

Cap Drop Offline Supply for E-Meters

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ABSTRACT

This design idea provides a simple non-isolated AC/DC power supply for low power applications such as smart grid E-meter applications. The design uses a "capacitive-dropper" front-end combined with a LM46000 SIMPLE SWITCHER® buck regulator from Texas Instruments. The circuit provides 3.3 V at a minimum of 50 mA from a line supply of 90 VAC to 265 VAC. Theory of operation as well as design equations and performance results are given.

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1 Introduction

Many times a simple off-line power supply is required for low power applications such as E-meters. Typically, the need is to convert the line voltage to a small DC value such as 3.3 V or 5 V. This can be done with a line frequency power transformer or a complex AC/DC off-line power supply. Both approaches have well known disadvantages of weight, size, and/or complexity. A better option is shown in [Figure 1](#).

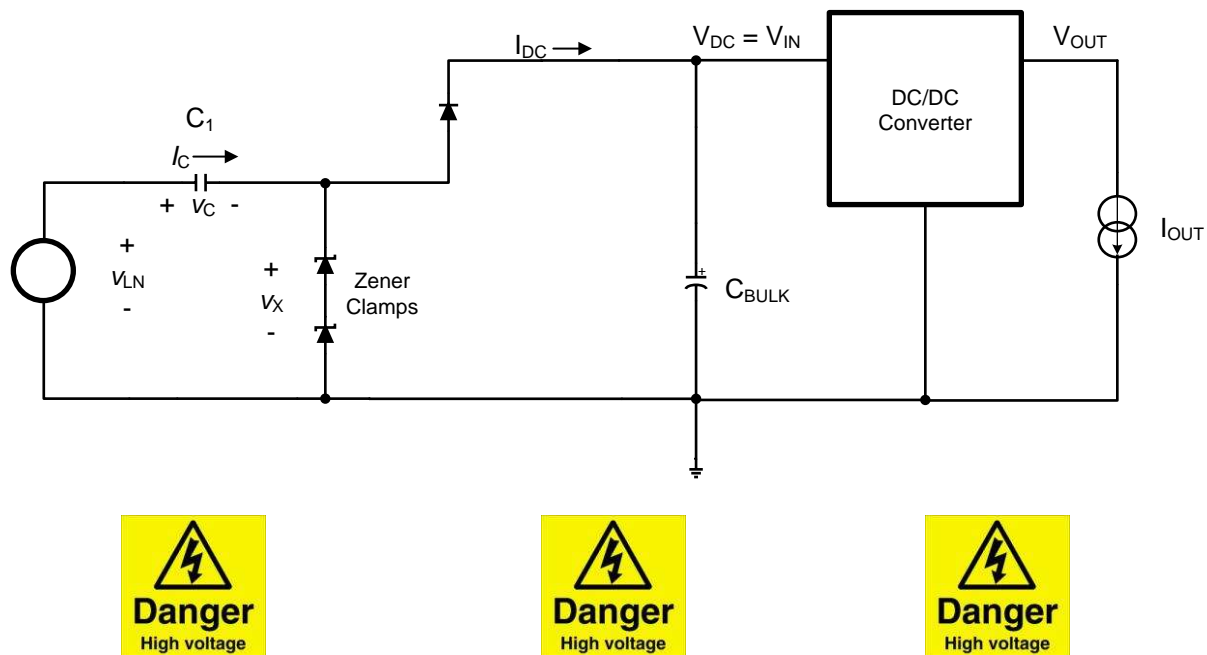


Figure 1. Basic Schematic

Here we first convert the line voltage to an intermediate unregulated DC rail (V_{DC}) and then use a Wide V_{IN} range DC/DC converter to supply the load. The front-end is the well-known "capacitive-dropper". The Zener diodes clamp the input voltage to the DC/DC converter under no-load conditions. The input voltage to the DC/DC converter ($V_{DC} = V_{IN}$) is set to a relatively high value, so that the current required from the "capacitive-dropper" can be kept low. For this design we use the [LM46000](#) as the DC/DC converter and V_{DC} is set to 48 V. The LM46000 converts this down to 3.3 V. For a line voltage range of 90 VAC to 265 VAC, this design can supply at least 50 mA to the 3.3 V load. The high step-down ratio, possible with the LM46000, allows the 50 mA load to appear as less as 5 mA load to the DC/DC. This permits a small value of C_1 to be used. See also [SNVA733](#).

