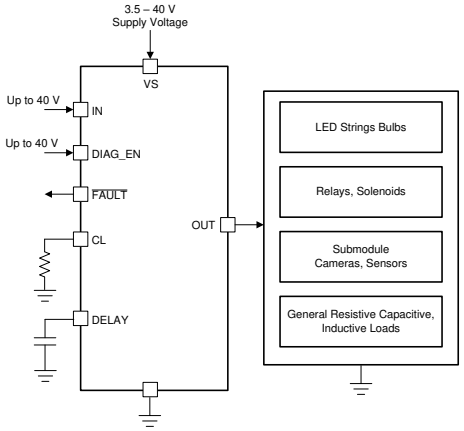
TPS1H200A-Q1 40V 200mΩ Single-Channel Smart High-Side Switch

1 Features

- Qualified for automotive applications
- AEC-Q100 qualified with the following results:
  - Device temperature grade 1: –40°C to +125°C ambient operating temperature range
  - Device HBM ESD Classification Level H2
  - Device CDM ESD Classification Level C4B
- Functional Safety-Capable
  - Documentation available to aid functional safety system design
- Single-channel 200mΩ smart high-side switch
- Wide operating voltage: 3.4V to 40V
- Ultra-low standby current, < 500nA
- Adjustable current limit with external resistor
  - ±15% when ≥ 500mA
  - ±10% when ≥ 1.5A
- Configurable behavior after current limit
  - Holding mode
  - Latch-off mode with adjustable delay time
  - Auto-retry mode
- Supports stand-alone operation without an MCU
- Protection:
  - Short-to-GND and overload protection
  - Thermal shutdown and thermal swing
  - Negative voltage clamp for inductive loads
  - Loss of GND and loss of battery protection
- Diagnostics:
  - Overload and short-to-GND detection
  - Open-load and short-to-battery detection in ON or OFF state
  - Thermal shutdown and thermal swing



Typical Block Diagram

2 Applications

- Body lighting
- Infotainment system
- Advanced Driver Assistance Systems (ADAS)
- Single-channel high-side switch for submodules
- General resistive, inductive, and capacitive loads

3 Description

The TPS1H200A-Q1 device is a fully protected single-channel high-side power switch with an integrated 200mΩ NMOS power FET.

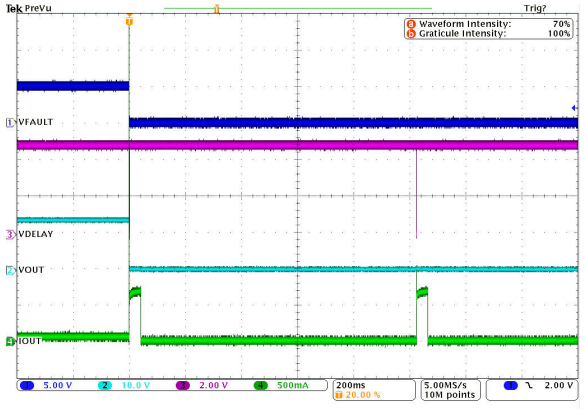
An adjustable current limit improves system reliability by limiting the inrush or overload current. The high accuracy of the current limit improves overload protection, simplifying the front-stage power design. Configurable features besides current limit provide design flexibility in functionality, cost, and thermal dissipation.

The device supports full diagnostics with the digital status output. Open-load detection is available in ON and OFF states. The device supports operation with or without an MCU. Stand-alone mode allows isolated systems to use the device.

Package Information

PART NUMBER	PACKAGE(1)	PACKAGE SIZE(2)
TPS1H200A-Q1	HVSSOP (8)	3.00mm × 3.00mm

(1) For all available packages, see the orderable addendum at the end of the datasheet.
(2) The package size (length × width) is a nominal value and includes pins, where applicable.



Current Limit Protection in Auto-Retry Mode

An IMPORTANT NOTICE at the end of this data sheet addresses availability, warranty, changes, use in safety-critical applications, intellectual property matters and other important disclaimers. PRODUCTION DATA.

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4 Pin Configuration and Functions

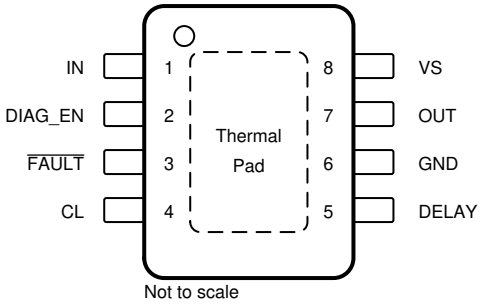


Figure 4-1. DGN PowerPAD™ Package 8-Pin HVSSOP With Exposed Thermal Pad Top View

Table 4-1. Pin Functions

PIN		TYPE(1)	DESCRIPTION
NAME	NO.		
CL	4	O	Adjustable current limit. Connect to device GND if external current limit is not used.
DELAY	5	I/O	Function configuration when current limit; internal pullup
DIAG_EN	2	I	Enable the diagnostic function
$\overline{\text{FAULT}}$	3	O	Open-drain diagnostic status output. Leave floating if not used.
GND	6	—	Ground
IN	1	I	Input control for output activation; internal pulldown
OUT	7	O	Output, source of the high-side switch, connected to the load
Thermal pad	—	—	Thermal pad. Connect to device GND or leave floating.
VS	8	I	Power supply, drain for the high-side switch

(1) I = input, O = output, I/O = bidirectional

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5 Specifications

5.1 Absolute Maximum Ratings

over operating ambient temperature range (unless otherwise noted)(1) (2)

		MIN	MAX	UNIT
Supply voltage VS pin	t < 400ms	—	42	V
Reverse polarity voltage (3)	t < 1 minute	–36	—	V
Current on GND	t < 2 minutes	–100	250	mA
Voltage on IN and DIAG_EN pins		–0.3	VS	V
Current on IN and DIAG_EN pins		–10	—	mA
Voltage on DELAY pin		–0.3	7	V
Current on DELAY pin		–60	—	mA
Voltage on $\overline{\text{FAULT}}$ pin		–0.3	7	V
Current on $\overline{\text{FAULT}}$ pin		–30	10	mA
Voltage on CL pin		–0.3	7	V
Current on CL pin		—	6	mA
Voltage on OUT pin		—	42	V
Inductive load switch-off energy dissipation single pulse(4)		—	40	mJ
Operating junction temperature, T_J		–40	150	°C
Storage temperature, T_{stg}		–65	150	°C

(1) Operation outside the *Absolute Maximum Ratings* may cause permanent device damage. *Absolute Maximum Ratings* do not imply functional operation of the device at these or any other conditions beyond those listed under *Recommended Operating Conditions*. If used outside the *Recommended Operating Conditions* but within the *Absolute Maximum Ratings*, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.
(2) All voltage values are with respect to ground.
(3) Reverse polarity condition: V_{IN} = 0V, reverse current < $I_{R(2)}$, GND pin 1kΩ resistor in parallel with diode.
(4) Test condition: V_{VS} = 13.5V, L = 8mH, T_J = 150°C. FR4 2s2p board, 2 × 70µm Cu, 2 × 35µm Cu. 600mm^2 thermal pad copper area.

5.2 ESD Ratings

				VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per AEC Q100-002(1)	All pins except VS, OUT, and GND	±2000	V
			Pins VS, OUT, and GND	±3000	
		Charged-device model (CDM), per AEC Q100-011		±750	

(1) AEC-Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specifications.

5.3 Recommended Operating Conditions

over operating ambient temperature range (unless otherwise noted)

		MIN	MAX	UNIT
V_S	Operating voltage	4	40	V
	Voltage on IN and DIAG_EN pins	0	40	V
	Voltage on $\overline{\text{FAULT}}$ pin	0	5	V
$I_{o,nom}$	Nominal DC load current	0	2.5	A
T_J	Operating junction temperature	–40	150	°C

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5.4 Thermal Information

THERMAL METRIC[1]		TPS1H200A-Q1 DGN (HVSSOP) 8 PINS	UNIT
$R_{\theta JA}$	Junction-to-ambient thermal resistance	47.4	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	67.8	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	18.7	°C/W
ψ_{JT}	Junction-to-top characterization parameter	3.1	°C/W
ψ_{JB}	Junction-to-board characterization parameter	18.7	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	5.6	°C/W

(1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application note.

5.5 Electrical Characteristics

over operating ambient temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
OPERATING VOLTAGE						
$V_{VS(nom)}$	Nominal operating voltage		4		40	V
$V_{VS(uvr)}$	Undervoltage restart	V_{VS} rising	3.5	3.7	4	V
$V_{VS(uvf)}$	Undervoltage shutdown	V_{VS} falling	3	3.2	3.4	V
$V_{(uv,hys)}$	Undervoltage shutdown, hysteresis			0.5		V
OPERATING CURRENT						
$I_{(op)}$	Nominal operating current	V_{VS} = 13.5V, V_{IN} = 5V V_{DIAG_EN} = X V, I_{OUT} = 0.5A I_{CL} = 2A			5	mA
$I_{(off)}$	Standby current	V_{VS} = 13.5V V_{IN} = V_{DIAG_EN} = V_{CL} = V_{OUT} = 0V T_J = 25°C			0.5	µA
		V_{VS} = 13.5V V_{IN} = V_{DIAG_EN} = V_{CL} = V_{OUT} = 0V T_J = 125°C			3	
$I_{(off,diag)}$	Standby current with diagnostics enabled	V_{VS} = 13.5V V_{IN} = 0V, V_{DIAG_EN} = 5V			3	mA
$t_{(off,deg)}$	Standby-mode deglitch time[1]	IN from high to low if deglitch time ≥ $t_{(off,deg)}$, then the device enters into standby mode.		12.5		ms
$I_{lkg(out)}$	Output leakage current in OFF state	V_{VS} = 13.5V V_{IN} = V_{DIAG_EN} = V_{OUT} = 0V			3	µA
POWER STAGE						
$r_{DS(on)}$	ON state resistance	V_{VS} ≥ 3.5V, T_J = 25°C		200		mΩ
		V_{VS} ≥ 3.5V, T_J = 150°C			400	
$I_{CL(int)}$	Internal current limit	CL pin connected to GND	3.5	4.8	6	A
$I_{CL(TSD)}$	Current-limit value percentage during thermal shutdown			60%		
$V_{DS(clamp)}$	Drain−to−source voltage internally clamped		45		65	V
OUTPUT DIODE CHARACTERISTICS						
V_F	Drain−to-source diode voltage	IN = 0, I_{OUT} = −0.15A	0.3	0.7	1	V

