

# TPD4S480 USB Type-C® 48V EPR Port Protector: Short-to-VBUS Overvoltage and IEC ESD Protection

## 1 Features

- 4-channels of short-to- $V_{BUS}$  overvoltage protection (CC1, CC2, SBU1, SBU2):  $63V_{DC}$  tolerant
- 4-channels of IEC 61000-4-2 ESD protection (CC1, CC2, SBU1, SBU2)
- CC1 and CC2 overvoltage protection FETs for passing  $V_{CONN}$  power
- $\pm 65V$  surge protection on CC pins
- $+65V/-35V$  surge protection on SBU pins
- Integrated VBUS divider circuit with enable for dividing down EPR level  $V_{BUS}$
- Integrated FET driver for control of external EPR blocking FET
- CC dead battery resistors integrated for handling dead battery use case in mobile devices
- $3mm \times 3mm$  QFN package

## 2 Applications

- Desktop PC/motherboard
- Standard notebook PC
- Chromebook and WOA
- Docking station
- Port/cable adapters and dongles
- Smartphones

## 3 Description

The TPD4S480 is a single-chip USB Type-C port protection device that provides 48V Short-to- $V_{BUS}$  overvoltage and IEC ESD protection.

Since the release of the USB Type-C connector, many products and accessories for USB Type-C have been released that do not meet the USB Type-C specification. One example of this is USB Type-C Power Delivery adapters that only place high voltage on the  $V_{BUS}$  line. Another concern for USB Type-C is that mechanical twisting and sliding of the connector could short pins due to the close proximity they have in this small connector. This can cause 48V  $V_{BUS}$  to be shorted to the CC and SBU pins. Also due to the proximity of the pins in the Type-C connector, there is a heightened concern that debris and moisture will cause the 48V  $V_{BUS}$  pin to be shorted to the CC and SBU pins.

These non-ideal equipment and mechanical events make it necessary for the CC and SBU pins to be 48V tolerant, even though the pins only operate at 5V or lower. The TPD4S480 enables the CC and SBU pins to be 48V tolerant without interfering with normal operation by providing over-voltage protection on the CC and SBU pins. The device places high voltage FETs in series on the SBU and CC lines. When a voltage above the OVP threshold is detected on these lines, the high voltage switches are opened up, isolating the rest of the system from the high voltage condition present on the connector.

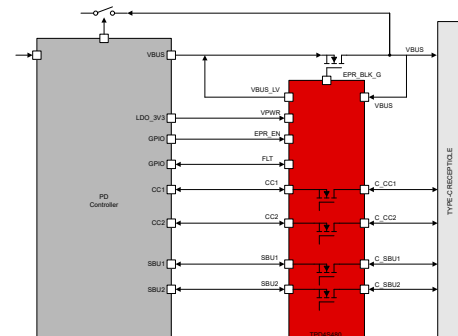
The integrated VBUS divider circuit and FET driver, allow PD controllers not rated for EPR operation to safely operate in EPR voltage ranges. When enabled by GPIO or automatically when  $V_{BUS}$  exceeds 24V, the TPD4S480 disables the optional external blocking FET and enables the voltage divider. This action protects 20V rated PD controllers and allows the use of existing  $V_{BUS}$  sense circuits.

Finally, most systems require IEC 61000-4-2 system level ESD protection for external pins. The TPD4S480 integrates IEC 61000-4-2 ESD protection for the CC1, CC2, SBU1, and SBU2 pins, eliminating the need to place high voltage TVS diodes externally on the connector.

### Package Information (1)

PART NUMBER	PACKAGE	BODY SIZE (NOM)
TPD4S480	RUK (WQFN, 20)	3mm $\times$ 3mm

- (1) For all available packages, see the orderable addendum at the end of the data sheet.

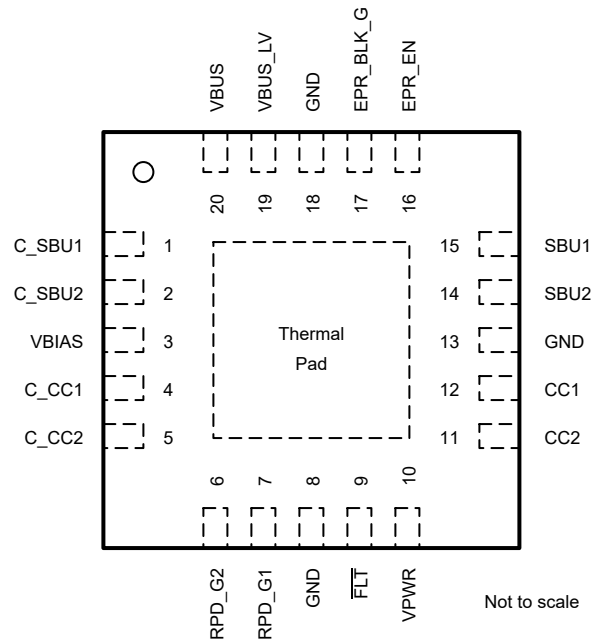


CC and SBU Overvoltage Protection

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## 4 Pin Configuration and Functions



**Figure 4-1. RUK Package 20-Pin QFN**

**Table 4-1. Pin Functions**

PIN		TYPE <sup>(1)</sup>	DESCRIPTION
NO.	NAME		
12	CC1	I/O	System side of the CC1 OVP FET. Connect to either CC pin of the CC/PD controller.
11	CC2	I/O	System side of the CC2 OVP FET. Connect to either CC pin of the CC/PD controller.
4	C_CC1	I/O	Connector side of the CC1 OVP FET. Connect to either CC pin of the USB Type-C connector.
5	C_CC2	I/O	Connector side of the CC2 OVP FET. Connect to either CC pin of the USB Type-C connector.
1	C_SBU1	I/O	Connector side of the SBU1 OVP FET. Connect to either SBU pin of the USB Type-C connector. Alternatively, connect to either USB2.0 pin of the USB Type-C connector to protect the USB2.0 pins instead of the SBU pins.
2	C_SBU2	I/O	Connector side of the SBU2 OVP FET. Connect to either SBU pin of the USB Type-C connector. Alternatively, connect to either USB2.0 pin of the USB Type-C connector to protect the USB2.0 pins instead of the SBU pins.
15	SBU1	I/O	System side of the SBU1 OVP FET. Connect to either SBU pin of the SBU MUX. Alternatively, connect to either USB2.0 pin of the USB2.0 Phy when protecting the USB2.0 pins instead of protecting the SBU pins.
14	SBU2	I/O	System side of the SBU2 OVP FET. Connect to either SBU pin of the SBU MUX. Alternatively, connect to either USB2.0 pin of the USB2.0 Phy when protecting the USB2.0 pins instead of protecting the SBU pins.
7	RPD_G1	I/O	Short to C_CC1 if dead battery resistors are needed. If dead battery resistors are not needed, short pin to GND.
6	RPD_G2	I/O	Short to C_CC2 if dead battery resistors are needed. If dead battery resistors are not needed, short pin to GND.
9	FLT	O	Open drain for fault reporting.
8, 13, 18	GND	GND	Ground

**Table 4-1. Pin Functions (continued)**

PIN		TYPE <sup>(1)</sup>	DESCRIPTION
NO.	NAME		
3	VBIAS	P	Pin for ESD support capacitor. Place a 0.1-μF capacitor on this pin to ground.
10	VPWR	P	2.7V to 4.5V power supply.
20	VBUS	I	Input for EPR VBUS divider. Tie to USB-C receptacle VBUS pins.
19	VBUS_LV	O	Output of EPR VBUS divider. When EPR_EN is asserted, VBUS_LV is divided down voltage from VBUS. When EPR_EN is de-asserted VBUS_LV is equal to VBUS.
16	EPR_EN	I	EPR mode enable input. When asserted EPR_BLK_G is disabled and VBUS_LV is divided VBUS.
17	EPR_BLK_G	O	Gate driver output to optional VBUS blocking FET. FET is enabled when in SPR mode and disabled in EPR mode.
-	Thermal Pad	GND	Internally connected to GND. Used as a heatsink. Connect to the PCB GND plane

(1) I = input, O = output, I/O = input and output, GND = ground, P = power

## 5 Specifications

### 5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

			MIN	MAX	UNIT
V <sub>I</sub>	Input voltage	EPR_EN	-0.3	6.0	V
V <sub>I</sub>	Input voltage	VPWR	-0.3	5	V
		RPD_G1, RPD_G2, VBUS	-0.3	63	V
V <sub>I</sub>	Input voltage	VBUS (VPWR = 0V)	-0.3	24	V
V <sub>I</sub>	Input voltage	VBUS (VPWR > 2.7V)	-0.3	63	V
V <sub>O</sub>	Output voltage	FLT	-0.3	6	V
		VBIAS	-0.3	63	V
V <sub>O</sub>	Output voltage	VBUS_LV	-0.3	24	V
V <sub>O</sub>	Output voltage	EPR_BLK_G	-0.3	30	V
V <sub>IO</sub>	I/O voltage	CC1, CC2, SBU1, SBU2	-0.3	6	V
		C_CC1, C_CC2, C_SBU1, C_SBU2	-0.3	63	V
t <sub>rise</sub>	Input voltage rise time (V <sub>I</sub> > 36V)	CC1, CC2, SBU1, SBU2	400		ns
T <sub>J</sub>	Operating junction temperature		-40	125	°C
T <sub>stg</sub>	Storage temperature		-65	150	°C

- (1) Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If used outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.

### 5.2 ESD Ratings—IEC Specification

			VALUE	UNIT
V <sub>(ESD)</sub>	Electrostatic discharge	IEC 61000-4-2, C_CC1, C_CC2	±8000	V
V <sub>(ESD)</sub>	Electrostatic discharge	IEC 61000-4-2, C_CC1, C_CC2	±15000	V
V <sub>(ESD)</sub>	Electrostatic discharge	IEC 61000-4-2, C_SBU1, C_SBU2	±8000	V
V <sub>(ESD)</sub>	Electrostatic discharge	IEC 61000-4-2, C_SBU1, C_SBU2	±15000	V
V <sub>(Surge)</sub>	Lightning and Surge	IEC 61000-4-5, C_CC1, C_CC2	±65	V
		IEC 61000-4-5, C_SBU1, C_SBU2	+65/-35	V

### 5.3 Recommended Operating Conditions

			MIN	NOM	MAX	UNIT
V <sub>I</sub>	Input voltage	VPWR	2.7	3.3	4.5	V
		RPD_G1, RPD_G2	0		5.5	V
		EPR_EN	0		VPWR	V
		VBUS	0		51	V
V <sub>O</sub>	Output voltage	FLT Pull-up resistor power rail	2.7		5.5	V
V <sub>IO</sub>	I/O voltage	CC1, CC2, C_CC1, C_CC2	0		5.5	V
		SBU1, SBU2, C_SBU1, C_SBU2	0		4.3	V
I <sub>VCONN</sub>	V <sub>CONN</sub> Current	Current flowing into CC1/2 and flowing out of C_CC1/2, T <sub>J</sub> ≤ 105 °C			600	mA
I <sub>VCONN</sub>	V <sub>CONN</sub> Current	Current flowing into CC1/2 and flowing out of C_CC1/2, T <sub>J</sub> ≤ 85 °C			1.25	A
T <sub>J</sub>	Operating Junction Temperature		-40		125	°C

		MIN	NOM	MAX	UNIT	
External Components <sup>(1)</sup>	External Components <sup>(1)</sup>	FLT Pull-up resistance	1.7		300	kΩ
		VBIAS capacitance <sup>(2)</sup>	0.04	0.1	0.14	μF
		VPWR Capacitance	0.3	1		μF
		VBUS_LV Capacitance		0.1		μF

- (1) For recommended values for capacitors and resistors, the typical values assume a component placed on the board near the pin. Minimum and maximum values listed are inclusive of manufacturing tolerances, voltage derating, board capacitance, and temperature variation. The effective value presented should be within the minimum and maximums listed in the table.
- (2) The VBIAS pin requires a minimum 63-VDC rated capacitor. A 100-VDC rated capacitor is recommended to reduce capacitance derating. See the VBIAS Capacitor Selection section for more information on selecting the VBIAS capacitor.

