







SGLS118E - DECEMBER 2001 - REVISED JULY 2024

TPS769-Q1

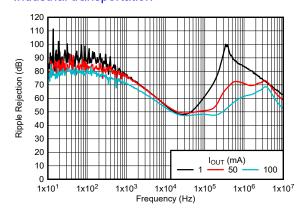
TPS769-Q1 100mA, 16V, Low-Dropout Linear Regulator

1 Features

- AEC-Q100 qualified for automotive applications:
 - Temperature grade 1: −40°C to 125°C, T_A
- Input voltage range:
 - Legacy chip: 2.7V to 10V (13.5V absolute max)
 - New chip: 2.5V to 16V (18V absolute max)
- Output voltage range (adjustable):
 - Legacy chip: 1.25V to 5.5V
 - New chip: 1.2V to 5.5V
- Output voltage range (fixed):
 - Legacy chip: 1.5V to 5V
- New chip: 1.2V to 5V
- High PSRR (new chip): 46dB at 1MHz
- Output accuracy:
 - Legacy chip: 3% over load and temperature
 - New chip: 1.2% over load and temperature
- Dropout voltage:
 - Legacy chip: 71mV (typ) at 100mA
 - New chip: 150mV (typ) at 100mA
- Integrated fault protection:
 - Thermal shutdown
 - Overcurrent protection
- Internal soft-start time (new chip): 750µs (typical)
- Output capacitor for stable operation:
 - Legacy chip: ≥ 4.7µF
 - New chip: ≥ 2.2µF
- Package: 5-pin SOT-23, $R_{\theta JA} = 178.6$ °C/W (new

2 Applications

- Hybrid, electric, and powertrain systems
- **ADAS** modules
- Infotainment and cluster
- Industrial transportation



TPS76933-Q1 PSRR vs Output Current (New Chip)

3 Description

The TPS769-Q1 is a low-dropout (LDO) linear voltage regulator. This device supports an input voltage range from 2.5V to 16V (new chip) and up to 100mA of load current. For the new chip, the supported output range is from 1.2V to 5.0V (fixed version) or from 1.2V to 5.5V (adjustable version).

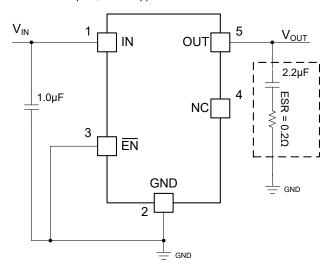
The wide input voltage range makes the device a good choice for operating from regulated rails (such as 10V or 12V). The voltage range is up to 16V for the new chip. This range allows the LDO to generate the bias voltage for a variety of applications. These applications include power microcontrollers (MCUs) and processors, as well as silicon carbide (SiC) gate drivers and microphones.

Wide bandwidth PSRR performance is greater than 70dB at 1kHz and 46dB at 1MHz (new chip). This performance helps attenuate the switching frequency of an upstream DC/DC converter and minimizes post regulator filtering. The new chip supports an internal soft-start circuit mechanism that reduces inrush current during start-up, thus allowing for smaller input capacitance.

Package Information

PART NUMBER	PACKAGE ⁽¹⁾	PACKAGE SIZE(2)	
TPS769-Q1	DBV (SOT-23, 5)	2.9mm × 2.8mm	

- (1) For more information, see the Mechanical, Packaging, and Orderable Information.
- (2)The package size (length × width) is a nominal value and includes pins, where applicable.



Typical Application Circuit



Table of Contents

1 Features1	6.4 Device Functional Modes	23
2 Applications1	7 Application and Implementation	
3 Description1	7.1 Application Information	25
4 Pin Configuration and Functions2	7.2 Typical Application	25
5 Specifications3	7.3 Power Supply Recommendations	33
5.1 Absolute Maximum Ratings3	7.4 Layout	33
5.2 ESD Ratings3	8 Device and Documentation Support	
5.3 Recommended Operating Conditions4	8.1 Device Support	
5.4 Thermal Information4	8.2 Documentation Support	
5.5 Dissipation Ratings (Legacy Chip)5	8.3 Receiving Notification of Documentation Updates	s34
5.6 Electrical Characteristics5	8.4 Support Resources	34
5.7 Typical Characteristics8	8.5 Trademarks	
5.8 Typical Characteristics: Supported ESR Range 16	8.6 Electrostatic Discharge Caution	35
6 Detailed Description20	8.7 Glossary	
6.1 Overview20	9 Revision History	
6.2 Functional Block Diagrams20	10 Mechanical, Packaging, and Orderable	
6.3 Feature Description21	Information	35

4 Pin Configuration and Functions

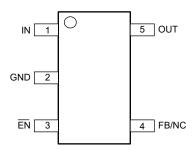


Figure 4-1. DBV Package, 5-Pin SOT-23 (Top View)

Table 4-1. Pin Functions

PIN		I/O	DESCRIPTION		
NO.	NAME	1/0	DESCRIPTION		
1	IN	I	Input pin. Use the recommended capacitor value as listed in the <i>Recommended Operating Conditions</i> . Place the input capacitor as close to the IN and GND pins of the device as possible. See the <i>Input and Output Capacitor Requirements</i> section for more information.		
2	GND	_	Ground.		
3	ĒN	I	Enable pin. Driving the enable pin low enables the device. Driving this pin high disables the device. Low and high thresholds are listed in the <i>Electrical Characteristics</i> table.		
4	FB/NC	I	Adjustable version (TPS76901-Q1): Feedback pin. Input to the control-loop error amplifier. This pin sets the output voltage of the device with external resistors. Do not float this pin. Fixed version: No connection (legacy chip).Do not connect (new chip).		
5	OUT	0	Output pin. Use the recommended capacitor value as listed in the <i>Recommended Operating Conditions</i> . Place the output capacitor as close to the OUT and GND pins of the device as possible. See the <i>Input and Output Capacitor Requirements</i> section for more information.		



5 Specifications

5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1) (2)

		MIN	MAX	UNIT
.,	Continuous input voltage (legacy chip)	-0.3	13.5	
V _{IN}	Continuous input voltage (new chip)	-0.3	18	
	Output voltage (legacy chip)	-0.3	7	
V _{OUT}	Output voltage (new chip)	-0.3	V _{IN} + 0.3 or 7 (whichever is smaller)	V
.,	FB pin voltage (legacy chip)	-0.3	7	
V _{FB}	FB pin voltage (new chip)	-0.3	3	
_	EN pin voltage (legacy chip)	-0.3	V _{IN} + 0.3	
$V_{\overline{EN}}$	EN pin voltage (new chip)	-0.3	18	
Current	Maximum output	Internally	limited	Α
	Operating junction temperature , T _J (legacy chip)	-40	150	
Temperature	Operating junction temperature , T _J (new chip)	-55	150	°C
	Storage, T _{stg}	-65	150	

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

5.2 ESD Ratings

			VALUE (Legacy Chip)	VALUE (New Chip)	UNIT
V	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±2000	±3000	\ \ \
V _(ESD) Electrostatic discharge		Charged device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾	N/A	±1000	V

⁽¹⁾ JEDEC document JEP155 states that 2kV HBM allows safe manufacturing with a standard ESD control process.

⁽²⁾ All voltages with respect to GND.

⁽²⁾ JEDEC document JEP157 states that 500V CDM allows safe manufacturing with a standard ESD control process.



5.3 Recommended Operating Conditions

		MIN	NOM	MAX	UNIT
V	Supply input voltage (legacy chip)	2.7		10	V
V_{IN}	Supply input voltage (new chip)	2.5		16	V
M	Output voltage (legacy chip)	1.25		5.5	V
V _{OUT}	Output voltage (new chip)	1.2		5.5	V
V	FB voltage (legacy chip)		1.224		V
V_{FB}	FB voltage (new chip)		1.2		V
M	Enable voltage (legacy chip)	0		V _{IN}	V
$V_{\overline{EN}}$	Enable voltage (new chip)	0		16	V
I _{OUT}	Output current	0		100	mA
C _{IN} ⁽¹⁾	Input capacitor		1		μF
C (1)	Output capacitor (legacy chip)	4.7			
C _{OUT} (1)	Output capacitor (new chip)	2.2		200	μF
ESR	ESR range (legacy chip)	0.2		10	0
ESK	ESR range (new chip)	0		3	Ω
T _J	Operating junction temperature	-40		125	°C

⁽¹⁾ All capacitor values are assumed to derate to 50% of the nominal capacitor value. Maintain an effective output capacitance of 1 μF minimum for stability.

5.4 Thermal Information

		Legacy chip	New chip	
	THERMAL METRIC (TPS769-Q1) (1) (2)	DBV (SOT23-5)	DBV (SOT23-5)	UNIT
		5 PINS	5 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	204.6	178.6	°C/W
R _{0JC(top)}	Junction-to-case (top) thermal resistance	117.5	77.9	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	34.4	47.2	°C/W
ΨЈТ	Junction-to-top characterization parameter	11.8	15.9	°C/W
ΨЈВ	Junction-to-board characterization parameter	33.5	46.9	°C/W

⁽¹⁾ For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application note.

⁽²⁾ Thermal performance results are based on the JEDEC standard of 2s2p PCB configuration. These thermal metric parameters can be further improved by 35-55% based on thermally optimized PCB layout designs. See the analysis of the Impact of board layout on LDO thermal performance application note.



5.5 Dissipation Ratings (Legacy Chip)

DISSIPATION RATINGS				
THERMAL METRIC	· · · · · · · · · · · · · · · · · · ·	DBV (SOT-23) 5 PINS		
	Low K ⁽¹⁾	High K (2)		
R _{θJC} (Junction-to-case thermal resistance)	65.8	65.8	°C/W	
R _{θJA} (Junction-to-ambient thermal resistance)	259	180	°C/W	
Derating factor above T _A = +25°C	3.9	5.6	mW/°C	
Power rating (T _A < 25°C)	386	555	mW	
Power rating (T _A = 70°C)	212	305	mW	
Power rating (T _A = 85°C)	154	222	mW	

- (1) The JEDEC Low K (1s) board design used to derive this data was a 3 inch x 3 inch, two layer board with 2 ounce copper traces on top of the board.
- (2) The JEDEC High K (2s2p) board design used to derive this data was a 3 inch x 3 inch, multilayer board with 1 ounce internal power and ground planes and 2 ounce copper traces on top and bottom of the board.

