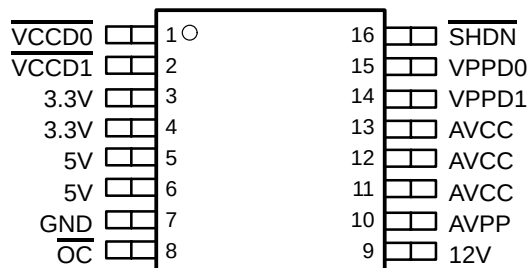


TPS2211 SINGLE-SLOT PC CARD POWER INTERFACE SWITCH FOR PARALLEL PCMCIA CONTROLLERS

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- Fully Integrated V_{CC} and V_{pp} Switching for Single-Slot PC Card™ Interface
- Low $r_{DS(on)}$ (90-m Ω 5-V V_{CC} Switch and 3.3-V V_{CC} Switch)
- Compatible With Controllers From Cirrus, Ricoh, O₂Micro, Intel, and Texas Instruments
- 3.3-V Low-Voltage Mode
- Meets PC Card Standards
- 12-V Supply Can Be Disabled Except During 12-V Flash Programming
- Short-Circuit and Thermal Protection
- Space-Saving 16-Pin SSOP (DB)
- Compatible With 3.3-V, 5-V, and 12-V PC Cards
- Break-Before-Make Switching

DB PACKAGE
(TOP VIEW)



description

The TPS2211 PC Card power-interface switch provides an integrated power-management solution for a single PC Card. All of the discrete power MOSFETs, a logic section, current limiting, and thermal protection for PC Card control are combined on a single integrated circuit, using the Texas Instruments LinBiCMOS™ process. The circuit allows the distribution of 3.3-V, 5-V, and/or 12-V card power, and is compatible with many PCMCIA controllers. The current-limiting feature eliminates the need for fuses, which reduces component count and improves reliability. Current-limit reporting can help the user isolate a system fault to the PC Card.

The TPS2211 features a 3.3-V low-voltage mode that allows for 3.3-V switching without the need for 5 V. Bias power can be derived from either the 3.3-V or 5-V inputs. This facilitates low-power system designs such as sleep mode and pager mode where only 3.3 V is available.

End equipment for the TPS2211 includes notebook computers, desktop computers, personal digital assistants (PDAs), digital cameras, and bar-code scanners.

AVAILABLE OPTIONS

T_A	PACKAGED DEVICE	CHIP FORM (Y)
	SMALL OUTLINE (DB)	
–40°C to 85°C	TPS2211IDBR	TPS2211Y

The DB package is only available taped and reeled, indicated by the R suffix on the device type.



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PC Card is a trademark of PCMCIA (Personal Computer Memory Card International Association).
LinBiCMOS is a trademark of Texas Instruments Incorporated.

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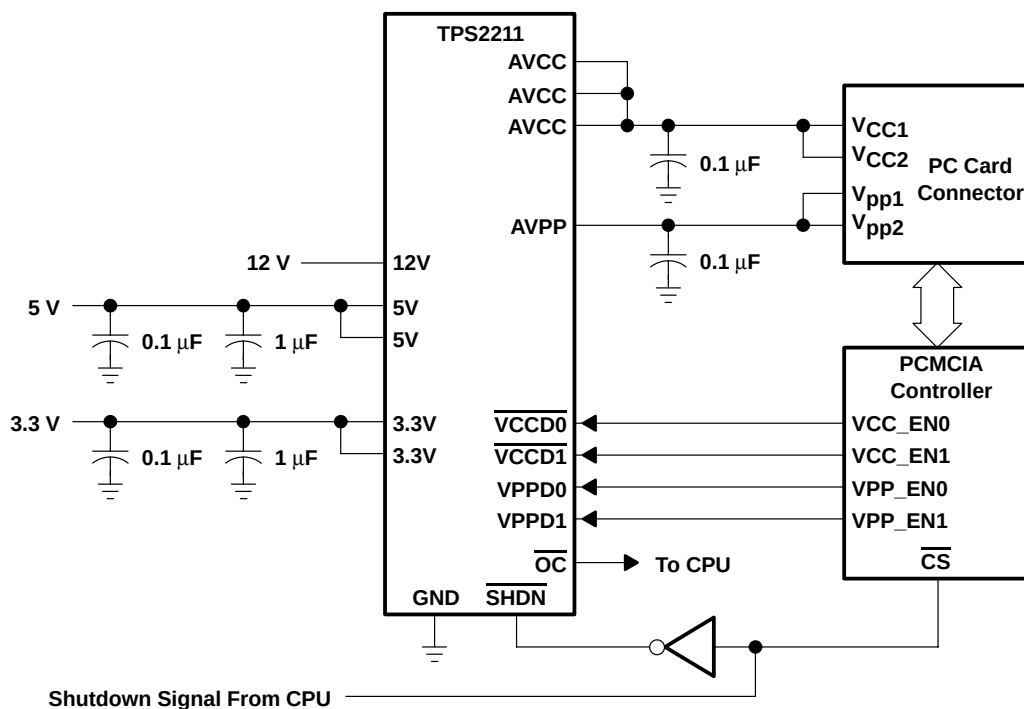
TPS2211

SINGLE-SLOT PC CARD POWER INTERFACE SWITCH

FOR PARALLEL PCMCIA CONTROLLERS

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typical PC-card power-distribution application

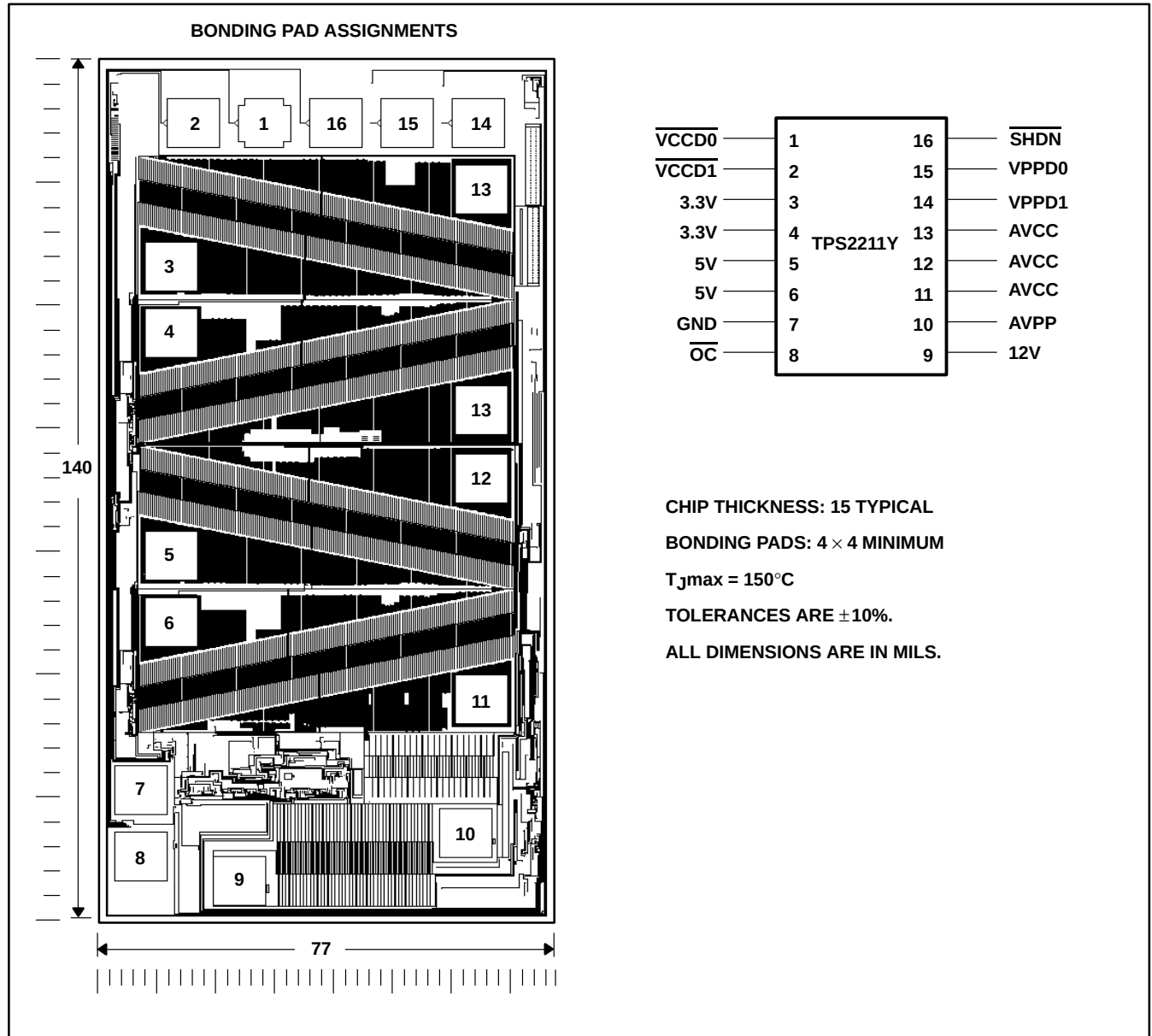


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TPS2211Y chip information

This chip, when properly assembled, displays characteristics similar to those of the TPS2211. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. The chips may be mounted with conductive epoxy or a gold-silicon preform.