### 5.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		TPD4S480	UNIT
		QFN	
		20 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	44.7	°C/W
R <sub>θJC(top)</sub>	Junction-to-case (top) thermal resistance	41.8	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	17.7	°C/W
ψ <sub>JT</sub>	Junction-to-top characterization parameter	0.7	°C/W
ψ <sub>JB</sub>	Junction-to-board characterization parameter	17.7	°C/W
R <sub>θJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	7.9	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

### 5.5 Electrical Characteristics

over operating junction temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT		
<b>CC OVP Switches</b>							
R <sub>ON</sub>	On Resistance of CC OVP FETs	CCx = 5.5V, T <sub>J</sub> ≤ 85 °C		272	420	mΩ	
C <sub>ON_CC</sub>	Equivalent on Capacitance	40	74	120		pF	
RD_DB	Dead Battery Pull-Down Resistors (only present when device is unpowered)	V <sub>C_CCx</sub> = 2.6V		4.1	5.1	6.1	kΩ
VTH_DB	Threshold voltage of the pull-down FET in series with RD during dead battery	I <sub>C_CCx</sub> = 80μA		0.5	0.9	1.2	V
V <sub>OVPCC</sub>	OVP Threshold on CC Pins	Place 5.5V on C <sub>_CCx</sub> . Step up C <sub>_CCx</sub> until FLT pin is asserted. Put 100mA load through the CC FET and see the FET shuts off.		5.6	5.9	6.2	V
V <sub>OVPCC_HYS</sub>	Hysteresis on CC OVP	Place 6.5 V on C <sub>_CCx</sub> . Step down the voltage on C <sub>_CCx</sub> until the FLT pin is deasserted. Measure difference between rising and falling OVP threshold for C <sub>_CCx</sub> .		50		mV	
BW <sub>ON</sub>	On Bandwidth Single Ended (-3dB)	Measure the -3 dB bandwidth from C <sub>_CCx</sub> to CCx. Single ended measurement, 50-Ω system. V <sub>cm</sub> = 0.1V to 1.2 V.		125		MHz	
V <sub>STBUS_CC</sub>	Short-to-VBUS tolerance on the CC pins	Hot-Plug C <sub>_CCx</sub> with a 1 meter USB Type C Cable, place a 30-Ω load on CCx		51		V	
V <sub>STBUS_CC_CLAMP</sub>	Short-to-VBUS System-Side Clamping Voltage on the CC pins (CCx)	Hot-Plug C <sub>_CCx</sub> with a 1 meter USB Type C Cable. Hot-Plug voltage C <sub>_CCx</sub> = 51 V. VPWR = 3.3 V. Place a 30-Ω load on CCx.		7		V	
<b>SBU OVP Switches</b>							
R <sub>ON</sub>	On Resistance of SBU OVP FETs	SBUx = 3.6 V. -40°C ≤ T <sub>J</sub> ≤ +85°C		4	6.8	Ω	

## 5.5 Electrical Characteristics (continued)

over operating junction temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
C <sub>ON_SBU</sub>	Equivalent on Capacitance	Capacitance from SBUx or C_SBUx to GND when device is powered. Measure at V <sub>C_SBUx</sub> /V <sub>SBUx</sub> = 0.3V to 4.0V.		6		pF
V <sub>OVP</sub> SBU	OVP Threshold on SBU Pins	Place 3.6V on C_SBUx. Step up C_SBUx until FLT pin is asserted.	4.0	4.2	4.41	V
V <sub>OVP</sub> SBU_HYS	Hysteresis on SBU OVP	Place 5 V on C_CCx. Step down the voltage on C_CCx until the FLT pin is deasserted. Measure difference between rising and falling OVP threshold for C_SBUx.		50		mV
BW <sub>ON</sub>	On Bandwidth Single Ended (-3dB)	Measure the -3 dB bandwidth from C_SBUx to SBUx. Single ended measurement, 50Ω system. V <sub>cm</sub> = 0.1V to 3.6V.	600	760		MHz
X <sub>TALK</sub>	Crosstalk	Measure crosstalk at f = 1 MHz from SBU1 to C_SBU2 or SBU2 to C_SBU1. V <sub>cm1</sub> = 3.6 V, V <sub>cm2</sub> = 0.3 V. Terminate open sides to 50Ω.		-70		dB
V <sub>STBUS_SBU</sub>	Short-to-VBUS tolerance on the SBU pins	Hot-Plug C_SBUx with a 1 meter USB Type C Cable. Put a 100-nF capacitor in series with a 40-Ω resistor to GND on SBUx.			51	V
V <sub>STBUS_SBU_CLAMP</sub>	Short-to-VBUS System-Side Clamping Voltage on the SBU pins (SBUx)	Hot-Plug C_SBUx with a 1 meter USB Type C Cable. Hot-Plug voltage C_SBUx = 51V. VPWR = 3.3 V. Put a 150-nF capacitor in series with a 40-Ω resistor to GND on SBUx.		7		V
<b>EPR Adapter</b>						
VBUS_DIV_SPR	VBUS_LV to VBUS divider ratio, SPR Mode	VBUS_LV/VBUS, EPR_EN = 0, VBUS = 4.5 - 21V, I_VBUS_LV = 0-20mA		1		V/V
VBUS_DIV_EPR	VBUS_LV to VBUS divider ratio, EPR Mode	VBUS_LV/VBUS, EPR_EN = 1, VBUS = 26.6-50.4, I_VBUS_LV = 0-20mA		0.42		V/V
I <sub>VBUSLV</sub>	Current from VBUS_LV				20	mA
V <sub>FWD_VBUSLV</sub>	VBUS to VBUS_LV forward voltage drop	I_VBUS_LV=20mA, VBUS=4.5V, EPR_EN=0		456	700	mV
V <sub>FWD_VBUSLV</sub>	VBUS to VBUS_LV forward voltage drop	I_VBUS_LV=20mA, VBUS=26V, EPR_EN=1		823	940	mV
V <sub>EPR_THRESH_R</sub>	Rising VBUS EPR enable threshold		22.7		24	V
V <sub>EPR_THRESH_F</sub>	Falling VBUS EPR enable threshold		22.4		23.4	V
V <sub>EPR_BLK_G</sub>	Gate Drive voltage for EPR_BLK_G	0 ≤ VBUS ≤ 22 V	5		12	V
I <sub>EPR_BLK_G</sub>	Gate Driver Sourcing Current	0 ≤ V <sub>EPR_BLK_G</sub> -V <sub>VBUS</sub> ≤ 5 V, 0 V ≤ V <sub>VBUS</sub> ≤ 22 V, measure I <sub>EPR_BLK_G</sub>		4		μA
V <sub>EPR_EN_V+</sub>	EPR_EN rising threshold			0.7*VPWR		V
V <sub>EPR_EN_V-</sub>	EPR_EN falling threshold		0.3*VPWR			V
<b>Power Supply and Leakage Currents</b>						
V <sub>PWR_UVLO</sub>	V <sub>PWR</sub> Under Voltage Lockout	Place 1 V on VPWR and raise voltage until SBU or CC FETs turn-on.	2.1	2.3	2.6	V
V <sub>PWR_UVLO_HYS</sub>	V <sub>PWR</sub> UVLO Hysteresis	Place 3 V on VPWR and lower voltage until SBU or CC FETs turnoff; measure difference between rising and falling UVLO to calculate hysteresis.	70	100	130	mV
I <sub>VPWR</sub>	V <sub>PWR</sub> supply current	VPWR = 3.3 V (typical), VPWR = 4.5 V (maximum). -40°C ≤ T <sub>J</sub> ≤ +85°C.		112	160	μA
I <sub>C_CC_LEAK</sub>	Leakage current for C_CCx pins when device is powered	VPWR = 3.3 V, V <sub>C_CCx</sub> = 3.6 V, CCx pins are floating, measure leakage current into C_CCx pins.			5	μA

## 5.5 Electrical Characteristics (continued)

over operating junction temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$I_{C\_SBU\_LEAK}$	Leakage current for C_SBUx pins when device is powered	VPWR = 3.3 V, $V_{C\_SBUx} = 3.6$ V, SBUx pins are floating, measure leakage current into C_SBUx pins. Result should be same if SBUx side is biased and C_SBUx is left floating. $-40^{\circ}\text{C} \leq T_J \leq +85^{\circ}\text{C}$			3.2	$\mu\text{A}$
$I_{C\_CC\_LEAK\_OVP}$	Leakage current for C_CCx pins when device is in OVP	VPWR = 0 V or 3.3 V, $V_{C\_CCx} = 51$ V, CCx pins are set to 0 V, measure leakage current into C_CCx pins.			1200	$\mu\text{A}$
$I_{C\_SBU\_LEAK\_OVP}$	Leakage current for C_SBUx pins when device is in OVP	VPWR = 0 V or 3.3 V, $V_{C\_SBUx} = 51$ V, SBUx pins are set to 0 V, measure leakage current into C_SBUx pins.			720	$\mu\text{A}$
$I_{CC\_LEAK\_OVP}$	Leakage current for CC pins when device is in OVP	VPWR = 0 V or 3.3 V, $V_{C\_CCx} = 51$ V, CCx pins are set to 0 V, measure leakage current out of CCx pins.			30	$\mu\text{A}$
$I_{SBU\_LEAK\_OVP}$	Leakage current for SBU pins when device is in OVP	VPWR = 0 V, $V_{C\_SBUx} = 51$ V, SBUx pins are set to 0 V, measure leakage current into SBUx pins.	-1		1	$\mu\text{A}$
<b>/FLT Pin</b>						
$V_{OL}$	Low-level output voltage	IOL = 3mA. Measure voltage at FLT pin.			0.4	V
<b>Over Temperature Protection</b>						
$T_{SD\_RISING}$	The rising over-temperature protection shutdown threshold		150	175		$^{\circ}\text{C}$
$T_{SD\_FALLING}$	The falling over-temperature protection shutdown threshold		130	140		$^{\circ}\text{C}$
$T_{SD\_HYST}$	The over-temperature protection shutdown threshold hysteresis			35		$^{\circ}\text{C}$

