

# OPAx171 36V, Single-Supply, SOT-553, General-Purpose Operational Amplifiers

## 1 Features

- Supply range: 2.7V to 36V,  $\pm 1.35V$  to  $\pm 18V$
- Low noise:  $14nV/\sqrt{Hz}$
- Low offset drift:  $\pm 0.3\mu V/^\circ C$  (Typical)
- RFI filtered inputs
- Input range includes the negative supply
- Input range operates to positive supply
- Rail-to-rail output
- Gain bandwidth: 3MHz
- Low quiescent current: 475 $\mu A$  per amplifier
- High common-mode rejection: 120dB (typical)
- Low-input bias current: 8pA
- Industry-standard packages:
  - 5-pin SOT-23
  - 8-pin SOIC
  - 14-pin TSSOP
- *micro*Packages:
  - Single in SOT-553
  - Dual in VSSOP-8

## 2 Applications

- [Tracking amplifier in power modules](#)
- [Merchant power supplies](#)
- [Transducer amplifiers](#)
- [Bridge amplifiers](#)
- [Temperature measurements](#)
- [Strain gauge amplifiers](#)
- [Precision integrators](#)
- [Battery-powered instruments](#)
- [Test equipment](#)

## 3 Description

The OPA171, OPA2171, and OPA4171 (OPAx171) are a family of 36V, single-supply, low-noise operational amplifiers with the ability to operate on supplies ranging from 2.7V ( $\pm 1.35V$ ) to 36V ( $\pm 18V$ ). These devices are available in micro-packages and offer low offset, drift, and bandwidth with low quiescent current. The single, dual, and quad versions all have identical specifications for maximum design flexibility.

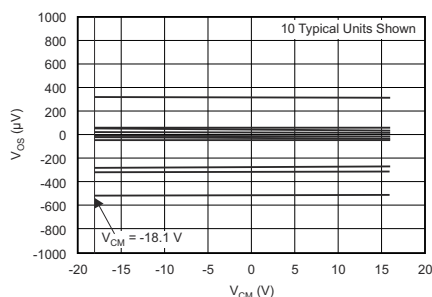
Unlike most operational amplifiers, which are specified at only one supply voltage, the OPAx171 family is specified from 2.7V to 36V. Input signals beyond the supply rails do not cause phase reversal. The OPAx171 family is stable with capacitive loads up to 300pF. The input can operate 100mV below the negative rail and within 2V of the top rail during normal operation. These devices can operate with full rail-to-rail input 100mV beyond the top rail, but with reduced performance within 2V of the top rail.

The OPAx171 series of operational amplifiers are specified from  $-40^\circ C$  to  $+125^\circ C$ .

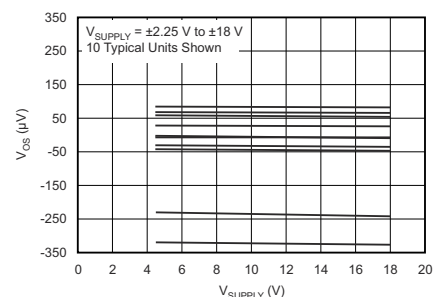
### Package Information

PART NUMBER	CHANNEL COUNT	PACKAGE <sup>(1)</sup>	PACKAGE SIZE <sup>(2)</sup>
OPA171	Single	D (SOIC, 8)	4.9mm × 6mm
		DBV (SOT-23, 5)	2.9mm × 2.8mm
		DRL (SOT-5X3, 5)	1.6mm × 1.6mm
OPA2171	Dual	D (SOIC, 8)	4.9mm × 6mm
		DCU (VSSOP, 8)	2mm × 3.1mm
		DGK (VSSOP, 8)	3mm × 4.9mm
OPA4171	Quad	PW (TSSOP, 14)	5mm × 6.4mm
		D (SOIC, 14)	8.65mm × 6mm

- (1) For all available packages, see the orderable addendum at the end of the data sheet.
- (2) The package size (length × width) is a nominal value and includes pins, where applicable.



Offset Voltage vs Common-Mode Voltage



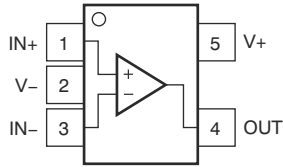
Offset Voltage vs Power Supply



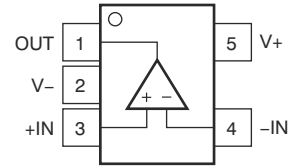
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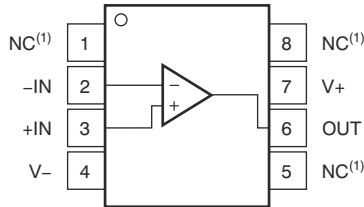
## 4 Pin Configuration and Functions



**Figure 4-1. OPA171 DRL Package: 5-Pin SOT-553 (Top View)**



**Figure 4-2. OPA171 DBV Package: 5-Pin SOT-23 (Top View)**

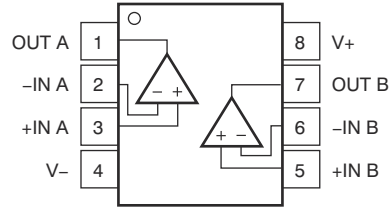


A. NC- no internal connection

**Figure 4-3. OPA171 D Package: 8-Pin SOIC (Top View)**

### Pin Functions: OPA171

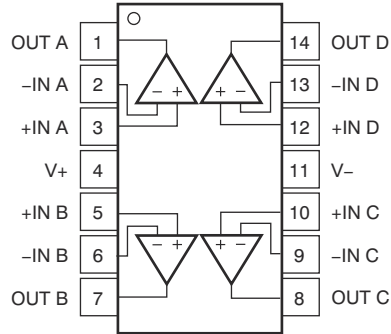
NAME	PIN			TYPE	DESCRIPTION
	DRL	DBV	D		
+IN	1	3	3	I	Noninverting input
-IN	3	4	2	I	Inverting input
OUT	4	1	6	O	Output
V+	5	5	7	—	Positive (highest) supply
V-	2	2	4	—	Negative (lowest) supply
NC	—	—	1, 5, 8	—	No internal connection (can be left floating)



**Figure 4-4. OPA2171 D, DCU, and DCK Packages: 8-Pin SO and VSSOP (Top View)**

**Table 4-1. Pin Functions: OPA2171**

PIN		TYPE	DESCRIPTION
NAME	NO.		
+IN A	3	I	Noninverting input
+IN B	5	I	Noninverting input
-IN A	2	I	Inverting input
-IN B	6	O	Inverting input
OUT A	1	O	Output
OUT B	7	—	Output
V+	8	—	Positive (highest) supply
V-	4	—	Negative (lowest) supply



**Figure 4-5. OPA4171 D and PW Packages: 14-Pin SO and TSSOP (Top View)**

**Table 4-2. Pin Functions: OPA4171**

PIN		TYPE	DESCRIPTION
NAME	NO.		
+IN A	3	I	Noninverting input
+IN B	5	I	Noninverting input
+IN C	10	I	Noninverting input
+IN D	12	I	Noninverting input
-IN A	2	I	Inverting input
-IN B	6	I	Inverting input
-IN C	9	I	Inverting input
-IN D	13	I	Inverting input
OUT A	1	O	Output
OUT B	7	O	Output
OUT C	8	O	Output
OUT D	14	O	Output
V+	4	—	Positive (highest) supply
V-	11	—	Negative (lowest) supply

## 5 Specifications

### 5.1 Absolute Maximum Ratings

over operating free-air temperature range, (unless otherwise noted) <sup>(1)</sup>

		MIN	MAX	UNIT
Supply voltage		0	40	V
Signal input terminals	Voltage	(V <sub>-</sub> ) – 0.5	(V <sub>+</sub> ) + 0.5	V
	Current	–10	10	mA
Output short circuit <sup>(2)</sup>		Continuous		
Operating temperature		–55	150	°C
Junction temperature			150	°C
Storage temperature		–65	150	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) Short-circuit to ground, one amplifier per package.

### 5.2 ESD Ratings

		VALUE	UNIT
V <sub>(ESD)</sub>	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±4000
		Charged-device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>	±750

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### 5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	NOM	MAX	UNIT
Supply voltage (V <sub>+</sub> – V <sub>-</sub> )	4.5 (±2.25)		36 (±18)	V
Specified temperature	–40		125	°C

## 5.4 Thermal Information: OPA171

THERMAL METRIC <sup>(1)</sup>		OPA171			UNIT
		D (SO)	DBV (SOT-23)	DRL (SOT-553)	
		8 PINS	5 PINS	5 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	149.5	245.8	208.1	°C/W
R <sub>θJC(top)</sub>	Junction-to-case(top) thermal resistance	97.9	133.9	0.1	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	87.7	83.6	42.4	°C/W
ψ <sub>JT</sub>	Junction-to-top characterization parameter	35.5	18.2	0.5	°C/W
ψ <sub>JB</sub>	Junction-to-board characterization parameter	89.5	83.1	42.2	°C/W
R <sub>θJC(bot)</sub>	Junction-to-case(bottom) thermal resistance	N/A	N/A	N/A	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

## 5.5 Thermal Information: OPA2171

THERMAL METRIC <sup>(1)</sup>		OPA2171			UNIT
		D (SO)	DGK (VSSOP)	DCU (VSSOP)	
		8 PINS	8 PINS	8 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	134.3	175.2	195.3	°C/W
R <sub>θJC(top)</sub>	Junction-to-case(top) thermal resistance	72.1	74.9	59.4	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	60.6	22.2	115.1	°C/W
ψ <sub>JT</sub>	Junction-to-top characterization parameter	18.2	1.6	4.7	°C/W
ψ <sub>JB</sub>	Junction-to-board characterization parameter	53.8	22.8	114.4	°C/W
R <sub>θJC(bot)</sub>	Junction-to-case(bottom) thermal resistance	N/A	N/A	N/A	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report .

## 5.6 Thermal Information: OPA4171

THERMAL METRIC <sup>(1)</sup>		OPA4171		UNIT
		D (SOIC)	PW (TSSOP)	
		14 PINS	14 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	93.2	106.9	°C/W
R <sub>θJC(top)</sub>	Junction-to-case(top) thermal resistance	51.8	24.4	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	49.4	59.3	°C/W
ψ <sub>JT</sub>	Junction-to-top characterization parameter	13.5	0.6	°C/W
ψ <sub>JB</sub>	Junction-to-board characterization parameter	42.2	54.3	°C/W
R <sub>θJC(bot)</sub>	Junction-to-case(bottom) thermal resistance	N/A	N/A	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

## 5.7 Electrical Characteristics

at  $T_A = 25^\circ\text{C}$ ,  $V_S = 2.7$  to  $36\text{ V}$ ,  $V_{CM} = V_{OUT} = V_S / 2$ , and  $R_{LOAD} = 10\text{ k}\Omega$  connected to  $V_S / 2$ , (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
OFFSET VOLTAGE						
$V_{OS}$	Input offset voltage			0.25	$\pm 1.8$	mV
	Over temperature	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$		0.3	$\pm 2$	mV
$dV_{OS}/dT$	Drift	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$		0.3	$\pm 2$	$\mu\text{V}/^\circ\text{C}$
	vs power supply	$V_S = 4$ to $36\text{ V}$ $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$		1	$\pm 3$	$\mu\text{V}/\text{V}$
	Channel separation, DC	DC		5		$\mu\text{V}/\text{V}$
INPUT BIAS CURRENT						
$I_B$	Input bias current			$\pm 8$	$\pm 15$	pA
	Over temperature	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			$\pm 3.5$	nA
$I_{OS}$	Input offset current			$\pm 4$		pA
	Over temperature	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			$\pm 3.5$	nA
NOISE						
	Input voltage noise	$f = 0.1\text{ Hz}$ to $10\text{ Hz}$		3		$\mu\text{V}_{PP}$
$e_n$	Input voltage noise density	$f = 100\text{ Hz}$		25		$\text{nV}/\sqrt{\text{Hz}}$
		$f = 1\text{ kHz}$		14		$\text{nV}/\sqrt{\text{Hz}}$
INPUT VOLTAGE						
$V_{CM}$	Common-mode voltage range		$(V-) - 0.1\text{ V}$		$(V+) - 2\text{ V}$	V
CMRR	Common-mode rejection ratio	$V_S = \pm 2\text{ V}$ $(V-) - 0.1\text{ V} < V_{CM} < (V+) - 2\text{ V}$ $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	90	104		dB
		$V_S = \pm 18\text{ V}$ $(V-) - 0.1\text{ V} < V_{CM} < (V+) - 2\text{ V}$ $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	104	120		dB
INPUT IMPEDANCE						
	Differential			100    3		$\text{M}\Omega$    pF
	Common-mode			6    3		$10^{12}\Omega$    pF
OPEN-LOOP GAIN						
$A_{OL}$	Open-loop voltage gain	$V_S = 4\text{ V}$ to $36\text{ V}$ $(V-) + 0.35\text{ V} < V_O < (V+) - 0.35\text{ V}$ $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	110	130		dB
FREQUENCY RESPONSE						
GBP	Gain bandwidth product			3		MHz
SR	Slew rate	$G = 1$		1.5		V/ $\mu\text{s}$
$t_s$	Settling time	To 0.1% $V_S = \pm 18\text{ V}$ , $G = 1$ 10-V step		6		$\mu\text{s}$
		To 0.01% (12 bit) $V_S = \pm 18\text{ V}$ , $G = 1$ 10-V step		10		$\mu\text{s}$
	Overload recovery time	$V_{IN} \times \text{gain} > V_S$		2		$\mu\text{s}$
THD+N	Total harmonic distortion + noise	$G = 1$ , $f = 1\text{ kHz}$ $V_O = 3\text{ V}_{RMS}$		0.0002%		
OUTPUT						



at  $T_A = 25^\circ\text{C}$ ,  $V_S = 2.7$  to  $36\text{ V}$ ,  $V_{CM} = V_{OUT} = V_S / 2$ , and  $R_{LOAD} = 10\text{ k}\Omega$  connected to  $V_S / 2$ , (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_O$	Voltage output swing from rail	$V_S = 5\text{ V}$ $R_L = 10\text{ k}\Omega$		30		mV
	Over temperature	$R_L = 10\text{ k}\Omega$ $A_{OL} \geq 110\text{ dB}$ $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	$(V-) + 0.35$	$(V+) - 0.35$		V
$I_{SC}$	Short-circuit current			+25/-35		mA
$C_{LOAD}$	Capacitive load drive			See <a href="#">Section 5.9</a>		pF
$R_O$	Open-loop output resistance	$f = 1\text{ MHz}$ $I_O = 0\text{ A}$		150		$\Omega$
POWER SUPPLY						
$V_S$	Specified voltage range		2.7		36	V
$I_Q$	Quiescent current per amplifier	$I_O = 0\text{ A}$		475	595	$\mu\text{A}$
	Over temperature	$I_O = 0\text{ A}$ $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			650	$\mu\text{A}$
TEMPERATURE						
	Specified range		-40		125	$^\circ\text{C}$
	Operating range		-55		150	$^\circ\text{C}$

