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**TPS22948** 

#### SLVSEZ7A-MARCH 2019-REVISED OCTOBER 2019

# TPS22948 5.5-V, 240-mA Current Limited Load Switch with Reverse Current Blocking

Technical

Documents

### 1 Features

- Input operating voltage range (V<sub>IN</sub>): 2.5 V to 5.5 V
- Output current limit (I<sub>LIMIT</sub>): 240 mA (typical)
- Thermal shutdown (TSD)
- ON-Resistance (R<sub>ON</sub>): 300 mΩ (typical)
- Slow Turn ON timing limits inrush current (typical):
  - Turn ON time (t<sub>ON</sub>): 820 us at 6.6 mV/µs
- Always-ON Reverse Current Blocking (RCB):
  - ON State activation current (I<sub>RCB</sub>): –200 mA (typical)
- Fault indication (FLT)
- Smart ON pin pull down (R<sub>PD.ON</sub>):
  - ON V<sub>IH</sub> (I<sub>ON</sub>): 25 nA (maximum)
  - ON  $V_{IL}$  ( $R_{PD,ON}$ ): 500 k $\Omega$  (typical)
- Low power consumption:
  - ON State (I<sub>Q</sub>): 50 uA (typical)
  - OFF State (I<sub>SD</sub>): 0.3 uA (typical)

### 2 Applications

- Personal electronics
- Set top box
- HDMI output ports
- Notebook, desktop PC
- Docking stations

## 3 Description

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The TPS22948 device is a small, single channel load switch with robust protection against fault cases with output current limiting, reverse current blocking, and thermal shutdown.

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2.0

The switch ON state is controlled by a digital input that is capable of interfacing directly with low-voltage control signals. When power is first applied, a smart pull down is used to keep the ON pin from floating until system sequencing is complete. Once the pin is deliberately driven high (>VIH), the smart pull down will be disconnected to prevent unnecessary power loss.

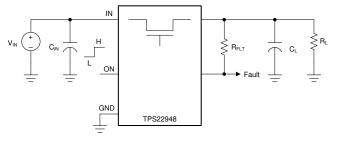
TPS22948 is available in a standard SC-70 package characterized for operation over a temperature range of  $-40^{\circ}$ C to  $125^{\circ}$ C.

#### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)		
TPS22948	SC-70 (6)	2.1 mm x 2.0 mm		

(1) For all available packages, see the orderable addendum at the end of the data sheet.

#### **Simplified Schematic**



Product Folder Links: TPS22948

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## 4 Revision History

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•	Changed from Advance Information to Production Data	······ ·	1
•	FIrst Public Release	····· · ·	1

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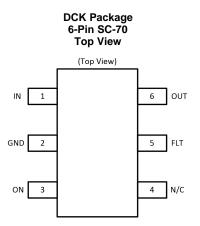
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# 5 Pin Configuration and Functions



#### **Pin Functions**

PIN NO. NAME		1/0	DESCRIPTION		
		1/0	DESCRIPTION		
1	1 IN I Switch input		Switch input		
2	GND	-	Device ground		
3	3 ON I Active high switch control input. Do not leave floating.		Active high switch control input. Do not leave floating.		
4	N/C – No connect pin, leave floating or GND		No connect pin, leave floating or GND		
5	FLT	0	Open-drain output, pulled low during thermal shutdown or reverse current-conditions.		
6 OUT O		0	Switch output		

## 6 Specifications

### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

		MIN	MAX	UNIT
V <sub>IN</sub>	Maximum Input Voltage Range	-0.3	6	V
V <sub>OUT</sub>	Maximum Output Voltage Range	-0.3	6	V
V <sub>ON</sub>	Maximum ON Pin Voltage Range	-0.3	6	V
V <sub>FLT</sub>	Maximum FLT Pin Voltage	-0.3	6	V
I <sub>MAX</sub>	Maximum Output Current	Internally Limite	əd	А
TJ	Junction temperature	Internally Limite	əd	°C
T <sub>STG</sub>	Storage temperature	-65	150	°C
T <sub>LEAD</sub>	Maximum Lead Temperature (10 s soldering time)		300	°C

(1) Stresses beyond those listed under Absolute Maximum Rating may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Condition. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### 6.2 ESD Ratings

			VALUE	UNIT
M	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, allpins <sup>(1)</sup>	±2000	N
V <sub>(ESD)</sub>	Electrostatic discharge	Charged device model (CDM), per JEDEC specificationJESD22-C101, all pins <sup>(2)</sup>	±500	V

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process. Manufacturing with less is possible with the necessary precautions. Pins listed may actually have higher performance.

### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	TYP	MAX	UNIT
V <sub>IN</sub>	Input Voltage Range	2.5		5.5	V
V <sub>OUT</sub>	Output Voltage Range	0		5.5	V
V <sub>IH</sub>	ON Pin High Voltage Range	1		5.5	V
VIL	ON Pin Low Voltage Range	0		0.35	V
I <sub>OUT</sub>	Output Current Range	0		130	mA
C <sub>OUT</sub> <sup>(1)</sup>	Output Capacitance		18		nF
T <sub>A</sub>	Ambient temperature	-40		125	°C

(1) The recommended output capacitance is the capacitance placed next to the output of the device that will provide optimal hard short performance across different load cable lengths.

### 6.4 Thermal Information

		TPS22948	
	THERMAL METRIC <sup>(1)</sup>	DCK (SC-70)	UNIT
		6 PINS	
$R_{\thetaJA}$	Junction-to-ambient thermal resistance	213.5	°C/W
R <sub>0JC(top)</sub>	Junction-to-case (top) thermal resistance	148.8	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	66.9	°C/W
$\Psi_{\text{JT}}$	Junction-to-top characterization parameter	50.0	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	66.7	°C/W

(1) For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.



### 6.5 Electrical Characteristics

Unless otherwise noted, the characteristics in the following table applies at 5 V with a load of  $C_L = 0.1 \ \mu\text{F}$ ,  $R_L = 100 \ \Omega$ . Typical Values are at 25°C.

	PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
Input Sup	ply (VIN)							
I <sub>Q, VIN</sub>	VIN Quiescent Current	$V_{ON} \ge V_{IH}, VOUT = Op$	ben	-40°C to 125°C		50	85	μA
I <sub>SD, VIN</sub>	VIN Shutdown Current	$V_{ON} \le V_{IL}$ , VOUT = GN	1D	-40°C to 125°C		0.3	5	μA
ON-Resist	ance (RON)							
				25°C		300	350	
R <sub>ON</sub>	ON-State Resistance	I <sub>OUT</sub> = -50 mA		-40°C to 85°C			450	mΩ
				-40°C to 125°C			500	
Output Cu	Irrent Limit (ILIM)							
I <sub>LIM</sub>	Output Current Limit			-40°C to 125°C	130	240	350	mA
t <sub>LIM</sub>	Current Limit Response Time	Output hard short (I <sub>OU</sub> -	Output hard short (I <sub>OUT</sub> > I <sub>LIM</sub> )			2		μs
Reverse C	urrent Blocking (RCB)			i.				
	Activation Threshold	V <sub>OUT</sub> Rising; V <sub>OUT</sub> > V	IN	-40°C to 125°C		60		mV
V <sub>RCB</sub>	Release Threshold	V <sub>OUT</sub> Falling; V <sub>OUT</sub> > V	IN	-40°C to 125°C		44		mV
t <sub>RCB</sub>	Response Time	$V_{OUT} = V_{IN} + 1V$		-40°C to 125°C		3		μs
I <sub>Q, RCB</sub>	RCB Quiescent Current (VIN)	$V_{ON} \le V_{IL}$ $V_{OUT} - V_{IN} = 1V$	$V_{ON} \le V_{IL}$ $V_{OUT} - V_{IN} = 1V$	-40°C to 125°C			15	μA
Fault Indi	cation (FLT)							
V <sub>OL, FLT</sub>	Output Low Voltage	I <sub>FLT</sub> = 1 mA		-40°C to 125°C			0.1	V
t <sub>DG,FLT</sub>	Fault Delay Time	V <sub>ON</sub> ≥ V <sub>IH</sub>		-40°C to 125°C		10		μs
I <sub>FLT</sub>	Off State Leakage	$V_{ON} \le V_{IL}$		-40°C to 125°C			25	nA
Enable Pi	n (ON)							
R <sub>PD, ON</sub>	Smart Pull Down Resistance	$V_{ON} \le V_{IL}$		-40°C to 85°C		500		kΩ
I <sub>ON</sub>	ON Pin Leakage	V <sub>ON</sub> ≥ V <sub>IH</sub>		-40°C to 125°C			25	nA
Thermal S	hutdown (TSD)							
TSD	Thermal Shutdown	Rising		N/A	130	150	170	°C
130		Falling (Hysteresis)		N/A	100	120	140	°C