## 5.6 Timing Requirements

		MIN	NOM	MAX	UNIT
<b>Power-On and Off Timings</b>					
$t_{ON\_FET}$	Time from Crossing Rising VPWR UVLO until CC and SBU OVP FETs are on.		1.3	3.5	ms
$t_{ON\_FET\_DB}$	Time from Crossing Rising VPWR UVLO until CC and SBU OVP FETs are on and the dead battery resistors are off.		5.7	9.5	ms
$dV_{PWR\_OFF}/dt$	Minimum slew rate for CC and FETs turn off during a power off.	-0.5			V/ $\mu\text{s}$
<b>Over Voltage Protection</b>					
$t_{OVP\_RESPONSE\_CC}$	OVP response time on the CCx pins. Time from OVP asserted until OVP FETs turn off.		70		ns
$t_{OVP\_RESPONSE\_SBU}$	OVP response time on the SBUx pins. Time from OVP asserted until OVP FETs turn off.		80		ns
$t_{OVP\_RECOVERY\_CC}$	OVP recovery time on the CCx pins. Once an OVP has occurred, the minimum time duration until the CC FETs turn back on. OVP must be removed for CC FETs to turn back on.		0.9	2.3	ms
$t_{OVP\_RECOVERY\_CC\_DB}$	OVP recovery time on the CCx pins. Once an OVP has occurred, the minimum time duration until the CC FETs turn back on and the dead battery resistors turn off. OVP must be removed for CC FETs to turn back on.		5		ms
$t_{OVP\_RECOVERY\_SBU}$	OVP recovery time on the SBUx pins. Once an OVP has occurred, the minimum time duration until the SBU FETs turn back on. OVP must be removed for SBU FETs to turn back on.		0.62		ms
$t_{OVP\_FLT\_ASSERTION}$	Time from OVP Asserted to /FLT assertion. FLT assertion is 10% of the maximum value. Set C_CCx or C_SBUx above the maximum OVP threshold. Start the time where it passes the typical OVP threshold value.		20		$\mu\text{s}$
$t_{OVP\_FLT\_DEASSERTION}$	Time from CC FET turn on after an OVP to FLT deassertion.		5		ms