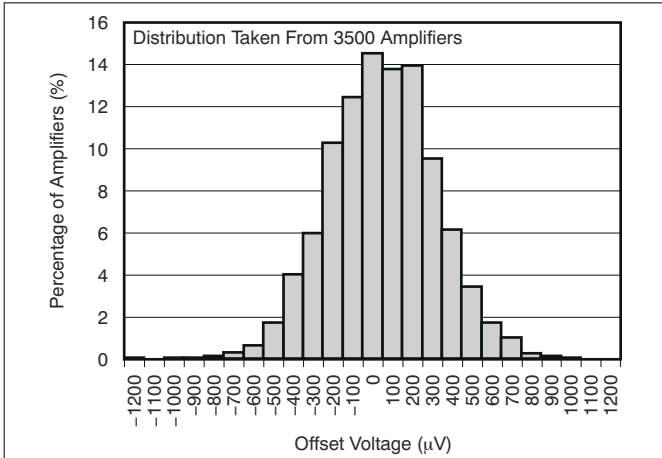
## 5.8 Typical Characteristics: Table of Graphs

**Table 5-1. Characteristic Performance Measurements**

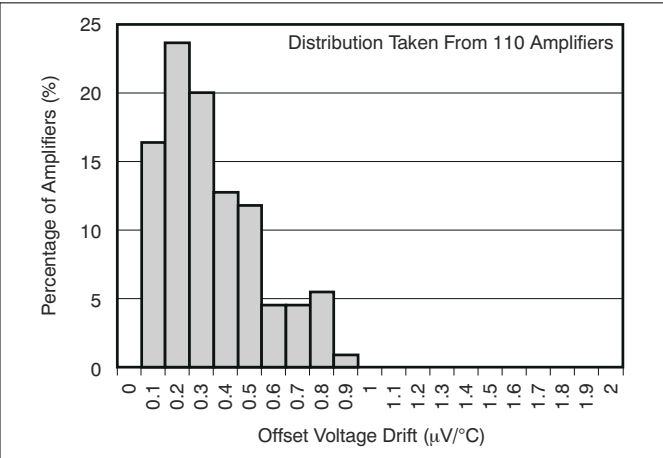
DESCRIPTION	FIGURE
Offset Voltage Production Distribution	<a href="#">Figure 5-1</a>
Offset Voltage Drift Distribution	<a href="#">Figure 5-2</a>
Offset Voltage vs Temperature	<a href="#">Figure 5-3</a>
Offset Voltage vs Common-Mode Voltage	<a href="#">Figure 5-4</a>
Offset Voltage vs Common-Mode Voltage (Upper Stage)	<a href="#">Figure 5-5</a>
Offset Voltage vs Power Supply	<a href="#">Figure 5-6</a>
$I_B$ and $I_{OS}$ vs Common-Mode Voltage	<a href="#">Figure 5-7</a>
Input Bias Current vs Temperature	<a href="#">Figure 5-8</a>
Output Voltage Swing vs Output Current (Maximum Supply)	<a href="#">Figure 5-9</a>
CMRR and PSRR vs Frequency (Referred-to Input)	<a href="#">Figure 5-10</a>
CMRR vs Temperature	<a href="#">Figure 5-11</a>
PSRR vs Temperature	<a href="#">Figure 5-12</a>
0.1-Hz to 10-Hz Noise	<a href="#">Figure 5-13</a>
Input Voltage Noise Spectral Density vs Frequency	<a href="#">Figure 5-14</a>
THD+N Ratio vs Frequency	<a href="#">Figure 5-15</a>
THD+N vs Output Amplitude	<a href="#">Figure 5-16</a>
Quiescent Current vs Temperature	<a href="#">Figure 5-17</a>
Quiescent Current vs Supply Voltage	<a href="#">Figure 5-18</a>
Open-Loop Gain and Phase vs Frequency	<a href="#">Figure 5-19</a>
Closed-Loop Gain vs Frequency	<a href="#">Figure 5-20</a>
Open-Loop Gain vs Temperature	<a href="#">Figure 5-21</a>
Open-Loop Output Impedance vs Frequency	<a href="#">Figure 5-22</a>
Small-Signal Overshoot vs Capacitive Load (100-mV Output Step)	<a href="#">Figure 5-23, Figure 5-24</a>
No Phase Reversal	<a href="#">Figure 5-25</a>
Positive Overload Recovery	<a href="#">Figure 5-26</a>
Negative Overload Recovery	<a href="#">Figure 5-27</a>
Small-Signal Step Response (100 mV)	<a href="#">Figure 5-28, Figure 5-29</a>
Large-Signal Step Response	<a href="#">Figure 5-30, Figure 5-31</a>
Large-Signal Settling Time (10-V Positive Step)	<a href="#">Figure 5-32</a>
Large-Signal Settling Time (10-V Negative Step)	<a href="#">Figure 5-33</a>
Short-Circuit Current vs Temperature	<a href="#">Figure 5-34</a>
Maximum Output Voltage vs Frequency	<a href="#">Figure 5-35</a>
Channel Separation vs Frequency	<a href="#">Figure 5-36</a>

### 5.9 Typical Characteristics

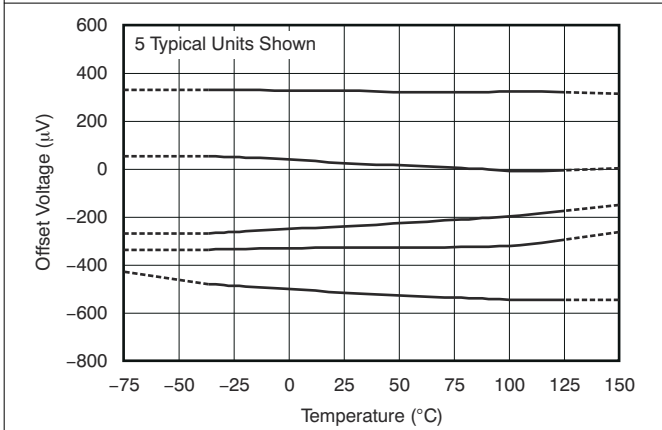
$V_S = \pm 18\text{ V}$ ,  $V_{CM} = V_S / 2$ ,  $R_{LOAD} = 10\text{ k}\Omega$  connected to  $V_S / 2$ , and  $C_L = 100\text{ pF}$ , (unless otherwise noted)



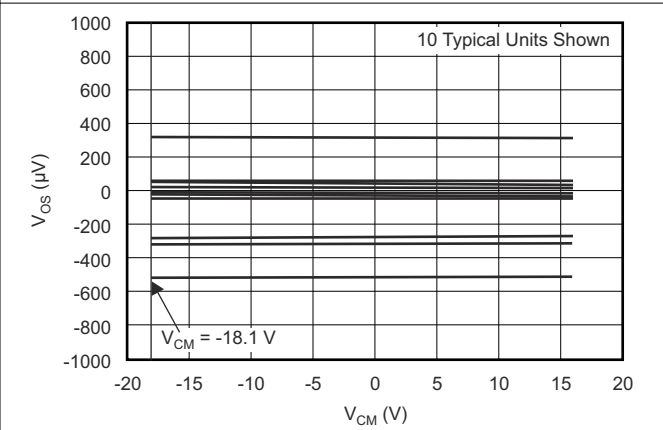
**Figure 5-1. Offset Voltage Production Distribution**



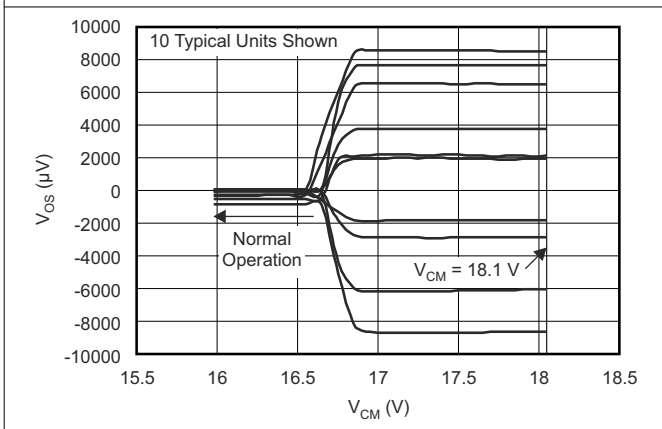
**Figure 5-2. Offset Voltage Drift Distribution**



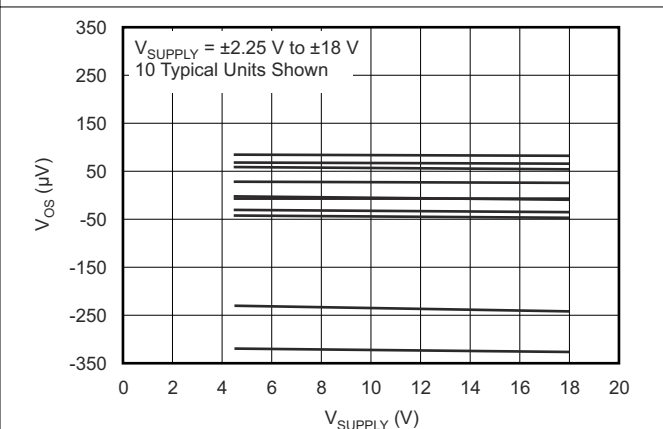
**Figure 5-3. Offset Voltage vs Temperature**



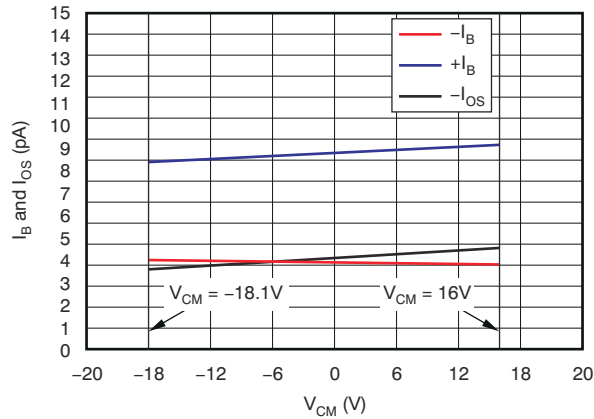
**Figure 5-4. Offset Voltage vs Common-Mode Voltage**



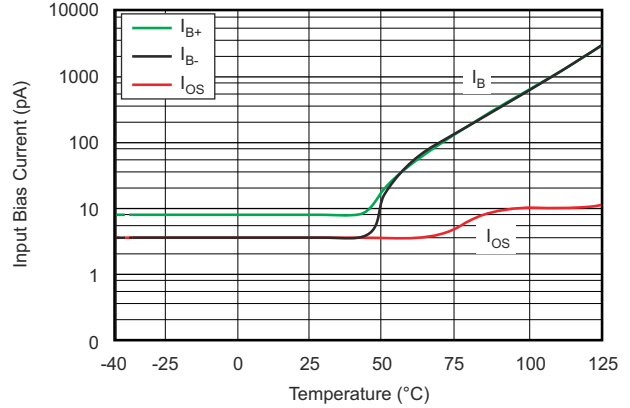
**Figure 5-5. Offset Voltage vs Common-Mode Voltage (Upper Stage)**



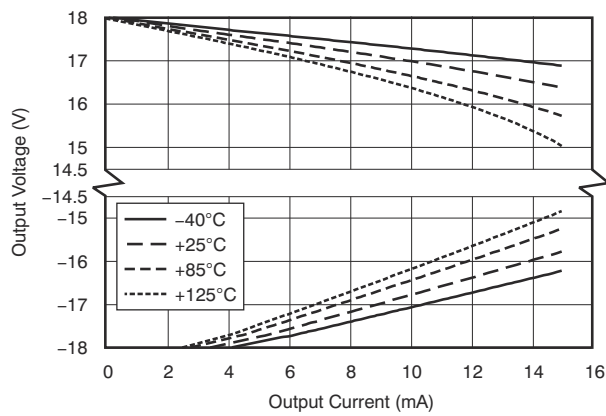
**Figure 5-6. Offset Voltage vs Power Supply**



