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- Floating Bootstrap or Ground-Reference High-Side Driver
- Adaptive Dead-Time Control
- 50-ns Max Rise/Fall Times and 100-ns Max Propagation Delay – 3.3-nF Load
- Ideal for High-Current Single or Multiphase Power Supplies
- 2.4-A Typical Peak Output Current
- 4.5-V to 15-V Supply Voltage Range
- Internal Schottky Bootstrap Diode
- Low Supply Current....3-mA Typical
- -40°C to 125°C Operating Virtual Junction Temperature
- Available in SOIC Package

description

The TPS2832 and TPS2833 are MOSFET drivers for synchronous-buck power stages. These devices are ideal for designing a high-performance power supply using switching controllers that do not have MOSFET drivers. The drivers are designed to deliver 2.4-A peak currents into large capacitive loads. The high-side driver can be configured as a ground-reference driver or as a floating bootstrap driver. An adaptive dead-time control circuit eliminates shoot-through currents through the main power FETs during switching transitions and provides high efficiency for the buck regulator.

The TPS2832 has a noninverting input. The TPS2833 has an inverting input. The TPS2832/33 drivers, available in 8-terminal SOIC packages, operate over a junction temperature range of –40°C to 125°C.

	PACKAGED DEVICES
ТJ	SOIC (D)
–40°C to 125°C	TPS2832D TPS2833D

AVAILABLE OPTIONS

The D package is available taped and reeled. Add R suffix to device type (e.g., TPS2832DR)

Related Synchronous MOSFET Drivers

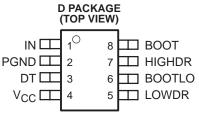
DEVICE NAME	ADDITIONAL FEATURES	INPUTS			
TPS2830		Noninverted			
TPS2831	ENABLE, SYNC and CROWBAR	CMOS	Inverted		
TPS2834			Noninverted		
TPS2835	ENABLE, SYNC and CROWBAR	TTL	Inverted		
TPS2836			Noninverted		
TPS2837	W/O ENABLE, SYNC and CROWBAR	TTL	Inverted		



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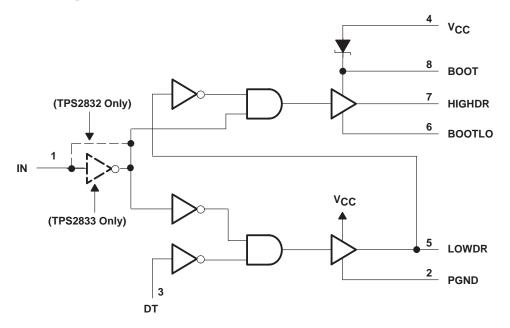
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TPS2832, TPS2833 FAST SYNCHRONOUS-BUCK MOSFET DRIVERS WITH DEAD-TIME CONTROL SLVS195C – FEBRUARY 1999 – REVISED JANUARY 2001

functional block diagram



Terminal Functions

TERMIN	NAL		DECODIDETION
NAME	NO.	1/0	DESCRIPTION
BOOT	8	I	Bootstrap terminal. A ceramic capacitor is connected between BOOT and BOOTLO terminals to develop the floating bootstrap voltage for the high-side MOSFET. The capacitor value is typically between 0.1 μ F and 1 μ F. A 1-M Ω resistor should be connected across the bootstrap capacitor to provide a discharge path when the driver has been powered down.
BOOTLO	6	0	This terminal connects to the junction of the high-side and low-side MOSFETs.
DT	3	Ι	Dead-time control terminal. Connect DT to the junction of the high-side and low-side MOSFETs
HIGHDR	7	0	Output drive for the high-side power MOSFET
IN	1	Т	Input signal to the MOSFET drivers (noninverting input for the TPS2832; inverting input for the TPS2833).
LOWDR	5	0	Output drive for the low-side power MOSFET
PGND	2		Power ground. Connect to the FET power ground.
VCC	4	Ι	Input supply. Recommended that a 1 μF capacitor be connected from $V_{\mbox{CC}}$ to PGND.



detailed description

low-side driver

The low-side driver is designed to drive low Rds(on) N-channel MOSFETs. The current rating of the driver is 2 A, source and sink.