5.6 Electrical Characteristics

specified at T_J = -40° C to 125° C, V_{IN} = $V_{OUT(nom)}$ + 1.0V or V_{IN} = 2.5V (whichever is greater), I_{OUT} = 10μ A, \overline{EN} = 0V, C_{IN} = 1.0μ F, C_{OUT} = 2.2μ F (unless otherwise noted); typical values are at T_J = 25° C

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
		Adjustable, legacy	1.2V ≤ V _{OUT} ≤ 5.5V, 10µA ≤ I _{OUT} ≤ 100mA, T _J = 25°C		V _{OUT}		
		chip	1.2V ≤ V_{OUT} ≤ 5.5V, $10\mu A$ ≤ I_{OUT} ≤ 100mA	0.97 × V _{OUT}		1.03 × V _{OUT}	1
V _{OUT}	Output voltage	Fixed, legacy chip	$10\mu A \le I_{OUT} \le 100mA$, $T_J = 25^{\circ}C$, $V_{OUT(nom)} + 1V < V_{IN} < 10V$		V _{OUT}		V
		Tixed, legacy criip		0.97 × V _{OUT}		1.03 × V _{OUT}	
		New chip	$10\mu A \le I_{OUT} \le 100mA, V_{OUT(nom)} + 1V$ < $V_{IN} < 16V$	0.988 × V _{OUT}		1.012 × V _{OUT}	1
V	Feedback voltage	Legacy chip			1.224		V
V _{FB}	r eedback voltage	New chip			1.2		\ \
	Quiescent current (GND current)	Legacy chip	$\overline{\text{EN}}$ = 0V, 0mA \leq I _{OUT} \leq 100mA, T _J = +25°C		17		Δ
			<u>EN</u> = 0V, I _{OUT} = 100mA			28	
I _Q		New chip	EN = 0V, I _{OUT} = 0mA (adjustable)		50	80	
			$\overline{\text{EN}}$ = 0V, I _{OUT} = 0mA (fixed)		55	95	
			<u>EN</u> = 0V, I _{OUT} = 100mA		620		
$\Delta V_{OUT(\Delta}$	Output voltage line regulation $(\Delta V_{OUT}/V_{OUT})$	Legacy chip	$V_{OUT(NOM)}$ +1.0V \leq V_{IN} \leq 10V, I_{OUT} = 100mA, T_{J} = 25°C		0.04	0.032	%/V
VOUT)		New chip	$V_{OUT(NOM)} + 1.0V \le V_{IN} \le 16V$, $I_{OUT} = 10\mu A$		-	0.032	
$\Delta V_{OUT(\Delta}$	Outrot valtana land namulation	Legacy chip	0 - A < 1		12		\/
IOUT)	Output voltage load regulation	New chip	0mA ≤ I _{OUT} ≤ 100mA, T _J = 25°C		20		mV
		Legacy chip	BW = 300Hz to 50kHz, C_{OUT} = 10 μ F, T_J = 25 $^{\circ}$ C		190		
V _n	Output noise voltage	New chip	BW = 300Hz to 50kHz, I _{OUT} = 100mA, C _{OUT} = 4.7µF		165		μV _R мs
		INEW CHIP	BW = 10Hz to 100kHz, I _{OUT} = 100mA, C _{OUT} = 4.7µF		195		



5.6 Electrical Characteristics (continued)

specified at T_J = -40° C to 125°C, V_{IN} = $V_{OUT(nom)}$ + 1.0V or V_{IN} = 2.5V (whichever is greater), I_{OUT} = 10μ A, \overline{EN} = 0V, C_{IN} = 1.0 μ F, C_{OUT} = 2.2 μ F (unless otherwise noted); typical values are at T_J = 25°C

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
T _{SD(shutdow}	Thermal shutdown temperature	New chip	Temperature increasing		173		°C	
T _{SD(reset)}	Thermal shutdown reset temperature	New chip	Temperature falling		157		°C	
1	Output current limit	Legacy chip	V _{OUT} = 0V		350	750	mA	
I _{CL}	Output current iirnit	New chip	VOUT - UV		370	450	mA	
		Legacy chip	EN = V _{IN} , 2.7V < V _{IN} < 10V		1			
1	Standby current	Legacy Chip	EN = V _{IN} , 2.7V < V _{IN} < 10V			2		
ISTANDBY	Standby current	New chip	EN = V _{IN} , 2.5V < V _{IN} < 16V		0.9		μA	
		New Chip	EN = V _{IN} , 2.5V < V _{IN} < 16V			2.75		
1	Feedback pin current	Legacy chip	V _{FB} = 1.224V	-1		1	μA	
I _{FB}		Newchip	V _{FB} = 1.2V	-0.1		0.1	μA	
	High level enable input voltage	Legacy chip	2.7V ≤ V _{IN} ≤ 10V	1.7				
ΕN	Low level enable input voltage					0.9) V	
EIN	High level enable input voltage	Name data	25/15/1 510/1	1.6				
	Low level enable input voltage	New chip	2.5V ≤ V _{IN} ≤ 16V			0.415		
PSRR	Dower cumply ripple rejection	Legacy chip	I_{OUT} = 100mA, f = 1kHz, C_{OUT} = 10 μ F, T_J = 25 °C		60		dB	
FORK	Power-supply ripple rejection	New chip	I_{OUT} = 100mA, f = 1kHz, C_{OUT} = 4.7 μ F, T_J = 25°C		58		ub	
		Logovy obin	EN = 0V	-1	0	1		
	In the second of	Legacy chip	Legacy Chip	EN = V _{IN}	-1		1	
I _{EN}	Input current (EN)	Now ohin	EN = 0V	-0.75	-0.4	0.02	μA	
		New chip $\overline{\overline{EN}} = 6V$	-0.01		0.01			



5.6 Electrical Characteristics (continued)

specified at T_J = -40° C to 125°C, V_{IN} = $V_{OUT(nom)}$ + 1.0V or V_{IN} = 2.5V (whichever is greater), I_{OUT} = 10μ A, \overline{EN} = 0V, C_{IN} = 1.0μ F, C_{OUT} = 2.2μ F (unless otherwise noted); typical values are at T_J = 25° C

	PARAMETER		TEST CONDITIONS	MIN TYP	MAX	UNI
			I _{OUT} = 50mA	60		
		TPS76928-Q1	I _{OUT} = 50mA, T _J = -40°C to 125°C		125	
		(legacy chip)	I _{OUT} = 100mA	122		
			I _{OUT} = 100mA, T _J = -40°C to 125°C		245	
			I _{OUT} = 50mA	120		
		TPS76928-Q1	I _{OUT} = 50mA, T _J = -40°C to 125°C		184	
		(new chip)	I _{OUT} = 100mA	150		
			I_{OUT} = 100mA, T_J = -40°C to 125°C		218	
			I _{OUT} = 50mA	57		
		TPS76930-Q1	I _{OUT} = 50mA, T _J = -40°C to 125°C		115	
		(legacy chip)	I _{OUT} = 100mA	115		
			I _{OUT} = 100mA, T _J = -40°C to 125°C		230	
			I _{OUT} = 50mA	120		
		TPS76930-Q1	$I_{OUT} = 50$ mA, $T_{J} = -40$ °C to 125°C		184	mV
	Dropout voltage	(new chip)	I _{OUT} = 100mA	150		
			I _{OUT} = 100mA, T _J = -40°C to 125°C		218	
DO			I _{OUT} = 50mA	48		
		TPS76933-Q1	$I_{OUT} = 50$ mA, $T_{J} = -40$ °C to 125°C		100	
		(legacy chip)	I _{OUT} = 100mA	98		
			$I_{OUT} = 100 \text{mA}, T_J = -40 ^{\circ}\text{C} \text{ to } 125 ^{\circ}\text{C}$		200	
			I _{OUT} = 50mA	120		
		TPS76933-Q1	$I_{OUT} = 50$ mA, $T_{J} = -40$ °C to 125°C		184	
		(new chip)	I _{OUT} = 100mA	150		
			$I_{OUT} = 100 \text{mA}, T_J = -40 ^{\circ}\text{C} \text{ to } 125 ^{\circ}\text{C}$		218	
			I _{OUT} = 50mA	35		
		TPS76950-Q1	$I_{OUT} = 50$ mA, $T_{J} = -40$ °C to 125°C		85	
		(legacy chip)	I _{OUT} = 100mA	71		
			$I_{OUT} = 100 \text{mA}, T_{J} = -40 ^{\circ}\text{C} \text{ to } 125 ^{\circ}\text{C}$		170)
			I _{OUT} = 50mA	120		
	TPS76950-Q1 (new chip)	TPS76950-Q1	$I_{OUT} = 50 \text{mA}, T_{J} = -40 ^{\circ}\text{C} \text{ to } 125 ^{\circ}\text{C}$		184	
		(new chip)	I _{OUT} = 100mA	150		
			I _{OUT} = 100mA, T _J = -40°C to 125°C		218	
UVLO+	Rising bias supply UVLO		V _{IN} rising, –40°C ≤ T _J ≤ 125°C	2.2	2.4	
UVLO-	Falling bias supply UVLO	TPS769-Q1 (new	V _{IN} falling, –40°C ≤ T _J ≤ 125°C	1.9 2.07		,
UVLO(HYS	UVLO hysteresis	chip)	-40°C ≤ T _J ≤ 125°C	0.130		<u>'</u>



5.7 Typical Characteristics

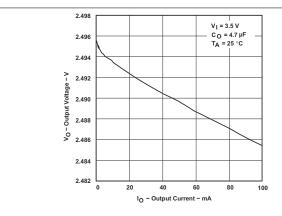


Figure 5-1. TPS76925-Q1 Output Voltage vs Output Current (Legacy Chip)

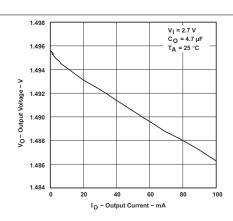


Figure 5-2. TPS76915-Q1 Output Voltage vs Output Current (Legacy Chip)

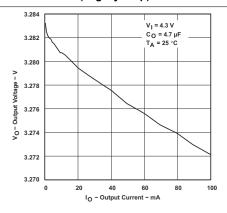


Figure 5-3. TPS76933-Q1 Output Voltage vs Output Current (Legacy Chip)

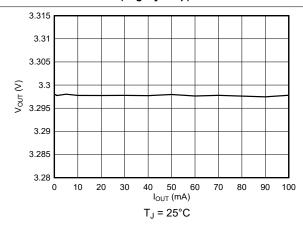


Figure 5-4. TPS76933-Q1 Output Voltage vs Output Current (New Chip)

