text{ACEx}}$ goes inactive after the final command has been issued (CEEXT = 0). For synchronous FIFO interface with glue, $\overline{\text{ACEx}}$ is active when $\overline{\text{ASOE}}$ is active (CEEXT = 1).
– Function of $\overline{\text{ASADS/ASRE}}$ (RENEN): For standard SBSRAM or ZBT SRAM interface, $\overline{\text{ASADS/ASRE}}$ acts as $\overline{\text{ASADS}}$ with deselect cycles (RENEN = 0). For FIFO interface, $\overline{\text{ASADS/ASRE}}$ acts as $\overline{\text{ASRE}}$ with NO deselect cycles (RENEN = 1).
– Synchronization clock (SNCCLK): Synchronized to AECLKOUT1 or AECLKOUT2

§ $\overline{\text{AARE/ASDCAS/ASADS/ASRE}}$, $\overline{\text{AAOE/ASDRAS/ASOE}}$, and $\overline{\text{AAWE/ASDWE/AsWE}}$ operate as $\overline{\text{ASADS/ASRE}}$, $\overline{\text{ASOE}}$, and $\overline{\text{ASWE}}$, respectively, during programmable synchronous interface accesses.

Figure 6–1. Programmable Synchronous Interface Read Timing for EMIFA
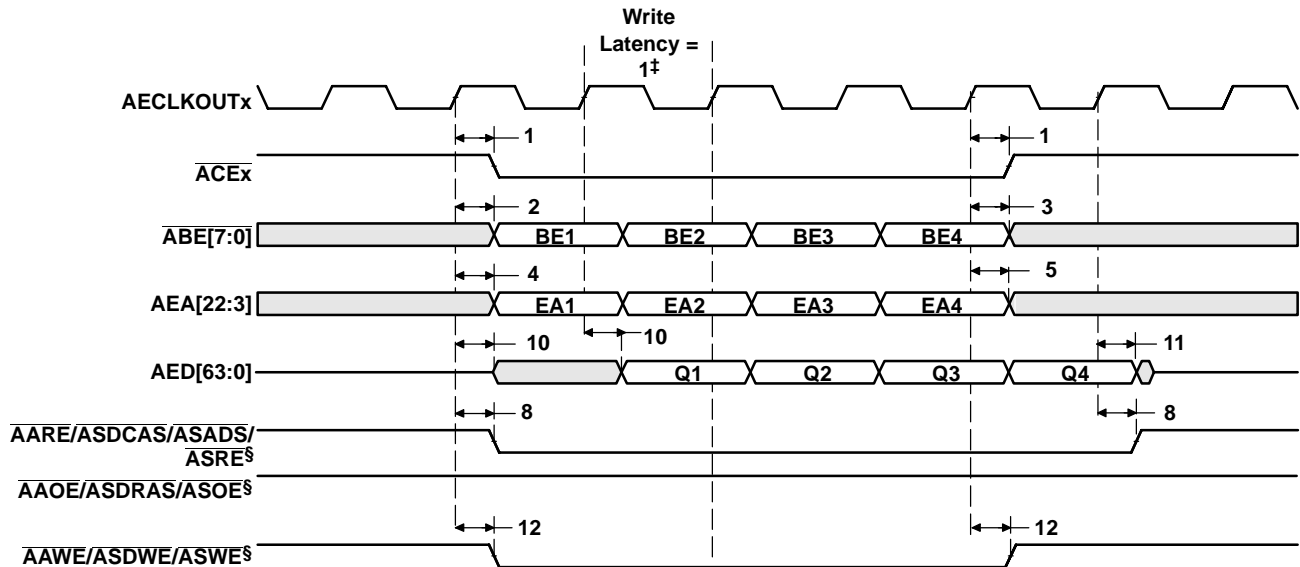**(With Read Latency = 2)†‡**

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† The write latency and the length of $\overline{\text{ACEx}}$ assertion are programmable via the SYNCWL and CEEXT fields, respectively, in the EMIFA CE Space Secondary Control register (CExSEC). In this figure, SYNCWL = 0 and CEEXT = 0.

‡ The following parameters are programmable via the EMIF CE Space Secondary Control register (CExSEC):
  – Read latency (SYNCRL): 0-, 1-, 2-, or 3-cycle read latency
  – Write latency (SYNCWL): 0-, 1-, 2-, or 3-cycle write latency
  – $\overline{\text{ACEx}}$ assertion length (CEEXT): For standard SBSRAM or ZBT SRAM interface, $\overline{\text{ACEx}}$ goes inactive after the final command has been issued (CEEXT = 0). For synchronous FIFO interface with glue, $\overline{\text{ACEx}}$ is active when $\overline{\text{ASOE}}$ is active (CEEXT = 1).
  – Function of $\overline{\text{ASADS}}/\overline{\text{ASRE}}$ (RENEN): For standard SBSRAM or ZBT SRAM interface, $\overline{\text{ASADS}}/\overline{\text{ASRE}}$ acts as $\overline{\text{ASADS}}$ with deselect cycles (RENEN = 0). For FIFO interface, $\overline{\text{ASADS}}/\overline{\text{ASRE}}$ acts as $\overline{\text{ASRE}}$ with NO deselect cycles (RENEN = 1).
  – Synchronization clock (SNCCLK): Synchronized to AECLKOUT1 or AECLKOUT2

§ $\overline{\text{AARE}}/\overline{\text{ASDCAS}}/\overline{\text{ASADS}}/\overline{\text{ASRE}}$, $\overline{\text{AAOE}}/\overline{\text{ASDRAS}}/\overline{\text{ASOE}}$, and $\overline{\text{AAWE}}/\overline{\text{ASDWE}}/\overline{\text{ASWE}}$ operate as $\overline{\text{ASADS}}/\overline{\text{ASRE}}$, $\overline{\text{ASOE}}$, and $\overline{\text{ASWE}}$, respectively, during programmable synchronous interface accesses.

Figure 6–2. Programmable Synchronous Interface Write Timing for EMIFA
(With Write Latency = 0)[†‡§]

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† The write latency and the length of $\overline{\text{ACEx}}$ assertion are programmable via the SYNCWL and CEEXT fields, respectively, in the EMIFA CE Space Secondary Control register (CExSEC). In this figure, SYNCWL = 1 and CEEXT = 0.

‡ The following parameters are programmable via the EMIF CE Space Secondary Control register (CExSEC):
  – Read latency (SYNCRL): 0-, 1-, 2-, or 3-cycle read latency
  – Write latency (SYNCWL): 0-, 1-, 2-, or 3-cycle write latency
  – $\overline{\text{ACEx}}$ assertion length (CEEXT): For standard SBSRAM or ZBT SRAM interface, $\overline{\text{ACEx}}$ goes inactive after the final command has been issued (CEEXT = 0). For synchronous FIFO interface with glue, $\overline{\text{ACEx}}$ is active when $\overline{\text{ASOE}}$ is active (CEEXT = 1).
  – Function of $\overline{\text{ASADS}}/\overline{\text{ASRE}}$ (RENEN): For standard SBSRAM or ZBT SRAM interface, $\overline{\text{ASADS}}/\overline{\text{ASRE}}$ acts as $\overline{\text{ASADS}}$ with deselect cycles (RENEN = 0). For FIFO interface, $\overline{\text{ASADS}}/\overline{\text{ASRE}}$ acts as $\overline{\text{ASRE}}$ with NO deselect cycles (RENEN = 1).
  – Synchronization clock (SNCCLK): Synchronized to AECLKOUT1 or AECLKOUT2

§ $\overline{\text{AARE}}/\overline{\text{ASDCAS}}/\overline{\text{ASADS}}/\overline{\text{ASRE}}$, $\overline{\text{AAOE}}/\overline{\text{ASDRAS}}/\overline{\text{ASOE}}$, and $\overline{\text{AAWE}}/\overline{\text{ASDWE}}/\overline{\text{ASWE}}$ operate as $\overline{\text{ASADS}}/\overline{\text{ASRE}}$, $\overline{\text{AsOE}}$, and $\overline{\text{ASWE}}$, respectively, during programmable synchronous interface accesses.

Figure 6–3. Programmable Synchronous Interface Write Timing for EMIFA (With Write Latency = 1)†‡

7 Synchronous DRAM Timing

Table 7−1. Timing Requirements for Synchronous DRAM Cycles for EMIFA Module (see Figure 7−1)

NO.			−500		−600 −720		UNIT
			MIN	MAX	MIN	MAX	
6	$t_{su(EDV-EKO1H)}$	Setup time, read AEDx valid before AECLKOUTx high	2.1		0.6		ns
7	$t_{h(EKO1H-EDV)}$	Hold time, read AEDx valid after AECLKOUTx high	2.8		2.1		ns

Table 7−2. Switching Characteristics Over Recommended Operating Conditions for Synchronous DRAM Cycles for EMIFA Module (see Figure 7−1−Figure 7−8)
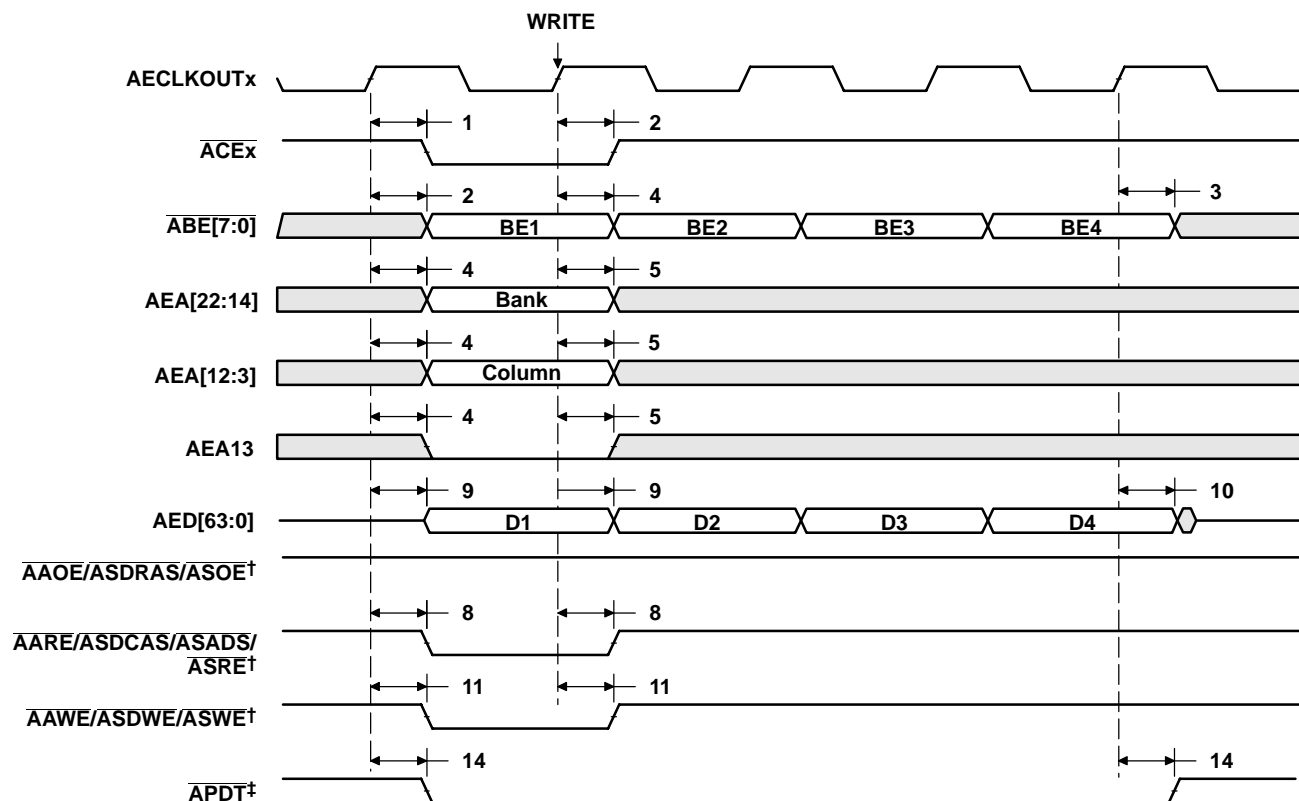
NO.		PARAMETER	−500		−600 −720		UNIT
			MIN	MAX	MIN	MAX	
1	$t_{d(EKO1H-CEV)}$	Delay time, AECLKOUTx high to $\overline{ACEx}$ valid	1.3	6.4	1.3	4.9	ns
2	$t_{d(EKO1H-BEV)}$	Delay time, AECLKOUTx high to $\overline{ABEx}$ valid		6.4		4.9	ns
3	$t_{d(EKO1H-BEIV)}$	Delay time, AECLKOUTx high to $\overline{ABEx}$ invalid	1.3		1.3		ns
4	$t_{d(EKO1H-EAV)}$	Delay time, AECLKOUTx high to AEAx valid		6.4		4.9	ns
5	$t_{d(EKO1H-EAIV)}$	Delay time, AECLKOUTx high to AEAx invalid	1.3		1.3		ns
8	$t_{d(EKO1H-CASV)}$	Delay time, AECLKOUTx high to $\overline{ASDCAS}$ valid	1.3	6.4	1.3	4.9	ns
9	$t_{d(EKO1H-EDV)}$	Delay time, AECLKOUTx high to AEDx valid		6.4		4.9	ns
10	$t_{d(EKO1H-EDIV)}$	Delay time, AECLKOUTx high to AEDx invalid	1.3		1.3		ns
11	$t_{d(EKO1H-WEV)}$	Delay time, AECLKOUTx high to $\overline{ASDWE}$ valid	1.3	6.4	1.3	4.9	ns
12	$t_{d(EKO1H-RAS)}$	Delay time, AECLKOUTx high to $\overline{ASDRAS}$ valid	1.3	6.4	1.3	4.9	ns
13	$t_{d(EKO1H-ACKEV)}$	Delay time, AECLKOUTx high to ASDCKE valid	1.3	6.4	1.3	4.9	ns
14	$t_{d(EKO1H-PDTV)}$	Delay time, AECLKOUTx high to $\overline{APDT}$ valid	1.3	6.4	1.3	4.9	ns
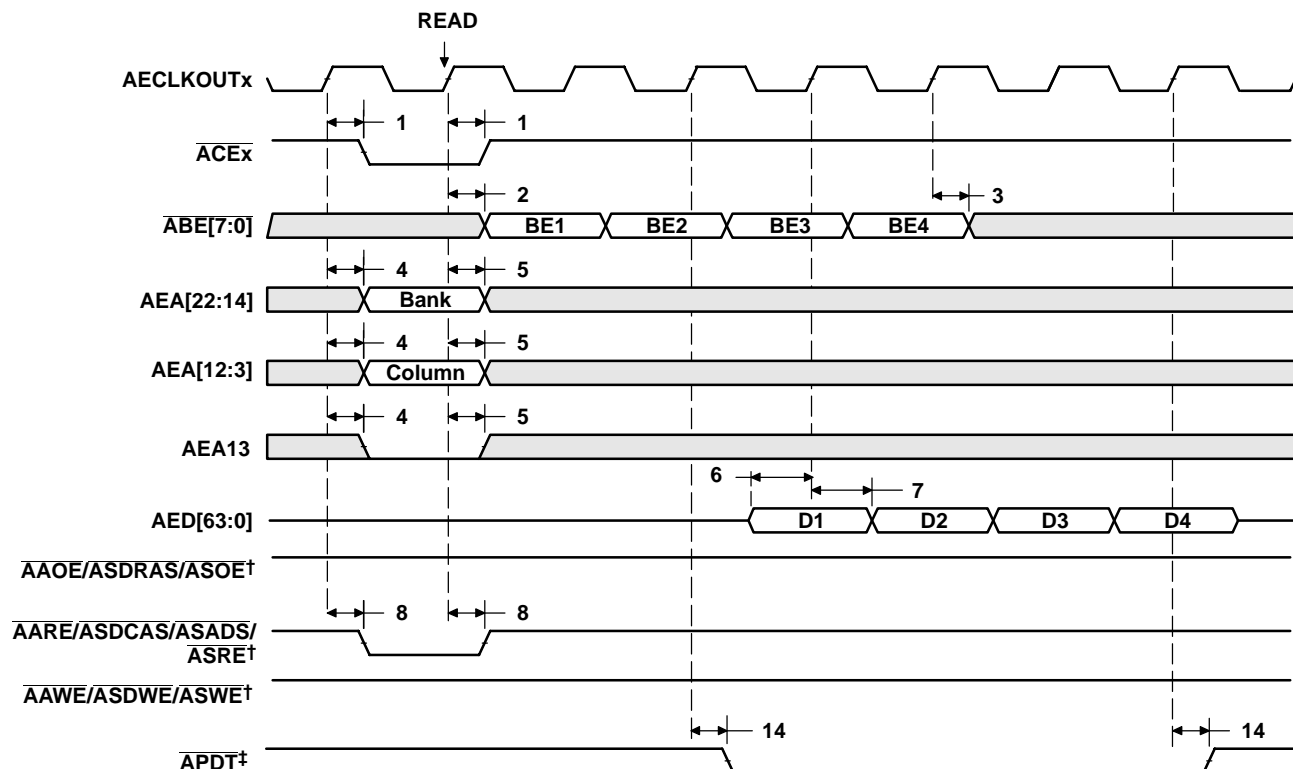
Figure 7–1. SDRAM Read Command (CAS Latency 3) for EMIFA

† $\overline{AARE}/\overline{ASDCAS}/\overline{ASADS}/\overline{ASRE}$, $\overline{AAWE}/\overline{ASDWE}/\overline{ASWE}$, and $\overline{AAOE}/\overline{ASDRAS}/\overline{ASOE}$ operate as $\overline{ASDCAS}$, $\overline{ASDWE}$, and $\overline{ASDRAS}$, respectively, during SDRAM accesses.

‡ $\overline{APDT}$ signal is only asserted when the EDMA is in PDT mode (set the PDTS bit to 1 in the EDMA options parameter RAM). For $\overline{APDT}$ read, data is not latched into EMIF. The PDTRL field in the PDT control register (PDTCTL) configures the latency of the $\overline{APDT}$ signal with respect to the data phase of a read transaction. The latency of the $\overline{APDT}$ signal for a read can be programmed to 0, 1, 2, or 3 by setting PDTRL to 00, 01, 10, or 11, respectively. PDTRL equals 00 (zero latency) in Figure 7–1.
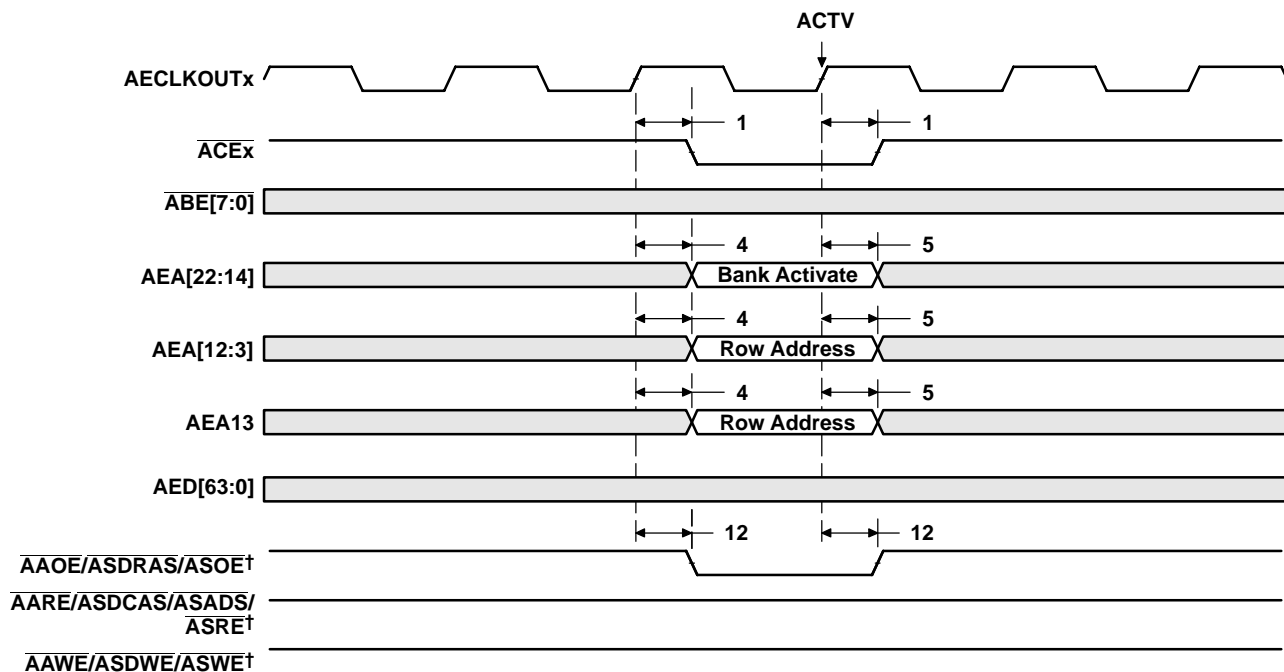
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Figure 7−2. SDRAM Write Command for EMIFA

† $\overline{AARE}/\overline{ASDCAS}/\overline{ASADS}/\overline{ASRE}$, $\overline{AAWE}/\overline{ASDWE}/\overline{ASWE}$, and $\overline{AAOE}/\overline{ASDRAS}/\overline{ASOE}$ operate as $\overline{ASDCAS}$, $\overline{ASDWE}$, and $\overline{ASDRAS}$, respectively, during SDRAM accesses.

