## AVC

Advanced Very-Low-Voltage CMOS Logic

## Data Book


General Information1
Widebus ${ }^{\text {™ }}$
Widebus ${ }^{\text {TM }}{ }^{\text {M }}$
Application Reports

# AVC Advanced Very-Low-Voltage CMOS Logic Data Book 

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## INTRODUCTION

The new Texas Instruments (TITM) AVC (Advanced Very-Low-Voltage CMOS) logic family provides designers the tools to create tomorrow's advanced high-speed systems with propagation delays of less than 2 ns. Although optimized for $2.5-\mathrm{V}$ systems, AVC logic supports operating voltages between 1.2 V and 3.6 V . The AVC family features TI's new Dynamic Output Control (DOCTM) circuitry, which dynamically lowers circuit output impedance during signal transition for fast rise and fall times, then raises the impedance after signal transition to reduce ringing.

Trends in digital electronics design emphasize lower power consumption, lower supply voltages, faster operating speeds, smaller timing budgets, and heavier loads. Many designs are making the transition from 3.3 V to 2.5 V , with bus speeds increasing beyond 100 MHz . Signal integrity need not be compromised to meet these design requirements. The TI AVC family is designed to meet the needs of these high-speed, low-voltage systems, including next-generation high-performance workstations, PCs, networking servers, and telecommunications switching equipment.
Key features are:

- Sub-2-ns maximum $t_{p d}$ at 2.5 V for AVC16245
- Designed for next-generation, high-performance PCs, workstations, and servers
- DOC circuitry enhances high-speed, low-noise operation.
- Supports mixed-voltage systems
- Optimized for 2.5 V ; operable from 1.2 V to 3.6 V
- Bus-hold option eliminates need for external resistors on unused input pins.
- $\mathrm{I}_{\text {off }}$ supports partial power down.

For more information on these or other TI products, please consult the TI Worldwide Technical Support list in the back of this data book, or visit the TI logic web site at http://www.ti.com/sc/logic.

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INSTRUMENTS

## INTRODUCTION

These symbols, terms, and definitions are in accordance with those currently agreed upon by the JEDEC Council of the Electronic Industries Association (EIA) for use in the USA and by the International Electrotechnical Commission (IEC) for international use.
operating conditions and characteristics (in sequence by letter symbols)
$\mathrm{C}_{\mathbf{i}} \quad$ Input capacitance
The capacitance of an input terminal of the device
$\mathrm{C}_{\mathrm{io}} \quad$ Input/output capacitance
The capacitance of an input/output (I/O) terminal of the device with the input conditions applied that, according to the product specification, establishes the high-impedance state at the output
$C_{0} \quad$ Output capacitance
The capacitance of an output terminal of the device with the input conditions applied that, according to the product specification, establishes the high-impedance state at the output
$\mathrm{C}_{\mathrm{pd}} \quad$ Power dissipation capacitance
Used to determine the no-load dynamic power dissipation per logic function (see individual circuit pages): $P_{D}=C_{p d} V_{C C}{ }^{2} f+I_{C C} V_{C C}$
$f_{\text {max }} \quad$ Maximum clock frequency
The highest rate at which the clock input of a bistable circuit can be driven through its required sequence while maintaining stable transitions of logic level at the output with input conditions established that should cause changes of output logic level in accordance with the specification
$\mathrm{I}_{\mathrm{BH}} \quad$ Bus-hold high sustaining current
The bus-hold circuit can source at least the minimum high sustaining current at $\mathrm{V}_{\mathrm{IH}}$ min. $\mathrm{I}_{\mathrm{BHH}}$ should be measured after raising $\mathrm{V}_{\mathrm{IN}}$ to $\mathrm{V}_{\mathrm{CC}}$ and then lowering it to $\mathrm{V}_{\mathrm{IH}}$ min.
IBHL Bus-hold low sustaining current
The bus-hold circuit can sink at least the minimum low sustaining current at $\mathrm{V}_{\mathrm{IL}}$ max. $\mathrm{I}_{\mathrm{BHL}}$ should be measured after lowering $\mathrm{V}_{\mathrm{IN}}$ to GND and then raising it to $\mathrm{V}_{\mathrm{IL}}$ max.
$\mathrm{I}_{\mathrm{BH}} \quad$ Bus-hold high overdrive current
An external driver must sink at least $I_{\mathrm{BHHO}}$ to switch this node from high to low.
IBHLO Bus-hold low overdrive current
An external driver must source at least $\mathrm{l}_{\mathrm{BHLO}}$ to switch this node from low to high.
ICC Supply current
The current into* the $\mathrm{V}_{\mathrm{CC}}$ supply terminal of an integrated circuit
$\Delta \mathrm{l} C \mathrm{Cupply}$ current change
The increase in supply current for each input that is at one of the specified TTL voltage levels rather than 0 V or $\mathrm{V}_{\mathrm{CC}}$
ICEX Output high leakage current
The maximum leakage current into* an output that is in a high state and $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{CC}}$
$I_{\text {(hold) }} \quad$ Input hold current
The input current that holds the input at the previous state when the driving device goes to the high-impedance state

| $\mathrm{IIH}^{\text {H}}$ | High-level input current |
| :---: | :---: |
|  | The current into* an input when a high-level voltage is applied to that input |
| IIL | Low-level input current |
|  | The current into* an input when a low-level voltage is applied to that input |
| $l_{\text {off }}$ | Input/output power-off leakage current |
|  | The maximum leakage current into* an input or output terminal of the device with the specified voltage applied to the terminal and $\mathrm{V}_{\mathrm{CC}}=0 \mathrm{~V}$ |
| IOH | High-level output current |
|  | The current into* an output with input conditions applied that, according to the product specification, establishes a high level at the output |
| IOHS | Static high-level output current |
|  | The static and testable current into* a Dynamic Output Control (DOC'TM) output with input conditions applied that, according to the product specifications, establishes a static high level at the output. The dynamic drive current is not specified for devices with DOC outputs because of its transient nature; however, it is similar to the dynamic drive current that is available from a high-drive (nondamping resistor) standard-output device. |
| $\mathrm{IOL}^{\text {O }}$ | Low-level output current |
|  | The current into* an output with input conditions applied that, according to the product specification, establishes a low level at the output |
| Iols | Static low-level output current |
|  | The static and testable current into* a Dynamic Output Control (DOC) output with input conditions applied that, according to the product specifications, establishes a static low level at the output. The dynamic drive current is not specified for devices with DOC outputs because of its transient nature; however, it is similar to the dynamic drive current that is available from a high-drive (nondamping resistor) standard-output device. |
| loz | Off-state (high-impedance state) output current (of a 3-state output) |
|  | The current flowing into* an output with the input conditions applied that, according to the product specification, establishes the high-impedance state at the output |
| IOZPD | Power-down off-state (high-impedance state) output current (of a 3-state output) |
|  | The current flowing into* an output that is switched to or held in the high-impedance state as the device is being powered down to $\mathrm{V}_{\mathrm{CC}}=0 \mathrm{~V}$ |
| IozPu | Power-up off-state (high-impedance state) output current (of a 3-state output) |
|  | The current flowing into* an output that is switched to or held in the high-impedance state as the device is being powered up from $\mathrm{V}_{\mathrm{CC}}=0 \mathrm{~V}$ |
| jitter | Jitter |
|  | Dispersion of a time parameter of the pulse waveforms in a pulse train with respect to a reference time, interval, or duration. Unless otherwise specified by a mathematical adjective, peak-to-peak jitter is assumed. |
| jitter(RMS) | RMS jitter |
|  | The root mean square jitter, one-sixth of the maximum peak-to-peak jitter |

SR

The time between the specified reference points on the input and output voltage waveforms with the output changing from one defined level (high or low) to the other defined level ( $t_{\mathrm{pd}}=\mathrm{t}_{\mathrm{PHL}}$ or $\mathrm{t}_{\mathrm{PLH}}$ )
tpHL Propagation delay time, high-to-low level output
The time between the specified reference points on the input and output voltage waveforms with the output changing from the defined high level to the defined low level

## $t_{\text {PHZ }} \quad$ Disable time (of a 3-state output) from high level

The time interval between the specified reference points on the input and the output voltage waveforms with the 3 -state output changing from the defined high level to the high-impedance (off) state

| $t_{\text {PLH }}$ | Propagation delay time, low-to-high level output |
| :---: | :---: |
|  | The time between the specified reference points on the input and output voltage waveforms with the output changing from the defined low level to the defined high level |
| $t_{\text {PLZ }}$ | Disable time (of a 3-state output) from low level |
|  | The time interval between the specified reference points on the input and the output voltage waveforms with the 3 -state output changing from the defined low level to the high-impedance (off) state |
| ${ }^{\text {tPZH }}$ | Enable time (of a 3-state output) to high level |
|  | The time interval between the specified reference points on the input and output voltage waveforms with the 3 -state output changing from the high-impedance (off) state to the defined high level |
| ${ }^{\text {t PZL }}$ | Enable time (of a 3-state output) to low level |
|  | The time interval between the specified reference points on the input and output voltage waveforms with the 3 -state output changing from the high-impedance (off) state to the defined low level |
| $t_{r}$ | Rise time |
|  | The time interval between two reference points ( $10 \%$ and $90 \%$, unless otherwise specified) on a waveform that is changing from the defined low level to the defined high level |
| $\mathrm{t}_{\text {sk( }}$ ( $)$ | Input skew |
|  | The difference between any two propagation delay times that originate at different inputs and terminate at a single output. Input skew describes the ability of a device to manipulate (stretch, shrink, or chop) a clock signal. This is typically accomplished with a multiple-input gate wherein one of the inputs acts as a controlling signal to pass the clock through. $\mathrm{t}_{\text {sk(i) }}$ describes the ability of the gate to shape the pulse to the same duration, regardless of the input used as the controlling input. |
| $\mathrm{t}_{\text {sk(I) }}$ | Limit skew |
|  | The difference between 1) the greater of the maximum specified values of $t_{P L H}$ and $t_{P H L}$ and 2) the lesser of the minimum specified values of $t_{\text {PLH }}$ and $t_{\text {PHL }}$. Limit skew is not directly observed on a device. It is calculated from the data-sheet limits for $t_{\text {PLH }}$ and $t_{\text {PHL }} \cdot \mathrm{t}_{\mathrm{sk}(\mathrm{I})}$ quantifies for the designer how much variation in propagation delay time is induced by operation over the entire ranges of supply voltage, temperature, output load, and other specified operating conditions. Specified as such, $\mathrm{t}_{\text {sk(I) }}$ also accounts for process variation. In fact, all other skew specifications $\left[\mathrm{t}_{\mathrm{sk}(\mathrm{o})}, \mathrm{t}_{\mathrm{sk}(\mathrm{i})}, \mathrm{t}_{\mathrm{sk}(\mathrm{p})}\right.$, and $\left.\mathrm{t}_{\mathrm{sk}(\mathrm{pr})}\right]$ are subsets of $t_{\mathrm{sk}(I)}$; they are never greater than $\mathrm{t}_{\mathrm{sk}(I)}$. |
| $\mathrm{t}_{\mathbf{s k}(0)}$ | Output skew |
|  | The skew between specified outputs of a single logic device with all driving inputs connected together and the outputs switching in the same direction while driving identical specified loads |
| $\mathrm{t}_{\mathbf{s k}(\mathrm{p})}$ | Pulse skew |
|  | The magnitude of the time difference between the propagation delay times, $\mathrm{t}_{\text {PHL }}$ and $\mathrm{t}_{\text {PLH }}$, when a single switching input causes one or more outputs to switch |
| $\mathrm{t}_{\text {sk }}$ (pr) | Process skew |
|  | The magnitude of the difference in propagation delay times between corresponding terminals of two logic devices when both logic devices operate with the same supply voltages, operate at the same temperature, and have identical package styles, identical specified loads, identical internal logic functions, and the same manufacturer |


| tsu |  |
| :--- | :--- |
| $\mathbf{t}_{\mathbf{s u}}$ | Setup time <br> The time interval between the application of a signal at a specified input terminal and a subsequent <br> active transition at another specified input terminal |
| NOTES: 1. The setup time is the actual time interval between two signal events and is determined by |  |
| the system in which the digital circuit operates. A minimum value is specified that is the |  |
| shortest interval for which correct operation of the digital circuit is specified. |  |
| 2. The setup time may have a negative value, in which case the minimum limit defines the |  |
| longest interval (between the active transition and the application of the other signal) for which |  |
| correct operation of the digital circuit is specified. |  |

The following symbols are used in function tables on TI data sheets:

| H | $=$ high level (steady state) |
| :---: | :---: |
| L | $=$ low level (steady state) |
| $\uparrow$ | $=$ transition from low to high level |
| $\downarrow$ | $=$ transition from high to low level |
| $\longrightarrow$ | $=$ value/level or resulting value/level is routed to indicated destination |
| $\sim$ | = value/level is re-entered |
| X | $=$ irrelevant (any input, including transitions) |
| Z | $=$ off (high-impedance) state of a 3-state output |
| a...h | $=$ the level of steady-state inputs A through H , respectively |
| $Q_{0}$ | level of $Q$ before the indicated steady-state input conditions were established |
| $\bar{Q}_{0}$ | $=$ complement of $Q_{0}$ or level of $\bar{Q}$ before the indicated steady-state input conditions were established |
| $Q_{n}$ | $=$ level of Q before the most recent active transition indicated by $\downarrow$ or $\uparrow$ |
| $\Omega$ | $=$ one high-level pulse |
| $\checkmark$ | = one low-level pulse |
| Toggle | $=$ each output changes to the complement of its previous level on each active transition indicated by $\downarrow$ or $\uparrow$ |

If, in the input columns, a row contains only the symbols $\mathrm{H}, \mathrm{L}$, and/or X , this means the indicated output is valid whenever the input configuration is achieved and regardless of the sequence in which it is achieved. The output persists so long as the input configuration is maintained.
If, in the input columns, a row contains $H, L$, and/or $X$ together with $\uparrow$ and/or $\downarrow$, this means the output is valid whenever the input configuration is achieved but the transition(s) must occur following the achievement of the steady-state levels. If the output is shown as a level $\left(H, L, Q_{0}\right.$, or $\left.\bar{Q}_{0}\right)$, it persists so long as the steady-state input levels and the levels that terminate indicated transitions are maintained. Unless otherwise indicated, input transitions in the opposite direction to those shown have no effect at the output. (If the output is shown as a pulse, $\measuredangle$ or $\checkmark$, the pulse follows the indicated input transition and persists for an interval dependent on the circuit.)

## EXPLANATION OF FUNCTION TABLES

Among the most complex function tables are those of the shift registers. These embody most of the symbols used in any of the function tables, plus more. Below is the function table of a 4-bit bidirectional universal shift register.

FUNCTION TABLE

|  |  |  |  | NPUTS |  |  |  |  |  |  | OUT | UTS |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CLEAR |  |  | CLOCK |  | IAL |  | AR | LEL |  |  |  |  |  |
|  | S1 | S0 | CLOCK | LEFT | RIGHT | A | B | C | D | $\mathrm{Q}_{\mathrm{A}}$ | $Q_{B}$ | $Q_{C}$ | $Q_{D}$ |
| L | X | X | X | X | X | X | X | X | X | L | L | L | L |
| H | X | X | L | X | X | X | X | X | X | $Q_{A 0}$ | QB0 | QC0 | QD0 |
| H | H | H | $\uparrow$ | X | X | a | b | c | d | a | b | c | d |
| H | L | H | $\uparrow$ | X | H | H | H | H | H | H | $Q_{\text {An }}$ | $Q_{B n}$ | $Q_{C n}$ |
| H | L | H | $\uparrow$ | X | L | L | L | L | L | L | $Q_{\text {An }}$ | $Q_{B n}$ | $Q_{C n}$ |
| H | H | L | $\uparrow$ | H | X | X | X | X | X | $Q_{B n}$ | $Q_{\text {Cn }}$ | QDn | H |
| H | H | L | $\uparrow$ | L | X | X | X | X | X | $Q_{B n}$ | QCn | QDn | L |
| H | L | L | X | X | X | X | X | X | X | QA0 | QB0 | QC0 | QD0 |

The first line of the table represents a synchronous clearing of the register and says that if clear is low, all four outputs will be reset low regardless of the other inputs. In the following lines, clear is inactive (high) and so has no effect.
The second line shows that so long as the clock input remains low (while clear is high), no other input has any effect and the outputs maintain the levels they assumed before the steady-state combination of clear high and clock low was established. Since on other lines of the table only the rising transition of the clock is shown to be active, the second line implicitly shows that no further change in the outputs occurs while the clock remains high or on the high-to-low transition of the clock.

The third line of the table represents synchronous parallel loading of the register and says that if S1 and S0 are both high then, without regard to the serial input, the data entered at $A$ is at output $Q_{A}$, data entered at $B$ is at $Q_{B}$, and so forth, following a low-to-high clock transition.
The fourth and fifth lines represent the loading of high- and low-level data, respectively, from the shift-right serial input and the shifting of previously entered data one bit; data previously at $Q_{A}$ is now at $Q_{B}$, the previous levels of $Q_{B}$ and $Q_{C}$ are now at $Q_{C}$ and $Q_{D}$, respectively, and the data previously at $Q_{D}$ is no longer in the register. This entry of serial data and shift takes place on the low-to-high transition of the clock when S 1 is low and S 0 is high and the levels at inputs $A$ through $D$ have no effect.

The sixth and seventh lines represent the loading of high- and low-level data, respectively, from the shift-left serial input and the shifting of previously entered data one bit; data previously at $Q_{B}$ is now at $Q_{A}$, the previous levels of $Q_{C}$ and $Q_{D}$ are now at $Q_{B}$ and $Q_{C}$, respectively, and the data previously at $Q_{A}$ is no longer in the register. This entry of serial data and shift takes place on the low-to-high transition of the clock when S1 is high and S0 is low and the levels at inputs A through D have no effect.
The last line shows that as long as both inputs are low, no other input has any effect and, as in the second line, the outputs maintain the levels they assumed before the steady-state combination of clear high and both mode inputs low was established.

The function table functional tests do not reflect all possible combinations or sequential modes.

It is normal TI practice to name the outputs and other inputs of a D-type flip-flop or latch and to draw its logic symbol based on the assumption of true data (D) inputs. Outputs that produce data in phase with the data inputs are called Q and those producing complementary data are called $\overline{\mathrm{Q}}$. An input that causes a Q output to go high or a $\overline{\mathrm{Q}}$ output to go low is called preset (PRE). An input that causes a $\bar{Q}$ output to go high or a Q output to go low is called clear (CLR). Bars are used over these pin names ( $\overline{\mathrm{PRE}}$ and $\overline{\mathrm{CLR}}$ ) if they are active low.
The devices on several data sheets are second-source designs, and the pin-name conventions used by the original manufacturers have been retained. That makes it necessary to designate the inputs and outputs of the inverting circuits $\overline{\mathrm{D}}$ and Q .
In some applications, it may be advantageous to redesignate the data input from $\bar{D}$ to $\overline{\mathrm{D}}$ or vice versa. In that case, all the other inputs and outputs should be renamed as shown below. Also shown are corresponding changes in the graphical symbols. Arbitrary pin numbers are shown.


Flip-Flop


Flip-Flop
The figures show that when Q and $\overline{\mathrm{Q}}$ exchange names, the preset and clear pins also exchange names. The polarity indicators ( $\triangle$ ) on $\overline{\text { PRE }}$ and $\overline{C L R}$ remain, as these inputs are still active low, but the presence or absence of the polarity indicator changes at $D($ or $\bar{D}), Q$, and $\bar{Q}$. Pin $5(Q$ or $\bar{Q})$ is still in phase with the data input ( $D$ or $\bar{D}$ ); their active levels change together.

## THERMAL INFORMATION

In digital-system design, consideration must be given to thermal management of components. The small size of the small-outline packages makes this even more critical. Figures $1-5$ show the high-effect (High-K) thermal resistance for the small-outline 14-, 16-, 20-, 24-, and 48-pin packages for various rates of airflow calculated in accordance with JESD 51-7.

The thermal resistances in Figures 1-5 can be used to approximate typical and maximum virtual junction temperatures. In general, the junction temperature for any device can be calculated using the following equation:

$$
T_{J}=R_{\theta J A} \times P_{T}+T_{A}
$$

where:
$\mathrm{T}_{\mathrm{J}} \quad=$ virtual junction temperature $\left({ }^{\circ} \mathrm{C}\right)$
$\mathrm{R}_{\theta J A}=$ thermal resistance, junction to free air $\left({ }^{\circ} \mathrm{C} / \mathrm{W}\right)$
$\mathrm{P}_{\mathrm{T}} \quad=$ total power dissipation of the device (W)
$\mathrm{T}_{\mathrm{A}}=$ free-air temperature $\left({ }^{\circ} \mathrm{C}\right)$
JUNCTION-TO-AMBIENT THERMAL RESISTANCE
vs
AIR VELOCITY


Figure 1


Figure 2

JUNCTION-TO-AMBIENT THERMAL RESISTANCE
vs AIR VELOCITY


Figure 4

JUNCTION-TO-AMBIENT THERMAL RESISTANCE vs AIR VELOCITY


Figure 3

JUNCTION-TO-AMBIENT THERMAL RESISTANCE VS AIR VELOCITY


Figure 5

## Example:





## 1 Standard Prefix

Example: SNJ - Conforms to MIL-PRF-38535 (QML)

## 2 Temperature Range

Examples: 54 - Military
74 - Commercial

## 3 Family

Examples: Blank - Transistor-Transistor Logic
ABT - Advanced BiCMOS Technology
ABTE - Advanced BiCMOS Technology/
Enhanced Transceiver Logic
AC/ACT - Advanced CMOS Logic
AHC/AHCT - Advanced High-Speed CMOS Logic
ALB - Advanced Low-Voltage BiCMOS
ALS - Advanced Low-Power Schottky Logic
ALVC - Advanced Low-Voltage CMOS Technology
AS - Advanced Schottky Logic
AVC - Advanced Very-Low-Voltage CMOS Logic
BCT - BiCMOS Bus-Interface Technology
CBT - Crossbar Technology
CBTLV - Low-Voltage Crossbar Technology
F - F Logic
FB - Backplane Transceiver Logic/Futurebus+
GTL - Gunning Transceiver Logic
HC/HCT - High-Speed CMOS Logic
HSTL - High-Speed Transceiver Logic
LS - Low-Power Schottky Logic
LV - Low-Voltage CMOS Technology
LVC - Low-Voltage CMOS Technology
LVT - Low-Voltage BiCMOS Technology
S - Schottky Logic
SSTL - Stub Series-Terminated Logic
TVC - Translation Voltage Clamp Logic

## 4 Special Features

Examples: Blank = No Special Features
D - Level-Shifting Diode (CBTD)
H - Bus Hold (ALVCH)
R - Damping Resistor on Inputs/Outputs (LVCR)
S - Schottky Clamping Diode (CBTS)

## 5 Bit Width

Examples: Blank = Gates, MSI, and Octals
1G - Single Gate
8 - Octal IEEE 1149.1 (JTAG)
16 - Widebus ${ }^{\text {TM }}$ ( 16,18 , and 20 bit)
18 - Widebus IEEE 1149.1 (JTAG)
32 - Widebus $+^{\text {TM }}$ (32 and 36 bit)

## 6 Options

Examples: Blank = No Options
2 - Series-Damping Resistor on Outputs
4 - Level Shifter
$25-25-\Omega$ Line Driver

## 7 Function

Examples: 244 - Noninverting Buffer/Driver
374 - D-Type Flip-Flop
573 - D-Type Transparent Latch
640 - Inverting Transceiver

## 8 Device Revision

Examples: Blank = No Revision
Letter Designator A-Z

## 9 Packages

Examples: D, DW - Small-Outline Integrated Circuit (SOIC)
DB, DL - Shrink Small-Outline Package (SSOP)
DBB, DGV - Thin Very Small-Outline Package (TVSOP)
DBQ - Quarter-Size Outline Package (QSOP)
DBV, DCK - Small-Outline Transistor Package (SOT)
DGG, PW - Thin Shrink Small-Outline Package (TSSOP)
FN - Plastic Leaded Chip Carrier (PLCC)
GKE, GKF - MicroStar BGA ${ }^{\text {TM }}$ Low-Profile Fine-Pitch
Ball Grid Array (LFBGA)
N, NP, NT - Plastic Dual-In-Line Package (PDIP)
NS, PS - Small-Outline Package (SOP)
PAG, PAH, PCA, PCB, PM, PN, PZ -
Thin Quad Flatpack (TQFP)
PH, PQ, RC - Quad Flatpack (QFP)

## 10 Tape and Reel

Devices in the DB and PW package types include the R designation for reeled product. Existing product inventory designated LE may remain, but all products are being converted to the R designation.
Examples:
Existing Nomenclature - SN74LVTxxxDBLE
New Nomenclature - SN74LVTxxxADBR
LE - Left Embossed (valid for DB and PW packages only)
$R$ - Standard (valid for all surface-mount packages)
There is no functional difference between LE and R designated products, with respect to the carrier tape, cover tape, or reels used.

## NOTIFICATION OF PACKAGE NOMENCLATURE ALIAS (for Standard Linear and Logic device names of greater than 18 characters)

TI is converting from its current order-entry system to a more advanced system. This conversion requires modifications, both internal and external, to Tl's current business processes. This new system will ultimately provide significant improvements to all facets of Tl's business - from production, to order entry, to logistics. One change required is a limitation of TI part numbers to no more than 18 characters in length. Based on customer inputs, Standard Linear and Logic determined the least disruptive implementations as outlined below:

1. Package alias

TI will use a package alias to denote specific package types for devices currently exceeding 18 characters in length. Table 1 shows a mapping of package codes to an alias single-character representation.

Table 1

| CURRENT <br> PACKAGE <br> CODE | ALIAS |
| :--- | :---: |
| DL | L |
| DGG/DBB | G |
| DGV | V |
| DLR | LR - tape/reel packing |
| DGGR/DBBR | GR - tape/reel packing |
| DGVR | VR - tape/reel packing |

Current: SN74 ALVCH 162269A DGGR
New: SN74 ALVCH 162269A GR

## 2. Resistor-option nomenclature

For devices greater than 18 characters with input and output resistors, TI will adopt a simplified nomenclature to designate the resistor option. This will eliminate the redundant " 2 " (designating output resistors) when the part number also contains an "R" (designating input/output resistors).


Current: SN74 ALVCH R 162245 A
New: SN74 ALVCH R16 245 A
There is no change to the device or data-sheet electrical parameters. The packages involved and the changes in nomenclature are noted in Table 1.

These nomenclature changes are being gradually implemented. The first customer-visible conversions for TI logic devices will be made to data sheets. Over the next few months, TI logic data sheets will be updated. These changes in device nomenclature do not reflect a change in device performance or process characteristics.
General Information
Widebus $^{\text {™ }}$
Widebus ${ }^{\text {TM }}{ }^{\text {M }}$3
Application Reports

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- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Less Than 2-ns Maximum Propagation Delay at $2.5-\mathrm{V}$ and $3.3-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- I Ioff Supports Partial-Power-Down Mode Operation
- ESD Protection Exceeds JESD 22
- 2000-V Human-Body Model (A114-A)
- 200-V Machine Model (A115-A)
- Latch-Up Performance Exceeds 100 mA Per JESD 78, Class II
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{IOL}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 16 -bit buffer/driver is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.
The SN74AVC16244 is designed specifically to improve the performance and density of 3-state memory address drivers, clock drivers, and bus-oriented receivers and transmitters.
The device can be used as four 4-bit buffers, two 8-bit buffers, or one 16-bit buffer. It provides true outputs and symmetrical active-low output-enable ( $\overline{\mathrm{OE} \text { ) inputs. }}$

## description (continued)

To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.
This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off. }}$ The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.
The SN74AVC16244 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments



FUNCTION TABLE
(each 4-bit buffer)

| INPUTS |  | OUTPUT |
| :---: | :---: | :---: |
| $\mathbf{O E}$ | $\mathbf{A}$ |  |
| L | L | L |
| L | $H$ | $H$ |
| $H$ | $X$ | $Z$ |

logic symbol $\dagger$

† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$



Voltage range applied to any output in the high-impedance or power-off state, $\mathrm{V}_{\mathrm{O}}$
(see Note 1)
-0.5 V to 4.6 V
Voltage range applied to any output in the high or low state, $\mathrm{V}_{\mathrm{O}}$
(see Notes 1 and 2) -0.5 V to $\mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}$


Continuous output current, IO . ...................................................................... $\pm 50 \mathrm{~mA}$

Package thermal impedance, $\theta_{\mathrm{JA}}$ (see Note 3): DGG package .................................. $70^{\circ} \mathrm{C} / \mathrm{W}$
DGV package ....................................... $58^{\circ} \mathrm{C} / \mathrm{W}$

$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed.
2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed.
3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage | Operating | 1.4 | 3.6 | V |
|  |  | Data retention only | 1.2 |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
| $\mathrm{V}_{\mathrm{IL}}$ | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\text {CC }}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\text {cC }}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
| $\mathrm{V}_{\mathrm{O}}$ | Output voltage | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | 3-state | 0 | 3.6 |  |
| l OHS | Static high-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
| Iols | Static low-level output current ${ }^{\dagger}$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta v$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| $\mathrm{T}_{\mathrm{A}}$ | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $I_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {M }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{C C}=1.2 \mathrm{~V}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| tpd | A | Y | 3.1 | 0.6 | 3.3 | 0.7 | 2.9 | 0.6 | 1.9 | 0.5 | 1.7 | ns |
| ten | $\overline{\mathrm{OE}}$ | Y | 7.6 | 1.4 | 8 | 1.3 | 6.8 | 0.9 | 4 | 0.7 | 3.5 | ns |
| ${ }^{\text {d }}$ dis | $\overline{\mathrm{OE}}$ | Y | 7.2 | 1.7 | 7.3 | 1.6 | 6.2 | 1 | 4.3 | 1 | 3.5 | ns |

operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{\text {CC }}=1.8 \mathrm{~V}$ | $\mathrm{V}_{C C}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\text {CC }}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  | 23 | 27 | 33 | pF |
|  |  | Outputs disabled | 0.1 | 0.1 |  |  | 0.1 |  |  |

## PARAMETER MEASUREMENT INFORMATION

$\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



VOLTAGE WAVEFORMS
SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES



[^0]NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 2. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{tPZH}$ | GND |



VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control.

Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: PRR $\leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. $t_{P L H}$ and $t_{P H L}$ are the same as $t_{p d}$.

Figure 3. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}_{\mathrm{PZH}}$ | GND |




VOLTAGE WAVEFORMS PULSE DURATION


NOTES: A. CL includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $\quad t P L Z$ and $t_{P H Z}$ are the same as $t_{d i s}$.
F. $t_{P Z L}$ and $t_{P Z H}$ are the same as $t_{e n}$.
G. $\mathrm{t}_{\mathrm{PLH}}$ and tPHL are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $$
\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}
$$



| TEST | S 1 |
| :---: | :---: |
| $\mathrm{t}_{\mathrm{pd}}$ | Open |
| $\mathrm{t}_{\mathrm{PLZ}} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{tPZH}^{2}$ | GND |



VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES

VOLTAGE WAVEFORMS PULSE DURATION


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tphZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {™ }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Less Than 2-ns Maximum Propagation Delay at $2.5-\mathrm{V}$ and $3.3-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- I Ioff Supports Partial-Power-Down Mode Operation
- Bus Hold on Data Inputs Eliminates the Need for External Pullup/Pulldown Resistors
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{l}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{l}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {™ }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 16 -bit buffer/driver is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.

The SN74AVCH16244 is designed specifically to improve the performance and density of 3-state memory address drivers, clock drivers, and bus-oriented receivers and transmitters.

The device can be used as four 4-bit buffers, two 8-bit buffers, or one 16 -bit buffer. It provides true outputs and symmetrical active-low output-enable ( $\overline{\mathrm{OE}})$ inputs.

## description (continued)

To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.
Active bus-hold circuitry is provided to hold unused or floating data inputs at a valid logic level.
This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off }}$. The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.

The SN74AVCH16244 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments



| FUNCTION TABLE <br> (each 4-bit buffer) |  |
| :---: | :---: |
| INPUTS  OUTPUT <br> $\overline{\text { OE }}$ A Y <br> L L L <br> L $H$ H <br> H X $Z$ |  |

## logic symbol $\dagger$


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)


absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$

Input voltage range, $\mathrm{V}_{\mathrm{I}}$ (see Note 1) ........................................................... -0.5 V to 4.6 V
Voltage range applied to any output in the high-impedance or power-off state, $\mathrm{V}_{\mathrm{O}}$
(see Note 1)
-0.5 V to 4.6 V
Voltage range applied to any output in the high or low state, $\mathrm{V}_{\mathrm{O}}$
(see Notes 1 and 2) -0.5 V to $\mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}$



Continuous current through each $\mathrm{V}_{\mathrm{CC}}$ or GND ................................................... $\pm 100 \mathrm{~mA}$
Package thermal impedance, $\theta_{\mathrm{JA}}$ (see Note 3): DGG package ................................... $70^{\circ} \mathrm{C} / \mathrm{W}$
DGV package ........................................ 58º $\mathrm{C} / \mathrm{W}$

$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed.
2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed.
3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)


$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}} \mathrm{vs} \mathrm{IOH}_{\mathrm{O}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {M }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused control inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
$\ddagger$ The bus-hold circuit can sink at least the minimum low sustaining current at $\mathrm{V}_{\mathrm{IL}}$ max. $\mathrm{I}_{\mathrm{BHL}}$ should be measured after lowering $\mathrm{V}_{\text {IN }}$ to $G N D$ and then raising it to $\mathrm{V}_{\text {IL }}$ max.
§ The bus-hold circuit can source at least the minimum high sustaining current at $\mathrm{V}_{\mathrm{IH}}$ min. $\mathrm{I}_{\mathrm{BHH}}$ should be measured after raising $\mathrm{V}_{I N}$ to $\mathrm{V}_{\mathrm{CC}}$ and then lowering it to $\mathrm{V}_{\mathrm{IH}}$ min.
II An external driver must source at least IBHLO to switch this node from low to high.

