Analytic Expressions for currents in the CCM PFC stage

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SLYY131

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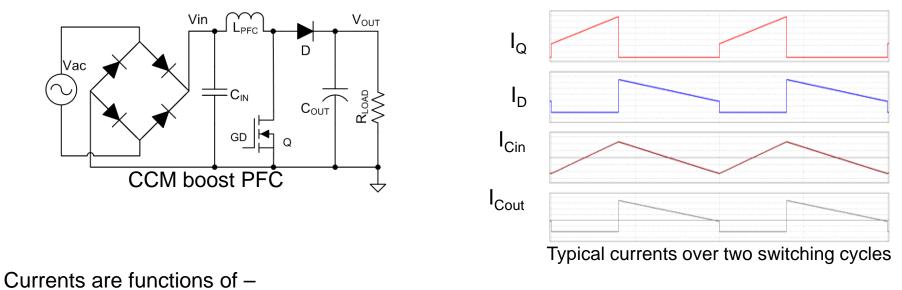


Agenda

- CCM Boost PFC stage with current waveforms
 - RMS calculation
 - RMS current expressions, Diode, MOSFET, C_{OUT} , C_{IN}
- Two Phase Interleaved CCM Boost PFC
 - RMS current expressions, Diode, MOSFET
 - RMS current calculation, C_{OUT}
 - RMS current expression, C_{OUT}
- Results
- Conclusions
- References

Note: currents are rms unless otherwise stated

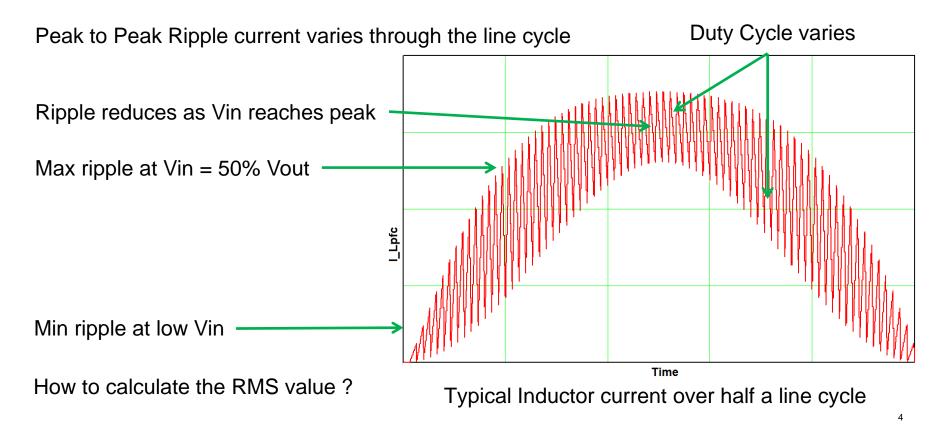
CCM Boost PFC Waveforms



Parameters: L_{PFC}, F_{sw}, V_{OUT}, I_{OUT} ← Assumed to be constant:

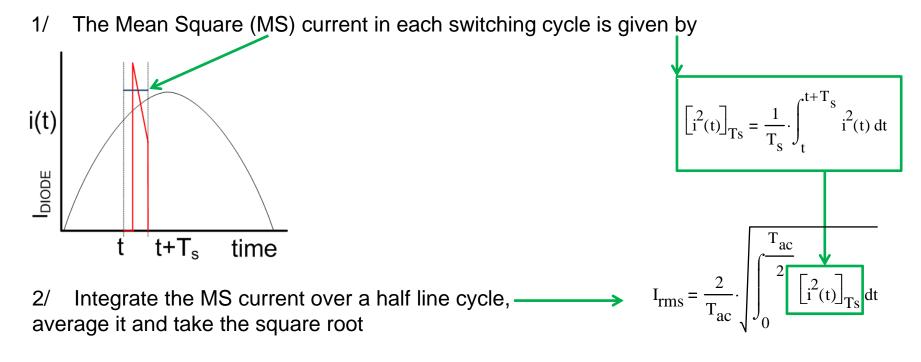
Variables: Vac, Duty Cycle (depends on Line phase angle)