## 5.7 Typical Characteristics

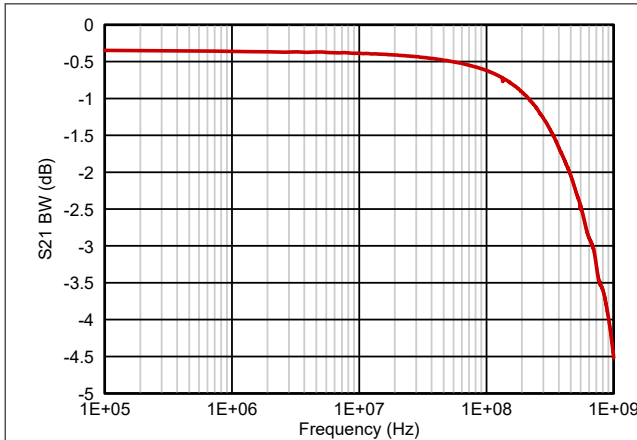


Figure 5-1. SBU Bandwidth

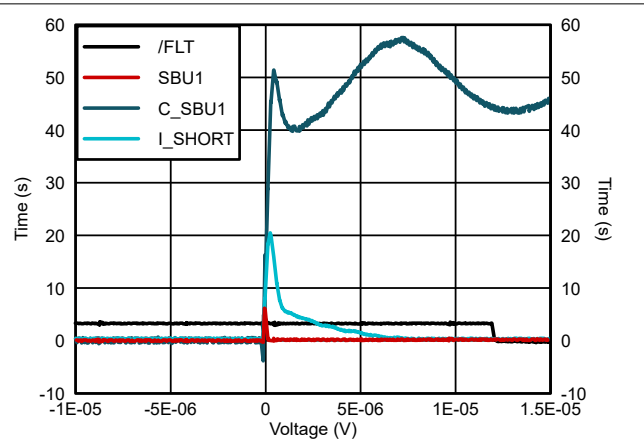


Figure 5-2. SBU Short-to-V<sub>BUS</sub> 48V

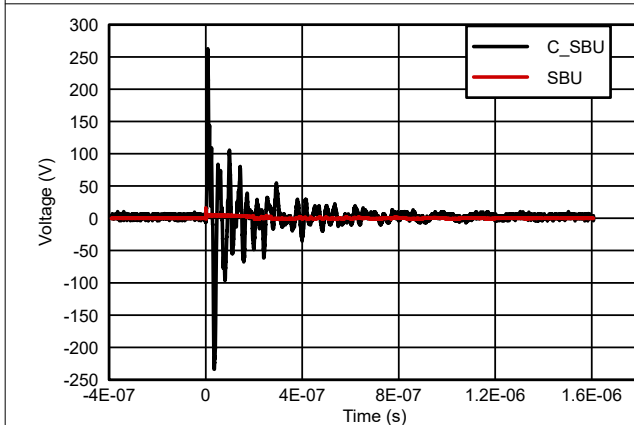


Figure 5-3. SBU IEC 61000-4-2 4kV Response Waveform

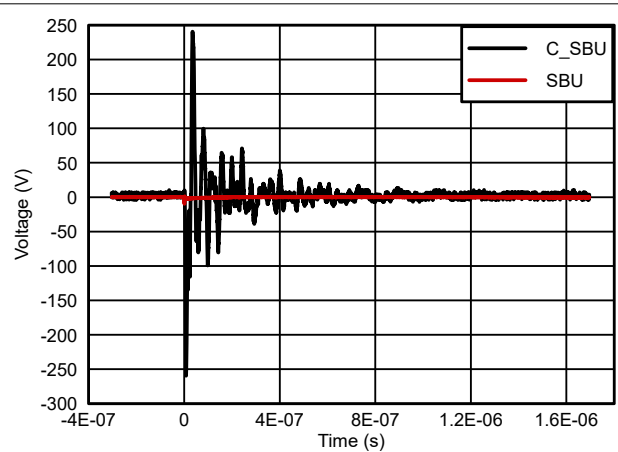


Figure 5-4. SBU IEC 61000-4-2 -4kV Response Waveform

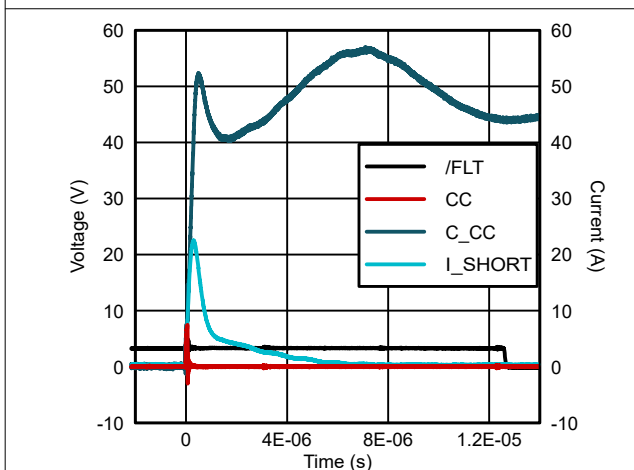


Figure 5-5. CC Short-to-V<sub>BUS</sub> 48V

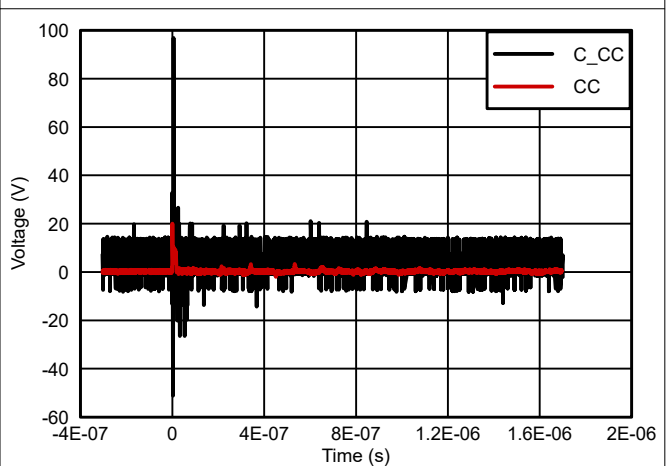


Figure 5-6. CC IEC 61000-4-2 8kV Response Waveform

### 5.7 Typical Characteristics (continued)

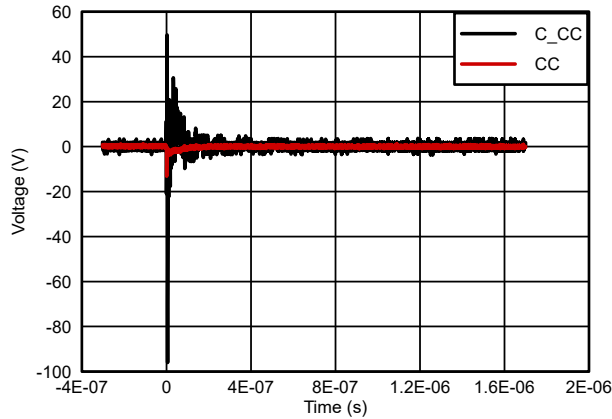


Figure 5-7. CC IEC 61000-4-2 -8kV Response Waveform

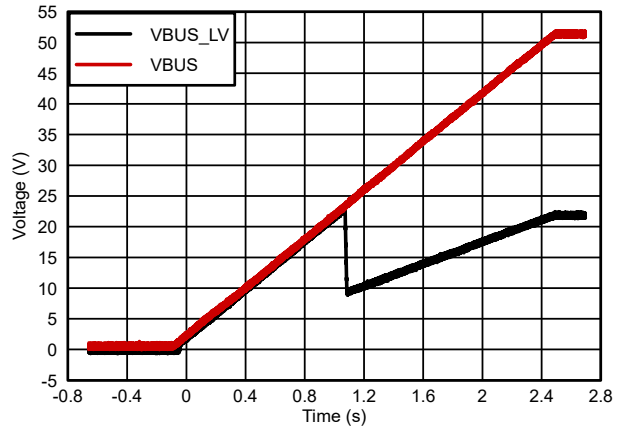


Figure 5-8. VBUS Sweep with EPR\_EN low

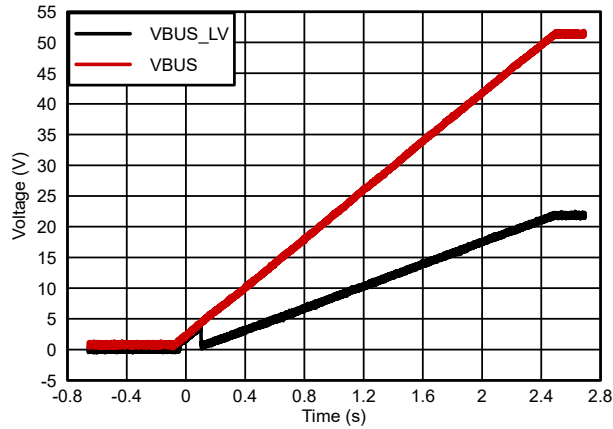


Figure 5-9. VBUS Sweep with EPR\_EN high

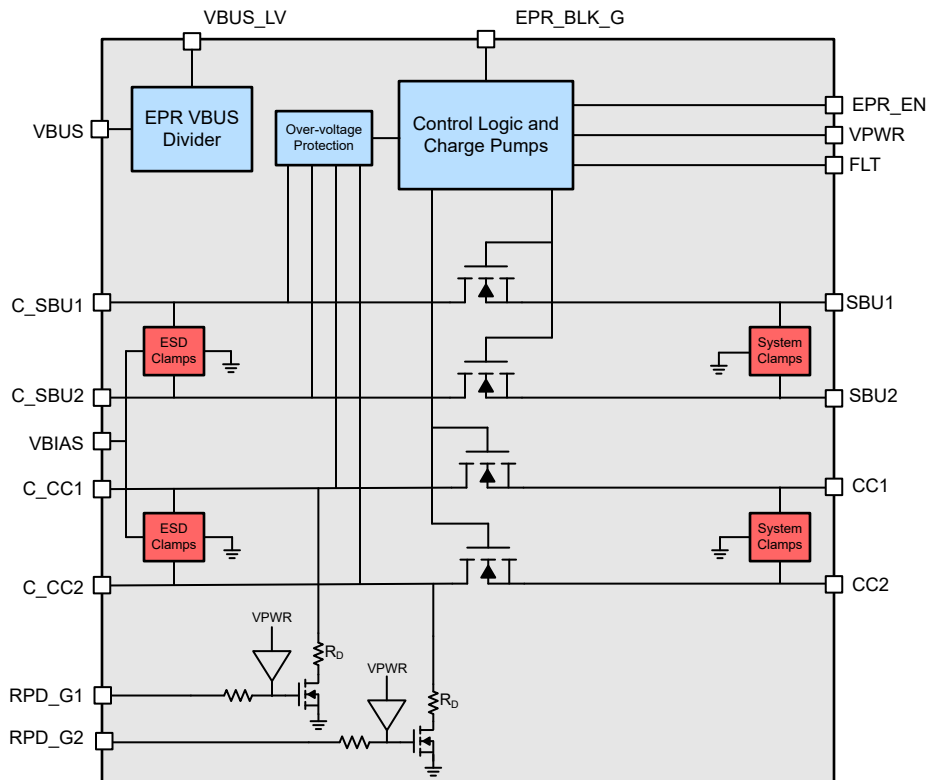
## 6 Detailed Description

### 6.1 Overview

The TPD4S480 is a single chip USB Type-C port protector that provides 48V Short-to- $V_{BUS}$  overvoltage and ESD protection. Due to the small pin pitch of the USB Type-C connector and non-compliant USB Type-C cables and accessories, the  $V_{BUS}$  pins can get shorted to the CC and SBU pins inside the USB Type-C connector. Because of this short-to- $V_{BUS}$  event, the CC and SBU pins need to be 48V tolerant, to support protection on the full USB PD-EPR voltage range. The TPD4S480 integrates four channels of 48V Short-to- $V_{BUS}$  overvoltage protection for the CC1, CC2, SBU1, and SBU2 pins of the USB Type-C connector.

Additionally, IEC 61000-4-2 system level ESD protection is required to protect a USB Type-C port from ESD strikes generated by end product users. The TPD4S480 integrates four channels of IEC61000-4-2 ESD protection for the CC1, CC2, SBU1, and SBU2 pins of the USB Type-C connector. This means ESD protection is provided for all of the low-speed pins on the USB Type-C connector in a single chip by the TPD4S480. Additionally, high-voltage ESD protection that is 55V DC tolerant is required for the CC and SBU lines to simultaneously support ESD and Short-to- $V_{BUS}$  protection. The TPD4S480 integrates a high-voltage ESD diode designed to work in conjunction with the overvoltage protection FETs inside the device.

### 6.2 Functional Block Diagram



### 6.3 Feature Description

#### 6.3.1 4-Channels of Short-to- $V_{BUS}$ Overvoltage Protection (CC1, CC2, SBU1, SBU2 Pins or CC1, CC2, DP, DM Pins): 63- $V_{DC}$ Tolerant

The TPD4S480 provides 4-channels of Short-to- $V_{BUS}$  Overvoltage Protection for the CC1, CC2, SBU1, and SBU2 pins (or the CC1, CC2, DP, and DM pins) of the USB Type-C connector. The TPD4S480 is able to handle 63 $V_{DC}$  on its C\_CC1, C\_CC2, C\_SBU1, and C\_SBU2 pins. This level of protection is necessary because according to the USB PD specification, with  $V_{BUS}$  set for 48-V operation, the  $V_{BUS}$  voltage is allowed to legally swing up to 50.4V and 50.9V on voltage transitions from a different USB PD  $V_{BUS}$  voltage. The TPD4S480 builds in tolerance up to 63 $V_{BUS}$  to provide margin above this 50.9V specification to be able to support inductive ringing that may occur during a short event.

When a short-to- $V_{BUS}$  event occurs, ringing happens due to the RLC elements in the hot-plug event. With very low resistance in this RLC circuit, ringing up to twice the settling voltage can appear on the connector. Ringing of more than twice the DC level can be generated if any capacitor on the line derates in capacitance value during the short-to- $V_{BUS}$  event. This behavior means that more than 90V could be seen on a USB Type-C pin during a Short-to- $V_{BUS}$  event. The TPD4S480 has built in circuit protection to handle this ringing. The diode clamps used for IEC ESD protection also clamp the ringing voltage during the short-to- $V_{BUS}$  event to limit the peak ringing to approximately 53V. Additionally, the overvoltage protection FETs integrated inside the TPD4S480 are 63V tolerant, therefore being capable of supporting the high-voltage ringing waveform that is experienced during the short-to- $V_{BUS}$  event. The well designed combination of voltage clamps and 63V tolerant OVP FETs insures the TPD4S480 can handle Short-to- $V_{BUS}$  hot-plug events with hot-plug voltages as high as  $51V_{DC}$ .

The TPD4S480 has an extremely fast turnoff time of 70ns typical. Furthermore, additional voltage clamps are placed after the OVP FET on the system side (CC1, CC2, SBU1, SBU2) pins of the TPD4S480, to further limit the voltage and current that are exposed to the USB Type-C CC/PD controller during the 70ns interval while the OVP FET is turning off. The combination of connector side voltage clamps, OVP FETs with extremely fast turnoff time, and system side voltage clamps all work together to insure the level of stress seen on a CC1, CC2, SBU1, or SBU2 pin during a short-to- $V_{BUS}$  event is less than or equal to an HBM event.

The SBU OVP FETs are designed with to be able to optionally protect the DP, DM (USB2.0) pins in lieu of the SBU pins. Some systems designers also prefer to protect the DP, DM pins from Short-to- $V_{BUS}$  events due to the potential for moisture/water in the connector to short the  $V_{BUS}$  pins to DP, DM pins. This can be especially applicable in cases where the end equipment with a USB Type-C connector is trying to be made water-proof. If desiring to protect the DP, DM pins on the USB Type-C connector from a Short-to- $V_{BUS}$  event, connect the C\_SBUx pins to the DP, DM pins on the USB Type-C connector, and the SBUx pins to the USB2.0 pins of the system device being protected from the Short-to- $V_{BUS}$  event.

### 6.3.2 CC1, CC2 Overvoltage Protection FETs 600-mA Capable for Passing VCONN Power

The CC pins on the USB Type-C connector serve many functions; one of the functions is to be a provider of power to active cables. Active cables are required when desiring to pass greater than 3 A of current on the  $V_{BUS}$  line or when the USB Type-C port uses the super-speed lines (TX1+, TX2-, RX1+, RX1-, TX2+, TX2-, RX2+, RX2-). When CC is configured to provide power, it is called VCONN. VCONN is a DC voltage source in the range of 3V to 5.5V. If supporting VCONN, a VCONN provider must be able to provide 1.5 W of power to a cable; this translates into a current range of 273mA to 500mA (depending on the VCONN voltage level).

When a USB Type-C port is configured for VCONN and using the TPD4S480, this VCONN current flows through the OVP FETs of the TPD4S480. Therefore, the TPD4S480 has been designed to handle these currents and have an RON low enough to provide a specification compliant VCONN voltage to the active cable.

### 6.3.3 CC Dead Battery Resistors Integrated for Handling the Dead Battery Use Case in Mobile Devices

An important feature of USB Type-C and USB PD is the ability for this connector to serve as the sole power source to mobile devices. With support up to 240W, the USB Type-C connector supporting USB PD can be used to power a whole new range of mobile devices not previously possible with legacy USB connectors.

When the USB Type-C connector is the sole power supply for a battery powered device, the device must be able to charge from the USB Type-C connector even when its battery is dead. In order for a USB Type-C power adapter to supply power on  $V_{BUS}$ , RD pulldown resistors must be exposed on the CC pins. These RD resistors are typically included inside a USB Type-C CC/PD controller. However, when the TPD4S480 is used to protect the USB Type-C port, the OVP FETs inside the device isolate these RD resistors in the CC/PD controller when the mobile device has no power. When the TPD4S480 has no power, the OVP FETs are turned off to provide overvoltage protection in a dead battery condition. Therefore, the TPD4S480 integrates high-voltage, dead battery RD pull-down resistors to allow dead battery charging simultaneously with high-voltage OVP protection.

If dead battery support is required, short the RPD\_G1 pin to the C\_CC1 pin, and short the RPD\_G2 pin to the C\_CC2 pin. This short connects the dead battery resistors to the connector CC pins. When the TPD4S480 is unpowered, and the RP pull-up resistor is connected from a power adapter, this RP pull-up resistor activates the RD resistor inside the TPD4S480, and enables  $V_{BUS}$  to be applied from the power adapter even in a dead

battery condition. Once power is restored back to the system and back to the TPD4S480 on its VPWR pin, the TPD4S480 turns ON its OVP FETs in 3.5ms and then turns OFF its dead battery RD. The TPD4S480 first turns ON its CC OVP FETs fully, and then removes its dead battery RDs to make sure the PD controller RD is fully exposed before removing the RD of the TPD4S480.

If desiring to power the CC/PD controller during dead battery mode and if the CC/PD Controller is configured as a DRP, it is critical that the TPD4S480 be powered before or at the same time that the CC/PD controller is powered. It is also critical that when unpowered, the CC/PD controller also expose its dead battery resistors. When the TPD4S480 gets powered, it exposes the CC pins of the CC/PD controller within 3.5ms, and then removes its own RD dead battery resistors. Once the TPD4S480 turns on, the RD pull-down resistors of the CC/PD controller must be present immediately, in order to maintain a connection. If the power adapter does not see RD present, it can disconnect  $V_{BUS}$ . This event removes power from the device with its battery still not sufficiently charged, which consequently removes power from the CC/PD controller and the TPD4S480. Then the RD resistors of the TPD4S480 are exposed again, and connects the  $V_{BUS}$  of the power adapter to start the cycle over.

