## AFE20408 8-Channel Power-Amplifier Monitor and Controller

## 1 Features

- Eight analog outputs
- Eight monotonic DACs: 1.22 mV resolution
- Automatically configured output ranges:
- Positive output voltage: 0 V to 10 V
- Negative output voltage: -10 V to 0 V
- High current drive capability
- High capacitive load tolerance
- Output on and off control switches
- Fast switching time
- Low resistance
- Multichannel ADC monitor
- Two high-voltage external inputs: 0 V to 85 V
- Two high-side current-sense amplifiers: up to 85 V common mode range
- Local temperature sensor: $\pm 2.5^{\circ} \mathrm{C}$ error
- Output sequence control for start-up and shutdown events
- Internal 2.5 V reference
- SPI and $\mathrm{I}^{2} \mathrm{C}$ interface: 1.65 V to 3.6 V operation
- SPI: 4-wire interface
- $I^{2} \mathrm{C}: 16$ target addresses
- Specified temperature range: $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$
- Operating temperature range $-55^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$


Simplified Schematic

## 2 Applications

- Radar
- Electronic warfare
- Software defined radio
- Seeker front end


## 3 Description

The AFE20408 is a highly integrated power-amplifier (PA) monitor and control device capable of temperature, current, and voltage supervision.
The AFE20408 bias controller is based around eight digital-to-analog converters (DACs) with programmable output ranges. The eight gate bias outputs are switched on and off through dedicated control pins. The gate bias switches are designed for fast response and enable correct power sequencing and protection of depletion-mode transistors, such as GaAs and GaN.

The AFE20408 supervisor is based around an accurate multichannel analog-to-digital converter (ADC). The device integrates two high-voltage inputs, two high-side current-sense amplifiers and an accurate on-chip temperature sensor.
The function integration and wide operating temperature range make the AFE20408 an excellent choice as an all-in-one, bias control circuit for power amplifiers. The flexible DAC output ranges and builtin sequencing features let the device be used as a biasing controller for a large variety of transistor technologies, such as LDMOS, GaAs, and GaN.

Device Information

| PART NUMBER | PACKAGE $^{(1)}$ | PACKAGE SIZE $^{(2)}$ |
| :--- | :--- | :--- |
| AFE20408 | RHB (VQFN, 32) | $5 \mathrm{~mm} \times 5 \mathrm{~mm}$ |

(1) For more information, see Section 11.
(2) The package size (length $\times$ width) is a nominal value and includes pins, where applicable.

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## 4 Pin Configuration and Functions



Figure 4-1. RHB Package, 32-Pin VQFN (Top View)

Table 4-1. Pin Functions

| PIN |  | TYPE | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| NO. | NAME |  |  |
| 1 | VCCB | Power | Group B output buffers positive analog power supply. |
| 2 | DACB0 | Output | DACB0 buffer output. |
| 3 | OUTB0 | Output | DACB0 switch output. |
| 4 | DACB1 | Output | DACB1 buffer output. |
| 5 | DACB2 | Output | DACB2 buffer output. |
| 6 | OUTB2 | Output | DACB2 switch output. |
| 7 | DACB3 | Output | DACB3 buffer output. |
| 8 | VSSB | Power | Group B output buffers negative analog power supply. |
| 9 | DRVENO | Input | Asynchronous switch control signal. |
| 10 | SDA/SCLK | Input/Output | ${ }^{2} \mathrm{C}$ : bidirectional data line. SPI: Clock input. |
| 11 | SCL/ $\overline{\text { CS }}$ | Input | ${ }^{2} \mathrm{C}$ : Clock input. <br> SPI: Active-low serial data enable. This input is the frame synchronization signal for the serial data. When the signal goes low, this pin enables the serial interface input shift register. |
| 12 | A0/SDI | Input | $I^{2} \mathrm{C}$ : Target address selector. <br> SPI: Data input. Data are clocked into the input shift register on each falling edge of the SCLK pin. |
| 13 | A1/SDO | Input/Output | ${ }^{1} \mathrm{C}$ : Target address selector. <br> SPI: Data output. Data are clocked out of the input shift register on either rising or falling edges of the SCLK pin as specified by the FSDO bit. |
| 14 | VIO | Power | IO supply voltage. This pin sets the I/O operating voltage for the device. |
| 15 | RESET/FLEXIO | Input | Active low reset input. Logic low on this pin causes the device to initiate a reset event. Also referred to as FLEXIO, as this pin can be configured as RESET (default), GPIO, $\overline{A L A R M O U T}$ output, ALARMIN input, DRVEN2, or LDAC. |
| 16 | DRVEN1 | Input | Asynchronous switch control signal. |
| 17 | VSSA | Power | Group A output buffers negative analog power supply. |
| 18 | DACA3 | Output | DACA3 buffer output. |
| 19 | OUTA2 | Output | DACA2 switch output. |
| 20 | DACA2 | Output | DACA2 buffer output. |
| 21 | DACA1 | Output | DACA1 buffer output. |
| 22 | OUTA0 | Output | DACA0 switch output. |
| 23 | DACAO | Output | DACA0 buffer output. |
| 24 | VCCA | Power | Group A output buffers positive analog power supply. |
| 25 | GND | Ground | Ground reference point for all circuitry on the device. |
| 26 | VDD | Power | Analog supply voltage. |
| 27 | PAON | Output | Synchronization signal for PA biasing. |
| 28 | SENSE1- | Input | Current sensor 1 connection. |
| 29 | $\begin{aligned} & \text { SENSE1+/ } \\ & \text { ADCHV1 } \end{aligned}$ | Input | Current sensor 1 connection. Alternatively can be used as a high voltage ADC analog input. |
| 30 | SENSE0- | Input | Current sensor 0 connections. |
| 31 | SENSE0+ | Input | Current sensor 0 connections. |
| 32 | ADCHV0 | Input | High voltage ADC analog input 0 . |
| Thermal Pad | Thermal Pad | - | The thermal pad is located on the package underside. Connect the thermal pad to any internal PCB ground plane through multiple vias for good thermal performance. |

## 5 Specifications

### 5.1 Absolute Maximum Ratings

all ratings over operating free-air temperature range (unless otherwise noted) ${ }^{(1)}$

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $V_{\text {DD }}$ to GND | -0.3 | 6 |  |
|  |  | $\mathrm{V}_{10}$ to GND | -0.3 | 6 |  |
|  | Supply voltage ${ }^{(2)}$ | $\mathrm{V}_{\mathrm{CC}[\mathrm{A}, \mathrm{B}]}$ to GND | -0.3 | 12 | V |
|  |  | $\mathrm{V}_{\text {SS }}$ A,B] to GND | -12 | 0.3 |  |
|  |  | $\mathrm{V}_{C C[A, B]}$ to $\mathrm{V}_{S S[A, B]}$ | -0.3 | 12 |  |
|  |  | DACA[0:3] output pins to GND | $\mathrm{V}_{\text {SSA }}-0.3$ | $\mathrm{V}_{\text {CCA }}+0.3$ |  |
|  |  | DACB[0:3] output pins to GND | $V_{\text {SSB }}-0.3$ | $\mathrm{V}_{\mathrm{CCB}}+0.3$ |  |
|  |  | OUTA $[0,2]$ output pins to GND | $\mathrm{V}_{\text {SSA }}-0.3$ | $\mathrm{V}_{\text {CCA }}+0.3$ |  |
|  | Pin voltage | OUTB[0,2] output pins to GND | $\mathrm{V}_{\text {SSB }}-0.3$ | $\mathrm{V}_{\text {CCB }}+0.3$ | V |
|  | Pinvolage | ADCHV0 and ADCHV1 input pins to GND | -0.3 | 85 |  |
|  |  | Digital pins to GND | -0.3 | $\mathrm{V}_{10}+0.3$ |  |
|  |  | SENSE0[+/-] and SENSE1[+/-] input pins to GND | -0.3 | 85 |  |
|  |  | SENSE[0:1]+ to SENSE[0:1]- differential input ${ }^{(3)}$ | -85 | 85 |  |
| $\mathrm{T}_{\mathrm{J}}$ | Junction tem | rature | -55 | 150 | C |
| $\mathrm{T}_{\text {stg }}$ | Storage tem | rature | -65 | 150 |  |

(1) Operation outside the Absolute Maximum Ratings can cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If used outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.
(2) The device can be configured to operate with mixed range output; that is, positive output on one group of DACs, and negative output range on the other group of DACs. In this case, the supply voltages for the device must be set up such that $\mathrm{V}_{\text {SSA }} \leqq \mathrm{V}_{\text {SSB }}$, and $\mathrm{V}_{\mathrm{CCA}} \leqq$ $\mathrm{V}_{\text {CCB }}$ (that is, DAC group B must operate in positive output range mode, while DAC group A must operate in negative output range mode).
(3) Do not maintain an -85 V differential input over a very extended period of time ( $\geq 5$ years) because doing so can degrade the internal clamp diodes that serve to prevent current from flowing directly from VCC to VSS (that is, to protect the device from short circuiting).

### 5.2 ESD Ratings

|  |  |  | VALUE | UNIT |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {(ESD) }}$ | Electrostatic discharge | Human body model (HBM), per ANSI/ESDA/ JEDEC JS-001, all pins ${ }^{(1)}$ | $\pm 1000$ | V |
|  |  | Charged device model (CDM), per ANSI/ESDA/ JEDEC JS-002, all pins ${ }^{(2)}$ | $\pm 500$ |  |

(1) JEDEC document JEP155 states that 500V HBM allows safe manufacturing with a standard ESD control process.
(2) JEDEC document JEP157 states that 250V CDM allows safe manufacturing with a standard ESD control process.

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### 5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

|  |  | MIN | NOM MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: |
| $V_{\text {DD }}$ | Analog supply voltage | 3.0 | 5.5 | V |
| $\mathrm{V}_{10}$ | Digital IO supply voltage | 1.65 | 3.6 | V |
| $\mathrm{V}_{\mathrm{CC}[\mathrm{A}, \mathrm{B}]}{ }^{(1)}$ | Output buffer positive supply voltage | 3.0 | 11.0 | V |
| $\mathrm{V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}{ }^{(2)}$ | Output buffer negative supply voltage | -11.0 | -3.0 | V |
| $\mathrm{V}_{\text {CCA }}-\mathrm{V}_{\text {SSA }}$ | Group A output buffer supply voltage range | 3.0 | 11.0 | V |
| $\mathrm{V}_{\text {CCB }}-\mathrm{V}_{\text {SSB }}$ | Group B output buffer supply voltage range | 3.0 | 11.0 | V |
| $\mathrm{V}_{\mathrm{CC}[\mathrm{A}, \mathrm{B}]}-\mathrm{V}_{\text {SS[B,A] }}$ | Mixed DAC range supply voltage range | 3.0 | 18.0 | V |
| $\mathrm{V}_{\mathrm{CM}}-\mathrm{V}_{\text {SS[A,B] }}$ | SENSE pins common-mode input range | -0.3 | 85 | V |
| $\mathrm{V}_{\mathrm{IN}}-\mathrm{V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}$ | ADCHV pins input voltage range | 0 | 85 | V |
| $\mathrm{T}_{J}$ | Specified junction temperature | -40 | 125 | ${ }^{\circ} \mathrm{C}$ |
|  | Operating junction temperature | -55 | 150 |  |

(1) $V_{C C[A, B]}$ must be connected to GND when the corresponding DAC group is configured for negative output voltage range operation.
(2) $\mathrm{V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}$ must be connected to GND when the corresponding DAC group is configured for positive output voltage range operation.

### 5.4 Thermal Information

| THERMAL METRIC ${ }^{(1)}$ |  | AFE20408 | UNIT |
| :---: | :---: | :---: | :---: |
|  |  | RHB (VQFN) |  |
|  |  | 32 PINS |  |
| $\theta_{J A}$ | Junction-to-ambient thermal resistance | 34.7 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\theta_{\text {JC(top) }}$ | Junction-to-case (top) thermal resistance | 19.5 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\theta_{\mathrm{JB}}$ | Junction-to-board thermal resistance | 14.6 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\Psi_{\text {JT }}$ | Junction-to-top characterization parameter | 0.3 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\Psi_{\text {JB }}$ | Junction-to-board characterization parameter | 14.5 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\theta_{\mathrm{JC} \text { (bot) }}$ | Junction-to-case (bottom) thermal resistance | 6.7 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |

(1) For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

### 5.5 Electrical Characteristics

all minimum and maximum values at $\mathrm{T}_{J}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$; all typical values at $\mathrm{T}_{J}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{1 \mathrm{O}}=1.65 \mathrm{~V}$ to 3.6 V , positive output range: $\mathrm{V}_{\mathrm{CC}[\mathrm{A}, \mathrm{B}]}=3.0 \mathrm{~V}$ to $11.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}=\mathrm{GND}$, negative output range: $\mathrm{V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}=-11.0 \mathrm{~V}$ to -3.0 V , $\mathrm{V}_{\mathrm{CC}[\mathrm{A}, \mathrm{B}]}=\mathrm{GND}$, and DAC outputs unloaded (unless otherwise noted)


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### 5.5 Electrical Characteristics (continued)

all minimum and maximum values at $\mathrm{T}_{J}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$; all typical values at $\mathrm{T}_{J}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{1 \mathrm{O}}=1.65 \mathrm{~V}$ to 3.6 V , positive output range: $\mathrm{V}_{\mathrm{CC}[\mathrm{A}, \mathrm{B}]}=3.0 \mathrm{~V}$ to $11.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}=\mathrm{GND}$, negative output range: $\mathrm{V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}=-11.0 \mathrm{~V}$ to -3.0 V , $V_{C C[A, B]}=G N D$, and DAC outputs unloaded (unless otherwise noted)

| PARAMETER |  | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OUTPUT SWITCH DC CHARACTERISTICS |  |  |  |  |  |  |
| $\mathrm{R}_{\mathrm{DAC}}$ | Resistance between DAC buffers and output pins | 1.5 V headroom from supply |  | 3 | 5 | $\Omega$ |
| Rvss | Resistance between $\mathrm{V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}$ and output pins |  |  | 4 | 7 | $\Omega$ |

SUPPLY MONITOR CHARACTERISTICS

| $\mathrm{V}_{\text {SSth }}$ | $\mathrm{V}_{\text {Ss }}$ threshold detector | Low $\mathrm{V}_{\text {SS }}$ supply failure detect | -2.2 | -1.7 | V |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Mid $\mathrm{V}_{\text {SS }}$ supply failure detect | -3.7 | -3.2 |  |
|  |  | High $\mathrm{V}_{\text {SS }}$ supply failure detect | -6.7 | -6.2 |  |
| $\mathrm{V}_{\text {cСт }}$ | $\mathrm{V}_{\text {CC }}$ threshold detector | $\mathrm{V}_{\text {CC }}$ supply failure detect | 1.7 | 2.2 | V |

HIGH-VOLTAGE SENSE CHARACTERISTICS

| $\mathrm{V}_{\mathrm{IN}}$ | Analog input range | $\mathrm{V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}=0 \mathrm{~V}$ | 0 | 85 | V |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Offset voltage | $\mathrm{V}_{\mathrm{IN}}=20 \mathrm{mV}$ | $\pm 1.5$ | $\pm 13$ | mV |
|  | Offset voltage ${ }^{(4)}$ | $\mathrm{V}_{\text {IN }}=20 \mathrm{mV}$, operation in extended temperature range ( $\mathrm{T}_{\mathrm{J}}=-55^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ ) | $\pm 5$ | $\pm 40$ | mV |
|  | Offset voltage drift |  | $\pm 20$ |  | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
|  | Gain error |  | $\pm 0.1$ | $\pm 1.0$ | \% |
|  | Gain error ${ }^{(4)}$ | Operation in extended temperature range $\left(\mathrm{T}_{J}=-55^{\circ} \mathrm{C} \text { to }+150^{\circ} \mathrm{C}\right)$ | $\pm 0.3$ | $\pm 3.0$ | \% |
|  | Gain error drift |  | $\pm 20$ |  | ppm/ ${ }^{\circ} \mathrm{C}$ |
|  | Input impedance | Active mode | 1 |  | $\mathrm{M} \Omega$ |
|  | Pin leakage current | Shutdown mode, $\mathrm{V}_{\mathrm{IN}}=85 \mathrm{~V}$ | 1 |  | $\mu \mathrm{A}$ |
|  | Resolution |  | 3.125 |  | mV |

CURRENT SENSE CHARACTERISTICS


### 5.5 Electrical Characteristics (continued)

all minimum and maximum values at $\mathrm{T}_{J}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$; all typical values at $\mathrm{T}_{J}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{1 \mathrm{O}}=1.65 \mathrm{~V}$ to 3.6 V , positive output range: $\mathrm{V}_{\mathrm{CC}[\mathrm{A}, \mathrm{B}]}=3.0 \mathrm{~V}$ to $11.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}=\mathrm{GND}$, negative output range: $\mathrm{V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}=-11.0 \mathrm{~V}$ to -3.0 V , $V_{C C[A, B]}=G N D$, and DAC outputs unloaded (unless otherwise noted)

|  | PARAMETER | TEST CONDITIONS | MIN | TYP MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| DIGITAL INPUT CHARACTERISTICS |  |  |  |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage |  | $0.7 \times \mathrm{V}_{10}$ |  | V |
| $\mathrm{V}_{\mathrm{IL}}$ | Low-level input voltage |  |  | $0.3 \times V_{10}$ | V |
|  | Input current |  |  | 2 | $\mu \mathrm{A}$ |
|  | Input pin capacitance |  |  | 8 | pF |


| $\mathrm{V}_{\mathrm{OH}}$ | High-level output voltage | $I_{\text {SOURCE }}=0.2 \mathrm{~mA}$ | $\mathrm{V}_{10}-0.4$ | V |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {OL }}$ | Low-level output voltage | $\mathrm{I}_{\text {SINK }}=-0.2 \mathrm{~mA}$ | 0.4 | V |
|  | Output pin capacitance |  | 8 | pF |
| VoL | Open-drain low-level output voltage | $\mathrm{I}_{\text {SINK }}=2 \mathrm{~mA}$ | 0.4 | V |
| POWER CONSUMPTION CHARACTERISTICS |  |  |  |  |
| IVDD | $\mathrm{V}_{\mathrm{DD}}$ supply current ${ }^{(4)}$ | Positive output range | 5 | mA |
|  |  | Negative output range | 6 |  |
|  |  | Positive output range, operating in extended temperature range ( $\mathrm{T}_{\mathrm{J}}=-55^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ ) | 10 |  |
|  |  | Negative output range, operating in extended temperature range ( $\mathrm{T}_{\mathrm{J}}=-55^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ ) | 12 |  |
| Ivcc | $\mathrm{V}_{\text {CC }}$ supply current ${ }^{(4)}$ | Positive output range, midscale output | 6 | mA |
|  |  | Positive output range, midscale output, operating in extended temperature range ( $\mathrm{T}_{\mathrm{J}}=-55^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ ) | 12 |  |
| Ivss | $\mathrm{V}_{\text {SS }}$ supply current ${ }^{(4)}$ | Negative output range, $1 / 4$ of full-scale output | 7 | mA |
|  |  | Negative output range, $1 / 4$ of full-scale output, operating in extended temperature range ( $\mathrm{T}_{J}=-55^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ ) | 15 |  |
| $\mathrm{I}_{\mathrm{VIO}}$ | $\mathrm{V}_{10}$ supply current ${ }^{(4)}$ |  | 5 | $\mu \mathrm{A}$ |
|  |  | Operating in extended temperature range ( $\mathrm{T}_{\mathrm{J}}=-55^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ ) | 10 |  |

(1) End point fit between codes 64 to 8128 for negative output range and 64 to 8128 for positive output range.
(2) Overload condition protection. Junction temperature can be exceeded during current limit. Operation greater than the specified maximum junction temperature can impair device reliability.
(3) No continuous oscillations when DAC transitions between codes.
(4) All values provided for extended temperature range $\left(-55^{\circ} \mathrm{C}\right.$ to $\left.+150^{\circ} \mathrm{C}\right)$ are specified by characterization

### 5.6 Timing Requirements

all minimum and maximum values at $\mathrm{T}_{J}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$; all typical values at $\mathrm{T}_{\mathrm{J}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{1 \mathrm{O}}=1.65 \mathrm{~V}$ to 3.6 V , positive output range: $\mathrm{V}_{\mathrm{CC}[\mathrm{A}, \mathrm{B}]}=3.0 \mathrm{~V}$ to $11.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}=\mathrm{GND}$, negative output range: $\mathrm{V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}=-11.0 \mathrm{~V}$ to -3.0 V , $V_{C C[A, B]}=G N D$, and DAC outputs unloaded (unless otherwise noted)


SPI TIMING REQUIREMENTS, FSDO $=0$

| $\mathrm{f}_{\text {(SCLK }}$ | SCLK frequency |  | 20 | MHz |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\text {(SCLKH) }}$ | SCLK high time | 23 |  | ns |
| $\mathrm{t}_{\text {(SCLKL) }}$ | SCLK low time | 23 |  | ns |
| $\mathrm{t}_{\text {(SDIS }}$ | SDI setup time | 7 |  | ns |
| ${ }^{\text {(SDIH }}$ ) | SDI hold time | 7 |  | ns |
| $\mathrm{t}^{\text {(SDOTOZ }}$ ) | SDO driven to tri-state condition | 0 | 17 | ns |
| $\mathrm{t}_{\text {(SDOTOD }}$ | SDO tri-state condition to driven | 0 | 21 | ns |
| $\mathrm{t}_{\text {(SDODLY }}$ | SDO output delay | 0 | 23 | ns |
| $\mathrm{t}_{\text {(CSS }}$ | $\overline{\text { CS setup time }}$ | 21 |  | ns |
| ${ }^{\text {t }}$ (CSH) | $\overline{\mathrm{CS}}$ hold time | 20 |  | ns |
| ${ }^{\text {t }}$ (CSHIGH) | $\overline{\mathrm{CS}}$ high time | 20 |  | ns |

SPI TIMING REQUIREMENTS, FSDO = 1

| $\mathrm{f}_{\text {(SCLK) }}$ | SCLK frequency |  | 25 | MHz |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\text {(SCLKH) }}$ | SCLK high time | 17 |  | ns |
| $\mathrm{t}_{\text {(SCLKL) }}$ | SCLK low time | 17 |  | ns |
| ${ }^{\text {(SDIS }}$ ) | SDI setup time | 7 |  | ns |
| $\mathrm{t}_{\text {(SDIH }}$ | SDI hold time | 7 |  | ns |
| $\mathrm{t}_{\text {(SDOTOZ }}$ | SDO driven to tri-state condition | 0 | 17 | ns |
| $\mathrm{t}_{\text {(SDOTOD }}$ | SDO tri-state condition to driven | 0 | 21 | ns |
| $\mathrm{t}_{\text {(SDODLY }}$ | SDO output delay | 3.5 | 32 | ns |
| $\mathrm{t}_{\text {(CSS }}$ | $\overline{\text { CS setup time }}$ | 21 |  | ns |
| $\mathrm{t}_{(\text {CSH }}$ | $\overline{\mathrm{CS}}$ hold time | 20 |  | ns |
| $\mathrm{t}_{\text {(CSHIGH) }}$ | $\overline{\mathrm{CS}}$ high time | 20 |  | ns |

### 5.7 Switching Characteristics

all minimum and maximum values at $\mathrm{T}_{J}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$; all typical values at $\mathrm{T}_{J}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{1 \mathrm{O}}=1.65 \mathrm{~V}$ to 3.6 V , positive output range: $\mathrm{V}_{\mathrm{CC}[\mathrm{A}, \mathrm{B}]}=3.0 \mathrm{~V}$ to $11.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}=\mathrm{GND}$, negative output range: $\mathrm{V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}=-11.0 \mathrm{~V}$ to -3.0 V , $V_{C C[A, B]}=G N D$, and DAC outputs unloaded (unless otherwise noted).

| PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
| :--- | :--- | :--- | :---: | :---: | :---: |
| OUTPUT SWITCH AC CHARACTERISTICS |  | 400 | ns |  |  |
| ton | On-time digital response time | Midscale code, measured from <br> DRVEN[0,1] trigger | 400 | ns |  |
| tofF | Off-time digital response time | Midscale code, measured from <br> DRVEN[0,1] trigger |  | 4 |  |

## PA_ON CHARACTERISTICS

| tPA_OFF | PA_ON turn-off time | Measured from an ALARMIN alarm event, <br> unloaded | 50 | ns |
| :--- | :--- | :--- | :--- | :---: |

### 5.8 Timing Diagrams



Figure 5-1. $\mathrm{I}^{2} \mathrm{C}$ Timing Diagram


Figure 5-2. SPI Timing Diagram

### 5.9 Typical Characteristics

at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}, \mathrm{~V}_{I O}=3.3 \mathrm{~V}$, negative output range: $\mathrm{V}_{\mathrm{CC}}=\mathrm{GND}, \mathrm{V}_{\mathrm{SS}}=-11 \mathrm{~V}$, and DAC outputs unloaded (unless otherwise noted)


Figure 5-3. DAC DNL vs Digital Input Code


Figure 5-5. DAC TUE vs Digital Input Code


Figure 5-7. DAC INL vs Temperature


Figure 5-4. DAC INL vs Digital Input Code


Figure 5-6. DAC DNL vs Temperature


Figure 5-8. DAC Total Unadjusted Error vs Temperature

### 5.9 Typical Characteristics (continued)

at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}, \mathrm{~V}_{I O}=3.3 \mathrm{~V}$, negative output range: $\mathrm{V}_{\mathrm{CC}}=\mathrm{GND}, \mathrm{V}_{\mathrm{SS}}=-11 \mathrm{~V}$, and DAC outputs unloaded (unless otherwise noted)


### 5.9 Typical Characteristics (continued)

at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}, \mathrm{~V}_{I O}=3.3 \mathrm{~V}$, negative output range: $\mathrm{V}_{\mathrm{CC}}=\mathrm{GND}, \mathrm{V}_{\mathrm{SS}}=-11 \mathrm{~V}$, and DAC outputs unloaded (unless otherwise noted)


Figure 5-15. DAC Headroom vs Normal-Mode Sourcing Current


DAC code $=0 \times 1$ FFF
Figure 5-17. DAC Headroom vs Low-Mode Sourcing Current


DAC code $=0 \times 1000$
Figure 5-19. Source and Sink Current Capability


Figure 5-16. DAC Headroom vs Normal-Mode Sinking Current


DAC code $=0 \times 0000$
Figure 5-18. DAC Headroom vs Low-Mode Sinking Current


DAC step size: -5 V to -2.5 V
Figure 5-20. DAC Settling Time vs Capacitive Load

### 5.9 Typical Characteristics (continued)

at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}, \mathrm{~V}_{I O}=3.3 \mathrm{~V}$, negative output range: $\mathrm{V}_{\mathrm{CC}}=\mathrm{GND}, \mathrm{V}_{\mathrm{SS}}=-11 \mathrm{~V}$, and DAC outputs unloaded (unless otherwise noted)


DAC code $=0 \times 1000$
Figure 5-21. DAC Output Noise, 0.1 Hz to 10 Hz


Figure 5-23. Switch Resistance vs Temperature


Figure 5-25. OUT Pin: DACA1 to VSS Switch Response


DAC code $=0 \times 1000$
Figure 5-22. DAC Output Noise Density vs Frequency


DACAO output: $-2.5 \mathrm{~V} \quad \mathrm{C}_{\mathrm{L}}=1 \mathrm{nF}$ VSS: -7V

Figure 5-24. OUT Pin: DAC to VSS Switch Response


DACAO output: $-2.5 \mathrm{~V} \quad \mathrm{C}_{\mathrm{L}}=1 \mathrm{nF}$
DACA1 output: -7 V
Figure 5-26. OUT Pin: DAC to DAC Switch Response

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### 5.9 Typical Characteristics (continued)

at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}, \mathrm{~V}_{I O}=3.3 \mathrm{~V}$, negative output range: $\mathrm{V}_{\mathrm{CC}}=\mathrm{GND}, \mathrm{V}_{\mathrm{SS}}=-11 \mathrm{~V}$, and DAC outputs unloaded (unless otherwise noted)

$-$
Figure 5-27. ADC Input Offset Error vs Temperature


Figure 5-29. Shunt Offset Error vs Temperature


Figure 5-31. Shunt Input Offset Error vs Common-Mode Voltage


Figure 5-28. ADC Input Gain Error vs Temperature


Figure 5-30. Shunt Gain Error vs Temperature


Figure 5-32. Shunt Input Gain Error vs Common-Mode Voltage

### 5.9 Typical Characteristics (continued)

at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}, \mathrm{~V}_{1 O}=3.3 \mathrm{~V}$, negative output range: $\mathrm{V}_{\mathrm{CC}}=\mathrm{GND}, \mathrm{V}_{\mathrm{SS}}=-11 \mathrm{~V}$, and DAC outputs unloaded (unless otherwise noted)


Figure 5-33. Local Temperature Sensor Error vs Temperature


Figure 5-35. $\mathrm{V}_{\text {SS }}$ AC Power Supply Rejection Ratio


Figure 5-37. $\mathrm{V}_{\mathrm{DD}} \mathrm{AC}$ Power Supply Rejection Ratio


Figure 5-34. Common-Mode Rejection Ratio vs Temperature


Figure 5-36. $\mathrm{V}_{\mathrm{CC}} \mathrm{AC}$ Power Supply Rejection Ratio

$V_{S S}=-10 \mathrm{~V}$
Figure 5-38. $\mathbf{V}_{\text {ss }}$ Supply Collapse Response

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### 5.9 Typical Characteristics (continued)

at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}, \mathrm{~V}_{I O}=3.3 \mathrm{~V}$, negative output range: $\mathrm{V}_{\mathrm{CC}}=\mathrm{GND}, \mathrm{V}_{\mathrm{SS}}=-11 \mathrm{~V}$, and DAC outputs unloaded (unless otherwise noted)


Figure 5-39. $\mathrm{V}_{\mathrm{CC}}$ Supply Collapse Response

$\mathrm{V}_{\mathrm{IO}}=1.8 \mathrm{~V}$
Figure 5-41. $\mathrm{V}_{10}$ Supply Collapse Response

$V_{D D}=5 \mathrm{~V}$
Figure 5-40. $\mathrm{V}_{\mathrm{DD}}$ Supply Collapse Response


Figure 5-42. PAON Supply Response (Device Start-Up, Positive Range)


Figure 5-43. PAON Supply Response (Device Start-Up, Negative Range)

## 6 Detailed Description

### 6.1 Overview

The AFE20408 is an eight-channel power amplifier (PA) controller with dedicated gate bias switch control, as well as voltage, current, and temperature supervision capabilities.

The AFE20408 eight gate bias outputs are separated into two groups, each with four digital-to-analog converters (DACs). The two output groups have dedicated supply inputs, enabling each group to operate with independent output voltage ranges. The eight gate bias outputs can be switched on and off through digital control pins or software. The output switching is designed for fast response and in combination with the device sequencing features enables a robust PA control system.
The AFE20408 external signal supervisor uses an accurate analog-to-digital converter (ADC). The supervisor is capable of monitoring two high-voltage external inputs, two high-common mode current sense inputs, and the device internal temperature. Communication to the device is performed through an SPI and $\mathrm{I}^{2} \mathrm{C}$ compatible interface.