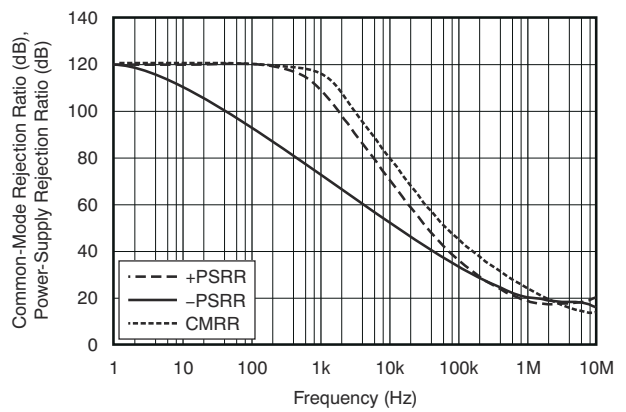
**Figure 5-7.  $I_B$  and  $I_{OS}$  vs Common-Mode Voltage**



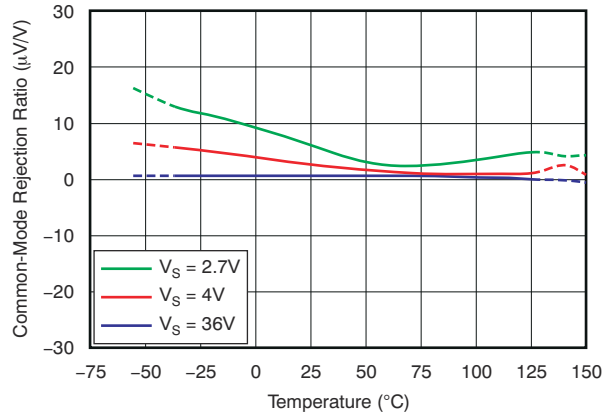
**Figure 5-8. Input Bias Current vs Temperature**



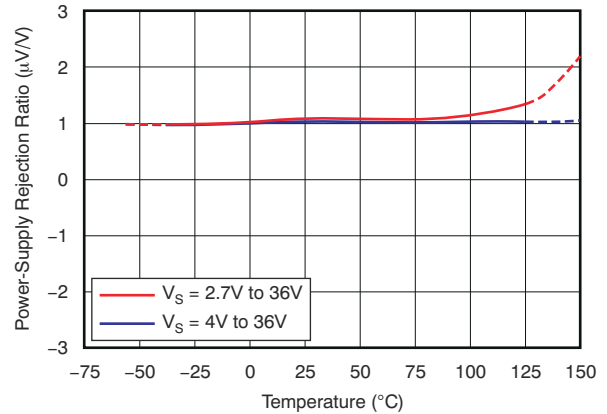
**Figure 5-9. Output Voltage Swing vs Output Current (Maximum Supply)**



**Figure 5-10. CMRR and PSRR vs Frequency (Referred-to Input)**



**Figure 5-11. CMRR vs Temperature**



**Figure 5-12. PSRR vs Temperature**

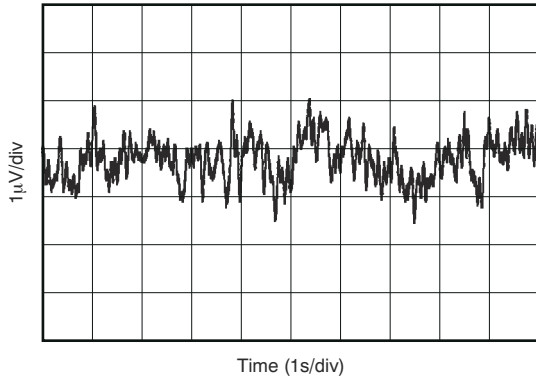


Figure 5-13. 0.1-Hz to 10-Hz Noise

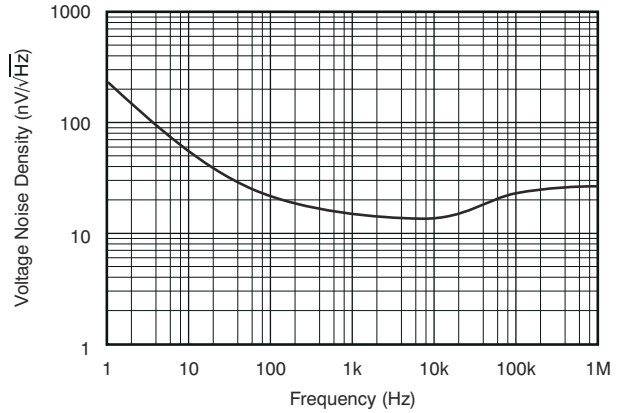


Figure 5-14. Input Voltage Noise Spectral Density vs Frequency

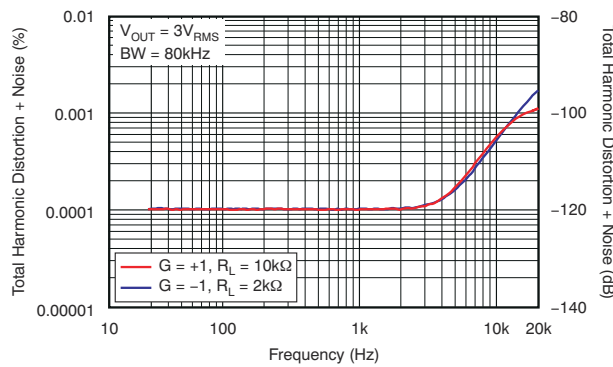


Figure 5-15. THD+N Ratio vs Frequency

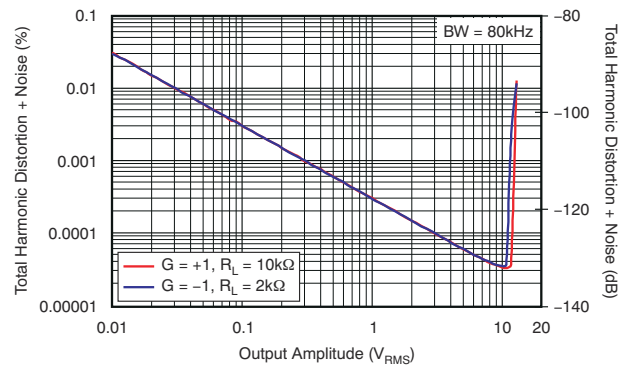


Figure 5-16. THD+N vs Output Amplitude

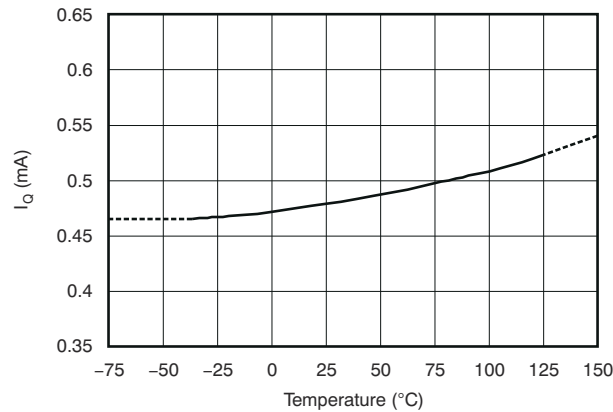


Figure 5-17. Quiescent Current vs Temperature

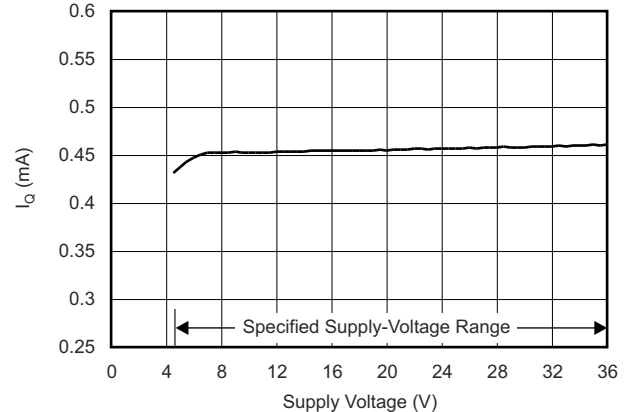
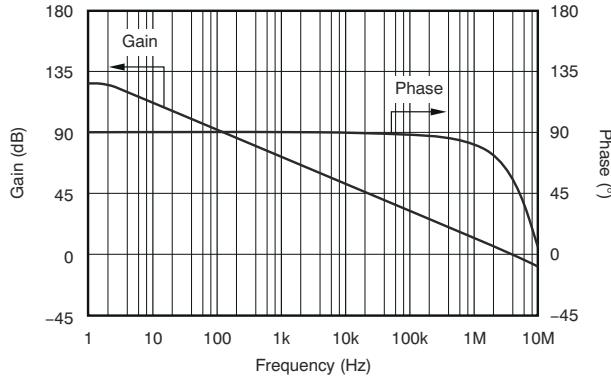
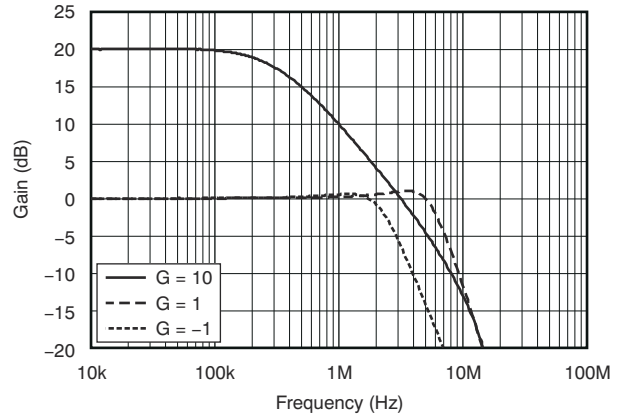


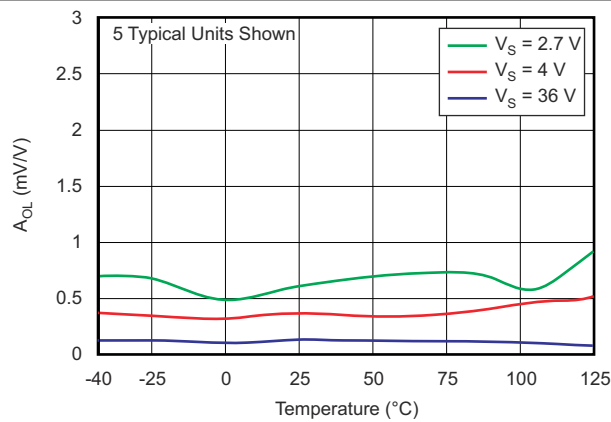
Figure 5-18. Quiescent Current vs Supply Voltage



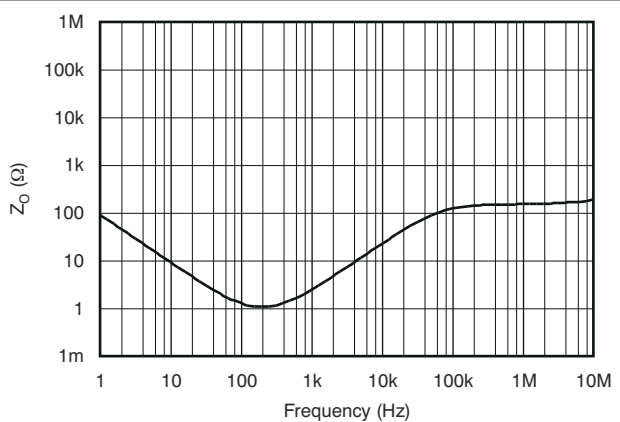
**Figure 5-19. Open-Loop Gain and Phase vs Frequency**



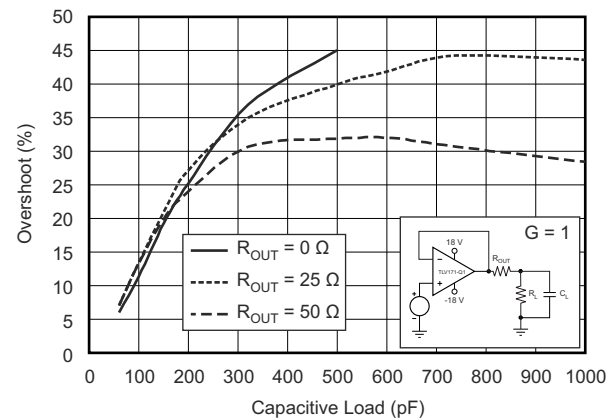
**Figure 5-20. Closed-Loop Gain vs Frequency**



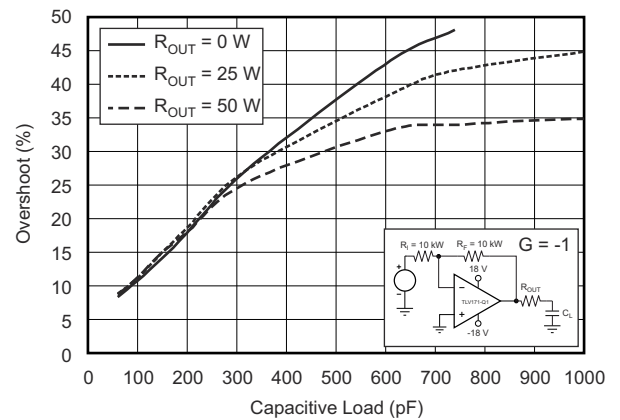
**Figure 5-21. Open-Loop Gain vs Temperature**



