

Data Converter Basics

How to Use and Test Data-Acquisition Products

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Texas Instruments, Inc.

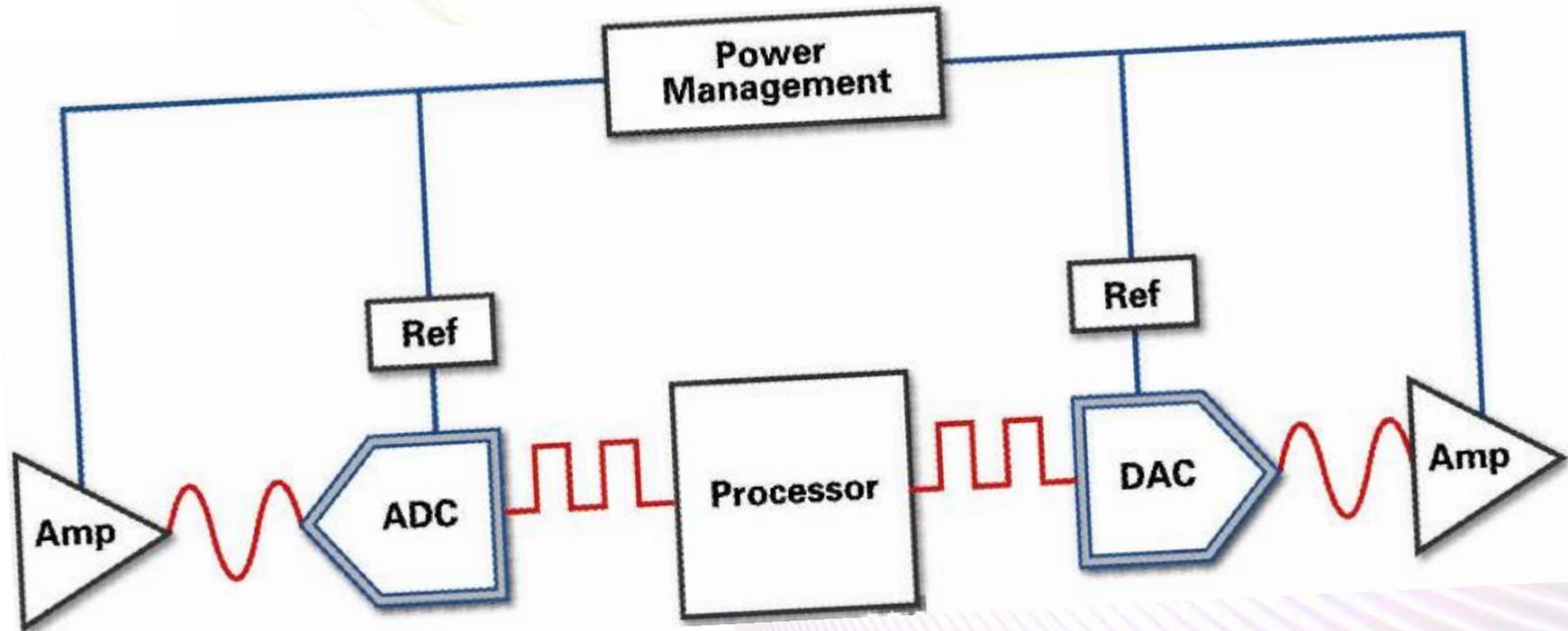


Workshop Agenda

- Data Converter Fundamentals
 - Key concerns when choosing a data converter
 - Data converter and analog design tools
- Design Considerations
 - Input drive considerations
 - Cost and performance tradeoffs
 - PCB layout
- Putting the Pieces Together...
 - Software tool demos
 - Hardware tool demos

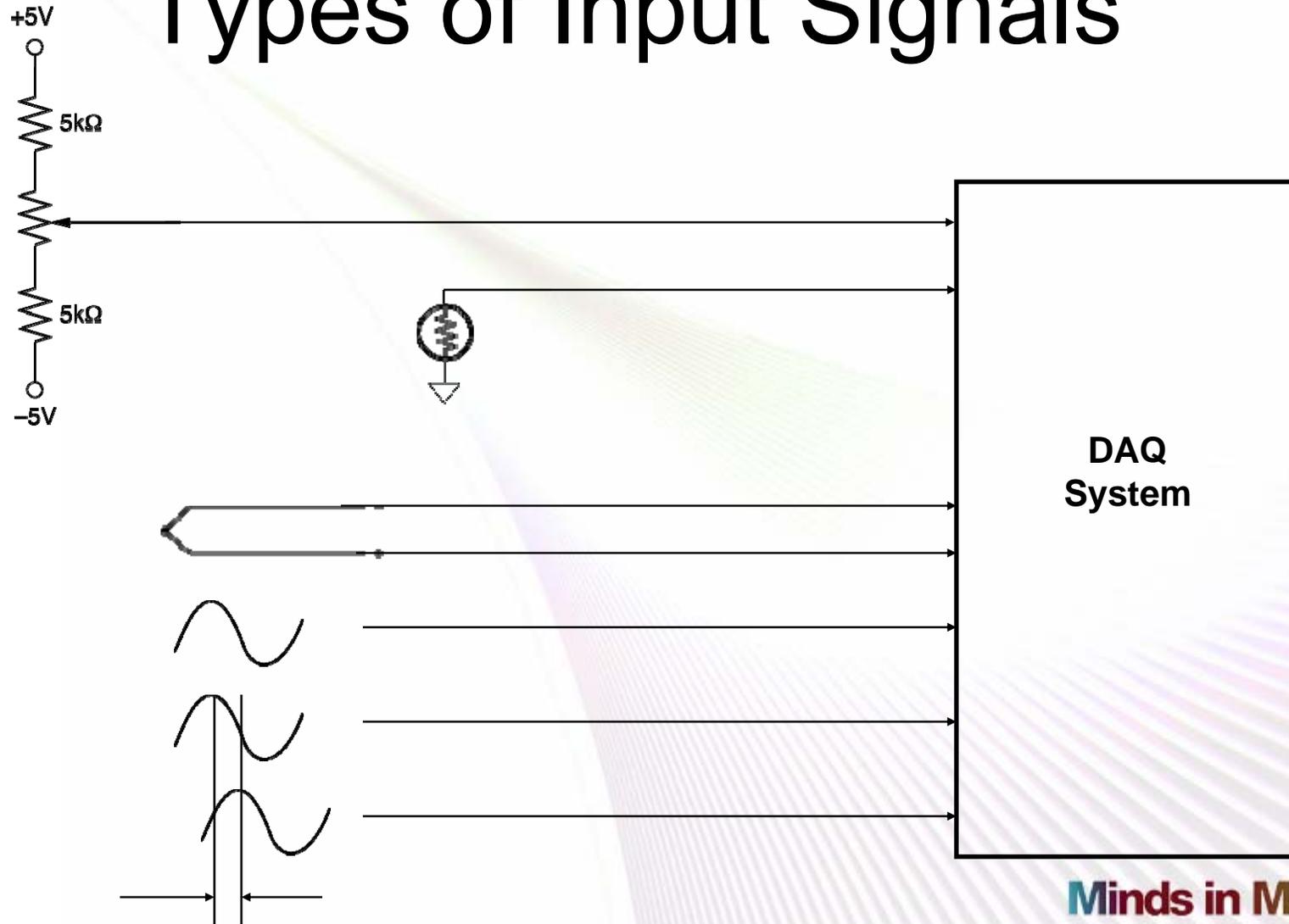
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Data Acquisition Components



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Types of Input Signals



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Simultaneous Sampling

- Why do simultaneous sampling?
- Applications for simultaneous sampling converters

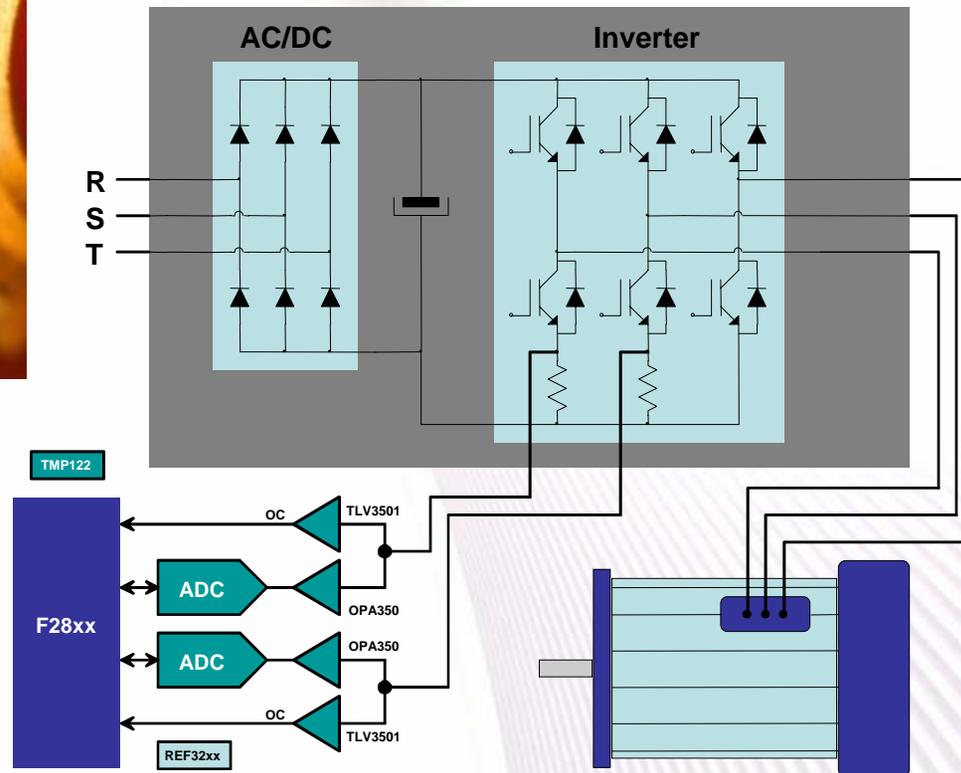
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Why Sample Simultaneously?

- Preservation of Phase Relationship
 - Speed/position sensors
 - Vibration analysis
- Relative Signals for Various Equations
 - Power
 - Torque, etc.

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Simultaneous Sampling Applications



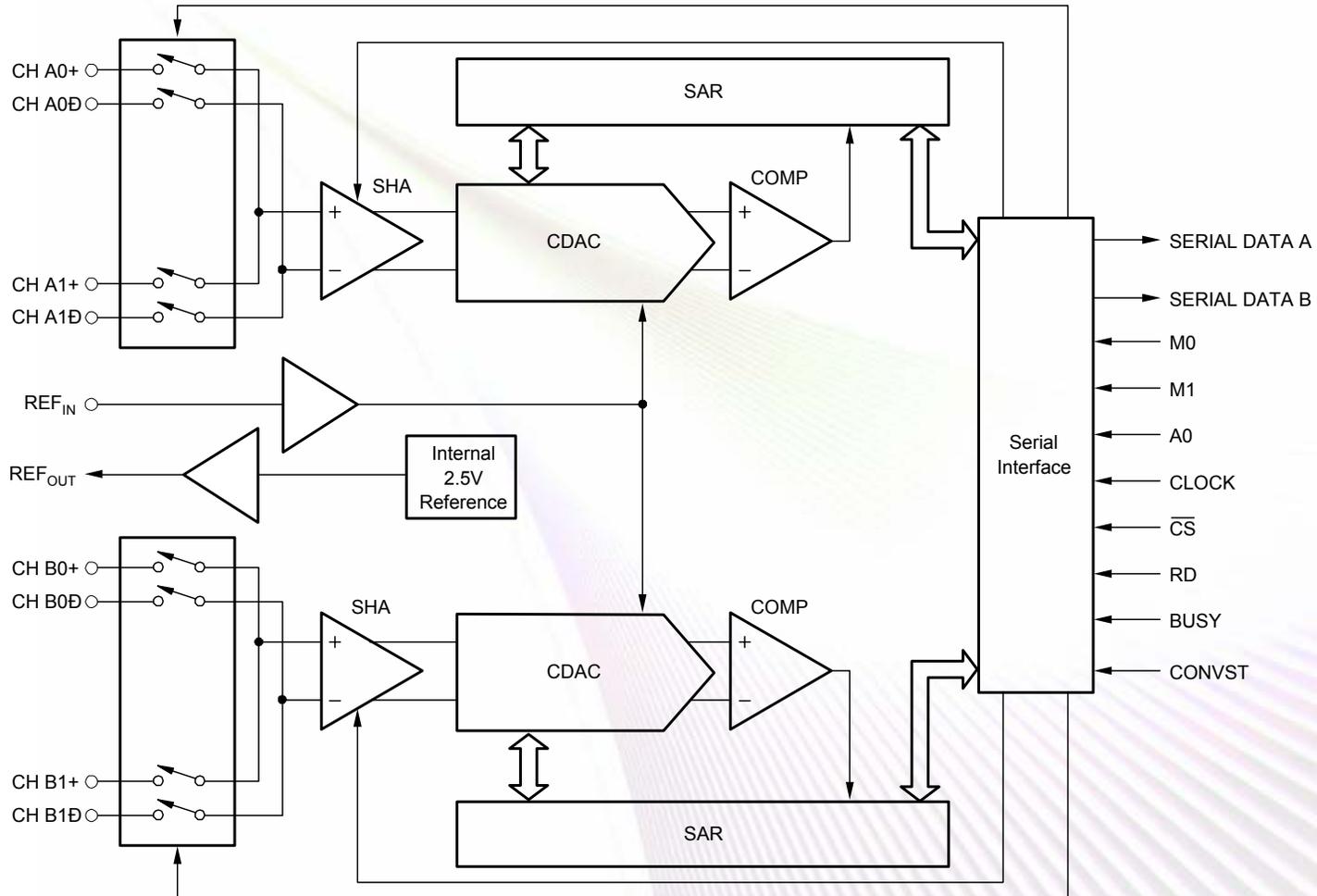
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Simultaneous Sampling ADCs

- 2x2 SAR: 2 ADCs with 2 S/H
 - ADS7861, ADS8361
 - ADS7862
- 3x2 SAR: 2 ADCs with 3 S/H
 - ADS7864
- 1x6 SAR: 6 ADCs with 1 S/H
 - ADS8364, ADS8365
- Pipeline
 - THS12xx

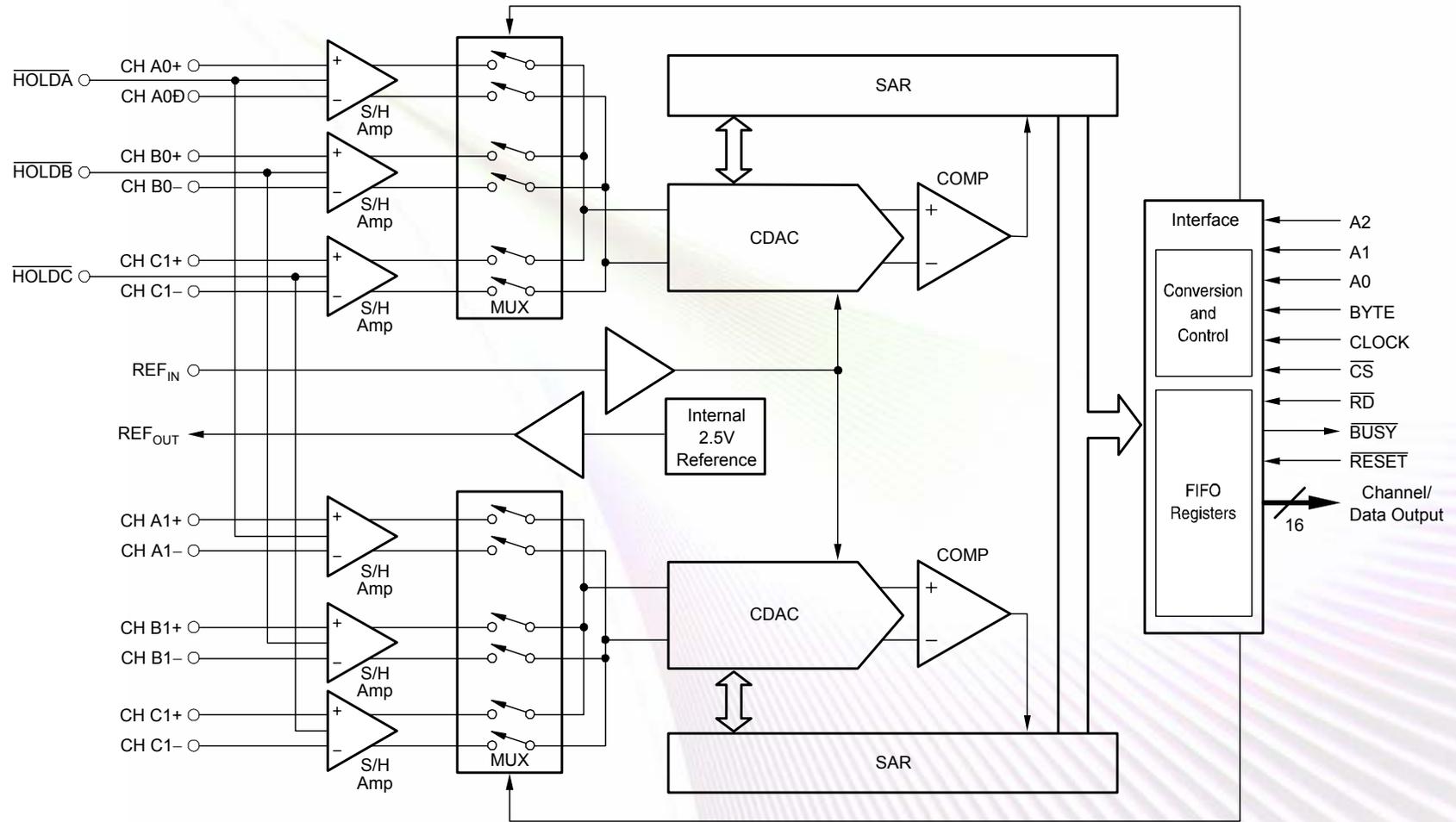
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2x2 SAR



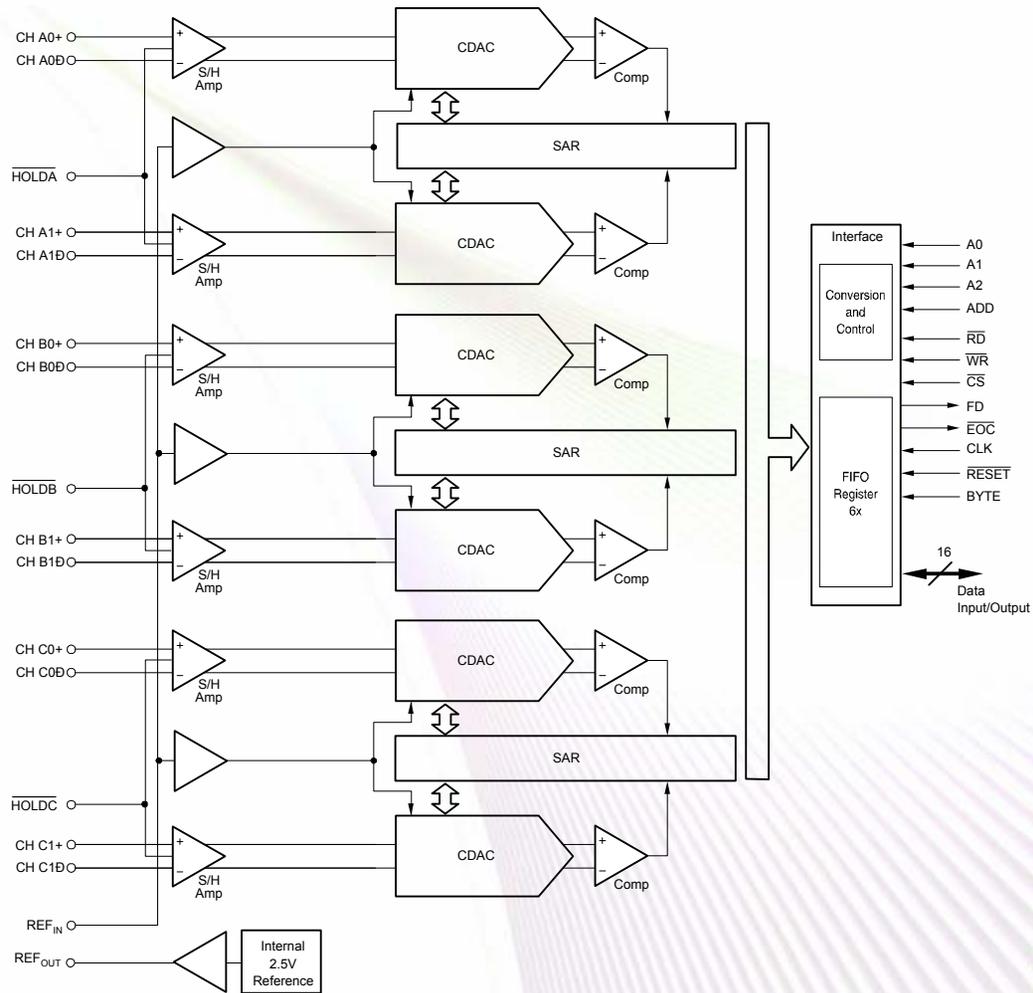
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3x2 SAR



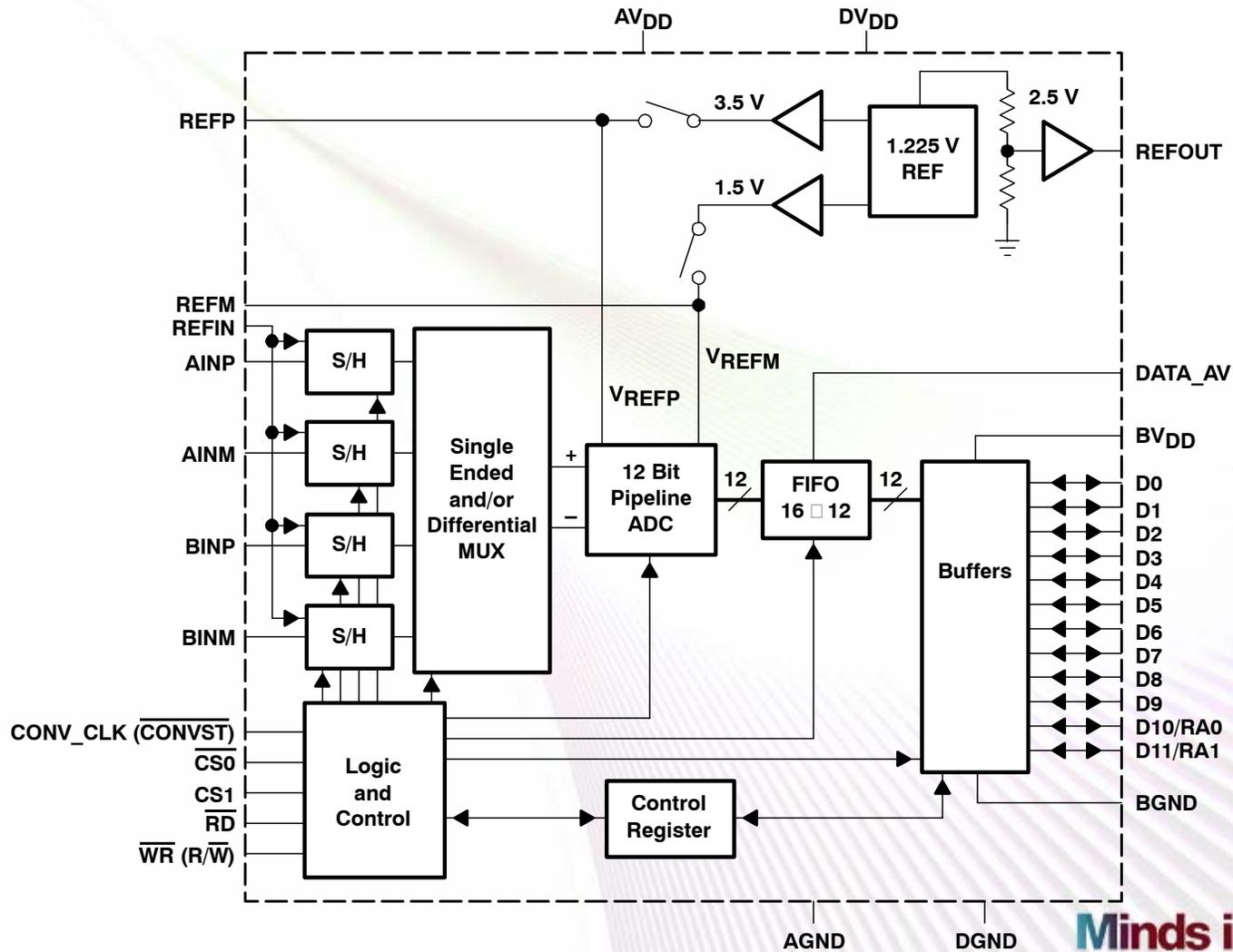
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1x6 SAR



