1 Features

- Ground Referenced Tachometer Input Interfaces Directly With Variable Reluctance Magnetic Pickups
- Op Amp Has Floating Transistor Output
- 50-mA Sink or Source to Operate Relays, Solenoids, Meters, or LEDs
- Frequency Doubling For Low Ripple
- Tachometer Has Built-In Hysteresis With Either Differential Input or Ground Referenced Input
- ±0.3% Linearity (Typical)
- Ground-Referenced Tachometer is Fully Protected From Damage Due to Swings Above $V_{CC}$ and Below Ground
- Output Swings to Ground For Zero Frequency Input
- Easy to Use; $V_{OUT} = f_{IN} \times V_{CC} \times R_1 \times C_1$
- Zener Regulator on Chip allows Accurate and Stable Frequency to Voltage or Current Conversion (LM2917)

2 Applications

- Over- and Under-Speed Sensing
- Frequency-to-Voltage Conversion (Tachometer)
- Speedometers
- Breaker Point Dwell Meters
- Hand-Held Tachometers
- Speed Governors
- Cruise Control
- Automotive Door Lock Control
- Clutch Control
- Horn Control
- Touch or Sound Switches

3 Description

The LM2907 and LM2917 devices are monolithic frequency-to-voltage converters with a high gain op amp designed to operate a relay, lamp, or other load when the input frequency reaches or exceeds a selected rate. The tachometer uses a charge pump technique and offers frequency doubling for low-ripple, full-input protection in two versions (8-pin LM2907 and LM2917), and its output swings to ground for a zero frequency input.

The op amp is fully compatible with the tachometer and has a floating transistor as its output. This feature allows either a ground or supply referred load of up to 50 mA. The collector may be taken above $V_{CC}$ up to a maximum $V_{CE}$ of 28 V.

The two basic configurations offered include an 8-pin device with a ground-referenced tachometer input and an internal connection between the tachometer output and the op amp noninverting input. This version is well suited for single speed or frequency switching or fully buffered frequency-to-voltage conversion applications.

Device Information

<table>
<thead>
<tr>
<th>PART NUMBER</th>
<th>PACKAGE</th>
<th>BODY SIZE (NOM)</th>
</tr>
</thead>
<tbody>
<tr>
<td>LM2907-N, LM2917-N</td>
<td>PDIP (8)</td>
<td>6.35 mm × 9.81 mm</td>
</tr>
<tr>
<td></td>
<td>PDIP (14)</td>
<td>6.35 mm × 19.177 mm</td>
</tr>
<tr>
<td></td>
<td>SOIC (8)</td>
<td>3.91 mm × 4.90 mm</td>
</tr>
<tr>
<td></td>
<td>SOIC (14)</td>
<td>3.91 mm × 8.65 mm</td>
</tr>
</tbody>
</table>

(1) For all available packages, see the orderable addendum at the end of the data sheet.
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2 Applications ............................................................ 1

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### Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

#### Changes from Revision C (March 2013) to Revision D

<table>
<thead>
<tr>
<th>Changes</th>
<th>Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>Added Device Information table, Pin Configuration and Functions section, ESD Ratings table, Recommended Operating Conditions table, Thermal Information table, Parameter Measurement Information section, Detailed Description section, Application and Implementation section, Power Supply Recommendations section, Layout section, Device and Documentation Support section, and Mechanical, Packaging, and Orderable Information section</td>
<td>1</td>
</tr>
<tr>
<td>Deleted Built-In Zener on LM2917 from Features</td>
<td>1</td>
</tr>
<tr>
<td>Deleted Only One RC Network Provides Frequency Doubling from Features</td>
<td>1</td>
</tr>
<tr>
<td>Added Thermal Information table</td>
<td>6</td>
</tr>
<tr>
<td>Changed tablenote text From: C2 = 0.22 mFd To: C2 = 0.22 µF</td>
<td>6</td>
</tr>
</tbody>
</table>

#### Changes from Revision B (March 2013) to Revision C

<table>
<thead>
<tr>
<th>Changes</th>
<th>Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>Changed layout of National Semiconductor Data Sheet to TI format</td>
<td>1</td>
</tr>
</tbody>
</table>
5 Description (continued)
The more versatile configurations provide differential tachometer input and uncommitted op amp inputs. With this version the tachometer input may be floated and the op amp becomes suitable for active filter conditioning of the tachometer output.

Both of these configurations are available with an active shunt regulator connected across the power leads. The regulator clamps the supply such that stable frequency-to-voltage and frequency-to-current operations are possible with any supply voltage and a suitable resistor.