### 6.2 Functional Block Diagram



### 6.3 Feature Description

### 6.3.1 Digital-to-Analog Converter (DAC) Overview

The device features eight analog control channels. Each control channel is centered on a DAC that operates from the device voltage reference. Four of these DACs can be used for setting the internal switches off voltages. The DACs in the device consist of a 13 -bit string DAC and an output voltage buffer. Figure 6-1 shows a block diagram of the DAC architecture.


Figure 6-1. DAC Block Diagram
The DACs can be configured for positive- or negative-output-range operation with identical voltage resolution. In positive-output-range operation, the full-scale range is 0 V to 10 V ; however, the output voltage is limited by $\mathrm{V}_{\mathrm{CC}}$. In negative-output-range operation, the full-scale range is -10 V to 0 V , and the output voltage is limited by $\mathrm{V}_{\mathrm{SS}}$.
After a reset event, all the DAC registers are set to zero-code, the DAC output amplifiers are powered down, and the DAC outputs are clamped to $\mathrm{V}_{\text {Ss }}$. Each DAC can be independently enabled through software, by writing a 1 to the appropriate bit of the PWR_EN register (located in the DAC Configuration page of the register map).

### 6.3.1.1 DAC Resistor String

The resistor string structure consists of a series of resistors, each of value R, as shown in Figure 6-2. The code loaded to the DAC determines at which node on the string the voltage is tapped off to be fed into the output amplifier. The voltage is tapped off by closing one or more of the switches connecting the string to the amplifier. This resistor string architecture has inherent monotonicity, voltage output, and low glitch.


Figure 6-2. DAC Resistor String

### 6.3.1.2 DAC Register Structure

The DAC produces unipolar output voltages proportional to a 13-bit input data code. Input data are written to the DAC data register in straight binary format for all output ranges.
The DAC transfer function is given by:

$$
\begin{equation*}
\mathrm{V}_{\mathrm{DAC}}=\left(\frac{\text { DACIN }}{2^{13}} \times F S R\right)+V_{M I N} \tag{1}
\end{equation*}
$$

where

- DACIN $=$ the decimal equivalent of the binary code loaded to the DAC register. DACIN range $=0$ to $2^{13}-1$.
- FSR = DAC full-scale range for the selected output range. FSR is 10 V for the 0 V to 10 V and -10 V to 0 V ranges.
- $\mathrm{V}_{\text {MIN }}=$ the lowest voltage for the selected DAC output range. Either 0 V for the 0 V to 10 V ranges, or -10 V for the -10 V to 0 V range.
The DAC output spans the voltage ranges shown in Table 6-1.
Table 6-1. DAC Data Format

| DAC DATA REGISTER |  | DAC OUTPUT VOLTAGE (V) |  |
| :---: | :---: | :---: | :---: |
| BINARY | HEX | $\begin{gathered} \hline \text { OV to 10V RANGE } \\ V_{\mathrm{cc}}=11 \mathrm{~V} \\ \mathrm{~V}_{\mathrm{SS}}=\mathrm{GND} \end{gathered}$ | $\begin{gathered} -10 \mathrm{~V} \text { to } 0 \mathrm{~V} \text { RANGE } \\ \mathrm{V}_{\mathrm{ss}}=-11 \mathrm{~V} \\ \mathrm{~V}_{\mathrm{cc}}=\mathrm{GND} \end{gathered}$ |
| 0000000000000000 | 0000 | 0 | -10 |
| 0000000000000001 | 0001 | 0.001221 | -9.998779 |
| 0001000000000000 | 1000 | 5 | -5 |
| 0001111111111110 | 1FFE | 9.997559 | -0.002441 |
| 0001111111111111 | 1FFF | 9.998779 | -0.001221 |

By setting the corresponding BCEN bits in the DAC_SYNC_CFG register (located in the DAC configuration register page), each DAC can be configured to operate in broadcast mode. When a value is written to the DAC_BCAST register (in the global register page), this value is automatically stored in the buffer and active data registers of all DACs operating in broadcast mode. Additionally, a DAC code limit feature is included, which can be used to digitally limit the DAC code to one of 64 different limits. When enabled, a limit is placed on the upper six bits of the DAC code written to the data registers. The limit is only enforced on the DAC active register, and on codes which are written after the DAC code limit has been set to a code less than full scale. The user needs to configure the DAC code limit register, and then subsequent DAC writes are subjected to the currently set DAC code limit. Code limits are specified by writing to the DAC_CODE_LIMIT registers in the DAC Configuration register page (see Section 7.5 for more details).

### 6.3.1.2.1 DAC Synchronous Operation

The update mode for each DAC channel is determined by the value of the corresponding SYNCEN bit in the DAC_SYNC_CFG register (in the DAC Configuration Register page). In asynchronous mode, a write to the DAC buffer data register results in an immediate update of the DAC active register and DAC outputs. In synchronous mode, writing to the DAC data register does not automatically update the DAC output. Instead, the update occurs only after a DAC trigger event. A DAC trigger signal is generated either by setting the DAC_TRIG bit in the TRIGGER register (located in the global register page) or by the FLEXIO pin when configured as LDAC (the LDAC pin can only be used to trigger DACAO and DACA2). The synchronous update mode enables simultaneous update of multiple DAC outputs.

### 6.3.1.3 DAC Buffer Amplifier

The DAC output buffer amplifiers are capable of rail-to-rail operation. The amplifier outputs are available at the DAC output pins. The buffer amplifiers for the two DAC groups are biased from dedicated supply rails: $\mathrm{V}_{\mathrm{CC}[\mathrm{A}, \mathrm{B}]}$ and $\mathrm{V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}$. The maximum DAC group output voltage range is limited by these supplies.
The output amplifier is designed to drive capacitive loads without oscillation. The output buffers are able to source and sink up to 120 mA . The device implements short-circuit protection for momentary output shorts to ground and either supply. The source and sink short-circuit current can be configured to either 30 mA for low-current mode, 90 mA for normal-current mode, or 120 mA for high-current mode. The desired current mode can be set by writing to the DAC_CURRENT register in the DAC Configuration register page.
The high output current of the device gives good slewing characteristics even with large capacitive loads. To estimate the positive and negative slew rates for large capacitive loads, divide the source and sink short-circuit current value by the capacitor.

After start up, the DAC outputs are set automatically into $\mathrm{V}_{\mathrm{SS}}$ clamp mode and the range for each group is configured automatically by the voltage present in the $\mathrm{V}_{\mathrm{SS}}[\mathrm{A}, \mathrm{B}]$ and $\mathrm{V}_{\mathrm{c}}[\mathrm{A}, \mathrm{B}]$ pins. In $\mathrm{V}_{\mathrm{SS}}$ clamp mode, the DAC output pins are internally connected to the $\mathrm{V}_{\mathrm{SS}}[\mathrm{A}, \mathrm{B}]$ pins through a current limited discharge path. The DAC outputs remain in $\mathrm{V}_{\mathrm{SS}}$ clamp mode until the DAC output buffers are powered up through the power enable registers.

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### 6.3.1.3.1 Autorange Detection

The DAC buffer amplifiers are automatically configured at start up for positive voltage operation when $\mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}$ and $3 \mathrm{~V} \leq \mathrm{V}_{\mathrm{CC}} \leq 11 \mathrm{~V}$. Alternatively, the amplifiers are configured at start up for negative voltage operation ( -10 V to 0 V , default) when $\mathrm{V}_{\mathrm{CC}}=0 \mathrm{~V}$ and $-3 \mathrm{~V} \leq \mathrm{V}_{\mathrm{SS}} \leq-11 \mathrm{~V}$. The autorange detector results for each DAC group are stored in the power status register (PWR_STATUS_0).

## Note

The clamp voltage of each DAC group is set by the corresponding VSS[A,B] pin. The autorange detector sets the DAC output range based on the clamp voltage. Make sure that the DAC output range setting matches the DAC clamp voltage. For positive output ranges the VSS[A,B] pins must be connected to GND, in which case the clamp voltage is 0 V . For negative output ranges, the VSS[A,B] pins must be connected to negative supply voltages, in which case the unloaded clamp voltage for each group is determined by the value of the negative supply voltage (see Figure 6-3).


Figure 6-3. DAC Clamp Output vs $\mathrm{V}_{\mathrm{ss}}$

### 6.3.1.3.2 Power-Supply Monitoring

The device continuously monitors the buffer amplifier supplies of each DAC group to provide proper operation. The valid supply range for each DAC group is shown in Table 6-2.

Table 6-2. Valid Supply Matrix

| DAC GROUP SUPPLY CONFIGURATION | SUPPLY |  |
| :---: | :---: | :---: |
|  | VCC[A,B] | VSS[A,B] |
| Invalid configuration | $0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{CC}}<3 \mathrm{~V}$ | $-3 \mathrm{~V}<\mathrm{V}_{\mathrm{SS}} \leq 0 \mathrm{~V}$ |
| $\mathrm{~V}_{\mathrm{CC}}$ configuration | $3 \mathrm{~V} \leq \mathrm{V}_{\mathrm{CC}} \leq 11 \mathrm{~V}$ | $\mathrm{~V}_{\mathrm{SS}}=0 \mathrm{~V}$ |
| Invalid configuration | $3 \mathrm{~V} \leq \mathrm{V}_{\mathrm{CC}} \leq 11 \mathrm{~V}$ | $\mathrm{~V}_{\mathrm{SS}}<0 \mathrm{~V}$ |
| $\mathrm{~V}_{\mathrm{SS}}$ configuration | $\mathrm{V}_{\mathrm{CC}}=0 \mathrm{~V}$ | $-11 \mathrm{~V} \leq \mathrm{V}_{\mathrm{SS}}<-3 \mathrm{~V}$ |
| Invalid configuration | $\mathrm{V}_{\mathrm{CC}}>0 \mathrm{~V}$ | $-11 \mathrm{~V} \leq \mathrm{V}_{\mathrm{SS}}<-3 \mathrm{~V}$ |

During operation, if $\mathrm{V}_{D D}$ drops below 3 V , or $\mathrm{V}_{I O}$ drops below 1.65 V , a power-on reset event is generated, and all DAC outputs return to the $\mathrm{V}_{\mathrm{SS}}$ clamp mode. If $\mathrm{V}_{\mathrm{CC}[\mathrm{A}, \mathrm{B}]}, \mathrm{V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}$ or the internal reference voltage fall below a specified threshold value, there is no power-on reset; however the corresponding alarm bits are activated in the ALARM_STATUS registers (all located within the global register page), which in turn can be used to automatically power down any DAC output.

### 6.3.2 Analog-to-Digital Converter (ADC)

The AFE20408 features a monitoring system centered on a 16 -bit delta-sigma ADC. The ADC measures shunt voltage, bus voltage, and internal temperature. Programmable registers allow flexible configuration for measurement precision as well as automatic or direct operation

### 6.3.2.1 Versatile High-VoItage Measurement Capability

The ADC can measure voltage and current on rails as high as 85 V . Measure the current by sensing the voltage drop across an external shunt resistor at the SENSE+ and SENSE- pins. The voltage drop across the SENSE pins can be measured by the ADC using one of two full-scale range settings. Configure these settings by writing to the SHUNT_RANGE bit in the ADC_GEN_CFG register (located in the ADC configuration register page). The input stage of the ADC is designed such that the input common-mode voltage can be higher than the device supply voltage. The supported common-mode voltage range at the input pins is -0.3 V to +85 V , which makes the device an excellent choice for both high-side and low-side current measurements. There are no special considerations for power-supply sequencing because the common-mode input range and device supply voltage are independent of each other; therefore, the bus voltage can be present with the supply voltage off, and vice-versa without damaging the device.

The device also measures the bus supply voltage through the ADC pins and temperature through the integrated temperature sensor. The differential shunt voltage is measured between the SENSE+ and SENSE- pins, while the bus voltage is measured with respect to device ground. Monitored bus voltages can range from 0 V to 85 V , while monitored temperatures can range from $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$. Figure $6-4$ shows how the shunt voltage, bus voltage, and temperature measurements are multiplexed internally to the ADCs.


Figure 6-4. High Voltage Input Multiplexer

### 6.3.2.2 High-Precision Delta-Sigma ADC

The integrated ADC is a high-performance, low-offset, low-drift, delta-sigma ADC designed to support bidirectional current flow at the shunt voltage measurement channel. The measured inputs are selected through the high-voltage input multiplexer to the ADC inputs. The ADC architecture enables lower drift measurement across temperature and consistent offset measurements across the common-mode voltage, temperature, and power supply variations. A low-offset ADC is preferred in current sensing applications to provide a near OV offset voltage that maximizes the useful dynamic range of the system.
ADC conversion time for each input can be set independently by the CONV_RATE_SENSE, CONV_RATE_ADC, and CONV_RATE_TMP bits in register ADC_CONV_CFG_0 (located in the ADC Configuration register page), in the range of $52 \mu \mathrm{~s}$ to 4.122 ms . Furthermore, a sample averaging function in the range of $1 \times$ to $1024 \times$ is implemented and can be selected by the AVG_SENSE, AVG_ADC, and AVG_TMP bits in ADC_CONV_CFG_1 register (also located in the ADC Configuration register page).

The device can measure the shunt voltage, bus voltage, and die temperature, or a combination of any based on the selected bits setting in the ADC_CCS_IDS registers (described further in Section 6.3.2.2.1). This permits selecting modes to convert only the shunt voltage or bus voltage to further allow the user to configure the monitoring function to fit the specific application requirements. When no averaging is selected, once an ADC conversion is completed, the converted values are independently updated in the corresponding registers and can be read through the digital interface at the time of conversion end. The conversion time for shunt voltage, bus voltage, and temperature inputs are set independently from $52 \mu \mathrm{~s}$ to 4.122 ms depending on the values programmed in the ADC_CONV_CFG_0 register. Enabled measurement inputs are converted sequentially so the total time to convert all inputs depends on the conversion time for each input and the number of inputs enabled. When averaging is used, the intermediate values are subsequently stored in an averaging accumulator, and the conversion sequence repeats until the number of averages is reached. After all of the averaging has been completed, the final values are updated in the corresponding registers that can then be read. These values remain in the data output registers until the next fully completed conversion results are ready. In this case, reading the data output registers does not affect a conversion in progress.

The ADC has two conversion modes-auto and direct-set by the CMODE bit in the ADC_GEN_CFG register. In auto-conversion mode, the ADC can continuously convert the input measurements and update the output registers in an indefinite loop. In direct-conversion mode, the ADC converts the input measurements, after which the ADC goes into shutdown mode until another single-shot trigger is generated by writing to the ADC_TRIG bit in the TRIGGER register (located in the global register page). Writing the ADC_TRIG bit interrupts and restarts auto or direct conversions that are in progress. Although the device can be read at any time, and the data from the last conversion remains available, the ADC not ready flag ( $\overline{A D C}$ _READY bit of the GEN_STATUS register, in the global register page) is provided to help coordinate triggered conversions in direct mode. This bit is cleared after all conversions and averaging are completed.

The data stored in each ADC data register can be manually overridden by a previously specified code, bypassing all ADC data conversions. This code can be set by writing to the ADC_BYP register in the ADC Configuration page of the register map. By writing a 1 to the ADC_BYP_EN bit in the GLOBAL_CFG register (located in the global register page), the codes stored in all ADC data registers are overwritten by the specified bypass value.

### 6.3.2.2.1 ADC Custom Channel Sequencer

The device incorporates an ADC custom channel sequencer. The ADC sequencer is used to specify which channels are converted and in which order. In this way, channels that are of greater importance can be converted more often than other lower-priority channels. The sequencer consists of 63 indexed slots that provide a high level of flexibility in the channel-order configuration. In addition, the sequencer also provides programmable start and stop index fields to configure the start and stop conversion points. In direct-mode conversion, the ADC converts from the start index to the stop index once and then stops. In auto-mode conversion, the ADC converts from the start to stop index repeatedly until the ADC is stopped.

The ADC input channel assignments for the sequencer are listed in Table 6-3.
Table 6-3. Custom Channel Sequence ADC Input
Assignments

| CCS_ID[2:0] | ADC INPUT |
| :---: | :---: |
| 000 | GND |
| 001 | SENSE0 |
| 010 | SENSE1 |
| 011 | ADC0 |
| 100 | ADC1 |
| 101 | TMP |

### 6.3.2.3 Low Latency Digital Filter

The device integrates a low-pass digital filter that performs both decimation and filtering on the ADC output data, which helps with noise reduction. The digital filter is automatically adjusted for the different output data rates and always settles within one conversion cycle. The user has the flexibility to choose different output conversion time periods ( $\mathrm{T}_{\mathrm{CT}}$ ) from $52 \mu \mathrm{~s}$ to 4.122 ms . With this configuration the first amplitude notch appears at the Nyquist frequency of the output signal which is determined by the selected conversion time period and defined as $\mathrm{f}_{\mathrm{NOTCH}}=1 /\left(2 \times \mathrm{T}_{\text {CT }}\right)$. This means that the filter cut-off frequency scales proportionally with the data output rate. Figure 6-5 shows the filter response when the 1.054 ms conversion time period is selected


Figure 6-5. ADC Frequency Response

### 6.3.2.4 Flexible Conversion Times and Averaging

ADC conversion times for shunt voltage, bus voltage and temperature can be set independently from $52 \mu \mathrm{~s}$ to 4.122 ms . The flexibility in conversion time allows for robust operation in a variety of noisy environments. The device also allows for programmable averaging times from a single conversion all the way to an average of 1024 conversions. The amount of averaging selected applies uniformly to all active measurement inputs. Figure 6-6 and Figure 6-7 illustrate the effect of conversion time and averaging on a constant input signal.


### 6.3.2.5 Integrated Precision Oscillator

The internal timebase of the device is provided by an internal oscillator that is trimmed to less than $0.5 \%$ tolerance at room temperature. The precision oscillator is the timing source for ADC conversions, as well as the time-count used for calculation of energy and charge. The digital filter response varies with conversion time; therefore, the precise clock maintains filter response and notch frequency consistency across temperature. At power up, the internal oscillator and ADC take roughly $300 \mu \mathrm{~s}$ to reach $<1 \%$ error stability.

### 6.3.3 Output Switch Overview

The AFE20408 facilitates rapid turn-on and turn-off of the voltage at the device OUT outputs. The OUTO and OUT2 outputs (from groups A and B) can be switched on or off by the DRVEN inputs or alternatively through software. The ON voltages are set by the DAC0 and DAC2 outputs of each respective DAC group, while the OFF voltages are set by either $\mathrm{V}_{\text {SS }}$ or a specified clamp voltage for each DAC. The OUT0 and OUT2 pins are driven by DACO and DAC2 when the corresponding switch control pin or bit is asserted high (drive enabled). Otherwise, the OUT pins are in drive disabled state and driven to either $\mathrm{V}_{\mathrm{SS}}$ or the corresponding clamp DAC.
Additionally, the DAC1 and DAC3 outputs from each group include a simplified switch network that facilitates fast turnoff. The DAC1 and DAC3 pins can be switched on or off, either through one of the DRVEN pins or through software. The DAC1 and DAC3 output pins are driven by the DAC1 and DAC3 buffers when on, and to $\mathrm{V}_{\mathrm{SS}}$ when off. While fast turnoff is possible, turn-on time is limited by the DAC1 and DAC3 buffer bandwidth, and also DAC1 and DAC3 have to exit the power-down state.

The switches are designed to be bidirectional, allowing for two-way current when powered ON and blocking voltage when powered OFF. The switch control is optimized for minimum delay between the DRVEN input and the output pins voltage switching. The switches default to the off (drive disabled) state at start-up or after an alarm event. Along with a $V_{D D}$ supply collapse, there are three additional reset events: a logic low on the RESET pin, a software reset command, or an $I^{2} \mathrm{C}$ general-call reset. All reset events generate a power-down, drive disable sequence. At reset, the DAC and OUT outputs go to $V_{\text {ss }}$.
Figure 6-8 shows the configuration of switching channels in the AFE20408, for both DAC output groups.


Figure 6-8. Switch Block Diagram

### 6.3.4 Drain Switch Control

The AFE20408 device includes an output-control voltage (PAON pin) capable of driving an external MOSFET switch that turns on and off the drain current to a PA FET. The use of this control signal in conjunction with the DAC clamp option allows control of the sequence in which the PA FET is powered up and powered down.
The PAON pin is disabled on start-up. After the device powers on, the PAON pin can be enabled by setting the PAON bit to 1, in the PWR_EN register (located in the global register page). During operation, the status of the PAON pin can be monitored by reading the PAON_STS bit in the GEN_STATUS register (located in the global register page). By default, the PAON pin is pulled to GND, and is in the OFF state.
The maximum output voltage is determined by the voltage at the VDD pin. When enabled, the PAON can be turned off by any alarm generated by the various monitoring circuits in the device, including thermal, supply, ADC, and reference alarms. This configuration is done by writing to the appropriate bits in the PAON_SRC_0 and PAON_SRC_1 registers (located in the general configuration register page).


Figure 6-9. PAON Operation
The PAON pin operates in push-pull mode by default. The PAON pin can be configured to operate in open-drain mode by setting the PAON_ODE bit in the GEN_CFG_0 register (located in the general configuration register page). In push-pull mode, the PAON pin is internally connected to VDD via a pullup resistor. As a result, the PAON pin outputs OV (or the voltage at GND) while in the OFF state, and $V_{D D}$ while in the ON state. In the open drain mode, there is no internal pullup resistor to the VDD pin, and the user must install an external pullup resistor to VDD. This is further described in Section 8.2.2.4.

Additionally, the PAON pin can be configured to invert the ON and OFF states (so that the high voltage is off and the low voltage is on) by setting the PAON_POL bit in the GEN_CFG_0 register (located in the general configuration register page).

For FETs requiring a negative bias voltage, such as GaN , making sure that the bias voltage remains within an acceptable range is crucial; otherwise, significant and irreversible damage to the FET can occur. The AFE20408 bipolar DAC operation and clamping mechanism rely on the $V_{D D}$ and $V_{S S}$ voltages for proper operation. For this reason, when either the $V_{D D}$ or $V_{S S}$ voltage falls outside the acceptable range, turning off the drain current to the FET is desirable.

### 6.3.5 FLEXIO Pin

The AFE20408 features a FLEXIO pin that can be independently configured as a GPIO or a special function pin. The function performed by the FLEXIO pin depends on the value written to the FLEXIO_FUNC field of the GEN_CFG_1 register (located in the General Configuration page of the register map).

On the AFE20408, the following functions are performed by the FLEXIO pin. To enable FLEXIO special functions on any DAC, the corresponding bits in the FLEXIO_EN register (located in the DAC Configuration register page) must be set.

1. RESET: When FLEXIO_FUNC is $0 \times 01$, the FLEXIO pin acts as an active-low external reset. This is the default function of the FLEXIO pin
2. ALARMOUT: When FLEXIO_FUNC is set to $0 \times 02$, the FLEXIO pin acts as an active-low alarm output. The ALARMOUT pin is by default active-low, push-pull, but both the active level and the drive type can be configured by writing to the FLEXIO_OUT_POL and FLEXIO_OUT_ODE bits; see also Section 7.2.1.4. The ALARMOUT_SRC registers (addresses $0 \times 48$ and $0 \times 49$ in the General Configuration register page) are used to configure the alarms that assert the pin; see also Section 7.2.
3. GPIO: When FLEXIO_FUNC is set to 0x04, the FLEXIO pin acts as a GPIO pin. The GPIO acts as an output during write operations, and as an input during read operations. When a GPIO pin acts as an output, the pin state can be set by writing to the GPIO bit in the GPIO_DATA register, located in the global register map. As with the ALARMOUT function, the GPIO output is by default active-low, push-pull, but both the active level and the drive type can be configured by writing to the FLEXIO_OUT_POL and FLEXIO_OUT_ODE bits. When a GPIO pin acts as input, the digital value on the pin is acquired by reading the GPIO_DATA register address. After a power-on reset or any forced reset, all GPIO_DATA bits are reset to 1 .
4. $\overline{\text { LDAC: When FLEXIO_FUNC is set to } 0 \times 08 \text {, the FLEXIO pin acts as a trigger input for DAC outputs DACAO }}$ and DACA2. Specifically, when these DACs are configured to operate in synchronous mode, the active data registers for these DACs only update once the pin has been pulled to logic 0 .
5. ALARMIN: When FLEXIO_FUNC is set to $0 \times 10$, the FLEXIO pin acts as an active-low alarm input pin. On the AFE20408, the ALARMIN inputs can trigger DAC auto-power-down and OUT pin auto-power-down events by setting the appropriate bits in the DAC_APD_SRC and OUT_APD_SRC registers, located in the DAC Configuration register page.
6. DRVEN: When FLEXIO_FUNC is set to $0 \times 20$, the FLEXIO pin acts as an additional switch control input DRVEN2, in addition to the existing DRVEN0 and DRVEN1 pins on the device. By writing to the respective bits in the FLEXIO_EN register (located in the DAC Configuration register map) DRVEN2 control can be enabled on any desired DAC.

### 6.3.6 Internal Temperature Sensor

The device has an on-chip temperature sensor that measures the device die temperature. The normal operating temperature range for the internal temperature sensor is limited by the operating temperature range of the device $\left(-55^{\circ} \mathrm{C}\right.$ to $\left.+150^{\circ} \mathrm{C}\right)$.
The temperature sensor has a resolution of 16 bits $\left(0.0078^{\circ} \mathrm{C}\right)$ and conversion is independent from the device SAR ADC. Temperature data results are represented in binary format, as shown in Table 6-4. The temperature data format allows for representation of negative temperatures using signed 2's complement representation. Make sure to observe the parameter values listed in the Absolute Maximum Ratings table.

Table 6-4. Temperature Data Format (ADC_TMP [15:0])

| TEMPERATURE <br> $\left({ }^{\circ} \mathrm{C}\right)$ | INTERNAL TEMPERATURE REGISTER VALUE |  |
| :---: | :---: | :---: |
|  | BINARY | HEX |
| -55 | 1111111111001001 | FFC9 |
| -25 | 1111111111100111 | FFE7 |
| -1 | 1111111111111111 | FFFF |
| 0 | 0000000000000000 | 0000 |
| 1 | 0000000010000000 | 0080 |
| 10 | 0000010100000000 | 0500 |
| 25 | 0000110010000000 | $0 C 80$ |
| 50 | 0001100100000000 | 1900 |
| 75 | 0010010110000000 | 2580 |
| 100 | 0011001000000000 | 3200 |
| 125 | 0011111010000000 | $3 E 80$ |
| 127 | 0011111110000000 | $3 F 80$ |
| 150 | 0100101100000000 | $4 B 00$ |

The temperature data registers are accessed by reading the ADC_TMP register, located in the global register page.

### 6.3.7 Programmable Out-of-Range Alarms

The AFE20408 is capable of continuously analyzing the supplies, reference, external ADC inputs, and internal temperature for normal operation. Normal operation for the conversion results is established through the lowerand upper-threshold registers. When any of the monitored inputs is out of the specified range, the corresponding alarm bit in the alarm status registers is set. In addition, the global alarm bit (GALR in the GEN_STATUS register) is also set.

All of the alarms can be set to activate the FLEXIO pin, when configured as ALARMOUT. Any alarm event can activate the pin as long as the alarm is not masked in the ALARMOUT_SRC registers. When an alarm event is masked, the occurrence of the event sets the corresponding status bit in the alarm status registers, but does not activate the ALARMOUT pin.

The ALARM_LATCH_DIS bit (located in the GEN_CFG_0 register, part of the General Configuration register page) sets the latching behavior for the internal device alarms, as well as the ALARMOUT pin. When the ALARM_LATCH_DIS bit is cleared to 0, the alarms are latched. The alarms are referred to as being latched because the GALR bit and ALARMOUT pin remain active until the GEN_STATUS register is read by software, even if the alarm condition subsides before the read. This design makes sure that out-of-limit events cannot be missed if the software is polling the device periodically. When the ALARM_LATCH_DIS bit is set to 1 , the alarm bits are not latched. In this case, the GALR bit and ALARMOUT pin are deactivated as soon as the error condition subsides, regardless of whether the GEN_STATUS register is read or not. Regardless of the ALARM_LATCH_DIS bit value, all bits in the alarm status registers are cleared only after a software read. Read the alarm status registers twice to confirm that the bits have cleared after the alarm condition subsides. These bits are reasserted if the out-of limit condition still exists on the next monitoring cycle.

In addition, all of the alarms can be set to force one or more DACs to the power-down state. To enable this functionality, the alarm event must be enabled as a power-down source by writing to the appropriate bits within the DAC_APD_SRC and OUT_APD_SRC registers (all located within the DAC Configuration register page). Additionally, the DAC outputs to be controlled by the alarm event must be specified. In this application, when a DAC control alarm event is detected, all the DACs that are set to power down in response to the alarm do so. When the alarm event is cleared, the DACs are reloaded with the contents of the DAC active registers, which allows the DAC outputs to return to the previous operating point without any additional commands.

### 6.3.7.1 Temperature Sensor Alarm Function

The AFE20408 continuously monitors the internal die temperature. The device includes a thermal error alarm bit (THERMERR_ALR) that is set when the die temperature exceeds $150^{\circ} \mathrm{C}$. A thermal error alarm can be configured to set the ALARMOUT pin, as well as configures all DAC outputs into the power-down state. If a power-down event occurs due to a thermal alarm, the DAC outputs remain in power-down mode even after the device temperature lowers below $150^{\circ} \mathrm{C}$. To resume normal operation, the thermal error alarm must be cleared while the DAC channels are in power-down mode. Apart from the thermal error alarm, the device also features a temperature alarm with a configurable threshold (written to the TMP_UP_THRESH register in the ADC Configuration register page). The TMP_ALR bit, located in the ALARM_STATUS_0 global register, is set when the temperature exceeds the threshold, and can be configured to set the ALARMOUT pin or trigger DAC power-down events.

### 6.3.7.2 Supply Out-of-Range Alarm Function

The AFE20408 is capable of monitoring all power supply voltages, including the internal reference. For VSS and VCC power supply pins, after the voltage supply reaches the power-on threshold, the corresponding bits in the Alarm Status registers are set if the magnitude of voltage at the respective supply pin is less than the supply collapse threshold. Table 6-5 shows the voltage thresholds for power-supply alarm activation.

Table 6-5. Supply Alarm Thresholds

| POWER SUPPLY | ALARM THRESHOLD (POWER-ON) | ALARM THRESHOLD (SUPPLY COLLAPSE) |
| :---: | :---: | :---: |
| VDD | 2.3 V | 2.6 V |
| VCCA/VCCB | 2.2 V | 1.7 V |
| VSSA/VSSB (low-range) | -2.2 V | -1.7 V |
| VSSA/VSSB (mid-range) | -3.7 V | -3.2 V |
| VSSA/VSSB (high-range) | -6.7 V | -6.2 V |

The alarm depends on voltage magnitude (not polarity); therefore, the VSSA and VSSB alarms are set when the respective pin voltages are less negative than the specified supply collapse thresholds. Additionally, the VSSA and VSSB alarm thresholds are determined based on the range selected for the respective DAC group; see also Section 7.2.1.5.

The device provides out-of-range detection for the high performance internal reference. If the internal reference voltage is less than 1.5 V (after initially reaching a power-on threshold of 2.0 V ), the reference alarm flag is set. Verify that the reference alarm condition has not been issued prior to powering up the DAC output buffers.

By setting the appropriate bits in the DAC_APD_SRC and OUT_APD_SRC registers, both the power supply and internal reference alarms can be configured to trigger the alarm pin, a DAC auto power-down event, or both.