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4X1 Pipeline



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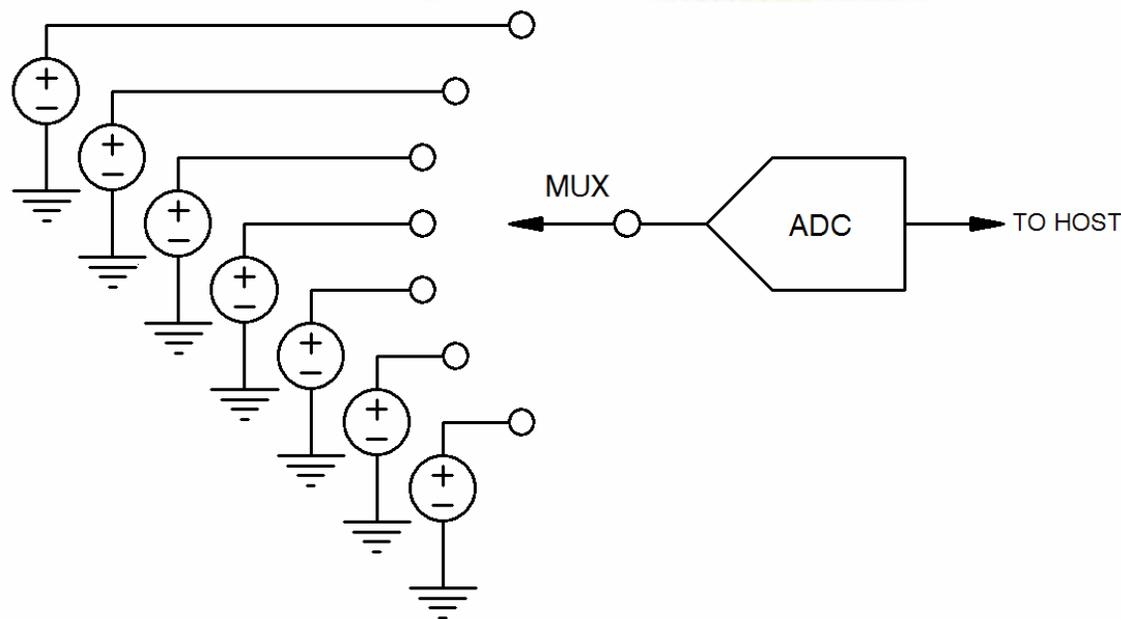
Multiplexed Sampling

- Are there advantages?
- What about using multiple ADCs?
- Applications for multiplexed inputs

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Multiplexed Advantages

- Fewer converters needed per channel – often only one needed.
- Often lower power
- Often lower cost



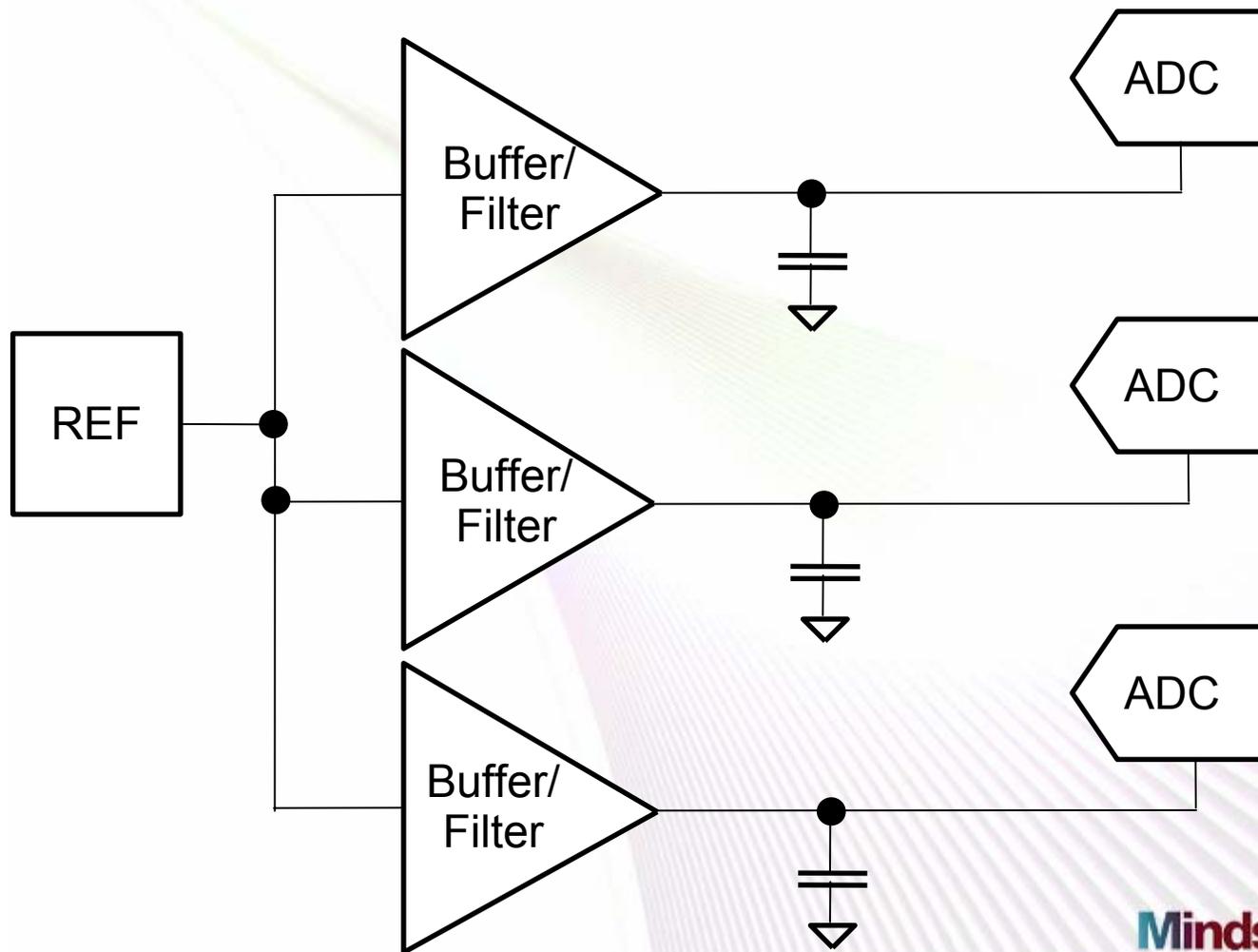
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Multiplexed vs. Multi-Chip

- Reference distribution issues
- Input signal issues: One channel has very different input needs from others.
 - Use a single channel device, MUX the others.
- Multiplexing implies time delays between channels – system issues? Simultaneous sampling better?
- Can the processor easily support multiple ADCs?

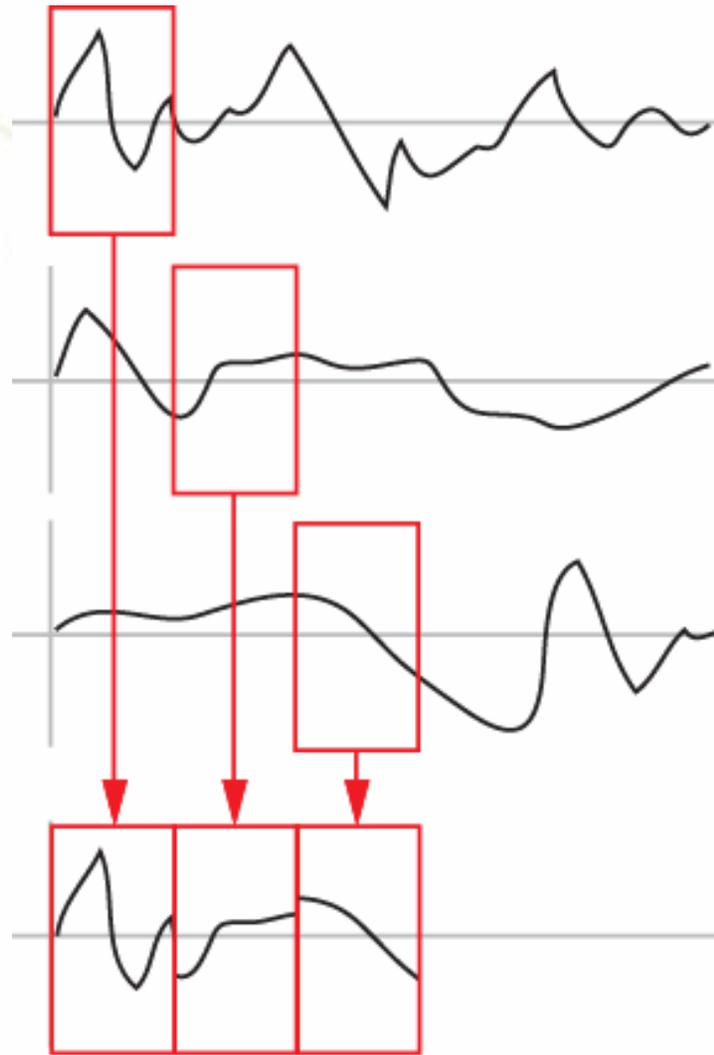
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Reference Distribution



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Multiplexing Basics



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Multiplexed Applications

- Temperature monitoring
- Multi-point weigh scales
- Pressure monitoring
- Control surface transducer readings (e.g., audio mixer control surface)
- General purpose low-speed sensor data acquisition

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Multiplexing Application Tips

- If converter is available with internal MUX, use it!
 - Available on both SAR and delta-sigma.
- Internal MUXes are tailored to match the ADC performance.
- Don't have to guess about switching and settling time.

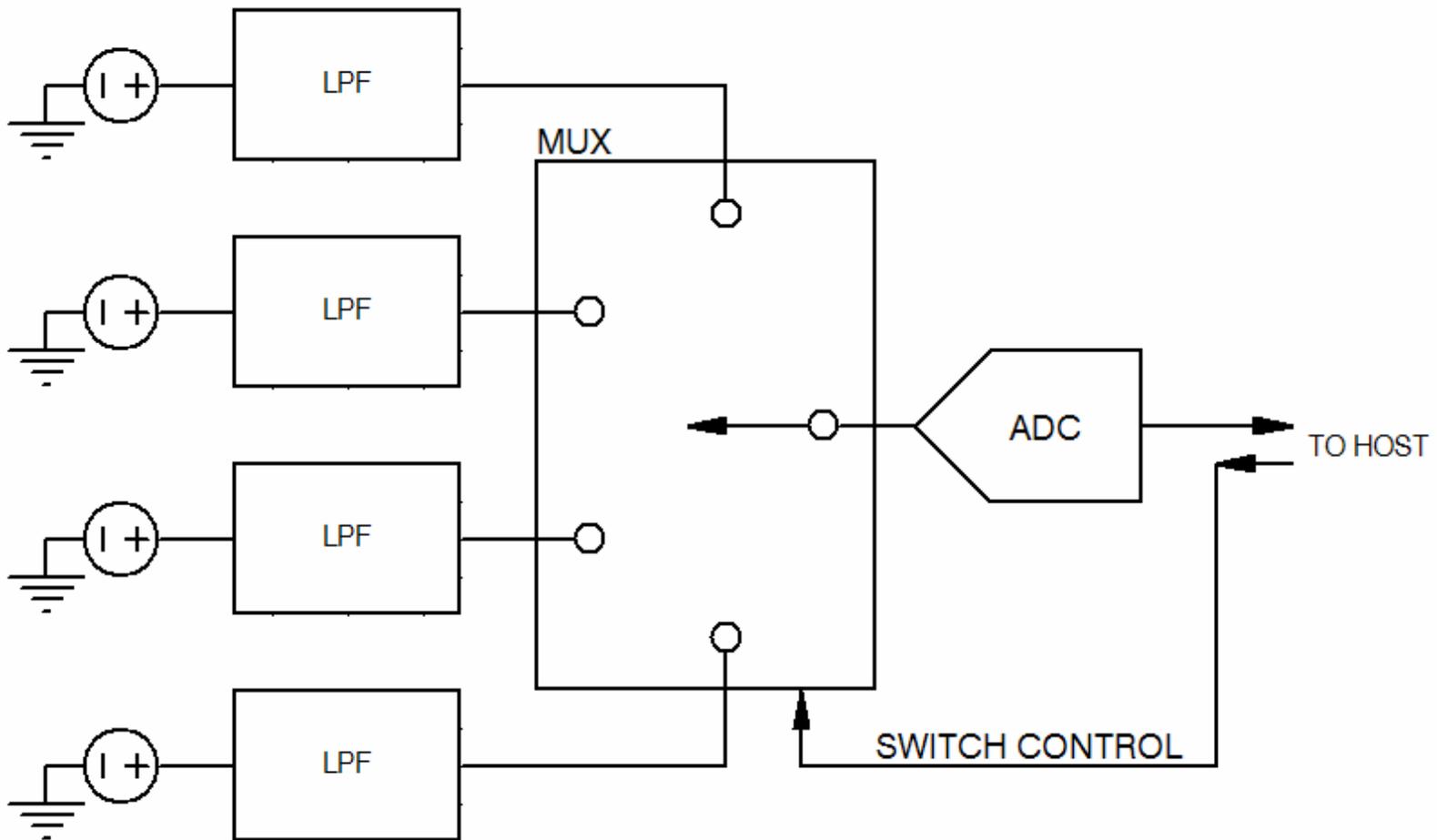
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Multiplexing SARs vs. $\Delta\Sigma$ s

- SAR-based systems often require low-pass filters on *each channel*, increasing power and cost.
- $\Delta\Sigma$ systems generally offer greater accuracy and lower noise floor.
- SAR designs with a post-multiplexer filter result in a system with settling-time issues almost identical to the $\Delta\Sigma$ design, but with higher noise.

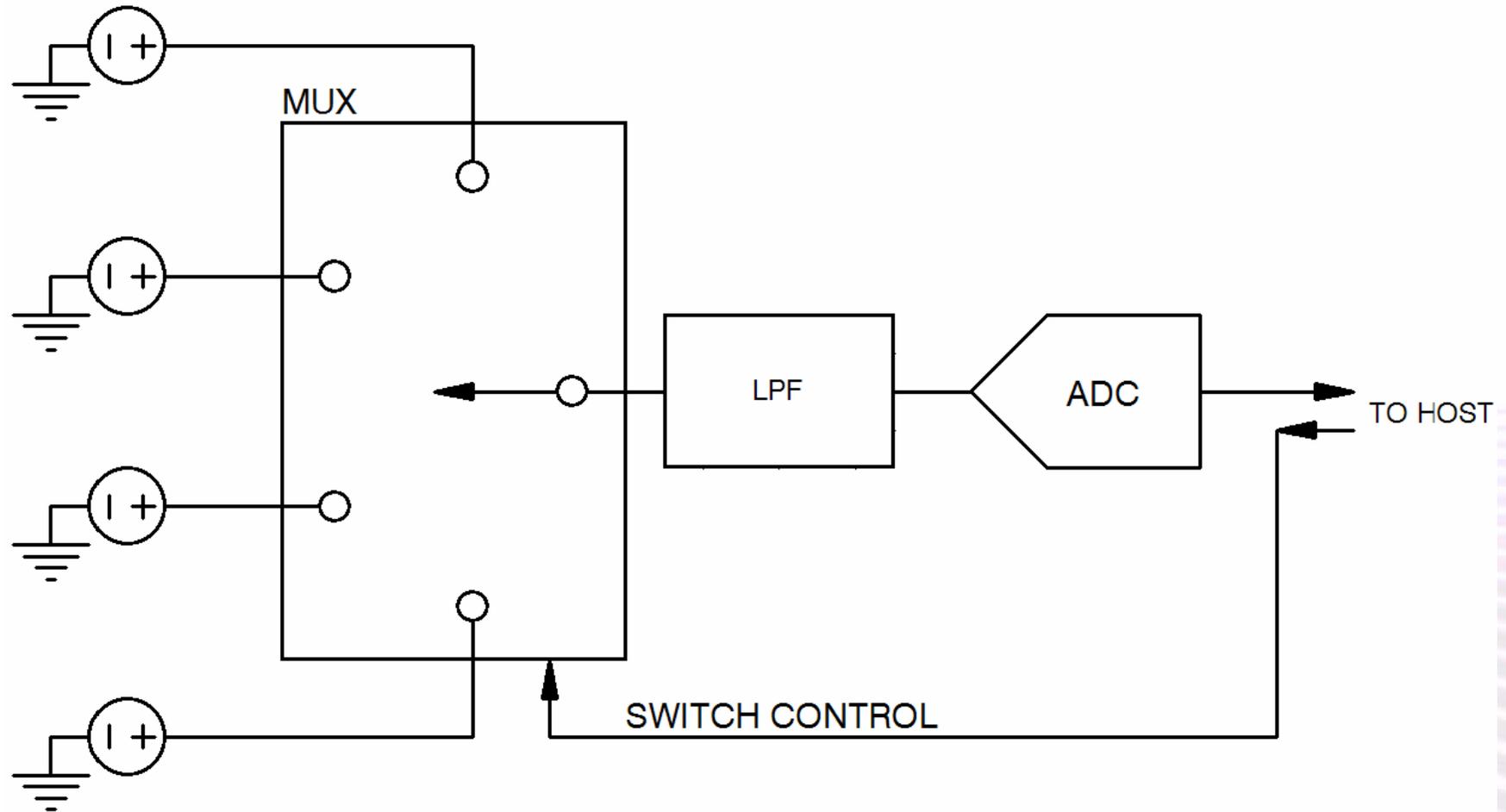
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Multiplexing with SARs



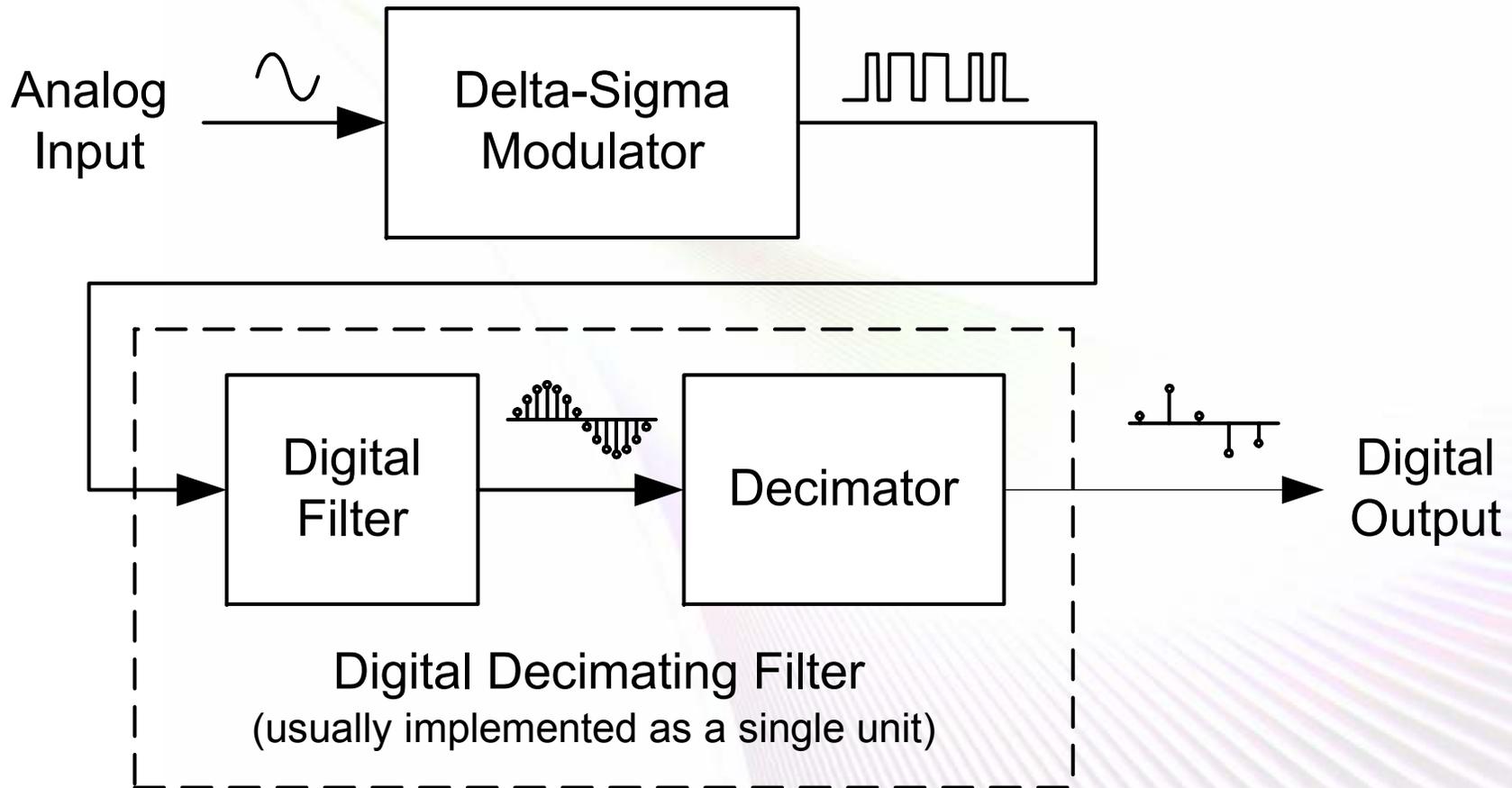
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Multiplexing with Delta-Sigmas



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The Delta-Sigma Architecture



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SAR vs. Delta-Sigma Conversions

- SAR conversions have Start Conversion Signal.
- Delta-Sigma is always sampling/converting.

SAR Converter

Start Conversion

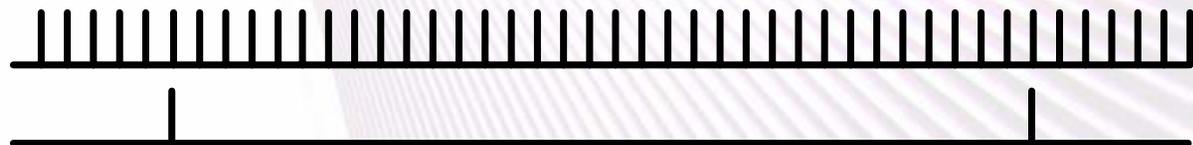
Conversion Done



Delta-Sigma Converter

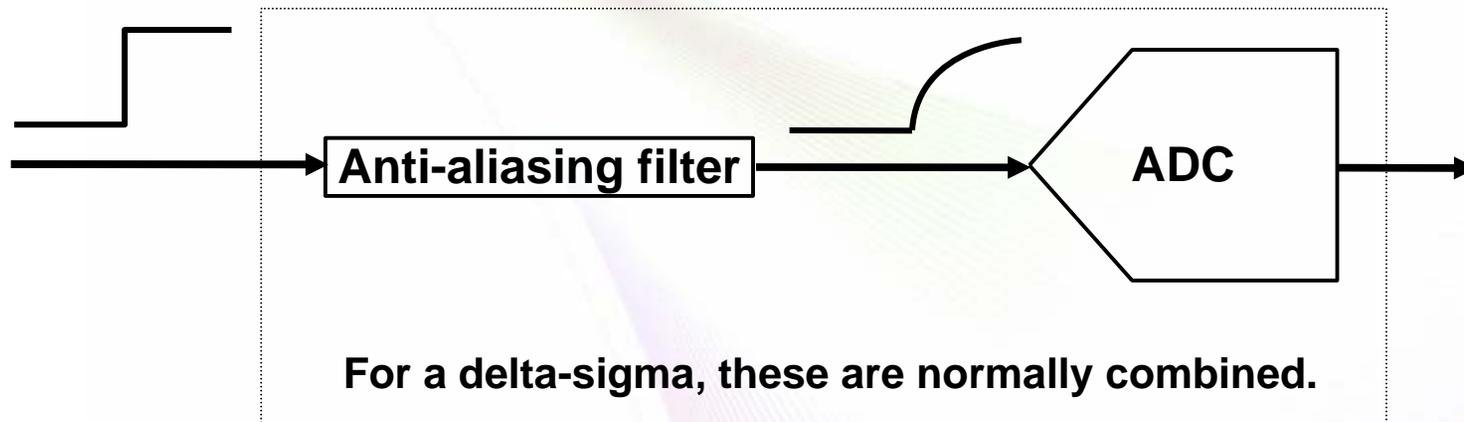
Input Sampling

Conversion Done



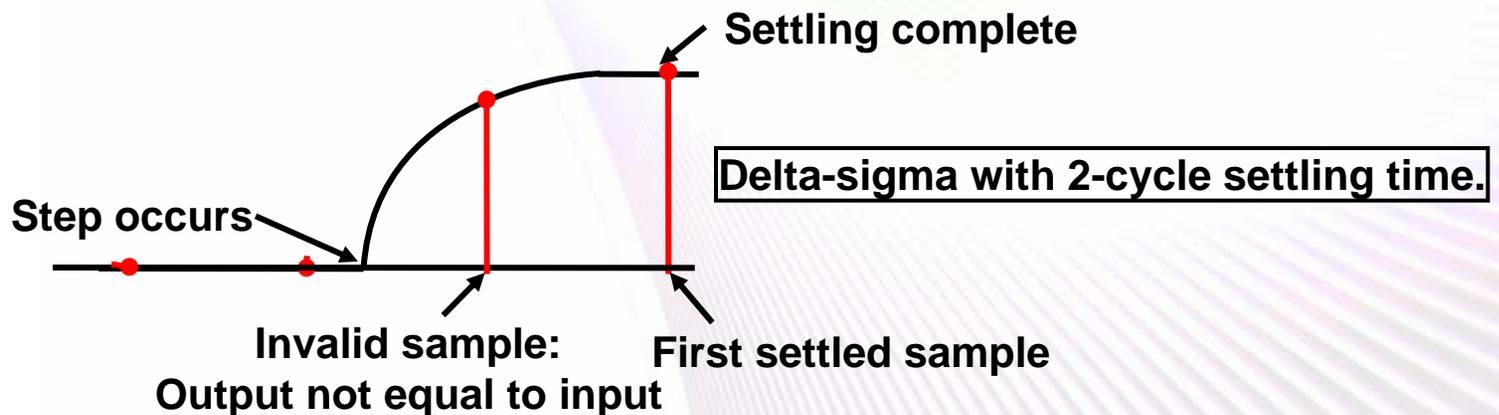
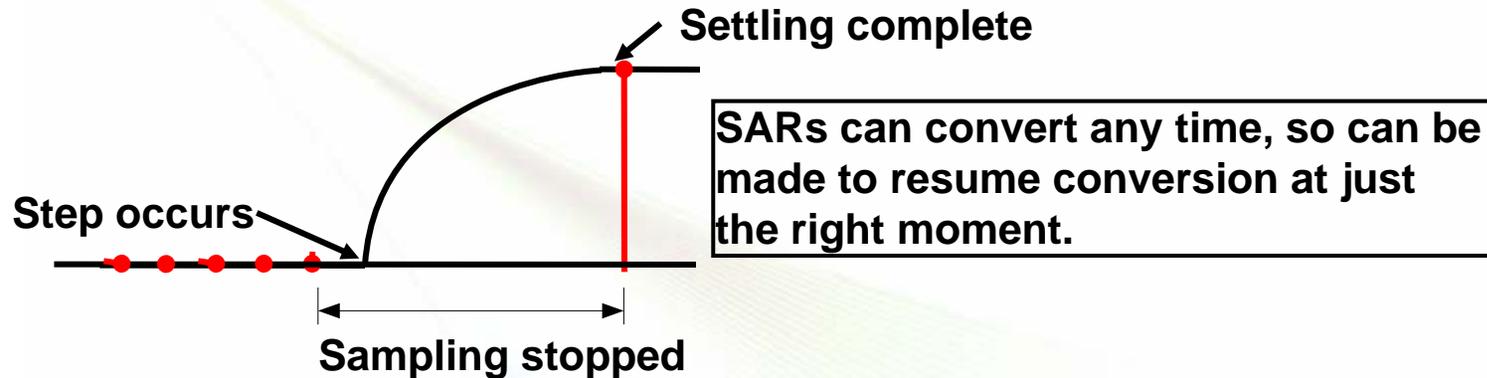
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Settling Time