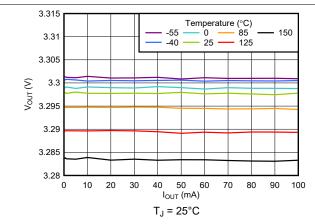


Figure 5-5. TPS76933-Q1 Output Voltage vs Output Current (New Chip)

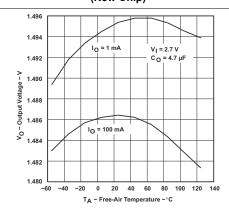


Figure 5-6. TPS76915-Q1 Output Voltage vs Free-Air Temperature (Legacy Chip)

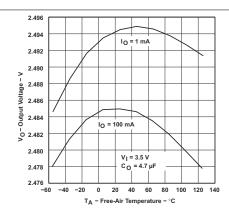


Figure 5-7. TPS76925-Q1 Output Voltage vs Free-Air Temperature (Legacy Chip)

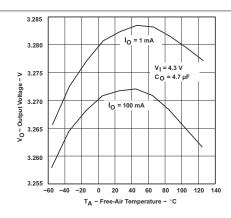


Figure 5-8. TPS76933-Q1 Output Voltage vs Free-Air Temperature (Legacy Chip)

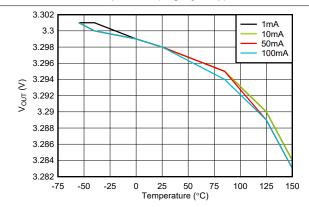


Figure 5-9. TPS76933-Q1 Output Voltage vs Free-Air Temperature (New Chip)

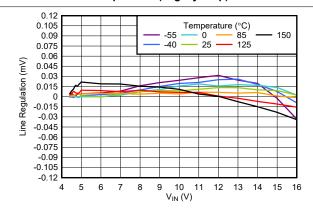


Figure 5-10. TPS76933-Q1 Line Regulation vs Free-Air Temperature (New Chip)

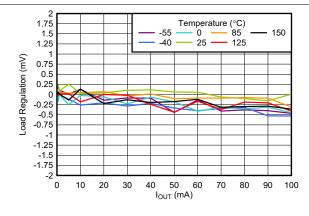


Figure 5-11. TPS76933-Q1 Load Regulation vs Free-Air Temperature (New Chip)

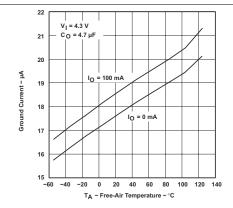
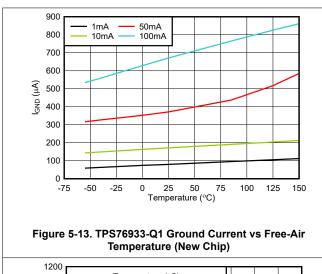


Figure 5-12. TPS76933-Q1 Ground Current vs Free-Air Temperature (Legacy Chip)





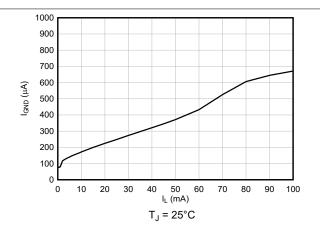
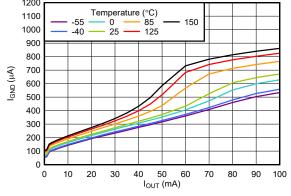


Figure 5-14. TPS76933-Q1 Ground Current vs Output Current (New Chip)



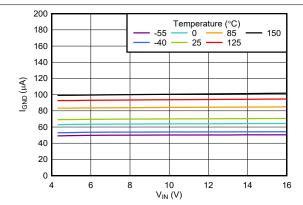
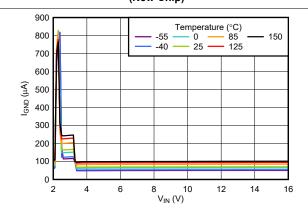


Figure 5-15. TPS76933-Q1 Ground Current vs Output Current (New Chip)

Figure 5-16. TPS76933-Q1 Ground Current vs Input Supply (New Chip)



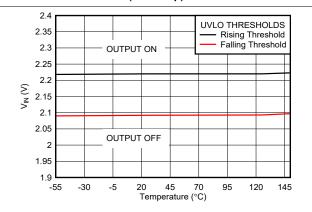


Figure 5-17. TPS76933-Q1 Ground Current vs Input Supply (New Chip)

Figure 5-18. TPS76933-Q1 UVLO Threshold vs Free-Air Temperature (New Chip)

Submit Document Feedback

Copyright © 2024 Texas Instruments Incorporated

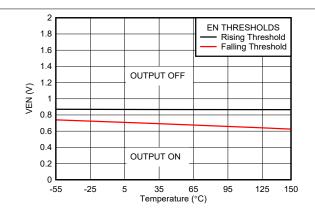


Figure 5-19. TPS76933-Q1 EN Threshold vs Free-Air Temperature (New Chip)

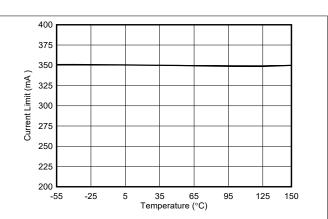


Figure 5-20. TPS76933-Q1 Current Limit vs Free-Air Temperature (New Chip)

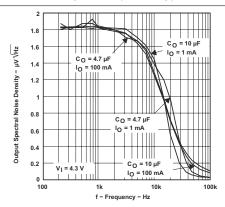


Figure 5-21. TPS76933-Q1 Output Spectral Noise Density vs Frequency (Legacy Chip)

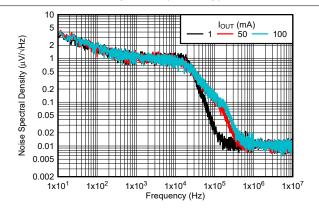


Figure 5-22. TPS76933-Q1 Output Spectral Noise Density vs Output Current (New Chip)

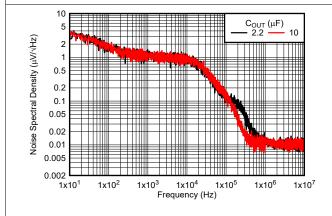


Figure 5-23. TPS76933-Q1 Output Spectral Noise Density vs Output Capacitor (New Chip)

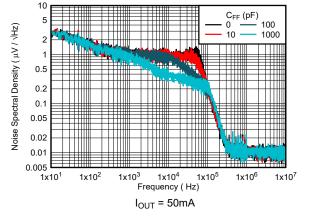


Figure 5-24. TPS76901-Q1 Output Spectral Noise Density vs Feed-Forward Capacitor (New Chip)



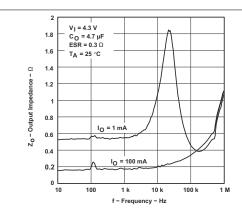


Figure 5-25. Output Impedance vs Frequency (Legacy Chip)

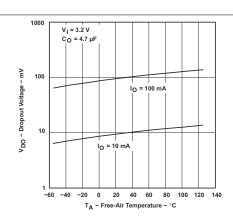


Figure 5-26. TPS76933-Q1 Dropout Voltage vs Free-Air Temperature (Legacy Chip)

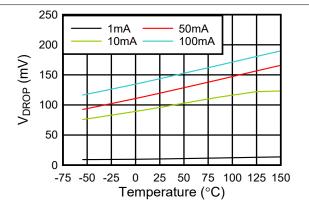


Figure 5-27. TPS76933-Q1 Dropout Voltage vs Free-Air Temperature (New Chip)

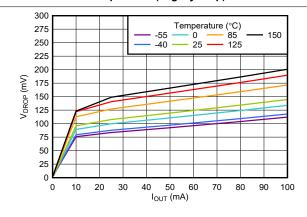


Figure 5-28. TPS76933-Q1 Dropout Voltage vs Output Current (New Chip)

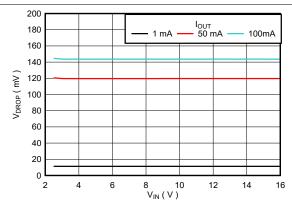


Figure 5-29. TPS76901-Q1 Dropout Voltage vs Input Supply (New Chip)

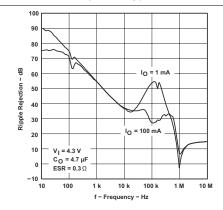
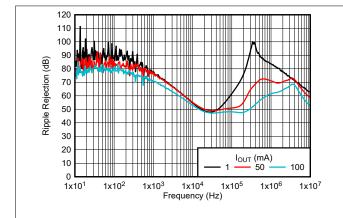


Figure 5-30. TPS76933-Q1 Ripple Rejection vs Frequency (Legacy Chip)

Submit Document Feedback

Copyright © 2024 Texas Instruments Incorporated



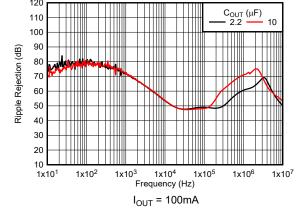
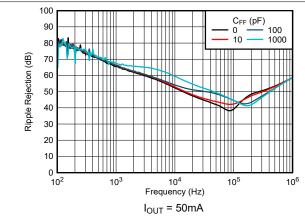


Figure 5-31. TPS76933-Q1 Ripple Rejection vs Output Current (New Chip)

Figure 5-32. TPS76933-Q1 Ripple Rejection vs Output Capacitor (New Chip)



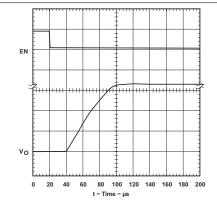
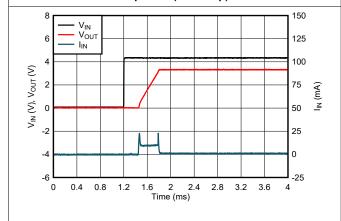


Figure 5-33. TPS76901-Q1 Ripple Rejection vs Feed-Forward Capacitor (New Chip)





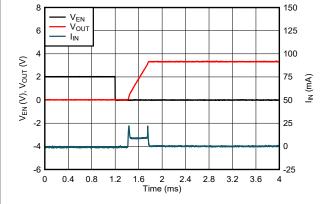


Figure 5-35. LDO Start-Up Time With Input Supply (New Chip)

Figure 5-36. LDO Start-Up Time With EN (New Chip)



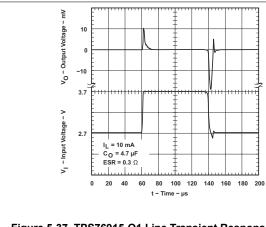


Figure 5-37. TPS76915-Q1 Line Transient Response (Legacy Chip)