### 6.3.7.3 ADC Alarm Function

The device provides independent out-of-range detection for each of the ADC inputs. Figure 6-10 shows the out-of-range detection block. When the measurement is out-of-range, the corresponding alarm bit in the alarm status register is set to flag the out-of-range condition. The values in the ADC high limit and low limit registers define the upper and lower bound thresholds for the ADC inputs.


Figure 6-10. ADC Inputs Out-of-Range Alarms
To prevent false alarms, an alarm event is only registered when the monitored signal is out of range for an $N$ number of consecutive conversions. If the monitored signal returns to the normal range before N consecutive conversions, an alarm event is not issued. The false alarm factor, $N$, for the ADC input alarms can be configured by writing to the FALR_ADC, FALR_SENSE and/or FALR_TMP fields in the ADC_GEN_CFG register (located in the ADC Configuration register page).
If an ADC input signal is out of range and the alarm is enabled, the corresponding alarm bit is set to 1 . However, the alarm condition is cleared only when the conversion result returns either a value lower than the high limit register setting or higher than the low limit register setting by the number of codes specified in the ADC hysteresis setting (see Figure 6-11). The hysteresis for ADC alarms can be set by writing to bits 7 through 0 in the ADC_HYST_0 register. Hysteresis can also be set for the SENSE input alarms, by writing to bits 7 through 0 in the ADC_HYST_1 register. In both these cases, the hysteresis is a programmable value between 0 LSB to 127 LSB.


Figure 6-11. ADC Alarm Hysteresis

### 6.4 Device Functional Modes

The DACs in the AFE20408 are split into groups A and B, each with four DAC channels and two OUT channels. The output range and clamp voltage for each DAC group is set independently, and thus, enables the device to operate in one of the following modes:

- All-positive DAC range mode
- All-negative DAC range mode
- Mixed DAC range mode


### 6.4.1 All-Positive DAC Range Mode

In the AFE20408 all-positive DAC range mode, the two DAC groups are set to a positive-voltage output range ( 0 V to 10 V ).
The minimum DAC output for each group cannot be less than the corresponding $\mathrm{V}_{\mathrm{SS}}$ voltage. In all-positive DAC range mode, the minimum DAC output for both groups is 0 V , and consequently, the VSSA and VSSB pins must be connected to GND.
The maximum DAC output for each group cannot be greater than the corresponding $\mathrm{V}_{\mathrm{CC}}$ voltage. In all-positive DAC range mode, the VCCA and VCCB pins must be connected to a positive supply voltage; however these pins are not required to be tied to the same potential. Typically, the positive voltage at each VCC pin is dictated by the desired positive-voltage output range, but this configuration is not required. In the case where the $\mathrm{V}_{\mathrm{CC}}$ supply voltage for a group is less than the positive full-scale range configuration, the maximum DAC voltage is limited to $\mathrm{V}_{\mathrm{CC}[\mathrm{A}, \mathrm{B}]}$. Table 6-6 lists the typical configurations for this mode.

Table 6-6. All-Positive DAC Range Mode Typical Configuration

| PIN | TEST CONDITIONS | TYPICAL CONNECTION |
| :---: | :---: | :---: |
| VDD |  | 4.5 V to 5.5 V |
| VIO | $\mathrm{V}_{\text {IO }} \leq \mathrm{V}_{\mathrm{DD}}$ | 1.65 V to 5.5 V |
| VCCA | $\mathrm{V}_{\mathrm{DACA}} \leq \mathrm{V}_{\mathrm{CCA}}$ | $3 \mathrm{~V} \leq \mathrm{V}_{\mathrm{CCA}} \leq 11 \mathrm{~V}$ |
| VCCB | $\mathrm{V}_{\mathrm{DACB}} \leq \mathrm{V}_{\mathrm{CCB}}$ | $3 \mathrm{~V} \leq \mathrm{V}_{\mathrm{CCB}} \leq 11 \mathrm{~V}$ |
| VSSA |  | GND |
| VSSB |  | GND |
| Thermal Pad |  | GND |

After a reset event, the output range for each DAC group is automatically set by the autorange detector.
The VSS[A,B] pins set the clamp voltage for each DAC group. The clamp voltage depends only on the voltage in the VSS[A,B] pins; therefore, changes to the DAC range registers do not affect the clamp setting. With both VSSA and VSSB pins connected to GND, the clamp voltage for all DACs is 0 V .

### 6.4.2 All-Negative DAC Range Mode

In the AFE20408 all-negative DAC range mode, the two DAC groups are set to a negative-voltage output range ( -10 V to 0 V ).
The maximum DAC output for each group cannot be less than the corresponding $\mathrm{V}_{\mathrm{CC}}$ voltage. In all-negative DAC range mode, the maximum DAC output for both groups is 0 V , and consequently, the VCCA and VCCB pins must be connected to GND.
The minimum DAC output for each group cannot be less than the corresponding $\mathrm{V}_{\mathrm{SS}}$ voltage. In all-negative DAC range mode, the VSSA and VSSB pins must be connected to a negative supply voltage; however, these pins are not required to be tied to the same potential. Specifically, the voltage at the VSSA pin must always be less than (more negative) or equal to the VSSB voltage. Typically, the negative voltage at each VSS pin is dictated by the desired negative-voltage output range, but this configuration is not required. In the case where the $V_{S S}$ supply voltage for a group is less than the negative full-scale range configuration, the minimum DAC voltage is limited to $\mathrm{V}_{\mathrm{SS}[\mathrm{A}, \mathrm{B}]}$. Table 6-7 lists the typical configurations for this mode.

Table 6-7. All-Negative DAC Range Mode Typical Configuration

| PIN | TEST CONDITIONS | TYPICAL CONNECTION |
| :---: | :---: | :---: |
| VDD |  | 4.5 V to 5.5 V |
| VIO | $\mathrm{V}_{\text {IO }} \leq \mathrm{V}_{\mathrm{DD}}$ | 1.65 V to 5.5 V |
| VCCA |  | GND |
| VCCB |  | GND |
| VSSA | $\mathrm{V}_{\mathrm{DACA}} \geq \mathrm{V}_{\text {SSA }}$ | $-11 \mathrm{~V} \leq \mathrm{V}_{\text {SSA }} \leq-3 \mathrm{~V}$ |
| VSSB | $\mathrm{V}_{\text {DACB }} \geq \mathrm{V}_{\text {SSB }}$ | $-11 \mathrm{~V} \leq \mathrm{V}_{\text {SSB }} \leq-3 \mathrm{~V}$ |
| Thermal Pad |  | GND |

After a reset event, the output range for each DAC group is automatically set by the autorange detector.
The VSS[A,B] pins set the clamp voltage for each DAC group. The clamp voltage depends only on the voltage in the VSS[A,B] pins; therefore, changes to the DAC range registers do not affect the clamp setting. With both VSSA and VSSB pins connected to negative supply voltages, the clamp voltage for DAC group $A$ is equal to $V_{S S A}$, and the clamp voltage for DAC group $B$ is equal to $V_{S S B}$.

### 6.4.3 Mixed DAC Range Mode

In the AFE20408 mixed DAC range mode, DAC group A is set to a negative-voltage output range ( -10 V to 0 V ) and DAC group B is set to a positive-voltage output range ( 0 V to 10 V ). DAC group B cannot be set to a negative-voltage output range if DAC group $A$ is set to a positive-voltage output range.
The VCC pin of DAC group B must be connected to a positive supply voltage. Typically, the positive voltage at the VCC pin is dictated by the desired positive-voltage output range, but this configuration is not required. In the case where the $\mathrm{V}_{\mathrm{CC}}$ supply voltage for the positive-voltage output range group is less than the positive-voltage full-scale-range configuration, the maximum DAC voltage is limited to $\mathrm{V}_{\mathrm{CC}}$. The VSS pin of DAC group B must be connected to GND.

The VSS pin of DAC group A must be connected to a negative supply voltage. Typically, the negative voltage at the VSS pin is dictated by the desired negative-voltage output range, but this configuration is not required. In the case where the $V_{S S}$ supply voltage for the negative output range group is less than the negative full-scale range configuration, the minimum DAC voltage is limited to $\mathrm{V}_{\mathrm{SS}}$. The VCC pin of DAC group A must be connected to GND. Table 6-8 lists the typical configurations for this mode.

Table 6-8. Mixed DAC Range Mode Typical Configuration

| PIN | TEST CONDITIONS | TYPICAL CONNECTION |
| :---: | :---: | :---: |
| VDD |  | 4.5 V to 5.5 V |
| VIO | $\mathrm{V}_{\mathrm{IO}} \leq \mathrm{V}_{\mathrm{DD}}$ | 1.65 V to 5.5 V |
| VCCA |  | GND |
| VCCB | $\mathrm{V}_{\mathrm{SS}} \leq \mathrm{V}_{\mathrm{DACB}} \leq \mathrm{V}_{\mathrm{CC}}$ | $3 \mathrm{~V} \leq \mathrm{V}_{\mathrm{CC}} \leq 11 \mathrm{~V}$ |
| VSSA | $\mathrm{V}_{\mathrm{SS}} \leq \mathrm{V}_{\mathrm{DACA}} \leq \mathrm{V}_{\mathrm{CC}}$ | $-11 \mathrm{~V} \leq \mathrm{V}_{\mathrm{SS}} \leq-3 \mathrm{~V}$ |
| VSSB |  | GND |
| Thermal Pad |  | GND |

The VSS[A,B] pins set the clamp voltage for each DAC group. The clamp voltage depends only on the voltage in the VSS[A,B] pins; therefore, changes to the DAC range registers do not affect the clamp setting. The clamp voltage for $D A C$ group $A$ is equal to $V_{S S A}$, and the clamp voltage for $D A C$ group $B$ is equal to $V_{S S B}$.

### 6.5 Programming

The device communicates with the system controller through a serial interface, which supports either an $I^{2} \mathrm{C}$ compatible two-wire bus, or an SPI-compatible bus. The device includes a robust mechanism that detects between an SPI-compatible or $\mathrm{I}^{2} \mathrm{C}$-compatible controller, and automatically configures the interface accordingly. The interface detection mechanism operates at start-up, thus preventing protocol change during normal operation.
Figure 6-12 shows that the device uses a paging system to organize registers by functionality.


Figure 6-12. Register Page System

In both SPI and $\mathrm{I}^{2} \mathrm{C}$ configurations, address $0 \times 01$ is used to select the different pages in the device. To read and write to one of the device registers, the page for that register must first be selected by writing the 5 -bit representation of the page number (PAGE[4:0]) to address 0x01, as shown in Figure 6-13. The page register holds the page value until a new page address is programmed to the device.


Figure 6-13. Page Access Format
Addresses $0 \times 00$ to $0 \times 3 F$ in each page are global registers, thus enabling access of these bits regardless of the page configuration.

### 6.5.1 $1^{2} \mathrm{C}$ Serial Interface

In $I^{2} \mathrm{C}$ mode, the device operates only as a target device on the two-wire bus. Connections to either bus are made using the open-drain I/O lines, SDA and SCL. The SDA and SCL pins feature integrated spike suppression filters and Schmitt triggers to minimize the effects of input spikes and bus noise. The device supports the transmission protocol for fast mode ( 1 kHz to 400 kHz ). All data bytes are transmitted MSB first.

### 6.5.1.1 $I^{2} C$ Bus Overview

The device is $1^{2} \mathrm{C}$ compatible. In $1^{2} \mathrm{C}$ protocol, the device that initiates the transfer is called a controller, and a device controlled by the controller is called a target. The bus must be controlled by a controller device that generates the serial clock (SCL), controls the bus access, and generates the START and STOP conditions.

To address a specific device, a START condition is initiated. A START condition is indicated by pulling the data line (SDA) from a high-to-low logic level while SCL is high. All targets on the bus receive the target address byte, with the last bit indicating whether a read or write operation is intended. During the ninth clock pulse, the target being addressed responds to the controller by generating an acknowledge bit and pulling SDA low.
Data transfer is then initiated and sent over eight clock pulses followed by an acknowledge bit. During data transfer, SDA must remain stable while SCL is high because any change in SDA while SCL is high is interpreted as a control signal.
After all data have been transferred, the controller generates a STOP condition. A STOP condition is indicated by pulling SDA from low to high, while SCL is high.

### 6.5.1.2 $I^{2} \mathrm{C}$ Bus Definitions

The device is $\mathrm{I}^{2} \mathrm{C}$-compatible and the bus definitions are listed in Table 6-9.
Table 6-9. $I^{2} \mathrm{C}$ Symbol Set

| CONDITION | SYMBOL | SOURCE | DESCRIPTION |
| :---: | :---: | :---: | :--- |
| START | S | Controller | Begins all bus transactions. A change in the state of the SDA line, from <br> high to low, while the SCL line is high, defines a START condition. Each <br> data transfer initiates with a START condition |
| STOP | P | Controller | Terminates all transactions and resets bus. A change in the state of the <br> SDA line from low to high while the SCL line is high defines a STOP <br> condition. Each data transfer terminates with a repeated START or STOP <br> condition. |
| IDLE | I | Controller | Bus idle. Both SDA and SCL lines remain high. |
| ACK (Acknowledge) | A | Controller/Target | Handshaking bit (low). Each receiving device, when addressed, is obliged <br> to generate an acknowledge bit. A device that acknowledges must pull <br> down the SDA line during the acknowledge clock pulse in such a way <br> that the SDA line is stable low during the high period of the acknowledge <br> clock pulse. Take setup and hold times into account. |
| NACK (Not Acknowledge) | A | Controller/Target | Handshaking bit (high). On a controller receive, data transfer termination <br> can be signaled by the controller generating a not-acknowledge on the <br> last byte that has been transmitted by the target. |
| READ | R | Controller | Active-high bit that follows immediately after the target address <br> sequence. Indicates that the controller is initiating the target-to-controller <br> data transfer. The number of data bytes transferred between a START <br> and a STOP condition is not limited and is determined by the controller <br> device. The receiver acknowledges data transfer. |
| WRITE | $\bar{W}$ | Controller | Active-low bit that follows immediately after the target address sequence. <br> Indicates that the controller is initiating the controller-to-target data <br> transfer. The number of data bytes transferred between a START and a <br> STOP condition is not limited and is determined by the controller device. <br> The receiver acknowledges data transfer. |
| REPEATED START | Sr | Controller | Generated by controller, same function as the START condition <br> (highlights the fact that STOP condition is not strictly necessary.) |
| BLOCK ACCESS | B | Controller | Active-high bit that indicates the controller is initiating a block access data <br> transfer. |
| R |  |  |  |

### 6.5.1.3 $I^{2} \mathrm{C}$ Target Address Selection

The $I^{2} \mathrm{C}$ bus target address is selected by installing shunts from the AO and A 1 address pins to the $\mathrm{V}_{10}$ or GND rails. The state of the address pins is tested after every occurrence of START condition on the $I^{2} \mathrm{C}$ bus. The device discerns between four possible options for each pin, shunt to $\mathrm{V}_{10}$ (logic 1), shunt to GND (logic 0), shunt to SDA, and shunt to SCL for a total of sixteen possible target addresses, as shown in Table 6-10.

Table 6-10. $\mathrm{I}^{2} \mathrm{C}$ Target Address Space

| DEVICE PINS |  | I$^{2}$ C TARGET ADDRESS |
| :---: | :---: | :---: |
| A1 | A0 | [A6:A0] |
| 0 | 0 | 1010000 |
| 0 | 1 | 1010001 |
| 0 | SDA | 1010010 |
| 0 | SCL | 1010011 |
| 1 | 0 | 1010100 |
| 1 | 1 | 1010101 |
| 1 | SDA | 1010110 |
| 1 | SCL | 1010111 |
| SDA | 0 | 1011000 |
| SDA | 1 | 1011001 |
| SDA | SDA | 1011010 |
| SDA | SCL | 1011011 |
| SCL | 0 | 1011100 |
| SCL | 1 | 1011101 |
| SCL | SDA | 1011110 |
| SCL | SCL | 1011111 |

### 6.5.1.4 $I^{2} C$ Read and Write Operations

When writing to the device, the value for the address register is the first byte transferred after the target address byte with the $R / \bar{W}$ bit low. Every write operation to the device requires a value for the address register, as shown in Figure 6-14.

| S | TargetAddr[6:0] | $\overline{\mathrm{W}}$ | A | B | RegAddr[6:0] | A | Data[15:8] | A | Data[7:0] | A |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| P |  |  |  |  |  |  |  |  |  |  |

From Controller to Target


Figure 6-14. ${ }^{2}$ C Write Access Protocol
When reading from the device, the last value stored in the address register by a write operation is used to determine which register is read by a read operation. To change which register is read for a read operation, a new value must be written to the address register. This transaction is accomplished by issuing a target address byte with the $R / \bar{W}$ bit low, followed by the address register byte; no additional data are required. The controller can then generate a START condition and send the target address byte with the $R / \bar{W}$ bit high to initiate the read command.

If repeated reads from the same register are desired, there is no need to continually send the address register bytes because the device retains the address register value until the value is changed by the next write operation. The register bytes are big endian and left justified.

Terminate read operations by issuing a not-acknowledge command at the end of the last byte to be read. The controller must leave the SDA line high during the acknowledge time of the last byte that is read from the target, as shown in Figure 6-15.

| S | TargetAddr[6:0] | W | A | B | RegAddr[6:0] | A | Sr | TargetAddr[6:0] | R | A | Data[15:8] | A | Data[7:0] | A | P |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

```
        From Controller to Target
From Target to Controller
```

Figure 6-15. $\mathrm{I}^{\mathbf{2}} \mathrm{C}$ Read Access Protocol
Block access functionality is provided to minimize the transfer overhead of large data sets. Block access enables multibyte transfers and is configured by setting the block access bit high. Until the transaction is terminated by the STOP condition, the device reads and writes the subsequent memory locations, as shown in Figure 6-16 and Figure 6-17. If the controller reaches address $0 x 7 \mathrm{~F}$ in a page, the device continues reading and writing from this address until the transaction is terminated.


Figure 6-16. $1^{2} \mathrm{C}$ Block Write Access


Figure 6-17. $\mathbf{I}^{2}$ C Block Read Access

### 6.5.1.5 $I^{2} C$ Timeout Function

The device resets the serial interface if either SCL or SDA are held low for 25 ms (typical) between a START and STOP condition. If the device is holding the bus low, the device releases the bus and waits for a START condition. After the bus is released, all previously received frames on the bus are discarded by the device, and any previous commands and acknowledgment requests are ignored. To avoid activating the timeout function, maintain a communication speed of at least 1 kHz for the SCL operating frequency. Figure 6-18 shows the logic diagram for the timeout feature, while Figure 6-19 shows the timing diagram.


Figure 6-18. I2C Timeout (Logic Diagram)

SCL (Host)


SCL (Gated)


SDA


Figure 6-19. I2C Timeout (Timing Diagram)

### 6.5.1.6 $I^{2} C$ General-Call Reset

The device supports reset using the two-wire general call address 00h (0000 0000b). The device acknowledges the general-call address, and responds to the second byte. If the second byte is 06 h ( 00000110 b ), the device executes a software reset. This software reset initiates a reset event. The device takes no action in response to other values in the second byte.

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### 6.5.2 Serial Peripheral Interface (SPI)

In SPI mode, the device is controlled through a flexible four-wire serial interface that is compatible with SPI-type interfaces used on many microcontrollers and DSP controllers. The interface provides access to the device registers.

### 6.5.2.1 SPI Bus Overview

A serial interface access cycle is initiated by asserting the $\overline{C S}$ pin low. The serial clock SCLK can be a continuous or gated clock. SDI data are clocked on SCLK falling edges. A regular serial interface access cycle is 24 bits long, thus the $\overline{C S}$ pin must stay low for at least 24 SCLK falling edges. The access cycle ends when the $\overline{\mathrm{CS}}$ pin is deasserted high. If the access cycle contains less than the minimum clock edges, the communication is ignored. If the access cycle contains more than the minimum clock edges, only the last 24 bits are used by the device. When $\overline{\mathrm{CS}}$ is high, the SCLK and SDI signals are blocked and the SDO pin is in a Hi-Z state.

In a serial interface access cycle, the first byte input to SDI is the instruction cycle that identifies the request as a read or write command, and the 7 -bit address to be accessed. The following bits in the cycle form the data cycle, as shown in Table 6-11.

Table 6-11. SPI Serial Interface Access Cycle

| BIT | FIELD | DESCRIPTION |
| :---: | :---: | :--- |
| 23 | RW | Identifies the communication as a read or write command to the addressed register. <br> RW $=0$ sets a write operation. <br> RW $=1$ sets a read operation. |
| $22: 16$ | A[6:0] | Register address. Specifies the register to be accessed during the read or write operation. |
| $15: 0$ | DI[15:0] | Data cycle bits. <br> If a write command, the data cycle bits are the values to be written to the register with address A[6:0]. <br> If a read command, the data cycle bits are don't care values. |

Read operations require that the SDO pin is first enabled by setting the SDO_EN bit. A read operation is initiated by issuing a read command access cycle. After the read command, a second access cycle must be issued to get the requested data, formatted as shown in Table 6-12. Data are clocked out on the SDO pin on SCLK rising or falling edges, according to the FSDO bit setting.

Table 6-12. SDO Output Access Cycle

| BIT | FIELD | DESCRIPTION |
| :---: | :---: | :--- |
| 23 | RW | Echo RW bit from previous access cycle. |
| $22: 16$ | STATUS[6:0] | Lower seven bits of the General Status (GEN_STATUS) register. |
| $15: 0$ | DO[15:0] | Readback data requested on previous access cycle. |

## 7 Register Maps

Table 7-1. Register Section/Block Access Type Codes

| Access Type | Code | Description |
| :--- | :--- | :--- |
| Read Type | R | Read |
| R | R | Read  <br> R-0  <br>   <br>   <br> Write Type  <br> W  <br> Reset or Default Value  <br> $-n$  |

### 7.1 Global Register Map

Table 7-2. Global Page: Global Register Map

| $\begin{aligned} & \text { ADDR } \\ & \text { (HEX) } \end{aligned}$ | REGISTER | TYPE | $\begin{aligned} & \text { RESET } \\ & \text { (HEX) } \end{aligned}$ | BIT DESCRIPTION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 00 | NOP_RESET | R/W | 0000 | SW_RST/NOP [15:0] |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 01 | PAGE | R/W | 0000 | RESERVED |  |  |  |  |  |  |  |  |  |  | PAGE[4:0] |  |  |  |  |
| 03 | GEN_STATUS | R | 4000 | RESERVED |  |  |  | GREFALR | GTHERM_ ALR | $\begin{aligned} & \text { GADC } \\ & \text { ALR } \end{aligned}$ | GSENSE_ ALR | $\frac{\overline{\text { ADC }}}{\text { READY }}$ | RESERVED | GVCCVSS _ALR | RESERVED | GALARM IN_ALR | PAON_STS | $\begin{aligned} & \text { GTMP } \\ & \text { ALR } \end{aligned}$ | GALR |
| 04 | ALARM STATUS_0 | R | 0000 | RESERVED |  |  |  |  |  |  | TMP_ALR | RESERVED |  | $\begin{gathered} \text { ADC1 } \\ \text { ALR } \end{gathered}$ | $\begin{gathered} \text { ADCO- } \\ \text { ALR } \end{gathered}$ | RESERVED |  | $\begin{gathered} \text { SENSE1 } \\ \text { ALR } \end{gathered}$ | $\begin{gathered} \text { SENSEO_ } \\ \text { ALR } \end{gathered}$ |
| 05 | ALARM STATUS_1 | R | 0000 | RESERVED |  | ALARMIN _ALR | $\begin{aligned} & \text { REF- } \\ & \text { ALRR } \end{aligned}$ | $\begin{gathered} \text { THERM } \\ \text { ERR_ALR } \end{gathered}$ | RESERVED |  |  |  |  | $\begin{aligned} & \text { VSSB } \\ & \text { ALR } \end{aligned}$ | $\begin{aligned} & \text { VSSA } \\ & \text { ALR } \end{aligned}$ | RESERVED |  | $\begin{aligned} & \text { VCCB } \\ & \text { ALR } \end{aligned}$ | $\begin{aligned} & \hline \text { VCCA } \\ & \text { ALR } \end{aligned}$ |
| 06 | PWR_STATUS <br> $\begin{array}{l}0\end{array}$ | R | 0001 | $\begin{aligned} & \text { VCCB_ } \\ & \text { STS } \end{aligned}$ |  |  |  | VCCA STS |  |  | VSSA LOW RANGE STS | RESERVED |  |  |  |  |  |  | VDDL_ STS |
| 07 | PWR_STATUS _1 | R | 0000 | DRVEN DACB3_ STS | $\begin{aligned} & \hline \text { DRVEN_ } \\ & \text { DACB2_ } \end{aligned}$ STS | $\begin{aligned} & \hline \text { DRVEN_ } \\ & \text { DACB1- } \end{aligned}$ STS | DRVEN DACBO_ STS | DRVEN DACA3 STS | DRVEN DACA2 STS | DRVEN <br> DACA1_ STS | DRVEN_ DACAO_ STS | PDACB3 STS | $\begin{aligned} & \text { PDACB2_ } \\ & \text { STS } \end{aligned}$ | $\begin{gathered} \text { PDACB1_ } \\ \text { STS } \end{gathered}$ | PDACB0 STS | PDACA3 STS | $\begin{gathered} \text { PDACA2_- } \\ \text { STS } \end{gathered}$ | PDACA1 STS | $\begin{aligned} & \text { PDACAO_ } \\ & \text { STS } \end{aligned}$ |
| 08 | PWR_EN | R/W | 0200 | RESERVED |  |  |  |  |  |  | PAON | PDACB3 | PDACB2 | PDACB1 | PDACBO | PDACA3 | PDACA2 | PDACA1 | PDACAO |
| 10 | TRIGGER | W | 0000 | RESERVED |  |  |  |  |  |  |  |  |  |  |  |  | ALARM_LA TCH_CLR | $\begin{aligned} & \text { DAC- } \\ & \text { TRIG } \end{aligned}$ | $\begin{aligned} & \hline \text { ADC_ } \\ & \text { TRIG } \end{aligned}$ |
| 11 | GPIO_DATA | R/W | 0001 | RESERVED |  |  |  |  |  |  |  |  |  |  |  |  |  |  | GPIO |
| 12 | $\underset{\text { EN }}{\text { DRVEN_SW_ }}$ | R/W | 00FF | RESERVED |  |  |  |  |  |  |  | $\begin{aligned} & \text { DRVEN_ } \\ & \text { SW_EN_ } \end{aligned}$ DACB3 | DRVEN <br> SW_EN <br> DACB2 | $\begin{aligned} & \text { DRVEN_- } \\ & \text { SW_EN- } \\ & \text { DACB1 } \end{aligned}$ | DRVEN <br> SW_EN DACB0 | DRVEN SW EN DACA3 | DRVEN SW EN DACA2 | DRVEN SW_EN DACA1 | DRVEN SW EN DACAO |
| 13 | DRVEN | R/W | 0000 | RESERVED |  |  |  |  |  |  |  | $\begin{aligned} & \text { DRVEN_- } \\ & \text { DACB3 } \end{aligned}$ | DRVEN_ DACB2 | DRVEN DACB1 | DRVEN_ DACBO | DRVEN DACA3 | DRVEN DACA2 | DRVEN DACA1 | DRVEN DACAO |
| 14 | DAC_BCAST | W | 0000 | RESERVED |  |  | DAC[12:0] |  |  |  |  |  |  |  |  |  |  |  |  |
| 17 | GLOBAL_CFG | R/W | 0000 | RESERVED |  |  |  |  |  |  |  |  |  |  |  |  |  | $\begin{gathered} \text { ADC_- } \\ \text { BYP_EN } \end{gathered}$ | ALARM BYP_EN |
| 18 | $\begin{gathered} \hline \text { ADC_-_ } \\ \text { SENSEE_0 } \end{gathered}$ | R | 0000 | ADC[15:0] |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 19 | $\begin{gathered} \text { ADC_-1 } \\ \text { SENSE__ } \end{gathered}$ | R | 0000 | ADC[15:0] |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1A | $\begin{aligned} & \text { ADC_- } \\ & { }_{2} \end{aligned}$ | R | 0000 | ADC[15:0] |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1B | $\begin{aligned} & \text { ADC_ } \\ & \text { ADC_1 } \end{aligned}$ | R | 0000 | ADC[15:0] |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1 C | $\begin{aligned} & \text { ADC- } \\ & \text { TMP } \end{aligned}$ | R | 0000 | ADC[15:0] |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

### 7.1.1 Global Registers: Global Page

### 7.1.1.1 NOP_RESET Register (address $=00 \mathrm{~h}$ ) [reset $=0000 \mathrm{~h}]$

Figure 7-1. NOP_RESET Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SW_RST[15:8]/NOP |  |  |  |  |  |  |  |
| R/W-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| SW_RST[7:0]/NOP |  |  |  |  |  |  |  |
| R/W-Oh |  |  |  |  |  |  |  |

Table 7-3. NOP_RESET Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| 0 | SW_RST/NOP | R/W | Oh | No operation (NOP), unless the data matches a specified value <br> below <br> Ox00AD : Software Reset. Executes a full power-on-reset. <br> Resets the device and all registers to the default power-on-reset <br> state. Auto clears with execution |

### 7.1.1.2 PAGE Register (address $=01 \mathrm{~h})$ [reset $=0000 \mathrm{~h}]$

Figure 7-2. PAGE Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| RESERVED |  |  | PAGE[4:0] |  |  |  |  |
| R-Oh |  |  | R/W-Oh |  |  |  |  |

Table 7-4. PAGE Register Field Descriptions
$\left.\begin{array}{|l|l|l|l|l|}\hline \text { Bit } & \text { Field } & \text { Type } & \text { Reset } & \text { Description } \\ \hline 4-0 & \text { PAGE } & \text { R/W } & \text { Oh } & \begin{array}{l}\text { Sets the page value. See the page map for more details. } \\ \text { Registers on the Global page are accessible from any page, } \\ \text { regardless of the page setting. } \\ \text { ox00: General Configuration Register Page }\end{array} \\ \text { 0x01: ADC Configuration Register Page } \\ \text { 0x02: ADC CCS Configuration Register Page } \\ \text { 0x03: DAC Configuration Register Page } \\ \text { Ox04: DAC Buffer Register Page } \\ \text { Ox06: DAC Active Register Page }\end{array}\right]$

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### 7.1.1.3 GEN_STATUS Register (address = 03h) [reset = 4000h]

Figure 7-3. GEN_STATUS Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  | GREF_ALR | GTHERM_ | GADC_ALR | GSENSE_ |
| R-4h |  |  |  | R-Oh | R-Oh | R-Oh | R-Oh |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| $\frac{\overline{\text { ADC }}}{\overline{\text { READ }}}$ | RESERVED | $\begin{gathered} \text { GVCCVSS_ } \\ \text { ALR } \end{gathered}$ | RESERVED | GALARMIN ALR | PAON_STS | GTMP_ALR | GALR |
| R-Oh | R-Oh | R-Oh | R-Oh | R-Oh | R-Oh | R-Oh | R-Oh |