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | то (OUTPUT) | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN MAX | MIN MAX | MIN MAX | MIN MAX |  |
| tpd | A | Y |  |  |  |  |  | ns |
| ten | $\overline{\mathrm{OE}}$ | Y |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | Y |  |  |  |  |  | ns |

operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{C C}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  |  |  |  | pF |
|  |  | Outputs disabled |  |  |  |  |  |  |  |

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{C}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{tpd}_{\mathrm{pd}} \\ \mathrm{tpLZ}^{\prime / \mathrm{tPZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{tpZH}^{2} \end{gathered}$ | $\begin{gathered} \text { Open } \\ 2 \times V_{\text {CC }} \\ \text { GND } \end{gathered}$ |




VOLTAGE WAVEFORMS PULSE DURATION


[^1]NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{tr}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.

Figure 2. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION

$$
V_{C C}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}_{\text {PZH }}$ | GND |



VOLTAGE WAVEFORMS
SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{tr}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 3. Load Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\mathrm{pd}}$ | Open |
| $\mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\mathrm{PZL}}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |



VOLTAGE WAVEFORMS PULSE DURATION


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: PRR $\leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. $t_{P L H}$ and $t_{P H L}$ are the same as $t_{p d}$.

Figure 4. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| ${ }^{\text {tpd }}$ | Open |
| ${ }^{\text {tPLZ }}$ /tPZL | $2 \times V_{C C}$ |
| tPHZ/tPZH | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS PROPAGATION DELAY TIMES

VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{tf}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. $t_{P L H}$ and $t_{P H L}$ are the same as $t_{p d}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Less Than 2-ns Maximum Propagation Delay at $2.5-\mathrm{V}$ and $3.3-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- I Ioff Supports Partial-Power-Down Mode Operation
- ESD Protection Exceeds JESD 22
- 2000-V Human-Body Model (A114-A)
- 200-V Machine Model (A115-A)
- Latch-Up Performance Exceeds 250 mA Per JESD 78
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{IOL}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 16 -bit (dual octal) noninverting bus transceiver is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.
The SN74AVC16245 is designed for asynchronous communication between data buses. The control-function implementation minimizes external timing requirements.
This device can be used as two 8 -bit transceivers or one 16 -bit transceiver. It allows data transmission from the $A$ bus to the $B$ bus or from the $B$ bus to the $A$ bus, depending on the logic level at the direction-control (DIR) input. The output-enable ( $\overline{\mathrm{OE}}$ ) input can be used to disable the device so that the buses are effectively isolated.

## description (continued)

To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.
This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off. }}$ The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.
The SN74AVC16245 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments

DGG OR DGV PACKAGE
(TOP VIEW)

## logic symbol $\dagger$


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.
logic diagram (positive logic)


To Seven Other Channels


To Seven Other Channels

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$

| Supply voltage range, $\mathrm{V}_{\mathrm{CC}}$ |  |
| :---: | :---: |
| Input voltage range, $\mathrm{V}_{\mathrm{I}}$ (see Note 1) | -0 |
| Voltage range applied to any input/output |  |
| Voltage range applied to any input/output when the output is in the high or low state, $\mathrm{V}_{\mathrm{O}}$ (see Notes 1 and 2) | -0.5 V to $\mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}$ |
|  |  |
| Output clamp current, $\mathrm{I}_{\mathrm{OK}}\left(\mathrm{V}_{\mathrm{O}}<0\right)$ |  |
| Continuous output current, $\mathrm{I}_{0}$ |  |
| Continuous current through each $\mathrm{V}_{\text {CC }}$ or GND ........................................... $\pm 100 \mathrm{~mA}$ |  |
| Package thermal impedance, $\theta_{\mathrm{JA}}$ (see Note 3): DGG package DGV package |  |
|  |  |
|  |  |
| esses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and ctional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not lied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability. |  |
| ES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed. <br> 2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed. <br> 3. The package thermal impedance is calculated in accordance with JESD 51 |  |
|  |  |

$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed.
2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed
3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)


$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}} \mathrm{vs} \mathrm{IOH}_{\mathrm{O}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
$\ddagger$ For I/O ports, the parameter loz includes the input leakage current.
switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| $t_{\text {pd }}$ | A or B | B or A | 3.9 | 0.8 | 4 | 0.7 | 3 | 0.6 | 1.9 | 0.5 | 1.7 | ns |
| ten | $\overline{\mathrm{OE}}$ | A or B | 8.4 | 1.5 | 9.2 | 1.4 | 7 | 1 | 4.3 | 0.7 | 3.7 | ns |
| ${ }^{\text {dis }}$ | $\overline{\mathrm{OE}}$ | A or B | 8.4 | 2.3 | 9.3 | 2.2 | 7 | 1.1 | 4 | 1.2 | 3.9 | ns |

operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{\text {CC }}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\text {CC }}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\text {pd }}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  | 35 | 38 | 44 | pF |
|  |  | Outputs disabled | 6 | 6 |  |  | 7 |  |  |

## PARAMETER MEASUREMENT INFORMATION

$\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |



VOLTAGE WAVEFORMS
SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES
NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tPLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tPZL and tpZH are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 2. Load Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| tpd | Open |
| tPLZ/tpzL | $2 \times \mathrm{V}_{\text {C }}$ |
| ${ }_{\text {tPHZ }} /{ }^{\text {P/PZH}}$ | GND |



NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 3. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



VOLTAGE WAVEFORMS
SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES
NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tphZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tPZL and tpZH are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION

$$
\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| tpLZ/tPZL | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


## VOLTAGE WAVEFORMS

 ENABLE AND DISABLE TIMESNOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tpHZ are the same as $t_{\text {dis }}$.
F. tpZL and tpZH are the same as ten.
G. $\mathrm{tPLH}^{\text {and }} \mathrm{tPHL}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {™ }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Less Than 2-ns Maximum Propagation Delay at $2.5-\mathrm{V}$ and $3.3-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- I ${ }_{\text {off }}$ Supports Partial-Power-Down Mode Operation
- Bus Hold on Data Inputs Eliminates the Need for External Pullup/Pulldown Resistors
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {¹ }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 16 -bit (dual-octal) noninverting bus transceiver is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.

The SN74AVCH16245 is designed for asynchronous communication between data buses. The control-function implementation minimizes external timing requirements.

This device can be used as two 8-bit transceivers or one 16 -bit transceiver. It allows data transmission from the A bus to the $B$ bus or from the $B$ bus to the $A$ bus, depending on the logic level at the direction-control (DIR) input. The output-enable ( $\overline{\mathrm{OE}}$ ) input can be used to disable the device so that the buses are effectively isolated.

## description (continued)

To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.
Active bus-hold circuitry is provided to hold unused or floating data inputs at a valid logic level.
This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off. }}$ The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.

The SN74AVCH16245 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments



| FUNCTION TABLE <br> (each 8-bit transceiver) |  |
| :---: | :---: |
| INPUTS  OPERATION <br> OE DIR  <br> L L B data to A bus <br> L H A data to B bus <br> H X Isolation |  |

## logic symbol $\dagger$


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.
logic diagram (positive logic)


To Seven Other Channels


To Seven Other Channels

# absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$ 

| Supply voltage range, $\mathrm{V}_{\mathrm{CC}}$ | -0.5 V to 4.6 V |
| :---: | :---: |
| Input voltage range, $\mathrm{V}_{\mathrm{I}}$ : Except I/O ports (see Note 1) | -0.5 V to 4.6 V |
| I/O ports (see Notes 1 and 2) | -0.5 V to $\mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}$ |
| Voltage range applied to any input/output |  |
| when the output is in the high-impedance or power-off state, $\mathrm{V}_{\mathrm{O}}$ (see Note 1 ) | -0.5 V to 4.6 V |
| Voltage range applied to any input/output |  |
| when the output is in the high or low state, $\mathrm{V}_{\mathrm{O}}$ (see Notes 1 and 2) | -0.5 V to $\mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}$ |
| Input clamp current, $\mathrm{I}_{\mathrm{I}}\left(\mathrm{V}_{\mathrm{I}}<0\right)$ | -50 mA |
| Output clamp current, $\mathrm{I}_{\mathrm{OK}}\left(\mathrm{V}_{\mathrm{O}}<0\right)$ | -50 mA |
| Continuous output current, lo | $\pm 50 \mathrm{~mA}$ |
| Continuous current through each $\mathrm{V}_{\text {CC }}$ or GND | $\pm 100 \mathrm{~mA}$ |
| Package thermal impedance, $\theta_{\text {JA }}$ (see Note 3): DGG package | $70^{\circ} \mathrm{C} / \mathrm{W}$ |
| DGV package | $58^{\circ} \mathrm{C} / \mathrm{W}$ |
| Storage temperature range, $\mathrm{T}_{\text {stg }}$ | $-65^{\circ} \mathrm{C}$ to $150^{\circ} \mathrm{C}$ |

$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed.
2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed.
3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $V_{C C}$ | Supply voltage | Operating | 1.4 | 3.6 | V |
|  |  | Data retention only | 1.2 |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
| $\mathrm{V}_{\text {IL }}$ | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\text {CC }}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
| $\mathrm{V}_{\mathrm{O}}$ | Output voltage | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | 3-state | 0 | 3.6 |  |
| IOHS | Static high-level output current ${ }^{\dagger}$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
| IoLs | Static low-level output current ${ }^{\dagger}$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta \mathrm{v}$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| TA | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}} \mathrm{vs} \mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {™ }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused control inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
$\ddagger$ The bus-hold circuit can sink at least the minimum low sustaining current at $\mathrm{V}_{I L}$ max. $\mathrm{I}_{\mathrm{BH}}$ should be measured after lowering $\mathrm{V}_{\text {IN }}$ to $G N D$ and then raising it to $\mathrm{V}_{\mathrm{IL}}$ max.
§ The bus-hold circuit can source at least the minimum high sustaining current at $\mathrm{V}_{\mathrm{IH}}$ min. $\mathrm{I}_{\mathrm{BH}}$ should be measured after raising $\mathrm{V}_{\mathrm{IN}}$ to $\mathrm{V}_{\mathrm{CC}}$ and then lowering it to $\mathrm{V}_{\mathrm{IH}}$ min.
II An external driver must source at least IBHLO to switch this node from low to high.
\# An external driver must sink at least $\mathrm{I}_{\mathrm{BH}} \mathrm{BO}$ to switch this node from high to low.
II For I/O ports, the parameter IOZ includes the input leakage current.
switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| $t_{\text {pd }}$ | A or B | B or A |  |  |  |  |  |  |  |  |  | ns |
| ten | $\overline{\mathrm{OE}}$ | A or B |  |  |  |  |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | A or B |  |  |  |  |  |  |  |  |  | ns |

operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C L=0$, | $f=10 \mathrm{MHz}$ |  |  |  | pF |
|  |  | Outputs disabled |  |  |  |  |  | F |

# PARAMETER MEASUREMENT INFORMATION <br> $$
\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V} \text { AND } 1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}
$$ 



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\mathbf{t} \text { tPZL }}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{tpZH}^{2}$ | GND |



NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tPLZ and $\mathrm{tPHZ}^{2}$ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tpZH are the same as ten.
G. $\operatorname{tPLH}$ and PPHL are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 2. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION

$$
V_{C C}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



VOLTAGE WAVEFORMS
SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 3. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| tpd | Open |
| tpLz/tPZL | $2 \times \mathrm{V}$ C |
| tPHZ/tPZH | GND |



Figure 4. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $t_{\text {pd }}$ | Open |
| tPLZ ${ }^{\text {/ }}$ PZL | $2 \times \mathrm{V}_{\text {c }}$ |
| tPHZ/tPZH | GND |



VOLTAGE WAVEFORMS PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. tPLL and $\mathrm{tPHL}^{\text {are }}$ the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- Ioff Supports Partial-Power-Down Mode Operation
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{IOL}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 12 -bit to 24 -bit registered bus exchanger is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.
The SN74AVC16269 is used in applications in which two separate ports must be multiplexed onto, or demultiplexed from, a single port. The device is particularly suitable as an interface between synchronous DRAMs and high-speed microprocessors.
Data is stored in the internal B-port registers on the low-to-high transition of the clock (CLK) input when the appropriate clock-enable (CLKENA) inputs are low. Proper control of these inputs allows two sequential 12-bit words to be presented as a 24 -bit word on the B port. For data transfer in the B-to-A direction, a single storage register is provided. The select ( $\overline{\mathrm{SEL}}$ ) line selects 1B or 2B data for the A outputs. The register on the A output permits the fastest possible data transfer, thus extending the period during which the data is valid on the bus.

DOC, EPIC, and Widebus are trademarks of Texas Instruments Incorporated.

## description (continued)

The control terminals are registered so that all transactions are synchronous with CLK. Data flow is controlled by the active-low output enables ( $\overline{\mathrm{OEA}}, \overline{\mathrm{OEB1}}, \overline{\mathrm{OEB2}}$ ).
To ensure the high-impedance state during power up or power down, a clock pulse should be applied as soon as possible, and $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver. Due to $\overline{\mathrm{OE}}$ being routed through a register, the active state of the outputs cannot be determined prior to the arrival of the first clock pulse.

This device is fully specified for partial-power-down applications using $\mathrm{l}_{\text {off }}$. The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.

The SN74AVC16269 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments



## Function Tables

OUTPUT ENABLE

| INPUTS |  |  | OUTPUTS |  |
| :---: | :---: | :---: | :---: | :---: |
| CLK | $\overline{\text { OEA }}$ | OEB | A | 1B, 2B |
| $\uparrow$ | H | H | Z | Z |
| $\uparrow$ | H | L | Z | Active |
| $\uparrow$ | L | H | Active | Z |
| $\uparrow$ | L | L | Active | Active |


| A-TO-B STORAGE $(\overline{\text { OEB }} \mathbf{~} \mathbf{L})$ |  |  |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: |
| INPUTS    OUTPUTS  <br> CLKENA1 CLKENA2 CLK A 1 B 2B <br> H H X X $1 \mathrm{~B}_{0}{ }^{\dagger}$ $2 \mathrm{~B}_{0}{ }^{\dagger}$ <br> L X $\uparrow$ L L X <br> L X $\uparrow$ H H X <br> X L $\uparrow$ L X L <br> X L $\uparrow$ H X H |  |  |  |  |  |

†Output level before the indicated steady-state input conditions were established

| B-TO-A STORAGE ( $\overline{\text { OEA }}=\mathrm{L}$ ) |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| INPUTS |  |  |  | OUTPUT <br> A |
| CLK | SEL | 1B | 2B |  |
| X | H | X | X | $\mathrm{A}_{0}{ }^{\dagger}$ |
| X | L | X | X | $\mathrm{A}_{0} \dagger$ |
| $\uparrow$ | H | L | X | L |
| $\uparrow$ | H | H | X | H |
| $\uparrow$ | L | X | L | L |
| $\uparrow$ | L | X | H | H |

[^2]logic diagram (positive logic)


# absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$ 



## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage | Operating | 1.4 | 3.6 | V |
|  |  | Data retention only | 1.2 |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
| VIL | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
| $\mathrm{V}_{\mathrm{O}}$ | Output voltage | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | 3-state | 0 | 3.6 |  |
| IOHS | Static high-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
| IoLs | Static low-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta v$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| TA | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{VOH}_{\mathrm{OH}} \mathrm{IOH}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
$\ddagger$ For I/O ports, the parameter loz includes the input leakage current.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | $\begin{gathered} \text { TO } \\ \text { (OUTPUT) } \end{gathered}$ | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{VCC}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| $\mathrm{f}_{\text {max }}$ |  |  |  |  |  |  |  |  |  |  |  | MHz |
| ${ }^{\text {tpd }}$ | CLK | B |  |  |  |  |  |  |  |  |  | ns |
|  |  | A |  |  |  |  |  |  |  |  |  |  |
| ten | CLK | B |  |  |  |  |  |  |  |  |  | ns |
|  |  | A |  |  |  |  |  |  |  |  |  |  |
| ${ }^{\text {d }}$ dis | CLK | B |  |  |  |  |  |  |  |  |  | ns |
|  |  | A |  |  |  |  |  |  |  |  |  |  |

switching characteristics, $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{C}_{\mathrm{L}}=0 \mathrm{pF} \dagger$

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN MAX |  |
| ${ }^{\text {tpd }}$ | CLK | B |  | ns |
|  |  | A |  |  |

† Texas Instruments SPICE simulation data
operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{\text {CC }}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\text {CC }}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  |  |  |  | pF |
|  |  | Outputs disabled |  |  |  |  |  |  |  |

## PARAMETER MEASUREMENT INFORMATION

$\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| ${ }^{\text {tpd }}$ | Open |
| tPLZ $/$ PPZL | $2 \times \mathrm{V}$ C |
| ${ }_{\text {tPHZ }} /{ }^{\text {P }}$ PZH | GND |

Timing
Input


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES

VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{tr}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{d i s}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 2. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{t}_{\mathrm{pd}} \\ \mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\mathrm{PZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}} \end{gathered}$ | Open $2 \times V_{C C}$ GND |



VOLTAGE WAVEFORMS
PULSE DURATION


[^3]NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tphZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 3. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $$
\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES

VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION

$$
\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}
$$




Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- I ${ }_{\text {off }}$ Supports Partial-Power-Down Mode Operation
- Ideal for Use in PC133 Registered DIMM Applications
- ESD Protection Exceeds JESD 22
- 2000-V Human-Body Model (A114-A)
- 200-V Machine Model (A115-A)
- Latch-Up Performance Exceeds 100 mA Per JESD 78, Class II
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{IOL}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 16 -bit universal bus driver is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to 3.6-V VCC operation.

Data flow from A to Y is controlled by the output-enable ( $\overline{\mathrm{OE}})$ input. The device operates in the transparent mode when the latch-enable ( $\overline{\mathrm{LE}}$ ) input is low. When $\overline{\mathrm{LE}}$ is high, the A data is latched if the clock (CLK) input is held at a high or low logic level. If $\overline{\mathrm{LE}}$ is high, the A data is stored in the latch/flip-flop on the low-to-high transition of CLK. When $\overline{\mathrm{OE}}$ is high, the outputs are in the high-impedance state.
To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{C}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.

## description (continued)

This device is fully specified for partial-power-down applications using $I_{\text {off }}$ The $I_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.
The SN74AVC16334 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments

| dgG or dg package (TOP VIEW) |  |
| :---: | :---: |
|  |  |
|  |  |
| 2 | 47 A1 |
| 3 | 46 A2 |
| ND 4 | 45 GN |
| 5 | 44 |
| [6 | 43 A4 |
| $\mathrm{V}_{\mathrm{CC}}{ }^{7}$ | ${ }^{42} \mathrm{~V}_{\mathrm{C}}$ |
| 8 | 41. A5 |
| Y6 9 | 40 A6 |
| GND 10 | $39]$ GN |
| 11 | 38 A7 |
| 12 | 37 A8 |
| 13 | A9 |
| 0[14 | 5 A10 |
| GND 15 | $34]$ GND |
| $1{ }^{16}$ | 33 A11 |
| 12.17 | $32]$ A12 |
| $\mathrm{V}_{\mathrm{CC}} 18$ | $31 . \mathrm{V}_{\mathrm{CC}}$ |
| Y13 19 | 30 A13 |
| Y14 20 | 29 A14 |
| GND 21 | 28 GND |
| Y15 22 | 27 A15 |
| Y16 23 | 26 A16 |
| [24 | $\sqrt{\overline{L E}}$ |

NC - No internal connection

FUNCTION TABLE
(each universal bus driver)

| INPUTS |  |  |  | $\begin{aligned} & \text { OUTPUT } \\ & \mathrm{Y} \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\overline{\mathrm{OE}}$ | $\overline{\text { LE }}$ | CLK | A |  |
| H | X | X | X | Z |
| L | L | X | L | L |
| L | L | X | H | H |
| L | H | $\uparrow$ | L | L |
| L | H | $\uparrow$ | H | H |
| L | H | L or H | X | $Y_{0}{ }^{\dagger}$ |

$\dagger$ Output level before the indicated steady-state input conditions were established

## logic symbol $\dagger$


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.
logic diagram (positive logic)


To 15 Other Channels

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$


$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed.
2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed
3. The package thermal impedance is calculated in accordance with JESD 51.

INSTRUMENTS

## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage | Operating | 1.4 | 3.6 | V |
|  |  | Data retention only | 1.2 |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
| $\mathrm{V}_{\mathrm{IL}}$ | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\text {CC }}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\text {cC }}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
| $\mathrm{V}_{\mathrm{O}}$ | Output voltage | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | 3-state | 0 | 3.6 |  |
| l OHS | Static high-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
| Iols | Static low-level output current ${ }^{\dagger}$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta v$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| $\mathrm{T}_{\mathrm{A}}$ | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $I_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC'M) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | то (OUTPUT) | $\mathrm{V}_{C C}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| ${ }_{\text {max }}$ |  |  |  |  |  | 150 |  | 150 |  | 150 |  | MHz |
| ${ }^{\text {tpd }}$ | A | Y | 5.3 | 1.2 | 6.2 | 1.5 | 4.9 | 1 | 3.2 | 0.9 | 2.5 | ns |
|  | $\overline{\text { LE }}$ |  | 7 | 2.2 | 9.7 | 1.8 | 7.5 | 1.5 | 4.9 | 0.8 | 4 |  |
|  | CLK |  | 6 | 1.9 | 7.8 | 1.6 | 6 | 1.1 | 3.7 | 1 | 3.1 |  |
| ten | OE | Y | 7.9 | 2.4 | 10.2 | 1.6 | 8.8 | 1.5 | 6.7 | 1 | 6.2 | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | Y | 7.7 | 2.1 | 10.3 | 1.5 | 8.4 | 1.2 | 5.3 | 1 | 5.3 | ns |

switching characteristics, $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{C}_{\mathrm{L}}=0 \mathrm{pF} \dagger$

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\begin{gathered} \hline \mathrm{VCC}=3.3 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | MAX |  |
| ${ }^{\text {tpd }}$ | A | Y | 0.6 | 1.3 | ns |
|  | CLK |  | 0.7 | 1.5 |  |

${ }^{\dagger}$ Texas Instruments SPICE simulation data
operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{\text {CC }}=1.8 \mathrm{~V}$ | $\mathrm{V}_{C C}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\text {CC }}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\text {pd }}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  | 45 | 48 | 52 | pF |
|  |  | Outputs disabled | 23 | 25 |  |  | 28 |  |  |

PARAMETER MEASUREMENT INFORMATION

$$
\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V} \text { AND } 1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{t}_{\mathrm{pd}} \\ \mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\mathrm{PZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}} \end{gathered}$ | Open $2 \times V_{C C}$ GND |



VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS
PULSE DURATION


[^4]NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: PRR $\leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $\mathrm{tPZH}^{2}$ are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 2. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\mathrm{pd}}$ | Open |
| $\mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\text {PZL }}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |

LOAD CIRCUIT


VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis. }}$
F. tpZL and tPZH are the same as ten.
G. $\mathrm{t}_{\mathrm{PLH}}$ and $\mathrm{t}_{\mathrm{PHL}}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 3. Load Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| tPLZ $^{\text {t PZL }}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{tPZH}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS PROPAGATION DELAY TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: PRR $\leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. $\quad$ tPLH and $t_{P H L}$ are the same as $t_{p d}$.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| tPLZ $^{\text {t }}$ PZL | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{tf}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. $t_{P L H}$ and $t_{P H L}$ are the same as $t_{p d}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {™ }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- $\mathrm{I}_{\text {off }}$ Supports Partial-Power-Down Mode Operation
- ESD Protection Exceeds JESD 22
- 2000-V Human-Body Model (A114-A)
- 200-V Machine Model (A115-A)
- Latch-Up Performance Exceeds 100 mA Per JESD 78, Class II
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 16 -bit transparent D -type latch is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.
The SN74AVC16373 is particularly suitable for implementing buffer registers, I/O ports, bidirectional bus drivers, and working registers. This device can be used as two 8 -bit latches or one 16 -bit latch. When the latch-enable (LE) input is high, the Q outputs follow the data (D) inputs. When LE is taken low, the Q outputs are latched at the levels set up at the D inputs.

## description (continued)

A buffered output-enable ( $\overline{\mathrm{OE}}$ ) input can be used to place the eight outputs in either a normal logic state (high or low logic levels) or the high-impedance state. In the high-impedance state, the outputs neither load nor drive the bus lines significantly. The high-impedance state and the increased drive provide the capability to drive bus lines without need for interface or pullup components. $\overline{\text { OE }}$ does not affect internal operations of the latch. Old data can be retained or new data can be entered while the outputs are in the high-impedance state.

To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.

This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off }}$ The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.

The SN74AVC16373 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments

| DGG ORD (TO | DGV PACKAGE P VIEW) |
| :---: | :---: |
| $1 \overline{O E}[1$ | $\cup_{48} 1 \mathrm{LE}$ |
| 1Q1 ${ }^{\text {d }}$ | 47 1D1 |
| 1Q2 3 | 46 1D2 |
| GND [4 | 45 GND |
| 1Q3 [5 | 44 1D3 |
| 1Q4 6 | 43 1D4 |
| $\mathrm{V}_{\mathrm{CC}}[7$ | $42] \mathrm{V}_{\mathrm{CC}}$ |
| 105 8 | 41 1D5 |
| 1Q6 [9 | 40 1D6 |
| GND 10 | 39 GND |
| 1 Q7 11 | 38 1D7 |
| 108 12 | 37 108 |
| 2Q1 13 | 36 2D1 |
| 2Q2 14 | 35 2D2 |
| GND 15 | 34 GND |
| 2Q3 16 | 33 2D3 |
| 2Q4 17 | $32] 2 \mathrm{~L}$ |
| $\mathrm{V}_{\mathrm{CC}}[18$ | $31] \mathrm{V}_{\mathrm{CC}}$ |
| 2Q5 19 | 30 2D5 |
| 2 Q 6 [20 | 29 206 |
| GND 21 | 28 GND |
| 2Q7 [22 | 27 2D7 |
| 2Q8 [23 | 26 2D8 |
| $2 \overline{\mathrm{OE}}$ [24 | 25 2LE |


| FUNCTION TABLE (each 8-bit latch) |  |  |  |
| :---: | :---: | :---: | :---: |
| INPUTS |  |  | OUTPUT |
| $\overline{\mathrm{OE}}$ | LE | D | Q |
| L | H | H | H |
| L | H | L | L |
| L | L | X | $Q_{0}$ |
| H | X | X | Z |

## logic symbol $\dagger$


$\dagger$ This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.
logic diagram (positive logic)


To Seven Other Channels

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$


$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed.
2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed
3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage | Operating | 1.4 | 3.6 | V |
|  |  | Data retention only | 1.2 |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
| $\mathrm{V}_{\mathrm{IL}}$ | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\text {CC }}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\text {cC }}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
| $\mathrm{V}_{\mathrm{O}}$ | Output voltage | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | 3-state | 0 | 3.6 |  |
| IOHS | Static high-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
| IoLs | Static low-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta \mathrm{v}$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| $\mathrm{T}_{\mathrm{A}}$ | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $I_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC' ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

| PARAMETER |  | TEST CONDITIONS |  | $\mathrm{V}_{\mathrm{CC}}$ | MIN TYP $\dagger$ | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{OH}}$ |  | $\mathrm{IOHS}=-100 \mu \mathrm{~A}$ |  | 1.4 V to 3.6 V | $\mathrm{V}_{\mathrm{CC}}-0.2$ |  | V |
|  |  | IOHS $=-2 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=0.91 \mathrm{~V}$ | 1.4 V | 1.05 |  |  |
|  |  | $\mathrm{I} \mathrm{OHS}=-4 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=1.07 \mathrm{~V}$ | 1.65 V | 1.2 |  |  |
|  |  | $\mathrm{I} \mathrm{OHS}=-8 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=1.7 \mathrm{~V}$ | 2.3 V | 1.75 |  |  |
|  |  | l OHS $=-12 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=2 \mathrm{~V}$ | 3 V | 2.3 |  |  |
| VOL |  | IOLS $=100 \mu \mathrm{~A}$ |  | 1.4 V to 3.6 V |  | 0.2 | V |
|  |  | IOLS $=2 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.49 \mathrm{~V}$ | 1.4 V |  | 0.4 |  |
|  |  | l OLS $=4 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.57 \mathrm{~V}$ | 1.65 V |  | 0.45 |  |
|  |  | IOLS = 8 mA , | $\mathrm{V}_{\mathrm{IL}}=0.7 \mathrm{~V}$ | 2.3 V |  | 0.55 |  |
|  |  | IOLS $=12 \mathrm{~mA}$, | $\mathrm{V}_{\text {IL }}=0.8 \mathrm{~V}$ | 3 V |  | 0.7 |  |
| 1 | Control inputs | $\mathrm{V}_{1}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 3.6 V |  | $\pm 2.5$ | $\mu \mathrm{A}$ |
| $l_{\text {off }}$ |  | $\mathrm{V}_{\text {I }}$ or $\mathrm{V}_{\mathrm{O}}=3.6 \mathrm{~V}$ |  | 0 |  | $\pm 10$ | $\mu \mathrm{A}$ |
| IOZ |  | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\text {CC }}$ or GND |  | 3.6 V |  | $\pm 10$ | $\mu \mathrm{A}$ |
| ${ }^{\text {ICC }}$ |  | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND, |  | 3.6 V |  | 40 | $\mu \mathrm{A}$ |
| $\mathrm{C}_{i}$ | Control inputs | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V | 3 |  | pF |
|  |  |  |  | 3.3 V | 3 |  |  |
|  | Data inputs | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V | 2.5 |  |  |
|  |  |  |  | 3.3 V | 2.5 |  |  |
| $\mathrm{C}_{0}$ | Outputs | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V | 6.5 |  | pF |
|  |  |  |  | 3.3 V | 6.5 |  |  |

$\dagger$ Typical values are measured at $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ and $3.3 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

|  |  | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{VCC}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| $t_{w}$ | Pulse duration, LE high |  |  |  |  | 2.2 |  | 2 |  | 1.8 |  | ns |
| $\mathrm{t}_{\text {su }}$ | Setup time, data before LE $\downarrow$ | 1.7 |  | 1.2 |  | 1.1 |  | 0.9 |  | 0.8 |  | ns |
| th | Hold time, data after LE $\downarrow$ | 2 |  | 1.1 |  | 1.1 |  | 1.1 |  | 1 |  | ns |

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{C C}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{VCC}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{V} \mathrm{CC}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| ${ }^{\text {tpd }}$ | D | Q | 5.8 | 1.2 | 6.8 | 1 | 5.7 | 0.8 | 3.3 | 0.7 | 2.8 | ns |
|  | LE |  | 7.2 | 1.4 | 8.3 | 1.1 | 6.6 | 0.8 | 4 | 0.7 | 3.2 |  |
| ten | $\overline{\mathrm{OE}}$ | Q | 7.4 | 1.6 | 8.8 | 1.6 | 6.7 | 1.4 | 4.3 | 0.7 | 3.4 | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | Q | 8.4 | 2.5 | 9.4 | 2.3 | 7.8 | 1.3 | 4.2 | 1.2 | 3.9 | ns |

operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\text {pd }}$ | Power dissipation capacitance | Outputs enabled |  |  | $C L=0$, | $\mathrm{f}=10 \mathrm{MHz}$ | 40 | 43 | 47 | pF |
|  |  | Outputs disabled | 20 | 22 |  |  | 24 |  |  |

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$






VOLTAGE WAVEFORMS PULSE DURATION


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tPLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. tPLH and tPHL are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 2. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| tpd | Open |
| tPLZ/tPZL | $2 \times \mathrm{V}$ CC |
| ${ }_{\text {tPHZ }} /$ tPZH | GND |



NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 3. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{tPHZ}^{\mathrm{t}} \mathrm{tPZH}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES

Input



VOLTAGE WAVEFORMS
PULSE DURATION


NOTES: A. $\mathrm{C}_{\mathrm{L}}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{d i s}$.
F. tpZL and tPZH are the same as ten.
G. $\quad \mathrm{TPLH}$ and $\mathrm{t}_{\mathrm{PHL}}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{t}^{\mathrm{tLZ}} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{tr}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tpZL and tpZH are the same as ten.
G. tPLH and tPHL are the same as $t_{p d}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {™ }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- I ${ }_{\text {off }}$ Supports Partial-Power-Down Mode Operation
- Bus Hold on Data Inputs Eliminates the Need for External Pullup/Pulldown Resistors
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 16 -bit transparent D -type latch is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.

The SN74AVCH16373 is particularly suitable for implementing buffer registers, I/O ports, bidirectional bus drivers, and working registers. This device can be used as two 8 -bit latches or one 16 -bit latch. When the latch-enable (LE) input is high, the Q outputs follow the data (D) inputs. When LE is taken low, the Q outputs are latched at the levels set up at the $D$ inputs.

## description (continued)

A buffered output-enable ( $\overline{\mathrm{OE}})$ input can be used to place the eight outputs in either a normal logic state (high or low logic levels) or the high-impedance state. In the high-impedance state, the outputs neither load nor drive the bus lines significantly. The high-impedance state and the increased drive provide the capability to drive bus lines without need for interface or pullup components. $\overline{\mathrm{OE}}$ does not affect internal operations of the latch. Old data can be retained or new data can be entered while the outputs are in the high-impedance state.

To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.