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Terminal Functions

TERMINAL NAME	NO.	I/O	DESCRIPTION
3.3V	3, 4	I	3.3-V V_{CC} input for card power and/or chip power if 5 V is not present
5V	5, 6	I	5-V V_{CC} input for card power and/or chip power
12V	9	I	12-V V_{pp} input card power
AVCC	11, 12, 13	O	Switched output that delivers 0 V, 3.3-V, 5-V, or high impedance to card
AVPP	10	O	Switched output that delivers 0 V 3.3-V, 5-V, 12-V, or high impedance to card
GND	7		Ground
$\overline{OC}$	8	O	Logic-level overcurrent reporting output that goes low when an overcurrent conditions exists
$\overline{SHDN}$	16	I	Logic input that shuts down the TPS2211 and sets all power outputs to high-impedance state
$\overline{VCCD0}$	1	I	Logic input that controls voltage of AVCC (see control-logic table)
$\overline{VCCD1}$	2	I	Logic input that controls voltage of AVCC (see control-logic table)
$\overline{VPPD0}$	15	I	Logic input that controls voltage of AVPP (see control-logic table)
$\overline{VPPD1}$	14	I	Logic input that controls voltage of AVPP (see control-logic table)

absolute maximum ratings over operating free-air temperature (unless otherwise noted)[†]

Input voltage range for card power:	$V_{I(5V)}$	–0.3 V to 7 V
	$V_{I(3.3V)}$	–0.3 V to 7 V
	$V_{I(12V)}$	–0.3 V to 14 V
Logic input voltage		–0.3 V to 7 V
Continuous total power dissipation		See Dissipation Rating Table
Output current (each card):	$I_{O(VCC)}$	internally limited
	$I_{O(VPP)}$	internally limited
Operating virtual junction temperature range, T_J		–40°C to 150°C
Operating free-air temperature range, T_A		–40°C to 85°C
Storage temperature range, T_{stg}		–55°C to 150°C
Lead temperature 1.6 mm (1/16 inch) from case for 10 seconds		260°C

[†] Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING
DB	775 mW	6.2 mW/°C	496 mW	403 mW

These devices are mounted on an FR4 board with no special thermal considerations.

recommended operating conditions

		MIN	MAX	UNIT
Input voltage, V_I	$V_{I(5V)}$	0	5.25	V
	$V_{I(3.3V)}$	0	5.25	V
	$V_{I(12V)}$	0	13.5	V
Output current	$I_{O(AVCC)}$		1	A
	$I_{O(AVPP)}$		150	mA
Operating virtual junction temperature, T_J		–40	125	°C



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electrical characteristics, $T_A = -40^{\circ}\text{C}$ to 85°C (unless otherwise noted)

power switch

PARAMETER		TEST CONDITIONS†	TPS2211			UNIT	
			MIN	TYP	MAX		
Switch resistance	5 V to AVCC	V _I (5V) = 5 V		50	90	mΩ	
	3.3 V to AVCC	V _I (5V) = 5 V, V _I (3.3V) = 3.3 V		48	90		
	3.3 V to AVCC	V _I (5V) = 0 V, V _I (3.3V) = 3.3 V		48	90		
	5 V to AVPP	T _J =25°C			6	Ω	
	3.3 V to AVPP	T _J =25°C			6		
	12 V to AVPP	T _J =25°C			1		
V _O (AVPP)	Clamp low voltage	I _{pp} at 10 mA			0.8	V	
V _O (AVCC)	Clamp low voltage	I _{CC} at 10 mA			0.8	V	
I _{lkg}	Leakage current	I _{pp} high-impedance state	T _A = 25°C		1	10	μA
			T _A = 85°C				
	I _{CC} high-impedance state	T _A = 25°C		1	10		
		T _A =85°C				50	
I _I	Input current	V _I (5V) = 5 V	V _O (AVCC) = 5 V, V _O (AVPP) = 12 V		40	150	μA
		V _I (5V) = 0 V, V _I (3.3V) = 3.3 V	V _O (AVCC) = 3.3 V, V _O (AVPP) = 12 V		40	150	
		Shutdown mode	V _O (AVCC) = V _O (AVPP) = Hi-Z			1	
I _{OS}	Short-circuit output-current limit	I _O (AVCC)	T _J = 85°C, output powered into a short to GND		1	2.2	A
		I _O (AVPP)			120	400	mA

[†] Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.

logic section

PARAMETER	TEST CONDITIONS†	TPS2211		UNIT
		MIN	MAX	
Logic input current			1	μA
Logic input high level			2	V
Logic input low level			0.8	V
Logic output high level	V _{I(5V)} = 5 V, I _O = 1 mA	V _{I(5V)} – 0.4		V
	V _{I(5V)} = 0 V, I _O = 1 mA, V _{I(3.3V)} = 3.3 V	V _{I(3.3V)} – 0.4		
Logic output low level	I _O = 1 mA		0.4	V

[†] Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.

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electrical characteristics, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

power switch

PARAMETER		TEST CONDITIONS†	TPS2211Y			UNIT
			MIN	TYP	MAX	
Switch resistance	5 V to AVCC	$V_{I(5V)} = 5\text{ V}$		50		$\text{m}\Omega$
	3.3 V to AVCC	$V_{I(5V)} = 5\text{ V}, V_{I(3.3V)} = 3.3\text{ V}$		48		
	3.3 V to AVCC	$V_{I(5V)} = 0\text{ V}, V_{I(3.3V)} = 3.3\text{ V}$		48		
	5 V to AVPP	$T_J = 25^\circ\text{C}$		4.3		Ω
	3.3 V to AVPP	$T_J = 25^\circ\text{C}$		4.3		
	12 V to AVPP	$T_J = 25^\circ\text{C}$		0.5		
$V_O(\text{AVPP})$	Clamp low voltage	I_{pp} at 10 mA		0.28		V
$V_O(\text{AVCC})$	Clamp low voltage	I_{pp} at 10 mA		0.28		V
I_{lkg}	Leakage current	I_{pp} high-impedance state		1		μA
		I_{CC} high-impedance state		1		
I_I	Input current	$V_I = 5\text{ V}$		42		μA
		$V_{I(5V)} = 5\text{ V}, V_{I(3.3V)} = 3.3\text{ V}$		42		
		$V_O(\text{AVCC}) = 3.3\text{ V}, V_O(\text{AVPP}) = 12\text{ V}$		42		

† Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.

switching characteristics‡

PARAMETER		TEST CONDITIONS§	TPS2211, TPS2211Y			UNIT
			MIN	TYP	MAX	
t_r	Rise times, output	$V_O(\text{AVCC})$		2.8		ms
		$V_O(\text{AVPP})$		6.4		
t_f	Fall times, output	$V_O(\text{AVCC})$		4.5		
		$V_O(\text{AVPP})$		12		
t_{pd}	Propagation delay (see Figure1)	$V_I(\text{VPPD0})$ to $V_O(\text{AVPP})$	t_{on}	6.8		ms
			t_{off}	18		
		$V_I(\overline{\text{VCCD1}})$ to $V_O(\text{AVCC})$ (3.3V)	t_{on}	4		
			t_{off}	17		
		$V_I(\overline{\text{VCCD0}})$ to $V_O(\text{AVCC})$ (5V)	t_{on}	6.6		
			t_{off}	17		

‡ Switching Characteristics are with $C_L = 150\text{ }\mu\text{F}$.

§ Refer to Parameter Measurement Information

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PARAMETER MEASUREMENT INFORMATION