6 Pin Configuration and Functions

<table>
<thead>
<tr>
<th>PIN</th>
<th>I/O</th>
<th>DESCRIPTION</th>
</tr>
</thead>
<tbody>
<tr>
<td>COL</td>
<td>5</td>
<td>The collector of the bipolar junction transistor</td>
</tr>
<tr>
<td>CP1</td>
<td>2</td>
<td>A capacitor placed on this pin will be charged up to ( V_{CC}/2 ) by a constant current source of 180 ( \mu )A typical at the start of every positive half cycle. At the beginning of negative half cycles this capacitor is discharged the same amount at the same rate.</td>
</tr>
<tr>
<td>CP2/IN+</td>
<td>3</td>
<td>I/O See pins CP1 and IN+. On 8-pin devices (8-pin LM2907 and LM2917) these two nodes share a pin and are internally connected.</td>
</tr>
<tr>
<td>EMIT</td>
<td>4</td>
<td>O The emitter of the bipolar junction transistor</td>
</tr>
<tr>
<td>GND</td>
<td>—</td>
<td>— No connect</td>
</tr>
<tr>
<td>IN+</td>
<td>—</td>
<td>I The noninverting input to the high gain op amp</td>
</tr>
<tr>
<td>IN–</td>
<td>7</td>
<td>I The inverting input to the high gain op amp</td>
</tr>
<tr>
<td>NC</td>
<td>—</td>
<td>— No connect</td>
</tr>
<tr>
<td>TACH+</td>
<td>1</td>
<td>I Positive terminal for the input signal that leads to the noninverting terminal of the internal Schmitt-Trigger comparator.</td>
</tr>
<tr>
<td>TACH–/GND</td>
<td>8</td>
<td>I Negative terminal for the input signal that leads to the noninverting terminal of the internal Schmitt-Trigger comparator. (NOTE: On 8-pin devices, LM2907 and LM2917, this pin is internally connected to ground and must be tied to ground externally to provide the reference voltage of the device).</td>
</tr>
<tr>
<td>V+</td>
<td>6</td>
<td>I Supply voltage</td>
</tr>
</tbody>
</table>
### Pin Functions: 14 Pins

<table>
<thead>
<tr>
<th>PIN</th>
<th>I/O</th>
<th>DESCRIPTION</th>
</tr>
</thead>
<tbody>
<tr>
<td>COL</td>
<td>8</td>
<td>I</td>
</tr>
<tr>
<td>CP1</td>
<td>2</td>
<td>O</td>
</tr>
<tr>
<td>CP2</td>
<td>3</td>
<td>O</td>
</tr>
<tr>
<td>EMIT</td>
<td>5</td>
<td>O</td>
</tr>
<tr>
<td>GND</td>
<td>12</td>
<td>G</td>
</tr>
<tr>
<td>IN+</td>
<td>4</td>
<td>I</td>
</tr>
<tr>
<td>IN–</td>
<td>10</td>
<td>I</td>
</tr>
<tr>
<td>NC</td>
<td>6, 7, 13, 14</td>
<td>—</td>
</tr>
<tr>
<td>TACH+</td>
<td>1</td>
<td>I</td>
</tr>
<tr>
<td>TACH–</td>
<td>11</td>
<td>I</td>
</tr>
<tr>
<td>V+</td>
<td>9</td>
<td>I</td>
</tr>
</tbody>
</table>
# 7 Specifications

## 7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)

<table>
<thead>
<tr>
<th></th>
<th>MIN</th>
<th>MAX</th>
<th>UNIT</th>
</tr>
</thead>
<tbody>
<tr>
<td>Supply voltage</td>
<td>28</td>
<td>28</td>
<td>V</td>
</tr>
<tr>
<td>Supply current (Zener options)</td>
<td>25</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td>Collector voltage</td>
<td>28</td>
<td>28</td>
<td>V</td>
</tr>
<tr>
<td>Differential input voltage</td>
<td>28</td>
<td>28</td>
<td>V</td>
</tr>
<tr>
<td>Input voltage</td>
<td>28</td>
<td>28</td>
<td>V</td>
</tr>
<tr>
<td>Power dissipation</td>
<td>0</td>
<td>28</td>
<td>V</td>
</tr>
<tr>
<td>Soldering information</td>
<td>260</td>
<td>1580</td>
<td>mW</td>
</tr>
<tr>
<td>Operating temperature, $T_J$</td>
<td>–40</td>
<td>85</td>
<td>°C</td>
</tr>
<tr>
<td>Storage temperature, $T_{Stg}$</td>
<td>–65</td>
<td>150</td>
<td>°C</td>
</tr>
</tbody>
</table>

(1) Stresses beyond those listed under **Absolute Maximum Ratings** may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under **Recommended Operating Conditions**. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/Distributors for availability and specifications.