## 6.6 Switching Characteristics

Unless otherwise noted, the typical characteristics in the following table applies at 5 V and 25°C

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>ON</sub>	Turn ON Time	$C_L$ = 18 nF, $R_L$ = 100 $\Omega$		820		μs
t <sub>R</sub>	Output Rise Time	$C_{L} = 18 \text{ nF}, R_{L} = 100 \Omega$		600		μs
SR <sub>ON</sub>	Turn ON Slew Rate	$C_{L} = 18 \text{ nF}, R_{L} = 100 \Omega$		6.6		mV/µs
t <sub>OFF</sub>	Turn OFF Time	$C_L$ = 18 nF, $R_L$ = 100 $\Omega$		15		μs
t <sub>FALL</sub>	Output Fall Time	$C_L$ = 18 nF, $R_L$ = 100 $\Omega$		6.9		μs

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#### 6.7 Typical Characteristics Tek PreVu Tek Sto VIN /nN NOUT **I**00 🔊 холт IOUT 400µs 250kS/s 1000 points **4** 96.0r 17 Sep 2019 06:38:59 1 5.00 V 10.0µs 100MS/s 10k point **4** J 1 44 21 Sep 2019 04:51:13 5.00 \ 5.00 \ 2 5.00 4 100m **2** 5.00 **4** 2.00 $V_{IN} = 5 V$ $T_A = 25 \ ^{\circ}C$ $V_{IN} = 5 V$ $T_A = 25 \ ^{\circ}C$ $V_{OUT} = 0 V$ $C_{OUT} = 18 \text{ nF}$ Figure 2. Hot Short on VOUT with a 1ft Cable Figure 1. Startup Into Short on VOUT Tek Stop Tok S VIN **IN** ΝΙνουτ IOUT-**IN** IOUT-NOUT ■ VOUT ■ VIN 1 5.00 V 3 5.00 V 100MS/s 10k point **4** / 1.44 / 10.0µs 21 Sep 2019 04:51:23 **4** -192m 25.0MS/s 10k points 1.00 \ 3 5.00 \ 21 Sep 2019 04:20:10 40.0µs á $V_{IN} = 5 V$ $T_A = 25 \ ^\circ C$ $C_{OUT} = 18 \text{ nF}$

Figure 3. Hot Short on VOUT with a 3ft Cable

 $V_{IN} = 5 V$ T<sub>A</sub> = 25 °C



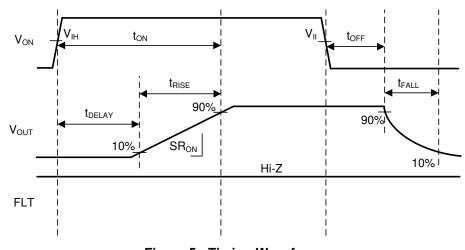
Figure 4. Reverse Current Blocking Behavior

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**ISTRUMENTS** 

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# 7 Parameter Measurement Information 7.1 Timing Waveform Diagram







### 8 Detailed Description

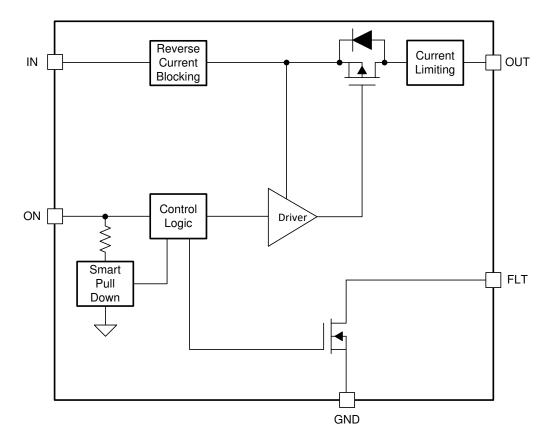
### 8.1 Overview

The TPS22948 device is a 5.5-V, 240-mA current limited load switch in a 6-pin SC-70 package. The 300-m $\Omega$  P-channel FET is used to switch power from input to output with minimal voltage drop across the device.