If the CC/PD Controller is configured for DRP and has started to DRP toggle before the TPD4S480 turns on, this DRP toggle is unable to maintain a connection with a power adapter. If the CC/PD controller is configured for DRP, the dead battery resistors of the PD controller need to be exposed as well, and that the resistors remain exposed until the TPD4S480 turns on. This behavior is typically accomplished by powering the TPD4S480 at the same time as the CC/PD controller when powering the CC/PD controller in dead battery operation.

If dead battery charging is not required in your application, connect the RPD\_G1 and RPD\_G2 pins to ground.

### 6.3.4 EPR Adapter

The TPD4S480 integrates additional circuitry that may be used to adapt a PD controller with pin tolerances below EPR levels for use in EPR applications. The EPR adapter consists of two components, the VBUS divider and the EPR blocking FET gate driver. The EPR adapter features are enabled by asserting the EPR\_EN pin or when the VBUS pin exceeds EPR\_THRESH\_R.

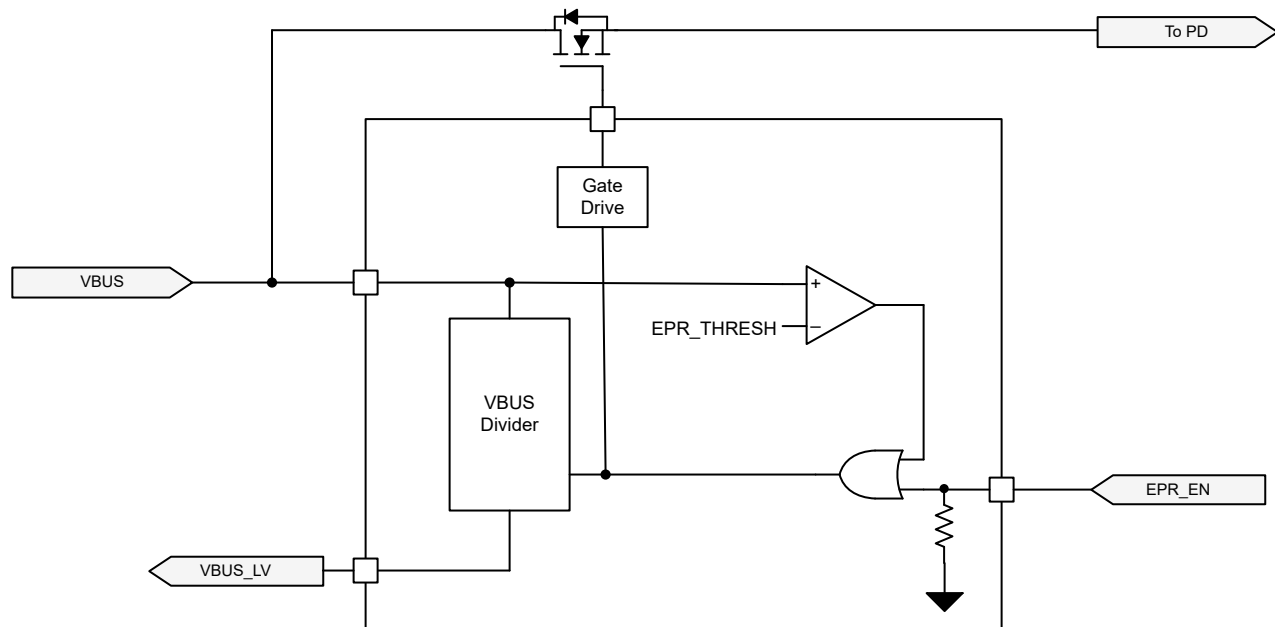


Figure 6-1. EPR Adapter

#### 6.3.4.1 VBUS Divider

The VBUS divider provides a divided down output of VBUS so that an attached PD controller can safely sense EPR voltages. Table 6-1 summarizes the operating states of the VBUS divider.

**Table 6-1. VBUS Divider States**

EPR_EN	VBUS	VBUS_LV Ratio (VBUS_LV / VBUS)	Description
0	< EPR_THRES_R	1	SPR Operation
1	X	0.42	EPR Operation
X	> EPR_THRESH_R	0.42	

**6.3.4.2 EPR Blocking FET Gate Driver**

An NFET gate driver is integrated for controlling an external blocking FET. When in EPR mode the gate driver is disabled, isolating any non-EPR tolerant circuitry from VBUS. When in SPR mode the gate driver is enabled connecting low voltage components to VBUS.

**Table 6-2. VBUS Divider States**

EPR_EN	VBUS	Gate Driver State	Description
0	< EPR_THRES_R	Enabled	SPR Operation
1	X	Disabled	EPR Operation
X	> EPR_THRESH_R	Disabled	

**6.4 Device Functional Modes**

Table 6-3 describes all of the functional modes for the TPD4S480. The "X" in the below table are "don't care" conditions, meaning any value can be present within the absolute maximum ratings of the data sheet and maintain that functional mode.

**Table 6-3. Device Mode Table**

Device Mode Table		Inputs					Outputs				
MODE		VPWR	C <sub>CCx</sub>	C <sub>SBUx</sub>	RPD <sub>Gx</sub>	T <sub>J</sub>	FLT	CC FETs	SBU FETs	VBUS_LV	EPR_BLK <sub>G</sub>
Normal Operating Conditions	Unpowered, no dead battery support	<UVL O	X	X	Grounded	X	High-Z	OFF	OFF	VBUS	Disabled
	Unpowered, dead battery support	<UVL O	X	X	Shorted to C <sub>CCx</sub>	X	High-Z	OFF	OFF	VBUS	Disabled
	Powered on, SPR mode	>UVL O	<OVP	<OVP	X, forced OFF	<TSD	High-Z	ON	ON	VBUS	Enabled
	Powered on, EPR mode	>UVL O	<OVP	<OVP	X, forced OFF	<TSD	High-Z	ON	ON	Divided VBUS	Disabled
Fault Conditions	Thermal shutdown	>UVL O	X	X	X, forced OFF	>TSD	Low (Fault Asserted)	OFF	OFF	Maintains EPR state	Maintains EPR state
	CC over voltage condition	>UVL O	>OVP	X	X, forced OFF	<TSD	Low (Fault Asserted)	OFF	OFF	Maintains EPR state	Maintains EPR state
	SBU over voltage condition	>UVL O	X	>OVP	X, forced OFF	<TSD	Low (Fault Asserted)	OFF	OFF	Maintains EPR state	Maintains EPR state
	IEC ESD generated over voltage condition <sup>(1)</sup>	>UVL O	X	X	R <sub>D</sub> ON if RPD <sub>Gx</sub> is shorted to C <sub>CCx</sub>	<TSD	Low (Fault Asserted)	OFF	OFF	Maintains current EPR state	Maintains current EPR state

(1) This row describes the state of the device while still in OVP after the IEC ESD strike which put the device into OVP is over, and the voltages on the C<sub>CCx</sub> and C<sub>SBUx</sub> pins have returned to their normal voltage levels.

## 7 Application and Implementation

### Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

### 7.1 Application Information

The TPD4S480 provides 4-channels of Short-to- $V_{BUS}$  overvoltage protection for the CC1, CC2, SBU1, and SBU2 pins of the USB Type-C connector. Care must be taken to insure that the TPD4S480 provides adequate system protection as well as insuring that proper system operation is maintained. The following application example explains how to properly design the TPD4S480 into a USB Type-C system.

### 7.2 Typical Application

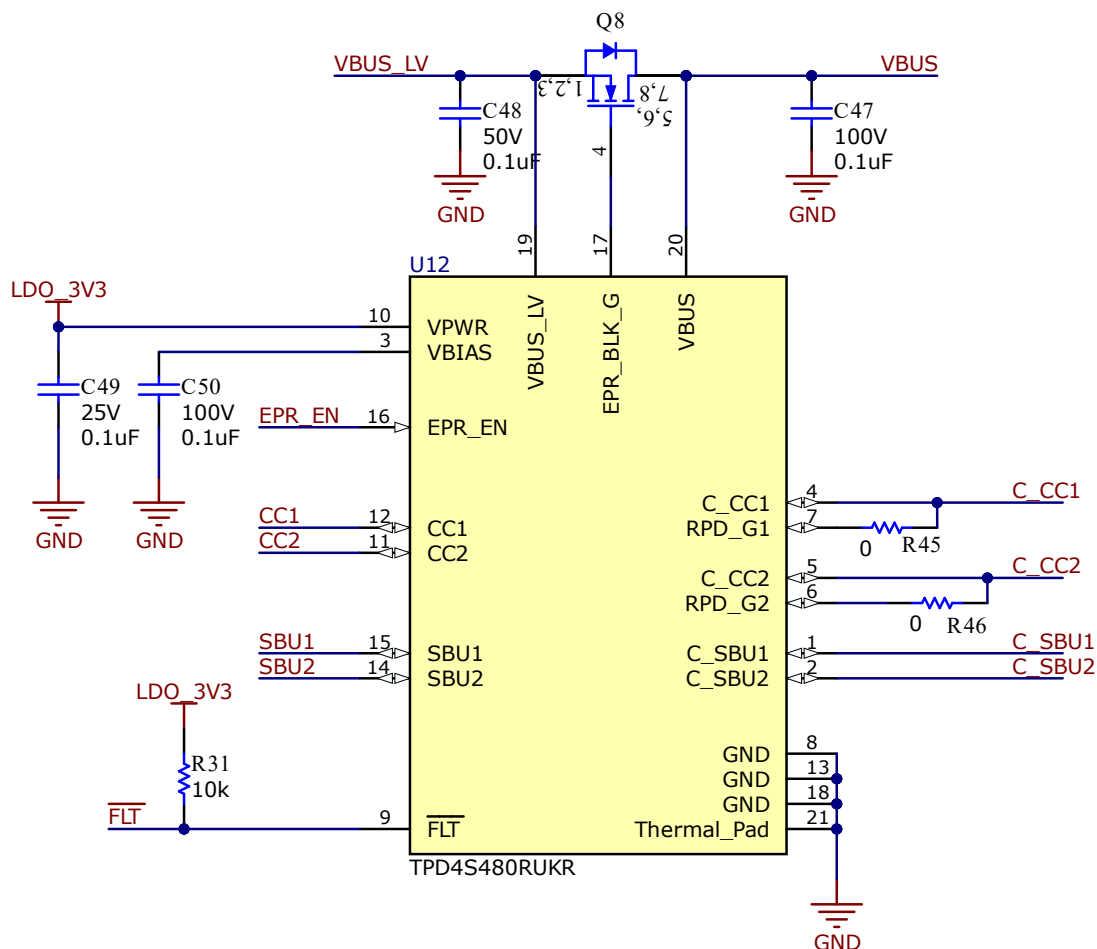


Figure 7-1. TPD4S480 Reference Schematic

#### 7.2.1 Design Requirements

In this application example we study the protection requirements for a USB Type-C DRP Port, fully equipped with USB-PD, and 240W charging. The [TPS26750](#) is used to easily enable a DRP port. Both the CC and SBU

pins are susceptible to shorting to the  $V_{BUS}$  pin. With 240W charging,  $V_{BUS}$  operates at 48V, requiring the CC and SBU pins to tolerate  $48V_{DC}$ . With these protection requirements present for the USB Type-C connector, the TPD4S480 is used.

