1 Features

- $V_{\text{IN}}$: 2.5 V to 16 V
- $V_{\text{OUT}}$:
  - 0.8 V to 14.6 V (adjustable)
  - 0.8 V to 6.6 V (fixed, 50-mV steps)
- 1% output accuracy over load and temperature
- Low $I_{\text{Q}}$: 50 µA (typical, ~1.5 µA in shutdown)
- Internal soft-start time: 500 µs (typical)
- Fold-back current limiting and thermal protection
- Stable with 1-µF ceramic capacitors
- High PSRR: 70 dB at 1 kHz, 46 dB at 1 MHz
- Temperature range: –40°C to +125°C

2 Applications

- Appliances
- TVs, monitors, and set top boxes
- Motion detectors (PIR, uWave, and so forth)
- Motor drives and control boards
- Printers and PC peripherals
- Wi-Fi access points and routers

3 Description

The TLV767 is a wide input linear voltage regulator supporting an input voltage range from 2.5 V to 16 V and up to 1 A of load current. The output range is from 0.8 V to 6.6 V or up to 14.6 V in the adjustable version.

Additionally, the TLV767 has a 1% output accuracy that can meet the needs of low voltage microcontrollers (MCUs) and processors.

The TLV767 is designed to have a much lower $I_{\text{Q}}$ than traditional wide-$V_{\text{IN}}$ regulators, thus making the device well positioned to meet the needs of increasingly stringent standby power requirements. When disabled, the TLV767 draws only 1.5 µA of $I_{\text{Q}}$.

The internal soft-start time and foldback current limit reduce inrush current during start up, thus minimizing input capacitance.

Wide bandwidth PSRR performance is greater than 70 dB at 1 kHz and 46 dB at 1 MHz, which helps attenuate the switching frequency of an upstream DC/DC converter and minimizes post regulator filtering. To allow for more flexibility, the TLV767 has both fixed and adjustable versions.

The TLV767 is available in a 6-pin, 2-mm × 2-mm WSON (DRV), an 8-pin 3-mm x 3-mm HVSSOP (DGN), and a 5-pin 2.9-mm x 1.6-mm SOT-23 (DBV) package.

Device Information

<table>
<thead>
<tr>
<th>PART NUMBER</th>
<th>PACKAGE</th>
<th>BODY SIZE (NOM)</th>
</tr>
</thead>
<tbody>
<tr>
<td>TLV767</td>
<td>WSON (6)</td>
<td>2.00 mm × 2.00 mm</td>
</tr>
<tr>
<td></td>
<td>HVSSOP (8)</td>
<td>3.00 mm x 3.00 mm</td>
</tr>
<tr>
<td></td>
<td>SOT-23 (5)</td>
<td>2.90 mm x 1.60 mm</td>
</tr>
</tbody>
</table>

(1) For all available packages, see the orderable addendum at the end of the data sheet.

Reduced Inrush Current With 22 µF at $C_{\text{OUT}}$

Typical Application Circuit
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4 Revision History
NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision C (June 2020) to Revision D (July 2021) Page
  • Added DBV (SOT-23) package to document .......................................................... 1
  • Changed VOUT adjustable Features bullet from 0.8 V to 13.6 V (adjustable) to 0.8 V to 14.6 V (adjustable) .... 1
  • Changed maximum output range for adjustable version from 13.6 V to 14.6 V ........................................ 1
  • Added DBV pinout and pin information to Pin Configuration and Functions section ..................................... 3
  • Added Layout Example for the Fixed DBV Version figure to the Layout Examples section ........................................ 24

Changes from Revision B (April 2019) to Revision C (June 2020) Page
  • Added DGN (HVSSOP) package to document .......................................................... 1
  • Changed Applications section .............................................................................. 1
  • Added DGN pinouts and pin information to Pin Configuration and Functions section ................................. 3
  • Added HVSSOP thermal information .................................................................... 6
  • Added Layout Example for the Fixed HVSSOP Version and Layout Example for the Adjustable HVSSOP Version figures to the Layout Examples section ........................................ 24
5 Pin Configuration and Functions

**Figure 5-1.** DRV Package (Adjustable), 6-Pin WSON, Top View

**Figure 5-2.** DRV Package (Fixed), 6-Pin WSON, Top View

**Figure 5-3.** DGN Package (Adjustable), 8-Pin HVSSOP, Top View

**Figure 5-4.** DGN Package (Fixed), 8-Pin HVSSOP, Top View

**Figure 5-5.** DBV Package (Fixed), 5-Pin SOT-23, Top View
<table>
<thead>
<tr>
<th>PIN</th>
<th>NAME</th>
<th>DRV (Adj)</th>
<th>DRV (Fixed)</th>
<th>DGN (Adj)</th>
<th>DGN (Fixed)</th>
<th>DBV (Fixed)</th>
<th>I/O</th>
<th>DESCRIPTION</th>
</tr>
</thead>
<tbody>
<tr>
<td>EN</td>
<td>4</td>
<td>4</td>
<td>5</td>
<td>5</td>
<td>3</td>
<td>I</td>
<td></td>
<td>Enable pin. Driving the enable pin high enables the device. Driving this pin low disables the device. High and low thresholds are listed in the Electrical Characteristics table. This pin has an internal pullup and can be left floating to enable the device or the pin can be connected to the input pin.</td>
</tr>
<tr>
<td>FB</td>
<td>2</td>
<td>—</td>
<td>2</td>
<td>—</td>
<td>—</td>
<td>I</td>
<td></td>
<td>Feedback pin. Input to the control-loop error amplifier. This pin is used to set the output voltage of the device with the use of external resistors. Do not float this pin. For adjustable-voltage version devices only.</td>
</tr>
<tr>
<td>GND</td>
<td>3, 5</td>
<td>3, 5</td>
<td>4, 6</td>
<td>4, 6</td>
<td>2</td>
<td>—</td>
<td></td>
<td>Ground pin. All ground pins must be grounded.</td>
</tr>
<tr>
<td>DNC</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>4</td>
<td>—</td>
<td></td>
<td>Do not connect to a biased voltage. Tie this pin to ground or leave floating</td>
</tr>
<tr>
<td>IN</td>
<td>6</td>
<td>6</td>
<td>8</td>
<td>8</td>
<td>1</td>
<td>I</td>
<td></td>
<td>Input pin. Use the recommended capacitor value as listed in the Recommended Operating Conditions table. Place the input capacitor as close to the IN and GND pins of the device as possible.</td>
</tr>
<tr>
<td>OUT</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>5</td>
<td>O</td>
<td></td>
<td>Output pin. Use the recommended capacitor value as listed in the Recommended Operating Conditions table. Place the output capacitor as close to the OUT and GND pins of the device as possible.</td>
</tr>
<tr>
<td>SNS</td>
<td>—</td>
<td>2</td>
<td>—</td>
<td>2</td>
<td>—</td>
<td>I</td>
<td></td>
<td>Output sense pin. Connect the SNS pin to the OUT pin, or to remotely sense the output voltage at the load, connect the SNS pin to the load. Do not float this pin. For fixed-voltage version devices only.</td>
</tr>
<tr>
<td>Thermal pad</td>
<td>Pad</td>
<td>Pad</td>
<td>Pad</td>
<td>Pad</td>
<td>—</td>
<td>—</td>
<td>Exposed pad of the package. Connect this pad to ground or leave floating. Connect the thermal pad to a large-area ground plane for best thermal performance.</td>
<td></td>
</tr>
</tbody>
</table>
6 Specifications