Active bus-hold circuitry is provided to hold unused or floating data inputs at a valid logic level.
This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off }}$. The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.
The SN74AVCH16373 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.
terminal assignments


| FUNCTION TABLE (each 8-bit latch) |  |  |  |
| :---: | :---: | :---: | :---: |
| INPUTS |  |  | OUTPUTQ |
| $\overline{\mathrm{OE}}$ | LE | D |  |
| L | H | H | H |
| L | H | L | L |
| L | L | X | $Q_{0}$ |
| H | X | X | Z |

## logic symbol $\dagger$


$\dagger$ This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.
logic diagram (positive logic)


# absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$ 


3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)


$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}} \mathrm{vs} \mathrm{IOH}_{\mathrm{O}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {M }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused control inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

| PARAMETER |  | TEST CONDITIONS |  | $\mathrm{V}_{\text {cc }}$ | MIN TYP $\dagger$ | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{OH}}$ |  | $\mathrm{I} \mathrm{OHS}=-100 \mu \mathrm{~A}$ |  | 1.4 V to 3.6 V | $\mathrm{V}_{\mathrm{CC}}-0.2$ |  | V |
|  |  | IOHS $=-2 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=0.91 \mathrm{~V}$ | 1.4 V | 1.05 |  |  |
|  |  | $\mathrm{I} \mathrm{OHS}=-4 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=1.07 \mathrm{~V}$ | 1.65 V | 1.2 |  |  |
|  |  | l OHS $=-8 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=1.7 \mathrm{~V}$ | 2.3 V | 1.75 |  |  |
|  |  | $\mathrm{l} \mathrm{OHS}=-12 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=2 \mathrm{~V}$ | 3 V | 2.3 |  |  |
| VoL |  | loLs $=100 \mu \mathrm{~A}$ |  | 1.4 V to 3.6 V | 0.2 |  | V |
|  |  | IOLS $=2 \mathrm{~mA}$, | $\mathrm{V}_{\text {IL }}=0.49 \mathrm{~V}$ | 1.4 V | 0.4 <br> 0.45 <br> 0.55 <br> 0.7 |  |  |
|  |  | IOLS $=4 \mathrm{~mA}$, | $\mathrm{V}_{\text {IL }}=0.57 \mathrm{~V}$ | 1.65 V |  |  |  |
|  |  | IOLS = 8 mA , | $\mathrm{V}_{\text {IL }}=0.7 \mathrm{~V}$ | 2.3 V |  |  |  |
|  |  | l OLS $=12 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.8 \mathrm{~V}$ | 3 V |  | 0.7 |  |
| 1 | Control inputs | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 3.6 V |  | $\pm 2.5$ | $\mu \mathrm{A}$ |
| ${ }^{1} \mathrm{BHL}^{\ddagger}$ |  | $\mathrm{V}_{\mathrm{I}}=0.57 \mathrm{~V}$ |  | 1.65 V | 25 |  | $\mu \mathrm{A}$ |
|  |  | $\mathrm{V}_{1}=0.7 \mathrm{~V}$ |  | 2.3 V | 45 |  |  |
|  |  | $\mathrm{V}_{1}=0.8 \mathrm{~V}$ |  | 3 V | 75 |  |  |
| ${ }^{1} \mathrm{BHH} \mathrm{S}^{\text {§ }}$ |  | $\mathrm{V}_{\mathrm{I}}=1.07 \mathrm{~V}$ |  | 1.65 V | -25 |  | $\mu \mathrm{A}$ |
|  |  | $\mathrm{V}_{1}=1.7 \mathrm{~V}$ |  | 2.3 V | -45 |  |  |
|  |  | $\mathrm{V}_{\mathrm{I}}=2 \mathrm{~V}$ |  | 3 V | -75 |  |  |
| IBHLO ${ }^{\text {d }}$ |  | $\mathrm{V}_{\mathrm{I}}=0$ to $\mathrm{V}_{\mathrm{CC}}$ |  | 1.95 V | 200 |  | $\mu \mathrm{A}$ |
|  |  | 2.7 V | 300 |  |  |  |  |
|  |  | 3.6 V | 500 |  |  |  |  |
| ${ }^{\text {IBHHO }}$ |  |  |  | $\mathrm{V}_{\mathrm{I}}=0$ to $\mathrm{V}_{\mathrm{CC}}$ |  | 1.95 V | -200 |  | $\mu \mathrm{A}$ |
|  |  | 2.7 V | -300 |  |  |  |  |  |
|  |  | 3.6 V | -500 |  |  |  |  |  |
| $\mathrm{l}_{\text {off }}$ |  |  |  | $\mathrm{V}_{1}$ or $\mathrm{V}_{\mathrm{O}}=3.6 \mathrm{~V}$ | 0 |  | $\pm 10$ | $\mu \mathrm{A}$ |  |
| IOZ |  | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\text {CC }}$ or GND |  |  |  | 3.6 V |  | $\pm 10$ | $\mu \mathrm{A}$ |
| ICC |  | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\text {CC }}$ or GND, | $\mathrm{I}=0$ |  |  | 3.6 V |  | 40 | $\mu \mathrm{A}$ |
| $\mathrm{C}_{\mathrm{i}}$ | Control inputs | $\mathrm{V}_{\mathrm{l}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V |  |  | pF |  |
|  |  |  |  | 3.3 V |  |  |  |  |
|  | Data inputs |  |  | 2.5 V |  |  |  |  |
|  |  |  |  | 3.3 V |  |  |  |  |
| $\mathrm{C}_{0}$ | Outputs | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V |  |  | pF |  |
|  |  |  |  | 3.3 V |  |  |  |  |

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
$\ddagger$ The bus-hold circuit can sink at least the minimum low sustaining current at $\mathrm{V}_{\mathrm{IL}}$ max. $\mathrm{I}_{\mathrm{BHL}}$ should be measured after lowering $\mathrm{V}_{\text {IN }}$ to $G N D$ and then raising it to $\mathrm{V}_{\text {IL }}$ max.
§ The bus-hold circuit can source at least the minimum high sustaining current at $\mathrm{V}_{I H}$ min. $I_{\mathrm{BHH}}$ should be measured after raising $\mathrm{V}_{I N}$ to $\mathrm{V}_{\mathrm{CC}}$ and then lowering it to $\mathrm{V}_{\mathrm{IH}}$ min.
II An external driver must source at least IBHLO to switch this node from low to high.
\# An external driver must sink at least $\mathrm{I}_{\mathrm{BH}} \mathrm{HO}$ to switch this node from high to low.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

|  |  | $\mathrm{V}_{C C}=1.2 \mathrm{~V}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| $t_{\text {w }}$ | Pulse duration, LE high or low |  |  |  |  |  |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {su }}$ | Setup time, data before LE $\downarrow$ |  |  |  |  |  |  |  |  |  |  | ns |
| th | Hold time, data after LE $\downarrow$ |  |  |  |  |  |  |  |  |  |  | ns |

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\begin{gathered} \hline \mathrm{V} \mathrm{CC}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{VCC}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN MAX | MIN MAX | MIN MAX | MIN MAX |  |
| $t_{\text {tpd }}$ | D | Q |  |  |  |  |  | ns |
|  | LE |  |  |  |  |  |  |  |
| $\mathrm{t}_{\text {en }}$ | $\overline{\mathrm{OE}}$ | Q |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | Q |  |  |  |  |  | ns |

operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{C C}=1.8 \mathrm{~V}$ | $\mathrm{V}_{C C}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\text {pd }}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  |  |  |  | pF |
|  |  | Outputs disabled |  |  |  |  |  |  |  |

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{C}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{tpd}_{\mathrm{pd}} \\ \mathrm{tpLZ}^{\prime / \mathrm{tPZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{tpZH}^{2} \end{gathered}$ | $\begin{gathered} \text { Open } \\ 2 \times V_{\text {CC }} \\ \text { GND } \end{gathered}$ |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES

VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. $\mathrm{tPLH}^{\text {and }} \mathrm{tPHL}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 2. Load Circuit and Voltage Waveforms


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 3. Load Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\mathrm{pd}}$ | Open |
| $\mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\mathrm{PZL}}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |



VOLTAGE WAVEFORMS PULSE DURATION


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tPLZ and tPHZ are the same as $t_{\text {dis }}$.
F. tpZL and tpZH are the same as ten.
G. $\mathrm{tPLH}^{\text {and }} \mathrm{tPHL}^{\text {are the same }}$ as $\mathrm{t}_{\mathrm{pd}}$.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{C}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{t}^{\mathrm{t} L Z} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{tf}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. $t_{P L H}$ and $t_{P H L}$ are the same as $t_{p d}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {™ }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- $I_{\text {off }}$ Supports Partial-Power-Down Mode Operation
- ESD Protection Exceeds JESD 22
- 2000-V Human-Body Model (A114-A)
- 200-V Machine Model (A115-A)
- Latch-Up Performance Exceeds 100 mA Per JESD 78, Class II
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 16 -bit edge-triggered D -type flip-flop is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.

## description (continued)

The SN74AVC16374 is particularly suitable for implementing buffer registers, I/O ports, bidirectional bus drivers, and working registers. It can be used as two 8 -bit flip-flops or one 16 -bit flip-flop. On the positive transition of the clock (CLK) input, the Q outputs of the flip-flop take on the logic levels at the data (D) inputs. $\overline{\mathrm{OE}}$ can be used to place the eight outputs in either a normal logic state (high or low logic levels) or the high-impedance state. In the high-impedance state, the outputs neither load nor drive the bus lines significantly. The high-impedance state and the increased drive provide the capability to drive bus lines without need for interface or pullup components.
$\overline{\mathrm{OE}}$ does not affect internal operations of the flip-flop. Old data can be retained or new data can be entered while the outputs are in the high-impedance state.
To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.
This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off }}$ The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.
The SN74AVC16374 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments



FUNCTION TABLE
(each 8-bit flip-flop)

| INPUTS |  |  | OUTPUT |
| :---: | :---: | :---: | :---: |
| $\mathbf{O E}$ | CLK | D | Q |
| L | $\uparrow$ | H | H |
| L | $\uparrow$ | L | L |
| L | H or L | X | $\mathrm{Q}_{0}$ |
| H | X | X | Z |

## logic symbol $\dagger$


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.
logic diagram (positive logic)


## absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$


$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed.
2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed
3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage | Operating | 1.4 | 3.6 | V |
|  |  | Data retention only | 1.2 |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
| $\mathrm{V}_{\mathrm{IL}}$ | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\text {CC }}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\text {cC }}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
| $\mathrm{V}_{\mathrm{O}}$ | Output voltage | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | 3-state | 0 | 3.6 |  |
| l OHS | Static high-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
| Iols | Static low-level output current ${ }^{\dagger}$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta v$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| $\mathrm{T}_{\mathrm{A}}$ | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $I_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC'M) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

| PARAMETER |  | TEST CONDITIONS |  | $\mathrm{V}_{\mathrm{Cc}}$ | MIN TYP $\dagger$ | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{OH}}$ |  | ${ }^{\text {I }} \mathrm{OHS}=-100 \mu \mathrm{~A}$ |  | 1.4 V to 3.6 V | $\mathrm{V}_{\mathrm{CC}}-0.2$ |  | V |
|  |  | IOHS $=-2 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=0.91 \mathrm{~V}$ | 1.4 V | 1.05 |  |  |
|  |  | IOHS $=-4 \mathrm{~mA}$, | $\mathrm{V}_{\text {IH }}=1.07 \mathrm{~V}$ | 1.65 V | 1.2 |  |  |
|  |  | $\mathrm{IOHS}=-8 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=1.7 \mathrm{~V}$ | 2.3 V | 1.75 |  |  |
|  |  | l OHS $=-12 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=2 \mathrm{~V}$ | 3 V | 2.3 |  |  |
| $\mathrm{V}_{\text {OL }}$ |  | loLS $=100 \mu \mathrm{~A}$ |  | 1.4 V to 3.6 V | 0.2 |  | V |
|  |  | loLs $=2 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.49 \mathrm{~V}$ | 1.4 V | 0.40.450.550.7 |  |  |
|  |  | loLs $=4 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.57 \mathrm{~V}$ | 1.65 V |  |  |  |
|  |  | IOLS = 8 mA , | $\mathrm{V}_{\text {IL }}=0.7 \mathrm{~V}$ | 2.3 V |  |  |  |
|  |  | IOLS $=12 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.8 \mathrm{~V}$ | 3 V | 0.7 |  |  |
| 1 | Control inputs | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 3.6 V |  | $\pm 2.5$ | $\mu \mathrm{A}$ |
| 1 lff |  | $\mathrm{V}_{\mathrm{I}}$ or $\mathrm{V}_{\mathrm{O}}=3.6 \mathrm{~V}$ |  | 0 |  | $\pm 10$ | $\mu \mathrm{A}$ |
| IOZ |  | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 3.6 V |  | $\pm 10$ | $\mu \mathrm{A}$ |
| ${ }^{\text {ICC }}$ |  | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND, | $\mathrm{I}=0$ | 3.6 V | 40 |  | $\mu \mathrm{A}$ |
| $\mathrm{C}_{\mathrm{i}}$ | Control inputs | $V_{I}=V_{C C}$ or GND |  | 2.5 V | 3 |  | pF |
|  |  |  |  | 3.3 V | 3 |  |  |
|  | Data inputs | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V | 2.5 |  |  |
|  |  |  |  | 3.3 V | 2.5 |  |  |
| $\mathrm{C}_{0}$ | Outputs | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V | 6.5 |  | pF |
|  |  |  |  | 3.3 V | 6.5 |  |  |

$\dagger$ Typical values are measured at $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ and $3.3 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

|  |  | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| $\mathrm{f}_{\text {clock }}$ | Clock frequency |  |  |  |  |  | 160 |  | 200 |  | 200 | MHz |
| $\mathrm{t}_{\mathrm{w}}$ | Pulse duration, CLK high or low |  |  |  |  | 3.1 |  | 2.5 |  | 2.5 |  | ns |
| $\mathrm{t}_{\text {su }}$ | Setup time, data before CLK $\uparrow$ | 4.1 |  | 2.7 |  | 1.9 |  | 1.4 |  | 1.4 |  | ns |
| th | Hold time, data after CLK $\uparrow$ | 1.7 |  | 1.3 |  | 1.2 |  | 1.1 |  | 1.1 |  | ns |

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| $f_{\text {max }}$ |  |  |  |  |  | 160 |  | 200 |  | 200 |  | MHz |
| $t_{\text {pd }}$ | CLK | Q | 7.3 | 1.5 | 8.4 | 1.2 | 6.7 | 0.8 | 4.1 | 0.7 | 3.3 | ns |
| ten | $\overline{\mathrm{OE}}$ | Q | 7.4 | 1.6 | 8.5 | 1.6 | 6.7 | 0.9 | 4.3 | 0.7 | 3.4 | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | Q | 8.4 | 2.5 | 9.4 | 2.3 | 7.8 | 1 | 4.2 | 1.5 | 3.9 | ns |

operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\text {CC }}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0$, | $\mathrm{f}=10 \mathrm{MHz}$ | 74 | 81 | 89 | pF |
|  |  | Outputs disabled | 52 | 57 |  |  | 63 |  |  |

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\mathrm{pd}}$ | Open |
| $\mathrm{tPLZ}^{\prime} \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |



Input


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tPLZ and $\mathrm{tPHZ}^{2}$ are the same as $\mathrm{t}_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. tPLH and tPHL are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 2. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| tpd | Open |
| tpLz/tPZL | $2 \times \mathrm{V}$ C |
| tPHZ/tPZH | GND |



VOLTAGE WAVEFORMS PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


[^5]NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tphZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.

Figure 3. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\mathrm{pd}}$ | Open |
| $\mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\text {PZL }}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES

Input


VOLTAGE WAVEFORMS
PULSE DURATION


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{d i s}$.
F. tPZL and tPZH are the same as ten.
G. $\mathrm{t}_{\mathrm{PLH}}$ and $\mathrm{t}_{\mathrm{PHL}}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $$
V_{C C}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{t}^{\mathrm{PLZ}} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{tPZH}^{2}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{tr}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{tf}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. tPLH and tPHL are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 5. Load Circuit and Voltage Waveforms
－Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
－EPIC ${ }^{\text {TM }}$（Enhanced－Performance Implanted CMOS）Submicron Process
－DOC ${ }^{\text {TM }}$（Dynamic Output Control）Circuit Dynamically Changes Output Impedance， Resulting in Noise Reduction Without Speed Degradation
－Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
－Overvoltage－Tolerant Inputs／Outputs Allow Mixed－Voltage－Mode Data Communications
－I Ioff Supports Partial－Power－Down Mode Operation
－Bus Hold on Data Inputs Eliminates the Need for External Pullup／Pulldown Resistors
－Package Options Include Plastic Thin Shrink Small－Outline（DGG）and Thin Very Small－Outline（DGV）Packages

## description

A Dynamic Output Control（DOC）circuit is implemented，which，during the transition，initially lowers the output impedance to effectively drive the load and，subsequently，raises the impedance to reduce noise．Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{IOH}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit．At the beginning of the signal transition，the DOC circuit provides a maximum dynamic drive that is equivalent to a high－drive standard－output device．For more information，refer to the TI application reports，AVC Logic Family Technology and Applications，literature number SCEA006，and Dynamic Output Control（DOC ${ }^{\text {M }}$ ） Circuitry Technology and Applications，literature number SCEA009．


Figure 1．Output Voltage vs Output Current
This 16 －bit edge－triggered D －type flip－flop is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ ，but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation．

The SN74AVCH16374 is particularly suitable for implementing buffer registers，I／O ports，bidirectional bus drivers，and working registers．It can be used as two 8 －bit flip－flops or one 16 －bit flip－flop．On the positive transition of the clock（CLK）input，the Q outputs of the flip－flop take on the logic levels at the data（D）inputs． $\overline{O E}$ can be used to place the eight outputs in either a normal logic state（high or low logic levels）or the high－impedance state．In the high－impedance state，the outputs neither load nor drive the bus lines significantly． The high－impedance state and the increased drive provide the capability to drive bus lines without need for interface or pullup components．

## description (continued)

$\overline{\mathrm{OE}}$ does not affect internal operations of the flip-flop. Old data can be retained or new data can be entered while the outputs are in the high-impedance state.
To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.

Active bus-hold circuitry is provided to hold unused or floating data inputs at a valid logic level.
This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off }}$. The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.

The SN74AVCH16374 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments



FUNCTION TABLE
(each 8-bit flip-flop)

| INPUTS |  |  | OUTPUT |
| :---: | :---: | :---: | :---: |
| $\mathbf{O E}$ | CLK | D | Q |
| $L$ | $\uparrow$ | $H$ | $H$ |
| $L$ | $\uparrow$ | $L$ | $L$ |
| $L$ | $H$ or L | $X$ | $Q_{0}$ |
| $H$ | $X$ | $X$ | $Z$ |

## logic symbol $\dagger$


$\dagger$ This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.
logic diagram (positive logic)


To Seven Other Channels


To Seven Other Channels

# absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$ 


3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)


$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}} \mathrm{vs} \mathrm{IOH}_{\mathrm{O}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEAOO9.
NOTE 4: All unused control inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
$\ddagger$ The bus-hold circuit can sink at least the minimum low sustaining current at $\mathrm{V}_{\mathrm{IL}}$ max. $\mathrm{I}_{\mathrm{BHL}}$ should be measured after lowering $\mathrm{V}_{\text {IN }}$ to $G N D$ and then raising it to $\mathrm{V}_{\text {IL }}$ max.
§ The bus-hold circuit can source at least the minimum high sustaining current at $\mathrm{V}_{\mathrm{IH}}$ min. $\mathrm{I}_{\mathrm{BHH}}$ should be measured after raising $\mathrm{V}_{I N}$ to $\mathrm{V}_{\mathrm{CC}}$ and then lowering it to $\mathrm{V}_{\mathrm{IH}}$ min.
II An external driver must source at least IBHLO to switch this node from low to high.
\# An external driver must sink at least $\mathrm{I}_{\mathrm{BH}} \mathrm{HO}$ to switch this node from high to low.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

|  |  | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| ${ }^{\text {f clock }}$ | Clock frequency |  |  |  |  |  |  |  |  |  |  | MHz |
| $\mathrm{t}_{\text {w }}$ | Pulse duration, CLK high or low |  |  |  |  |  |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {su }}$ | Setup time, data before CLK $\uparrow \uparrow$ |  |  |  |  |  |  |  |  |  |  | ns |
| th | Hold time, data after CLK $\uparrow$ |  |  |  |  |  |  |  |  |  |  | ns |

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{V} \mathrm{CC}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN MAX | MIN MAX | MIN MAX | MIN MAX |  |
| $f_{\text {max }}$ |  |  |  |  |  |  |  | MHz |
| $t_{\text {pd }}$ | CLK | Q |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {en }}$ | OE | Q |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | OE | Q |  |  |  |  |  | ns |

operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{C C}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  |  |  |  | pF |
|  |  | Outputs disabled |  |  |  |  |  |  |  |

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{C}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{tpd}_{\mathrm{pd}} \\ \mathrm{tpLZ}^{\prime / \mathrm{tPZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{tpZH}^{2} \end{gathered}$ | $\begin{gathered} \text { Open } \\ 2 \times V_{\text {CC }} \\ \text { GND } \end{gathered}$ |



Figure 2. Load Circuit and Voltage Waveforms


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 3. Load Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| tPLZ $^{\text {t PZL }}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{tPZH}^{2}$ | GND |



VOLTAGE WAVEFORMS PULSE DURATION


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tpHZ are the same as $t_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. $\mathrm{tPLH}^{\text {and }} \mathrm{tPHL}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| tPLZ $^{\text {t }}$ PZL | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES

VOLTAGE WAVEFORMS PULSE DURATION

－Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
－EPIC ${ }^{\text {TM }}$（Enhanced－Performance Implanted CMOS）Submicron Process
－UBT ${ }^{\mathrm{TM}}$（Universal Bus Transceiver） Combines D－Type Latches and D－Type Flip－Flops for Operation in Transparent， Latched，or Clocked Mode
－DOC ${ }^{\text {TM }}$（Dynamic Output Control）Circuit Dynamically Changes Output Impedance， Resulting in Noise Reduction Without Speed Degradation
－Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
－Overvoltage－Tolerant Inputs／Outputs Allow Mixed－Voltage－Mode Data Communications
－Ioff Supports Partial－Power－Down Mode Operation
－Package Options Include Plastic Thin Shrink Small－Outline（DGG）and Thin Very Small－Outline（DGV）Packages

## description

A Dynamic Output Control（DOC）circuit is implemented，which，during the transition，initially lowers the output impedance to effectively drive the load and，subsequently，raises the impedance to reduce noise．Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit．At the beginning of the signal transition，the DOC circuit provides a maximum dynamic drive that is equivalent to a high－drive standard－output device．For more information，refer to the TI application reports，AVC Logic Family Technology and Applications，literature number SCEA006，and Dynamic Output Control（DOC ${ }^{\text {TM }}$ ） Circuitry Technology and Applications，literature number SCEA009．


Figure 1．Output Voltage vs Output Current
This 18 －bit universal bus transceiver is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ ，but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation．
Data flow in each direction is controlled by output－enable（OEAB and $\overline{O E B A}$ ），latch－enable（LEAB and LEBA）， and clock（CLKAB and CLKBA）inputs．For A－to－B data flow，the device operates in the transparent mode when LEAB is high．When LEAB is low，the A data is latched if CLKAB is held at a high or low logic level．If LEAB is low，the A data is stored in the latch／flip－flop on the low－to－high transition of CLKAB．When OEAB is high，the outputs are active．When OEAB is low，the outputs are in the high－impedance state．
Data flow for $B$ to $A$ is similar to that of $A$ to $B$ ，but uses $\overline{O E B A}$ ，LEBA，and CLKBA．The output enables are complementary（OEAB is active high and OEBA is active low）．

## description (continued)

To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OEBA}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor and OEAB should be tied to GND through a pulldown resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.
This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off. }}$ The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.

The SN74AVC16501 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments


(each universal bus transceiver)

| INPUTS |  |  |  | OUTPUT <br> B |
| :---: | :---: | :---: | :---: | :---: |
| OEAB | LEAB | CLKAB | A |  |
| L | X | X | X | Z |
| H | H | X | L | L |
| H | H | X | H | H |
| H | L | $\uparrow$ | L | L |
| H | L | $\uparrow$ | H | H |
| H | L | L or H | X | $\mathrm{B}_{0} \ddagger$ |

$\dagger$ A-to- B data flow is shown: B -to-A flow is similar but uses OEBA, LEBA, and CLKBA.
$\ddagger$ Output level before the indicated steady-state input conditions were established, provided that CLKAB is high before LEAB goes low

## logic symbol§


§ This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$


Input voltage range, $\mathrm{V}_{\mathrm{I}}$ (see Note 1) ............................................................. -0.5 V to 4.6 V
Voltage range applied to any input/output when the output is in the high-impedance or power-off state, $\mathrm{V}_{\mathrm{O}}$ (see Note 1) ............................. 0.5 V to 4.6 V
Voltage range applied to any input/output when the output



Continuous output current, $\mathrm{I}_{0}$. ............................................................................ $\pm 50 \mathrm{~mA}$

Package thermal impedance, $\theta_{\mathrm{JA}}$ (see Note 3): DGG package . ...................................... 64² $\mathrm{C} / \mathrm{W}$
DGV package ...................................... $48^{\circ} \mathrm{C} / \mathrm{W}$

$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input negative-voltage and output voltage ratings may be exceeded if the input and output current ratings are observed.
2. This value is limited to 4.6 V maximum.
3. The package thermal impedance is calculated in accordance with JESD 51 .

## recommended operating conditions (see Note 4)


$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}} \mathrm{vs} \mathrm{IOH}_{\mathrm{O}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ and $3.3 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
$\ddagger$ For I/O ports, the parameter loz includes the input leakage current.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| $f_{\text {max }}$ |  |  |  |  |  |  |  |  |  |  |  | MHz |
| $t_{\text {tpd }}$ | A or B | B or A |  |  |  |  |  |  |  |  |  | ns |
|  | LE | A or B |  |  |  |  |  |  |  |  |  |  |
|  | CLK |  |  |  |  |  |  |  |  |  |  |  |
| ten | OEAB | B |  |  |  |  |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | OEAB | B |  |  |  |  |  |  |  |  |  | ns |
| ten | OEBA | A |  |  |  |  |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\text { OEBA }}$ | A |  |  |  |  |  |  |  |  |  | ns |

operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{Cc}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  |  |  |  | pF |
|  |  | Outputs disabled |  |  |  |  |  |  |  |

PARAMETER MEASUREMENT INFORMATION

$$
\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V} \text { AND } 1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{tpd}_{\mathrm{pd}} \\ \mathrm{tpLZ}^{\prime / \mathrm{tPZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{tpZH}^{2} \end{gathered}$ | $\begin{gathered} \text { Open } \\ 2 \times V_{\text {CC }} \\ \text { GND } \end{gathered}$ |




VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 2. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION

$$
V_{C C}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



VOLTAGE WAVEFORMS
SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 3. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION 

$$
\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{t}_{\mathrm{pd}} \\ \mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\mathrm{PZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{t}} \end{gathered}$ | Open $2 \times V_{C C}$ GND |



Figure 4. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| t PLZ $^{\prime}$ tPZL | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{tPZH}^{2}$ | GND |



VOLTAGE WAVEFORMS PROPAGATION DELAY TIMES

VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tphZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. tPLL and $\mathrm{tPHL}^{\text {are }}$ the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- UBT ${ }^{\mathrm{TM}}$ (Universal Bus Transceiver) Combines D-Type Latches and D-Type Flip-Flops for Operation in Transparent, Latched, or Clocked Mode
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{cc}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- I Ioff Supports Partial-Power-Down Mode Operation
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {™ }}$ ) Circuitry Technology and Applications, literature number SCEA009.


This 18 -bit universal bus transceiver is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.
The SN74AVC16601 combines D-type latches and D-type flip-flops to allow data flow in transparent, latched, and clocked modes.

Figure 1. Output Voltage vs Output Current

## description (continued)

Data flow in each direction is controlled by output-enable ( $\overline{\mathrm{OEAB}}$ and $\overline{\mathrm{OEBA}}$ ), latch-enable (LEAB and LEBA), and clock (CLKAB and CLKBA) inputs. The clock can be controlled by the clock-enable ( $\overline{C L K E N A B}$ and CLKENBA) inputs. For A-to-B data flow, the device operates in the transparent mode when LEAB is high. When LEAB is low, the A data is latched if CLKAB is held at a high or low logic level. If LEAB is low, the A data is stored in the latch/flip-flop on the low-to-high transition of CLKAB. Output enable $\overline{O E A B}$ is active low. When $\overline{O E A B}$ is low, the outputs are active. When $\overline{O E A B}$ is high, the outputs are in the high-impedance state.
Data flow for $B$ to $A$ is similar to that of $A$ to $B$, but uses $\overline{O E B A}$, LEBA, CLKBA, and CLKENBA.
To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.
This device is fully specified for partial-power-down applications using $I_{\text {off }}$ The $I_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.
The SN74AVC16601 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments

| DGG OR DGV PACKAGE (TOP VIEW) |  |  |
| :---: | :---: | :---: |
| OEAB 1 | $v_{56}$ | $]$ CLKENAB |
| LEAB [2 | 55 | ] CLKAB |
| A1 3 | 54 | B1 |
| GND [4 | 53 | GND |
| A2 5 | 52 | B2 |
| A3 6 | 51 | B3 |
| $\mathrm{v}_{\mathrm{CC}} \mathrm{S}^{7}$ | 50 | $\mathrm{V}_{\mathrm{CC}}$ |
| A4 ${ }^{\text {8 }}$ | 49 | B4 |
| A5 9 | 48 | B5 |
| A6 10 | 47 | B6 |
| GND [11 | 46 | GND |
| A7 12 | 45 | B7 |
| A8 13 | 44 | B8 |
| A9 14 | 43 | B9 |
| A10 15 | 42 | B10 |
| A11 16 | 41 | B11 |
| A12 17 | 40 | B12 |
| GND [18 | 39 | ] GND |
| A13 19 | 38 | B13 |
| A14 [20 | 37 | B14 |
| A15 21 | 36 | B15 |
| $\mathrm{V}_{\mathrm{CC}}[22$ | 35 | $\mathrm{V}_{\mathrm{CC}}$ |
| A16 23 | 34 | B16 |
| A17 24 | 33 | B17 |
| GND 25 | 32 | ] GND |
| A18 26 | 31 | B18 |
| OEBA 27 | 30 | ClkBA |
| LEBA 28 |  | CLKENBA |

FUNCTION TABLE $\dagger$
(each universal bus transceiver)

| INPUTS |  |  |  |  | OUTPUT <br> B |
| :---: | :---: | :---: | :---: | :---: | :---: |
| CLKENAB | OEAB | LEAB | CLKAB | A |  |
| X | H | X | X | X | Z |
| X | L | H | X | L | L |
| X | L | H | X | H | H |
| H | L | L | X | X | $\mathrm{B}_{0} \ddagger$ |
| H | L | L | X | X | $\mathrm{B}_{0} \ddagger$ |
| L | L | L | $\uparrow$ | L | L |
| L | L | L | $\uparrow$ | H | H |
| L | L | L | L or H | X | $\mathrm{B}_{0} \ddagger$ |

$\dagger$ A-to-B data flow is shown: B-to-A flow is similar, but uses $\overline{O E B A}$, LEBA, CLKBA, and CLKENBA.
$\ddagger$ Output level before the indicated steady-state input conditions were established
logic diagram (positive logic)


# absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$ 

| Supply voltage range, VCC | 4.6 V |
| :---: | :---: |
| Input voltage range, $\mathrm{V}_{\mathrm{I}}$ (see Note 1) | -0.5 V to 4.6 V |
| Voltage range applied to any input/output when the output is in the high-impedance or power-off state, $\mathrm{V}_{\mathrm{O}}$ (see Note 1) | $.5 \mathrm{~V} \text { to } 4.6 \mathrm{~V}$ |
| Voltage range applied to any input/output when the output is in the high or low state, $\mathrm{V}_{\mathrm{O}}$ (see Notes 1 and 2) | $\mathrm{V} \text { to } \mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}$ |
| Input clamp current, $\mathrm{I}_{\text {IK }}\left(\mathrm{V}_{\mathrm{I}}<0\right)$ | 50 mA |
| Output clamp current, $\mathrm{I}_{\mathrm{OK}}\left(\mathrm{V}_{\mathrm{O}}<0\right)$ | -50 mA |
| Continuous output current, lo | $\pm 50 \mathrm{~mA}$ |
| Continuous current through each $\mathrm{V}_{\mathrm{CC}}$ or GND | $\pm 100 \mathrm{~mA}$ |
| Package thermal impedance, $\theta_{\text {JA }}$ (see Note 3): DGG package DGV package | $\begin{aligned} & 64^{\circ} \mathrm{C} / \mathrm{W} \\ & 48^{\circ} \mathrm{C} / \mathrm{W} \end{aligned}$ |
| Storage temperature range, $\mathrm{T}_{\text {stg }}$ | $150^{\circ} \mathrm{C}$ |
| es beyond those listed under "absolute maximum ratings" may cause permane <br> nal operation of the device at these or any other conditions beyond those in <br> d. Exposure to absolute-maximum-rated conditions for extended periods may | damage to the device. These are stress ratings only, and ated under "recommended operating conditions" is not ect device reliability. |
| ES: 1. The input negative-voltage and output voltage ratings may be exceeded <br> 2. This value is limited to 4.6 V maximum. <br> 3. The package thermal impedance is calculated in accordance with JES | he input and output current ratings are observed. |

3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)


$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}} \mathrm{vs} \mathrm{IOH}_{\mathrm{O}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.