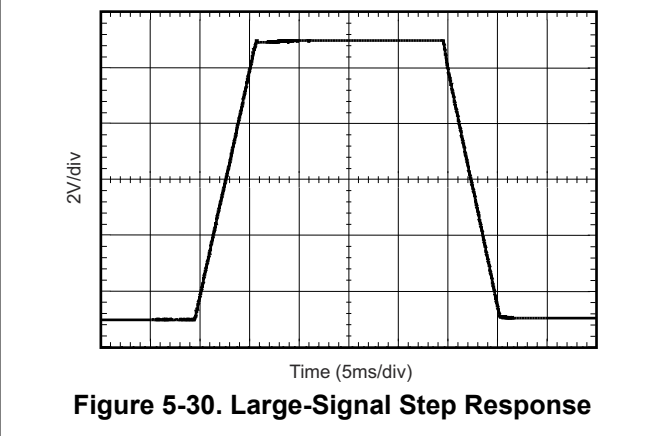
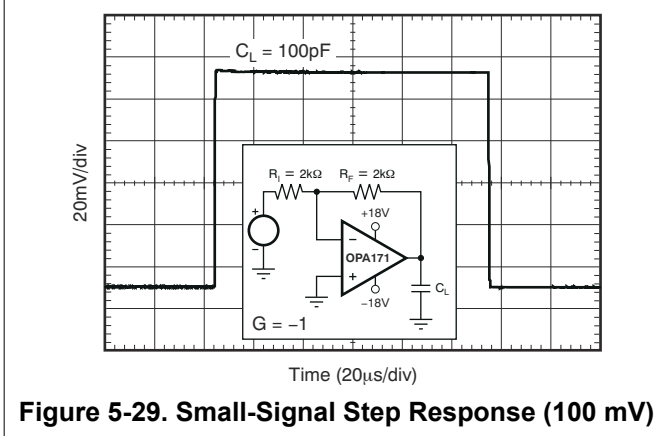
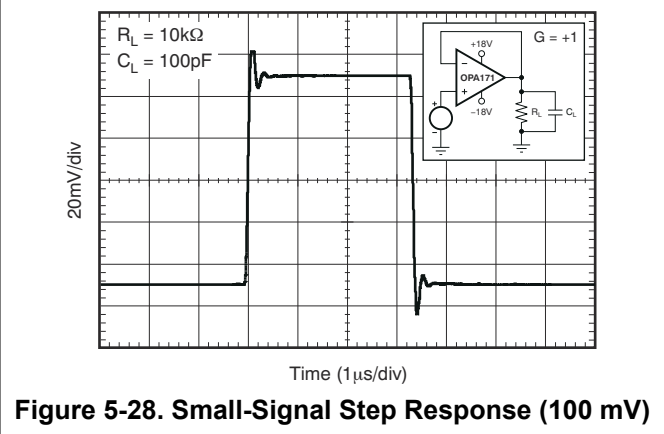
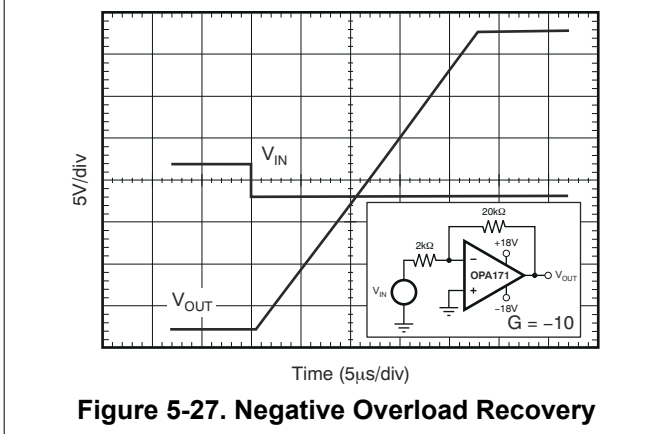
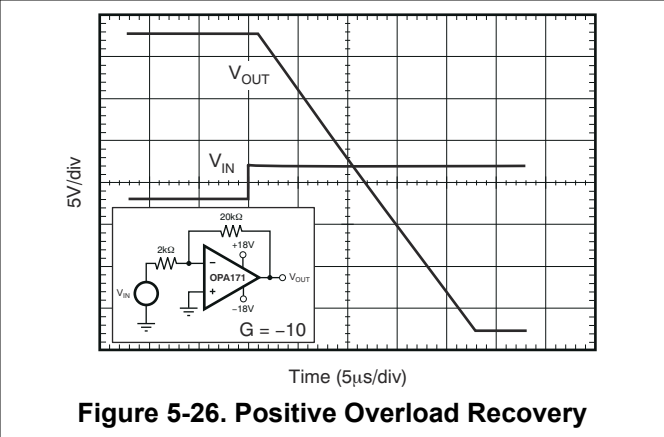
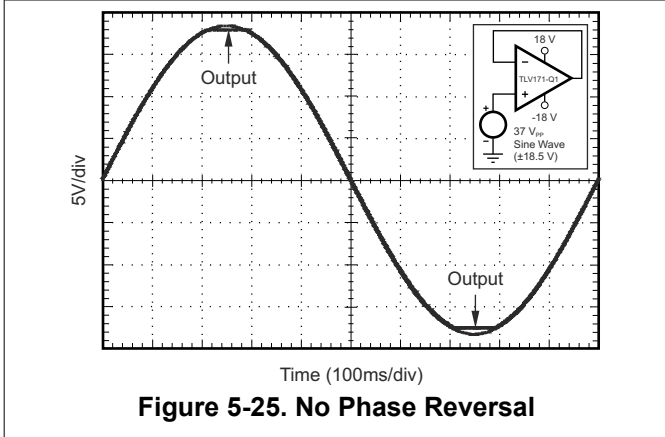
**Figure 5-22. Open-Loop Output Impedance vs Frequency**

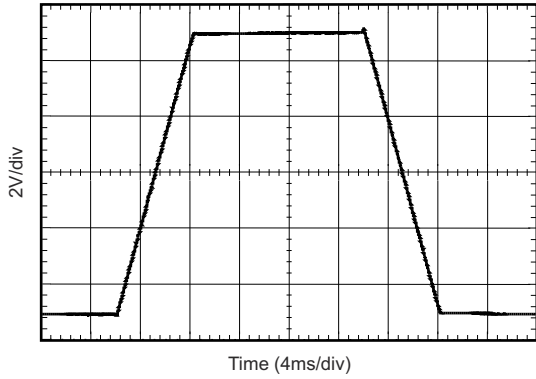


**Figure 5-23. Small-Signal Overshoot vs Capacitive Load (100-mV Output Step)**

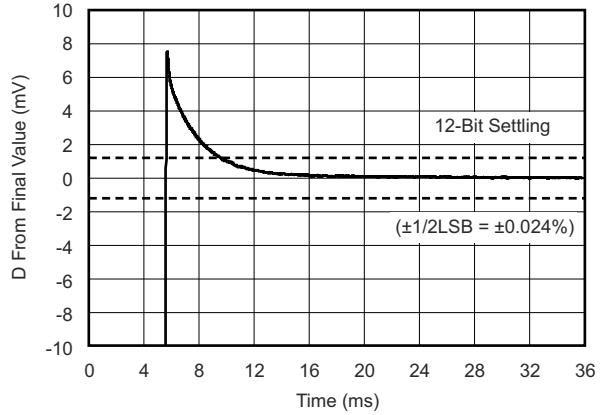


**Figure 5-24. Small-Signal Overshoot vs Capacitive Load (100-mV Output Step)**

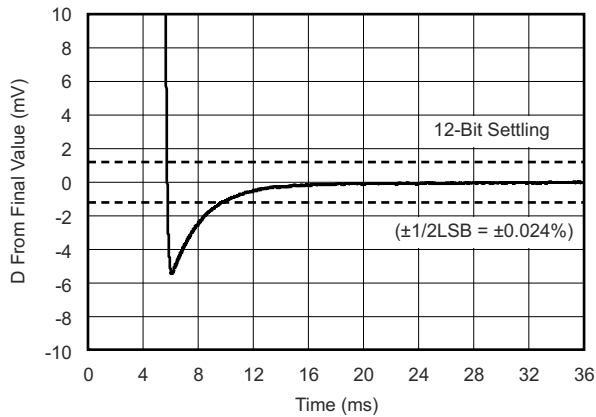




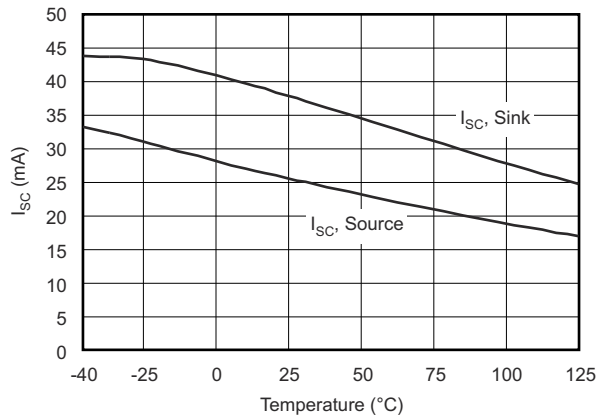
**Figure 5-31. Large-Signal Step Response**



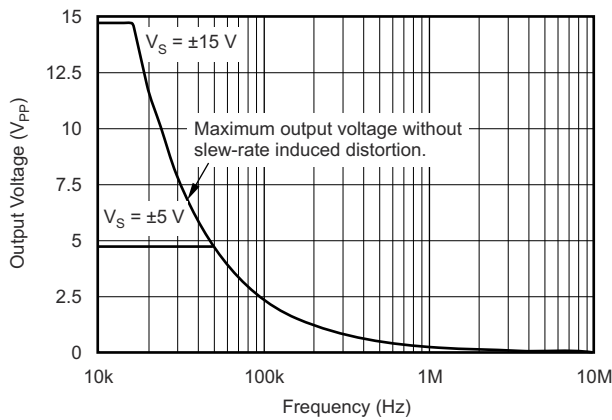
**Figure 5-32. Large-Signal Settling Time (10-V Positive Step)**



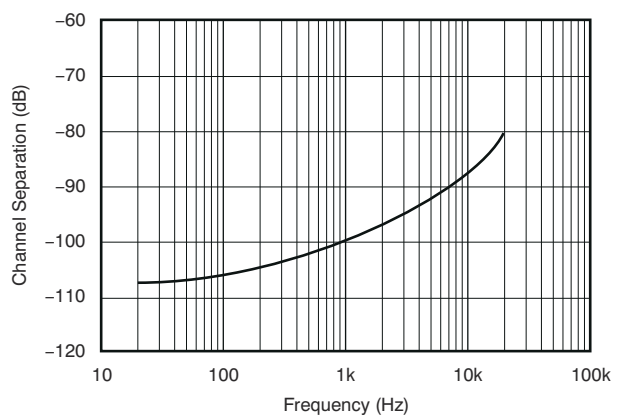
**Figure 5-33. Large-Signal Settling Time (10-V Negative Step)**