6.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)(1)

<table>
<thead>
<tr>
<th>Voltage(2)</th>
<th>MIN</th>
<th>MAX</th>
<th>UNIT</th>
</tr>
</thead>
<tbody>
<tr>
<td>V\text{IN}</td>
<td>–0.3</td>
<td>18</td>
<td>V</td>
</tr>
<tr>
<td>V\text{OUT}(3)</td>
<td>–0.3</td>
<td>V\text{IN} + 0.3</td>
<td>V</td>
</tr>
<tr>
<td>V\text{SNS}(3)</td>
<td>–0.3</td>
<td>V\text{IN} + 0.3</td>
<td>V</td>
</tr>
<tr>
<td>V\text{FB}</td>
<td>–0.3</td>
<td>3</td>
<td></td>
</tr>
<tr>
<td>V\text{EN}</td>
<td>–0.3</td>
<td>18</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Current</th>
<th>Maximum output current</th>
<th>Internally Limited</th>
<th>A</th>
</tr>
</thead>
</table>

<table>
<thead>
<tr>
<th>Temperature</th>
<th>Operating junction (T\text{J})</th>
<th>–50</th>
<th>150</th>
<th>°C</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Storage (T\text{STG})</td>
<td>–65</td>
<td>150</td>
<td></td>
</tr>
</tbody>
</table>

(1) Stresses beyond those listed under Absolute Maximum Rating may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Condition. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltages with respect to GND.

(3) V\text{IN} + 0.3 V or 18 V (whichever is smaller)

6.2 ESD Ratings

<table>
<thead>
<tr>
<th>V_{\text{ESD}}</th>
<th>Electrostatic discharge</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Human body model (HBM), per ANSI/ESDA/ JEDEC JS-001, all pins(1)</td>
</tr>
<tr>
<td></td>
<td>Charged device model (CDM), per JEDEC specification JESD22-C101, all pins(2)</td>
</tr>
</tbody>
</table>

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

<table>
<thead>
<tr>
<th>V\text{IN}</th>
<th>Input voltage</th>
<th>2.5</th>
<th>16</th>
<th>V</th>
</tr>
</thead>
<tbody>
<tr>
<td>V\text{EN}</td>
<td>Enable voltage</td>
<td>0</td>
<td>16</td>
<td>V</td>
</tr>
<tr>
<td>V\text{OUT}</td>
<td>Output voltage</td>
<td>0.8</td>
<td>14.6</td>
<td>V</td>
</tr>
<tr>
<td>I\text{OUT}</td>
<td>Output current (2.5 V ≤ V\text{IN} &lt; 3 V)</td>
<td>0</td>
<td>0.8</td>
<td>A</td>
</tr>
<tr>
<td>I\text{OUT}</td>
<td>Output current (V\text{IN} ≥ 3 V)</td>
<td>0</td>
<td>1</td>
<td>A</td>
</tr>
<tr>
<td>C\text{OUT}</td>
<td>Output capacitor(1)</td>
<td>1</td>
<td>2.2</td>
<td>220</td>
</tr>
<tr>
<td>C\text{OUT ESR}</td>
<td>Output capacitor ESR</td>
<td>2</td>
<td>500</td>
<td>mΩ</td>
</tr>
<tr>
<td>C\text{IN}</td>
<td>Input capacitor</td>
<td>1</td>
<td></td>
<td>µF</td>
</tr>
<tr>
<td>C\text{FF}</td>
<td>Feed-forward capacitor (optional(2), for adjustable device only)</td>
<td>10</td>
<td></td>
<td>pF</td>
</tr>
<tr>
<td>I_{FB_DIVIDER}</td>
<td>Feedback divider current(3), (adjustable device only)</td>
<td>5</td>
<td></td>
<td>µA</td>
</tr>
<tr>
<td>T\text{J}</td>
<td>Junction temperature</td>
<td>–40</td>
<td>125</td>
<td>°C</td>
</tr>
</tbody>
</table>

(1) Effective output capacitance of 0.5 µF minimum required for stability.

(2) C\text{FF} required for stability if the feedback divider current < 5 µA. Feedback divider current = V\text{OUT} / (R_1 + R_2). See Feed-Forward Capacitor (C\text{FF}) section for details.
6.4 Thermal Information

<table>
<thead>
<tr>
<th>THERMAL METRIC(1)</th>
<th>TLV767</th>
<th>UNIT</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>DBV (SOT23)</td>
<td>DGN (HVSSOP)</td>
</tr>
<tr>
<td><strong>RθJA</strong></td>
<td>Junction-to-ambient thermal resistance</td>
<td>165.7</td>
</tr>
<tr>
<td><strong>RθJC(top)</strong></td>
<td>Junction-to-case (top) thermal resistance</td>
<td>61.6</td>
</tr>
<tr>
<td><strong>RθJB</strong></td>
<td>Junction-to-board thermal resistance</td>
<td>37.9</td>
</tr>
<tr>
<td><strong>ΨJT</strong></td>
<td>Junction-to-top characterization parameter</td>
<td>11.6</td>
</tr>
<tr>
<td><strong>ΨJB</strong></td>
<td>Junction-to-board characterization parameter</td>
<td>37.6</td>
</tr>
<tr>
<td><strong>RθJC(bot)</strong></td>
<td>Junction-to-case (bottom) thermal resistance</td>
<td>N/A</td>
</tr>
</tbody>
</table>

(1) For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

6.5 Electrical Characteristics

Specified at $T_J = -40°C$ to 125°C, $V_IN = V_OUT(nom) + 1.5 V$ or $V_IN = 2.5 V$ (whichever is greater), FB/SNS tied to OUT, $I_{OUT} = 10 mA$, $V_EN = 2 V$, $C_IN = 1.0 \mu F$, $C_OUT = 1.0 \mu F$ (unless otherwise noted). Typical values are at $T_J = 25ºC$.