The TPS22948 device has a slow slew rate which helps reduce or eliminate power supply droop because of large inrush currents. During shutdown, the device has very low leakage currents, thereby reducing unnecessary leakages for downstream modules during standby. Integrated control logic, and driver eliminates the need for any external components which reduces solution size and bill of materials (BOM) count.

The TPS22948 load switch also provides protection features such as reverse current blocking, output current limiting and thermal shutdown.

### 8.2 Functional Block Diagram



#### 8.3 Feature Description

#### 8.3.1 On and Off Control

The ON pin controls the state of the switch. The ON pin is compatible with standard GPIO logic threshold so it can be used in a wide variety of applications. When power is first applied to VIN, a smart pull down is used to keep the ON pin from floating until system sequencing is complete. Once the ON pin is deliberately driven high  $(\geq V_{IH})$ , the smart pull down is disconnected to prevent unnecessary power loss. See Table 1 when the ON pin smart pull down is active.

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VON	Pull Down				
≤ V <sub>IL</sub>	Connected				
≥ V <sub>IH</sub>	Disconnected				

#### 8.3.2 Fault Indication (FLT)

The FLT pin is an open drain output that acts as a status indication for the device. It is pulled low during thermal shutdown or reverse-current events. The behavior of the FLT pin is shown in Figure 6.

#### 8.3.3 Current Limiting (V<sub>sc</sub>)

The TPS22948 responds to overcurrent conditions by limiting its output current to the ILIM level shown in Figure 6.

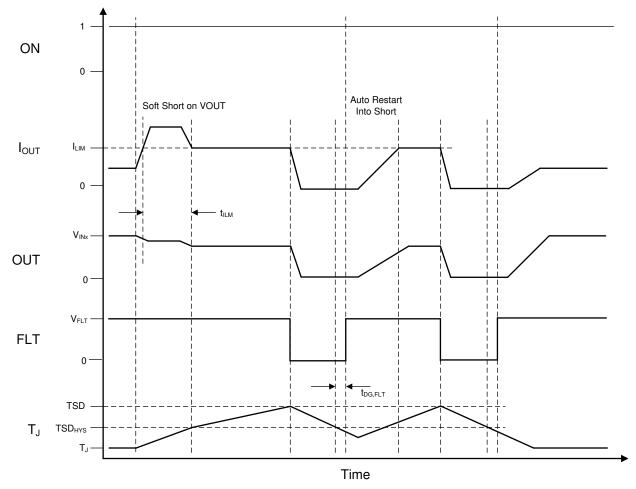


Figure 6. TPS22948 Current Limiting Behavior

When an overcurrent condition is detected, the device maintains a constant output current and reduces the output voltage accordingly. Two possible overload conditions can occur.

The first condition is when a short circuit or partial short circuit is present on the output and the ON pin is toggled high, turning the device on. The output voltage is held near zero potential with respect to ground and the TPS22948 ramps the output current to  $I_{LIM}$ . The TPS22948 device will limit the current to  $I_{LIM}$  until the overload condition is removed or the internal junction temperature of the device reaches thermal shutdown and the device turns itself off. The device remains off until the junction temperature has lowered by TSD<sub>HYS</sub>, and the device will turn itself back on. This will cycle until the overload condition is removed.



The second condition is when a short circuit, partial short circuit, or transient overload occurs after the device has been fully powered on. The device responds to the overcurrent condition within time  $t_{LIM}$  (see Figure 7), and before this time, the current is able to exceed  $I_{LIM}$ . In the case of a fast transient, the current-sense amplifier is overdriven and momentarily disables the internal power FET. The current-sense amplifier recovers and limits the output current to  $I_{LIM}$ . Similar to the previous case, the TPS22948 limits the current to  $I_{LIM}$  until the overload condition is removed or the internal junction temperature of the device reaches thermal shutdown and begins thermally cycling on and off.

