# AC-coupled transimpedance amplifier circuit



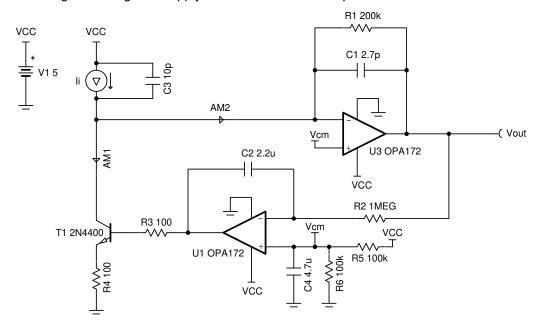
**Amplifiers** 

## **Design Goals**

Input Current		Ambient light ourrent	Output voltage		Target Bandwidth	Supply	
I <sub>iMin</sub>	I <sub>iMax</sub>	- Ambient light current	V <sub>oMin</sub>	V <sub>oMax</sub>	raiget Baildwidtii	V <sub>cc</sub>	V <sub>ee</sub>
–10µA	10µA	100µA	0.5V	4.5V	300kHz	5V	0V

#### **Design Description**

This circuit uses an op amp configured as a transimpedance amplifier to amplify the AC signal of a photodiode (modeled by  $I_i$  and  $C_3$ ). The circuit rejects DC signals using a transistor to sink DC current out of the photodiode through the use of an integrator in a servo loop. The bias voltage applied to the non-inverting input prevents the output from saturating to the negative supply rail in the absence of input current.



#### **Design Notes**

- 1. Use a JFET or CMOS input op amp with low-bias current to reduce DC errors.
- 2. A capacitor placed in parallel with the feedback resistor will limit bandwidth, improve stability and help reduce noise.
- The junction capacitance of photodiode changes with reverse bias voltage which will influence the stability of the circuit.
- 4. Reverse-biasing the photodiode can reduce the effects of dark current.
- 5. A resistor, R<sub>3</sub>, may be needed on the output of the integrator amplifier.
- An emitter degeneration resistor, R<sub>4</sub>, should be used to help stabilize the BJT.
- Use the op amp in a linear operating region. Linear output swing is usually specified under the A<sub>OL</sub> test conditions.



#### **Design Steps**

The transfer function of the circuit is:

$$V_{out} = -I_i \times R_1$$

1. Calculate the value of the feedback resistor, R<sub>1</sub>, to produce the desired output swing.

$$R_1 = \frac{V_{0Max} - V_{0Min}}{I_{1Max} - I_{1Min}} = \frac{4.5V - 0.5V}{10\mu A - (-10\mu A)} = 200k\Omega$$

2. Calculate the feedback capacitor to limit the signal bandwidth.

$$\text{C}_1 \!\!=\!\! \frac{1}{2\pi \times \text{R}_1 \! \times \text{f}_p} \!\!=\!\! \frac{1}{2\pi \times 200 \text{k}\Omega \times 300 \text{kHz}} \!\!= 2.65 \text{pF} \approx 2.7 \text{pF (Standard Value)}$$

3. Calculate the gain bandwidth of the amplifier needed for the circuit to be stable.

$$GBW = \frac{C_i + C_1}{2\pi \times R_1 \times C_1^2} = \frac{23pF + 2.7pF}{2\pi \times 200k\Omega \times (2.7pF)^2} = 2.97MHz$$

Where:

$$C_i = C_{pd} + C_b + C_d + C_{cm} = 10pF + 5pF + 4pF + 4pF = 23pF$$

Given:

- C<sub>pd</sub>: Junction capacitance of photodiode
- C<sub>b</sub>: Output capacitance of BJT
- C<sub>d</sub>: Differential input capacitance of the amplifier
- C<sub>cm</sub>: Common-mode input capacitance of the inverting input
- 4. Set the cutoff frequency of the integrator circuit,  $f_l$ , to 0.1Hz to only allow signals near DC to be subtracted from the photodiode output current. The cutoff frequency is set by  $R_2$  and  $C_2$ . Select  $R_2$  as  $1M\Omega$ .

$$C_2 = \frac{1}{2\pi \times R_2 \times f_1} = \frac{1}{2\pi \times 1M\Omega \times 0.1Hz} = 1.59 \mu F \approx 2.2 \mu F \text{ (Standard Value)}$$

- 5. Select  $R_3$  as  $100\Omega$  to isolate the capacitance of the BJT from op amp and stabilize the amplifier. For more information on stability analysis, see the Design References section [2].
- 6. Bias the output of the circuit by setting the input common mode voltage of the integrator circuit to mid-supply. Select  $R_5$  and  $R_6$  as  $100k\Omega$ .