<table>
<thead>
<tr>
<th>PARAMETER</th>
<th>TEST CONDITIONS</th>
<th>MIN</th>
<th>TYP</th>
<th>MAX</th>
<th>UNIT</th>
</tr>
</thead>
<tbody>
<tr>
<td>$V_{OUT}$</td>
<td>Nominal output accuracy</td>
<td>$T_J = 25ºC$</td>
<td>–0.5</td>
<td>0.5</td>
<td>%</td>
</tr>
<tr>
<td>$V_{OUT}$</td>
<td>Output accuracy over temperature</td>
<td>$V_IN \geq 3.0 V$, $1 mA \leq I_{OUT} \leq 1 A$</td>
<td>–1</td>
<td>1</td>
<td>%</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$2.5 V \leq V_IN &lt; 3.0 V$, $1 mA \leq I_{OUT} \leq 800 mA$</td>
<td>–1</td>
<td>1</td>
<td>%</td>
</tr>
<tr>
<td>$V_{FB}$</td>
<td>Feedback voltage</td>
<td></td>
<td>0.8</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>$V_{REF}$</td>
<td>Internal reference (adjustable device)</td>
<td>$T_J = 25ºC$</td>
<td>–0.5</td>
<td>0.5</td>
<td>%</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>–1</td>
<td>1</td>
<td>%</td>
</tr>
<tr>
<td>$I_{FB}$</td>
<td>Feedback pin current</td>
<td>$V_{FB} = 1 V$</td>
<td>10</td>
<td>50</td>
<td>nA</td>
</tr>
<tr>
<td>$\Delta V_{OUT(VIN)}$</td>
<td>Line regulation(1)</td>
<td>$V_{OUT(nom)} + 1.5 V \leq V_IN \leq 16 V$, $I_{OUT} = 10 mA$</td>
<td>0.02</td>
<td>%/V</td>
<td></td>
</tr>
<tr>
<td>$\Delta V_{OUT(IOUT)}$</td>
<td>Load regulation</td>
<td>$1 mA \leq I_{OUT} \leq 1 A$, $V_IN \geq 3.0 V$</td>
<td>0.1</td>
<td>0.5</td>
<td>%/A</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$1 mA \leq I_{OUT} \leq 800 mA$, $2.5 V \leq V_IN &lt; 3.0 V$</td>
<td>0.1</td>
<td>0.5</td>
<td>%/A</td>
</tr>
<tr>
<td>$V_{DO}$</td>
<td>Dropout voltage(2)</td>
<td>$V_IN \geq 3.0 V$, $I_{OUT} = 1 A$, DGN package</td>
<td>0.9</td>
<td>1.5</td>
<td>V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$V_IN \geq 3.0 V$, $I_{OUT} = 1 A$, DRV package</td>
<td>0.9</td>
<td>1.4</td>
<td>V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$2.5 V \leq V_IN &lt; 3.0 V$, $I_{OUT} = 800 mA$</td>
<td>0.8</td>
<td>1.3</td>
<td>V</td>
</tr>
<tr>
<td>$I_{CL}$</td>
<td>Output current limit</td>
<td>$V_{OUT} = 0.9 x V_{OUT(nom)}$, $V_IN \geq 3.0 V$</td>
<td>1.1</td>
<td>1.6</td>
<td>A</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$V_{OUT} = 0.9 x V_{OUT(nom)}$, $2.5 V \leq V_IN &lt; 3.0 V$</td>
<td>0.81</td>
<td>1.6</td>
<td>A</td>
</tr>
<tr>
<td>$I_{SC}$</td>
<td>Short-circuit current limit</td>
<td>$V_{OUT} = 0 V$, DGN package</td>
<td>250</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>$V_{OUT} = 0 V$, DRV package</td>
<td>150</td>
<td>250</td>
<td>350</td>
</tr>
<tr>
<td>$I_Q$</td>
<td>Quiescent current</td>
<td>$I_{OUT} = 0 mA$</td>
<td>50</td>
<td>80</td>
<td>µA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Fixed output devices, $I_{OUT} = 0 mA$</td>
<td>60</td>
<td>95</td>
<td>µA</td>
</tr>
<tr>
<td>$I_{GND}$</td>
<td>Ground current</td>
<td>$I_{OUT} = 1 A$, $V_IN \geq 3.0 V$</td>
<td>1.5</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td>$I_{SHUTDOWN}$</td>
<td>Shutdown current</td>
<td>$V_EN \leq 0.4 V$, $V_IN = 16 V$</td>
<td>1.5</td>
<td>3</td>
<td>µA</td>
</tr>
<tr>
<td>$V_{EN(HIGH)}$</td>
<td>Enable pin logic high</td>
<td>$2.5 V \leq V_IN \leq 16 V$</td>
<td>1.2</td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>$V_{EN(LOW)}$</td>
<td>Enable pin logic low</td>
<td>$2.5 V \leq V_IN \leq 16 V$</td>
<td>0.4</td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>$I_{EN}$</td>
<td>Enable pullup current</td>
<td>$V_EN = 0 V$</td>
<td>400</td>
<td>nA</td>
<td></td>
</tr>
<tr>
<td>$I_{PULLDOWN}$</td>
<td>Output pulldown current</td>
<td>$V_IN = 16 V$, $V_{OUT} = 2.5 V$, $V_{EN}=0V$</td>
<td>1.2</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td>$PSRR$</td>
<td>Power-supply rejection ratio</td>
<td>$V_IN = 3.3 V$, $V_{OUT} = 1.8 V$, $I_{OUT} = 300 mA$, $f = 120 Hz$</td>
<td>70</td>
<td>dB</td>
<td></td>
</tr>
<tr>
<td>$V_n$</td>
<td>Output noise voltage</td>
<td>$BW = 10 Hz$ to $100 kHz$, $V_IN = 3.3 V$, $V_{OUT} = 0.8 V$, $I_{OUT} = 100 mA$</td>
<td>60</td>
<td>µV RMS</td>
<td></td>
</tr>
<tr>
<td>$V_{UVLO+}$</td>
<td>UVLO threshold rising</td>
<td>$V_IN$ rising</td>
<td>2.2</td>
<td>2.4</td>
<td>V</td>
</tr>
<tr>
<td>$V_{UVLO(HYS)}$</td>
<td>UVLO hysteresis</td>
<td></td>
<td>130</td>
<td>mV</td>
<td></td>
</tr>
</tbody>
</table>
6.5 Electrical Characteristics (continued)

Specified at $T_J = -40°C$ to 125°C, $V_{IN} = V_{OUT(nom)} + 1.5$ V or $V_{IN} = 2.5$ V (whichever is greater), FB/SNS tied to OUT, $I_{OUT} = 10$ mA, $V_{EN} = 2$ V, $C_{IN} = 1.0$ μF, $C_{OUT} = 1.0$ μF (unless otherwise noted). Typical values are at $T_J = 25°C$.