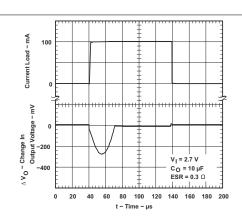


Figure 5-38. TPS76915-Q1 Load Transient Response (Legacy Chip)

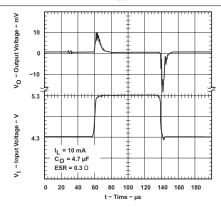


Figure 5-39. TPS76933-Q1 Line Transient Response (Legacy Chip)

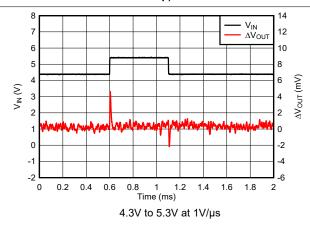


Figure 5-40. TPS76933-Q1 Line Transient Response (New Chip)

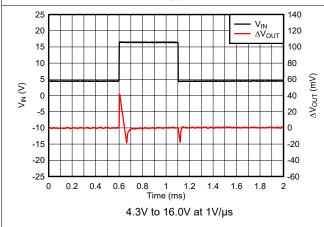


Figure 5-41. TPS76933-Q1 Line Transient Response (New Chip)

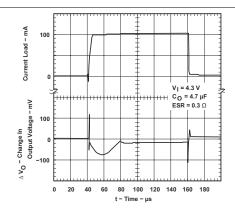
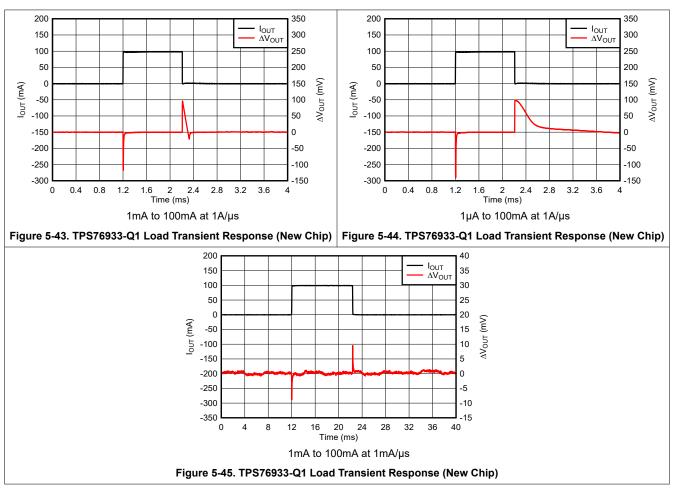


Figure 5-42. TPS76933-Q1 Load Transient Response (Legacy Chip)

Submit Document Feedback

Copyright © 2024 Texas Instruments Incorporated





5.8 Typical Characteristics: Supported ESR Range

Equivalent series resistance (ESR) refers to the total series resistance. This resistance includes the ESR of the capacitor, any series resistance added externally, and PCB trace resistance to C_O.

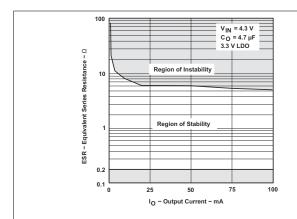


Figure 5-46. TPS76933-Q1 Typical Regions of Stability ESR vs Output Current (Legacy Chip)

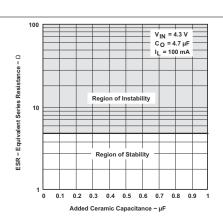


Figure 5-47. TPS76933-Q1 Typical Regions of Stability ESR vs Added Ceramic Capacitance (Legacy Chip)

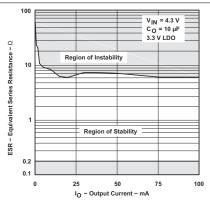


Figure 5-48. TPS76933-Q1 Typical Regions of Stability ESR vs Output Current (Legacy Chip)

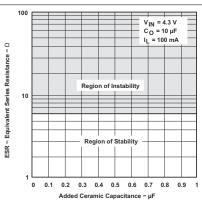


Figure 5-49. TPS76933-Q1 Typical Regions of Stability ESR vs Added Ceramic Capacitance (Legacy Chip)

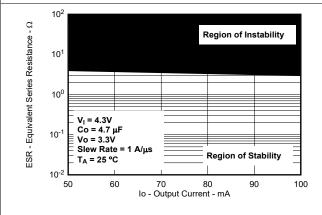


Figure 5-50. Typical Region of Stability ESR vs Output Current (New Chip)

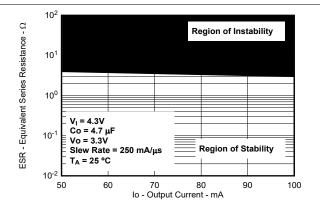


Figure 5-51. Typical Region of Stability ESR vs Output Current (New Chip)

Submit Document Feedback

Copyright © 2024 Texas Instruments Incorporated

5.8 Typical Characteristics: Supported ESR Range (continued)

Equivalent series resistance (ESR) refers to the total series resistance. This resistance includes the ESR of the capacitor, any series resistance added externally, and PCB trace resistance to C_O.

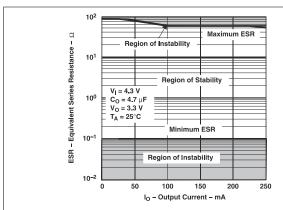


Figure 5-52. Typical Region of Stability ESR vs Output Current (Legacy Chip)

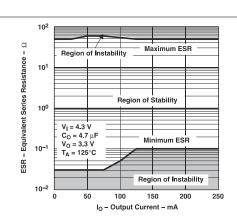


Figure 5-53. Typical Region of Stability ESR vs Output Current (Legacy Chip)

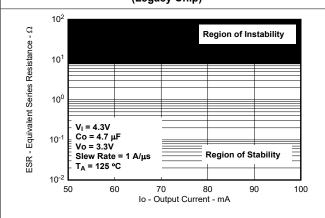


Figure 5-54. Typical Region of Stability ESR vs Output Current (New Chip)

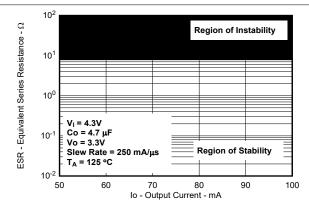


Figure 5-55. Typical Region of Stability ESR vs Output Current (New Chip)

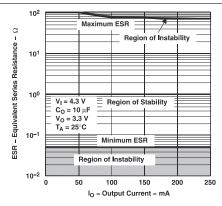


Figure 5-56. Typical Region of Stability ESR vs Output Current (Legacy Chip)

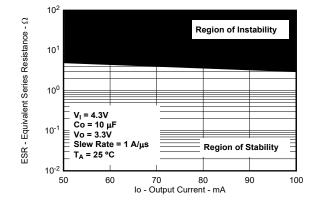
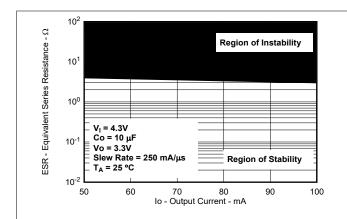


Figure 5-57. Typical Region of Stability ESR vs Output Current (New Chip)



5.8 Typical Characteristics: Supported ESR Range (continued)

Equivalent series resistance (ESR) refers to the total series resistance. This resistance includes the ESR of the capacitor, any series resistance added externally, and PCB trace resistance to C_O.



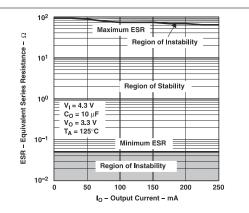
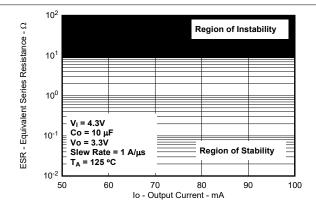


Figure 5-58. Typical Region of Stability ESR vs Output Current (New Chip)

Figure 5-59. Typical Region of Stability ESR vs Output Current (Legacy Chip)



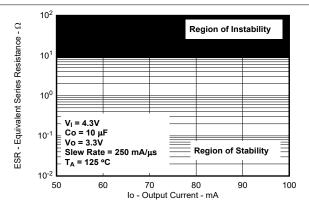
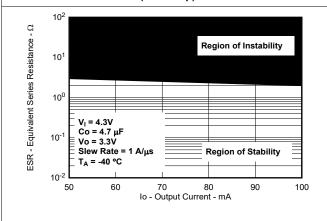


Figure 5-60. Typical Region of Stability ESR vs Output Current (New Chip)

Figure 5-61. Typical Region of Stability ESR vs Output Current (New Chip)



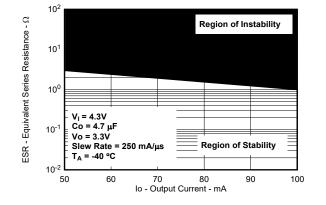


Figure 5-62. Typical Region of Stability ESR vs Output Current (New Chip)

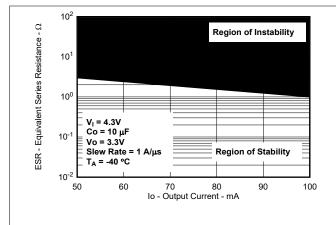
Figure 5-63. Typical Region of Stability ESR vs Output Current (Legacy Chip)

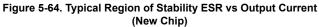
Submit Document Feedback

Copyright © 2024 Texas Instruments Incorporated

5.8 Typical Characteristics: Supported ESR Range (continued)

Equivalent series resistance (ESR) refers to the total series resistance. This resistance includes the ESR of the capacitor, any series resistance added externally, and PCB trace resistance to C_O.





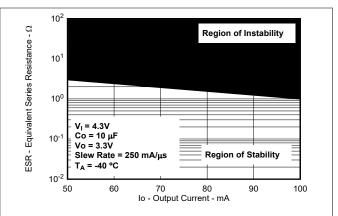


Figure 5-65. Typical Region of Stability ESR vs Output Current (New Chip)



6 Detailed Description

6.1 Overview

The TPS769-Q1 is a low quiescent current, high PSRR, low-dropout (LDO) voltage regulator capable of handling up to 100mA of the load current. The TPS769-Q1 is optimized for use in battery-powered and automotive applications.

The TPS769-Q1 features an integrated overcurrent limit, thermal shutdown, output enable, internal output pulldown, and undervoltage lockout (UVLO for the new chip). This device delivers excellent line and load transient performance and supports a wide range of ESR (up to 2Ω for the new chip). The operating ambient temperature range of the device is -40° C to $+125^{\circ}$ C.