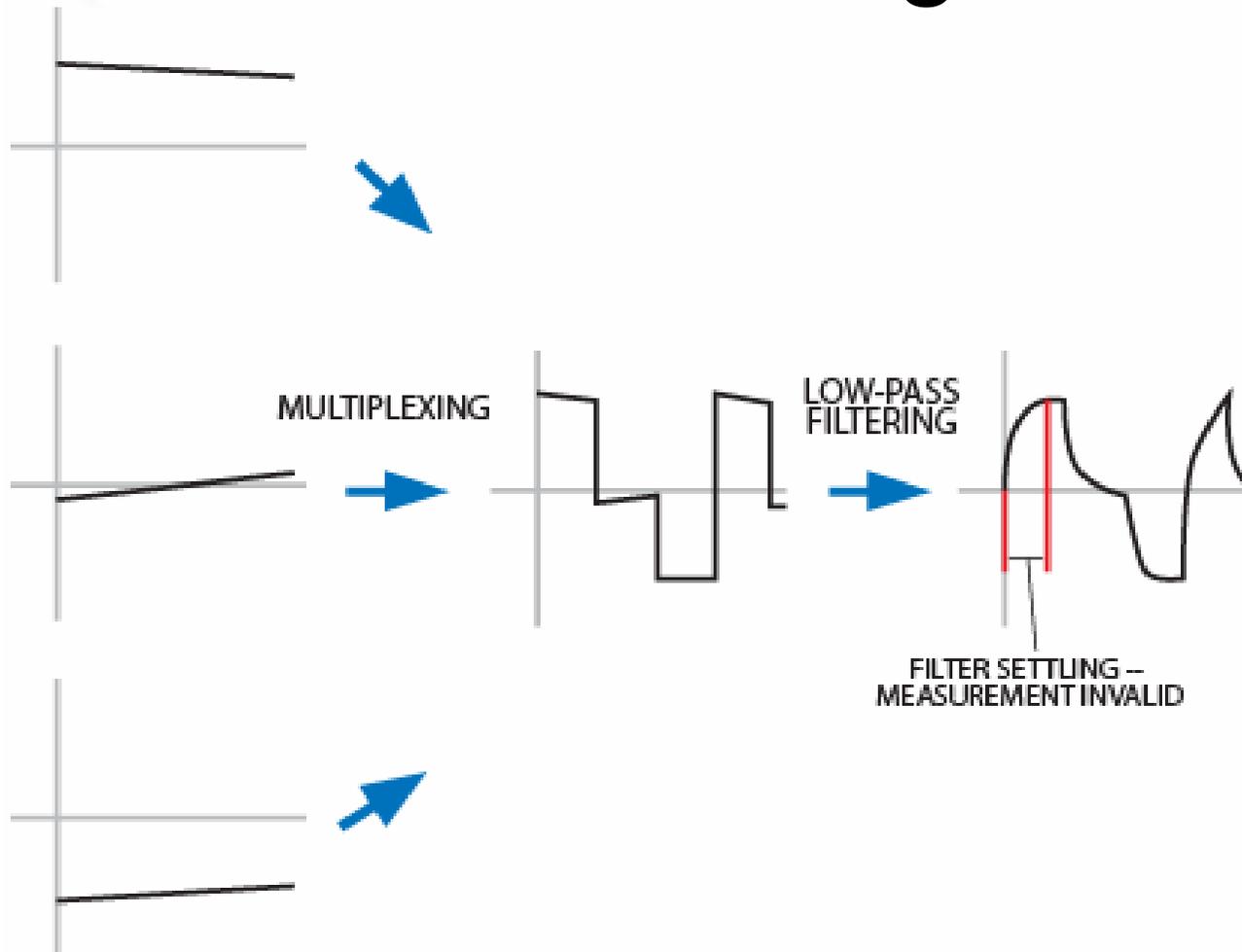
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Settling Time: SAR vs. $\Delta\Sigma$



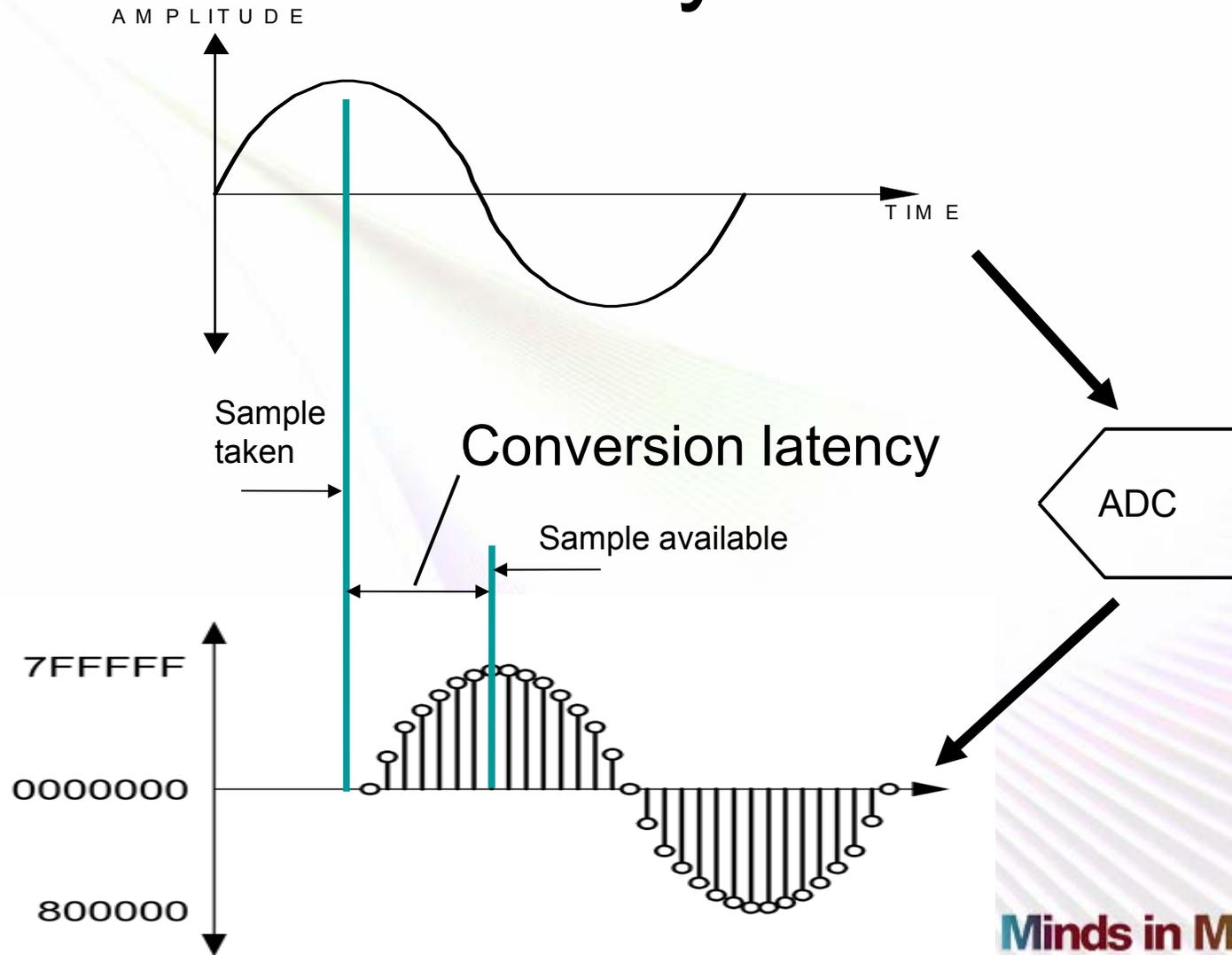
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Effects of Settling Time



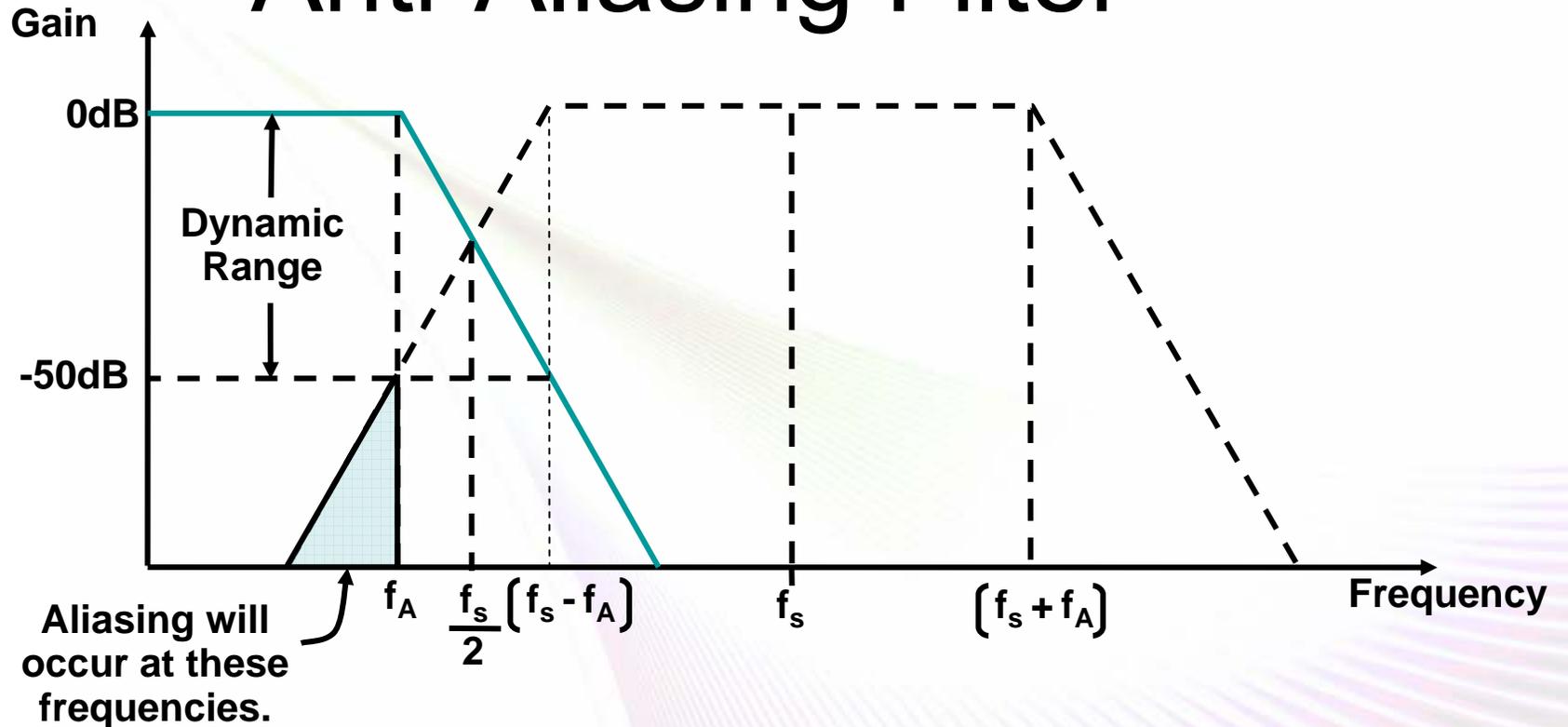
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Latency



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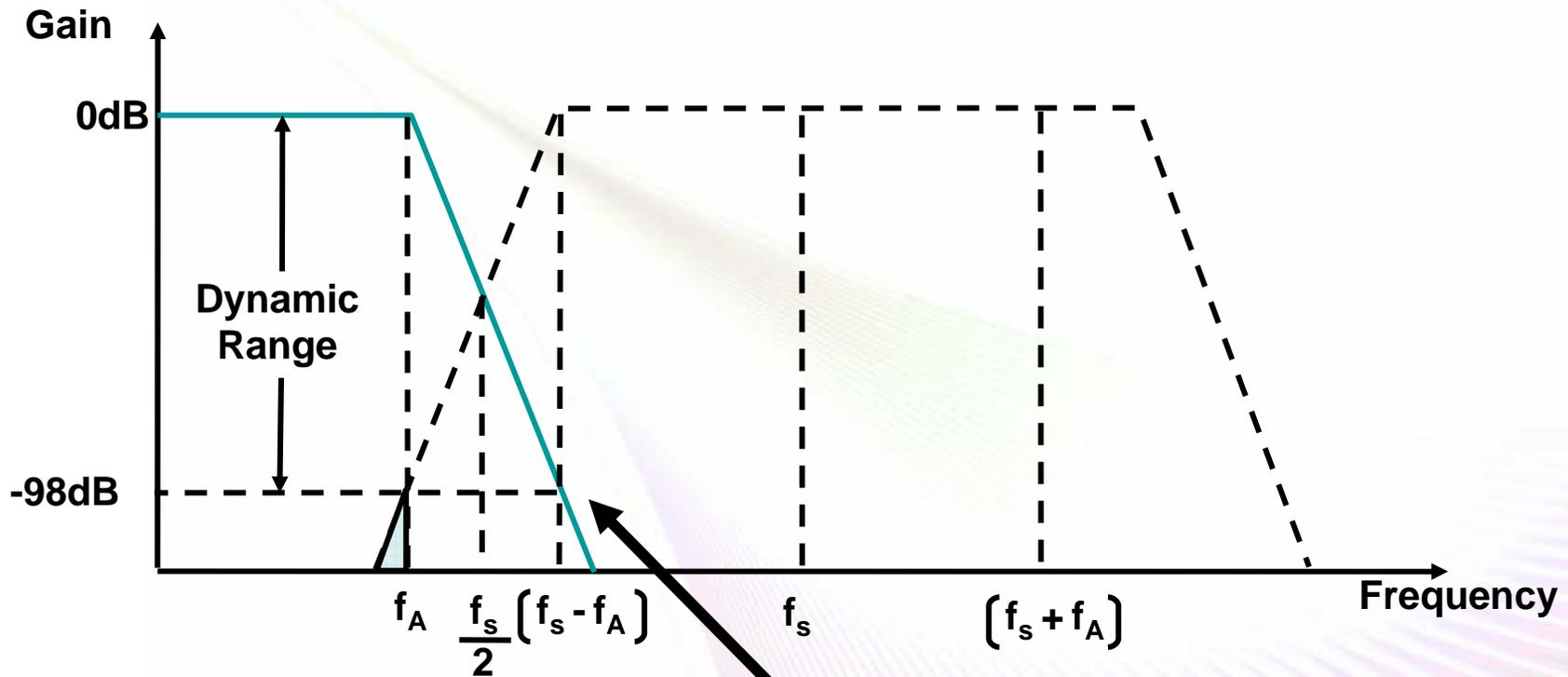
Anti-Aliasing Filter



50-dB, alias-free range is insufficient for 16 bits.

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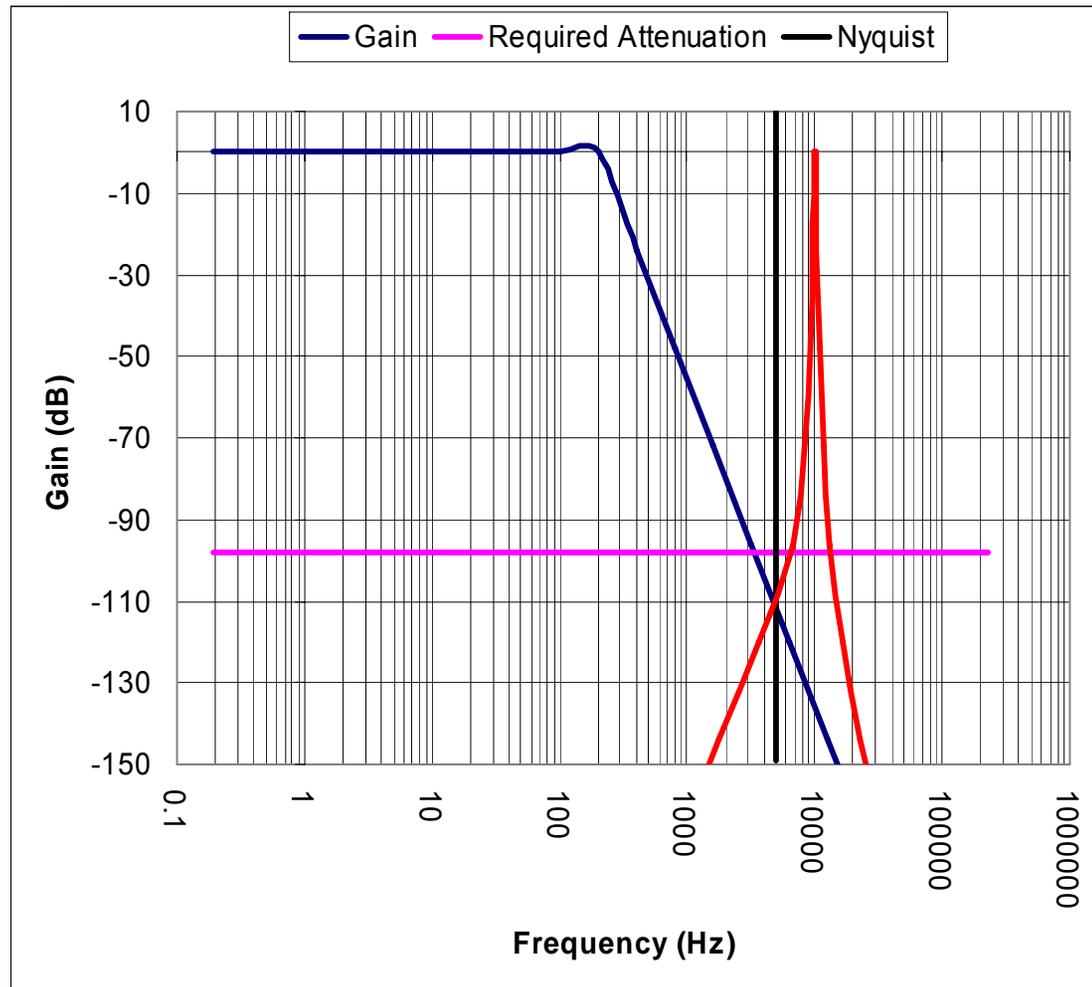
Improved Anti-Aliasing Filter



An 8-pole filter increases dynamic range to 98 dB!

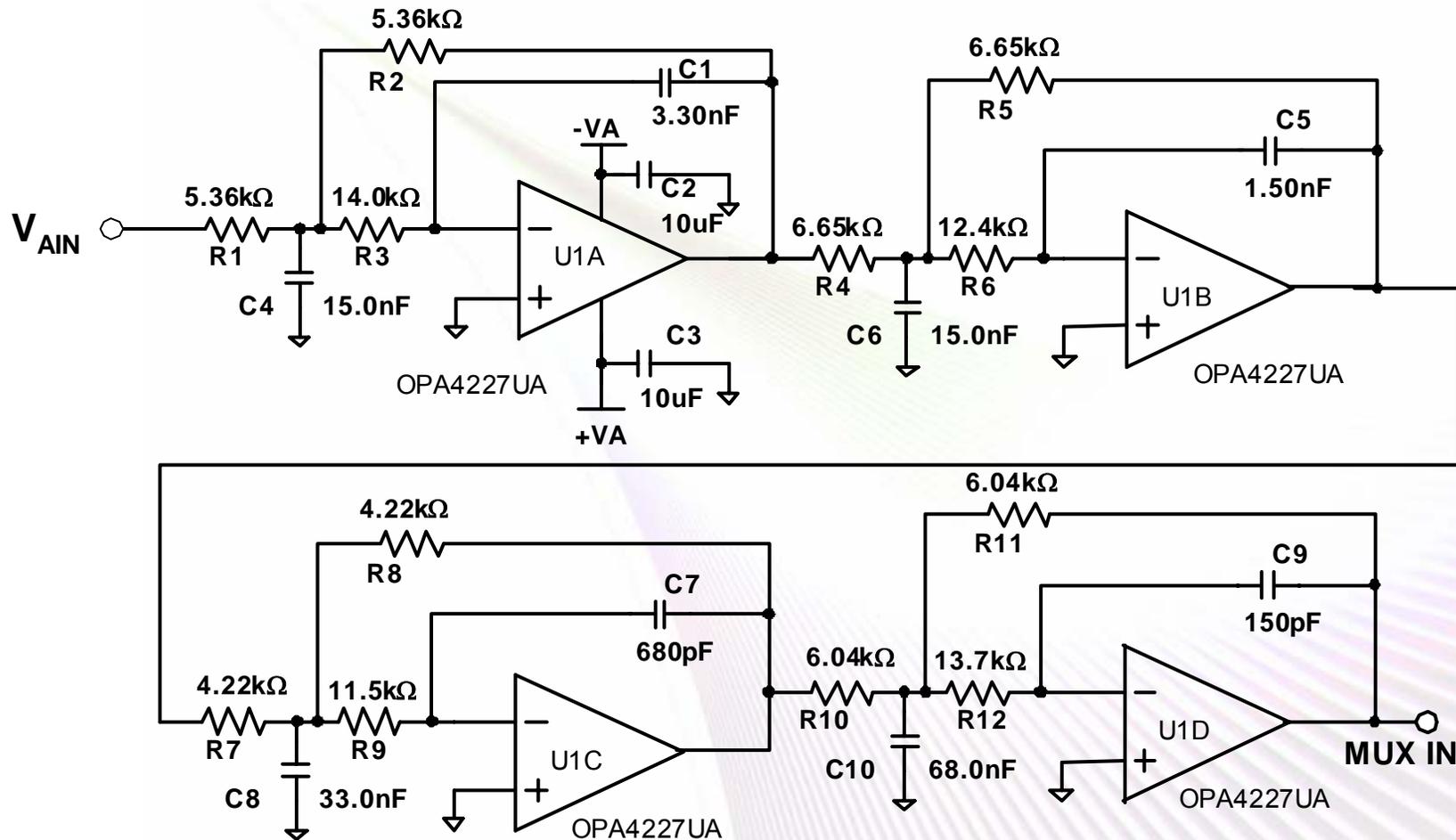
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Anti-Aliasing Filter Calculator



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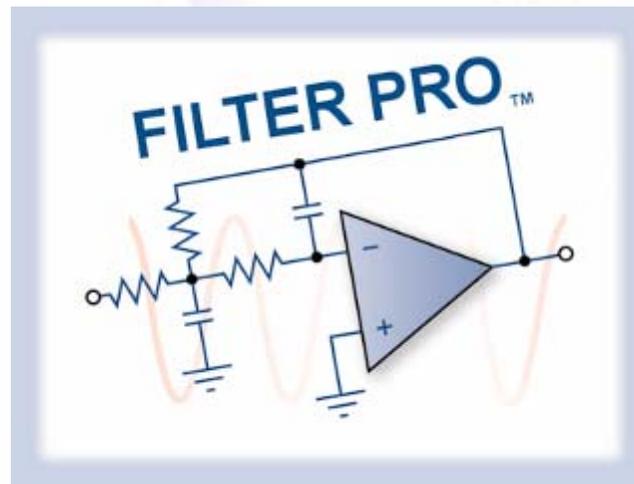
8th Order Filter Example



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Filter Pro™

- A filter synthesis tool for designing multi-section, low-pass and high-pass active filters.
- Supports 2nd to 10th order, multiple-feedback (MFB) and Sallen-Key filter topologies.



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The Filter Pro Worksheet

Topology info: The Multiple Feedback (MFB) or Infinite Gain topology places two feedback paths around an op amp. This filter type is less sensitive to component values than the Sallen-Key topology.

Response plot: A graph showing magnitude response (dB) vs. frequency (Hz) on a log scale. The plot shows a low-pass filter response with a cutoff frequency of 10.0 kHz. The magnitude is 0 dB at low frequencies and drops to -60 dB at 100 kHz. The phase response is shown as a red line, starting at 0° and reaching -630° at 100 kHz.

Filter topology: MFB

Pass band: Low Pass

No. of poles: 4

Ripple (Chebychev): 0.5 dB

Cutoff freq.: 10.0 k Hz

Response freq.: 10.0 k Hz

Gain: 4.993

Filter schematic with component values: The schematic shows two op-amp stages, Section A and Section B. Section A uses resistors R1 (3.48 k), R2 (6.98 k), R3 (10.2 k) and capacitor C1 (1.00 nF). Section B uses resistors R1 (3.48 k), R2 (8.66 k), R3 (12.4 k) and capacitor C1 (100 pF). Both sections also have a capacitor C2 (22.0 nF).