<table>
<thead>
<tr>
<th>PARAMETER</th>
<th>TEST CONDITIONS</th>
<th>MIN</th>
<th>TYP</th>
<th>MAX</th>
<th>UNIT</th>
</tr>
</thead>
<tbody>
<tr>
<td>$V_{UVLO}$</td>
<td>UVLO threshold falling</td>
<td>1.9</td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>$T_{SD(shutdown)}$</td>
<td>Thermal shutdown temperature</td>
<td></td>
<td></td>
<td>180</td>
<td>ºC</td>
</tr>
<tr>
<td>$T_{SD(reset)}$</td>
<td>Thermal shutdown reset temperature</td>
<td></td>
<td></td>
<td>160</td>
<td>ºC</td>
</tr>
</tbody>
</table>

(1) Line regulation is measured with $V_{IN} = V_{OUT(nom)} + 1.5$ V or 2.5 V (whichever is greater)
(2) $V_{DO}$ is measured with $V_{IN} = 0.95 \times V_{OUT(nom)}$ for fixed output devices. $V_{DO}$ is not measured for fixed output devices when $V_{OUT} < 2.5$ V. For adjustable output device, $V_{DO}$ is measured with $V_{FB} = 0.95 \times V_{FB(nom)}$.
7 Typical Characteristics

at operating temperature $T_J = 25°C$, $V_{IN} = V_{OUT(NOM)} + 1.5$ V or $2.5$ V (whichever is greater), $I_{OUT} = 10$ mA, $V_{EN} = 2.0$ V, $C_{IN} = 1.0$ µF, and $C_{OUT} = 1.0$ µF (unless otherwise noted)

![Graphs showing output voltage accuracy vs output current for different input voltages and temperature conditions.](image-url)

**Figure 7-1. $V_{OUT}$ Accuracy vs $I_{OUT}$**

![Graph showing quiescent current in shutdown vs input voltage for different temperature conditions.](image-url)

**Figure 7-4. $I_{SHUTDOWN}$ vs $V_{IN}$**

![Graph showing quiescent current vs input voltage for different temperature conditions.](image-url)

**Figure 7-5. $I_Q$ vs $V_{IN}$**

![Graph showing quiescent current vs temperature for different voltage versions.](image-url)

**Figure 7-6. $I_Q$ vs Temperature**
7 Typical Characteristics (continued)

at operating temperature $T_J = 25°C$, $V_{IN} = V_{OUT(NOM)} + 1.5$ V or 2.5 V (whichever is greater), $I_{OUT} = 10$ mA, $V_{EN} = 2.0$ V, $C_{IN} = 1.0$ µF, and $C_{OUT} = 1.0$ µF (unless otherwise noted)

Figure 7-7. $I_{GND}$ vs $I_{OUT}$

Figure 7-8. $I_{GND}$ vs $I_{OUT}$

Figure 7-9. $I_Q$ Increase Below Minimum $V_{IN}$

Figure 7-10. $I_{OUT}$ Transient From 0 mA to 100 mA

Figure 7-11. $I_{OUT}$ Transient From 1 mA to 1 A

Figure 7-12. $I_{OUT}$ Transient From 250 mA to 850 mA

$V_{IN} = 5$ V, $V_{OUT} = 3.3$ V, $C_{PF} = 10$ pF, ramp rate = 0.4 A/µs

$V_{IN} = 5$ V, $V_{OUT} = 3.3$ V, ramp rate = 0.8 A/µs

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Product Folder Links: TLV767
7 Typical Characteristics (continued)

at operating temperature $T_J = 25^\circ C$, $V_{IN} = V_{OUT(NOM)} + 1.5 \text{ V or 2.5 V (whichever is greater)}$, $I_{OUT} = 10 \text{ mA}$, $V_{EN} = 2.0 \text{ V}$, $C_{IN} = 1.0 \mu\text{F}$, and $C_{OUT} = 1.0 \mu\text{F}$ (unless otherwise noted)

$V_{OUT} = 3.3 \text{ V}, I_{OUT} = 1 \text{ A}, V_{IN}$ ramp rate = 0.6 V/µs

**Figure 7-13. $V_{IN}$ Transient in Dropout From 4 V to 13 V**

$V_{OUT} = 3.3 \text{ V}, I_{OUT} = 33 \mu\text{A}, V_{IN}$ ramp rate = 1.6 V/µs

**Figure 7-14. $V_{IN}$ Transient From 5 V to 16 V**

$V_{IN} = 3.0 \text{ V}$

**Figure 7-17. $V_{DO}$ vs $I_{OUT}$**

$V_{IN} = 2.5 \text{ V}$

**Figure 7-18. $V_{DO}$ vs $I_{OUT}$**
7 Typical Characteristics (continued)

at operating temperature $T_J = 25^\circ C$, $V_{IN} = V_{OUT}(\text{NOM}) + 1.5 \text{ V or 2.5 V (whichever is greater)}$, $I_{OUT} = 10 \text{ mA}$, $V_{EN} = 2.0 \text{ V}$, $C_{IN} = 1.0 \mu \text{F}$, and $C_{OUT} = 1.0 \mu \text{F}$ (unless otherwise noted)

![Figure 7-19. Foldback Current Limit vs Temperature](image)

![Figure 7-20. Foldback Current Limit vs Temperature](image)

![Figure 7-21. Startup With Separate $V_{EN}$ and $V_{IN}$](image)

![Figure 7-22. Startup With $V_{EN}$ Floating](image)

![Figure 7-23. $V_{EN}$ Thresholds vs Temperature](image)

![Figure 7-24. $V_{EN}$ Thresholds vs Temperature](image)
7 Typical Characteristics (continued)

at operating temperature $T_J = 25°C$, $V_{IN} = V_{OUT(\text{NOM})} + 1.5 \text{ V or 2.5 V (whichever is greater)}$, $I_{OUT} = 10 \text{ mA}$, $V_{EN} = 2.0 \text{ V}$, $C_{IN} = 1.0 \mu\text{F}$, and $C_{OUT} = 1.0 \mu\text{F}$ (unless otherwise noted)