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5.5 Electrical Characteristics (continued)

over operating ambient temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$I_{R(1)}$	Continuous reverse current from source to drain during a short-to-battery condition(1)	t < 60s, V_{IN}= 0V, T_J = 25°C.			2	A
$I_{R(2)}$	Continuous reverse current from source to drain during a reverse-polarity condition(1)	t < 60s, V_{IN}= 0V, T_J = 25°C GND pin 1kΩ resistor in parallel with diode.			2	A
LOGIC INPUT (IN, DIAG_EN)						
V_{IH}	Logic high-level voltage		1.8			V
V_{IL}	Logic low-level voltage				0.8	V
$R_{pd,in}$	Logic-pin pulldown resistor	IN. V_{IN} = 5V	150		400	kΩ
		DIAG_EN. V_{VS} = V_{DIAG_EN} = 5V	350		850	
DIAGNOSTICS						
$I_{lkg(loss,GND)}$	Loss of ground output leakage current				100	µA
$t_{d(ol,on)}$	Open-load deglitch time in ON state	V_{IN} = 5V, V_{DIAG_EN} = 5V when I_{OUT} < $I_{(ol,on)}$, duration longer than $t_{d(ol,on)}$, open load is detected.	200	300	450	µs
$I_{(ol,on)}$	Open-load detection threshold in ON state	V_{IN} = 5V, V_{DIAG_EN} = 5V when I_{OUT} < $I_{(ol,on)}$ duration longer than $t_{d(ol,on)}$ open load is detected.		10	20	mA
$V_{(ol,off)}$	Open-load detection threshold in OFF state	V_{IN} = 0V, V_{DIAG_EN} = 5V when V_{VS} – V_{OUT} < $V_{(ol,off)}$ duration longer than $t_{d(ol,off)}$ open load is detected.	1.4		2.6	V
$t_{d(ol,off)}$	Open-load deglitch time in OFF state	V_{IN} = 0V, V_{DIAG_EN} = 5V when V_{VS} – V_{OUT} < $V_{(ol,off)}$ duration longer than $t_{d(ol,off)}$ open load is detected.	200	300	450	µs
$I_{(ol,off)}$	OFF state output sink current	V_{IN} = 0V, V_{DIAG_EN} = 5V V_{VS} = V_{OUT} = 13.5V	–75			µA
$V_{\overline{FAULT}}$	$\overline{FAULT}$ low output voltage	$I_{\overline{FAULT}}$ = 2mA			0.2	V
$t_{\overline{FAULT}}$	$\overline{FAULT}$ signal holding time(1)			8.5		ms
$T_{(SD)}$	Thermal shutdown threshold(1)			175		°C
$T_{(SD,rst)}$	Thermal shutdown status reset(1)			155		°C
$T_{(sw)}$	Thermal swing shutdown threshold(1)			60		°C
$T_{(hys)}$	Hysterisis for resetting the thermal shutdown and swing(1)			10		°C
CURRENT LIMIT AND DELAY CONFIGURATION						
$K_{(CL)}$	Current-limit current ratio(1)			2500		
$V_{CL(th)}$	Current-limit internal threshold voltage(1)			0.8		V
$dK_{(CL)}$ / $K_{(CL)}$	External current limit accuracy (I_{OUT} – I_{CL} × $K_{(CL)}$ × 100 / (I_{CL} × $K_{(CL)}$)	I_{limit} ≥ 0.25A, V_{VS} – V_{OUT} ≥ 2.5V	–20%		20%	
		I_{limit} ≥ 0.5A, V_{VS} – V_{OUT} ≥ 2.5V	–15%		15%	
		I_{limit} ≥ 1.5A, I_{limit} < 5A V_{VS} – V_{OUT} ≥ 2.5V	–10%		10%	
$I_{dl(chg)}$	Delay pin charging current in latch-off mode(1)			4.5		µA
$V_{dl(th)}$	Pulling up threshold in auto-retry mode		2.7			V
$V_{dl(ref)}$	Internal reference voltage in latch-off mode			1.45		V
t_{dl1}	Internal fixed delay time(1)		300	400	500	µs

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5.5 Electrical Characteristics (continued)

over operating ambient temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{dl2}	Adjustable delay time by external capacitor on DELAY pin(1)	Connect with 3.3 µF capacitor as the maximum value.			1000	ms
$t_{CL(deg)}$	Deglitch time when current limit (1)	IN low to high or IN keeps high but thermal shutdown recovery, V_{DIAG_EN} = 5V the deglitch time from IN rising edge to FAULT reporting out.	300		550	µs
		IN keeps high, V_{DIAG_EN} = 5V the deglitch time from CL start-point to FAULT reporting out.	80		200	
$t_{hic(on)}$	On-time when in auto-retry mode(1)		35	40	45	ms
$t_{hic(off)}$	Off-time when in auto-retry mode(1)		0.8	1	1.2	s

(1) Value specified by design, not subject to production test.

5.6 Switching Characteristics

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{d(on)}$	Turnon delay time IN rising edge to 10% of V_{OUT} (1)	V_{VS} = 13.5V, V_{DIAG_EN} = 5V, I_{OUT} = 0.1A	20	50	90	µs
$t_{d(off)}$	Turnoff delay time IN falling edge to 90% of V_{OUT} (1)	V_{VS} = 13.5V, V_{DIAG_EN} = 5V, I_{OUT} = 0.1A	20	50	90	µs
$dV/dt_{(on)}$	Slew rate on V_{OUT} from 10% to 90% (1)	V_{VS} = 13.5V, V_{DIAG_EN} = 5V, I_{OUT} = 0.1A	0.1	0.3	0.6	V/µs
$dV/dt_{(off)}$	Slew rate off V_{OUT} from 90% to 10% (1)	V_{VS} = 13.5V, V_{DIAG_EN} = 5V, I_{OUT} = 0.1A	0.1	0.35	0.6	V/µs

(1) Value specified by design, not subject to production test.

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5.7 Typical Characteristics

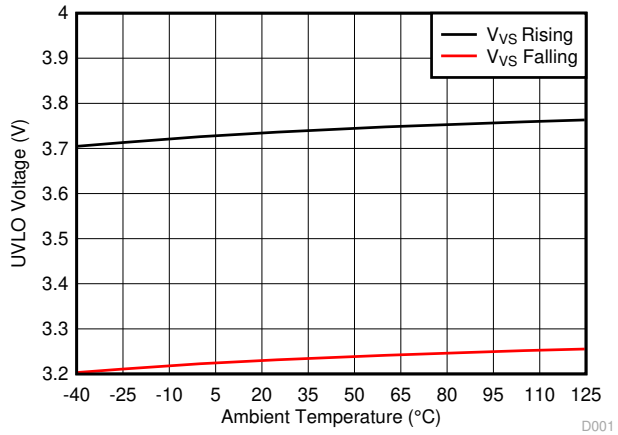


Figure 5-1. UVLO Voltage Threshold

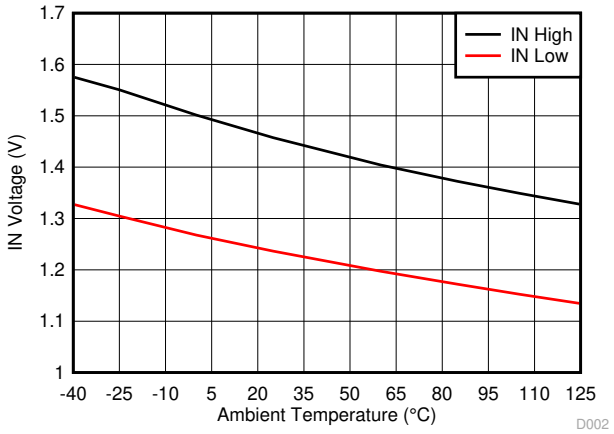


Figure 5-2. IN Voltage Threshold

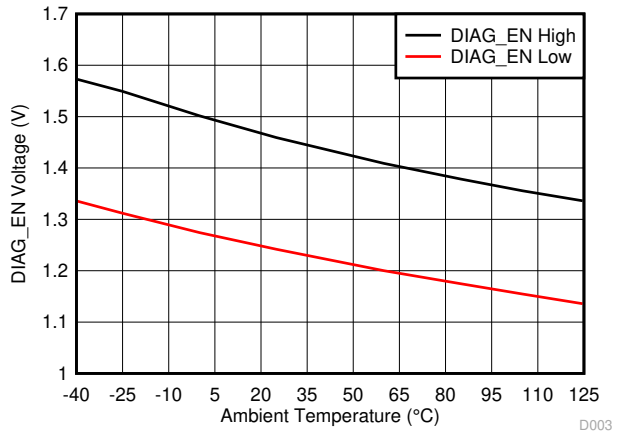


Figure 5-3. DIAG_EN Voltage Threshold

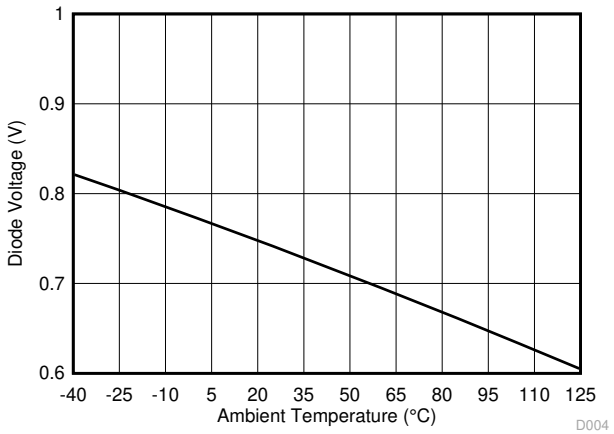


Figure 5-4. Body-Diode Forward Voltage

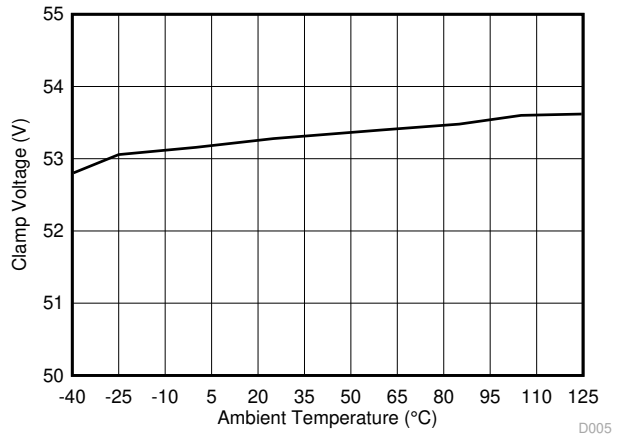


Figure 5-5. Drain-to-Source Clamp Voltage

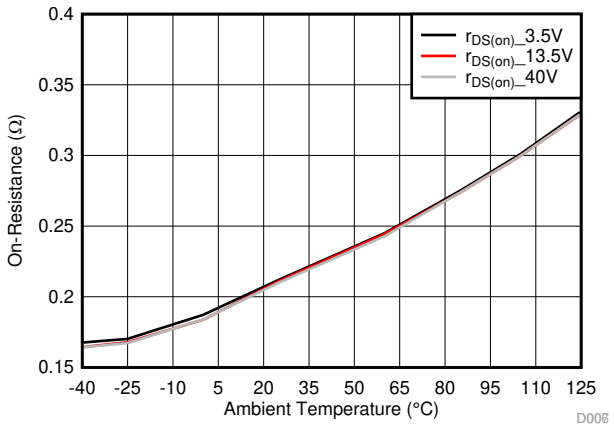


Figure 5-6. FET On-Resistance

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5.7 Typical Characteristics (continued)