Table 7-5. GEN_STATUS Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 11 | GREF_ALR | R | Oh | Global reference status bit. This bit is the OR function of all REF alarm status bits. <br> $0=$ No alarm condition <br> 1 = Alarm condition present. |
| 10 | GTHERM_ALR | R | Oh | Global thermal error status bit. This bit is the OR function of all thermal alarm status bits. <br> $0=$ No alarm condition <br> 1 = Alarm condition present. |
| 9 | GADC_ALR | R | Oh | Global ADC status bit for all ADC inputs. This bit is the OR function of all ADC alarm status bits. <br> $0=$ No alarm condition <br> 1 = Alarm condition present. |
| 8 | GSENSE_ALR | R | Oh | Global sense voltage status bit for all SENSE voltage input pins. This bit is the OR function of all SENSE alarm status bits. $0=$ No alarm condition 1 = Alarm condition present. |
| 7 | ADC_READY | R | Oh | ADC ready indicator (active low). <br> $0=A D C$ is ready for trigger to start. <br> 1 = ADC is not ready. |
| 5 | GVCCVSS_ALR | R | Oh | Global VCC or VSS status bit. This bit is the OR function of all VCC and VSS alarm status bits. <br> $0=$ No alarm condition <br> 1 = Alarm condition present. |
| 3 | GALARMIN_ALR | R | Oh | Global ALARMIN status bit. <br> $0=$ No alarm condition <br> 1 = Alarm condition present. |
| 2 | PAON_STS | R | Oh | PAON status bit. Read to confirm whether PAON is active, or turned off. If any alarm event is configured to turn PAON off, this bit follows the latching behavior of the originating alarm (based on the ALARM_LATCH_DIS setting) <br> $0=$ PAON is inactive <br> $1=$ PAON is active |
| 1 | GTMP_ALR | R | Oh | Global temperature sensor status bit. <br> $0=$ No alarm condition <br> 1 = Alarm condition present. |
| 0 | GALR | R | Oh | Global alarm bit. This bit represents the OR function of all individual alarm statuses, and is set to 1 if any alarm condition is present. |

### 7.1.1.4 ALARM_STATUS_0 Register (address = 04h) [reset = 0000h]

Figure 7-4. ALARM_STATUS_0 Register


Table 7-6. ALARM_STATUS_0 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| 8 | TMP_ALR | R | Oh | Out-of-range alarm status for temperature sensor, defined by the <br> corresponding threshold registers. <br> $0=$ Temperature is in the specified range <br> $1=$ Temperature is out-of-range |
| 5 | ADC1_ALR | R | Oh | Out-of-range alarm status for ADC1, defined by the <br> corresponding threshold registers. <br> $0=$ ADC1 channel is in the specified range <br> $1=$ ADC1 channel is out-of-range |
| 4 | ADC0_ALR | R | Oh | Out-of-range alarm status for ADC0, defined by the <br> corresponding threshold registers. <br> $0=$ ADC0 channel is in the specified range <br> $1=$ ADC0 channel is out-of-range |
| 1 | SENSE1_ALR | R | Oh | Out-of-range alarm status for SENSE1, defined by the <br> corresponding threshold registers. <br> $0=$ SENSE1 channel is in the specified range <br> $1=$ SENSE1 channel is out-of-range |
| 0 | SENSE0_ALR | R | Oh | Out-of-range alarm status for SENSE0, defined by the <br> corresponding threshold registers. <br> $0=$ SENSEO channel is in the specified range <br> $1=$ SENSE0 channel is out-of-range |

7.1.1.5 ALARM_STATUS_1 Register (address = 05h) [reset = 0000h]

Figure 7-5. ALARM_STATUS_1 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  | $\begin{gathered} \text { ALARMIN_ } \\ \text { ALR } \end{gathered}$ | REF_ALR | $\underset{\text { ALR }}{\text { THERMERR_ }}$ |  | RESERVED |  |
| R-Oh |  | R-Oh | R-Oh | R-Oh |  | R-Oh |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| RESERVED |  | VSSB_ALR | VSSA_ALR | RESERVED |  | VCCB_ALR | VCCA_ALR |
| R-Oh |  | R-Oh | R-Oh | R-Oh |  | R-Oh | R-Oh |

Table 7-7. ALARM_STATUS_1 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| 13 | ALARMIN_ALR | R | Oh | ALARMIN alarm status. <br> $0=\overline{A L A R M I N ~ h a s ~ n o t ~ t r i g g e r e d . ~}$ <br> $1=\overline{A L A R M I N}$ has triggered. |
| 12 | REF_ALR | R | Oh | Reference alarm status. <br> $0=$ Internal reference voltage is valid <br> $1=$ Internal reference voltage is less than minimum reference <br> threshold voltage. |
| 11 | THERMERR_ALR | R | Oh | Thermal error alarm status. <br> $0=$ Die temperature is less than $150^{\circ} \mathrm{C}$ (typical) <br> $1=$ Operating temperature greater than or equal to $150^{\circ} \mathrm{C}$ |
| 5 | VSSB_ALR | VSSA_ALR | R | Oh |
| 4 | VCCB_ALR | VSSB alarm status. <br> $0=$ VSSB is greater than the minimum VSS threshold voltage <br> $1=$ VSSB is less than the minimum VSS threshold voltage |  |  |
| 1 | R | VSSA alarm status. <br> $0=$ VSSA is less than the minimum VSS threshold voltage <br> $1=$ VSSA is greater than the minimum VSS threshold voltage |  |  |
| 0 | VCCA_ALR | R | VCCB alarm status. <br> $0=$ VCCB is greater than the minimum VCC threshold voltage <br> $1=$ VCCB is less than the minimum VCC threshold voltage |  |

### 7.1.1.6 PWR_STATUS_0 Register (address = 06h) [reset = 0001h]

Figure 7-6. PWR_STATUS_0 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| VCCB_STS | $\begin{gathered} \text { VSSB_ } \\ \text { HIGHRANGE_ } \\ \text { STS } \end{gathered}$ | $\begin{aligned} & \text { VSSB } \\ & \text { MIDRANGE_ } \\ & \text { STS } \end{aligned}$ | $\begin{aligned} & \text { VSSB_ } \\ & \text { LOWRANGE_ } \\ & \text { STS } \end{aligned}$ | VCCA_STS | VSSA HIGHRANGE_ STS | $\qquad$ | $\begin{gathered} \text { VSSA } \\ \text { LOWRANGE } \\ \text { STS } \end{gathered}$ |
| R-Oh | R-Oh | R-Oh | R-Oh | R-Oh | R-Oh | R-Oh | R-Oh |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| RESERVED |  |  |  |  |  |  | VDDL_STS |
| R-Oh |  |  |  |  |  |  | R-1h |

Table 7-8. PWR_STATUS_0 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 15 | VCCB_STS | R | Oh | Supply detection result for VCCB. <br> $0=$ VCCB is less than the minimum VCC threshold voltage <br> $1=$ VCCB has exceeded the minimum VCC threshold voltage |
| 14 | VSSB_HIGHRANGE_STS | R | Oh | Supply detection result for VSSB. <br> $0=\mathrm{VSSB}$ is between V and the high range VSS threshold voltage. <br> $1=$ VSSB has exceeded (is more negative than) the high range VSS threshold voltage |
| 13 | VSSB_MIDRANGE_STS | R | Oh | Supply detection result for VSSB. <br> $0=\mathrm{VSSB}$ is between OV and the mid range VSS threshold voltage. <br> $1=$ VSSB has exceeded (is more negative than) the mid range VSS threshold voltage |
| 12 | VSSB_LOWRANGE_STS | R | Oh | Supply detection result for VSSB. <br> $0=\mathrm{VSSB}$ is between 0 V and the low range VSS threshold voltage. <br> 1 = VSSB has exceeded (is more negative than) the low range VSS threshold voltage |
| 11 | VCCA_STS | R | Oh | Supply detection result for VCCA. <br> $0=$ VCCA is less than the minimum VCC threshold voltage <br> 1 = VCCA has exceeded the minimum VCC threshold voltage. |
| 10 | VSSA_HIGHRANGE_STS | R | Oh | Supply detection result for VSSA. $0=$ VSSA is between OV and the high range VSS threshold voltage. <br> $1=$ VSSA has exceeded (is more negative than) the high range VSS threshold voltage |
| 9 | VSSA_MIDRANGE_STS | R | Oh | Supply detection result for VSSA. $0=$ VSSA is between OV and the mid range VSS threshold voltage. <br> $1=$ VSSA has exceeded (is more negative than) the mid range VSS threshold voltage |
| 8 | VSSA_LOWRANGE_STS | R | Oh | Supply detection result for VSSA. <br> $0=$ VSSA is between $O V$ and the low range VSS threshold voltage. <br> $1=$ VSSA has exceeded (is more negative than) the low range VSS threshold voltage |
| 0 | VDDL_STS | R | 1h | Supply detection result for VDDL. <br> $0=$ VDDL is less than the minimum threshold voltage. <br> 1 = VDDL has exceeded the minimum threshold voltage. |

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### 7.1.1.7 PWR_STATUS_1 Register (address = 07h) [reset $=0000 \mathrm{~h}]$

Figure 7-7. PWR_STATUS_1 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{gathered} \hline \text { DRVEN } \\ \text { DACB3 } \\ \text { STS } \end{gathered}$ | $\begin{gathered} \text { DRVEN_ } \\ \text { DACB2_ } \\ \text { STS } \end{gathered}$ | $\begin{gathered} \text { DRVEN } \\ \text { DACB1- } \\ \text { STS } \end{gathered}$ | $\begin{aligned} & \text { DRVEN_ } \\ & \text { DACB0_ } \\ & \text { STS } \end{aligned}$ | $\begin{gathered} \hline \text { DRVEN } \\ \text { DACA3- } \\ \text { STS } \end{gathered}$ | $\begin{aligned} & \text { DRVEN } \\ & \text { DACA2- } \\ & \text { STS } \end{aligned}$ | DRVEN DACA1_ STS | $\begin{gathered} \text { DRVEN_- } \\ \text { DACAO_ } \\ \text { STS } \end{gathered}$ |
| R-Oh | R-Oh | R-Oh | R-Oh | R-Oh | R-Oh | R-Oh | R-Oh |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| $\begin{gathered} \text { PDACB3_ } \\ \text { STS } \end{gathered}$ | $\begin{gathered} \text { PDACB2 } \\ \text { STS } \end{gathered}$ | $\begin{gathered} \text { PDACB1_ } \\ \text { STS } \end{gathered}$ | $\begin{gathered} \text { PDACB0_ } \\ \text { STS } \end{gathered}$ | $\begin{gathered} \text { PDACA3 } \\ \text { STS } \end{gathered}$ | $\begin{gathered} \text { PDACA2 } \\ \text { STS } \end{gathered}$ | $\begin{gathered} \text { PDACA1 } \\ \text { STS } \end{gathered}$ | $\begin{gathered} \text { PDACAO_ } \\ \text { STS } \end{gathered}$ |
| R-Oh | R-Oh | R-Oh | R-Oh | R-Oh | R-Oh | R-Oh | R-Oh |

Table 7-9. PWR_STATUS_1 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 15 | DRVEN_DACB3_STS | R | Oh | DACB3 drive enable status. <br> 0 : DRVEN = 0 (DACB3 disabled, outputs forced to VSS). <br> 1: DRVEN $=1$. |
| 14 | DRVEN_DACB2_STS | R | Oh | DACB2 drive enable status. <br> 0 : DRVEN = 0 (DACB2 disabled, outputs forced to VSS). <br> 1: DRVEN $=1$. |
| 13 | DRVEN_DACB1_STS | R | Oh | DACB1 drive enable status. <br> 0 : DRVEN = 0 (DACB1 disabled, outputs forced to VSS). <br> 1: DRVEN $=1$. |
| 12 | DRVEN_DACB0_STS | R | Oh | DACB0 drive enable status. <br> 0 : DRVEN = 0 (DACB0 disabled, outputs forced to VSS). <br> 1: DRVEN $=1$. |
| 11 | DRVEN_DACA3_STS | R | Oh | DACA3 drive enable status. <br> 0 : DRVEN = 0 (DACA3 disabled, outputs forced to VSS). <br> 1: DRVEN $=1$. |
| 10 | DRVEN_DACA2_STS | R | Oh | DACA2 drive enable status. <br> 0 : DRVEN = 0 (DACA2 disabled, outputs forced to VSS). <br> 1: DRVEN $=1$. |
| 9 | DRVEN_DACA1_STS | R | Oh | DACA1 drive enable status. <br> 0 : DRVEN = 0 (DACA1 disabled, outputs forced to VSS). <br> 1: DRVEN $=1$. |
| 8 | DRVEN_DACAO_STS | R | Oh | DACAO drive enable status. <br> 0 : DRVEN = 0 (DACAO disabled, outputs forced to VSS). <br> 1: DRVEN $=1$. |
| 7 | PDACB3_STS | R | Oh | DACB3 power status. <br> 0: DACB3 disabled in low-power mode. <br> 1: DACB3 is on. |
| 6 | PDACB2_STS | R | Oh | DACB2 power status. <br> 0: DACB2 disabled in low-power mode. <br> 1: DACB2 is on. |
| 5 | PDACB1_STS | R | Oh | DACB1 power status. <br> 0: DACB1 disabled in low-power mode. <br> 1: DACB1 is on. |
| 4 | PDACB0_STS | R | Oh | DACBO power status. <br> 0: DACB0 disabled in low-power mode. <br> 1: DACB0 is on. |
| 3 | PDACA3_STS | R | Oh | DACA3 power status. <br> 0: DACA3 disabled in low-power mode. <br> 1: DACA3 is on. |
| 2 | PDACA2_STS | R | Oh | DACA2 power status. <br> 0: DACA2 disabled in low-power mode. <br> 1: DACA2 is on. |
| 1 | PDACA1_STS | R | Oh | DACA1 power status. <br> 0: DACA1 disabled in low-power mode. <br> 1: DACA1 is on. |
| 0 | PDACA0_STS | R | Oh | DACAO power status. <br> 0: DACA0 disabled in low-power mode. <br> 1: DACAO is on. |

### 7.1.1.8 PWR_EN Register (address $=08 \mathrm{~h}$ ) [reset $=0200 \mathrm{~h}]$

Figure 7-8. PWR_EN Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  |  | PAON |
| R -1h |  |  |  |  |  |  | R/W-Oh |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| PDACB3 | PDACB2 | PDACB1 | PDACB0 | PDACA3 | PDACA2 | PDACA1 | PDACAO |
| R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh |

Table 7-10. PWR_EN Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 8 | PAON | R/W | Oh | PAON power enable bit. <br> 0 : PAON is not enabled <br> 1 : PAON is enabled from power setting (note that this bit remains at 1 if an alarm event turns PAON off; read PAON_STS in the GEN_STATUS register to confirm if PAON is active or not). |
| 7 | PDACB3 | R/W | Oh | DACB3 enabled status. <br> 0 : DACB3 disabled. <br> 1: DACB3 is enabled. |
| 6 | PDACB2 | R/W | Oh | DACB2 enabled status. <br> 0 : DACB2 disabled. <br> 1: DACB2 is enabled. |
| 5 | PDACB1 | R/W | Oh | DACB1 enabled status. <br> 0 : DACB1 disabled. <br> 1: DACB1 is enabled. |
| 4 | PDACB0 | R/W | Oh | DACB0 enabled status. <br> 0 : DACBO disabled. <br> 1: DACBO is enabled. |
| 3 | PDACA3 | R/W | Oh | DACA3 enabled status. <br> 0 : DACA3 disabled. <br> 1: DACA3 is enabled. |
| 2 | PDACA2 | R/W | Oh | DACA2 enabled status. <br> 0 : DACA2 disabled. <br> 1: DACA2 is enabled. |
| 1 | PDACA1 | R/W | Oh | DACA1 enabled status. <br> 0 : DACA1 disabled. <br> 1: DACA1 is enabled. |
| 0 | PDACA0 | R/W | Oh | DACAO enabled status. <br> 0 : DACAO disabled. <br> 1: DACAO is enabled. |

### 7.1.1.9 TRIGGER Register (address $=10 \mathrm{~h})$ [reset $=0000 \mathrm{~h}]$

Figure 7-9. TRIGGER Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  |  | RESERVED |  |  | ALARM_LATCH _CLR | DAC_TRIG | ADC_TRIG |
| R-Oh |  |  |  |  | W-Oh | W-Oh | W-Oh |

Table 7-11. TRIGGER Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| 2 | ALARM_LATCH_CLR | W | Oh | Manually clear registers which are latching the alarm condition. <br> If an alarm condition is still present, the corresponding alarm <br> latches and alarm state are set again. This bit self-clears <br> $0=$ No action. <br> $1=$ Clear alarm bits. |
| 1 | DAC_TRIG | W | Oh | Software LDAC trigger. This bit self-clears. <br> $0=$ No action. <br> $1=$ Initiate data transfer from DAC buffer registers to active <br> registers. |
| 0 | ADC_TRIG | W | Oh | ADC conversion trigger. Set this bit to 1 to start the ADC <br> conversions. In direct-mode, this bit self-clears back to 0 after <br> all conversions are completed. In auto-mode, this bit remains <br> set and the ADC continuously converts until the user manually <br> clears the bit back to 0, stopping auto-mode. Before setting <br> ADC_TRIG to 1, confirm the ADC is ready by reading the |
| ADC_READY status bit as 0 twice in succession. <br> $0=$ Stop ADC conversions. <br> $1=$ Start ADC conversions |  |  |  |  |

### 7.1.1.10 GPIO_DATA Register (address $=11 \mathrm{~h})$ [reset $=0001 \mathrm{~h}]$

Figure 7-10. GPIO_DATA Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| RESERVED |  |  |  |  |  |  | GPIO |
| R-Oh R/W-1h |  |  |  |  |  |  |  |

Table 7-12. GPIO_DATA Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 0 | GPIO | R/W | 1h | For read operations the GPIO pin operates as an input. Read to receive the status of the corresponding GPIO pin. For write operations, the GPIO pin operates as an output based on the value written to this register, as follows <br> 0 : Set GPIO to logic low. <br> 1: Set GPIO to logic high (when FLEXIO_OUT_ODE $=0$ ) or a high impedance state (when FLEXIO_OUT_ODE = 1) |

### 7.1.1.11 DRVEN_SW_EN Register (address = 12h) [reset = 00FFh]

Figure 7-11. DRVEN_SW_EN Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  |  |  |
| R -Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| DRVEN <br> SW EN <br> DACB3 | DRVEN <br> SW EN <br> DACB2 | DRVEN <br> SW EN <br> DACB1 | DRVEN <br> SW EN <br> DACBO | DRVEN SW EN DACA3 | DRVEN SW EN DACA2 | DRVEN SW EN DACA1 | DRVEN SW EN DACAO |
| R/W-1h | R/W-1h | R/W-1h | R/W-1h | R/W-1h | R/W-1h | R/W-1h | R/W-1h |

Table 7-13. DRVEN_SW_EN Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 7 | DRVEN_SW_EN_DACB3 | R/W | 1h | Enables software operation of DACB3 channel switch (DRVEN). <br> 0 : Software control disabled. <br> 1: Software control enabled. |
| 6 | DRVEN_SW_EN_DACB2 | R/W | 1h | Enables software operation of DACB2 channel switch (DRVEN). <br> 0 : Software control disabled. <br> 1: Software control enabled. |
| 5 | DRVEN_SW_EN_DACB1 | R/W | 1h | Enables software operation of DACB1 channel switch (DRVEN). <br> 0 : Software control disabled. <br> 1: Software control enabled. |
| 4 | DRVEN_SW_EN_DACBO | R/W | 1h | Enables software operation of DACBO channel switch (DRVEN). <br> 0 : Software control disabled. <br> 1: Software control enabled. |
| 3 | DRVEN_SW_EN_DACA3 | R/W | 1h | Enables software operation of DACA3 channel switch (DRVEN). <br> 0 : Software control disabled. <br> 1: Software control enabled. |
| 2 | DRVEN_SW_EN_DACA2 | R/W | 1h | Enables software operation of DACA2 channel switch (DRVEN). <br> 0 : Software control disabled. <br> 1: Software control enabled. |
| 1 | DRVEN_SW_EN_DACA1 | R/W | 1h | Enables software operation of DACA1 channel switch (DRVEN). <br> 0: Software control disabled. <br> 1: Software control enabled. |
| 0 | DRVEN_SW_EN_DACAO | R/W | 1h | Enables software operation of DACAO channel switch (DRVEN). <br> 0: Software control disabled. <br> 1: Software control enabled. |

### 7.1.1.12 DRVEN Register (address = 13h) [reset = 0000h]

Figure 7-12. DRVEN Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  |  |  |
| R -Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| DRVEN <br> DACB3 | DRVEN DACB2 | DRVEN <br> DACB1 | $\begin{aligned} & \hline \text { DRVEN } \\ & \text { DACBO } \end{aligned}$ | DRVEN DACA3 | DRVEN DACA2 | DRVEN DACA1 | DRVEN <br> DACAO |
| R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-0h | R/W-Oh | R/W-0h | R/W-Oh |

Table 7-14. DRVEN Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 7 | DRVEN_DACB3 | R/W | Oh | Software drive enable value, when software control is enabled on DRVEN_SW_EN_DACB3. <br> 0 : DACB3 drive disabled, internally connected to VSSB. <br> 1: DACB3 drive enabled. |
| 6 | DRVEN_DACB2 | R/W | Oh | Software drive enable value, when software control is enabled on DRVEN_SW_EN_DACB2. <br> 0: OUTB2 drive disabled, internally connected to VSSB or DACB3, depending on CLAMP_SEL_OUTB2. <br> 1: OUTB2 drive enabled, connected to DACB2. |
| 5 | DRVEN_DACB1 | R/W | Oh | Software drive enable value, when software control is enabled on DRVEN_SW_EN_DACB1. <br> 0 : DACB1 drive disabled, internally connected to VSSB. <br> 1: DACB1 drive enabled. |
| 4 | DRVEN_DACBO | R/W | Oh | Software drive enable value, when software control is enabled on DRVEN_SW_EN_DACBO. <br> 0: OUTB0 drive disabled, internally connected to VSSB or DACB1, depending on CLAMP_SEL_OUTB0. <br> 1: OUTBO drive enabled, connected to DACBO. |
| 3 | DRVEN_DACA3 | R/W | Oh | Software drive enable value, when software control is enabled on DRVEN_SW_EN_DACA3. <br> 0 : DACA3 drive disabled, internally connected to VSSA. <br> 1: DACA3 drive enabled. |
| 2 | DRVEN_DACA2 | R/W | Oh | Software drive enable value, when software control is enabled on DRVEN_SW_EN_DACA2. <br> 0: OUTA2 drive disabled, internally connected to VSSA or DACA3, depending on CLAMP_SEL_OUTA2. <br> 1: OUTA2 drive enabled, connected to DACA2. |
| 1 | DRVEN_DACA1 | R/W | Oh | Software drive enable value, when software control is enabled on DRVEN_SW_EN_DACA1. <br> 0 : DACA1 drive disabled, internally connected to VSSA. <br> 1: DACA1 drive enabled. |
| 0 | DRVEN_DACAO | R/W | Oh | Software drive enable value, when software control is enabled on DRVEN_SW_EN_DACAO. <br> 0: OUTAO drive disabled, internally connected to VSSA or DACA1, depending on CLAMP_SEL_OUTAO. <br> 1: OUTAO drive enabled, connected to DACAO. |

### 7.1.1.13 DAC_BCAST Register (address = 14h) [reset = 0000h]

Figure 7-13. DAC_BCAST Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  | DAC[12:8] |  |  |  |  |
| R-Oh |  |  | W-Oh |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| DAC[7:0] |  |  |  |  |  |  |  |
| W-Oh |  |  |  |  |  |  |  |

Table 7-15. DAC_BCAST Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $12-0$ | DAC | W | Oh | Write to this register sets all DAC buffer and active data <br> register values to the specified code, on channels for which <br> the respective Broadcast Enable (BCEN) bit is set (see Section <br> 7.5.1.2). Otherwise that channel buffer and active registers are <br> unchanged. |

7.1.1.14 GLOBAL_CFG Register (address = 17h) [reset = 0000h]

Figure 7-14. GLOBAL_CFG Register


Table 7-16. GLOBAL_CFG Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| 1 | ADC_BYP_EN | R/W | Oh | ADC data bypass enable. Bypasses all ADC conversion results. <br> 0: Bypass disabled. <br> $1:$ All ADC conversion data are bypassed, with data registers <br> forced to the value stored in ADC_BYP (see Section 7.3.1.4). |
| 0 | ALARM_BYP_EN | R/W | Oh | Internal alarm bypass. <br> 0: Bypass disabled. <br> 1: All alarm condition states and alarm status bits are bypassed. |

### 7.1.1.15 ADC_SENSEO Register (address $=18 \mathrm{~h})$ [reset $=0000 \mathrm{~h}]$

Figure 7-15. ADC_SENSEO Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADC[15:8] |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ADC[7:0] |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |

Table 7-17. ADC_SENSEO Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $15-0$ | ADC | R | Oh | Differential voltage measured across the shunt output. 2's <br> complement value. When SHUNT_RANGE $=0$, the conversion <br> factor is $5 \mu \mathrm{~V} / \mathrm{LSB}$, and when SHUNT_RANGE $=1$, the <br> conversion factor is $1.25 \mu \mathrm{~V} / \mathrm{LSB}$. |

### 7.1.1.16 ADC_SENSE1 Register (address $=19 \mathrm{~h})$ [reset $=0000 \mathrm{~h}]$

Figure 7-16. ADC_SENSE1 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADC[15:8] |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ADC[7:0] |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |

Table 7-18. ADC_SENSE1 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $15-0$ | ADC | R | Oh | Differential voltage measured across the shunt output. 2's <br> complement value. When SHUNT_RANGE $=0$, the conversion <br> factor is $5 \mu \mathrm{~V} / \mathrm{LSB}$, and when SHUNT_RANGE $=1$, the <br> conversion factor is $1.25 \mu \mathrm{~V} / \mathrm{LSB}$. |

### 7.1.1.17 ADC_ADC0 Register (address = 1Ah) [reset = 0000h]

Figure 7-17. ADC_ADC0 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADC[15:8] |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ADC[7:0] |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |

Table 7-19. ADC_ADC0 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $15-0$ | ADC | R | Oh | ADCHV voltage output, 2's complement value (always positive). <br> Conversion factor: $3.125 \mathrm{mV} / \mathrm{LSB}$. |

### 7.1.1.18 ADC_ADC1 Register (address $=1 B h$ ) [reset $=0000 \mathrm{~h}]$

Figure 7-18. ADC_ADC1 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADC[15:8] |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ADC[7:0] |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |

Table 7-20. ADC_ADC1 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $15-0$ | ADC | R | Oh | ADCHV voltage output, 2's complement value (always positive). <br> Conversion factor: $3.125 \mathrm{mV} / \mathrm{LSB}$. |

7.1.1.19 ADC_TMP Register (address $=1 \mathrm{Ch}$ ) [reset $=0000 \mathrm{~h}]$

Figure 7-19. ADC_TMP Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADC[15:8] |  |  |  |  |  |  |  |
| $\mathrm{R}-\mathrm{Oh}$ |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ADC[7:0] |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |

Table 7-21. ADC_TMP Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $15-0$ | ADC | R | Oh | Internal die temperature measurement. 2's complement value. <br> Conversion factor: $7.8125 \mathrm{~m}^{\circ} \mathrm{C} / \mathrm{LSB}$. |

### 7.2 General Configuration Register Map

Table 7-22. Page 0: General Configuration Register Map

| ADDR <br> (HEX) | REGISTER | TYPE | RESET <br> (HEX) | BIT DESCRIPTION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | $7{ }^{7} \mathbf{6}$ | 5 | 4 | 3 | 2 | 1 | 0 |
| 40 | CHIP_ID | R | 2480 | CHIP_ID[15:0] |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 41 | CHIP_VER | R | 0000 | RESERVED |  |  |  |  |  |  |  |  |  |  | VERSION_ID[3:0] |  |  |  |
| 42 | SDO_EN | R/W | 0000 | RESERVED |  |  |  |  |  |  |  |  |  |  |  |  | FSDO | SDO_EN |
| 44 | GEN_CFG_0 | R/W | 0010 | $\begin{aligned} & \text { PAON } \\ & \text { POL }^{-} \end{aligned}$ | $\begin{aligned} & \text { PAON_- } \\ & \text { ODE }^{-} \end{aligned}$ |  |  |  |  |  |  |  | $\begin{gathered} \hline \text { FLEXIO_ } \\ \text { OUT_- } \\ \text { POL- } \end{gathered}$ | $\begin{aligned} & \text { FLEXIO_ } \\ & \text { OUT_- } \\ & \text { ODE } \end{aligned}$ | ALARM LATCH DIS | RESERVED |  |  |
| 45 | GEN_CFG_1 | R/W | 1101 | RESERVED | $\begin{gathered} \text { VSSB } \\ \text { RANGE[2:0] } \end{gathered}$ |  |  | RESERVED | VSSA RANGE[2:0] |  |  | RESERVED | FLEXIO FUNC[5:0] |  |  |  |  |  |
| 48 | $\begin{gathered} \hline \text { ALARMOUT_- } \\ \text { SRC_0 } \end{gathered}$ | R/W | 0000 | RESERVED |  |  |  |  |  |  | $\begin{gathered} \text { TMP } \\ \text { ALR_OUT } \end{gathered}$ | RESERVED | ADC1 ALR_OUT | ADCO ALR_OUT | RESERVED |  | SENSE1 ALR_OUT | SENSEO ALR_OUT |
| 49 | ALARMOUT_ SRC_1 | R/W | 1833 |  | RESERVED |  | $\begin{gathered} \text { REF_ }_{-} \\ \text {ALR_OUT } \end{gathered}$ | THERM ERR_ ALR_OUT | RESERVED |  |  |  | $\stackrel{\text { VSSB }}{\text { ALR_OUTT }}$ | VSSA ALR_OŪT | RESERVED |  | $\underset{\text { VCCB_- }}{\text { ALR }}$ | VCCA ALR_OUT |
| 4 C | ALARM STATUS 0_BYP | R/W | 0000 | RESERVED |  |  |  |  |  |  | TMP ALR BYP | RESERVED | $\begin{gathered} \text { ADC1 } \\ \text { ALR_BȲP } \end{gathered}$ | ADCO ALR_BYP | RESERVED |  | SENSE1ALR_BYP | SENSEOALR_BYP |
| 4D | ALARM STATUS 1_BYP | R/W | 0000 | RESERVED |  |  | $\underset{\text { REF_}}{\text { ALR_BYP }}$ | THERM ERR ALR_BYP | RESERVED |  |  |  | VSSB <br> ALR_BYP | VSSA ALR_BYP | RESERVED |  | VCCB <br> ALR_BYP | VCCA <br> ALR_BYP |
| 50 | $\begin{aligned} & \text { PAON } \\ & \text { SRC_ } \end{aligned}$ | R/W | 0000 | RESERVED |  |  |  |  |  |  | TMP PAON_ OUT | RESERVED | ADC1_ PAON OUT | ADCO PAONOUT | RESERVED |  | SENSE1 PAON OUT | SENSEO PAON OUT |
| 51 | $\begin{aligned} & \text { PAON } \\ & \text { SRC_ } \end{aligned}$ | R/W | 1833 | RESERVED |  |  | REF PAON_ OUT | $\begin{gathered} \text { THERM } \\ \text { ERR } \\ \text { PAON_OUT } \end{gathered}$ | RESERVED[4:0] |  |  |  | vSSB PAON OUT | VSSA PAON OUT | RESERVED |  | vCCB PAON OUT | VCCA PAON OUT |
| 70 | $\begin{aligned} & \text { RESET- } \\ & \text { FLAGS } \end{aligned}$ | W | 000F | RESERVED |  |  |  |  |  |  |  |  |  |  | VDD COLLAP̄SE _FLAG | $\begin{aligned} & \text { RSTPIN } \\ & \text { FLAG } \end{aligned}$ | $\begin{aligned} & \text { VIO } \\ & \text { FLAG } \end{aligned}$ | PORBASE <br> _FLAG |