## 7.2 ESD Ratings

<table>
<thead>
<tr>
<th>Electrostatic discharge</th>
<th>VALUE</th>
<th>UNIT</th>
</tr>
</thead>
<tbody>
<tr>
<td>Human-body model (HBM), per ANSI/ESDA/JESD22-A114$^{(1)}$</td>
<td>±1000</td>
<td>V</td>
</tr>
<tr>
<td>Charged-device model (CDM), per JEDEC specification JESD22-C101$^{(2)}$</td>
<td>±250</td>
<td>V</td>
</tr>
</tbody>
</table>

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

## 7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

<table>
<thead>
<tr>
<th></th>
<th>MIN</th>
<th>NOM</th>
<th>MAX</th>
<th>UNIT</th>
</tr>
</thead>
<tbody>
<tr>
<td>Input voltage</td>
<td>–28</td>
<td>28</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td></td>
<td>0</td>
<td>28</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>Output sink current</td>
<td>50</td>
<td>mA</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.
7.4 Thermal Information

<table>
<thead>
<tr>
<th>THERMAL METRIC(1)</th>
<th>LM2907, LM2917</th>
<th></th>
<th></th>
<th></th>
<th>UNIT</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>P (PDIP)</td>
<td>D (SOIC)</td>
<td>NFF (PDIP)</td>
<td>D (SOIC)</td>
<td></td>
</tr>
<tr>
<td></td>
<td>8 PINS</td>
<td>8 PINS</td>
<td>14 PINS</td>
<td>14 PINS</td>
<td></td>
</tr>
<tr>
<td>Junction-to-ambient thermal resistance</td>
<td>77.6</td>
<td>110</td>
<td>69.1</td>
<td>83.7</td>
<td>°C/W</td>
</tr>
<tr>
<td>Junction-to-case (top) thermal resistance</td>
<td>80.5</td>
<td>53.9</td>
<td>64.8</td>
<td>42.1</td>
<td>°C/W</td>
</tr>
<tr>
<td>Junction-to-board thermal resistance</td>
<td>54.8</td>
<td>50.4</td>
<td>49.1</td>
<td>38</td>
<td>°C/W</td>
</tr>
<tr>
<td>Junction-to-top characterization parameter</td>
<td>37.6</td>
<td>9.1</td>
<td>35.1</td>
<td>7.7</td>
<td>°C/W</td>
</tr>
<tr>
<td>Junction-to-board characterization parameter</td>
<td>54.8</td>
<td>49.9</td>
<td>49</td>
<td>37.7</td>
<td>°C/W</td>
</tr>
<tr>
<td>Junction-to-case (bottom) thermal resistance</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>°C/W</td>
</tr>
</tbody>
</table>

(1) For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

7.5 Electrical Characteristics

\( V_{CC} = 12 \) \( V_{DC} \), \( T_A = 25^\circ C \), see test circuit

<table>
<thead>
<tr>
<th>PARAMETER</th>
<th>TEST CONDITIONS</th>
<th>MIN</th>
<th>TYP</th>
<th>MAX</th>
<th>UNIT</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>TACHOMETER</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Input thresholds</td>
<td>( V_N = 250 ) mVp-p at 1 kHz(1)</td>
<td>±10</td>
<td>±25</td>
<td>±40</td>
<td>mV</td>
</tr>
<tr>
<td>Hysteresis</td>
<td>( V_N = 250 ) mVp-p at 1 kHz(1)</td>
<td>30</td>
<td></td>
<td></td>
<td>mV</td>
</tr>
<tr>
<td>LM29x7 offset voltage</td>
<td>( V_N = 250 ) mVp-p at 1 kHz(1)</td>
<td>3.5</td>
<td>10</td>
<td></td>
<td>mV</td>
</tr>
<tr>
<td></td>
<td>( V_N = 250 ) mVp-p at 1 kHz (8-pin LM29x7)(1)</td>
<td>5</td>
<td>15</td>
<td></td>
<td>mV</td>
</tr>
<tr>
<td>Input bias current</td>
<td>( V_N = \pm 50 ) mVDC</td>
<td>0.1</td>
<td>1</td>
<td></td>
<td>µA</td>
</tr>
<tr>
<td>High level output voltage</td>
<td>For CP1, ( V_N = 125 ) mVDC(2)</td>
<td>8.3</td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>Low level output voltage</td>
<td>For CP1, ( V_N = 125 ) mVDC(2)</td>
<td>2.3</td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>( I_2, I_3 )</td>
<td>Output current</td>
<td>V2 = V3 = 6 V(3)</td>
<td>140</td>
<td>180</td>
<td>240</td>
</tr>
<tr>
<td>( I_3 )</td>
<td>Leakage current</td>
<td>( I_2 = 0, V_3 = 0 )</td>
<td>0.1</td>
<td></td>
<td></td>
</tr>
<tr>
<td>( K )</td>
<td>Gain constant</td>
<td>See(2)</td>
<td>0.9</td>
<td>1</td>
<td>1.1</td>
</tr>
<tr>
<td>Linearity</td>
<td>( I_{IN} = 1 ) kHz, 5 kHz, or 10 kHz(4)</td>
<td>–1%</td>
<td>0.3%</td>
<td>1%</td>
<td></td>
</tr>
</tbody>
</table>