6.2 Functional Block Diagrams

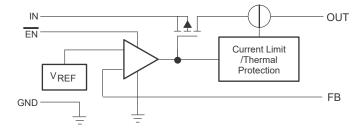


Figure 6-1. TPS76901-Q1 Functional Block Diagram (Legacy Chip)

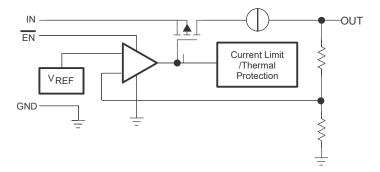


Figure 6-2. TPS769-Q1 Functional Block Diagram (Legacy Chip)

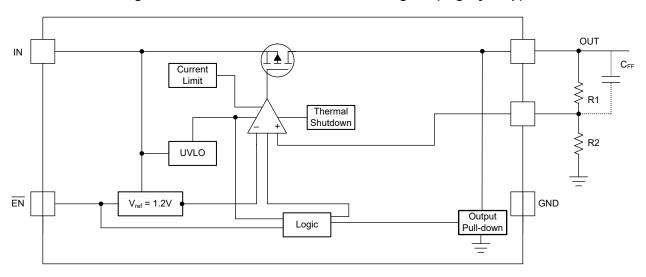


Figure 6-3. TPS76901-Q1 Functional Block Diagram (New Chip)

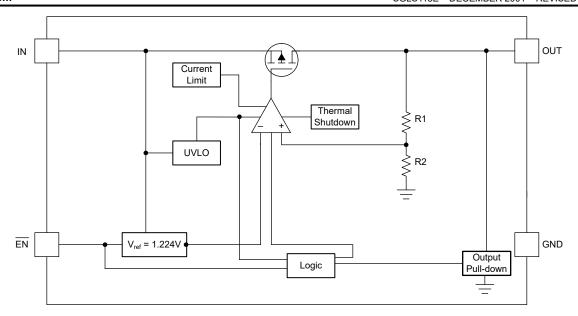


Figure 6-4. TPS769-Q1 Functional Block Diagram (New Chip)

6.3 Feature Description

6.3.1 Output Enable

The enable pin for the device is an active-low pin. The output voltage is enabled when the voltage of the enable pin is lower than the low-level input voltage of the $\overline{\text{EN}}$ pin. The output voltage is disabled when the enable pin voltage is higher than the high-level input voltage of the $\overline{\text{EN}}$ pin. If $\overline{\text{EN}}$ functionality is not needed, connect the enable pin to the GND of the device.

For the new chip, there is an internal pullup current on the \overline{EN} pin. Therefore, leave the \overline{EN} pin floating. If the \overline{EN} pin is left floating, the LDO is disabled.

In the new chip, the device has an internal output pulldown circuit that activates when the device is disabled to actively discharge the output voltage. See the *Output Pulldown* section for further information.

6.3.2 Dropout Voltage

Dropout voltage (V_{DO}) is defined as $V_{IN} - V_{OUT}$ at the rated output current (I_{RATED}), where the pass transistor is fully on. V_{IN} is the input voltage, V_{OUT} is the output voltage, and I_{RATED} is the maximum I_{OUT} listed in the *Recommended Operating Conditions* table. At this operating point, the pass transistor is driven fully on. Dropout voltage indirectly specifies a minimum input voltage greater than the nominal programmed output voltage where the output voltage is expected to stay in regulation. If the input voltage falls to less than the nominal output regulation, then the output voltage falls as well.

For a CMOS regulator, the dropout voltage is determined by the drain-source, on-state resistance ($R_{DS(ON)}$) of the pass transistor. Therefore, if the linear regulator operates at less than the rated current, the dropout voltage for that current scales accordingly. The following equation calculates the $R_{DS(ON)}$ of the device.

$$R_{DS(ON)} = \frac{V_{DO}}{I_{RATED}}$$
 (1)

6.3.3 Current Limit

The device has an internal current limit circuit that protects the regulator during transient high-load current faults or shorting events. The current limit is a brick-wall scheme. In a high-load current fault, the brick-wall scheme limits the output current to the current limit (I_{Cl}). I_{Cl} is listed in the *Electrical Characteristics* table.

The output voltage is not regulated when the device is in current limit. When a current limit event occurs, the device begins to heat up because of the increase in power dissipation. When the device is in brick-wall current limit, the pass transistor dissipates power $[(V_{IN} - V_{OUT}) \times I_{CL}]$. For more information on current limits, see the *Know Your Limits* application note.

Figure 6-5 shows a diagram of the current limit.

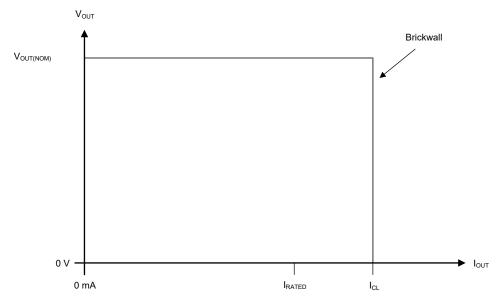


Figure 6-5. Current Limit

6.3.4 Undervoltage Lockout (UVLO)

The device has an independent undervoltage lockout (UVLO) circuit that monitors the input voltage. Thus, allowing a controlled and consistent turn on and off of the output voltage. The UVLO circuit has hysteresis functionality to prevent the device from turning off if the input drops during turn on.

6.3.5 Output Pulldown

The device (new chip) has an output pulldown circuit. The output pulldown circuit activates under the following conditions:

- The device is disabled with EN logic
- 1.0V < V_{IN} < V_{UVLO}

The output pulldown resistance for this device is $1.5k\Omega$ (typ), as listed in the *Electrical Characteristics* table.

Reverse current flows from the output to the input. Thus, do not rely on the output pulldown circuit for discharging a large amount of output capacitance after the input supply collapses. This reverse current flow potentially causes damage to the device. See the *Reverse Current* section for more details.

6.3.6 Thermal Shutdown

The device contains a thermal shutdown protection circuit to disable the device when the junction temperature (T_J) of the pass transistor rises to $T_{SD(shutdown)}$ (typical). Thermal shutdown hysteresis makes sure that the device resets (turns on) when the temperature falls to $T_{SD(reset)}$ (typical).

The thermal time-constant of the semiconductor die is fairly short. Thus the device cycles on and off when thermal shutdown is reached until power dissipation is reduced. Power dissipation during start-up is high from large $V_{\text{IN}} - V_{\text{OUT}}$ voltage drops across the device or from high inrush currents charging large output capacitors. Under some conditions, the thermal shutdown protection disables the device before start-up completes.

For reliable operation, limit the junction temperature to the maximum listed in the *Recommended Operating Conditions* table. Operation above this maximum temperature causes the device to exceed operational specifications. Although the device internal protection circuitry is designed to protect against thermal overload conditions, this circuitry is not intended to replace proper heat sinking. Continuously running the device into thermal shutdown or above the maximum recommended junction temperature reduces long-term reliability.

6.4 Device Functional Modes

Table 6-1 shows the conditions that lead to the different modes of operation. See the *Electrical Characteristics* table for parameter values.

OPERATING MODE	PARAMETER					
OPERATING WIDDE	V _{IN}	V _{EN}	I _{OUT}	TJ		
Normal operation	$V_{IN} > V_{OUT(nom)} + V_{DO}$ and $V_{IN} > V_{IN(min)}$	V _{EN} < V _{EN(LOW)}	I _{OUT} < I _{OUT(max)}	$T_J < T_{SD(shutdown)}$		
Dropout operation	$V_{IN(min)} < V_{IN} < V_{OUT(nom)} + V_{DO}$	V _{EN} < V _{EN(LOW)}	I _{OUT} < I _{OUT(max)}	T _J < T _{SD(shutdown)}		
Disabled (any true condition disables the device)	V _{IN} < V _{UVLO}	V _{EN} > V _{EN(HI)}	Not applicable	$T_J > T_{SD(shutdown)}$		

Table 6-1. Device Functional Mode Comparison

6.4.1 Normal Operation

The device regulates to the nominal output voltage when the following conditions are met:

- The input voltage is greater than the nominal output voltage plus the dropout voltage (V_{OUT(nom)} + V_{DO}).
- The current sourced from OUT is less than the current limit (I_{OUT} < I_{CL(OUT)}).
- The device junction temperature is less than the thermal shutdown temperature $(T_J < T_{SD})$.
- The enable voltage has previously receded the enable low level threshold voltage and has not yet increased higher than the enable high level threshold. Or, the EN pin is connected to ground.

6.4.2 Dropout Operation

If the input voltage is lower than the nominal output voltage plus the specified dropout voltage, the device operates in dropout mode. In this mode, the output voltage tracks the input voltage. In this mode, the transient performance of the device becomes significantly degraded. During this mode, the pass transistor is driven fully on. Line or load transients in dropout potentially result in large output voltage deviations.



When the device is in a steady dropout state, the pass transistor is driven fully on. This state is defined as when the device is in dropout, directly after being in a normal regulation state, but *not* during start-up. Dropout occurs when $V_{IN} < V_{OUT(NOM)} + V_{DO}$. When the regulator exits dropout, the input voltage returns to a value $\geq V_{OUT(NOM)} + V_{DO}$. During this time, the output voltage potentially overshoots for a short period of time. $V_{OUT(NOM)}$ is the nominal output voltage and V_{DO} is the dropout voltage. During dropout exit, the device pulls the pass transistor back from being driven fully on.

6.4.3 Disabled

Shutdown the device output by forcing the enable pin voltage to less than the maximum EN pin low-level input voltage (see the *Electrical Characteristics* table). When disabled, the pass transistor turns off and internal circuits shut down. The output voltage is also actively discharged to ground by an internal discharge circuit from the output to ground.

7 Application and Implementation

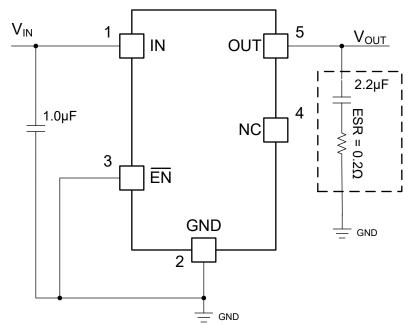
Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

7.1 Application Information

The TPS769-Q1 LDO provides a very accurate output with high PSRR and excellent line and load transient performance. The device is capable of handling up to 100mA of load current. Quiescent current consumption for the TPS769-Q1 is constant from no-load to maximum load. The TPS769-Q1 low dropout at a full 100mA load helps extend the battery operation range.