Section Info: G, f_n , "Q"

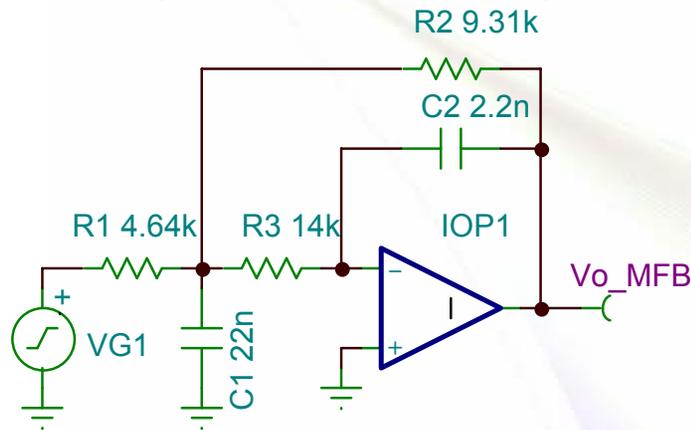
	Passband Gain (Vout/Vin)	Fn	Q	Response at 10.0 k Hz. Gain	Phase*	Req. GBP
A	2.006	5.9700	705.11 m	-1.56 dB	-127.4°	844
B	2.489	10.313	2.9406	15.59 dB	-78.3°	7.55
	4.993	Totals	2.0734	14.03 dB	-205.8°	

MFB, 4-Pole Low-Pass Chebychev, 0.5dB Ripple: 10.0 kHz Cutoff, Passband Gain of 4.993

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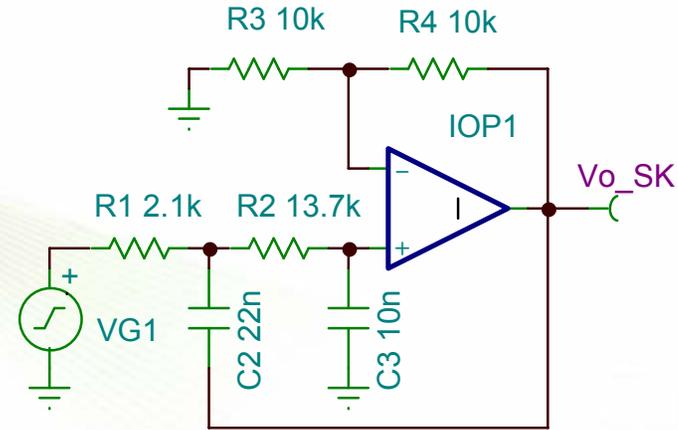
Filter Pro Topologies

Multiple Feedback (MFB)



- Inverted stage
- No common-mode error
- Low component sensitivity

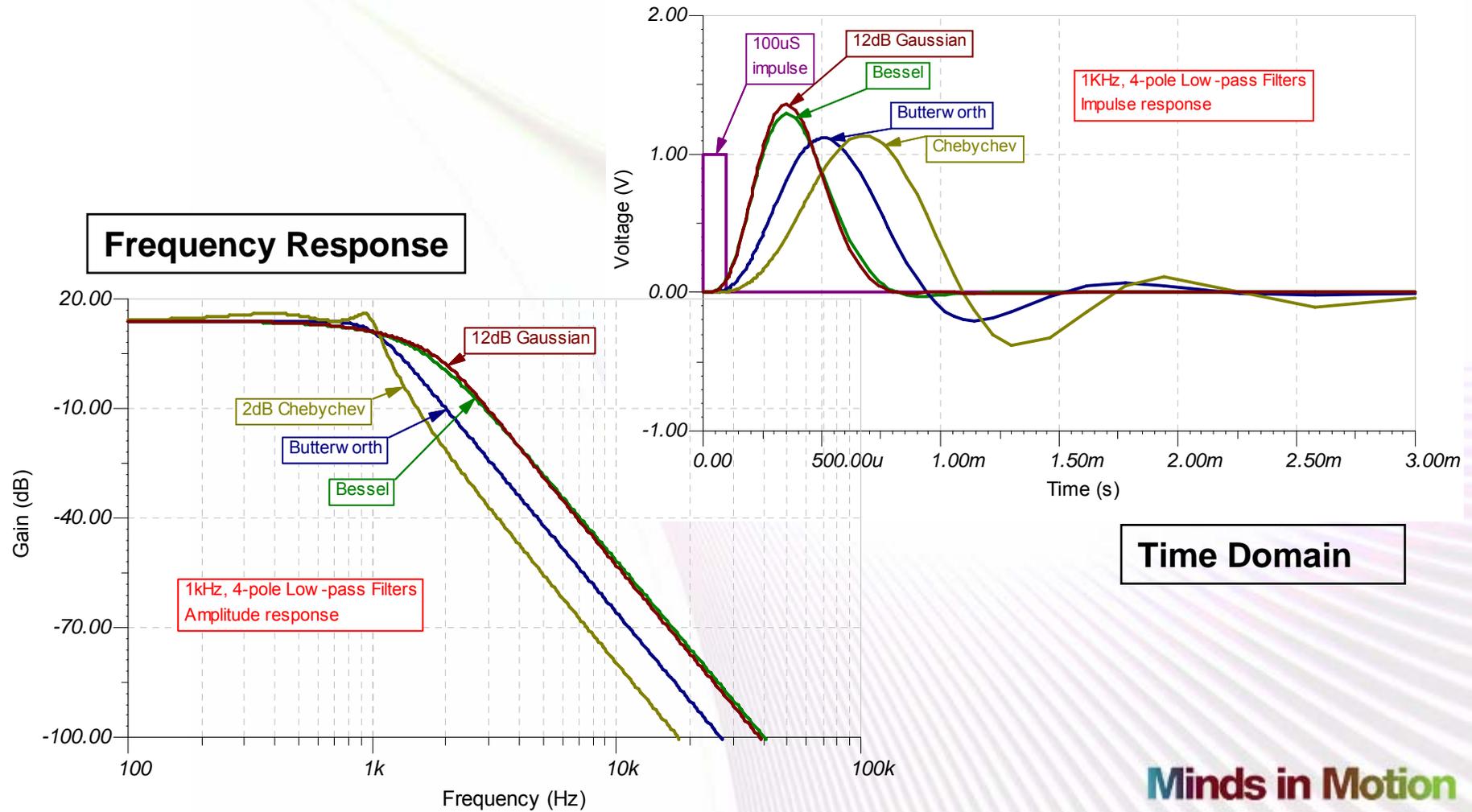
Sallen-Key



- Non-inverting stage
- Common-mode input C adds to C3
- $G = +1V/V$, reduces to follower
- High "Q" sensitivity

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Response Characteristics



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TINA-TI™

DesignSoft
Excellence in Computer Aided Design and Educational Software

ANALOG SIMULATION

HOME
TINA PRO
Circuit Entry
Components
Analyses
Analog
Digital
Mixed
Symbolic
Spice
Noise
Tolerance
S-parameters
Network Analysis
Spectrum Analysis
Optimization
Presentation
Virtual Instruments
Real Time T&M
Education
Documentation
TINALab
EDISON
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Demos | Registration | Order | Distributors | Events | FAQ | Technical support | Company

English

English...Deutsch...Español...

Analog Simulation

Amplifiers and Linear : Tina - Analog Circuit Simulation From Texas Instruments

TINA™

TINA is an easy-to-use, but powerful, circuit simulation program based on a SPICE engine. TINA-TI™ is a fully functional version of TINA, loaded with a library of TI macromodels plus passive and active models. TINA-TI is limited to circuits with two ICs and up to 20 additional nodes.

Screenshot of TINA™ output graphics

AC Analysis with parameter stepping

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Analysis

The image shows a screenshot of a schematic editor window titled "Wein_bridge_02 - Schematic Editor". The window displays a circuit diagram of a Wein bridge oscillator. The circuit includes an operational amplifier (U1 OPA743), two diodes (D1, D2 1N4148), and several resistors (R1 4.7k, R2 18.7k, R3 3.4k, R4 3.4k, R5 15k) and capacitors (C1 47n, C2 47n). The output of the oscillator is labeled "Vosc".

The "Analysis" menu is open, showing options for DC Analysis, AC Analysis, Transient..., Fourier Analysis, Noise Analysis..., and Options... The "Transient..." option is highlighted, and a red arrow points from it to the "Transient Analysis" dialog box.

The "Transient Analysis" dialog box is open, showing the following settings:

- Start display: 0 [s]
- End display: 10m [s]
- Calculate operating point:
- Use initial conditions:
- Zero initial values:
- Draw excitation:
- Integration method: Euler Gear

The dialog box has buttons for OK, Cancel, and Help. A red arrow points from the dialog box to the transient analysis plot.

The plot shows the output voltage "Vosc(V)" versus "Time (s)". The plot displays a sinusoidal waveform that starts at 0V, reaches a peak of approximately 4.5V, and then settles into a steady-state oscillation with a peak-to-peak amplitude of approximately 4.5V. The x-axis ranges from 0 to 10m seconds, and the y-axis ranges from -5.00 to 5.00 V.

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Test and Measurement

The image shows a screenshot of a schematic editor window titled "Wein_bridge_02 - Schematic Editor". The window displays a circuit diagram of a Wein-bridge Oscillator. A red box highlights the text "Wein-bridge Oscillator 1kHz". The circuit includes an OPA743 operational amplifier (U1) configured as a voltage follower. The feedback network consists of two resistors (R1, R2) and two capacitors (C1, C2) in a bridge configuration. The input network consists of two resistors (R3, R4) and two capacitors (C2, C1) in a bridge configuration. The output of the oscillator is labeled "Vosc".

Below the schematic, a virtual oscilloscope window titled "Oscilloscope - Virtual" is shown. The oscilloscope displays a sine wave on a grid. The vertical axis is labeled "Vosc: 2V" and the horizontal axis is labeled "Time/Div". The oscilloscope settings are as follows:

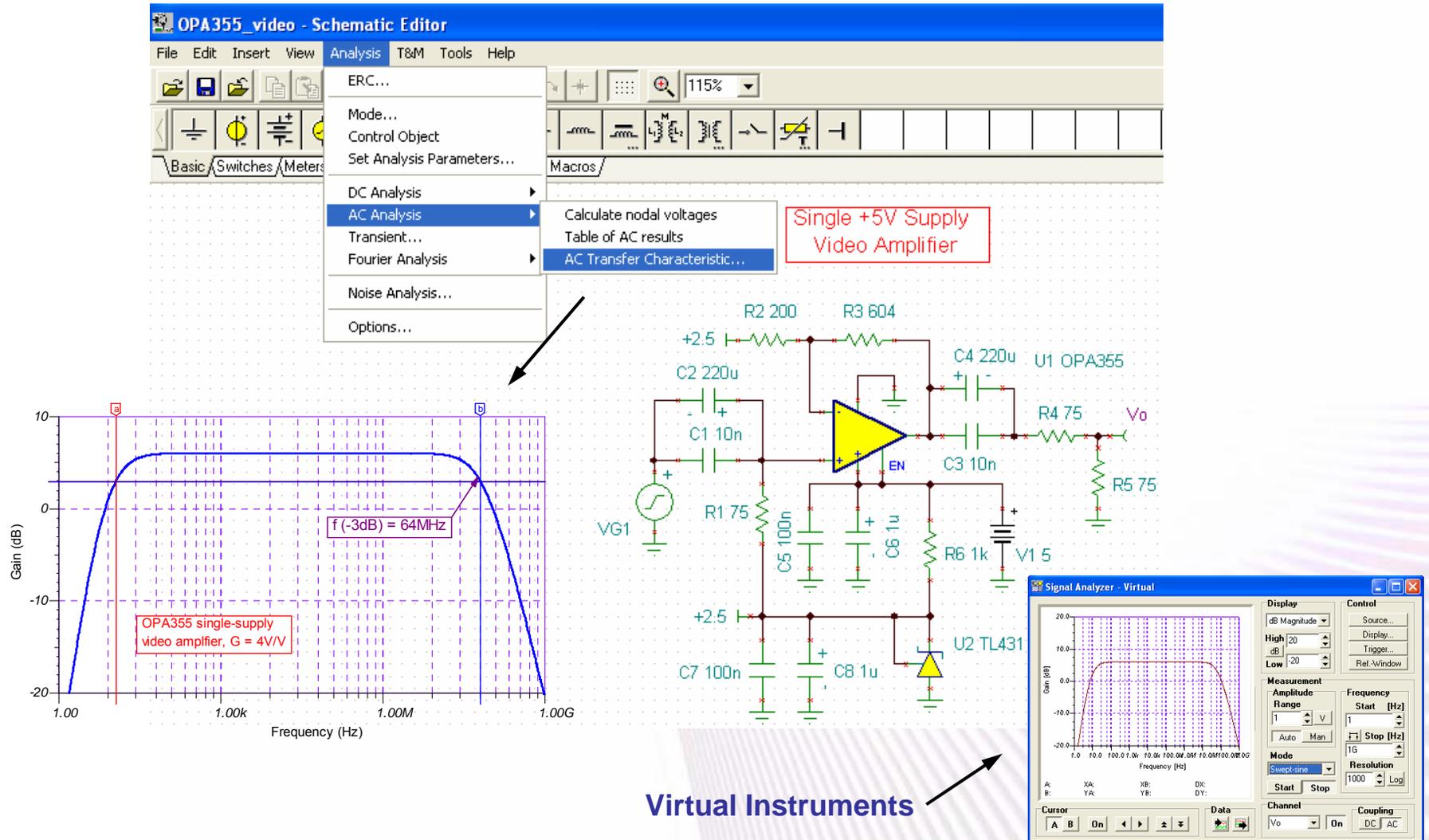
- Trigger: Mode: Auto, Source: Vosc, Level: 0
- Horizontal: Time/Div: 200u, Position: 0, Mode: Y/T, X Source: Vosc
- Channel: Vosc, Coupling: AC
- Vertical: Volts/Div: 2, Position: 0

The oscilloscope also shows a cursor and data table:

Cursor	A	B	On	Delta	Delta	Delta	Delta
XA:							
YA:							
XB:	498.19u						
YB:	34.86m						
DX:							
DY:							

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TI Example Circuits



For More on TINA-TI

Where to get TINA-TI:

<http://focus.ti.com/docs/toolsw/folders/print/tina-ti.html>

Or simply go to www.ti.com
and enter **TINA-TI** in the **keyword search** box.

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In Summary

- Know Your Input Signals
 - Speed, bandwidth
 - Time relationships to each other
- Choose the Right Scheme
 - Multiplexing
 - Simultaneous sampling
 - Multiple ADCs
- Choose the Right Converter Architecture
 - SAR
 - Input buffering and drive requirements
 - Delta Sigma
 - Settling time of filters

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Design Considerations

- ADC Input Drive
- Cost and Performance Tradeoffs
 - Improving embedded ADC performance.
- PCB Layout Considerations
 - Analog and auto routers don't mix.

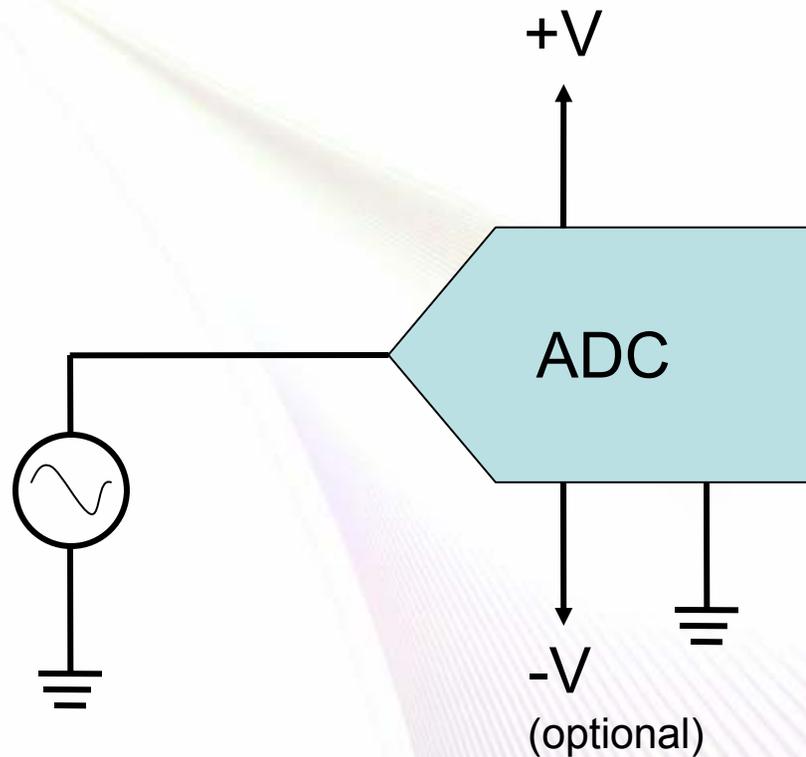
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Input Drive Circuitry

- Input Types
 - Single ended
 - Pseudo-differential
 - Differential
- Buffer Op Amp
 - Rail-to-rail considerations
- RC Circuit
 - Establishing starting criteria
- Modeling the ADC Input Interface

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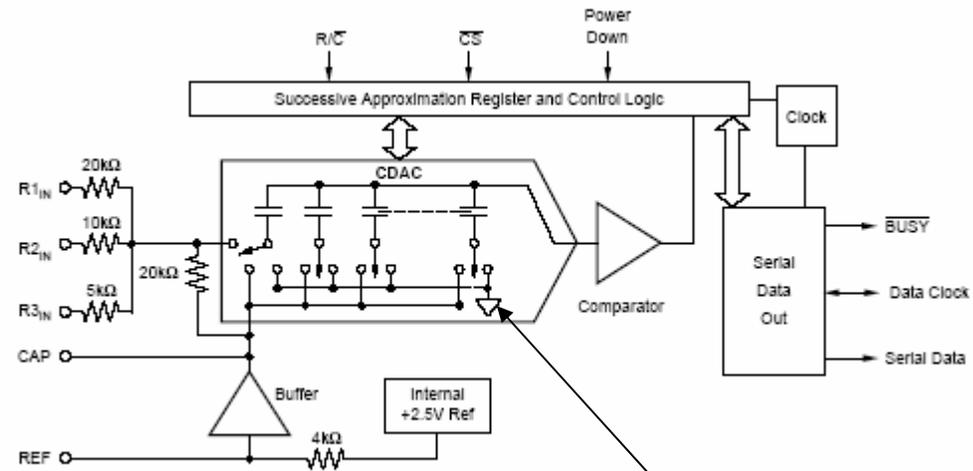
Single-Ended Inputs



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Single-Ended Inputs

- A single input pin for a particular channel.
- Input signals are referred to ground.



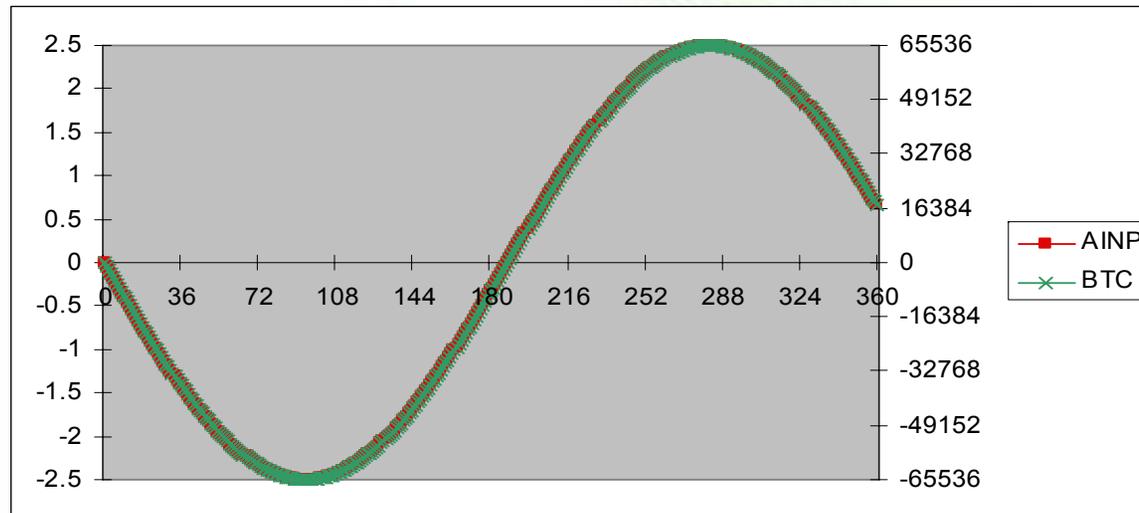
Note: input is referred to GROUND!

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True Bipolar Inputs

Devices with up to +/-10VDC inputs

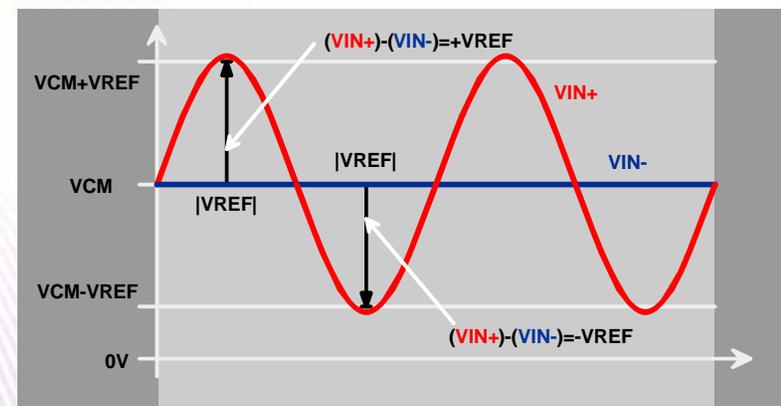
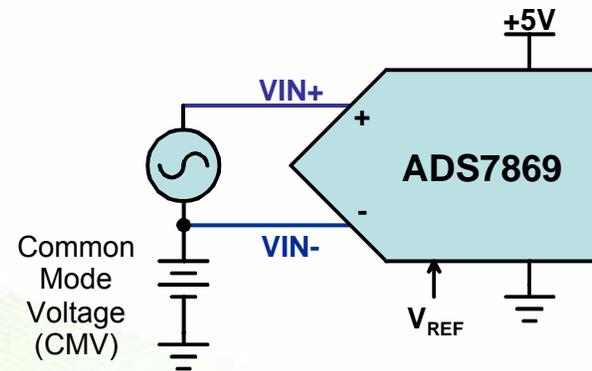
- ADS78xx and ADS85xx Series
- TLC257x and TLC357x Series



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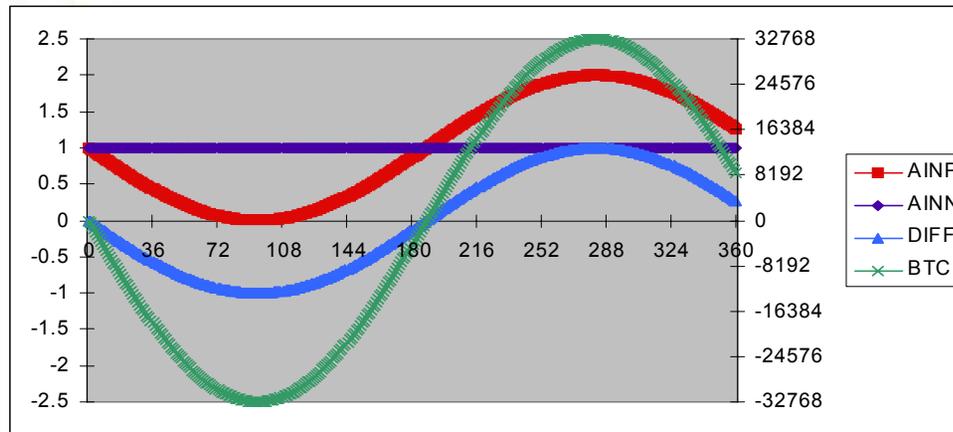
Pseudo Differential Inputs

- More like a single-ended input than differential, but with some advantages.
- IN- pin can move, but in a limited range.
- Good for removing common-mode voltages, offsets, etc.
- Provides a “clean” signal reference point.



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Pseudo Differential Example



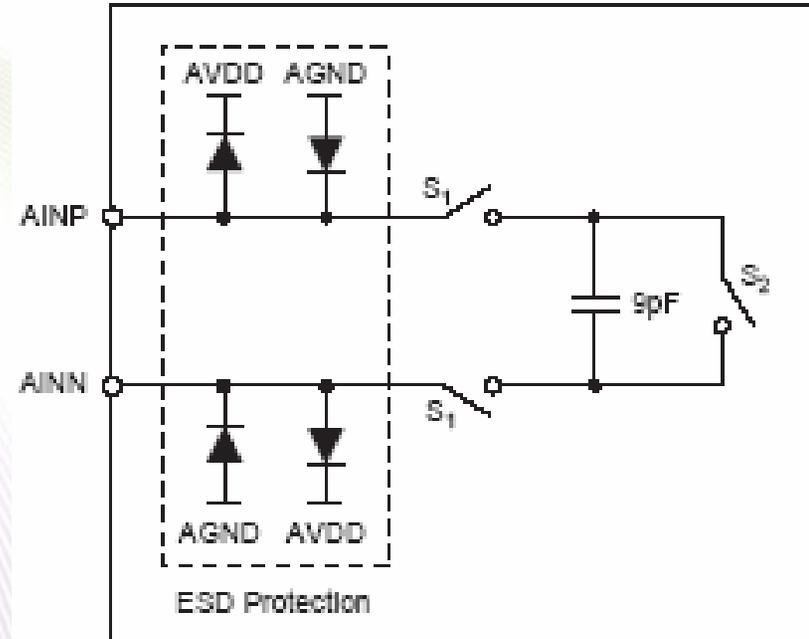
- Example: 0V to 2V sine wave; 1V common mode voltage.
- ADC sees the difference between the two inputs (“differential”).
- Output code may be unipolar or bipolar (bipolar shown here).