**OP AMP AND COMPARATOR**

| \( V_{OS} \)                  | \( V_N = 6 \) V | 3   | 10  |     | mV   |
| **\( I_{BIAS} \) **            | \( V_N = 6 \) V | 50  | 500 |     | nA   |
| Input common-mode voltage     | 0               | \( V_{CC}−1.5 \) V |     |     |      |
| Voltage gain                  | 200             | V/mV|     |     |      |
| Output sink current           | \( V_C = 1 \)   | 40  | 50  |     | mA   |
| Output source current         | \( V_E = V_{CC}−2 \) | 10  |     |     | mA   |
| Saturation voltage            | \( I_{SINK} = 5 \) mA | 0.1 | 0.5 | V    |
|                            | \( I_{SINK} = 20 \) mA | 1   |     | V    |
|                            | \( I_{SINK} = 50 \) mA | 1   | 1.5 | V    |

**ZENER REGULATOR**

| \( R_{DROP} \)                  | 470 \( \Omega \) | 7.56 |     |     | V    |
| Series resistance             | 10.5            | 15  |     |     | \( \Omega \) |
| Temperature stability         | 1               |     |     |     | mV/°C|
| Total supply current          | 3.8             | 6   |     |     | mA   |

(1) Hysteresis is the sum \( V_{TH} – (–V_{TH}) \), offset voltage is their difference. See test circuit.

(2) \( V_{OH} = 0.75 \times V_{CC} – 1 \) \( V_{BE} \) and \( V_{OL} = 0.25 \times V_{CC} – 1 \) \( V_{BE} \), therefore \( V_{OH} – V_{OL} = V_{CC} / 2 \). The difference \( (V_{OH} – V_{OL}) \) and the mirror gain \( (I_2 / I_3) \) are the two factors that cause the tachometer gain constant to vary from 1.

(3) Ensure that when choosing the time constant \( R_1 \times C_1 \) that the maximum anticipated output voltage at \( CP2/IN+ \) can be reached with \( I_3 \) × \( R_1 \). The maximum value for \( R_1 \) is limited by the output resistance of \( CP2/IN+ \) which is greater than 10 M\( \Omega \) typically.

(4) Nonlinearity is defined as the deviation of \( V_{OUT} \) (at \( CP2/IN+ \)) for \( f_{IN} = 5 \) kHz from a straight line defined by the \( V_{OUT} \) at 1 kHz and \( V_{OUT} \) at 10 kHz. \( C_1 = 1000 \) pF, \( R_1 = 68 \) k\( \Omega \) and \( C_2 = 0.22 \) \( \mu F \).
7.6 Typical Characteristics

![Figure 1. Tachometer Linearity vs Temperature](image1)

![Figure 2. Tachometer Linearity vs Temperature](image2)

![Figure 3. Total Supply Current](image3)

![Figure 4. Zener Voltage vs Temperature](image4)

![Figure 5. Normalized Tachometer Output (K) vs Temperature](image5)

![Figure 6. Normalized Tachometer Output (K) vs Temperature](image6)
Typical Characteristics (continued)

Figure 7. Tachometer Currents $I_2$ and $I_3$ vs Supply Voltage

Figure 8. Tachometer Currents $I_2$ and $I_3$ vs Temperature

Figure 9. Tachometer Linearity vs R1

Figure 10. Tachometer Input Hysteresis vs Temperature

Figure 11. Op Amp Output Transistor Characteristics

Figure 12. Op Amp Output Transistor Characteristics
8 Parameter Measurement Information

![Test Circuit Diagram](image)

Figure 13. Test Circuit

![Threshold Measurement Diagram](image)

Figure 14. Tachometer Input Threshold Measurement
9 Detailed Description

9.1 Overview

The LM29x7 frequency-to-voltage converter features two separate inputs to monitor the signal. In the 8-pin devices, one of these inputs is internally grounded and therefore it monitors the remaining input for zero crossings. In the 14-pin devices, both of these inputs are open and it instead detects whenever the differential voltage switches polarity. Therefore, the input comparator outputs a square wave of equal frequency to the input.