It is easy to see that this circuit is connected directly to the line supply and is not isolated. **EXTREME CAUTION** must be used when experimenting with this design. The user must ensure that the intended application for this power supply, including the load on the LM46000, is completely isolated from any contact with grounded entities; including people, animals and test equipment. All safety precautions must be observed when taking measurements. Test equipment with grounded inputs can not be used with this circuit without proper isolation. The user is also responsible for any fusing, transient protection, and/or EMI filtering required on the input to this circuit.

2 Theory of Operation

The idea behind this circuit is that the series capacitor C_1 acts as a lossless resistance and the reactance of the capacitor will set the maximum current that can be provided. Since a normal electrolytic capacitor cannot handle the stresses resulting from the line voltage, we use "X"-type capacitors which would be rated for the maximum line voltage in our range. From Figure 1 we can understand that the current i_C through the cap C_1 would be flowing when there is a voltage differential across the capacitor. The capacitor current would steadily increase while v_{LN} is increasing. When the line voltage reaches the peak voltage, C_1 stops charging, because the slope of the differential voltage across it goes to zero. Figure 2 shows the relevant waveforms; where we have the following definitions:

v_{LN} = line voltage

v_C = C_1 voltage

i_C = C_1 current

v_X = Voltage at input of bridge rectifier

V_{LN} = RMS line voltage

$V_{DC} = V_{IN}$ = DC intermediate bus voltage and input voltage to DC/DC

V_{OUT} = Output voltage of DC/DC

I_{OUT} = Output current of DC/DC = user load current

V_D = One diode drop

F = line frequency = $1/T$

η = Efficiency of LM46000

T_1 = Time during which the diode is forward biased.

