

TLC3704, TLC3704M QUAD MICROPOWER LinCMOS™ VOLTAGE COMPARATORS

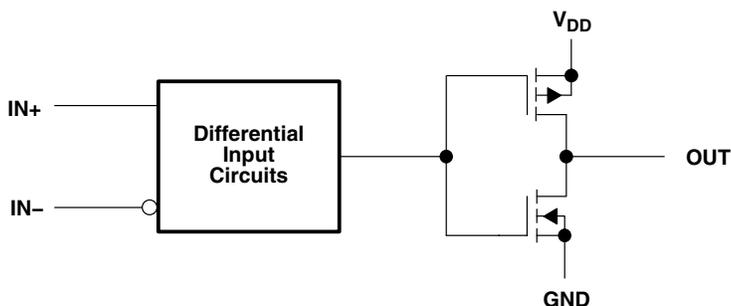
SLCS117C – NOVEMBER 1986 – REVISED NOVEMBER 2009

AVAILABLE OPTIONS

T _A	V _{IOmax} at 25°C	PACKAGE				
		SMALL OUTLINE (D)	CERAMIC (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)	TSSOP (PW)
0°C to 70°C	5 mV	TLC3704CD	—	—	TLC3704CN	TLC3704CPW
–40°C to 85°C	5 mV	TLC3704ID	—	—	TLC3704IN	TLC3704IPW
–55°C to 125°C	5 mV	TLC3704MD	TLC3704MFK	TLC3704MJ	—	—
–40°C to 125°C	5 mV	—	—	TLC3704QJ	—	—

The D and PW packages are available taped and reeled. Add R suffix to the device type (e.g., TLC3704CDR).

functional block diagram (each comparator)



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage range, V _{DD} (see Note 1)	–0.3 V to 18 V
Differential input voltage, V _{ID} (see Note 2)	± 18 V
Input voltage range, V _I	–0.3 to V _{DD}
Output voltage range, V _O	–0.3 to V _{DD}
Input current, I _I	±5 mA
Output current, I _O (each output)	±20 mA
Total supply current into V _{DD}	40 mA
Total current out of GND	60 mA
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, T _A : TLC3704C	0 to 70°C
TLC3704I	–40°C to 85°C
TLC3704M	–55°C to 125°C
TLC3704Q	–40°C to 125°C
Storage temperature range	–65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or N package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.
2. Differential voltages are at IN+ with respect to IN–.

TLC3704, TLC3704M

QUAD MICROPOWER LinCMOS™ VOLTAGE COMPARATORS

SLCS117C – NOVEMBER 1986 – REVISED NOVEMBER 2009

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	N/A
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
N	1150 mW	9.2 mW/°C	736 mW	598 mW	N/A
PW	675 mW	5.4 mW/°C	432 mW	351 mW	N/A

recommended operating conditions

	TLC3704C			UNIT
	MIN	NOM	MAX	
Supply voltage, V_{DD}	3	5	16	V
Common-mode input voltage, V_{IC}	-0.2	$V_{DD} - 1.5$		V
High-level output current, I_{OH}	-20			mA
Low-level output current, I_{OL}	20			mA
Operating free-air temperature, T_A	0	70		°C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS†	T_A	TLC3704C			UNIT
			MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{DD} = 5\text{ V to }10\text{ V}$, $V_{IC} = V_{ICRmin}$, See Note 3	25°C	1.2		5	mV
		0°C to 70°C	6.5			
I_{IO} Input offset current	$V_{IC} = 2.5\text{ V}$	25°C	1		-	pA
		70°C	0.3		-	nA
I_{IB} Input bias current	$V_{IC} = 2.5\text{ V}$	25°C	5		-	pA
		70°C	0.6		-	nA
V_{ICR} Common-mode input voltage range		25°C	0 to $V_{DD} - 1$		-	V
		0°C to 70°C	0 to $V_{DD} - 1.5$		-	
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84		-	dB
		70°C	84		-	
		0°C	84		-	
k_{SVR} Supply-voltage rejection ratio	$V_{DD} = 5\text{ V to }10\text{ V}$	25°C	85		-	dB
		70°C	85		-	
		0°C	85		-	
V_{OH} High-level output voltage	$V_{ID} = 1\text{ V}$, $I_{OH} = -4\text{ mA}$	25°C	4.5	4.7	-	V
		70°C	4.3		-	
V_{OL} Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OH} = 4\text{ mA}$	25°C	210	300	-	mV
		70°C	375		-	
I_{DD} Supply current (all four comparators)	Outputs low, No load	25°C	35	80	-	µA
		0°C to 70°C	100		-	

† All characteristics are measured with zero common-mode voltage unless otherwise noted.

NOTE 3: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V.



TLC3704, TLC3704M QUAD MICROPOWER LinCMOS™ VOLTAGE COMPARATORS

SLCS117C – NOVEMBER 1986 – REVISED NOVEMBER 2009

recommended operating conditions

	TLC3704I			UNIT
	MIN	NOM	MAX	
Supply voltage, V_{DD}	3	5	16	V
Common-mode input voltage, V_{IC}	-0.2		$V_{DD} - 1.5$	V
High-level output current, I_{OH}			-20	mA
Low-level output current, I_{OL}			20	mA
Operating free-air temperature, T_A	-40		85	°C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5\text{ V}$, $V_{IC} = 0$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A	TLC3704I			UNIT
			MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{DD} = 5\text{ V to }10\text{ V}$, $V_{IC} = V_{ICRmin}$, See Note 3	25°C		1.2	5	mV
		-40°C to 85°C			7	
I_{IO} Input offset current	$V_{IC} = 2.5\text{ V}$	25°C		1		pA
		85°C			1	nA
I_{IB} Input bias current	$V_{IC} = 2.5\text{ V}$	25°C		5		pA
		85°C			2	nA
V_{ICR} Common-mode input voltage range		25°C	0 to $V_{DD} - 1$		V	
		-40°C to 85°C	0 to $V_{DD} - 1.5$			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84		dB	
		85°C	84			
		-40°C	83			
k_{SVR} Supply-voltage rejection ratio	$V_{DD} = 5\text{ V to }10\text{ V}$	25°C	85		dB	
		85°C	85			
		-40°C	83			
V_{OH} High-level output voltage	$V_{ID} = 1\text{ V}$, $I_{OH} = -4\text{ mA}$	25°C	4.5	4.7	V	
		85°C	4.3			
V_{OL} Low-level output voltage	$V_{ID} = -1\text{ V}$, $I_{OH} = 4\text{ mA}$	25°C	210	300	mV	
		85°C	400			
I_{DD} Supply current (all four comparators)	Outputs low, No load	25°C	35	80	µA	
		-40°C to 85°C	125			

NOTE 3: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V.



TLC3704, TLC3704M

QUAD MICROPOWER LinCMOS™ VOLTAGE COMPARATORS

SLCS117C – NOVEMBER 1986 – REVISED NOVEMBER 2009

recommended operating conditions

	TLC3704M			UNIT
	MIN	NOM	MAX	
Supply voltage, V_{DD}	4	5	16	V
Common-mode input voltage, V_{IC}	0	$V_{DD} - 1.5$		V
High-level output current, I_{OH}	– 20			mA
Low-level output current, I_{OL}	20			mA
Operating free-air temperature, T_A	– 55	125		°C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5$ V, $V_{IC} = 0$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A	TLC3704M			UNIT
			MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{DD} = 5$ V to 10 V, $V_{IC} = V_{ICRmin}$, See Note 3	25°C	1.2		5	mV
		–55°C to 125°C			10	
I_{IO} Input offset current	$V_{IC} = 2.5$ V	25°C	1			pA
		125°C			15	nA
I_{IB} Input bias current	$V_{IC} = 2.5$ V	25°C	5			pA
		125°C			30	nA
V_{ICR} Common-mode input voltage range		25°C	0 to $V_{DD} - 1$			V
		–55°C to 125°C	0 to $V_{DD} - 1.5$			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84			dB
		125°C	83			
		–55°C	82			
k_{SVR} Supply-voltage rejection ratio	$V_{DD} = 5$ V to 10 V	25°C	85			dB
		125°C	85			
		–55°C	82			
V_{OH} High-level output voltage	$V_{ID} = 1$ V, $I_{OH} = -4$ mA	25°C	4.5	4.7		V
		125°C	4.2			
V_{OL} Low-level output voltage	$V_{ID} = -1$ V, $I_{OH} = 4$ mA	25°C	210		300	mV
		125°C			500	
I_{DD} Supply current (all four comparators)	Outputs low, No load	25°C	35		80	μA
		–55°C to 125°C			175	

NOTE 3: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V.



TLC3704, TLC3704M QUAD MICROPOWER LinCMOS™ VOLTAGE COMPARATORS

SLCS117C – NOVEMBER 1986 – REVISED NOVEMBER 2009

recommended operating conditions

	TLC3704Q			UNIT
	MIN	NOM	MAX	
Supply voltage, V_{DD}	3	5	16	V
Common-mode input voltage, V_{IC}	-0.2		$V_{DD} - 1.5$	V
High-level output current, I_{OH}			-20	mA
Low-level output current, I_{OL}			20	mA
Operating free-air temperature, T_A	-40		125	°C

electrical characteristics at specified operating free-air temperature, $V_{DD} = 5$ V, $V_{IC} = 0$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A	TLC3704Q			UNIT
			MIN	TYP	MAX	
V_{IO} Input offset voltage	$V_{DD} = 5$ V to 10 V, $V_{IC} = V_{ICRmin}$, See Note 3	25°C		1.2	5	mV
		-40°C to 125°C			7	
I_{IO} Input offset current	$V_{IC} = 2.5$ V	25°C		1		pA
		125°C			15	nA
I_{IB} Input bias current	$V_{IC} = 2.5$ V	25°C		5		pA
		125°C			30	nA
V_{ICR} Common-mode input voltage range		25°C	0 to $V_{DD} - 1$			V
		-40°C to 125°C	0 to $V_{DD} - 1.5$			
CMRR Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$	25°C	84			dB
		125°C	83			
		-40°C	83			
k_{SVR} Supply-voltage rejection ratio	$V_{DD} = 5$ V to 10 V	25°C	85			dB
		125°C	85			
		-40°C	83			
V_{OH} High-level output voltage	$V_{ID} = 1$ V, $I_{OH} = -4$ mA	25°C	4.5	4.7		V
		125°C	4.2			
V_{OL} Low-level output voltage	$V_{ID} = -1$ V, $I_{OH} = 4$ mA	25°C	210		300	mV
		125°C	500			
I_{DD} Supply current (all four comparators)	Outputs low, No load	25°C	35		80	μA
		-40°C to 125°C	175			

NOTE 3: The offset voltage limits given are the maximum values required to drive the output up to 4.5 V or down to 0.3 V.



TLC3704, TLC3704M

QUAD MICROPOWER LinCMOS™ VOLTAGE COMPARATORS

SLCS117C – NOVEMBER 1986 – REVISED NOVEMBER 2009

switching characteristics, $V_{DD} = 5\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS		TLC3704C, TLC3704I TLC3704M, TLC3704Q			UNIT
			MIN	TYP	MAX	
t_{PLH} Propagation delay time, low-to-high-level output†	$f = 10\text{ kHz}$, $C_L = 50\text{ pF}$	Overdrive = 2 mV	4.5		μs	
		Overdrive = 5 mV	2.7			
		Overdrive = 10 mV	1.9			
		Overdrive = 20 mV	1.4			
		Overdrive = 40 mV	1.1			
	$V_I = 1.4\text{-V}$ step at $IN+$	1.1				
t_{PHL} Propagation delay time, high-to-low-level output†	$f = 10\text{ kHz}$, $C_L = 50\text{ pF}$	Overdrive = 2 mV	4		μs	
		Overdrive = 5 mV	2.3			
		Overdrive = 10 mV	1.5			
		Overdrive = 20 mV	0.95			
		Overdrive = 40 mV	0.65			
	$V_I = 1.4\text{-V}$ step at $IN+$	0.15				
t_f Fall time	$f = 10\text{ kHz}$, $C_L = 50\text{ pF}$	Overdrive = 50 mV	50		ns	
t_r Rise time	$f = 10\text{ kHz}$, $C_L = 50\text{ pF}$	Overdrive = 50 mV	125		ns	

† Simultaneous switching of inputs causes degradation in output response.

TLC3704, TLC3704M QUAD MICROWPOWER LinCMOS™ VOLTAGE COMPARATORS

SLCS117C – NOVEMBER 1986 – REVISED NOVEMBER 2009

PRINCIPLES OF OPERATION

LinCMOS process

The LinCMOS process is a linear polysilicon-gate CMOS process. Primarily designed for single-supply applications, LinCMOS products facilitate the design of a wide range of high-performance analog functions from operational amplifiers to complex mixed-mode converters.

This short guide is intended to answer the most frequently asked questions related to the quality and reliability of LinCMOS products. Direct further questions to the nearest TI field sales office.

electrostatic discharge

CMOS circuits are prone to gate oxide breakdown when exposed to high voltages even if the exposure is only for very short periods of time. Electrostatic discharge (ESD) is one of the most common causes of damage to CMOS devices. It can occur when a device is handled without proper consideration for environmental electrostatic charges, e.g., during board assembly. If a circuit in which one amplifier from a dual op amp is being used and the unused pins are left open, high voltages tends to develop. If there is no provision for ESD protection, these voltages may eventually punch through the gate oxide and cause the device to fail. To prevent voltage buildup, each pin is protected by internal circuitry.

Standard ESD-protection circuits safely shunt the ESD current by providing a mechanism whereby one or more transistors break down at voltages higher than the normal operating voltages but lower than the breakdown voltage of the input gate. This type of protection scheme is limited by leakage currents which flow through the shunting transistors during normal operation after an ESD voltage has occurred. Although these currents are small, on the order of tens of nanoamps, CMOS amplifiers are often specified to draw input currents as low as tens of picoamps.

To overcome this limitation, TI design engineers developed the patented ESD-protection circuit shown in Figure 1. This circuit can withstand several successive 2-kV ESD pulses, while reducing or eliminating leakage currents that may be drawn through the input pins. A more detailed discussion of the operation of the TI ESD-protection circuit is presented on the next page.

