Analog Engineer's Circuit Bidirectional Current Sensing Circuit With Unidirectional Current Shunt Monitor

TEXAS INSTRUMENTS

Amplifiers

Design Goals

The design goal is the realization of a bidirectional current sensing circuit with a unidirectional current sense amplifier.

Input		Output		Supply		Error	
I _{Min}	I _{Max}	V _{out, min}	V _{out, max}	V _{cc}	V _{ee}	Over input current range	
-10 A	+10 A	About 100 mV	About 4 V	5 V	0 V	≤ 3% at room temperature ≤ 5% from 0°C to 85°C	

Design Description

This design utilizes a unidirectional current shunt monitor to measure a bidirectional load current. This concept only works for low-side current sensing applications. The load current range is from -10 A to +10 A, the output range is from 100 mV to 4 V, and the supply voltages are 5 V and ground.

Design Notes

- 1. Analyze unidirectional current measurement circuit
- Change the circuit to measure bidirectional current. Use 0.1% tolerance resistors for R1 and R2 with a temperature coefficient of 10 ppm. An example, is the TNPW0603e3 resistor from Vishay[®]
- 3. Error calculation over input current and temperature
- 4. As an example, the INA293A1 amplifier is used in this design note, but the concept is valid for many different current shunt monitors with input-bias-current information available and capable of low-side measurement
- 5. Be sure to Kelvin-connect the shunt resistor
- 6. Conduct a 1-point offset calibration at 25°C to compensate for resistor tolerances



Design Steps

1. Unidirectional current measurement standard circuit:



Standard Unidirectional Current Sensing - PSpice® for TI Simulation

Reversing the current in the previous example results in a negative voltage at the IN+ pin of the INA293A1 which will not disturb the part but results in an output voltage close to 0 V.

 Biasing the input voltage of the INA293A1 by 100 mV makes it possible to measure a bidirectional current from –10 A to 10 A with an input voltage from about 0 to 200 mV and an output voltage from close to 0 V to 4 V:



Modified Circuit With Biased Shunt Monitor Input Voltage

Choosing R1 with 2 k Ω results in an error current of I_{error} = (5 V – 0.201 V) / 2 k Ω = 2.4 mA which can be compensated after digitization in μ C SW. To calculate R2, the input bias current of the INA293A1 must be considered. The input bias current into IN+ is taken from the data sheet: At 200 mV the current is typically about 60 μ A (see the red dotted lines in *INA239x1 Input Bias Current vs V*_{SENSE}), but could vary by ±20%.



The simulation in the previous circuit shows the typical value. Because the IN– pin is connected to ground, this current can be ignored.



INA239x1 Input Bias Current vs VSENSE

R2 = 100 mV / (2.4 mA – 60 μ A) = 42.7 Ω . The closest E96 value is 43.2 Ω which results in a slightly higher input voltage of 201.1 mV as the *Error Calculations R1 MAX R2 MIN* simulation shows. Choosing the higher value for R2 will also prevent the input going negative in case a current of –10 A is applied.

3. R1 and R2 are the only additional error contributors compared to a unidirectional measurement. To get the worst-case values for the input voltage of the current shunt monitor in the *Error Calculations R1 MAX R2 MIN* schematics, a DC sweep from –10 A to +10 A is performed using *PSPICE-FOR-TI*.

Design Results

To calculate the worst-case error over temperature, check these two scenarios:

1. R1max in combination with R2min



Error Calculations R1 MAX R2 MIN



Error Calculations: All cases are simulated: I N + 200mV V o l t g e 160mV 120mV 80mV 40mV 0V--2A 10A -10A -8A □ ◊ ⊽ △ ◊ + V(U1:IN+) -6A -4A 4A 6A 8A 2A 0A Input Current

IN+ Voltage vs Input Current

Measurement Table

Measurements (mV)	(1)	(2)	(3)	(4)	(5)	(6)
Max(V(U1:IN+))	201.16	201.36	201.11	201.31	200.99	201.19
Min(V(U1:IN+))	7.01	7.22	6.96	7.17	6.84	7.04

(1) Nominal Values at 0°C

(2) Worst-Case Values at 0°
(3) Nominal Values at 27°C Worst-Case Values at 0°C

Worst-Case Values at 27°C

(4) (5) Nominal Values at 85°C

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(6) Worst-Case Values at 85°C



2. R1min in combination with R2max



R1 MIN R2 MAX

The input-voltage vs. input-current plot looks almost the same as under *R1max in combination with R2min*.

Measurement Table

Measurements (mV)	(1)	(2)	(3)	(4)	(5)	(6)
Max(V(U1:IN+))	201.06	201.25	201.11	201.31	201.23	201.42
Min(V(U1:IN+))	6.90	7.11	6.96	7.17	7.08	7.29

(1) Nominal Values at 0°C

(2) Worst-Case Values at 0°C

(3) Nominal Values at 27°C

(4) Worst-Case Values at 27°C

(5) Nominal Values at 85°C

(6) Worst-Case Values at 85°C



Error at 27°C Room Temperature

Because the difference between the worst-case (column 4) and nominal (column 3) curve is constant, this results in an error caused by R1 and R2 at 27°C of

201.31 mV / 201.11 mV – 1 = 0.1% at +10 A and 7.17 mV / 6.96 mV – 1 = 3% at –10 A.

Error Over Temperature Range

From the measurement tables, the minimum and maximum input voltages at IN+ are:

- -10 A: 6.84 mV and 7.29 mV, highlighted in green in the measurement tables
- +10 A: 200.99 mV and 201.42 mV, highlighted in blue in the measurement tables

Summary of Errors					
Input Current	Error at Room Temperature 27°C	Error from 0°C to 85°C			
-10 A	+3%	+4.47% -1.72%			
+10 A	+0.1%	+0.15% -0.06%			

Because the delta of the worst-case voltages at V_{IN} stays constant over the whole current range, the error becomes smaller with higher input voltage.

 V_{OUT} can be calculated with V_{IN} multiplied by 20, the gain of the INA293A1:

 $V_{out,min}$ = 6.84 mV × 20 = 137 mV and $V_{out,max}$ = 201.42 mV × 20 = 4.03 V which meets the design goal.



Design References

See Analog Engineer's Circuit Cookbooks for TI's comprehensive circuit library.

Download the PSpice files for this circuit - www.ti.com/lit/zip/SBOA529.

For more information on the INA293 device, see the INA293 –4-V to 110-V, 1-MHz, High-Precision Current Sense Amplifier data sheet.

Additional Resources

• PSpice[®] for TI design and simulation tool: *PSPICE-FOR-TI*

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