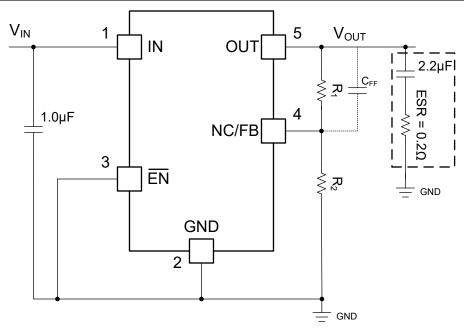
7.2 Typical Application



For fixed output voltage options only.

Figure 7-1. Typical Application Circuit (Fixed-Voltage Option)





For adjustable output voltage options only. Dotted lines indicate an optional CFF capacitor (new chip). See the Feed-Forward Capacitor (CFF) section.

Figure 7-2. Typical Application Circuit (Adjustable-Voltage Option)

Table 7-1 lists the R₁ and R₂ resistor values for the adjustable-voltage version.

Table 7-1. Adjustable Output Voltage for Resistors R₁ and R₂

	,	
OUTPUT VOLTAGE	R ₁ (kΩ)	$R_2 \; (k\Omega)$
2.5V	174	169
3.3V	287	169
3.6V	324	169
4.0V	383	169
5.0V	523	169

Product Folder Links: TPS769-Q1

Submit Document Feedback

7.2.1 Design Requirements

Table 7-2 lists the design parameters for this example.

Table 7-2. Design Parameters

PARAMETER	EXAMPLE VALUE				
Input voltage range	4V to 10V				
Output voltage	2.5V to 5V				
Output current rating	100mA				
Output capacitor	4.7μF to 10μF				
Output capacitor ESR range	200mΩ to 2Ω				

7.2.2 Detailed Design Procedure

7.2.2.1 Adjustable Device Feedback Resistors

The adjustable-version device requires external feedback divider resistors to set the output voltage. V_{OUT} is set using the feedback divider resistors, R_1 and R_2 , according to the following equation:

$$V_{OUT} = V_{REF} \times (1 + R_1 / R_2)$$
 (2)

where:

V_{REF} = 1.205V (typ) for the internal reference voltage (for the new chip)

To ignore the FB pin current error term in the V_{OUT} equation, set the feedback divider current to 100 times the FB pin current. This current is listed in the *Electrical Characteristics* table. This setting provides the maximum feedback divider series resistance, as shown in the following equation:

$$R_1 + R_2 \le V_{OUT} / (I_{FB} \times 100)$$
 (3)

In Table 7-1, examples of R_1 and R_2 values are given for different output voltage options with a feedback divider current designed at $7\mu A$.

7.2.2.2 Recommended Capacitor Types

The device (new chip) is designed to be stable using low equivalent series resistance (ESR) ceramic capacitors at the input and output. Multilayer ceramic capacitors have become the industry standard for these types of applications and are recommended, but use good judgment. Ceramic capacitors that employ X7R-, X5R-, and C0G-rated dielectric materials provide relatively good capacitive stability across temperature. However, using Y5V-rated capacitors is discouraged because of large variations in capacitance.

Regardless of the ceramic capacitor type selected, the effective capacitance varies with operating voltage and temperature. Generally, expect the effective capacitance to decrease by as much as 50%. The input and output capacitors listed in the *Recommended Operating Conditions* table account for an effective capacitance of approximately 50% of the nominal value.

7.2.2.3 Input and Output Capacitor Requirements

Although an input capacitor is not required for stability, good analog design practice is to connect a capacitor from IN to GND. This capacitor counteracts reactive input sources and improves transient response, input ripple, and PSRR. Use an input capacitor if the source impedance is more than 0.5Ω . Use a higher value capacitor if large, fast rise-time, load, or line transients are anticipated. Additionally, use a higher-value capacitor if the device is located several inches from the input power source.

As with most low-dropout regulators, the TPS769-Q1 requires an output capacitor connected between OUT and GND to stabilize the internal control loop.

For the legacy chip, the device requires an output capacitor connected between OUT and GND to stabilize the internal control loop. The minimum recommended capacitance is $4.7\mu F$. Make sure the equivalent series resistance (ESR) of the capacitor is between 0.2Ω and 10Ω to provide stability. Capacitor values larger than $4.7\mu F$ are acceptable, and allow the use of smaller ESR values. Capacitances less than $4.7\mu F$ are not

Copyright © 2024 Texas Instruments Incorporated

recommended because these components require careful selection of ESR to provide stability. Solid tantalum electrolytic, aluminum electrolytic, and multilayer ceramic capacitors are all suitable, provided these capacitors meet the described requirements. Most of the commercially available 4.7µF surface-mount solid tantalum capacitors, including devices from Sprague, Kemet, and Nichico, meet the ESR requirements previously stated. Multilayer ceramic capacitors potentially have very small equivalent series resistances and therefore require the addition of a low value series resistor to provide stability.

For the new chip, the device is designed to be stable using low ESR ceramic capacitors at the input and output. The minimum recommended capacitance value is $2.2\mu F$ and the ESR range is up to 2Ω . The supported ESR range depends on the output capacitance, operating junction temperature, and load current conditions. The Section 5.8 describes the supported ESR range in regards to the output capacitance across temperature for the supported load current range.

Dynamic performance of the device is improved by using an output capacitor. Use an output capacitor within the range specified in the *Recommended Operating Conditions* table for stability.

7.2.2.4 Reverse Current

Excessive reverse current potentially damages this device. Reverse current flows through the intrinsic body diode of the pass transistor instead of the normal conducting channel. At high magnitudes, this current flow degrades the long-term reliability of the device.

Conditions where reverse current occur are outlined in this section, all of which potentially exceed the absolute maximum rating of $V_{OUT} \le V_{IN} + 0.3V$.

- If the device has a large C_{OUT} and the input supply collapses with little or no load current
- The output is biased when the input supply is not established
- The output is biased above the input supply

If reverse current flow is expected in the application, use external protection to protect the device. Reverse current is not limited in the device, so use external limiting if extended reverse voltage operation is anticipated.

Figure 7-3 shows one approach for protecting the device.

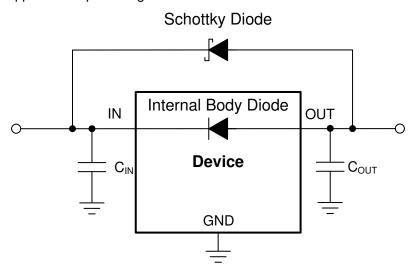


Figure 7-3. Example Circuit for Reverse Current Protection Using a Schottky Diode

7.2.2.5 Feed-Forward Capacitor (CFF)

For the adjustable-voltage version device, connect a feed-forward capacitor (C_{FF}) from the OUT pin to the FB pin. C_{FF} improves transient, noise, and PSRR performance, but is not required for regulator stability. Recommended C_{FF} values are listed in the *Recommended Operating Conditions* table. If a higher capacitance C_{FF} is used, the start-up time increases. For a detailed description of C_{FF} tradeoffs, see the *Pros and Cons of Using a Feedforward Capacitor with a Low-Dropout Regulator* application note.

 C_{FF} and R_1 form a zero in the loop gain at frequency f_Z . C_{FF} , R_1 , and R_2 form a pole in the loop gain at frequency f_P . Calculate the C_{FF} zero and pole frequencies from the following equations:

$$f_Z = 1 / (2 \times \pi \times C_{FF} \times R_1) \tag{4}$$

$$f_P = 1 / (2 \times \pi \times C_{FF} \times (R_1 || R_2))$$
 (5)

 $C_{FF} \ge 10 pF$ is required for stability if the feedback divider current is less than 5µA. The following equation calculates the feedback divider current.

$$I_{FB Divider} = V_{OUT} / (R_1 + R_2)$$
 (6)

To avoid start-up time increases from CFF, limit the product CFF × R1 to less than 50µs.

For an output voltage of 1.2V (for new chip) with the FB pin tied to the OUT pin, no CFF is used.

7.2.2.6 Power Dissipation (P_D)

Circuit reliability requires consideration of the device power dissipation, location of the circuit on the printed circuit board (PCB), and correct thermal plane sizing. Place few or no other heat-generating devices that cause added thermal stress in the PCB area around the regulator.

To first-order approximation, power dissipation in the regulator depends on the input-to-output voltage difference and load conditions. The following equation calculates power dissipation (P_D).

$$P_{D} = (V_{IN} - V_{OUT}) \times I_{OUT}$$
(7)

Note

Minimize power dissipation, and therefore achieve greater efficiency, by correctly selecting the system voltage rails. For the lowest power dissipation use the minimum input voltage required for correct output regulation.

For devices with a thermal pad, the primary heat conduction path for the device package is through the thermal pad to the PCB. Solder the thermal pad to a copper pad area under the device. This pad area contains an array of plated vias that conduct heat to additional copper planes for increased heat dissipation.

The maximum power dissipation determines the maximum allowable ambient temperature (T_A) for the device. Power dissipation and junction temperature are most often related by the $R_{\theta JA}$ of the combined PCB, device package, and the ambient air temperature (T_A). $R_{\theta JA}$ is the junction-to-ambient thermal resistance. The following equation calculates this relationship.

$$T_{J} = T_{A} + (R_{\theta,JA} \times P_{D}) \tag{8}$$

Thermal resistance $(R_{\theta JA})$ is highly dependent on the heat-spreading capability built into the particular PCB design. $R_{\theta JA}$ therefore varies according to the total copper area, copper weight, and location of the planes. The junction-to-ambient thermal resistance listed in the *Thermal Information* table is determined by the JEDEC standard PCB and copper-spreading area. $R_{\theta JA}$ is used as a relative measure of package thermal performance. $R_{\theta JA}$ is improved by 35% to 55% compared to the *Thermal Information* table value by optimizing the PCB board layout. See the *An empirical analysis of the impact of board layout on LDO thermal performance* application note for further information.