A charge pump system is used to translate the frequency of this square wave to a voltage. At the start of every positive half cycle of the input signal a 180-µA constant current charges C1 until its voltage has increased by \( V_{CC}/2 \). The capacitor is held at that voltage until the input signal begins a negative half cycle. Then the 180-µA constant current discharges capacitor C1 until its voltage has dropped by \( V_{CC}/2 \). This voltage is held until the next positive half cycle and the process repeats. This generates pulses of current flowing into and out of capacitor C1 at the same frequency as the input signal. For every full cycle, the charge pump mirrors both current pulses as positive current pulses into the parallel combination of resistor R1 and capacitor C2. Therefore every full cycle, the amount of charge leaving pin 3 is equal to the sum of the charge entering C1 and leaving C1. Because the voltage at pin 3 is equal to \( I_3(\text{avg}) \times R1 \), \( I(\text{avg}) \) is calculated in Equation 1.

\[
I_3(\text{avg}) = \frac{Q}{t} = \frac{Q_{\text{charge}} + Q_{\text{discharge}}}{(1 / f)} = 2 \times Q \times f \times C_1 \times \frac{V_{CC}}{2} \times f = C_1 \times V_{CC} \times f
\]  

(1)

This average current will be flowing across R1, giving the output voltage in Equation 2.

\[
V_o = R_1 \times C_1 \times V_{CC} \times f
\]

(2)

C2 acts as a filter to smooth the pulses of current and does not affect the output voltage. However, the size of C2 determines both the output response time for changes in frequency and the amount of output voltage ripple.

The voltage generated is then fed in a high gain op amp. This op amp drives a bipolar transistor whose collector and emitter are each broken out to a pin. The LM29x7 has the flexibility to be configured a variety of ways to meet system requirements including voltage output, driving loads, operating a relay, and more.
9.2 Functional Block Diagram

![Functional Block Diagram](image)

*This connection made on 8-pin LM2907 and LM2917 only.

**This connection made on LM2917 and 8-pin LM2917 only.

9.3 Feature Description

9.3.1 Differential Input

This device features a Schmitt-Trigger comparator that is the first stage in converting the input signal. Every time the output of the comparator flips between high and low correlates to a half cycle elapsing on the input signal. On the LM29x7-8 devices, one terminal of this comparator is internally connected to ground. This requires that the input signal cross zero volts in order for device to detect the frequency. On the LM29x7 devices, the input terminals to the Schmitt-Trigger comparator are both available for use. This open terminal allows the potential at which the comparator’s output is flipped to be applied externally. This allows the device to accept signals with DC offset or compare differential inputs.

9.3.2 Configurable

While the ratio of output voltage to input frequency is dependent on supply voltage, it is easily adjusted through the combination of one resistor and one capacitor, R1 and C1. The formula for calculating the expected output voltage is in **Equation 3**.

\[
V_{\text{OUT}} = V_{\text{CC}} \times f \times C1 \times R1.
\] (3)

The sizes of R1 and C1 have other effects on the system such as maximum frequency and output linearity. See **Choosing R1 and C1** for detailed instructions on sizing components.
Feature Description (continued)

9.3.3 Output Stage

The output voltage generated by the charge pump is fed in the noninverting terminal of a high gain op amp. This op amp then drives and uncommitted bipolar junction transistor. This allows the LM2907 to be configured a variety of ways to meet system needs. The output voltage can be buffered and used to drive a load (see Figure 15) or an output threshold can be given to trigger a load switch (see Figure 18).

9.4 Device Functional Modes

9.4.1 Grounded Input Devices (8-Pin LM2907 and LM2917)

These devices have one of the two Schmitt-Trigger comparator inputs internally grounded and must be externally connected to the system ground as well. This configuration monitors the remaining terminal for zero crossings.