CCM Boost PFC Waveforms



RMS Calculation – PFC Diode Current

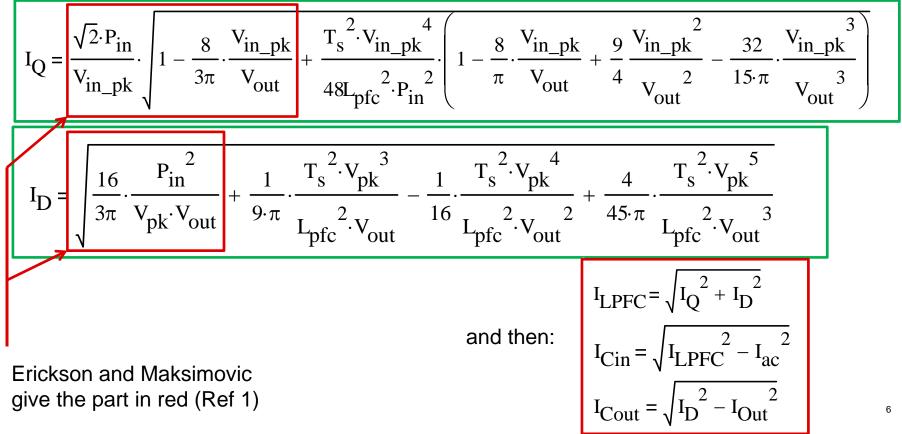
I²(t) is first averaged over a switching cycle then averaged over the AC line period



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RMS Currents, MOSFET and Diode



RMS Currents, C_{OUT}

The total current in C_{OUT} has two components.

A LF component at twice line frequency Ref (3)

$$I_{\text{Cout_LF}} = \frac{P_{\text{in}}}{\eta \cdot \sqrt{2} \cdot V_{\text{out}}}$$

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A HF component at the switching frequency and its harmonics. This current is $I_{Diode} - AVG(I_{Diode})$ – the calculation will be outlined later (slide 16)

$$I_{\text{Cout_HF_1Ph}} \coloneqq \sqrt{\frac{16}{3 \cdot \pi} \cdot \frac{P_{\text{in}}^{2}}{V_{\text{in_pk}} \cdot V_{\text{out}}} - \frac{3}{8} \cdot \frac{4 \cdot P_{\text{in}}^{2}}{V_{\text{out}}^{2}} + \frac{4}{3 \cdot \pi} \cdot \frac{T_{s}^{2} \cdot V_{\text{in_pk}}^{3}}{12 \cdot L_{\text{pfc}}^{2} \cdot V_{\text{out}}} - \frac{3}{8} \cdot \frac{T_{s}^{2} \cdot V_{\text{in_pk}}^{4}}{6 \cdot L_{\text{pfc}}^{2} \cdot V_{\text{out}}^{2}} + \frac{16}{15 \cdot \pi} \cdot \frac{T_{s}^{2} \cdot V_{\text{in_pk}}^{5}}{12 \cdot L_{\text{pfc}}^{2} \cdot V_{\text{out}}^{3}}$$
The total current in C_{OUT} is then
$$I_{\text{Cout}} = \sqrt{I_{\text{Cout_HF}}^{2} + I_{\text{Cout_LF}}^{2}}$$

RMS Currents, C_{IN}

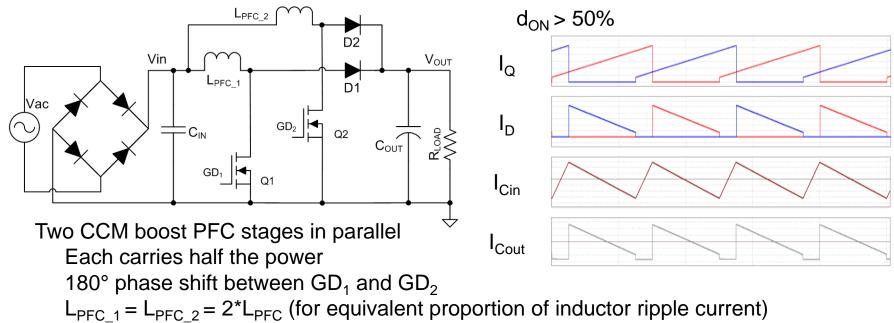
The total ripple current in C_{IN} is

$$I_{\rm Cin} = \sqrt{I_{\rm LPFC}^2 - I_{\rm ac}^2}$$

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 C_{IN} is usually small and there is effectively no line frequency current in this capacitor The RMS current in this capacitor is high frequency only.