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Differential Inputs

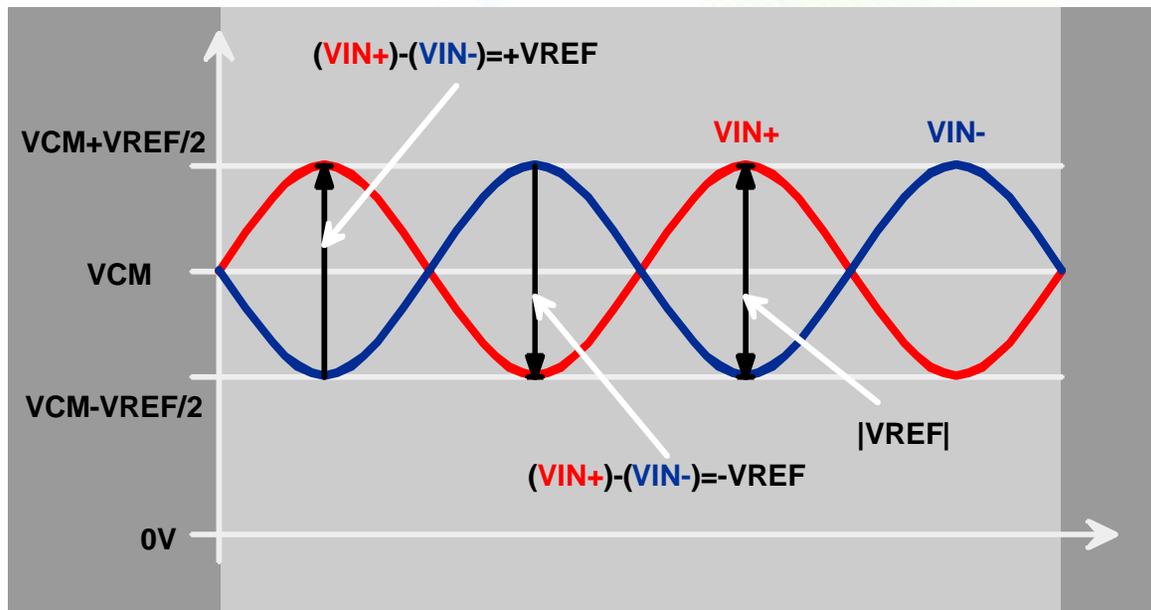
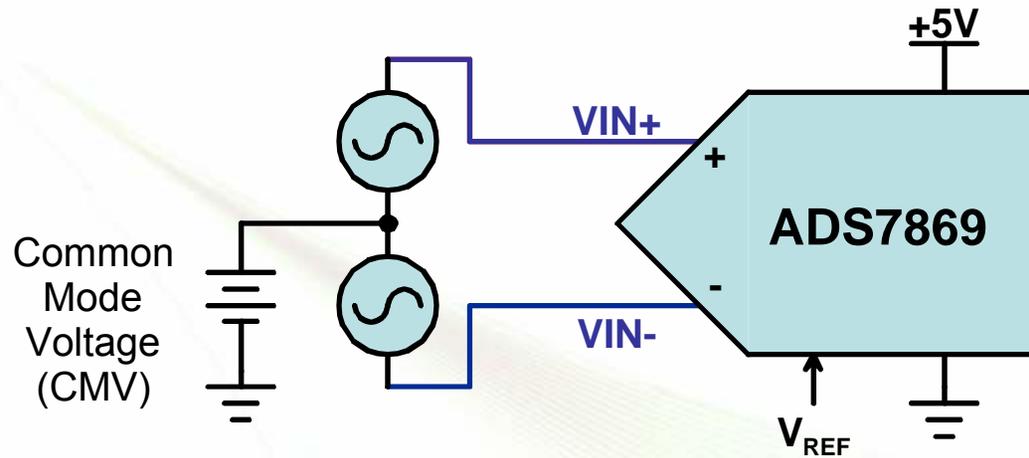
- Sometimes called “fully differential.”
- ADC sees AINP-AINN as input.
- Both inputs can swing from 0V to the full scale – but NOT below ground (in most cases).
- Typically move in a “balanced” fashion.

Example: ADS1271



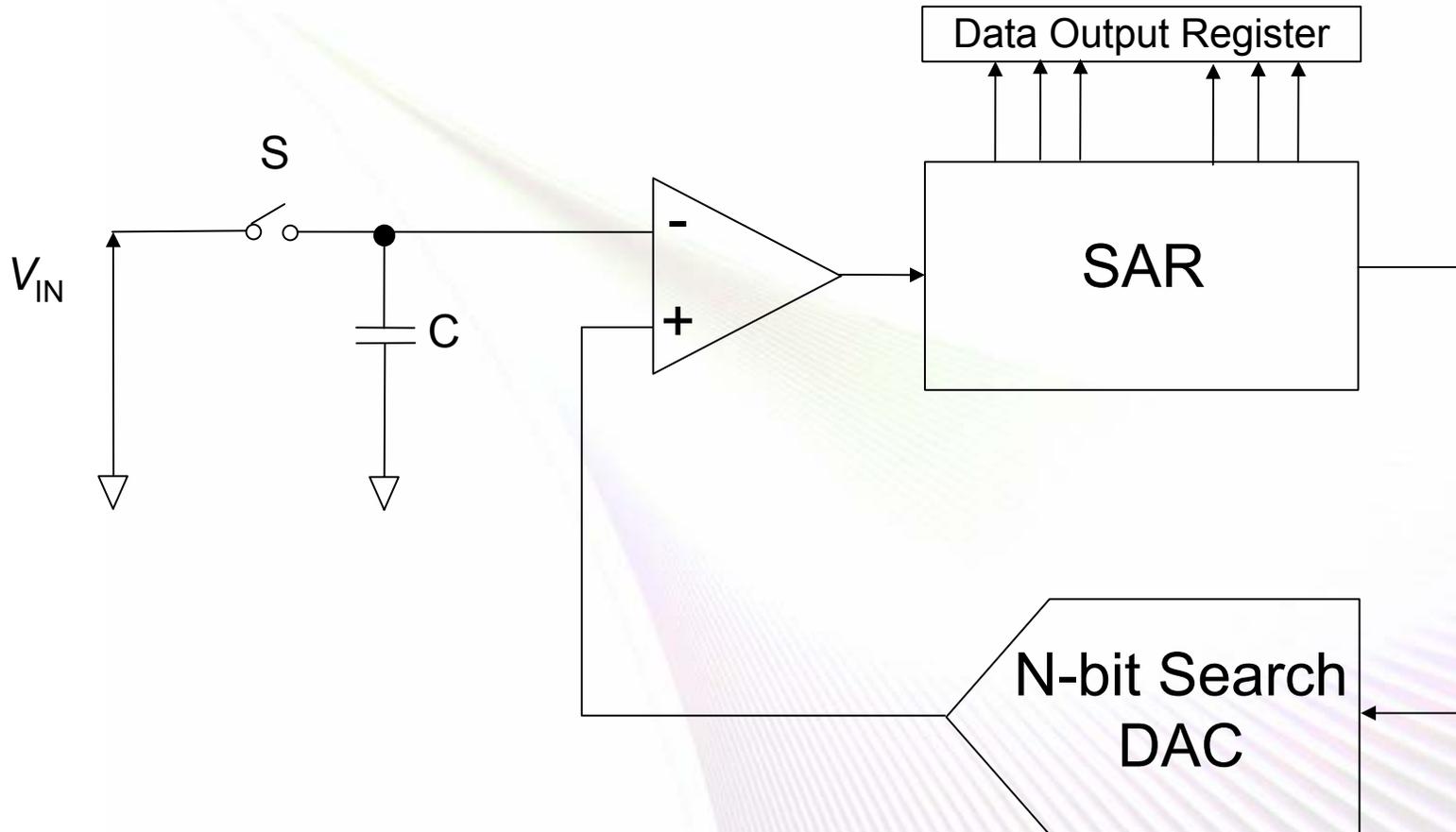
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Differential Input Example



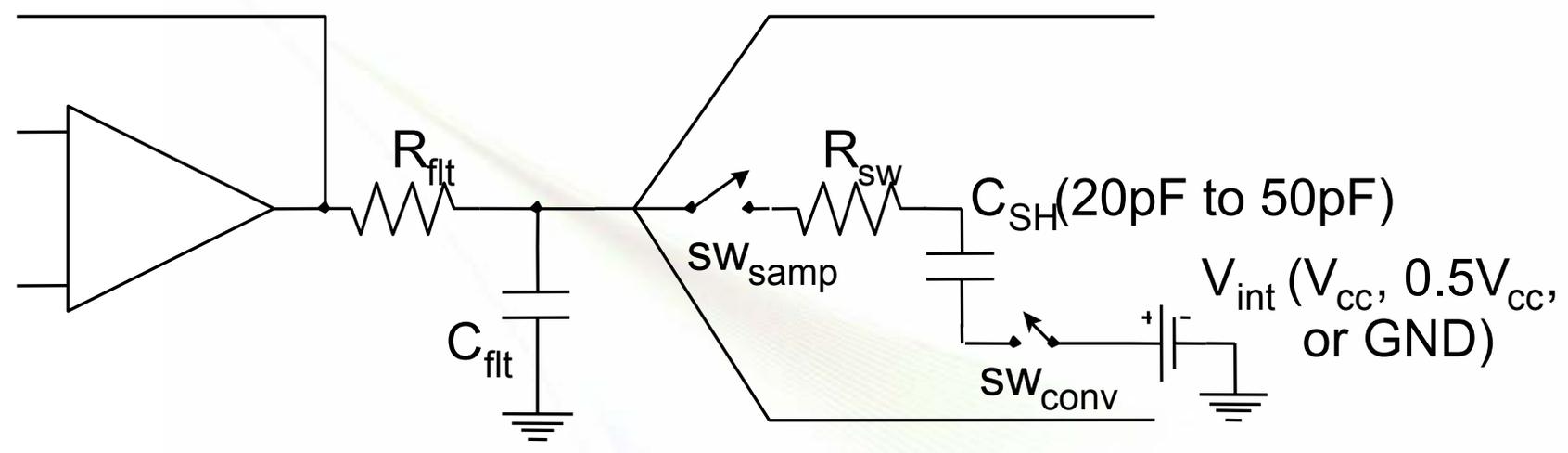
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Driving SAR Inputs



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Interface Circuit



Op Amp

CMV range, V_{OS} vs CMV
 RR out-swing to the rail
 Slew rate, Signal BW,
 Load transient, Settling time,
 Output impedance

Filter

Charge bucket
 Filtering,
 C_{load} isolation,

A/D

Acquisition time
 Input circuit parameters
 Initial voltage on C_{SH}

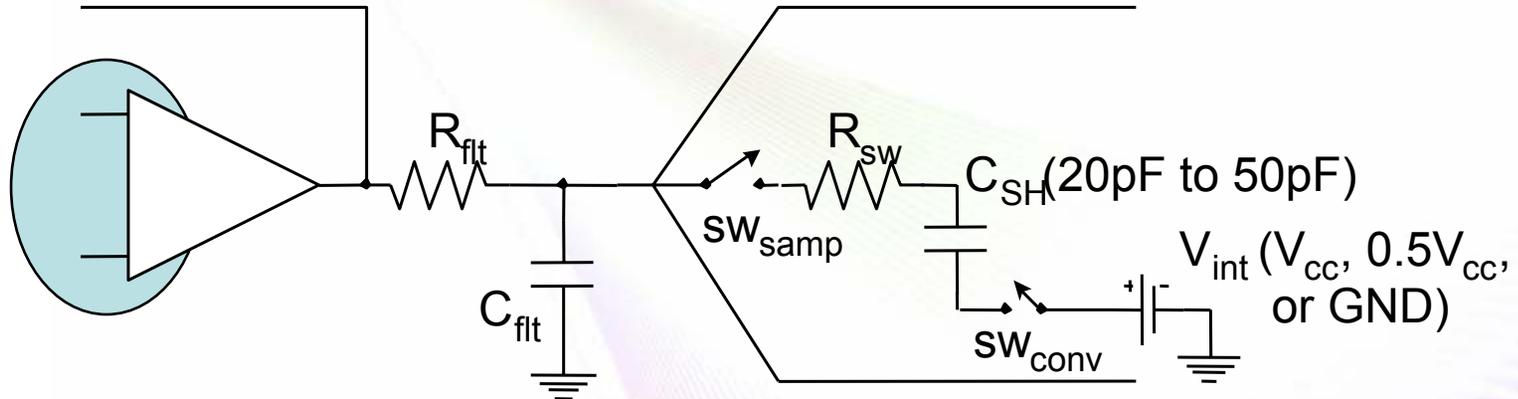
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Design Checklist Items

- Op Amp Common Mode Voltage
- Op Amp Output Swing to Rail
- Op Amp Settling Time
- Filter Capacitor Type
- Filter Component Values
- Op Amp Specifications

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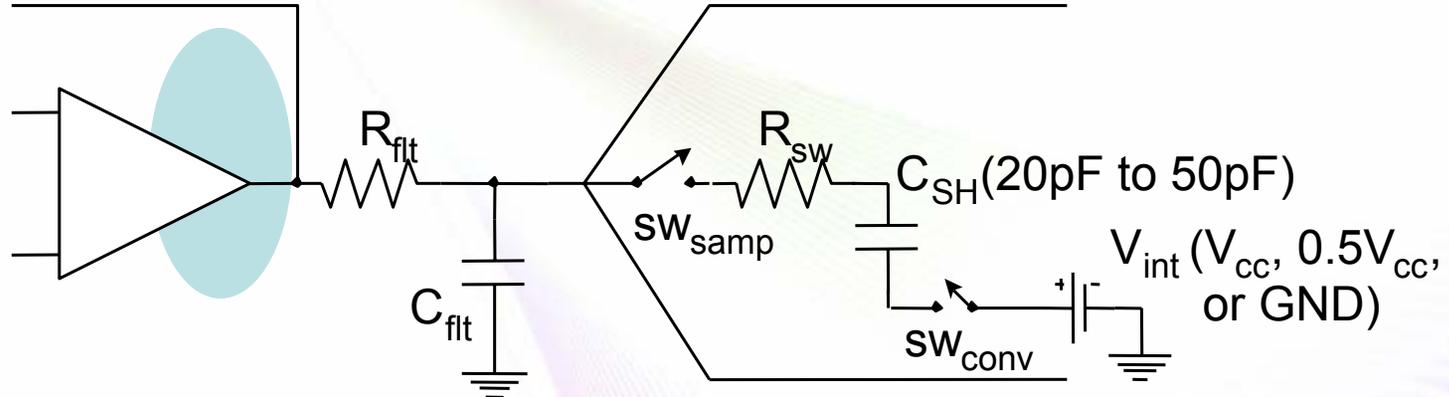
Amplifier Input Concerns



Rail-to-Rail Input

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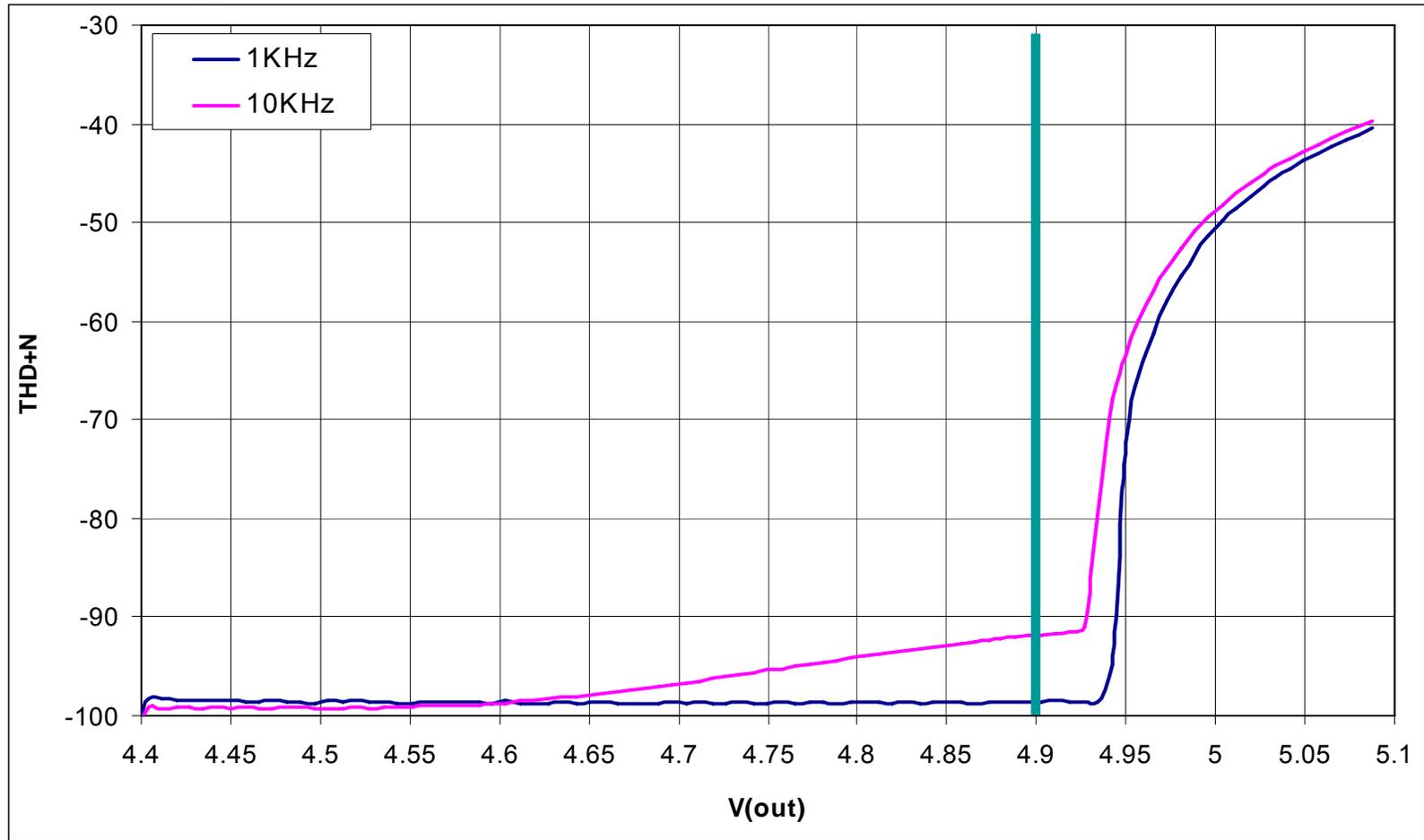
Drive Stage Concerns



Rail-to-Rail Output

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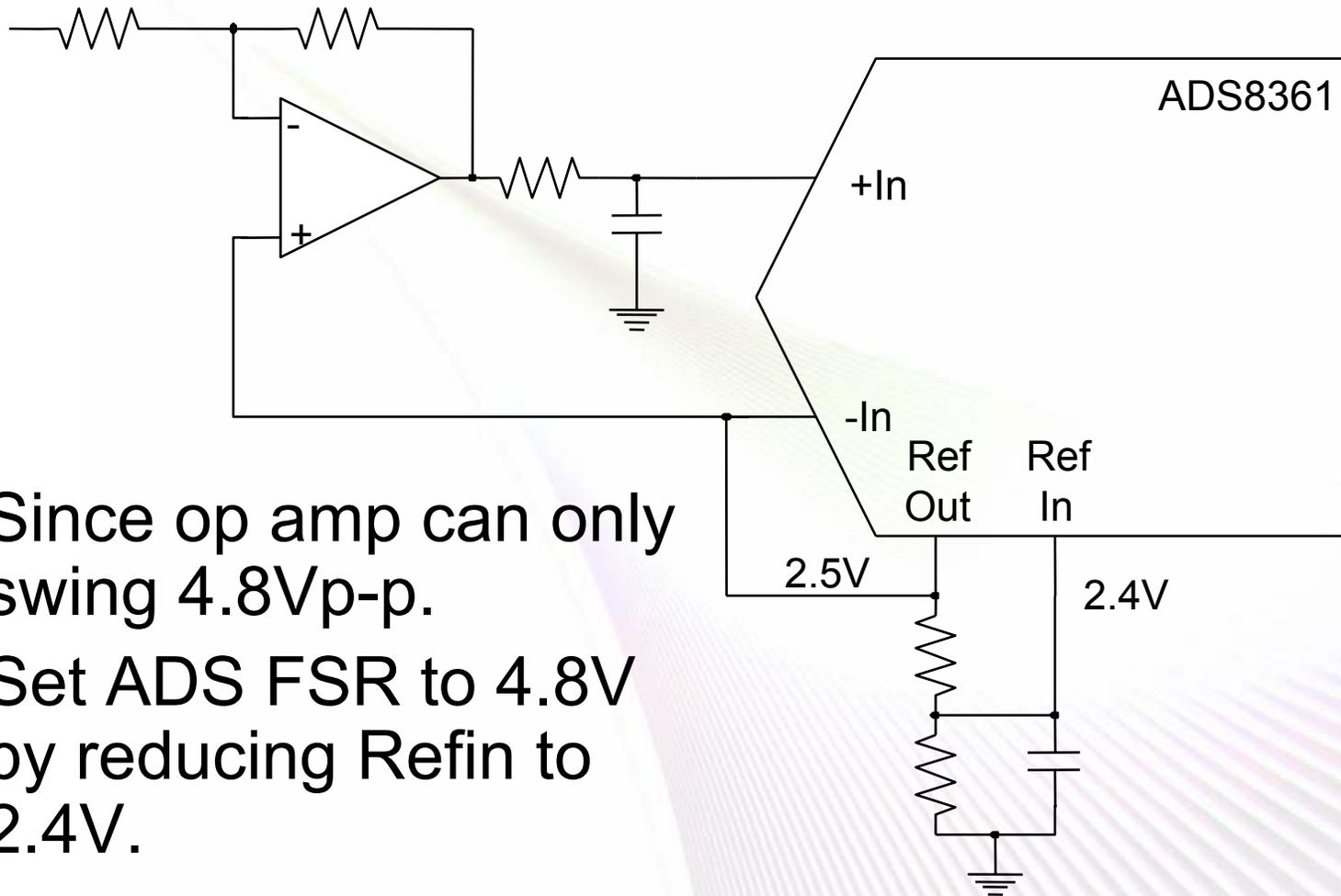
Rail-to-Rail Limitations



Example: OPA350

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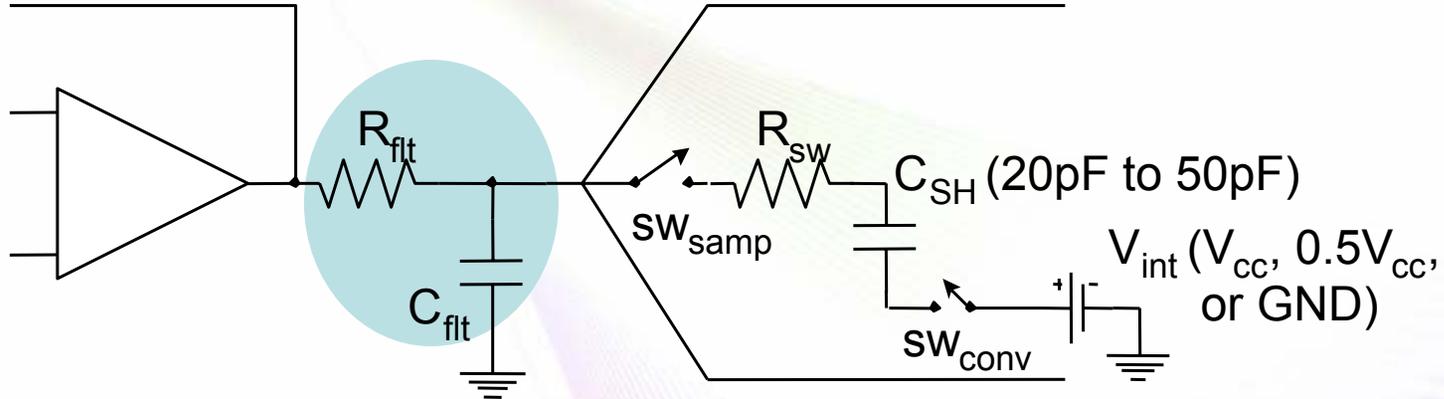
A Simple Solution



- Since op amp can only swing 4.8Vp-p.
- Set ADS FSR to 4.8V by reducing Refin to 2.4V.