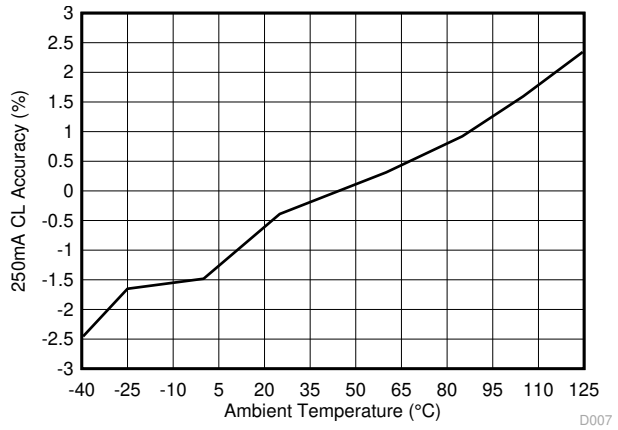


Figure 5-7. Current-Limit Accuracy at 250mA

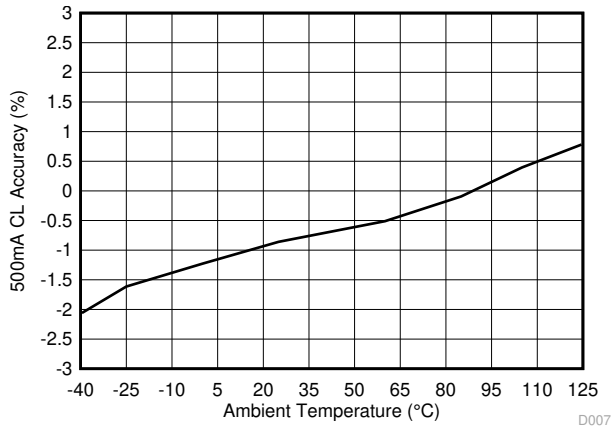


Figure 5-8. Current-Limit Accuracy at 500mA

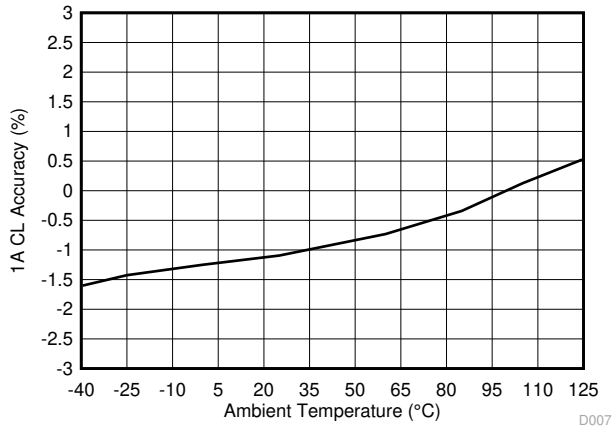


Figure 5-9. Current-Limit Accuracy at 1A

6 Detailed Description

6.1 Overview

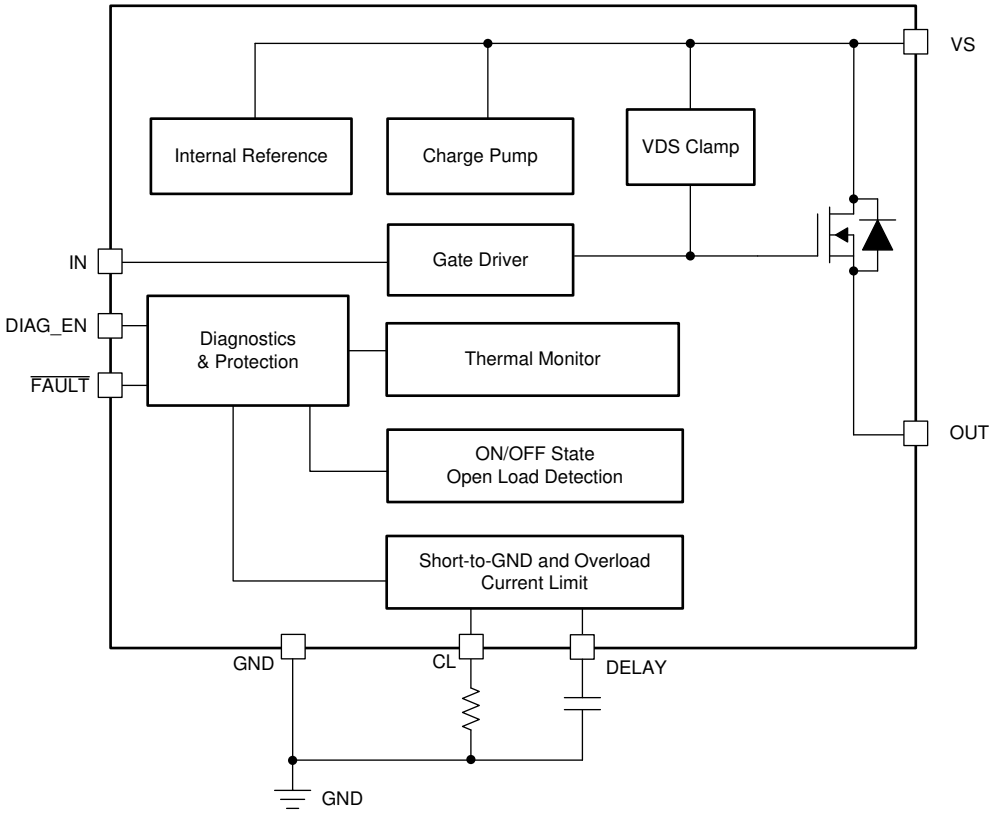
The TPS1H200A-Q1 device is a smart high-side switch with an internal charge pump and single-channel integrated NMOS power FET. The adjustable current limit function improves the reliability of the whole system. Full diagnostic features enable intelligent control of the load.

The external high-accuracy current limit sets the current limit value for the application. When overcurrent occurs, the device improves system reliability by clamping the inrush current effectively. The device saves system cost by reducing the size of PCB traces and connectors, and the capacity of the preceding power stage. The TPS1H200A-Q1 device allows three modes when a current limit occurs. Users can set the output to consistently hold the current, to immediately latch off, or to automatically retry through the configuration on the DELAY pin. The configurable behaviors during a current limit provide design flexibility. This includes functionality, cost, and thermal dissipation.

This device supports full diagnostics with the digital status output. High-accuracy and low-threshold open-load detection enables real-time ON state monitoring. The device supports operation without an MCU (stand-alone mode) which allows the system to locally implement full functionality.

The TPS1H200A-Q1 device is a smart high-side switch for a wide variety of resistive, inductive, and capacitive loads, including LEDs, bulbs, relays, solenoids, and sub-modules.

6.2 Functional Block Diagram



6.3 Feature Description

6.3.1 Current limit

A high-accuracy current limit allows high reliability of the design. The current limit protects the load and the power supply from over-stressing during short-circuit-to-GND or power-up conditions. The current limit can also save system cost by reducing the size of PCB traces and connectors, and the capacity of the preceding power stage.

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When a current limit threshold is hit, a closed loop immediately activates. The output current is clamped at the set value, and a fault is reported. The device heats up because of high power dissipation on the power FET.

The device has two current limit thresholds.

- Internal current limit: The internal current limit is fixed at $I_{CL(int)}$. Tie the CL pin directly to the device GND for large-transient-current applications.
- External adjustable current limit: An external resistor is used to set the current limit threshold. Use Equation 1 to calculate $R_{(CL)}$. The external adjustable current limit allows the flexibility to set the current limit value by application.

$$R_{CL} = \frac{V_{CL(th)} \times K_{(CL)}}{I_{OUT}} \quad (1)$$

where

- $V_{CL(th)}$ is the internal band-gap voltage.
- $K_{(CL)}$ is the ratio of the output current and the current limit set value.
- $K_{(CL)}$ is constant across temperature and supply voltage.

Note

When a GND network is used, that causes a level shift between the device GND and board GND, so the CL pin must be connected to the device GND.

For better protection from a hard short-to-GND condition (when the IN pin is enabled, a short-to-GND occurs suddenly), the device implements a fast-trip protection to turn off the output before the current limit closed loop is set up. Typically, the fast-trip response time is less than 1µs. With a fast response like this, the device can achieve a better inrush current-suppression performance.

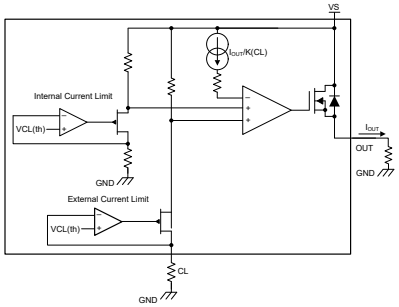


Figure 6-1. Current Limit

6.3.2 DELAY Pin Configuration

When a current limit occurs, the TPS1H200A-Q1 device supports three different outcomes of the output. Table 6-1 lists the current limit configurations and these outcomes behaviors.

Table 6-1. Current Limit Configurations

MODE	DELAY CONFIGURATION	OUTPUT CURRENT BEHAVIOR	$\overline{\text{FAULT}}$ RECOVERY
Holding	Connects to GND directly	When hitting a current limit, the output current holds at the setting current. The device enters into thermal shutdown mode when $T_J > T_{(SD)}$.	$\overline{\text{FAULT}}$ clears when IN turns low for a duration of time longer than $t_{\overline{\text{FAULT}}}$ or when the current limit is removed when IN is high.