high-side driver

The high-side driver is designed to drive low Rds(on) N-channel MOSFETs. The current rating of the driver is 2 A, source and sink. The high-side driver can be configured as a ground-reference driver or a floating bootstrap driver. The internal bootstrap diode, is a Schottky for improved drive efficiency. The maximum voltage that can be applied between the BOOT terminal and ground is 30 V.

dead-time (DT) control[†]

Dead-time control prevents shoot through current from flowing through the main power FETs during switching transitions by controlling the turn-on times of the MOSFET drivers. The high-side driver is not allowed to turn on until the gate drive voltage to the low-side FET is low, and the low-side driver is not allowed to turn on until the voltage at the junction of the power FETs (Vdrain) is low; the DT terminal connects to the junction of the power FETs.

IN[†]

The IN terminal is a digital terminal that is the input control signal for the drivers. The TPS2832 has a noninverting input; the TPS2833 has an inverting input.

[†]High-level input voltages on IN and DT must be greater than or equal to $0.7 V_{CC}$.



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absolute maximum ratings over operating free-air temperature (unless otherwise noted)[†]

Supply voltage range, V _{CC} (see Note 1)
Input voltage range: BOOT to PGND (high-side driver ON)
BOOTLO to PGND
BOOT to BOOTLO
IN (see Note 2)
DT (see Note 2)
Continuous total power dissipation
Operating virtual junction temperature range, T _J
Storage temperature range, T _{stg}
Lead temperature soldering 1,6 mm (1/16 inch) from case for 10 seconds

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTES: 1. Unless otherwise specified, all voltages are with respect to PGND.

2. High-level input voltages on the IN and DT terminals must be less than or equal to V_{CC}.

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C	DERATING FACTOR	T _A = 70°C	T _A = 85°C
	POWER RATING	ABOVE T _A = 25°C	POWER RATING	POWER RATING
D	600 mW	6.0 mW/°C	330 mW	240 mW

recommended operating conditions

		MIN	NOM MAX	UNIT
Supply voltage,	Vcc	4.5	15	V
Input voltage	BOOT to PGND	4.5	28	V

electrical characteristics over recommended operating virtual junction temperature range, $V_{CC} = 6.5 \text{ V}$, $C_L = 3.3 \text{ nF}$ (unless otherwise noted)

supply current

	PARAMETER	TEST CO	MIN	TYP	MAX	UNIT	
	Supply voltage range			4.5		15	V
		V _{CC} =15 V				100	μΑ
Vcc		V _{CC} =12 V, f _{SWX} = 200 kHz, C _{HIGHDR} = 50 pF,	BOOTLO grounded, C _{LOWDR} = 50 pF, See Note 3		3		mA

NOTE 3: Ensured by design, not production tested.



electrical characteristics over recommended operating virtual junction temperature range, $V_{CC} = 6.5 \text{ V}$, $C_L = 3.3 \text{ nF}$ (unless otherwise noted) (continued)