‡ $\overline{APDT}$ signal is only asserted when the EDMA is in PDT mode (set the PDTD bit to 1 in the EDMA options parameter RAM). For $\overline{APDT}$ write, data is not driven (in High-Z). The PDTWL field in the PDT control register (PDTCTL) configures the latency of the $\overline{APDT}$ signal with respect to the data phase of a write transaction. The latency of the $\overline{APDT}$ signal for a write transaction can be programmed to 0, 1, 2, or 3 by setting PDTWL to 00, 01, 10, or 11, respectively. PDTWL equals 00 (zero latency) in Figure 7−2.

† $\overline{AARE}/\overline{ASDCAS}/\overline{ASADS}/\overline{ASRE}$, $\overline{AAWE}/\overline{ASDWE}/\overline{ASWE}$, and $\overline{AAOE}/\overline{ASDRAS}/\overline{ASOE}$ operate as $\overline{ASDCAS}$, $\overline{ASDWE}$, and $\overline{ASDRAS}$, respectively, during SDRAM accesses.

Figure 7–3. SDRAM ACTV Command for EMIFA
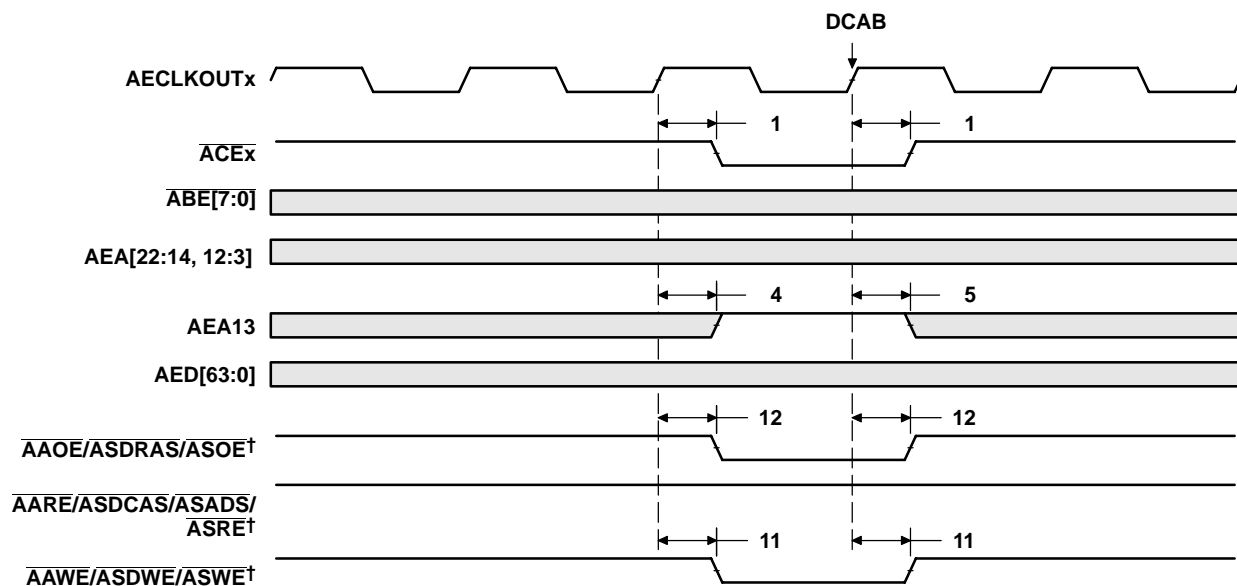
† $\overline{AARE}/\overline{ASDCAS}/\overline{ASADS}/\overline{ASRE}$, $\overline{AAWE}/\overline{ASDWE}/\overline{ASWE}$, and $\overline{AAOE}/\overline{ASDRAS}/\overline{ASOE}$ operate as $\overline{ASDCAS}$, $\overline{ASDWE}$, and $\overline{ASDRAS}$, respectively, during SDRAM accesses.

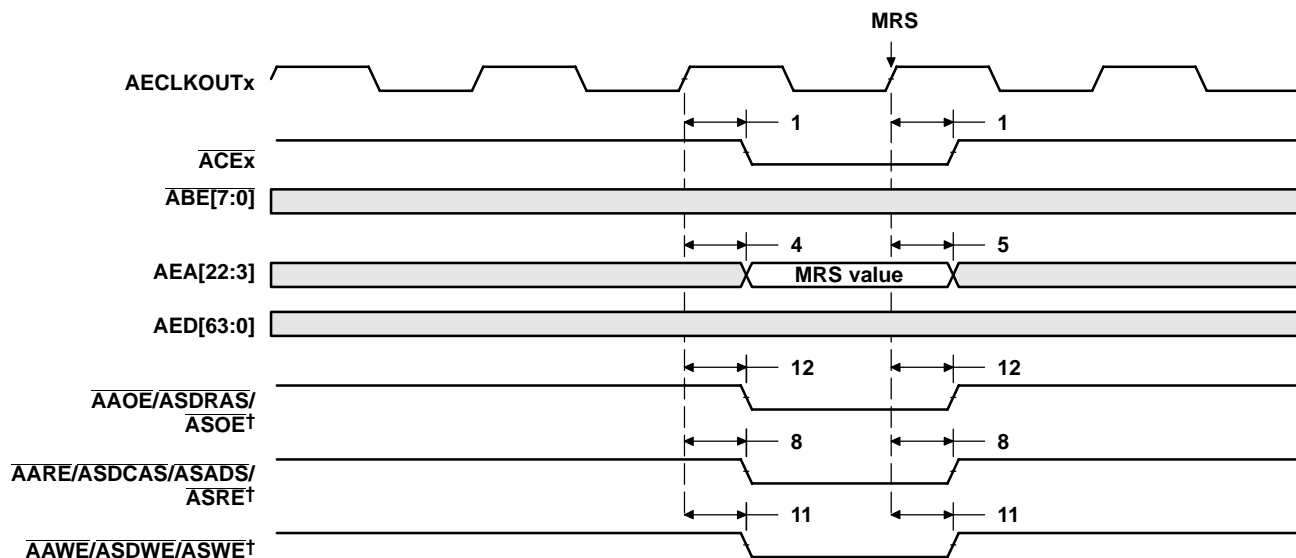**Figure 7–4. SDRAM DCAB Command for EMIFA**
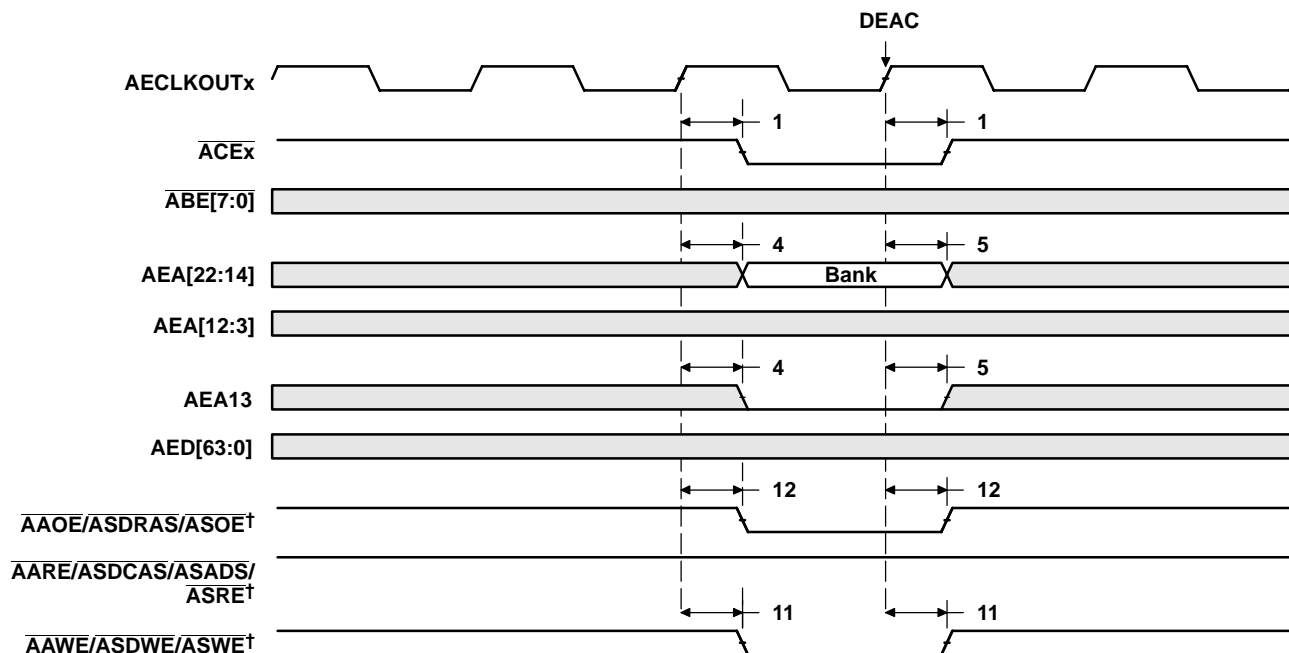
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† AARE/ASDCAS/ASADS/ASRE, AAWE/ASDWE/ASWE, and AAOE/ASDRAS/ASOE operate as ASDCAS, ASDWE, and ASDRAS, respectively, during SDRAM accesses.
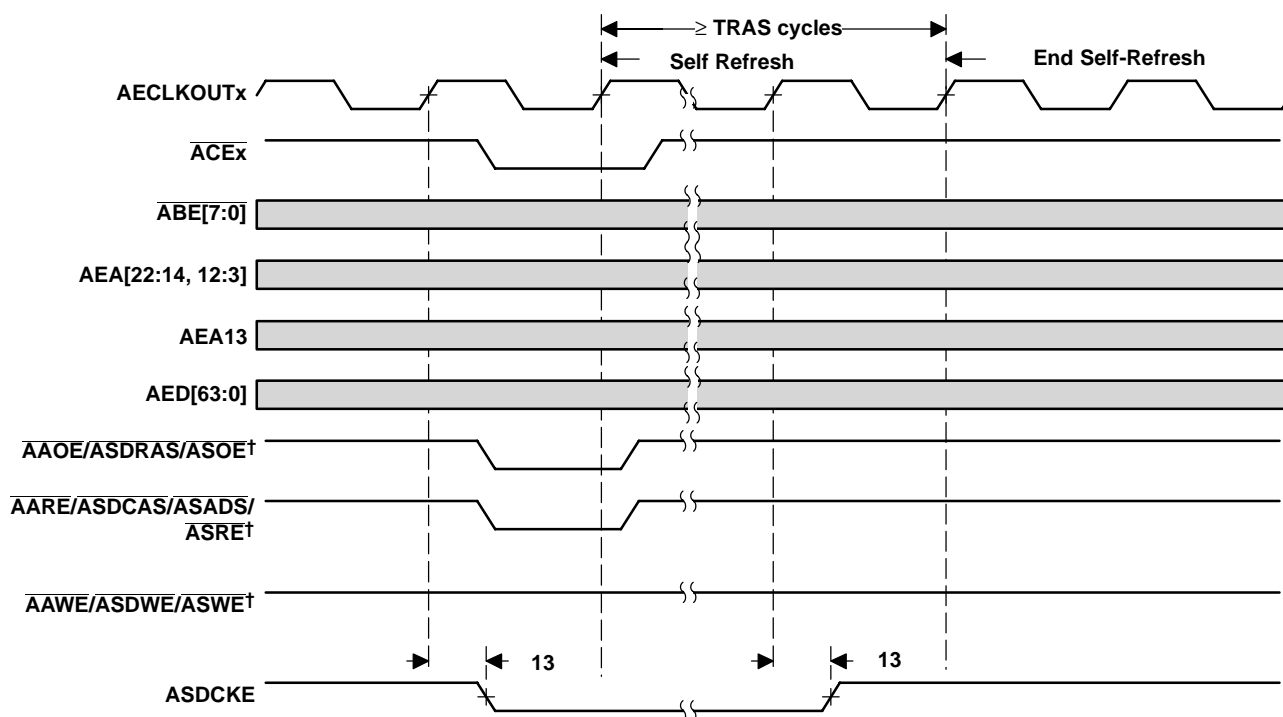
Figure 7−5. SDRAM DEAC Command for EMIFA
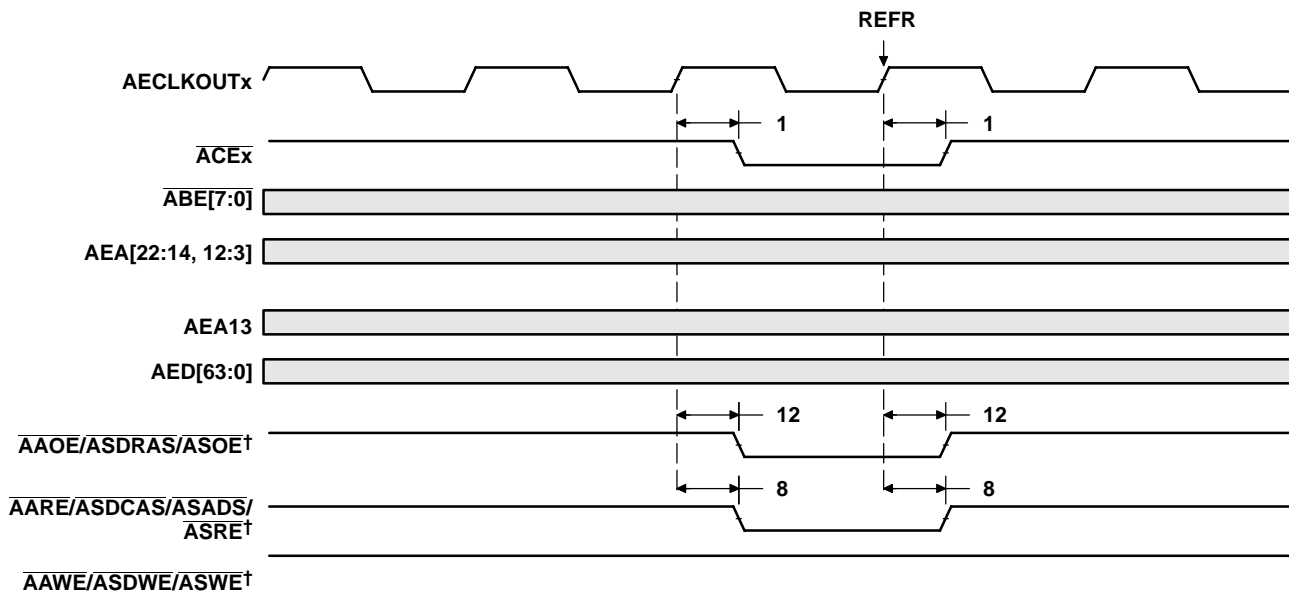
† AARE/ASDCAS/ASADS/ASRE, AAWE/ASDWE/ASWE, and AAOE/ASDRAS/ASOE operate as ASDCAS, ASDWE, and ASDRAS, respectively, during SDRAM accesses.

Figure 7−6. SDRAM REFR Command for EMIFA

† $\overline{AARE}/\overline{ASDCAS}/\overline{ASADS}/\overline{ASRE}$, $\overline{AAWE}/\overline{ASDWE}/\overline{ASWE}$, and $\overline{AAOE}/\overline{ASDRAS}/\overline{ASOE}$ operate as $\overline{ASDCAS}$, $\overline{ASDWE}$, and $\overline{ASDRAS}$, respectively, during SDRAM accesses.

Figure 7−7. SDRAM MRS Command for EMIFA

† $\overline{AARE}/\overline{ASDCAS}/\overline{ASADS}/\overline{ASRE}$, $\overline{AAWE}/\overline{ASDWE}/\overline{ASWE}$, and $\overline{AAOE}/\overline{ASDRAS}/\overline{ASOE}$ operate as $\overline{ASDCAS}$, $\overline{ASDWE}$, and $\overline{ASDRAS}$, respectively, during SDRAM accesses.

Figure 7−8. SDRAM Self-Refresh Timing for EMIFA

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8   HOLD/HOLDA Timing

Table 8–1. Timing Requirements for the HOLD/HOLDA Cycles for EMIFA Module[†] (see Figure 8–1)

NO.			–500		–600 –720		UNIT
			MIN	MAX	MIN	MAX	
3	$t_{h(HOLDAL-HOLDL)}$	Hold time, HOLD low after HOLDA low	E		E		ns

[†] E = the EMIF input clock (AECLKIN, CPU/4 clock, or CPU/6 clock) period in ns for EMIFA.

Table 8–2. Switching Characteristics Over Recommended Operating Conditions for the HOLD/HOLDA Cycles for EMIFA Module[†‡§] (see Figure 8–1)

NO.	PARAMETER		–500		–600 –720		UNIT
			MIN	MAX	MIN	MAX	
1	$t_{d(HOLDL-EMHZ)}$	Delay time, HOLD low to EMIFA Bus high impedance	2E	¶	2E	¶	ns
2	$t_{d(EMHZ-HOLDAL)}$	Delay time, EMIF Bus high impedance to HOLDA low	0	2E	0	2E	ns
4	$t_{d(HOLDH-EMLZ)}$	Delay time, HOLD high to EMIF Bus low impedance	2E	7E	2E	7E	ns
5	$t_{d(EMLZ-HOLDAH)}$	Delay time, EMIFA Bus low impedance to HOLDA high	0	2E	0	2E	ns
6	$t_{d(HOLDL-EKOHZ)}$	Delay time, HOLD low to AECLKOUTx high impedance	2E	¶	2E	¶	ns
7	$t_{d(HOLDH-EKOLZ)}$	Delay time, HOLD high to AECLKOUTx low impedance	2E	7E	2E	7E	ns

[†] E = the EMIF input clock (AECLKIN, CPU/4 clock, or CPU/6 clock) period in ns for EMIFA.
[‡] EMIFA Bus consists of: ACE[3:0], ABE[7:0], AED[63:0], AEA[22:3], AARE/ASDCAS/ASADS/ASRE, AAOE/ASDRAS/ASOE, and AAWE/ASDWE/ASWE, ASDCKE, ASOE3, and APDT.
[§] The EKxHZ bits in the EMIF Global Control register (GBLCTL) determine the state of the AECLKOUTx signals during HOLDA. If EKxHZ = 0, AECLKOUTx continues clocking during Hold mode. If EKxHZ = 1, AECLKOUTx goes to high impedance during Hold mode, as shown in Figure 8–1.
[¶] All pending EMIF transactions are allowed to complete before HOLDA is asserted. If no bus transactions are occurring, then the minimum delay time can be achieved. Also, bus hold can be indefinitely delayed by setting NOHOLD = 1.
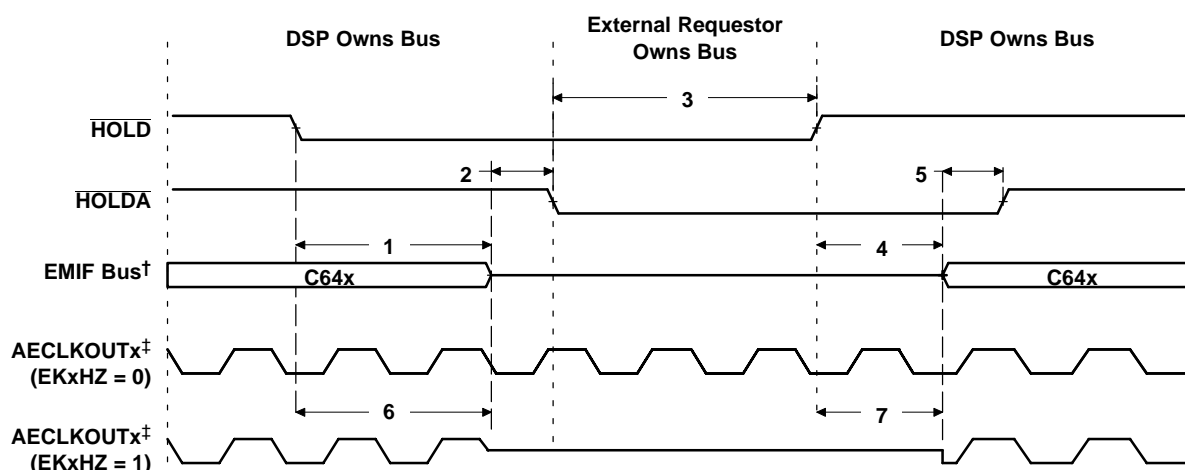
[†] EMIFA Bus consists of: ACE[3:0], ABE[7:0], AED[63:0], AEA[22:3], AARE/ASDCAS/ASADS/ASRE, AAOE/ASDRAS/ASOE, and AAWE/ASDWE/ASWE, ASDCKE, ASOE3, and APDT.
[‡] The EKxHZ bits in the EMIF Global Control register (GBLCTL) determine the state of the AECLKOUTx signals during HOLDA. If EKxHZ = 0, AECLKOUTx continues clocking during Hold mode. If EKxHZ = 1, AECLKOUTx goes to high impedance during Hold mode, as shown in Figure 8–1.