9.4.2 Differential Input Devices (LM2907 and LM2917)

These devices have both inputs to the Schmitt-Trigger comparator available and broken out to pins 1 and 11. This configuration allows a new switching threshold provided in the case of signals with DC offset or to intake a differential pair and switch based on voltage difference.
10 Application and Implementation

NOTE
Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI’s customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

10.1 Application Information

The LM2907 series of tachometer circuits is designed for minimum external part count applications and maximum versatility. To fully exploit its features and advantages, first examine its theory of operation. The first stage of operation is a differential amplifier driving a positive feedback flip-flop circuit. The input threshold voltage is the amount of differential input voltage at which the output of this stage changes state. Two options (8-pin LM2907 and LM2917) have one input internally grounded so that an input signal must swing above and below ground and exceed the input thresholds to produce an output. This is offered specifically for magnetic variable reluctance pickups which typically provide a single-ended AC output. This single input is also fully protected against voltage swings to ±28 V, which are easily attained with these types of pickups.

The differential input options (LM2907, LM2917) give the user the option of setting his own input switching level and still have the hysteresis around that level for excellent noise rejection in any application. Of course to allow the inputs to attain common-mode voltages above ground, input protection is removed and neither input should be taken outside the limits of the supply voltage being used. It is very important that an input not go below ground without some resistance in its lead to limit the current that will then flow in the epi-substrate diode.

Following the input stage is the charge pump where the input frequency is converted to a DC voltage. To do this requires one timing capacitor, one output resistor, and an integrating or filter capacitor. When the input stage changes state (due to a suitable zero crossing or differential voltage on the input) the timing capacitor is either charged or discharged linearly between two voltages whose difference is \( V_{CC}/2 \). Then in one half cycle of the input frequency or a time equal to \( 1/2 f_{IN} \) the change in charge on the timing capacitor is equal to \( V_{CC}/2 \times C1 \). The average amount of current pumped into or out of the capacitor is shown in Equation 4.

\[
\Delta Q/T = i_{c(AVG)} = C1 \times \frac{V_{CC}}{2} \times (2f_{IN}) = V_{CC} \times f_{IN} \times C1
\]

(4)

The output circuit mirrors this current very accurately into the load resistor \( R1, \) connected to ground, such that if the pulses of current are integrated with a filter capacitor, then \( V_{O} = i_{c} \times R1, \) and the total conversion formula becomes Equation 5.

\[
V_{O} = V_{CC} \times f_{IN} \times C1 \times R1 \times K
\]

where

- \( K \) is the gain constant (typically 1)

(5)

The size of \( C2 \) is dependent only on the amount of ripple voltage allowable and the required response time.
10.2 Typical Applications

10.2.1 Minimum Component Tachometer

\[ V = \frac{C_1 \times f \times V_{O} \times \text{pk-pk}}{C_2} \]

\[ \text{Figure 15. Minimum Component Tachometer Diagram} \]

10.2.1.1 Design Requirements

- C1: This capacitor is charged and discharged every cycle by a 180-µA typical current source. Smaller capacitors can be charged quicker therefore increasing the maximum readable frequency. However, lower capacitors values reduce the output voltage produced for a given frequency. C1 must not be sized lower than 500-pF due to its role in internal compensation.

- R1: This resistor produces the output voltage from current pulses source by the internal charge pump. Higher values increase the output voltage for a given frequency, but too large will degrade the output's linearity. Because the current pulses are a fixed magnitude of 180 µA typical, R1 must be big enough to produce the maximum desired output voltage at maximum input frequency. At maximum input frequency the pulse train duty cycle is 100%, therefore the average current is 180 µA and \( R_1 = \frac{V_{O\text{max}}}{180 \ \mu A} \).

- C2: This capacitor filters the ripple produced by the current pulses sourced by the charge pump. Large values reduce the output voltage ripple but increase the output's response time to changes in input frequency.

- Rload: The load resistance must be large enough that at maximum output voltage, the current is under the rated value of 50 mA.

10.2.1.2 Detailed Design Procedure

10.2.1.2.1 Choosing R1 and C1

There are some limitations on the choice of R1 and C1 which should be considered for optimum performance. The timing capacitor also provides internal compensation for the charge pump and must be kept larger than 500 pF for very accurate operation. Smaller values can cause an error current on R1, especially at low temperatures. Several considerations must be met when choosing R1. The output current at pin 3 is internally fixed and therefore \( V_{O}/R_1 \) must be less than or equal to this value. If R1 is too large, it can become a significant fraction of the output impedance at pin 3 which degrades linearity. Also output ripple voltage must be considered and the size of C2 is affected by R1. An expression that describes the ripple content on pin 3 for a single R1C2 combination is in Equation 6.

\[ V_{\text{RIPPLE}} = \frac{V_{CC}}{2} \times \frac{C_1}{C_2} \times \left( 1 - \frac{V_{CC} \times f_{IN} \times C_1}{I_2} \right) \text{pk-pk} \]

(6)

R1 can be chosen independent of ripple. However, response time, or the time it takes \( V_{OUT} \) to stabilize at a new voltage, increases as the size of C2 increases, so a compromise between ripple, response time, and linearity must be chosen carefully.