## 18-BIT UNIVERSAL BUS TRANSCEIVER

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ and $3.3 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
$\ddagger$ For I/O ports, the parameter loz includes the input leakage current.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | то (OUTPUT) | $\mathrm{V}_{C C}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| ${ }_{\text {max }}$ |  |  |  |  |  |  |  |  |  |  |  | MHz |
| ${ }^{\text {tpd }}$ | A or B | B or A |  |  |  |  |  |  |  |  |  | ns |
|  | LEAB or LEBA | A or B |  |  |  |  |  |  |  |  |  |  |
|  | CLKAB or CLKBA |  |  |  |  |  |  |  |  |  |  |  |
| ten |  | A or B |  |  |  |  |  |  |  |  |  | ns |
| ${ }_{\text {dis }}$ | $\overline{\text { OEAB or }} \overline{\text { OEBA }}$ | A or B |  |  |  |  |  |  |  |  |  | ns |

operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  |  | 10 MHz |  |  |  | F |
|  |  | Outputs disabled |  | z |  |  |  |  |

PARAMETER MEASUREMENT INFORMATION

$$
\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V} \text { AND } 1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{tpd}_{\mathrm{pd}} \\ \mathrm{tpLZ}^{\prime / \mathrm{tPZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{tpZH}^{2} \end{gathered}$ | $\begin{gathered} \text { Open } \\ 2 \times V_{\text {CC }} \\ \text { GND } \end{gathered}$ |




VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.

Figure 2. Load Circuit and Voltage Waveforms


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{d i s}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 3. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION 

$$
\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{t}_{\text {PLZ }} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{tPZH}^{2}$ | GND |



Figure 4. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| t PLZ $^{\prime}$ tPZL | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{tPZH}^{2}$ | GND |



VOLTAGE WAVEFORMS PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tphZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. tPLL and $\mathrm{tPHL}^{\text {are }}$ the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- I Ioff Supports Partial-Power-Down Mode Operation
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{IOL}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 16 -bit bus transceiver and register is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.
The SN74AVC16646 can be used as two 8 -bit transceivers or one 16-bit transceiver. Data on the A or B bus is clocked into the registers on the low-to-high transition of the appropriate clock (CLKAB or CLKBA) input. Figure 2 illustrates the four fundamental bus-management functions that can be performed with the SN74AVC16646.
Output-enable ( $\overline{\mathrm{OE}}$ ) and direction-control (DIR) inputs are provided to control the transceiver functions. In the transceiver mode, data present at the high-impedance port may be stored in either register or in both. The select-control (SAB and SBA) inputs can multiplex stored and real-time (transparent mode) data.

## description (continued)

The circuitry used for select control eliminates the typical decoding glitch that occurs in a multiplexer during the transition between stored and real-time data. DIR determines which bus receives data when $\overline{\mathrm{OE}}$ is low. In the isolation mode ( $\overline{\mathrm{OE}}$ high), A data may be stored in one register and/or B data may be stored in the other register.

When an output function is disabled, the input function is still enabled and may be used to store and transmit data. Only one of the two buses, A or B, can be driven at a time.

To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.

This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off• }}$. The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.

The SN74AVC16646 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments

DGG OR DGV PACKAGE
(TOP VIEW)
(each 8-bit transceiver/register)

| INPUTS |  |  |  |  |  | DATA I/Os |  | OPERATION OR FUNCTION |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{O E}$ | DIR | CLKAB | CLKBA | SAB | SBA | A1-A8 | B1-B8 |  |
| X | X | $\uparrow$ | X | X | X | Input | Unspecified $\dagger$ | Store A, B unspecified $\dagger$ |
| X | X | X | $\uparrow$ | X | X | Unspecified $\dagger$ | Input | Store B, A unspecified $\dagger$ |
| H | X | $\uparrow$ | $\uparrow$ | X | X | Input | Input | Store A and B data |
| H | X | H or L | H or L | X | X | Input disabled | Input disabled | Isolation, hold storage |
| L | L | X | X | X | L | Output | Input | Real-time B data to A bus |
| L | L | X | H or L | X | H | Output | Input | Stored $B$ data to $A$ bus |
| L | H | X | X | L | X | Input | Output | Real-time A data to B bus |
| L | H | H or L | X | H | X | Input | Output | Stored A data to B bus |

$\dagger$ The data-output functions may be enabled or disabled by various signals at $\overline{O E}$ and DIR. Data-input functions are always enabled; i.e., data at the bus terminals is stored on every low-to-high transition of the clock inputs.


Figure 2. Bus-Management Functions

## logic symbol $\dagger$


$\dagger$ This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.
logic diagram (positive logic)


# absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$ 

| Supply voltage range, $\mathrm{V}_{\mathrm{CC}}$ |  |
| :---: | :---: |
| Input voltage range, $\mathrm{V}_{\mathrm{I}}$ (see Note 1) | 0.5 V to 4 |
| Voltage range applied to any input/output when the output |  |
| Voltage range applied to any input/output when the output |  |
|  |  |
|  |  |
| Continuous output current, lo |  |
| Continuous current through each $\mathrm{V}_{\mathrm{CC}}$ or GND |  |
|  |  |
|  |  |
| esses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and ctional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not plied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability. |  |
| ES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed. |  |
|  |  |

## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage | Operating | 1.4 | 3.6 | V |
|  |  | Data retention only | 1.2 |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
| VIL | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
| $\mathrm{V}_{\mathrm{O}}$ | Output voltage | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | 3-state | 0 | 3.6 |  |
| IOHS | Static high-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
| IoLs | Static low-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta v$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| TA | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{VOH}_{\mathrm{OH}} \mathrm{IOH}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
$\ddagger$ For I/O ports, the parameter loz includes the input leakage current.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 3 through 6)

|  |  |  | $\mathrm{V}_{C C}=1.2 \mathrm{~V}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| ${ }_{\text {f clock }}$ | Clock frequency |  |  |  |  |  |  |  |  |  |  |  | MHz |
| ${ }^{\text {w }}$ w | Pulse duration | CLKAB or CLKBA high or low |  |  |  |  |  |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {su }}$ | Setup time | A before CLKAB $\uparrow$ or $B$ before CLKBA $\uparrow$ |  |  |  |  |  |  |  |  |  |  | ns |
| th | Hold time | A after CLKAB $\uparrow$ or B after CLKBA $\uparrow$ |  |  |  |  |  |  |  |  |  |  | ns |

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 3 through 6)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{C C}=1.2 \mathrm{~V}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN MAX | MIN MAX | MIN MAX | MIN MAX |  |
| $f_{\text {max }}$ |  |  |  |  |  |  |  | MHz |
| ${ }^{\text {tpd }}$ | A or B | B or A |  |  |  |  |  | ns |
|  | CLKAB or CLKBA | A or B |  |  |  |  |  |  |
|  | SAB or SBA |  |  |  |  |  |  |  |
| $\mathrm{t}_{\text {en }}$ | $\overline{\mathrm{OE}}$ | A or B |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | A or B |  |  |  |  |  | ns |
| ten | DIR | A or B |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | DIR | A or B |  |  |  |  |  | ns |

operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{C C}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\text {pd }}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  |  |  |  | pF |
|  |  | Outputs disabled |  |  |  |  |  |  |  |

## PARAMETER MEASUREMENT INFORMATION

$\mathrm{V}_{\mathrm{C}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |

Timing
Input


VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES



VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES
MヨI^ヨપd IOnOOपd

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{tr}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tPLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tPZL and tpZH are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 3. Load Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{t}_{\mathrm{pd}} \\ \mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\mathrm{PZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}} \end{gathered}$ | Open $2 \times V_{C C}$ GND |



Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $$
\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{t}^{\mathrm{tLZ}} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 5. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION

$$
V_{C C}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}
$$




Figure 6. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- I Ioff Supports Partial-Power-Down Mode Operation
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 20 -bit flip-flop is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.
The 20 flip-flops of the AVC16721 are edge-triggered D-type flip-flops with clock-enable (CLKEN) input. On the positive transition of the clock (CLK) input, the device provides true data at the $Q$ outputs if CLKEN is low. If CLKEN is high, no data is stored.
A buffered output-enable ( $\overline{\mathrm{OE}})$ input places the 20 outputs in either a normal logic state (high or low) or the high-impedance state. In the high-impedance state, the outputs neither load nor drive the bus lines significantly. The high-impedance state and increased drive provide the capability to drive bus lines without need for interface or pullup components. $\overline{\mathrm{OE}}$ does not affect the internal operation of the flip-flops. Old data can be retained or new data can be entered while the outputs are in the high-impedance state.

## description (continued)

To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.
This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off }}$. The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.
The SN74AVC16721 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments



NC - No internal connection

| FUNCTION TABLE (each flip-flop) |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| INPUTS |  |  |  | OUTPUT |
| $\overline{O E}$ | CLKEN | CLK | D | Q |
| L | H | X | X | $Q_{0}$ |
| L | L | $\uparrow$ | H | H |
| L | L | $\uparrow$ | L | L |
| L | L | L or H | X | $Q_{0}$ |
| H | X | X | X | Z |

## logic diagram (positive logic)



To 19 Other Channels
absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$

Input voltage range, $\mathrm{V}_{\mathrm{I}}$ (see Note 1) ........................................................... -0.5 V to 4.6 V
Voltage range applied to any output in the high-impedance or power-off state, $\mathrm{V}_{\mathrm{O}}$ (see Note 1)
-0.5 V to 4.6 V
Voltage range applied to any output in the high or low state, $\mathrm{V}_{\mathrm{O}}$



Continuous output current, IO . ......................................................................... $\pm 50 \mathrm{~mA}$
Continuous current through each $\mathrm{V}_{\mathrm{CC}}$ or GND ................................................... $\pm 100 \mathrm{~mA}$
Package thermal impedance, $\theta_{\mathrm{JA}}$ (see Note 3): DGG package ................................... $64^{\circ} \mathrm{C} / \mathrm{W}$ DGV package ...................................... $48^{\circ} \mathrm{C} / \mathrm{W}$
Storage temperature range, $\mathrm{T}_{\text {stg }}$
$-65^{\circ} \mathrm{C}$ to $150^{\circ} \mathrm{C}$
$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed.
2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed.
3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage | Operating | 1.4 | 3.6 | V |
|  |  | Data retention only | 1.2 |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
| VIL | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
| $\mathrm{V}_{\mathrm{O}}$ | Output voltage | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | 3-state | 0 | 3.6 |  |
| IOHS | Static high-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
| IoLs | Static low-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta v$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| TA | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{VOH}_{\mathrm{OH}} \mathrm{IOH}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{C C}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{V} \mathrm{CC}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN MAX | MIN MAX | MIN MAX | MIN MAX |  |
| $\mathrm{f}_{\text {max }}$ |  |  |  |  |  |  |  | MHz |
| $\mathrm{t}_{\mathrm{pd}}$ | CLK | Q |  |  |  |  |  | ns |
| $t_{\text {en }}$ | $\overline{\mathrm{OE}}$ | Q |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | Q |  |  |  |  |  | ns |

switching characteristics, $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{C}_{\mathrm{L}}=0 \mathrm{pF} \dagger$

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\begin{gathered} \hline \mathrm{VCC}=3.3 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN MAX |  |
| ${ }_{\text {tpd }}$ | CLK | Q |  | ns |

† Texas Instruments SPICE simulation data
operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{C C}=1.8 \mathrm{~V}$ | $\mathrm{V}_{C C}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  |  |  |  | pF |
|  |  | Outputs disabled |  |  |  |  |  |  |  |

## PARAMETER MEASUREMENT INFORMATION

$\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}_{\text {PZH }}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{tr}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $\quad t P Z L$ and $t P Z H$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 2. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{t}_{\mathrm{pd}} \\ \mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\mathrm{PZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}} \end{gathered}$ | Open $2 \times V_{C C}$ GND |



Figure 3. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION

$\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\mathrm{pd}}$ | Open |
| $\mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\text {PZL }}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES

Input


VOLTAGE WAVEFORMS
PULSE DURATION


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis. }}$
F. $\quad \mathrm{P} Z \mathrm{ZL}$ and tPZH are the same as ten.
G. $\mathrm{t}_{\mathrm{PLH}}$ and $\mathrm{t}_{\mathrm{PHL}}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{tf}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $\quad t P L Z$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. $t_{P L H}$ and $\mathrm{tPHL}^{2}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- Ioff Supports Partial-Power-Down Mode Operation
- Packaged in Thin Shrink Small-Outline Package


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {™ }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 22-bit flip-flop is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.
The 22 flip-flops of the SN74AVC16722 are edge-triggered D-type flip-flops with clock-enable ( $\overline{\text { CLKEN }}$ ) input. On the positive transition of the clock (CLK) input, the device stores data into the flip-flops if CLKEN is low. If CLKEN is high, no data is stored.
A buffered output-enable ( $\overline{\mathrm{OE}})$ input places the 22 outputs in either a normal logic state (high or low) or the high-impedance state. In the high-impedance state, the outputs neither load nor drive the bus lines significantly. $\overline{\mathrm{OE}}$ does not affect the internal operation of the flip-flops. Old data can be retained or new data can be entered while the outputs are in the high-impedance state.
To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{Cc}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.

## description (continued)

This device is fully specified for partial-power-down applications using $I_{\text {off }}$ The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.
The SN74AVC16722 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.
terminal assignments

| DGG PACKAGE (TOP VIEW) |  |
| :---: | :---: |
| OE 1 | $\mathrm{U}_{64} 7 \mathrm{CLK}$ |
| Q1 ${ }^{2}$ | 63 D1 |
| Q2[3 | 62 D2 |
| GND 4 | 61 GND |
| Q3 [5 | 60 D3 |
| Q4 6 | 59 D4 |
| $\mathrm{v}_{\mathrm{CC}} \mathrm{Cl}_{7}$ | 58 V CC |
| Q5 8 | 57 D5 |
| Q6[9 | 56 D6 |
| Q7-10 | 55 D7 |
| GND 11 | 54 GND |
| Q8 12 | 53 D8 |
| Q9 13 | 52 D9 |
| Q10 14 | 51 D10 |
| Q11 15 | 50 D11 |
| Q12 16 | 49 D12 |
| Q13 17 | 48 D13 |
| GND 18 | 47 GND |
| Q14 19 | 46 D14 |
| Q15 20 | 45 D15 |
| Q16 21 | 44 D16 |
| $\mathrm{v}_{\mathrm{CC}}{ }^{22}$ | ${ }^{43} \mathrm{~V}_{\mathrm{CC}}$ |
| Q17 23 | 42 D17 |
| Q18[24 | 41 D18 |
| GND 25 | 40 GND |
| Q19 26 | 39 D19 |
| Q20 27 | 38 D20 |
| $\mathrm{v}_{\mathrm{CC}}[28$ | 37 V $\mathrm{C}_{\mathrm{CC}}$ |
| Q21 29 | 36 D21 |
| Q22 30 | 35 D22 |
| GND 31 | 34 GND |
| NC[32 | 33 CLKEN |

NC - No internal connection

| FUNCTION TABLE (each flip-flop) |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| INPUTS |  |  |  | OUTPUT |
| $\overline{\mathrm{OE}}$ | CLKEN | CLK | D | Q |
| L | H | X | X | $\mathrm{Q}_{0}$ |
| L | L | $\uparrow$ | H | H |
| L | L | $\uparrow$ | L | L |
| L | L | L or H | X | $Q_{0}$ |
| H | X | X | X | Z |

## logic diagram (positive logic)



To 21 Other Channels
absolute maximum ratings over operating free-air temperature range (unless otherwise noted) ${ }^{\dagger}$
Supply voltage range, $\mathrm{V}_{\mathrm{C}}$.......................................................................... -0.5 V to 4.6 V
Input voltage range, $\mathrm{V}_{\mathrm{I}}$ (see Note 1) ........................................................... -0.5 V to 4.6 V
Voltage range applied to any output in the high-impedance or power-off state, $\mathrm{V}_{\mathrm{O}}$ (see Note 1)
-0.5 V to 4.6 V
Voltage range applied to any output in the high or low state, $\mathrm{V}_{\mathrm{O}}$



Continuous output current, Io $\ldots \ldots \ldots$................................................................... $\pm 50 \mathrm{~mA}$
Continuous current through each $\mathrm{V}_{\text {CC }}$ or GND .................................................. $\pm 100 \mathrm{~mA}$
Package thermal impedance, $\theta_{\mathrm{JA}}$ (see Note 3) ....................................................... $55^{\circ} \mathrm{C} / \mathrm{W}$
Storage temperature range, $\mathrm{T}_{\text {stg }} \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots . .65^{\circ} \mathrm{C}$ to $150^{\circ} \mathrm{C}$
$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed.
2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed.
3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Supply voltage | Operating | 1.4 | 3.6 | V |
| $V_{\text {cc }}$ | Supply voriage | Data retention only | 1.2 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
| $V_{\text {IL }}$ | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
|  | ut voltag | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  | tpurvolage | 3-state | 0 | 3.6 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 |  |
| lohs | Static high-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 | mA |
| IOHS | Staic high leve output curnt | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 |  |
|  | Static low-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 | m |
| IoLs | Staic low-level outpur curent | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta v$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| $\mathrm{T}_{\mathrm{A}}$ | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}} \mathrm{vs} \mathrm{I}_{\mathrm{OH}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {M }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{C C}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{V} \mathrm{CC}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN MAX | MIN MAX | MIN MAX | MIN MAX |  |
| $\mathrm{f}_{\text {max }}$ |  |  |  |  |  |  |  | MHz |
| $\mathrm{t}_{\mathrm{pd}}$ | CLK | Q |  |  |  |  |  | ns |
| $t_{\text {en }}$ | $\overline{\mathrm{OE}}$ | Q |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | Q |  |  |  |  |  | ns |

switching characteristics, $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{C}_{\mathrm{L}}=0 \mathrm{pF} \dagger$

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\begin{gathered} \hline \mathrm{VCC}=3.3 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN MAX |  |
| ${ }_{\text {t }}$ d | CLK | Q |  | ns |

† Texas Instruments SPICE simulation data
operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{C C}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  |  |  |  | pF |
|  |  | Outputs disabled |  |  |  |  |  |  |  |

## PARAMETER MEASUREMENT INFORMATION

$\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}_{\text {PZH }}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{tr}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 2. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{t}_{\mathrm{pd}} \\ \mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\mathrm{PZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}} \end{gathered}$ | Open $2 \times V_{C C}$ GND |



Figure 3. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{C}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\mathrm{pd}}$ | Open |
| $\mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\text {PZL }}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES

Input


VOLTAGE WAVEFORMS
PULSE DURATION


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis. }}$
F. tPZL and tPZH are the same as ten.
G. $t_{P L H}$ and $t_{P H L}$ are the same as $t_{p d}$.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



VOLTAGE WAVEFORMS PULSE DURATION


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{tr}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tpZL and tpZH are the same as ten.
G. tPLH and tPHL are the same as $t_{p d}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- Ioff Supports Partial-Power-Down Mode Operation
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{IOL}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 10 -bit flip-flop is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.
The flip-flops of the SN74AVC16820 are edge-triggered D-type flip-flops. On the positive transition of the clock (CLK) input, the device provides true data at the Q outputs.
A buffered output-enable ( $\overline{\mathrm{OE}}$ ) input can be used to place the ten outputs in either a normal logic state (high or low logic level) or the high-impedance state. In the high-impedance state, the outputs neither load nor drive the bus lines significantly. The high-impedance state and increased drive provide the capability to drive bus lines without need for interface or pullup components.
$\overline{\mathrm{OE}}$ input does not affect the internal operation of the flip-flops. Old data can be retained or new data can be entered while the outputs are in the high-impedance state.

## description (continued)

To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.
This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off. }}$ The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.
The SN74AVC16820 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments


FUNCTION TABLE
(each flip-flop)

| INPUTS |  |  |  |
| :---: | :---: | :---: | :---: |
| OUTPUT |  |  |  |
| $\overline{O E}_{\mathbf{n}} \dagger$ | CLK | D | Q $_{\mathbf{n}}{ }^{\dagger}$ |
| L | $\uparrow$ | H | H |
| L | $\uparrow$ | L | L |
| L | L | X | $\mathrm{Q}_{0}$ |
| H | X | X | Z |

$\dagger \mathrm{n}=1,2$

## logic diagram (positive logic)



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\ddagger$

> Supply voltage range, $\mathrm{V}_{\mathrm{CC}}$
> -0.5 V to 4.6 V
> Input voltage range, $\mathrm{V}_{1}$ (see Note 1) -0.5 V to 4.6 V
> Voltage range applied to any output in the high-impedance or power-off state, $\mathrm{V}_{\mathrm{O}}$
> (see Note 1)
> -0.5 V to 4.6 V

Voltage range applied to any output in the high or low state, $\mathrm{V}_{\mathrm{O}}$
(see Notes 1 and 2) -0.5 V to $\mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}$



Continuous current through each $\mathrm{V}_{\mathrm{CC}}$ or GND .................................................. $\pm 100 \mathrm{~mA}$
Package thermal impedance, $\theta_{\mathrm{JA}}$ (see Note 3): DGG package ................................... 640 $\mathrm{C} / \mathrm{W}$
DGV package ....................................... $48^{\circ} \mathrm{C} / \mathrm{W}$
Storage temperature range, $T_{\text {stg }}$
$\ddagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed.
2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed.
3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Supply voltage | Operating | 1.4 | 3.6 | V |
| $V_{\text {cc }}$ | Supply voriage | Data retention only | 1.2 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
| $V_{\text {IL }}$ | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
|  | ut voltag | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  | tpurvolage | 3-state | 0 | 3.6 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 |  |
| lohs | Static high-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 | mA |
| IOHS | Staic high leve output curnt | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 |  |
|  | Static low-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 | m |
| IoLs | Staic low-level outpur curent | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta v$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| $\mathrm{T}_{\mathrm{A}}$ | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}} \mathrm{vs} \mathrm{I}_{\mathrm{OH}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC'M) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.

## SN74AVC16820 10-BIT FLIP-FLOP WITH DUAL OUTPUTS AND 3-STATE OUTPUTS

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

|  |  | $\mathrm{V}_{C C}=1.2 \mathrm{~V}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V} \text { CC }=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| $\mathrm{f}_{\text {clock }}$ | Clock frequency |  |  |  |  |  |  |  |  |  |  | MHz |
| $\mathrm{t}_{\mathrm{w}}$ | Pulse duration, CLK high or low |  |  |  |  |  |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {su }}$ | Setup time, data before CLK $\uparrow$ |  |  |  |  |  |  |  |  |  |  | ns |
| th | Hold time, data after CLK $\uparrow$ |  |  |  |  |  |  |  |  |  |  | ns |

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\begin{gathered} \hline \mathrm{V} C=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN MAX | MIN MAX | MIN MAX | MIN MAX |  |
| $f_{\text {max }}$ |  |  |  |  |  |  |  | MHz |
| tpd | CLK | Q |  |  |  |  |  | ns |
| ten | $\overline{\mathrm{OE}}$ | Q |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | Q |  |  |  |  |  | ns |

switching characteristics, $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{C}_{\mathrm{L}}=0 \mathrm{pF} \dagger$

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\begin{gathered} \hline \mathrm{VCC}=3.3 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN MAX |  |
| ${ }_{\text {tpd }}$ | CLK | Q |  | ns |

$\dagger$ Texas Instruments SPICE simulation data
operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{C C}=1.8 \mathrm{~V}$ | $\mathrm{V}_{C C}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  |  |  |  | pF |
|  |  | Outputs disabled |  |  |  |  |  |  |  |

## PARAMETER MEASUREMENT INFORMATION

$\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\mathrm{PZL}}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}_{\text {PZH }}$ | GND |

Timing
Input


VOLTAGE WAVEFORMS
SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES



VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES
MヨI^ヨપ્d $\perp$ IOnaOYd

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tPLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tPZL and tpZH are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 2. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{t}_{\mathrm{tLZ}} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{tPZH}$ | GND |



PROPAGATION DELAY TIMES
NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control.

Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tpHZ are the same as $t_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. $\mathrm{tPLH}^{\text {and }} \mathrm{tPHL}^{\text {are the same }}$ as $\mathrm{t}_{\mathrm{pd}}$.

Figure 3. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}_{\text {PZH }}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION

$$
\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}
$$




NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. $\mathrm{tPLH}^{\text {and }} \mathrm{tPHL}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- I Ioff Supports Partial-Power-Down Mode Operation
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{IOL}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 20-bit bus-interface flip-flop is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to 3.6-V VCC operation.

The SN74AVC16821 can be used as two 10-bit flip-flops or one 20-bit flip-flop. The 20 flip-flops are edge-triggered D-type flip-flops. On the positive transition of the clock (CLK) input, the device provides true data at the $Q$ outputs.
A buffered output-enable ( $\overline{\mathrm{OE}}$ ) input can be used to place the ten outputs in either a normal logic state (high or low logic levels) or the high-impedance state. In the high-impedance state, the outputs neither load nor drive the bus lines significantly. The high-impedance state and increased drive provide the capability to drive bus lines without need for interface or pullup components.
$\overline{\mathrm{OE}}$ does not affect the internal operation of the flip-flops. Old data can be retained or new data can be entered while the outputs are in the high-impedance state.

## description (continued)

To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.
This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off }}$. The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.
The SN74AVC16821 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments

| dgG or dgv package (TOP VIEW) |  |
| :---: | :---: |
|  |  |
| E1 | 56 1CLK |
| 1Q1 2 | 551 1D1 |
| 1Q2 3 | 541 D 2 |
| GND [4 | 53 GND |
| 1Q3 [5 | 521 D 3 |
| 1Q4 6 | 51 1D4 |
| $\mathrm{v}_{\mathrm{CC}}[7$ | $50 . \mathrm{V}_{\mathrm{CC}}$ |
| 1Q5 8 | 491 D5 |
| 1Q6 [9 | $481 \mathrm{D6}$ |
| 1Q7 10 | 471 107 |
| GND [11 | 46 GND |
| 1Q8 12 | 451 108 |
| 1Q9 13 | $441 \mathrm{D9}$ |
| 1Q10 14 | 43 1D10 |
| 2Q1 ${ }^{15}$ | 42 2D1 |
| 2Q2 16 | 41 2D2 |
| 2Q3 17 | 40 2D3 |
| GND 18 | $39]$ GND |
| 2Q4 [19 | 38 2D4 |
| 2Q5 [20 | 37 2D5 |
| 2Q6 21 | 36 2D6 |
| $\mathrm{V}_{\text {CC }}[22$ | ${ }^{35}$ - $\mathrm{V}_{\mathrm{CC}}$ |
| 2Q7 [23 | 34 2D7 |
| 2Q8 24 | 33 2D8 |
| GND [25 | $32]$ GND |
| 2Q9 [26 | 31 20 9 |
| 2Q10 27 | $30] 2 \mathrm{D} 10$ |
| $2 \overline{\mathrm{OE}}$ [28 | 29] 2CLK |

FUNCTION TABLE
(each 10-bit flip-flop)

| INPUTS |  |  | OUTPUT |
| :---: | :---: | :---: | :---: |
| $\mathbf{O E}$ | CLK | $\mathbf{D}$ | Q |
| L | $\uparrow$ | $H$ | $H$ |
| L | $\uparrow$ | L | L |
| L | H or L | X | $\mathrm{Q}_{0}$ |
| H | X | X | Z |

## logic symbol $\dagger$


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.
logic diagram (positive logic)



# absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$ 



## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage | Operating | 1.4 | 3.6 | V |
|  |  | Data retention only | 1.2 |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
| VIL | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
| $\mathrm{V}_{\mathrm{O}}$ | Output voltage | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | 3-state | 0 | 3.6 |  |
| IOHS | Static high-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
| IoLs | Static low-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta v$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| TA | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{VOH}_{\mathrm{OH}} \mathrm{IOH}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {M }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

|  |  | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| $\mathrm{f}_{\text {clock }}$ | Clock frequency |  |  |  |  |  |  |  |  |  |  | MHz |
| $\mathrm{t}_{\mathrm{w}}$ | Pulse duration, CLK high or low |  |  |  |  |  |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {su }}$ | Setup time, data before CLK $\uparrow$ |  |  |  |  |  |  |  |  |  |  | ns |
| th | Hold time, data after CLK $\uparrow$ |  |  |  |  |  |  |  |  |  |  | ns |

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | $\begin{gathered} \text { TO } \\ \text { (OUTPUT) } \end{gathered}$ | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{VCC}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{VCC}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | MIN MAX | MIN MAX | MIN MAX | MIN MAX |  |
| $\mathrm{f}_{\text {max }}$ |  |  |  |  |  |  |  | MHz |
| tpd | CLK | Q |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {en }}$ | $\overline{\mathrm{OE}}$ | Q |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | OE | Q |  |  |  |  |  | ns |

## SN74AVC16821

20-BIT BUS-INTERFACE FLIP-FLOP
WITH 3-STATE OUTPUTS
SCES175F - DECEMBER 1998 - REVISED FEBRUARY 2000
switching characteristics, $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{C}_{\mathrm{L}}=0 \mathrm{pF} \dagger$

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\begin{gathered} \hline \mathrm{VCC}=3.3 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN MAX |  |
| $\mathrm{t}_{\mathrm{pd}}$ | CLK | Q |  | ns |

† Texas Instruments SPICE simulation data
operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{\text {CC }}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  |  |  |  | pF |
|  |  | Outputs disabled |  |  |  |  |  |  |  |

## PARAMETER MEASUREMENT INFORMATION

$\mathrm{V}_{\mathrm{C}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\mathrm{PZL}}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}_{\text {PZH }}$ | GND |



VOLTAGE WAVEFORMS
SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 2. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{tPZH}$ | GND |



VOLTAGE WAVEFORMS
PULSE DURATION


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control.

Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tpHZ are the same as $t_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. $\mathrm{tPLH}^{\text {and }} \mathrm{tPHL}^{\text {are the same }}$ as $\mathrm{t}_{\mathrm{pd}}$.

Figure 3. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION

$\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| ${ }^{\mathrm{t}_{\mathrm{pd}}}$ | Open |
| $\mathrm{t}^{\mathrm{tPLZ}} / \mathrm{t}_{\text {PZL }}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |



NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{d i s}$.
F. tpZL and tPZH are the same as ten.
G. $t_{P L H}$ and $t_{P H L}$ are the same as $t_{p d}$.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION

$$
V_{C C}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}
$$




VOLTAGE WAVEFORMS PROPAGATION DELAY TIMES

| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| tpLZ $^{\text {t/ }} \mathrm{tZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| tPHZ $^{\text {tPZH }}$ | GND |



VOLTAGE WAVEFORMS
PULSE DURATION


NOTES: A. $\mathrm{C}_{\mathrm{L}}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. tpLH and tPHL are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Less Than 2-ns Maximum Propagation Delay at $2.5-\mathrm{V}$ and $3.3-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- I Ioff Supports Partial-Power-Down Mode Operation
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 20 -bit non-inverting buffer/driver is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.
The SN74AVC16827 is composed of two 10-bit sections with separate output-enable signals. For either 10-bit buffer section, the two output-enable ( $1 \overline{\mathrm{OE}} 1$ and $1 \overline{\mathrm{OE} 2}$ or $2 \overline{\mathrm{OE} 1}$ and $2 \overline{\mathrm{OE} 2}$ ) inputs must both be low for the corresponding Y outputs to be active. If either output-enable input is high, the outputs of that 10-bit buffer section are in the high-impedance state.
To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.
This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off. }}$. The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.

## SN74AVC16827

20-BIT BUFFER/DRIVER
WITH 3-STATE OUTPUTS
SCES176F - DECEMBER 1998 - REVISED FEBRUARY 2000

## description (continued)

The SN74AVC16827 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments


FUNCTION TABLE
(each 10-bit buffer/driver)

| INPUTS |  |  | OUTPUT |
| :---: | :---: | :---: | :---: |
| $\overline{\text { OE1 }}$ | $\overline{\text { OE2 }}$ | A | Y |
| L | L | L | L |
| L | L | H | H |
| H | X | X | Z |
| X | H | X | Z |

logic symbol $\dagger$

$\dagger$ This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.
logic diagram (positive logic)


To Nine Other Channels


To Nine Other Channels

# absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$ 


3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)


$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{IOH}_{\mathrm{OH}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\top \mathrm{M}}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

| PARAMETER |  | TEST CONDITIONS |  | $\mathrm{V}_{\mathrm{CC}}$ | MIN TYP $\dagger$ | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{OH}}$ |  | $\mathrm{IOHS}=-100 \mu \mathrm{~A}$ |  | 1.4 V to 3.6 V | $\mathrm{V}_{\mathrm{CC}}{ }^{-0.2}$ |  | V |
|  |  | l OHS $=-2 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=0.91 \mathrm{~V}$ | 1.4 V | 1.05 |  |  |
|  |  | $\mathrm{I} \mathrm{OHS}=-4 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=1.07 \mathrm{~V}$ | 1.65 V | 1.2 |  |  |
|  |  | $\mathrm{l} \mathrm{OHS}=-8 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=1.7 \mathrm{~V}$ | 2.3 V | 1.75 |  |  |
|  |  | $\mathrm{I} \mathrm{OHS}=-12 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=2 \mathrm{~V}$ | 3 V | 2.3 |  |  |
| VOL |  | IOLS $=100 \mu \mathrm{~A}$ |  | 1.4 V to 3.6 V |  | 0.2 | V |
|  |  | IOLS $=2 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.49 \mathrm{~V}$ | 1.4 V |  | 0.4 |  |
|  |  | IOLS $=4 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.57 \mathrm{~V}$ | 1.65 V |  | 0.45 |  |
|  |  | IOLS = 8 mA , | $\mathrm{V}_{\mathrm{IL}}=0.7 \mathrm{~V}$ | 2.3 V |  | 0.55 |  |
|  |  | IOLS $=12 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.8 \mathrm{~V}$ | 3 V |  | 0.7 |  |
| 1 | Control inputs | $\mathrm{V}_{1}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 3.6 V |  | $\pm 2.5$ | $\mu \mathrm{A}$ |
| Ioff |  | $\mathrm{V}_{\mathrm{I}}$ or $\mathrm{V}_{\mathrm{O}}=3.6 \mathrm{~V}$ |  | 0 |  | $\pm 10$ | $\mu \mathrm{A}$ |
| IOZ |  | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\text {CC }}$ or GND |  | 3.6 V |  | $\pm 10$ | $\mu \mathrm{A}$ |
| ICC |  | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\text {CC }}$ or GND, | $\mathrm{I}=0$ | 3.6 V |  | 40 | $\mu \mathrm{A}$ |
| $\mathrm{C}_{i}$ | Control inputs | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V |  |  | pF |
|  |  |  |  | 3.3 V |  |  |  |
|  | Data inputs |  |  | 2.5 V |  |  |  |
|  |  |  |  | 3.3 V |  |  |  |
| $\mathrm{C}_{0}$ | Outputs | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V |  |  | pF |
|  |  |  |  | 3.3 V |  |  |  |

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{C C}=1.2 \mathrm{~V}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{VCC}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN MAX | MIN MAX | MIN MAX | MIN MAX |  |
| tpd | A | Y |  |  |  |  |  | ns |
| ten | $\overline{\mathrm{OE}}$ | Y |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | Y |  |  |  |  |  | ns |

switching characteristics, $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{C}_{\mathrm{L}}=0 \mathrm{pF} \ddagger$

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\begin{gathered} \hline \mathrm{VCC}=3.3 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN MAX |  |
| $\mathrm{t}_{\mathrm{pd}}$ | A | Y |  | ns |

$\ddagger$ Texas Instruments SPICE simulation data
operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V} \mathrm{CC}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0$, | $\mathrm{f}=10 \mathrm{MHz}$ |  |  |  |  |
|  |  | Outputs disabled |  |  |  |  |  | pF |

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\mathrm{pd}}$ | Open |
| $\mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\mathrm{PZL}}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{tPZH}$ | GND |



VOLTAGE WAVEFORMS ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $\mathrm{tPHZ}^{2}$ are the same as $\mathrm{t}_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. $\mathrm{t}_{\mathrm{PLH}}$ and $\mathrm{t}_{\mathrm{PHL}}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 2. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{tPZH}$ | GND |



NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control.

Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tpHZ are the same as $t_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. $\mathrm{tPLH}^{\text {and }} \mathrm{tPHL}^{\text {are the same }}$ as $\mathrm{t}_{\mathrm{pd}}$.

Figure 3. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{t}_{\mathrm{PLZ}} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}_{\text {PZH }}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$



| TEST | S 1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



VOLTAGE WAVEFORMS ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{tr}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{tf}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. $\quad \mathrm{P} P \mathrm{LH}$ and tPHL are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- Ioff Supports Partial-Power-Down Mode Operation
- Packaged in Thin Very Small-Outline Package


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOCㅗ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 9-bit 1-to-4 address register/driver is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.

The device is ideal for use in applications in which a single address bus is driving four separate memory locations. The SN74AVC16831 can be used as a buffer or a register, depending on the logic level of the select (SEL) input.
When $\overline{\text { SEL }}$ is logic high, the device is in the buffer mode. The outputs follow the inputs and are controlled by the two output-enable ( $\overline{\mathrm{OE}}$ ) controls. Each $\overline{\mathrm{OE}}$ controls two groups of nine outputs.
When $\overline{\mathrm{SEL}}$ is logic low, the device is in the register mode. The register is an edge-triggered D -type flip-flop. On the positive transition of the clock (CLK) input, data set up at the A inputs is stored in the internal registers. $\overline{\mathrm{OE}}$ controls operate the same as in buffer mode.

## description (continued)

When $\overline{\mathrm{OE}}$ is logic low, the outputs are in a normal logic state (high or low logic level). When $\overline{\mathrm{OE}}$ is logic high, the outputs are in the high-impedance state.
$\overline{\mathrm{SEL}}$ and $\overline{\mathrm{OE}}$ do not affect the internal operation of the flip-flops. Old data can be retained or new data can be entered while the outputs are in the high-impedance state.
To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.

This device is fully specified for partial-power-down applications using $I_{\text {off }}$ The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.
The SN74AVC16831 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments



NC - No internal connection

FUNCTION TABLE

| INPUTS |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| OUTPUT |  |  |  |  |
| OE | SEL | CLK | A | Y |
| H | X | X | X | Z |
| L | H | X | L | L |
| L | H | X | H | H |
| L | L | $\uparrow$ | L | L |
| L | L | $\uparrow$ | H | H |

## logic diagram (positive logic)


absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$

$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed.
2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed.
3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)


$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}} \mathrm{vs} \mathrm{IOH}_{\mathrm{O}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.

## 9-BIT 1-TO-4 ADDRESS REGISTER/DRIVER <br> WITH 3-STATE OUTPUTS

SCES179F - DECEMBER 1998 - REVISED FEBRUARY 2000
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

| PARAMETER |  | TEST CONDITIONS |  | $\mathrm{V}_{\mathrm{CC}}$ | MIN TYP $\dagger$ | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{OH}}$ |  | IOHS $=-100 \mu \mathrm{~A}$ |  | 1.4 V to 3.6 V | $\mathrm{V}_{\mathrm{CC}}-0.2$ |  | V |
|  |  | $\mathrm{I}^{\mathrm{O}} \mathrm{OHS}=-2 \mathrm{~mA}$, | $\mathrm{V}_{\text {IH }}=0.91 \mathrm{~V}$ | 1.4 V | 1.05 |  |  |
|  |  | $\mathrm{I} \mathrm{OHS}=-4 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=1.07 \mathrm{~V}$ | 1.65 V | 1.2 |  |  |
|  |  | $\mathrm{l} \mathrm{OHS}=-8 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=1.7 \mathrm{~V}$ | 2.3 V | 1.75 |  |  |
|  |  | $\mathrm{IOHS}=-12 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=2 \mathrm{~V}$ | 3 V | 2.3 |  |  |
| $\mathrm{V}_{\text {OL }}$ |  | loLs $=100 \mu \mathrm{~A}$ |  | 1.4 V to 3.6 V |  | 0.2 | V |
|  |  | IOLS $=2 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.49 \mathrm{~V}$ | 1.4 V |  | 0.4 |  |
|  |  | IOLS $=4 \mathrm{~mA}$, | $\mathrm{V}_{\text {IL }}=0.57 \mathrm{~V}$ | 1.65 V |  | 0.45 |  |
|  |  | IOLS = 8 mA , | $\mathrm{V}_{\mathrm{IL}}=0.7 \mathrm{~V}$ | 2.3 V |  | 0.55 |  |
|  |  | IOLS $=12 \mathrm{~mA}$, | $\mathrm{V}_{\text {IL }}=0.8 \mathrm{~V}$ | 3 V |  | 0.7 |  |
| 1 | Control inputs | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 3.6 V |  | $\pm 2.5$ | $\mu \mathrm{A}$ |
| $l_{\text {off }}$ |  | $\mathrm{V}_{\text {I }}$ or $\mathrm{V}_{\mathrm{O}}=3.6 \mathrm{~V}$ |  | 0 |  | $\pm 10$ | $\mu \mathrm{A}$ |
| IOZ |  | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\text {CC }}$ or GND |  | 3.6 V |  | $\pm 10$ | $\mu \mathrm{A}$ |
| ICC |  | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND, | $\mathrm{I}=0$ | 3.6 V |  | 40 | $\mu \mathrm{A}$ |
| $\mathrm{Ci}_{i}$ | Control inputs | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V |  |  | pF |
|  |  |  |  | 3.3 V |  |  |  |
|  | Data inputs |  |  | 2.5 V |  |  |  |
|  |  |  |  | 3.3 V |  |  |  |
| $\mathrm{C}_{0}$ | Outputs | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V |  |  | pF |
|  |  |  |  | 3.3 V |  |  |  |

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

|  |  | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| $\mathrm{f}_{\text {clock }}$ | Clock frequency |  |  |  |  |  |  |  |  |  |  | MHz |
| $\mathrm{t}_{\text {w }}$ | Pulse duration, CLK high or low |  |  |  |  |  |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {su }}$ | Setup time, A data before CLK $\uparrow$ |  |  |  |  |  |  |  |  |  |  | ns |
| th | Hold time, A data after CLK $\uparrow$ |  |  |  |  |  |  |  |  |  |  | ns |

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM(INPUT) | TO(OUTPUT) | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{v}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{v}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{v}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | min max | min MAX | min max | min max |  |
| ${ }_{\text {max }}$ |  |  |  |  |  |  |  | MHz |
| ${ }_{\text {tpd }}$ | A | Y |  |  |  |  |  | ns |
|  | CLK |  |  |  |  |  |  |  |
|  | SEL |  |  |  |  |  |  |  |
| ten | $\overline{\mathrm{OE}}$ | Y |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | OE | Y |  |  |  |  |  | ns |

switching characteristics, $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{C}_{\mathrm{L}}=0 \mathrm{pF} \dagger$

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\begin{gathered} \mathrm{VCC}=3.3 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN MAX |  |
| ${ }^{\text {tpd }}$ | A | Y |  | ns |
|  | CLK |  |  |  |

† Texas Instruments SPICE simulation data
operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{C C}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  |  |  |  | pF |
|  |  | Outputs disabled |  |  |  |  |  |  |  |

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{C}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{tpd}_{\mathrm{pd}} \\ \mathrm{tpLZ}^{\prime / \mathrm{tPZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{tpZH}^{2} \end{gathered}$ | $\begin{gathered} \text { Open } \\ 2 \times V_{\text {CC }} \\ \text { GND } \end{gathered}$ |




VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.

Figure 2. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $$
V_{C C}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| ${ }^{\mathrm{t}_{\mathrm{pd}}}$ | Open |
| $\mathrm{t}^{\mathrm{tPLZ}} / \mathrm{t}_{\text {PZL }}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |



NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis. }}$
F. $t_{P Z L}$ and tPZH are the same as ten.
G. $\mathrm{t}_{\mathrm{PLH}}$ and $\mathrm{t}_{\mathrm{PHL}}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 3. Load Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION

$$
V_{C C}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{tPZH}$ | GND |



VOLTAGE WAVEFORMS PULSE DURATION


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tphZ are the same as $t_{\text {dis }}$.
F. tpZL and tpZH are the same as ten.
G. $\mathrm{tPLH}^{\text {and }} \mathrm{tPHL}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{C}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| tpLZ $^{\prime}$ tpZL | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| tPHZ $^{\text {tpZH }}$ | GND |



VOLTAGE WAVEFORMS PROPAGATION DELAY TIMES


## VOLTAGE WAVEFORMS <br> ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t_{P Z H}$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and $\mathrm{t}_{\mathrm{PHL}}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- I ${ }_{\text {off }}$ Supports Partial-Power-Down Mode Operation
- Ideal for Use in PC133 Registered DIMM Applications
- ESD Protection Exceeds JESD 22
- 2000-V Human-Body Model (A114-A)
- 200-V Machine Model (A115-A)
- Latch-Up Performance Exceeds 100 mA Per JESD 78, Class II
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{IOL}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 18 -bit universal bus driver is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to 3.6-V VCC operation.

Data flow from A to Y is controlled by the output-enable $(\overline{\mathrm{OE}})$ input. The device operates in the transparent mode when the latch-enable ( $\overline{\mathrm{LE}}$ ) input is low. The A data is latched if the clock (CLK) input is held at a high or low logic level. If $\overline{\mathrm{LE}}$ is high, the A data is stored in the latch/flip-flop on the low-to-high transition of CLK. When $\overline{\mathrm{OE}}$ is high, the outputs are in the high-impedance state.
To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.

## SN74AVC16834

18-BIT UNIVERSAL BUS DRIVER
WITH 3-STATE OUTPUTS
SCES183F - DECEMBER 1998 - REVISED FEBRUARY 2000

## description (continued)

This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off }}$. The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.

The SN74AVC16834 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments

| DGG OR DGV PACKAGE (TOP VIEW) |  |
| :---: | :---: |
| NC [1 | $\left.\cup_{56}\right]$ GND |
| NC[2 | 55 J NC |
| Y1 ${ }^{3}$ | 54 A1 |
| GND [4 | 53 GND |
| Y2 [5 | 52 A2 |
| Y3 [6 | 51.1 A3 |
| $\mathrm{v}_{\mathrm{CC}}[7$ | 50 V ClC |
| Y4 [8 | 49 A4 |
| Y5 ${ }^{\text {¢ }}$ | 48 A5 |
| Y6 [10 | 47 A6 |
| GND 11 | 46 GND |
| Y7 [12 | 45 A7 |
| Y8-13 | 44 A8 |
| Y9 14 | 43 A9 |
| Y10 15 | 42 A10 |
| Y11 16 | 41 A11 |
| Y12 17 | 40 A12 |
| GND 18 | 39 GND |
| Y13 [19 | 38 A13 |
| Y14 20 | 37 - ${ }^{\text {A14 }}$ |
| Y15 21 | 36 A15 |
| $\mathrm{V}_{\mathrm{CC}}$ [ $^{22}$ | $35] \mathrm{V}_{\mathrm{CC}}$ |
| Y16 23 | 34 A16 |
| Y17 24 | 33 A17 |
| GND 25 | 32 GND |
| Y18 26 | 31 A18 |
| OE 27 | 30 J CLK |
| LE 428 | 29 GND |

FUNCTION TABLE
(each universal bus driver)

| INPUTS |  |  |  | OUTPUT <br> Y |
| :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { OE }}$ | $\overline{\text { LE }}$ | CLK | A |  |
| H | X | X | X | Z |
| L | L | X | L | L |
| L | L | X | H | H |
| L | H | $\uparrow$ | L | L |
| L | H | $\uparrow$ | H | H |
| L | H | H | X | $Y_{0}{ }^{\dagger}$ |
| L | H | L | X | $\mathrm{Y}_{0} \ddagger$ |

$\dagger$ Output level before the indicated steady-state input conditions were established, provided that CLK is high before LE goes high
$\ddagger$ Output level before the indicated steady-state input conditions were established

## logic symbol§


§ This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



To 17 Other Channels
absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$

Input voltage range, $\mathrm{V}_{\mathrm{I}}$ (see Note 1) ........................................................... 0.5 V to 4.6 V
Voltage range applied to any output in the high-impedance or power-off state, $\mathrm{V}_{\mathrm{O}}$ (see Note 1)
-0.5 V to 4.6 V
Voltage range applied to any output in the high or low state, $\mathrm{V}_{\mathrm{O}}$ (see Notes 1 and 2) -0.5 V to $\mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}$




Package thermal impedance, $\theta_{\mathrm{JA}}$ (see Note 3): DGG package ................................... $64^{\circ} \mathrm{C} / \mathrm{W}$ DGV package ....................................... $48^{\circ} \mathrm{C} / \mathrm{W}$

$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed.
2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed.
3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage | Operating | 1.4 | 3.6 | V |
|  |  | Data retention only | 1.2 |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
| $\mathrm{V}_{\mathrm{IL}}$ | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\text {CC }}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\text {cC }}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
| $\mathrm{V}_{\mathrm{O}}$ | Output voltage | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | 3-state | 0 | 3.6 |  |
| IOHS | Static high-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
| IoLs | Static low-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta \mathrm{v}$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| $\mathrm{T}_{\mathrm{A}}$ | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $I_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {™ }}$ ) Circuitry Technology and Applications, literature number SCEA009
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

| PARAMETER |  | TEST CONDITIONS |  | $\mathrm{V}_{\mathrm{CC}}$ | MIN TYP $\dagger$ | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{OH}}$ |  | ${ }^{\text {I OHS }}=-100 \mu \mathrm{~A}$ |  | 1.4 V to 3.6 V | $\mathrm{V}_{\mathrm{CC}}-0.2$ |  | V |
|  |  | IOHS $=-2 \mathrm{~mA}$, | $\mathrm{V}_{\text {IH }}=0.91 \mathrm{~V}$ | 1.4 V | 1.05 |  |  |
|  |  | $\mathrm{I} \mathrm{OHS}=-4 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=1.07 \mathrm{~V}$ | 1.65 V | 1.2 |  |  |
|  |  | IOHS $=-8 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=1.7 \mathrm{~V}$ | 2.3 V | 1.75 |  |  |
|  |  | $\mathrm{l} \mathrm{OHS}=-12 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=2 \mathrm{~V}$ | 3 V | 2.3 |  |  |
| $\mathrm{V}_{\mathrm{OL}}$ |  | loLS $=100 \mu \mathrm{~A}$ |  | 1.4 V to 3.6 V |  | 0.2 | V |
|  |  | loLs = 2 mA , | $\mathrm{V}_{\mathrm{IL}}=0.49 \mathrm{~V}$ | 1.4 V |  | 0.4 |  |
|  |  | l OLS $=4 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.57 \mathrm{~V}$ | 1.65 V |  | 0.45 |  |
|  |  | IOLS = 8 mA , | $\mathrm{V}_{\text {IL }}=0.7 \mathrm{~V}$ | 2.3 V |  | 0.55 |  |
|  |  | IOLS $=12 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.8 \mathrm{~V}$ | 3 V |  | 0.7 |  |
| 1 | Control inputs | $\mathrm{V}_{1}=\mathrm{V}_{\text {CC }}$ or GND |  | 3.6 V |  | $\pm 2.5$ | $\mu \mathrm{A}$ |
| loff |  | $\mathrm{V}_{\mathrm{I}}$ or $\mathrm{V}_{\mathrm{O}}=3.6 \mathrm{~V}$ |  | 0 |  | $\pm 10$ | $\mu \mathrm{A}$ |
| IOZ |  | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\text {CC }}$ or GND |  | 3.6 V |  | $\pm 10$ | $\mu \mathrm{A}$ |
| ICC |  | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\text {CC }}$ or GND, $\quad \mathrm{IO}=0$ |  | 3.6 V |  | 40 | $\mu \mathrm{A}$ |
| $\mathrm{C}_{\mathrm{i}}$ | CLK input | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V | 4 |  | pF |
|  |  |  |  | 3.3 V | 4 |  |  |
|  | Control inputs | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V | 4 |  |  |
|  |  |  |  | 3.3 V | 4 |  |  |
|  | Data inputs | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V | 2.5 |  |  |
|  |  |  |  | 3.3 V | 2.5 |  |  |
| $\mathrm{C}_{0}$ | Outputs | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V | 6.5 |  | pF |
|  |  |  |  | 3.3 V | 6.5 |  |  |

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | то (OUTPUT) | $\mathrm{V}_{C C}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| ${ }_{\text {max }}$ |  |  |  |  |  | 150 |  | 150 |  | 150 |  | MHz |
| ${ }^{\text {tpd }}$ | A | Y | 5.3 | 1.2 | 6.2 | 1.5 | 4.9 | 1 | 3.2 | 0.9 | 2.5 | ns |
|  | $\overline{\text { LE }}$ |  | 7 | 2.2 | 9.7 | 1.8 | 7.5 | 1.5 | 4.9 | 0.8 | 4 |  |
|  | CLK |  | 6 | 1.9 | 7.8 | 1.6 | 6 | 1.1 | 3.7 | 1 | 3.1 |  |
| ten | OE | Y | 7.9 | 2.4 | 10.2 | 1.6 | 8.8 | 1.5 | 6.7 | 1 | 6.2 | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | Y | 7.7 | 2.1 | 10.3 | 1.5 | 8.4 | 1.2 | 5.3 | 1 | 5.3 | ns |

switching characteristics, $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{C}_{\mathrm{L}}=0 \mathrm{pF} \dagger$

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\begin{gathered} \hline \mathrm{VCC}=3.3 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | MAX |  |
| ${ }^{\text {tpd }}$ | A | Y | 0.6 | 1.3 | ns |
|  | CLK |  | 0.7 | 1.5 |  |

${ }^{\dagger}$ Texas Instruments SPICE simulation data
operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{C C}=1.8 \mathrm{~V}$ | $\mathrm{V}_{C C}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\text {CC }}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\text {pd }}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  | 45 | 48 | 52 | pF |
|  |  | Outputs disabled | 23 | 25 |  |  | 28 |  |  |

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{C}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{t}_{\mathrm{pd}} \\ \mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\mathrm{PZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}} \end{gathered}$ | Open $2 \times V_{C C}$ GND |



VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS
PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $\mathrm{tPZH}^{2}$ are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as $t_{p d}$.

Figure 2. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\mathrm{pd}}$ | Open |
| $\mathrm{tPLZ}^{\prime} \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |

LOAD CIRCUIT


VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis. }}$
F. tpZL and tPZH are the same as ten.
G. $\mathrm{t}_{\mathrm{PLH}}$ and $\mathrm{t}_{\mathrm{PHL}}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 3. Load Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| tPLZ $^{\prime}$ tPZL | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{tPZH}^{2}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $\quad t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis. }}$.
F. $t_{P Z L}$ and $t_{P Z H}$ are the same as ten.
G. $\quad$ tPLH and $t_{P H L}$ are the same as $t_{p d}$.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}_{\text {PZH }}$ | GND |



VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{tf}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. $t_{P L H}$ and $t_{P H L}$ are the same as $t_{p d}$.

Figure 5. Load Circuit and Voltage Waveforms

- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- $\mathrm{I}_{\text {off }}$ Supports Partial-Power-Down Mode Operation
- Ideal for Use in PC133 Registered DIMM Applications
- ESD Protection Exceeds JESD 22
- 2000-V Human-Body Model (A114-A)
- 200-V Machine Model (A115-A)
- Latch-Up Performance Exceeds 100 mA Per JESD 78, Class II
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to TI application reports AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 18 -bit universal bus driver is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to 3.6-V VCC operation.

Data flow from A to Y is controlled by the output-enable ( $\overline{\mathrm{OE}}$ ) input. The device operates in the transparent mode when the latch-enable (LE) input is high. The A data is latched if the clock (CLK) input is held at a high or low logic level. If LE is low, the A data is stored in the latch/flip-flop on the low-to-high transition of CLK. When $\overline{\mathrm{OE}}$ is high, the outputs are in the high-impedance state.
To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.

## SN74AVC16835

18-BIT UNIVERSAL BUS DRIVER
WITH 3-STATE OUTPUTS
SCES168H - DECEMBER 1998 - REVISED FEBRUARY 2000

## description (continued)

This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off }}$. The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.

The SN74AVC16835 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments



FUNCTION TABLE
(each universal bus driver)

| INPUTS |  |  |  | OUTPUT |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{O E}$ | LE | CLK | A | Y |
| $H$ | X | X | X | Z |
| L | H | X | L | L |
| L | H | X | H | H |
| L | L | $\uparrow$ | L | L |
| L | L | $\uparrow$ | H | H |
| L | L | L or H | X | $\mathrm{Y}_{0}{ }^{\dagger}$ |

$\dagger$ Output level before the indicated steady-state input conditions were established, provided that CLK is high before LE goes low

## logic symbol $\ddagger$


$\ddagger$ This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

## logic diagram (positive logic)



To 17 Other Channels
absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$

Input voltage range, $\mathrm{V}_{\mathrm{I}}$ (see Note 1) ........................................................... 0.5 V to 4.6 V
Voltage range applied to any output in the high-impedance or power-off state, $\mathrm{V}_{\mathrm{O}}$ (see Note 1)
-0.5 V to 4.6 V
Voltage range applied to any output in the high or low state, $\mathrm{V}_{\mathrm{O}}$ (see Notes 1 and 2) -0.5 V to $\mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}$




Package thermal impedance, $\theta_{\mathrm{JA}}$ (see Note 3): DGG package ................................... $64^{\circ} \mathrm{C} / \mathrm{W}$ DGV package ....................................... $48^{\circ} \mathrm{C} / \mathrm{W}$

$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed.
2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed.
3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage | Operating | 1.4 | 3.6 | V |
|  |  | Data retention only | 1.2 |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
| $\mathrm{V}_{\mathrm{IL}}$ | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
| $\mathrm{V}_{\mathrm{O}}$ | Output voltage | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | 3-state | 0 | 3.6 |  |
| IOHS | Static high-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
| Iols | Static low-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta v$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| TA | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $I_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ characteristics. Refer to TI application reports AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to TI application report Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | то (OUTPUT) | $\mathrm{V}_{C C}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| ${ }_{\text {max }}$ |  |  |  |  |  | 150 |  | 150 |  | 150 |  | MHz |
| ${ }^{\text {tpd }}$ | A | Y | 4.5 | 1.2 | 6.2 | 1.3 | 5.5 | 1 | 3.1 | 0.9 | 2.5 | ns |
|  | LE |  | 6.2 | 1.6 | 9.4 | 1.3 | 7.2 | 1.1 | 4.7 | 0.9 | 3.8 |  |
|  | CLK |  | 5.2 | 1.6 | 7.8 | 1.5 | 6 | 1 | 3.7 | 0.8 | 3.1 |  |
| $\mathrm{t}_{\text {en }}$ | OE | Y | 7.1 | 2.4 | 10.2 | 2.2 | 8.8 | 1.5 | 6.7 | 1.2 | 6.2 | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | Y | 6.9 | 2.2 | 10.3 | 2 | 8.4 | 1.2 | 5.3 | 1.1 | 5.3 | ns |

switching characteristics, $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{C}_{\mathrm{L}}=0 \mathrm{pF} \dagger$

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | MAX |  |
| ${ }^{\text {tpd }}$ | A | Y | 0.6 | 1.3 | ns |
|  | CLK |  | 0.7 | 1.5 |  |

† Texas Instruments SPICE simulation data
operating characteristics, $\mathrm{T}_{\mathrm{A}}=\mathbf{2 5}{ }^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{\text {CC }}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0$, | $\mathrm{f}=10 \mathrm{MHz}$ | 45 | 48 | 52 | pF |
|  |  | Outputs disabled | 23 | 25 |  |  | 28 |  |  |

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{C}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{t}_{\mathrm{pd}} \\ \mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\mathrm{PZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}} \end{gathered}$ | Open $2 \times V_{C C}$ GND |



VOLTAGE WAVEFORMS PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


[^6]NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. $t_{P Z L}$ and $t P Z H$ are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 2. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\mathrm{pd}}$ | Open |
| $\mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\text {PZL }}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis. }}$
F. tpZL and tPZH are the same as ten.
G. $\mathrm{t}_{\mathrm{PLH}}$ and $\mathrm{t}_{\mathrm{PHL}}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 3. Load Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\mathrm{pd}}$ | Open |
| $\mathrm{t}^{\mathrm{t} L Z} / \mathrm{t}_{\mathrm{PZL}}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


NOTES: A. $\mathrm{C}_{\mathrm{L}}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tpHZ are the same as $t_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. $\mathrm{tPLH}^{\text {and }} \mathrm{tPHL}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| tPLZ $^{\text {t }}$ PZL | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{tf}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. $t_{P L H}$ and $t_{P H L}$ are the same as $t_{p d}$.

Figure 5. Load Circuit and Voltage Waveforms

## - Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family

- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- I Ioff Supports Partial-Power-Down Mode Operation
- Ideal for Use in PC133 Registered DIMM Applications
- Package Options Include Plastic Thin Shrink Small-Outline (DGG) and Thin Very Small-Outline (DGV) Packages


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {M }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 20-bit universal bus driver is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.
Data flow from A to Y is controlled by the output-enable ( $\overline{\mathrm{OE}})$ input. The device operates in the transparent mode when the latch-enable ( $\overline{\mathrm{LE}}$ ) input is low. When $\overline{\mathrm{LE}}$ is high, the A data is latched if the clock (CLK) input is held at a high or low logic level. If $\overline{\mathrm{LE}}$ is high, the A data is stored in the latch/flip-flop on the low-to-high transition of CLK. When $\overline{O E}$ is high, the outputs are in the high-impedance state.
To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.
This device is fully specified for partial-power-down applications using $I_{\text {off. }}$. The $I_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.

## SN74AVC16836

20-BIT UNIVERSAL BUS DRIVER
WITH 3-STATE OUTPUTS
SCES170F - DECEMBER 1998 - REVISED FEBRUARY 2000

## description (continued)

The SN74AVC16836 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

## terminal assignments



FUNCTION TABLE
(each universal bus driver)

| INPUTS |  |  |  | $\begin{gathered} \text { OUTPUT } \\ \mathbf{Y} \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { OE }}$ | $\overline{\text { LE }}$ | CLK | A |  |
| H | X | X | X | Z |
| L | L | X | L | L |
| L | L | X | H | H |
| L | H | $\uparrow$ | L | L |
| L | H | $\uparrow$ | H | H |
| L | H | L or H | X | $\mathrm{Y}_{0}{ }^{\dagger}$ |

$\dagger$ Output level before the indicated steady-state input conditions were established

## logic symbol $\dagger$


† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.
logic diagram (positive logic)


# absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$ 


3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage | Operating | 1.4 | 3.6 | V |
|  |  | Data retention only | 1.2 |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
| VIL | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
| $\mathrm{V}_{\mathrm{O}}$ | Output voltage | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | 3-state | 0 | 3.6 |  |
| IOHS | Static high-level output current ${ }^{\dagger}$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
| loLs | Static low-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta \mathrm{v}$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| TA | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}} \mathrm{vs} \mathrm{IOH}_{\mathrm{O}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{C}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

| PARAMETER |  | TEST CONDITIONS |  | $\mathrm{V}_{\mathrm{CC}}$ | MIN TYP $\dagger$ | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{OH}}$ |  | $\mathrm{IOHS}=-100 \mu \mathrm{~A}$ |  | 1.4 V to 3.6 V | $\mathrm{V}_{\mathrm{CC}}-0.2$ |  | V |
|  |  | l OHS $=-2 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=0.91 \mathrm{~V}$ | 1.4 V | 1.05 |  |  |
|  |  | l OHS $=-4 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=1.07 \mathrm{~V}$ | 1.65 V | 1.2 |  |  |
|  |  | l OHS $=-8 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=1.7 \mathrm{~V}$ | 2.3 V | 1.75 |  |  |
|  |  | $\mathrm{I} \mathrm{OHS}=-12 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=2 \mathrm{~V}$ | 3 V | 2.3 |  |  |
| $\mathrm{V}_{\mathrm{OL}}$ |  | IOLS $=100 \mu \mathrm{~A}$ |  | 1.4 V to 3.6 V |  | 0.2 | V |
|  |  | IOLS $=2 \mathrm{~mA}$, | $\mathrm{V}_{\text {IL }}=0.49 \mathrm{~V}$ | 1.4 V |  | 0.4 |  |
|  |  | IOLS $=4 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.57 \mathrm{~V}$ | 1.65 V |  | 0.45 |  |
|  |  | IOLS = 8 mA , | $\mathrm{V}_{\mathrm{IL}}=0.7 \mathrm{~V}$ | 2.3 V |  | 0.55 |  |
|  |  | IOLS $=12 \mathrm{~mA}$, | $\mathrm{V}_{\text {IL }}=0.8 \mathrm{~V}$ | 3 V |  | 0.7 |  |
| 1 | Control inputs | $\mathrm{V}_{1}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 3.6 V |  | $\pm 2.5$ | $\mu \mathrm{A}$ |
| loff |  | $\mathrm{V}_{1}$ or $\mathrm{V}_{\mathrm{O}}=3.6 \mathrm{~V}$ |  | 0 |  | $\pm 10$ | $\mu \mathrm{A}$ |
| IOZ |  | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 3.6 V |  | $\pm 10$ | $\mu \mathrm{A}$ |
| ICC |  | $\mathrm{V}_{\mathrm{l}}=\mathrm{V}_{\mathrm{CC}}$ or GND, | $\mathrm{I}=0$ | 3.6 V |  | 40 | $\mu \mathrm{A}$ |
| $\mathrm{C}_{i}$ | Control inputs | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V |  |  | pF |
|  |  |  |  | 3.3 V |  |  |  |
|  | Data inputs |  |  | 2.5 V |  |  |  |
|  |  |  |  | 3.3 V |  |  |  |
| $\mathrm{C}_{0}$ | Outputs | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{CC}}$ or GND |  | 2.5 V |  |  | pF |
|  |  |  |  | 3.3 V |  |  |  |

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

|  |  |  |  | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| $\mathrm{f}_{\text {clock }}$ | Clock frequency |  |  |  |  |  |  |  |  |  |  |  |  | MHz |
| ${ }^{\text {w }}$ w | Pulse duration | LE low |  |  |  |  |  |  |  |  |  |  |  |  |
|  |  | CLK high or low |  |  |  |  |  |  |  |  |  |  |  |  |
| $\mathrm{t}_{\text {su }}$ | Setup time | Data before CLK $\uparrow$ |  |  |  |  |  |  |  |  |  |  |  | ns |
|  |  | Data before $\overline{L E} \uparrow$ | CLK high |  |  |  |  |  |  |  |  |  |  |  |
|  |  |  | CLK low |  |  |  |  |  |  |  |  |  |  |  |
| th | Hold time | Data after CLK $\uparrow$ |  |  |  |  |  |  |  |  |  |  |  | ns |
|  |  | Data after $\overline{\mathrm{LE}} \uparrow$ | CLK high or low |  |  |  |  |  |  |  |  |  |  |  |

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | то (OUTPUT) | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| ${ }_{\text {max }}$ |  |  |  |  |  |  |  |  |  |  |  | MHz |
| ${ }^{\text {tpd }}$ | A | Y |  |  |  |  |  |  |  |  |  | ns |
|  | $\overline{\text { LE }}$ |  |  |  |  |  |  |  |  |  |  |  |
|  | CLK |  |  |  |  |  |  |  |  |  |  |  |
| ten | $\overline{\mathrm{OE}}$ | Y |  |  |  |  |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | Y |  |  |  |  |  |  |  |  |  | ns |

switching characteristics, $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{C}_{\mathrm{L}}=0 \mathrm{pF} \dagger$

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN MAX |  |
| ${ }^{\text {tpd }}$ | A | Y |  | ns |
|  | CLK |  |  |  |

† Texas Instruments SPICE simulation data

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{C}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{tpd}_{\mathrm{pd}} \\ \mathrm{tpLZ}^{\prime / \mathrm{tPZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{tpZH}^{2} \end{gathered}$ | $\begin{gathered} \text { Open } \\ 2 \times V_{\text {CC }} \\ \text { GND } \end{gathered}$ |




VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 2. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION

$$
V_{C C}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| tPLZ $^{\text {t }}$ PZL | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}$ PZH | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis. }}$
F. tpZL and tPZH are the same as ten.
G. $\mathrm{t}_{\mathrm{PLH}}$ and $\mathrm{t}_{\mathrm{PHL}}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 3. Load Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\mathrm{pd}}$ | Open |
| $\mathrm{t}^{\mathrm{t} L Z} / \mathrm{t}_{\mathrm{PZL}}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |



VOLTAGE WAVEFORMS PULSE DURATION


NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tpHZ are the same as $t_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. $\mathrm{tPLH}^{\text {and }} \mathrm{tPHL}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 4. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| tPLZ $^{\text {t }}$ PZL | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}$ PZH | GND |



> VOLTAGE WAVEFORMS
> PROPAGATION DELAY TIMES

VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $\mathrm{C}_{\mathrm{L}}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{tf}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. $t_{P L H}$ and $t_{P H L}$ are the same as $t_{p d}$.