All input and output pins on LinCMOS and Advanced LinCMOS products have associated ESD-protection circuitry that undergoes qualification testing to withstand 2000 V discharged from a 100-pF capacitor through a 1500-Ω resistor (human body model) and 200 V from a 100-pF capacitor with no current-limiting resistor (charged device model). These tests simulate both operator and machine handling of devices during normal test and assembly operations.

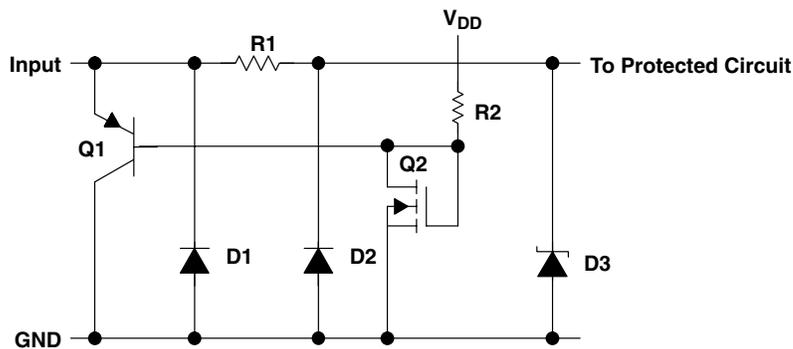


Figure 1. LinCMOS ESD-Protection Schematic

PRINCIPLES OF OPERATION

input protection circuit operation

Texas Instruments patented protection circuitry allows for both positive- and negative-going ESD transients. These transients are characterized by extremely fast rise times and usually low energies, and can occur both when the device has all pins open and when it is installed in a circuit.

positive ESD transients

Initial positive charged energy is shunted through Q1 to V_{SS} . Q1 turns on when the voltage at the input rises above the voltage on the V_{DD} pin by a value equal to the V_{BE} of Q1. The base current increases through R2 with input current as Q1 saturates. The base current through R2 forces the voltage at the drain and gate of Q2 to exceed its threshold level ($V_T \sim 22$ to 26 V) and turn Q2 on. The shunted input current through Q1 to V_{SS} is now shunted through the n-channel enhancement-type MOSFET Q2 to V_{SS} . If the voltage on the input pin continues to rise, the breakdown voltage of the zener diode D3 is exceeded, and all remaining energy is dissipated in R1 and D3. The breakdown voltage of D3 is designed to be 24 to 27 V, which is well below the gate-oxide voltage of the circuit to be protected.

negative ESD transients

The negative charged ESD transients are shunted directly through D1. Additional energy is dissipated in R1 and D2 as D2 becomes forward biased. The voltage seen by the protected circuit is -0.3 V to -1 V (the forward voltage of D1 and D2).

circuit-design considerations

LinCMOS products are being used in actual circuit environments that have input voltages that exceed the recommended common-mode input voltage range and activate the input protection circuit. Even under normal operation, these conditions occur during circuit power up or power down, and in many cases, when the device is being used for a signal conditioning function. The input voltages can exceed V_{ICR} and not damage the device only if the inputs are current limited. The recommended current limit shown on most product data sheets is ± 5 mA. Figures 2 and 3 show typical characteristics for input voltage versus input current.

Normal operation and correct output state can be expected even when the input voltage exceeds the positive supply voltage. Again, the input current should be externally limited even though internal positive current limiting is achieved in the input protection circuit by the action of Q1. When Q1 is on, it saturates and limits the current to approximately 5-mA collector current by design. When saturated, Q1 base current increases with input current. This base current is forced into the V_{DD} pin and into the device I_{DD} or the V_{DD} supply through R2 producing the current limiting effects shown in Figure 2. This internal limiting lasts only as long as the input voltage is below the V_T of Q2.

When the input voltage exceeds the negative supply voltage, normal operation is affected and output voltage states may not be correct. Also, the isolation between channels of multiple devices (duals and quads) can be severely affected. External current limiting must be used since this current is directly shunted by D1 and D2 and no internal limiting is achieved. If normal output voltage states are required, an external input voltage clamp is required (see Figure 4).

TLC3704, TLC3704M QUAD MICROWPOWER LinCMOS™ VOLTAGE COMPARATORS

SLCS117C – NOVEMBER 1986 – REVISED NOVEMBER 2009

PRINCIPLES OF OPERATION

circuit-design considerations (continued)

INPUT CURRENT
vs
INPUT VOLTAGE

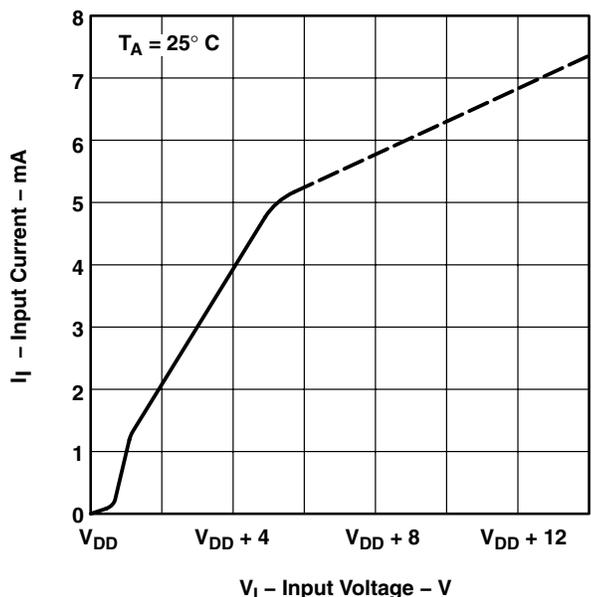


Figure 2

INPUT CURRENT
vs
INPUT VOLTAGE

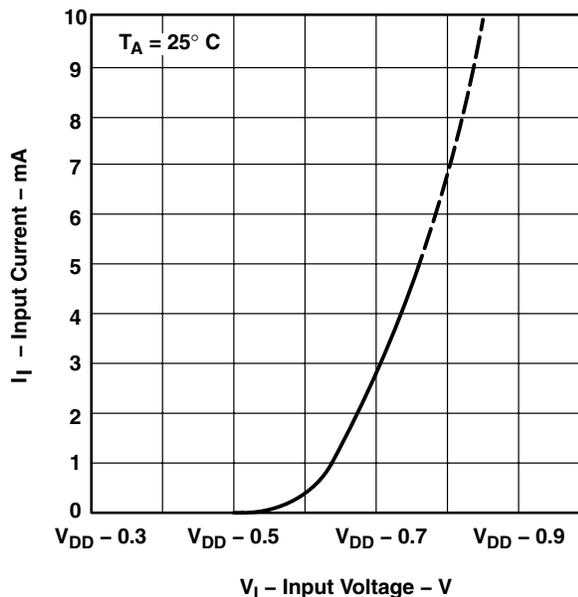
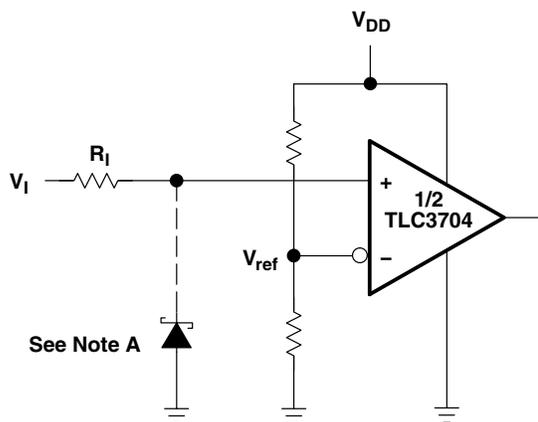


Figure 3



Positive Voltage Input Current Limit :

$$R_I = \frac{V_I - V_{DD} - 0.3 \text{ V}}{5 \text{ mA}}$$

Negative Voltage Input Current Limit :

$$R_I = \frac{-V_I - V_{DD} - (-0.3 \text{ V})}{5 \text{ mA}}$$

NOTE A: If the correct input state is required when the negative input exceeds GND, a Schottky clamp is required.

Figure 4. Typical Input Current-Limiting Configuration for a LinCMOS Comparator

PARAMETER MEASUREMENT INFORMATION

The TLC3704 contains a digital output stage which, if held in the linear region of the transfer curve, can cause damage to the device. Conventional operational amplifier/comparator testing incorporates the use of a servo loop which is designed to force the device output to a level within this linear region. Since the servo-loop method of testing cannot be used, we offer the following alternatives for measuring parameters such as input offset voltage, common-mode rejection, etc.

To verify that the input offset voltage falls within the limits specified, the limit value is applied to the input as shown in Figure 5(a). With the noninverting input positive with respect to the inverting input, the output should be high. With the input polarity reversed, the output should be low.

A similar test can be made to verify the input offset voltage at the common-mode extremes. The supply voltages can be slewed as shown in Figure 5(b) for the V_{ICR} test, rather than changing the input voltages, to provide greater accuracy.

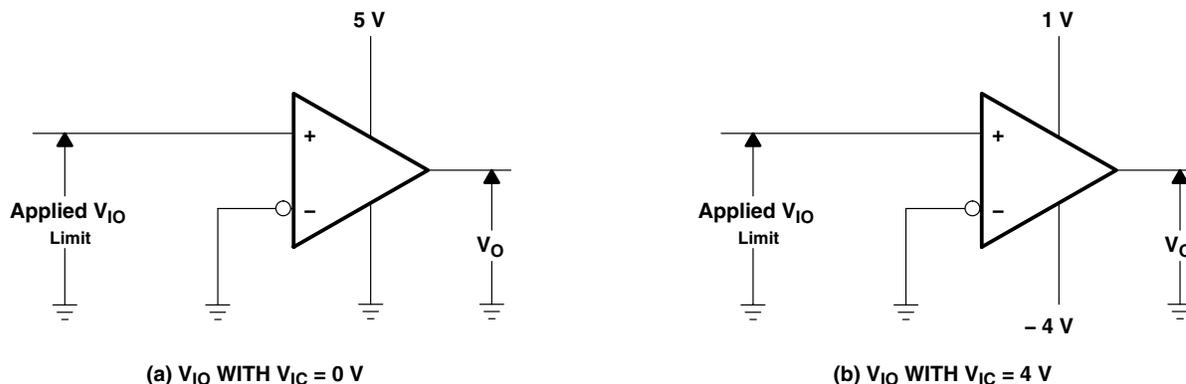


Figure 5. Method for Verifying That Input Offset Voltage Is Within Specified Limits

A close approximation of the input offset voltage can be obtained by using a binary search method to vary the differential input voltage while monitoring the output state. When the applied input voltage differential is equal, but opposite in polarity, to the input offset voltage, the output changes states.

Figure 6 illustrates a practical circuit for direct dc measurement of input offset voltage that does not bias the comparator in the linear region. The circuit consists of a switching mode servo loop in which IC1a generates a triangular waveform of approximately 20-mV amplitude. IC1b acts as a buffer, with C2 and R4 removing any residual d.c. offset. The signal is then applied to the inverting input of the comparator under test, while the noninverting input is driven by the output of the integrator formed by IC1c through the voltage divider formed by R8 and R9. The loop reaches a stable operating point when the output of the comparator under test has a duty cycle of exactly 50%, which can only occur when the incoming triangle wave is sliced symmetrically or when the voltage at the noninverting input exactly equals the input offset voltage.

Voltage divider R8 and R9 provides an increase in the input offset voltage by a factor of 100 to make measurement easier. The values of R5, R7, R8, and R9 can significantly influence the accuracy of the reading; therefore, it is suggested that their tolerance level be one percent or lower.

Measuring the extremely low values of input current requires isolation from all other sources of leakage current and compensation for the leakage of the test socket and board. With a good picoammeter, the socket and board leakage can be measured with no device in the socket. Subsequently, this open socket leakage value can be subtracted from the measurement obtained with a device in the socket to obtain the actual input current of the device.

TLC3704, TLC3704M QUAD MICROPOWER LinCMOS™ VOLTAGE COMPARATORS

SLCS117C – NOVEMBER 1986 – REVISED NOVEMBER 2009

PARAMETER MEASUREMENT INFORMATION

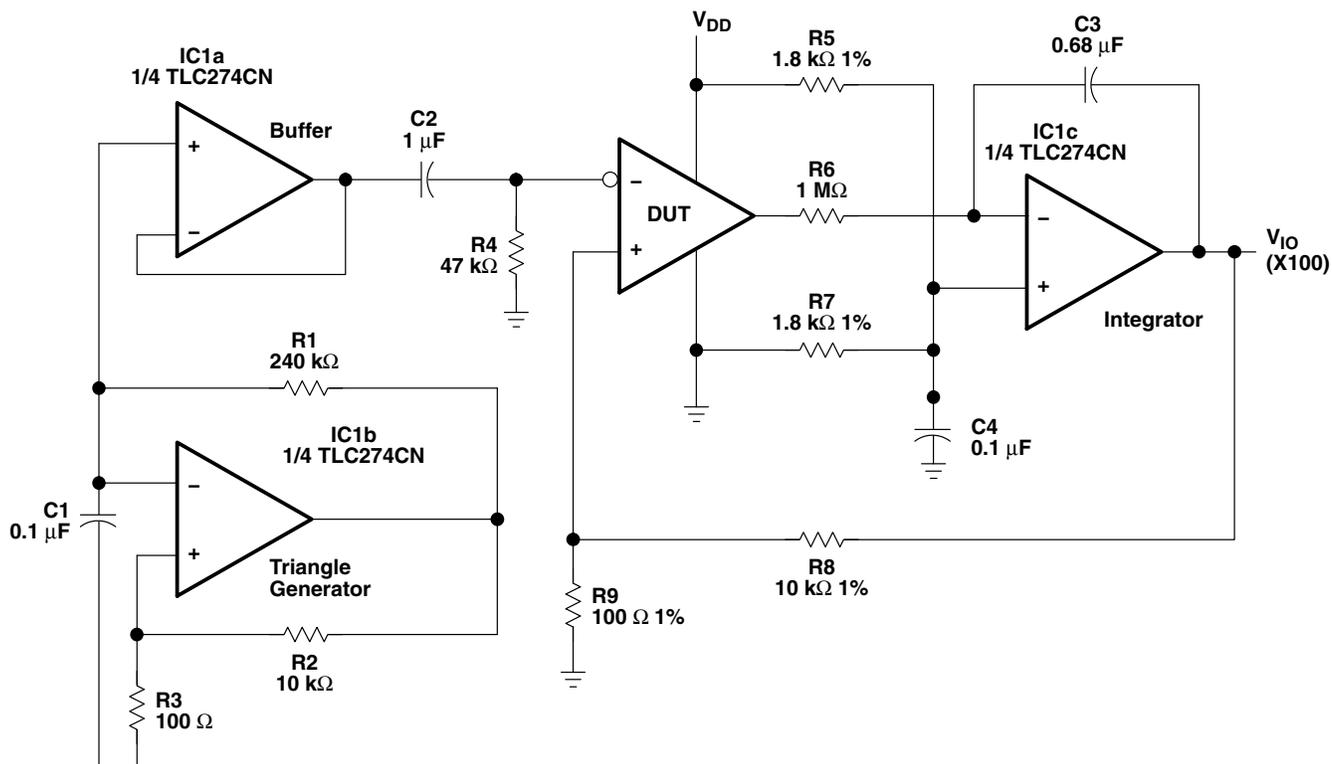
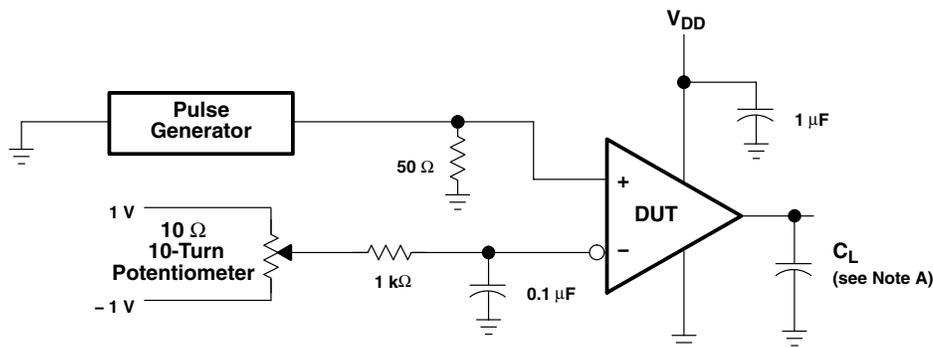