As a final consideration, the maximum attainable input frequency is determined by \( V_{CC} \), C1, and I2 in Equation 7.
Typical Applications (continued)

\[ f_{\text{MAX}} = \frac{I_2}{C_1 \times V_{\text{CC}}} \]  

(7)

10.2.1.2.2 Using Zener Regulated Options (LM2917)

For those applications where an output voltage or current must be obtained independent of supply voltage variations, the LM2917 is offered. The most important consideration in choosing a dropping resistor from the unregulated supply to the device is that the tachometer and op amp circuitry alone require about 3 mA at the voltage level provided by the Zener. At low supply voltages there must be some current flowing in the resistor above the 3-mA circuit current to operate the regulator. As an example, if the raw supply varies from 9 V to 16 V, a resistance of 470 Ω minimizes the Zener voltage variation to 160 mV. If the resistance goes under 400 Ω or over 600 Ω, the Zener variation quickly rises above 200 mV for the same input variation.

10.2.1.3 Application Curves

![Graphs showing output response and voltage ripple](image)

**Figure 16. Output Response to an Increase in Input Frequency**

**Figure 17. Output Voltage Ripple**

10.2.2 Other Application Circuits

This section shows application circuit examples using the LM2907-N and LM2917-N devices. Customers must fully validate and test these circuits before implementing a design based on these examples.

![Diagram of a speed switch](image)

**Figure 18. Speed Switch**
Typical Applications (continued)

Figure 19. Zener Regulated Frequency to Voltage Converter

Figure 20. Breaker Point Dwell Meter
Typical Applications (continued)

Figure 21. Voltage Driven Meter Indicating Engine RPM

Figure 22. Current Driven Meter Indicating Engine RPM
Typical Applications (continued)

\[ V_{OUT} = 1 \text{ V to } 10 \text{ V for } CX = 0.01 \text{ to } 0.1 \text{ mF}\text{d and } R = 111 \text{ k}\Omega \]

**Figure 23. Capacitance Meter**

\[ \text{CURRENT \ SENSE} \]

**Figure 24. Two-Wire Remote Speed Switch**
Typical Applications (continued)

V3 steps up in voltage by the amount of \((V_{CC} \times C1) / C2\), for each complete input cycle (2 zero crossings).
For example: if \(C2 = 200 \times C1\) after 100 consecutive input cycles, then \(V3 = 1/2 \times V_{CC}\).

**Figure 25. 100 Cycle Delay Switch**

Flashing begins when \(f_{IN} \geq 100\) Hz
Flash rate increases with input frequency increase beyond trip point.

**Figure 26. Flashing LED Indicates Over-Speed**
\[ f_{\text{POLE}} = \frac{0.707}{(2 \times \pi \times RC)} \]
\[ \tau_{\text{RESPONSE}} = \frac{2.57}{(2 \times \pi \times f_{\text{POLE}})} \]

**Figure 27. Frequency to Voltage Converter With 2 Pole Butterworth Filter to Reduce Ripple**

### 10.2.2.1 Variable Reluctance Magnetic Pickup Buffer Circuits

Precision two-shot output frequency is twice the input frequency

Pulse width = \( \frac{(V_{\text{CC}})}{2} \times (C1 / 12) \)

Pulse height = \( V_{\text{ZENER}} \)

**Figure 28. Magnetic Pickup Buffer With Zener Regulated Output Magnitude**
10.2.2.2 Finger Touch or Contact Switch

Figure 29. Magnetic Pickup Buffer With 1/2 \( V_{CC} \) Output Magnitude

Figure 30. Finger Touch or Contact Switch Diagram

Figure 31. System Output in Response to Consecutive Button Presses
10.2.2.3 Over-Speed Latch

Output latches when \( f_{IN} = \left( \frac{R_2}{R_1 + R_2} \right) \times \left( \frac{1}{RC} \right) \). Output is reset by removing \( V_{CC} \).

Figure 32. Over-Speed Latch Circuit Diagram

Figure 33. \( V_{OUT} \) vs \( F_{IN} \)
10.2.2.4 Frequency Switch Applications

Some frequency switch applications may require hysteresis in the comparator function which can be implemented in several ways. Example circuits are shown in Figure 34 to Figure 36.