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### 7.2.1.1 CHIP_ID Register (address $=40 \mathrm{~h}$ ) [reset $=2480 \mathrm{~h}]$

Figure 7-20. CHIP_ID Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CHIP_ID[15:8] |  |  |  |  |  |  |  |
| R-24h |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| CHIP_ID[7:0] |  |  |  |  |  |  |  |
| R-80h |  |  |  |  |  |  |  |

Table 7-23. CHIP_ID Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $15-0$ | CHIP_ID | R | 2480 h | Chip identification code |

7.2.1.2 CHIP_VER Register (address $=41 \mathrm{~h}$ ) [reset $=0000 \mathrm{~h}]$

Figure 7-21. CHIP_VER Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| RESERVED |  |  |  | VERSION[3:0] |  |  |  |
| R-Oh |  |  |  | R -Oh |  |  |  |

Table 7-24. CHIP_VER Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :--- | :--- | :--- | :--- | :--- |
| $3-0$ | VERSION[3:0] | R | Oh | Chip version ID. <br> $0 \times 0: R e v ~ A . ~$ <br> $0 \times 1: ~ R e v ~ B . ~$ <br> $0 x 2: ~ R e v ~ C . ~$ <br> 0x3: Rev D. |

### 7.2.1.3 SDO_EN Register (address = 42h) [reset = 0000h]

Figure 7-22. SDO_EN Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| RESERVED |  |  |  |  |  | FSDO SDO_EN <br> R/W-Oh R/W-0h |  |
| R-Oh |  |  |  |  |  |  |  |

Table 7-25. SDO_EN Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| 1 | FSDO | R/W | Oh | Enables faster SPI bus speeds by sending the SDO data <br> out one SCLK half-cycle earlier (FSDO mode). Ignored when <br> SDO_EN = 0 <br> $0=$ FSDO disabled, SDO drives MSB when chip select goes low <br> and then updates on each SCLK rising edge (opposite edge of <br> SDI latching edge) <br> $1=$ FSDO enabled, SDO drives MSB when chip select goes low <br> and then updates on each SCLK falling edge (same edge as <br> SDI latching edge) |
| 0 | SDO_EN | R/W | Oh | SDO Enable. SDO is enabled for read and write operations <br> whenever the SPI CS pin is low. SDO is always disabled in $I^{2} \mathrm{C}$ <br> mode regardless of this bit setting. <br> $0=$ SDO disabled <br> $1=$ SDO enabled during read/write operations |

### 7.2.1.4 GEN_CFG_0 Register (address $=$ 44h) [reset $=0010 h]$

Table 7-26. GEN_CFG_0 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PAON_POL | PAON_ODE |  |  | RES |  |  |  |
| R/W-Oh | R/W-Oh | R-Oh |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| RESERVED |  | $\underset{\text { POL }}{\text { FLEXIO_OUT_ }}$ | $\begin{gathered} \text { FLEXIO_OUT_ } \\ \text { ODE } \end{gathered}$ | ALARM_LATCH_ DIS |  | RESERVED |  |
| R-Oh |  | R/W-Oh | R/W-1h | R/W-Oh | R-Oh |  |  |

Table 7-27. GEN_CFG_0 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| 15 | PAON_POL | R/W | Oh | PAON polarity <br> 0: PAON pin follows PWR_EN PAON bit setting, with high <br> voltage being on and low voltage being off <br> $1:$ PAON pin is inverted from PWR_EN PAON bit setting, with <br> high voltage being off and low voltage being on. |
| 14 | PAON_ODE | R/W | Oh | PAON open-drain enable <br> 0: PAON pin is a push-pull output (default) <br> $1:$ PAON pin is an open-drain output |
| 5 | FLEXIO_OUT_POL | R/W | Oh | FLEXIO pin polarity, when configured as digital output <br> $0=$ FLEXIO is active low <br> $1=$ FLEXIO is active high |
| 4 | FLEXIO_OUT_ODE | R/W | 1 h | FLEXIO pin drive mode, when configured as digital output <br> $0=$ FLEXIO pin is a push-pull output <br> $1=$ FLEXIO pin is an open-drain output |
| 3 | ALARM_LATCH_DIS | R/W | Oh | Alarm latch status <br> $0=$ Alarm state is latched. Global alarm bit GALR and FLEXIO <br> pin (if configured as ALARMOUT) only return to default state <br> if the GALR bit is read 0 after corresponding alarm condition <br> subsides <br> $1=$ Alarm state is not latched. Global alarm bit and FLEXIO pin <br> (if configured as ALARMOUT) return to default state as soon as <br> alarm condition subsides. |

### 7.2.1.5 GEN_CFG_1 Register (address $=45 \mathrm{~h}$ ) [reset = 1101h]

Figure 7-23. GEN_CFG_1 Register


Table 7-28. GEN_CFG_1 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 14-12 | VSSB_RANGE | R/W | 1h | Configure VSS range for DAC group B; at any voltage outside this range, VSSB_ALR is set <br> 001: Low-range VSS ( -11 V to -3 V ) <br> 010: Mid-range VSS ( -11 V to -4.5 V ) <br> 100: High-range VSS ( -11 V to -7.5 V ) |
| 10-8 | VSSA_RANGE | R/W | 1h | Configure VSS range for DAC group A; at any voltage outside this range, VSSA_ALR is set <br> 001: Low-range VSS ( -11 V to -3 V ) <br> 010: Mid-range VSS ( -11 V to -4.5 V ) <br> 100: High-range VSS ( -11 V to -7.5 V ) |
| 5-0 | FLEXIO_FUNC | R/W | 1h | Sets function for FLEXIO pin 0x01: RESET <br> 0x02: ALARMOUT <br> 0x04: GPIO pin <br> $0 \times 08$ : LDAC <br> $0 \times 10$ : ALARMIN <br> 0x20: DRVEN2 |

### 7.2.1.6 ALARMOUT_SRC_0 Register (address = 48h) [reset = 0000h]

Figure 7-24. ALARMOUT_SRC_0 Register


Table 7-29. ALARMOUT_SRC_0 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 8 | TMP_ALR_OUT | R/W | Oh | 0 : Temperature alarm is not a source for ALARMOUT pin assertion. <br> 1: Temperature alarm is a source for ALARMOUT pin assertion. |
| 5 | ADC1_ALR_OUT | R/W | Oh | 0 : ADC1 alarm is not a source for ALARMOUT pin assertion. <br> 1: ADC1 alarm is a source for ALARMOUT pin assertion. |
| 4 | ADCO_ALR_OUT | R/W | Oh | 0 : ADC0 alarm is not a source for $\overline{\text { ALARMOUT }}$ pin assertion. <br> 1: ADC0 alarm is a source for ALARMOUT pin assertion. |
| 1 | SENSE1_ALR_OUT | R/W | Oh | 0 : SENSE1 channel alarm is not a source for ALARMOUT pin assertion <br> 1: SENSE1 channel is a source for ALARMOUT pin assertion. |
| 0 | SENSEO_ALR_OUT | R/W | Oh | 0 : SENSE0 channel alarm is not a source for ALARMOUT pin assertion 1: SENSE0 channel is a source for ALARMOUT pin assertion. |

### 7.2.1.7 ALARMOUT_SRC_1 Register (address = 49h) [reset = 1833h]

Figure 7-25. ALARMOUT_SRC_1 Register


| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 12 | REF_ALR_OUT | R/W | 1h | 0 : Reference alarm is not a source for ALARMOUT pin assertion. <br> 1: Reference alarm is a source for ALARMOUT pin assertion. |
| 11 | THERMERR_ALR_OUT | R/W | 1h | 0 : Thermal alarm is not a source for ALARMOUT pin assertion. <br> 1: Thermal alarm is a source for ALARMOUT pin assertion. |
| 5 | VSSB_ALR_OUT | R/W | 1h | 0 : VSSB alarm is not a source for ALARMOUT pin assertion. <br> 1: VSSB alarm is a source for ALARMOUT pin assertion. |
| 4 | VSSA_ALR_OUT | R/W | 1h | 0 : VSSA alarm is not a source for ALARMOUT pin assertion. <br> 1: VSSA alarm is a source for ALARMOUT pin assertion. |
| 1 | VCCB_ALR_OUT | R/W | 1h | 0 : VCCB alarm is not a source for ALARMOUT pin assertion. <br> 1: VCCB alarm is a source for ALARMOUT pin assertion. |
| 0 | VCCA_ALR_OUT | R/W | 1h | 0 : VCCA alarm is not a source for ALARMOUT pin assertion. <br> 1: VCCA alarm is a source for ALARMOUT pin assertion. |

### 7.2.1.8 ALARM_STATUS_0_BYP Register (address = 4Ch) [reset = 0000h]

Figure 7-26. ALARM_STATUS_0_BYP Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  |  | TMP ALR_BYP |
| R-Oh |  |  |  |  |  |  | R/W-Oh |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| RESERVED |  | ADC1 ALR BȲP | ADC0 ALR BȲ |  |  | SENSE1 <br> ALR BYP | SENSEO <br> ALR BYP |
| R-Oh |  | R/W-Oh | R/W-Oh |  |  | R/W-Oh | R/W-Oh |

Table 7-31. ALARM_STATUS_0_BYP Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 8 | TMP_ALR_BYP | R/W | Oh | Temperature alarm bypass command (when ALARM_BYP_EN = 1) <br> 0 : Temperature alarm status is forced to 0 <br> 1: Temperature alarm status is forced to 1 |
| 5 | ADC1_ALR_BYP | R/W | Oh | ADC1 alarm bypass command (when ALARM_BYP_EN = 1) <br> 0 : ADC1 alarm status is forced to 0 <br> 1: ADC1 alarm status is forced to 1 |
| 4 | ADCO_ALR_BYP | R/W | Oh | ADC0 alarm bypass command (when ALARM_BYP_EN = 1) <br> 0 : ADC0 alarm status is forced to 0 <br> 1: ADC0 alarm status is forced to 1 |
| 1 | SENSE1_ALR_BYP | R/W | Oh | SENSE1 alarm bypass command (when ALARM_BYP_EN = 1) <br> 0 : SENSE1 alarm status is forced to 0 <br> 1: SENSE1 alarm status is forced to 1 |
| 0 | SENSEO_ALR_BYP | R/W | Oh | SENSE0 alarm bypass command (when ALARM_BYP_EN = 1) <br> 0 : SENSE0 alarm status is forced to 0 <br> 1: SENSE0 alarm status is forced to 1 |

### 7.2.1.9 ALARM_STATUS_1_BYP Register (address = 4Dh) [reset = 0000h]

Figure 7-27. ALARM_STATUS_1_BYP Register


Table 7-32. ALARM_STATUS_1_BYP Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 13 | ALARMIN_ALR_BYP | R/W | Oh | ALARMIN bypass command (when ALARM_BYP_EN = 1) <br> 0 : ALARMIN alarm status is forced to 0 <br> 1: ALARMIN alarm status is forced to 1 |
| 12 | REF_ALR_BYP | R/W | Oh | Reference alarm bypass command (when ALARM_BYP_EN = 1) <br> 0 : Reference alarm status is forced to 0 <br> 1: Reference alarm status is forced to 1 |
| 11 | THERMERR_ALR_BYP | R/W | Oh | Thermal alarm bypass command (when ALARM_BYP_EN = 1) <br> 0 : Thermal alarm status is forced to 0 <br> 1: Thermal alarm status is forced to 1 |
| 5 | VSSB_ALR_BYP | R/W | Oh | VSSB alarm bypass command (when ALARM_BYP_EN = 1) <br> 0 : VSSB alarm status is forced to 0 <br> 1: VSSB alarm status is forced to 1 |
| 4 | VSSA_ALR_BYP | R/W | Oh | VSSA alarm bypass command (when ALARM_BYP_EN = 1) <br> 0 : VSSA alarm status is forced to 0 <br> 1: VSSA alarm status is forced to 1 |
| 1 | VCCB_ALR_BYP | R/W | Oh | VCCB alarm bypass command (when ALARM_BYP_EN = 1) <br> 0 : VCCB alarm status is forced to 0 <br> 1: VCCB alarm status is forced to 1 |
| 0 | VCCA_ALR_BYP | R/W | Oh | VCCA alarm bypass command (when ALARM_BYP_EN = 1) <br> 0 : VCCA alarm status is forced to 0 <br> 1: VCCA alarm status is forced to 1 |

### 7.2.1.10 PAON_SRC_0 Register (address = 50h) [reset = 0000h]

Figure 7-28. PAON_SRC_0 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | RESERVED |  |  |  |  |

Table 7-33. PAON_SRC_0 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| 8 | TMP_PAON_OUT | R/W | Oh | 0: Temperature alarm event does not affect PAON pin. <br> 1: Temperature alarm event turns off PAON pin. |
| 5 | ADC1_PAON_OUT | R/W | Oh | 0: ADC1 alarm event does not affect PAON pin. <br> 1: ADC1 alarm event turns off PAON pin. |
| 4 | ADC0_PAON_OUT | R/W | Oh | 0: ADC0 alarm event does not affect PAON pin. <br> 1: ADC0 alarm event turns off PAON pin. |
| 1 | SENSE1_PAON_OUT | R/W | Oh | 0: SENSE1 alarm event does not affect PAON pin. <br> 1: SENSE1 alarm event turns off PAON pin. |
| 0 | SENSE0_PAON_OUT | R/W | Oh | 0: SENSE0 alarm event does not affect PAON pin. <br> 1: SENSE0 alarm event turns off PAON pin. |

### 7.2.1.11 PAON_SRC_1 Register (address = 51h) [reset = 1833h]

Figure 7-29. PAON_SRC_1 Register


Table 7-34. PAON_SRC_1 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| 12 | REF_PAON_OUT | R/W | 1h | 0: Reference alarm event does not affect PAON pin. <br> $1:$ Reference alarm event turns off PAON pin. |
| 11 | THERMERR_PAON_OUT | R/W | 1h | 0: Thermal error alarm event does not affect PAON pin. <br> $1:$ Thermal error alarm event turns off PAON pin. |
| 5 | VSSB_PAON_OUT | R/W | 1h | 0: VSSB alarm event does not affect PAON pin. <br> $1:$ VSSB alarm event turns off PAON pin. |
| 4 | VSSA_PAON_OUT | R/W | 1h | 0: VSSA alarm event does not affect PAON pin. <br> $1:$ VSSA alarm event turns off PAON pin. |
| 1 | VCCB_PAON_OUT | R/W | 1h | 0: VCCB alarm event does not affect PAON pin. <br> $1:$ VCCB alarm event turns off PAON pin. |
| 0 | VCCA_PAON_OUT | R/W | 1 h | 0: VCCA alarm event does not affect PAON pin. <br> $1:$ VCCA alarm event turns off PAON pin. |

7.2.1.12 RESET_FLAGS Register (Offset $=70 h$ ) [Reset $=000 \mathrm{Fh}]$

Figure 7-30. RESET_FLAGS Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  |  |  |  | $\begin{aligned} & \text { VDD_ } \\ & \text { COLLAPSE_ } \\ & \text { FLAG } \end{aligned}$ | RSTPIN_FLAG | VIO_FLAG | PORBASE_FLA G |
| R-Oh |  |  |  | W-1h | W-1h | W-1h | W-1h |

Table 7-35. RESET_FLAGS Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| 3 | VDD_COLLAPSE_FLAG | W | 1 h | VDD collapse flag. <br> Write to 0 to detect a VDD collapse event, at which time this flag is <br> automatically set to 1. VDD collapse occurs when VDD reaches to <br> within 1V of the VREF voltage. |
| 2 | RSTPIN_FLAG | W | 1 h | RESET pin reset flag. <br> Write to 0 to detect a RESET pin reset event, at which time this flag <br> is automatically set to 1. |
| 1 | VIO_FLAG | W | 1 h | VIO reset flag. <br> Write to 0 to detect a VIO reset event, at which time this flag is <br> automatically set to 1. VIO reset event occurs as a result of VIO <br> dropping below the POR threshold voltage. |
| 0 | PORBASE_FLAG | W | 1 h | POR base flag. <br> Write to 0 to detect a POR-base reset event, at which time this flag <br> is automatically set to 1. POR-base reset event occurs as a result of <br> VDD dropping below the POR threshold voltage. |

### 7.3 ADC Configuration Register Map

Table 7-36. Page 1: ADC Configuration Register Map


### 7.3.1 ADC Configuration Registers: Page 1

### 7.3.1.1 ADC_GEN_CFG Register (address = 40h) [reset = 3334h]

Figure 7-31. ADC_GEN_CFG Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  | FALR_ADC[2:0] |  | RESERVED | FALR_SENSE[2:0] |  |  |
| R-Oh | R/W-3h |  |  | R-Oh |  | R/W-3h |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| RESERVED |  | FALR_TMP[2:0] |  | RESERVED | CMODE | SHUNT RANGE | RESERVED |
| R-Oh |  | R/W-3h |  | R-Oh | R/W-1h | R/W-Oh | R-Oh |

Table 7-37. ADC_GEN_CFG Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 14-12 | FALR_ADC | R/W | 3h | False alarm factor for external input (ADC) channels. <br> 000: 1 out-of-range conversion required to trigger alarm. <br> 001: 4 consecutive out-of-range conversions required to trigger alarm. 010: 8 consecutive out-of-range conversions required to trigger alarm. 011: 16 consecutive out-of-range conversions required to trigger alarm. 100: 32 consecutive out-of-range conversions required to trigger alarm. 101: 64 consecutive out-of-range conversions required to trigger alarm. 110: 128 consecutive out-of-range conversions required to trigger alarm. 111: 256 consecutive out-of-range conversions required to trigger alarm. |
| 10-8 | FALR_SENSE | R/W | 3h | False alarm factor for SENSE channels. <br> 000: 1 out-of-range conversion required to trigger alarm. <br> 001: 4 consecutive out-of-range conversions required to trigger alarm. 010: 8 consecutive out-of-range conversions required to trigger alarm. <br> 011: 16 consecutive out-of-range conversions required to trigger alarm. 100: 32 consecutive out-of-range conversions required to trigger alarm. <br> 101: 64 consecutive out-of-range conversions required to trigger alarm. <br> 110: 128 consecutive out-of-range conversions required to trigger alarm. <br> 111: 256 consecutive out-of-range conversions required to trigger alarm. |
| 6-4 | FALR_TMP | R/W | 3h | False alarm factor for temperature measurements. <br> 000: 1 out-of-range conversion required to trigger alarm. <br> 001: 4 consecutive out-of-range conversions required to trigger alarm. <br> 010: 8 consecutive out-of-range conversions required to trigger alarm. <br> 011: 16 consecutive out-of-range conversions required to trigger alarm. <br> 100: 32 consecutive out-of-range conversions required to trigger alarm. <br> 101: 64 consecutive out-of-range conversions required to trigger alarm. <br> 110: 128 consecutive out-of-range conversions required to trigger alarm. <br> 111: 256 consecutive out-of-range conversions required to trigger alarm. |
| 2 | CMODE | R/W | 1h | ADC conversion mode bit. This bit selects the ADC conversion mode. 0 : Direct-mode. The analog inputs specified in the device sequencer are converted sequentially one time. When one set of conversions is complete the ADC is idle and waits for a new trigger. <br> 1: Auto-mode. The analog inputs specified in the device sequencer are converted sequentially and repeatedly. When one set of conversions is complete the ADC sequencer returns to the start index and repeats the sequence. |
| 1 | SHUNT RANGE | R/W | Oh | Shunt voltage range selection bit for SENSE input channels $0: \pm 163.84 \mathrm{mV}$ range <br> $1: \pm 40.96 \mathrm{mV}$ range |

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7.3.1.2 ADC_CONV_CFG_0 Register (address = 41h) [reset $=0555 \mathrm{~h}]$

Figure 7-32. ADC_CONV_CFG_0 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  | CONV_RATE_TMP[2:0] |  |  |
| R-Oh |  |  |  |  | R/W-5h |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| RESERVED |  | CONV_RATE_ADC[2:0] |  | RESERVED | CONV_RATE_SENSE[2:0] |  |  |
| R-Oh | R/W-5h |  |  | R-Oh | R/W-5h |  |  |

Table 7-38. ADC_CONV_CFG_0 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 10-8 | CONV_RATE_TMP | R/W | 5h | Total acquisition + conversion time for temperature measurements with no averaging. <br> 000: $52 \mu \mathrm{~s}$ <br> 001: $86 \mu \mathrm{~s}$ <br> 010: $152 \mu \mathrm{~s}$ <br> 011: $282 \mu \mathrm{~s}$ <br> 100: $542 \mu \mathrm{~s}$ <br> 101: 1054 us <br> 110: $2076 \mu \mathrm{~s}$ <br> 111: $4122 \mu \mathrm{~s}$ |
| 6-4 | CONV_RATE_ADC | R/W | 5h | ```Total acquisition + conversion time for ADC voltage measurements with no averaging. 000: \(52 \mu \mathrm{~s}\) 001: \(86 \mu \mathrm{~s}\) 010: \(152 \mu \mathrm{~s}\) 011: \(282 \mu \mathrm{~s}\) 100: \(542 \mu \mathrm{~s}\) 101: 1054 us 110: \(2076 \mu \mathrm{~s}\) 111: \(4122 \mu \mathrm{~s}\)``` |
| 2-0 | CONV_RATE_SENSE | R/W | 5h | Total acquisition + conversion time for SENSE shunt voltage measurements with no averaging. $\begin{aligned} & \text { 000: } 52 \mu \mathrm{~s} \\ & 001: 86 \mu \mathrm{~s} \\ & 010: 152 \mu \mathrm{~s} \\ & 011: 282 \mu \mathrm{~s} \\ & \text { 100: } 542 \mu \mathrm{~s} \\ & \text { 101: } 1054 \mu \mathrm{~s} \\ & 110: 2076 \mu \mathrm{~s} \\ & 111: 4122 \mu \mathrm{~s} \end{aligned}$ |

### 7.3.1.3 ADC_CONV_CFG_1 Register (address = 42h) [reset = 0000h]

Figure 7-33. ADC_CONV_CFG_1 Register


Table 7-39. ADC_CONV_CFG_1 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 10-8 | AVG_TMP | R/W | Oh | Averaging setting for temperature measurements. The device reports and acts upon averaged result. <br> 000: 1 sample <br> 001: 4 samples <br> 010: 16 samples <br> 011: 64 samples <br> 100: 128 samples <br> 101: 256 samples <br> 110: 512 samples <br> 111: 1024 samples |
| 6-4 | AVG_ADC | R/W | Oh | Averaging setting for ADC voltage measurements. The device reports and acts upon averaged result. <br> 000: 1 sample <br> 001: 4 samples <br> 010: 16 samples <br> 011: 64 samples <br> 100: 128 samples <br> 101: 256 samples <br> 110: 512 samples <br> 111: 1024 samples |
| 2-0 | AVG_SENSE | R/W | Oh | Averaging setting for SENSE shunt voltage measurements. The device reports and acts upon averaged result. <br> 000: 1 sample <br> 001: 4 samples <br> 010: 16 samples <br> 011: 64 samples <br> 100: 128 samples <br> 101: 256 samples <br> 110: 512 samples <br> 111: 1024 samples |

### 7.3.1.4 ADC_BYP Register (address $=44 \mathrm{~h})$ [reset $=0000 \mathrm{~h}]$

Figure 7-34. ADC_BYP Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADC_BYP[15:8] |  |  |  |  |  |  |  |
| R/W-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ADC_BYP[7:0] |  |  |  |  |  |  |  |
| R/W-Oh |  |  |  |  |  |  |  |

Table 7-40. ADC_BYP Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $15-0$ | ADC_BYP | R/W | Oh | ADC data bypass value. Only used when ADC_BYP_EN is set <br> to 1 |

### 7.3.1.5 ADC_HYST_0 Register (address $=46 \mathrm{~h}$ ) [reset $=0808 \mathrm{~h}]$

Figure 7-35. ADC_HYST_0 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| HYST_TMP[7:0] |  |  |  |  |  |  |  |
| R/W-8h |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| HYST_ADC[7:0] |  |  |  |  |  |  |  |
| R/W-8h |  |  |  |  |  |  |  |

Table 7-41. ADC_HYST_0 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $15-8$ | HYST_TMP | R/W | 8 h | Hysteresis setting for temperature measurements. 1 LSB per <br> step. |
| $7-0$ | HYST_ADC | R/W | 8 h | Hysteresis setting for ADC voltage measurements. 1 LSB per <br> step. |

7.3.1.6 ADC_HYST_1 Register (address = 47h) [reset = 0008h]

Figure 7-36. ADC_HYST_1 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| HYST_SENSE[7:0] |  |  |  |  |  |  |  |
| R/W-8h |  |  |  |  |  |  |  |

Table 7-42. ADC_HYST_1 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :--- | :--- | :--- | :--- | :--- |
| $7-0$ | HYST_SENSE | R/W | 8 h | Hysteresis setting for shunt voltage measurements. 1 LSB per <br> step. |

7.3.1.7 SENSEO_UP_THRESH Register (address = 50h) [reset = 7FFFh]

Figure 7-37. SENSE0_UP_THRESH Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| THRU[15:8] |  |  |  |  |  |  |  |
| R/W-7Fh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| THRU[7:0] |  |  |  |  |  |  |  |
| R/W-FFh |  |  |  |  |  |  |  |

Table 7-43. SENSE0_UP_THRESH Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $15-0$ | THRU | R/W | 7FFFh | Upper threshold for shunt voltage measurements, 2's <br> complement value. Corresponding alarm status bit is activated <br> if (channel code > UP thresh) or (channel code < LO thresh), <br> and is cleared if (channel code $\leq$ UP thresh - hyst) and (channel <br> code $\geq$ LO thresh + hyst). Upper threshold minus hysteresis <br> must always be greater than lower threshold plus hysteresis. <br> When SHUNT_RANGE =0, the conversion factor is 5 $\mu \mathrm{V} / \mathrm{LSB}$, <br> and when SHUNT_RANGE =1, the conversion factor is 1.25 $\mu \mathrm{V} /$ <br> LSB. |

7.3.1.8 SENSEO_LO_THRESH Register (address = 51h) [reset = 8000h]

Figure 7-38. SENSE0_LO_THRESH Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| THRL[15:8] |  |  |  |  |  |  |  |
| R/W-80h |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| THRL[7:0] |  |  |  |  |  |  |  |
| R/W-00h |  |  |  |  |  |  |  |

Table 7-44. SENSEO_LO_THRESH Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $15-0$ | THRL | R/W | 8000 h | Lower threshold for shunt voltage measurements, 2's <br> complement value. Corresponding alarm status bit is activated <br> if (channel code > UP thresh) or (channel code < LO thresh), <br> and is cleared if (channel code $\leq$ UP thresh - hyst) and (channel <br> code $\geq$ LO thresh + hyst). Upper threshold minus hysteresis <br> must always be greater than lower threshold plus hysteresis. <br> When SHUNT_RANGE $=0$, the conversion factor is 5 $\mu \mathrm{V} / \mathrm{LSB}$, <br> and when SHUNT_RANGE $=1$, the conversion factor is $1.25 \mu \mathrm{~V} /$ <br> LSB. |

7.3.1.9 SENSE1_UP_THRESH Register (address = 52h) [reset = 7FFFh]

Figure 7-39. SENSE1_UP_THRESH Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| THRU[15:8] |  |  |  |  |  |  |  |
| R/W-7Fh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| THRU[7:0] |  |  |  |  |  |  |  |
| R/W-FFh |  |  |  |  |  |  |  |

Table 7-45. SENSE1_UP_THRESH Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $15-0$ | THRU | R/W | 7FFFh | Upper threshold for shunt voltage measurements, 2's <br> complement value. Corresponding alarm status bit is activated <br> if (channel code > UP thresh) or (channel code < LO thresh), <br> and is cleared if (channel code $\leq$ UP thresh - hyst) and (channel <br> code $\geq$ LO thresh + hyst). Upper threshold minus hysteresis <br> must always be greater than lower threshold plus hysteresis. <br> When SHUNT_RANGE $=0$, the conversion factor is 5 $\mu \mathrm{V} / \mathrm{LSB}$, <br> and when SHUNT_RANGE $=1$, the conversion factor is $1.25 \mu \mathrm{VI} /$ <br> LSB. |

### 7.3.1.10 SENSE1_LO_THRESH Register (address = 53h) [reset = 8000h]

Figure 7-40. SENSE1_LO_THRESH Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| THRL[15:8] |  |  |  |  |  |  |  |
| R/W-80h |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| THRL[7:0] |  |  |  |  |  |  |  |
| R/W-00h |  |  |  |  |  |  |  |

Table 7-46. SENSE1_LO_THRESH Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $15-0$ | THRL | R/W | 8000 h | Lower threshold for shunt voltage measurements, 2's <br> complement value. Corresponding alarm status bit is activated <br> if (channel code > UP thresh) or (channel code < LO thresh), <br> and is cleared if (channel code $\leq$ UP thresh - hyst) and (channel <br> code $\geq$ LO thresh + hyst). Upper threshold minus hysteresis <br> must always be greater than lower threshold plus hysteresis. <br> When SHUNT_RANGE $=0$, the conversion factor is 5 $\mu \mathrm{V} / \mathrm{LSB}$, <br> and when SHUNT_RANGE $=1$, the conversion factor is $1.25 \mu \mathrm{~V} /$ <br> LSB. |

### 7.3.1.11 ADCO_UP_THRESH Register (address = 54h) [reset = 7FFFh]

Figure 7-41. ADC0_UP_THRESH Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED | THRU[14:8] |  |  |  |  |  |  |
| R-Oh | R/W-7Fh |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| THRU[7:0] |  |  |  |  |  |  |  |
| R/W-FFh |  |  |  |  |  |  |  |

Table 7-47. ADC0_UP_THRESH Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $14-0$ | THRU | R/W | 7FFFh | Upper threshold for ADC voltage measurements, unsigned <br> positive value. Corresponding alarm status bit is activated if <br> (channel code > UP thresh) or (channel code < LO thresh), and <br> is cleared if (channel code $\leq$ UP thresh - hyst) and (channel <br> code $\geq$ LO thresh + hyst). Upper threshold minus hysteresis <br> must always be greater than lower threshold plus hysteresis. <br> Conversion factor is 3.125mV/LSB. |

7.3.1.12 ADCO_LO_THRESH Register (address = 55h) [reset $=0000 \mathrm{~h}]$

Figure 7-42. ADC0_LO_THRESH Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED | THRL[14:8] |  |  |  |  |  |  |
| R-Oh | R/W-Oh |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| THRL[7:0] |  |  |  |  |  |  |  |
| R/W-00h |  |  |  |  |  |  |  |

Table 7-48. ADC0_LO_THRESH Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $14-0$ | THRL | R/W | 0000 h | Lower threshold for ADC voltage measurements, unsigned <br> positive value. Corresponding alarm status bit is activated if <br> (channel code $>$ UP thresh) or (channel code $<$ LO thresh), and <br> is cleared if (channel code $\leq$ UP thresh - hyst) and (channel <br> code $\geq$ LO thresh + hyst). Upper threshold minus hysteresis <br> must always be greater than lower threshold plus hysteresis. <br> Conversion factor is $3.125 \mathrm{mV} / \mathrm{LSB}$. |

7.3.1.13 ADC1_UP_THRESH Register (address = 56h) [reset = 7FFFh]

Figure 7-43. ADC1_UP_THRESH Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED | THRU[14:8] |  |  |  |  |  |  |
| R-Oh | R/W-7Fh |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| THRU[7:0] |  |  |  |  |  |  |  |
| R/W-FFh |  |  |  |  |  |  |  |