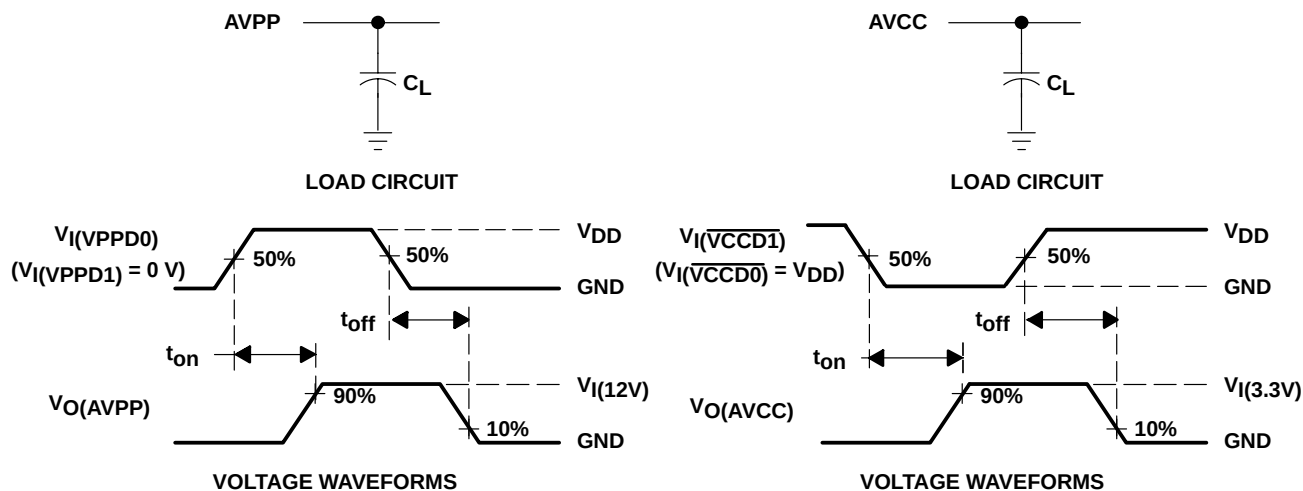


Figure 1. Test Circuits and Voltage Waveforms

Table of Timing Diagrams

	FIGURE
AVCC Propagation Delay and Rise Time With 1- μF Load, 3.3-V Switch	2
AVCC Propagation Delay and Fall Time With 1- μF Load, 3.3-V Switch	3
AVCC Propagation Delay and Rise Time With 150- μF Load, 3.3-V Switch	4
AVCC Propagation Delay and Fall Time With 150- μF Load, 3.3-V Switch	5
AVCC Propagation Delay and Rise Time With 1- μF Load, 5-V Switch	6
AVCC Propagation Delay and Fall Time With 1- μF Load, 5-V Switch	7
AVCC Propagation Delay and Rise Time With 150- μF Load, 5-V Switch	8
AVCC Propagation Delay and Fall Time With 150- μF Load, 5-V Switch	9
AVPP Propagation Delay and Rise Time With 1- μF Load, 12-V Switch	10
AVPP Propagation Delay and Fall Time With 1- μF Load, 12-V Switch	11
AVPP Propagation Delay and Rise Time With 150- μF Load, 12-V Switch	12
AVPP Propagation Delay and Fall Time With 150- μF Load, 12-V Switch	13

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PARAMETER MEASUREMENT INFORMATION