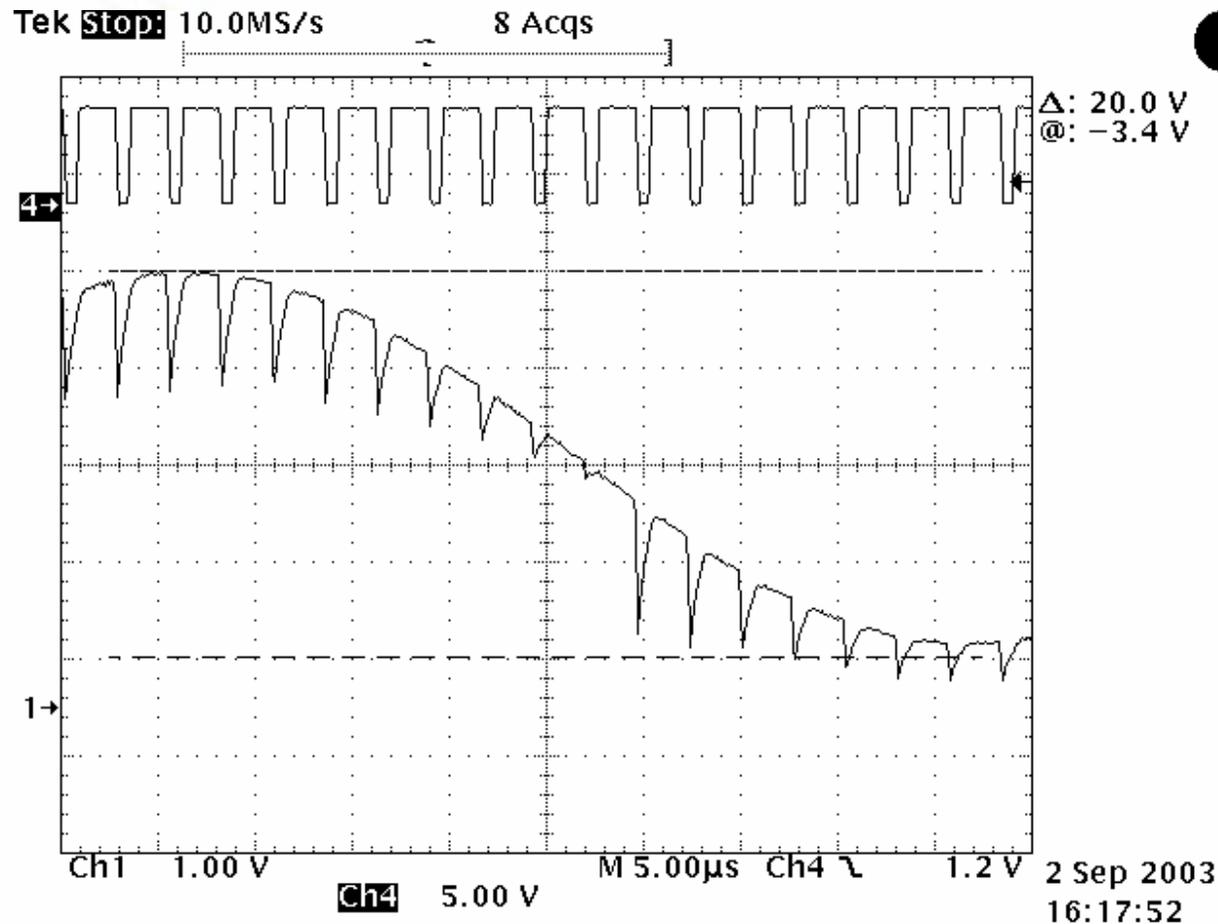
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Why the R/C Filter?



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Charge Transients



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R/C Component Selection

- Pick $C_{ft} = 20 * C_{SH}$
- R_{ft} Calculation
 - $t_{ft_settle} = t_{ACQ} = 12 \tau_{FLT}$
Theoretical Minimum

Practical Results:

- Use $t = 18 \tau_{FLT}$ margin for:
 - Op amp output load transient
 - Op amp output small signal settling time
- $R_{ft} = t_{ACQ} / 18 C_{ft}$

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Settling Time vs. Resolution

Number of bits	0.5LSB	Time Constants
10	0.0488281%	8
12	0.0122070%	9
14	0.0030518%	11
16	0.0007629%	12
18	0.0001907%	13
20	0.0000477%	15
22	0.0000119%	17
24	0.0000030%	18

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In Summary

- Choose the right ADC for your system input requirements
 - Single-ended
 - Pseudo-differential
 - Differential
- Buffer op amp
 - Rail-to-Rail considerations
 - Settling times and drive capability
- RC circuit
 - Use the starting criteria established here.
 - Experiment to optimize.

Minds in Motion

Balancing Cost and Performance

- What does the system really need?
- But I already have an on-board ADC...
- How to improve performance of an embedded converter.

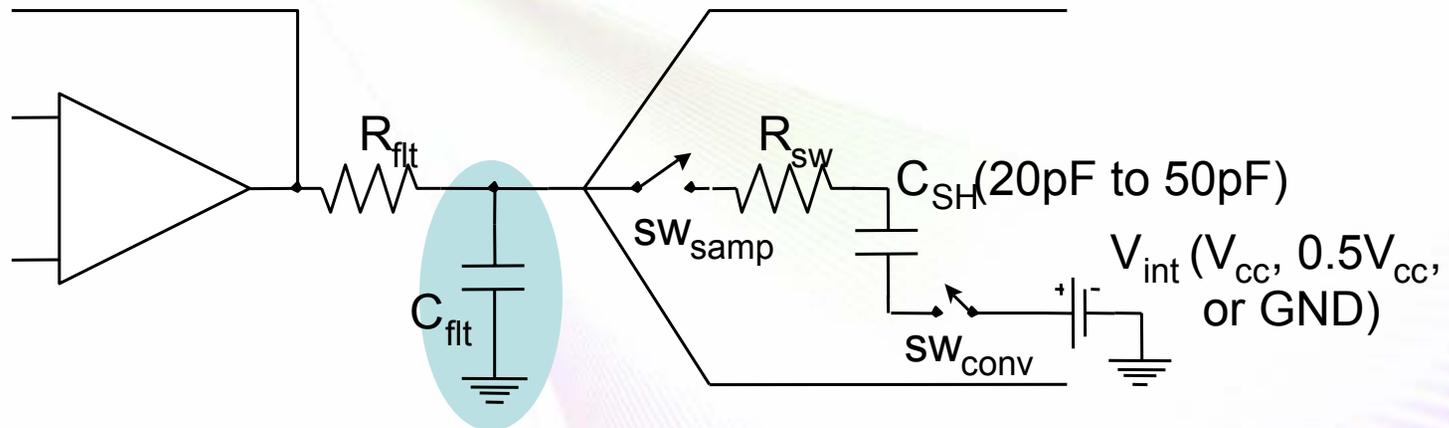
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System Input Requirements

- Component selection: choice does matter!
- What performance level are you looking for?
- Does the embedded ADC give the required accuracy?
- Does the on-board reference support the required accuracy?

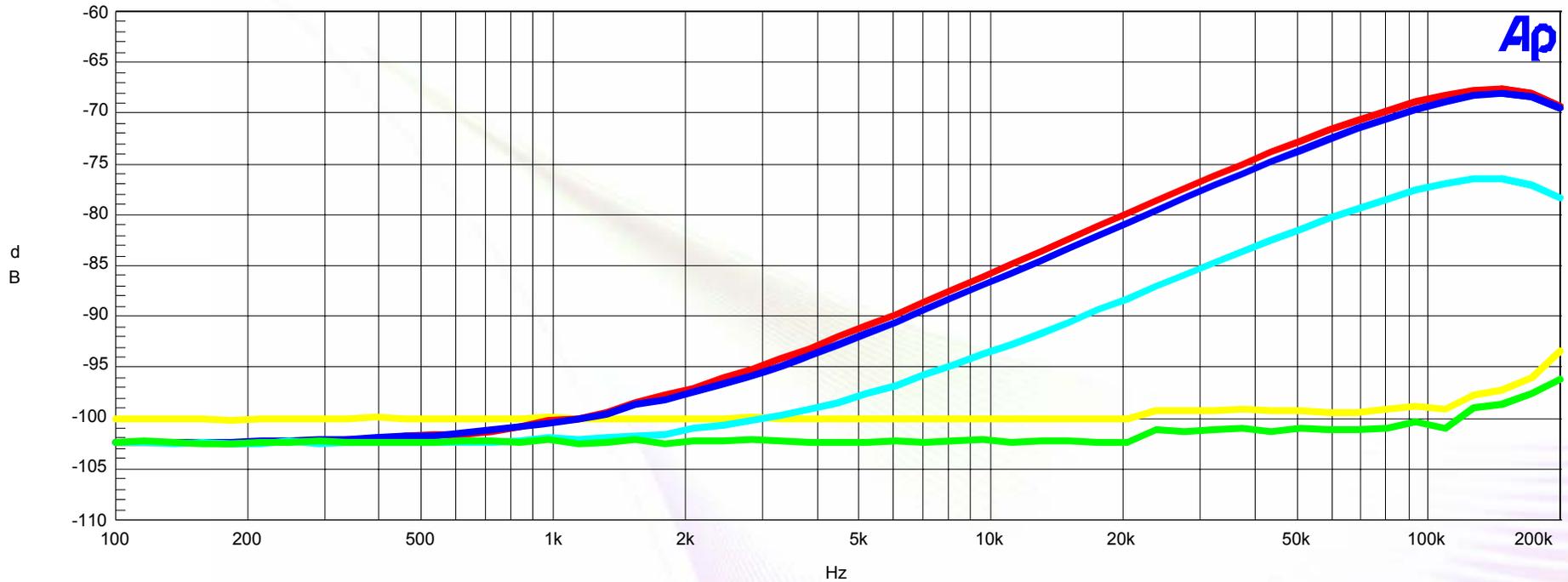
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Capacitor Choice Is Important!



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THD+N vs. Frequency

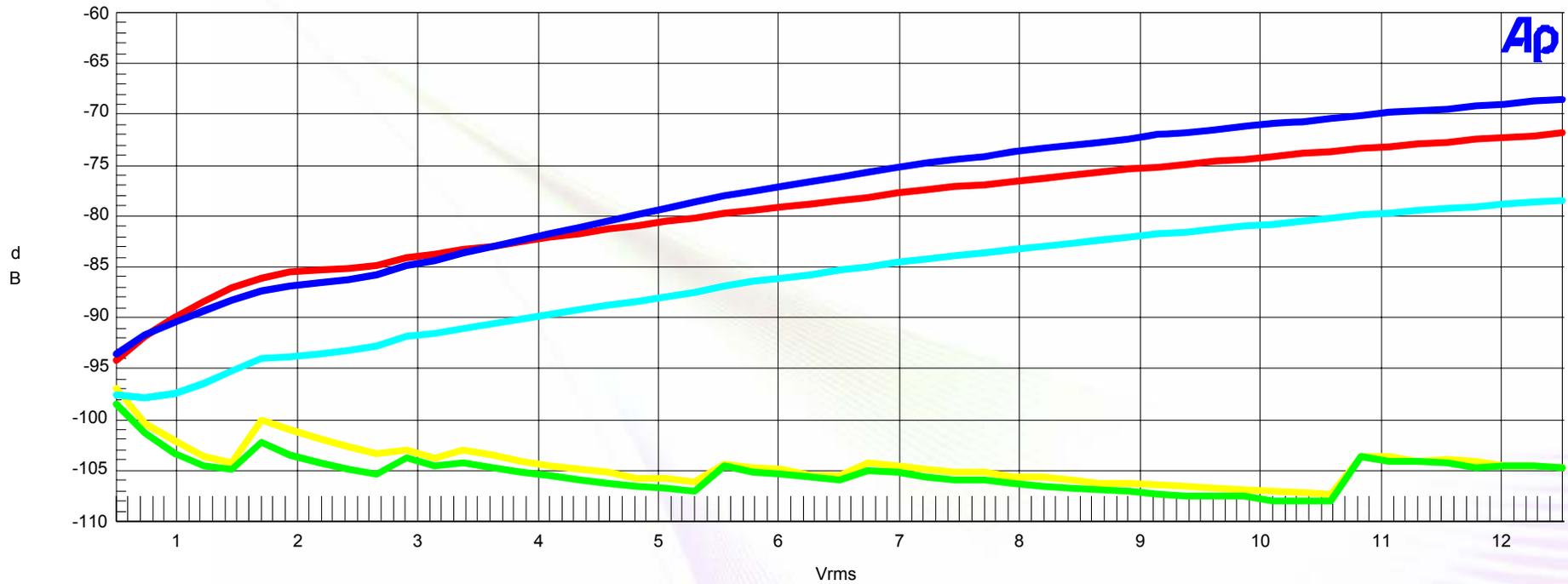


Sweep	Trace	Color	Line Style	Thick	Data	Axis	Comment
1	1	Yellow	Solid	1	Anlr.THd+N Ratio	Left	C = 0
2	1	Red	Solid	1	Anlr.THd+N Ratio	Left	C1
3	1	Blue	Solid	1	Anlr.THd+N Ratio	Left	C2
4	1	Cyan	Solid	1	Anlr.THd+N Ratio	Left	C3
5	1	Green	Solid	1	Anlr.THd+N Ratio	Left	C4

R = 100 ohm, C = 3.3 nF, $V_{p-p} = 5V$ $f(-3dB) = 482$ kHz

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THD+N vs. Voltage



Sweep	Trace	Color	Line Style	Thick	Data	Axis	Comment
1	1	Yellow	Solid	1	Anlr.THd+N Ratio	Left	C = 0
2	1	Red	Solid	1	Anlr.THd+N Ratio	Left	C1
3	1	Blue	Solid	1	Anlr.THd+N Ratio	Left	C2
4	1	Cyan	Solid	1	Anlr.THd+N Ratio	Left	C3
5	1	Green	Solid	1	Anlr.THd+N Ratio	Left	C4

R = 100 ohm, C = 3.3 nF, f = 10 kHz, f(-3dB) = 482 kHz

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What 'C' Type Should I Use?

- Best performance:
 - Silver Mica or C0G(NPO)
- Avoid others.
- Others may cost less and be smaller but can distort the input signal.

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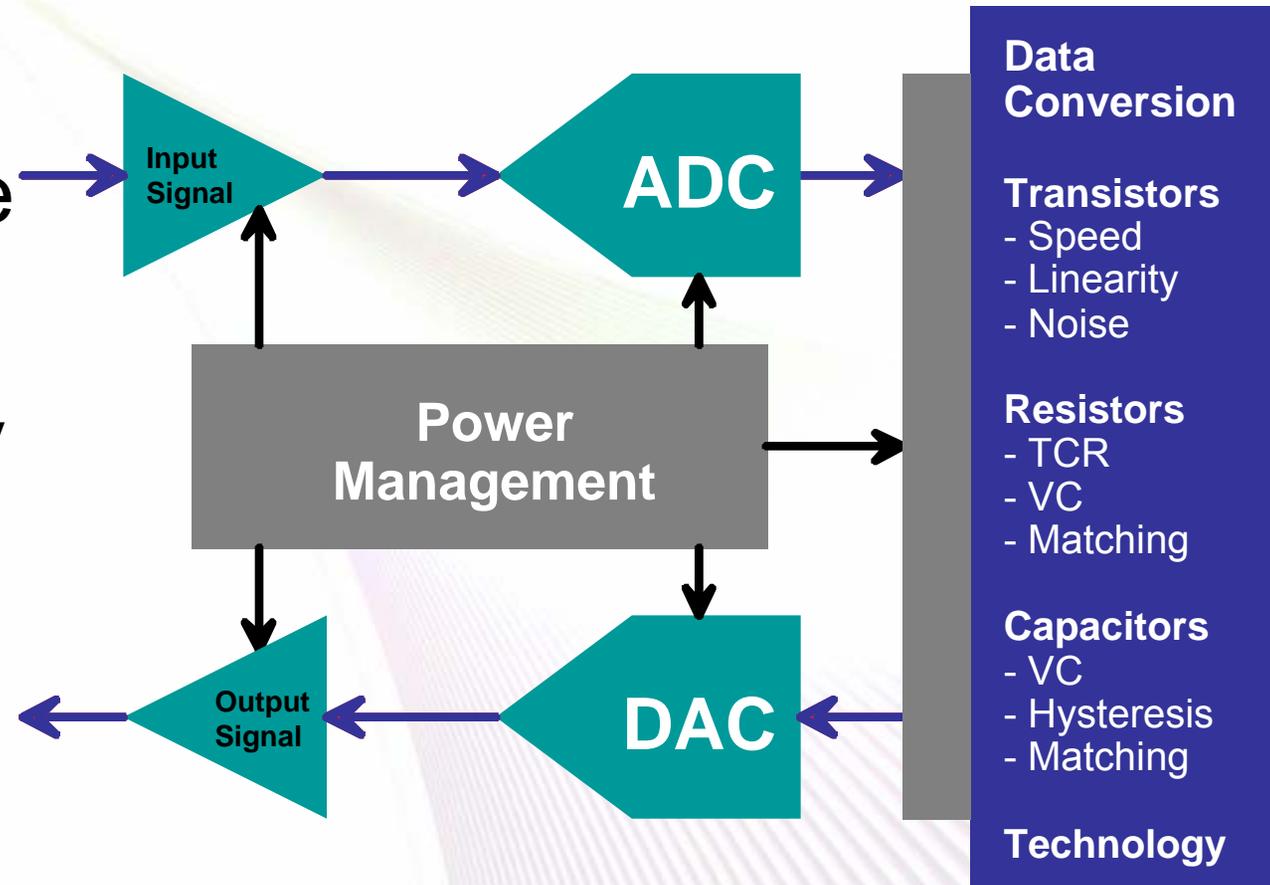
Embedded Processor ADCs

- Performance vs. cost
- Process issues
 - Digital process, analog application...
- How to improve on-board converter accuracy
 - Try an external precision reference!

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DAQ Signal Chain

- Temp
- Pressure
- Speed
- Humidity
- Position
- Flow
- Light



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Specifications to Watch for

- Speed: How fast is the converter?
- Resolution: What is the ENOB?
- INL: Integral non-linearity
- DNL: Differential non-linearity
- Offset: Error (in LSBs) regarding offset
- Gain: Error (in %FSR) regarding gain

Minds in Motion

Important Parameters

		ADS7829		ADS7866		ADS7886		56F807	
		Typ	Max	Typ	Max	Typ	Max	Typ	Max
Speed	kSPS		125		100		1000		714*
Resolution		12		12		12		12	
INL (@12bit)	LSB	±0.4	±1.25	±1	±1.5	±0.6	±1.5	±2.5	±4
DNL	LSB	±0.4	+1.25/-1	±0.8	+1.5/-0.9	±0.75	+1.5/-0.9	±0.9	±1
Offset	mV	±0.18	±1.83	±0.5	±0.87	±0.31	±0.92	-25	+10/-90
Gain	%	±0.007	±0.05	±0.007	±0.05	±0.012	±0.04	±0	+8/-7

ELECTRICAL CHARACTERISTICS

Over recommended operating free-air temperature range at -40°C to 85°C , $V_{\text{DD}} = 5\text{V}$, $BV_{\text{DD}} = 3\text{V}$, $V_{\text{REF}} = \text{internal } +2.5\text{V}$, $f_{\text{CLK}} = 10\text{MHz}$, and $f_{\text{SAMPLE}} = 500\text{ kSPS}$, unless otherwise noted.

PARAMETER	CONDITIONS	ADS8361			UNITS
		MIN	TYP ⁽¹⁾	MAX	
ANALOG INPUT					
Full-Scale Range ⁽²⁾	(FSR) +IN – (–IN)			$\pm V_{\text{REF}}$	V
Operating Common-Mode Signal		2.2		2.8	V
Input Switch Resistance	–IN = V_{REF}		20		Ω
Input Capacitance	–IN = V_{REF}		25		nF

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Datasheet Specs...

ELECTRICAL CHARACTERISTICS

Over recommended operating free-air temperature range at -40°C to 85°C , $\text{AV}_{\text{DD}} = 5\text{V}$, $\text{BV}_{\text{DD}} = 3\text{V}$, $\text{V}_{\text{REF}} = \text{internal} + 2.5\text{V}$, $f_{\text{CLK}} = 10\text{MHz}$, and $f_{\text{SAMPLE}} = 500\text{KSPS}$, unless otherwise noted.

PARAMETER	CONDITIONS	ADS8361			UNITS
		MIN	TYP ⁽¹⁾	MAX	
ANALOG INPUT					
Full-Scale Range ⁽²⁾ (FSR)	+IN – (–IN)	2.2		$\pm\text{V}_{\text{REF}}$	V
Operating Common-Mode Signal			20	2.8	V
Input Switch Resistance	–IN = V_{REF}		25		Ω
Input Capacitance	–IN = V_{REF}		25		pF
Input Leakage Current	–IN = V_{REF}		± 1		nA
Differential Input Switch Resistance			40		Ω
Differential Input Capacitance			15		pF
Common-Mode Rejection Ratio (CMRR)	At DC		84		dB
	$\text{V}_{\text{IN}} = \pm 1.25\text{Vp-p}$ at 50kHz		80		dB
DC ACCURACY					
Resolution		16			Bits
No Missing Code (NMC)		14			Bits
Integral Linearity Error (INL)			± 3	± 8	LSB ⁽³⁾
Integral Linearity Match	Channel 0/1, Same A/D		4		LSB
Differential Nonlinearity (DNL)			$+1.5^{(4)}$		LSB
Bipolar Offset Error (V_{OS})	Channel 0/1, Same A/D		± 0.5	± 2	mV
Bipolar Offset Error Match			0.5	1	mV
Bipolar Offset Error Drift (TCV_{OS})			0.4		ppm/ $^{\circ}\text{C}$
Gain Error ⁽⁶⁾ (G_{ERR})			± 0.05	± 0.5	%
Gain Error Match			0.05	0.15	%
Gain Error Drift (TCG_{ERR})			20		ppm/ $^{\circ}\text{C}$
Noise			60		μV_{rms}
Power-Supply Rejection Ratio (PSRR)	$4.75\text{V} < \text{AV}_{\text{DD}} < 5.25\text{V}$, with External Reference, at DC		–70		dB
SAMPLING DYNAMICS					
Conversion Time per A/D (t_{CONV})	$100\text{kHz} \leq f_{\text{CLK}} \leq 10\text{MHz}$	1.6		160	μs
Acquisition Time (t_{AQ})	$f_{\text{CLK}} = 10\text{MHz}$	400			ns
Throughput Rate				500	kSPS

NOTES: (1) All Values are at $T_A = 25^{\circ}\text{C}$. (2) Ideal input span; does not include gain or offset error. (3) LSB means Least Significant Bit, with V_{REF} equal to +2.5V; 1LSB = $76\mu\text{V}$. (4) Specified for 14-bit no missing code. (5) Specified for 15-bit no missing code. (6) Measured relative to an ideal, full-scale input (+IN – (–IN)) of 4.9999V. Thus, gain error does not include the error of the internal voltage reference.

ADC's performance over temperature at specified speed

Ensured values based on tests in production

Clear test conditions for specified parameters

No hidden and unclear limitations

Read the Fine Print!