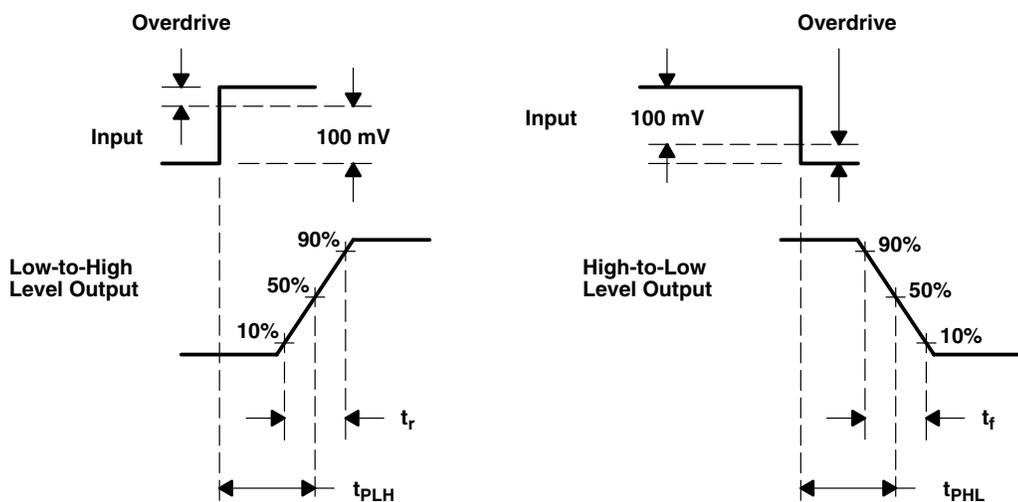
Figure 6. Circuit for Input Offset Voltage Measurement

Response time is defined as the interval between the application of an input step function and the instant when the output reaches 50% of its maximum value. Response time for the low-to-high-level output is measured from the leading edge of the input pulse, while response time for the high-to-low-level output is measured from the trailing edge of the input pulse. Response time measurement at low input signal levels can be greatly affected by the input offset voltage. The offset voltage should be balanced by the adjustment at the inverting input as shown in Figure 7, so that the circuit is just at the transition point. A low signal, for example 105-mV or 5-mV overdrive, causes the output to change state.

PARAMETER MEASUREMENT INFORMATION



TEST CIRCUIT



VOLTAGE WAVEFORMS

NOTE A: C_L includes probe and jig capacitance.

Figure 7. Response, Rise, and Fall Times Circuit and Voltage Waveforms

TLC3704, TLC3704M QUAD MICROWPOWER LinCMOS™ VOLTAGE COMPARATORS

SLCS117C – NOVEMBER 1986 – REVISED NOVEMBER 2009

TYPICAL CHARACTERISTICS

Table of Graphs

			FIGURE
V_{IO}	Input offset voltage	Distribution	8
I_{IB}	Input bias current	vs Free-air temperature	9
CMRR	Common-mode rejection ratio	vs Free-air temperature	10
k_{SVR}	Supply-voltage rejection ratio	vs Free-air temperature	11
V_{OH}	High-level output current	vs Free-air temperature	12
		vs High-level output current	13
V_{OL}	Low-level output voltage	vs Low-level output current	14
		vs Free-air temperature	15
t_t	Output transition time	vs Load capacitance	16
	Supply current response to an output voltage transition		17
	Low-to-high-level output response for various input overdrives		18
	High-to-low-level output response for various input overdrives		19
t_{PLH}	Low-to-high-level output response time	vs Supply voltage	20
t_{PHL}	High-to-low-level output response time	vs Supply voltage	21
I_{DD}	Supply current	vs Frequency	22
		vs Supply voltage	23
		vs Free-air temperature	24

DISTRIBUTION OF INPUT
OFFSET VOLTAGE†

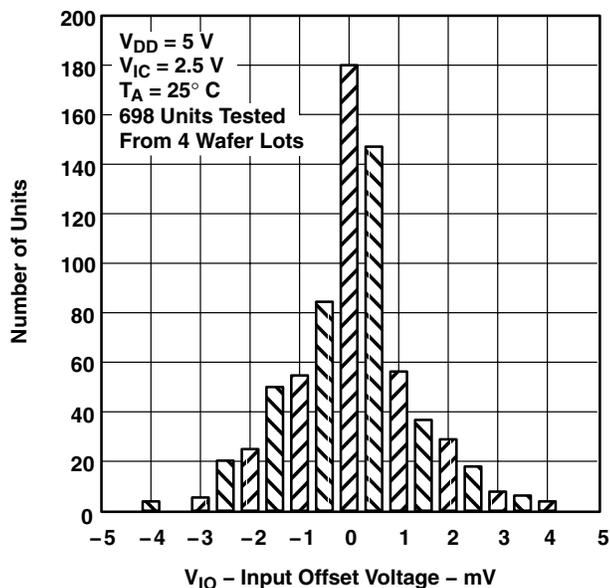


Figure 8

INPUT BIAS CURRENT
VS
FREE-AIR TEMPERATURE†

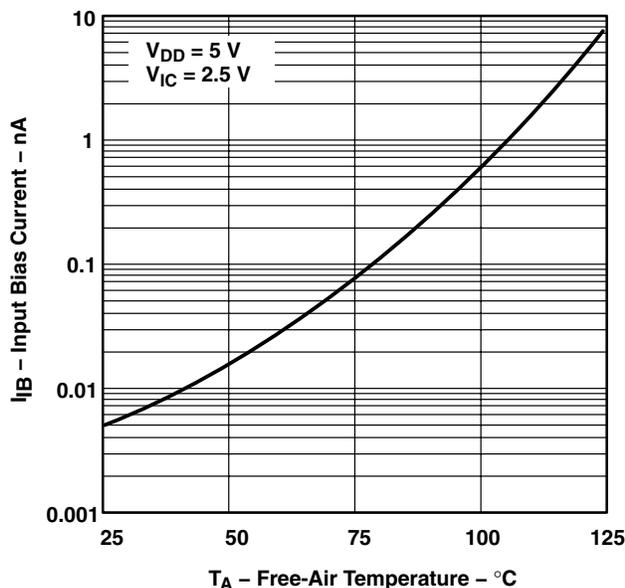
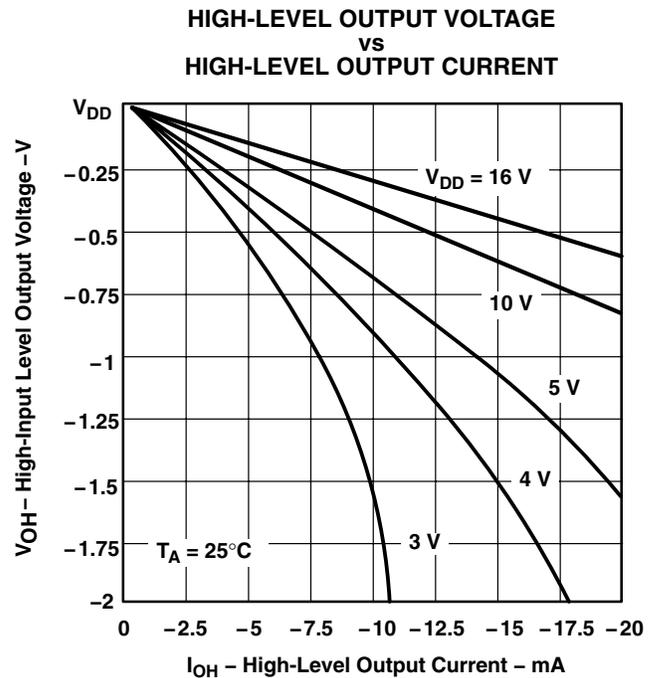
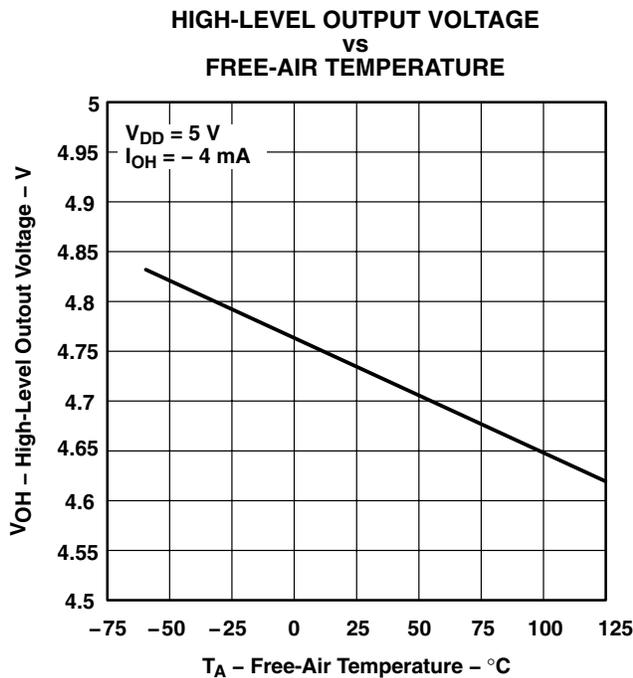
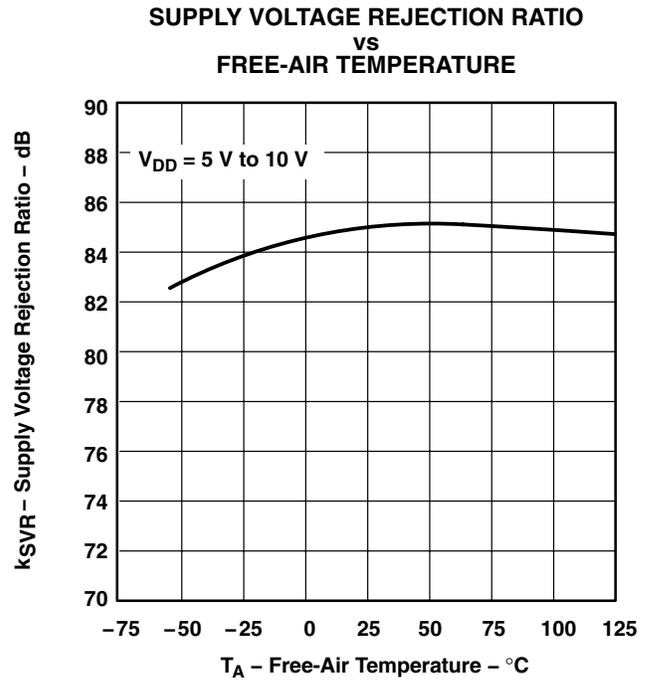
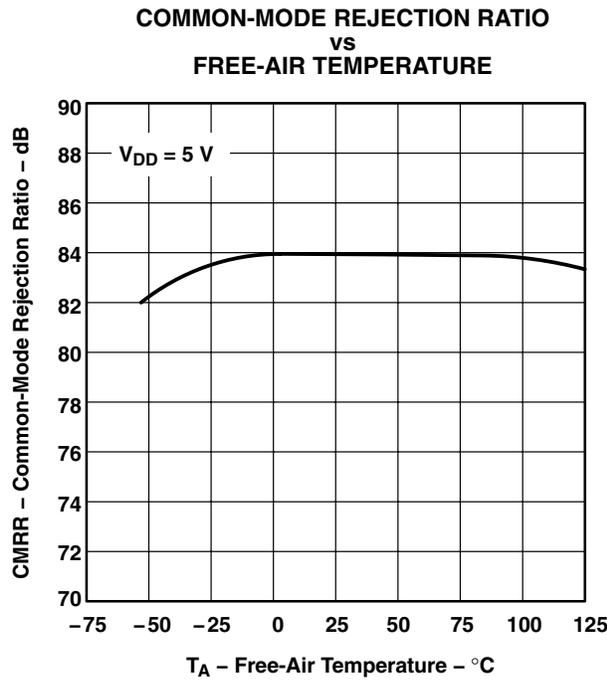


Figure 9

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS†



† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC3704, TLC3704M QUAD MICROPOWER LinCMOS™ VOLTAGE COMPARATORS