Table 7-1 lists the TPD4S480 design parameters.

**Table 7-1. Design Parameters**

DESIGN PARAMETER	EXAMPLE VALUE
$V_{BUS}$ nominal operating voltage	48V
Short-to- $V_{BUS}$ tolerance for the CC and SBU pins	63V
VBIAS nominal capacitance	0.1 $\mu$ F
Dead battery charging	240W
Maximum ambient temperature requirement	85°C

The recommended MOSFET settings for Q8, as shown in Table 7-2, are as follows:

**Table 7-2. MOSFET Selection**

VDS (V)	VGS (V)	Type	RDS (on)
$\geq 30V$	$\geq 15V$	N-channel	$\leq 10m\Omega$ (3A max Q8 current) $\leq 20m\Omega$ (1.5A max Q8 current)

### 7.2.1.1 EPR Design Requirements

The TPD4S480 works in conjunction with the PD controller to provide the following functionality in USB-PD EPR:

- Short to VBUS protection for direct shorts to CC1 and CC2 pins of the Type-C connector.
- Short to VBUS protection for the liquid detection circuitry that is connected to the SBU1 and SBU2 pins of the Type-C connector, if the liquid detection feature is implemented.
- Voltage level translation from up the EPR maximum voltage down to the operation range of the VBUS pins of the PD controller.
- Gate drive for a high voltage NMOS transistor to allow the internal 5V power path to be used to source 5V in systems that only require a 5V output.

## 7.2.2 Detailed Design Procedure

### 7.2.2.1 VBIAS Capacitor Selection

As noted in the [Section 5.3](#) table, a minimum of  $63V_{BUS}$  rated capacitor is required for the VBIAS pin, and a  $100V_{BUS}$  capacitor is recommended. The VBIAS capacitor is in parallel with the central diode clamp integrated inside the TPD4S480. A forward biased hiding diode connects the VBIAS pin to the C\_CCx and C\_SBUx pins. Therefore, when a Short-to- $V_{BUS}$  event occurs at 48V,  $48V_{BUS}$  minus a forward biased diode drop is exposed to the VBIAS pin. Additionally, during the short-to- $V_{BUS}$  event, ringing can occur almost double the settling voltage of 48V, allowing a potential 96V to be exposed to the C\_CCx and C\_SBUx pins. However, the internal diode clamps limit the voltage exposed to the C\_CCx and C\_SBUx pins to around 63V. Therefore, at least 63V capacitor is required to insure the VBIAS capacitor does not get destroyed during Short-to- $V_{BUS}$  events.

A 100V, X7R capacitor is recommended to further improve the derating performance of the capacitors. When the voltage across a real capacitor is increased, the capacitance value derates. The more the capacitor derates, the larger the ringing in the short-to- $V_{BUS}$  RLC circuit. The 100V X7R capacitors have great derating performance, allowing for the best short-to- $V_{BUS}$  performance of the TPD4S480.

### 7.2.2.2 Dead Battery Operation

For this application, we want to support 240W dead battery operation; when the device is out of battery, we still want to charge the laptop at 48V and 5A. This means that the USB PD Controller must receive power in dead battery mode. The TPS26750 has a built in LDO to supply the TPS26750 power from  $V_{BUS}$  in a dead battery condition.

The OVP FETs of the TPD4S480 remain OFF when unpowered to provide protection in dead battery or unpowered situations. However, when the OVP FETs are OFF, this isolates the TPS26750s dead battery resistors from the USB Type-C ports CC pins. A USB Type-C power adapter must see the RD pull-down dead battery resistors on the CC pins to provide power on  $V_{BUS}$ . Since the dead battery resistors of the TPS26750 are isolated from the USB Type-C connector CC pins, the built-in, dead battery resistors of the TPD4S480 must be connected. Short the RPD\_G1 pin to the C\_CC1 pin, and short the RPD\_G2 pin to the C\_CC2 pin.

Once the power adapter sees the dead battery resistors of the TPD4S480, 5V is applied on the  $V_{BUS}$  pin. This provides power to the TPS26750, turning the PD controller on, and allowing the battery to begin to charge. However, this application requires 240W charging in dead battery mode, so  $V_{BUS}$  at 48V and 5A is required. USB PD negotiation is required to accomplish this, so the TPS26750 needs to be able to communicate on the CC pins. The TPD4S480 needs to be turned on in dead battery mode as well so the PD controller can be exposed to the CC lines. To turn on the device, the TPD4S480 is powered by the internal LDO of the TPS26750, the LDO\_3V3 pin. When the TPS26750 receives power on  $V_{BUS}$ , the TPD4S480 is turned on simultaneously.

The dead battery resistors of the PD controller also need to be present so the PD controller properly boots up in dead battery operation with the correct voltages on the CC pins.

Once this process has occurred, the TPS26750 can start negotiating with the power adapter through USB PD for higher power levels, allowing 240W operation in dead battery mode.

For more information on the TPD4S480 dead battery operation, see the [CC Dead Battery Resistors Integrated for Handling the Dead Battery Use Case in Mobile Devices](#) section of the data sheet.

### 7.2.2.3 CC Line Capacitance

USB PD has a specification for the total amount of capacitance that is required for proper USB PD BMC operation on the CC lines.

**Table 7-3. USB PD cReceiver Specification**

NAME	DESCRIPTION	MIN	MAX	UNIT	COMMENT
cReceiver	CC receiver capacitance	200	600	pF	The DFP or UFP system shall have capacitance within this range when not transmitting on the line

The capacitance on the CC lines must stay in between 200pF and 600pF when USB PD is being used. Therefore, the combination of capacitances added to the system by the TPS26750, the TPD4S480, and any external capacitor must fall within these limits.

#### 7.2.2.4 Additional ESD Protection on CC and SBU Lines

If additional IEC ESD protection is desired to be placed on either the CC or SBU lines, it is important that high-voltage ESD protection diodes be used. The maximum DC voltage that can be seen in USB PD is 50.4V, with 50.9V allowed during voltage transitions. Therefore, an ESD protection diode must have a reverse stand off voltage higher than 50.9V to prevent the diode from breakdown during a short-to- $V_{BUS}$  event.

Due to the fact that the Short-to- $V_{BUS}$  event applies a DC voltage to the CC and SBU pins, a deep-snap-back diode cannot be used unless its minimum trigger voltage is above 96V. During a Short-to- $V_{BUS}$  event, RLC ringing of up to twice the settling voltage can be exposed to CC and SBU, allowing for up to 48V to be exposed. Furthermore, if any capacitor derates on the CC or SBU line, greater ringing can occur. Since this ringing is hard to bound, it is recommended to not use deep-snap-back diodes. If the deep-snap-back diode triggers during the short-to- $V_{BUS}$  hot-plug event, it begins to operate in its conduction region. With a 48V  $V_{BUS}$  source present on the CC or SBU line, this allows the diode to conduct indefinitely.

#### 7.2.2.5 $\overline{FLT}$ Pin Operation

Once a Short-to- $V_{BUS}$  occurs on the C\_CCx or C\_SBUx pins, the  $\overline{FLT}$  pin is asserted in 20 $\mu$ s (typical) so the PD controller can be notified quickly. If  $V_{BUS}$  is being shorted to CC or SBU, it is recommended to respond to the event by forcing a detach in the USB PD controller to remove  $V_{BUS}$  from the port. Although the USB Type-C port using the TPD4S480 is not damaged, as the TPD4S480 provides protection from these events, the other device connected through the USB Type-C Cable or any active circuitry in the cable can be damaged. Although shutting the  $V_{BUS}$  off through a detach does not ensure it stops the other device or cable from being damaged, it can mitigate any high current paths from causing further damage after the initial damage takes place. Additionally, even if the active cable or other device does have proper protection, the short-to- $V_{BUS}$  event may corrupt a configuration in an active cable or in the other PD controller, so it is best to detach and reconfigure the port.

#### 7.2.2.6 How to Connect Unused Pins

If either the RPD\_Gx pins are unused in a design, they must be connected to GND.

### 7.2.3 EPR Application Curves

Figure 7-2 shows the  $V_{BUS}$  and  $V_{BUS\_LV}$  pins during a 5V to 48V voltage transition.