Table 6-1. Current Limit Configurations (continued)

MODE	DELAY CONFIGURATION	OUTPUT CURRENT BEHAVIOR	$\overline{\text{FAULT}}$ RECOVERY
Latch-off	Connects to GND through a capacitor	When hitting a current limit, the output current holds at the setting current, but latches off after a preset DELAY time (t_{dl1}+ t_{dl2}). t_{dl1} is the default delay time, and t_{dl2} is a capacitor-configurable delay time. The output stays latched off regardless of whether the current limit is removed. The output recovers only when IN is toggling.	$\overline{\text{FAULT}}$ clears when IN turns low for a duration of time longer than t $_{\overline{\text{FAULT}}}$.
Auto-retry	External pullup	When hitting a current limit, the output current holds at the setting current, but periodically comes on for $t_{hic(on)}$ and turns off for $t_{hic(off)}$.	$\overline{\text{FAULT}}$ clears when IN turns low for a duration of time longer than t $_{\overline{\text{FAULT}}}$ OR when the current limit is removed for $t_{hic(on)}$.

6.3.2.1 Holding Mode

Holding mode is active when the DELAY pin connects directly to GND. When a current limit is reached, the output current holds at the setting current. The device then enters thermal shutdown mode when $T_J > T_{(SD)}$.

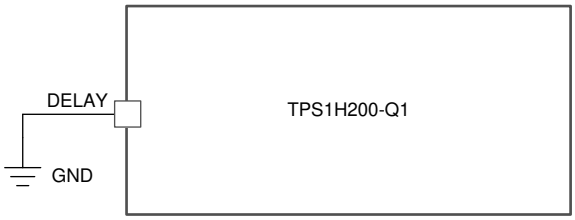


Figure 6-2. Holding Mode Connection

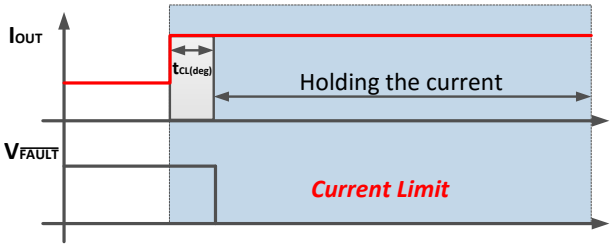


Figure 6-3. Holding Mode Example

6.3.2.2 Latch-Off Mode

Latch-off mode is active when the DELAY pin connects to GND through a capacitor. When a current limit is reached, the output current holds at the setting current, but latches off after a preset DELAY time (t_{dl1}+ t_{dl2}). t_{dl1} is the default delay time, and t_{dl2} is a configurable delay time set by a capacitor. Regardless of whether the current limit is removed or not, the output remains latched off. The output only recovers when IN is toggling.

Use Equation 2to calculate t_{dl2}.

$$C_{DELAY} = \frac{I_{dl(chg)} \times t_{dl2}}{V_{dl(ref)}} \quad (2)$$

where

- C_{DELAY} is the capacitor connected on the DELAY pin
- The $I_{dl(chg)}$ is the device that charges the current in latch-off mode
- t_{dl2} is the user-setting delay time
- $V_{dl(ref)}$ is the internal reference voltage in latch-off mode

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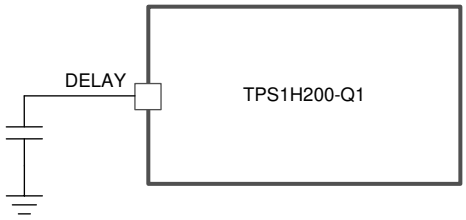


Figure 6-4. Latch-Off-Mode Connection

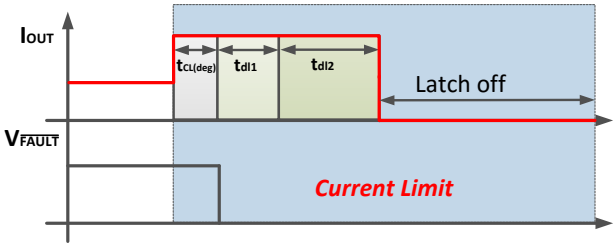


Figure 6-5. Latch-Off-Mode Example

6.3.2.3 Auto-Retry Mode

Auto-retry mode is active when the DELAY pin is externally pulled up. The pullup voltage must be higher than $V_{dl(th)}$. When the current limit is reached, the output current holds at the setting current, but periodically turns on for $t_{hic(on)}$ and turns off for $t_{hic(off)}$. The device checks the current limit status at the falling edge of $t_{hic(on)}$ clock. If current limit status is captured, the device shuts down for $t_{hic(off)}$. If the current limit status is not captured because of the off window during the thermal conditions, the device keeps turning on for additional $t_{hic(on)}$ or more until the current limit status is captured.

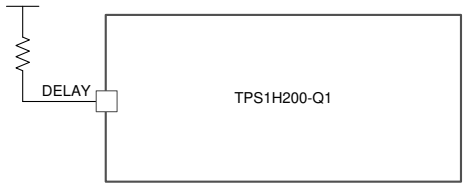


Figure 6-6. Auto-Retry-Mode Connection

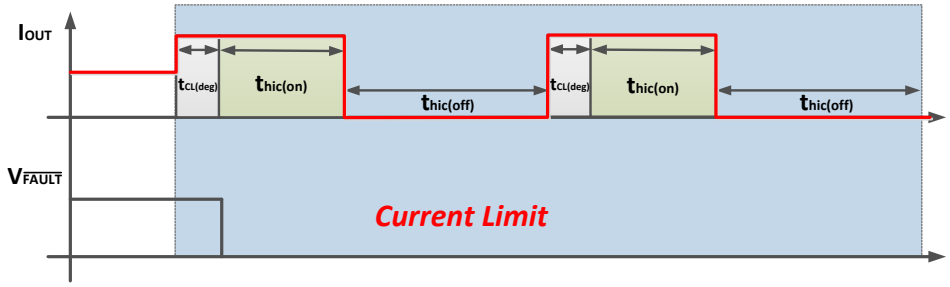


Figure 6-7. Auto-Retry-Mode Example

6.3.3 Stand-alone Operation

In a typical application, the TPS1H200A-Q1 device is controlled by a microcontroller. The device also supports stand-alone operation. IN and DIAG_EN have a 40V maximum DC rating, and can directly connect to the VS pin directly. When in auto-retry mode, the DELAY pin is connected to the VS pin through a 100kΩ resistor.

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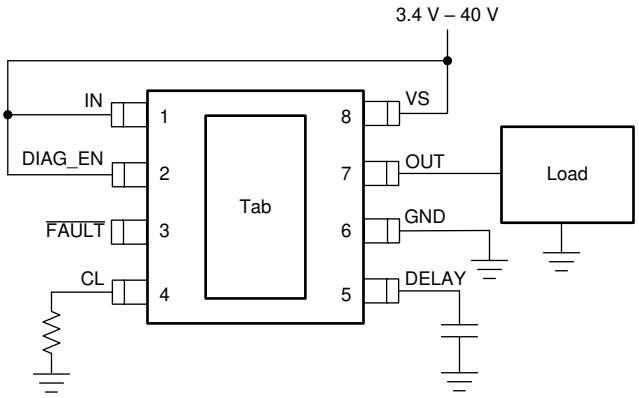


Figure 6-8. Stand-Alone Operation in Latch-Off Mode

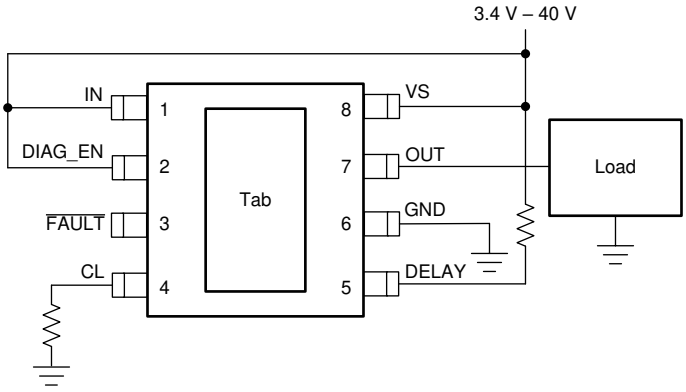


Figure 6-9. Stand-Alone Operation in Auto-Retry Mode

6.3.4 Fault Truth Table

The DIAG_EN pin enables or disables the diagnostic functions. If multiple devices are used, but the ADC resource is limited in the microcontroller, the microcontroller uses GPIOs to set DIAG_EN high to enable the diagnostics of one device, and disables the diagnostics of the other devices by setting DIAG_EN low. Additionally, the device can keep power consumption to a minimum by setting DIAG_EN and IN low.

Table 6-2 applies when the DIAG_EN pin is enabled, and Table 6-3 applies when the DIAG_EN pin is disabled.

Table 6-2. Fault Truth Table

CONDITION	IN	OUT	CRITERION	$\overline{\text{FAULT}}$	FAULT RECOVERY
Normal	L	L	N/A	H	N/A
	H	H	N/A	H	
Overload or short to GND	H	L	current limit triggered	L	See Table 6-1.
Open load or short to battery	H	H	$I_{OUT} < I_{(ol,on)}$	L	$\overline{\text{FAULT}}$ clears when IN turns low for a duration longer than $t_{\overline{\text{FAULT}}}$. OR $\overline{\text{FAULT}}$ clears when the open load is removed.
	L(1)	H	$V_{VS} - V_{OUT} < V_{(ol,off)}$	L	$\overline{\text{FAULT}}$ clears when IN is toggling OR $\overline{\text{FAULT}}$ clears when the open load is removed.
Thermal shutdown	H	N/A	Thermal shutdown triggered	L	$\overline{\text{FAULT}}$ clears when IN turns low for a duration longer than $t_{\overline{\text{FAULT}}}$. OR $\overline{\text{FAULT}}$ clears when thermal shutdown quits.

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Table 6-2. Fault Truth Table (continued)

CONDITION	IN	OUT	CRITERION	$\overline{\text{FAULT}}$	FAULT RECOVERY
Thermal swing	H	N/A	Thermal swing triggered	L	$\overline{\text{FAULT}}$ clears when IN turns low for a duration longer than $t_{\overline{\text{FAULT}}}$. OR $\overline{\text{FAULT}}$ clears when thermal swing quits.

(1) An external pullup is required for open-load detection.

Table 6-3. DIAG_EN Disabled Condition

DIAG_EN	IN CONDITION	PROTECTIONS AND DIAGNOSTICS
LOW	ON	Diagnostics disabled, full protections
	OFF	Diagnostics disabled, no protection

6.3.5 Full Diagnostics

6.3.5.1 Short-to-GND and Overload Detection

When the output is on, a short-to-GND or overload condition causes an overcurrent. If the overcurrent triggers the internal or external current limit threshold, the fault condition is reported as $\overline{\text{FAULT}}$ pin = low.