Figure 8–1. HOLD/HOLDA Timing for EMIFA

9   BUSREQ Timing

Table 9–1. Switching Characteristics Over Recommended Operating Conditions for the BUSREQ Cycles for EMIFA Module (see Figure 9–1)

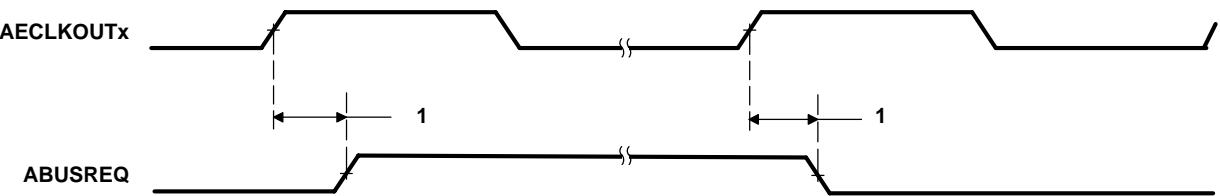| NO.		PARAMETER	−500		−600 −720		UNIT
			MIN	MAX	MIN	MAX	
1	$t_{d(AEKO1H-ABUSRV)}$	Delay time, AECLKOUTx high to ABUSREQ valid	0.6	7.1	1	5.5	ns

Figure 9–1. BUSREQ Timing for EMIFA

10 Reset Timing

Table 10–1. Timing Requirements for Reset (see Figure 10–1)

NO.			–500 –600 –720		UNIT
			MIN	MAX	
1	$t_{w(RST)}$	Width of the $\overline{RESET}$ pulse	250		μs
16	$t_{su(boot)}$	Setup time, boot configuration bits valid before $\overline{RESET}$ high[†]	4E or 4C[‡]		ns
17	$t_{h(boot)}$	Hold time, boot configuration bits valid after $\overline{RESET}$ high[†]	4P[§]		ns
18	$t_{su(PCLK-RSTH)}$	Setup time, PCLK active before $\overline{RESET}$ high[¶]	32N		ns

[†] AEA[22:19], LENDIAN, PCIEEAI, and HD5/AD5 are the boot configuration pins during device reset.
[‡] E = 1/AECLKIN clock frequency in ns. C = 1/CLKIN clock frequency in ns.
  Select the MIN parameter value, whichever value is larger.
[§] P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
[¶] N = the PCI input clock (PCLK) period in ns. When PCI is enabled (PCI_EN = 1), this parameter *must* be met.

Table 10–2. Switching Characteristics Over Recommended Operating Conditions During Reset[§#‖]
(see Figure 10–1)

NO.		PARAMETER	–500 –600 –720		UNIT
			MIN	MAX	
2	$t_{d(RSTL-ECKI)}$	Delay time, $\overline{RESET}$ low to AECLKIN synchronized internally	2E	3P + 20E	ns
3	$t_{d(RSTH-ECKI)}$	Delay time, $\overline{RESET}$ high to AECLKIN synchronized internally	2E	8P + 20E	ns
4	$t_{d(RSTL-ECKO1HZ)}$	Delay time, $\overline{RESET}$ low to AECLKOUT1 high impedance	2E		ns
5	$t_{d(RSTH-ECKO1V)}$	Delay time, $\overline{RESET}$ high to AECLKOUT1 valid		8P + 20E	ns
6	$t_{d(RSTL-EMIFZHZ)}$	Delay time, $\overline{RESET}$ low to EMIF Z high impedance	2E	3P + 4E	ns
7	$t_{d(RSTH-EMIFZV)}$	Delay time, $\overline{RESET}$ high to EMIF Z valid	16E	8P + 20E	ns
8	$t_{d(RSTL-EMIFHIV)}$	Delay time, $\overline{RESET}$ low to EMIF high group invalid	2E		ns
9	$t_{d(RSTH-EMIFHV)}$	Delay time, $\overline{RESET}$ high to EMIF high group valid		8P + 20E	ns
10	$t_{d(RSTL-EMIFLIV)}$	Delay time, $\overline{RESET}$ low to EMIF low group invalid	2E		ns
11	$t_{d(RSTH-EMIFLV)}$	Delay time, $\overline{RESET}$ high to EMIF low group valid		8P + 20E	ns
12	$t_{d(RSTL-LOWIV)}$	Delay time, $\overline{RESET}$ low to low group invalid	0		ns
13	$t_{d(RSTH-LOWV)}$	Delay time, $\overline{RESET}$ high to low group valid		11P	ns
14	$t_{d(RSTL-ZHZ)}$	Delay time, $\overline{RESET}$ low to Z group high impedance	0		ns
15	$t_{d(RSTH-ZV)}$	Delay time, $\overline{RESET}$ high to Z group valid	2P	8P	ns

[§] P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
[#] E = the EMIF input clock (AECLKIN, CPU/4 clock, or CPU/6 clock) period in ns for EMIFA.
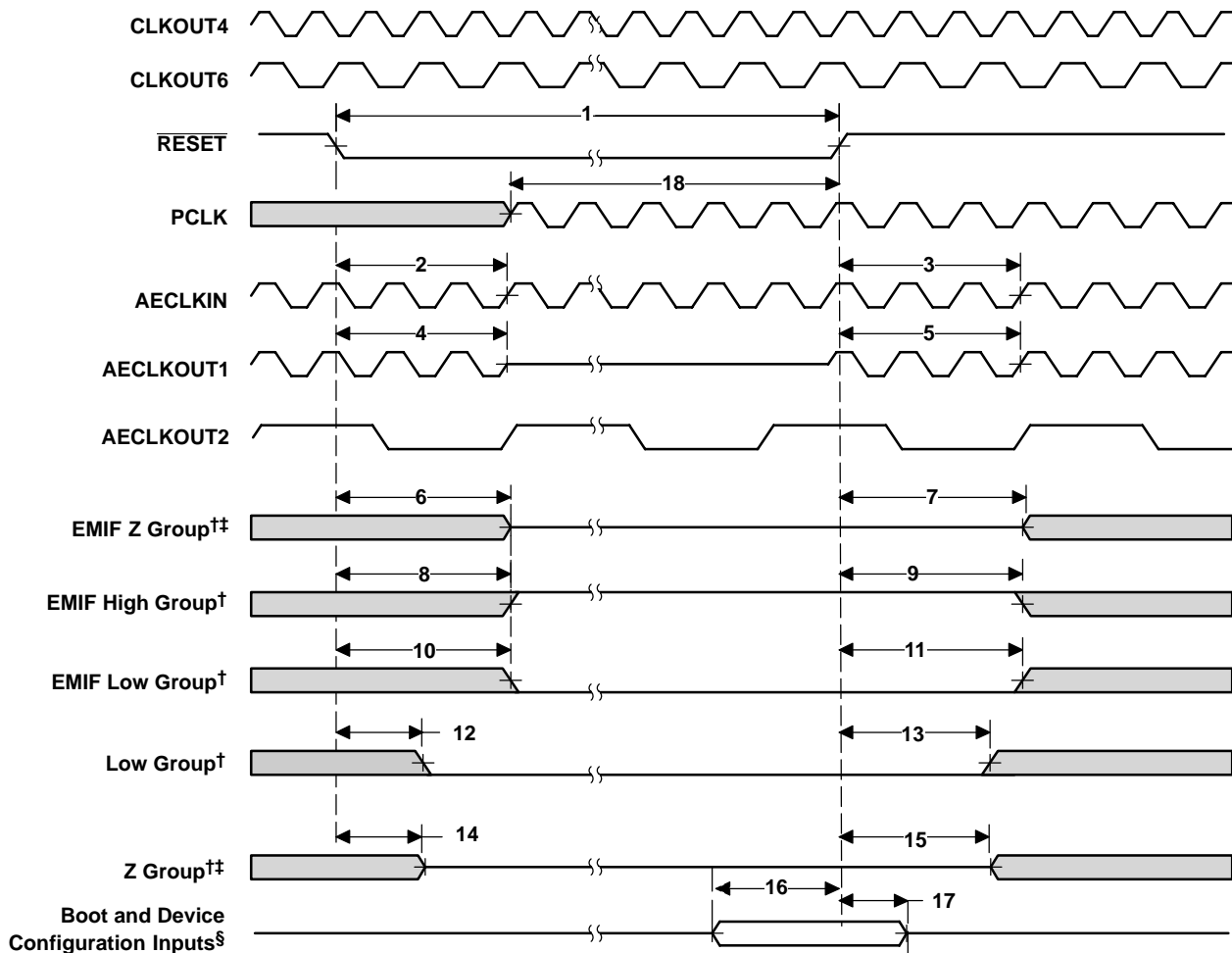[‖] EMIF Z group consists of:　AEA[22:3], AED[63:0], $\overline{ACE[3:0]}$, $\overline{ABE[7:0]}$, $\overline{AARE}/\overline{ASDCAS}/\overline{ASADS}/\overline{ASRE}$, $\overline{AAWE}/\overline{ASDWE}/\overline{ASWE}$,
  　and $\overline{AAOE}/\overline{ASDRAS}/\overline{ASOE}$, $\overline{ASOE3}$, ASDCKE, and $\overline{APDT}$.
  EMIF high group consists of: $\overline{AHOLDA}$ (when the corresponding $\overline{HOLD}$ input is high)
  EMIF low group consists of: ABUSREQ; $\overline{AHOLDA}$ (when the corresponding $\overline{HOLD}$ input is low)
  Low group consists of:　XSP_CS, XSP_CLK/MDCLK, and XSP_DO/MDIO; all of which apply only when PCI EEPROM is enabled
  　(with PCI_EN = 1). Otherwise, the XSP_CLK/MDCLK and XSP_DO/MDIO pins are in the
  　Z group. For more details on the PCI configuration pins, see the Device Configurations section of this data sheet.
  Z group consists of:　HD[31:0]/AD[31:0] and the muxed EMAC output pins, XSP_CLK/MDCLK, XSP_DO/MDIO, CLKX0,
  　CLKX1, FSX0, FSX1, DX0, DX1, CLKR0, CLKR1, FSR0, FSR1, TOUT0, TOUT1,
  　GP0[8]/$\overline{PCI66}$, GP0[7:0], GP0[10]/$\overline{PCBE3}$, HR/$\overline{W}$/$\overline{PCBE2}$, $\overline{HDS2}$/$\overline{PCBE1}$, $\overline{PCBE0}$, GP0[13]/$\overline{PINTA}$,
  　GP0[11]/$\overline{PREQ}$, HDS1/$\overline{PSERR}$, $\overline{HCS}$/$\overline{PPERR}$, HCNTL1/$\overline{PDEVSEL}$, $\overline{HAS}$/PPAR, HCNTL0/$\overline{PSTOP}$,
  　$\overline{HHWIL}$/$\overline{PTRDY}$ (16-bit HPI mode only), $\overline{HRDY}$/$\overline{PIRDY}$, and $\overline{HINT}$/$\overline{PFRAME}$.

† EMIF Z group consists of: AEA[22:3], AED[63:0], $\overline{ACE}$[3:0], $\overline{ABE}$[7:0], $\overline{AARE}$/ASDCAS/ASADS/ASRE,$\overline{AAWE}$/ASDWE/ASWE, and $\overline{AAOE}$/ASDRAS/ASOE, ASOE3, ASDCKE, and $\overline{APDT}$.

EMIF high group consists of: $\overline{AHOLDA}$ (when the corresponding $\overline{HOLD}$ input is high)

EMIF low group consists of: ABUSREQ; $\overline{AHOLDA}$ (when the corresponding $\overline{HOLD}$ input is low)

Low group consists of: XSP_CS, XSP_CLK/MDCLK, and XSP_DO/MDIO; all of which apply only when PCI EEPROM is enabled (with PCI_EN = 1). Otherwise, the XSP_CLK/MDCLK and XSP_DO/MDIO pins are in the Z group. For more details on the PCI configuration pins, see the Device Configurations section of this data sheet.

Z group consists of: HD[31:0]/AD[31:0] and the muxed EMAC output pins, XSP_CLK/MDCLK, XSP_DO/MDIO, CLKX0, CLKX1, FSX0, FSX1, DX0, DX1, CLKR0, CLKR1, FSR0, FSR1, TOUT0, TOUT1, GP0[8]/$\overline{PCI66}$, GP0[7:0], GP0[10]/$\overline{PCBE3}$, HR/$\overline{W}$/PCBE2, $\overline{HDS2}$/PCBE1, $\overline{PCBE0}$, GP0[13]/$\overline{PINTA}$, GP0[11]/$\overline{PREQ}$, $\overline{HDS1}$/PSERR, $\overline{HCS}$/PPERR, HCNTL1/PDEVSEL, $\overline{HAS}$/PPAR, HCNTL0/$\overline{PSTOP}$, HHWIL/$\overline{PTRDY}$ (16-bit HPI mode only), $\overline{HRDY}$/PIRDY, and $\overline{HINT}$/PFRAME.

‡ If AEA[22:19], LENDIAN, PCIEEAI, MAC_EN, and HD5/AD5 pins are actively driven, care must be taken to ensure no timing contention between parameters 6, 7, 14, 15, 16, and 17.

§ Boot and Device Configurations Inputs (during reset) include: AEA[22:19],LENDIAN, PCIEEAI, MAC_EN, and HD5/AD5. The LENDIAN and MAC_EN configuration inputs are muxed with timer output pins and driven low after reset; therefore, it is recommended that external pullup/pulldown resistors be used to configure these pins during reset and that these pins **not** be driven through external logic. The PCI_EN pin **must** be valid at all times and the user **must not** switch values throughout device operation.

Figure 10–1. Reset Timing†

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11 External Interrupt Timing

Table 11–1. Timing Requirements for External Interrupts† (see Figure 11–1)

NO.			−500 −600 −720		UNIT
			MIN	MAX	UNIT
1	$t_{w(ILOW)}$	Width of the NMI interrupt pulse low	4P		ns
		Width of the EXT_INT interrupt pulse low	8P		ns
2	$t_{w(IHIGH)}$	Width of the NMI interrupt pulse high	4P		ns
		Width of the EXT_INT interrupt pulse high	8P		ns

† P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
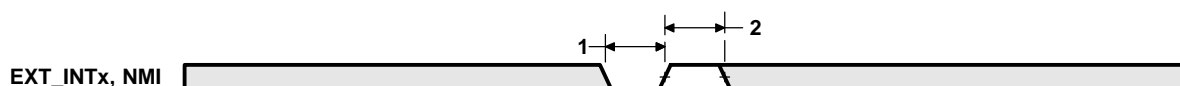
Figure 11–1. External/NMI Interrupt Timing

12   Inter-Integrated Circuits (I2C) Timing

Table 12–1. Timing Requirements for I2C Timings[†] (see Figure 12–1)

NO.			−500 −600 −720				UNIT
			STANDARD MODE		FAST MODE		
			MIN	MAX	MIN	MAX	
1	$t_{c(SCL)}$	Cycle time, SCL	10		2.5		µs
2	$t_{su(SCLH\text{-}SDAL)}$	Setup time, SCL high before SDA low (for a repeated START condition)	4.7		0.6		µs
3	$t_{h(SCLL\text{-}SDAL)}$	Hold time, SCL low after SDA low (for a START and a repeated START condition)	4		0.6		µs
4	$t_{w(SCLL)}$	Pulse duration, SCL low	4.7		1.3		µs
5	$t_{w(SCLH)}$	Pulse duration, SCL high	4		0.6		µs
6	$t_{su(SDAV\text{-}SDLH)}$	Setup time, SDA valid before SCL high	250		100[‡]		ns
7	$t_{h(SDA\text{-}SDLL)}$	Hold time, SDA valid after SCL low (For I2C bus™ devices)	0[§]		0[§]	0.9[¶]	µs
8	$t_{w(SDAH)}$	Pulse duration, SDA high between STOP and START conditions	4.7		1.3		µs
9	$t_{r(SDA)}$	Rise time, SDA		1000	$20 + 0.1C_b$[#]	300	ns
10	$t_{r(SCL)}$	Rise time, SCL		1000	$20 + 0.1C_b$[#]	300	ns
11	$t_{f(SDA)}$	Fall time, SDA		300	$20 + 0.1C_b$[#]	300	ns
12	$t_{f(SCL)}$	Fall time, SCL		300	$20 + 0.1C_b$[#]	300	ns
13	$t_{su(SCLH\text{-}SDAH)}$	Setup time, SCL high before SDA high (for STOP condition)	4		0.6		µs
14	$t_{w(SP)}$	Pulse duration, spike (must be suppressed)			0	50	ns
15	C_b[#]	Capacitive load for each bus line	400			400	pF

[†] The I2C pins SDA and SCL do not feature fail-safe I/O buffers. These pins could potentially draw current when the device is powered down.

[‡] A Fast-mode I2C-bus™ device can be used in a Standard-mode I2C-bus™ system, but the requirement $t_{su(SDA\text{-}SCLH)} \geq 250$ ns must then be met. This will automatically be the case if the device does not stretch the LOW period of the SCL signal. If such a device does stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line t_r max + $t_{su(SDA\text{-}SCLH)}$ = 1000 + 250 = 1250 ns (according to the Standard-mode I2C-Bus Specification) before the SCL line is released.

[§] A device must internally provide a hold time of at least 300 ns for the SDA signal (referred to the V_{IHmin} of the SCL signal) to bridge the undefined region of the falling edge of SCL.

[¶] The maximum $t_{h(SDA\text{-}SCLL)}$ has only to be met if the device does not stretch the low period [$t_{w(SCLL)}$] of the SCL signal.