Table 7-49. ADC1_UP_THRESH Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $14-0$ | THRU | R/W | 7FFFh | Upper threshold for ADC voltage measurements, unsigned <br> positive value. Corresponding alarm status bit is activated if <br> (channel code > UP thresh) or (channel code < LO thresh), and <br> is cleared if (channel code $\leq$ UP thresh - hyst) and (channel <br> code $\geq$ LO thresh + hyst). Upper threshold minus hysteresis <br> must always be greater than lower threshold plus hysteresis. <br> Conversion factor is 3.125mV/LSB. |

7.3.1.14 ADC1_LO_THRESH Register (address $=57 \mathrm{~h})$ [reset $=0000 \mathrm{~h}]$

Figure 7-44. ADC1_LO_THRESH Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED | THRL[14:8] |  |  |  |  |  |  |
| R-Oh | R/W-0h |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| THRL[7:0] |  |  |  |  |  |  |  |
| R/W-00h |  |  |  |  |  |  |  |

Table 7-50. ADC1_LO_THRESH Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $14-0$ | THRL | R/W | 0000 h | Lower threshold for ADC voltage measurements, unsigned <br> positive value. Corresponding alarm status bit is activated if <br> (channel code $>$ UP thresh) or (channel code $<$ LO thresh), and <br> is cleared if (channel code $\leq$ UP thresh - hyst) and (channel <br> code $\geq$ LO thresh + hyst). Upper threshold minus hysteresis <br> must always be greater than lower threshold plus hysteresis. <br> Conversion factor is $3.125 \mathrm{mV} / \mathrm{LSB}$. |

7.3.1.15 TMP_UP_THRESH Register (address = 58h) [reset = 7FFFh]

Figure 7-45. TMP_UP_THRESH Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| THRU[15:8] |  |  |  |  |  |  |  |
| R/W-7Fh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| THRU[7:0] |  |  |  |  |  |  |  |
| R/W-FFh |  |  |  |  |  |  |  |

Table 7-51. TMP_UP_THRESH Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $15-0$ | THRU | R/W | 7FFFh | Upper threshold for temperature measurements, in 2's <br> complement representation (use positive values only). <br> Corresponding alarm status bit is activated if (channel code > <br> UP thresh), and is cleared if (channel code $\leq$ UP thresh - hyst). <br> Upper threshold minus hysteresis must always be positive. <br> Conversion factor is $7.8125 \mathrm{~m}^{\circ} \mathrm{C} / L S B$. |

### 7.4 ADC Custom Channel Sequencer Configuration Register Map

Table 7-52. Page 2: ADC Custom Channel Sequencer Configuration Register Map


Table 7-52. Page 2: ADC Custom Channel Sequencer Configuration Register Map (continued)

| ADDR <br> (HEX) | REGISTER | TYPE | $\begin{aligned} & \text { RESET } \\ & \text { (HEX) } \end{aligned}$ | BIT DESCRIPTION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 5A | $\begin{aligned} & \text { ADC_CCS_IDS_ } \\ & 26 \end{aligned}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_53[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_52[2:0] |  |  |
| 5B | $\operatorname{ADC}_{27}{ }_{27}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_55[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_54[2:0] |  |  |
| 5C | $\begin{gathered} \text { ADC_CCS_IDS_ } \\ 28 \end{gathered}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_57[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_56[2:0] |  |  |
| 5D | $\begin{aligned} & \text { ADC_CCS_IDS_ } \\ & 29 \end{aligned}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_59[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_58[2:0] |  |  |
| 5E | $\begin{gathered} \text { ADC_CCS_IDS_ } \\ 30 \end{gathered}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_61[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_60[2:0] |  |  |
| 5 F | $\begin{gathered} \text { ADC_CCS_IDS_ } \\ 31 \end{gathered}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_63[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_62[2:0] |  |  |
| 60 | ADC_CCs_IDS_ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_65[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_64[2:0] |  |  |
| 61 | $\underset{33}{ }{ }^{\text {ADC_CCS_IDS_ }}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_67[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_66[2:0] |  |  |
| 62 | $\text { ADC_CCS_IDS_ }_{34}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_69[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_68[2:0] |  |  |
| 63 | $\text { ADC_CCS_IDS_ }_{35}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_71[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_70[2:0] |  |  |
| 64 | ${ }_{36}{ }^{\text {ADC_CCS_IDS_ }}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_73[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_72[2:0] |  |  |
| 65 | ${ }_{37} \text { ADC_CCS_IDS_ }^{2}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_75[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_74[2:0] |  |  |
| 66 | $\text { ADC_CCS_IDS_ }_{38}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_77[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_76[2:0] |  |  |
| 67 | $\begin{gathered} \text { ADC_CCS_IDS_ } \\ 39 \end{gathered}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_79[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_78[2:0] |  |  |
| 68 | $\text { ADC_CCS_IDS_ }_{40}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_81[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_80[2:0] |  |  |
| 69 | $\text { ADC_CCS_IDS_ }_{41}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_83[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_82[2:0] |  |  |
| 6A | $\text { ADC_CCS_IDS_ }_{42}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_85[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_84[2:0] |  |  |
| 6B | $\underset{43}{ } \mathrm{ADC}_{2} \mathrm{CCS}_{-}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_87[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_86[2:0] |  |  |
| 6 C | $\underset{44}{ } \mathrm{ADC}_{-} \mathrm{CCS}_{-} \mathrm{IDS}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_89[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_88[2:0] |  |  |
| 6D | $\text { ADC_CCS_IDS_ }_{45}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_91[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_90[2:0] |  |  |
| 6 E | $\underset{46}{ } \mathrm{ADC}_{-} \mathrm{CCS} \text { IDS_ }$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_93[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_92[2:0] |  |  |
| 6 F | $\mathrm{ADC}_{47} \mathrm{CCS} \text { IDS_ }$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_95[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_94[2:0] |  |  |
| 70 | $\underset{48}{ }{ }^{\text {ADC_CCS_IDS_ }}$ | R/W | 0000 | RESERVED |  |  |  |  | CCS_ID_97[2:0] |  |  | RESERVED |  |  |  |  | CCS_ID_96[2:0] |  |  |

Table 7-52. Page 2: ADC Custom Channel Sequencer Configuration Register Map (continued)


### 7.4.1 ADC CCS Registers: Page 3

7.4.1.1 ADC_CCS_IDS_n Registers (address = 40h to 7Eh) [reset = see Section 7.4]

Figure 7-46. ADC_CCS_IDS_n Register


Table 7-53. ADC_CCS_IDS_n Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $10-8$ | CCS_ID_a | R/W | see <br> Section 7.4 | ADC custom channel sequence index setting <br> 000: GND <br> 001: SENSE0 |
| $2-0$ | CCS_ID_b | R/W | see <br> Section 7.4 | 010: SENSE1 <br> 011: ADC0 <br> 100: ADC1 <br> 101: TMP |

## Note

CCS_ID_a refers to odd-indexed CCS ID registers, and CCS_ID_b refers to even-indexed CCS ID registers.
7.4.1.2 ADC_CCS_CFG_0 Register (address = 7Fh) [reset $=0004 \mathrm{~h}]$

Figure 7-47. ADC_CCS_CFG_0 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  | CCS_START_INDEX[6:0] |  |  |  |  |
| R-Oh |  | 5 | R/W-Oh |  |  |  |  |
| 7 | 6 | 4 | 3 | 2 | 1 | 0 |  |
| RESERVED |  |  | CCS_STOP_INDEX[6:0] |  |  |  |  |
| R-Oh |  | R/W-4h |  |  |  |  |  |

Table 7-54. ADC_CCS_CFG_0 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $14-8$ | CCS_START_INDEX[6:0] | R/W | Oh | Starting index pointer |
| $6-0$ | CCS_STOP_INDEX[6:0] | R/W | 4 h | Stopping index pointer. Must not be less than the <br> CCS_START_INDEX. |

### 7.5 DAC Configuration Register Map

Table 7-55. Page 3: DAC Configuration Register Map

| $\begin{aligned} & \text { ADDR } \\ & \text { (HEX) } \end{aligned}$ | REGISTER | TYPE | $\begin{aligned} & \text { RESET } \\ & \text { (HEX) } \end{aligned}$ | BIT DESCRIPTION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 40 | DAC CURRENT | R/W | 0000 | DACB3_CURRENT[1:0] |  | DACB2_CURRENT[1:0] |  | DACB1_CURRENT[1:0] |  | DACB0_CURRENT[1:0] |  | DACA3_CURRENT[1:0] |  | DACA2_CURRENT[1:0] |  | DACA1_CURRENT[1:0] |  | DACAO_CURRENT[1:0] |  |
| 41 | $\begin{gathered} \text { DAC_SYNC_ } \\ \text { CFG } \end{gathered}$ | R/W | 0000 | $\begin{aligned} & \text { BCEN } \\ & \text { DACB } \end{aligned}$ | $\begin{aligned} & \text { BCEN } \\ & \text { DACB2 } \end{aligned}$ | $\begin{aligned} & \text { BCEN } \\ & \text { DACB1 } \end{aligned}$ | $\begin{aligned} & \text { BCEN } \\ & \text { DACBO } \end{aligned}$ | $\begin{aligned} & \text { BCEN- } \\ & \text { DACA } \end{aligned}$ | $\begin{aligned} & \text { BCEN } \\ & \text { DACÁa } \end{aligned}$ | $\begin{aligned} & \text { BCEN } \\ & \text { DACA1 } \end{aligned}$ | $\begin{aligned} & \text { BCEN } \\ & \text { DACAO } \end{aligned}$ | SYNCEN DACB3 | SYNCEN DACB2 | $\begin{gathered} \text { SYNCEN_- } \\ \text { DACB1 } \end{gathered}$ | SYNCEN_ DACB0 | $\begin{gathered} \text { SYNCEN_ } \\ \text { DACA3 } \end{gathered}$ | SYNCEN DACA2 | $\begin{gathered} \text { SYNCEN_ } \\ \text { DACA1 } \end{gathered}$ | SYNCEN DACAO |
| 42 | DAC_CFG | R/W | 0000 | RESERVED | DACB BIPOLAR | RESERVED | DACA BIPOLAR | RESERVED |  |  |  |  |  |  |  | CLAMP SEL OUTB2 | $\begin{aligned} & \text { CLAMP_ }_{\text {SEL }} \\ & \text { OUTBO } \end{aligned}$ | CLAMP SEL OUTA2 | $\begin{aligned} & \text { CLAMP_ }_{\text {SEL }} \\ & \text { OUTAO } \end{aligned}$ |
| 43 | DAC_APD_EN | R/W | AAFF | $\begin{gathered} \text { APD } \\ \text { EN_OUTB2[1:0] } \end{gathered}$ |  | APDEN_OUTB0[1:0] |  | $\begin{gathered} \text { APD } \\ \text { EN_OUTAㄹ[1:0] } \end{gathered}$ |  | $\begin{gathered} \text { APD } \\ \text { EN_OUTAO[1:0] } \end{gathered}$ |  | $\begin{gathered} \hline \text { APD_EN } \\ \text { DACB3 } \end{gathered}$ | APD_EN_ DACB2 | $\begin{gathered} \hline \text { APD_EN } \\ \text { DACB1 } \end{gathered}$ | $\begin{gathered} \text { APD_EN_ } \\ \text { DACBO } \end{gathered}$ | $\begin{gathered} \hline \text { APD_EN } \\ \text { DACA3 } \end{gathered}$ | $\begin{gathered} \text { APD_EN_ } \\ \text { DACA2 } \end{gathered}$ | $\begin{aligned} & \hline \text { APD_EN } \\ & \text { DACA1 } \end{aligned}$ | APD EN DACAO |
| 44 | DACA_APD_ SRC﹎﹎ | R/W | 0000 | RESERVED |  |  |  |  |  |  | TMP ALR_APD | RESERVED |  | ADC1 ALR_APD | $\begin{gathered} \text { ADCO_- } \\ \text { ALR_APD } \end{gathered}$ | RESERVED |  | SENSE1 ALR_APD | $\begin{aligned} & \hline \text { SENSEO- } \\ & \text { ALR_APD } \end{aligned}$ |
| 45 | DACA_APD_ SRC_1 | R/W | 1833 | RESER | VED | ALARMIN ALR_APD | $\stackrel{R E F}{\text { ALR_APD }}$ | $\begin{aligned} & \text { THERM } \\ & \text { ERR } \\ & \text { ALR_APD } \end{aligned}$ |  | RESERVED |  |  |  | VSSB ALR_APD | VSSA ALR_APD | RESERVED |  | VCCB ALR_APD | VCCA ALR_APD |
| 46 | $\begin{aligned} & \hline \text { OUTA_APD_ } \\ & \text { SRC_0 } \end{aligned}$ | R/W | 0000 | RESERVED |  |  |  |  |  |  | $\begin{aligned} & \text { TMP_̄ } \\ & \text { ALR_APD } \end{aligned}$ | RESERVED |  | ADC1 ALR_APD | ADCO ALR_APD | RESERVED |  | SENSE1 ALR_APD | $\begin{aligned} & \hline \text { SENSEO- } \\ & \text { ALR_APD } \end{aligned}$ |
| 47 | OUTA_APD_ SRC_1 | R/W | 1833 | RESER | VED | ALARMIN_ ALR_APD | $\begin{aligned} & \text { REF_-̄ } \\ & \text { ALR_APD } \end{aligned}$ | $\begin{aligned} & \text { THERM } \\ & \text { ERR } \\ & \text { ALR_APD } \end{aligned}$ |  | RESERVED |  |  |  | VSSB ALR_APD | VSSA ALR_APD | RESERVED |  | VCCB ALR_APD | VCCA ALR_APD |
| 48 | $\begin{gathered} \hline \text { DACB_APD_ } \\ \text { SRC_0 } \end{gathered}$ | R/W | 0000 | RESERVED |  |  |  |  |  |  | TMP ALR_APD | RESERVED |  | $\begin{aligned} & \text { ADC1 } \\ & \text { ALR_APD } \end{aligned}$ | $\begin{aligned} & \text { ADCO_- } \\ & \text { ALR_APD } \end{aligned}$ | RESERVED |  | $\begin{aligned} & \text { SENSE1 } \\ & \text { ALR_APD } \end{aligned}$ | $\begin{aligned} & \text { SENSEO_ } \\ & \text { ALR_APD } \end{aligned}$ |
| 49 | $\begin{gathered} \text { DACB_APD_ } \\ \text { SRC_1 } \end{gathered}$ | R/W | 1833 | RESER | VED | ALARMIN ALR_APD | $\stackrel{R E F}{\text { ALR_APD }}$ | $\begin{aligned} & \text { THERM } \\ & \text { ERR } \\ & \text { ALR_APD } \end{aligned}$ |  | RESERVED |  |  |  | VSSB ALR_APD | VSSA ALR_APD | RESERVED |  | VCCB ALR_APD | VCCA ALR_APD |
| 4A | $\begin{aligned} & \hline \text { OUTB_APD_ } \\ & \text { SRC_0 } \end{aligned}$ | R/W | 0000 | RESERVED |  |  |  |  |  |  | TMP ALR_APD | RESERVED |  | ADC1 ALR_APD | ADCO ALR_APD | RESERVED |  | SENSE1 ALR_APD | SENSEO <br> ALR_APD |
| 4B | OUTB_APD_ SRC_1 | R/W | 1833 | RESER | VED | ALARMIN_ ALR_APD | $\begin{aligned} & \text { REF_- } \\ & \text { ALR_APD } \end{aligned}$ | $\begin{aligned} & \text { THERM } \\ & \text { ERR- } \\ & \text { ALR_APD } \end{aligned}$ |  | RESERVED |  |  |  | VSSB ALR_APD | VSSA ALR_APD | RESERVED |  | VCCB ALR_APD | VCCA ALR_APD |
| 4C | $\begin{gathered} \hline \text { DAC_CODE_ } \\ \text { LIMIT_0 } \end{gathered}$ | R/W | 3F3F | RESER | VED | DACA1_LIMIT[5:0] |  |  |  |  |  | RESERVED |  | DACAO_LIMIT[5:0] |  |  |  |  |  |
| 4D | DAC_CODE_ LIMIT_1 | R/W | 3F3F | RESER | VED | DACA3_LIMIT[5:0] |  |  |  |  |  | RESERVED |  | DACA2_LIMITT5:0] |  |  |  |  |  |
| 4E | DAC_CODE LIMIT_2 | R/W | 3F3F | RESER | VED | DACB1_LIMIT[5:0] |  |  |  |  |  | RESERVED |  | DACBO_LIMIT[5:0] |  |  |  |  |  |
| 4F | $\begin{gathered} \hline \text { DAC_CODE_- } \\ \text { LIMIT_3 } \end{gathered}$ | R/W | 3F3F | RESER | VED | DACB3_LIMIT[5:0] |  |  |  |  |  | RESERVED |  | DACB2_LIMIT[5:0] |  |  |  |  |  |
| 50 | DRVENO_EN | R/W | 0000 | RESERVED |  |  |  |  |  |  |  | $\begin{gathered} \text { DRVENO_ } \\ \text { EN } \\ \text { DACB̄3 } \end{gathered}$ | $\begin{aligned} & \text { DRVENO_ } \\ & \text { EN- } \\ & \text { DACB2 } 2 \end{aligned}$ | $\begin{aligned} & \text { DRVENO_ } \\ & \text { EN } \\ & \text { DACB } 1 \end{aligned}$ | $\begin{aligned} & \text { DRVENO_ } \\ & \text { EN- } \\ & \text { DACBO } \end{aligned}$ | $\begin{gathered} \text { DRVENO_ } \\ \text { EN- } \\ \text { DACA } 3 \end{gathered}$ | $\begin{aligned} & \text { DRVENO_ } \\ & \text { EN- } \\ & \text { DACAC2 } \end{aligned}$ | $\begin{gathered} \text { DRVENO_ } \\ \text { EN- } \\ \text { DACA } 1 \end{gathered}$ | $\begin{aligned} & \text { DRVENO_ } \\ & \text { EN_ } \\ & \text { DACAOO } \end{aligned}$ |
| 51 | DRVEN1_EN | R/W | 0000 | RESERVED |  |  |  |  |  |  |  | $\begin{gathered} \hline \text { DRVEN1_ } \\ \text { EN- } \\ \text { DACB3 } \end{gathered}$ | $\begin{gathered} \hline \text { DRVEN1_ } \\ \text { EN }-\overline{3} \text { 2 } \\ \text { DACB2 } \end{gathered}$ | $\begin{gathered} \hline \text { DRVEN1_ } \\ \text { EN }-\overline{\text { DACB1 }} \end{gathered}$ | $\begin{gathered} \hline \text { DRVEN1_ } \\ \text { EN }-\bar{B} 0 \\ \text { DACB0 } \end{gathered}$ | $\begin{gathered} \hline \text { DRVEN1_ } \\ \text { EN- } \\ \text { DACAA } \end{gathered}$ | $\begin{gathered} \hline \text { DRVEN1_ } \\ \text { EN }-\overline{2} \\ \text { DACA2 } \end{gathered}$ | $\begin{gathered} \hline \text { DRVEN1_ } \\ \text { EN- } \\ \text { DACAA1 } \end{gathered}$ | $\begin{gathered} \hline \text { DRVEN1_ } \\ \text { EN }-\overline{1} \\ \text { DACAO } \end{gathered}$ |
| 52 | FLEXIO_EN | R/W | 0000 | RESERVED |  |  |  |  |  |  |  | $\begin{gathered} \hline \text { FLEXIO_ } \\ \text { EN_- } \\ \text { DACB3 } \end{gathered}$ | $\begin{gathered} \hline \text { FLEXIO_ } \\ \text { EN- } \\ \text { DACB2 } \end{gathered}$ | $\begin{gathered} \hline \text { FLEXIO_ } \\ \text { EN- } \\ \text { DACB1 } \end{gathered}$ | $\begin{gathered} \hline \text { FLEXIO_ } \\ \text { EN- } \\ \text { DACB̄O } \end{gathered}$ | $\begin{gathered} \hline \text { FLEXIO_ } \\ \text { EN_- } \\ \text { DACAB } \end{gathered}$ | $\begin{gathered} \hline \text { FLEXIO_ } \\ \text { EN_- } \\ \text { DACA2 } \end{gathered}$ | $\begin{gathered} \hline \text { FLEXIO_ } \\ \text { EN_- } \\ \text { DACA1 } \end{gathered}$ | $\begin{gathered} \hline \text { FLEXIO_ } \\ \text { EN_- } \\ \text { DACAOO } \end{gathered}$ |

### 7.5.1 DAC Configuration Registers: Page 3

### 7.5.1.1 DAC_CURRENT Register (address $=40 \mathrm{~h})$ [reset $=0000 \mathrm{~h}]$

Figure 7-48. DAC_CURRENT Register

| 15 | 14 | 13 | 12 | 10 | 9 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| DACB3_CURRENT[1:0] | DACB2_CURRENT[1:0] | DACB1_CURRENT[1:0] | DACB0_CURRENT[1:0] |  |  |
| R/W-Oh | R/W-0h | R/W-0h | R/W-Oh |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 |
| DACA3_CURRENT[1:0] | DACA2_CURRENT[1:0] | DACA1_CURRENT[1:0] | DACAO_CURRENT[1:0] |  |  |
| R/W-Oh | R/W-0h | R/W-0h | R/W-0h |  |  |

Table 7-56. DAC_CURRENT Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 15-14 | DACB3_CURRENT | R/W | Oh | DAC output current mode selection. <br> 00: Start-up, 15mA <br> 01: Low current mode, 30mA <br> 10: Normal current mode, 90 mA <br> 11: High current mode, 120 mA |
| 13-12 | DACB2_CURRENT | R/W | Oh | DAC output current mode selection. <br> 00: Start-up, 15mA <br> 01: Low current mode, 30 mA <br> 10: Normal current mode, 90 mA <br> 11: High current mode, 120 mA |
| 11-10 | DACB1_CURRENT | R/W | Oh | DAC output current mode selection. <br> 00: Start-up, 15mA <br> 01: Low current mode, 30mA <br> 10: Normal current mode, 90 mA <br> 11: High current mode, 120 mA |
| 9-8 | DACBO_CURRENT | R/W | Oh | DAC output current mode selection. <br> 00: Start-up, 15mA <br> 01: Low current mode, 30 mA <br> 10: Normal current mode, 90 mA <br> 11: High current mode, 120 mA |
| 7-6 | DACA3_CURRENT | R/W | Oh | DAC output current mode selection. <br> 00: Start-up, 15mA <br> 01: Low current mode, 30mA <br> 10: Normal current mode, 90 mA <br> 11: High current mode, 120 mA |
| 5-4 | DACA2_CURRENT | R/W | Oh | DAC output current mode selection. <br> 00: Start-up, 15mA <br> 01: Low current mode, 30 mA <br> 10: Normal current mode, 90 mA <br> 11: High current mode, 120 mA |
| 3-2 | DACA1_CURRENT | R/W | Oh | DAC output current mode selection. <br> 00: Start-up, 15mA <br> 01: Low current mode, 30mA <br> 10: Normal current mode, 90 mA <br> 11: High current mode, 120 mA |
| 1-0 | DACAO_CURRENT | R/W | Oh | DAC output current mode selection. <br> 00: Start-up, 15mA <br> 01: Low current mode, 30 mA <br> 10: Normal current mode, 90 mA <br> 11: High current mode, 120 mA |

### 7.5.1.2 DAC_SYNC_CFG Register (address = 41h) [reset = 0000h]

Figure 7-49. DAC_SYNC_CFG Register

| 15 | 1413 |  | 12 | 1110 |  | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BCEN_DACB3 | BCEN_DACB2 | BCEN_DACB1 | BCEN_DACB0 | BCEN_DACA3 | BCEN_DACA2 | BCEN_DACA1 | BCEN_DACA0 |
| R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| SYNCEN DACB3 | SYNCEN DACB2 | SYNCEN DACB1 | SYNCEN DACB0 | SYNCEN DACA3 | SYNCEN DACA2 | SYNCEN DACA1 | SYNCEN DACAO |
| R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh |

Table 7-57. DAC_SYNC_CFG Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 15 | BCEN_DACB3 | R/W | Oh | DAC broadcast enable. <br> 0: Ignores broadcast writes on this DAC <br> 1: Allow broadcast writes on this DAC |
| 14 | BCEN_DACB2 | R/W | Oh |  |
| 13 | BCEN_DACB1 | R/W | Oh |  |
| 12 | BCEN_DACB0 | R/W | Oh |  |
| 11 | BCEN_DACA3 | R/W | Oh |  |
| 10 | BCEN_DACA2 | R/W | Oh |  |
| 9 | BCEN_DACA1 | R/W | Oh |  |
| 8 | BCEN_DACAO | R/W | Oh |  |
| 7 | SYNCEN_DACB3 | R/W | Oh | DAC synchronous configuration. <br> 0: Set DAC into asynchronous mode. <br> 1: Set DAC into synchronous mode. |
| 6 | SYNCEN_DACB2 | R/W | Oh |  |
| 5 | SYNCEN_DACB1 | R/W | Oh |  |
| 4 | SYNCEN_DACB0 | R/W | Oh |  |
| 3 | SYNCEN_DACA3 | R/W | Oh |  |
| 2 | SYNCEN_DACA2 | R/W | Oh |  |
| 1 | SYNCEN_DACA1 | R/W | Oh |  |
| 0 | SYNCEN_DACAO | R/W | Oh |  |

7.5.1.3 DAC_CFG Register (address $=42 \mathrm{~h}$ ) [reset $=0000 \mathrm{~h}]$

Figure 7-50. DAC_CFG Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED | $\begin{aligned} & \text { DACB_ }_{-} \\ & \text {BIPOLAR } \end{aligned}$ | RESERVED | $\begin{aligned} & \text { DACA } \\ & \text { BIPOLAR } \end{aligned}$ | RESERVED |  |  |  |
| R-Oh | R/W-Oh | R-Oh | R/W-Oh | R-Oh |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| RESERVED |  |  |  | CLAMP_SEL OUTB2 | $\begin{gathered} \text { CLAMP_SEL_ } \\ \text { OUTBO } \end{gathered}$ | $\begin{aligned} & \text { CLAMP_SEL_ } \\ & \text { OUTA2 } \end{aligned}$ | $\begin{gathered} \text { CLAMP_SEL_ } \\ \text { OUTAO } \end{gathered}$ |
| R-Oh |  |  |  | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh |

Table 7-58. DAC_CFG Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| 14 | DACB_BIPOLAR | R/W | Oh | Used to configure DAC group B for bipolar operation. <br> o: Unipolar operation <br> 1: Bipolar operation |
| 12 | DACA_BIPOLAR | R/W | Oh | Used to configure DAC group A for bipolar operation. <br> 0: Unipolar operation <br> 1: Bipolar operation |
| 3 | CLAMP_SEL_OUTB2 | R/W | Oh | Clamp voltage selection for OUTB2. <br> 0: Clamp voltage is VSSB <br> 1: Clamp voltage is DACB3 |
| 2 | CLAMP_SEL_OUTB0 | R/W | Oh | Clamp voltage selection for OUTB0. <br> 0: Clamp voltage is VSSB <br> 1: Clamp voltage is DACB1 |
| 1 | CLAMP_SEL_OUTA2 | R/W | Oh | Clamp voltage selection for OUTA2. <br> 0: Clamp voltage is VSSA <br> 1: Clamp voltage is DACA3 |
| 0 | CLAMP_SEL_OUTA0 | R/W | Oh | Clamp voltage selection for OUTA0. <br> 0: Clamp voltage is VSSA <br> 1: Clamp voltage is DACA1 |

### 7.5.1.4 DAC_APD_EN Register (address = 43h) [reset = AAFFh]

Figure 7-51. DAC_APD_EN Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| APD_E | [2[1:0] | APD_EN_OUTB0[1:0] |  | APD_EN_OUTA2[1:0] |  | APD_EN_OUTA0[1:0] |  |
| R/W-2h |  | R/W-2h |  | R/W-2h |  | R/W-2h |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | $\begin{gathered} \mathrm{APD}_{-} \\ \mathrm{EN}-\overline{\mathrm{C}} 2 \\ \mathrm{DAC} \end{gathered}$ |  |  |  |  |  | $\begin{aligned} & \mathrm{APD}_{-} \\ & \mathrm{EN} \\ & \mathrm{DACA} 0 \end{aligned}$ |
| R/W-1h | R/W-1h | R/W-1h | R/W-1h | R/W-1h | R/W-1h | R/W-1h | R/W-1h |

Table 7-59. DAC_APD_EN Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 15-14 | APD_EN_OUTB2 | R/W | 2h | OUTB pin auto-power-down enable <br> 00: Ignore auto-power-down events on the OUTB pin. <br> 10: Disable OUTB pin drive channel, and connect to VSSB during an auto-power-down event. <br> 11: Disable OUTB pin drive channel, and connect to VSSB or DAC output (depending on clamp setting) during an auto-powerdown event. |
| 13-12 | APD_EN_OUTB0 | R/W | 2h |  |
| 11-10 | APD_EN_OUTA2 | R/W | 2h | OUTA pin auto-power-down enable <br> 00: Ignore auto-power-down events on the OUTA pin. <br> 10: Disable OUTA pin drive channel, and connect to VSSA during an auto-power-down event. <br> 11: Disable OUTA pin drive channel, and connect to VSSA or DAC output (depending on clamp setting) during an auto-powerdown event. |
| 9-8 | APD_EN_OUTA0 | R/W | 2h |  |
| 7 | APD_EN_DACB3 | R/W | 1h | DAC pin auto-power-down enable <br> 0 : Ignore auto-power-down events on the DAC. <br> 1: Force this DAC to power-down during an auto-power-down event. |
| 6 | APD_EN_DACB2 | R/W | 1h |  |
| 5 | APD_EN_DACB1 | R/W | 1h |  |
| 4 | APD_EN_DACB0 | R/W | 1h |  |
| 3 | APD_EN_DACA3 | R/W | 1h |  |
| 2 | APD_EN_DACA2 | R/W | 1h |  |
| 1 | APD_EN_DACA1 | R/W | 1h |  |
| 0 | APD_EN_DACA0 | R/W | 1h |  |

### 7.5.1.5 DACA_APD_SRC_0 Register (address = 44h) [reset = 0000h]

Figure 7-52. DACA_APD_SRC_0 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | RESERVED |  |  |  |  |

Table 7-60. DACA_APD_SRC_0 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| 8 | TMP_ALR_APD | R/W | Oh | This bit determines if group A DACs are forced into a power- <br> down state by this alarm. The respective DACA channels must <br> be enabled in the DAC_APD_EN register. <br> 0: Temperature alarm does not trigger DACA auto-power-down <br> event <br> 1: Temperature alarm triggers DACA auto-power-down event |
| 5 | ADC1_ALR_APD | R/W | Oh | This bit determines if group A DACs are forced into a power- <br> down state by this alarm. The respective DACA channels must <br> be enabled in the DAC_APD_EN register. <br> 0: ADC1 alarm does not trigger DACA auto-power-down event <br> 1: ADC1 alarm triggers DACA auto-power-down event |
| 4 | ADC0_ALR_APD | R/W | Oh | This bit determines if group A DACs are forced into a power- <br> down state by this alarm. The respective DACA channels must <br> be enabled in the DAC_APD_EN register. <br> 0: ADC0 alarm does not trigger DACA auto-power-down event <br> 1: ADC0 alarm triggers DACA auto-power-down event |
| 1 | SENSE1_ALR_APD | R/W | Oh | This bit determines if group A DACs are forced into a power- <br> down state by this alarm. The respective DACA channels must <br> be enabled in the DAC_APD_EN register. <br> 0: SENSE1 alarm does not trigger DACA auto-power-down <br> event <br> 1: SENSE1 alarm triggers DACA auto-power-down event |
| 0 | SENSEO_ALR_APD | R/W | Oh | This bit determines if group A DACs are forced into a power- <br> down state by this alarm. The respective DACA channels must <br> be enabled in the DAC_APD_EN register. <br> 0: SENSEO alarm does not trigger DACA auto-power-down <br> event <br> 1: SENSE0 alarm triggers DACA auto-power-down event |