SLCS117C – NOVEMBER 1986 – REVISED NOVEMBER 2009

TYPICAL CHARACTERISTICS†

LOW-LEVEL OUTPUT VOLTAGE
vs
LOW-LEVEL OUTPUT CURRENT

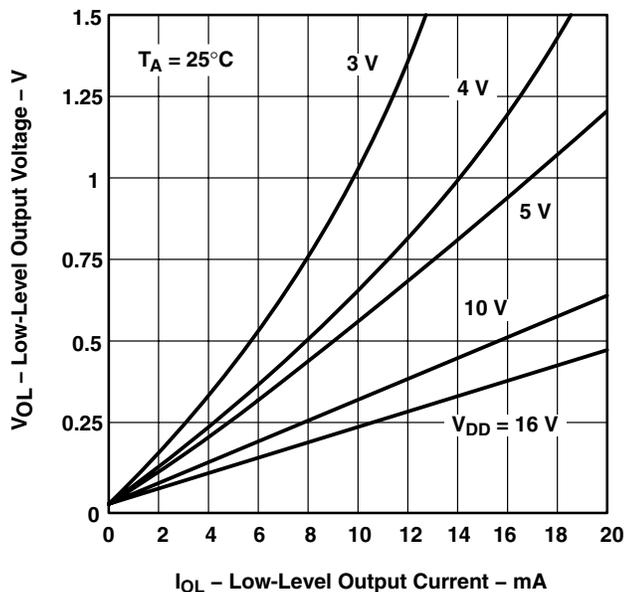


Figure 14

LOW-LEVEL OUTPUT VOLTAGE
vs
FREE-AIR TEMPERATURE

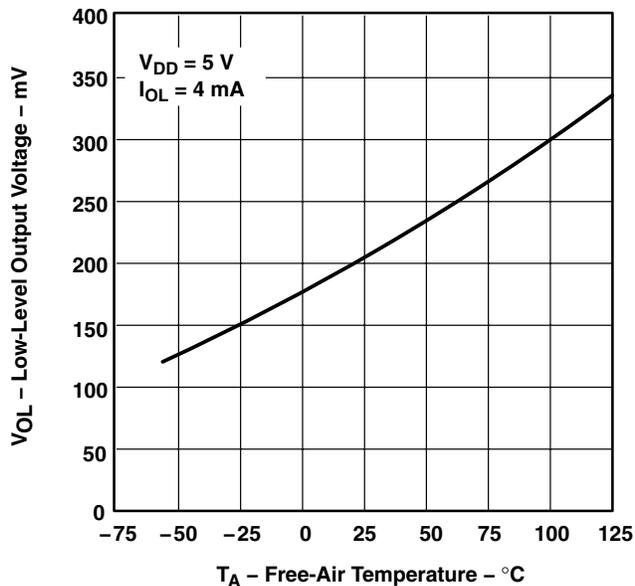


Figure 15

OUTPUT TRANSITION TIME
vs
LOAD CAPACITANCE

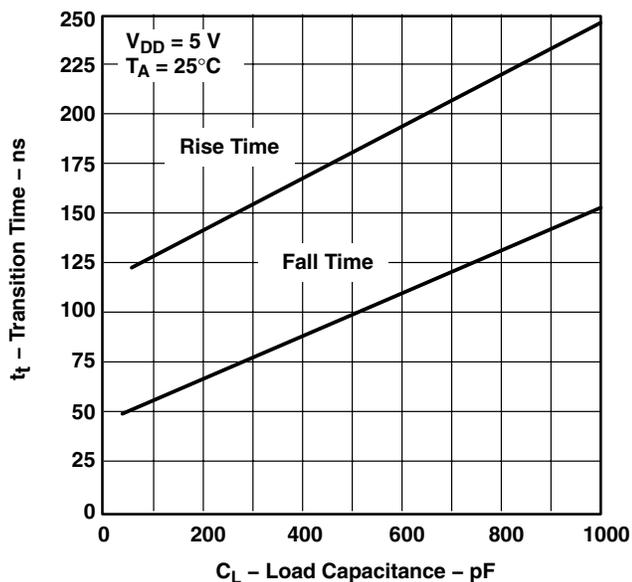


Figure 16

SUPPLY CURRENT RESPONSE
TO AN OUTPUT VOLTAGE TRANSITION

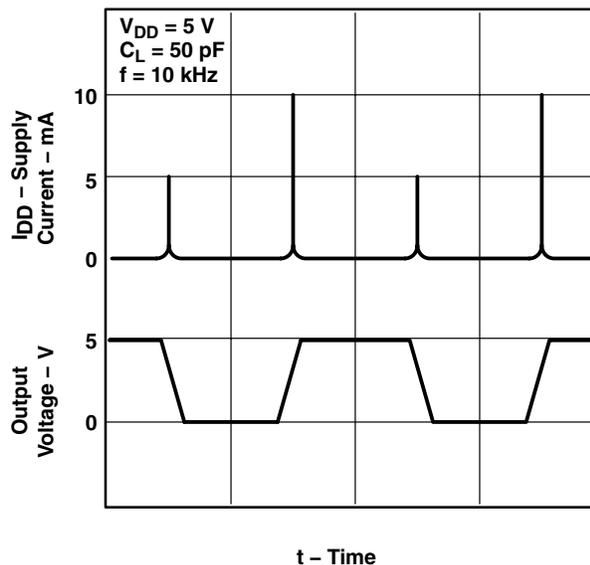


Figure 17

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS

LOW-TO-HIGH-LEVEL OUTPUT RESPONSE
 FOR VARIOUS INPUT OVERDRIVES

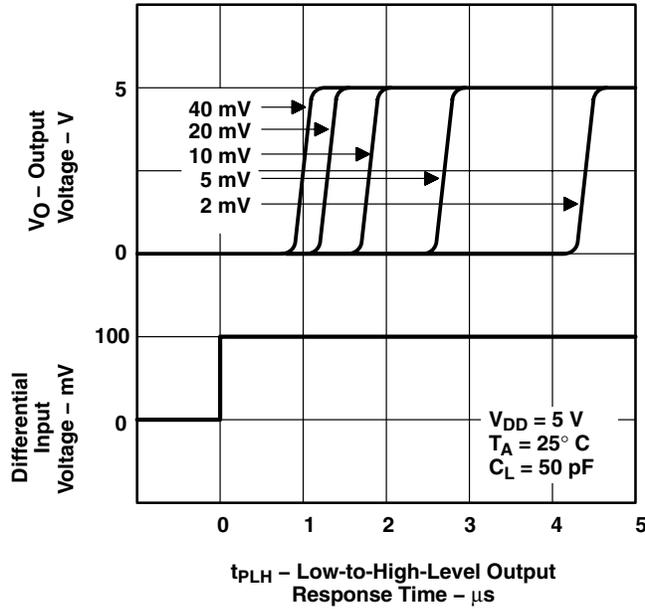


Figure 18

HIGH-TO-LOW-LEVEL OUTPUT RESPONSE
 FOR VARIOUS INPUT OVERDRIVES

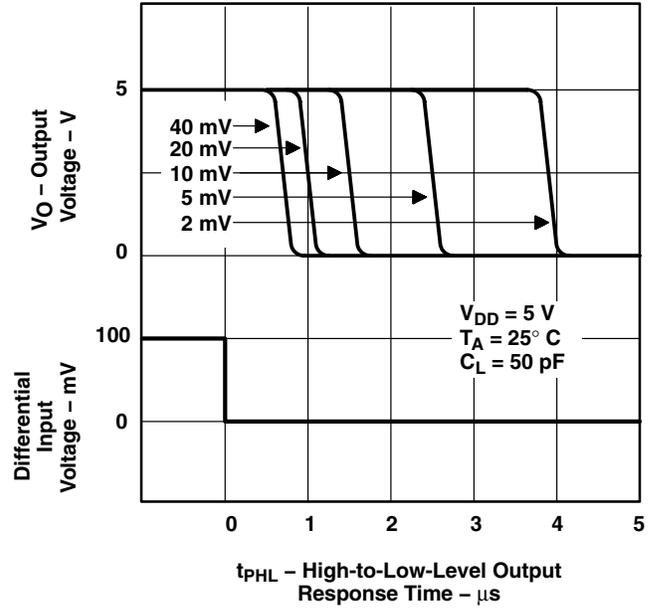


Figure 19

LOW-TO-HIGH-LEVEL
 OUTPUT RESPONSE TIME
 vs
 SUPPLY VOLTAGE

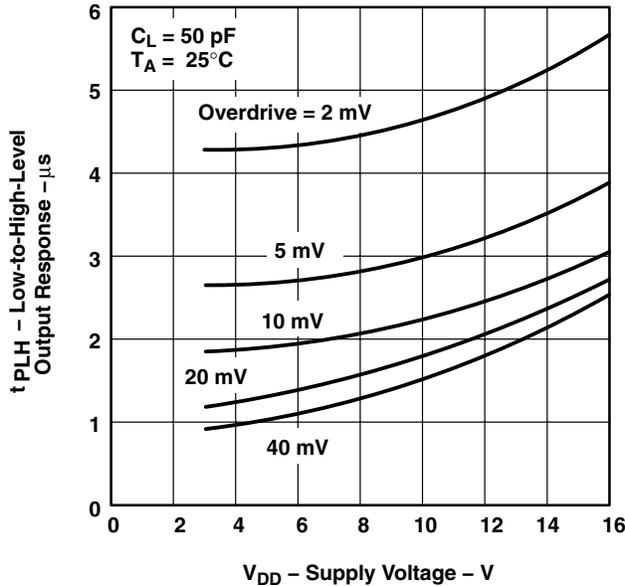


Figure 20

HIGH-TO-LOW-LEVEL
 OUTPUT RESPONSE TIME
 vs
 SUPPLY VOLTAGE

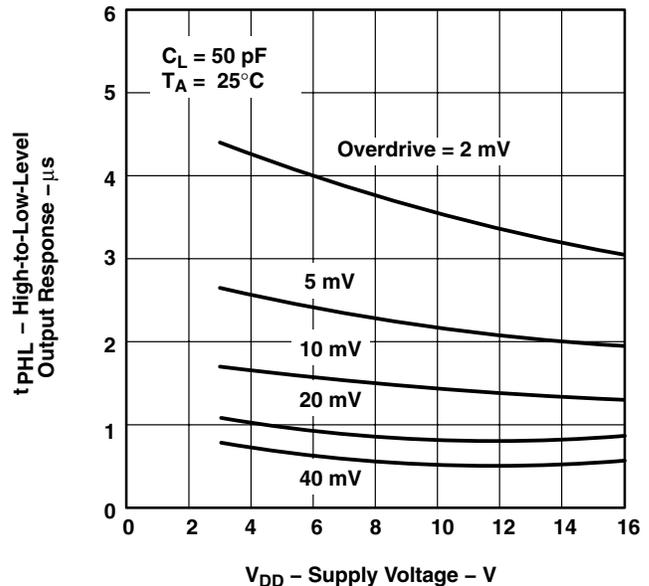


Figure 21

TLC3704, TLC3704M QUAD MICROWPOWER LinCMOS™ VOLTAGE COMPARATORS

SLCS117C – NOVEMBER 1986 – REVISED NOVEMBER 2009

TYPICAL CHARACTERISTICS†

**AVERAGE SUPPLY CURRENT
(PER COMPARATOR)
VS
FREQUENCY**

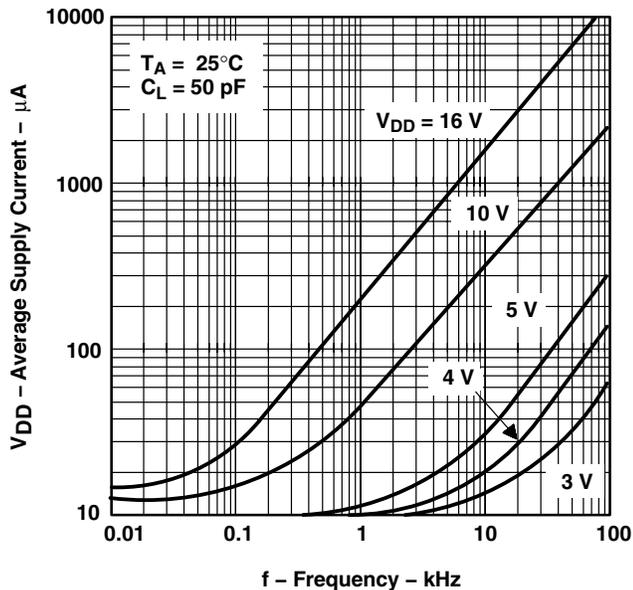


Figure 22

**SUPPLY CURRENT
VS
SUPPLY VOLTAGE**

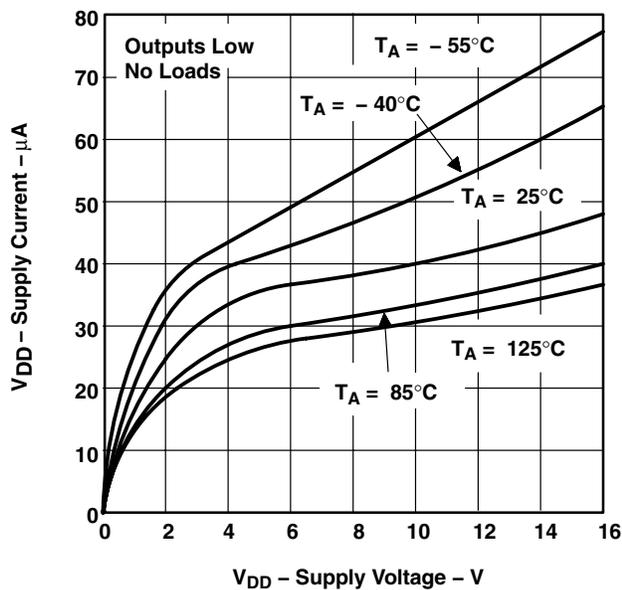


Figure 23

**SUPPLY CURRENT
VS
FREE-AIR TEMPERATURE**

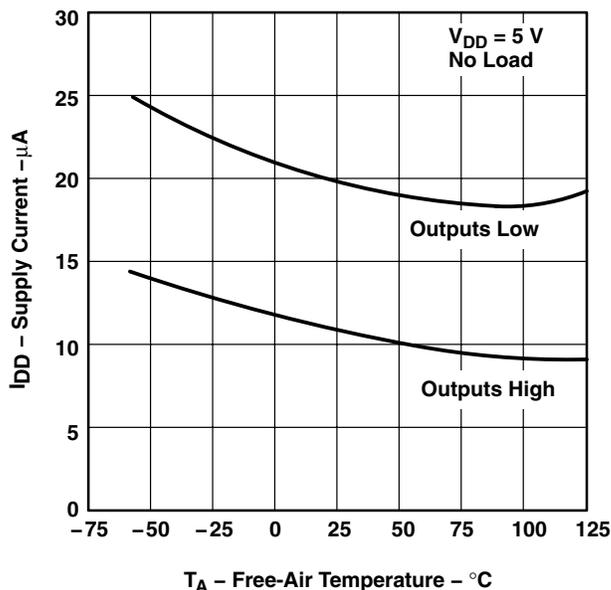


Figure 24

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.