Table 35. ADC Characteristics

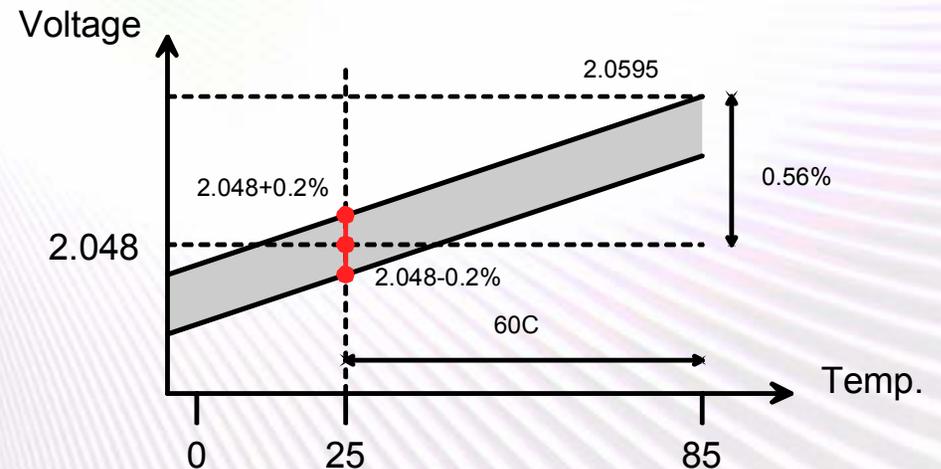
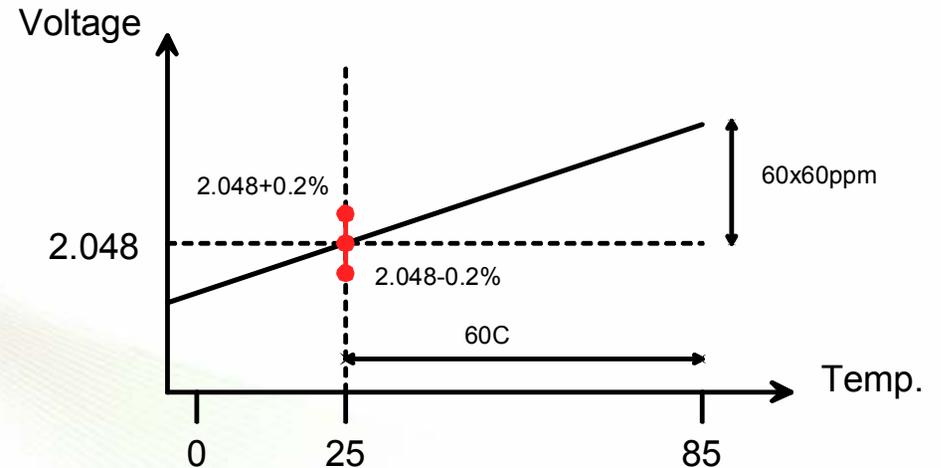
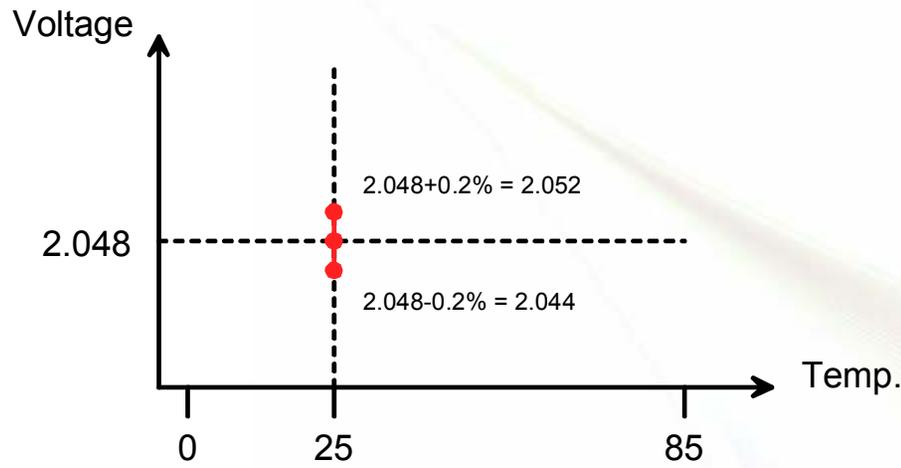
Operating Conditions: $V_{SS} = V_{SSA} = 0\text{ V}$, $V_{DD} = V_{DDA} = 3.0\text{--}3.6\text{ V}$, $V_{REF} = V_{DD} - 0.3\text{ V}$, ADCDIV = 4, 9, or 14, (for optimal performance), **ADC clock = 4MHz**, $3.0\text{--}3.6\text{ V}$, $T_A = -40^\circ\text{ to }+85^\circ\text{C}$, $C_L \leq 50\text{ pF}$, $f_{OP} = 80\text{ MHz}$

Characteristic	Symbol	Min	Typ	Max	Unit
ADC input voltages	V_{ADCIN}	0 ¹	—	V_{REF} ²	V
Resolution	R_{ES}	12	—	12	Bits
Integral Non-Linearity ³	INL	—	+/- 2.5	+/- 4	LSB ⁴
Differential Non-Linearity	DNL	—	+/- 0.9	+/- 1	LSB ⁴
Monotonicity	GUARANTEED				
ADC internal clock ⁵	f_{ADIC}	0.5	—	5	MHz
Conversion range	R_{AD}	V_{SSA}	—	V_{DDA}	V
Conversion time	t_{ADC}	—	6	—	t_{AIC} cycles ⁶
Sample time	t_{ADS}	—	1	—	t_{AIC} cycles ⁶
Input capacitance	C_{ADI}	—	5	—	pF ⁶
Gain Error (transfer gain) ⁵	E_{GAIN}	0.93	1.00	1.08	—

- For optimum ADC performance, **keep the minimum V_{ADCIN} value $\geq 25\text{ mV}$** . Inputs less than 25mV may convert to a digital output code of 0.
- V_{REF} must be equal to or less than V_{DDA} and must be greater than 2.7V. For optimal ADC performance, set V_{REF} to $V_{DDA} - 0.3\text{ V}$.
- Measured in 10-90% range.**
- LSB = Least Significant Bit.
- Guaranteed by characterization.**

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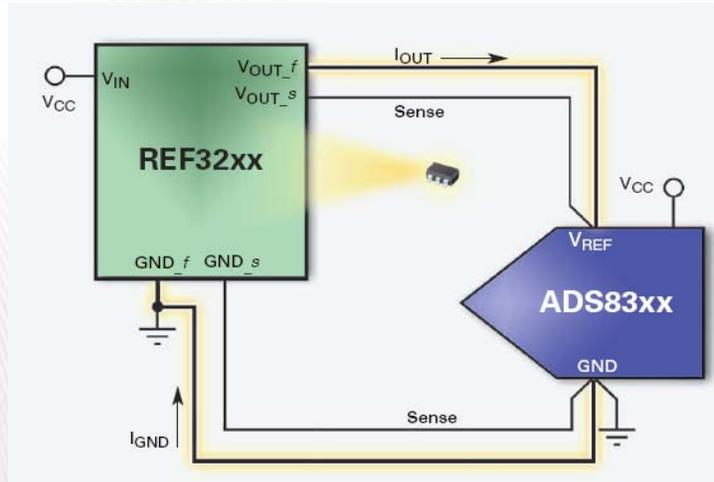
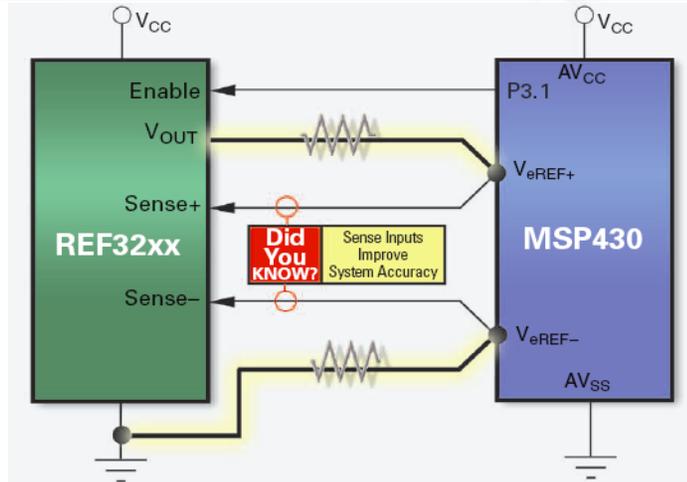
Reference Voltage Accuracy



- Initial accuracy: $\pm 0.2\%$
- Temp. coefficient: $60\text{ppm}/^\circ\text{C}$
- Range of interest: 60°C
- Insured accuracy: $\pm 0.56\%$
 $0.2\% + (60 \times 60)\text{ppm} =$
 $0.2\% + 0.36\% = 0.56\%$

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Device	Initial Accuracy (%)		Tmp.Coef. (ppm/°C)		Total Accuracy (%)	
	Max	Typ	Max	Typ	Max	Typ
MSP430F167	±4	—	±100*	—	±4.60	—
56F807	—	—	—	—	—	—
F2808	—	±50	—	—	?	—
ADS7829	—	—	—	—	—	—
ADS7866	—	—	—	—	—	—
ADS1203	±2	±20	—	—	±2.12	—
ADS8361	±1	±20	—	—	±1.12	—
REF29xx	±2	±35	±100	—	±2.21	±2.60
REF30xx	±0.2	±20	±50	—	±0.32	±0.50
REF31xx	±0.2	±5	±15	—	±0.23	±0.29
REF32xx	±0.2	±4	±7	—	±0.22	±0.24



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PCB Layout Issues

- LSB size
 - Affects your layout decisions.
- Basic PCB Design Principles
 - Care must be exercised for high resolution!
- PCB Layout Reviews
 - Learn from others! Ask your co-workers what worked and what did not. Better yet: Ask us!

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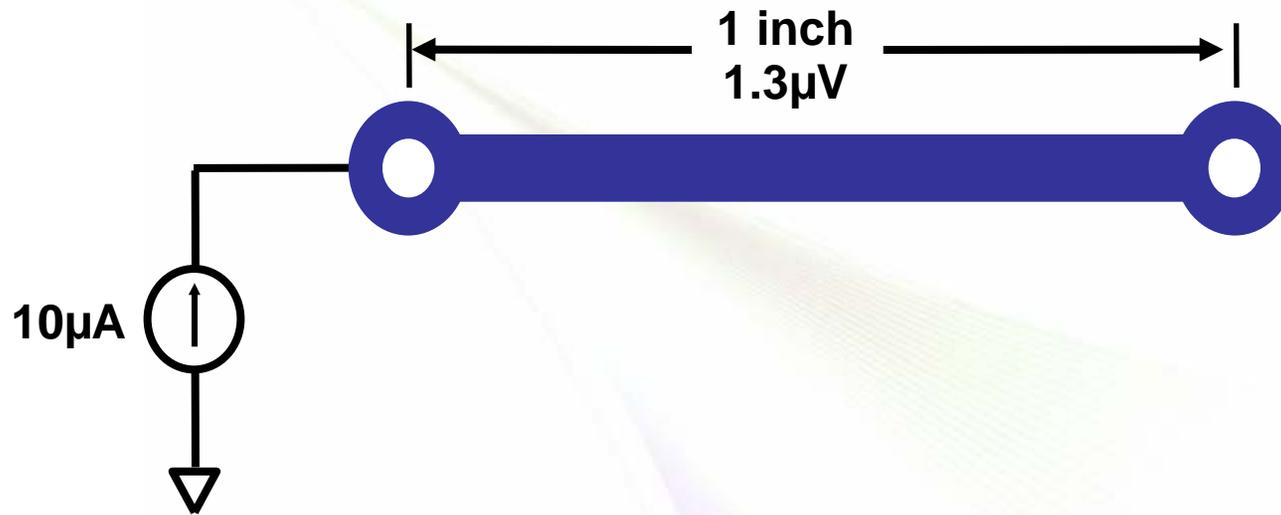
LSB Size

Signal range is critical

- $\pm 10V$ is a FSR of 20V
 - 16 bits: $20V/65,536 = 305\mu V$ per LSB
 - 24 bits: $20V/16,777,216 = 1,192nV$ per LSB
- $\pm 2.5V$ is a FSR of 5V
 - 16 bits: $5V/65,536 = 76.3\mu V$ per LSB
 - 24 bits: $5V/16,777,216 = 298nV$ per LSB
- $\pm 0.020V$ is a FSR of 0.040V
 - 16 bits: $0.040V/65,536 = 0.610\mu V$ per LSB
 - 24 bits: $0.040V/16,777,216 = 2nV$ per LSB

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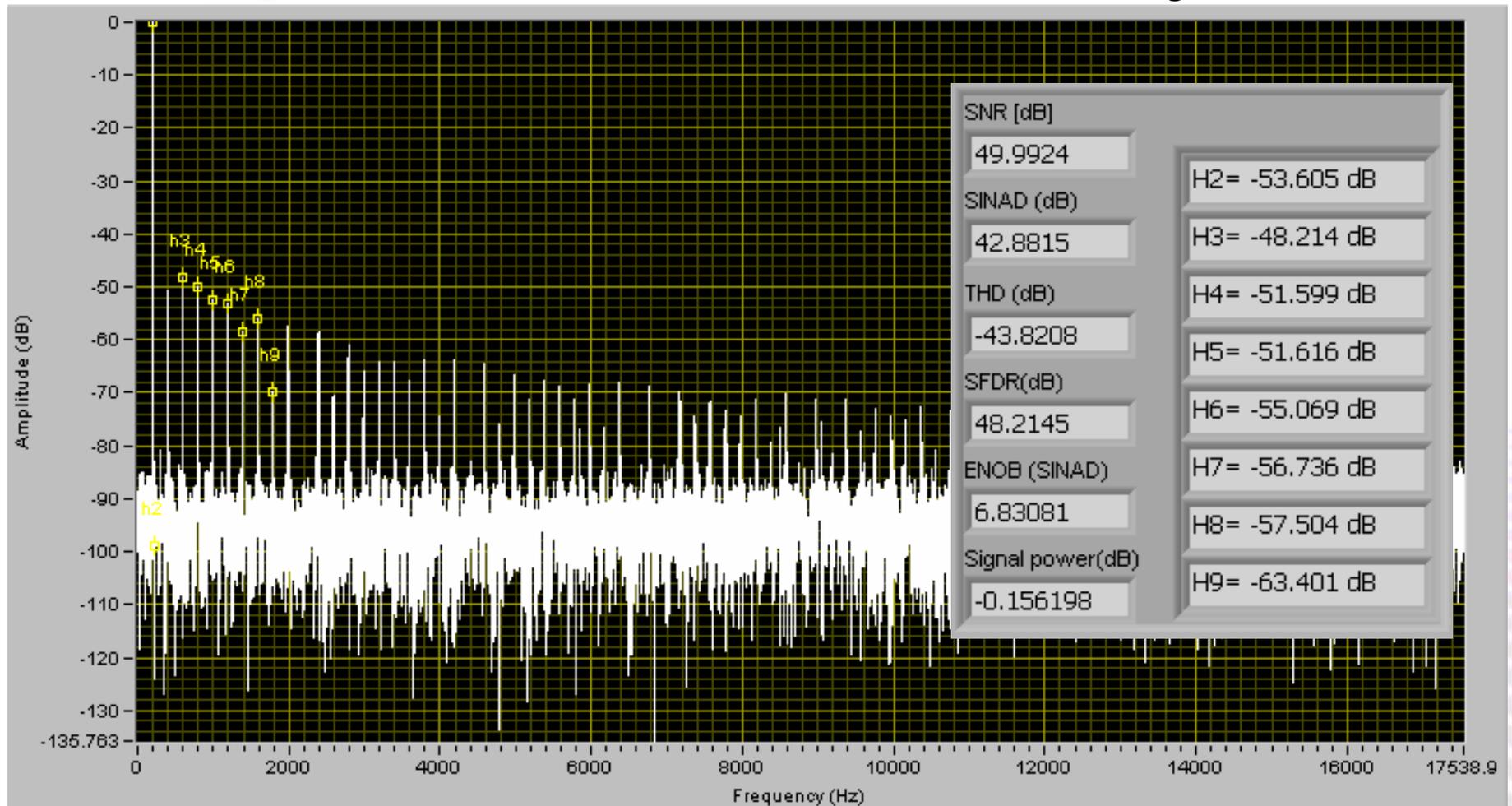
Trace Length/Width Concerns



- 1 inch (7 mil) trace of 1/2 oz copper with $10\mu\text{A}$ of current => voltage drop of $1.3\mu\text{V}$.
- This is 4 LSBs at 24 bits!

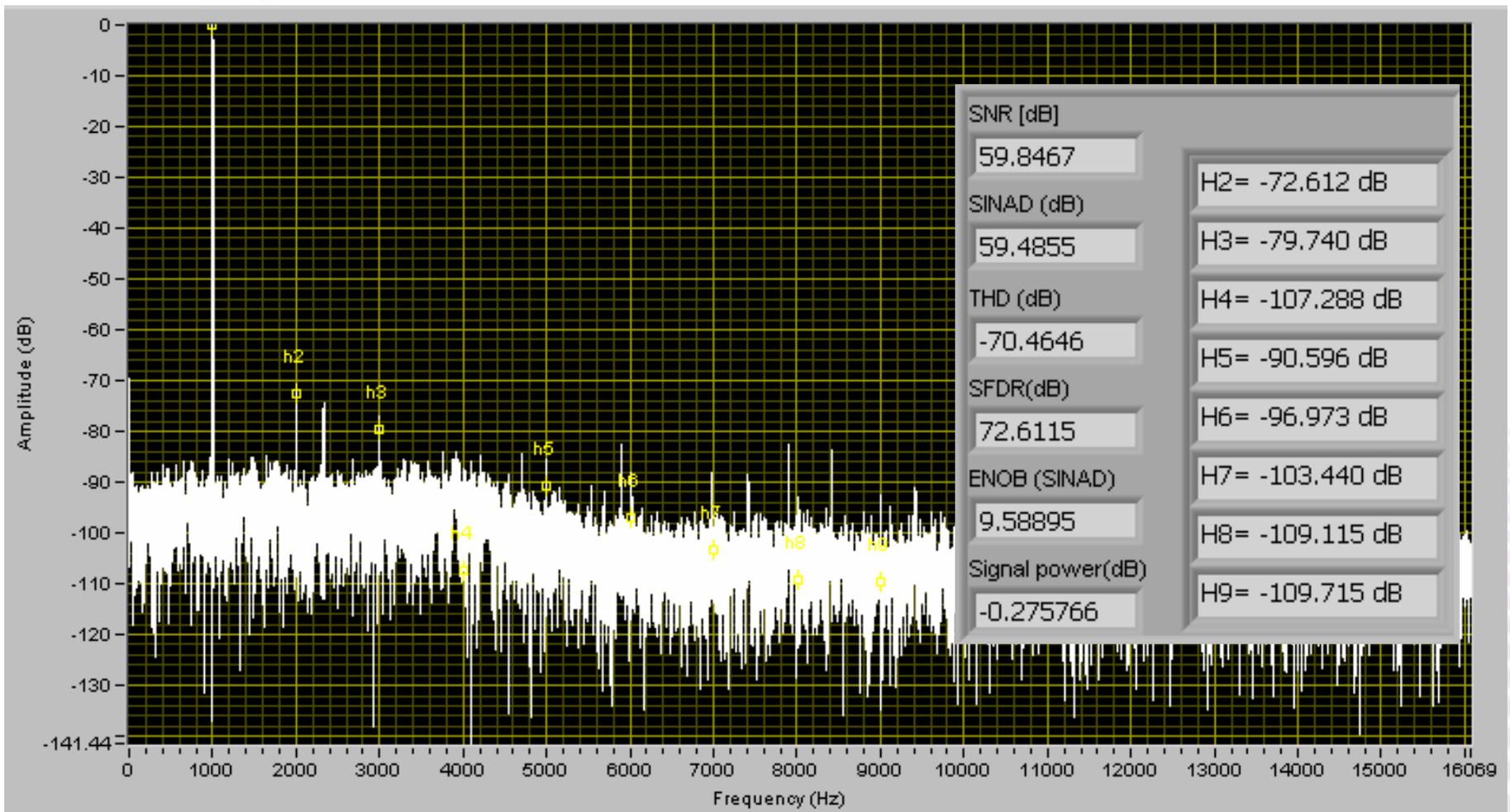
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AC Performance vs. Layout



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With Short Traces



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Basic Layout Issues

- Separate analog and digital signals
- AGND and DGND connected at ADC/DAC
- Provide good ground return paths
- High-frequency bypassing
- Minimize inductance
- Differential signal measurements
 - Route in pairs

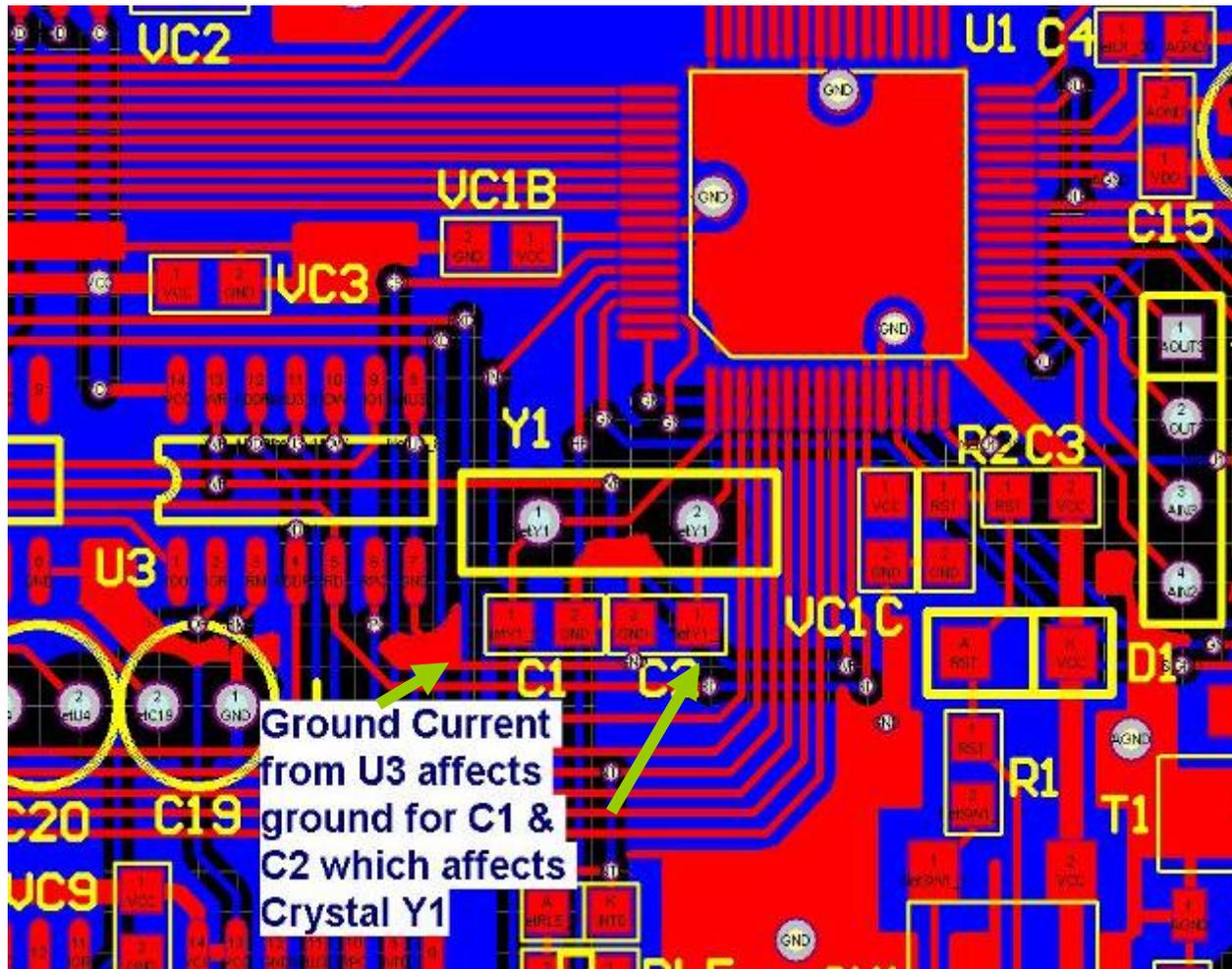
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Avoid 'Auto Routing'

- Auto routers do not handle analog signals very well!
- Route critical signals by hand for best performance – this is imperative.
- Route “like” channels in similar fashion.
- Remember to tie grounds locally at the most critical circuit – usually the ADC.

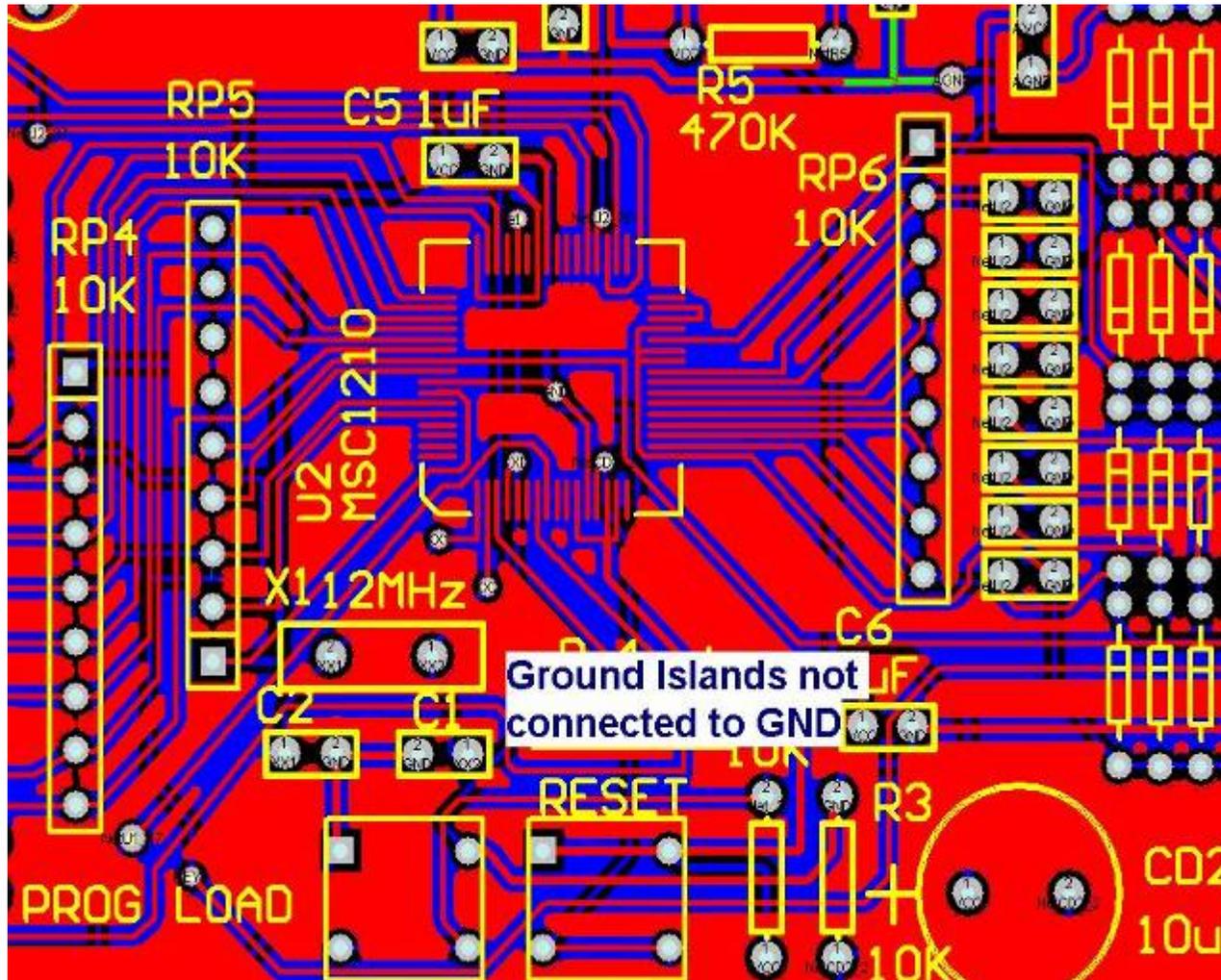
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PCB Layout