**Figure 5-34. Short-Circuit Current vs Temperature**



**Figure 5-35. Maximum Output Voltage vs Frequency**



**Figure 5-36. Channel Separation vs Frequency**

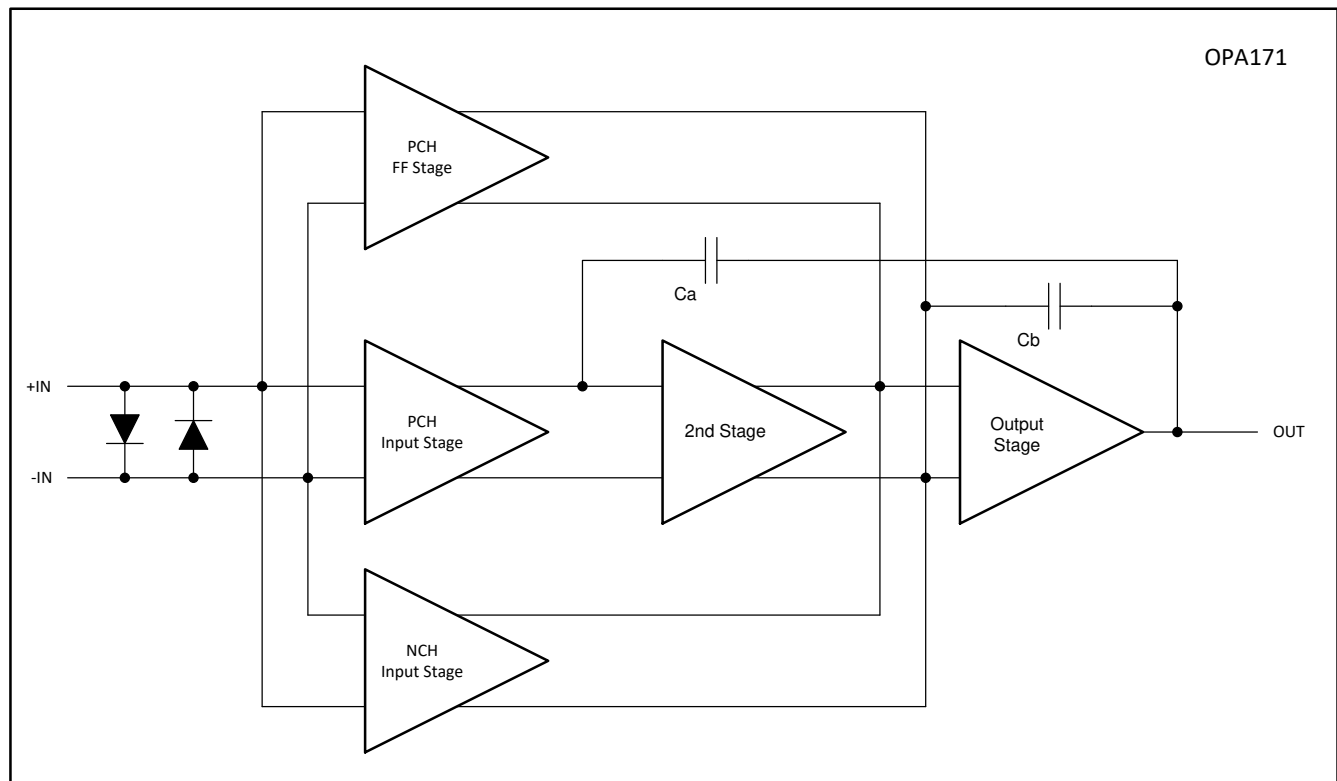


## 6 Detailed Description

### 6.1 Overview

The OPAx171 operational amplifiers provide high overall performance, and are designed for many general-purpose applications. The excellent offset drift of only  $2 \mu\text{V}/^\circ\text{C}$  provides excellent stability over the entire temperature range. In addition, the series offers good overall performance with high CMRR, PSRR, and  $A_{OL}$ . As with all amplifiers, applications with noisy or high-impedance power supplies require decoupling capacitors close to the device pins. In most cases,  $0.1\text{-}\mu\text{F}$  capacitors are adequate.

### 6.2 Functional Block Diagram



### 6.3 Feature Description

#### 6.3.1 Operating Characteristics

The OPAx171 family of amplifiers is specified for operation from 2.7 to 36 V ( $\pm 1.35$  to  $\pm 18$  V). Many of the specifications apply from  $-40^\circ\text{C}$  to  $+125^\circ\text{C}$ . Parameters that can exhibit significant variance with regard to operating voltage or temperature are presented in [Section 5.9](#).

#### 6.3.2 Common-Mode Voltage Range

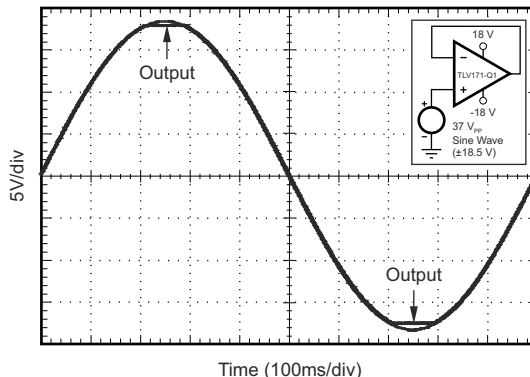
The input common-mode voltage range of the OPAx171 series extends 100 mV below the negative rail and within 2 V of the top rail for normal operation.

This family can operate with full rail-to-rail input 100 mV beyond the top rail, but with reduced performance within 2 V of the top rail. The typical performance in this range is summarized in [Table 6-1](#).

#### 6.3.3 Phase-Reversal Protection

The OPAx171 family has an internal phase-reversal protection. Many operational amplifiers exhibit a phase reversal when the input is driven beyond the linear common-mode range. This condition is most often encountered in noninverting circuits when the input is driven beyond the specified common-mode voltage range, causing the output to reverse into the opposite rail. The input of the OPAx171 prevents phase reversal with

excessive common-mode voltage. Instead, the output limits into the appropriate rail. **Figure 6-1** shows the performance.



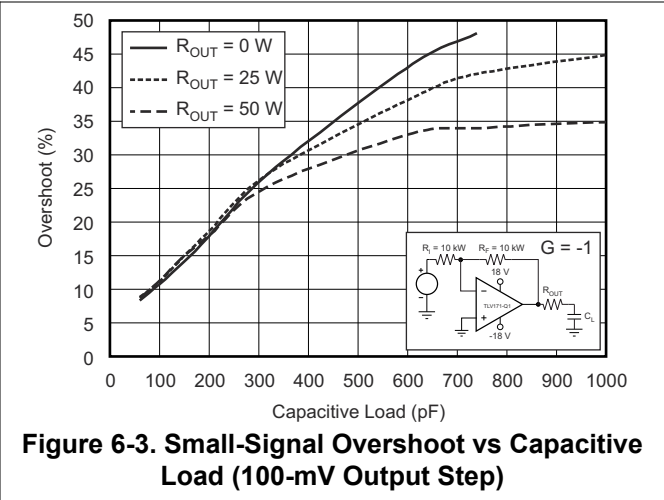
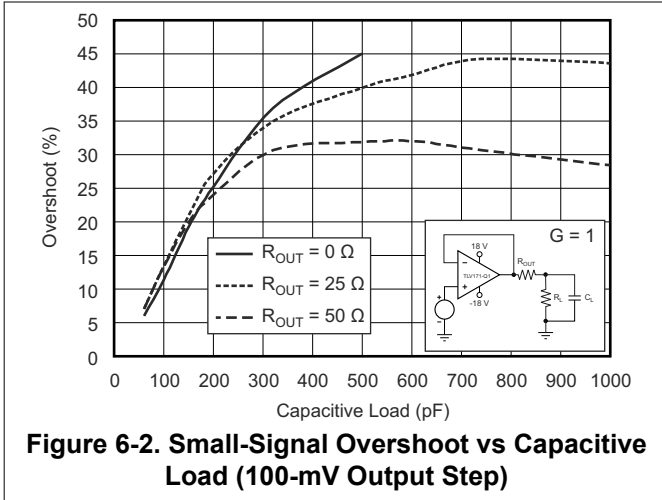
**Figure 6-1. No Phase Reversal**

**Table 6-1. Typical Performance Range**

PARAMETER	MIN	TYP	MAX	UNIT
Input common-mode voltage	$(V+) - 2$		$(V+) + 0.1$	V
Offset voltage		7		mV
vs temperature		12		$\mu\text{V}/^\circ\text{C}$
Common-mode rejection		65		dB
Open-loop gain		60		dB
GBW		0.7		MHz
Slew rate		0.7		$\text{V}/\mu\text{s}$
Noise at $f = 1 \text{ kHz}$		30		$\text{nV}/\sqrt{\text{Hz}}$

### 6.3.4 Capacitive Load and Stability

The dynamic characteristics of the OPAx171-Q1 family of devices have been optimized for commonly encountered operating conditions. The combination of low closed-loop gain and high capacitive loads decreases the phase margin of the amplifier and can lead to gain peaking or oscillations. As a result, heavier capacitive loads must be isolated from the output. The simplest way to achieve this isolation is to add a small resistor (for example,  $R_{\text{OUT}}$  equal to  $50 \Omega$ ) in series with the output. **Figure 6-2** and **Figure 6-3** show small-signal overshoot versus capacitive load for several values of  $R_{\text{OUT}}$ . For details of analysis techniques and application circuits, see [Applications Bulletin AB-028](#), available for download from [TI.com](#).



## 6.4 Device Functional Modes

### 6.4.1 Common-Mode Voltage Range

The input common-mode voltage range of the OPAx171 family extends 100 mV below the negative rail and within 2 V of the top rail for normal operation.

These devices can operate with full rail-to-rail input 100 mV beyond the top rail, but with reduced performance within 2 V of the top rail. The typical performance in this range is summarized in [Table 6-1](#).

## 7 Application and Implementation

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### Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

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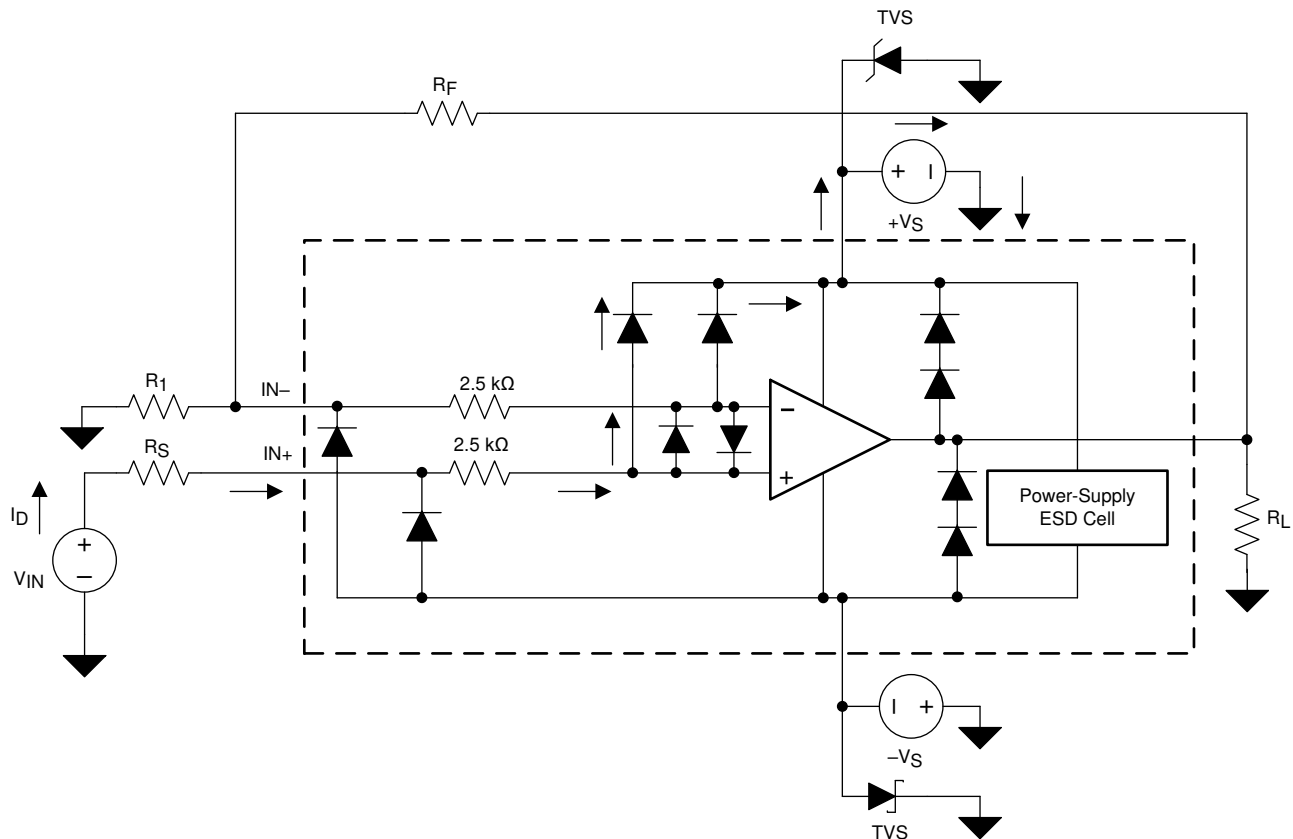
### 7.1 Application Information

The OPAx171 operational amplifiers provide high overall performance, and are designed for many general-purpose applications. The excellent offset drift of only 2  $\mu\text{V}/^\circ\text{C}$  provides excellent stability over the entire temperature range. In addition, the series offers good overall performance with high CMRR, PSRR, and  $A_{OL}$ . As with all amplifiers, applications with noisy or high-impedance power supplies require decoupling capacitors close to the device pins. In most cases, 0.1- $\mu\text{F}$  capacitors are adequate.

#### 7.1.1 Electrical Overstress

Designers often ask questions about the capability of an operational amplifier to withstand electrical overstress. These questions tend to focus on the device inputs, but can involve the supply voltage pins or even the output pin. Each of these different pin functions have electrical stress limits determined by the voltage breakdown characteristics of the particular semiconductor fabrication process and specific circuits connected to the pin. Additionally, internal electrostatic discharge (ESD) protection is built into these circuits for protection from accidental ESD events both before and during product assembly.

A good understanding of this basic ESD circuitry and the relevance to an electrical overstress event is helpful. [Figure 7-1](#) shows the ESD circuits contained in the OPAx171 (indicated by the dashed line area). The ESD protection circuitry involves several current-steering diodes connected from the input and output pins and routed back to the internal power supply lines, where the diodes meet at an absorption device internal to the operational amplifier. This protection circuitry is intended to remain inactive during normal circuit operation.



**Figure 7-1. Equivalent Internal ESD Circuitry Relative to a Typical Circuit Application**

An ESD event produces a short duration, high-voltage pulse that is transformed into a short duration, high-current pulse when discharging through a semiconductor device. The ESD protection circuits are designed to provide a current path around the operational amplifier core to prevent damage. The energy absorbed by the protection circuitry is then dissipated as heat.

When an ESD voltage develops across two or more amplifier device pins, current flows through one or more steering diodes. Depending on the path that the current takes, the absorption device can activate. The absorption device contains a trigger (or threshold voltage) that is above the normal operating voltage of the OPAx171 but below the device breakdown level. When this threshold is exceeded, the absorption device quickly activates and clamps the voltage across the supply rails to a safe level.