---

**Figure 7-25. UVLO Thresholds vs Temperature**

$V_{OUT} = 1.8 \text{ V}$, $V_{IN} = 3.3 \text{ V}$, $C_{FF} = 1 \text{ nF}$

**Figure 7-26. PSRR vs $I_{OUT}$**

$V_{OUT} = 3.3 \text{ V}$, $V_{IN} = 4.8 \text{ V}$, $I_{OUT} = 0.33 \text{ A}$

**Figure 7-27. PSRR vs $V_{IN}$**

$V_{OUT} = 1.8 \text{ V}$, $I_{OUT} = 0.55 \text{ A}$, $C_{FF} = 1 \text{ nF}$

**Figure 7-28. PSRR vs $C_{FF}$**

$V_{OUT} = 1.8 \text{ V}$, RMS Noise = 66.4 $\mu\text{V Rms}$

$V_{OUT} = 3.3 \text{ V}$, RMS Noise = 216.5 $\mu\text{V Rms}$

---

**Figure 7-29. Output Noise ($V_n$) vs $V_{OUT}$**

$C_{FF} = 0 \text{ nF}$, $I_{OUT} = 0.1 \text{ A}$, RMS noise BW = 10 Hz to 100 kHz

---

**Figure 7-30. $I_{EN}$ vs $V_{IN}$**

$V_{EN} = 0 \text{ V}$

---
7 Typical Characteristics (continued)

at operating temperature $T_J = 25^\circ C$, $V_{IN} = V_{OUT(NOM)} + 1.5\, V$ or $2.5\, V$ (whichever is greater), $I_{OUT} = 10\, mA$, $V_{EN} = 2.0\, V$, $C_{IN} = 1.0\, \mu F$, and $C_{OUT} = 1.0\, \mu F$ (unless otherwise noted)

---

**Figure 7-31. $I_{PULLDOWN}$ vs $V_{IN}$**

**Figure 7-32. $I_{FB}$ vs Temperature**

**Figure 7-33. Startup Inrush Current With $C_{OUT} = 22\, \mu F$**

**Figure 7-34. Startup Inrush Current With $C_{OUT} = 47\, \mu F$**
8 Detailed Description
8.1 Overview

The TLV767 is a low quiescent current, high PSRR linear regulator capable of handling up to 1 A of load current. Unlike typical high current linear regulators, the TLV767 consumes significantly less quiescent current. This device is ideal for high current applications that require very sensitive power-supply rails.

This device features integrated foldback current limit, thermal shutdown, output enable, internal output pulldown and undervoltage lockout (UVLO). This device delivers excellent line and load transient performance. This device is low noise and exhibits a very good PSRR. The operating ambient temperature range of the device is −40°C to +125°C.

8.2 Functional Block Diagrams

Figure 8-1. Adjustable Version Block Diagram
8.3 Feature Description

8.3.1 Output Enable

The enable pin for the device is an active-high pin. The output voltage is enabled when the voltage of the enable pin is greater than the high-level input voltage of the EN pin and disabled with the enable pin voltage is less than the low-level input voltage of the EN pin. If independent control of the output voltage is not needed, connect the enable pin to the input of the device.

This device has an internal pullup current on the EN pin. The EN pin can be left floating to enable the device. The device has an internal pulldown circuit that activates when the device is disabled to actively discharge the output voltage.

8.3.2 Dropout Voltage

Dropout voltage ($V_{DO}$) is defined as the input voltage minus the output voltage ($V_{IN} - V_{OUT}$) at the rated output current ($I_{RATED}$), where the pass transistor is fully on. $I_{RATED}$ is the maximum $I_{OUT}$ listed in the Recommended Operating Conditions table. The pass transistor is in the ohmic or triode region of operation, and acts as a switch. The dropout voltage indirectly specifies a minimum input voltage greater than the nominal programmed output voltage at which the output voltage is expected to stay in regulation. If the input voltage falls to less than the nominal output regulation, then the output voltage falls as well.

For a CMOS regulator, the dropout voltage is determined by the drain-source on-state resistance ($R_{DS(ON)}$) of the pass transistor. Therefore, if the linear regulator operates at less than the rated current, the dropout voltage for that current scales accordingly. The following equation calculates the $R_{DS(ON)}$ of the device.

$$R_{DS(ON)} = \frac{V_{DO}}{I_{RATED}}$$  

(1)

8.3.3 Foldback Current Limit

The device has an internal current limit circuit that protects the regulator during transient high-load current faults or shorting events. The current limit is a hybrid brickwall-foldback scheme. The current limit transitions from a
brickwall scheme to a foldback scheme at the foldback voltage \((V_{FOLDBACK})\). In a high-load current fault with
the output voltage above \(V_{FOLDBACK}\), the brickwall scheme limits the output current to the current limit \((I_{CL})\).
When the voltage drops below \(V_{FOLDBACK}\), a foldback current limit activates that scales back the current as the
output voltage approaches GND. When the output is shorted, the device supplies a typical current called the
short-circuit current limit \((I_{SC})\). \(I_{CL}\) and \(I_{SC}\) are listed in the Electrical Characteristics table.

For this device, \(V_{FOLDBACK} = 50\% \times V_{OUT(nom)}\).

The output voltage is not regulated when the device is in current limit. When a current limit event occurs, the
device begins to heat up because of the increase in power dissipation. When the device is in brickwall current
limit, the pass transistor dissipates power \([V_{IN} - V_{OUT}] \times I_{CL}\). When the device output is shorted and the output
is below \(V_{FOLDBACK}\), the pass transistor dissipates power \([V_{IN} - V_{OUT}] \times I_{SC}\). If thermal shutdown is triggered,
the device turns off. After the device cools down, the internal thermal shutdown circuit turns the device back on.
If the output current fault condition continues, the device cycles between current limit and thermal shutdown. For
more information on current limits, see the Know Your Limits application report.

Figure 8-3 shows a diagram of the foldback current limit.

![Figure 8-3. Foldback Current Limit](image)

8.3.4 Undervoltage Lockout (UVLO)

The device has an independent undervoltage lockout (UVLO) circuit that monitors the input voltage, allowing a
controlled and consistent turn on and off of the output voltage. To prevent the device from turning off if the input
drops during turn on, the UVLO has hysteresis as specified in the Electrical Characteristics table.

8.3.5 Output Pulldown

The device has an output pulldown circuit. \(V_{OUT}\) pulldown sink to ground capability is listed in the Electrical
Characteristics table. The output pulldown activates under the following conditions:

- Device disabled
- \(1.0 \, V < V_{IN} < V_{UVLO}\)

The output pulldown current for this device is 1.2 mA typical, as listed in the Electrical Characteristics table.
Do not rely on the output pulldown circuit for discharging a large amount of output capacitance after the input supply has collapsed because reverse current can flow from the output to the input. This reverse current flow can cause damage to the device. See the Reverse Current section for more details.