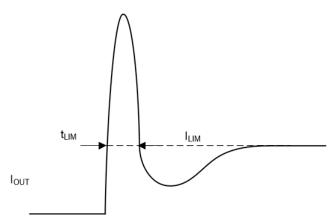


Figure 7. Transient Current Limit Waveform

#### 8.3.4 Reverse Current Blocking (RCB)

In a scenario where the device is enabled and VOUT is greater than VIN, there is potential for reverse current to flow through the pass FET or the body diode. When the reverse current threshold is exceeded (about 200 mA), there is a delay time ( $t_{RCB}$ ) before the switch turns off to stop the current flow. The switch will remain off and block reverse current as long as the reverse voltage condition exists. Once VOUT has dropped below the release voltage threshold ( $V_{RCB}$ ) the device will turn back on. When the ON pin is pulled low, the device will constantly block reverse current.

### 8.4 Device Functional Modes

Table 2 describes the connection of the VOUT pin depending on the state of the ON pin.

### Table 2. VOUT Connection

ON	TPS22919 VOUT				
L	Open				
Н	VIN				



### 9 Application and Implementation

#### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 9.1 Application Information

This section highlights some of the design considerations when implementing this device in various applications.

### 9.2 Typical Application

This typical application demonstrates how the TPS22948 device can be used to power downstream modules.

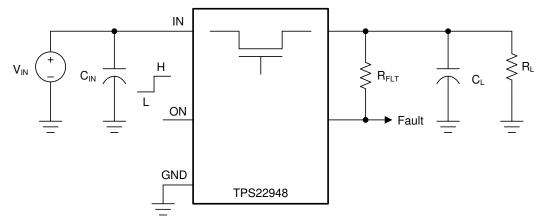


Figure 8. Typical Application Schematic

#### 9.2.1 Design Requirements

For this design example, use the values listed in Table 3 as the design parameters:

**Table 3. Design Parameters** 

DESIGN PARAMETER	EXAMPLE VALUE
Input Voltage (V <sub>IN</sub> )	5 V
Load Current / Resistance (RL)	1 kΩ
Load Capacitance (C <sub>L</sub> )	10 µF
Maximum Inrush Current (I <sub>INRUSH</sub> )	100 mA

Although the load capacitance is 10  $\mu$ F, this is assumed to be at the end of a cable or closer to the load. An 18nF capacitance close to the output of the device is recommended for optimal performance during short circuit conditions.

### 9.2.2 Detailed Design Procedure

#### 9.2.2.1 Limiting Inrush Current

Use Equation 1 to find the maximum output capacitance for a given inrush current requirement.

 $C_{L} = I_{INRUSH} \times t_{R} \div (0.8 \times V_{IN})$ 

where

- C<sub>L</sub> = capacitance on VOUT (μF)
- I<sub>INRUSH</sub> = maximum acceptable inrush current (A)
- $t_R$  = rise time of the TPS22948 (µs)



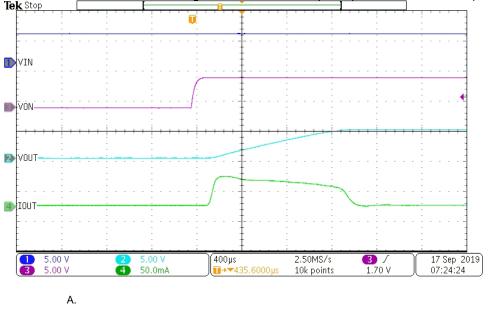
(1)

#### • V<sub>IN</sub> = input voltage (V)

Based on Equation 1, the maximum output capacitance that limits the inrush current to 100 mA is 12.5  $\mu$ F. Therefore, the desired 10- $\mu$ F load capacitance will not exceed the inrush current design requirement during turn on.

#### 9.2.3 Application Curves

The below scope shot shows the inrush current generated from a 10-µF capacitance on the output.