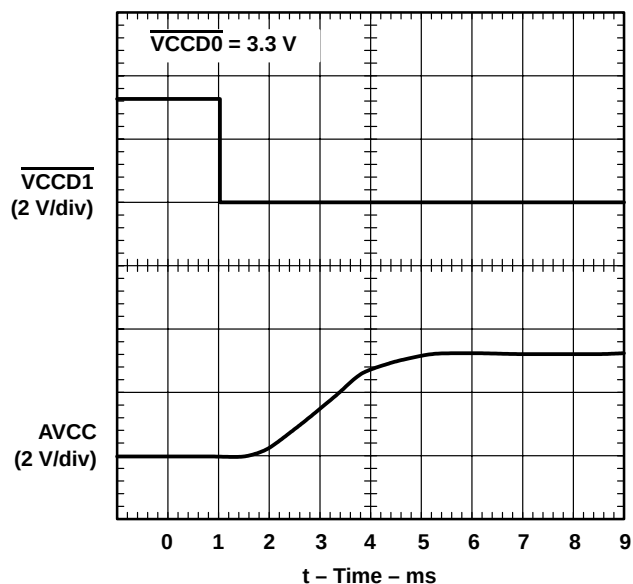


Figure 2. AVCC Propagation Delay and Rise Time With 1-μF Load, 3.3-V Switch

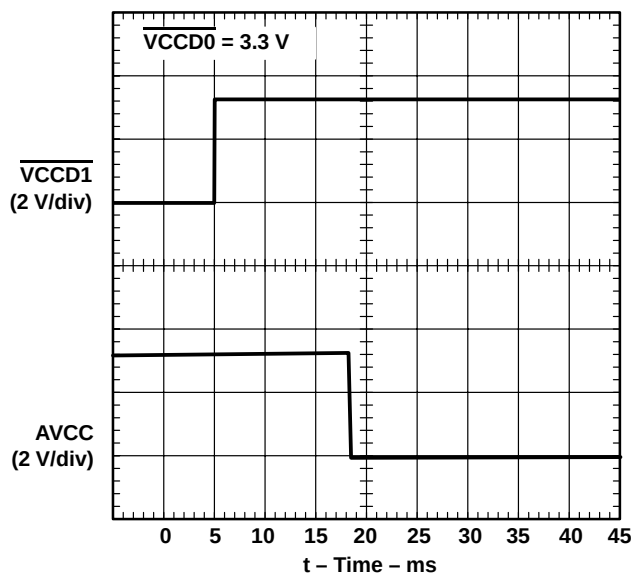


Figure 3. AVCC Propagation Delay and Fall Time With 1-μF Load, 3.3-V Switch

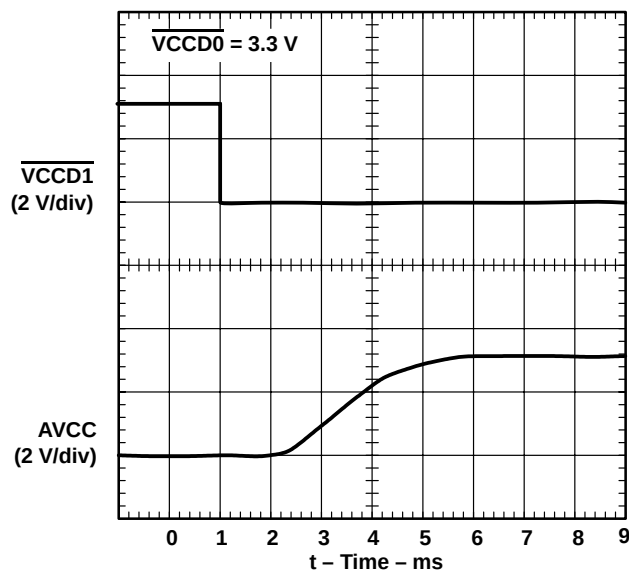


Figure 4. AVCC Propagation Delay and Rise Time With 150-μF Load, 3.3-V Switch

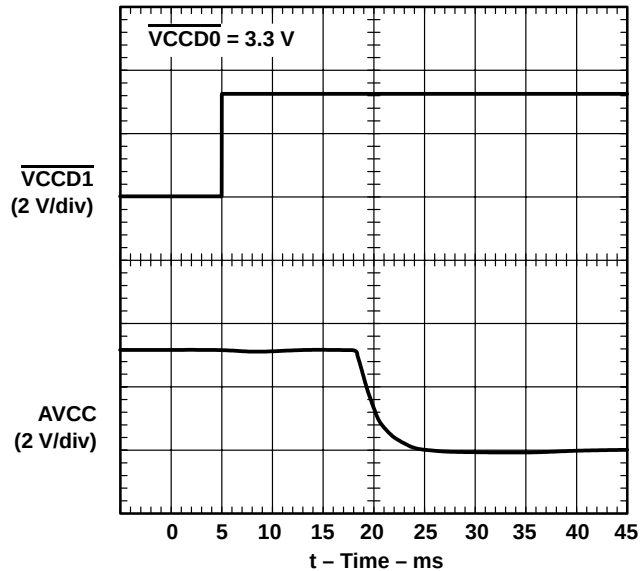


Figure 5. AVCC Propagation Delay and Fall Time With 150-μF Load, 3.3-V Switch

PARAMETER MEASUREMENT INFORMATION

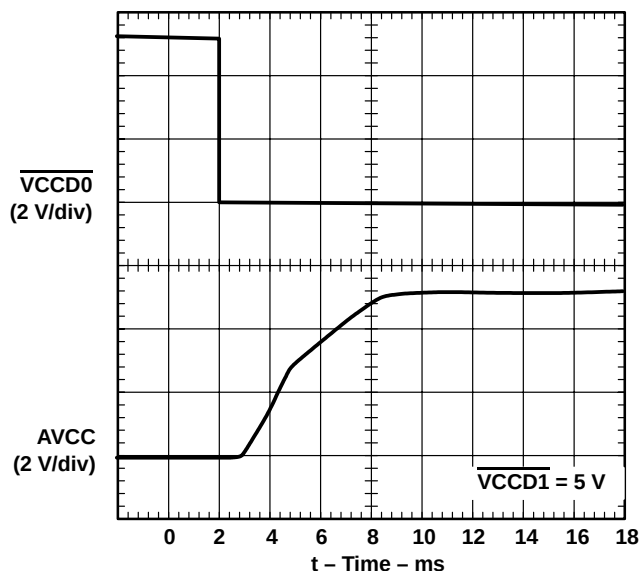


Figure 6. AVCC Propagation Delay and Rise Time With 1- μ F Load, 5-V Switch

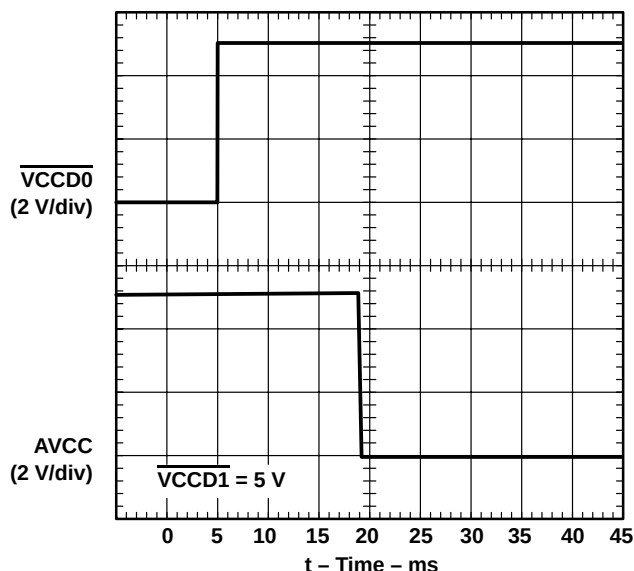


Figure 7. AVCC Propagation Delay and Fall Time With 1- μ F Load, 5-V Switch

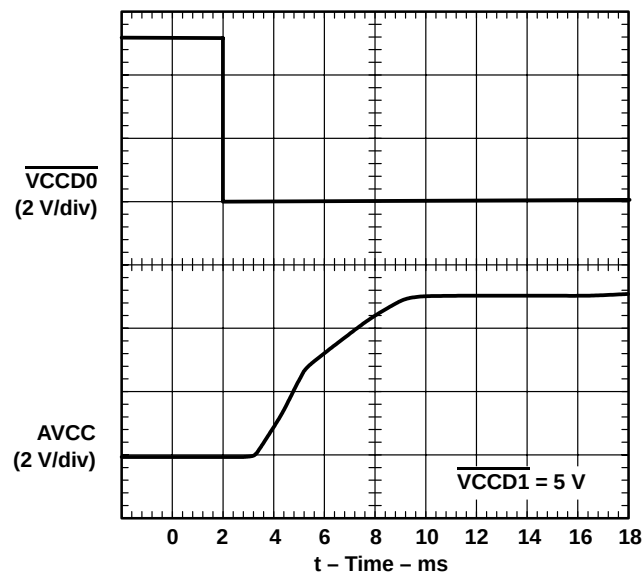


Figure 8. AVCC Propagation Delay and Rise Time With 150- μ F Load, 5-V Switch

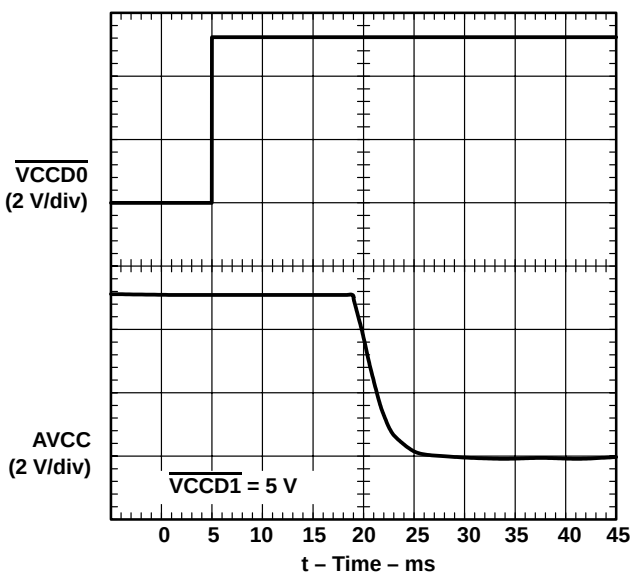


Figure 9. AVCC Propagation Delay and Fall Time With 150- μ F Load, 5-V Switch

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PARAMETER MEASUREMENT INFORMATION

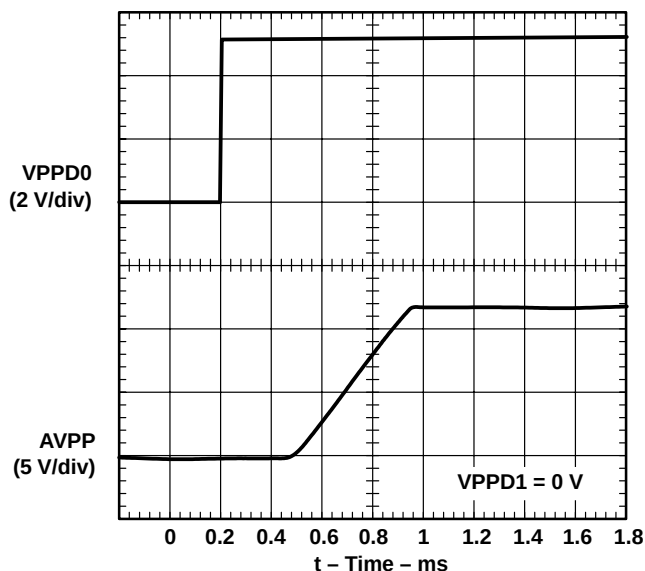


Figure 10. AVPP Propagation Delay and Rise Time With 1- μ F Load, 12-V Switch

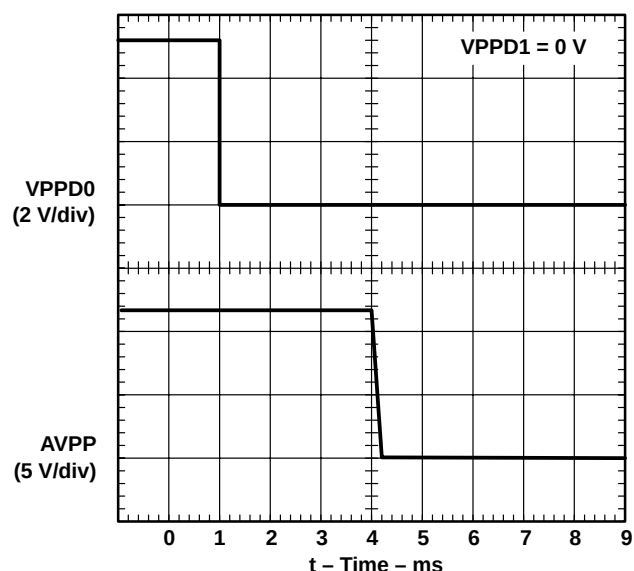


Figure 11. AVPP Propagation Delay and Fall Time With 1- μ F Load, 12-V Switch

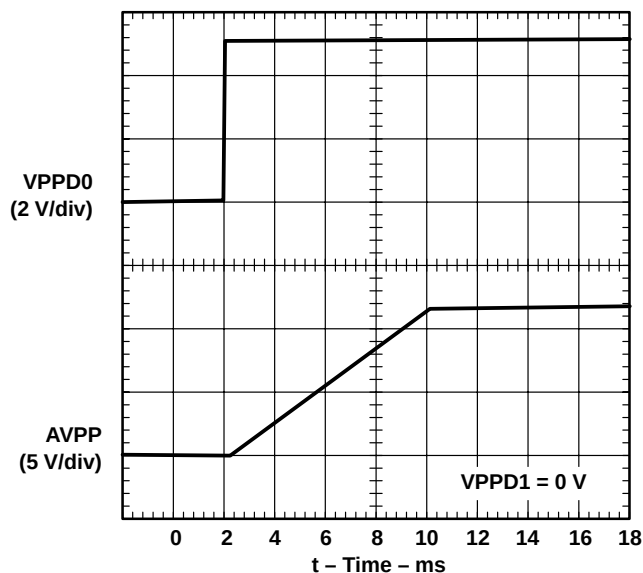


Figure 12. AVPP Propagation Delay and Rise Time With 150- μ F Load, 12-V Switch

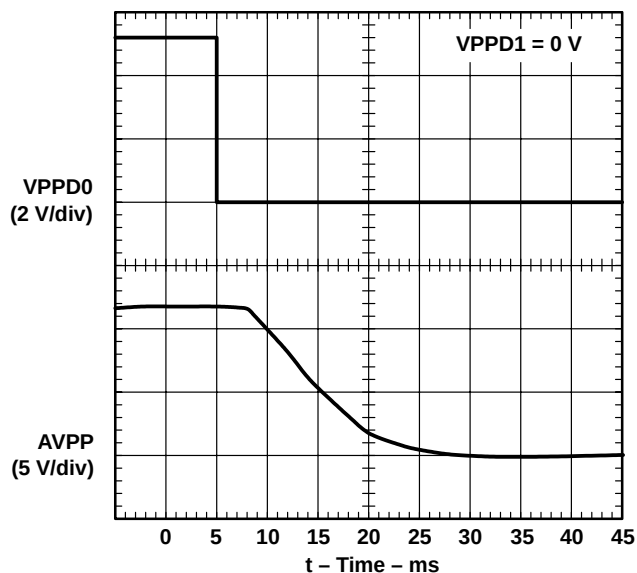


Figure 13. AVPP Propagation Delay and Fall Time With 150- μ F Load, 12-V Switch

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TYPICAL CHARACTERISTICS

Table of Graphs

			FIGURE
$I_{CC}(5V)$	Supply current	vs Junction temperature	14
$I_{CC}(3.3V)$	Supply current	vs Junction temperature	15
$r_{DS(on)}$	Static drain-source on-state resistance, 5-V VCC switch	vs Junction temperature	16
$r_{DS(on)}$	Static drain-source on-state resistance, 3.3-V VCC switch	vs Junction temperature	17
$r_{DS(on)}$	Static drain-source on-state resistance, 12-V VPP switch	vs Junction temperature	18
$V_O(AVCC)$	Output voltage, 5-V VCC switch	vs Output current	19
$V_O(AVCC)$	Output voltage, 3.3-V VCC switch	vs Output current	20
$V_O(AVPP)$	Output voltage, 12-V VPP switch	vs Output current	21
$I_{OS}(AVCC)$	Short-circuit current, 5-V VCC switch	vs Junction temperature	22
$I_{OS}(AVCC)$	Short-circuit current, 3.3-V VCC switch	vs Junction temperature	23
$I_{OS}(AVPP)$	Short-circuit current, 12-V VPP switch	vs Junction temperature	24