6.3.5.2 Open-Load Detection

6.3.5.2.1 Output On

When the output is on, the device recognizes an open-load fault if the current flowing through the output $I_{OUT} < I_{(ol,on)}$. For open-load detection when output is on, no external circuitry is required.

6.3.5.2.2 Output Off

When the output is off, the output is pulled down to GND if a load is connected. But if an open load occurs, the output voltage is close to the supply voltage ($V_{VS} - V_{OUT} < V_{(ol,off)}$), and the device recognizes an open-load fault.

There is always a leakage current $I_{(ol,off)}$ on the output due to the internal logic control path or external humidity, corrosion, and so forth. As a result, TI recommends using an external pullup resistor to offset the leakage current when an open load is detected. The recommended pullup resistance is 15kΩ.

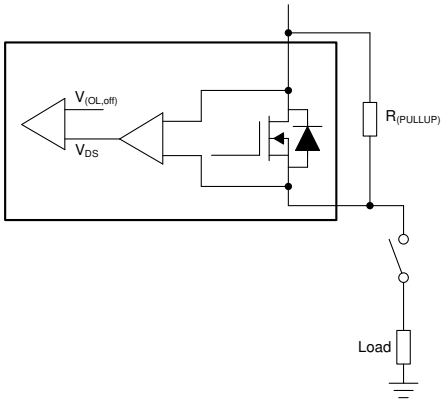


Figure 6-10. Open-Load Detection in Output OFF State

6.3.5.3 Short-to-Battery Detection

Short-to-battery has the same detection mechanism and behavior as open-load detection in the ON state and the OFF state.

6.3.5.4 Thermal Fault Detection

To protect the device in severe power stressing cases, the device implements two types of thermal fault detection, absolute temperature protection (thermal shutdown) and dynamic temperature protection (thermal swing).

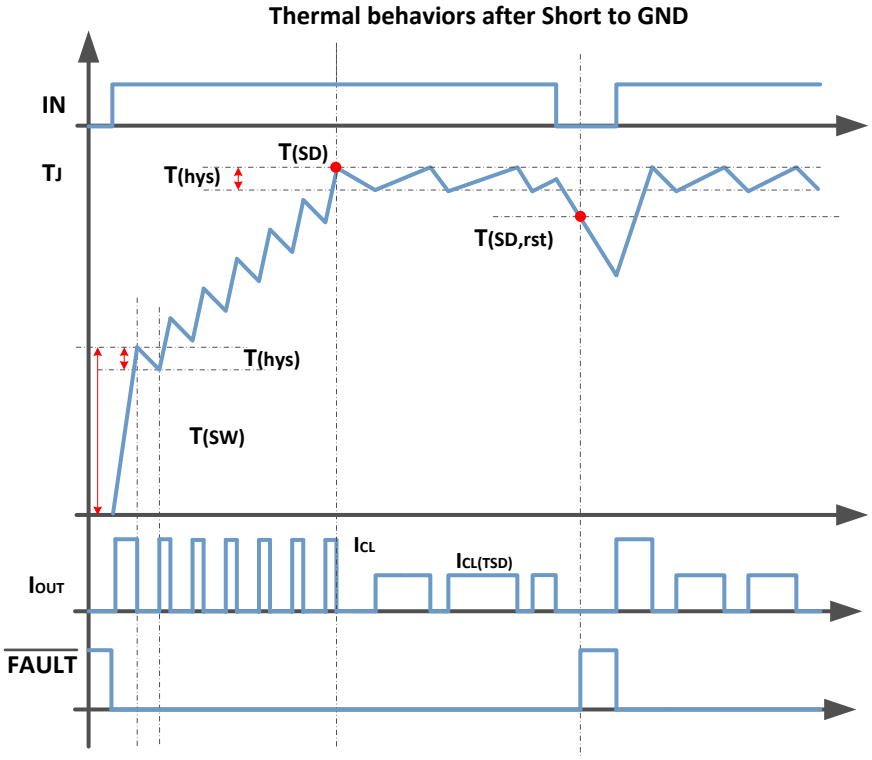


Figure 6-11. Thermal Behavior Diagram

6.3.5.4.1 Thermal Shutdown

Thermal shutdown is active when the absolute temperature $T_J > T_{(SD)}$. When thermal shutdown occurs, the output turns off.

6.3.5.4.2 Thermal Swing

Thermal swing activates when the power FET temperature sharply increases, that is, when Equation 3, then the output turns off.

$$\Delta T = T_{(FET)} - T_{(Logic)} > T_{(sw)} \quad (3)$$

The output automatically recovers and the fault signal clears when

$$\Delta T = T_{(FET)} - T_{(Logic)} < T_{(sw)} - T_{(hys)} \quad (4)$$

The thermal swing function improves the reliability of the device when subjected to repetitively fast thermal variation.

6.3.5.4.3 Fault Report Holding

When using PWM dimming, $\overline{\text{FAULT}}$ is easily cleared by the PWM falling edge. Even if the fault condition remains all the time, FAULT is discontinuous. To avoid this unexpected fault report behavior, the device implements fault report holding time. Figure 6-12 shows an issue that typically occurs during PWM dimming, the $\overline{\text{FAULT}}$ is cleared unexpectedly even when the short-to-GND still exists. The TPS1H200A-Q1 device with fault-report holding function allows the right behavior as shown in Figure 6-13.

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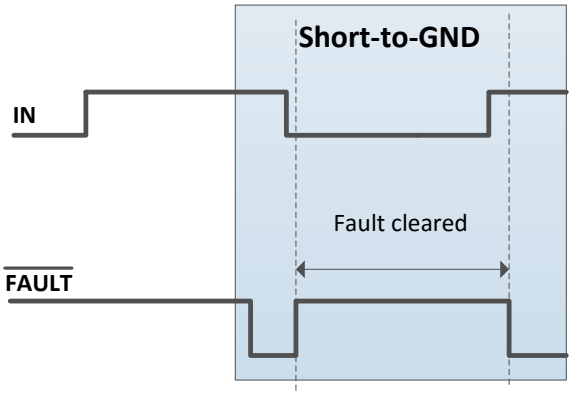


Figure 6-12. Without Fault-Report Holding

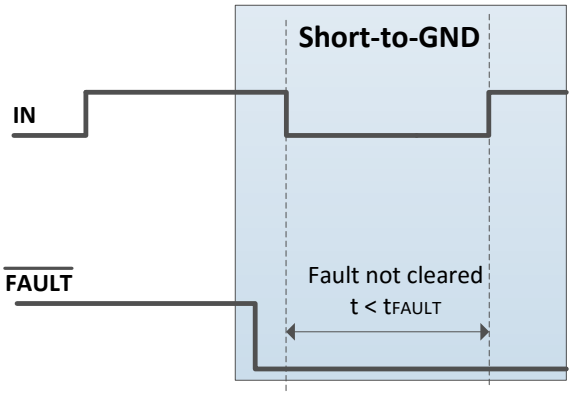


Figure 6-13. With Fault-Report Holding

6.3.6 Full Protections

6.3.6.1 UVLO Protection

The device monitors the supply voltage, V_{VS}, to prevent unpredicted behaviors when V_{VS} is too low. When V_{VS} drops down to $V_{VS(uvf)}$, the device shuts down. When V_{VS} rises up to $V_{VS(uvr)}$, the device turns on.

6.3.6.2 Inductive Load Switching Off Clamp

When an inductive load is switched off, the inductive reactance pulls the output voltage negative. However, excessive negative voltage can cause the power FET to break down. To protect the power FET from breaking down, an internal clamp ($V_{DS(clamp)}$) is implemented.

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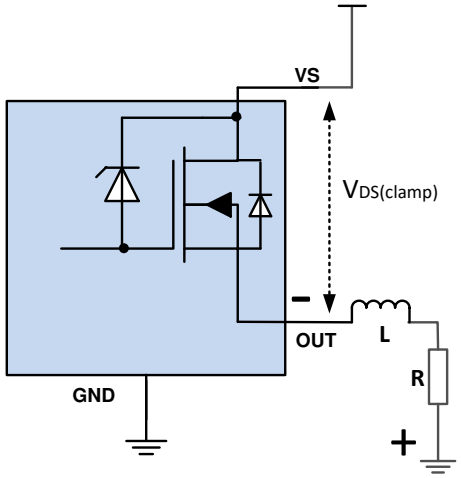


Figure 6-14. Drain-to-Source Clamping Structure

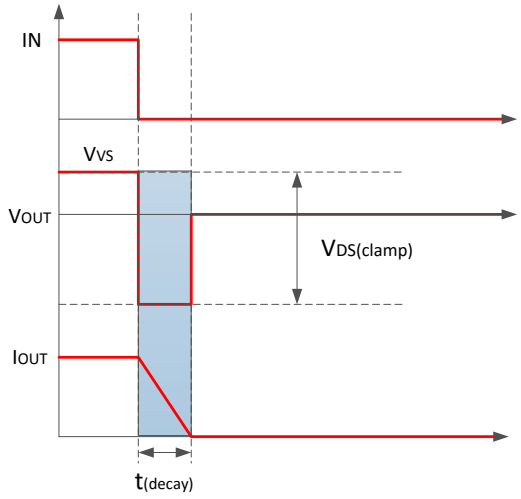


Figure 6-15. Inductive-Load Switching-Off Diagram

6.3.6.3 Loss-of-GND Protection

When a loss-of-GND occurs, the output shuts down, regardless of whether the IN pin is high or low. The device can protect against two ground-loss conditions, loss of device GND and loss of module GND.

6.3.6.4 Loss-of-Power-Supply Protection

When a loss-of-power-supply occurs, the output shuts down, regardless of whether the IN pin is high or low. For a resistive or a capacitive load, the loss-of-power-supply has no risk. For a charged inductive load, the current is driven from all the logic control pins to maintain the inductance current. To protect the system in this condition, TI recommends protection with an external free-wheeling diode.