### **10 Power Supply Recommendations**

The device is designed to operate with a VIN range of 2.5 V to 5.5 V. The VIN power supply must be well regulated and placed as close to the device terminal as possible. The power supply must be able to withstand all transient load current steps. In most situations, using an input capacitance (CIN) of 1  $\mu$ F is sufficient to prevent the supply voltage from dipping when the switch is turned on. In cases where the power supply is slow to respond to a large transient current or large load current step, additional bulk capacitance may be required on the input. A 18nF capacitance close to the output of the device is recommended for optimal performance during short circuit conditions.



### 11 Layout

### 11.1 Layout Guidelines

For best performance, all traces must be as short as possible. To be most effective, the input and output capacitors must be placed close to the device to minimize the effects that parasitic trace inductances may have on normal operation. Using wide traces for VIN, VOUT, and GND helps minimize the parasitic electrical effects.

### 11.2 Layout Example

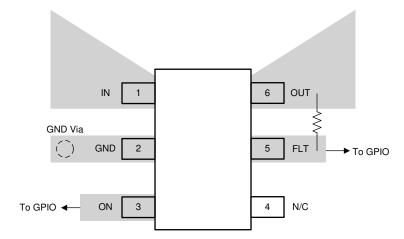


Figure 10. Recommended Board Layout

## **12 Device and Documentation Support**

### 12.1 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

### 12.2 Support Resources

TI E2E<sup>™</sup> support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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#### 12.3 Trademarks

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#### 12.4 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 12.5 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

### 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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**STRUMENTS** 

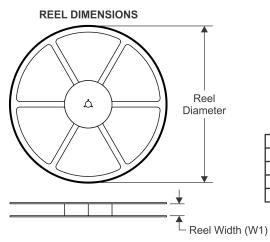
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# PACKAGE MATERIALS INFORMATION

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## TAPE AND REEL INFORMATION





### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



1	*All dimensions are nominal												
	Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
	TPS22948DCKR	SC70	DCK	6	3000	178.0	9.0	2.4	2.5	1.2	4.0	8.0	Q3

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# PACKAGE MATERIALS INFORMATION

11-Oct-2019



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS22948DCKR	SC70	DCK	6	3000	180.0	180.0	18.0

DCK (R-PDSO-G6)

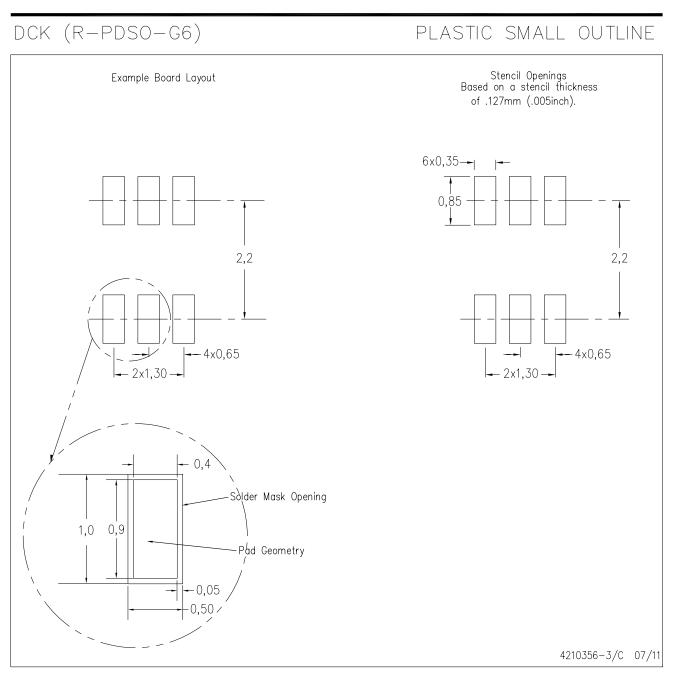
PLASTIC SMALL-OUTLINE PACKAGE



- NOTES: A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.
  - D. Falls within JEDEC MO-203 variation AB.



# LAND PATTERN DATA



NOTES:

- A. All linear dimensions are in millimeters.B. This drawing is subject to change without notice.
- C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
- D. Publication IPC-7351 is recommended for alternate designs.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.



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