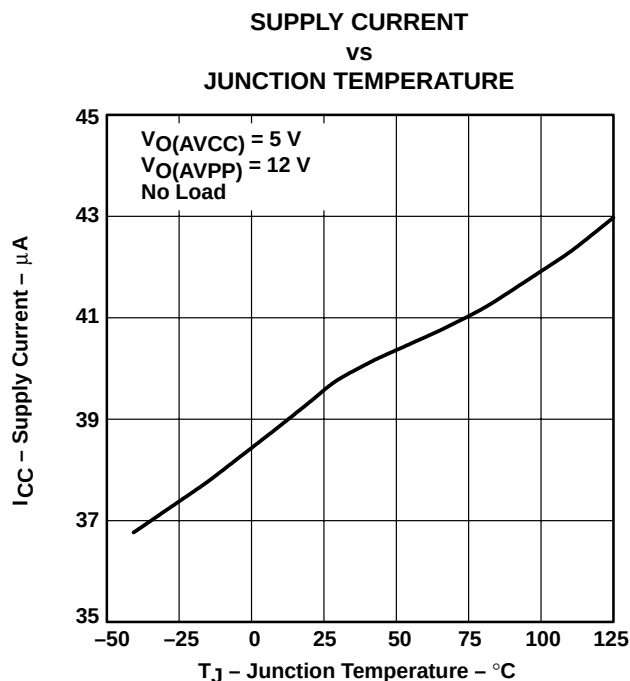


Figure 14

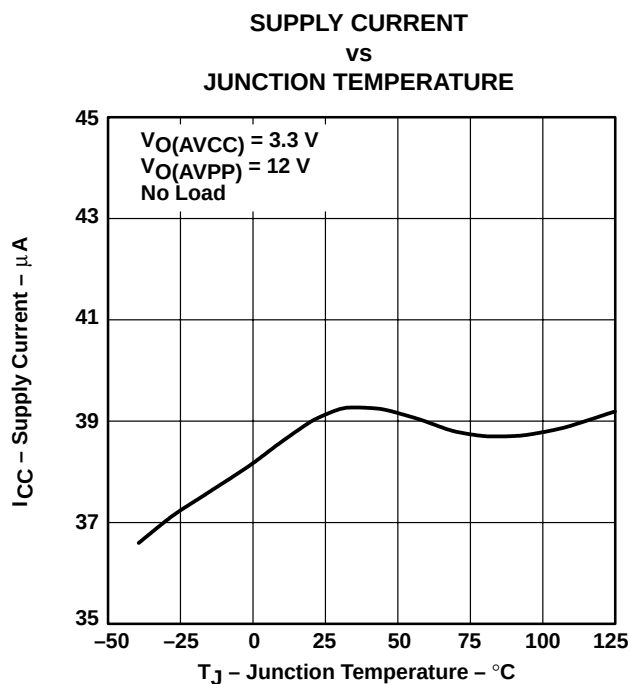


Figure 15

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TYPICAL CHARACTERISTICS

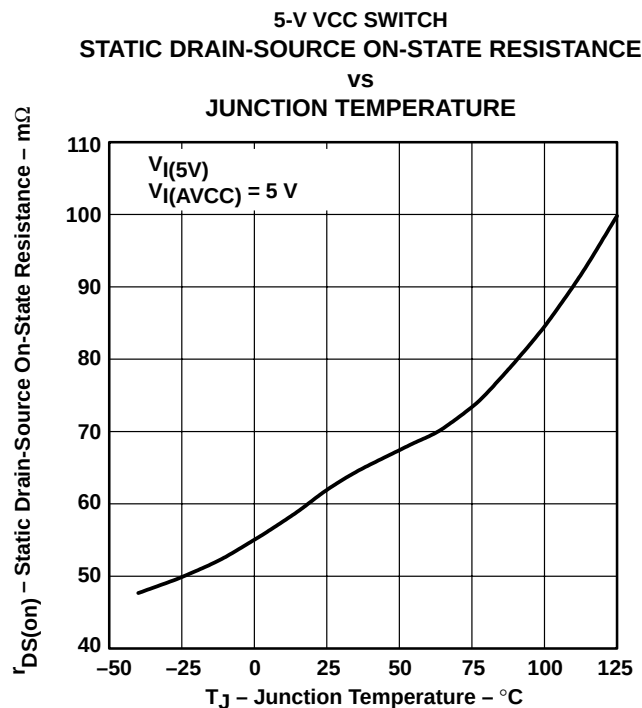


Figure 16

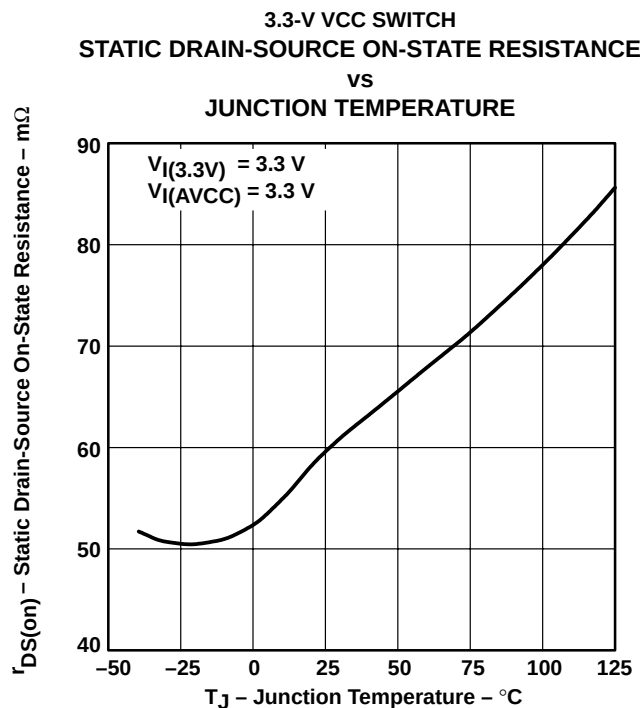


Figure 17

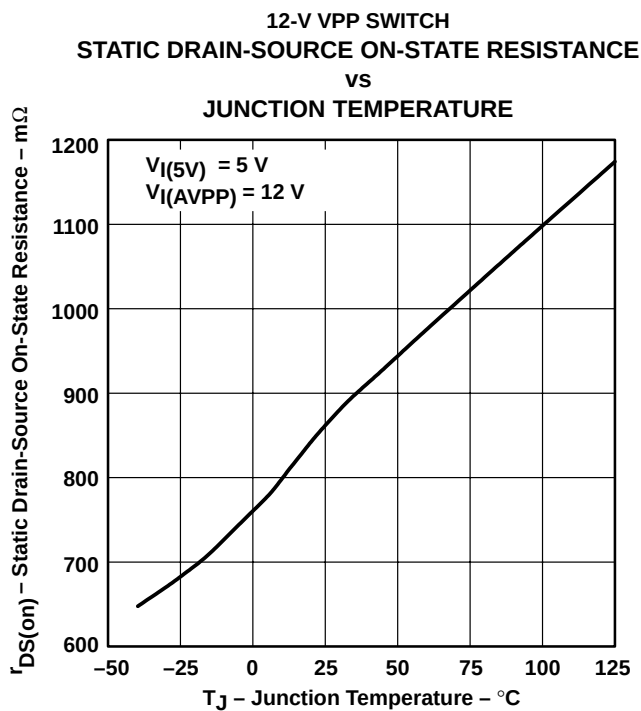


Figure 18

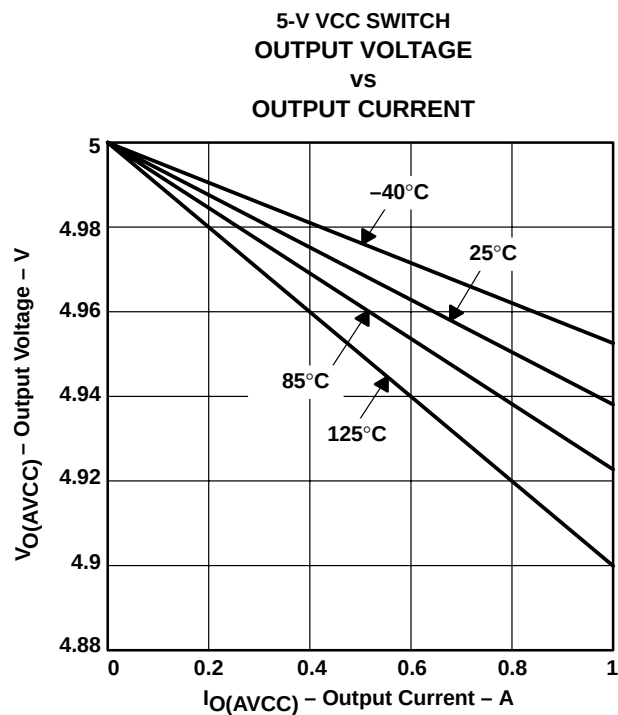


Figure 19

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TYPICAL CHARACTERISTICS

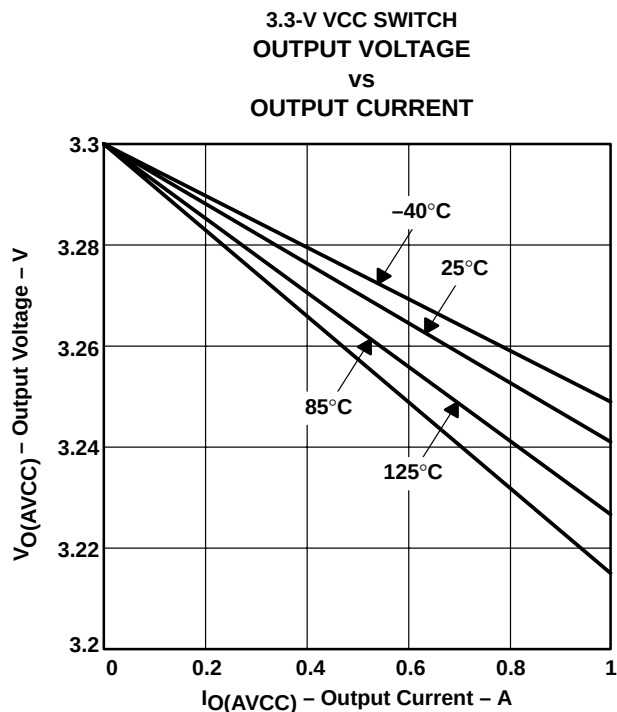


Figure 20

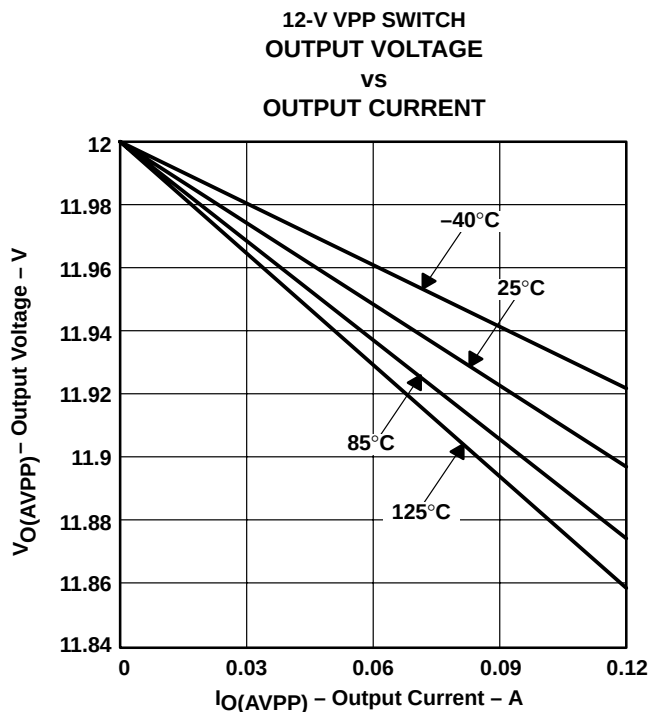


Figure 21

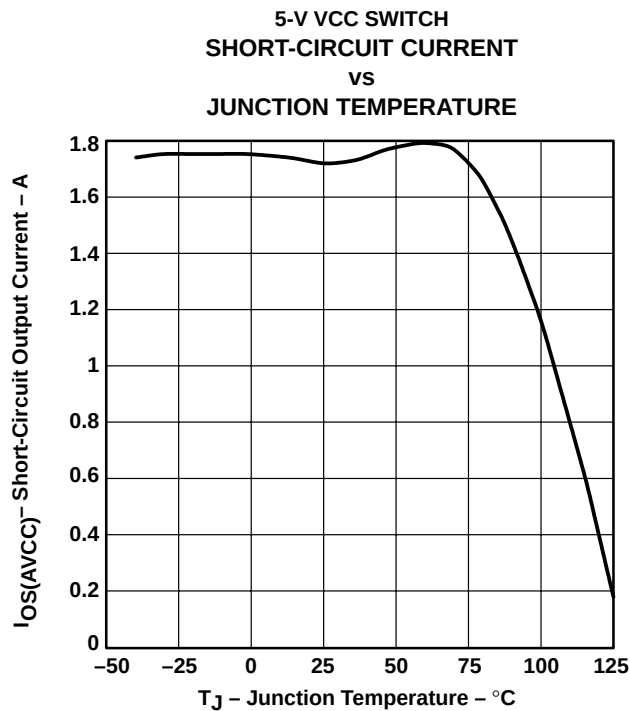


Figure 22

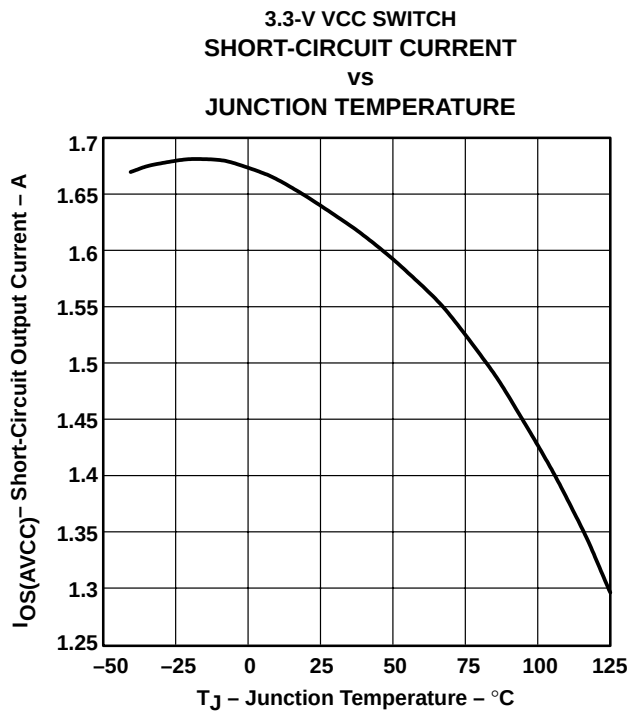


Figure 23

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TYPICAL CHARACTERISTICS

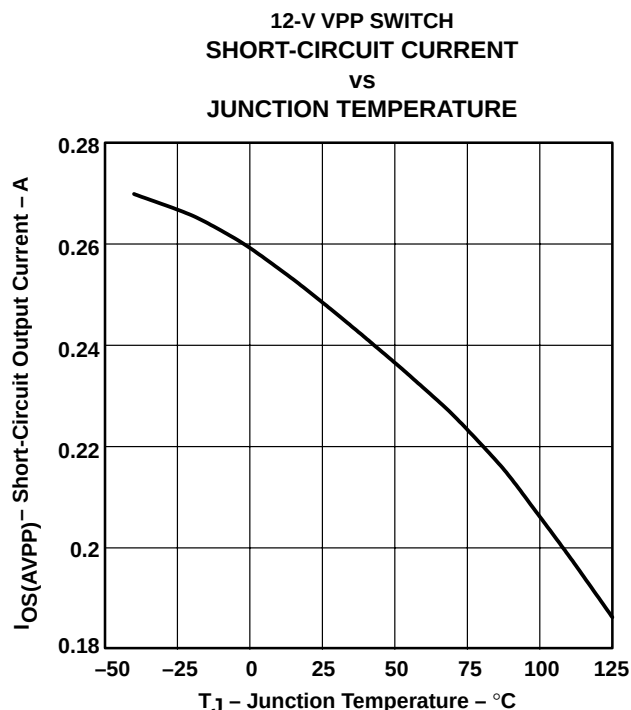


Figure 24

APPLICATION INFORMATION

overview

PC Cards were initially introduced as a means to add EEPROM (flash memory) to portable computers with limited onboard memory. The idea of add-in cards quickly took hold; modems, wireless LANs, GPS systems, multimedia, and hard-disk versions were soon available. As the number of PC Card applications grew, the engineering community quickly recognized the need for a standard to ensure compatibility across platforms. To this end, the PCMCIA (Personal Computer Memory Card International Association) was established, comprised of members from leading computer, software, PC Card, and semiconductor manufacturers. One key goal was to realize the *plug-and-play* concept, i.e. cards and hosts from different vendors should be compatible.