APPLICATION INFORMATION

The inputs should always remain within the supply rails in order to avoid forward biasing the diodes in the electrostatic discharge (ESD) protection structure. If either input exceeds this range, the device is not damaged as long as the input is limited to less than 5 mA. To maintain the expected output state, the inputs must remain within the common-mode range. For example, at 25°C with $V_{DD} = 5\text{ V}$, both inputs must remain between -0.2 V and 4 V to ensure proper device operation. To ensure reliable operation, the supply should be decoupled with a capacitor ($0.1\text{ }\mu\text{F}$) that is positioned as close to the device as possible.

Output and supply current limitations should be watched carefully since the TLC3704 does not provide current protection. For example, each output can source or sink a maximum of 20 mA; however, the total current to ground can only be an absolute maximum of 60 mA. This prohibits sinking 20 mA from each of the four outputs simultaneously since the total current to ground would be 80 mA.

The TLC3704 has internal ESD-protection circuits that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

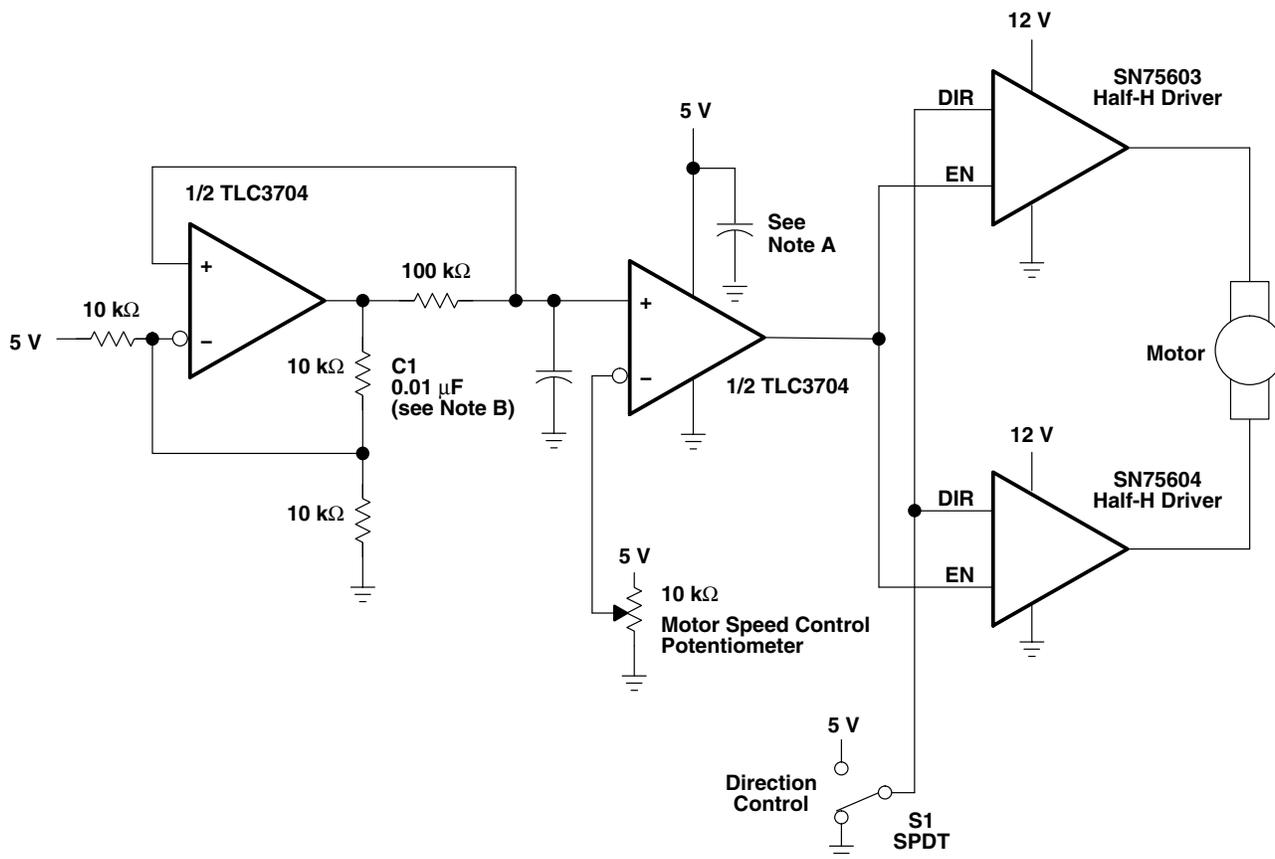
Table of Applications

	FIGURE
Pulse-width-modulated motor speed controller	25
Enhanced supply supervisor	26
Two-phase nonoverlapping clock generator	27
Micropower switching regulator	28

TLC3704, TLC3704M QUAD MICROPOWER LinCMOS™ VOLTAGE COMPARATORS

SLCS117C – NOVEMBER 1986 – REVISED NOVEMBER 2009

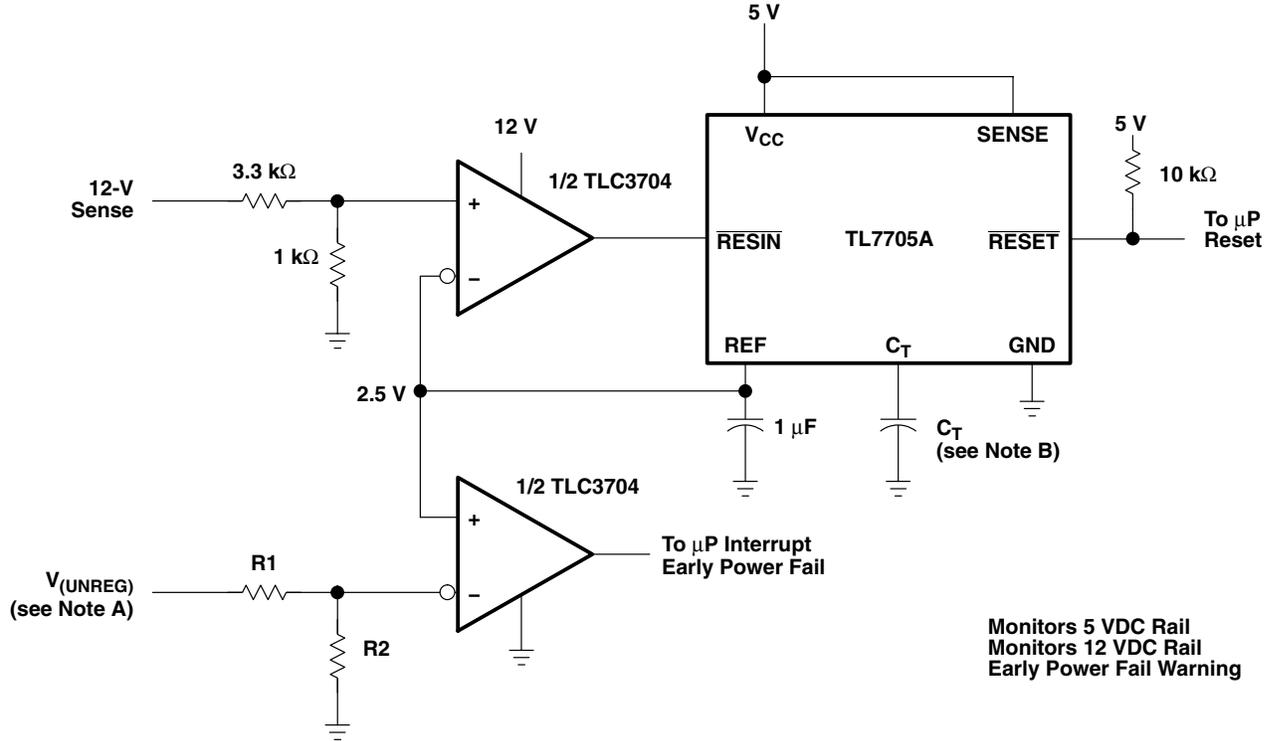
APPLICATION INFORMATION



NOTES: A. The recommended minimum capacitance is 10 μF to eliminate common ground switching noise.
 B. Adjust C1 for change in oscillator frequency

Figure 25. Pulse-Width-Modulated Motor Speed Controller

APPLICATION INFORMATION



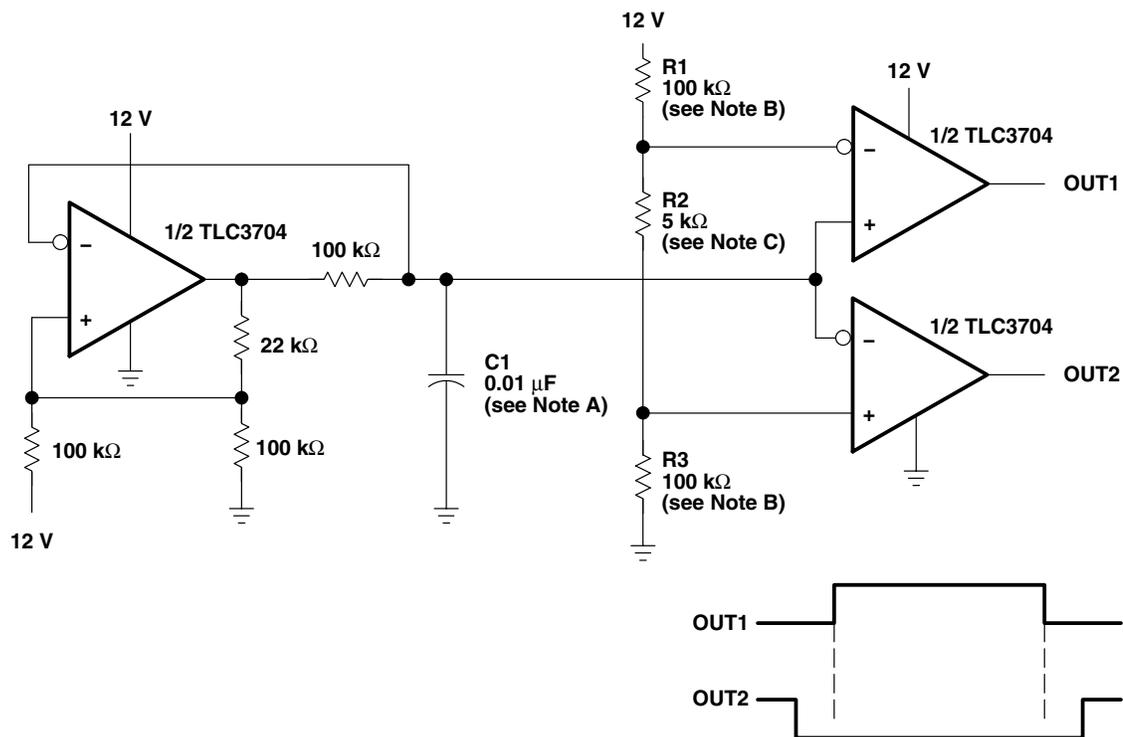
- NOTES: A. $V_{(UNREG)} = 2.5 \frac{(R1 + R2)}{R2}$
 B. The value of C_T determines the time delay of reset.

Figure 26. Enhanced Supply Supervisor

TLC3704, TLC3704M QUAD MICROPOWER LinCMOS™ VOLTAGE COMPARATORS

SLCS117C – NOVEMBER 1986 – REVISED NOVEMBER 2009

APPLICATION INFORMATION



- NOTES: A. Adjust C1 for a change in oscillator frequency where:
 $1/f = 1.85(100 \text{ k}\Omega)C1$
 B. Adjust R1 and R3 to change duty cycle
 C. Adjust R2 to change deadtime