**Figure 34. Frequency Switch With Resistor Divider Threshold**

**Figure 35. Frequency Switch With Added Hysteresis**
Figure 36. Frequency Switch With Added Hysteresis

10.2.2.4.1 Application Curves

Figure 37. $V_{OUT}$ vs $V_3$

Figure 38. $V_{OUT}$ vs $V_3$

10.2.2.5 Anti-Skid Circuits

10.2.2.5.1 Select-Low Circuit

Figure 39. Select-Low Circuit Diagram
\( V_{\text{OUT}} \) is proportional to the lower of the two input wheel speeds

**Figure 40.** \( V_{\text{OUT}} \) vs Wheel Speed

### 10.2.2.5.2 Select-High Circuit

\( V_{\text{OUT}} \) is proportional to the higher of the two input wheel speeds

**Figure 42.** \( V_{\text{OUT}} \) vs Wheel Speed
10.2.2.5.3 Select-Average Circuit

Figure 43. Select-Average Circuit Diagram

10.2.2.6 Changing the Output Voltage for an Input Frequency of Zero

Figure 44. Tachometer With Adjustable Zero Speed Voltage Output

Figure 45. \( V_{OUT} \) vs Frequency Plot Shifted to Produce 1 V at Zero Speed
### 10.2.2.7 Changing Tachometer Gain Curve or Clamping the Minimum Output Voltage

![Diagram of Tachometer With Output Clamped at Low Speeds]

**Figure 46.** Tachometer With Output Clamped at Low Speeds

![Plot of V_OUT vs Frequency With Output Clamped at Low Speeds]

**Figure 47.** $V_{OUT}$ vs Frequency Plot With Output Clamped at Low Speeds

### 11 Power Supply Recommendations

This family of devices is designed to operate from a supply voltage up to 28 V. For the 8-pin LM2907 and LM2917, devices with a fixed ground reference for TACH–, the tachometer inputs can intake voltages between ±28 V. LM2907 and LM2917 devices have both tachometer inputs available at the cost of input protection features. This means neither input should be taken outside of the supply voltage range without additional precautions (see [Application Information](#)).
12 Layout

12.1 Layout Guidelines

- Bypass capacitors must be placed as close as possible to the supply pin. When using a through-hole package, it is acceptable to place the bypass capacitor on the bottom layer. All other components must be placed as close to the device as possible.
- Use of a ground plane is recommended to provide a low-impedance ground across the circuit.
- Feedback loops must use short and wide traces.

12.2 Layout Example

Figure 48. Layout Recommendation
13 Device and Documentation Support

13.1 Related Links
The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

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13.2 Receiving Notification of Documentation Updates
To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on Alert me to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

13.3 Community Resources
The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community TI's Engineer-to-Engineer (E2E) Community. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support TI's Design Support Quickly find helpful E2E forums along with design support tools and contact information for technical support.

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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

13.6 Glossary
SLYZ022 — Ti Glossary.
This glossary lists and explains terms, acronyms, and definitions.

14 Mechanical, Packaging, and Orderable Information
The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.
## PACKAGING INFORMATION

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(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".
RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.
Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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**TAPE AND REEL INFORMATION**

**REEL DIMENSIONS**
- Reel Diameter
- Reel Width (W1)

**TAPE DIMENSIONS**
- A0: Dimension designed to accommodate the component width
- B0: Dimension designed to accommodate the component length
- K0: Dimension designed to accommodate the component thickness
- W: Overall width of the carrier tape
- P1: Pitch between successive cavity centers

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**
- Pocket Quadrants
- Sprocket Holes
- User Direction of Feed

*All dimensions are nominal*

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## TAPE AND REEL BOX DIMENSIONS

*All dimensions are nominal

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*All dimensions are nominal*
NOTES:
A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
   ▶ Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0.15) each side.
   ▶ Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0.43) each side.
E. Reference JEDEC MS-012 variation AB.
NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Publication IPC-7351 is recommended for alternate designs.
D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.
NOTES:

1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches.
   Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
4. This dimension does not include interlead flash.
5. Reference JEDEC registration MS-012, variation AA.
EXAMPLE BOARD LAYOUT

NOTES: (continued)
6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.
NOTES:
A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. Falls within JEDEC MS-001 variation BA.
NOTES:
A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
△ Falls within JEDEC MS-001, except 18 and 20 pin minimum body length (Dim A).
△ The 20 pin end lead shoulder width is a vendor option, either half or full width.
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