output drivers

High-side sink (see Note 4)Duty cycle < 2%, tpw < 100 μ s (see Note 3)High-side sourceDuty cycle < 2%, tpw < 100 μ s (see Note 3)			TEST CONDIT	IONS	MIN	TYP	MAX	UNIT	
		Duty cycle < 2%.	VBOOT - VBOOTLO = 4.5 V	VHIGHDR = 4 V	0.7	1.1			
	0	t _{pw} < 100 μs	VBOOT - VBOOTLO = 6.5 V	VHIGHDR = 5 V	1.1	1.5		А	
		(see Note 3)	V _{BOOT} – V _{BOOTLO} = 12 V,	V _{HIGHDR} = 10.5 V	2	2.4			
	Hiah-side	Dutv cvcle < 2%.	V _{BOOT} - V _{BOOTLO} = 4.5 V	V _{HIGHDR} = 0.5V	1.2	1.4			
			V _{BOOT} - V _{BOOTLO} = 6.5 V	V _{HIGHDR} = 1.5 V	1.3	1.6		А	
Peak output-	(see Note 4)	(see Note 3)	$V_{BOOT} - V_{BOOTLO} = 12 V,$	V _{HIGHDR} = 1.5 V	2.3	2.7			
current	I and at the state	Duty cycle < 2%,	V _{CC} = 4.5 V,	$V_{LOWDR} = 4 V$	1.3	1.8			
	Low-side sink (see Note 4)	t _{pw} < 100 μs	$V_{CC} = 6.5 V,$	V _{LOWDR} = 5 V	2	2.5		А	
		(see Note 3)	V _{CC} = 12 V,	$V_{LOWDR} = 10.5 V$	3	3.5			
	Low-side source (see Note 4)	Duty cycle < 2%,	V _{CC} = 4.5 V,	$V_{LOWDR} = 0.5V$	1.4	1.7			
		t _{pw} < 100 μs	$V_{CC} = 6.5 V,$	$V_{LOWDR} = 1.5 V$	2	2.4		А	
		(see Note 3)	V _{CC} = 12 V,	$V_{LOWDR} = 1.5 V$	2.5	3			
			VBOOT - VBOOTLO = 4.5 V	VHIGHDR = 0.5 V			5		
	High-side sink (see Note 4)		VBOOT - VBOOTLO = 6.5 V	VHIGHDR = 0.5 V	VHIGHDR = 0.5 V 5				
			VBOOT - VBOOTLO = 12 V,	V _{HIGHDR} = 0.5 V			5		
			VBOOT - VBOOTLO = 4.5 V	VHIGHDR = 4 V			75		
	High-side source	(see Note 4)	$V_{BOOT} - V_{BOOTLO} = 6.5 V_{e}$	V _{HIGHDR} = 6 V			75	Ω	
Output			$V_{BOOT} - V_{BOOTLO} = 12 V,$	V _{HIGHDR} =11.5 V			75		
resistance			V _{DRV} = 4.5 V,	$V_{LOWDR} = 0.5 V$			9		
	Low-side sink (se	ee Note 4)	V _{DRV} = 6.5 V	$V_{LOWDR} = 0.5 V$			7.5	Ω	
			V _{DRV} = 12 V,	$V_{LOWDR} = 0.5 V$			6		
			V _{DRV} = 4.5 V,	$V_{LOWDR} = 4 V$			75		
	Low-side source	(see Note 4)	V _{DRV} = 6.5 V,	$V_{LOWDR} = 6 V$			75	Ω	
			V _{DRV} = 12 V,	V _{LOWDR} = 11.5 V			75	<u> </u>	

NOTES: 3. Ensured by design, not production tested.

4. The pull-up/pull-down circuits of the drivers are bipolar and MOSFET transistors in parallel. The peak output current rating is the combined current from the bipolar and MOSFET transistors. The output resistance is the Rds(on) of the MOSFET transistor when the voltage on the driver output is less than the saturation voltage of the bipolar transistor.

dead time

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{IH}	High-level input voltage			0.7V _{CC}			N
V_{IL}	Low-level input voltage	LOWDR	Over the V _{CC} range (see Note 3)			1	V
VIH	High-level input voltage	DT	Over the V _{CC} range	0.7V _{CC}			V
VIL	Low-level input voltage	יטן				1	v

NOTE 3: Ensured by design, not production tested.

digital control terminals

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
VIH	High-level input voltage		0.7V _{CC}			V
V_{IL}	Low-level input voltage	Over the V _{CC} range			1	V



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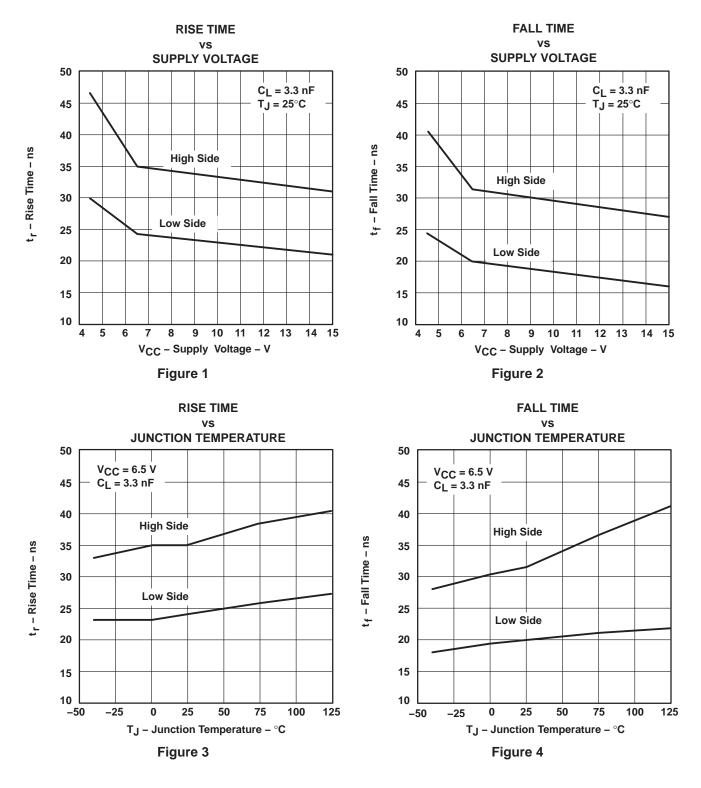
switching characteristics over recommended operating virtual junction temperature range, C_L = 3.3 nF (unless otherwise noted)