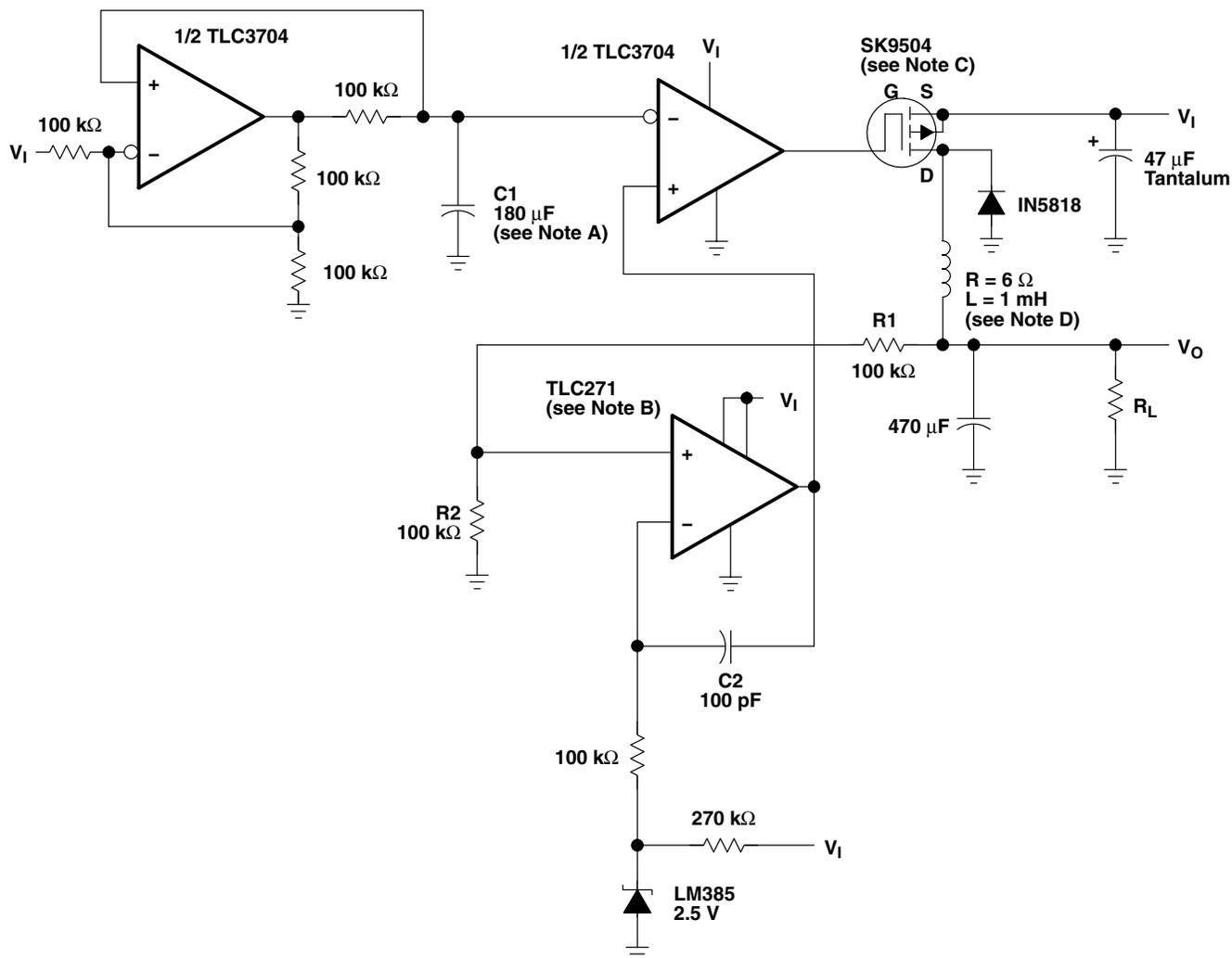
Figure 27. Two-Phase Nonoverlapping Clock Generator

APPLICATION INFORMATION

$$V_I = 6 \text{ V to } 16 \text{ V}$$

$$I_L = 0.01 \text{ mA to } 0.25 \text{ mA}$$

$$V_O = 2.5 \frac{(R1 + R2)}{R2}$$



- NOTES: A. Adjust $C1$ for a change in oscillator frequency
 B. TLC271 – Tie pin 8 to pin 7 for low bias operation
 C. SK9504 – $V_{DS} = 40 \text{ V}$
 $I_{DS} = 1 \text{ A will}$
 D. To achieve microampere current drive, the inductance of the circuit must be increased.

Figure 28. Micropower Switching Regulator

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
5962-9096901M2A	Active	Production	LCCC (FK) 20	55 TUBE	No	SNPB	N/A for Pkg Type	-55 to 125	5962-9096901M2A TLC3704 MFKB
5962-9096901MCA	Active	Production	CDIP (J) 14	25 TUBE	No	SNPB	N/A for Pkg Type	-55 to 125	5962-9096901MCA A TLC3704MJB
TLC3704CD	Obsolete	Production	SOIC (D) 14	-	-	Call TI	Call TI	0 to 70	TLC3704C
TLC3704CDR	Active	Production	SOIC (D) 14	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	TLC3704C
TLC3704CDR.A	Active	Production	SOIC (D) 14	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	TLC3704C
TLC3704CN	Active	Production	PDIP (N) 14	25 TUBE	Yes	NIPDAU	N/A for Pkg Type	0 to 70	TLC3704CN
TLC3704CN.A	Active	Production	PDIP (N) 14	25 TUBE	Yes	NIPDAU	N/A for Pkg Type	0 to 70	TLC3704CN
TLC3704CNSR	Active	Production	SOP (NS) 14	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	TLC3704
TLC3704CNSR.A	Active	Production	SOP (NS) 14	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	TLC3704
TLC3704CPW	Obsolete	Production	TSSOP (PW) 14	-	-	Call TI	Call TI	0 to 70	P3704
TLC3704CPWR	Active	Production	TSSOP (PW) 14	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	P3704
TLC3704CPWR.A	Active	Production	TSSOP (PW) 14	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	P3704
TLC3704ID	Obsolete	Production	SOIC (D) 14	-	-	Call TI	Call TI	-40 to 85	TLC3704I
TLC3704IDR	Active	Production	SOIC (D) 14	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TLC3704I
TLC3704IDR.A	Active	Production	SOIC (D) 14	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TLC3704I
TLC3704IDRG4	Active	Production	SOIC (D) 14	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-55 to 125	TLC3704I
TLC3704IDRG4.A	Active	Production	SOIC (D) 14	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-55 to 125	TLC3704I
TLC3704IN	Active	Production	PDIP (N) 14	25 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	TLC3704IN
TLC3704IN.A	Active	Production	PDIP (N) 14	25 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	TLC3704IN
TLC3704IPW	Obsolete	Production	TSSOP (PW) 14	-	-	Call TI	Call TI	-40 to 85	P3704I
TLC3704IPWR	Active	Production	TSSOP (PW) 14	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	P3704I
TLC3704IPWR.A	Active	Production	TSSOP (PW) 14	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	P3704I
TLC3704MD	Obsolete	Production	SOIC (D) 14	-	-	Call TI	Call TI	-55 to 125	

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
TLC3704MFKB	Active	Production	LCCC (FK) 20	55 TUBE	No	SNPB	N/A for Pkg Type	-55 to 125	5962-9096901M2A TLC3704 MFKB
TLC3704MFKB.A	Active	Production	LCCC (FK) 20	55 TUBE	No	SNPB	N/A for Pkg Type	-55 to 125	5962-9096901M2A TLC3704 MFKB
TLC3704MJB	Active	Production	CDIP (J) 14	25 TUBE	No	SNPB	N/A for Pkg Type	-55 to 125	5962-9096901MC A TLC3704MJB
TLC3704MJB.A	Active	Production	CDIP (J) 14	25 TUBE	No	SNPB	N/A for Pkg Type	-55 to 125	5962-9096901MC A TLC3704MJB

(1) **Status:** For more details on status, see our [product life cycle](#).

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "-" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative

and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

OTHER QUALIFIED VERSIONS OF TLC3704, TLC3704M :

- Catalog : [TLC3704](#)
- Automotive : [TLC3704-Q1](#), [TLC3704-Q1](#)
- Military : [TLC3704M](#)

NOTE: Qualified Version Definitions:

- Catalog - TI's standard catalog product
- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects
- Military - QML certified for Military and Defense Applications

TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

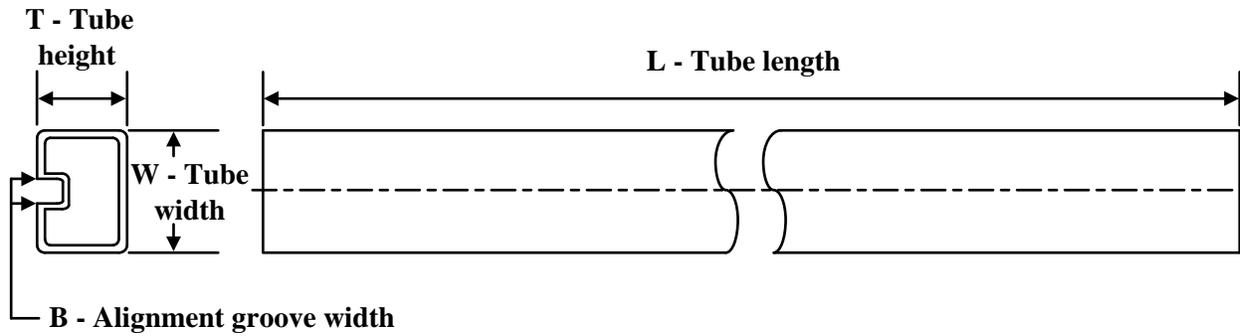

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLC3704CDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC3704CNSR	SOP	NS	14	2000	330.0	16.4	8.45	10.55	2.5	12.0	16.2	Q1
TLC3704CPWR	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TLC3704IDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC3704IDRG4	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC3704IPWR	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLC3704CDR	SOIC	D	14	2500	353.0	353.0	32.0
TLC3704CNSR	SOP	NS	14	2000	353.0	353.0	32.0
TLC3704CPWR	TSSOP	PW	14	2000	353.0	353.0	32.0
TLC3704IDR	SOIC	D	14	2500	353.0	353.0	32.0
TLC3704IDRG4	SOIC	D	14	2500	353.0	353.0	32.0
TLC3704IPWR	TSSOP	PW	14	2000	353.0	353.0	32.0

TUBE


*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
5962-9096901M2A	FK	LCCC	20	55	506.98	12.06	2030	NA
TLC3704CN	N	PDIP	14	25	506	13.97	11230	4.32
TLC3704CN.A	N	PDIP	14	25	506	13.97	11230	4.32
TLC3704IN	N	PDIP	14	25	506	13.97	11230	4.32
TLC3704IN.A	N	PDIP	14	25	506	13.97	11230	4.32
TLC3704MFKB	FK	LCCC	20	55	506.98	12.06	2030	NA
TLC3704MFKB.A	FK	LCCC	20	55	506.98	12.06	2030	NA

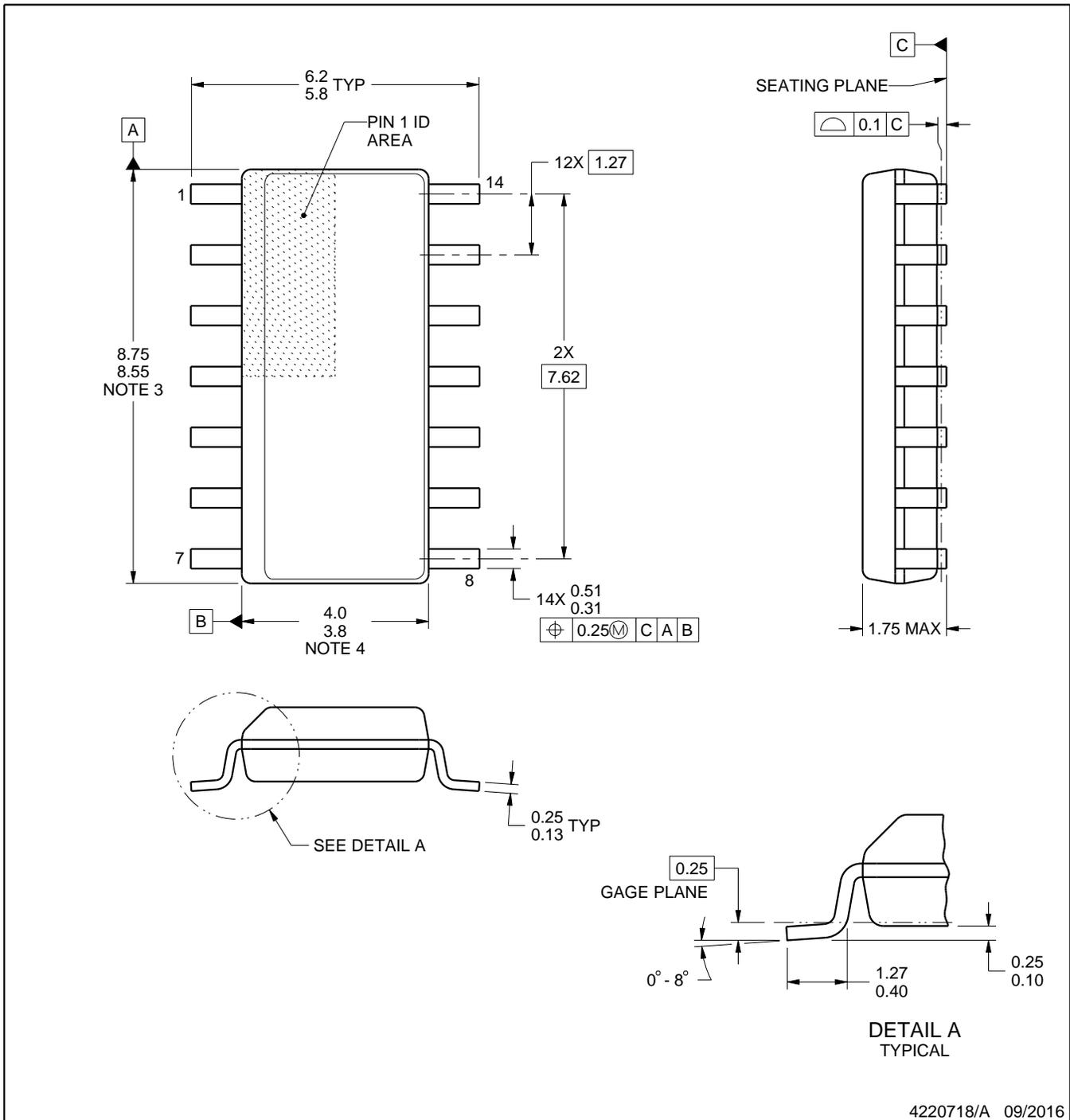
D0014A



PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES:

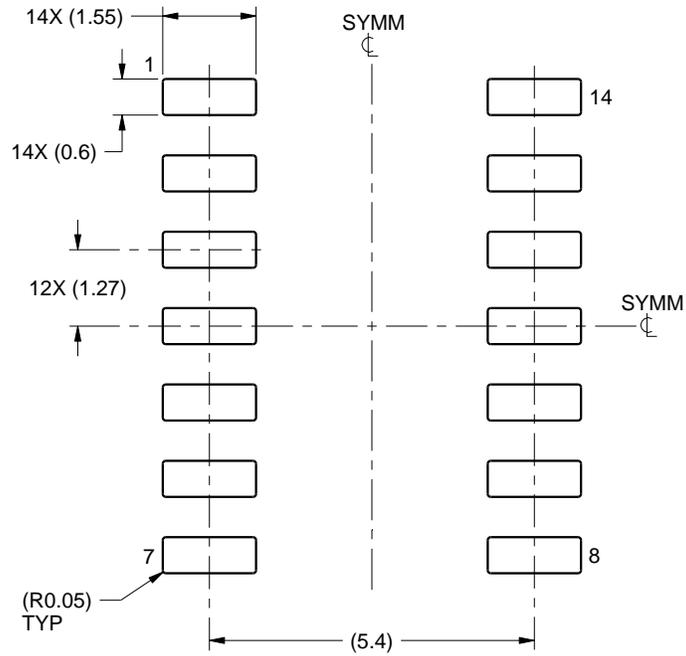
1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.43 mm, per side.
5. Reference JEDEC registration MS-012, variation AB.