[#] C_b = total capacitance of one bus line in pF. If mixed with HS-mode devices, faster fall-times are allowed.
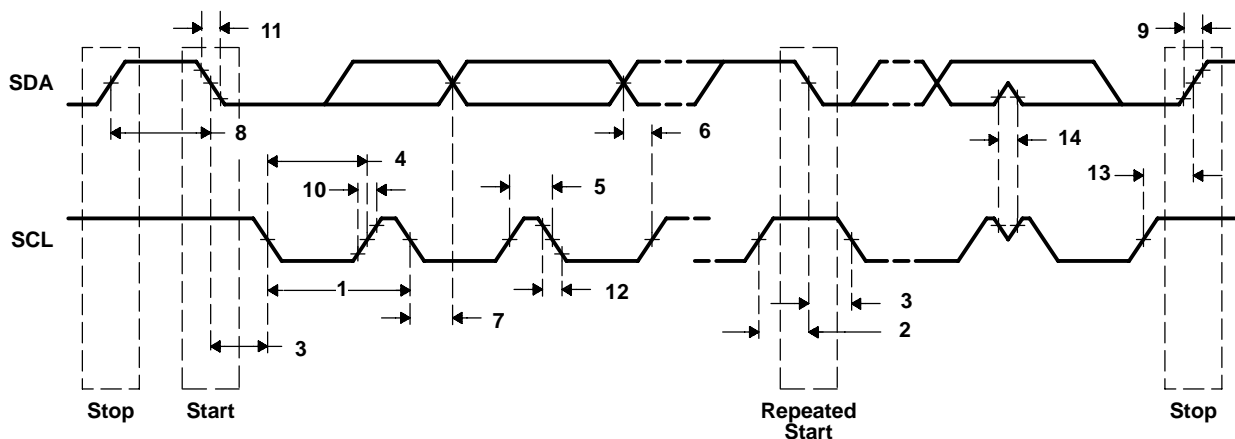

Figure 12–1. I2C Receive Timings

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Table 12–2. Switching Characteristics for I2C Timings† (see Figure 12–2)

NO.	PARAMETER		–500 –600 –720				UNIT
			STANDARD MODE		FAST MODE		
			MIN	MAX	MIN	MAX	
16	$t_{c(SCL)}$	Cycle time, SCL	10		2.5		µs
17	$t_{d(SCLH\text{-}SDAL)}$	Delay time, SCL high to SDA low (for a repeated START condition)	4.7		0.6		µs
18	$t_{d(SDAL\text{-}SCLL)}$	Delay time, SDA low to SCL low (for a START and a repeated START condition)	4		0.6		µs
19	$t_{w(SCLL)}$	Pulse duration, SCL low	4.7		1.3		µs
20	$t_{w(SCLH)}$	Pulse duration, SCL high	4		0.6		µs
21	$t_{d(SDAV\text{-}SDLH)}$	Delay time, SDA valid to SCL high	250		100		ns
22	$t_{v(SDLL\text{-}SDAV)}$	Valid time, SDA valid after SCL low (For I2C bus™ devices)	0		0	0.9	µs
23	$t_{w(SDAH)}$	Pulse duration, SDA high between STOP and START conditions	4.7		1.3		µs
24	$t_{r(SDA)}$	Rise time, SDA		1000	$20 + 0.1C_b$†	300	ns
25	$t_{r(SCL)}$	Rise time, SCL		1000	$20 + 0.1C_b$†	300	ns
26	$t_{f(SDA)}$	Fall time, SDA		300	$20 + 0.1C_b$†	300	ns
27	$t_{f(SCL)}$	Fall time, SCL		300	$20 + 0.1C_b$†	300	ns
28	$t_{d(SCLH\text{-}SDAH)}$	Delay time, SCL high to SDA high (for STOP condition)	4		0.6		µs
29	C_p	Capacitance for each I2C pin		10		10	pF

† C_b = total capacitance of one bus line in pF. If mixed with HS-mode devices, faster fall-times are allowed.
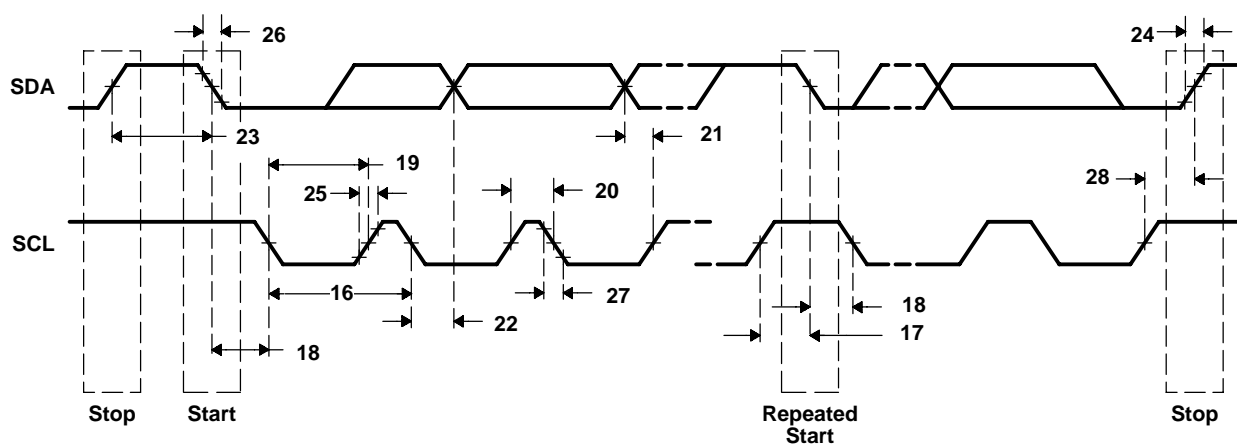

Figure 12–2. I2C Transmit Timings

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13 Host-Port Interface (HPI) Timing

Table 13–1. Timing Requirements for Host-Port Interface Cycles[†‡] (see Figure 13–1 through Figure 13–8)

NO.			−500 −600 −720		UNIT
			MIN	**MAX**	
1	$t_{su(SELV-HSTBL)}$	Setup time, select signals[§] valid before $\overline{HSTROBE}$ low	5		ns
2	$t_{h(HSTBL-SELV)}$	Hold time, select signals[§] valid after $\overline{HSTROBE}$ low	2.4		ns
3	$t_{w(HSTBL)}$	Pulse duration, $\overline{HSTROBE}$ low	4P[¶]		ns
4	$t_{w(HSTBH)}$	Pulse duration, $\overline{HSTROBE}$ high between consecutive accesses	4P		ns
10	$t_{su(SELV-HASL)}$	Setup time, select signals[§] valid before $\overline{HAS}$ low	5		ns
11	$t_{h(HASL-SELV)}$	Hold time, select signals[§] valid after $\overline{HAS}$ low	2		ns
12	$t_{su(HDV-HSTBH)}$	Setup time, host data valid before $\overline{HSTROBE}$ high	5		ns
13	$t_{h(HSTBH-HDV)}$	Hold time, host data valid after $\overline{HSTROBE}$ high	2.8		ns
14	$t_{h(HRDYL-HSTBL)}$	Hold time, $\overline{HSTROBE}$ low after $\overline{HRDY}$ low. $\overline{HSTROBE}$ should not be inactivated until $\overline{HRDY}$ is active (low); otherwise, HPI writes will not complete properly.	2		ns
18	$t_{su(HASL-HSTBL)}$	Setup time, $\overline{HAS}$ low before $\overline{HSTROBE}$ low	2		ns
19	$t_{h(HSTBL-HASL)}$	Hold time, $\overline{HAS}$ low after $\overline{HSTROBE}$ low	2.1		ns

[†] $\overline{HSTROBE}$ refers to the following logical operation on $\overline{HCS}$, $\overline{HDS1}$, and $\overline{HDS2}$: [NOT($\overline{HDS1}$ XOR $\overline{HDS2}$)] OR $\overline{HCS}$.
[‡] P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
[§] Select signals include: HCNTL[1:0] and HR/$\overline{W}$. For HPI16 mode only, select signals also include HHWIL.
[¶] Select the parameter value of 4P or 12.5 ns, whichever is larger.

Table 13–2. Switching Characteristics Over Recommended Operating Conditions During Host-Port Interface Cycles[†‡] (see Figure 13–1 through Figure 13–8)

NO.		PARAMETER	−500 −600 −720		UNIT
			MIN	**MAX**	
6	$t_{d(HSTBL-HRDYH)}$	Delay time, $\overline{HSTROBE}$ low to $\overline{HRDY}$ high[#]	1.3	4P + 8	ns
7	$t_{d(HSTBL-HDLZ)}$	Delay time, $\overline{HSTROBE}$ low to HD low impedance for an HPI read	2		ns
8	$t_{d(HDV-HRDYL)}$	Delay time, HD valid to $\overline{HRDY}$ low	−3		ns
9	$t_{oh(HSTBH-HDV)}$	Output hold time, HD valid after $\overline{HSTROBE}$ high	1.5		ns
15	$t_{d(HSTBH-HDHZ)}$	Delay time, $\overline{HSTROBE}$ high to HD high impedance		12	ns
16	$t_{d(HSTBL-HDV)}$	Delay time, $\overline{HSTROBE}$ low to HD valid (HPI16 mode, 2nd half-word only)		4P + 8	ns

[†] $\overline{HSTROBE}$ refers to the following logical operation on $\overline{HCS}$, $\overline{HDS1}$, and $\overline{HDS2}$: [NOT($\overline{HDS1}$ XOR $\overline{HDS2}$)] OR $\overline{HCS}$.
[‡] P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
[#] This parameter is used during HPID reads and writes. For reads, at the beginning of a word transfer (HPI32) or the first half-word transfer (HPI16) on the falling edge of $\overline{HSTROBE}$, the HPI sends the request to the EDMA internal address generation hardware, and $\overline{HRDY}$ remains high until the EDMA internal address generation hardware loads the requested data into HPID. For writes, $\overline{HRDY}$ goes high if the internal write buffer is full.
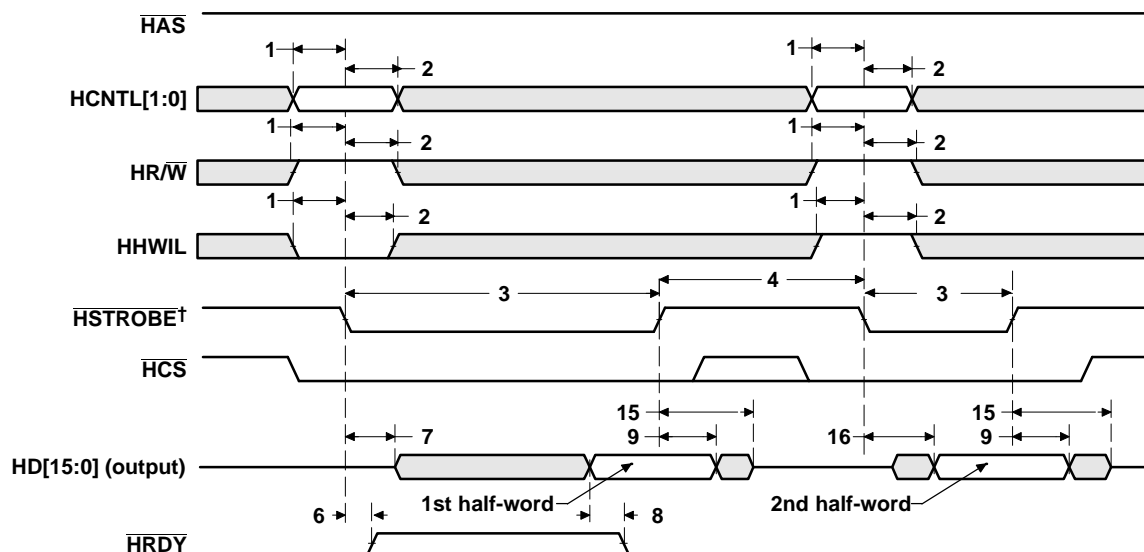
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† HSTROBE refers to the following logical operation on HCS, HDS1, and HDS2: [NOT(HDS1 XOR HDS2)] OR HCS.

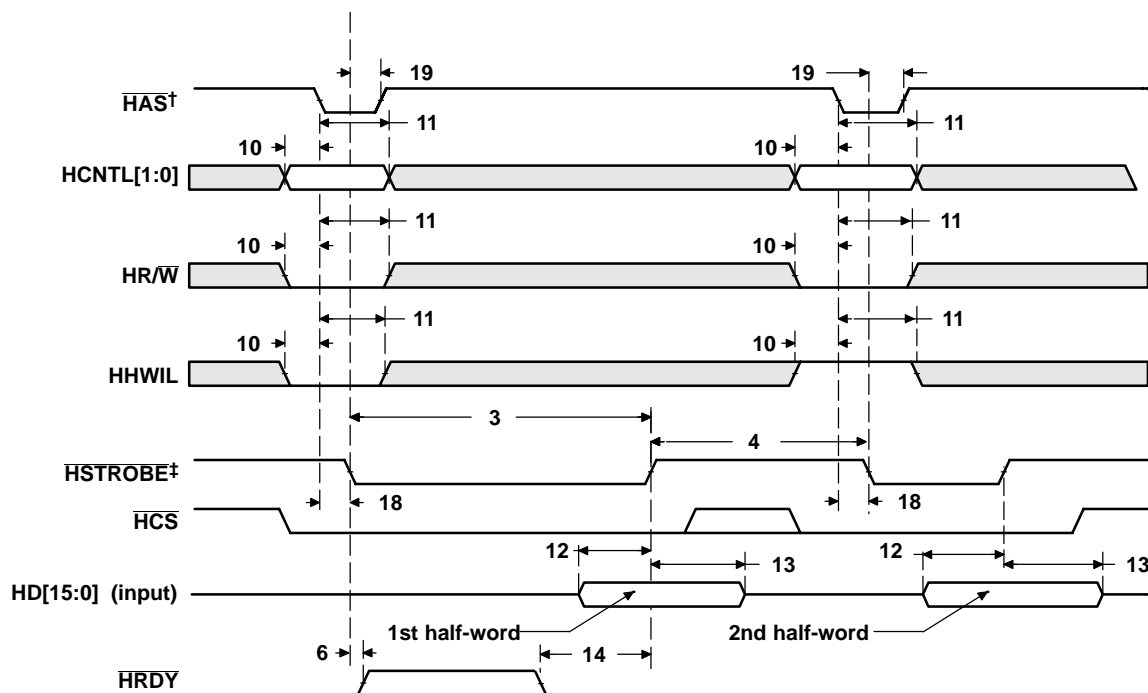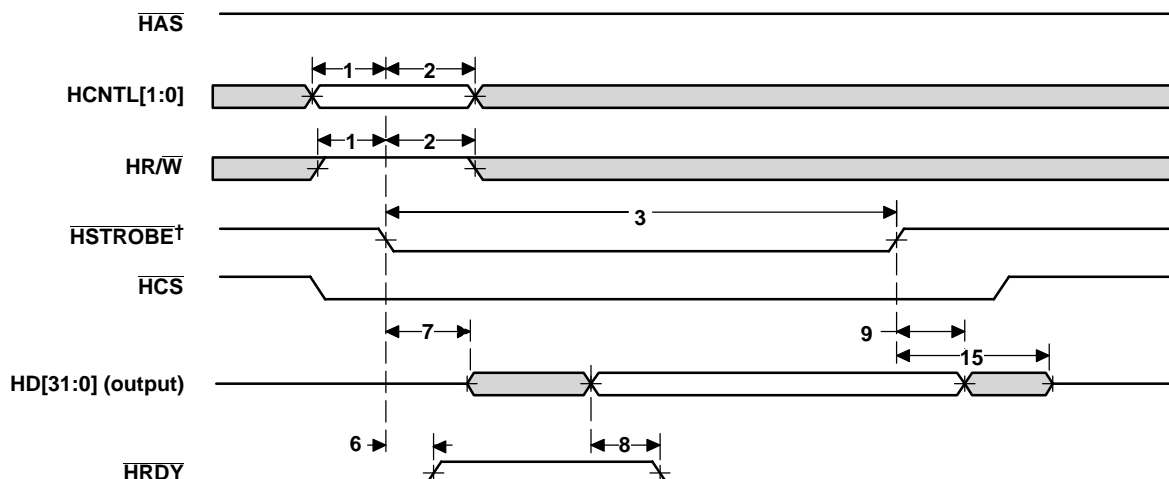**Figure 13–1. HPI16 Read Timing (HAS Not Used, Tied High)**
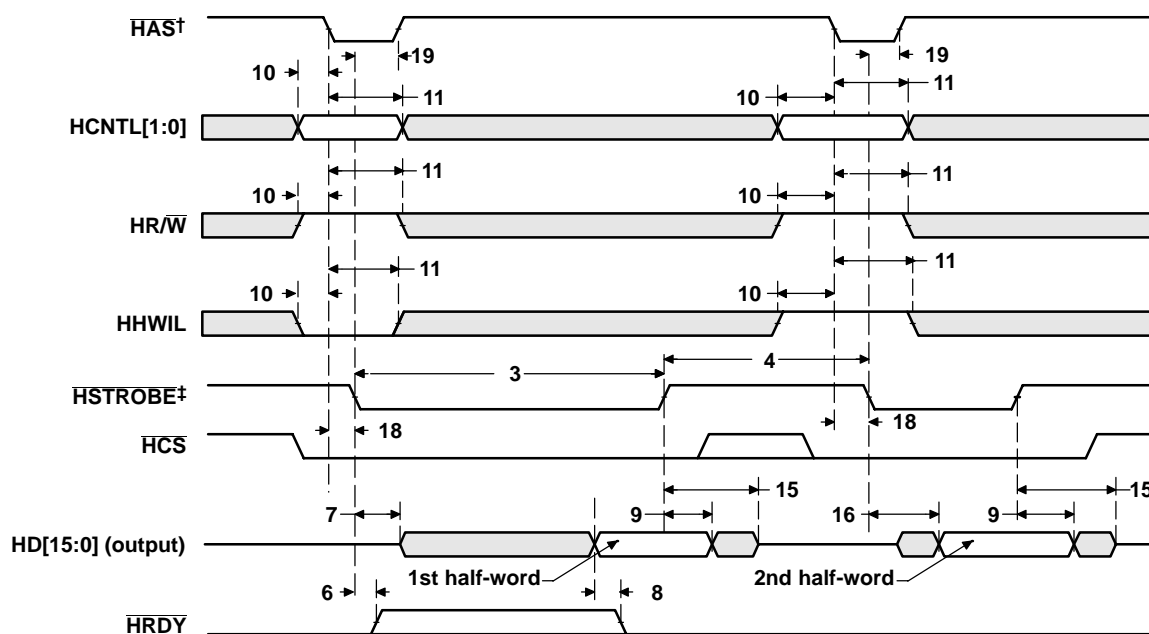
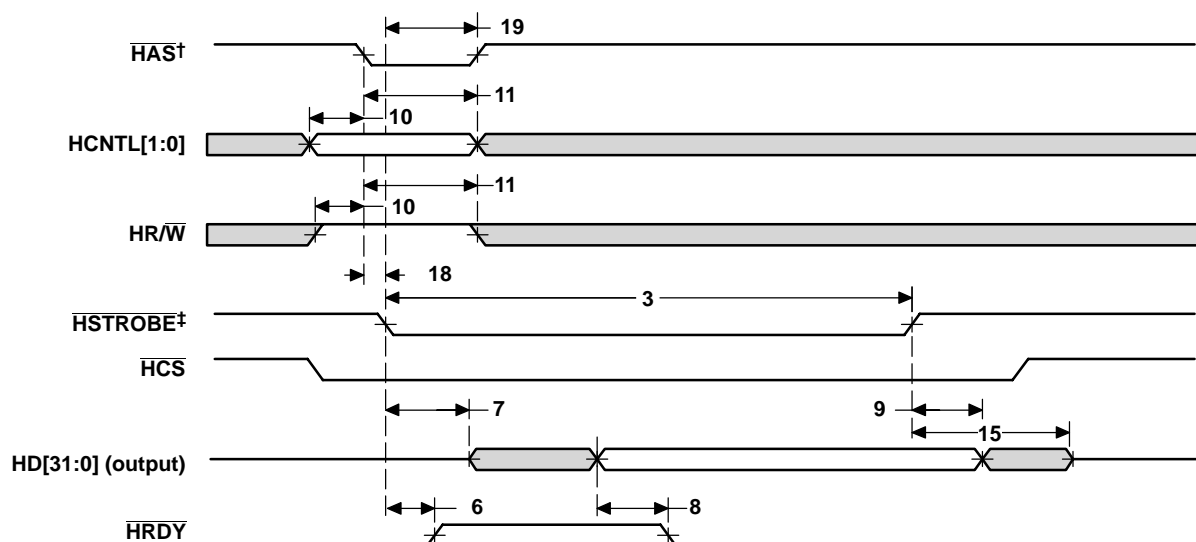† For correct operation, strobe the HAS signal only once per HSTROBE active cycle.
‡ HSTROBE refers to the following logical operation on HCS, HDS1, and HDS2: [NOT(HDS1 XOR HDS2)] OR HCS.