PC Card power specification

System compatibility also means power compatibility. The most current set of specifications (PC Card Standard) set forth by the PCMCIA committee states that power is to be transferred between the host and the card through eight of the 68 terminals of the PC Card connectors. This power interface consists of two V_{CC} , two V_{pp} , and four ground terminals. Multiple V_{CC} and ground terminals minimize connector-terminal and line resistance. The two V_{pp} terminals were originally specified as separate signals but are commonly tied together in the host to form a single node to minimize voltage losses. Card primary power is supplied through the V_{CC} terminals; flash-memory programming and erase voltage is supplied through the V_{pp} terminals.

APPLICATION INFORMATION

designing for voltage regulation

The current PCMCIA specification for output voltage regulation of the 5-V output is 5% (250 mV). In a typical PC power-system design, the power supply will have an output voltage regulation ($V_{PS(reg)}$) of 2% (100 mV). Also, a voltage drop from the power supply to the PC Card will result from resistive losses (V_{PCB}) in the PCB traces and the PCMCIA connector. A typical design would limit the total of these resistive losses to less than 1% (50 mV) of the output voltage. Therefore, the allowable voltage drop (V_{DS}) for the TPS2211 is the PCMCIA voltage regulation less the power supply regulation and less the PCB and connector resistive drops:

$$V_{DS} = V_{O(reg)} - V_{PS(reg)} - V_{PCB}$$

Typically, this would leave 100 mV for the allowable voltage drop across the TPS2211. The voltage drop is the output current multiplied by the switch resistance of the TPS2211. Therefore, the maximum output current that can be delivered to the PC Card in regulation is the allowable voltage drop across the TPS2211 divided by the output switch resistance.

$$I_{Omax} = \frac{V_{DS}}{r_{DS(on)}}$$

The AVCC outputs deliver 1 A continuous at 5 V and 3.3 V within regulation over the operating temperature range. Using the same equations, the PCMCIA specification for output voltage regulation of the 3.3-V output is 300 mV. Using the voltage drop percentages for power supply regulation (2%) and PCB resistive loss (1%), the allowable voltage drop for the 3.3-V switch is 200 mV. The 12-V outputs (AVPP) of the TPS2211 can deliver 150 mA continuously.

overcurrent and overtemperature protection

PC Cards are inherently subject to damage from mishandling. Host systems require protection against short-circuited cards that could lead to power supply or PCB trace damage. Even systems sufficiently robust to withstand a short circuit would still undergo rapid battery discharge into the damaged PC Card, resulting in a sudden loss of system power. Most hosts include fuses for protection. The reliability of fused systems is poor and requires troubleshooting and repair, usually by the manufacturer, when fuses are blown.

The TPS2211 uses sense FETs to check for overcurrent conditions in each of the AVCC and AVPP outputs. Unlike sense resistors or polyfuses, these FETs do not add to the series resistance of the switch; therefore voltage and power losses are reduced. Overcurrent sensing is applied to each output separately. When an overcurrent condition is detected, only the power output affected is limited; all other power outputs continue to function normally. The $\overline{OC}$ indicator, normally a logic high, is a logic low when an overcurrent condition is detected providing for initiation of system diagnostics and/or sending a warning message to the user.

During power up, the TPS2211 controls the rise time of the AVCC and AVPP outputs and limits the current into a faulty card or connector. If a short circuit is applied after power is established (e.g., hot insertion of a bad card), current is initially limited only by the impedance between the short and the power supply. In extreme cases, as much as 10 A to 15 A may flow into the short before the current limiting of the TPS2211 engages. If the AVCC or AVPP outputs are driven below ground, the TPS2211 may latch nondestructively in an off state. Cycling power will reestablish normal operation.

Overcurrent limiting for the AVCC outputs is designed to activate if powered up into a short in the range of 1 A to 2.2 A, typically at about 1.6 A. The AVPP outputs limit from 120 mA to 400 mA, typically around 280 mA. The protection circuitry acts by linearly limiting the current passing through the switch rather than initiating a full shutdown of the supply. Shutdown occurs only during thermal limiting.

Thermal limiting prevents destruction of the IC from overheating if the package power dissipation ratings are exceeded. Thermal limiting disables power output until the device has cooled.

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12-V supply not required

Most PC Card switches use the externally supplied 12 V to power gate drive and other chip functions, which require that power be present at all times. The TPS2211 offers considerable power savings by using an internal charge pump to generate the required higher voltages from the 5-V input. Therefore, the external 12-V supply can be disabled except when needed for flash-memory functions, thereby extending battery lifetime. Do not ground the 12-V switch inputs when the 12-V input is not used. Additional power savings are realized by the TPS2211 during a software shutdown in which quiescent current drops to a maximum of 1 μ A.

3.3-V low-voltage mode

The TPS2211 will operate in a 3.3-V low-voltage mode when 3.3 V is the only available input voltage ($V_{I(5V)} = 0$). This allows host and PC Cards to be operated in low-power 3.3-volts-only modes such as sleep or pager modes. Note that in these operation modes, the TPS2211 will derive its bias current from the 3.3-V input pin and only 3.3 V can be delivered to the PC Card.

voltage transitioning requirement

PC Cards are migrating from 5 V to 3.3 V to minimize power consumption, optimize board space, and increase logic speeds. The TPS2211 meets all combinations of power delivery as currently defined in the PCMCIA standard. The latest protocol accommodates mixed 3.3-V/5-V systems by first powering the card with 5 V, then polling it to determine its 3.3-V compatibility. The PCMCIA specification requires that the capacitors on 3.3-V compatible cards be discharged to below 0.8 V before applying 3.3-V power. This functions as a power reset and ensures that sensitive 3.3-V circuitry is not subjected to any residual 5-V charge. The TPS2211 offers a selectable V_{CC} and V_{pp} ground state, in accordance with PCMCIA 3.3-V/5-V switching specifications.

output ground switches

PC Card specification requires that V_{CC} be discharged within 100 ms. PC Card resistance can not be relied on to provide a discharge path for voltages stored on PC Card capacitance because of possible high-impedance isolation by power-management schemes.

power-supply considerations

The TPS2211 has multiple pins for each of its 3.3-V and 5-V power inputs and for the switched AVCC outputs. Any individual pin can conduct the rated input or output current. Unless all pins are connected in parallel, the series resistance is significantly higher than that specified, resulting in increased voltage drops and lost power. It is recommended that all input and output power pins be paralleled for optimum operation.

To increase the noise immunity of the TPS2211, the power supply inputs should be bypassed with a 1- μ F electrolytic or tantalum capacitor paralleled by a 0.047- μ F to 0.1- μ F ceramic capacitor. It is strongly recommended that the switched outputs be bypassed with a 0.1- μ F, or larger, ceramic capacitor; doing so improves the immunity of the TPS2211 to electrostatic discharge (ESD). Care should be taken to minimize the inductance of PCB traces between the TPS2211 and the load. High switching currents can produce large negative voltage transients, which forward biases substrate diodes, resulting in unpredictable performance. Similarly, no pin should be taken below -0.3 V.



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calculating junction temperature

The switch resistance, $r_{DS(on)}$, is dependent on the junction temperature, T_J , of the die and the current through the switch. To calculate T_J , first find $r_{DS(on)}$ from Figures 16 through 18 using an initial temperature estimate about 50°C above ambient. Then calculate the power dissipation for each switch, using the formula:

$$P_D = r_{DS(on)} \times I^2$$

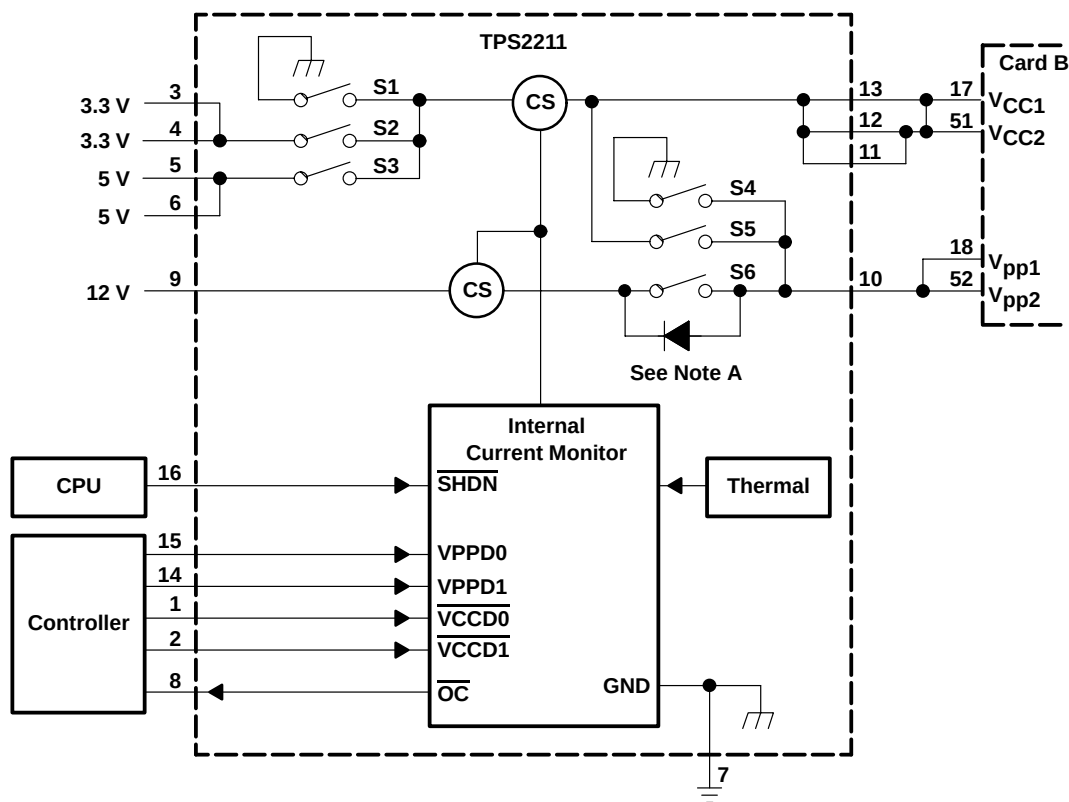
Next, sum the power dissipation and calculate the junction temperature:

$$T_J = \left(\sum P_D \times R_{\theta JA} \right) + T_A, \quad R_{\theta JA} = 108^\circ\text{C/W}$$

Compare the calculated junction temperature with the initial temperature estimate. If the temperatures are not within a few degrees of each other, recalculate using the calculated temperature as the initial estimate.