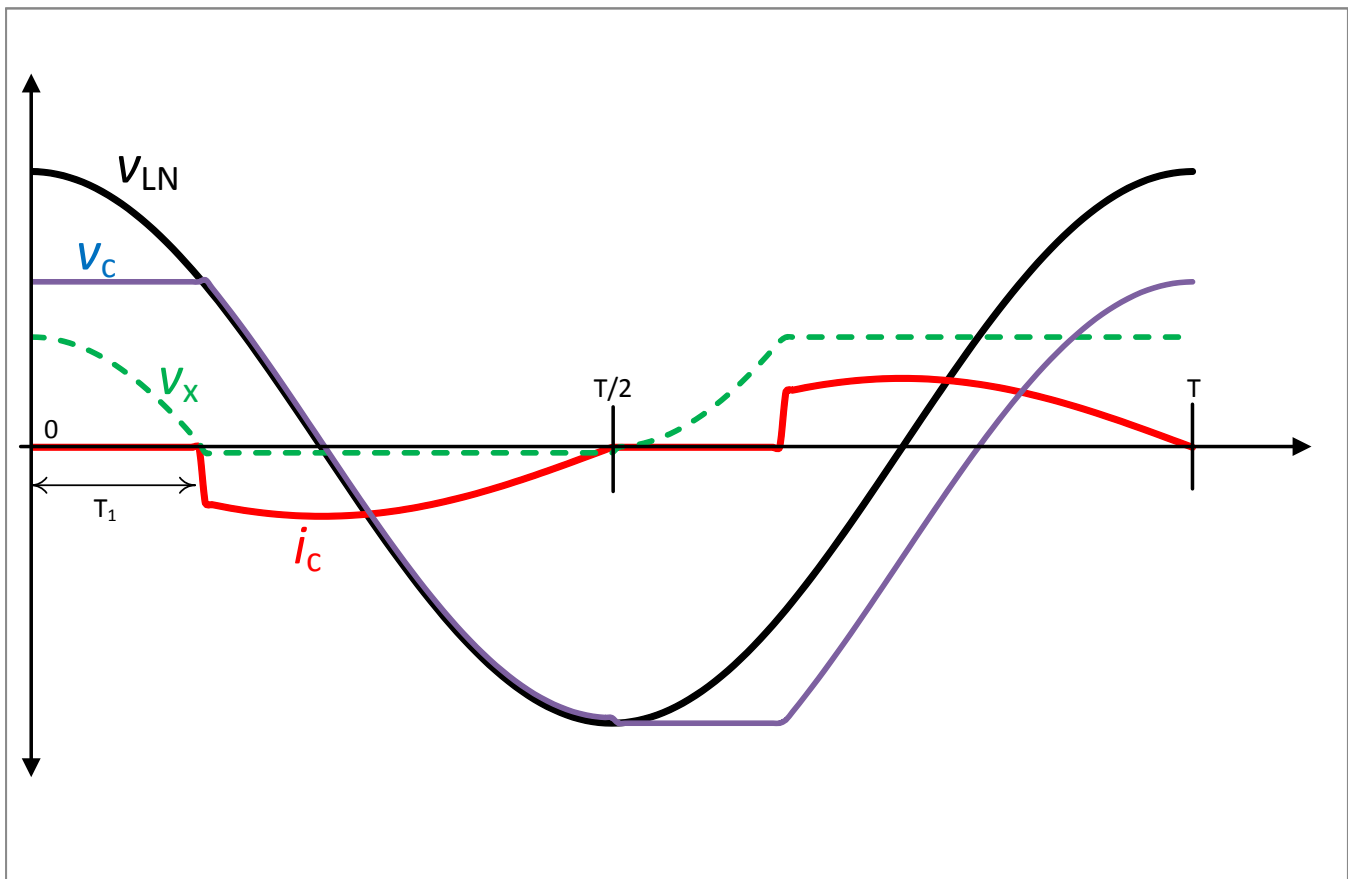


Figure 2. Voltage and Current Waveforms

The peak capacitor current can be obtained as follows:

$$I_P = 2\pi \cdot F \cdot C_1 \cdot V_{LN} \cdot \sqrt{2} \quad (1)$$

Where,

I_P = Peak C_1 current

In the half wave implementation of the capacitive dropper circuit, the diode D1 will be turned off during the negative half cycle of the line voltage. During that time the Zener diode will act as a regular diode and will allow current flow. During the positive half cycle, the Zener diode will clamp the line voltage which is held by the bulk capacitor. Since the negative half cycle of the line voltage is completely ignored, the current delivered to the DC/DC regulator would be half that of the full wave implementation. E-Meter applications using the cap drop implementation have an upper limit on the apparent power being pulled from the mains. This app note uses 8 VA as the upper limit. The approximate value of capacitor C_1 can be obtained from that relationship as shown:

$$C_1 \cong \frac{VA}{2\pi \cdot F \cdot V_{LN}^2}$$

where

- $VA = V_{RMS} \cdot I_{RMS}$ (2)

At time T_1 the Zener conducts current in the opposite direction acting as a regular diode. From observing the waveforms in [Figure 2](#), we can evaluate time T_1 as follows:

$$T_1 = \frac{1}{2\pi \cdot F} \cdot \cos^{-1} \left(1 - \frac{(V_{DC} + V_D)}{V_{LN} \cdot \sqrt{2}} \right) \quad (3)$$

I_{RMS} is the AC current flowing through the C_1 capacitor. Substituting that and using the time T_1 we can calculate a more accurate equation for capacitor C_1 as shown:

$$C_1 = \frac{VA}{2 \cdot \sqrt{2} \cdot F \cdot \pi \cdot V_{LN}^2 \cdot \sqrt{0.5 - T_1 \cdot F + \frac{1}{4\pi} \cdot \sin(4\pi \cdot T_1 \cdot F)}} \quad (4)$$