Figure 13–2. HPI16 Read Timing (HAS Used)
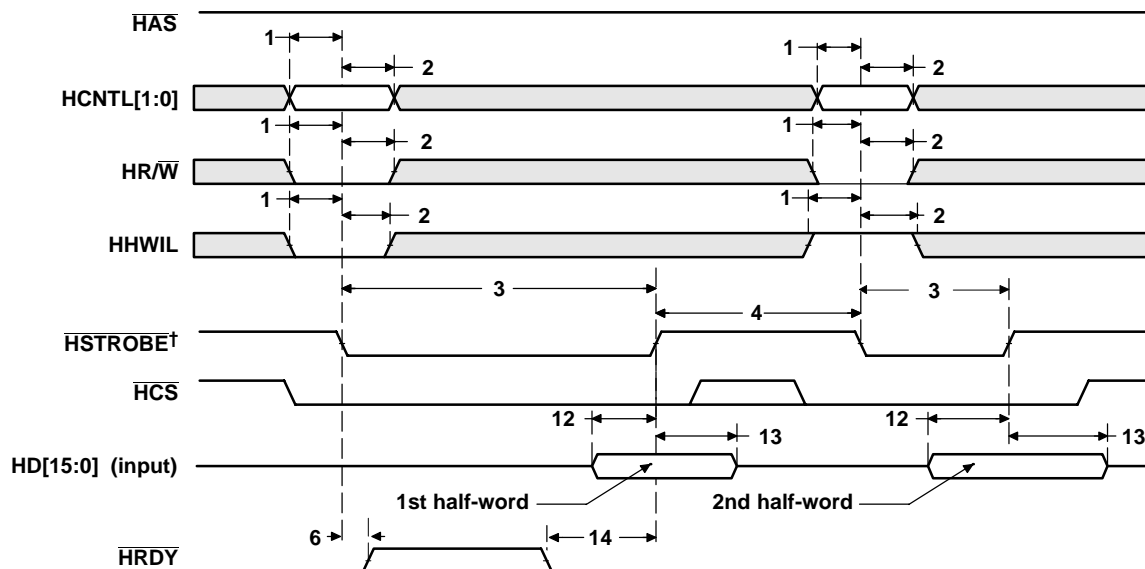
† $\overline{\text{HSTROBE}}$ refers to the following logical operation on $\overline{\text{HCS}}$, $\overline{\text{HDS1}}$, and $\overline{\text{HDS2}}$: [NOT($\overline{\text{HDS1}}$ XOR $\overline{\text{HDS2}}$)] OR $\overline{\text{HCS}}$.

Figure 13–3. HPI16 Write Timing ($\overline{\text{HAS}}$ Not Used, Tied High)

† For correct operation, strobe the $\overline{\text{HAS}}$ signal only once per $\overline{\text{HSTROBE}}$ active cycle.
‡ $\overline{\text{HSTROBE}}$ refers to the following logical operation on $\overline{\text{HCS}}$, $\overline{\text{HDS1}}$, and $\overline{\text{HDS2}}$: [NOT($\overline{\text{HDS1}}$ XOR $\overline{\text{HDS2}}$)] OR $\overline{\text{HCS}}$.

Figure 13–4. HPI16 Write Timing ($\overline{\text{HAS}}$ Used)

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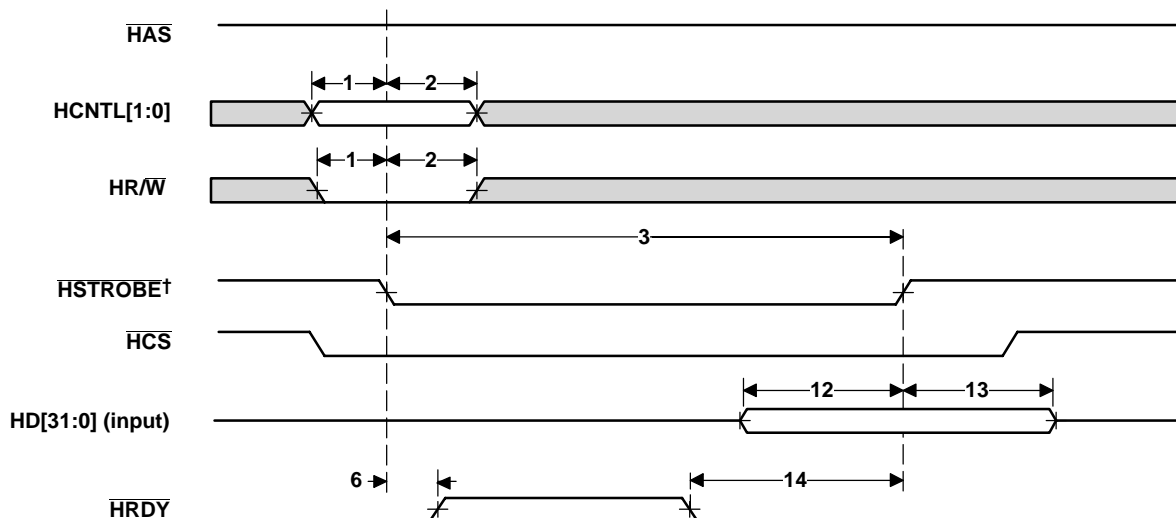† HSTROBE refers to the following logical operation on HCS, HDS1, and HDS2: [NOT(HDS1 XOR HDS2)] OR HCS.

Figure 13−5. HPI32 Read Timing (HAS Not Used, Tied High)

† For correct operation, strobe the HAS signal only once per HSTROBE active cycle.
‡ HSTROBE refers to the following logical operation on HCS, HDS1, and HDS2: [NOT(HDS1 XOR HDS2)] OR HCS.

Figure 13−6. HPI32 Read Timing (HAS Used)

† HSTROBE refers to the following logical operation on HCS, HDS1, and HDS2: [NOT(HDS1 XOR HDS2)] OR HCS.

Figure 13–7. HPI32 Write Timing (HAS Not Used, Tied High)

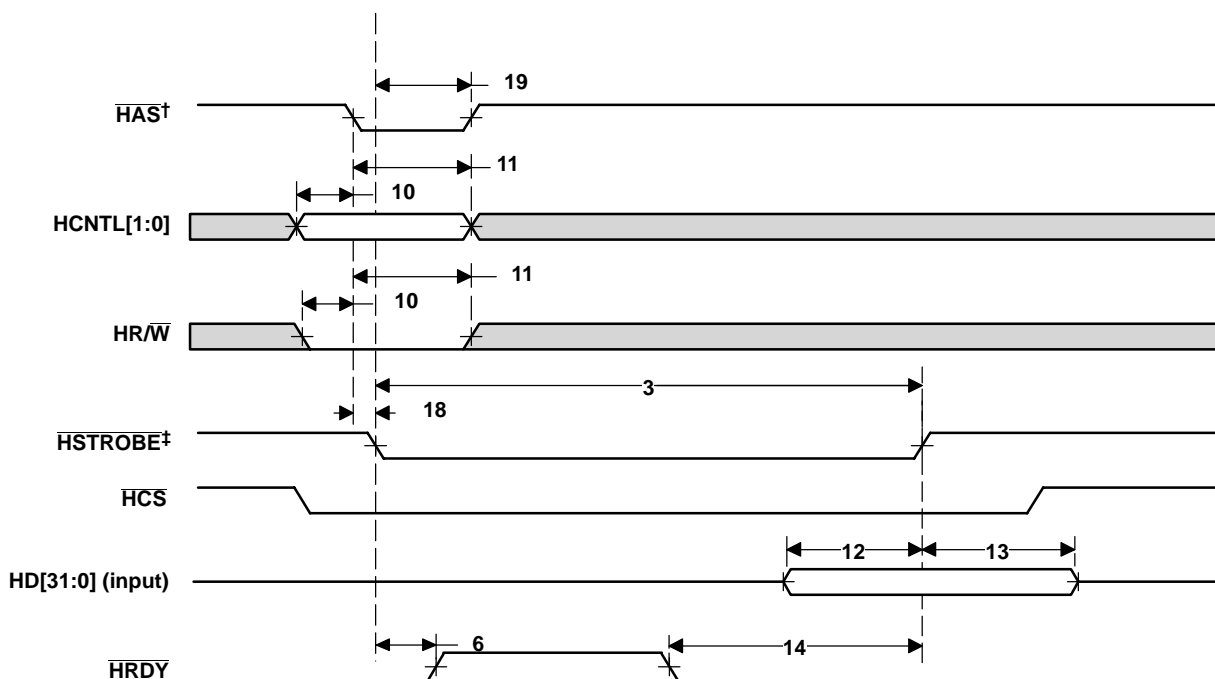† For correct operation, strobe the HAS signal only once per HSTROBE active cycle.
‡ HSTROBE refers to the following logical operation on HCS, HDS1, and HDS2: [NOT(HDS1 XOR HDS2)] OR HCS.

Figure 13–8. HPI32 Write Timing (HAS Used)

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14   Peripheral Component Interconnect (PCI) Timing

Table 14–1. Timing Requirements for PCLK†‡ (see Figure 14–1)

NO.			−500 [33 MHz]		−600, −720 [66 MHz]		UNIT
			MIN	**MAX**	**MIN**	**MAX**	
1	$t_{c(PCLK)}$	Cycle time, PCLK	30 (or 4P§)		15 (or 4P§)		ns
2	$t_{w(PCLKH)}$	Pulse duration, PCLK high	11		6		ns
3	$t_{w(PCLKL)}$	Pulse duration, PCLK low	11		6		ns
4	$t_{sr(PCLK)}$	Δv/Δt slew rate, PCLK	1	4	1.5	4	V/ns

† For 3.3-V operation, the reference points for the rise and fall transitions are measured at V_{ILP} MAX and V_{IHP} MIN.
‡ P = 1/CPU clock frequency in ns. For example when running parts at 600 MHz, use P = 1.67 ns.
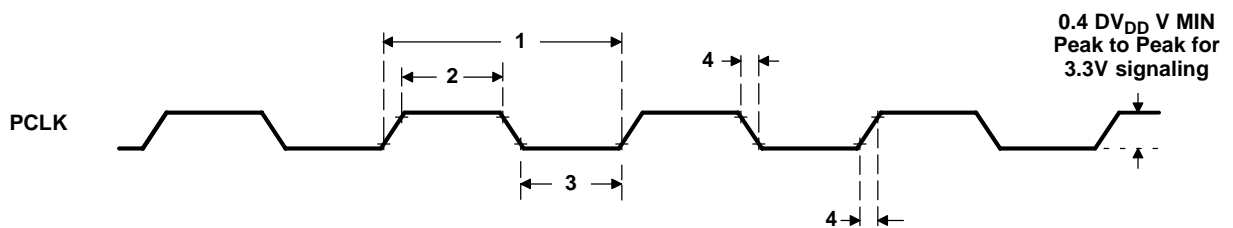§ Select the parameter value, whichever is larger.

Figure 14–1. PCLK Timing

Table 14–2. Timing Requirements for PCI Reset (see Figure 14–2)

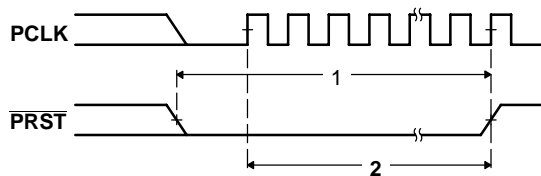| NO.			−500 −600 −720		UNIT
			MIN	**MAX**	
1	$t_{w(PRST)}$	Pulse duration, $\overline{PRST}$	1		ms
2	$t_{su(PCLKA-PRSTH)}$	Setup time, PCLK active before $\overline{PRST}$ high	100		µs

Figure 14–2. PCI Reset ($\overline{PRST}$) Timing

Table 14–3. Timing Requirements for PCI Inputs (see Figure 14–3)

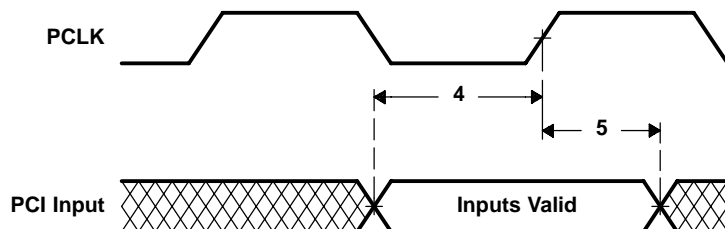| NO.			–500		–600 –720		UNIT
			33 MHz		66 MHz		
			MIN	MAX	MIN	MAX	
4	$t_{su(IV-PCLKH)}$	Setup time, input valid before PCLK high	7		3		ns
5	$t_{h(IV-PCLKH)}$	Hold time, input valid after PCLK high	0		0		ns

Figure 14–3. PCI Input Timing (33-/66-MHz)

Table 14–4. Switching Characteristics Over Recommended Operating Conditions for PCI Outputs (see Figure 14–4)

NO.	PARAMETER		–500		–600 –720		UNIT
			33 MHz		66 MHz		
			MIN	MAX	MIN	MAX	
1	$t_{d(PCLKH-OV)}$	Delay time, PCLK high to output valid	2	11	2	6	ns
2	$t_{d(PCLKH-OLZ)}$	Delay time, PCLK high to output low impedance	2		2		ns
3	$t_{d(PCLKH-OHZ)}$	Delay time, PCLK high to output high impedance		28		14	ns
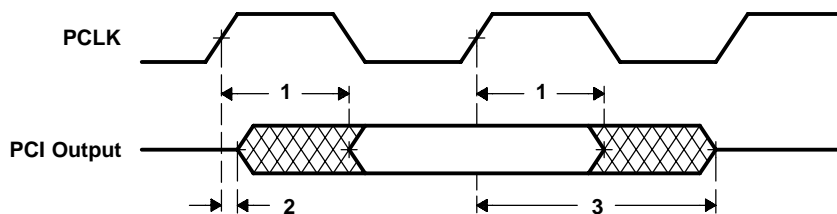
Figure 14–4. PCI Output Timing (33-/66-MHz)

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Table 14–5. Timing Requirements for Serial EEPROM Interface (see Figure 14–5)

NO.			–500 –600 –720		UNIT
			MIN	MAX	
8	$t_{su(DIV-CLKH)}$	Setup time, XSP_DI valid before XSP_CLK high	50		ns
9	$t_{h(CLKH-DIV)}$	Hold time, XSP_DI valid after XSP_CLK high	0		ns

Table 14–6. Switching Characteristics Over Recommended Operating Conditions for Serial EEPROM Interface[†] (see Figure 14–5)

NO.		PARAMETER	–500 –600 –720			UNIT
			MIN	TYP	MAX	
1	$t_{w(CSL)}$	Pulse duration, XSP_CS low		4092P		ns
2	$t_{d(CLKL-CSL)}$	Delay time, XSP_CLK low to XSP_CS low		0		ns
3	$t_{d(CSH-CLKH)}$	Delay time, XSP_CS high to XSP_CLK high		2046P		ns
4	$t_{w(CLKH)}$	Pulse duration, XSP_CLK high		2046P		ns
5	$t_{w(CLKL)}$	Pulse duration, XSP_CLK low		2046P		ns
6	$t_{osu(DOV-CLKH)}$	Output setup time, XSP_DO valid before XSP_CLK high		2046P		ns
7	$t_{oh(CLKH-DOV)}$	Output hold time, XSP_DO valid after XSP_CLK high		2046P		ns

[†] P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
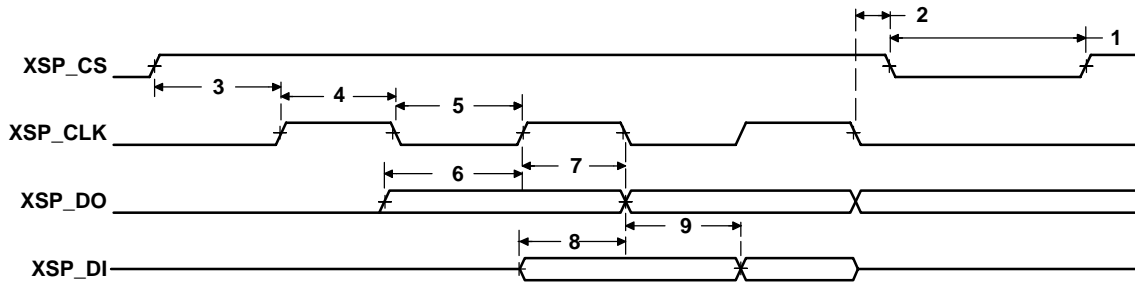
Figure 14–5. PCI Serial EEPROM Interface Timing

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15 Multichannel Buffered Serial Port (McBSP) Timing

Table 15–1. Timing Requirements for McBSP† (see Figure 15–1)

NO.				–500 –600 –720 MIN	MAX	UNIT
2	$t_{c(CKRX)}$	Cycle time, CLKR/X	CLKR/X ext	4P or 6.67‡§		ns
3	$t_{w(CKRX)}$	Pulse duration, CLKR/X high or CLKR/X low	CLKR/X ext	$0.5t_{c(CKRX)} - 1$¶		ns
5	$t_{su(FRH-CKRL)}$	Setup time, external FSR high before CLKR low	CLKR int	9		ns
			CLKR ext	1.3		
6	$t_{h(CKRL-FRH)}$	Hold time, external FSR high after CLKR low	CLKR int	6		ns
			CLKR ext	3		
7	$t_{su(DRV-CKRL)}$	Setup time, DR valid before CLKR low	CLKR int	8		ns
			CLKR ext	0.9		
8	$t_{h(CKRL-DRV)}$	Hold time, DR valid after CLKR low	CLKR int	3		ns
			CLKR ext	3.1		
10	$t_{su(FXH-CKXL)}$	Setup time, external FSX high before CLKX low	CLKX int	9		ns
			CLKX ext	1.3		
11	$t_{h(CKXL-FXH)}$	Hold time, external FSX high after CLKX low	CLKX int	6		ns
			CLKX ext	3		

† CLKRP = CLKXP = FSRP = FSXP = 0. If polarity of any of the signals is inverted, then the timing references of that signal are also inverted.
‡ P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
§ Minimum CLKR/X cycle times *must* be met, even when CLKR/X is generated by an internal clock source. The minimum CLKR/X cycle times are based on internal logic speed; the maximum usable speed may be lower due to EDMA limitations and AC timing requirements.
¶ This parameter applies to the maximum McBSP frequency. Operate serial clocks (CLKR/X) in the reasonable range of 40/60 duty cycle.