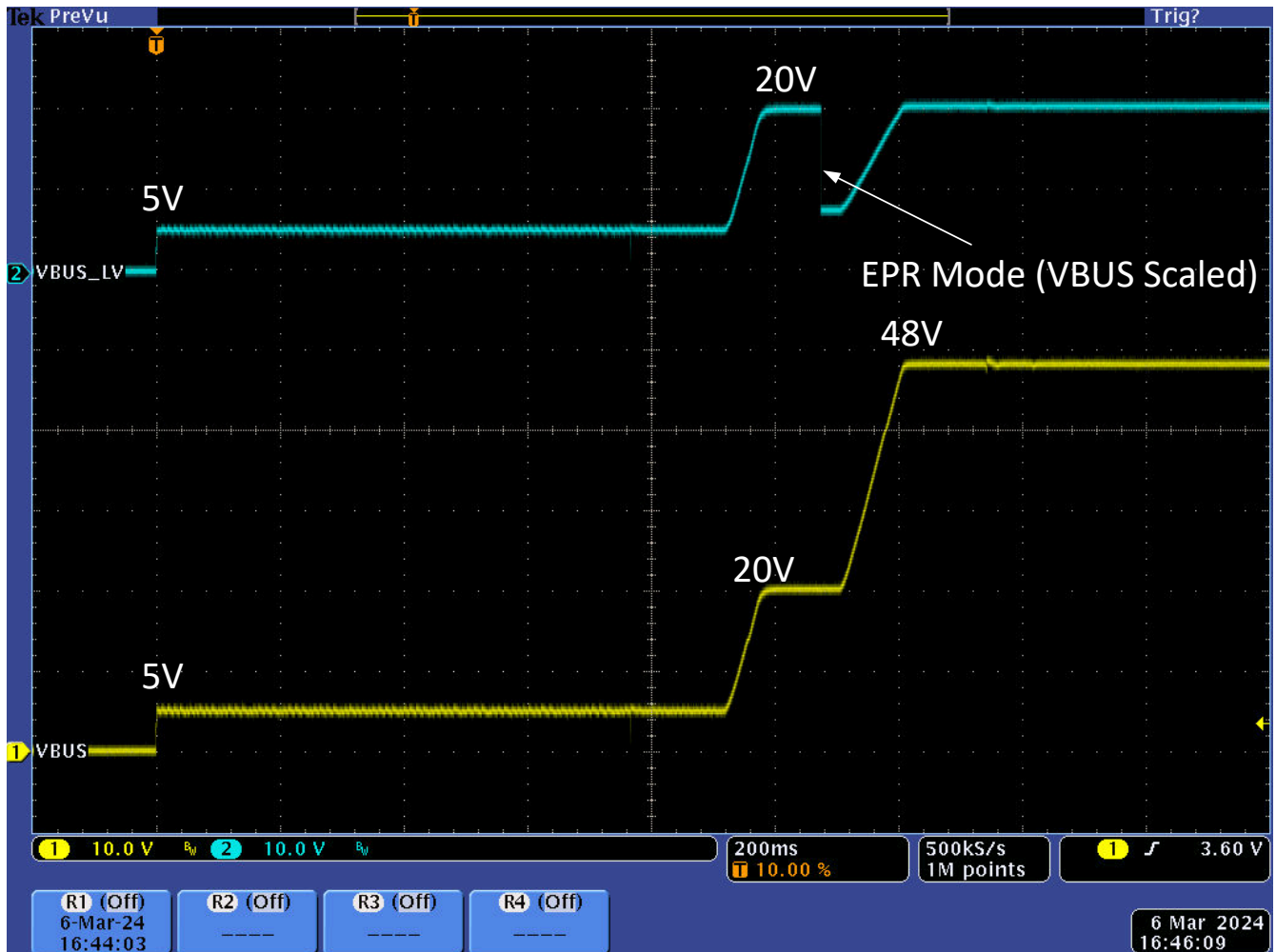


Figure 7-2. 48V EPR Contract Negotiation VBUS and VBUS LV

### 7.3 Power Supply Recommendations

The  $V_{PWR}$  pin provides power to all the circuitry in the TPD4S480. It is recommended a 1- $\mu$ F decoupling capacitor is placed as close as possible to the  $V_{PWR}$  pin. If USB PD is desired to be operated in dead battery conditions, it is critical that the TPD4S480 share the same power supply as the PD controller in dead battery boot-up (such as sharing the same dead battery LDO). See the [CC Dead Battery Resistors Integrated for Handling the Dead Battery Use Case in Mobile Devices](#) section for more details.

### 7.4 Layout

#### 7.4.1 Layout Guidelines

Proper routing and placement is important to maintain the signal integrity the USB2.0, SBU, CC line signals. The following guidelines apply to the TPD4S480 device:

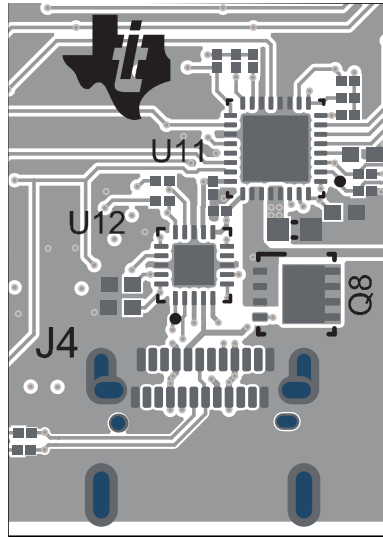
- Place the bypass capacitors as close as possible to the  $V_{PWR}$  pin, and ESD protection capacitor as close as possible to the  $V_{BIAS}$  pin. Capacitors must be attached to a solid ground. This minimizes voltage disturbances during transient events such as short-to- $V_{BUS}$  and ESD strikes.
- The USB2.0 and SBU lines must be routed as straight as possible and any sharp bends must be minimized.

Standard ESD recommendations apply to the C\_CC1, C\_CC2, C\_SBU1, and C\_SBU2 as well:

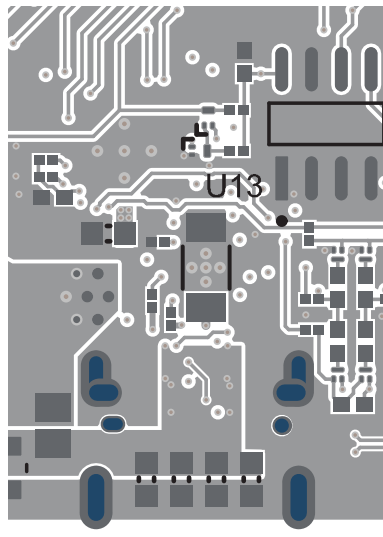
- The optimum placement for the device is as close to the connector as possible:

- EMI during an ESD event can couple from the trace being struck to other nearby unprotected traces, resulting in early system failures.
- The PCB designer must minimize the possibility of EMI coupling by keeping any unprotected traces away from the protected traces which are between the TPD4S480 device and the connector.
- Route the protected traces as straight as possible.
- Eliminate any sharp corners on the protected traces between the TVS and the connector by using rounded corners with the largest radii possible.
  - Electric fields tend to build up on corners, increasing EMI coupling.

**7.4.2 Layout Example**



**Figure 7-3. TPS26750 PCB Layout - Top Composite**



**Figure 7-4. TPS26750 PCB Layout - Bottom Composite**

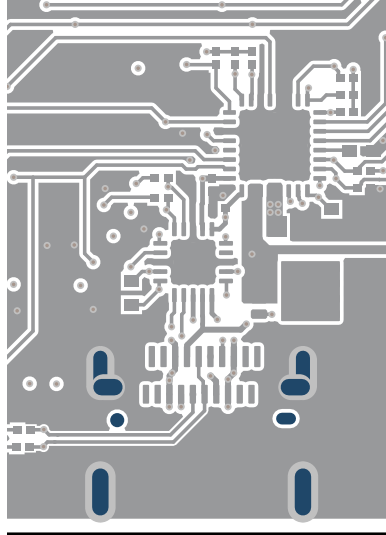


Figure 7-5. TPS26750 PCB Layout - Top Layer 1



Figure 7-6. TPS26750 PCB Layout - GND Layer 2

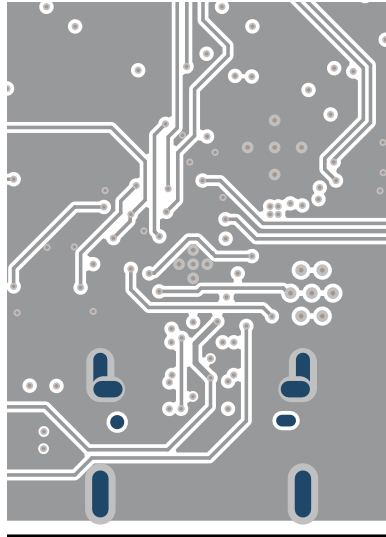


Figure 7-7. TPS26750 PCB Layout - Signal Layer 3

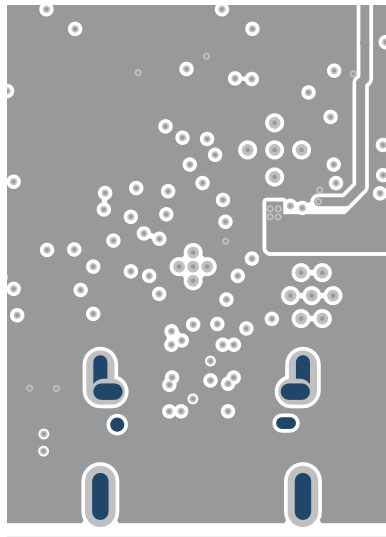
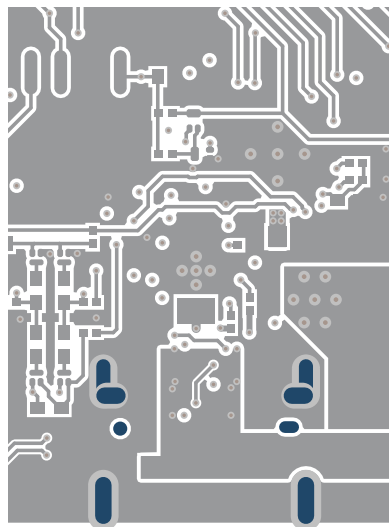


Figure 7-8. TPS26750 PCB Layout - Signal Layer 4



**Figure 7-9. TPS26750 PCB Layout - GND Layer 5**



**Figure 7-10. TPS26750 PCB Layout - Bottom Layer 6**

## 8 Device and Documentation Support

### 8.1 Documentation Support

#### 8.1.1 Related Documentation

For related documentation see the following:

[TPS26750 USB Type-C® and USB PD Controller With Integrated Power Switches Optimized for Power Applications](#)

### 8.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on [ti.com](https://www.ti.com). Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

### 8.3 Support Resources

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

### 8.4 Trademarks

TI E2E™ is a trademark of Texas Instruments.  
All trademarks are the property of their respective owners.

### 8.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 8.6 Glossary

[TI Glossary](#) This glossary lists and explains terms, acronyms, and definitions.

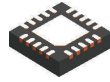
## 9 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

DATE	REVISION	NOTES
November 2024	*	Initial Release

## 10 Mechanical, Packaging, and Orderable Information

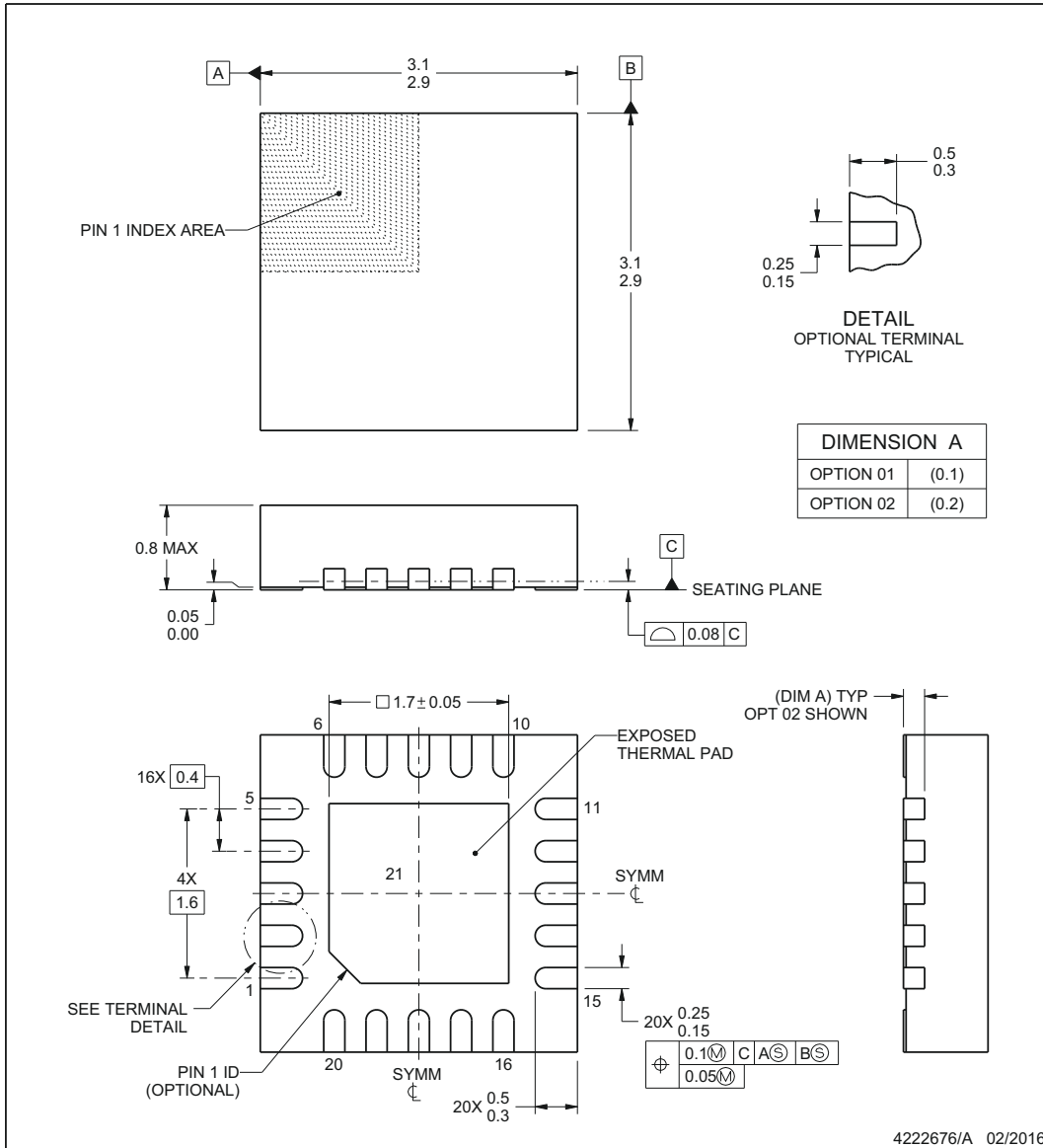
The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