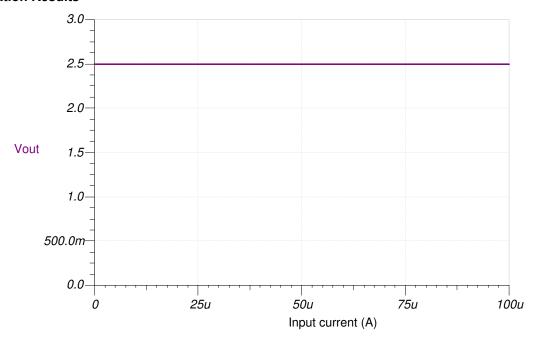
$$Vcm = \frac{R_6}{R_5 + R_6} \times Vcc = \frac{100k\Omega}{100k\Omega + 100k\Omega} \times 5V = 2.5V$$

7. Calculate capacitor C<sub>2</sub> to filter the power supply and resistor noise. Set the cutoff frequency to 1Hz.

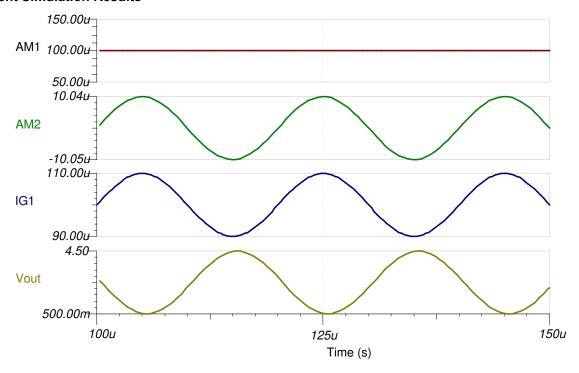
$$C_2 = \frac{1}{2\pi \times (R_2||R_3) \times 1Hz} = \frac{1}{2\pi \times (100k\Omega\,||\,100k\Omega) \times 1Hz} = 3.183 \mu F \approx 4.7 \mu F$$

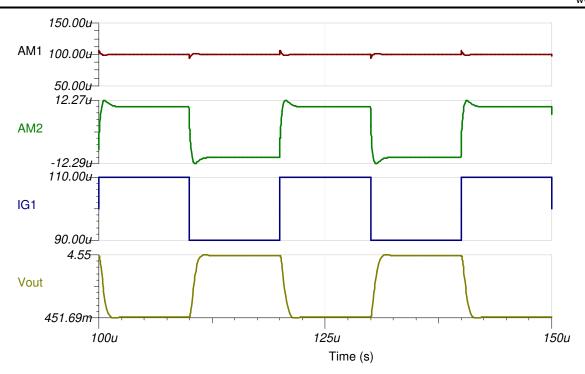
## **Design Simulations**

## **DC Simulation Results**

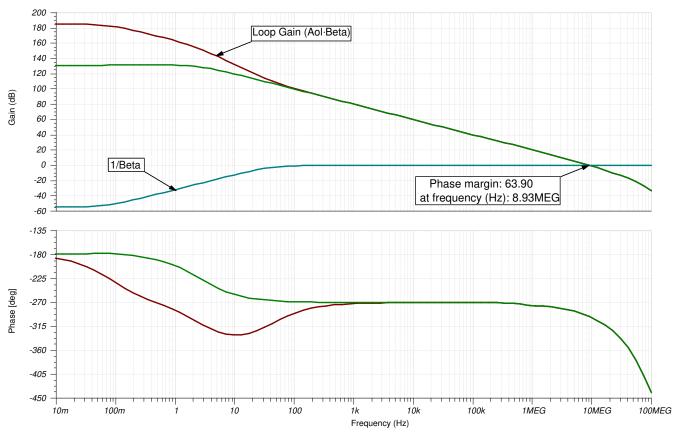


### **Transient Simulation Results**



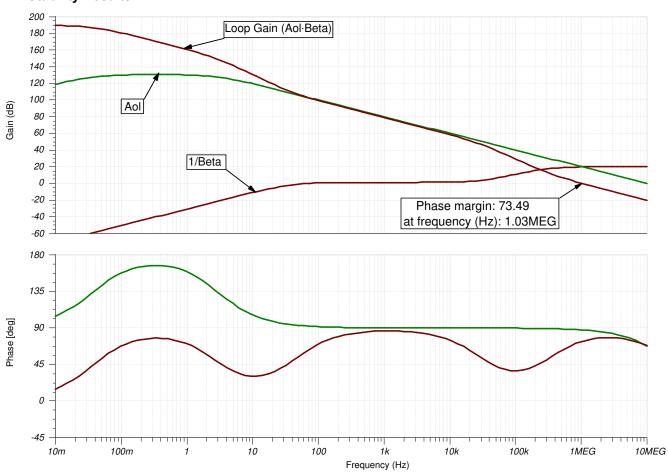


# **Integrator Open Loop Stability**





# **TIA Stability Results**



### **Design References**

- 1. See Analog Engineer's Circuit Cookbooks for TI's comprehensive circuit library.
- 2. TI Precision Labs

### **Design Featured Op Amp**

OPA172				
V <sub>cc</sub>	±2.25V to ±18V, 4.5V to 36V			
V <sub>inCM</sub>	(V–) – 0.1V to (V+) – 2V			
V <sub>out</sub>	Rail-to-rail			
V <sub>os</sub>	0.2mV			
Iq	1.6mA			
l <sub>b</sub>	8pA			
UGBW	10MHz			
SR	10V/µs			
#Channels	1,2,4			
www.ti.com/product/OPA172				

## **Design Alternate Op Amps**

	OPA2991	TLV9042
V <sub>ss</sub>	±1.35V to ±20V, 2.7V to 40V	±0.6V to ±2.75V, 1.2V to 5.5V
V <sub>inCM</sub>	Rail-to-rail	Rail-to-rail
V <sub>out</sub>	Rail-to-rail	Rail-to-rail
V <sub>os</sub>	125µV	0.6mV
Iq	560µV	10uA
I <sub>b</sub>	1pA	1pA
UGBW	4.5MHz	350kHz
SR	20V/μs	0.2V/us
#Channels	1, 2, 4	1, 2, 4
	www.ti.com/product/OPA2991	www.ti.com/product/TLV9042

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