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Table 15–2. Switching Characteristics Over Recommended Operating Conditions for McBSP†‡

NO.	PARAMETER		−500 −600 −720		UNIT
			MIN	MAX	
1	$t_{d(CKSH-CKRXH)}$	Delay time, CLKS high to CLKR/X high for internal CLKR/X generated from CLKS input	1.4	10	ns
2	$t_{c(CKRX)}$	Cycle time, CLKR/X	CLKR/X int	4P or 6.67§¶#		ns
3	$t_{w(CKRX)}$	Pulse duration, CLKR/X high or CLKR/X low	CLKR/X int	C − 1‖	C + 1‖	ns
4	$t_{d(CKRH-FRV)}$	Delay time, CLKR high to internal FSR valid	CLKR int	−2.1	3	ns
9	$t_{d(CKXH-FXV)}$	Delay time, CLKX high to internal FSX valid	CLKX int	−1.7	3	ns
			CLKX ext	1.7	9	
12	$t_{dis(CKXH-DXHZ)}$	Disable time, DX high impedance following last data bit from CLKX high	CLKX ext	−3.9	4	ns
				−2.1	9	
13	$t_{d(CKXH-DXV)}$	Delay time, CLKX high to DX valid	CLKX int	−3.9 + D1☆	4 + D2☆	ns
			CLKX ext	−2.1 + D1☆	9 + D2☆	
14	$t_{d(FXH-DXV)}$	Delay time, FSX high to DX valid	FSX int	−2.3 + D1□	5.6 + D2□	ns
		ONLY applies when in data delay 0 (XDATDLY = 00b) mode	FSX ext	1.9 + D1□	9 + D2□	

† CLKRP = CLKXP = FSRP = FSXP = 0. If polarity of any of the signals is inverted, then the timing references of that signal are also inverted.
‡ Minimum delay times also represent minimum output hold times.
§ Use whichever value is greater. Minimum CLKR/X cycle times *must* be met, even when CLKR/X is generated by an internal clock source. The minimum CLKR/X cycle times are based on internal logic speed; the maximum usable speed may be lower due to EDMA limitations and AC timing requirements.
¶ P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
‖ C = H or L
   S = sample rate generator input clock = 4P if CLKSM = 1 (P = 1/CPU clock frequency)
      = sample rate generator input clock = P_clks if CLKSM = 0 (P_clks = CLKS period)
   H = CLKX high pulse width = (CLKGDV/2 + 1) * S if CLKGDV is even
                            = (CLKGDV + 1)/2 * S if CLKGDV is odd or zero
   L = CLKX low pulse width = (CLKGDV/2) * S if CLKGDV is even
                            = (CLKGDV + 1)/2 * S if CLKGDV is odd or zero
   CLKGDV should be set appropriately to ensure the McBSP bit rate does not exceed the maximum limit (see ¶ footnote above).
☆ Extra delay from CLKX high to DX valid **applies *only* to the first data bit of a device**, if and only if DXENA = 1 in SPCR.
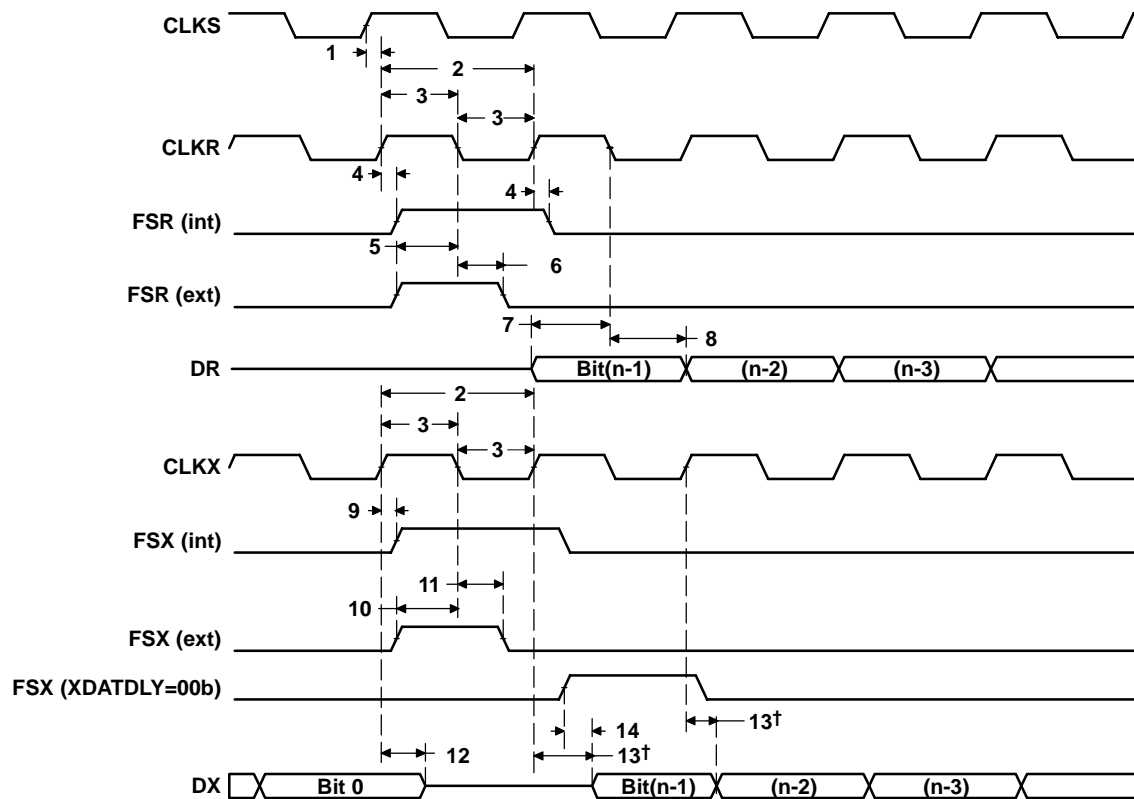   if DXENA = 0, then D1 = D2 = 0
   if DXENA = 1, then D1 = 4P, D2 = 8P
□ Extra delay from FSX high to DX valid **applies *only* to the first data bit of a device**, if and only if DXENA = 1 in SPCR.
   if DXENA = 0, then D1 = D2 = 0
   if DXENA = 1, then D1 = 4P, D2 = 8P

† Parameter No. 13 applies to the first data bit *only* when XDATDLY ≠ 0.

Figure 15–1. McBSP Timing

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Table 15–3. Timing Requirements for FSR When GSYNC = 1 (see Figure 15–2)

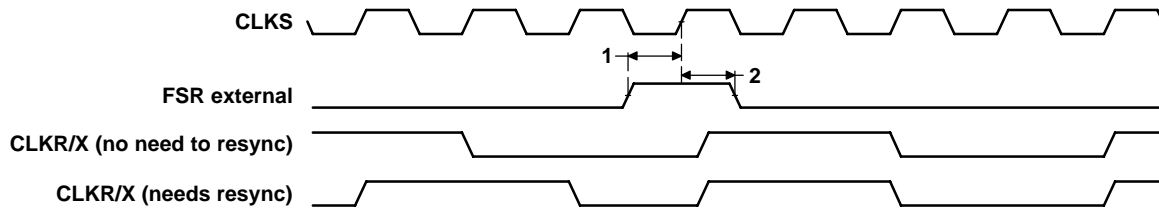| NO.			−500 −600 −720		UNIT
			MIN	MAX	
1	$t_{su(FRH-CKSH)}$	Setup time, FSR high before CLKS high	4		ns
2	$t_{h(CKSH-FRH)}$	Hold time, FSR high after CLKS high	4		ns

Figure 15–2. FSR Timing When GSYNC = 1

Table 15–4. Timing Requirements for McBSP as SPI Master or Slave:
CLKSTP = 10b, CLKXP = 0†‡ (see Figure 15–3)

NO.			−500, −600 −720				UNIT
			MASTER		**SLAVE**		
			MIN	**MAX**	**MIN**	**MAX**	
4	$t_{su(DRV-CKXL)}$	Setup time, DR valid before CLKX low	12		2 – 12P		ns
5	$t_{h(CKXL-DRV)}$	Hold time, DR valid after CLKX low	4		5 + 24P		ns

† P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
‡ For all SPI Slave modes, CLKG is programmed as 1/4 of the CPU clock by setting CLKSM = CLKGDV = 1.

Table 15–5. Switching Characteristics Over Recommended Operating Conditions for McBSP as
SPI Master or Slave: CLKSTP = 10b, CLKXP = 0†‡ (see Figure 15–3)

NO.		PARAMETER	−500, −600 −720				UNIT
			MASTER§		**SLAVE**		
			MIN	**MAX**	**MIN**	**MAX**	
1	$t_{h(CKXL-FXL)}$	Hold time, FSX low after CLKX low¶	T – 2	T + 3			ns
2	$t_{d(FXL-CKXH)}$	Delay time, FSX low to CLKX high#	L – 2.5	L + 3			ns
3	$t_{d(CKXH-DXV)}$	Delay time, CLKX high to DX valid	−2	4	12P + 2.8	20P + 17	ns
6	$t_{dis(CKXL-DXHZ)}$	Disable time, DX high impedance following last data bit from CLKX low	L – 2	L + 3			ns
7	$t_{dis(FXH-DXHZ)}$	Disable time, DX high impedance following last data bit from FSX high			4P + 3	12P + 17	ns
8	$t_{d(FXL-DXV)}$	Delay time, FSX low to DX valid			8P + 1.8	16P + 17	ns

† P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
‡ For all SPI Slave modes, CLKG is programmed as 1/4 of the CPU clock by setting CLKSM = CLKGDV = 1.
§ S = Sample rate generator input clock = 4P if CLKSM = 1 (P = 1/CPU clock frequency)
   = Sample rate generator input clock = P_clks if CLKSM = 0 (P_clks = CLKS period)
 T = CLKX period = (1 + CLKGDV) * S
 H = CLKX high pulse width = (CLKGDV/2 + 1) * S if CLKGDV is even
                            = (CLKGDV + 1)/2 * S if CLKGDV is odd or zero
 L = CLKX low pulse width   = (CLKGDV/2) * S if CLKGDV is even
                            = (CLKGDV + 1)/2 * S if CLKGDV is odd or zero
¶ FSRP = FSXP = 1. As a SPI Master, FSX is inverted to provide active-low slave-enable output. As a Slave, the active-low signal input on FSX and FSR is inverted before being used internally.
  CLKXM = FSXM = 1, CLKRM = FSRM = 0 for Master McBSP
  CLKXM = CLKRM = FSXM = FSRM = 0 for Slave McBSP
# FSX should be low before the rising edge of clock to enable Slave devices and then begin a SPI transfer at the rising edge of the Master clock (CLKX).
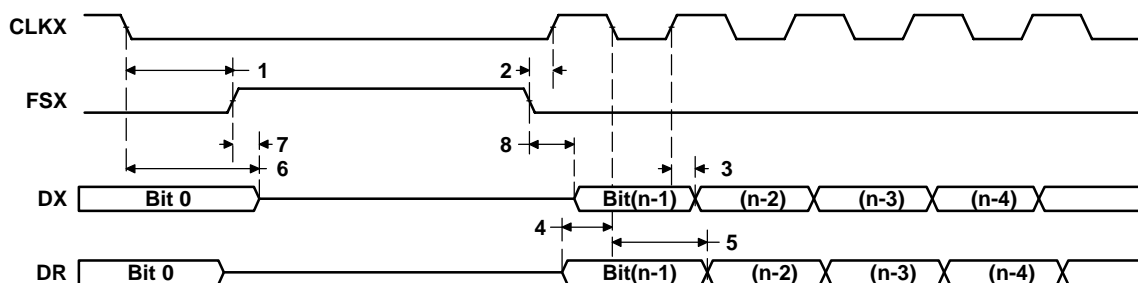

Figure 15–3. McBSP Timing as SPI Master or Slave: CLKSTP = 10b, CLKXP = 0

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Table 15–6. Timing Requirements for McBSP as SPI Master or Slave:
CLKSTP = 11b, CLKXP = 0[†‡] (see Figure 15–4)

NO.			−500 −600 −720				UNIT
			MASTER		SLAVE		
			MIN	MAX	MIN	MAX	
4	$t_{su(DRV-CKXH)}$	Setup time, DR valid before CLKX high	12		2 − 12P		ns
5	$t_{h(CKXH-DRV)}$	Hold time, DR valid after CLKX high	4		5 + 24P		ns

† P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
‡ For all SPI Slave modes, CLKG is programmed as 1/4 of the CPU clock by setting CLKSM = CLKGDV = 1.

Table 15–7. Switching Characteristics Over Recommended Operating Conditions for McBSP as
SPI Master or Slave: CLKSTP = 11b, CLKXP = 0[†‡] (see Figure 15–4)

NO.	PARAMETER		−500 −600 −720				UNIT
			MASTER[§]		SLAVE		
			MIN	MAX	MIN	MAX	
1	$t_{h(CKXL-FXL)}$	Hold time, FSX low after CLKX low[¶]	L − 2	L + 3			ns
2	$t_{d(FXL-CKXH)}$	Delay time, FSX low to CLKX high[#]	T − 2.5	T + 3			ns
3	$t_{d(CKXL-DXV)}$	Delay time, CLKX low to DX valid	−2	4	12P + 3	20P + 17	ns
6	$t_{dis(CKXL-DXHZ)}$	Disable time, DX high impedance following last data bit from CLKX low	−2	4	12P + 3	20P + 17	ns
7	$t_{d(FXL-DXV)}$	Delay time, FSX low to DX valid	H − 2	H + 4	8P + 2	16P + 17	ns

† P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
‡ For all SPI Slave modes, CLKG is programmed as 1/4 of the CPU clock by setting CLKSM = CLKGDV = 1.
§ S = Sample rate generator input clock = 4P if CLKSM = 1 (P = 1/CPU clock frequency)
　 = Sample rate generator input clock = P_clks if CLKSM = 0 (P_clks = CLKS period)
 T = CLKX period = (1 + CLKGDV) * S
 H = CLKX high pulse width = (CLKGDV/2 + 1) * S if CLKGDV is even
　　　　　　　　　　　　 = (CLKGDV + 1)/2 * S if CLKGDV is odd or zero
 L = CLKX low pulse width = (CLKGDV/2) * S if CLKGDV is even
　　　　　　　　　　　　 = (CLKGDV + 1)/2 * S if CLKGDV is odd or zero
¶ FSRP = FSXP = 1. As a SPI Master, FSX is inverted to provide active-low slave-enable output. As a Slave, the active-low signal input on FSX and FSR is inverted before being used internally.
 CLKXM = FSXM = 1, CLKRM = FSRM = 0 for Master McBSP
 CLKXM = CLKRM = FSXM = FSRM = 0 for Slave McBSP
# FSX should be low before the rising edge of clock to enable Slave devices and then begin a SPI transfer at the rising edge of the Master clock (CLKX).
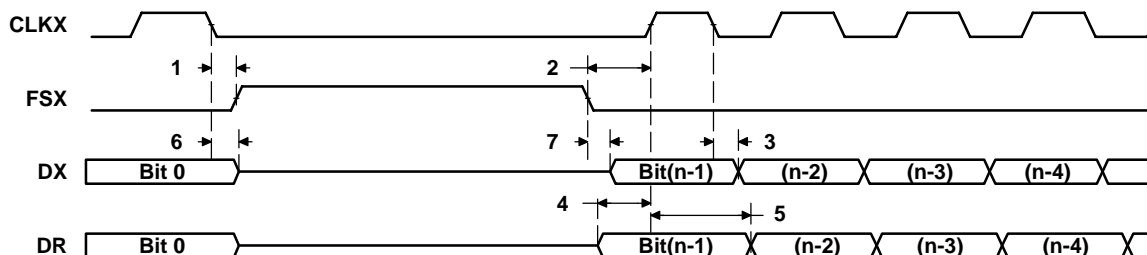

Figure 15–4. McBSP Timing as SPI Master or Slave: CLKSTP = 11b, CLKXP = 0

Table 15–8. Timing Requirements for McBSP as SPI Master or Slave:
CLKSTP = 10b, CLKXP = 1†‡ (see Figure 15–5)

NO.			–500, –600 –720				UNIT
			MASTER		SLAVE		
			MIN	MAX	MIN	MAX	
4	$t_{su(DRV\text{-}CKXH)}$	Setup time, DR valid before CLKX high	12		2 – 12P		ns
5	$t_{h(CKXH\text{-}DRV)}$	Hold time, DR valid after CLKX high	4		5 + 24P		ns

† P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
‡ For all SPI Slave modes, CLKG is programmed as 1/4 of the CPU clock by setting CLKSM = CLKGDV = 1.

Table 15–9. Switching Characteristics Over Recommended Operating Conditions for McBSP as
SPI Master or Slave: CLKSTP = 10b, CLKXP = 1†‡ (see Figure 15–5)

NO.		PARAMETER	–500, –600 –720				UNIT
			MASTER§		SLAVE		
			MIN	MAX	MIN	MAX	
1	$t_{h(CKXH\text{-}FXL)}$	Hold time, FSX low after CLKX high¶	T – 2	T + 3			ns
2	$t_{d(FXL\text{-}CKXL)}$	Delay time, FSX low to CLKX low#	H – 2.5	H + 3			ns
3	$t_{d(CKXL\text{-}DXV)}$	Delay time, CLKX low to DX valid	–2	4	12P + 3	20P + 17	ns
6	$t_{dis(CKXH\text{-}DXHZ)}$	Disable time, DX high impedance following last data bit from CLKX high	H – 2	H + 3			ns
7	$t_{dis(FXH\text{-}DXHZ)}$	Disable time, DX high impedance following last data bit from FSX high			4P + 3	12P + 17	ns
8	$t_{d(FXL\text{-}DXV)}$	Delay time, FSX low to DX valid			8P + 2	16P + 17	ns

† P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
‡ For all SPI Slave modes, CLKG is programmed as 1/4 of the CPU clock by setting CLKSM = CLKGDV = 1.
§ S = Sample rate generator input clock = 4P if CLKSM = 1 (P = 1/CPU clock frequency)
  = Sample rate generator input clock = P_clks if CLKSM = 0 (P_clks = CLKS period)
 T = CLKX period = (1 + CLKGDV) * S
 H = CLKX high pulse width = (CLKGDV/2 + 1) * S if CLKGDV is even
                       = (CLKGDV + 1)/2 * S if CLKGDV is odd or zero
 L = CLKX low pulse width = (CLKGDV/2) * S if CLKGDV is even
                     = (CLKGDV + 1)/2 * S if CLKGDV is odd or zero
¶ FSRP = FSXP = 1. As a SPI Master, FSX is inverted to provide active-low slave-enable output. As a Slave, the active-low signal input on FSX and FSR is inverted before being used internally.
 CLKXM = FSXM = 1, CLKRM = FSRM = 0 for Master McBSP
 CLKXM = CLKRM = FSXM = FSRM = 0 for Slave McBSP
# FSX should be low before the rising edge of clock to enable Slave devices and then begin a SPI transfer at the rising edge of the Master clock (CLKX).
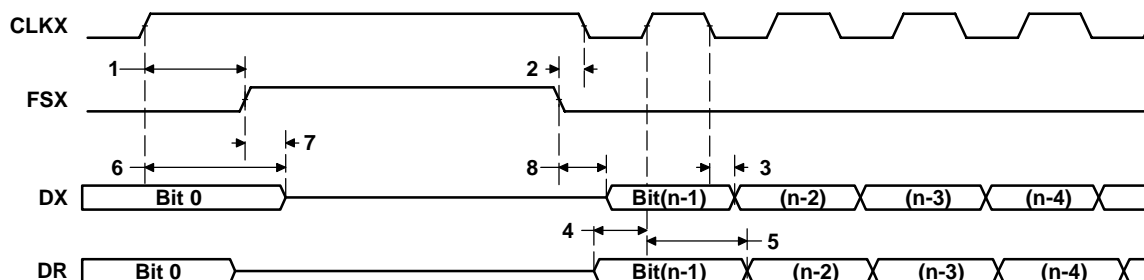

Figure 15–5. McBSP Timing as SPI Master or Slave: CLKSTP = 10b, CLKXP = 1

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Table 15–10. Timing Requirements for McBSP as SPI Master or Slave:
CLKSTP = 11b, CLKXP = 1[†‡] (see Figure 15–6)

NO.			−500 −600 −720				UNIT
			MASTER		**SLAVE**		
			MIN	MAX	MIN	MAX	
4	$t_{su(DRV-CKXH)}$	Setup time, DR valid before CLKX high	12		2 − 12P		ns
5	$t_{h(CKXH-DRV)}$	Hold time, DR valid after CLKX high	4		5 + 24P		ns

[†] P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
[‡] For all SPI Slave modes, CLKG is programmed as 1/4 of the CPU clock by setting CLKSM = CLKGDV = 1.