Since the value of C_1 is limited, the max current that can be delivered to the input of the DC/DC regulator will also be limited. It is shown as follows:

$$I_{MAX} = 2 \cdot \sqrt{2} \cdot F \cdot C_1 \cdot V_{LN} \cdot \left(1 - \frac{V_{DC}}{2 \cdot \sqrt{2} \cdot V_{LN}} \right) \quad (5)$$

3 Application Circuit and Plots

WARNING

CAUTION MUST BE USED IN THE CONSTRUCTION, TESTING, AND USE OF THE CIRCUITS FOUND IN THIS DOCUMENT.

LETHAL VOLTAGES ARE PRESENT IN THESE CIRCUITS THAT MAY CAUSE INJURY.

THE USER MUST ENSURE THAT SAFETY PROCEDURES ARE FOLLOWED WHEN WORKING ON THESE CIRCUITS.

Let's look over some BOM calculations for the capacitor dropping circuit.

3.1 *Dropping Capacitor*

The dropping capacitor C_1 is sized for the lowest line voltage thus ensuring that the load current is maintained even at the worst case. For our design requirements and from [Equation 4](#) the cap C_1 is sized to be about 0.39 μF rated for 375 VAC. Care must be taken to not oversize this capacitor. Oversizing this capacitor would increase the apparent power drawn from the mains. This capacitor must be rated for the highest peak line voltage.

3.2 *Zener Diodes*

In the half wave implementation as shown in the schematic, The LM46000 is rated for a maximum input voltage of 60 V and a load current of 500 mA. Therefore V_{DC} can be clamped at a high voltage of 48 V. The Zener voltage established V_{DC} . As shown in the schematic two Zener diodes of 24 V each have been used in series to obtain a clamping voltage of 48 V. It is important to size the Zener diodes for the right power requirement.

In this implementation, the Zener current will be equal to the I_{MAX} current. At higher line voltages, the I_{MAX} increases and so does the Zener current. The power dissipating in the Zener will be a product of I_{MAX} and V_{DC} .

3.3 *Bulk Capacitor*

A bulk electrolytic capacitor of 680 μF is used to hold the 48 V with low ripple voltage. Keeping the ripple voltage on the intermediate rail low will also help with keeping the output voltage ripple low. Having enough bulk capacitance is also important to maintain enough voltage at the input of converter in case of a fast load transient at the output of the converter. A range of 470 μF to 680 μF was tested to be appropriate. [Figure 3](#) shows the voltage ripple at the input of the LM46000. The 680 μF cap results in about a 200 mV ripple at 60 Hz. Since the ripple at V_{DC} is at a relatively low frequency, it is important to keep the ripple low because it cannot be filtered effectively by the inductor and the output capacitor of the LM46000.

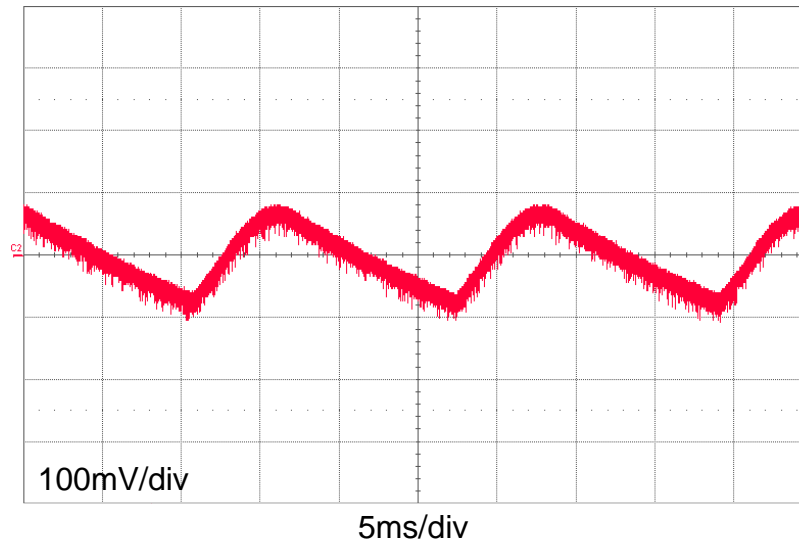


Figure 3. Voltage Ripple at V_{DC}

The newly released LM46000 Wide V_{IN} DC/DC converter was interfaced with the "capacitive drop" front-end to obtain the schematic as shown in Figure 4.