When the operational amplifier connects into a circuit (as shown in [Figure 7-1](#)), the ESD protection components are intended to remain inactive and do not become involved in the application circuit operation. However, circumstances may arise when an applied voltage exceeds the operating voltage of a given pin. If this condition occurs, there is a risk that some internal ESD protection circuits can turn on and conduct current. Any such current flow occurs through steering-diode paths and rarely involves the absorption device.

[Figure 7-1](#) shows a specific example where the input voltage ( $V_{IN}$ ) exceeds the positive supply voltage ( $V+$ ) by 500 mV or more. Much of what happens in the circuit depends on the supply characteristics. If  $V+$  can sink the current, one of the upper steering diodes conducts and directs current to  $V+$ . Excessively high current levels can flow with increasingly higher  $V_{IN}$ . As a result, the data sheet specifications recommend that applications limit the input current to 10 mA.

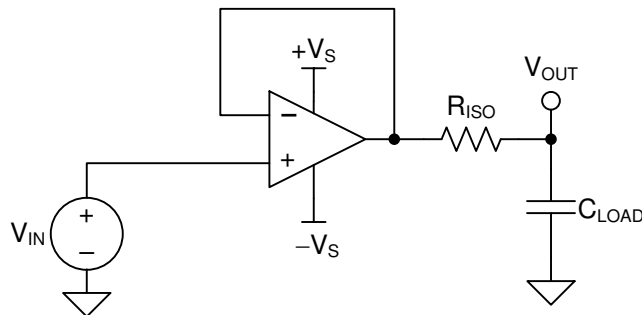
If the supply is not capable of sinking the current,  $V_{IN}$  begins sourcing current to the operational amplifier and then take over as the source of positive supply voltage. The danger in this case is that the voltage can rise to levels that exceed the operational amplifier absolute maximum ratings.

Another common question involves what happens to the amplifier if an input signal is applied to the input when the power supplies ( $V_+$  or  $V_-$ ) are at 0 V. This question depends on the supply characteristic when at 0 V, or at a level below the input signal amplitude. If the supplies appear to be high-impedance, then the input source supplies the operational amplifier current through the current-steering diodes. This state is not a normal bias condition. Most likely, the amplifier does not operate normally. If the supplies are low-impedance, then the current through the steering diodes can be quite high. The current level depends on the ability of the input source to deliver current and any resistance in the input path.

If there is any uncertainty about the ability of the supply to absorb this current, add external Zener diodes to the supply pins; see [Figure 7-1](#). Select the Zener voltage so that the diode does not turn on during normal operation. However, the Zener voltage must be low enough so that the Zener diode conducts if the supply pin begins to rise above the safe operating, supply-voltage level.

The OPAx171 input pins are protected from excessive differential voltage with back-to-back diodes; see [Figure 7-1](#). In most circuit applications, the input protection circuitry does not affect the application. However, in low gain or  $G = 1$  circuits, fast-ramping input signals can forward bias these diodes because the output of the amplifier cannot respond rapidly enough to the input ramp. If the input signal is fast enough to create this forward-bias condition, limit the input signal current to 10 mA or less. If the input signal current is not inherently limited, an input series resistor can be used to limit the input signal current. This input series resistor degrades the low noise performance of the OPAx171. [Figure 7-1](#) shows an example configuration that implements a current-limiting feedback resistor.

## 7.2 Typical Application



**Figure 7-2. Unity-Gain Buffer With  $R_{ISO}$  Stability Compensation**

### 7.2.1 Design Requirements

The design requirements are:

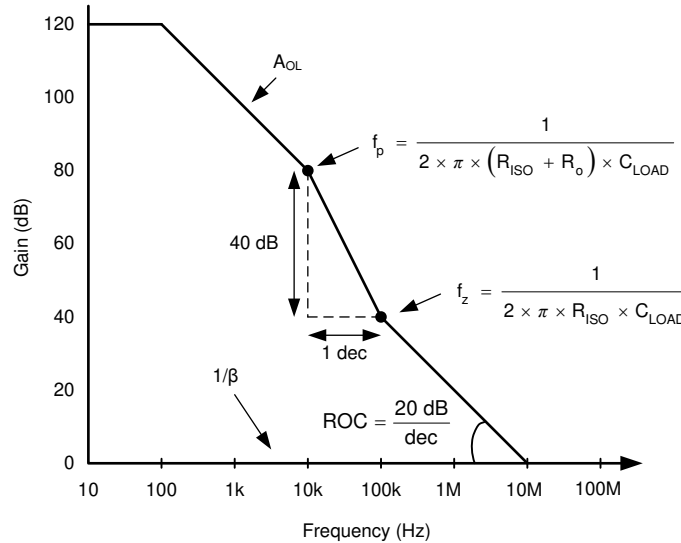
- Supply voltage: 30 V ( $\pm 15$  V)
- Capacitive loads: 100 pF, 1000 pF, 0.01  $\mu$ F, 0.1  $\mu$ F, and 1  $\mu$ F
- Phase margin: 45° and 60°

### 7.2.2 Detailed Design Procedure

[Figure 7-3](#) shows a unity-gain buffer driving a capacitive load. [Equation 1](#) shows the transfer function for the circuit in [Figure 7-3](#). Not shown in [Figure 7-3](#) is the open-loop output resistance of the operational amplifier,  $R_o$ .

$$T(s) = \frac{1 + C_{LOAD} \times R_{ISO} \times s}{1 + (R_o + R_{ISO}) \times C_{LOAD} \times s} \quad (1)$$

The transfer function in [Equation 1](#) contains a pole and a zero. The frequency of the pole ( $f_p$ ) is determined by  $(R_o + R_{ISO})$  and  $C_{LOAD}$ . Components  $R_{ISO}$  and  $C_{LOAD}$  determine the frequency of the zero ( $f_z$ ). Select  $R_{ISO}$  such that the rate of closure (ROC) between the open-loop gain ( $A_{OL}$ ) and  $1/\beta$  is 20 dB/decade to obtain a stable system. [Figure 7-3](#) shows the concept. The  $1/\beta$  curve for a unity-gain buffer is 0 dB.



**Figure 7-3. Unity-Gain Amplifier With  $R_{ISO}$  Compensation**

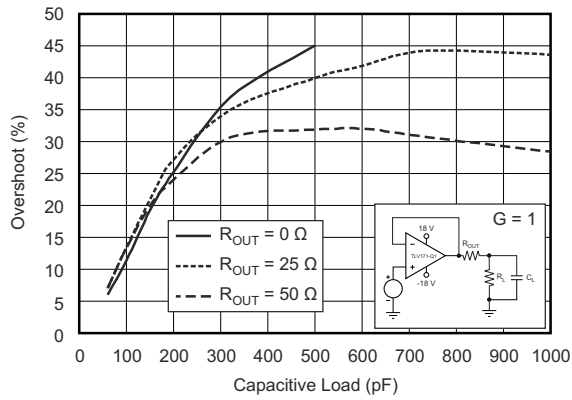
ROC stability analysis is typically simulated. The validity of the analysis depends on multiple factors, especially the accurate modeling of  $R_o$ . In addition to simulating the ROC, a robust stability analysis includes a measurement of overshoot percentage and AC gain peaking of the circuit using a function generator, oscilloscope, and gain and phase analyzer. Phase margin is then calculated from these measurements. [Table 7-1](#) shows the overshoot percentage and AC gain peaking that correspond to phase margins of 45° and 60°. For more details on this design and other alternative devices that can be used in place of the OPAx171, see [Capacitive Load Drive Solution using an Isolation Resistor](#).

**Table 7-1. Phase Margin vs Overshoot and AC Gain Peaking**

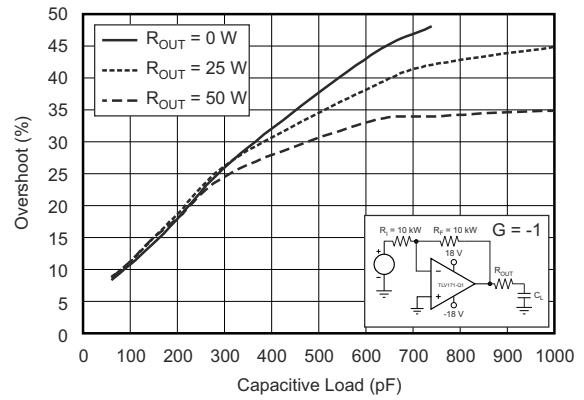
PHASE MARGIN	OVERSHOOT	AC GAIN PEAKING
45°	23.3%	2.35 dB
60°	8.8%	0.28 dB

### 7.2.2.1 Capacitive Load and Stability

The dynamic characteristics of the OPAx171 are optimized for commonly encountered operating conditions. The combination of low closed-loop gain and high capacitive loads decreases the phase margin of the amplifier and can lead to gain peaking or oscillations. As a result, heavier capacitive loads must be isolated from the output. The simplest way to achieve this isolation is to add a small resistor (for example,  $R_{OUT}$  equal to 50  $\Omega$ ) in series with the output. [Figure 6-2](#) and [Figure 6-3](#) illustrate graphs of small-signal overshoot versus capacitive load for several values of  $R_{OUT}$ . See [Applications Bulletin AB-028](#), available for download from the TI website for details of analysis techniques and application circuits.



**Figure 7-4. Small-Signal Overshoot vs Capacitive Load (100-mV Output Step)**



**Figure 7-5. Small-Signal Overshoot vs Capacitive Load (100-mV Output Step)**



### 7.2.3 Application Curve

The OPAx171 meets the supply voltage requirements of 30 V. The OPAx171 is tested for various capacitive loads and RISO is adjusted to get an overshoot corresponding to Table 7-1. The results of these tests are summarized in Figure 7-6.

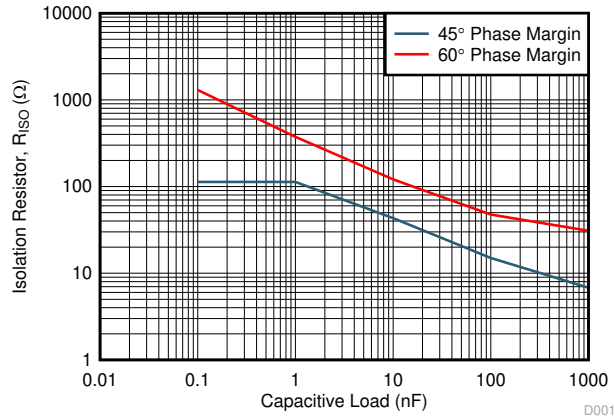


Figure 7-6. RISO vs CLOAD

## 7.3 Power Supply Recommendations

The OPAx171 family is specified for operation from 4.5 V to 36 V ( $\pm 2.25$  V to  $\pm 18$  V); many specifications apply from  $-40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ . Parameters that can exhibit significant variance with regard to operating voltage or temperature are presented in the Section 5 section.

### CAUTION

Supply voltages larger than 40 V can permanently damage the device; see the Section 5.1 table.

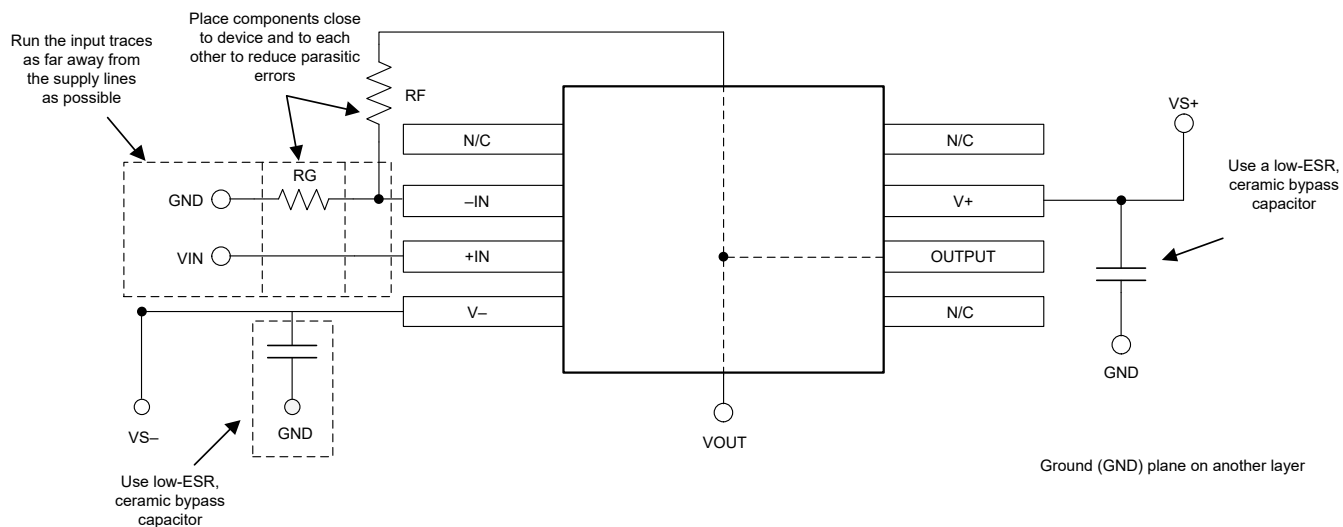
Place 0.1- $\mu\text{F}$  bypass capacitors close to the power-supply pins to reduce errors coupling in from noisy or high-impedance power supplies. For detailed information on bypass capacitor placement, see the Section 7.4.1 section.