Table 15–11. Switching Characteristics Over Recommended Operating Conditions for McBSP as
SPI Master or Slave: CLKSTP = 11b, CLKXP = 1[†‡] (see Figure 15–6)

NO.		PARAMETER	−500 −600 −720				UNIT
			MASTER[§]		**SLAVE**		
			MIN	MAX	MIN	MAX	
1	$t_{h(CKXH-FXL)}$	Hold time, FSX low after CLKX high[¶]	H − 2	H + 3			ns
2	$t_{d(FXL-CKXL)}$	Delay time, FSX low to CLKX low[#]	T − 2.5	T + 1.5			ns
3	$t_{d(CKXH-DXV)}$	Delay time, CLKX high to DX valid	−2	4	12P + 3	20P + 17	ns
6	$t_{dis(CKXH-DXHZ)}$	Disable time, DX high impedance following last data bit from CLKX high	−2	4	12P + 3	20P + 17	ns
7	$t_{d(FXL-DXV)}$	Delay time, FSX low to DX valid	L − 2	L + 4	8P + 2	16P + 17	ns

[†] P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
[‡] For all SPI Slave modes, CLKG is programmed as 1/4 of the CPU clock by setting CLKSM = CLKGDV = 1.
[§] S = Sample rate generator input clock = 4P if CLKSM = 1 (P = 1/CPU clock frequency)
    = Sample rate generator input clock = P_clks if CLKSM = 0 (P_clks = CLKS period)
  T = CLKX period = (1 + CLKGDV) * S
  H = CLKX high pulse width = (CLKGDV/2 + 1) * S if CLKGDV is even
                             = (CLKGDV + 1)/2 * S if CLKGDV is odd or zero
  L = CLKX low pulse width = (CLKGDV/2) * S if CLKGDV is even
                             = (CLKGDV + 1)/2 * S if CLKGDV is odd or zero
[¶] FSRP = FSXP = 1. As a SPI Master, FSX is inverted to provide active-low slave-enable output. As a Slave, the active-low signal input on FSX and FSR is inverted before being used internally.
  CLKXM = FSXM = 1, CLKRM = FSRM = 0 for Master McBSP
  CLKXM = CLKRM = FSXM = FSRM = 0 for Slave McBSP
[#] FSX should be low before the rising edge of clock to enable Slave devices and then begin a SPI transfer at the rising edge of the Master clock (CLKX).
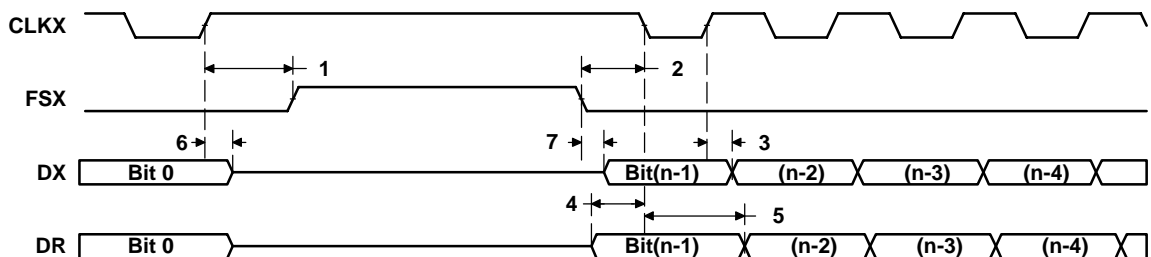

Figure 15–6. McBSP Timing as SPI Master or Slave: CLKSTP = 11b, CLKXP = 1

16 Ethernet Media Access Controller (EMAC) Timing

Table 16–1. Timing Requirements for MRCLK (see Figure 16–1)

NO.			−500 −600 −720		UNIT
			MIN	**MAX**	
1	$t_{c(MRCLK)}$	Cycle time, MRCLK	40		ns
2	$t_{w(MRCLKH)}$	Pulse duration, MRCLK high	14		ns
3	$t_{w(MRCLKL)}$	Pulse duration, MRCLK low	14		ns
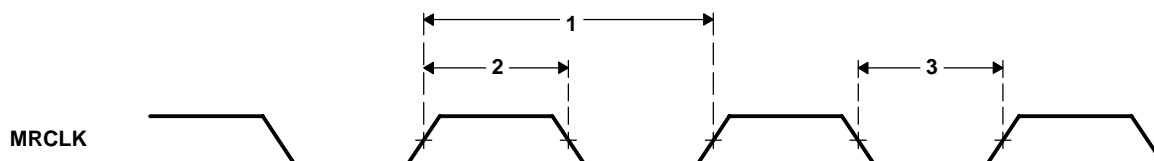
Figure 16–1. MRCLK Timing (EMAC – Receive)

Table 16–2. Timing Requirements for MTCLK (see Figure 16–1)

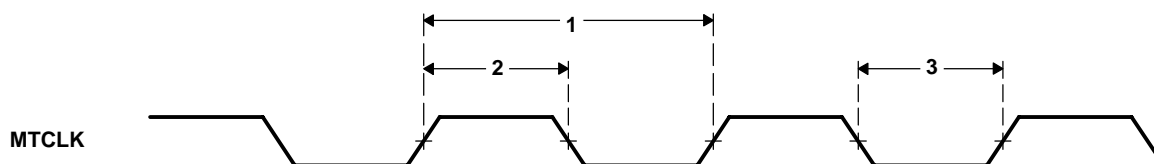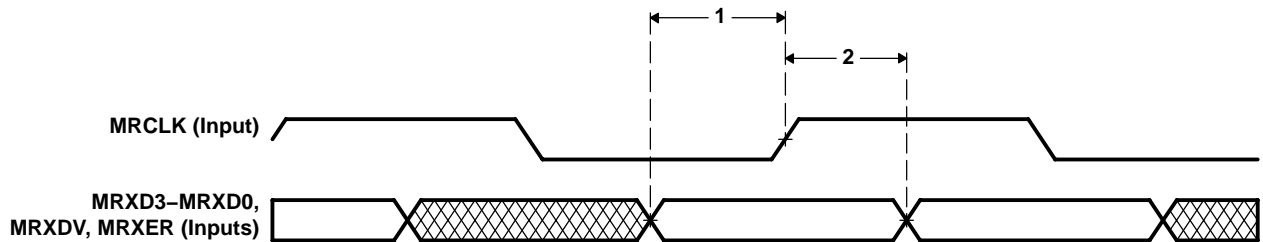| NO.			−500 −600 −720		UNIT
			MIN	**MAX**	
1	$t_{c(MTCLK)}$	Cycle time, MTCLK	40		ns
2	$t_{w(MTCLKH)}$	Pulse duration, MTCLK high	14		ns
3	$t_{w(MTCLKL)}$	Pulse duration, MTCLK low	14		ns

Figure 16–2. MTCLK Timing (EMAC – Transmit)

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Table 16–3. Timing Requirements for EMAC MII Receive 10/100 Mbit/s[†] (see Figure 16–3)
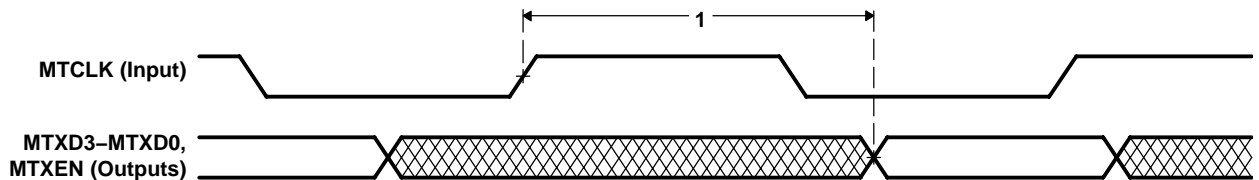
NO.			−500 −600 −720		UNIT
			MIN	MAX	
1	$t_{su(MRXD-MRCLKH)}$	Setup time, receive selected signals valid before MRCLK high	8		ns
2	$t_{h(MRCLKH-MRXD)}$	Hold time, receive selected signals valid after MRCLK high	8		ns

[†] Receive selected signals include: MRXD3–MRXD0, MRXDV, and MRXER.

Figure 16–3. EMAC Receive Interface Timing

Table 16–4. Switching Characteristics Over Recommended Operating Conditions for EMAC MII Transmit 10/100 Mbit/s[‡] (see Figure 16–4)

NO.			−500 −600 −720		UNIT
			MIN	MAX	
1	$t_{d(MTCLKH-MTXD)}$	Delay time, MTCLK high to transmit selected signals valid	5	25	ns

[‡] Transmit selected signals include: MTXD3–MTXD0, and MTXEN.

Figure 16–4. EMAC Transmit Interface Timing

17   Management Data Input/Output (MDIO) Timing

Table 17–1. Timing Requirements for MDIO Input (see Figure 17–1)

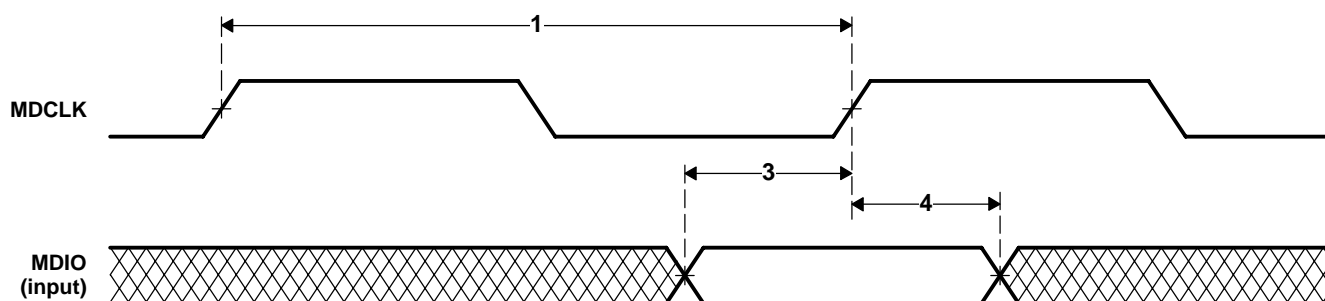| NO.			−500 −600 −720		UNIT
			MIN	MAX	
1	$t_{c(MDCLK)}$	Cycle time, MDCLK	400		ns
2	$t_{w(MDCLK)}$	Pulse duration, MDCLK high/low	180		ns
3	$t_{su(MDIO-MDCLKH)}$	Setup time, MDIO data input valid before MDCLK high	10		ns
4	$t_{h(MDCLKH-MDIO)}$	Hold time, MDIO data input valid after MDCLK high	0		ns

Figure 17–1. MDIO Input Timing

Table 17–2. Switching Characteristics Over Recommended Operating Conditions for MDIO Output (see Figure 17–2)

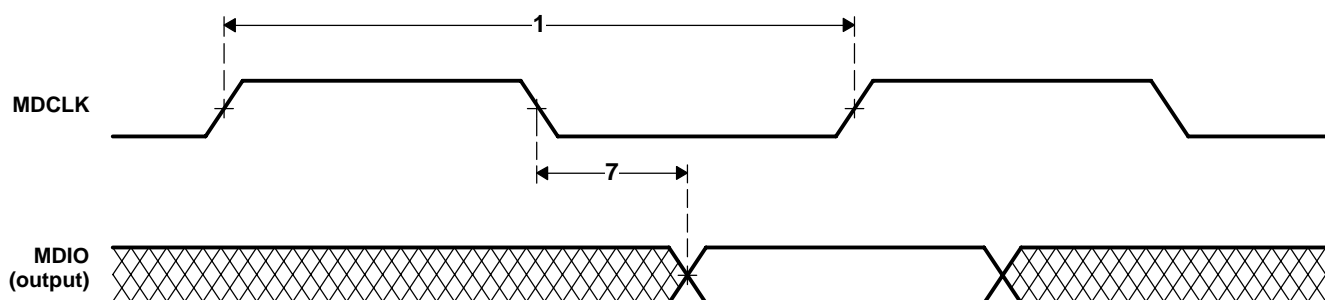| NO.			−500 −600 −720		UNIT
			MIN	MAX	
7	$t_{d(MDCLKL-MDIO)}$	Delay time, MDCLK low to MDIO data output valid	−10	100	ns

Figure 17–2. MDIO Output Timing

18   Timer Timing

Table 18–1. Timing Requirements for Timer Inputs† (see Figure 18–1)

NO.			−500 −600 −720 MIN	−500 −600 −720 MAX	UNIT
1	$t_{w(TINPH)}$	Pulse duration, TINP high	8P		ns
2	$t_{w(TINPL)}$	Pulse duration, TINP low	8P		ns

† P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.

**Table 18–2. Switching Characteristics Over Recommended Operating Conditions for Timer Outputs†
(see Figure 18–1)**

NO.	PARAMETER		−500 −600 −720 MIN	−500 −600 −720 MAX	UNIT
3	$t_{w(TOUTH)}$	Pulse duration, TOUT high	8P − 3		ns
4	$t_{w(TOUTL)}$	Pulse duration, TOUT low	8P − 3		ns

† P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
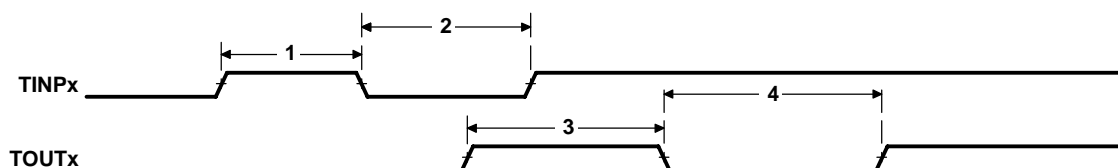
Figure 18–1. Timer Timing

19   General-Purpose Input/Output (GPIO) Port Timing

Table 19–1. Timing Requirements for GPIO Inputs[†][‡] (see Figure 19–1)

NO.			−500 −600 −720		UNIT
			MIN	**MAX**	
1	$t_{w(GPIH)}$	Pulse duration, GPIx high	8P		ns
2	$t_{w(GPIL)}$	Pulse duration, GPIx low	8P		ns

[†] P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
[‡] The pulse width given is sufficient to generate a CPU interrupt or an EDMA event. However, if a user wants to have the DSP recognize the GPIx changes through software polling of the GPIO register, the GPIx duration must be extended to at least 12P to allow the DSP enough time to access the GPIO register through the CFGBUS.

Table 19–2. Switching Characteristics Over Recommended Operating Conditions for GPIO Outputs[†]
(see Figure 19–1)

NO.		PARAMETER	−500 −600 −720		UNIT
			MIN	**MAX**	
3	$t_{w(GPOH)}$	Pulse duration, GPOx high	24P − 8[‡]		ns
4	$t_{w(GPOL)}$	Pulse duration, GPOx low	24P − 8[‡]		ns

[†] P = 1/CPU clock frequency in ns. For example, when running parts at 600 MHz, use P = 1.67 ns.
[‡] This parameter value should not be used as a maximum performance specification. Actual performance of back-to-back accesses of the GPIO is dependent upon internal bus activity.
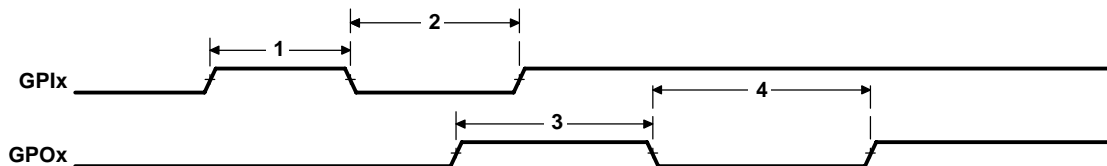

Figure 19–1. GPIO Port Timing

20 JTAG Test-Port Timing

Table 20–1. Timing Requirements for JTAG Test Port (see Figure 20–1)

NO.			−500 −600 −720		UNIT
			MIN	MAX	
1	$t_{c(TCK)}$	Cycle time, TCK	35		ns
3	$t_{su(TDIV\text{-}TCKH)}$	Setup time, TDI/TMS/$\overline{TRST}$ valid before TCK high	10		ns
4	$t_{h(TCKH\text{-}TDIV)}$	Hold time, TDI/TMS/$\overline{TRST}$ valid after TCK high	9		ns

Table 20–2. Switching Characteristics Over Recommended Operating Conditions for JTAG Test Port (see Figure 20–1)

NO.	PARAMETER		−500 −600 −720		UNIT
			MIN	MAX	
2	$t_{d(TCKL\text{-}TDOV)}$	Delay time, TCK low to TDO valid	0	18	ns

Figure 20–1. JTAG Test-Port Timing

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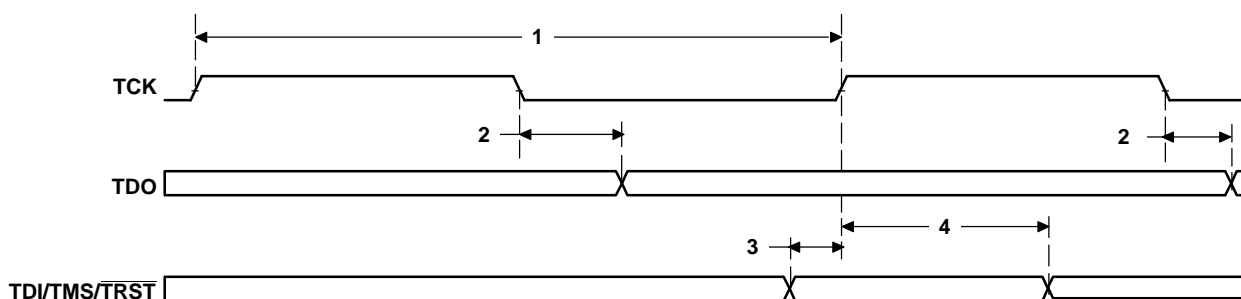
21   Mechanical Data

21.1   Thermal Data

The following tables show the thermal resistance characteristics for the GDK, ZDK, GNZ and ZNZ mechanical packages.

Table 21–1. Thermal Resistance Characteristics (S-PBGA Package) [GDK]

NO			°C/W	Air Flow (m/s[†])
1	RΘ$_{JC}$	Junction-to-case	3.3	N/A
2	RΘ$_{JB}$	Junction-to-board	7.92	N/A
3			18.2	0.00
4	RΘ$_{JA}$	Junction-to-free air	15.3	0.5
5			13.7	1.0
6			12.2	2.00
7			0.37	0.00
8	Psi$_{JT}$	Junction-to-package top	0.47	0.5
9			0.57	1.0
10			0.7	2.00
11			11.4	0.00
12	Psi$_{JB}$	Junction-to-board	11	0.5
13			10.7	1.0
14			10.2	2.00

[†] m/s = meters per second

Table 21–2. Thermal Resistance Characteristics (S-PBGA Package) [ZDK]

NO			°C/W	Air Flow (m/s[†])
1	RΘ$_{JC}$	Junction-to-case	3.3	N/A
2	RΘ$_{JB}$	Junction-to-board	7.92	N/A
3			18.2	0.00
4	RΘ$_{JA}$	Junction-to-free air	15.3	0.5
5			13.7	1.0
6			12.2	2.00
7			0.37	0.00
8	Psi$_{JT}$	Junction-to-package top	0.47	0.5
9			0.57	1.0
10			0.7	2.00
11			11.4	0.00
12	Psi$_{JB}$	Junction-to-board	11	0.5
13			10.7	1.0
14			10.2	2.00

[†] m/s = meters per second

Table 21−3. Thermal Resistance Characteristics (S-PBGA Package) [GNZ]

NO			°C/W	Air Flow (m/s[†])
1	$R\Theta_{JC}$	Junction-to-case	3.3	N/A
2	$R\Theta_{JB}$	Junction-to-board	7.46	N/A
3			17.4	0.00
4	$R\Theta_{JA}$	Junction-to-free air	14.0	0.5
5			12.3	1.0
6			10.8	2.00
7			0.37	0.00
8	Psi_{JT}	Junction-to-package top	0.47	0.5
9			0.57	1.0
10			0.7	2.00
11			11.4	0.00
12	Psi_{JB}	Junction-to-board	11	0.5
13			10.7	1.0
14			10.2	2.00

[†] m/s = meters per second

Table 21−4. Thermal Resistance Characteristics (S-PBGA Package) [ZNZ]

NO			°C/W	Air Flow (m/s[†])
1	$R\Theta_{JC}$	Junction-to-case	3.3	N/A
2	$R\Theta_{JB}$	Junction-to-board	7.46	N/A
3			17.4	0.00
4	$R\Theta_{JA}$	Junction-to-free air	14.0	0.5
5			12.3	1.0
6			10.8	2.00
7			0.37	0.00
8	Psi_{JT}	Junction-to-package top	0.47	0.5
9			0.57	1.0
10			0.7	2.00
11			11.4	0.00
12	Psi_{JB}	Junction-to-board	11	0.5
13			10.7	1.0
14			10.2	2.00

[†] m/s = meters per second

21.2 Packaging Information

The following packaging information reflects the most current released data available for the designated device(s). This data is subject to change without notice and without revision of this document.

Figure 21–1 shows some examples of the types of C6412 package symbolization for –500 MHz and –600 MHz devices. Pin A1 is always located at the top-left corner when you can read the silkscreening/laser-etching and view the TI logo properly.
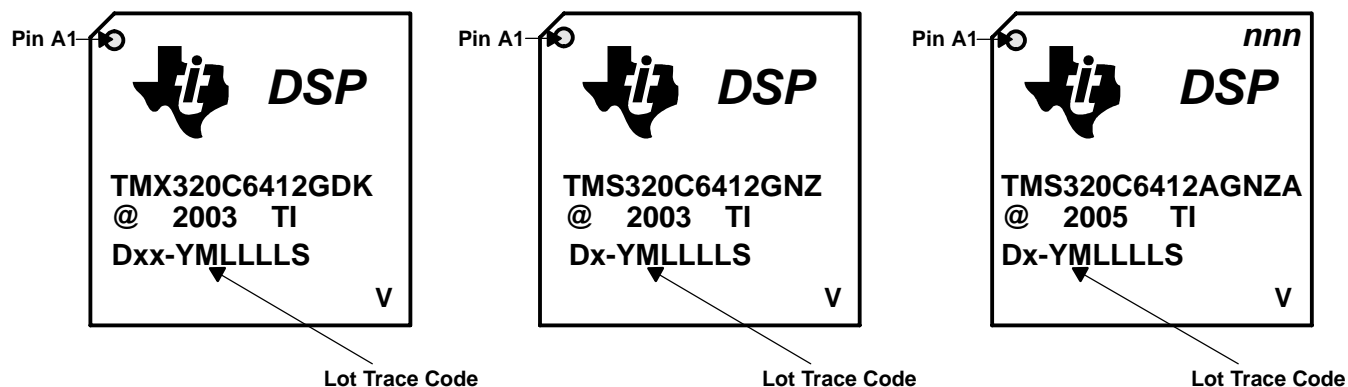

Figure 21–1. Example, Lot Trace Codes for TMX320C6412 and TMS320C6142 (GDK and GNZ Packages)

For more details on package markings, see the *TMS320C6412 Digital Signal Processor Silicon Errata* (SPRZ199).