ESD protection

All TPS2211 inputs and outputs incorporate ESD-protection circuitry designed to withstand a 2-kV human-body-model discharge as defined in MIL-STD-883C, Method 3015. The AVCC and AVPP outputs can be exposed to potentially higher discharges from the external environment through the PC Card connector. Bypassing the outputs with 0.1-μF capacitors protects the devices from discharges up to 10 kV.



NOTE A: MOSFET switch S6 has a back-gate diode from the source to the drain. Unused switch inputs should never be grounded.

Figure 25. Internal Switching Matrix, TPS2211 Control Logic

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TPS2211 control logic

AVPP

CONTROL SIGNALS			INTERNAL SWITCH SETTINGS			OUTPUT
$\overline{\text{SHDN}}$	VPPD0	VPPD1	S4	S5	S6	AVPP
1	0	0	CLOSED	OPEN	OPEN	0 V
1	0	1	OPEN	CLOSED	OPEN	AVCC [†]
1	1	0	OPEN	OPEN	CLOSED	VPP (12 V)
1	1	1	OPEN	OPEN	OPEN	Hi-Z
0	X	X	OPEN	OPEN	OPEN	Hi-Z

[†] Output depends on AVCC

AVCC

CONTROL SIGNALS			INTERNAL SWITCH SETTINGS			OUTPUT
$\overline{\text{SHDN}}$	$\overline{\text{VCCD1}}$	$\overline{\text{VCCD0}}$	S1	S2	S3	AVCC
1	0	0	CLOSED	OPEN	OPEN	0 V
1	0	1	OPEN	CLOSED	OPEN	3.3 V
1	1	0	OPEN	OPEN	CLOSED	5 V
1	1	1	CLOSED	OPEN	OPEN	0 V
0	X	X	OPEN	OPEN	OPEN	Hi-Z

12-V flash memory supply

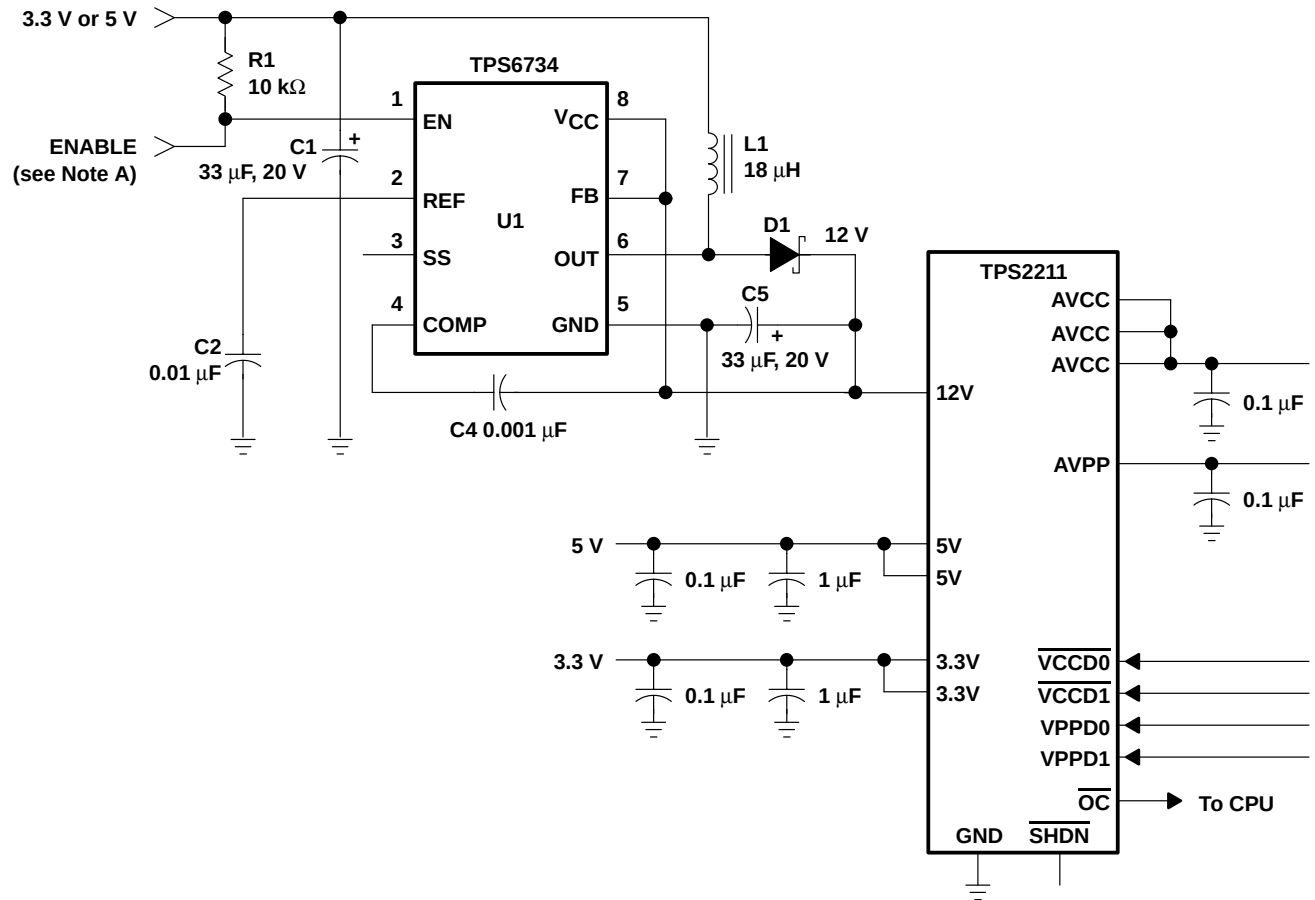
The TPS6734 is a fixed 12-V output boost converter capable of delivering 120 mA from inputs as low as 2.7 V. The device is pin-for-pin compatible with the MAX734 regulator and offers the following advantages: lower supply current, wider operating input-voltage range, and higher output currents. As shown in Figure 1, the only external components required are: an inductor, a Schottky rectifier, an output filter capacitor, an input filter capacitor, and a small capacitor for loop compensation. The entire converter occupies less than 0.7 in² of PCB space when implemented with surface-mount components. An enable input is provided to shut the converter down and reduce the supply current to 3 μ A when 12 V is not needed.

The TPS6734 is a 170-kHz current-mode PWM (pulse-width modulation) controller with an n-channel MOSFET power switch. Gate drive for the switch is derived from the 12-V output after start-up to minimize the die area needed to realize the 0.7- Ω MOSFET and improve efficiency at input voltages below 5 V. Soft start is accomplished with the addition of one small capacitor. A 1.22-V reference (pin 2) is brought out for external use. For additional information, see the TPS6734 data sheet (SLVS127).

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NOTE A: The enable terminal can be tied to a general-purpose I/O terminal on the PCMCIA controller or tied high.

Figure 26. TPS2211 With TPS6734 12-V, 120-mA Supply

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
TPS2211IDB	Active	Production	SSOP (DB) 16	80 TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	-	PU2211
TPS2211IDB.A	Active	Production	SSOP (DB) 16	80 TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	PU2211
TPS2211IDBG4	Active	Production	SSOP (DB) 16	80 TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	See TPS2211IDB	PU2211
TPS2211IDBR	Active	Production	SSOP (DB) 16	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	PU2211
TPS2211IDBR.A	Active	Production	SSOP (DB) 16	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	PU2211

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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TAPE AND REEL INFORMATION



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS2211IDBR	SSOP	DB	16	2000	330.0	16.4	8.35	6.6	2.4	12.0	16.0	Q1

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS2211IDBR	SSOP	DB	16	2000	353.0	353.0	32.0

TUBE



*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
TPS2211IDB	DB	SSOP	16	80	530	10.5	4000	4.1
TPS2211IDB.A	DB	SSOP	16	80	530	10.5	4000	4.1
TPS2211IDBG4	DB	SSOP	16	80	530	10.5	4000	4.1

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