2 Phase Interleaved boost PFC



 I_{Cout} Ripple at twice line frequency is unaffected by interleaving HF ripple is at twice the switching frequency Switching frequency ripple reduction in $C_{\rm IN}$ and $C_{\rm OUT}$ is a function of the duty cycle

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RMS Currents, 2 Phase, MOSFET and Diode

Calculation for MOSFET and diode currents is the same as in the 1 Phase case

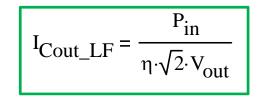
The input power is shared between the two phases so P_{in} is replaced by $P_{in}/2$ There is no ripple current cancellation in the MOSFET or Diodes

$$I_{Q} = \frac{\sqrt{2} \cdot \frac{P_{in}}{2}}{V_{in_pk}} \cdot \sqrt{1 - \frac{8}{3\pi} \cdot \frac{V_{in_pk}}{V_{out}} + \frac{T_{s}^{2} \cdot V_{in_pk}}{48L_{pfc}^{2} \cdot P_{in}^{2}}} \cdot \left(1 - \frac{8}{\pi} \cdot \frac{V_{in_pk}}{V_{out}} + \frac{9}{4} \frac{V_{in_pk}}{V_{out}^{2}} - \frac{32}{15 \cdot \pi} \cdot \frac{V_{in_pk}}{V_{out}^{3}}\right)$$

$$I_{D} = \sqrt{\frac{16}{3\pi} \cdot \frac{\left(\frac{P_{in}}{2}\right)^{2}}{V_{pk} \cdot V_{out}} + \frac{1}{9 \cdot \pi} \cdot \frac{T_{s}^{2} \cdot V_{pk}^{3}}{L_{pfc}^{2} \cdot V_{out}} - \frac{1}{16} \cdot \frac{T_{s}^{2} \cdot V_{pk}^{4}}{L_{pfc}^{2} \cdot V_{out}^{2}} + \frac{4}{45 \cdot \pi} \cdot \frac{T_{s}^{2} \cdot V_{pk}^{5}}{L_{pfc}^{2} \cdot V_{out}^{3}}}$$

2 Phase, I_{Cout_rms} Calculation Outline

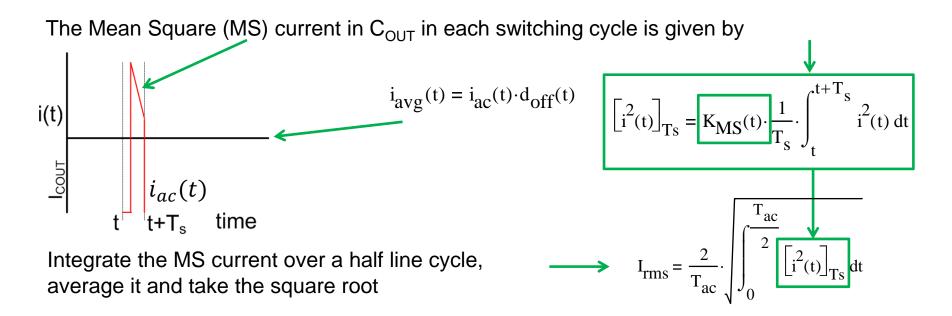
• The LF current ripple in C_{OUT} is unchanged by interleaving



- Calculate the 1 Ph HF current in C_{OUT}
- Determine a correction factor (K_{MS}(t))
- Apply the correction factor to the 1 Ph HF solution to get the 2 Ph HF solution
- Add the 2 Ph HF current to the LF current to get the total

RMS Calculation – Calculate the HF ripple in Cout

I²(t) is first averaged over a switching cycle then averaged over the AC line period