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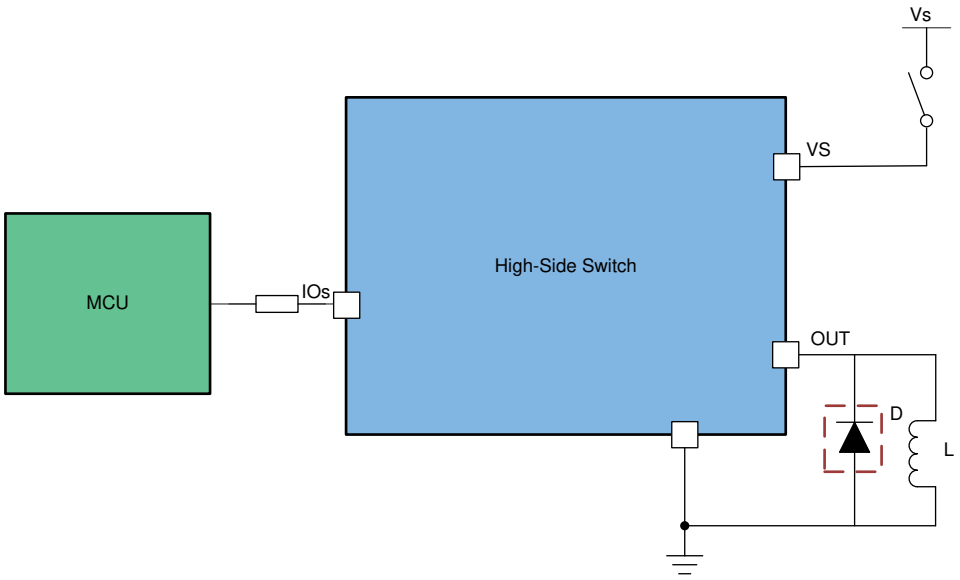


Figure 6-16. Protection for Loss-of-Power-Supply

6.3.6.5 Reverse-Current Protection

Reverse current occurs in two conditions: short to supply and reverse polarity.

- When a short to the supply occurs, there is only reverse current through the body diode. $I_{R(1)}$ specifies the limit of the reverse current.
- In a reverse-polarity condition, there are reverse currents through the body diode and the device GND pin. $I_{R(2)}$ specifies the limit of the reverse current.

To protect the device, TI recommends using two types of external circuitry.

- Adding a blocking diode (method 1). The device and load are protected when in reverse polarity.
- Adding a GND network (method 2). The reverse current through the device GND is blocked. The reverse current through the FET is limited by the load itself. TI recommends a resistor in parallel with the diode as a GND network. The recommended configuration is a 1kΩ resistor in parallel with a diode that is less than 100mA.

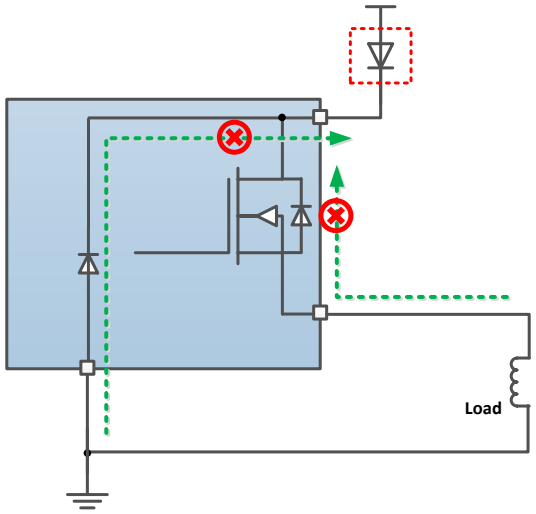


Figure 6-17. Reverse-Current External Protection Method 1

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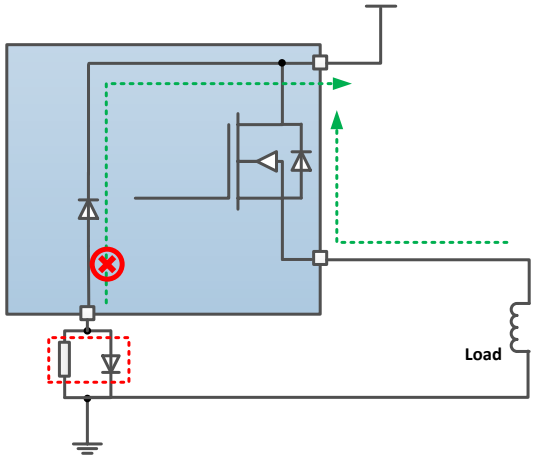


Figure 6-18. Reverse-Current External Protection Method 2

6.3.6.6 MCU I/O Protection

TI recommends using series resistors to protect the microcontroller, for example, 4.7kΩ when using a 3.3V microcontroller and 10 kΩ for a 5V microcontroller.

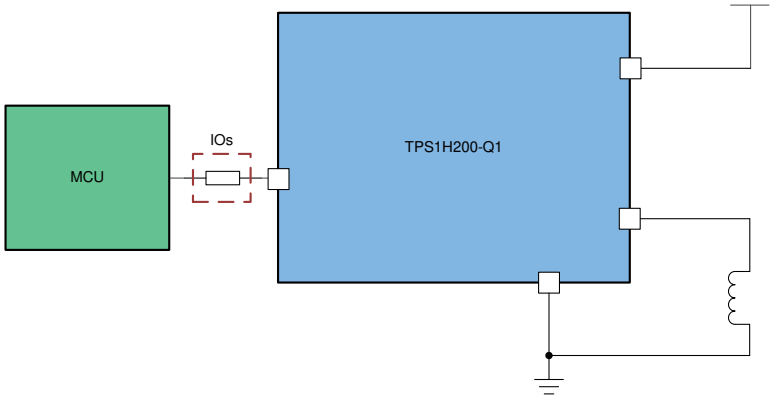


Figure 6-19. MCU I/O External Protection

6.4 Device Functional Modes

6.4.1 Working Modes

The device has three working modes: the normal mode, the standby mode, and the standby mode with diagnostics, as shown in Figure 6-20.

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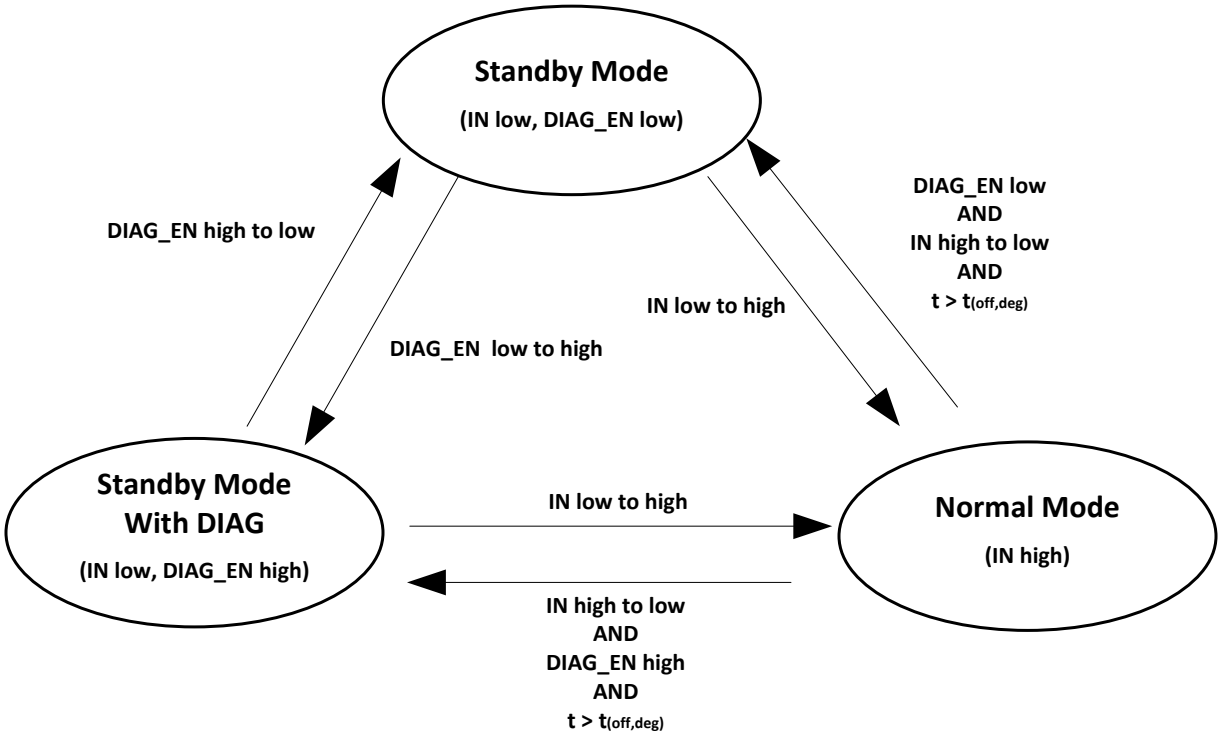


Figure 6-20. Working Modes

6.4.1.1 Normal Mode

When IN is high, the device enters normal mode.

6.4.1.2 Standby Mode

When IN is low and DIAG_EN is low, the device enters standby mode with ultra-low power consumption.

6.4.1.3 Standby Mode With Diagnostics

When IN is low and DIAG_EN is high, the device enters standby mode with diagnostics. The device still supports open-load and short-to-battery detection even when IN is low.

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7 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

7.1 Application Information

The TPS1H200A-Q1 device is a smart high-side switch, with an internal charge pump and single-channel integrated NMOS power FET. The adjustable current limit function greatly improves the reliability of the whole system. Full diagnostic features enable intelligent control of the load. The TPS1H200A-Q1 device applies for a wide variety of resistive, inductive, and capacitive loads, including LEDs, relays, and submodules.

7.2 Typical Application

Figure 7-1 shows an example of how to design the external circuitry parameters.