	PARAMETER	TEST CO	ONDITIONS	MIN	TYP	MAX	UNIT			
		V _{BOOT} = 4.5 V,	V _{BOOTLO} = 0 V			60				
Rise time	HIGHDR output (see Note 3)	V _{BOOT} = 6.5 V,	V _{BOOTLO} = 0 V			50	ns			
		V _{BOOT} = 12 V,	V _{BOOTLO} = 0 V			50				
Rise time		$V_{CC} = 4.5 V$				40				
	LOWDR output (see Note 3)	$V_{CC} = 6.5 V$				30	ns			
		V _{CC} = 12 V				30				
		V _{BOOT} = 4.5 V,	V _{BOOTLO} = 0 V			60				
	HIGHDR output (see Note 3)	V _{BOOT} = 6.5 V,	V _{BOOTLO} = 0 V			50	ns			
		V _{BOOT} = 12 V,	V _{BOOTLO} = 0 V			50				
Fall time		$V_{CC} = 4.5 V$				40				
	LOWDR output (see Note 3)	$V_{CC} = 6.5 V$				30	ns			
		V _{CC} = 12 V				30				
		V _{BOOT} = 4.5 V,	V _{BOOTLO} = 0 V			130				
	HIGHDR going low (excluding dead time) (see Note 3)	V _{BOOT} = 6.5 V,	V _{BOOTLO} = 0 V			100	ns			
Dranagation dalow time		V _{BOOT} = 12 V,	V _{BOOTLO} = 0 V			75				
Propagation delay time		V _{BOOT} = 4.5 V,	V _{BOOTLO} = 0 V			80				
	LOWDR going high (excluding dead time) (see Note 3)	V _{BOOT} = 6.5 V,	V _{BOOTLO} = 0 V			70	ns			
		V _{BOOT} = 12 V,	V _{BOOTLO} = 0 V			60				
		$V_{CC} = 4.5 V$				80				
Propagation delay time	LOWDR going low (excluding dead time) (see Note 3)	$V_{CC} = 6.5 V$				70	ns			
		V _{CC} = 12 V				60				
		$V_{CC} = 4.5 V$		40		170				
Driver nonoverlap time	DT to LOWDR and LOWDR to HIGHDR (see Note 3)	$V_{CC} = 6.5 V$		25		135	ns			
		V _{CC} = 12 V		15		85				

NOTE 3: Ensured by design, not production tested.