## 7.4 Layout

### 7.4.1 Layout Guidelines

For best operational performance of the devices, good printed circuit board (PCB) layout practices are recommended. Low-loss, 0.1- $\mu\text{F}$  bypass capacitors must be connected between each supply pin and ground, placed as close to the devices as possible. A single bypass capacitor from V+ to ground is applicable to single-supply applications.

### 7.4.2 Layout Example



**Figure 7-7. Operational Amplifier Board Layout for Noninverting Configuration**

## 8 Device and Documentation Support

### 8.1 Support Resources

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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### 8.2 Support Resources

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

### 8.3 Trademarks

TI E2E™ is a trademark of Texas Instruments.

All trademarks are the property of their respective owners.

### 8.4 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 8.5 Glossary

[TI Glossary](#) This glossary lists and explains terms, acronyms, and definitions.

## 9 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision G (May 2020) to Revision H (June 2024)	Page
• Changed MSOP-8 package to VSSOP-8 throughout the data sheet.....	1
• Added the 5-pin SOT-23 package to the industry-standard package list in <i>Features</i> .....	1
• Removed the 8-pin MSOP package from the industry-standard package list in <i>Features</i> .....	1
• Updated the numbering format for tables, figures, and cross-references throughout the document.....	1
• Changed Device Information table to Package Information.....	1

Changes from Revision F (April 2018) to Revision G (May 2020)	Page
• Added links to <i>Applications</i> .....	1
• Changed graphs with incorrect units (mV to $\mu$ V) in the Typical Characteristics section.....	11

Changes from Revision E (April 2015) to Revision F (April 2018)	Page
• Changed minimum supply voltage value from $\pm 20$ V to 0 V in <i>Absolute Maximum Ratings</i> table .....	6
• Added maximum supply voltage value of 40 V to <i>Absolute Maximum Ratings</i> table .....	6
• Rewrote <i>Electrical Overstress</i> subsection content in <i>Application Information</i> section .....	20

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**Changes from Revision D (September 2012) to Revision E (April 2015)** **Page**

- Changed device title (removed "Value Line Series")..... 1
  - Added *ESD Ratings* table, *Feature Description* section, *Device Functional Modes, Application and Implementation* section, *Power Supply Recommendations* section, *Layout* section, *Device and Documentation Support* section, and *Mechanical, Packaging, and Orderable Information* section ..... 1
- 

**Changes from Revision C (June 2011) to Revision D (September 2012)** **Page**

- Added "Value Line Series" to title..... 1
- 

**Changes from Revision B (November 2010) to Revision June 2011 C ( )** **Page**

- Added MSOP-8 package to device graphic..... 1
  - Added MSOP-8 package to Features bullets..... 1
  - Added MSOP-8 package to Product Family table..... 1
  - Updated pinout configurations for OPA2171 and OPA4171..... 3
  - Added MSOP-8 package to OPA2171 Thermal Information table..... 7
  - Added new row for Voltage Output Swing from Rail parameter to *Output* subsection of *Electrical Characteristics* ..... 8
  - Changed Voltage Output Swing from Rail parameter to over temperature in *Output* subsection of *Electrical Characteristics* ..... 8
  - Changed [Figure 5-9](#) ..... 11
- 

**Changes from Revision A (November 2010) to Revision B (November 2010)** **Page**

- Changed input offset voltage specification..... 8
  - Changed input offset voltage, over temperature specification..... 8
  - Changed quiescent current per amplifier, over temperature specification..... 8
- 

## 10 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
OPA171AID	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	O171A	<a href="#">Samples</a>
OPA171AIDBVR	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OSUI	<a href="#">Samples</a>
OPA171AIDBVT	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OSUI	<a href="#">Samples</a>
OPA171AIDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	O171A	<a href="#">Samples</a>
OPA171AIDRLR	ACTIVE	SOT-5X3	DRL	5	4000	RoHS & Green	NIPDAUAG	Level-1-260C-UNLIM	-40 to 125	DAP	<a href="#">Samples</a>
OPA171AIDRLT	ACTIVE	SOT-5X3	DRL	5	250	RoHS & Green	NIPDAUAG	Level-1-260C-UNLIM	-40 to 125	DAP	<a href="#">Samples</a>
OPA2171AID	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	2171A	<a href="#">Samples</a>
OPA2171AIDCUR	ACTIVE	VSSOP	DCU	8	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	OPOC	<a href="#">Samples</a>
OPA2171AIDCUT	ACTIVE	VSSOP	DCU	8	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	OPOC	<a href="#">Samples</a>
OPA2171AIDGK	ACTIVE	VSSOP	DGK	8	80	RoHS & Green	NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	OPMI	<a href="#">Samples</a>
OPA2171AIDGKR	ACTIVE	VSSOP	DGK	8	2500	RoHS & Green	NIPDAU   SN   NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	OPMI	<a href="#">Samples</a>
OPA2171AIDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	2171A	<a href="#">Samples</a>
OPA4171AID	ACTIVE	SOIC	D	14	50	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 125	OPA4171	<a href="#">Samples</a>
OPA4171AIDR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 125	OPA4171	<a href="#">Samples</a>
OPA4171AIPW	ACTIVE	TSSOP	PW	14	90	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA4171	<a href="#">Samples</a>
OPA4171AIPWR	ACTIVE	TSSOP	PW	14	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA4171	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of  $\leq 1000$ ppm threshold. Antimony trioxide based flame retardants must also meet the  $\leq 1000$ ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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**OTHER QUALIFIED VERSIONS OF OPA171, OPA2171, OPA4171 :**

● Automotive : [OPA171-Q1](#), [OPA2171-Q1](#), [OPA4171-Q1](#)

● Enhanced Product : [OPA2171-EP](#)

NOTE: Qualified Version Definitions:

- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects
- Enhanced Product - Supports Defense, Aerospace and Medical Applications

**TAPE AND REEL INFORMATION**

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
OPA171AIDBVR	SOT-23	DBV	5	3000	180.0	8.4	3.23	3.17	1.37	4.0	8.0	Q3
OPA171AIDBVR	SOT-23	DBV	5	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
OPA171AIDBVT	SOT-23	DBV	5	250	180.0	8.4	3.23	3.17	1.37	4.0	8.0	Q3
OPA171AIDBVT	SOT-23	DBV	5	250	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
OPA171AIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
OPA171AIDRLR	SOT-5X3	DRL	5	4000	180.0	8.4	1.98	1.78	0.69	4.0	8.0	Q3
OPA171AIDRLT	SOT-5X3	DRL	5	250	180.0	8.4	1.98	1.78	0.69	4.0	8.0	Q3
OPA2171AIDCUR	VSSOP	DCU	8	3000	180.0	8.4	2.25	3.35	1.05	4.0	8.0	Q3
OPA2171AIDCUT	VSSOP	DCU	8	250	180.0	8.4	2.25	3.35	1.05	4.0	8.0	Q3
OPA2171AIDGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA2171AIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
OPA4171AIDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
OPA4171AIPWR	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1

**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
OPA171AIDBVR	SOT-23	DBV	5	3000	213.0	191.0	35.0
OPA171AIDBVR	SOT-23	DBV	5	3000	195.0	200.0	45.0
OPA171AIDBVT	SOT-23	DBV	5	250	223.0	270.0	35.0
OPA171AIDBVT	SOT-23	DBV	5	250	213.0	191.0	35.0
OPA171AIDR	SOIC	D	8	2500	356.0	356.0	35.0
OPA171AIDRLR	SOT-5X3	DRL	5	4000	202.0	201.0	28.0
OPA171AIDRLT	SOT-5X3	DRL	5	250	202.0	201.0	28.0
OPA2171AIDCUR	VSSOP	DCU	8	3000	202.0	201.0	28.0
OPA2171AIDCUT	VSSOP	DCU	8	250	202.0	201.0	28.0
OPA2171AIDGKR	VSSOP	DGK	8	2500	356.0	356.0	35.0
OPA2171AIDR	SOIC	D	8	2500	356.0	356.0	35.0
OPA4171AIDR	SOIC	D	14	2500	356.0	356.0	35.0
OPA4171AIPWR	TSSOP	PW	14	2000	356.0	356.0	35.0



**TUBE**


\*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
OPA171AID	D	SOIC	8	75	506.6	8	3940	4.32
OPA2171AID	D	SOIC	8	75	506.6	8	3940	4.32
OPA2171AIDGK	DGK	VSSOP	8	80	330	6.55	500	2.88
OPA4171AID	D	SOIC	14	50	506.6	8	3940	4.32
OPA4171AIPW	PW	TSSOP	14	90	530	10.2	3600	3.5

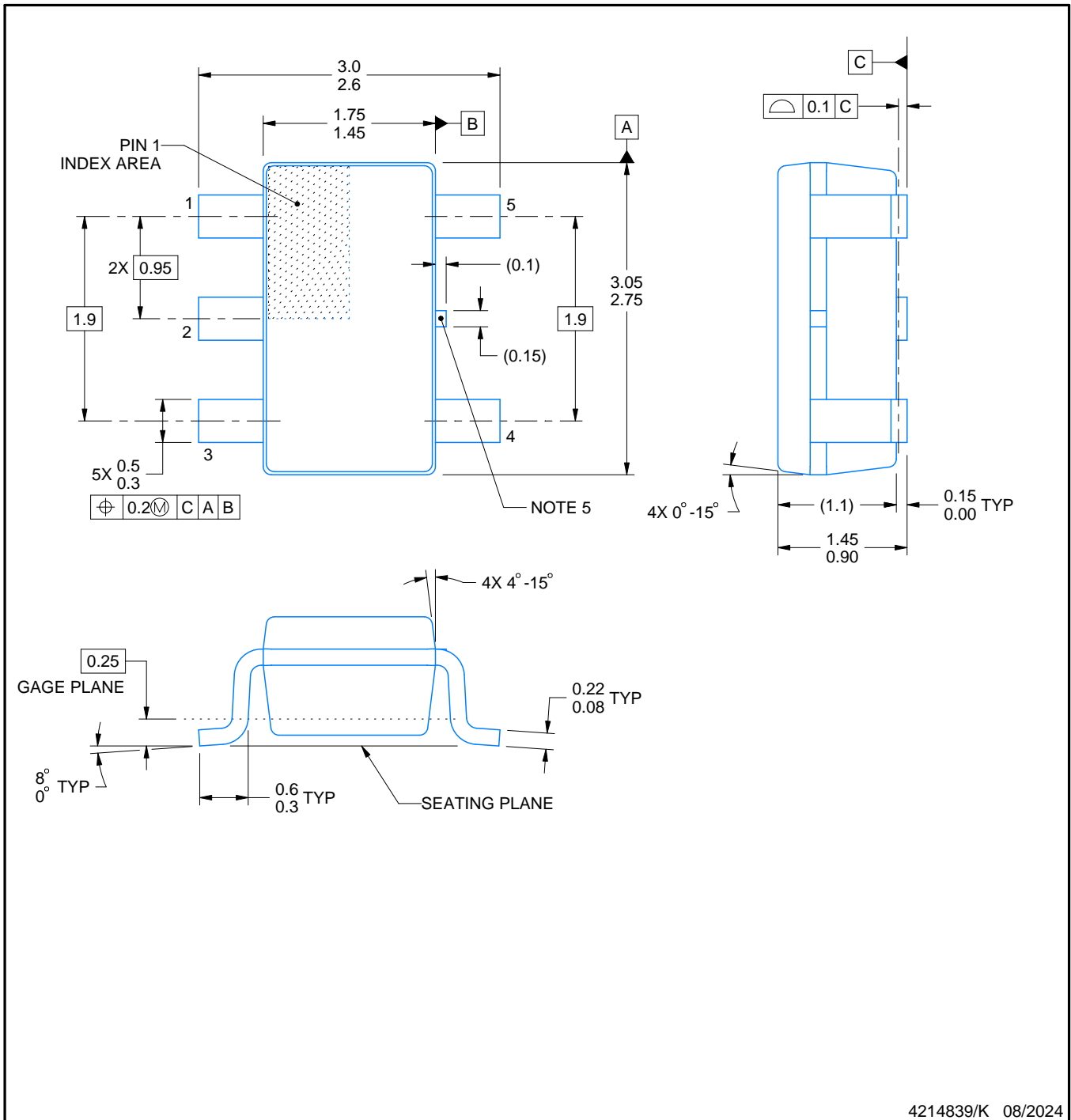
# DBV0005A



# PACKAGE OUTLINE

## SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



4214839/K 08/2024

### NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. Reference JEDEC MO-178.
4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.
5. Support pin may differ or may not be present.

# EXAMPLE BOARD LAYOUT

DBV0005A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:15X



SOLDER MASK DETAILS

4214839/K 08/2024

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DBV0005A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL  
SCALE:15X

4214839/K 08/2024

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

# DGK0008A



# PACKAGE OUTLINE

VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



4214862/A 04/2023

**NOTES:**

PowerPAD is a trademark of Texas Instruments.

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-187.

# EXAMPLE BOARD LAYOUT

DGK0008A

™ VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE: 15X



SOLDER MASK DETAILS

4214862/A 04/2023

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
8. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.
9. Size of metal pad may vary due to creepage requirement.