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7.5.1.6 DACA_APD_SRC_1 Register (address = 45h) [reset = 1833h]

Figure 7-53. DACA_APD_SRC_1 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  | ALARMIN ALR_APD | REF <br> ALR_APD | THERMERR_ ALR_APD |  | RESERVED |  |
| R-Oh |  | R/W-Oh | R/W-1h | R/W-1h |  | R-Oh |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| RESERVED |  | $\begin{aligned} & \text { VSSB_- } \\ & \text { ALR_APD } \end{aligned}$ | $\begin{aligned} & \text { VSSA } \\ & \text { ALR_APD } \end{aligned}$ | RESERVED |  | $\begin{aligned} & \text { VCCB_ } \\ & \text { ALR_APD } \end{aligned}$ | $\begin{aligned} & \text { VCCA } \\ & \text { ALR_APD } \end{aligned}$ |
| R-Oh |  | R/W-1h | R/W-1h | R-Oh |  | R/W-1h | R/W-1h |

Table 7-61. DACA_APD_SRC_1 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 13 | ALARMIN_ALR_APD | R/W | Oh | This bit determines if group A DACs are forced into a powerdown state by this alarm. The respective DACA channels must be enabled in the DAC_APD_EN register. <br> 0: $\overline{\text { ALARMIN }}$ alarm does not trigger DACA auto-power-down event <br> 1: $\overline{\text { ALARMIN }}$ alarm triggers DACA auto-power-down event |
| 12 | REF_ALR_APD | R/W | 1h | This bit determines if group A DACs are forced into a powerdown state by this alarm. The respective DACA channels must be enabled in the DAC_APD_EN register. <br> 0: Reference alarm does not trigger DACA auto-power-down event <br> 1: Reference alarm triggers DACA auto-power-down event |
| 11 | THERMERR_ALR_APD | R/W | 1h | This bit determines if group A DACs are forced into a powerdown state by this alarm. The respective DACA channels must be enabled in the DAC_APD_EN register. <br> 0 : Thermal error alarm does not trigger DACA auto-power-down event <br> 1: Thermal error alarm triggers DACA auto-power-down event |
| 5 | VSSB_ALR_APD | R/W | 1h | This bit determines if group A DACs are forced into a powerdown state by this alarm. The respective DACA channels must be enabled in the DAC_APD_EN register. <br> 0: VSSB alarm does not trigger DACA auto-power-down event <br> 1: VSSB alarm triggers DACA auto-power-down event |
| 4 | VSSA_ALR_APD | R/W | 1h | This bit determines if group A DACs are forced into a powerdown state by this alarm. The respective DACA channels must be enabled in the DAC_APD_EN register. <br> 0: VSSA alarm does not trigger DACA auto-power-down event <br> 1: VSSA alarm triggers DACA auto-power-down event |
| 1 | VCCB_ALR_APD | R/W | 1h | This bit determines if group A DACs are forced into a powerdown state by this alarm. The respective DACA channels must be enabled in the DAC_APD_EN register.0: VCCB alarm does not trigger DACA auto-power-down event <br> 1: VCCB alarm triggers DACA auto-power-down event |
| 0 | VCCA_ALR_APD | R/W | 1h | This bit determines if group A DACs are forced into a powerdown state by this alarm. The respective DACA channels must be enabled in the DAC_APD_EN register. <br> 0: VCCA alarm does not trigger DACA auto-power-down event <br> 1: VCCA alarm triggers DACA auto-power-down event |

### 7.5.1.7 OUTA_APD_SRC_0 Register (address = 46h) [reset = 0000h]

Figure 7-54. OUTA_APD_SRC_0 Register

| 15 | 14 | 13 | 12 | 11 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  | $\begin{gathered} \text { TMP } \\ \text { ALR_APD } \end{gathered}$ |
| R-Oh |  |  |  |  |  | R/W-Oh |
| 7 | 6 | 5 | 4 | 3 | 1 | 0 |
| RESERVED |  | $\begin{gathered} \text { ADC1 } \\ \text { ALR_APD } \end{gathered}$ | $\begin{aligned} & \text { ADC0 } \\ & \text { ALR_APD } \end{aligned}$ |  | SENSE1 ALR_APD | $\begin{aligned} & \text { SENSEO } \\ & \text { ALR_APD } \end{aligned}$ |
| R-Oh |  | R/W-Oh | R/W-Oh |  | R/W-Oh | R/W-Oh |

Table 7-62. OUTA_APD_SRC_0 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| 8 | TMP_ALR_APD | R/W | Oh | This bit determines if group A OUT pins are forced into a power- <br> down state by this alarm. The respective OUTA pins must be <br> enabled in the DAC_APD_EN register. <br> 0: Temperature alarm does not trigger OUTA pin auto-power- <br> down event <br> 1: Temperature alarm triggers OUTA pin auto-power-down event |
| 5 | ADC1_ALR_APD |  | R/W | Oh |
| 4 | ADC0_ALR_APD | This bit determines if group A OUT pins are forced into a power- <br> down state by this alarm. The respective OUTA pins must be <br> enabled in the DAC_APD_EN register. <br> 0: ADC1 alarm does not trigger OUTA pin auto-power-down <br> event <br> 1: ADC1 alarm triggers OUTA pin auto-power-down event |  |  |
| 1 | SENSE1_ALR_APD | R/W | Oh | This bit determines if group A OUT pins are forced into a power- <br> down state by this alarm. The respective OUTA pins must be <br> enabled in the DAC_APD_EN register. <br> 0: ADC0 alarm does not trigger OUTA pin auto-power-down <br> event <br> 1: ADC0 alarm triggers OUTA pin auto-power-down event |
| 0 | SENSEO_ALR_APD | R/W | Oh | This bit determines if group A OUT pins are forced into a power- <br> dewn state by this alarm. The respective OUTA pins must be <br> enabled in the DAC_APD_EN register. <br> 0: SENSE1 alarm does not trigger OUTA pin auto-power-down <br> event <br> 1: SENSE1 alarm triggers OUTA pin auto-power-down event |

### 7.5.1.8 OUTA_APD_SRC_1 Register (address = 47h) [reset = 1833h]

Figure 7-55. OUTA_APD_SRC_1 Register

| 15 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | ALARMIN ALR_APD | REF ALR_APD | THERMERR_ ALR_APD |  | RESERVED |  |
| R-Oh | R/W-Oh | R/W-1h | R/W-1h |  | R-Oh |  |
| 7 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | $\begin{aligned} & \text { VSSB_} \\ & \text { ALR_APD } \end{aligned}$ | $\begin{aligned} & \text { VSSA } \\ & \text { ALR_APD } \end{aligned}$ | RESERVED |  | $\begin{aligned} & \text { VCCB } \\ & \text { ALR_APD } \end{aligned}$ | $\begin{aligned} & \text { VCCA } \\ & \text { ALR_APD } \end{aligned}$ |
| R-Oh | R/W-1h | R/W-1h | R-Oh |  | R/W-1h | R/W-1h |

Table 7-63. OUTA_APD_SRC_1 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 13 | ALARMIN_ALR_APD | R/W | Oh | This bit determines if group A OUT pins are forced into a powerdown state by this alarm. The respective OUTA pins must be enabled in the DAC_APD_EN register. <br> 0 : $\overline{\text { ALARMIN }}$ alarm does not trigger OUTA pin auto-power-down event <br> 1: $\overline{\text { ALARMIN }}$ alarm triggers OUTA pin auto-power-down event |
| 12 | REF_ALR_APD | R/W | 1h | This bit determines if group A OUT pins are forced into a powerdown state by this alarm. The respective OUTA pins must be enabled in the DAC_APD_EN register. <br> 0 : Reference alarm does not trigger OUTA pin auto-power-down event <br> 1: Reference alarm triggers OUTA pin auto-power-down event |
| 11 | THERMERR_ALR_APD | R/W | 1h | This bit determines if group A OUT pins are forced into a powerdown state by this alarm. The respective OUTA pins must be enabled in the DAC_APD_EN register. <br> 0 : Thermal error alarm does not trigger OUTA pin auto-powerdown event <br> 1: Thermal error alarm triggers OUTA pin auto-power-down event |
| 5 | VSSB_ALR_APD | R/W | 1h | This bit determines if group A OUT pins are forced into a powerdown state by this alarm. The respective OUTA pins must be enabled in the DAC_APD_EN register. <br> 0 : VSSB alarm does not trigger OUTA pin auto-power-down event <br> 1: VSSB alarm triggers OUTA pin auto-power-down event |
| 4 | VSSA_ALR_APD | R/W | 1h | This bit determines if group A OUT pins are forced into a powerdown state by this alarm. The respective OUTA pins must be enabled in the DAC_APD_EN register. <br> 0: VSSA alarm does not trigger OUTA pin auto-power-down event <br> 1: VSSA alarm triggers OUTA pin auto-power-down event |
| 1 | VCCB_ALR_APD | R/W | 1h | This bit determines if group A OUT pins are forced into a powerdown state by this alarm. The respective OUTA pins must be enabled in the DAC_APD_EN register. <br> 0: VCCB alarm does not trigger OUTA pin auto-power-down event <br> 1: VCCB alarm triggers OUTA pin auto-power-down event |
| 0 | VCCA_ALR_APD | R/W | 1h | This bit determines if group A OUT pins are forced into a powerdown state by this alarm. The respective OUTA pins must be enabled in the DAC_APD_EN register. <br> 0: VCCA alarm does not trigger OUTA pin auto-power-down event <br> 1: VCCA alarm triggers OUTA pin auto-power-down event |

7.5.1.9 DACB_APD_SRC_0 Register (address = 48h) [reset $=0000 \mathrm{~h}]$

Figure 7-56. DACB_APD_SRC_0 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | RESERVED |  |  |  |  |

Table 7-64. DACB_APD_SRC_0 Register Field Descriptions
$\left.\begin{array}{|c|l|l|l|l|}\hline \text { Bit } & \text { Field } & \text { Type } & \text { Reset } & \text { Description } \\ \hline 8 & \text { TMP_ALR_APD } & \text { R/W } & \text { Oh } & \begin{array}{l}\text { This bit determines if group B DAC pins are forced into a power- } \\ \text { down state by this alarm. The respective DACB pins must be } \\ \text { enabled in the DAC_APD_EN register. } \\ \text { 0: Temperature alarm does not trigger DACB auto-power-down } \\ \text { event } \\ \text { 1: Temperature alarm triggers DACB auto-power-down event }\end{array} \\ \hline 5 & \text { ADC1_ALR_APD } & & \text { R/W } & \text { Oh }\end{array} \begin{array}{l}\text { This bit determines if group B DAC pins are forced into a power- } \\ \text { down state by this alarm. The respective DACB pins must be } \\ \text { enabled in the DAC_APD_EN register. } \\ \text { 0: ADC1 alarm does not trigger DACB auto-power-down event } \\ \text { 1: ADC1 alarm triggers DACB auto-power-down event }\end{array}\right]$

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7.5.1.10 DACB_APD_SRC_1 Register (address = 49h) [reset = 1833h]

Figure 7-57. DACB_APD_SRC_1 Register


Table 7-65. DACB_APD_SRC_1 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 13 | ALARMIN_ALR_APD | R/W | Oh | This bit determines if group B DACs are forced into a powerdown state by this alarm. The respective DACB channels must be enabled in the DAC_APD_EN register. <br> 0: $\overline{\text { ALARMIN }}$ alarm does not trigger DACB auto-power-down event <br> 1: $\overline{\text { ALARMIN alarm triggers DACB auto-power-down event }}$ |
| 12 | REF_ALR_APD | R/W | 1h | This bit determines if group B DACs are forced into a powerdown state by this alarm. The respective DACB channels must be enabled in the DAC_APD_EN register. <br> 0: Reference alarm does not trigger DACB auto-power-down event <br> 1: Reference alarm triggers DACB auto-power-down event |
| 11 | THERMERR_ALR_APD | R/W | 1h | This bit determines if group B DACs are forced into a powerdown state by this alarm. The respective DACB channels must be enabled in the DAC_APD_EN register. <br> 0 : Thermal error alarm does not trigger DACB auto-power-down event <br> 1: Thermal error alarm triggers DACB auto-power-down event |
| 5 | VSSB_ALR_APD | R/W | 1h | This bit determines if group B DACs are forced into a powerdown state by this alarm. The respective DACB channels must be enabled in the DAC_APD_EN register. <br> 0: VSSB alarm does not trigger DACB auto-power-down event <br> 1: VSSB alarm triggers DACB auto-power-down event |
| 4 | VSSA_ALR_APD | R/W | 1h | This bit determines if group B DACs are forced into a powerdown state by this alarm. The respective DACB channels must be enabled in the DAC_APD_EN register. <br> 0: VSSA alarm does not trigger DACB auto-power-down event <br> 1: VSSA alarm triggers DACB auto-power-down event |
| 1 | VCCB_ALR_APD | R/W | 1h | This bit determines if group B DACs are forced into a powerdown state by this alarm. The respective DACB channels must be enabled in the DAC_APD_EN register. <br> 0 : VCCB alarm does not trigger DACB auto-power-down event <br> 1: VCCB alarm triggers DACB auto-power-down event |
| 0 | VCCA_ALR_APD | R/W | 1h | This bit determines if group B DACs are forced into a powerdown state by this alarm. The respective DACB channels must be enabled in the DAC_APD_EN register. <br> 0: VCCA alarm does not trigger DACB auto-power-down event <br> 1: VCCA alarm triggers DACB auto-power-down event |

### 7.5.1.11 OUTB_APD_SRC_0 Register (address = 4Ah) [reset = 0000h]

Figure 7-58. OUTB_APD_SRC_0 Register

| 15 | 14 | 13 | 12 | 11 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  | $\begin{gathered} \text { TMP } \\ \text { ALR_APD } \end{gathered}$ |
| R-Oh |  |  |  |  |  | R/W-Oh |
| 7 | 6 | 5 | 4 | 3 | 1 | 0 |
| RESERVED |  | $\begin{gathered} \text { ADC1 } \\ \text { ALR_APD } \end{gathered}$ | $\begin{aligned} & \text { ADC0 } \\ & \text { ALR_APD } \end{aligned}$ |  | SENSE1 ALR_APD | $\begin{aligned} & \text { SENSEO } \\ & \text { ALR_APD } \end{aligned}$ |
| R-Oh |  | R/W-Oh | R/W-Oh |  | R/W-Oh | R/W-Oh |

Table 7-66. OUTB_APD_SRC_0 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| 8 | TMP_ALR_APD | R/W | Oh | This bit determines if group B OUT pins are forced into a power- <br> down state by this alarm. The respective OUTB pins must be <br> enabled in the DAC_APD_EN register. <br> 0: Temperature alarm does not trigger OUTB pin auto-power- <br> down event <br> 1: Temperature alarm triggers OUTB pin auto-power-down event |
| 5 | ADC1_ALR_APD |  | R/W | Oh |
| 4 | ADC0_ALR_APD | This bit determines if group B OUT pins are forced into a power- <br> down state by this alarm. The respective OUTB pins must be <br> enabled in the DAC_APD_EN register. <br> 0: ADC1 alarm does not trigger OUTB pin auto-power-down <br> event <br> 1: ADC1 alarm triggers OUTB pin auto-power-down event |  |  |
| 1 | SENSE1_ALR_APD | R/W | Oh | This bit determines if group B OUT pins are forced into a power- <br> down state by this alarm. The respective OUTB pins must be <br> enabled in the DAC_APD_EN register. <br> 0: ADC0 alarm does not trigger OUTB pin auto-power-down <br> event <br> 1: ADC0 alarm triggers OUTB pin auto-power-down event |
| 0 | SENSEO_ALR_APD | R/W | Oh | This bit determines if group B OUT pins are forced into a power- <br> dewn state by this alarm. The respective OUTB pins must be <br> enabled in the DAC_APD_EN register. <br> 0: SENSE1 alarm does not trigger OUTB pin auto-power-down <br> event <br> 1: SENSE1 alarm triggers OUTB pin auto-power-down event |

7.5.1.12 OUTB_APD_SRC_1 Register (address = 4Bh) [reset = 1833h]

Figure 7-59. OUTB_APD_SRC_1 Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  | ALARMIN ALR_APD ${ }^{-}$ | REF <br> ALR_APD | THERMERR_ ALR_APD |  | RESERVED |  |
| R-Oh |  | R/W-Oh | R/W-1h | R/W-1h |  | R-Oh |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| RESERVED |  | VSSB <br> ALR_APD | $\begin{aligned} & \text { VSSA } \\ & \text { ALR_A디 } \end{aligned}$ | RESERVED |  | $\begin{gathered} \text { VCCB } \\ \text { ALR_APD } \end{gathered}$ | VCCA ALR_APD |
| R-Oh |  | R/W-1h | R/W-1h | R-Oh |  | R/W-1h | R/W-1h |

Table 7-67. OUTB_APD_SRC_1 Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 13 | ALARMIN_ALR_APD | R/W | Oh | This bit determines if group B OUT pins are forced into a powerdown state by this alarm. The respective OUTB pins must be enabled in the DAC_APD_EN register. <br> 0 : $\overline{\text { ALARMIN }}$ alarm does not trigger OUTB pin auto-power-down event <br> 1: $\overline{\text { ALARMIN }}$ alarm triggers OUTB pin auto-power-down event |
| 12 | REF_ALR_APD | R/W | 1h | This bit determines if group B OUT pins are forced into a powerdown state by this alarm. The respective OUTB pins must be enabled in the DAC_APD_EN register. <br> 0 : Reference alarm does not trigger OUTB pin auto-power-down event <br> 1: Reference alarm triggers OUTB pin auto-power-down event |
| 11 | THERMERR_ALR_APD | R/W | 1h | This bit determines if group B OUT pins are forced into a powerdown state by this alarm. The respective OUTB pins must be enabled in the DAC_APD_EN register. <br> 0 : Thermal error alarm does not trigger OUTB pin auto-powerdown event <br> 1: Thermal error alarm triggers OUTB pin auto-power-down event |
| 5 | VSSB_ALR_APD | R/W | 1h | This bit determines if group B OUT pins are forced into a powerdown state by this alarm. The respective OUTB pins must be enabled in the DAC_APD_EN register. <br> 0 : VSSB alarm does not trigger OUTB pin auto-power-down event <br> 1: VSSB alarm triggers OUTB pin auto-power-down event |
| 4 | VSSA_ALR_APD | R/W | 1h | This bit determines if group B OUT pins are forced into a powerdown state by this alarm. The respective OUTB pins must be enabled in the DAC_APD_EN register. <br> 0 : VSSA alarm does not trigger OUTB pin auto-power-down event <br> 1: VSSA alarm triggers OUTB pin auto-power-down event |
| 1 | VCCB_ALR_APD | R/W | 1h | This bit determines if group B OUT pins are forced into a powerdown state by this alarm. The respective OUTB pins must be enabled in the DAC_APD_EN register. <br> 0 : VCCB alarm does not trigger OUTB pin auto-power-down event <br> 1: VCCB alarm triggers OUTB pin auto-power-down event |
| 0 | VCCA_ALR_APD | R/W | 1h | This bit determines if group B OUT pins are forced into a powerdown state by this alarm. The respective OUTB pins must be enabled in the DAC_APD_EN register. <br> 0: VCCA alarm does not trigger OUTB pin auto-power-down event <br> 1: VCCA alarm triggers OUTB pin auto-power-down event |

7.5.1.13 DAC_CODE_LIMIT_0 Register (address = 4Ch) [reset = 3F3Fh]

Figure 7-60. DAC_CODE_LIMIT_0 Register

| 15 |  | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  | DACA1_LIMITS[5:0] |  |  |  |  |  |
| R-Oh |  |  | R/W-3Fh |  |  |  |  |  |
| 7 | 6 |  | 5 | 4 | 3 | 2 | 1 | 0 |
|  | RESERVED |  | DACA0_LIMITS[5:0] |  |  |  |  |  |
| R-Oh |  |  | R/W-3Fh |  |  |  |  |  |

Table 7-68. DAC_CODE_LIMIT_0 Register Field Descriptions

| Bit | Field | Type | Reset | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 13-8 | DACA1_LIMITS | R/W | 3Fh | DAC active register latch code limit; off by default. Program these bits with the following values to achieve the limit specified for the upper six MSBs of the DAC codes. |  |  |
| 5-0 | DACA0_LIMITS | R/W | 3 Fh |  |  |  |
|  |  |  |  | 00h: 007Fh | 16h: 0B7Fh | 2Bh: 15FFh |
|  |  |  |  | 01h: 00FFh | 17h: OBFFh | 2Ch: 167Fh |
|  |  |  |  | 02h: 017Fh | 18h: 0C7Fh | 2Dh: 16FFh |
|  |  |  |  | 03h: 01FFh | 19h: OCFFh | 2Eh: 177Fh |
|  |  |  |  | 04h: 027Fh | 1Ah: 0D7Fh | 2Fh: 17FFh |
|  |  |  |  | 05h: 02FFh | 1Bh: ODFFh | 30h: 187Fh |
|  |  |  |  | 06h: 037Fh | 1Ch: 0E7Fh | 31h: 18FFh |
|  |  |  |  | 07h: 03FFh | 1Dh: 0EFFh | 32h: 197Fh |
|  |  |  |  | 08h: 047Fh | 1Eh: 0F7Fh | 33h: 19FFh |
|  |  |  |  | 09h: 04FFh | 1Fh: 0FFFh | 34h: 1A7Fh |
|  |  |  |  | 0Ah: 057Fh | 20h: 107Fh | 35h: 1AFFh |
|  |  |  |  | 0Bh: 05FFh | 21h: 10FFh | 36h: 1B7Fh |
|  |  |  |  | 0Ch: 067Fh | 22h: 117Fh | 37h: 1BFFh |
|  |  |  |  | 0Dh: 06FFh | 23h: 11FFh | 38h: 1C7Fh |
|  |  |  |  | 0Eh: 077Fh | 24h: 127Fh | 39h: 1CFFh |
|  |  |  |  | 0Fh: 07FFh | 25h: 12FFh | 3Ah: 1D7Fh |
|  |  |  |  | 10h: 087Fh | 26h: 137Fh | 3Bh: 1DFFh |
|  |  |  |  | 11h: 08FFh | 27h: 13FFh | 3Ch: 1E7Fh |
|  |  |  |  | 12h: 097Fh | 28h: 147Fh | 3Dh: 1EFFh |
|  |  |  |  | 13h: 09FFh | 29h: 14FFh | 3Eh: 1F7Fh |
|  |  |  |  | 14h: 0A7Fh | 2Ah: 157Fh | 3Fh: 1FFFh |
|  |  |  |  | 15h: OAFFh |  |  |

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### 7.5.1.14 DAC_CODE_LIMIT_1 Register (address = 4Dh) [reset = 3F3Fh]

Figure 7-61. DAC_CODE_LIMIT_1 Register

| 15 |  | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  | DACA3_LIMITS[5:0] |  |  |  |  |  |
| R-Oh |  |  | R/W-3Fh |  |  |  |  |  |
| 7 | 6 |  | 5 | 4 | 3 | 2 | 1 | 0 |
|  | RESERVED |  | DACA2_LIMITS[5:0] |  |  |  |  |  |
| R-Oh |  |  | R/W-3Fh |  |  |  |  |  |

Table 7-69. DAC_CODE_LIMIT_1 Register Field Descriptions

| Bit | Field | Type | Reset | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 13-8 | DACA3_LIMITS | R/W | 3Fh | DAC active register latch code limit; off by default. Program these bits with the following values to achieve the limit specified for the upper six MSBs of the DAC codes. |  |  |
| 5-0 | DACA2_LIMITS | R/W | 3 Fh |  |  |  |
|  |  |  |  | 00h: 007Fh | 16h: 0B7Fh | 2Bh: 15FFh |
|  |  |  |  | 01h: 00FFh | 17h: OBFFh | 2Ch: 167Fh |
|  |  |  |  | 02h: 017Fh | 18h: 0C7Fh | 2Dh: 16FFh |
|  |  |  |  | 03h: 01FFh | 19h: OCFFh | 2Eh: 177Fh |
|  |  |  |  | 04h: 027Fh | 1Ah: 0D7Fh | 2Fh: 17FFh |
|  |  |  |  | 05h: 02FFh | 1Bh: ODFFh | 30h: 187Fh |
|  |  |  |  | 06h: 037Fh | 1Ch: 0E7Fh | 31h: 18FFh |
|  |  |  |  | 07h: 03FFh | 1Dh: 0EFFh | 32h: 197Fh |
|  |  |  |  | 08h: 047Fh | 1Eh: 0F7Fh | 33h: 19FFh |
|  |  |  |  | 09h: 04FFh | 1Fh: 0FFFh | 34h: 1A7Fh |
|  |  |  |  | 0Ah: 057Fh | 20h: 107Fh | 35h: 1AFFh |
|  |  |  |  | 0Bh: 05FFh | 21h: 10FFh | 36h: 1B7Fh |
|  |  |  |  | 0Ch: 067Fh | 22h: 117Fh | 37h: 1BFFh |
|  |  |  |  | 0Dh: 06FFh | 23h: 11FFh | 38h: 1C7Fh |
|  |  |  |  | 0Eh: 077Fh | 24h: 127Fh | 39h: 1CFFh |
|  |  |  |  | 0Fh: 07FFh | 25h: 12FFh | 3Ah: 1D7Fh |
|  |  |  |  | 10h: 087Fh | 26h: 137Fh | 3Bh: 1DFFh |
|  |  |  |  | 11h: 08FFh | 27h: 13FFh | 3Ch: 1E7Fh |
|  |  |  |  | 12h: 097Fh | 28h: 147Fh | 3Dh: 1EFFh |
|  |  |  |  | 13h: 09FFh | 29h: 14FFh | 3Eh: 1F7Fh |
|  |  |  |  | 14h: 0A7Fh | 2Ah: 157Fh | 3Fh: 1FFFh |
|  |  |  |  | 15h: OAFFh |  |  |

7.5.1.15 DAC_CODE_LIMIT_2 Register (address = 4Eh) [reset = 3F3Fh]

Figure 7-62. DAC_CODE_LIMIT_2 Register

| 15 |  | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  | DACB1_LIMITS[5:0] |  |  |  |  |  |
| R-Oh |  |  | R/W-3Fh |  |  |  |  |  |
| 7 | 6 |  | 5 | 4 | 3 | 2 | 1 | 0 |
|  | RESERVED |  | DACB0_LIMITS[5:0] |  |  |  |  |  |
| R-Oh |  |  | R/W-3Fh |  |  |  |  |  |

Table 7-70. DAC_CODE_LIMIT_2 Register Field Descriptions

| Bit | Field | Type | Reset | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 13-8 | DACB1_LIMITS | R/W | 3Fh | DAC active register latch code limit; off by default. Program these bits with the following values to achieve the limit specified for the upper six MSBs of the DAC codes. |  |  |
| 5-0 | DACB0_LIMITS | R/W | 3 Fh |  |  |  |
|  |  |  |  | 00h: 007Fh | 16h: 0B7Fh | 2Bh: 15FFh |
|  |  |  |  | 01h: 00FFh | 17h: OBFFh | 2Ch: 167Fh |
|  |  |  |  | 02h: 017Fh | 18h: 0C7Fh | 2Dh: 16FFh |
|  |  |  |  | 03h: 01FFh | 19h: OCFFh | 2Eh: 177Fh |
|  |  |  |  | 04h: 027Fh | 1Ah: 0D7Fh | 2Fh: 17FFh |
|  |  |  |  | 05h: 02FFh | 1Bh: ODFFh | 30h: 187Fh |
|  |  |  |  | 06h: 037Fh | 1Ch: 0E7Fh | 31h: 18FFh |
|  |  |  |  | 07h: 03FFh | 1Dh: 0EFFh | 32h: 197Fh |
|  |  |  |  | 08h: 047Fh | 1Eh: 0F7Fh | 33h: 19FFh |
|  |  |  |  | 09h: 04FFh | 1Fh: 0FFFh | 34h: 1A7Fh |
|  |  |  |  | 0Ah: 057Fh | 20h: 107Fh | 35h: 1AFFh |
|  |  |  |  | 0Bh: 05FFh | 21h: 10FFh | 36h: 1B7Fh |
|  |  |  |  | 0Ch: 067Fh | 22h: 117Fh | 37h: 1BFFh |
|  |  |  |  | 0Dh: 06FFh | 23h: 11FFh | 38h: 1C7Fh |
|  |  |  |  | 0Eh: 077Fh | 24h: 127Fh | 39h: 1CFFh |
|  |  |  |  | 0Fh: 07FFh | 25h: 12FFh | 3Ah: 1D7Fh |
|  |  |  |  | 10h: 087Fh | 26h: 137Fh | 3Bh: 1DFFh |
|  |  |  |  | 11h: 08FFh | 27h: 13FFh | 3Ch: 1E7Fh |
|  |  |  |  | 12h: 097Fh | 28h: 147Fh | 3Dh: 1EFFh |
|  |  |  |  | 13h: 09FFh | 29h: 14FFh | 3Eh: 1F7Fh |
|  |  |  |  | 14h: 0A7Fh | 2Ah: 157Fh | 3Fh: 1FFFh |
|  |  |  |  | 15h: OAFFh |  |  |

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### 7.5.1.16 DAC_CODE_LIMIT_3 Register (address = 4Fh) [reset = 3F3Fh]

Figure 7-63. DAC_CODE_LIMIT_3 Register

| 15 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  | DACB3_LIMITS[5:0] |  |  |  |
| R-Oh |  |  | R/W-3Fh |  |  |  |
| 7 | 5 | 4 | 3 | 2 | 1 | 0 |
| RESERVED |  |  | DACB2_LIMITS[5:0] |  |  |  |
|  |  |  |  |  |  |  |