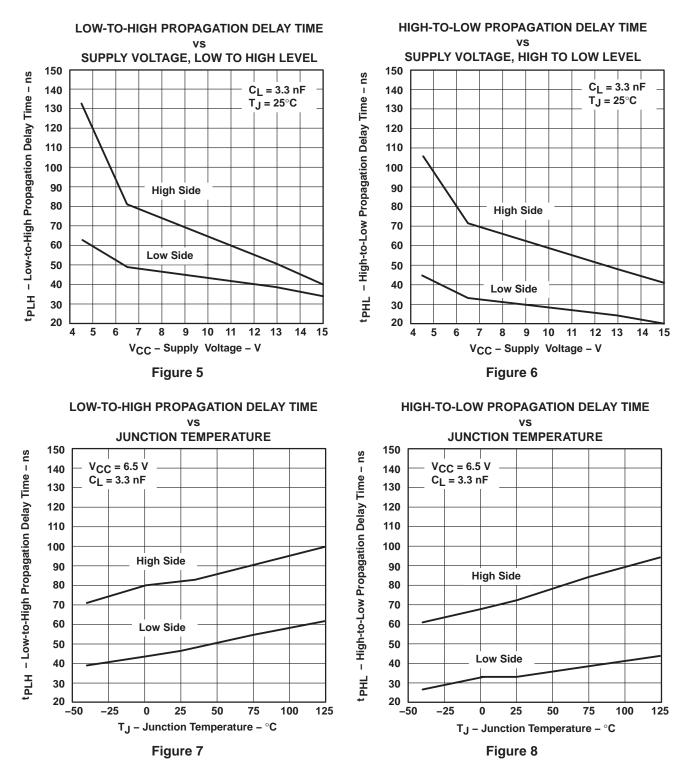


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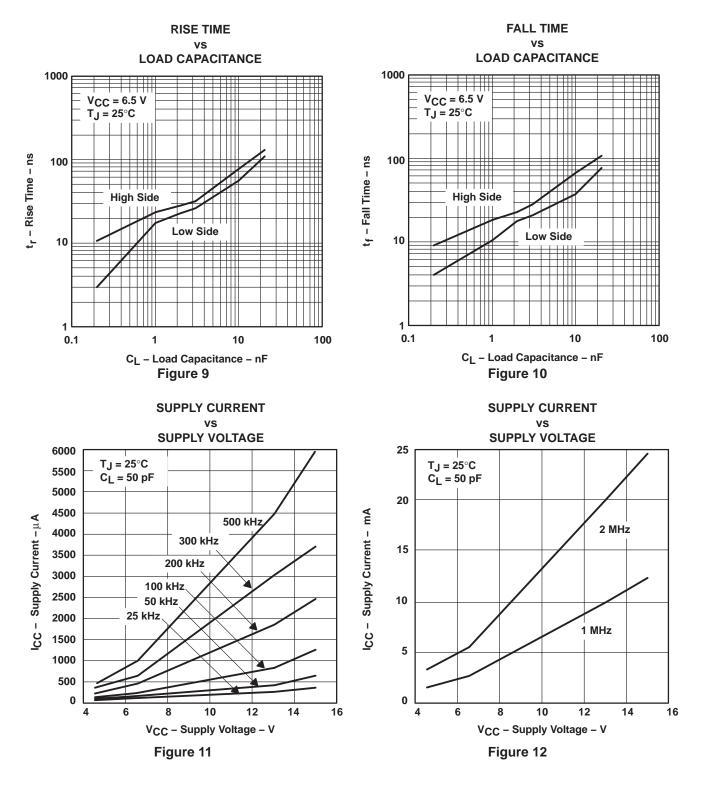


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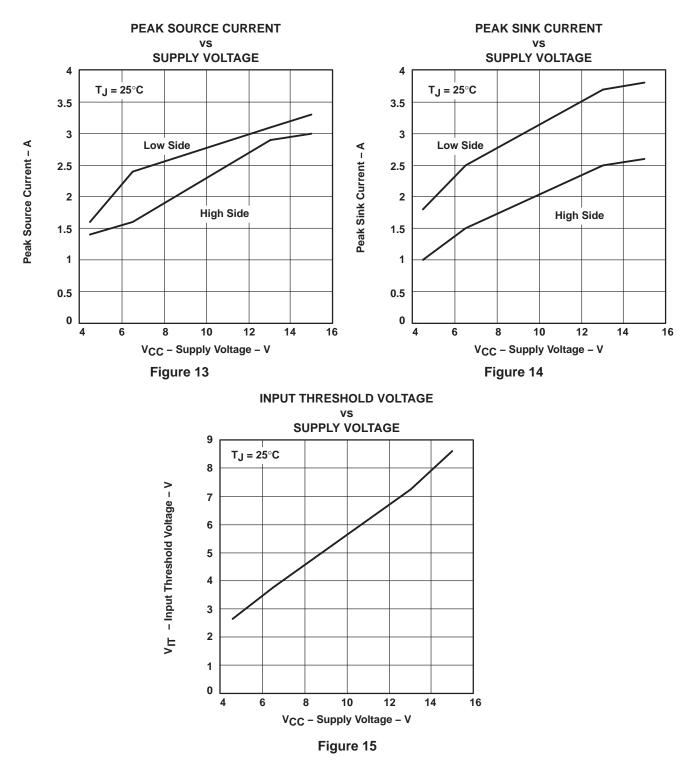