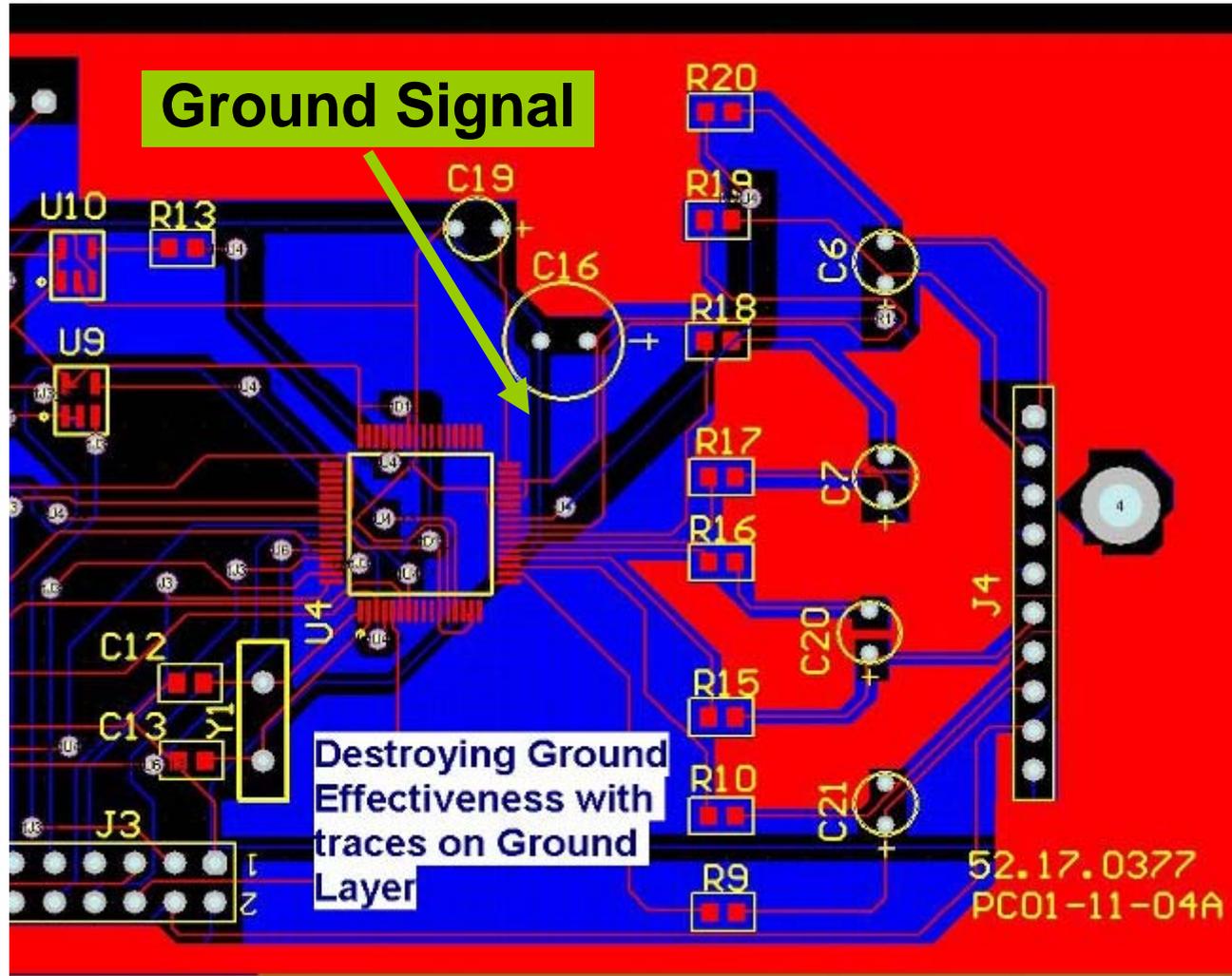
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PCB Layout



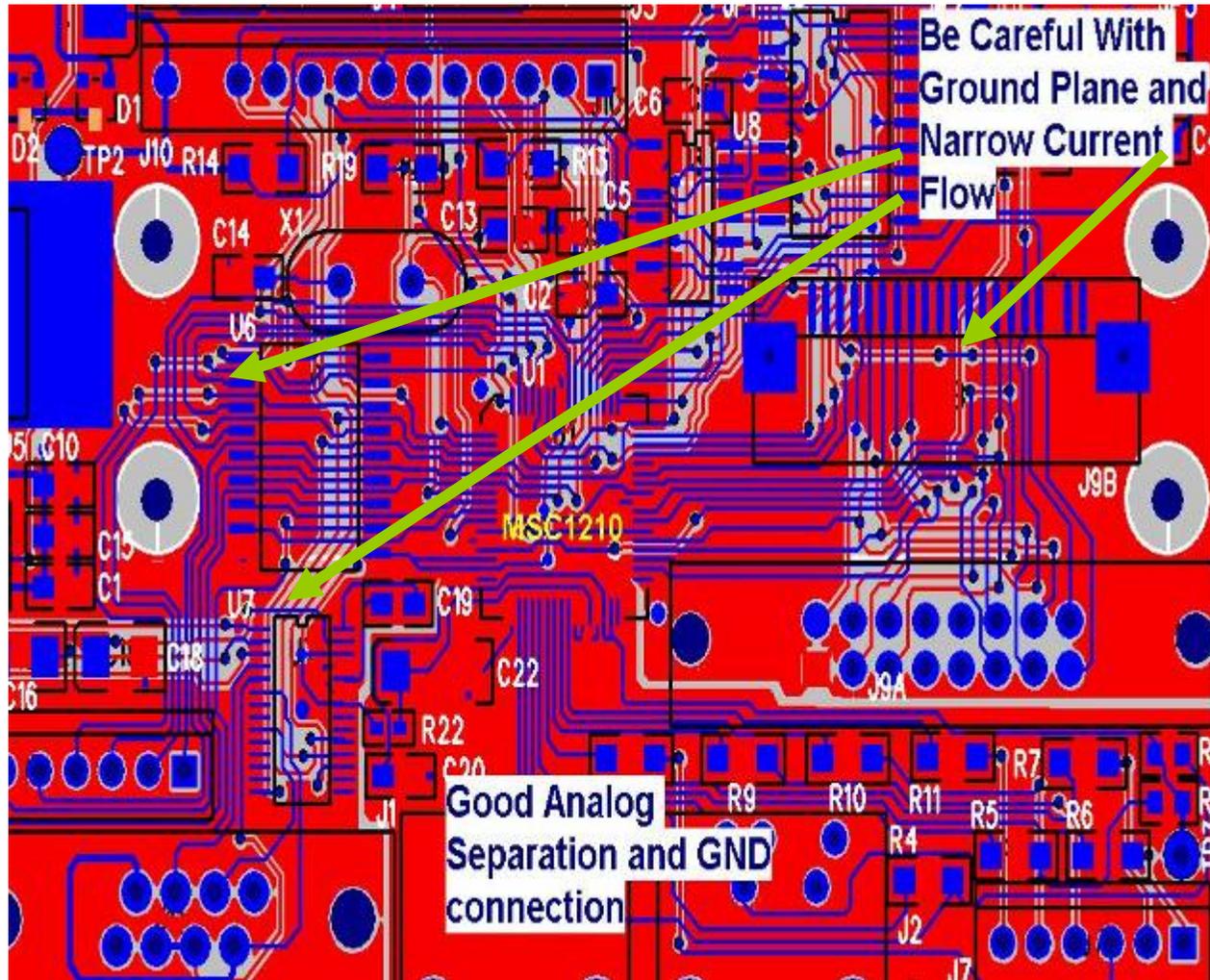
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PCB Layout



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PCB Layout



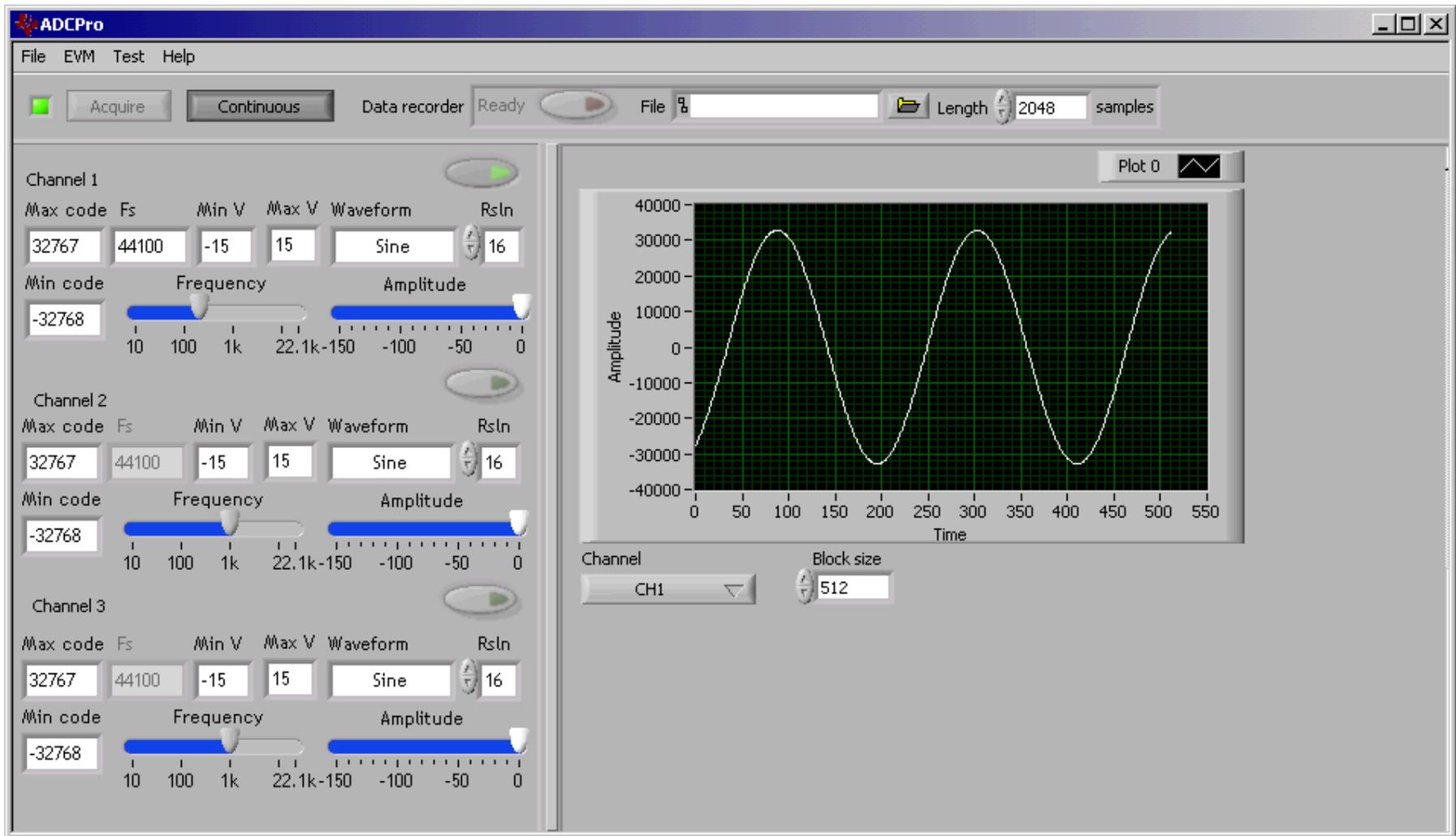
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Demo Time!

- Software Tools
 - ADC Pro
 - Code Composer Studio™, Embedded Workbench™
- Hardware Tools
 - Analog EVMs
 - DSP starter kits
- Where to get more
 - App notes
 - Code examples

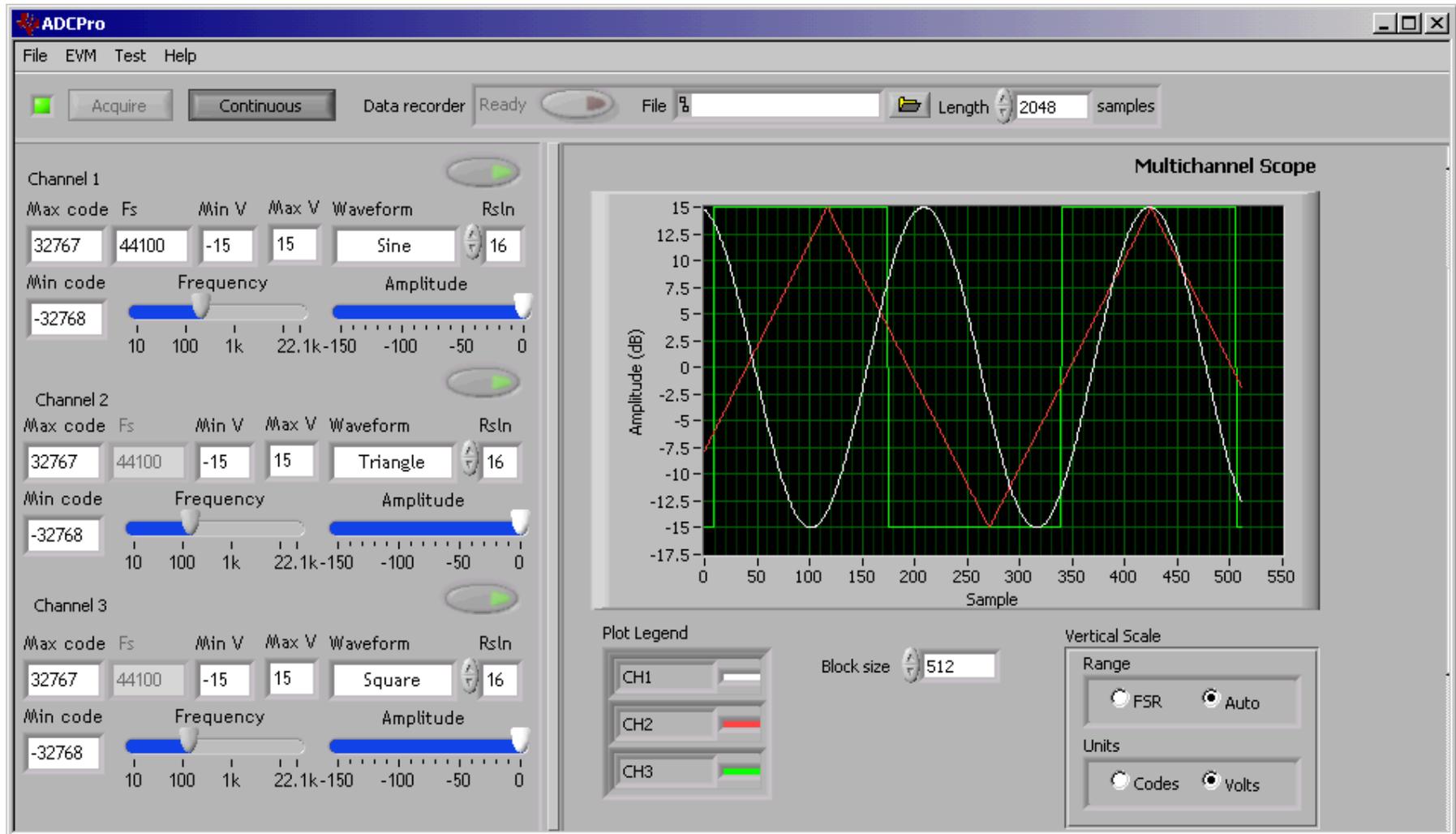
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ADC Pro



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ADC Pro



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Writing Application Code

- Code Composer Studio IDE
 - Application design, code and build, debug
 - TMS320 and TMS470
- Code Composer Essentials
 - C compiler, assembler and linker
 - MSP430 ultra-low-power controllers
- IAR Embedded Workbench
 - IDE for embedded applications
 - MSP430 and TMS470

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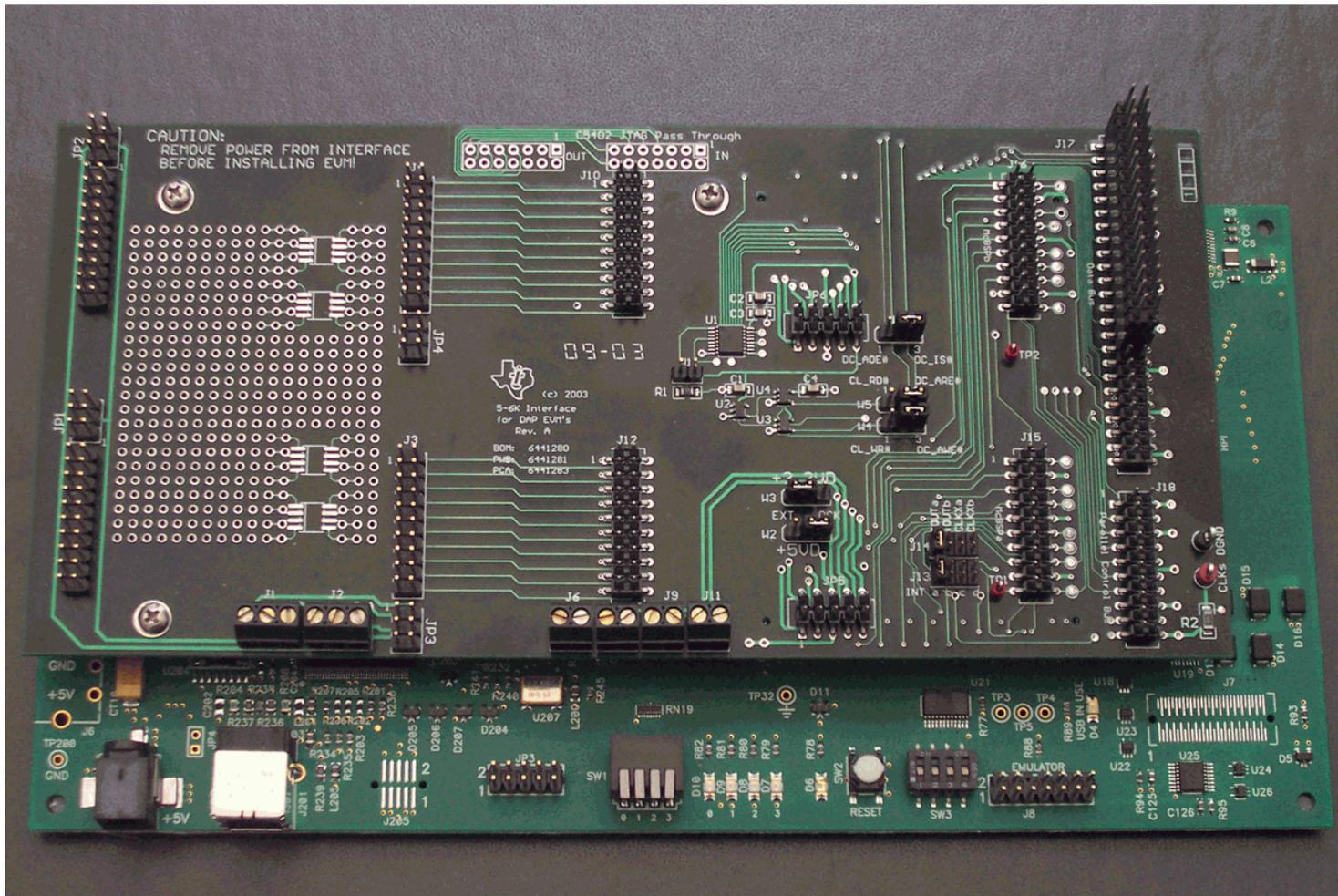
Code-Generation Tools

- Data converter “plug-in” tool for CCS
- Supports:
 - ADCs
 - DACs
 - CODECs
 - AMC
 - PCM



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TMS320C6713 Hardware

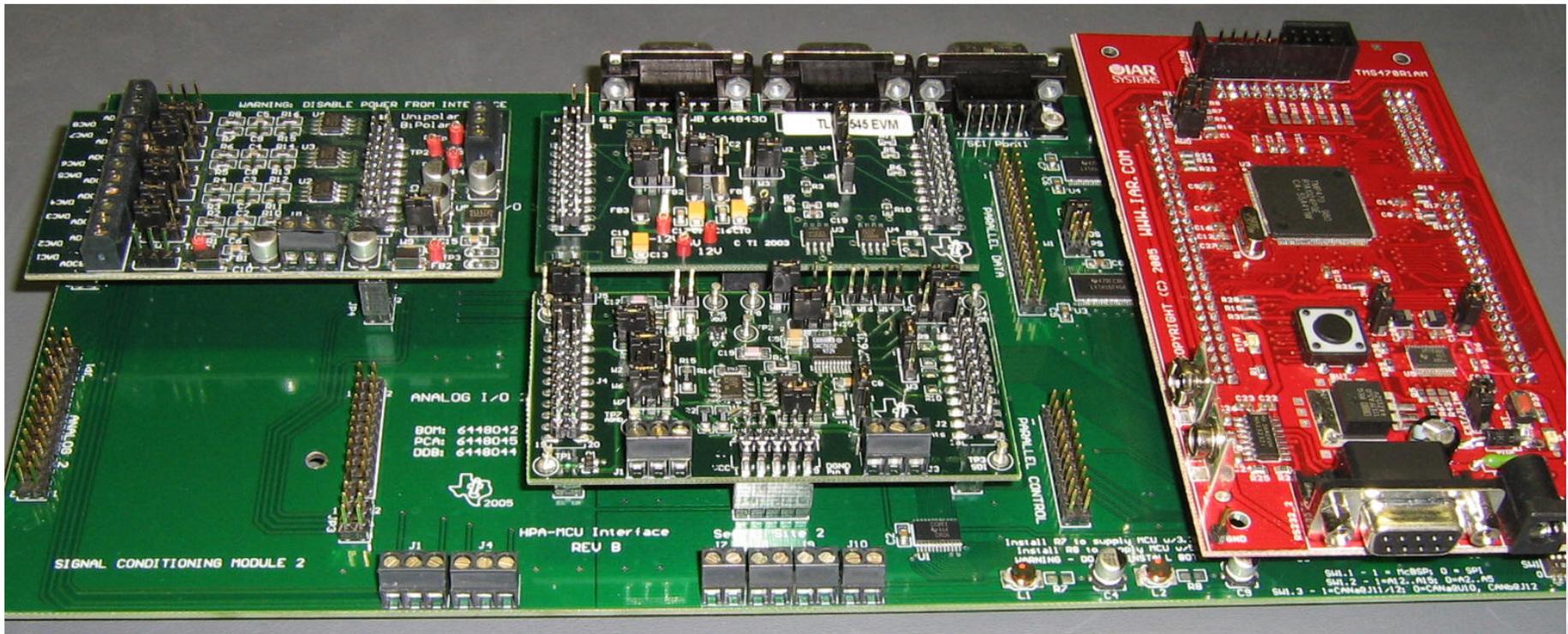


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Technology for Innovators™

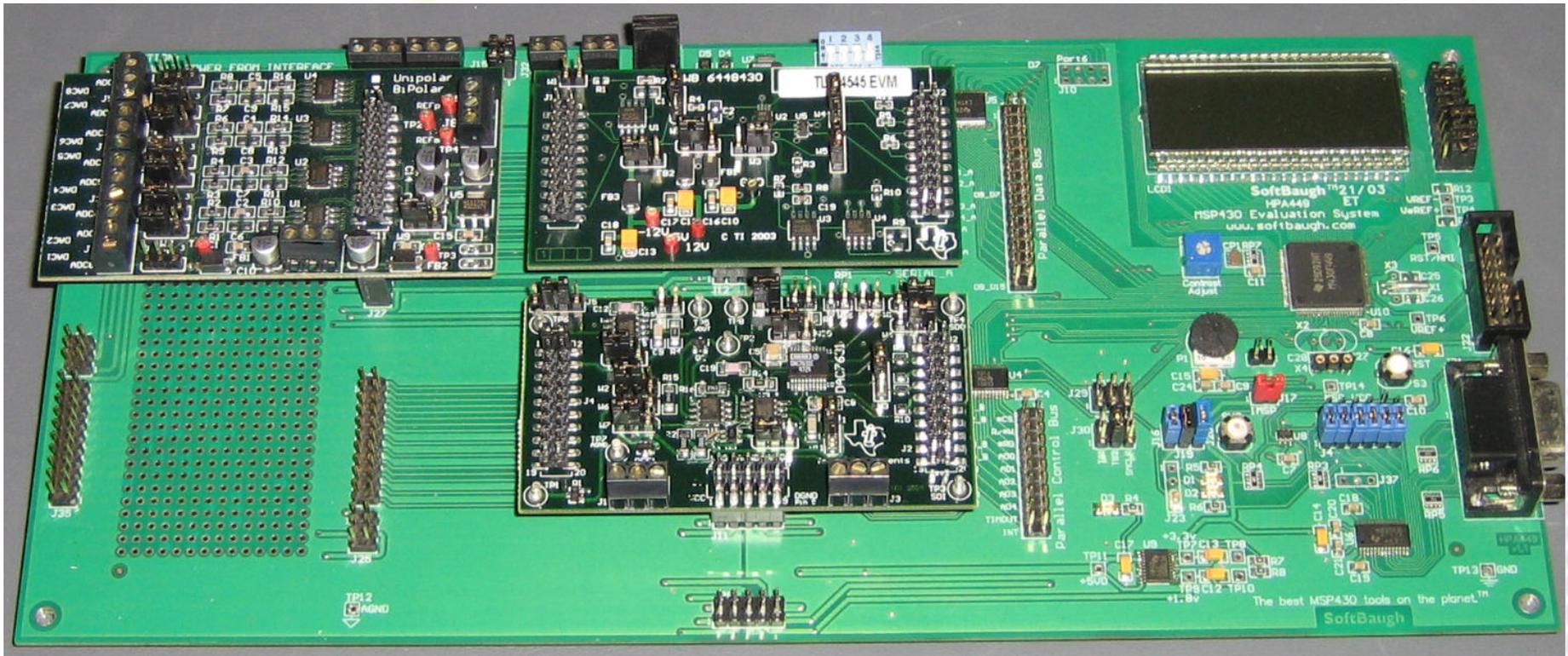
TEXAS INSTRUMENTS

TMS470R1B1M Hardware



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MSP430F449 Hardware



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Evaluating Circuit Sensitivity in Analog Filters

Session ID: S284574
Wednesday at 4:10 p.m.

Mark Fortunato
Texas Instruments, Inc.

Real World Analog Solutions for Your Processor Applications

Session ID: S283976
Thursday at 4:10 p.m.

Bonnie Baker
Texas Instruments, Inc.



Data Converter Basics

How to Use and Test Data-Acquisition Products

Tom Hendrick
t-hendrick@ti.com

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