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Correction Factor

Plot

1 Ph and 2 Ph I_{cout} currents simulated (PSIM) Mean Square values taken across d_{on} range

md = 1.2 cd = -0.6 If d_{on} > 50% md = -1.2 cd = 0.6 If d_{on} < 50%

Linearise $K_{MS}(d_{on}) = md \ d_{on} + cd$ ------

Duty cycle d_{on} is a function of time Restate in terms of t

 $\frac{I_{2Ph}}{I_{1Ph}}$ results against d_{on}

 $K_{MS}(t) = mt \; \frac{V_{in_pk}}{V_{out}} \sin(\omega t) + ct$

mt = -1.2 ct = 0.6 If d_{on} > 50% mt = 1.2 ct = -0.6 If d_{on} < 50%

$$d_{on}(t) = 1 - \frac{V_{pk}}{V_{out}} \cdot \sin(\omega \cdot t)$$

C_{OUT} – Switching Cycle HF Mean Square Current

 $I_{MS_Trap} = d \cdot I_{mid}^2 + \frac{d}{3} \cdot \Delta I(t)^2$ $-i_{ava}(t)$ During Ton current is During Toff current is $i_{ac}(t) - i_{avg}(t)$ Cou ¢∆Ι Imid $\left[i(t)^{2}\right]_{\text{Ts}} = i_{\text{avg}}(t)^{2} \cdot \left(1 - d_{\text{off}}\right) + d_{\text{off}} \cdot i_{\text{mid}}(t)^{2} + \frac{d_{\text{off}}}{3} \cdot \Delta i(t)^{2}$ So Multiply by $K_{MS}(t)$ for the interleaved case t off t on Τs $\left[\left[i_{Cout}^{2}(t)\right]_{T_{s}} = K_{MS}(t) \cdot \left[\left(i_{avg}(t)\right)^{2} \cdot \left(1 - d_{off}(t)\right) + d_{off}(t) \cdot \left(i_{mid}(t)\right)^{2} + \frac{d_{off}(t)}{3} \cdot \Delta i(t)^{2}\right]\right]$ Now we need expressions for the functions of t in this equation 14

Analytic Expressions for currents in the CCM PFC stage

Diode and MOSFET currents are trapezoidal

Substitutions

We have the following

 $d_{off}(t) = \frac{V_{in_pk}}{V_{out}} \cdot \sin(\omega \cdot t)$ $\Delta i(t) = \frac{T_{s} \cdot V_{in_pk}}{2 \cdot L_{nfc}} \cdot \left(1 - \frac{V_{in_pk}}{V_{out}} \cdot \sin(\omega \cdot t)\right) \cdot \sin(\omega \cdot t)$ $i_{ac}(t) = \frac{\sqrt{2} \cdot P_{in}}{V_{in nk}} \cdot \sin(\omega \cdot t)$ $I_{mid}(t) = I_{ac}(t) - I_{avg}(t)$ $K_{MS}(t) = mt \cdot \frac{V_{pk}}{V_{out}} \cdot \sin(\omega \cdot t) + ct$ 15

C_{OUT} – Switching Cycle HF Mean Square Current

$$\left[i_{\text{Cout}}^{2}(t)\right]_{\text{Ts}} = K_{\text{MS}}(t) \cdot \left[\left(i_{\text{avg}}(t)\right)^{2} \cdot \left(1 - d_{\text{off}}(t)\right) + d_{\text{off}}(t) \cdot \left(i_{\text{mid}}(t)\right)^{2} + \frac{d_{\text{off}}(t)}{3} \cdot \Delta i(t)^{2}\right]$$

Set $K_{MS}(t)$ to 1 for the 1Ph solution

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• Making the substitutions and simplifying gives

$$\begin{bmatrix} i_{\text{Cout}}^{2}(t) \end{bmatrix} = \text{mt} \cdot \frac{4P_{\text{in}}^{2}}{V_{\text{out}}^{2}} \cdot \left(\sin(\omega \cdot t)^{4} - \frac{V_{\text{pk}} \cdot \sin(\omega \cdot t)^{5}}{V_{\text{out}}^{2}} \right) + \text{ct} \cdot \frac{4 \cdot P_{\text{in}}^{2}}{V_{\text{out}}} \cdot \left(\frac{\sin(\omega \cdot t)^{3}}{V_{\text{pk}}} - \frac{\sin(\omega \cdot t)^{4}}{V_{\text{out}}} \right)$$

• Integrating gives the rms value
$$I_{RMS} = \frac{2}{T_{AC}} \sqrt{\int_{0}^{\frac{T_{AC}}{2}} [i^{2}(t)]_{T_{SW}}} dt$$

$$\boxed{I_{\text{Cout}_\text{HF}_2Ph} = \sqrt{\text{mt}} \left(\frac{3}{2} \cdot \frac{P_{\text{in}}^{2}}{V_{\text{out}}^{2}} - \frac{64}{15 \cdot \pi} \cdot \frac{P_{\text{in}}^{2} \cdot V_{\text{in}_pk}}{V_{\text{out}}^{3}} \right) + \text{ct} \cdot \left(\frac{16}{3 \cdot \pi} \cdot \frac{P_{\text{in}}^{2}}{V_{\text{in}_pk} \cdot V_{\text{out}}} - \frac{3}{2} \cdot \frac{P_{\text{in}}^{2}}{V_{\text{out}}^{2}} \right)}{V_{\text{out}}^{2}} \right)}$$