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APPLICATION INFORMATION

Figure 16 shows the circuit schematic of a 100-kHz synchronous-buck converter implemented with a TL5001A pulse-width-modulation (PWM) controller and a TPS2833 driver. The converter operates over an input range from 4.5 V to 12 V and has a 3.3 V output. The circuit can supply 3 A continuous load and the transient load is 5 A. The converter achieves an efficiency of 94% for $V_{IN} = 5$ V, $I_{Ioad} = 1$ A, and 93% for $V_{in} = 5$ V, $I_{Ioad} = 3$ A.

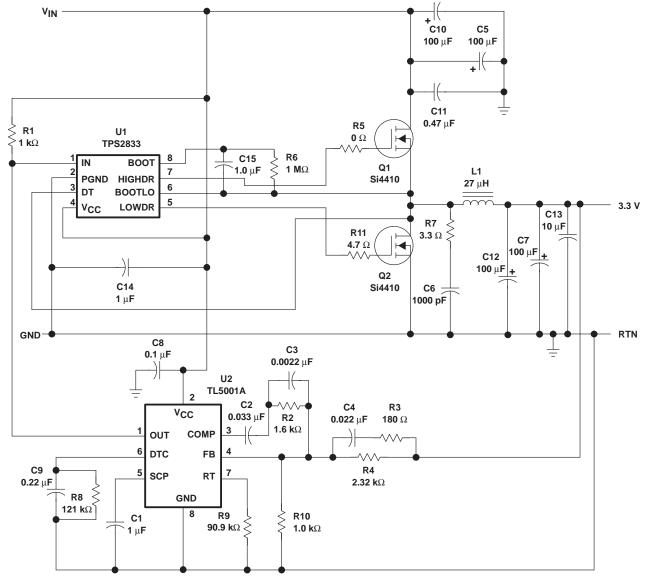


Figure 16. 3.3 V 3 A Synchronous-Buck Converter Circuit



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APPLICATION INFORMATION

Great care should be taken when laying out the pc board. The power-processing section is the most critical and will generate large amounts of EMI if not properly configured. The junction of Q1, Q2, and L1 should be very tight. The connection from Q1 drain to the positive sides of C5, C10, and C11 and the connection from Q2 source to the negative sides of C5, C10, and C11 should be as short as possible. The negative terminals of C7 and C12 should also be connected to Q2 source.

Next, the traces from the MOSFET driver to the power switches should be considered. The BOOTLO signal from the junction of Q1 and Q2 carries the large gate drive current pulses and should be as heavy as the gate drive traces. The bypass capacitor (C14) should be tied directly across V_{CC} and PGND.

The next most sensitive node is the FB node on the controller (terminal 4 on the TL5001A) This node is very sensitive to noise pickup and should be isolated from the high-current power stage and be as short as possible. The ground around the controller and low-level circuitry should be tied to the power ground as the output. If these three areas are properly laid out, the rest of the circuit should not have any other EMI problems and the power supply will be relatively free of noise.





PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TPS2832DR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2832	Samples
TPS2833D	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2833	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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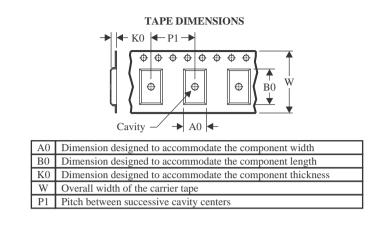
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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

ſ	Device	Package Type	Package Drawing			Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
	TPS2832DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1



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PACKAGE MATERIALS INFORMATION

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS2832DR	SOIC	D	8	2500	340.5	338.1	20.6

TEXAS INSTRUMENTS

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TUBE



- B - Alignment groove width

*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
TPS2832D	D	SOIC	8	75	507	8	3940	4.32
TPS2832D	D	SOIC	8	75	505.46	6.76	3810	4
TPS2833D	D	SOIC	8	75	507	8	3940	4.32

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