8.3.6 Thermal Shutdown

The device contains a thermal shutdown protection circuit to disable the device when the junction temperature \( T_J \) of the pass transistor rises to \( T_{SD(shutdown)} \) (typical). Thermal shutdown hysteresis assures that the device resets (turns on) when the temperature falls to \( T_{SD(reset)} \) (typical).

The thermal time-constant of the semiconductor die is fairly short, thus the device may cycle on and off when thermal shutdown is reached until power dissipation is reduced. Power dissipation during start up can be high from large \( V_{IN} - V_{OUT} \) voltage drops across the device or from high inrush currents charging large output capacitors. Under some conditions, the thermal shutdown protection disables the device before start up completes.

For reliable operation, limit the junction temperature to the maximum listed in the Recommended Operating Conditions table. Operation above this maximum temperature causes the device to exceed its operational specifications. Although the internal protection circuitry of the device is designed to protect against thermal overall conditions, this circuitry is not intended to replace proper heat sinking. Continuously running the device into thermal shutdown or above the maximum recommended junction temperature reduces long-term reliability.
8.4 Device Functional Modes

8.4.1 Device Functional Mode Comparison

Table 8-1 shows the conditions that lead to the different modes of operation. See the Electrical Characteristics table for parameter values.

<table>
<thead>
<tr>
<th>OPERATING MODE</th>
<th>PARAMETER</th>
<th>V\textsubscript{IN}</th>
<th>V\textsubscript{EN}</th>
<th>I\textsubscript{OUT}</th>
<th>T\textsubscript{J}</th>
</tr>
</thead>
<tbody>
<tr>
<td>Normal operation</td>
<td>V\textsubscript{IN} &gt; V\textsubscript{OUT(nom)} + V\textsubscript{DO} and V\textsubscript{IN} &gt; V\textsubscript{IN(min)}</td>
<td>V\textsubscript{EN} &gt; V\textsubscript{EN(HI)}</td>
<td>I\textsubscript{OUT} &lt; I\textsubscript{OUT(max)}</td>
<td>T\textsubscript{J} &lt; T\textsubscript{SD(shutdown)}</td>
<td></td>
</tr>
<tr>
<td>Dropout operation</td>
<td>V\textsubscript{IN(min)} &lt; V\textsubscript{IN} &lt; V\textsubscript{OUT(nom)} + V\textsubscript{DO}</td>
<td>V\textsubscript{EN} &gt; V\textsubscript{EN(HI)}</td>
<td>I\textsubscript{OUT} &lt; I\textsubscript{OUT(max)}</td>
<td>T\textsubscript{J} &lt; T\textsubscript{SD(shutdown)}</td>
<td></td>
</tr>
<tr>
<td>Disabled (any true condition disables the device)</td>
<td>V\textsubscript{IN} &lt; V\textsubscript{UVLO}</td>
<td>V\textsubscript{EN} &lt; V\textsubscript{EN(LOW)}</td>
<td>Not applicable</td>
<td>T\textsubscript{J} &gt; T\textsubscript{SD(shutdown)}</td>
<td></td>
</tr>
</tbody>
</table>

8.4.2 Normal Operation

The device regulates to the nominal output voltage when the following conditions are met:

- The input voltage is greater than the nominal output voltage plus the dropout voltage (V\textsubscript{OUT(nom)} + V\textsubscript{DO})
- The output current is less than the current limit (I\textsubscript{OUT} < I\textsubscript{CL})
- The device junction temperature is less than the thermal shutdown temperature (T\textsubscript{J} < T\textsubscript{SD})
- The enable voltage has previously exceeded the enable rising threshold voltage and has not yet decreased to less than the enable falling threshold

8.4.3 Dropout Operation

If the input voltage is lower than the nominal output voltage plus the specified dropout voltage, but all other conditions are met for normal operation, the device operates in dropout mode. In this mode, the output voltage tracks the input voltage. During this mode, the transient performance of the device becomes significantly degraded because the pass transistor is in the ohmic or triode region, and acts as a switch. Line or load transients in dropout can result in large output-voltage deviations.

When the device is in a steady dropout state (defined as when the device is in dropout, V\textsubscript{IN} < V\textsubscript{OUT(NOM)} + V\textsubscript{DO}, directly after being in a normal regulation state, but not during startup), the pass transistor is driven into the ohmic or triode region. When the input voltage returns to a value greater than or equal to the nominal output voltage plus the dropout voltage (V\textsubscript{OUT(NOM)} + V\textsubscript{DO}), the output voltage can overshoot for a short period of time while the device pulls the pass transistor back into the linear region.

8.4.4 Disabled

The output of the device can be shutdown by forcing the voltage of the enable pin to less than the maximum EN pin low-level input voltage (see the Electrical Characteristics table). When disabled, the pass transistor is turned off, internal circuits are shutdown, and the output voltage is actively discharged to ground by an internal discharge circuit from the output to ground.
9 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI’s customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

9.1 Application Information

9.1.1 Adjustable Device Feedback Resistors

The adjustable-version device requires external feedback divider resistors to set the output voltage. \( V_{OUT} \) is set using the feedback divider resistors, \( R_1 \) and \( R_2 \), according to the following equation:

\[
V_{OUT} = V_{FB} \times \left(1 + \frac{R_1}{R_2}\right)
\]  

(2)

To ignore the FB pin current error term in the \( V_{OUT} \) equation, set the feedback divider current to 100x the FB pin current listed in the Electrical Characteristics table. This setting provides the maximum feedback divider series resistance, as shown in the following equation:

\[
R_1 + R_2 \leq \frac{V_{OUT}}{(I_{FB} \times 100)}
\]  

(3)

9.1.2 Recommended Capacitor Types

The device is designed to be stable using low equivalent series resistance (ESR) ceramic capacitors at the input and output. Multilayer ceramic capacitors have become the industry standard for these types of applications and are recommended, but must be used with good judgment. Ceramic capacitors that employ X7R-, X5R-, and C0G-rated dielectric materials provide relatively good capacitive stability across temperature, whereas the use of Y5V-rated capacitors is discouraged because of large variations in capacitance.

Regardless of the ceramic capacitor type selected, the effective capacitance varies with operating voltage and temperature. As a rule of thumb, expect the effective capacitance to decrease by as much as 50%. The input and output capacitors recommended in the Recommended Operating Conditions table account for an effective capacitance of approximately 50% of the nominal value.