Figure 5. Load Circuit and Voltage Waveforms
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- Member of the Texas Instruments Widebus ${ }^{\text {TM }}$ Family
- EPIC ${ }^{\text {TM }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- Ioff Supports Partial-Power-Down Mode Operation
- Packaged in Plastic Fine-Pitch Ball Grid Array Package


## description

A Dynamic Output Control (DOC) circuit is implemented, which, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC ${ }^{\text {TM }}$ ) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 32-bit (dual-octal) noninverting bus transceiver is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.
The SN74AVC32245 is designed for asynchronous communication between data buses. The control-function implementation minimizes external timing requirements.
This device can be used as four 8-bit transceivers, two 16-bit transceivers, or one 32-bit transceiver. It allows data transmission from the $A$ bus to the $B$ bus or from the $B$ bus to the $A$ bus, depending on the logic level at the direction-control (DIR) input. The output-enable ( $\overline{\mathrm{OE}}$ ) input can be used to disable the device so that the buses are effectively isolated.
To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OE}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.

## description (continued)

This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off }}$. The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.

The SN74AVC32245 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.


## terminal assignments

| 6 | 1A2 | 1A4 | 1A6 | 1A8 | 2 A 2 | 2A4 | 2A6 | 2A7 | 3A2 | 3A4 | 3A6 | 3A8 | 4A2 | 4A4 | 4A6 | 4A7 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 5 | 1A1 | 1A3 | 1A5 | 1A7 | 2A1 | 2 A 3 | 2A5 | 2A8 | 3A1 | 3A3 | 3A5 | 3A7 | 4A1 | 4A3 | 4A5 | 4A8 |
| 4 | 1 $\overline{O E}$ | GND | $\mathrm{V}_{\mathrm{CC}}$ | GND | GND | $\mathrm{V}_{\mathrm{CC}}$ | GND | 2 $\overline{\mathrm{OE}}$ | $3 \overline{\mathrm{OE}}$ | GND | $\mathrm{V}_{\mathrm{CC}}$ | GND | GND | $\mathrm{V}_{\mathrm{CC}}$ | GND | 4 $\overline{\mathrm{OE}}$ |
| 3 | 1DIR | GND | $\mathrm{V}_{\mathrm{CC}}$ | GND | GND | $\mathrm{V}_{\mathrm{CC}}$ | GND | 2DIR | 3DIR | GND | $\mathrm{V}_{\mathrm{CC}}$ | GND | GND | $\mathrm{V}_{\mathrm{CC}}$ | GND | 4DIR |
| 2 | 1B1 | 1B3 | 1B5 | 1B7 | 2B1 | 2B3 | 2B5 | 2B8 | 3B1 | 3B3 | 3B5 | 3B7 | 4B1 | 4B3 | 4B5 | 4B8 |
| 1 | 1B2 | 1B4 | 1B6 | 1B8 | 2B2 | 2B4 | 2B6 | 2B7 | 3B2 | 3B4 | 3B6 | 3B8 | 4B2 | 4B4 | 4B6 | 4B7 |
|  | A | B | C | D | E | F | G | H | J | K | L | M | N | P | R | T |

FUNCTION TABLE
(each 8-bit transceiver)

| INPUTS |  | OPERATION |
| :---: | :---: | :---: |
| $\overline{\text { OE }}$ | DIR |  |
| L | L | B data to A bus |
| L | H | A data to B bus |
| H | X | Isolation |

logic diagram (positive logic)


To Seven Other Channels


To Seven Other Channels


To Seven Other Channels


To Seven Other Channels
absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$

Input voltage range, $\mathrm{V}_{\text {I }}$ : Except I/O ports (see Note 1) ......................................... 0.5 V to 4.6 V

Voltage range applied to any output in the high-impedance or power-off state, $\mathrm{V}_{\mathrm{O}}$
(see Note 1)
-0.5 V to 4.6 V
Voltage range applied to any output in the high or low state, $\mathrm{V}_{\mathrm{O}}$
(see Notes 1 and 2) ................................................................... 0.5 V to $\mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}$


Continuous output current, IO . ........................................................................... $\pm 50 \mathrm{~mA}$
Continuous current through each $\mathrm{V}_{\text {CC }}$ or GND ....................................................... $\pm 100 \mathrm{~mA}$
Package thermal impedance, $\theta_{\mathrm{JA}}$ (see Note 3) ....................................................... 40² $\mathrm{C} / \mathrm{W}$
Storage temperature range, $\mathrm{T}_{\text {stg }} \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots \ldots . .65^{\circ} \mathrm{C}$ to $150^{\circ} \mathrm{C}$
$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input and output negative-voltage ratings may be exceeded if the input and output current ratings are observed.
2. The output positive-voltage rating may be exceeded up to 4.6 V maximum if the output current rating is observed.
3. The package thermal impedance is calculated in accordance with JESD 51.

## recommended operating conditions (see Note 4)


$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC'M) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
$\ddagger$ For I/O ports, the parameter loz includes the input leakage current.
switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN MAX | MIN MAX | MIN MAX | MIN MAX |  |
| tpd | A or B | B or A |  |  |  |  |  | ns |
| ten | $\overline{\mathrm{OE}}$ | A or B |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {dis }}$ | $\overline{\mathrm{OE}}$ | A or B |  |  |  |  |  | ns |

operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{Cc}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\text {pd }}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  |  |  |  | pF |
|  |  | Outputs disabled |  |  |  |  |  |  |  |

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{C}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{tpd}_{\mathrm{pd}} \\ \mathrm{tpLZ}^{\prime / \mathrm{tPZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{tpZH}^{2} \end{gathered}$ | $\begin{gathered} \text { Open } \\ 2 \times V_{\text {CC }} \\ \text { GND } \end{gathered}$ |




VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 2. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION

$$
V_{C C}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}_{\text {PZH }}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{tr}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. $\quad \mathrm{tPLH}$ and tPHL are the same as tpd.

Figure 3. Load Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION

$$
\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| tpd | Open |
| tpLz/tPZL | $2 \times \mathrm{V}$ C |
| tPHZ/tPZH | GND |



Figure 4. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$ 



| TEST | Si |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| tPLZ $/$ tPZL | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHI }} / \mathrm{tPZH}$ | GND |



VOLTAGE WAVEFORMS PROPAGATION DELAY TIMES



VOLTAGE WAVEFORMS
PULSE DURATION


## VOLTAGE WAVEFORMS

ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tPLZ and $\mathrm{tPHZ}^{2}$ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. tPLL and $\mathrm{tPHL}^{\text {are }}$ the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure 5. Load Circuit and Voltage Waveforms

## - Member of the Texas Instruments Widebus $+^{\text {TM }}$ Family

- EPIC ${ }^{\text {™ }}$ (Enhanced-Performance Implanted CMOS) Submicron Process
- UBT ${ }^{\text {™ }}$ (Universal Bus Transceiver) Combines D-Type Latches and D-Type Flip-Flops for Operation in Transparent, Latched, or Clocked Mode
- DOC ${ }^{\text {TM }}$ (Dynamic Output Control) Circuit Dynamically Changes Output Impedance, Resulting in Noise Reduction Without Speed Degradation
- Less Than 2-ns Maximum Propagation Delay at $2.5-\mathrm{V}$ and $3.3-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Dynamic Drive Capability Is Equivalent to Standard Outputs With $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{IOL}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$
- Overvoltage-Tolerant Inputs/Outputs Allow Mixed-Voltage-Mode Data Communications
- I ${ }_{\text {off }}$ Supports Partial-Power-Down Mode Operation
- Packaged in Plastic Fine-Pitch Ball Grid Array Package


## description

A Dynamic Output Control (DOC) circuit is implemented that, during the transition, initially lowers the output impedance to effectively drive the load and, subsequently, raises the impedance to reduce noise. Figure 1 shows typical $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves to illustrate the output impedance and drive capability of the circuit. At the beginning of the signal transition, the DOC circuit provides a maximum dynamic drive that is equivalent to a high-drive standard-output device. For more information, refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC) Circuitry Technology and Applications, literature number SCEA009.


Figure 1. Output Voltage vs Output Current
This 36 -bit universal bus transceiver is operational at $1.2-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, but is designed specifically for $1.65-\mathrm{V}$ to $3.6-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation.
Data flow in each direction is controlled by output-enable (OEAB and $\overline{\mathrm{OEBA}}$ ), latch-enable (LEAB and LEBA), and clock (CLKAB and CLKBA) inputs. For A-to-B data flow, the device operates in the transparent mode when LEAB is high. When LEAB is low, the A data is latched if CLKAB is held at a high or low logic level. If LEAB is low, the A data is stored in the latch/flip-flop on the low-to-high transition of CLKAB. When OEAB is high, the outputs are active. When OEAB is low, the outputs are in the high-impedance state.

Data flow for $B$ to $A$ is similar to that of $A$ to $B$, but uses $\overline{O E B A}$, LEBA, and CLKBA. The output enables are complementary (OEAB is active high and OEBA is active low).

## description (continued)

To ensure the high-impedance state during power up or power down, $\overline{\mathrm{OEBA}}$ should be tied to $\mathrm{V}_{\mathrm{CC}}$ through a pullup resistor, and OEAB should be tied to GND through a pulldown resistor; the minimum value of the resistor is determined by the current-sinking capability of the driver.
This device is fully specified for partial-power-down applications using $\mathrm{I}_{\text {off }}$. The $\mathrm{I}_{\text {off }}$ circuitry disables the outputs, preventing damaging current backflow through the device when it is powered down.

The SN74AVC32501 is characterized for operation from $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$.

terminal assignments

|  | $\mathbf{1}$ | $\mathbf{2}$ | $\mathbf{3}$ | $\mathbf{4}$ | $\mathbf{5}$ | $\mathbf{6}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{A}$ | 1 A 2 | 1 A 1 | 1 LEAB | 1 CLKAB | 1 B 1 | 1 B 2 |
| $\mathbf{B}$ | 1 A 4 | 1 A 3 | 1 OEAB | 1 GND | 1 B 3 | 1 B 4 |
| $\mathbf{C}$ | 1 A 6 | 1 A 5 | 1 GND | 1 GND | 1 B 5 | 1 B 6 |
| $\mathbf{D}$ | 1 A 8 | 1 A 7 | $1 \mathrm{~V}_{\mathrm{CC}}$ | $1 \mathrm{~V}_{\mathrm{CC}}$ | 1 B 7 | 1 B 8 |
| $\mathbf{E}$ | 1 A 10 | 1 A 9 | 1 GND | 1 GND | 1 B 9 | 1 B 10 |
| $\mathbf{F}$ | 1 A 12 | 1 A 11 | 1 GND | 1 GND | 1 B 11 | 1 B 12 |
| $\mathbf{G}$ | 1 A 14 | 1 A 13 | $1 \mathrm{~V}_{\mathrm{CC}}$ | $1 \mathrm{~V}_{\mathrm{CC}}$ | 1 B 13 | 1 B 14 |
| $\mathbf{H}$ | 1 A 15 | 1 A 16 | 1 GND | 1 GND | 1 B 16 | 1 B 15 |
| $\mathbf{J}$ | 1 A 17 | 1 A 18 | $1 \overline{\mathrm{OEBA}}$ | 1 CLKBA | 1 B 18 | 1 B 17 |
| $\mathbf{K}$ | NC | 2 LEAB | 1 LEBA | 1 GND | 2 CLKAB | NC |
| $\mathbf{L}$ | 2 A 2 | 2 A 1 | 2 OEAB | 2 GND | 2 B 1 | 2 B 2 |
| $\mathbf{M}$ | 2 A 4 | 2 A 3 | 2 GND | 2 GND | 2 B 3 | 2 B 4 |
| $\mathbf{N}$ | 2 A 6 | 2 A 5 | $2 \mathrm{~V}_{\mathrm{CC}}$ | $2 \mathrm{~V}_{\mathrm{CC}}$ | 2 B 5 | 2 B 6 |
| $\mathbf{P}$ | 2 A 8 | 2 A 7 | 2 GND | 2 GND | 2 B 7 | 2 B 8 |
| $\mathbf{R}$ | 2 A 10 | 2 A 9 | 2 GND | 2 GND | 2 B 9 | 2 B 10 |
| $\mathbf{T}$ | 2 A 12 | 2 A 11 | $2 \mathrm{~V}_{\mathrm{CC}}$ | $2 \mathrm{~V}_{\mathrm{CC}}$ | 2 B 11 | 2 B 12 |
| $\mathbf{U}$ | 2 A 14 | 2 A 13 | 2 GND | 2 GND | 2 B 13 | 2 B 14 |
| $\mathbf{V}$ | 2 A 15 | 2 A 16 | $2 \overline{\mathrm{OEBA}}$ | 2 CLKBA | 2 B 16 | 2 B 15 |
| $\mathbf{W}$ | 2 A 17 | 2 A 18 | 2 LEBA | 2 GND | 2 B 18 | 2 B 17 |

NC - No internal connection

FUNCTION TABLE $\dagger$
(each 18-bit universal bus transceiver)

| INPUTS |  |  |  | OUTPUT |
| :---: | :---: | :---: | :---: | :---: |
| OEAB | LEAB | CLKAB | A |  |
| L | X | X | X | Z |
| H | $H$ | $X$ | L | L |
| $H$ | $H$ | $X$ | $H$ | $H$ |
| $H$ | L | $\uparrow$ | L | L |
| H | L | $\uparrow$ | $H$ | $H$ |
| $H$ | L | L or H | X | $B_{0} \ddagger$ |

† A-to-B data flow is shown. B-to-A flow is similar but uses OEBA, LEBA, and CLKBA.
$\ddagger$ Output level before the indicated steady-state input conditions were established, provided that CLKAB is high before LEAB goes low

## logic diagram (positive logic)



## logic diagram (positive logic) (continued)



## absolute maximum ratings over operating free-air temperature range (unless otherwise noted) $\dagger$


Input voltage range, $\mathrm{V}_{1}$ : Except I/O ports (see Note 1) ........................................ -0.5 V to 4.6 V

Voltage range applied to any output in the high-impedance or power-off state, $\mathrm{V}_{\mathrm{O}}$ (see Note 1)
-0.5 V to 4.6 V
Voltage range applied to any output in the high or low state, $\mathrm{V}_{\mathrm{O}}$
(see Notes 1 and 2)
-0.5 V to $\mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}$


Continuous output current, IO . ............................................................................. $\pm 50 \mathrm{~mA}$
Continuous current through each $\mathrm{V}_{\text {CC }}$ or $\operatorname{GND}$....................................................... $\pm 100 \mathrm{~mA}$
Package thermal impedance, $\theta_{\mathrm{JA}}$ (see Note 3) ........................................................... 39³ $\mathrm{C} / \mathrm{W}$

$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTES: 1. The input negative-voltage and output voltage ratings may be exceeded if the input and output current ratings are observed.
2. This value is limited to 4.6 V maximum.
3. The package thermal impedance is calculated in accordance with JESD 51 .

## recommended operating conditions (see Note 4)

|  |  |  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | Supply voltage | Operating | 1.4 | 3.6 | V |
|  |  | Data retention only | 1.2 |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | $0.65 \times \mathrm{V}_{\mathrm{CC}}$ |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 1.7 |  |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 2 |  |  |
| VIL | Low-level input voltage | $\mathrm{V}_{\mathrm{CC}}=1.2 \mathrm{~V}$ |  | GND | V |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | $0.35 \times \mathrm{V}_{\text {CC }}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | $0.35 \times \mathrm{V}_{\mathrm{CC}}$ |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 0.7 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 0.8 |  |
| $\mathrm{V}_{1}$ | Input voltage |  | 0 | 3.6 | V |
| $\mathrm{V}_{\mathrm{O}}$ | Output voltage | Active state | 0 | $\mathrm{V}_{\mathrm{CC}}$ | V |
|  |  | 3-state | 0 | 3.6 |  |
| IOHS | Static high-level output current ${ }^{\dagger}$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | -2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | -4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | -12 |  |
| Iols | Static low-level output current $\dagger$ | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 1.6 V |  | 2 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V |  | 4 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V |  | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V |  | 12 |  |
| $\Delta t / \Delta v$ | Input transition rise or fall rate | $\mathrm{V}_{\mathrm{CC}}=1.4 \mathrm{~V}$ to 3.6 V |  | 5 | ns/V |
| TA | Operating free-air temperature |  | -40 | 85 | ${ }^{\circ} \mathrm{C}$ |

$\dagger$ Dynamic drive capability is equivalent to standard outputs with $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24 \mathrm{~mA}$ at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. See Figure 1 for $\mathrm{V}_{\mathrm{OL}} \mathrm{vs} \mathrm{I}_{\mathrm{OL}}$ and $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ characteristics. Refer to the TI application reports, AVC Logic Family Technology and Applications, literature number SCEA006, and Dynamic Output Control (DOC) Circuitry Technology and Applications, literature number SCEA009.
NOTE 4: All unused inputs of the device must be held at $\mathrm{V}_{\mathrm{CC}}$ or GND to ensure proper device operation. Refer to the TI application report, Implications of Slow or Floating CMOS Inputs, literature number SCBA004.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

$\dagger$ Typical values are measured at $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ and $3.3 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
$\ddagger$ For I/O ports, the parameter loz includes the input leakage current.
timing requirements over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

switching characteristics over recommended operating free-air temperature range (unless otherwise noted) (see Figures 2 through 5)

| PARAMETER | FROM (INPUT) | TO (OUTPUT) | $\mathrm{V}_{\mathrm{Cc}}=1.2 \mathrm{~V}$ | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=1.5 \mathrm{~V} \\ \pm 0.1 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ \pm 0.15 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ \pm 0.2 \mathrm{~V} \end{gathered}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ \pm 0.3 \mathrm{~V} \end{gathered}$ |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | MIN | MAX | MIN | MAX | MIN | MAX | MIN | MAX |  |
| $f_{\text {max }}$ |  |  |  |  |  |  |  |  |  |  |  | MHz |
| ${ }^{\text {tpd }}$ | A or B | B or A |  |  |  |  |  |  |  |  |  | ns |
|  | LE | A or B |  |  |  |  |  |  |  |  |  |  |
|  | CLK |  |  |  |  |  |  |  |  |  |  |  |
| $\mathrm{t}_{\mathrm{en}}$ | OEAB | B |  |  |  |  |  |  |  |  |  | ns |
| ${ }_{\text {dis }}$ | OEAB | B |  |  |  |  |  |  |  |  |  | ns |
| $\mathrm{t}_{\text {en }}$ | OEBA | A |  |  |  |  |  |  |  |  |  | ns |
| ${ }_{\text {dis }}$ | $\overline{O E B A}$ | A |  |  |  |  |  |  |  |  |  | ns |

operating characteristics, $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER |  |  | TEST CONDITIONS |  | $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ | $\mathrm{V}_{\mathrm{Cc}}=3.3 \mathrm{~V}$ | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | TYP | TYP | TYP |  |
| $\mathrm{C}_{\mathrm{pd}}$ | Power dissipation capacitance | Outputs enabled |  |  | $C_{L}=0, \quad f=10 \mathrm{MHz}$ |  |  |  |  | pF |
|  |  | Outputs disabled |  |  |  |  |  |  |  |

PARAMETER MEASUREMENT INFORMATION
$\mathrm{V}_{\mathrm{C}}=1.2 \mathrm{~V}$ AND $1.5 \mathrm{~V} \pm 0.1 \mathrm{~V}$


| TEST | S1 |
| :---: | :---: |
| $\begin{gathered} \mathrm{t}_{\mathrm{pd}} \\ \mathrm{t}_{\mathrm{PLZ}} / \mathrm{t}_{\mathrm{PZL}} \\ \mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}} \end{gathered}$ | Open $2 \times V_{C C}$ GND |



NOTES: A. $\mathrm{C}_{\mathrm{L}}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tPHZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.

Figure 2. Load Circuit and Voltage Waveforms

## PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \pm 0.15 \mathrm{~V}$



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| $\mathrm{tPLZ}^{\prime} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{t}_{\text {PZH }}$ | GND |



VOLTAGE WAVEFORMS SETUP AND HOLD TIMES

Input


VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tplz and tphz are the same as $t_{\text {dis }}$.
F. tPZL and tpZH are the same as ten.
G. tPLH and tPHL are the same as tpd.

Figure 3. Load Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION

$$
V_{C C}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}
$$



| TEST | S1 |
| :---: | :---: |
| tpd | Open |
| tpLz/tPZL | $2 \times \mathrm{V}$ C |
| tPHZ/tPZH | GND |



Figure 4. Load Circuit and Voltage Waveforms

# PARAMETER MEASUREMENT INFORMATION <br> $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$ 



| TEST | S1 |
| :---: | :---: |
| $\mathrm{t}_{\text {pd }}$ | Open |
| t PLZ $^{\prime}$ tPZL | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\text {PHZ }} / \mathrm{tPZH}^{2}$ | GND |



VOLTAGE WAVEFORMS
PROPAGATION DELAY TIMES



VOLTAGE WAVEFORMS
PULSE DURATION
VOLTAGE WAVEFORMS
PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tphZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tPZH are the same as ten.
G. tPLL and $\mathrm{tPHL}^{\text {are }}$ the same as $\mathrm{t}_{\mathrm{pd}}$.

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# AVC Logic Family Technology and Applications 

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#### Abstract

Texas Instruments ( $\mathrm{TI}^{T M}$ ) announces the industry's first logic family to achieve maximum propagation delays of less than 2 ns at 2.5 V . TI's next-generation logic is the Advanced Very-low-voltage CMOS (AVC) family. Although optimized for $2.5-\mathrm{V}$ systems, AVC logic supports mixed-voltage systems because it is compatible with $3.3-\mathrm{V}$ and $1.8-\mathrm{V}$ devices. The AVC family features TI's Dynamic Output Control ( $\mathrm{DOC}^{\top M}$ ) circuit (patent pending). The DOC circuit provides enough current to achieve high signaling speeds, but automatically lowers the output impedance of the circuit during a signal transition and subsequently increases the impedance to reduce the overshoot and undershoot noise that is often found in high-speed logic. This feature of AVC logic eliminates the need for series damping resistors. AVC logic also has a power-off feature that disables outputs from the device when no power is applied.


## Introduction

Current trends in advanced digital electronics design continue to include lower power consumption, lower supply voltages, faster operating speeds, smaller timing budgets, and heavier loads. Many designs are making the transition from 3.3 V to 2.5 V , and bus speeds are increasing beyond 100 MHz . Encompassing all these goals makes the requirement of signal integrity more difficult to achieve. For designs that require very-low-voltage logic and bus-interface functions, TI produces a new logic family that designers of next-generation high-performance workstations, PCs, networking, and telecommunications equipment find particularly useful.

## AVC Family

TI's next-generation logic family is AVC (see Figure 1). As part of TI's Widebus ${ }^{\text {TM }}$ and Widebus ${ }^{\text {TM }}$ families, these devices give designers an easy migration path to higher performance and lower voltages. Also offered in the AVC family are a broad line of logic gates and octal bus-interface functions. The devices in TI's AVC family are available in multiple JEDEC-standard advanced packages to provide maximum flexibility in board layout and cost.

DOC, TI, Widebus, and Widebus+ are trademarks of Texas Instruments Incorporated.


Figure 1. Low-Voltage Logic Family Performance Positioning

## Unparalleled Performance

TI's AVC family is the industry's first logic family to achieve maximum propagation delays of less than 2 ns at 2.5 V . This premier performance is achieved through a combination of advances. The family was designed for high performance, incorporating several novel circuit structures and changes to conventional logic-circuit designs. TI's advanced 0.5-micron Enhanced-Performance Implanted CMOS (EPIC ${ }^{\top M}$ ) fabrication process is used to produce the new devices.

## Novel Output Structure

The AVC family features TI's DOC circuit, which changes output impedance during switching (see Figure 2). The DOC circuit allows a single device to have the desirable characteristics of reduced noise, similar to damping-resistor outputs during static conditions, and high drive similar to a low-impedance output during dynamic conditions. The DOC circuit controls overshoots and undershoots and limits noise, which are inherent in high-speed, high-current devices.

EPIC is a trademark of Texas Instruments Incorporated.


Figure 2. Impedance Changes Through Switching Transitions

## Mixed-Voltage Mode and Power Off

The AVC family is optimized for low-power $2.5-\mathrm{V}$ systems and effectively supports mixed-voltage systems because it is compatible with $3.3-\mathrm{V}$ and $1.8-\mathrm{V}$ devices. AVC device inputs and outputs are $3.6-\mathrm{V}$ tolerant at $2.5-\mathrm{V}$ and $1.8-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$. This provides a bidirectional data path between 3.3-V LVTTL and $2.5-\mathrm{V}$ CMOS, and a one-way data path from 3.3-V LVTTL or 2.5-V CMOS to $1.8-\mathrm{V}$ CMOS. AVC logic also has a power-off isolation feature that disables outputs from the device during system partial power down.

## Design Issues and AVC Family Solutions

## Low Power (Optimized for 2.5 V )

Perhaps one of the most pervasive trends in advanced digital-electronics design is lower power consumption. Lower power consumption is especially important to extend battery life of portable equipment. Reduced heat dissipation from lower power consumption simplifies the measures necessary to remove heat and decrease the necessary packaging area, leading to production of smaller and less expensive products. One of the most effective ways to reduce power dissipation is to decrease integrated-circuit operating voltages. The AVC family, designed to operate at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, enables high-performance, low-power, and advanced designs. Not simply a scaled-down 3.3-V family, AVC is the first logic family conceived and designed for optimized performance at 2.5 V .

## Unused and Undriven Inputs (Bus Hold)

A circuit element that must be addressed when designing with a CMOS family, such as AVC, is circuit inputs. With the totem-pole structure (see Figure 3) that characterizes the inputs of CMOS devices, the input node must be held as close to the $\mathrm{V}_{\mathrm{CC}}$ or GND rails as possible.


Figure 3. Totem-Pole Input Structure
Precautions should be taken to prevent the input voltage from floating near the threshold voltage because this biases both input transistors on and creates undesirably high $\mathrm{I}_{\mathrm{CC}}$ currents at the $\mathrm{V}_{\mathrm{CC}}$ pin of the device. Under certain conditions, this can damage the device. One way to address this concern is to place external pullup resistors at any input that might be in a high-impedance, undriven state. This is costly in terms of component count, reliability, and board area. An alternative solution is to employ the devices in the AVC family that utilize the optional bus-hold circuit at the inputs (see Figure 4). AVC devices with bus-hold circuitry are designated as AVCH.


Figure 4. Typical Bus-Hold Cell
The bus-hold circuit consists of two series inverters with the output fed back to the input through a resistor. This provides a weak positive feedback by sinking or sourcing current to the input node. The bus-hold cell holds the input at its last-known valid logic state until forcibly changed by a driving circuit. Figure 5 shows the input characteristics of bus hold as the input voltage is swept from 0 V to 2.5 V . These characteristics are similar to a weak bistable latch. The bus-hold cell sinks current when the input is low, and sources current when the input is high. When the input voltage is near the threshold, the circuit sinks or sources maximum current to force the input node toward either the $\mathrm{V}_{\mathrm{CC}}$ or GND rail.


Figure 5. Bus Hold Across $\mathrm{V}_{\mathrm{CC}}$
Generally, pullup and pulldown resistors should not be used on the inputs of devices with bus hold. In applications that require pullup or pulldown resistors to hold the inputs at a specific logic level, the $\mathrm{I}_{\mathrm{I}(\mathrm{hold})}$ maximum specification should be considered. The resistor value should be chosen to overcome bus hold under worst-case conditions. The resistor must supply enough current so that the input is pulled through the threshold to the desired logic level. If the current supplied is too weak, the input node could be held near the threshold, causing a high $\mathrm{I}_{\mathrm{CC}}$ that could damage the part.

## Partial Power-Down and Mixed-Voltage-Mode Data Communication

The inputs and outputs of the AVC family have been designed with all reverse-current paths to $\mathrm{V}_{\mathrm{CC}}$ blocked. This low $\mathrm{I}_{\mathrm{OFF}}$ current feature allows the device to remain electrically connected to a bus during partial power down without loading the remaining live circuits. This feature also allows the use of this family in a mixed-voltage environment. If the inputs or outputs are at a voltage greater than the $\mathrm{V}_{\mathrm{CC}}$ of the device, there is no current sourcing back through the device from the higher voltage node to the lower-voltage $\mathrm{V}_{\mathrm{CC}}$ supply.

With a bidirectional AVC transceiver powered with $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$, two-way data communication between 3.3-V LVTTL devices and $2.5-\mathrm{V}$ CMOS devices can occur (see Figure 6). The inputs of the AVC part are 3.6-V tolerant and accept the LVTTL switching levels. The outputs of the AVC part, when powered at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ under worst-case conditions, are accepted as valid switching levels at the input of a 3.3-V LVTTL device.

With a unidirectional AVC driver powered with $1.8-\mathrm{VV}_{\mathrm{CC}}$, data communication from $2.5-\mathrm{V}$ or $3.3-\mathrm{V}$ signal levels to $1.8-\mathrm{V}$ devices can occur (see Figure 7). The inputs of the AVC part are tolerant of the higher voltages and accept the higher switching levels. The outputs of the AVC driver are valid $1.8-\mathrm{V}$ signal levels.


Figure 6. Device at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{Cc}}$ With $3.3-\mathrm{V}$ I/Os on One Side and $2.5-\mathrm{V} \mathrm{I} / \mathrm{Os}$ on the Other, Showing Switching Levels


Figure 7. Device at $1.8-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ With $2.5-\mathrm{V}$ Inputs or $3.3-\mathrm{V}$ Inputs, Showing Switching Levels

## Device Characteristics

To facilitate a preliminary analysis of the characteristics of the AVC family, SPICE analysis graphs from TI's initial AVC-family device, the SN74AVC16245 16-bit bus transceiver with 3-state outputs are shown in Figures 8 through 22. These analyses are the outputs of SPICE simulations using standard loads specified in the parameter measurement information illustrations in Appendix A, unless otherwise noted.

## Power Consumption

Figure 8 presents SPICE information about the device dynamic power consumption across the operating frequencies. Table 1 shows modeled values of power dissipation capacitance $\left(\mathrm{C}_{\mathrm{pd}}\right)$. The $\mathrm{C}_{\mathrm{pd}}$ data were obtained using an input edge rate of 1 ns $(0 \%-100 \%)$, open-circuit load on the output, and one output switching with a 48-pin TSSOP (DGG) package.


Figure 8. Icc vs Frequency With 1, 8, or 16 Outputs Switching
Table 1. $\mathrm{C}_{\mathrm{pd}}$ for Various Conditions, One Output Switching

| PARAMETER | TEST CONDITIONS $C_{L}=0, f=10 \mathrm{MHz}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{CC}}=1.8 \mathrm{~V} \\ & \pm 0.15 \mathrm{~V} \text { TYP } \end{aligned}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \\ & \pm 0.2 \mathrm{~V} \text { TYP } \end{aligned}$ | $\begin{aligned} & \hline \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V} \\ & \pm 0.3 \mathrm{~V} \text { TYP } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{C}_{\mathrm{pd}}$ | Outputs enabled | 15.9 pF | 18.1 pF | 21.1 pF |
| $\mathrm{C}_{\text {pd }}$ | Outputs disabled | $\sim 1 \mathrm{pF}$ | $\sim 1 \mathrm{pF}$ | $\sim 1 \mathrm{pF}$ |

## Input Characteristics

Figures 9 and 10 present SPICE information about the device static behavior. Figure 9 shows the device supply-current requirements across input voltage and Figure 10 shows the output-voltage versus input-voltage transfer curves.


Figure 9. $\mathrm{I}_{\mathrm{CC}}$ vs $\mathrm{V}_{\mathrm{I}}$


Figure 10. $\mathrm{V}_{\mathrm{O}}$ vs $\mathrm{V}_{\mathrm{I}}$

## Switching Performance

Figures 11 through 16 present SPICE models of the device dynamic behavior. Propagation delay times across various conditions of ambient temperature, load capacitance with one output switching, and load capacitance with 16 outputs switching are shown.


Figure 11. $\mathrm{t}_{\mathrm{PHL}}$ vs $\mathrm{T}_{\mathbf{J}}$


Figure 12. $\mathrm{t}_{\mathrm{PLH}}$ vs $\mathrm{T}_{\mathrm{J}}$


Figure 13. t $_{\text {PHL }}$ vs Load Capacitance, One Output Switching


Figure 14. $\mathrm{t}_{\text {PLH }}$ vs Load Capacitance, One Output Switching


Figure 15. $\mathrm{t}_{\mathrm{PHL}}$ vs Load Capacitance, 16 Outputs Switching


Figure 16. t PLH vs Load Capacitance, 16 Outputs Switching

## Signal Integrity

Perhaps the most important measure of a device's performance in the dynamic domain is the effect of varying conditions upon signal integrity. Figures 17 through 20 show SPICE simulations of the device dynamic behavior. The effect of multiple outputs switching simultaneously on one that is held at a valid logic level is shown (see Figures 17 and 18). The effects of slow input-transition time (see Figure 19), and pin-to-pin skew (see Figure 20) are shown.