# EXAMPLE STENCIL DESIGN

DGK0008A

™ VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE

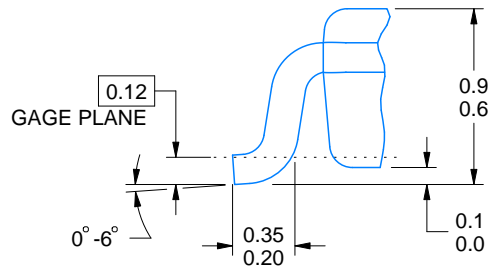
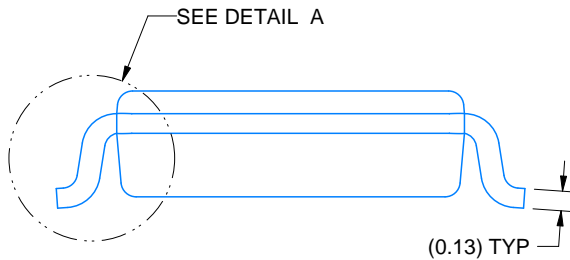
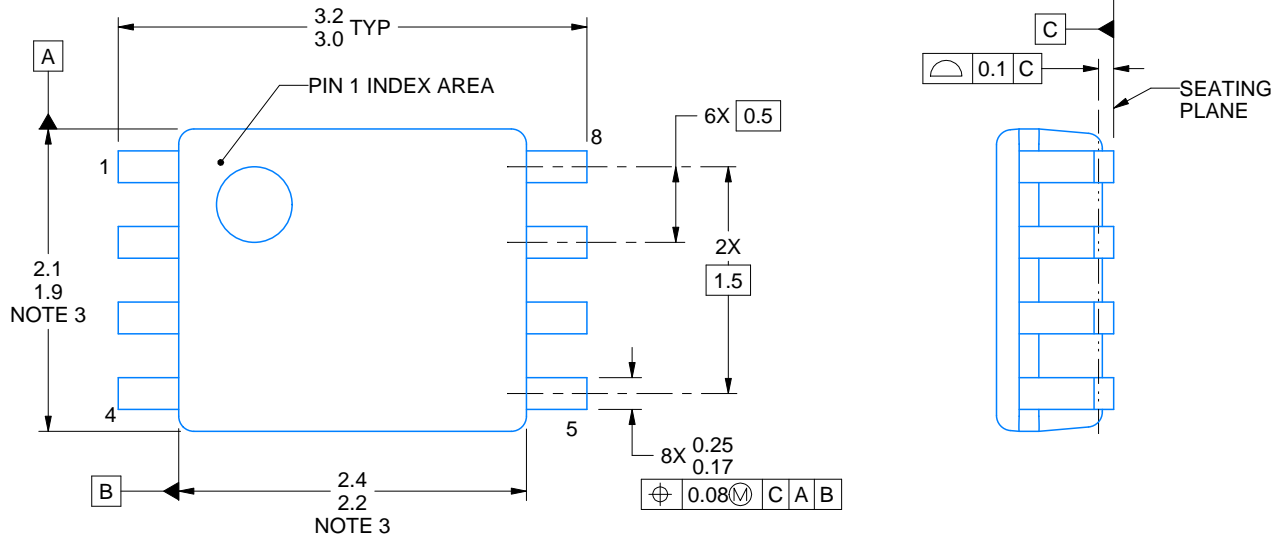


SOLDER PASTE EXAMPLE  
SCALE: 15X

4214862/A 04/2023

NOTES: (continued)

11. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
12. Board assembly site may have different recommendations for stencil design.



4225266/A 09/2014

NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. Reference JEDEC registration MO-187 variation CA.



# EXAMPLE BOARD LAYOUT

DCU0008A

VSSOP - 0.9 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE: 25X



4225266/A 09/2014

NOTES: (continued)

- 5. Publication IPC-7351 may have alternate designs.
- 6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DCU0008A

VSSOP - 0.9 mm max height

SMALL OUTLINE PACKAGE

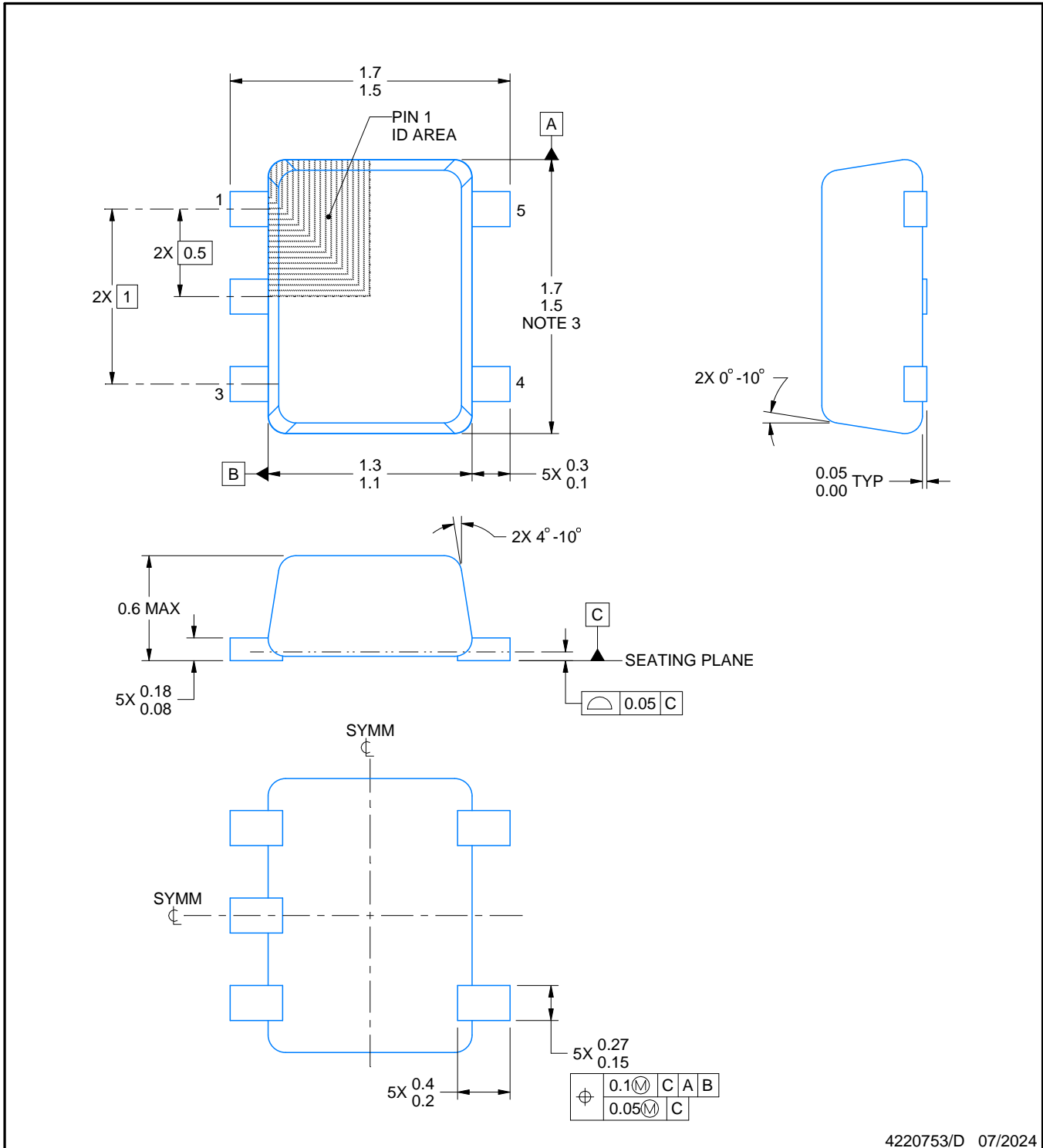
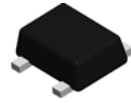


SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL  
SCALE: 25X

4225266/A 09/2014

NOTES: (continued)

7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
8. Board assembly site may have different recommendations for stencil design.



4220753/D 07/2024

NOTES:

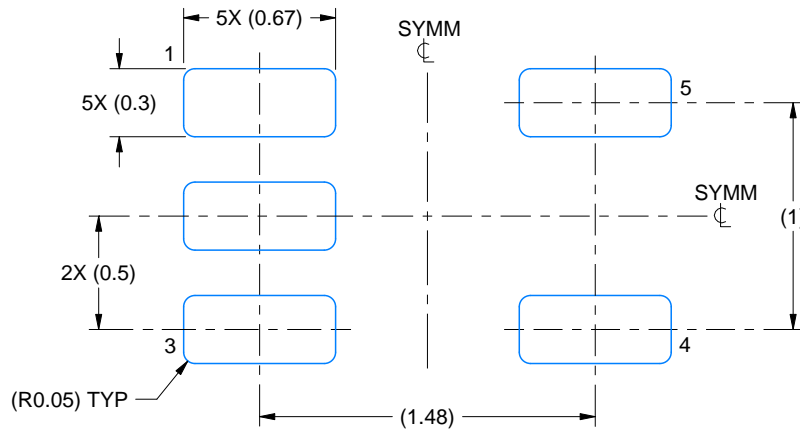
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. Reference JEDEC registration MO-293 Variation UAAD-1

# EXAMPLE BOARD LAYOUT

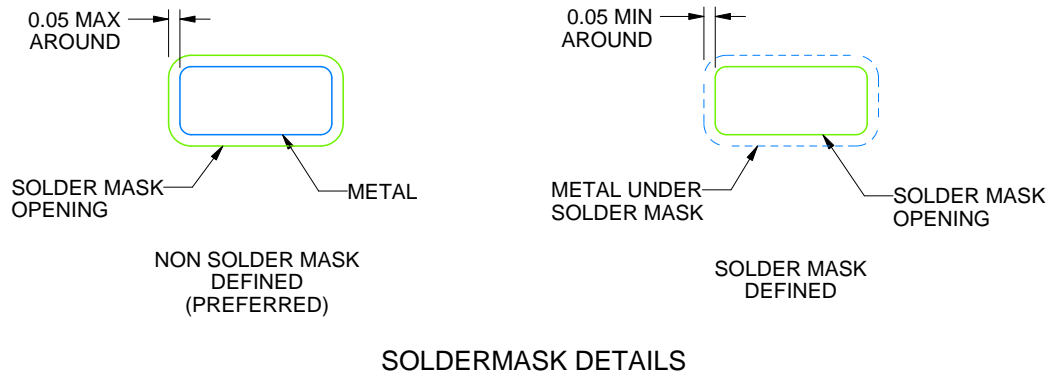
DRL0005A

SOT - 0.6 mm max height

PLASTIC SMALL OUTLINE



LAND PATTERN EXAMPLE  
SCALE:30X



SOLDERMASK DETAILS

4220753/D 07/2024

NOTES: (continued)

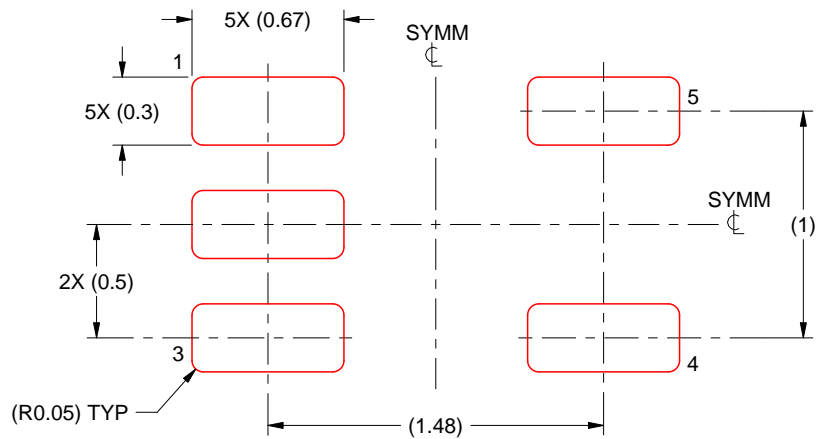
- 5. Publication IPC-7351 may have alternate designs.
- 6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DRL0005A

SOT - 0.6 mm max height

PLASTIC SMALL OUTLINE



SOLDER PASTE EXAMPLE  
BASED ON 0.1 mm THICK STENCIL  
SCALE:30X

4220753/D 07/2024

NOTES: (continued)

7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
8. Board assembly site may have different recommendations for stencil design.

D (R-PDSO-G14)

PLASTIC SMALL OUTLINE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
  - B. This drawing is subject to change without notice.
  -  Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
  -  Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
  - E. Reference JEDEC MS-012 variation AB.

D (R-PDSO-G14)

PLASTIC SMALL OUTLINE



4211283-3/E 08/12

- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Publication IPC-7351 is recommended for alternate designs.
  - D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
  - E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

PW (R-PDSO-G14)

PLASTIC SMALL OUTLINE



4040064-3/G 02/11

- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
  - B. This drawing is subject to change without notice.
  - C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.
  - D. Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.
  - E. Falls within JEDEC MO-153





D0008A

# PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



4214825/C 02/2019

### NOTES:

- Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- This drawing is subject to change without notice.
- This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed  $.006$  [0.15] per side.
- This dimension does not include interlead flash.
- Reference JEDEC registration MS-012, variation AA.

# EXAMPLE BOARD LAYOUT

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:8X



SOLDER MASK DETAILS

4214825/C 02/2019

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE  
BASED ON .005 INCH [0.125 MM] THICK STENCIL  
SCALE:8X

4214825/C 02/2019

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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