C_{OUT} – **RMS** Current

We have

$$I_{\text{Cout_HF}_2\text{Ph}} = \sqrt{\text{mt}} \left(\frac{3}{2} \cdot \frac{P_{\text{in}}^2}{V_{\text{out}}^2} - \frac{64}{15 \cdot \pi} \cdot \frac{P_{\text{in}}^2 \cdot V_{\text{in}_\text{pk}}}{V_{\text{out}}^3} \right) + \text{ct}} \left(\frac{16}{3 \cdot \pi} \cdot \frac{P_{\text{in}}^2}{V_{\text{in}_\text{pk}} \cdot V_{\text{out}}} - \frac{3}{2} \cdot \frac{P_{\text{in}}^2}{V_{\text{out}}^2} \right)$$

$$P_{\text{in}}$$

And
$$I_{\text{Cout_LF}} = \frac{\Gamma_{\text{in}}}{\eta \cdot \sqrt{2} \cdot V_{\text{out}}}$$

So

$$I_{\text{Cout}_2\text{Ph}} = \sqrt{I_{\text{Cout}_H\text{F}}^2 + I_{\text{Cout}_L\text{F}}^2}$$

Analytic Expressions for currents in the CCM PFC stage

Valid for V_{in_pk} < Vout/2

If V_{in_pk} > Vout/2 then the MS calculation and the integral have to be split. One for 0 < Vin < Vout/2 One for Vout/2 < Vin < Vout/2 And the results added

Results*

Taken on a Texas Instruments UCC28070 EVM - 2Ph interleaved CCM Boost PFC

ICout_2Ph				
Vin	Pin	Meas, mA	Calc, mA	Error
120V	300W	817	732	10.4%
120V	200W	702	610	13.1%
120V	100W	595	488	18.0%
90V	300W	976	946	3.1%
90V	200W	821	788	4.0%
90V	100W	684	631	7.7%

C_{OUT} , Ripple reduction example

Icout (Calc, mA)			
	1Ph	2Ph	Ratio
300W, 120V	1407	732	52%
300W, 90V	1633	946	58%

* With thanks to Sonal Singh for taking the measurements

Analytic Expressions for currents in the CCM PFC stage

IQ_2Ph				
Vin	Pin	Meas, mA	Calc, mA	Error
120V	300W	1185	1170	1.3%
120V	200W	1029	1035	0.6%
120V	100W	880	910	3.4%
90V	300W	1581	1525	3.5%
90V	200W	1355	1309	3.4%
90V	100W	1128	1102	2.3%

ID_2Ph				
V_in	Pin	Meas, mA	Calc, mA	Error
120V	300W	788	895	13.6%
120V	200W	670	788	17.6%
120V	100W	555	690	24.3%
90V	300W	880	949	7.8%
90V	200W	741	814	9.9%
90V	100W	602	684	13.6%

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Conclusions

• Equations for the RMS currents in the CCM boost PFC developed and presented

Diode

MOSFET

Input Capacitor

Output Capacitor

- Significant reduction in RMS currents in the interleaved CCM Boost PFC
- Output capacitor High Frequency ripple current significantly reduced by interleaving
- Output capacitor Low Frequency ripple current unaffected by interleaving
- Output capacitor total ripple current reduced by interleaving
- Good agreement with experimental results

References

Ref (1): 'Fundamentals of Power Electronics, Erickson and Maksimovic; Springer 2001, Table 18.3, summary of rectifier current stresses.

Ref (2): SLUP279 An Interleaving PFC Pre-Regulator for High-Power Converters. Mike O'Loughlin

Ref (3): Capacitor Ripple current in an interleaved PFC converter, Pratt and Jinsong,

IEEE transactions on Power Electronics, Vol 24, No 6 June 2009.

UCC28070: http://www.ti.com/product/UCC28070

UCC28180: http://www.ti.com/product/UCC28180

Useful Integrals
$$\frac{1}{\pi} \cdot \int_{0}^{\pi} \sin(\theta) \, d\theta = \frac{2}{\pi} \qquad \frac{1}{\pi} \cdot \int_{0}^{\pi} (\sin(\theta))^{2} \, d\theta = \frac{1}{2} \qquad \frac{1}{\pi} \cdot \int_{0}^{\pi} (\sin(\theta))^{3} \, d\theta = \frac{4}{3 \cdot \pi}$$
$$\frac{1}{\pi} \cdot \int_{0}^{\pi} (\sin(\theta))^{4} \, d\theta = \frac{3}{8} \qquad \frac{1}{\pi} \cdot \int_{0}^{\pi} (\sin(\theta))^{5} \, d\theta = \frac{16}{15 \cdot \pi} \qquad \frac{1}{\pi} \cdot \int_{0}^{\pi} (\sin(\theta))^{6} \, d\theta = \frac{15}{48}$$

Thank You

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