Table 7-71. DAC_CODE_LIMIT_3 Register Field Descriptions

| Bit | Field | Type | Reset | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 13-8 | DACB3_LIMITS | R/W | 3Fh | DAC active register latch code limit; off by default. Program these bits with the following values to achieve the limit specified for the upper six MSBs of the DAC codes. |  |  |
| 5-0 | DACB2_LIMITS | R/W | 3 Fh |  |  |  |
|  |  |  |  | 00h: 007Fh | 16h: 0B7Fh | 2Bh: 15FFh |
|  |  |  |  | 01h: 00FFh | 17h: OBFFh | 2Ch: 167Fh |
|  |  |  |  | 02h: 017Fh | 18h: 0C7Fh | 2Dh: 16FFh |
|  |  |  |  | 03h: 01FFh | 19h: OCFFh | 2Eh: 177Fh |
|  |  |  |  | 04h: 027Fh | 1Ah: 0D7Fh | 2Fh: 17FFh |
|  |  |  |  | 05h: 02FFh | 1Bh: ODFFh | 30h: 187Fh |
|  |  |  |  | 06h: 037Fh | 1Ch: 0E7Fh | 31h: 18FFh |
|  |  |  |  | 07h: 03FFh | 1Dh: 0EFFh | 32h: 197Fh |
|  |  |  |  | 08h: 047Fh | 1Eh: 0F7Fh | 33h: 19FFh |
|  |  |  |  | 09h: 04FFh | 1Fh: 0FFFh | 34h: 1A7Fh |
|  |  |  |  | 0Ah: 057Fh | 20h: 107Fh | 35h: 1AFFh |
|  |  |  |  | 0Bh: 05FFh | 21h: 10FFh | 36h: 1B7Fh |
|  |  |  |  | 0Ch: 067Fh | 22h: 117Fh | 37h: 1BFFh |
|  |  |  |  | 0Dh: 06FFh | 23h: 11FFh | 38h: 1C7Fh |
|  |  |  |  | 0Eh: 077Fh | 24h: 127Fh | 39h: 1CFFh |
|  |  |  |  | 0Fh: 07FFh | 25h: 12FFh | 3Ah: 1D7Fh |
|  |  |  |  | 10h: 087Fh | 26h: 137Fh | 3Bh: 1DFFh |
|  |  |  |  | 11h: 08FFh | 27h: 13FFh | 3Ch: 1E7Fh |
|  |  |  |  | 12h: 097Fh | 28h: 147Fh | 3Dh: 1EFFh |
|  |  |  |  | 13h: 09FFh | 29h: 14FFh | 3Eh: 1F7Fh |
|  |  |  |  | 14h: 0A7Fh | 2Ah: 157Fh | 3Fh: 1FFFh |
|  |  |  |  | 15h: OAFFh |  |  |

7.5.1.17 DRVENO_EN Register (address = 50h) [reset = 0000h]

Figure 7-64. DRVENO_EN Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| DRVEN0 EN DACB3 | DRVEN0 EN_DACB2 | DRVEN0 <br> EN DACB1 | DRVENO EN DACBO | DRVEN0 <br> EN DACĀ3 | DRVEN0 EN DACĀ2 | DRVEN0 <br> EN_DACĀ1 | DRVENO <br> EN DACĀO |
| R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh |

Table 7-72. DRVENO_EN Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 7 | DRVEN0_EN_DACB3 | R/W | Oh | 0: Ignore DRVEN0 on DACB3 <br> 1: DRVEN0 enabled for DACB3 |
| 6 | DRVEN0_EN_DACB2 | R/W | Oh | 0: Ignore DRVEN0 on DACB2 <br> 1: DRVEN0 enabled for DACB2 |
| 5 | DRVEN0_EN_DACB1 | R/W | Oh | 0: Ignore DRVEN0 on DACB1 <br> 1: DRVEN0 enabled for DACB1 |
| 4 | DRVENO_EN_DACBO | R/W | Oh | 0: Ignore DRVEN0 on DACB0 <br> 1: DRVEN0 enabled for DACB0 |
| 3 | DRVEN0_EN_DACA3 | R/W | Oh | 0: Ignore DRVEN0 on DACA3 <br> 1: DRVEN0 enabled for DACA3 |
| 2 | DRVEN0_EN_DACA2 | R/W | Oh | 0: Ignore DRVEN0 on DACA2 <br> 1: DRVEN0 enabled for DACA2 |
| 1 | DRVEN0_EN_DACA1 | R/W | Oh | 0: Ignore DRVEN0 on DACA1 <br> 1: DRVEN0 enabled for DACA1 |
| 0 | DRVEN0_EN_DACAO | R/W | Oh | 0: Ignore DRVENO on DACAO <br> 1: DRVENO enabled for DACAO |

### 7.5.1.18 DRVEN1_EN Register (address = 51h) [reset = 0000h]

Figure 7-65. DRVEN1_EN Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  |  |  |
| R -Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| DRVEN1 <br> EN_DACB3 | DRVEN1 EN_DACB2 | DRVEN1 EN_DACB1 | DRVEN1 EN_DACBO | DRVEN1 <br> EN_DACĀ3 | DRVEN1 EN_DACĀ2 | DRVEN1 <br> EN_DACĀ1 | DRVEN1 <br> EN_DACĀO |
| R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh |

Table 7-73. DRVEN1_EN Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 7 | DRVEN1_EN_DACB3 | R/W | Oh | 0: Ignore DRVEN1 on DACB3 <br> 1: DRVEN1 enabled for DACB3 |
| 6 | DRVEN1_EN_DACB2 | R/W | Oh | 0: Ignore DRVEN1 on DACB2 <br> 1: DRVEN1 enabled for DACB2 |
| 5 | DRVEN1_EN_DACB1 | R/W | Oh | 0: Ignore DRVEN1 on DACB1 <br> 1: DRVEN1 enabled for DACB1 |
| 4 | DRVEN1_EN_DACB0 | R/W | Oh | 0 : Ignore DRVEN1 on DACB0 <br> 1: DRVEN1 enabled for DACB0 |
| 3 | DRVEN1_EN_DACA3 | R/W | Oh | 0: Ignore DRVEN1 on DACA3 <br> 1: DRVEN1 enabled for DACA3 |
| 2 | DRVEN1_EN_DACA2 | R/W | Oh | 0: Ignore DRVEN1 on DACA2 <br> 1: DRVEN1 enabled for DACA2 |
| 1 | DRVEN1_EN_DACA1 | R/W | Oh | 0: Ignore DRVEN1 on DACA1 <br> 1: DRVEN1 enabled for DACA1 |
| 0 | DRVEN1_EN_DACA0 | R/W | Oh | 0: Ignore DRVEN1 on DACA0 <br> 1: DRVEN1 enabled for DACA0 |

### 7.5.1.19 FLEXIO_EN Register (address = 52h) [reset = 0000h]

Figure 7-66. FLEXIO_EN Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| $\begin{aligned} & \text { FLEXIO } \\ & \text { EN_DACB3 } \end{aligned}$ | $\begin{aligned} & \text { FLEXIO } \\ & \text { EN_DACB2 } \end{aligned}$ | $\begin{aligned} & \text { FLEXIO } \\ & \text { EN_DACB1 } \end{aligned}$ | $\begin{aligned} & \text { FLEXIO-} \\ & \text { EN_DACB̄O } \end{aligned}$ | $\begin{aligned} & \text { FLEXIO- } \\ & \text { EN_DACA } 3 \end{aligned}$ | $\begin{aligned} & \text { FLEXIO_ } \\ & \text { EN_DACA } 2 \end{aligned}$ | $\begin{aligned} & \text { FLEXIO_ } \\ & \text { EN_DACĀ1 } \end{aligned}$ | $\begin{aligned} & \text { FLEXIO_ } \\ & \text { EN_DACA } \end{aligned}$ |
| R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh | R/W-Oh |

Table 7-74. FLEXIO_EN Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :---: | :---: | :---: | :---: |
| 7 | FLEXIO_EN_DACB3 | R/W | Oh | 0: Ignore FLEXIO on DACB3 <br> 1: FLEXIO enabled for DACB3 |
| 6 | FLEXIO_EN_DACB2 | R/W | Oh | 0: Ignore FLEXIO on DACB2 <br> 1: FLEXIO enabled for DACB2 |
| 5 | FLEXIO_EN_DACB1 | R/W | Oh | 0: Ignore FLEXIO on DACB1 <br> 1: FLEXIO enabled for DACB1 |
| 4 | FLEXIO_EN_DACB0 | R/W | Oh | 0: Ignore FLEXIO on DACB0 <br> 1: FLEXIO enabled for DACBO |
| 3 | FLEXIO_EN_DACA3 | R/W | Oh | 0: Ignore FLEXIO on DACA3 <br> 1: FLEXIO enabled for DACA3 |
| 2 | FLEXIO_EN_DACA2 | R/W | Oh | 0: Ignore FLEXIO on DACA2 <br> 1: FLEXIO enabled for DACA2 |
| 1 | FLEXIO_EN_DACA1 | R/W | Oh | 0: Ignore FLEXIO on DACA1 <br> 1: FLEXIO enabled for DACA1 |
| 0 | FLEXIO_EN_DACAO | R/W | Oh | 0: Ignore FLEXIO on DACA0 <br> 1: FLEXIO enabled for DACAO |

### 7.6 DAC Buffer Register Map

Table 7-75. Page 4: DAC Buffer Register Map


### 7.6.1 DAC Buffer Data Registers: Page 4

### 7.6.1.1 DACA/Bn Buffer Registers (address $=40 \mathrm{~h}$ to 47h) [reset $=0000 \mathrm{~h}]$

Figure 7-67. DACA/Bn Buffer Register

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  | DAC[12:8] |  |  |  |  |
| R-Oh |  |  | R/W-Oh |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| DAC[7:0] |  |  |  |  |  |  |  |
| R/W-Oh |  |  |  |  |  |  |  |

Table 7-76. DACA/Bn Buffer Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $12-0$ | DAC | R/W | Oh | Stores 13-bit data to be loaded to DACn active register, in MSB- <br> aligned, unipolar binary format. |

### 7.7 DAC Active Register Map

Table 7-77. Page 6: DAC Active Register Map

| ADDR <br> (HEX) | REGISTER | TYPE | RESET (HEX) <br> (HEX) | BIT DESCRIPTION |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 40 | DACAO | R | 0000 | RESERVED |  |  | DAC[12:0] |  |  |  |  |  |  |  |  |  |  |  |  |
| 41 | DACA1 | R | 0000 | RESERVED |  |  | DAC[12:0] |  |  |  |  |  |  |  |  |  |  |  |  |
| 42 | DACA2 | R | 0000 | RESERVED |  |  | DAC[12:0] |  |  |  |  |  |  |  |  |  |  |  |  |
| 43 | DACA3 | R | 0000 | RESERVED |  |  | DAC[12:0] |  |  |  |  |  |  |  |  |  |  |  |  |
| 44 | DACB0 | R | 0000 | RESERVED |  |  | DAC[12:0] |  |  |  |  |  |  |  |  |  |  |  |  |
| 45 | DACB1 | R | 0000 | RESERVED |  |  | DAC[12:0] |  |  |  |  |  |  |  |  |  |  |  |  |
| 46 | DACB2 | R | 0000 | RESERVED |  |  | DAC[12:0] |  |  |  |  |  |  |  |  |  |  |  |  |
| 47 | DACB3 | R | 0000 | RESERVED |  |  | DAC[12:0] |  |  |  |  |  |  |  |  |  |  |  |  |

### 7.7.1 DAC Active Data Registers: Page 4

### 7.7.1.1 DACA/Bn Active Register (address $=40 \mathrm{~h}$ to 47h) [reset $=0000 \mathrm{~h}]$

## Figure 7-68. DACA/Bn Active Registers

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RESERVED |  |  | DAC[12:8] |  |  |  |  |
| R-Oh |  |  | R-Oh |  |  |  |  |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| DAC[7:0] |  |  |  |  |  |  |  |
| R-Oh |  |  |  |  |  |  |  |

Table 7-78. DACA/Bn Active Register Field Descriptions

| Bit | Field | Type | Reset | Description |
| :---: | :--- | :--- | :--- | :--- |
| $12-0$ | DAC | R | Oh | Stores 13-bit data to be loaded to DACn channel in MSB- <br> aligned, unipolar binary format. |

## 8 Application and Implementation

## Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. Tl's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

### 8.1 Application Information

The primary application of the AFE20408 device is to provide power amplifier (PA) gate-bias control. The integrated switches allow the gate bias to be switched between a temperature-adjusted on voltage and a static, lower-potential off voltage.

In addition, the AFE20408 has features to detect alarm conditions, and in response, lower the gate voltages and turn off the PA during these events.

### 8.1.1 Output Switching Timing

The externally applied output capacitors allow for noise filtering, and enable fast switching on the output channels of the device. Large capacitors can be connected to the output of the static channels: DACAO, DACA1, DACA2, DACA3 on group A, and DACB0, DACB1, DACB2, DACB3 on group B. Capacitors of lower values can be connected to the dynamic channels, OUTA0, OUTA2, OUTB0, and OUTB2. This capacitor arrangement means that the larger capacitors can quickly charge the smaller capacitors instead of relying on the DAC output buffers.

Figure 8-1 shows a simplified model of switch arrangement for the OUTAO channel. The on-resistance of the switches are represented by $\mathrm{R}_{\mathrm{SW} 1}$ and $\mathrm{R}_{\mathrm{SW} 2}$. These resistors primarily serve to limit the settling time of $\mathrm{V}_{\text {OUTA1 }}$ after a switching event, as the settling time is essentially an RC function.


Figure 8-1. Switching Transients

For example, consider the case where DRVENO changes from a low-state to a high-state. The steady-state of $V_{\text {DACAO }}$ is equal to $V_{\text {DACA1 }}$ before the switch event. After the DRVEN pin goes high, SW2 closes, connecting $C_{\text {OUTA1 }}$ and $C_{\text {DACAO }}$ to each other. As these capacitors are now in parallel, the voltages across each equalize to a new voltage. This voltage, described as $\mathrm{V}_{\text {CDAC||COUT }}$ in the following equation, can be calculated by finding the charge stored in each capacitor. The total charge on the two capacitors in parallel is equal to the sum of the charge of each capacitor.

$$
\begin{align*}
& \mathrm{Q}_{\text {CDAC } \mid \text { cout }}=\mathrm{Q}_{\text {CDAC }}+\mathrm{Q}_{\text {cout }}  \tag{2}\\
& \mathrm{V}_{\text {CDAC } \| \text { COUT }}\left(\mathrm{C}_{\text {DACA1 }}+\mathrm{C}_{\text {OUTA } 0}\right)=\mathrm{V}_{\text {DACA1 }} \times \mathrm{C}_{\text {DACA1 }}+\mathrm{V}_{\text {OUTA } 0} \times \mathrm{C}_{\text {OUTA }}  \tag{3}\\
& \mathrm{V}_{\text {CDAC||COUT }}=\frac{\mathrm{V}_{\text {DACA1 }} \times \mathrm{C}_{\text {DACA1 }}+\mathrm{V}_{\text {OUTA }} \times \mathrm{C}_{\text {OUTA }}}{\left(\mathrm{C}_{\text {DACA1 }}+\mathrm{C}_{\text {OUTA }}\right)} \tag{4}
\end{align*}
$$

The time required for the two output to equalize, described as the Capacitive Settling Period, is calculated using the equation below. As DACAO is lower potential than DACA1, $\mathrm{V}_{\text {OUTA0 }}$ can be expressed as a charging function.

$$
\begin{equation*}
\mathrm{V}_{\text {OUTA0 }}(\mathrm{t})=\left(\mathrm{V}_{\text {CDAC } \| \text { COUT }}-\mathrm{V}_{\text {OUTA0 }}\left(\mathrm{t}_{0}\right)\right)\left(1-\mathrm{e}^{\frac{-\mathrm{t}}{\mathrm{R}_{\text {SW }} \times \mathrm{C}_{\text {OUTA }}}}\right)+\mathrm{V}_{\text {OUTA } 0}\left(\mathrm{t}_{0}\right) \tag{5}
\end{equation*}
$$

During the capacitive settling period, $\mathrm{V}_{\mathrm{DACA} 1}$ is expressed as a discharging RC function.

$$
\begin{equation*}
\mathrm{V}_{\mathrm{DACA} 1}(\mathrm{t})=\mathrm{V}_{\mathrm{DACA} 1}\left(\mathrm{t}_{0}\right)-\left(\mathrm{V}_{\text {DACA1 }}\left(\mathrm{t}_{0}\right)-\mathrm{V}_{\mathrm{CDAC}| | \operatorname{COUT}}\right)\left(1-\mathrm{e}^{\frac{-\mathrm{t}}{\mathrm{R}_{\mathrm{SW} 1} \times \mathrm{C}_{\text {OUTA }}}}\right) \tag{6}
\end{equation*}
$$

Connecting the capacitors together allows the output to change to $\mathrm{V}_{\text {CDAC||COUT }}$ quickly, but after that period, the DAC output buffer continues to charge $\mathrm{C}_{\text {OUTA1 }}$ to the $\mathrm{V}_{\text {DACAO }}$ value. The settling time for that final transition depends on the RC function formed by the series resistance on the DAC output, the switch resistance, and the capacitive load on the DAC. In addition, the output current of the DAC is limited.

Figure $8-2$ shows the switch response for the OUTAO pin when switching from a static DAC channel to VSS, while Figure $8-3$ shows the switch response of the OUTAO signal when switching between static DAC outputs.


Figure 8-2. DAC-to-VSS Switch Response


Figure 8-3. DAC-to-DAC Switch Response

### 8.2 Typical Application

Figure 8-4 shows an example schematic for PA biasing applications, using a single AFE20408 device to bias GaN and LDMOS PAs simultaneously. In this application, DAC group A is configured in a negative output range, while DAC group B is configured in positive output range.


Figure 8-4. Power Amplifier Biasing Application

### 8.2.1 Design Requirements

The example schematic uses the majority of the design parameters listed in Table 8-1. The power supplies and DAC outputs are configured for the mixed output range.

Table 8-1. Design Parameters

| DESIGN PARAMETER | EXAMPLE VALUE |
| :--- | :--- |
| $\mathrm{V}_{\mathrm{CCA}}$ | Grounded |
| $\mathrm{V}_{\mathrm{CCB}}$ | 11 V |
| $\mathrm{~V}_{\mathrm{SSA}}$ | -11 V |
| $\mathrm{~V}_{\text {SSB }}$ | Grounded |
| $\mathrm{V}_{\mathrm{DD}}$ | 5 V |
| $\mathrm{~V}_{\text {IO }}$ | 1.8 V |
| DAC outputs | Group A selectable ranges: -10 V to 0 V |
|  | Group B selectable ranges: <br> OV to 10 V |

### 8.2.2 Detailed Design Procedure

### 8.2.2.1 ADC Input Conditioning

The ADC inputs feature an input range that can be configured as either 0 V to 2.5 V or 0 V to 5 V .
To reduce ADC sample glitch, place a 470 pF capacitor on the ADC input. By adding small series resistors (in series with the ADC inputs) a low-pass noise filter can be implemented, as shown in Figure 8-5.


Figure 8-5. ADC Input Conditioning

### 8.2.2.2 Quiescent Current and Total Power Consumption

Calculating the total power consumption of the device requires all of the supply inputs and DAC loads to be known. Equation 7 calculates the total power, Each component is the power contributed by a supply or DAC loads.

$$
\begin{equation*}
\mathrm{P}_{\mathrm{TOTAL}}=\mathrm{P}_{\mathrm{IO}}+\mathrm{P}_{\mathrm{DD}}+\mathrm{P}_{\mathrm{CC}}+\mathrm{P}_{\mathrm{SS}}+\mathrm{P}_{\mathrm{DAC}}-\text { LOAD } \tag{7}
\end{equation*}
$$

where

- $P_{1 O}$ is the power consumed by the device from the $V_{1 O}$ supply:

$$
\begin{equation*}
\mathrm{P}_{\mathrm{IO}}=\mathrm{V}_{\mathrm{IO}} \times \mathrm{I}_{\mathrm{IO}}-\text { quiescent } \tag{8}
\end{equation*}
$$

- $P_{D D}$ is the power consumed by the device from the $V_{D D}$ supply:

$$
\begin{equation*}
P_{D D}=V_{D D} \times I_{D D}-\text { quiescent } \tag{9}
\end{equation*}
$$

- $P_{C C}$ is the power consumed by the device from the $V_{C C}$ supply:

$$
\begin{equation*}
\mathrm{P}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \times \mathrm{I}_{\mathrm{CC}}-\text { quiescent } \tag{10}
\end{equation*}
$$

- $P_{S S}$ is the power consumed by the device from the $V_{S S}$ supply:

$$
\begin{equation*}
\mathrm{P}_{\mathrm{SS}}=\mathrm{V}_{\mathrm{SS}} \times \mathrm{I}_{\mathrm{SS}}-\text { quiescent } \tag{11}
\end{equation*}
$$

- $P_{\text {DAC-LOAD }}$ is the power consumed by the device as a result of the DAC loads from the sourcing or sinking supply. The power of each DAC channel can be calculated separately, then summed to find the total power of the DAC loads. The power depends not only on the voltage of the DAC output, but also on the difference between the current sourcing or sinking supply and the DAC output voltage. The following equation shows how to calculate $\mathrm{P}_{\mathrm{DAC}}$ LOAD:

$$
\begin{equation*}
\mathrm{P}_{\mathrm{DAC}}-\mathrm{LOAD}=\sum_{\text {channel } n}^{n}=0 V_{S U P P L Y-L O A D} \times I_{L O A D} \tag{12}
\end{equation*}
$$

Figure 8-6 shows the load configuration in both the positive output range and negative output range.


Figure 8-6. DAC Output Load

When the device is in the positive output range, the device is likely sourcing current. While in the negative range, the device is likely sinking current. The difference between the supply voltage and the DAC output voltage is $\mathrm{V}_{\text {SUPPLY-LOAD }}$, as shown in the following equations.
When the device is in the positive output range, $\mathrm{V}_{\text {SUPPLY-LOAD }}$ can be calculated as:

$$
\begin{equation*}
\mathrm{V}_{\text {SUPPLY }}-\mathrm{LOAD}=\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{DAC}} \tag{13}
\end{equation*}
$$

When the device is in the negative output range, $\mathrm{V}_{\text {SUPPLY-LOAD }}$ can be calculated as:

$$
\begin{equation*}
\mathrm{V}_{S U P P L Y}-\mathrm{LOAD}=\mathrm{V}_{\mathrm{SS}}-\mathrm{V}_{\mathrm{DAC}} \tag{14}
\end{equation*}
$$

### 8.2.2.2.1 Maximum VCC/VSS Supply Current Transients

In many applications, the DAC outputs of the device have a capacitive load. When the DAC outputs transition from one output voltage to another, the short-circuit limit protection can be triggered. If the DAC output buffer reaches the short-circuit current limit of the amplifier, significant current is drawn from the output amplifier supply. Equation 15 shows how to calculate the estimated maximum current that is demanded of the supply during the transition.

$$
\begin{equation*}
\mathrm{I}_{\mathrm{VCC}}-\mathrm{MAX}=I_{V C C}-\text { quiescent }+\sum_{\text {channel } n=0}^{n} I_{S H O R T}-\text { CIRCUIT }- \text { LIMIT } \tag{15}
\end{equation*}
$$

### 8.2.2.2.2 DAC Load Stability

Figure 8-7 shows the required configuration when capacitive loads are present on the DAC output. No series resistor is required on the DAC output, as the DAC is able to prevent oscillation issues on the output amplifier.


Figure 8-7. DAC Output Load

### 8.2.2.3 Disabling PA Drain Voltage

The PAVDD voltage is separated from the drain voltage of the power amplifier with a series PMOS transistor. The activation of the PMOS transistor connects the PAVDD voltage supply to the drain pin of the power amplifier. The PMOS transistor is driven with a voltage divider that swings from PAVDD to PAVDD(R2 / (R1 + R2)). The NMOS transistor shown in Figure 8-4 is connected to the PAON output of the AFE20408. When PAON is low, the PMOS gate is equal to the PAVDD voltage, disconnecting the PAVDD voltage from the PA. When PAON is high, the voltage divider turns on and enables the PMOS, connecting the PAVDD voltage to the PA.

### 8.2.2.4 PAON External Circuit

During start-up, the AFE20408 PAON is Hi-Z until all power supplies are established. When operating in pushpull mode, a pull-down resistor to ground is recommended to keep the PAON output from floating. When operating in open-drain mode, the PAON output is to be isolated from the pullup resistor to $\mathrm{V}_{\mathrm{DD}}$, until all power has been applied. Figure $8-8$ shows a dual NMOS circuit that grounds the PAON output until both $V_{D D}$ and $V_{10}$ are powered to an operating voltage, after which the PAON output operates as a pullup to $V_{D D}$.


Figure 8-8. PAON Open-Drain Circuit

### 8.2.3 Application Curves

### 8.2.3.1 DAC Load Stability

Figure $8-9$ shows the DAC output response when the DAC is unloaded, and when the DAC is loaded with a capacitor, respectively.


Figure 8-9. DAC Settling Time vs Load Capacitance

### 8.2.3.2 Start-Up Behavior

The AFE20408 is designed to minimize DAC output glitch during power supply transients at power-on and power-down. Figure 8-10 to Figure 8-13 detail this behavior.


Figure 8-10. DAC Output During $\mathbf{V}_{\mathrm{CC}}$ Power-On Transient


Figure 8-12. DAC Output During $\mathbf{V}_{\mathrm{cc}}$ Power-Down Transient


Figure 8-11. DAC Output During $\mathrm{V}_{\text {ss }}$ Power-On Transient


Figure 8-13. DAC Output During Vss Power-Down Transient

### 8.3 Initialization Setup

After power-up, the device can be configured over the serial interface. The following steps can be used in a typical configuration.

1. After the supplies have ramped to the final output voltage, issue a hardware or software reset to make sure the device is in a known state. Allow approximately 5 ms for the device registers to initialize after the reset event
2. Write to the ADC_CONV_CFG_0 and ADC_GEN_CFG registers (in the ADC configuration register page) to set the ADC conversion rate, conversion mode, and shunt range.
3. Configure the ADC inputs and custom channel sequencer (CCS), by writing to the ADC_CCS_IDS registers and the ADC_CCS_CFG_0 register (located in the ADC CCS Configuration register page).
4. Set the DAC current limits by writing to the DAC_CURRENT register in the DAC Configuration register page.
5. Initialize the DACs by configuring the DAC_APD_SRC registers, the OUT_APD_SRC registers, the ALARMOUT_SRC registers, the DAC DRVEN_EN registers, and the DAC_CODE_LIMIT registers (all located in the DAC Configuration register page).
6. Set the ADC and temperature sensor alarm limits by writing to the ADCn_UP_THRESH, ADCn_LOW_THRESH and TMP_UP_THRESH registers in the ADC Configuration register page (where $n=0,1$ ).
7. Write the initial DAC output values by writing to the DACAn or DACBn data registers (where $n=0,1,2,3$ ).
8. Enable the DACs by writing to the power enable (PWR_EN) register in the global register page.
9. Initiate a single (or multiple) ADC conversion by writing to the ADC_TRIG bit in the TRIGGER register (located in the global register page).
10. Update the DAC output values by writing to the DAC_TRIG bit in the TRIGGER register, if using synchronous mode on the respective DAC.

### 8.4 Power Supply Recommendations

There is no required supply sequence, but be aware that the device stays in the reset state until all supplies reach the power-good threshold. Also, a hardware or software reset to the device is recommended after the supplies reach the power-good threshold, so that the device can initialize in a known state. Following this reset (or any reset event) wait at least 5 ms so that the device registers can properly initialize.

In applications where a negative voltage is applied to $\mathrm{V}_{S S}$ first, some small negative voltages can be present at other supply pins, such as the $\mathrm{V}_{10}$ and $\mathrm{V}_{\mathrm{DD}}$. The negative voltages at the supply pins can exceed the values listed in Section 5.1, but because these voltages are created from intrinsic circuitry, the voltage levels are safe for operation.

### 8.5 Layout

### 8.5.1 Layout Guidelines

- Bypass all power supply pins to ground with a low-ESR ceramic bypass capacitor. Bypass capacitors on the $\mathrm{V}_{\mathrm{CCx}}$ and $\mathrm{V}_{\mathrm{SSx}}$ inputs are recommended to be three to four times the total capacitance on the respective group DAC outputs to make sure the inrush current does not cause localized supply collapse when the outputs transition to different voltage output. The typical recommended bypass capacitor has a value of $1 \mu \mathrm{~F}$ and is ceramic with X7R or NPO dielectric.
- Place capacitors on the DAC[0:3], OUT0, and OUT2 pins as close to the device as possible. This placement reduces the impact of parasitic inductance and resistance from the switching path. Parasitic inductance and resistance delays the output settling time.
- Connect the thermal pad on the device to a large copper area, preferably a ground plane.
- When using the local temperature sensor for the output bias voltage temperature compensations, place the device geographically close to the PA, preferably sharing a solid ground plane for thermal conduction.


### 8.5.2 Layout Diagram



Figure 8-14. AMC20408 Layout

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## 9 Device and Documentation Support

### 9.1 Documentation Support

### 9.1.1 Related Documentation

For related documentation see the following:

- Texas Instruments, AFE20408EVM user's guide


### 9.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on Notifications to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

### 9.3 Support Resources

TI E2E ${ }^{\text {TM }}$ support forums are an engineer's go-to source for fast, verified answers and design help - straight from the experts. Search existing answers or ask your own question to get the quick design help you need.
Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect Tl's views; see TI's Terms of Use.

### 9.4 Trademarks

TI E2E ${ }^{\text {TM }}$ is a trademark of Texas Instruments.
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### 9.5 Electrostatic Discharge Caution

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.
ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 9.6 Glossary

TI Glossary This glossary lists and explains terms, acronyms, and definitions.

## 10 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

| DATE | REVISION | NOTES |
| :---: | :---: | :---: |
| April 2024 | ${ }^{*}$ | Initial release. |

## 11 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated device(s). These data are subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

InSTRUMENTS

## PACKAGING INFORMATION

| Orderable Device | Status <br> (1) | Package Type | Package Drawing | Pins | Package Qty | Eco Plan <br> (2) | Lead finish/ Ball material <br> (6) | MSL Peak Temp <br> (3) | Op Temp ( ${ }^{\circ} \mathrm{C}$ ) | Device Marking <br> (4/5) | Samples |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| AFE20408RHBT | ACTIVE | VQFN | RHB | 32 | 250 | RoHS \& Green | NIPDAUAG | Level-3-260C-168 HR | -40 to 125 | $\begin{aligned} & \hline \text { AFE } \\ & 20408 \end{aligned}$ | Samples |

${ }^{(1)}$ The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.
${ }^{(2)}$ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed $0.1 \%$ by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. Tl may reference these types of products as "Pb-Free".
RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption
Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the $<=1000 \mathrm{ppm}$ threshold requirement.
${ }^{(3)}$ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
${ }^{(4)}$ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
${ }^{(5)}$ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
${ }^{(6)}$ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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TAPE AND REEL INFORMATION


TAPE DIMENSIONS


| A0 | Dimension designed to accommodate the component width |
| :--- | :--- |
| B0 | Dimension designed to accommodate the component length |
| K0 | Dimension designed to accommodate the component thickness |
| W | Overall width of the carrier tape |
| P1 | Pitch between successive cavity centers |

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

*All dimensions are nominal

| Device | Package <br> Type | Package <br> Drawing | Pins | SPQ | Reel <br> Diameter <br> $(\mathbf{m m})$ | Reel <br> Width <br> W1 $(\mathbf{m m})$ | A0 <br> $(\mathbf{m m})$ | B0 <br> $(\mathbf{m m})$ | K0 <br> $(\mathbf{m m})$ | P1 <br> $(\mathbf{m m})$ | W <br> $(\mathbf{m m})$ | Pin1 <br> Quadrant |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| AFE20408RHBT | VQFN | RHB | 32 | 250 | 180.0 | 12.4 | 5.25 | 5.25 | 1.1 | 8.0 | 12.0 | Q2 |


*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Length (mm) | Width (mm) | Height (mm) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| AFE20408RHBT | VQFN | RHB | 32 | 250 | 213.0 | 191.0 | 35.0 |



Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.


NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.


NOTES: (continued)
4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.


SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
EXPOSED PAD 33:
75\% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE
SCALE:20X

NOTES: (continued)
6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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