EXAMPLE BOARD LAYOUT

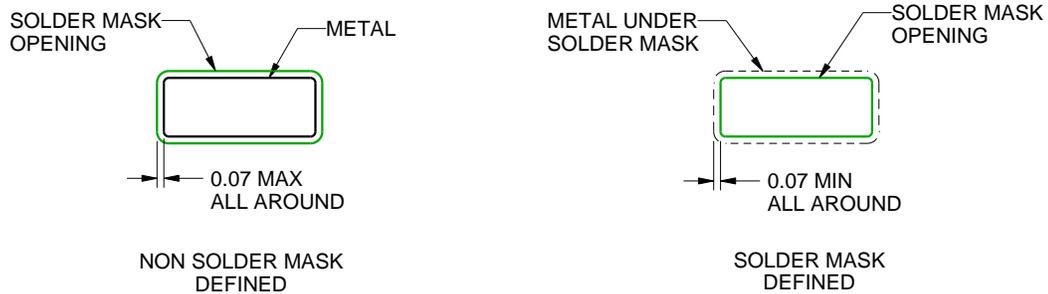
D0014A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE
SCALE:8X



SOLDER MASK DETAILS

4220718/A 09/2016

NOTES: (continued)

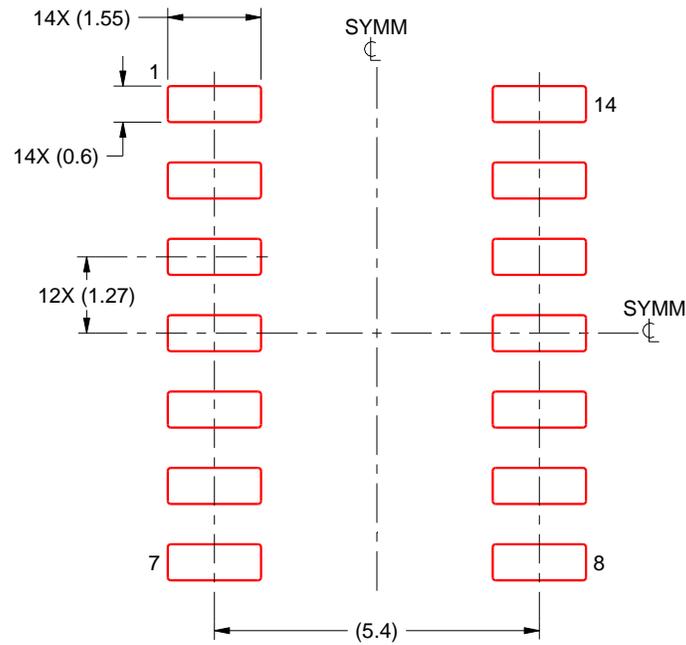
- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

D0014A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE:8X

4220718/A 09/2016

NOTES: (continued)

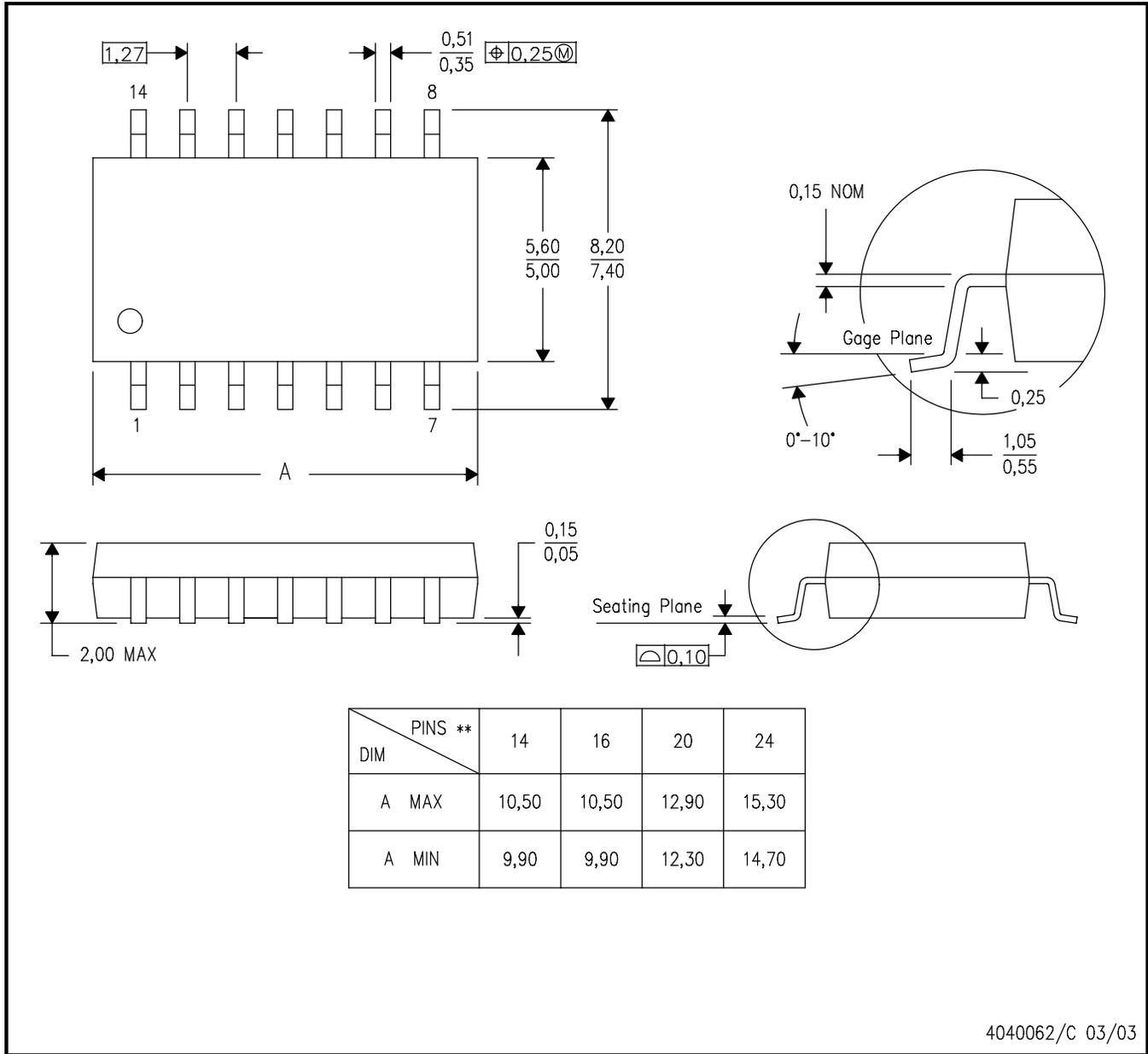
8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

MECHANICAL DATA

NS (R-PDSO-G**)

PLASTIC SMALL-OUTLINE PACKAGE

14-PINS SHOWN



- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Body dimensions do not include mold flash or protrusion, not to exceed 0,15.

GENERIC PACKAGE VIEW

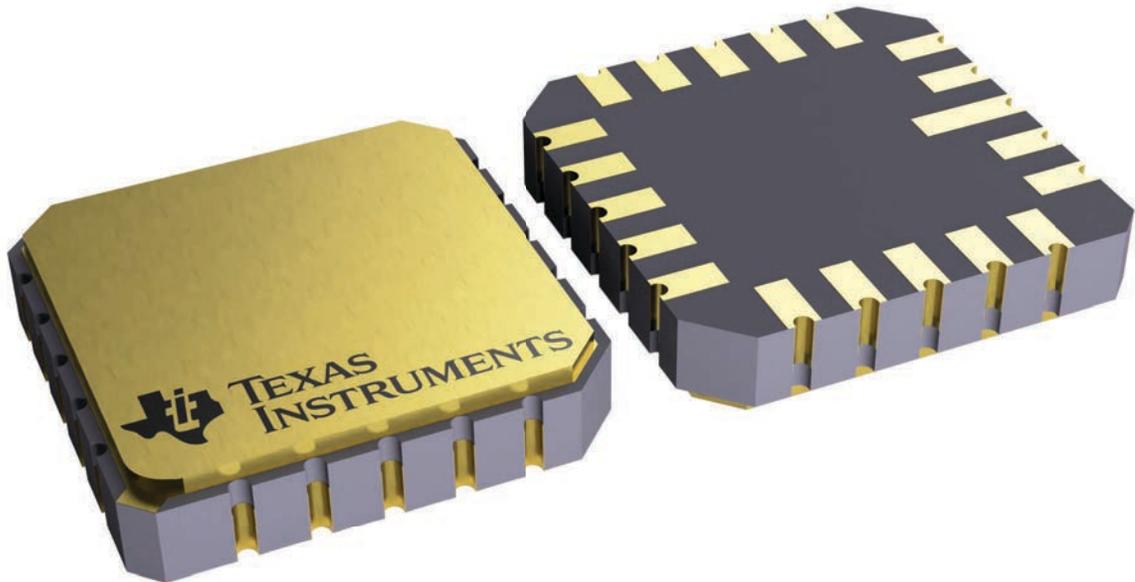
FK 20

LCCC - 2.03 mm max height

8.89 x 8.89, 1.27 mm pitch

LEADLESS CERAMIC CHIP CARRIER

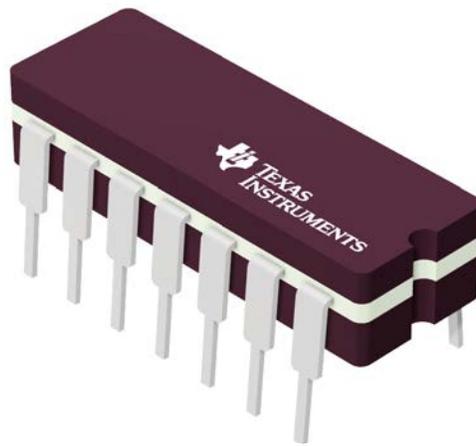
This image is a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.



4229370VA\

J 14

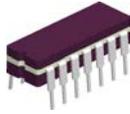
GENERIC PACKAGE VIEW
CDIP - 5.08 mm max height
CERAMIC DUAL IN LINE PACKAGE



Images above are just a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.

4040083-5/G

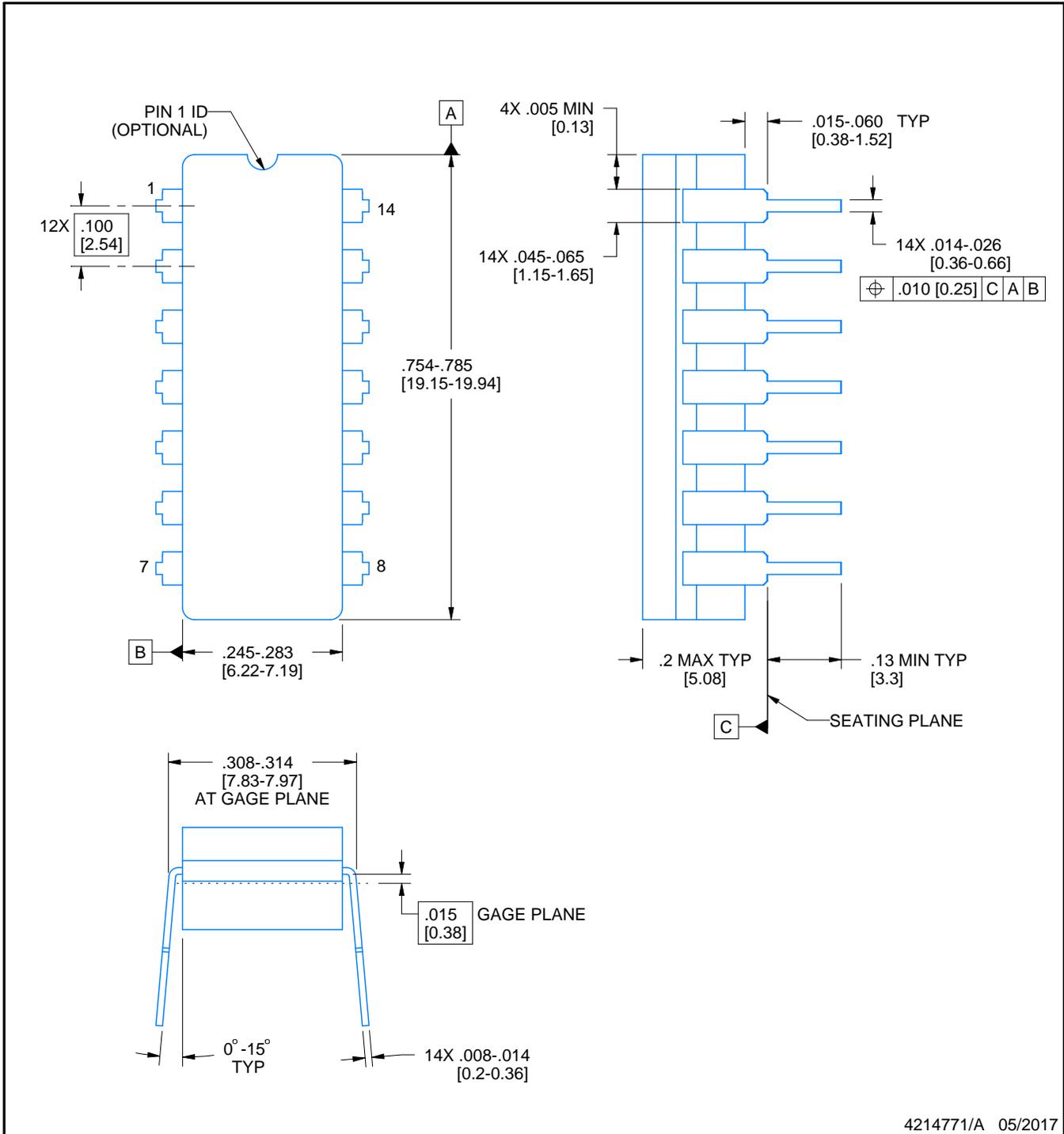
J0014A



PACKAGE OUTLINE

CDIP - 5.08 mm max height

CERAMIC DUAL IN LINE PACKAGE



4214771/A 05/2017

NOTES:

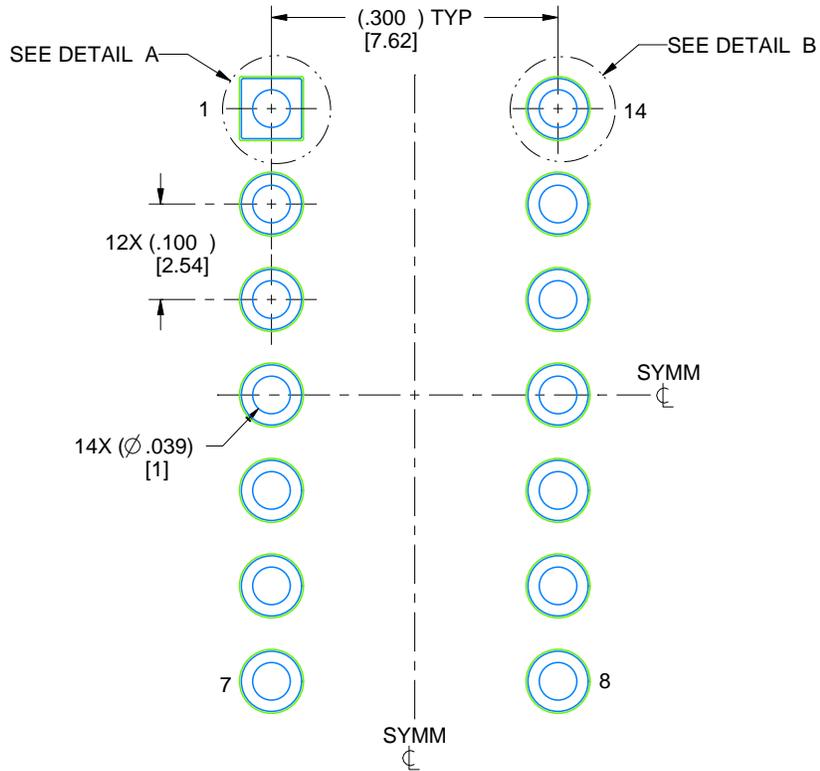
1. All controlling linear dimensions are in inches. Dimensions in brackets are in millimeters. Any dimension in brackets or parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This package is hermetically sealed with a ceramic lid using glass frit.
4. Index point is provided on cap for terminal identification only and on press ceramic glass frit seal only.
5. Falls within MIL-STD-1835 and GDIP1-T14.

EXAMPLE BOARD LAYOUT

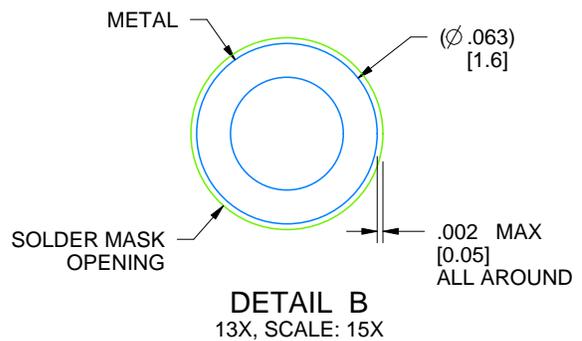
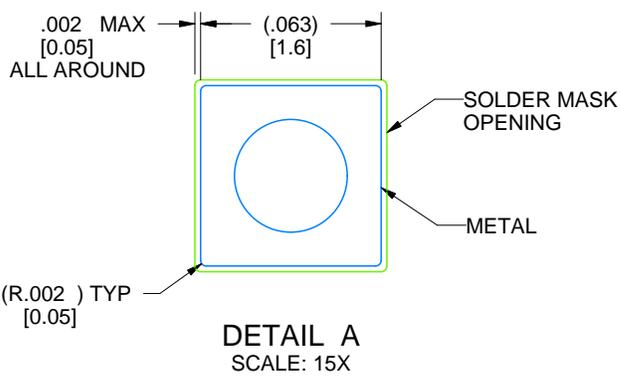
J0014A

CDIP - 5.08 mm max height

CERAMIC DUAL IN LINE PACKAGE



LAND PATTERN EXAMPLE
NON-SOLDER MASK DEFINED
SCALE: 5X

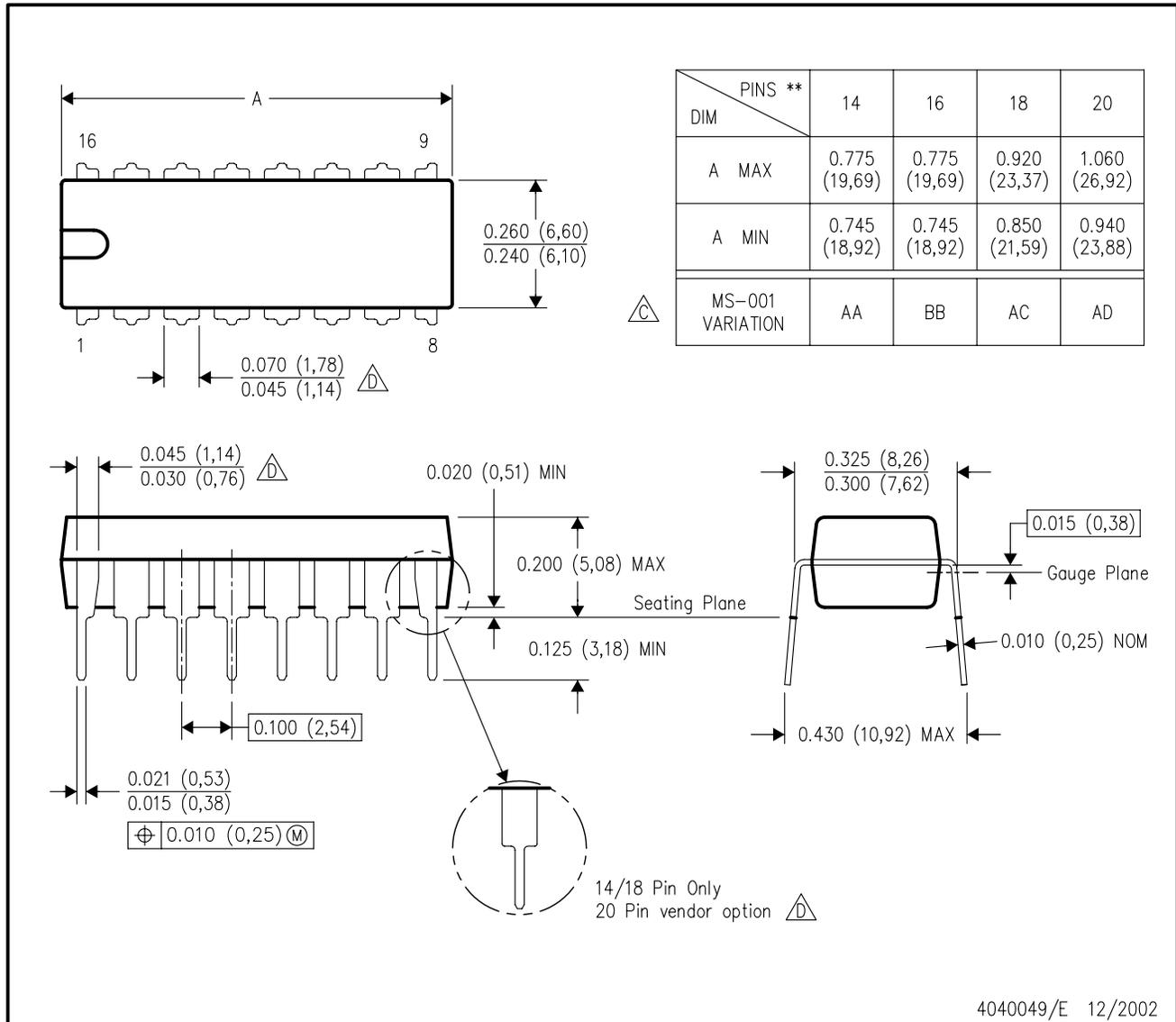


4214771/A 05/2017

N (R-PDIP-T**)

PLASTIC DUAL-IN-LINE PACKAGE

16 PINS SHOWN



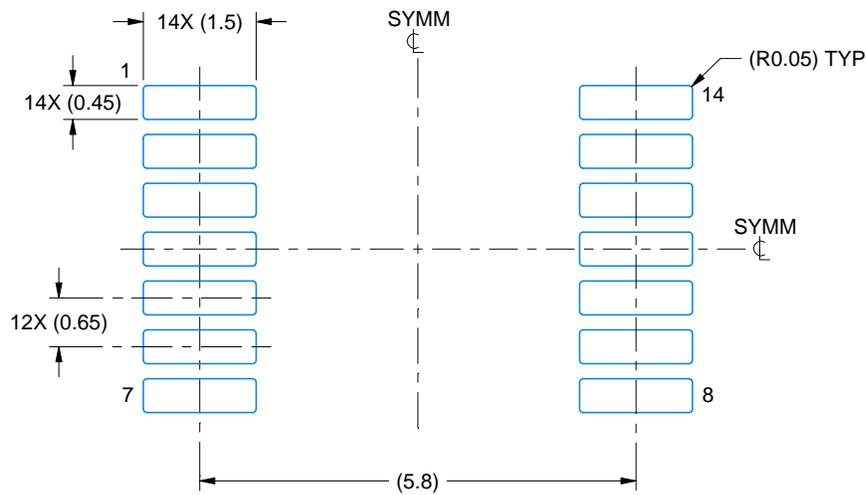
- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - $\triangle C$ Falls within JEDEC MS-001, except 18 and 20 pin minimum body length (Dim A).
 - $\triangle D$ The 20 pin end lead shoulder width is a vendor option, either half or full width.

EXAMPLE BOARD LAYOUT

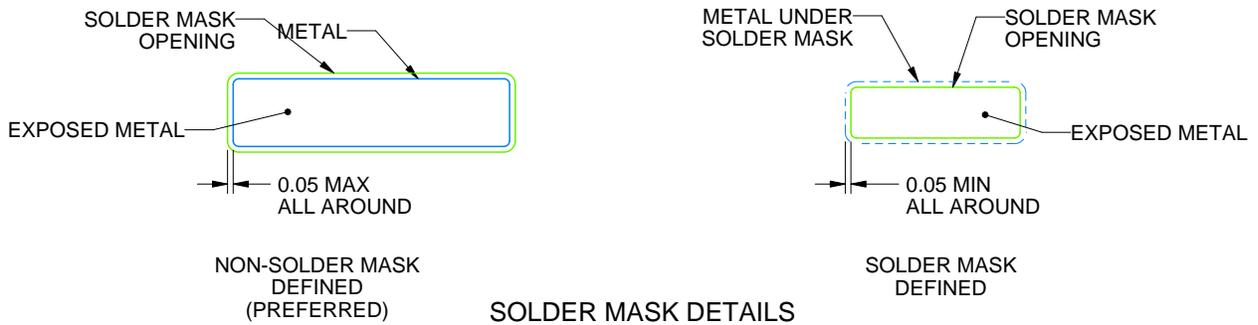
PW0014A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 10X



4220202/B 12/2023

NOTES: (continued)

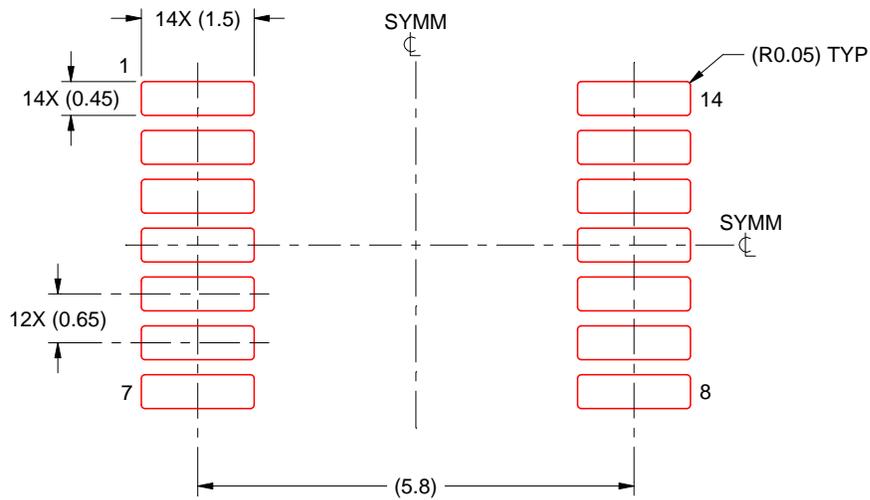
- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

PW0014A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE: 10X

4220202/B 12/2023

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

IMPORTANT NOTICE AND DISCLAIMER

TI PROVIDES TECHNICAL AND RELIABILITY DATA (INCLUDING DATASHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS AND IMPLIED, INCLUDING WITHOUT LIMITATION ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE OR NON-INFRINGEMENT OF THIRD PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for skilled developers designing with TI products. You are solely responsible for (1) selecting the appropriate TI products for your application, (2) designing, validating and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, regulatory or other requirements.

These resources are subject to change without notice. TI grants you permission to use these resources only for development of an application that uses the TI products described in the resource. Other reproduction and display of these resources is prohibited. No license is granted to any other TI intellectual property right or to any third party intellectual property right. TI disclaims responsibility for, and you fully indemnify TI and its representatives against any claims, damages, costs, losses, and liabilities arising out of your use of these resources.

TI's products are provided subject to [TI's Terms of Sale](#), [TI's General Quality Guidelines](#), or other applicable terms available either on ti.com or provided in conjunction with such TI products. TI's provision of these resources does not expand or otherwise alter TI's applicable warranties or warranty disclaimers for TI products. Unless TI explicitly designates a product as custom or customer-specified, TI products are standard, catalog, general purpose devices.

TI objects to and rejects any additional or different terms you may propose.

Copyright © 2026, Texas Instruments Incorporated

Last updated 10/2025