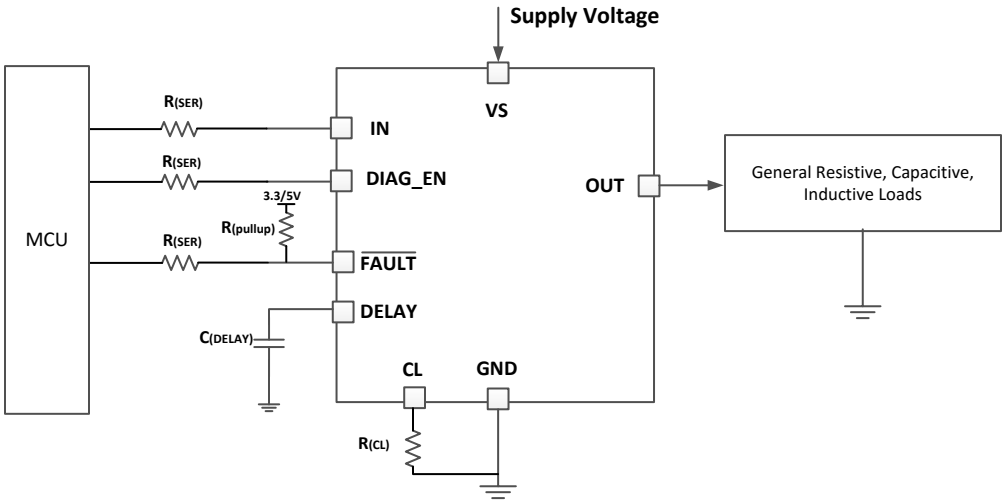


Figure 7-1. Typical Application Circuitry

7.2.1 Design Requirements

- V_{VS} range from 6V to 18V
- Nominal current of 500mA
- Expected current limit value of 2A
- Thermal sensitive system. When current limit occurs, the output latches off after 0.2 seconds. The 0.2 seconds is to establish the safe start-up for a capacitive load, clamping the inrush current but without latch-off during start-up.
- Full diagnostics with 5V MCU, including ON state open-load detection, short-to-GND, or overcurrent detection, and thermal shutdown detection

7.2.2 Detailed Design Procedure

To set the adjustable current limit value at 2A, calculate $R_{(CL)}$ as follows:

$$R_{(CL)} = \frac{V_{CL(th)} \times K_{(CL)}}{I_{OUT}} = \frac{0.8 \times 2500}{2} = 1000\Omega \quad (5)$$

To set the adjustable latch-off delay at 0.2s, calculate $C_{(DELAY)}$ as follows:

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$$t_{dl} = t_{CL(deg)} + t_{dl1} + t_{dl2} = 0.2 \gg t_{dl2}$$
$$C_{DELAY} = \frac{I_{dl(chg)} \times t_{dl2}}{V_{dl(rerf)}} = \frac{4.5 \times 0.2}{1.45} \times 10^{-6} = 0.62\mu F \quad (6)$$

TI recommends $R_{(SER)}$ = 10kΩ for a 5V MCU, and $R_{(pullup)}$ = 10kΩ as the pullup resistor.

7.2.3 Application Curves

The following curves are test examples of hard-short conditions. The load is 0.1A and the current limit value is 0.6A. Figure 7-2 shows a waveform of the latch-off mode. Figure 7-3 shows a waveform of the auto-retry mode.

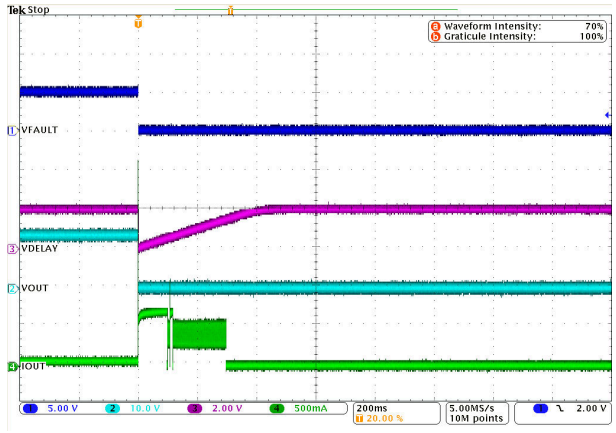


Figure 7-2. Hard-Short Condition in Latch-Off Mode

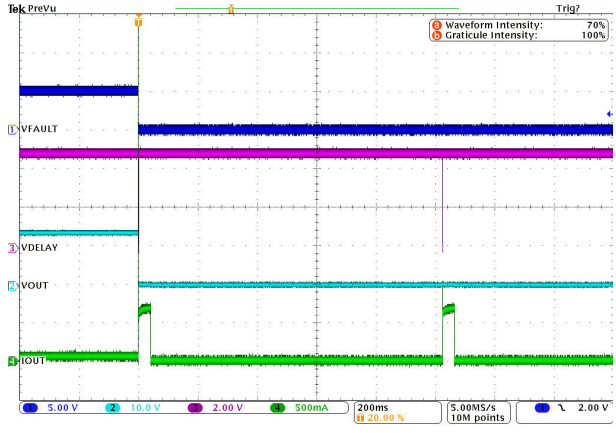


Figure 7-3. Hard-Short Condition in Auto-Retry Mode

7.3 Power Supply Recommendations

The device applies to 12V and 24V applications. The normal power supply connection is a 12V or 24V system.

7.4 Layout

7.4.1 Layout Guidelines

To prevent thermal shutdown, T_J must be less than 175°C. If the output current is high, the power dissipation can be large. However, the PCB layout is very important. A good PCB design optimizes heat transfer, which is essential for the long-term reliability of the device.

- Maximize the copper coverage on the PCB to increase the thermal conductivity of the board. The major heat-flow path from the package to the ambient is through the copper on the PCB. Maximum copper is extremely important when no heat sinks are attached to the PCB on the other side of the board opposite the package.
- Add as many thermal vias as possible directly under the package ground pad to optimize the thermal conductivity of the board.
- All thermal vias must either be plated shut or plugged and capped on both sides of the board to prevent solder voids. To ensure reliability and performance, the solder coverage must be at least 85%.

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7.4.2 Layout Example

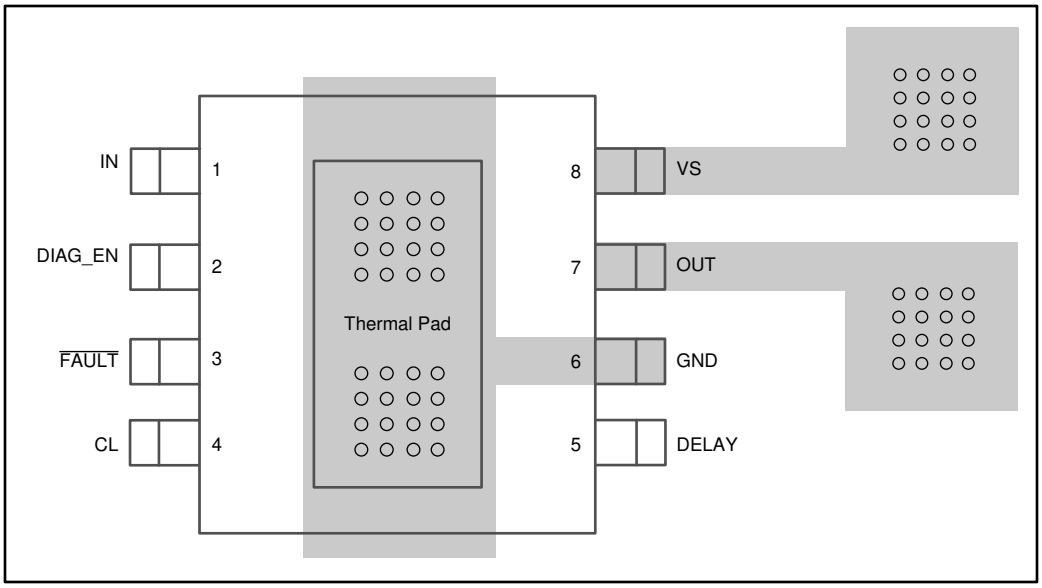


Figure 7-4. Layout Example

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8 Device and Documentation Support

8.1 Documentation Support

8.1.1 Related Documentation

For related documentation see the following:

- Texas Instruments, *TPS1H000-Q1 Evaluation Module (EVM)* User's Guide

8.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

8.3 Support Resources

TI E2E™ support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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8.4 Trademarks

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8.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

8.6 Glossary

TI Glossary	This glossary lists and explains terms, acronyms, and definitions.

9 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision D (September 2021) to Revision E (May 2026)	Page
• Updated the numbering format for tables, figures, and cross-references throughout the document	1
• Changed $R_{\theta JC(top)}$ from 49.2°C/W to 67.8°C/W	5
• Changed $R_{\theta JB(top)}$ from 18.3°C/W to 18.7°C/W	5
• Changed ψ_{JT} from 0.8°C/W to 3.1°C/W	5
• Changed ψ_{JB} from 18.4°C/W to 18.7°C/W	5
• Updated exposed pad dimensions to relax minimum limit	26

Changes from Revision C (December 2019) to Revision D (September 2021)	Page
• Changed the nominal operating current ($I_{(OP)}$) V_{DIAG_EN} variable to "X" for do not care in the *Electrical Characteristics* table	5

10 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

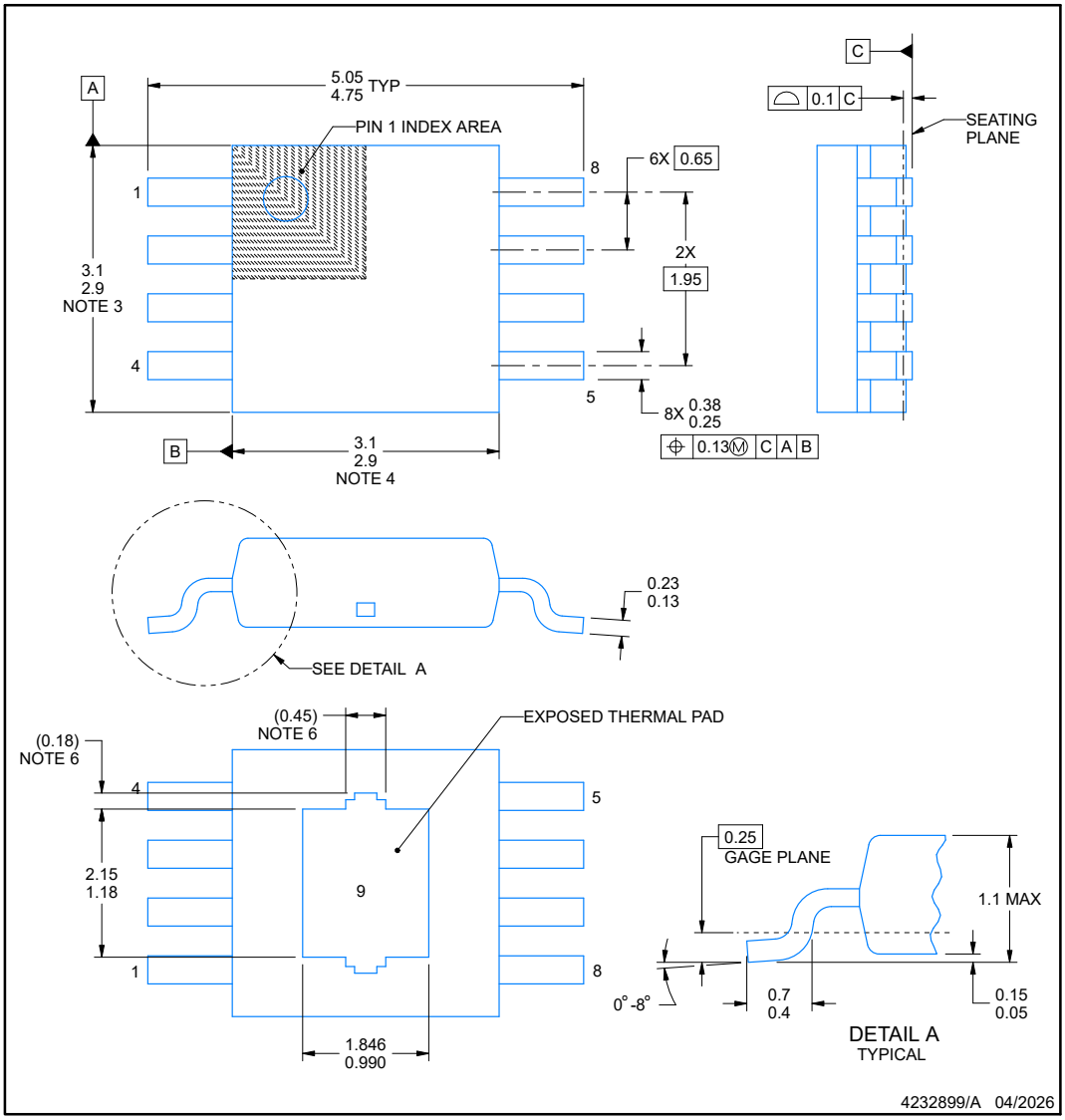
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10.1 Mechanical Data

PACKAGE OUTLINE

DGN0008K-C01  **PowerPAD™ VSSOP - 1.1 mm max height**

SMALL OUTLINE PACKAGE



PowerPAD is a trademark of Texas Instruments.

NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-187.
6. Features may differ or may not be present.

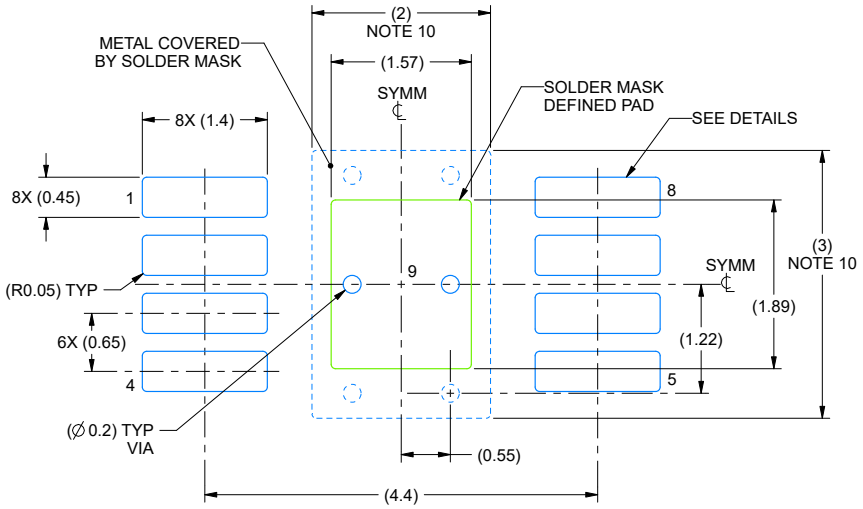
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EXAMPLE BOARD LAYOUT

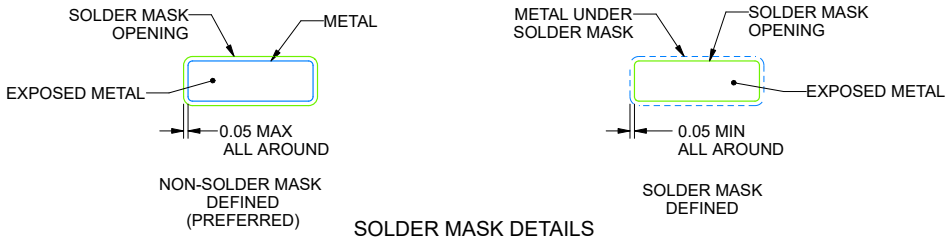
DGN0008K-C01

PowerPAD™ VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 15X



SOLDER MASK DETAILS

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NOTES: (continued)

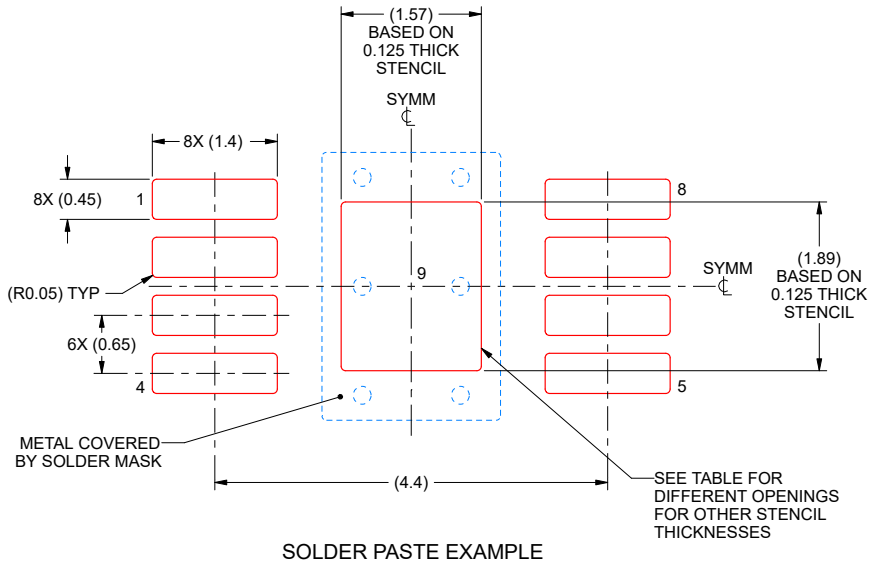
7. Publication IPC-7351 may have alternate designs.
8. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
9. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.
10. Size of metal pad may vary due to creepage requirement.

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EXAMPLE STENCIL DESIGN

DGN0008K-C01 **PowerPAD™ VSSOP - 1.1 mm max height**

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE
EXPOSED PAD 9:
100% PRINTED SOLDER COVERAGE BY AREA
SCALE: 15X

STENCIL THICKNESS	SOLDER STENCIL OPENING
0.1	1.76 X 2.11
0.125	1.57 X 1.89 (SHOWN)
0.15	1.43 X 1.73
0.175	1.33 X 1.60

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NOTES: (continued)

11. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
12. Board assembly site may have different recommendations for stencil design.

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PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package | Pins	Package qty | Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
TPS1H200AQDGNRQ1	Active	Production	HVSSOP (DGN) | 8	2500 | LARGE T&R	Yes	NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	1EWX
TPS1H200AQDGNRQ1.A	Active	Production	HVSSOP (DGN) | 8	2500 | LARGE T&R	Yes	NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	1EWX

(1) **Status:** For more details on status, see our product life cycle.

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

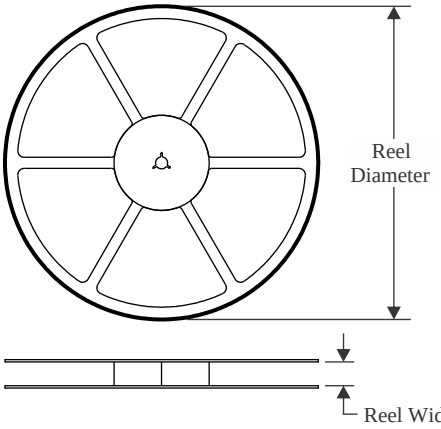
Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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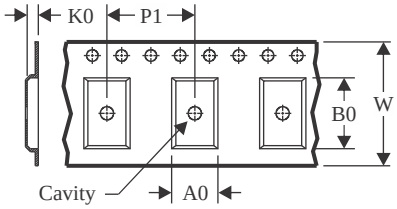
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TAPE AND REEL INFORMATION

REEL DIMENSIONS

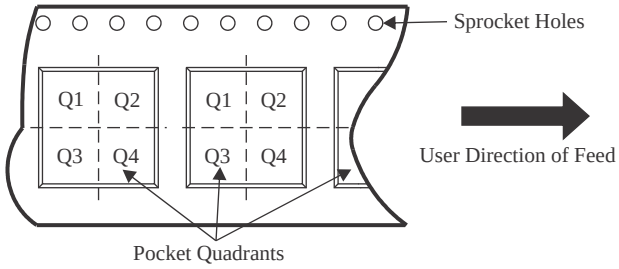


TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

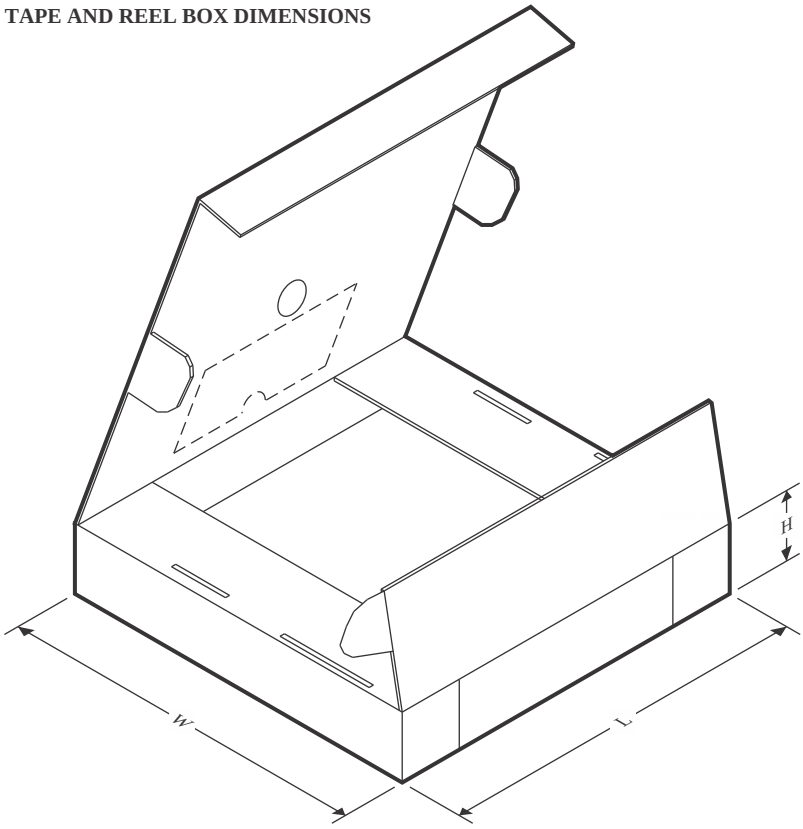
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS1H200AQDGNRQ1	HVSSOP	DGN	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS1H200AQDGNRQ1	HVSSOP	DGN	8	2500	366.0	364.0	50.0

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