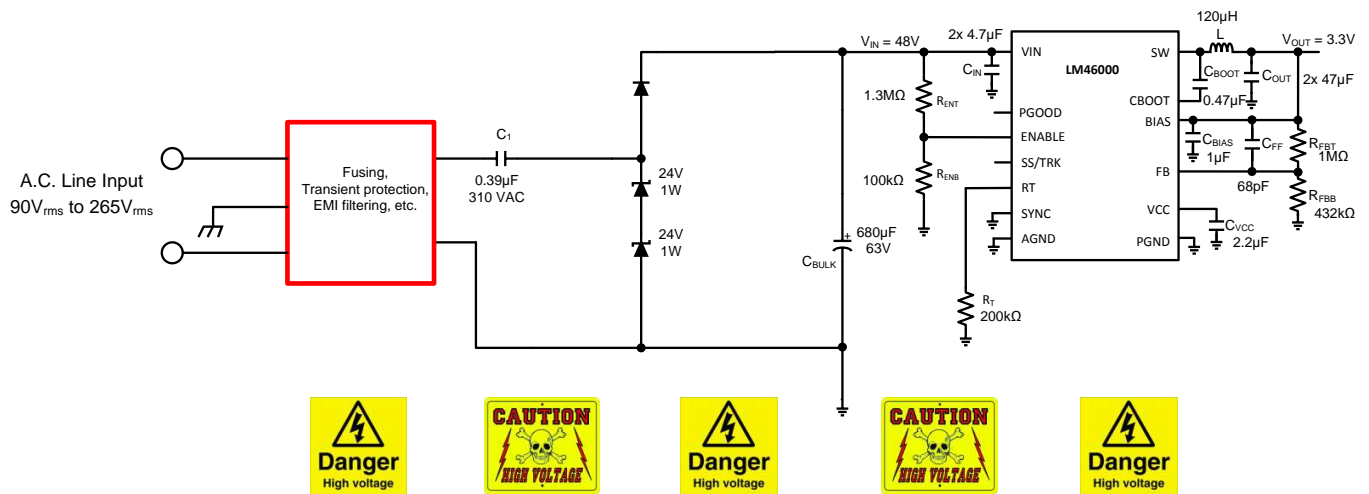


Figure 4. Application Schematic

Table 1. Application Requirements

Parameter	Value
V_{LN}	90 VAC to 265 VAC
V_{OUT}	3.3 V
I_{OUT}	50 mA at 120 VAC
η at 120 VAC and 50 mA	53 %
I_{RMS} from line at 120 VAC	16.5 mA

The BOM for the [LM46000](#) can be calculated for V_{IN} of 48 V to V_{OUT} of 3.3 V. The design can be obtained from the datasheet for LM46000. The datasheet has detailed calculations for the entire BOM. The rising UVLO threshold on the LM46000 was set to about 30 V. This helps with limiting the inrush currents and potential voltage crash at the input of the converter. The resulting falling UVLO threshold is about 25 V.

For the application circuit shown in [Figure 4](#) load regulation test was performed at 120 VAC line voltage. At light loads, the LM46000 enters the PFM mode. In this mode the switching frequency is folded back to improve the efficiency. In PFM operation, a small positive DC offset is required at the output voltage to activate the PFM detector. This can be seen in [Figure 5](#). Please refer to the datasheet for more information.