**RUK0020B**

**PACKAGE OUTLINE**  
**WQFN - 0.8 mm max height**

PLASTIC QUAD FLATPACK - NO LEAD



**NOTES:**

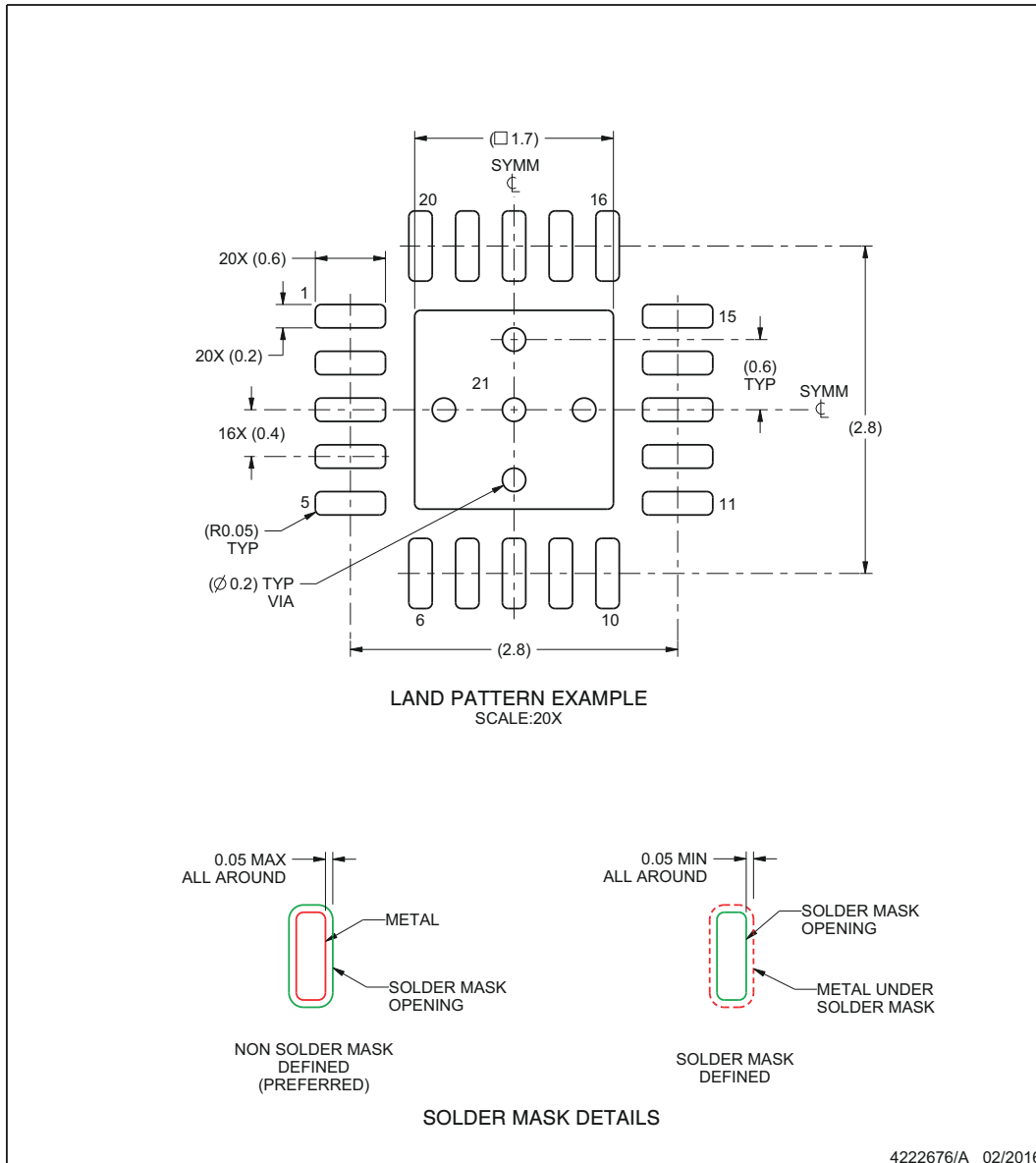
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

## EXAMPLE BOARD LAYOUT

**RUK0020B**

**WQFN - 0.8 mm max height**

PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

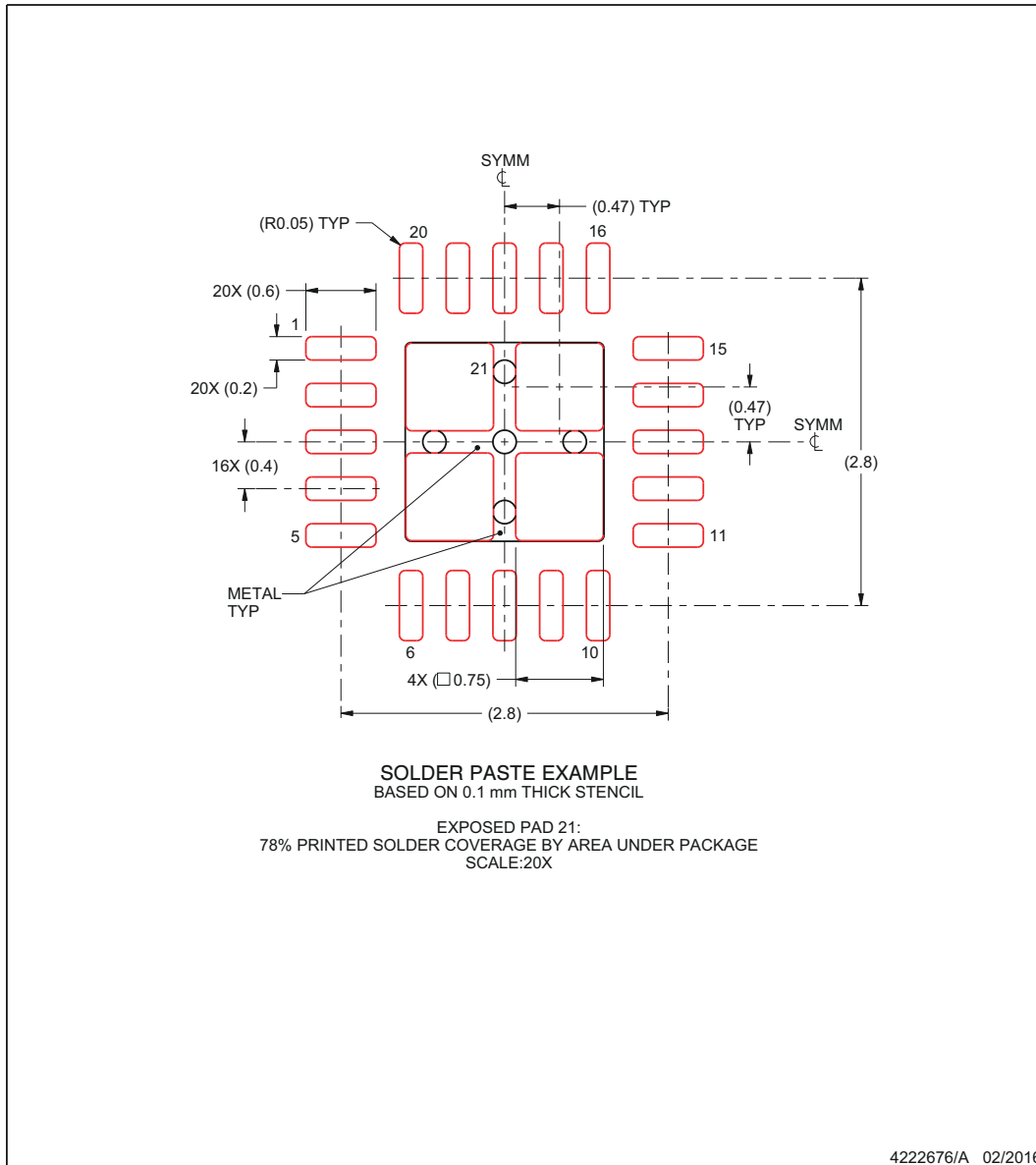
4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 ([www.ti.com/lit/sluea271](http://www.ti.com/lit/sluea271)).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

## EXAMPLE STENCIL DESIGN

**RUK0020B**

**WQFN - 0.8 mm max height**

PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

**PACKAGING INFORMATION**

Orderable part number	Status (1)	Material type (2)	Package   Pins	Package qty   Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
<a href="#">TPD4S480RUKR</a>	Active	Production	WQFN (RUK)   20	5000   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	4S480
TPD4S480RUKR.A	Active	Production	WQFN (RUK)   20	5000   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	4S480

(1) **Status:** For more details on status, see our [product life cycle](#).

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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**OTHER QUALIFIED VERSIONS OF TPD4S480 :**

- Automotive : [TPD4S480-Q1](#)

NOTE: Qualified Version Definitions:

- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

**TAPE AND REEL INFORMATION**

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPD4S480RUKR	WQFN	RUK	20	5000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
TPD4S480RUKR	WQFN	RUK	20	5000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2

**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPD4S480RUKR	WQFN	RUK	20	5000	367.0	367.0	35.0
TPD4S480RUKR	WQFN	RUK	20	5000	346.0	346.0	33.0

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Last updated 10/2025