Figure 17. Simultaneous-Switching Voltage ( $\mathrm{V}_{\text {OLP }}, \mathrm{V}_{\text {OLV }}$ ) vs Time


Figure 18. Simultaneous-Switching Voltage ( $\mathrm{V}_{\mathrm{OHP}}, \mathrm{V}_{\mathrm{OHV}}$ ) vs Time


Figure 19. Slow Input-Transition Time


Figure 20. Pin-to-Pin Skew ( $\mathrm{t}_{\text {PHL }}, \mathrm{t}_{\text {PLH }}$ ) (<100 ps nominal)

## Output Characteristics With DOC

Selecting a component with improved output drive characteristics simplifies the design engineer's job of ensuring signal integrity and meeting timing requirements. For signal integrity, the output must have an output impedance that minimizes overshoots and undershoots. A component with $26-\Omega$ series damping resistors on the output ports was sometimes necessary to improve the match of the impedance with the transmission-line load on the output of the buffer. The opposing characteristic that must be considered is having sufficient drive to meet the timing requirements. The AVC family features TI's DOC circuit that automatically lowers the output impedance of the circuit during a signal transition and subsequently raises the impedance to reduce overshoot and undershoot. Figures 21 and 22 contain typical voltage and current curves that illustrate the operation of the circuit as it transitions from one state to another.


Figure 21. $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{loL}_{\mathrm{OL}}$


Figure 22. $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$
The DOC circuitry provides enough drive current to achieve faster slew rates and meet timing requirements, but quickly switches the impedance level to reduce the overshoot and undershoot noise that is often found in high-speed logic. This feature of AVC logic eliminates the need for damping resistors in the output circuit, which are often used in series, and sometimes integrated with logic devices, to limit electrical noise. Damping resistors reduce the noise, but increase propagation delay due to the decreased drive current.

Because of the excellent signal integrity characteristics of the DOC output, transmission-line termination typically is unnecessary. Due to the high-impedance drive characteristics of the output in the static state, the use of dc termination is specifically discouraged. The output current that is required to bias a dc termination network could exceed the static-state output-drive capabilities of the device. AVC with DOC circuitry is ideally suited for any high-speed, point-to-point application or unterminated distributed load, such as high-speed memory interfacing.

## Design Support

Examination of the characteristics of the device is a critical portion of a successful design. To aid the design engineer in analysis of device characteristics, the latest versions of IBIS models can be obtained from TI's website at http://www.ti.com. SPICE models are also available from TI. Please contact your local TI field sales representative for more information.

## Features and Benefits

Table 2 provides selected AVC family features and benefits.
Table 2. Selected AVC Family Features and Benefits

| FEATURES | BENEFITS |
| :--- | :--- |
| Optimized for 2.5-V VCC | Enables low-power designs |
| Broad product offerings | Simplifies component choice |
| Advanced EPIC fabrication process; turbo-circuit design | Sub-2-ns (maximum) speeds at 2.5 V. <br> Easier to meet timing windows <br> in advanced high-speed designs |
| DOC outputs do not require series damping resistors internally or externally | Reduced ringing without series output resistors, <br> increased performance and cost savings |
| Bus-hold option | Eliminates pullup or pulldown resistors on inputs |
| IOFF - reverse-current paths to $\mathrm{V}_{\text {CC }}$ blocked on the inputs and outputs | Outputs disabled during power off for use in <br> partial power down and mixed-voltage designs |

## Conclusion

For designs that require $1.8-\mathrm{V}, 2.5-\mathrm{V}$, and $3.3-\mathrm{V}$ logic functions with the highest performance, the AVC family provides the fastest, quietest logic devices optimized for $2.5-\mathrm{V}$ and unterminated load conditions. AVC offers a broad line of Widebus and Widebus+ functions, logic gates, and octal bus-interface functions.

## Acknowledgment

The authors of this application report are Stephen M. Nolan and Tim Ten Eyck.

## Glossary

## A

| AVC | Advanced very-low-voltage CMOS |
| :--- | :--- |
| CMOS | Complementary metal-oxide semiconductor |
| DOC | Dynamic output control (patent pending) |
| E |  |
| EPIC | Enhanced-performance implanted CMOS |
| IBIS | I/O buffer information specification |
| $\mathrm{I}_{\mathrm{I}}$ | Input current |
| $\mathrm{I}_{\text {I(hold) }}$ | Input current (bus hold) <br> $\mathrm{I}_{\mathrm{OH}}$ |
| $\mathrm{I}_{\mathrm{OL}}$ | High-level output current |
| Low-level output current |  |

LVTTL Low-voltage TTL (3.3-V power supply and interface levels)

## P

PC Personal computer

SPICE Simulation program with integrated-circuit emphasis

## T

$t_{\text {pd }} \quad$ Propagation delay time
${ }^{t}$ PHL $\quad$ Propagation delay time, high- to low-level output
tpLH Propagation delay time, low- to high-level output

TSSOP Thin shrink small-outline package

TTL Transistor-transistor logic
V
$\mathrm{V}_{\mathrm{OH}} \quad$ High-level output voltage
VOL Low-level output voltage
$\mathrm{V}_{\text {OHP }} \quad$ High-level output voltage peak
V OHV High-level output voltage valley
V OLP Low-level output voltage peak

VOLV Low-level output voltage valley

## Appendix A - Parameter Measurement Information



| TEST | S1 |
| :---: | :---: |
| ${ }^{\text {tpd }}$ | Open |
| tPLZ/tPZL | $2 \times \mathrm{V}$ C $C$ |
| ${ }^{\text {tPHZ }}$ / ${ }^{\text {PRZH }}$ | GND |



NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{p L Z}$ and tphZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tpZH are the same as ten.
G. $\mathrm{tpLH}^{\text {and }} \mathrm{tPHL}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure A-1. AVC Parameter Measurement Information (1.8 $\mathrm{V} \pm 0.15 \mathrm{~V}$ )


| TEST | S 1 |
| :---: | :---: |
| $\mathrm{t}_{\mathrm{pd}}$ | Open |
| $\mathrm{t}_{\mathrm{PLZ}} / \mathrm{tPZL}$ | $2 \times \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{t}_{\mathrm{PHZ}} / \mathrm{t}_{\mathrm{PZH}}$ | GND |



VOLTAGE WAVEFORMS PROPAGATION DELAY TIMES


VOLTAGE WAVEFORMS PULSE DURATION


VOLTAGE WAVEFORMS
ENABLE AND DISABLE TIMES

NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $\quad t P L Z$ and $t P H Z$ are the same as $t_{d i s}$.
F. $t_{P Z L}$ and $t_{P Z H}$ are the same as ten.
G. $t_{P L H}$ and $t_{P H L}$ are the same as $t_{p d}$.

Figure A-2. AVC Parameter Measurement Information ( $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$ )


NOTES: A. $\mathrm{C}_{\mathrm{L}}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. $t_{P L Z}$ and $t_{P H Z}$ are the same as $t_{\text {dis }}$.
F. tPZL and tPZH are the same as ten.
G. tPLH and tPHL are the same as $t_{p d}$.

Figure A-3. AVC Parameter Measurement Information ( $\mathrm{V}_{\mathrm{Cc}}=3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$ )

# Dynamic Output Control (DOC ${ }^{\text {™ }}$ ) Circuitry Technology and Applications 

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#### Abstract

Texas Instruments ( $\mathrm{TI}^{\mathrm{TM}}$ ) next-generation logic is called the Advanced Very-low-voltage CMOS (AVC) family. The AVC family features TI's Dynamic Output Control ( $\mathrm{DOC}^{\text {TM }}$ ) circuit (patent pending). DOC circuitry automatically lowers the output impedance of the circuit at the beginning of a signal transition, providing enough current to achieve high signaling speeds, then subsequently raises the impedance to limit the overshoot and undershoot noise inherent in high-speed, high-current devices. This allows a single device to have characteristics similar to both series-damping-resistor outputs during static conditions and to high-current outputs during dynamic conditions, eliminating the need for series damping resistors. Due to the characteristics of the DOC output, the dc drive-current specifications for DOC devices are not useable as a relative indicator of the dynamic performance. A thorough understanding of static and dynamic drive-current conditions is required to design with the DOC feature of AVC logic.


## Introduction

## Performance

Trends in advanced digital electronics design continue to include lower power consumption, lower supply voltages, faster operating speeds, smaller timing budgets, and heavier loads. Many designs are making the transition from 3.3 V to 2.5 V , and bus speeds are increasing beyond 100 MHz . Trying to meet all of these goals makes the requirement of signal integrity harder to achieve. For designs that require very-low-voltage logic and bus-interface functions, TI announces the AVC family featuring TI's DOC circuit. The DOC circuit limits overshoot and undershoot noise inherent in high-speed, high-current devices, while still providing propagation delays of less than 2 ns , maximum, at 2.5 V .

## Impedance Matching

The design engineer must carefully consider a logic component's output characteristics to ensure signal integrity and meet timing requirements. The output must have an impedance that minimizes overshoots and undershoots for signal integrity. The opposing characteristic that must be considered is having sufficient drive to meet the timing requirements. In the past, the selection of a component with integrated $26-\Omega$ series damping resistors on the output ports or the use of external resistors was sometimes necessary. These resistors improve the impedance match of the driver output with the impedance of the transmission-line load and limit overshoot and undershoot noise. Damping resistors reduce the noise, but decrease slew rate and increase propagation delay due to the decreased drive current.

TI's DOC circuitry provides enough drive current to achieve fast slew rates and meet timing requirements, but quickly changes the output impedance level during the output transition to reduce the overshoot and undershoot noise that often is found in high-speed logic. This feature of AVC logic eliminates the need for series damping resistors in the output circuit, thereby improving the output slew rate and propagation-delay characteristics.
The dynamic drive current varies through the transition due to the dynamically changing output impedance. The static on-resistance $\left(\mathrm{R}_{\mathrm{ON}}\right)$ of the output can be calculated from the $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ curves (see Figure 1 and Figure 2). At any specific point on the $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves, $\mathrm{R}_{\mathrm{ON}}=\left(\mathrm{V}_{\mathrm{OH}}-\mathrm{V}_{\mathrm{CC}}\right) / \mathrm{I}_{\mathrm{OH}}$. At any specific point on the $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ curves, $\mathrm{R}_{\mathrm{ON}}=\mathrm{V}_{\mathrm{OL}} / \mathrm{I}_{\mathrm{OL}}$. The impedance during dynamic conditions is characterized by the slope of the $\mathrm{V}_{\mathrm{O}}$ vs $\mathrm{I}_{\mathrm{O}}$ line at any specific point on the graph.

The $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ curves (see Figure 1) illustrate the impedance characteristics of the output in the low state. The curves represent the amount of sink current available (at a given $\mathrm{V}_{\mathrm{CC}}$ ) to drive the load, as the output voltage decreases from $\mathrm{V}_{\mathrm{CC}}$ to 0 V when the output is sinking current (i.e., driving low). The $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ curve for $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ has two distinct regions of sink current availability. At the beginning of the transition from high to low, the portion of the output from $2.5-\mathrm{V}$ to $1.5-\mathrm{V}$ has a high amount of sink current available. In that region, the curve has characteristics that are similar to a circuit with an output resistance of approximately $20 \Omega$. Then, during the transition through 1.5 V , there is a steep drop in the drive current available. In the region from 1.5 V to ground, the curve has characteristics that are similar to a circuit with an output resistance of approximately $50 \Omega$. The $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ curves for $1.8-\mathrm{V}$ and $3.3-\mathrm{V}_{\mathrm{CC}}$ have similar characteristics.


Figure 1. $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{l}_{\mathrm{OL}}$
The $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves (see Figure 2) illustrate the impedance characteristics of the output in the high state. The curves represent the amount of source current available (at a given $\mathrm{V}_{\mathrm{CC}}$ ) to drive the load, as the output voltage increases from 0 V to $\mathrm{V}_{\mathrm{CC}}$ when the output is sourcing current (i.e., driving high). The operation of the output in the high state is similar to the operation in the low state. There are two distinct regions of source current availability, each with an output resistance (at $2.5-\mathrm{V}$ $\mathrm{V}_{\mathrm{CC}}$ ) of approximately $30 \Omega$ and $50 \Omega$, respectively. The $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves for $1.8-\mathrm{V}$ and $3.3-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ have similar characteristics.


Figure 2. $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{IOH}_{\mathrm{OH}}$
The dual-impedance regions of the DOC output allow a single device to have characteristics similar to a $\pm 24$-mA high-drive device, providing fast edge rates and propagation-delay times. During the latter portion of the transition and during static conditions, the device has the characteristics of a series-damping-resistor part, with reduced ringing. Figure 3 illustrates the dual-impedance nature of the DOC output as compared to the fixed-impedance outputs of both a high-drive part and a series-damping-resistor part by showing the $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ curves of all three.


Figure 3. DOC Output Curve Superimposed on Resistor-Output and High-Drive-Output Curves

## Output Circuitry

## What Happens at the Output in the Transition

A standard device with a fixed low-impedance output delivers high current to the load during the entire transition. At the top of the transition from low to high, high-drive circuits can experience a tremendous overshoot and ringing due to the fast slew rate (see Figure 4). The DOC circuit counteracts this by switching to a higher output impedance, thereby slowing the slew rate as the output approaches the top of the transition.


Figure 4. Switching Transition of a Fixed Low-Impedance Driver
Figure 5 illustrates the output of the DOC driver in the transition from low to high. Initially, the output is at a static low level. The $2.5-\mathrm{V}_{\mathrm{OL}}$ vs I $\mathrm{OL}_{\mathrm{OL}}$ impedance-characteristic curve (see Figure 1) shows that, with an output at 0 V , the output resistance in the low state is approximately $50 \Omega$. When the transition from low to high begins, the $2.5-\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curve (see Figure 2 ) illustrates the impedance characteristics of the output. Initially, the output resistance is approximately $30 \Omega$. Under typical conditions, this low-impedance output can deliver nearly 84 mA to the load, providing a very fast slew rate. After the output voltage passes through the threshold $(1.5 \mathrm{~V})$ in the transition from low to high, the output resistance is switched from approximately $30 \Omega$ to approximately $50 \Omega$. This increase in output resistance reduces the amount of drive current available. This decreases the slew rate and rolls off the transition, producing a smooth knee at the top and reducing overshoot or ringing. When the final output voltage is reached, due to the high output resistance, the amount of drive current available to hold the output voltage at a valid logic level is at a minimum, providing relatively low static-state power levels.


Figure 5. Impedance Through Switching Transitions
A transition from high to low behaves in a similar manner and can be understood by the same principles. When the transition from high to low begins, the $2.5-\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ curve (see Figure 1) illustrates the impedance characteristics of the output. Initially, the output resistance is approximately $20 \Omega$. Under typical conditions, this low output impedance can deliver nearly $105-\mathrm{mA}$ to the load. Then, as the output voltage passes through the threshold ( 1.5 V ), the output resistance is switched from approximately $20 \Omega$ to approximately $50 \Omega$. This results in minimal, or no undershoot.

## DOC Circuit Description

Figure 6 shows a simplified output stage of a typical logic circuit. When the input is low, the $n$-channel transistor $\left(\mathrm{Q}_{\mathrm{n}}\right)$ turns off and the p-channel transistor $\left(Q_{p}\right)$ turns on and begins to conduct, and the output voltage $\mathrm{V}_{\mathrm{O}}$ is pulled high. Conversely, when the input is high, $\mathrm{Q}_{\mathrm{p}}$ turns off, $\mathrm{Q}_{\mathrm{n}}$ begins to conduct, and $\mathrm{V}_{\mathrm{O}}$ is pulled low. This action is similar to an inverter, and several of these inverting stages typically are cascaded in series to form a buffer/driver.


Figure 6. Simplified Totem-Pole Output Stage

The sizes of the output transistors $\mathrm{Q}_{\mathrm{p}}$ and $\mathrm{Q}_{\mathrm{n}}$ determine the output impedance. The transistors are designed with the sizes of the n-channel FET and p-channel FET selected to provide an output impedance of a specific design value. The sizes can be selected so that the on-resistance of the output is, for example, characteristically approximately $25 \Omega$, which is the typical output impedance of a conventional low-voltage CMOS logic device. Figure 7 illustrates a driver with the output transistors sized to provide a $25-\Omega$ output. The driver is shown driving a transmission-line load consisting of a length of transmission line that is terminated into a capacitor. The waveform showing the signal incident at the capacitor depicts the fast slew rates and small propagation delays that are characteristic of low-impedance drivers. The fast edge rates create large overshoots and unacceptable ringing.


Figure 7. 25- $\Omega$ Driver Driving Transmission-Line Load and Waveform at the Load
One method of reducing the ringing and electrical noise is to slow down the edge rates. This can be accomplished by the addition of a damping resistor in series with the output. This creates a high-impedance low-drive output. Figure 8 illustrates a driver with a $25-\Omega$ output and a series $26-\Omega$ damping resistor driving the transmission-line load. The resultant signal is much cleaner, but the slower edge rate increases the propagation delay time. Depending on the total timing budget available, this could be an unacceptable solution. Series resistors also can raise the dc low-voltage level of a signal. This reduces noise immunity of the receiving logic. Finally, series damping resistors should be used only on point-to-point nets, and never with distributed loads, because of the half voltage that propagates down the transmission line due to incident wave switching.


Figure 8. $25-\Omega$ Driver and $26-\Omega$ Series Resistor Driving Transmission-Line Load and Waveform at the Load

Another method that can be used to improve the impedance match of the output with the load is to reduce the size of the output transistors. If their sizes are decreased, the output impedance increases. This provides a low-drive output. Figure 9 illustrates a driver with the output transistor sizes selected to provide a $50-\Omega$ output. The driver is shown driving the same transmission-line load and the resultant waveform at the load exhibits similar characteristics to the series-damping-resistor version.


Figure 9. 50- $\Omega$ Driver Driving Transmission-Line Load and Waveform at the Load
It is also interesting to explore the attributes of two drivers in parallel. Figure 10 represents two $50-\Omega$ drivers in parallel. The resultant waveform at the load exhibits characteristics similar to the single $25-\Omega$ driver. In fact, the parallel combination of the two has the same output impedance as a single $25-\Omega$ impedance driver. This effectively creates a low-impedance high-drive output.


Figure 10. Two 50- $\Omega$ Drivers In Parallel, Driving Transmission-Line Load and Waveform at the Load
Increasing the output impedance reduces overshoots and undershoots, but at the cost of increased propagation delays. Decreasing the output impedance decreases propagation delays, but at the cost of increased overshoots and undershoots. A desirable circuit would have a low output impedance for the beginning portion of the output transition and a high output impedance for the latter portion of the output transition. This would provide fast propagation delays, with minimal, or no overshoot or undershoot.

Figure 11 is a block diagram of the DOC circuit, which consists of a fixed driver with a nominal $50-\Omega$ on-resistance. The $50-\Omega$ driver functions like a typical high-impedance low-drive output, with good electrical and noise characteristics. In parallel with the $50-\Omega$ driver is a controllable $50-\Omega$ nominal on-resistance driver, with an output that can be enabled or disabled similar to the output of a 3-state device. When a device is disabled, its output is in a very high-impedance state and contributes nothing to the drive or to the loading of the output. When it is enabled, the parallel combination of the $50-\Omega$ drivers has the same output characteristics as a single 25- $\Omega$ impedance driver. This effectively creates a low-impedance high-drive output. The impedance control circuit (ZCC) enables and disables the controllable driver by controlling its ON signal. The ZCC monitors the output and controls the controllable driver at the appropriate times during the signal transition to achieve a high-drive, fast slew-rate transition.


Figure 11. DOC Circuit Driving Transmission-Line Load and Waveform at the Load
The operation of the DOC begins with the output in a static state, for example, at a logic low state. In the static low state, the ZCC has the controllable $50-\Omega$ driver disabled and the n channel of the fixed $50-\Omega$ driver sinks current to ground from the output. When the input transitions from low to high, the n-channel transistor in the fixed $50-\Omega$ driver turns off, and the p channel turns on, sourcing current to the output and beginning the output transition from low to high. Simultaneously, the ZCC enables the p channel in the controllable $50-\Omega$ driver. The parallel p channels of the drivers have a combined on-resistance of approximately $25 \Omega$. This low impedance provides a high drive current to cause a fast slew-rate signal transition. The ZCC senses the output voltage, and as the voltage passes through threshold in the transition from low to high, the ZCC disables the output p channel of the controllable $50-\Omega$ driver. The increase in output impedance decreases the slope and rolls off the output signal, reducing the overshoot.

The operation of the high-to-low transition is similar.

## AC Dynamic Drive vs DC Static Drive

The dc drive-current ratings in the recommended operating-conditions table of a device data sheet typically are selected to show the static-drive capability of a device when the output voltage is at a worst-case valid logic level, such as $\mathrm{V}_{\mathrm{OH}(\mathrm{MIN})}$ or $\mathrm{V}_{\mathrm{OL}(\mathrm{MAX})}$. Historically, these dc drive-current ratings were used as a relative measure of a component's ac dynamic-drive performance. For a device with a fixed output on-resistance, this was an acceptable method, because the dc current at a given logic level could be extrapolated to determine the amount of ac drive current available through the transition.

With DOC circuitry, the output impedance characteristics change dynamically during a transition. The dc drive-current specification is not a useable indicator of the devices' dynamic performance capability. The dc output ratings of DOC devices (see Table 1) can be used loosely as a relative comparison to the dc output ratings of devices with integral series damping resistors (see Table 2), and this is a good indication of the DOC circuit's excellent low-noise and low-power characteristics. However, unlike a part with a fixed low-drive output, the DOC circuitry provides good ac performance. The DOC output provides a very strong ac drive during dynamic conditions, capable of driving very heavily capacitive CMOS loads.

Table 1. Recommended Static Output Current for DOC Circuits ${ }^{1}$

|  |  |  | MIN MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: |
| IOHS | Static high-level output current | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | -4 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | -8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | -12 |  |
| Iols | Static low-level output current | $\mathrm{V}_{\mathrm{CC}}=1.65 \mathrm{~V}$ to 1.95 V | 4 | mA |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 8 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 12 |  |

Table 2. Recommended Output Current for ALVC Device With Damping Resistor ${ }^{2}$

|  |  | MIN MAX | UNIT |
| :---: | :---: | :---: | :---: |
| $\mathrm{IOH} \quad$ High-level output current | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ | -6 | mA |
|  | $\mathrm{V}_{\mathrm{CC}}=2.7 \mathrm{~V}$ | -8 |  |
|  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ | -12 |  |
| Low-level output current | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ | 6 | mA |
|  | $\mathrm{V}_{\mathrm{CC}}=2.7 \mathrm{~V}$ | 8 |  |
|  | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ | 12 |  |

The DOC device performs like a high-drive part during signal transition. Under typical conditions at 2.5-V $V_{C C}$, the drive current that is available during the beginning of a transition from low to high is about 84 mA , and from high to low is about 105 mA . Figure 12 illustrates the output current of the DOC circuit driving a standard load through the low-to-high and high-to-low transitions. Note the large peak currents during the transition.


Figure 12. DOC Device Output Current Through the Transition
The dynamic drive current is not specified on the data sheet for devices with DOC outputs because of its transient nature, but it is similar to the dynamic drive current that is available from a $\pm 24-\mathrm{mA}$ (at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ ) high-drive standard-output device (see Figure 13).


Figure 13. Output Current Through the Transition, $\pm 24$-mA High-Drive Standard-Output Device
Because a typical CMOS load is purely capacitive, with very little bias (leakage) current necessary to hold a valid static logic level, the amount of dc drive required of most drivers is small. The dc drive is specified on the data sheet of DOC output devices. The output parameters are static and testable values that are enumerated in terms of minimum and maximum output voltages at specific output currents (see Table 3).
Table 3. Output Voltage Characteristics Over Recommended Operating Free-Air Temperature Range ${ }^{1}$

| PARAMETER | TEST CONDITIONS |  | $\mathrm{V}_{\text {cc }}$ | MIN | TYP MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{OH}}$ | $\mathrm{I} \mathrm{OH}=-100 \mu \mathrm{~A}$ |  | 1.65 V to 3.6 V | $\mathrm{V}_{\mathrm{CC}}-0.2$ |  | V |
|  | $\mathrm{IOH}=-4 \mathrm{~mA}$, | $\mathrm{V}_{\text {IH }}=1.07 \mathrm{~V}$ | 1.65 V | 1.2 |  |  |
|  | $\mathrm{IOH}=-8 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=1.7 \mathrm{~V}$ | 2.3 V | 1.75 |  |  |
|  | $\mathrm{I}^{\mathrm{OH}}=-12 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IH}}=2 \mathrm{~V}$ | 3 V | 2.3 |  |  |
|  | $\mathrm{l} \mathrm{OL}=100 \mu \mathrm{~A}$ |  | 1.65 V to 3.6 V |  | 0.2 |  |
|  | $\mathrm{l} \mathrm{OL}=4 \mathrm{~mA}$, | $\mathrm{V}_{\text {IL }}=0.57 \mathrm{~V}$ | 1.65 V |  | 0.45 |  |
|  | $\mathrm{IOL}=8 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.7 \mathrm{~V}$ | 2.3 V |  | 0.55 |  |
|  | $\mathrm{IOL}=12 \mathrm{~mA}$, | $\mathrm{V}_{\mathrm{IL}}=0.8 \mathrm{~V}$ | 3 V |  | 0.7 |  |

## Termination (AC vs DC)

Because of the excellent signal-integrity characteristics of the DOC output, transmission-line termination typically is unnecessary. Due to the high-impedance characteristics of the output in the static state, the use of dc termination is specifically discouraged. The output current that is required to bias a dc termination network could exceed the static-state output-drive capabilities of the device. AVC family devices with DOC circuitry are suited ideally for any high-speed, point-to-point application or unterminated distributed load, such as high-speed memory interfaces.

## Waveforms - Comparison of ALVCH Standard and Resistor Outputs

Figures 14 and 15 show the SPICE results comparing SN74AVC16827 with SN74ALVCH16827 and SN74ALVCH162827 into a standard lumped load (see Appendix A) for $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ and $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$, respectively. The results show the relative propagation delay and noise performance of the DOC circuit.


Figure 14. Outputs Driving a Standard Lumped Load, $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$


Figure 15. Outputs Driving a Standard Lumped Load, $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$

Figures 16 and 17 show the SPICE modeling of the SN74AVC16827 with the DOC circuit, an SN74ALVCH16827 with low-impedance output circuit, and an SN74ALVCH162827 with series damping resistors driving a PC100 DQM load for $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$ and $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$, respectively. The DQM load is defined in the Intel ${ }^{\text {TM }}$ PCSDRAM Registered DIMM Specification, Revision 1.0, February 19983. For this example, the $256-\mathrm{Mbyte}$ load was used. The transmission lines have a characteristic impedance of $70 \Omega$ The lengths of the transmission lines are specified in the PC100 specification; series resistor R1 was specified as zero. This resistor is not necessary when using the DOC circuit. The six SDRAM loads were modeled by the circuit shown in Figure 18.

The waveforms shown in Figures 16 and 17 were measured at the input to the memory devices. The low-impedance driver exhibits excessive overshoots and undershoots, while the DOC circuit and the driver with series damping resistors does not. The DOC circuit is faster than the series-damping-resistor circuit. This improvement in speed is more pronounced when the simulations are run under worst-case weak conditions.


Figure 16. Outputs Driving a PC100 Load Network, $\mathrm{V}_{\mathrm{CC}}=2.5 \mathrm{~V}$

[^7]

Figure 17. Outputs Driving a PC100 Load Network, $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}$


Figure 18. SDRAM Load Model

Features and Benefits
Table 4 summarizes DOC circuit features and some of the benefits of those features.
Table 4. Features and Benefits of DOC Circuitry

| FEATURES | BENEFITS |
| :--- | :--- |
| Optimized for 2.5-V V CC . No damping resistors | Enables low-power designs |
| Low-impedance, high-drive output during the beginning of a signal transition | Fast edge-rates and small propagation delays |
| High output impedance for the later portion of the ouput transition | Minimal, or no overshoot or undershoot |
| High-impedance, low-drive steady-state output after signal transition | Enables low-power designs |
| DOC outputs do not require series damping resistors internally or externally | Reduced ringing without series output resistors; <br> increased performance; cost savings |
| IOFF - reverse-current paths to $\mathrm{V}_{\mathrm{CC}}$ blocked | Outputs disabled during power off for use in partial <br> power-down designs |

## Conclusion

The DOC circuitry provides a low-impedance, high-drive output during the beginning of a signal transition, to provide fast edge rates and small propagation delays. Then, as the output passes through the threshold, the DOC switches to a high-impedance, low-drive output to roll off the signal and reduce ringing. The amount of static dc drive current specified in the data sheets of devices with DOC features does not reflect the large amount of dynamic current that is available to drive a typical large capacitive CMOS load.

## Frequently Asked Questions

1. Q : What is DOC?

A: DOC is the Dynamic Output Control circuit (patent pending). It is the output circuit of TI's AVC family of devices that changes the output impedance during the signal transition.
2. Q: Why use DOC output?

A: During the beginning of the signal transition, DOC output provides the desirable characteristics of high drive to supply fast edge-rates and small propagation delays. As the signal passes through the threshold, the DOC output decreases the drive to roll off the signal and reduce ringing without the use of damping resistors.
3. Q: How does DOC work?

A: The DOC output has an impedance-control circuit that monitors the output signal. When a transition begins, the impedance-control circuit enables the outputs of two parallel drivers to provide a low-impedance, high-drive output. As the output passes through the threshold, the impedance-control circuit disables the output of one of the drivers, providing a high-impedance, low-drive output.
4. Q: Should I use series damping resistors on the output of DOC devices?

A: It is not necessary to use series damping resistors to reduce ringing because the DOC output provides a high-impedance, low-drive output at the end of the signal transition. Using series damping resistors would defeat the high-drive benefit of the DOC output.
5. Q: Can I use de termination on the output of DOC devices?

A: Do not use dc termination. The use of dc termination could exceed the static-drive capability of the DOC output. Due to the excellent signal-integrity characteristics of the DOC output, termination should be unnecessary.
6. Q: What is the maximum drive-current capability of the DOC output?

A: The DOC output has $\pm 8-\mathrm{mA}$ dc static-drive current capability at $2.5-\mathrm{V}_{\mathrm{CC}}$. Under typical conditions at $2.5-\mathrm{V}$ $\mathrm{V}_{\mathrm{CC}}$, the amount of ac dynamic-drive current that the DOC output can supply varies from a maximum of about 84 mA at the beginning of the transition from low to high. At the beginning of the transition from high to low, it varies from a maximum of about 105 mA .
7. Q: What is the output impedance of a DOC circuit?

A: The impedance during dynamic conditions is characterized by the slope of the $\mathrm{V}_{\mathrm{O}} \mathrm{vS}_{\mathrm{O}}$ line, at any specific point on the graph. The output $\mathrm{R}_{\mathrm{ON}}$ can be calculated from the $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ curves (see Figure 1 and Figure 2). At any specific point on the $\mathrm{V}_{\mathrm{OH}}$ vs $\mathrm{I}_{\mathrm{OH}}$ curves, $\mathrm{R}_{\mathrm{ON}}=\left(\mathrm{V}_{\mathrm{OH}}-\mathrm{V}_{\mathrm{CC}}\right) / \mathrm{I}_{\mathrm{OH}}$. At any specific point on the $\mathrm{V}_{\mathrm{OL}}$ vs $\mathrm{I}_{\mathrm{OL}}$ curves, $\mathrm{R}_{\mathrm{ON}}=\mathrm{V}_{\mathrm{OL}} / \mathrm{I}_{\mathrm{OL}}$. In the high state, the output $\mathrm{R}_{\mathrm{ON}}$ varies from approximately $50 \Omega$ in the high-impedance mode to approximately $30 \Omega$ in the low-impedance mode. In the low state, the output $\mathrm{R}_{\mathrm{ON}}$ varies from approximately $50 \Omega$ in the high-impedance mode to approximately $20 \Omega$ in the low-impedance mode.
8. Q: Are devices with DOC output circuitry fast?

A: Yes, the DOC output provides a very fast edge-rate to decrease the propagation delay times, while maintaining the excellent signal-integrity characteristics associated with the slower series-damping-resistor parts.
9. Q: Why aren't ac dynamic-drive specifications included in the data sheet?

A: The dynamic-drive current is not specified on the data sheet for devices with DOC outputs because of its transient nature, but it is similar to the drive current available from a standard-output device with an $\mathrm{I}_{\mathrm{OH}}$ and $\mathrm{I}_{\mathrm{OL}}$ of $\pm 24$ mA at $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$.
10. Q: In data sheets for devices with DOC outputs, is the dc static-drive specification an indicator of the devices' dynamic performance?

A: No. The devices perform like high-drive devices during signal transition. This is not reflected in the dc static-drive specification on the data sheet.
11. Q: Since the DOC output provides high-drive, does it suffer from poor simultaneous switching performance? How does its simultaneous switching performance compare to standard and resistor devices?

A: At $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ with output into a standard load, SPICE analysis shows that the SN74AVC16245 DOC outputs have a maximum $\mathrm{V}_{\text {OLV }}=-165 \mathrm{mV}$, standard outputs have a maximum $\mathrm{V}_{\mathrm{OLV}}=-574 \mathrm{mV}$, and resistor outputs have a maximum $\mathrm{V}_{\text {OLV }}=-36 \mathrm{mV}$ ( 15 outputs switching, one steady-state low).
12. Q: Do DOC outputs contribute to a device's low-power performance?