9.1.3 Input and Output Capacitor Requirements

Although an input capacitor is not required for stability, good analog design practice is to connect a capacitor from IN to GND. This capacitor counteracts reactive input sources and improves transient response, input ripple, and PSRR. An input capacitor is recommended if the source impedance is more than 0.5 Ω. A higher value capacitor may be necessary if large, fast rise-time load or line transients are anticipated or if the device is located several inches from the input power source.

Dynamic performance of the device is improved with the use of an output capacitor. Use an output capacitor within the range specified in the Recommended Operating Conditions table for stability.

9.1.4 Reverse Current

Excessive reverse current can damage this device. Reverse current flows through the intrinsic body diode of the pass transistor instead of the normal conducting channel. At high magnitudes, this current flow degrades the long-term reliability of the device.

Conditions where reverse current can occur are outlined in this section, all of which can exceed the absolute maximum rating of \( V_{OUT} \leq V_{IN} + 0.3 \) V.

• If the device has a large \( C_{OUT} \) and the input supply collapses with little or no load current
• The output is biased when the input supply is not established
• The output is biased above the input supply
If reverse current flow is expected in the application, external protection is recommended to protect the device. Reverse current is not limited in the device, so external limiting is required if extended reverse voltage operation is anticipated.

Figure 9-1 shows one approach for protecting the device.

![Figure 9-1. Example Circuit for Reverse Current Protection Using a Schottky Diode](image)

9.1.5 Feed-Forward Capacitor (C_FF)

For the adjustable-voltage version device, a feed-forward capacitor (C_FF) can be connected from the OUT pin to the FB pin. C_FF improves transient, noise, and PSRR performance, but is not required for regulator stability. Recommended C_FF values are listed in the Recommended Operating Conditions table. A higher capacitance C_FF can be used; however, the start-up time increases. For a detailed description of C_FF tradeoffs, see the Pros and Cons of Using a Feedforward Capacitor with a Low-Dropout Regulator application report.

C_FF and R_1 form a zero in the loop gain at frequency f_Z, while C_FF, R_1, and R_2 form a pole in the loop gain at frequency f_P. C_FF zero and pole frequencies can be calculated from the following equations:

\[ f_Z = \frac{1}{2 \pi \times C_{FF} \times R_1} \]  
\[ f_P = \frac{1}{2 \pi \times C_{FF} \times (R_1 || R_2)} \]

C_FF ≥ 10 pF is required for stability if the feedback divider current is less than 5 µA. Equation 6 calculates the feedback divider current.

\[ I_{FB Divider} = \frac{V_{OUT}}{R_1 + R_2} \]  

To avoid start-up time increases from C_FF, limit the product C_FF × R_1 < 50 µs.

For an output voltage of 0.8 V with the FB pin tied to the OUT pin, no C_FF is used.

9.1.6 Power Dissipation (P_D)

Circuit reliability requires consideration of the device power dissipation, location of the circuit on the printed circuit board (PCB), and correct sizing of the thermal plane. The PCB area around the regulator must have few or no other heat-generating devices that cause added thermal stress.

To first-order approximation, power dissipation in the regulator depends on the input-to-output voltage difference and load conditions. The following equation calculates power dissipation (P_D).

\[ P_D = (V_{IN} - V_{OUT}) \times I_{OUT} \]  

**Note**

Power dissipation can be minimized, and therefore greater efficiency can be achieved, by correct selection of the system voltage rails. For the lowest power dissipation use the minimum input voltage required for correct output regulation.

For devices with a thermal pad, the primary heat conduction path for the device package is through the thermal pad to the PCB. Solder the thermal pad to a copper pad area under the device. This pad area must contain an array of plated vias that conduct heat to additional copper planes for increased heat dissipation.

The maximum power dissipation determines the maximum allowable ambient temperature (T_A) for the device. According to the following equation, power dissipation and junction temperature are most often related by the
junction-to-ambient thermal resistance ($R_{θJA}$) of the combined PCB and device package and the temperature of the ambient air ($T_A$).

$$T_J = T_A + (R_{θJA} \times P_D)$$

(8)

Thermal resistance ($R_{θJA}$) is highly dependent on the heat-spreading capability built into the particular PCB design, and therefore varies according to the total copper area, copper weight, and location of the planes. The junction-to-ambient thermal resistance listed in the *Thermal Information* table is determined by the JEDEC standard PCB and copper-spreading area, and is used as a relative measure of package thermal performance.

**9.1.7 Estimating Junction Temperature**

The JEDEC standard now recommends the use of psi ($Ψ$) thermal metrics to estimate the junction temperatures of the linear regulator when in-circuit on a typical PCB board application. These metrics are not thermal resistance parameters and instead offer a practical and relative way to estimate junction temperature. These psi metrics are determined to be significantly independent of the copper area available for heat-spreading. The *Thermal Information* table lists the primary thermal metrics, which are the junction-to-top characterization parameter ($ψ_{JT}$) and junction-to-board characterization parameter ($ψ_{JB}$). These parameters provide two methods for calculating the junction temperature ($T_J$), as described in the following equations. Use the junction-to-top characterization parameter ($ψ_{JT}$) with the temperature at the center-top of device package ($T_T$) to calculate the junction temperature. Use the junction-to-board characterization parameter ($ψ_{JB}$) with the PCB surface temperature 1 mm from the device package ($T_B$) to calculate the junction temperature.

$$T_J = T_T + ψ_{JT} \times P_D$$

(9)

where:

- $P_D$ is the dissipated power
- $T_T$ is the temperature at the center-top of the device package

$$T_J = T_B + ψ_{JB} \times P_D$$

(10)

where

- $T_B$ is the PCB surface temperature measured 1 mm from the device package and centered on the package edge

For detailed information on the thermal metrics and how to use them, see the *Semiconductor and IC Package Thermal Metrics* application report.
9.2 Typical Application

This section discusses implementing this device for a typical application. Figure 9-2 shows the application circuit.

![Figure 9-2. Typical Application Circuit](image)

9.2.1 Design Requirements

Table 9-1 summarizes the design requirements for this application.

<table>
<thead>
<tr>
<th>PARAMETER</th>
<th>DESIGN REQUIREMENT</th>
</tr>
</thead>
<tbody>
<tr>
<td>Input voltage</td>
<td>5 V</td>
</tr>
<tr>
<td>Output voltage</td>
<td>3.3 V</td>
</tr>
<tr>
<td>Output current</td>
<td>100 mA</td>
</tr>
</tbody>
</table>

9.2.2 Detailed Design Procedure

9.2.2.1 Transient Response

As with any regulator, increasing the size of the output capacitor reduces overshoot and undershoot magnitude. If load transients are expected with ramp rates greater than 0.5 A/µs, use a 2.2-µF or larger output capacitor.