7.2.2.7 Estimating Junction Temperature

The JEDEC standard now recommends the use of psi (Ψ) thermal metrics. These metrics estimate the junction temperatures of the linear regulator when in-circuit on a typical PCB board application. These metrics are not thermal resistance parameters and instead offer a practical and relative way to estimate junction temperature. These psi metrics are determined to be significantly independent of the copper area available for heat spreading. The *Thermal Information* table lists the primary thermal metrics, which are the junction-to-top characterization

Copyright © 2024 Texas Instruments Incorporated

Submit Document Feedback



parameter (ψ_{JT}) and junction-to-board characterization parameter (ψ_{JB}) . These parameters provide two methods for calculating the junction temperature (T_J) , as described in the following equations. Use the junction-to-top characterization parameter (ψ_{JT}) with the temperature at the center-top of device package (T_T) to calculate the junction temperature. Use the junction-to-board characterization parameter (ψ_{JB}) with the PCB surface temperature 1mm from the device package (T_B) to calculate the junction temperature.

$$T_{J} = T_{T} + \psi_{JT} \times P_{D} \tag{9}$$

where:

- P_D is the dissipated power
- T_T is the temperature at the center-top of the device package

$$T_{J} = T_{B} + \psi_{JB} \times P_{D} \tag{10}$$

where:

• T_B is the PCB surface temperature measured 1mm from the device package and centered on the package edge

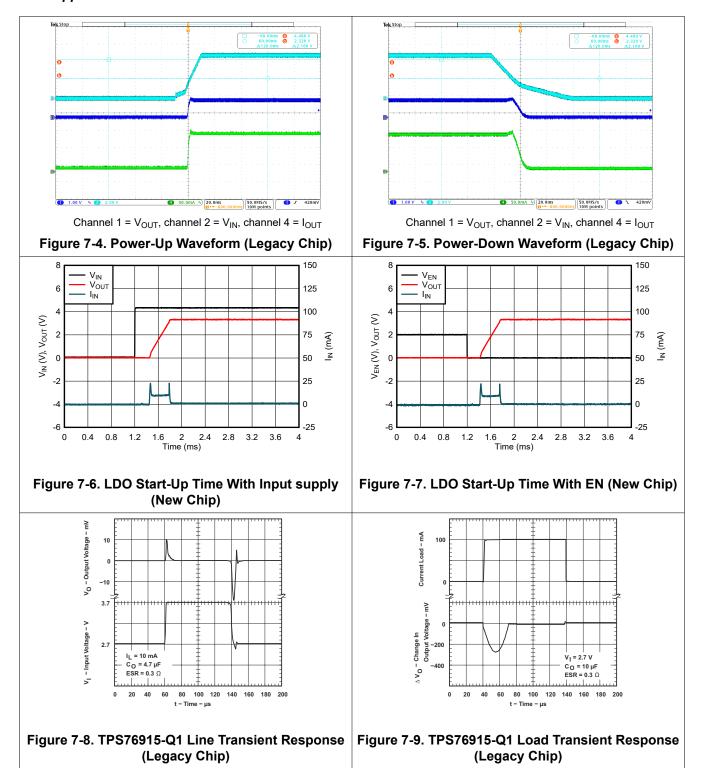
For detailed information on the thermal metrics and how to use them, see the *Semiconductor and IC Package Thermal Metrics* application note.

Submit Document Feedback

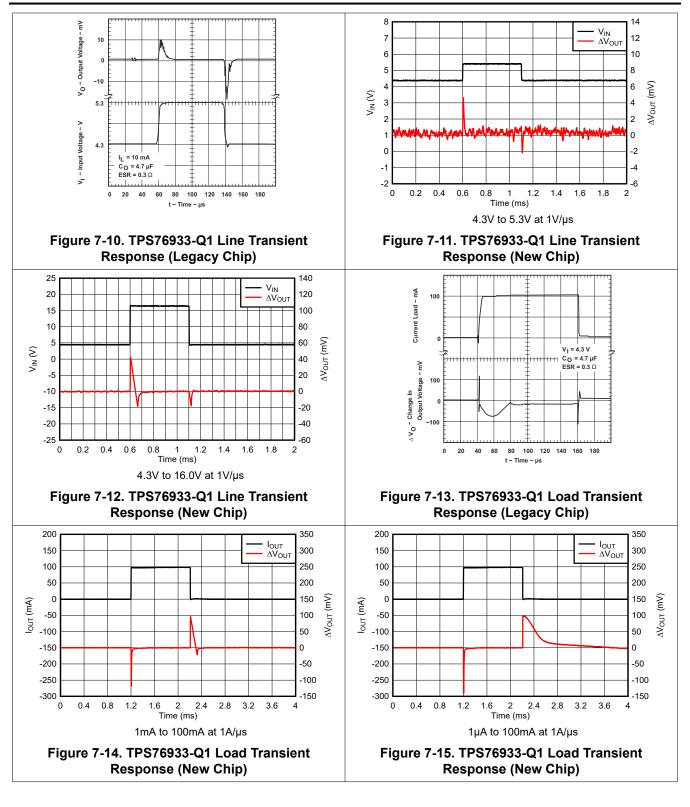
Copyright © 2024 Texas Instruments Incorporated

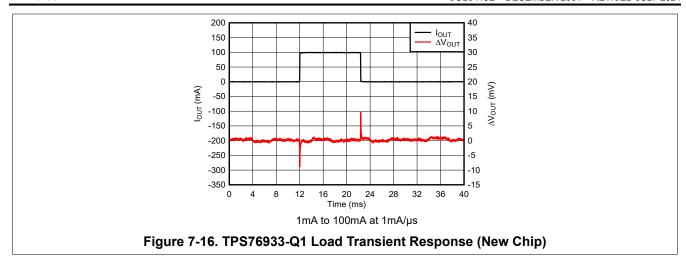


7.2.3 Application Curves









7.3 Power Supply Recommendations

The TPS769-Q1 is designed to operate from an input voltage supply range between 2.5V and 16V (new chip). The input voltage range provides adequate headroom for the device to have a regulated output. If the input supply is noisy, additional input capacitors with low ESR help improve the output noise performance.

7.4 Layout

7.4.1 Layout Guidelines

For the LDO power supply, especially high voltage and large current supplies, layout is an important step. If layout is not carefully designed, the regulator potentially does not deliver enough output current because of thermal limitation. Spread the GND layer as large as possible and place enough thermal vias on the thermal pad. These steps help improve device thermal performance, and maximize the current output at high ambient temperature. *Figure 7-17* shows an example layout.

7.4.2 Layout Example

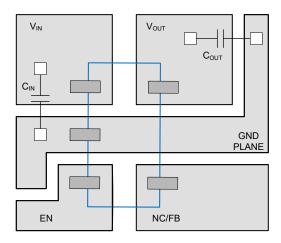


Figure 7-17. Layout Recommendation

8 Device and Documentation Support

8.1 Device Support

8.1.1 Development Support

8.1.1.1 Evaluation Module

An evaluation module (EVM) is available to assist in the initial circuit performance evaluation using the TPS769-Q1. Request the TPS76901EVM-127 evaluation module (and related user's guide) at the TI website through the product folders or purchase directly from the TI eStore.

8.1.1.2 Spice Models

Computer simulation of circuit performance using SPICE is often useful when analyzing the performance of analog circuits and systems. A SPICE model for the TPS769-Q1 is available through the product folders under *Tools & Software*.

8.1.1.3 Device Nomenclature

Table 8-1. Device Nomenclature

PRODUCT ⁽¹⁾	V _{OUT}
TPS769 xxyyyz Legacy chip	xx is the nominal output voltage (for example, 28 = 2.8V, 285 = 2.85V, and 01 = Adjustable). yyy is the package designator. z is the package quantity.
TPS769 xxyyyzM3 New chip	 xx is the nominal output voltage (for example, 28 = 2.8V, 285 = 2.85V, and 01 = Adjustable). yyy is the package designator. z is the package quantity. M3 is a suffix designator for new chip redesigns on the latest TI process technology.

⁽¹⁾ For the most current package and ordering information see the Package Option Addendum at the end of this document, or see the TI website at www.ti.com.

8.2 Documentation Support

8.2.1 Related Documentation

For related documentation see the following:

- Texas Instruments, TPS709-Q1 150-mA, 30-V, 1-μA IQ Voltage Regulators With Enable data sheet
- Texas Instruments. Universal LDO Evaluation Module user guide
- Texas Instruments, Pros and Cons of Using a Feedforward Capacitor with a Low-Dropout Regulator application note
- Texas Instruments, Know Your Limits application note

8.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

8.4 Support Resources

TI E2E[™] support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

8.5 Trademarks

TI E2E™ is a trademark of Texas Instruments.

All trademarks are the property of their respective owners.

8.6 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

8.7 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

9 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes fr	om Revision D (September 2016) to Revision E (July 2024)	Page
 Updated 	the numbering format for tables, figures, and cross-references throughout the	e document1
 Changed 	d entire document to align with current family format	1
 Added M 	13 devices to document	1
Changes fr	rom Revision C (June 2012) to Revision D (September 2016)	Page
Applicati and Doc	SD Ratings table, Thermal Information table, Feature Description section, Deformentation section, Power Supply Recommendations section, Lagumentation Support section, and Mechanical, Packaging, and Orderable Infor	yout section, Device
• Domovo	d Ordering Information table, see POA at the end of the data sheet	

10 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

www.ti.com

8-Nov-2025

PACKAGING INFORMATION

Orderable part number	Status	Material type	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
TPS76901QDBVRG4MQ1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCFQ
TPS76901QDBVRG4MQ1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCFQ
TPS76901QDBVRG4Q1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCFQ
TPS76901QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCFQ
TPS76901QDBVRM3Q1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	PCFQ
TPS76901QDBVRM3Q1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	PCFQ
TPS76901QDBVRQ1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCFQ
TPS76901QDBVRQ1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCFQ
TPS76915QDBVRG4Q1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCHQ
TPS76915QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCHQ
TPS76918QDBVRG4Q1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCIQ
TPS76918QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCIQ
TPS76918QDBVRQ1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCIQ
TPS76918QDBVRQ1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCIQ
TPS76925QDBVRG4Q1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCJQ
TPS76925QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCJQ
TPS76925QDBVRQ1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCJQ
TPS76925QDBVRQ1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCJQ
TPS76927QDBVRG4Q1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCKQ
TPS76927QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCKQ
TPS76928QDBVRG4Q1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCLQ
TPS76928QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCLQ
TPS76930QDBVRG4Q1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCMQ
TPS76930QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCMQ
TPS76930QDBVRQ1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCMQ
TPS76930QDBVRQ1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCMQ
TPS76933QDBVRG4Q1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCNQ
TPS76933QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCNQ
TPS76933QDBVRM3Q1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	PCNQ