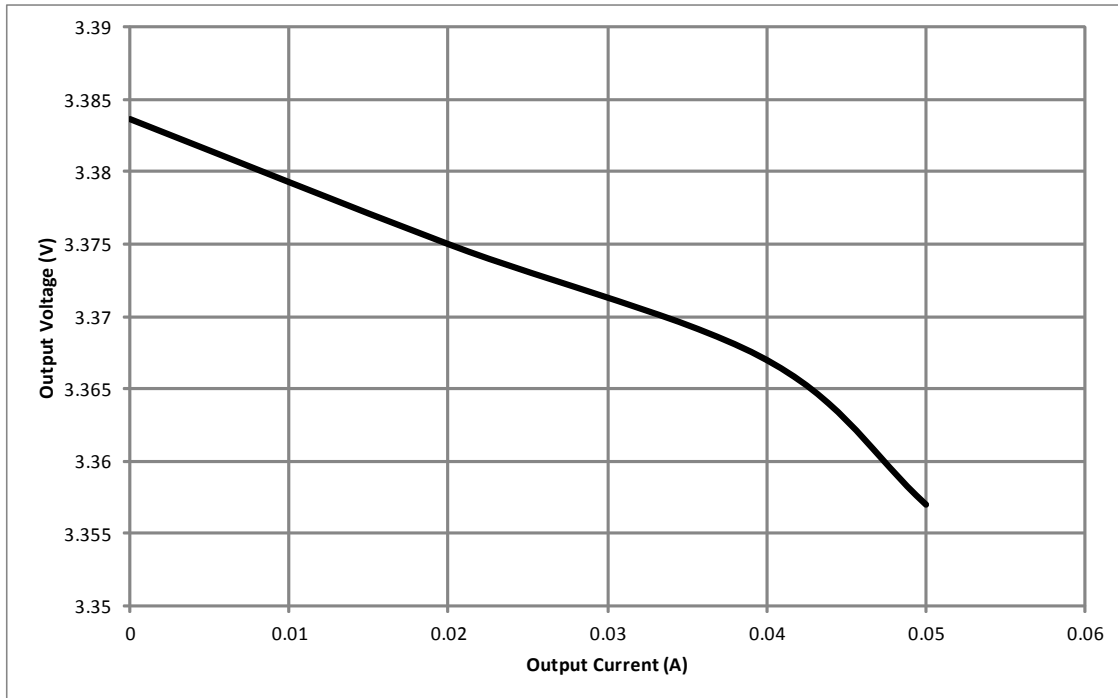


Figure 5. Load Regulation

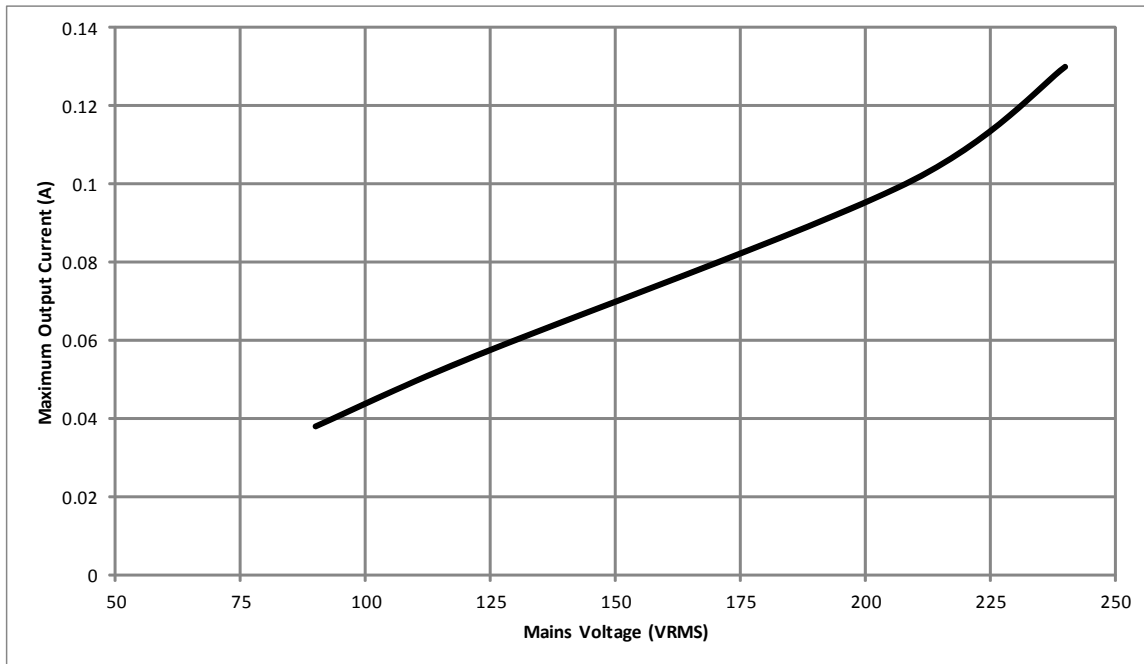


Figure 6. Max Output Current vs Line Voltage

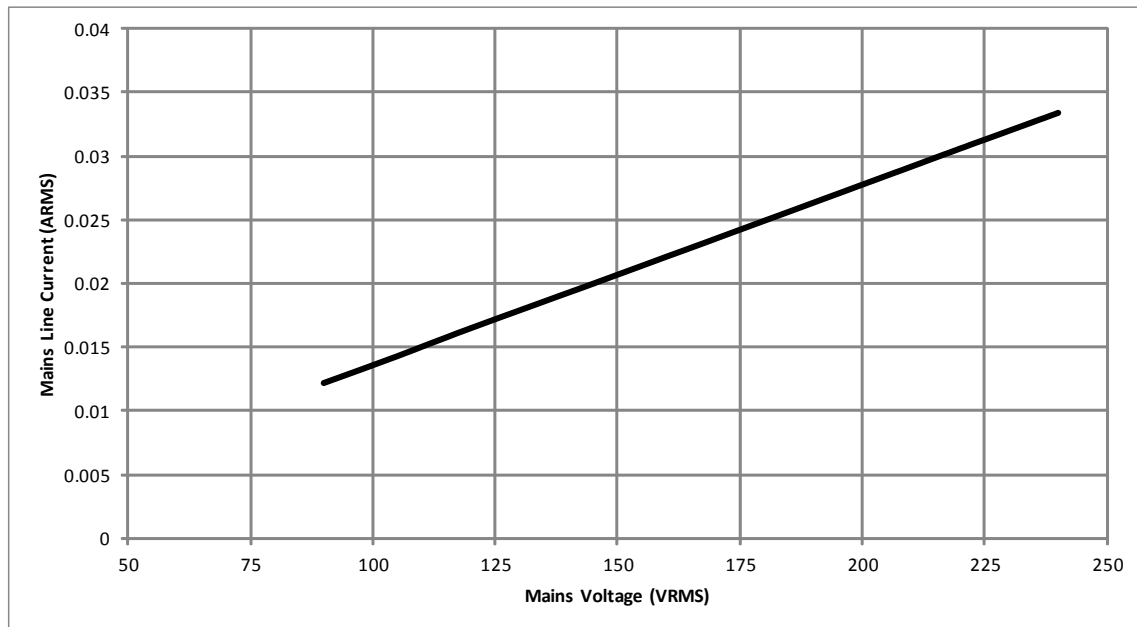


Figure 7. Line Current Vs Line Voltage

Because the capacitor C_1 value is limited the max load that can be pulled is also limited. Figure 6 shows the chart for the max load current capability of the design. Figure 8 shows the actual power and the apparent power drawn from the mains for this design.