A: Compared to a damping-resistor output where a portion of the output drive is dissipated in the resistor and not delivered to the load, the DOC output offers better low-power performance. The devices in the AVC family that feature DOC outputs are designed for $2.5-\mathrm{V} \mathrm{V}_{\mathrm{CC}}$ operation, enabling low-power designs.

## Acknowledgment

The authors of this application report are Stephen M. Nolan and Tim Ten Eyck.

## References

1. TI SN74AVC16245 16-Bit Bus Transceiver With 3-State Outputs, literature number SCES142.
2. TI SN74ALVC162245 16-Bit Bus Transceiver With 3-State Outputs, literature number SCES064.
3. Intel PC SDRAM Registered DIMM Specification, Revision 1.0, February 1998.

## Glossary

A

| A | Amperes |
| :--- | :--- |
| ac | Alternating current |
| ALVC | Advanced Low-Voltage CMOS |
| AVC | Advanced Very-low-voltage CMOS |
| B |  |
| B | Byte |
| C |  |
| CMOS | Complementary metal-oxide semiconductor |
| CD |  |
| D |  |


| dc | Direct current |
| :--- | :--- |
| DIMM | Dual-inline memory module |
| DOC | Dynamic output control (patent pending) |
| DQM | Data mask |
| DRAM | Dynamic random-access memory |
| F |  |
| F | Farad |
| FET | Field-effect transistor |

H

H
Henry

| IBIS | I/O buffer information specification |
| :---: | :---: |
| $\mathrm{I}_{\mathrm{I}}$ | Input current |
| $\mathrm{I}_{\text {OFF }}$ | Current into a pin when $\mathrm{V}_{\mathrm{CC}}=0 \mathrm{~V}$ |
| $\mathrm{I}_{\mathrm{OH}}$ | High-level output current |
| $\mathrm{I}_{\mathrm{OHS}}$ | Static high-level output current |
| $\mathrm{I}_{\mathrm{OL}}$ | Low-level output current |
| $\mathrm{I}_{\text {OLS }}$ | Static low-level output current |
| IV | Current vs voltage |
| M |  |
| Max | Maximum |
| Min | Minimum |
| PC | Personal computer |
| $\mathrm{R}_{\mathrm{ON}}$ | On-state output resistance |
| S | Seconds |
| SDRAM | Synchronous DRAM |
| SPICE | Simulation program with integrated-circuit emphasis |
| TI | Texas Instruments |

V Volts
$\mathrm{V}_{\mathrm{CC}} \quad$ Supply voltage
$\mathrm{V}_{\mathrm{O}} \quad$ Output voltage
$\mathrm{V}_{\mathrm{OH}} \quad$ High-level output voltage
VOL Low-level output voltage
$\mathrm{V}_{\text {OHP }} \quad$ High-level output voltage peak
VOHV High-level output voltage valley
$V_{\text {OLP }} \quad$ Low-level output voltage peak
VOLV Low-level output voltage valley
Z

ZCC Impedance control circuit

## Appendix A - Parameter Measurement Information



> VOLTAGE WAVEFORMS
> PROPAGATION DELAY TIMES

| TEST | S1 |
| :---: | :---: |
| ${ }^{\text {t }}$ d | Open |
| tpLz/tpZL | $2 \times \mathrm{V}$ C |
| tPHZ/tPZH | GND |



VOLTAGE WAVEFORMS PULSE DURATION


[^8]NOTES: A. $C_{L}$ includes probe and jig capacitance.
B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
C. All input pulses are supplied by generators having the following characteristics: $\mathrm{PRR} \leq 10 \mathrm{MHz}, \mathrm{Z}_{\mathrm{O}}=50 \Omega, \mathrm{t}_{\mathrm{r}} \leq 2 \mathrm{~ns}, \mathrm{t}_{\mathrm{f}} \leq 2 \mathrm{~ns}$.
D. The outputs are measured one at a time with one transition per measurement.
E. tpLZ and tphZ are the same as $\mathrm{t}_{\text {dis }}$.
F. tpZL and tpZH are the same as ten.
G. $\mathrm{tPLH}^{\text {and }} \mathrm{tPHL}$ are the same as $\mathrm{t}_{\mathrm{pd}}$.

Figure A-1. AVC Load Circuit and Voltage Waveforms ( $\mathrm{V}_{\mathrm{Cc}}=2.5 \mathrm{~V} \pm 0.2 \mathrm{~V}$ )

# Implications of Slow or Floating CMOS Inputs 

February 1998

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## Introduction

In recent years, CMOS (AC/ACT, AHC/AHCT, ALVC, CBT, CBTLV, HC/HCT, LVC, LV/LV-A) and BiCMOS (ABT, ALVT, BCT, FB, GTL, and LVT) logic families have further strengthened their position in the semiconductor market. New designs have adopted both technologies in almost every system that exists, whether it is a PC, a workstation, or a digital switch. The reason is obvious: power consumption is becoming a major issue in today's market. However, when designing systems using CMOS and BiCMOS devices, one must understand the characteristics of these families and the way inputs and outputs behave in systems. It is very important for the designer to follow all rules and restrictions that the manufacturer requires, as well as to design within the data-sheet specifications. Because data sheets do not cover the input behavior of a device in detail, this application report explains the input characteristics of CMOS and BiCMOS families in general. It also explains ways to deal with issues when designing with families in which floating inputs are a concern. Understanding the behavior of these inputs results in more robust designs and better reliability.

## Characteristics of Slow or Floating CMOS Inputs

Both CMOS and BiCMOS families have a CMOS input structure. This structure is an inverter consisting of a p-channel to $\mathrm{V}_{\mathrm{CC}}$ and an n-channel to GND as shown in Figure 1. With low-level input, the p-channel transistor is on and the n -channel is off, causing current to flow from $\mathrm{V}_{\mathrm{CC}}$ and pulling the node to a high state. With high-level input, the n -channel transistor is on, the p-channel is off, and the current flows to GND, pulling the node low. In both cases, no current flows from $\mathrm{V}_{\mathrm{CC}}$ to GND. However, when switching from one state to another, the input crosses the threshold region, causing the n-channel and the p-channel to turn on simultaneously, generating a current path between $\mathrm{V}_{\mathrm{CC}}$ and GND. This current surge can be damaging, depending on the length of time that the input is in the threshold region ( 0.8 to 2 V ). The supply current $\left(\mathrm{I}_{\mathrm{CC}}\right)$ can rise to several milliamperes per input, peaking at approximately $1.5-\mathrm{V}_{\mathrm{I}}$ (see Figure 2). This is not a problem when switching states within the data-sheet-specified input transition time limit specified in the recommended operating conditions table for the specific devices. Examples are shown in Figure 3.


Figure 1. Input Structures of ABT and LVT/LVC Devices


Figure 2. Supply Current Versus Input Voltage (One Input)
recommended operating conditions ${ }^{\dagger}$

|  |  |  | MIN MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: |
|  | ABT octals |  | 5 |  |
|  | ABT Wide |  | 10 |  |
|  | AHC, AHC |  | 20 |  |
|  | FB |  | 10 |  |
| $\Delta t / \Delta v \quad$ Input transition rise or fall rate | LVT, LVC, |  | 10 | $\mathrm{ns} / \mathrm{V}$ |
|  | LV |  | 100 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2.3 \mathrm{~V}$ to 2.7 V | 200 |  |
|  | LV-A | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V | 100 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=4.5 \mathrm{~V}$ to 5.5 V | 20 |  |
|  |  | $\mathrm{V}_{\mathrm{CC}}=2 \mathrm{~V}$ | 1000 |  |
| $\mathrm{t}_{\mathrm{t}} \quad$ Input transition (rise and fall) time | HC, HCT | $\mathrm{V}_{\mathrm{CC}}=4.5 \mathrm{~V}$ | 500 | ns |
|  |  | $\mathrm{V}_{\mathrm{CC}}=6 \mathrm{~V}$ | 400 |  |

$\dagger$ Refer to the latest TI data sheets for device specifications.
Figure 3. Input Transition Rise or Fall Rate as Specified in Data Sheets

## Slow Input Edge Rate

With increased speed, logic devices have become more sensitive to slow input edge rates. A slow input edge rate, coupled with the noise generated on the power rails when the output switches, can cause excessive output errors or oscillations. Similar situations can occur if an unused input is left floating or is not actively held at a valid logic level.

These functional problems are due to voltage transients induced on the device's power system as the output load current $\left(\mathrm{I}_{\mathrm{O}}\right)$ flows through the parasitic lead inductances during switching (see Figure 4). Because the device's internal power-supply nodes are used as voltage references throughout the integrated circuit, inductive voltage spikes, $\mathrm{V}_{\mathrm{GND}}$, affect the way signals appear to the internal gate structures. For example, as the voltage at the device's ground node rises, the input signal, $\mathrm{V}_{\mathrm{I}}{ }^{\prime}$, appears to decrease in magnitude. This undesirable phenomenon can then erroneously change the output if a threshold violation occurs.
In the case of a slowly rising input edge, if the change in voltage at GND is large enough, the apparent signal, $\mathrm{V}_{\mathrm{I}}^{\prime}$, at the device appears to be driven back through the threshold and the output starts to switch in the opposite direction. If worst-case conditions prevail (simultaneously switching all of the outputs with large transient load currents), the slow input edge is repeatedly driven back through the threshold, causing the output to oscillate. Therefore, the maximum input transition time of the device should not be violated, so no damage to the circuit or the package occurs.


Figure 4. Input/Output Model

## Floating Inputs

If a voltage between 0.8 V and 2 V is applied to the input for a prolonged period of time, this situation becomes critical and should not be ignored, especially with higher bit count and more dense packages (SSOP, TSSOP). For example, if an 18-bit transceiver has $36 \mathrm{I} / \mathrm{O}$ pins floating at the threshold, the current from $\mathrm{V}_{\mathrm{CC}}$ can be as high as 150 mA to 200 mA . This is approximately 1 W of power consumed by the device, which leads to a serious overheating problem. This continuous overheating of the device affects its reliability. Also, because the inputs are in the threshold region, the outputs tend to oscillate, resulting in damage to the internal circuit over a long period of time. The data sheet shows the increase in supply current $\left(\Delta \mathrm{I}_{\mathrm{CC}}\right)$ when the input is at a TTL level [for $\mathrm{ABT} \mathrm{V}_{\mathrm{I}}=3.4 \mathrm{~V}, \Delta \mathrm{I}_{\mathrm{CC}}=1.5 \mathrm{~mA}$ (see Figure 5)]. This becomes more critical when the input is in the threshold region as shown in Figure 6.
These characteristics are typical for all CMOS input circuits, including microprocessors and memories.
For CBT or CBTLV devices, this applies to the control inputs. For FB and GTL devices, this applies to the control inputs and the TTL ports only.
electrical characteristics over recommended operating free-air temperature range (unless otherwise noted) $\dagger$

| PARAMETER |  | TEST CONDITIONS |  |  | MIN MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ${ }^{\prime} \mathrm{CCC}^{\ddagger}$ | ABT, AHCT | $\mathrm{V}_{\mathrm{CC}}=5.5 \mathrm{~V}$, | One input at 3.4 V, | Other inputs at $\mathrm{V}_{\mathrm{CC}}$ or GND | 1.5 | mA |
|  | CBT <br> Control inputs | $\mathrm{V}_{\mathrm{CC}}=5.5 \mathrm{~V}$, | One input at 3.4 V , | Other inputs at $\mathrm{V}_{\mathrm{CC}}$ or GND | 2.5 |  |
| ${ }^{\text {I }} \mathrm{CC}{ }^{\ddagger}$ | CBTLV <br> Control inputs | $\mathrm{V}_{\mathrm{CC}}=3.6 \mathrm{~V}$, | One input at 3 V , | Other inputs at $\mathrm{V}_{\mathrm{CC}}$ or GND | 750 | $\mu \mathrm{A}$ |
| ${ }^{1} \mathrm{lcC}{ }^{\ddagger}$ | LVT | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to 3.6 V , | One input at $\mathrm{V}_{\mathrm{CC}}-0.6 \mathrm{~V}$, | Other inputs at $\mathrm{V}_{\mathrm{CC}}$ or GND | 0.2 | mA |
|  | LVC, ALVC, LV |  |  |  | 0.5 |  |

$\dagger$ Refer to the latest TI data sheets for device specifications.
$\ddagger$ This is the increase in supply current for each input that is at the specified TTL voltage level rather than $\mathrm{V}_{\mathrm{CC}}$ or GND.
Figure 5. Examples of Supply-Current Change of the Input at TTL Level as Specified in Data Sheets


Figure 6. Supply Current Versus Input Voltage ( 36 Inputs)
As long as the driver is active in a transmission path or bus, the receiver's input is always in a valid state. No input specification is violated as long as the rise and fall times are within the data-sheet limits. However, when the driver is in a high-impedance state, the receiver input is no longer at a defined level and tends to float. This situation can worsen when several transceivers share the same bus. Figure 7 is an example of a typical bus system. When all transceivers are inactive, the bus-line levels are undefined. When a voltage that is determined by the leakage currents of each component on the bus is reached, the condition is known as a floating state. The result is a considerable increase in power consumption and a risk of damaging all components on the bus. Holding the inputs or I/O pins at a valid logic level when they are not being used or when the part driving them is in the high-impedance state is recommended.


Figure 7. Typical Bidirectional Bus

## Recommendations for Designing More-Reliable Systems

## Bus Control

The simplest way to avoid floating inputs in a bus system is to ensure that the bus always is either active or inactive for a limited time when the voltage buildup does not exceed the maximum $\mathrm{V}_{\mathrm{IL}}$ specification ( 0.8 V for TTL-compatible input). At this voltage, the corresponding $\mathrm{I}_{\mathrm{CC}}$ value is too low and the device operates without any problem or concern (see Figures 2 and 4).

To avoid damaging components, the designer must know the maximum time the bus can float. First, assuming that the maximum leakage current is $\mathrm{I}_{\mathrm{OZ}}=50 \mu \mathrm{~A}$ and the total capacitance ( $\mathrm{I} / \mathrm{O}$ and line capacitance) is $\mathrm{C}=20 \mathrm{pF}$, the change in voltage with respect to time on an inactive line that exceeds the $0.8-\mathrm{V}$ level can be calculated as shown in equation 1 .

$$
\begin{equation*}
\Delta \mathrm{V} / \Delta \mathrm{t}=\frac{\mathrm{I}_{\mathrm{OZ}}}{\mathrm{C}}=\frac{50 \mu \mathrm{~A}}{20 \mathrm{pF}}=2.5 \mathrm{~V} / \mu \mathrm{s} \tag{1}
\end{equation*}
$$

The permissible floating time for the bus in this example should be reduced to 320 ns maximum, which ensures that the bus does not exceed the $0.8-\mathrm{V}$ level specified. The time constant does not change when multiple components are involved because their leakage currents and capacitances are summed.

The advantage of this method is that it requires no additional cost for adding special components. Unfortunately, this method does not always apply because buses are not always active.

## Pullup or Pulldown Resistors

When buses are disabled for more than the maximum allowable time, other ways should be used to prevent components from being damaged or overheated. A pullup or a pulldown resistor to $\mathrm{V}_{\mathrm{CC}}$ or GND, respectively, should be used to keep the bus in a defined state. The size of the resistor plays an important role and, if its resistance is not chosen properly, a problem may occur. Usually, a $1-\mathrm{k} \Omega$ to $10-\mathrm{k} \Omega$ resistor is recommended. The maximum input transition time must not be violated when selecting pullup or pulldown resistors (see Figure 3). Otherwise, components may oscillate, or device reliability may be affected.


Figure 8. Inactive-Bus Model With a Defined Level
Assume that an active-low bus goes to the high-impedance state as modeled in Figure 8. $\mathrm{C}_{\mathrm{T}}$ represents the device plus the bus-line capacitance and R is a pullup resistor to $\mathrm{V}_{\mathrm{CC}}$. The value of the required resistor can be calculated as shown in equation 2 .

$$
\begin{equation*}
\mathrm{V}(\mathrm{t})=\mathrm{V}_{\mathrm{CC}}-\left[\mathrm{e}^{-\mathrm{t} / \mathrm{RC}_{\mathrm{T}}}\left(\mathrm{~V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{i}}\right)\right] \tag{2}
\end{equation*}
$$

Where:

$$
\mathrm{V}(\mathrm{t})=2 \mathrm{~V} \text {, minimum voltage at time } \mathrm{t}
$$

$\mathrm{V}_{\mathrm{i}}=0.5 \mathrm{~V}$, initial voltage
$\mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V}$
$\mathrm{C}_{\mathrm{T}}=$ total capacitance
$\mathrm{R}=$ pullup resistor
$\mathrm{t} \quad=$ maximum input rise time as specified in the data sheets (see Figure 3).

Solving for R, the equation becomes:

$$
\begin{equation*}
\mathrm{R}=\frac{\mathrm{t}}{0.4 \times \mathrm{C}_{\mathrm{T}}} \tag{3}
\end{equation*}
$$

For multiple transceivers on a bus:

$$
\begin{equation*}
\mathrm{R}=\frac{\mathrm{t}}{0.4 \times \mathrm{C} \times \mathrm{N}} \tag{4}
\end{equation*}
$$

Where:
$\mathrm{C}=$ individual component and trace capacitance
$\mathrm{N}=$ number of components connected to the bus

Assuming that there are two components connected to the bus, each with a capacitance $\mathrm{C}=15 \mathrm{pF}$, requiring a maximum rise time of $10 \mathrm{~ns} / \mathrm{V}$ and $\mathrm{t}=15-\mathrm{ns}$ total rise time for the input $(2 \mathrm{~V})$, the maximum resistor size can be calculated:

$$
\begin{equation*}
\mathrm{R}=\frac{15 \mathrm{~ns}}{0.4 \times 15 \mathrm{pF} \times 2}=1.25 \mathrm{k} \Omega \tag{5}
\end{equation*}
$$

This pullup resistor method is recommended for ac-powered systems; however, it is not recommended for battery-operated equipment because power consumption is critical. Instead, use the bus-hold feature that is discussed in the next section. The overall advantage of using pullup resistors is that they ensure defined levels when the bus is floating and help eliminate some of the line reflections, because resistors also can act as bus terminations.

## Bus-Hold Circuits

The most effective method to provide defined levels for a floating bus is to use Texas Instruments ( $\mathrm{TI}^{\mathrm{TM}}$ ) built-in bus-hold feature on selected families or as an external component like the SN74ACT1071 and SN74ACT1073 (refer to Table 1).

Table 1. Devices With Bus Hold

| DEVICE TYPE | BUS HOLD INCORPORATED |
| :--- | :--- |
| SN74ACT1071 | 10-bit bus hold with clamping diodes |
| SN74ACT1073 | 16-bit bus hold with clamping diodes |
| ABT Widebus+ (32 and 36 bit) | All devices |
| ABT Octals and Widebus | Selected devices only |
| AHC/AHCT Widebus | TBA (Selected devices only) |
| Low Voltage (LVT and ALVC) | All devices |
| LVC Widebus | All devices |

Bus-hold circuits are used in selected TI families to help solve the floating-input problem and eliminate the need for pullup and pulldown resistors. Bus-hold circuits consist of two back-to-back inverters with the output fed back to the input through a resistor (see Figure 9). To understand how the bus-hold circuit operates, assume that an active driver has switched the line to a high level. This results in no current flowing through the feedback circuit. Now, the driver goes to the high-impedance state and the bus-hold circuit holds the high level through the feedback resistor. The current requirement of the bus-hold circuit is determined only by the leakage current of the circuit. The same condition applies when the bus is in the low state and then goes inactive.


Figure 9. Typical Bus-Hold Circuit

As mentioned previously in this section, TI offers the bus-hold capability as stand-alone 10-bit and 16-bit devices (SN74ACT1071 and SN74ACT1073) with clamping diodes to $\mathrm{V}_{\mathrm{CC}}$ and GND for added protection against line reflections caused by impedance mismatch on the bus. Because purely ohmic resistors cannot be implemented easily in CMOS circuits, a configuration known as a transmission gate is used as the feedback element (see Figure 10). An n-channel and a p-channel are arranged in parallel between the input and the output of the buffer stage. The gate of the n -channel transistor is connected to $\mathrm{V}_{\mathrm{CC}}$ and the gate of the p-channel is connected to GND. When the output of the buffer is high, the p-channel is on, and when the output is low, the $n$-channel is on. Both channels have a relatively small surface area - the on-state resistance from drain to source, $\mathrm{R}_{\mathrm{dson}}$, is about $5 \mathrm{k} \Omega$.


Figure 10. Stand-Alone Bus-Hold Circuit (SN74ACT107x)
Assume that in a practical application the leakage current of a driver on a bus is $\mathrm{I}_{\mathrm{OZ}}=10 \mu \mathrm{~A}$ and the voltage drop across the $5-\mathrm{k} \Omega$ resistance is $\mathrm{V}_{\mathrm{D}}=0.8 \mathrm{~V}$ (this value is assumed to ensure a defined logic level). Then, the maximum number of components that a bus-hold circuit can handle is calculated as follows:

$$
\begin{equation*}
\mathrm{N}=\frac{\mathrm{V}_{\mathrm{D}}}{\mathrm{I}_{\mathrm{OZ}} \times \mathrm{R}}=\frac{0.8 \mathrm{~V}}{10 \mu \mathrm{~A} \times 5 \mathrm{k} \Omega}=16 \text { components } \tag{6}
\end{equation*}
$$

The 74ACT1071 and 74ACT1073 also provide clamping diodes as an added feature to the bus-hold circuit. These diodes are useful for clamping any overshoot or undershoot generated by line reflections. Figure 11 shows the characteristics of the diodes when the input voltage is above $\mathrm{V}_{\mathrm{CC}}$ or below GND. At $\mathrm{V}_{\mathrm{I}}=-1 \mathrm{~V}$, the diode can source about 50 mA , which can help eliminate undershoots. This can be very useful when noisy buses are a concern.


Figure 11. Diode Characteristics (SN74ACT107x)

TI also offers the bus-hold circuit as a feature added to some of the advanced-family drivers and receivers. This circuit is similar to the stand-alone circuit, with a diode added to the drain of the second inverter (ABT and LVT only, see Figure 12). The diode blocks the overshoot current when the input voltage is higher than $V_{C C}\left(V_{I}>V_{C C}\right)$, so only the leakage current is present. This circuit uses the device's input stage as its first inverter; a second inverter creates the feedback feature. The calculation of the maximum number of components that the bus-hold circuit can handle is similar to the previous example. However, the advantage of this circuit over the stand-alone bus-hold circuit is that it eliminates the need for external components or resistors that occupy more area on the board. This becomes critical for some designs, especially when wide buses are used. Also, because cost and board-dimension restrictions are a major concern, designers prefer the easy fix: drop-in replaceable parts. TI offers this feature in most of the commonly used functions in several families (refer to Table 1 for more details).


Figure 12. Input Structure of ABT/LVT and ALVC/LVC Families With Bus-Hold Circuit
Figure 13 shows the input characteristics of the bus-hold circuit at $3.3-\mathrm{V}$ and $5-\mathrm{V}$ operations, as the input voltage is swept from 0 to 5 V . These characteristics are similar in behavior to a weak driver. This driver sinks current into the part when the input is low and sources current out of the part when the input is high. When the voltage is near the threshold, the circuit tries to switch to the other state, always keeping the input at a valid level. This is the result of the internal feedback circuit. The plot also shows that the current is at its maximum when the input is near the threshold. $\mathrm{I}_{\mathrm{I}(\text { hold })}$ maximum is approximately $25 \mu \mathrm{~A}$ for $3.3-\mathrm{V}$ input and $400 \mu \mathrm{~A}$ for $5-\mathrm{V}$ input.


Figure 13. Bus-Hold Input Characteristics
When multiple devices with bus-hold circuits are driven by a single driver, there may be some concern about the ac switching capability of the driver becoming weaker. As small drivers, bus-hold circuits require an ac current to switch them. This current is not significant when using TI CMOS and BiCMOS families. Figure 14 shows a $4-\mathrm{mA}$ buffer driving six LVTH16244 devices. The trace is a $75-\Omega$ transmission line. The receivers are separated by 1 cm , with the driver located in the center of the trace. Figure 15 shows the bus-hold loading effect on the driver when connected to six receivers switching low or high. It also shows the same system with the bus-hold circuit disconnected from the receivers. Both plots show the effect of bus hold on the driver's rise and fall times. Initially, the bus-hold circuit tries to counteract the driver, causing the rise or fall time to increase. Then, the bus-hold circuit changes states (note the crossover point), which helps the driver switch faster, decreasing the rise or fall time.


Figure 14. Driver and Receiver System


Figure 15. Output Waveforms of Driver With and Without Receiver Bus-Hold Circuit
Figure 16 shows the supply current $\left(\mathrm{I}_{\mathrm{CC}}\right)$ of the bus-hold circuit as the input is swept from 0 to 5 V . The spike at about $1.5-\mathrm{V}$ $\mathrm{V}_{\mathrm{I}}$ is due to both the n -channel and the p-channel conducting simultaneously. This is one of the CMOS transistor characteristics.


Figure 16. Bus-Hold Circuit Supply Current Versus Input Voltage

The power consumption of the bus-hold circuit is minimal when switching the input at higher frequencies. Figure 17 shows the power consumed by the input at different frequencies, with or without bus hold. The increase in power consumption of the bus-hold circuit at higher frequencies is not significant enough to be considered in power calculations.


Figure 17. Input Power With and Without Bus Hold at Different Frequencies

Figure 18 shows the data-sheet dc specifications for bus hold. The first test condition is the minimum current required to hold the bus at 0.8 V or 2 V . These voltages meet the specified low and high levels for TTL inputs. The second test condition is the maximum current that the bus-hold circuit sources or sinks at any input voltage between 0 V and 3.6 V (for low-voltage families) or between 0 V and 5.5 V (for ABT ). The bus-hold current becomes minimal as the input voltage approaches the rail voltage. The output leakage currents, $\mathrm{I}_{\mathrm{OZH}}$ and $\mathrm{I}_{\mathrm{OZL}}$, are insignificant for transceivers with bus hold because a true leakage test cannot be performed due to the existence of the bus-hold circuit. Because the bus-hold circuit behaves as a small driver, it tends to source or sink a current that is opposite in direction to the leakage current. This situation is true for transceivers with the bus-hold feature only and does not apply to buffers. All LVT, ABT Widebus+, and selected ABT octal and Widebus devices have the bus-hold feature (refer to Table 1 or contact the local TI sales office for more information).
electrical characteristics over recommended operating free-air temperature range (for families with bus-hold feature) $\dagger$

| PARAMETER |  |  | TEST CONDITIONS |  | MIN MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $1 /$ (hold) | Data inputs or I/Os | LVT, LVC, ALVC | $\mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ | $\mathrm{V}_{1}=0.8 \mathrm{~V}$ | 75 | $\mu \mathrm{A}$ |
|  |  |  |  | $\mathrm{V}_{\mathrm{I}}=2 \mathrm{~V}$ | -75 |  |
|  |  | LVC, ALVC | $\mathrm{V}_{\mathrm{CC}}=3.6 \mathrm{~V}$, | $\mathrm{V}_{\mathrm{I}}=0$ to 3.6 V | $\pm 500$ |  |
|  |  | ABT Widebus+ and selected ABT | $\mathrm{V}_{\mathrm{CC}}=4.5 \mathrm{~V}$ | $\mathrm{V}_{1}=0.8 \mathrm{~V}$ | 100 |  |
|  |  |  |  | $\mathrm{V}_{\mathrm{I}}=2 \mathrm{~V}$ | -100 |  |
| ${ }^{\text {l }}$ OZH/ $/ \mathrm{l}$ OZL | Transceivers with bus hold | ABT | This test is not a true loz test because bus hold always is active on an I/O pin. Bus hold tends to supply a current that is opposite in direction to the output leakage current. |  |  | $\mu \mathrm{A}$ |
|  |  | LVT, LVC, ALVC |  |  |  |  |
|  | Buffers with bus hold | ABT | This test is a true IOZ test since bus hold does not exist on an output pin. |  | $\pm 10$ |  |
|  |  | LVT, LVC, ALVC |  |  | $\pm 5$ |  |

$\dagger$ Refer to the latest TI data sheets for device specifications.
Figure 18. Example of Data-Sheet Minimum Specification for Bus Hold

## Summary

Floating inputs and slow rise and fall times are important issues to consider when designing with CMOS and advanced BiCMOS families. It is important to understand the complications associated with floating inputs. Terminating the bus properly plays a major role in achieving reliable systems. The three methods recommended in this application report should be considered. If it is not possible to control the bus directly, and adding pullup or pulldown resistors is impractical due to power-consumption and board-space limitations, bus hold is the best choice. TI designed bus hold to reduce the need for resistors used in bus designs, thus reducing the number of components on the board and improving the overall reliability of the system.
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$\square$
$\square$
$\square$-

Electrical characteristics presented in this data book, unless otherwise noted, apply for the circuit type(s) listed in the page heading, regardless of package. The availability of a circuit function in a particular package is denoted by an alphabetical reference above the pin-connection diagram(s). These alphabetical references refer to mechanical outline drawings shown in this section.
Factory orders for circuits described in this data book should include a three-part type number as explained in the following example.


Valid for surface-mount packages only. All orders for tape and reel must be for whole reels.
$R=$ Standard tape and reel [required for DGG (or G) and DGV (or V)] ${ }^{\ddagger}$
The purpose of tape-and-reel packing is to position components so they can be placed automatically. Components such as, but not limited to, diodes, capacitors, resistors, transistors, inductors, and integrated circuits can be packed in this manner.
The packing materials include a carrier tape, cover tape, and a reel. The normal dimensions for these items are listed in Table 1.
$\dagger$ TI is changing the nomenclature for select logic devices. For details, see Device Names and Package Designators in Section 1.
$\ddagger$ All reeled material previously designated LE will continue to be reeled left embossed, but an R designator will be used.

Table 1. Normal Dimensions of Packing Materials

| CARRIER-TAPE <br> WIDTH <br> $(\mathbf{m m})$ | COVER-TAPE <br> WIDTH <br> $(\mathbf{m m})$ | REEL <br> WIDTH <br> $(\mathbf{m m})$ | REEL <br> DIAMETER <br> $(\mathbf{m m})$ |
| :---: | :---: | :---: | :---: |
| 8 | 5.4 | 9.0 | 178 |
| 12 | 9.2 | 12.4 | 330 |
| 16 | 13.3 | 16.4 | 330 |
| 24 | 21.0 | 24.4 | 330 |
| 32 | 25.5 | 32.4 | 330 |
| 44 | 37.5 | 44.4 | 330 |
| 56 | 49.5 | 56.4 | 330 |

All material meets or exceeds industry guidelines for ESD protection.
Dimensions are selected based on package size and design configurations. All dimensions are established to be within the recommendations of the Electronics Industry Association Standard EIA-481-1,2,3.
Common dimensions of particular interest to the end user are carrier-tape width, pocket pitch, and quantity per reel (see Figure 1 and Table 2).


Figure 1. Typical Carrier-Tape Design

Table 2. Selected Tape-and-Reel Specifications

| PACKAGE |  | NO. OF <br> PINS | CARRIER-TAPE <br> WIDTH <br> $(\mathbf{m m})$ | POCKET <br> PITCH <br> $(\mathbf{m m})$ | QTY/REEL |
| :--- | :--- | :---: | :---: | :---: | :---: |
| LFBGA | GKE | 96 | 24.00 | 8.00 | 1000 |
|  | GKF | 114 | 24.00 | 8.00 | 1000 |
| TSSOP | DGG | 48 | 24.00 | 12.00 | 2000 |
|  | DBB | 80 | 24.00 | 12.00 | 2000 |
|  |  | 14 | 16.00 | 8.00 | 2000 |
|  |  | 16 | 16.00 | 8.00 | 2000 |
|  |  | 16.00 | 8.00 | 2000 |  |
|  |  | 48 | 16.00 | 8.00 | 2000 |

DBB (R-PDSO-G**)
80 PINS SHOWN


| DIM | $\mathbf{8 0}$ | $\mathbf{1 0 0}$ |
| :---: | :---: | :---: |
| A MAX | 17,10 | 20,90 |
| A MIN | 16,90 | 20,70 |

NOTES: A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. The 80-pin package falls within JEDEC MO-153 and the 100-pin package falls within JEDEC MO-194.

48 PINS SHOWN


NOTES: A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Body dimensions do not include mold protrusion not to exceed 0,15 .
D. Falls within JEDEC MO-153


| PIM | $\mathbf{1 4}$ | $\mathbf{1 6}$ | $\mathbf{2 0}$ | $\mathbf{2 4}$ | $\mathbf{4 8}$ | $\mathbf{5 6}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A MAX | 3,70 | 3,70 | 5,10 | 5,10 | 9,80 | 11,40 |
| A MIN | 3,50 | 3,50 | 4,90 | 4,90 | 9,60 | 11,20 |

NOTES: A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Body dimensions do not include mold flash or protrusion, not to exceed 0,15 per side.
D. Falls within JEDEC: $24 / 48$ Pins - MO-153

14/16/20/56 Pins - MO-194


NOTES: A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. MicroStar BGA ${ }^{T M}$ configuration


NOTES: A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. MicroStar BGA ${ }^{\text {TM }}$ configuration

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[^0]:    VOLTAGE WAVEFORMS
    ENABLE AND DISABLE TIMES

[^1]:    VOLTAGE WAVEFORMS
    ENABLE AND DISABLE TIMES

[^2]:    $\dagger$ Output level before the indicated steady-state input conditions were established

[^3]:    VOLTAGE WAVEFORMS
    ENABLE AND DISABLE TIMES

[^4]:    VOLTAGE WAVEFORMS
    ENABLE AND DISABLE TIMES

[^5]:    VOLTAGE WAVEFORMS
    ENABLE AND DISABLE TIMES

[^6]:    VOLTAGE WAVEFORMS
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