8-Nov-2025



www.ti.com

Orderable part number	Status	Material type	Package Pins	Package qty Carrier	RoHS	Lead finish/	MSL rating/	Op temp (°C)	Part marking
	(1)	(2)			(3)	Ball material	Peak reflow		(6)
						(4)	(5)		
TPS76933QDBVRM3Q1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	PCNQ
TPS76933QDBVRQ1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCNQ
TPS76933QDBVRQ1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCNQ
TPS76950QDBVRG4MQ1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCOQ
TPS76950QDBVRG4MQ1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCOQ
TPS76950QDBVRG4Q1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCOQ
TPS76950QDBVRG4Q1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCOQ
TPS76950QDBVRM3Q1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	PCOQ
TPS76950QDBVRM3Q1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	PCOQ
TPS76950QDBVRQ1	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCOQ
TPS76950QDBVRQ1.A	Active	Production	SOT-23 (DBV) 5	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PCOQ

⁽¹⁾ Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative

⁽²⁾ Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

⁽⁴⁾ Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

PACKAGE OPTION ADDENDUM

www.ti.com 8-Nov-2025

and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

OTHER QUALIFIED VERSIONS OF TPS769-Q1:

Catalog: TPS769

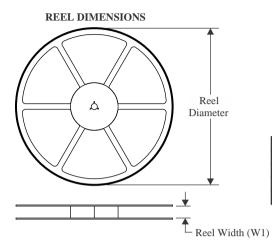
NOTE: Qualified Version Definitions:

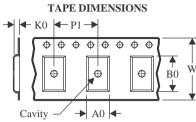
• Catalog - TI's standard catalog product



www.ti.com 6-Dec-2025

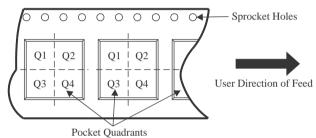
TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS76901QDBVRG4MQ1	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPS76901QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76901QDBVRM3Q1	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPS76901QDBVRQ1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76915QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76918QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76918QDBVRQ1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76925QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76925QDBVRQ1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76927QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76928QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76930QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76930QDBVRQ1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76933QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76933QDBVRM3Q1	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPS76933QDBVRQ1	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3



PACKAGE MATERIALS INFORMATION

www.ti.com 6-Dec-2025

Device	Package Type	Package Drawing			Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS76933QDBVRQ1	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS76950QDBVRG4MQ1	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPS76950QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPS76950QDBVRG4Q1	SOT-23	DBV	5	3000	178.0	9.0	3.3	3.2	1.4	4.0	8.0	Q3
TPS76950QDBVRM3Q1	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPS76950QDBVRQ1	SOT-23	DBV	5	3000	178.0	9.0	3.3	3.2	1.4	4.0	8.0	Q3
TPS76950QDBVRQ1	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3



www.ti.com 6-Dec-2025



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS76901QDBVRG4MQ1	SOT-23	DBV	5	3000	210.0	185.0	35.0
TPS76901QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76901QDBVRM3Q1	SOT-23	DBV	5	3000	210.0	185.0	35.0
TPS76901QDBVRQ1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76915QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76918QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76918QDBVRQ1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76925QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76925QDBVRQ1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76927QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76928QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76930QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76930QDBVRQ1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76933QDBVRG4Q1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76933QDBVRM3Q1	SOT-23	DBV	5	3000	210.0	185.0	35.0
TPS76933QDBVRQ1	SOT-23	DBV	5	3000	210.0	185.0	35.0
TPS76933QDBVRQ1	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS76950QDBVRG4MQ1	SOT-23	DBV	5	3000	210.0	185.0	35.0



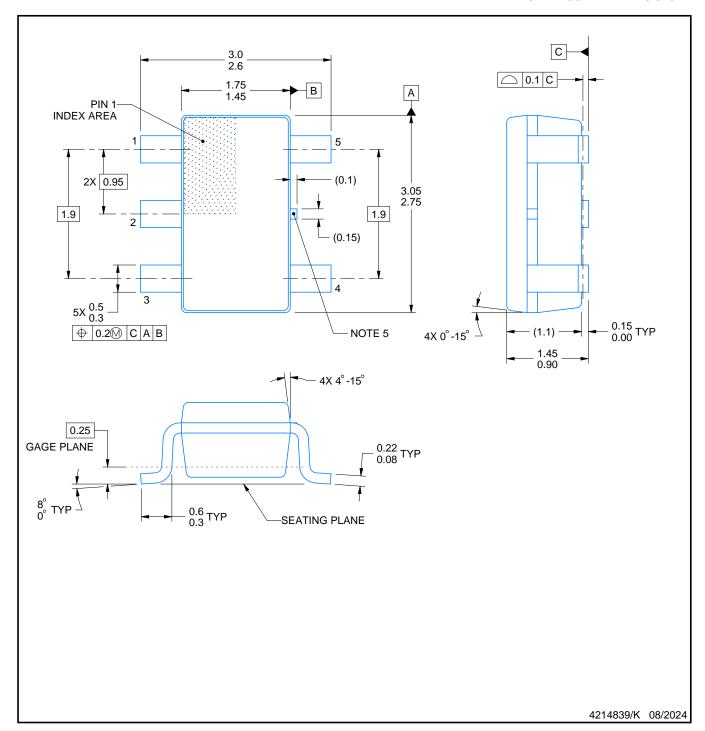
PACKAGE MATERIALS INFORMATION

www.ti.com 6-Dec-2025

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS76950QDBVRG4Q1	SOT-23	DBV	5	3000	210.0	185.0	35.0
TPS76950QDBVRG4Q1	SOT-23	DBV	5	3000	180.0	180.0	18.0
TPS76950QDBVRM3Q1	SOT-23	DBV	5	3000	210.0	185.0	35.0
TPS76950QDBVRQ1	SOT-23	DBV	5	3000	180.0	180.0	18.0
TPS76950QDBVRQ1	SOT-23	DBV	5	3000	210.0	185.0	35.0



SMALL OUTLINE TRANSISTOR



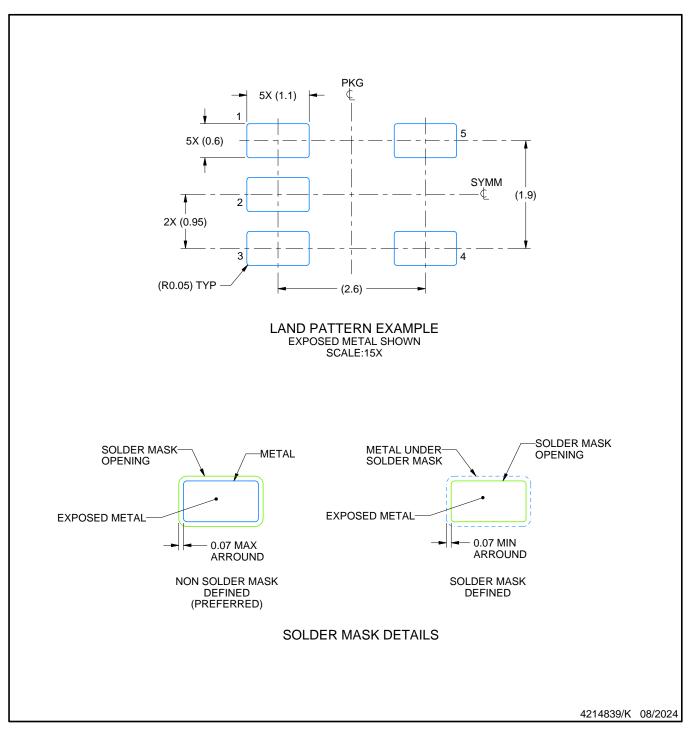
NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
 2. This drawing is subject to change without notice.
 3. Reference JEDEC MO-178.

- 4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.
- 5. Support pin may differ or may not be present.



SMALL OUTLINE TRANSISTOR



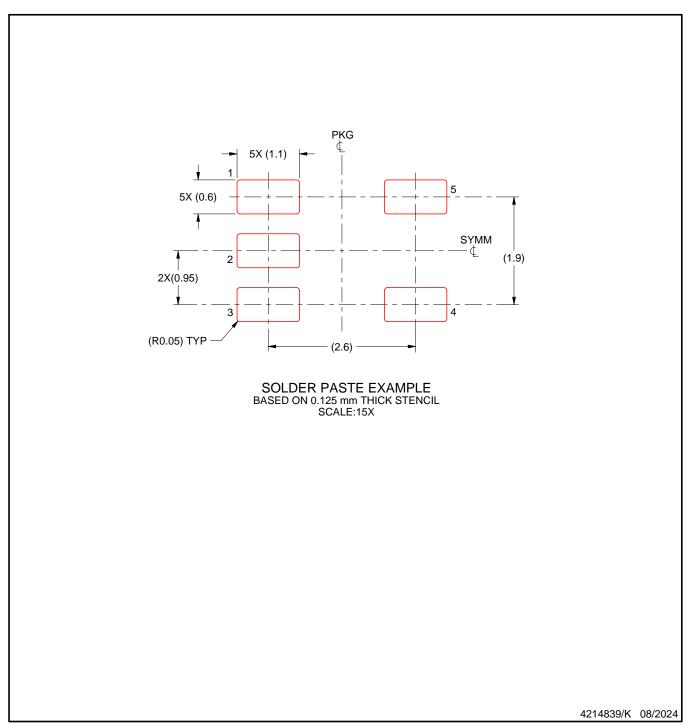
NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE TRANSISTOR



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



IMPORTANT NOTICE AND DISCLAIMER

TI PROVIDES TECHNICAL AND RELIABILITY DATA (INCLUDING DATASHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS AND IMPLIED, INCLUDING WITHOUT LIMITATION ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE OR NON-INFRINGEMENT OF THIRD PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for skilled developers designing with TI products. You are solely responsible for (1) selecting the appropriate TI products for your application, (2) designing, validating and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, regulatory or other requirements.

These resources are subject to change without notice. TI grants you permission to use these resources only for development of an application that uses the TI products described in the resource. Other reproduction and display of these resources is prohibited. No license is granted to any other TI intellectual property right or to any third party intellectual property right. TI disclaims responsibility for, and you fully indemnify TI and its representatives against any claims, damages, costs, losses, and liabilities arising out of your use of these resources.

TI's products are provided subject to TI's Terms of Sale, TI's General Quality Guidelines, or other applicable terms available either on ti.com or provided in conjunction with such TI products. TI's provision of these resources does not expand or otherwise alter TI's applicable warranties or warranty disclaimers for TI products. Unless TI explicitly designates a product as custom or customer-specified, TI products are standard, catalog, general purpose devices.

TI objects to and rejects any additional or different terms you may propose.

Copyright © 2025, Texas Instruments Incorporated

Last updated 10/2025