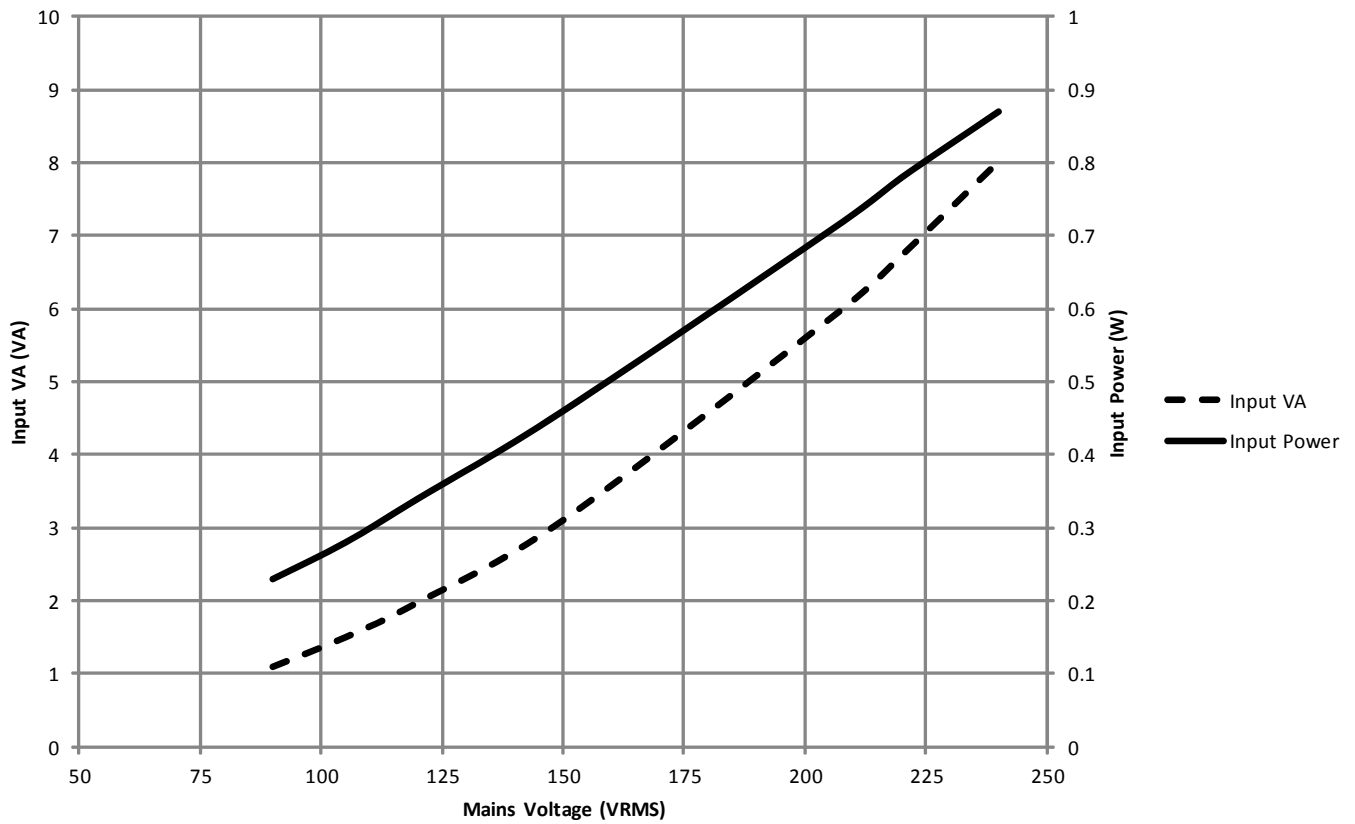


Figure 8. Actual Power and Apparent Power Vs Line Voltage

4 Conclusion

The cap drop circuit is an easy cost effective approach for low load AC-to-DC conversion. Interfacing with a Wide V_{IN} DC/DC converter can be further useful to draw relatively higher loads at the output while keeping the current drawn from the line low. A maximum of 130 mA can be obtained from the output of the LM46000 at 240 VAC line voltage. While this circuit is easy to make, utmost care should be taken to create a bench prototype and appropriate filtering and protection circuit should be added.

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