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package | Pins	Package qty | Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
TMS320C6412AGDKA5	Active	Production	FCCSP (GDK) | 548	60 | JEDEC TRAY (5+1)	No	SNPB	Level-4-220C-72 HR	-40 to 105	TMS320C6412A @ 2003 TI GDK A5
TMS320C6412AGDKA5.A	Active	Production	FCCSP (GDK) | 548	60 | JEDEC TRAY (5+1)	No	SNPB	Level-4-220C-72 HR	-40 to 105	TMS320C6412A @ 2003 TI GDK A5
TMS320C6412AZDKA5	Last Time Buy	Production	FCBGA (ZDK) | 548	60 | JEDEC TRAY (5+1)	ROHS Exempt	SNAGCU	Level-4-260C-72HR	-40 to 105	TMS320C6412 @ 2003 TI AZDK A5
TMS320C6412AZDKA5.A	Last Time Buy	Production	FCBGA (ZDK) | 548	60 | JEDEC TRAY (5+1)	ROHS Exempt	SNAGCU	Level-4-260C-72HR	-40 to 105	TMS320C6412 @ 2003 TI AZDK A5
TMS320C6412AZNZ5	Active	Production	FCBGA (ZNZ) | 548	40 | JEDEC TRAY (5+1)	ROHS Exempt	SNAGCU	Level-4-260C-72HR	0 to 90	TMS320C6412A @ 2003 TI ZNZ 500
TMS320C6412AZNZ5.A	Active	Production	FCBGA (ZNZ) | 548	40 | JEDEC TRAY (5+1)	ROHS Exempt	SNAGCU	Level-4-260C-72HR	0 to 90	TMS320C6412A @ 2003 TI ZNZ 500
TMS320C6412AZNZ6	Active	Production	FCBGA (ZNZ) | 548	40 | JEDEC TRAY (5+1)	ROHS Exempt	SNAGCU	Level-4-260C-72HR	0 to 90	TMS320C6412A @ 2003 TI ZNZ
TMS320C6412AZNZ6.A	Active	Production	FCBGA (ZNZ) | 548	40 | JEDEC TRAY (5+1)	ROHS Exempt	SNAGCU	Level-4-260C-72HR	0 to 90	TMS320C6412A @ 2003 TI ZNZ
TMS320C6412AZNZ7	Active	Production	FCBGA (ZNZ) | 548	40 | JEDEC TRAY (5+1)	ROHS Exempt	SNAGCU	Level-4-260C-72HR	0 to 90	TMS320C6412A @ 2003 TI ZNZ 720

Orderable part number	Status (1)	Material type (2)	Package | Pins	Package qty | Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
TMS320C6412AZNZ7.A	Active	Production	FCBGA (ZNZ) | 548	40 | JEDEC TRAY (5+1)	ROHS Exempt	SNAGCU	Level-4-260C-72HR	0 to 90	TMS320C6412A @ 2003 TI ZNZ 720
TMS320C6412AZNZA5	Last Time Buy	Production	FCBGA (ZNZ) | 548	40 | JEDEC TRAY (5+1)	ROHS Exempt	SNAGCU	Level-4-260C-72HR	-40 to 105	TMS320C6412A @ 2003 TI ZNZ A5
TMS320C6412AZNZA5.A	Last Time Buy	Production	FCBGA (ZNZ) | 548	40 | JEDEC TRAY (5+1)	ROHS Exempt	SNAGCU	Level-4-260C-72HR	-40 to 105	TMS320C6412A @ 2003 TI ZNZ A5

(1) **Status:** For more details on status, see our product life cycle.

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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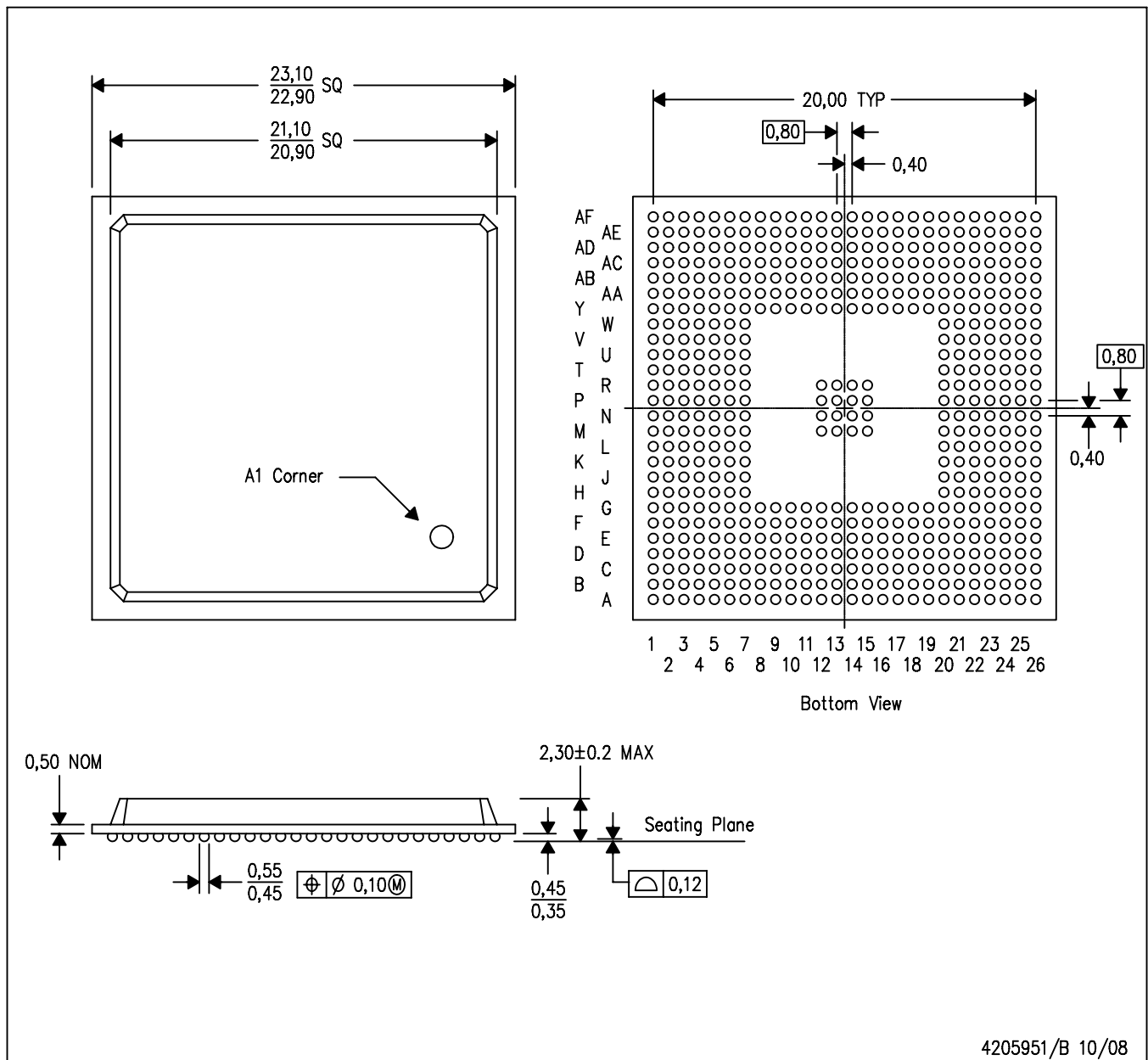
TRAY

L - Outer tray length without tabs

KO - Outer tray height

W - Outer tray width

P1 - Tray unit pocket pitch

CW - Measurement for tray edge (Y direction) to corner pocket center

CL - Measurement for tray edge (X direction) to corner pocket center

Chamfer on Tray corner indicates Pin 1 orientation of packed units.

*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	Unit array matrix	Max temperature (°C)	L (mm)	W (mm)	K0 (µm)	P1 (mm)	CL (mm)	CW (mm)
TMS320C6412AGDKA5	GDK	FCCSP	548	60	5 X 12	150	315	135.9	7620	25.5	17.25	16.95
TMS320C6412AGDKA5.A	GDK	FCCSP	548	60	5 X 12	150	315	135.9	7620	25.5	17.25	16.95
TMS320C6412AZDKA5	ZDK	FCBGA	548	60	5 X 12	150	315	135.9	7620	25.5	17.25	16.95
TMS320C6412AZDKA5.A	ZDK	FCBGA	548	60	5 X 12	150	315	135.9	7620	25.5	17.25	16.95
TMS320C6412AZNZ5	ZNZ	FCBGA	548	40	4x10	150	315	135.9	7620	29.2	26.1	24.15
TMS320C6412AZNZ5.A	ZNZ	FCBGA	548	40	4x10	150	315	135.9	7620	29.2	26.1	24.15
TMS320C6412AZNZ6	ZNZ	FCBGA	548	40	4x10	150	315	135.9	7620	29.2	26.1	24.15
TMS320C6412AZNZ6.A	ZNZ	FCBGA	548	40	4x10	150	315	135.9	7620	29.2	26.1	24.15
TMS320C6412AZNZ7	ZNZ	FCBGA	548	40	4x10	150	315	135.9	7620	29.2	26.1	24.15
TMS320C6412AZNZ7.A	ZNZ	FCBGA	548	40	4x10	150	315	135.9	7620	29.2	26.1	24.15
TMS320C6412AZNZA5	ZNZ	FCBGA	548	40	4x10	150	315	135.9	7620	29.2	26.1	24.15
TMS320C6412AZNZA5.A	ZNZ	FCBGA	548	40	4x10	150	315	135.9	7620	29.2	26.1	24.15

ZDK (S−PBGA−N548)   PLASTIC BALL GRID ARRAY

23,10 / 22,90 SQ

21,10 / 20,90 SQ

20,00 TYP

0,80

0,40

0,80

0,40

A1 Corner

AF
AD
AB
Y
V
T
P
M
K
H
F
D
B

AE
AC
AA
W
U
R
N
L
J
G
E
C
A

1 3 5 7 9 11 13 15 17 19 21 23 25
2 4 6 8 10 12 14 16 18 20 22 24 26

Bottom View

0,50 NOM

0,55 / 0,45   ⊕ Ø 0,10 Ⓜ

2,30±0.2 MAX

Seating Plane

0,45 / 0,35   △ 0,12

4205951/B 10/08

NOTES: A.   All linear dimensions are in millimeters.
           B.   This drawing is subject to change without notice.
           C.   Flip chip application only.
           D.   This package is lead−free.

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ZNZ (S–PBGA–N548)

PLASTIC BALL GRID ARRAY

$\dfrac{27,20}{26,80}$ SQ

$\dfrac{25,20}{24,80}$ SQ

25,00 TYP

1,00

0,50

1,00

0,50

A1 Corner

AF
AE
AD
AC
AB
AA
Y
W
V
U
T
R
P
N
M
L
K
J
H
G
F
E
D
C
B
A

1 3 5 7 9 11 13 15 17 19 21 23 25
 2 4 6 8 10 12 14 16 18 20 22 24 26

Bottom View

0,50 NOM

2,80 MAX

Seating Plane

$\dfrac{0,70}{0,50}$

⊕ Ø 0,10 Ⓜ

$\dfrac{0,60}{0,40}$

0,15

4206182–5/A 06/04
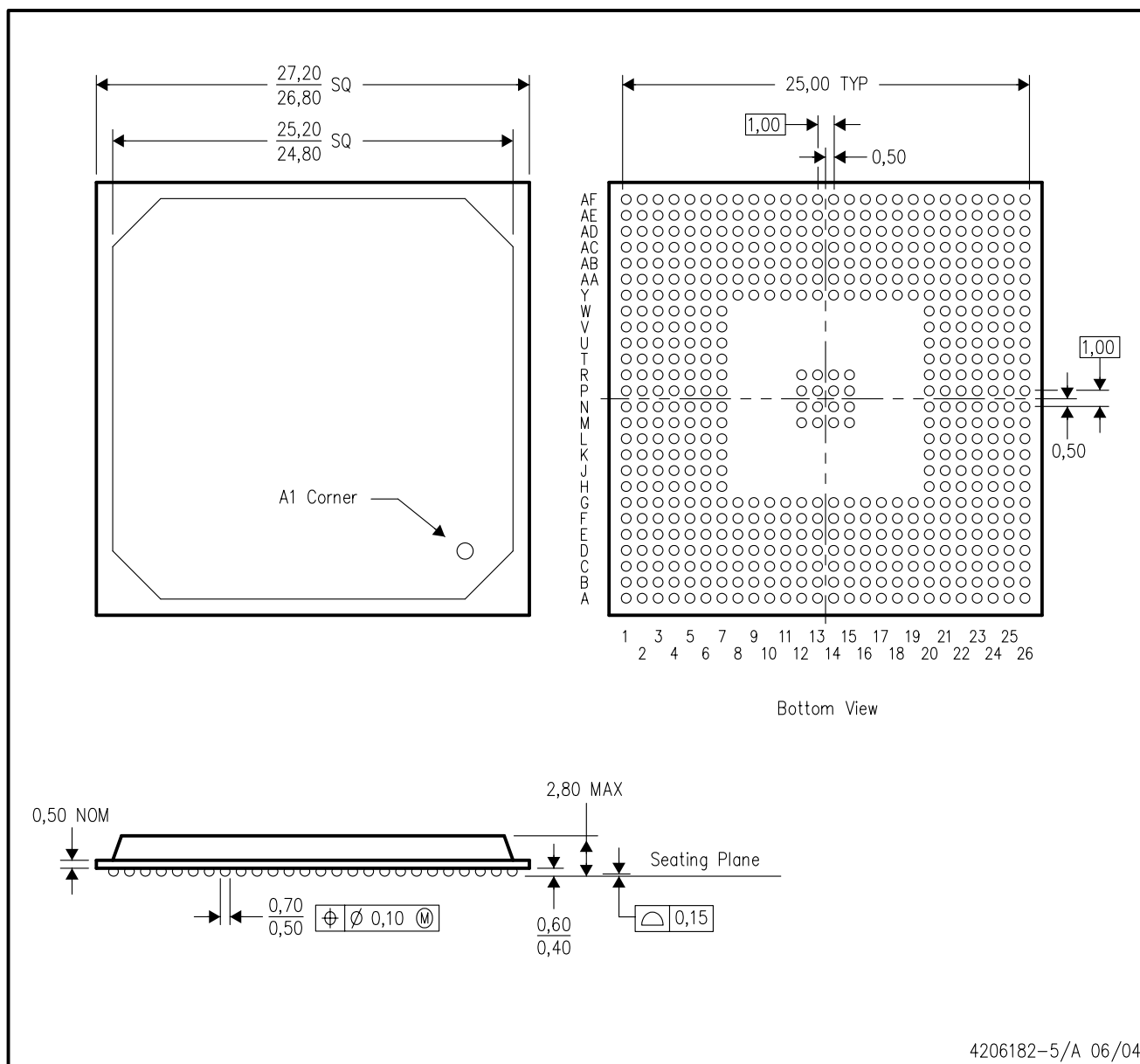
NOTES: A.   All linear dimensions are in millimeters.
         B.   This drawing is subject to change without notice.
         C.   Flip chip application only.
         D.   Substrate color may vary.
         E.   This package is lead–free.

GDK (S–PBGA–N548)   **PLASTIC BALL GRID ARRAY**
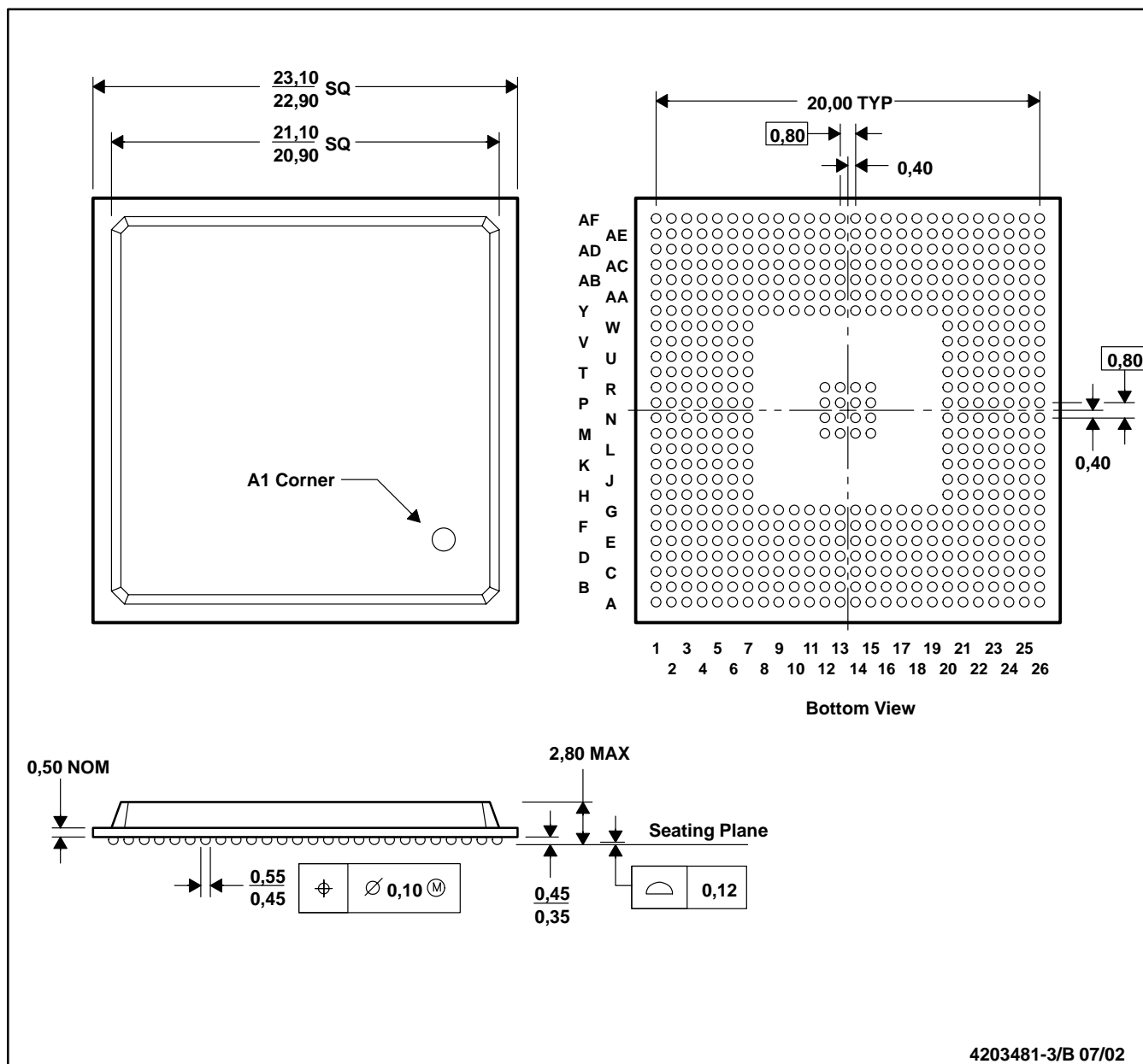
23,10 / 22,90 SQ

21,10 / 20,90 SQ

A1 Corner

20,00 TYP

0,80

0,40

0,80

0,40

AF
AD
AB
Y
V
T
P
M
K
H
F
D
B

AE
AC
AA
W
U
R
N
L
J
G
E
C
A

1   3   5   7   9   11 13 15 17 19 21 23 25
 2   4   6   8   10 12 14 16 18 20 22 24 26

Bottom View

0,50 NOM

2,80 MAX

Seating Plane

0,55 / 0,45

⊕ ⌀ 0,10 Ⓜ

0,45 / 0,35

⌒ 0,12

4203481-3/B 07/02

NOTES: A. All linear dimensions are in millimeters.
        B. This drawing is subject to change without notice.
        C. Flip chip application only.

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