9.2.2.2 Choose Feedback Resistors

For this design example, \( V_{OUT} \) is set to 3.3 V. Equation 11 and Equation 12 set the feedback divider resistors for the desired output voltage:

\[
V_{OUT} = V_{FB} \times (1 + \frac{R_1}{R_2}) \tag{11}
\]

\[
R_1 + R_2 \leq \frac{V_{OUT}}{(I_{FB} \times 100)} \tag{12}
\]

For improved output accuracy, use Equation 12 and \( I_{FB} = 50 \text{ nA} \) as listed in the Electrical Characteristics table to calculate the upper limit for series feedback resistance (\( R_1 + R_2 \leq 660 \text{ kΩ} \)).

The control-loop error amplifier drives the FB pin to the same voltage as the internal reference (\( V_{FB} = 0.8 \text{ V} \), as listed in the Electrical Characteristics table). Use Equation 11 to determine the ratio of \( R_1 / R_2 = 3.125 \). Use this ratio and solve Equation 12 for \( R_2 \). Now calculate the upper limit for \( R_2 \leq 160 \text{ kΩ} \). Select a standard value resistor for \( R_2 = 160 \text{ kΩ} \).

Reference Equation 11 and solve for \( R_1 \):

\[
R_1 = \frac{V_{OUT}}{V_{FB}} - 1 \times R_2 \tag{13}
\]

From Equation 13, \( R_1 = 500 \text{ kΩ} \) can be determined. Select a standard value resistor for \( R_1 = 499 \text{ kΩ} \). \( V_{OUT} = 3.3 \text{ V} \) (as determined by Equation 11).
9.2.3 Application Curves

![Application Curves Diagram]

\[ V_{\text{IN}} = 5 \, \text{V}, \quad V_{\text{OUT}} = 3.3 \, \text{V}, \quad C_{\text{OUT}} = 1 \, \mu\text{F}, \quad C_{\text{FF}} = 10 \, \text{pF} \]

**Figure 9-3. Load Transient Response, \( I_{\text{OUT}} \) 0 mA to 100 mA**

10 Power Supply Recommendations

This device is designed to operate from an input supply voltage range of 2.5 V to 16 V. To ensure that the output voltage is well regulated and dynamic performance is optimum, the input supply must be at least \( V_{\text{OUT(nom)}} + 1.5 \, \text{V} \). For 1-A output current operation, the input supply must be 3 V or greater. Connect a low output impedance power supply directly to the input pin of the TLV767.
11 Layout

11.1 Layout Guidelines

- Place input and output capacitors as close to the device as possible
- Use copper planes for device connections to IN, OUT, and GND pins to optimize thermal performance
- Place thermal vias around the device to distribute heat

11.2 Layout Examples

![Figure 11-1. Layout Example for the Adjustable WSON Version](image1)

![Figure 11-2. Layout Example for the Fixed WSON Version](image2)

![Figure 11-3. Layout Example for the Fixed HVSSOP Version](image3)

![Figure 11-4. Layout Example for the Adjustable HVSSOP Version](image4)

![Figure 11-5. Layout Example for the Fixed DBV Version](image5)
12 Device and Documentation Support

12.1 Device Support

12.1.1 Device Nomenclature

Table 12-1. Available Options

<table>
<thead>
<tr>
<th>PRODUCT</th>
<th>VOUT</th>
</tr>
</thead>
<tbody>
<tr>
<td>TLV767xx(x)yyy</td>
<td></td>
</tr>
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</table>

- xx(x) is nominal output voltage. For output voltages with a resolution of 100 mV, two digits are used in the ordering number; otherwise, three digits are used (for example, 33 = 3.3 V; 125 = 1.25 V). 01 indicates adjustable output version.
- yyy is package designator.
- z is package quantity. R is for large quantity reel, T is for small quantity reel.

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or visit the device product folder at www.ti.com.

12.2 Documentation Support

12.2.1 Related Documentation

For related documentation see the following:

- Texas Instruments, TLV767EVM-014 Evaluation module user's guide
- Texas Instruments, Pros and cons of using a feedforward capacitor with a low-dropout regulator application report
- Texas Instruments, Know your limits application report
- Texas Instruments, Universal low-dropout (LDO) linear voltage regulator MultiPkgLDOEVM-823 evaluation module user's guide

12.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on Subscribe to updates to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

12.4 Support Resources

TI E2E™ support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

12.5 Trademarks

TI E2E™ is a trademark of Texas Instruments.

All trademarks are the property of their respective owners.

12.6 Electrostatic Discharge Caution

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

12.7 Glossary

TI Glossary This glossary lists and explains terms, acronyms, and definitions.

Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.
## PACKAGING INFORMATION

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<th>Package Drawing</th>
<th>Pins</th>
<th>Package Qty</th>
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The marketing status values are defined as follows:

**ACTIVE**: Product device recommended for new designs.

**LIFEBUY**: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND**: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW**: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE**: TI has discontinued the production of the device.

**RoHS**: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt**: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green**: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

**MSL, Peak Temp.**: The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

**Important Information and Disclaimer**: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

**OTHER QUALIFIED VERSIONS OF TLV767**:

- Automotive : TLV767-Q1

**NOTE**: Qualified Version Definitions:

- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects
**TAPE AND REEL INFORMATION**

**REEL DIMENSIONS**

- Reel Diameter
- Reel Width (W1)

**TAPE DIMENSIONS**

- A0: Dimension designed to accommodate the component width
- B0: Dimension designed to accommodate the component length
- K0: Dimension designed to accommodate the component thickness
- W: Overall width of the carrier tape
- P1: Pitch between successive cavity centers

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**

- Cavity
- Pocket Quadrants
- Sprocket Holes
- User Direction of Feed

*All dimensions are nominal*

<table>
<thead>
<tr>
<th>Device</th>
<th>Package Type</th>
<th>Package Drawing</th>
<th>Pins</th>
<th>SPQ</th>
<th>Reel Diameter (mm)</th>
<th>Reel Width W1 (mm)</th>
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## TAPE AND REEL BOX DIMENSIONS

*All dimensions are nominal*

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This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.
NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-187.
NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
8. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.
9. Size of metal pad may vary due to creepage requirement.
NOTES: (continued)

10. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

11. Board assembly site may have different recommendations for stencil design.
NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.
5. Support pin may differ or may not be present.
NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

9. Board assembly site may have different recommendations for stencil design.
Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.
NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.
NOTES: (continued)

4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).

5. Vias are optional depending on application, refer to device data sheet. If some or all are implemented, recommended via locations are shown.
NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
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