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About This Manual

This manual describes the modules and peripherals of the MSP430F2xx and MSP430G2xx microcontrollers (MCUs). Each chapter describes all of the features and functions of the module or peripheral, but not all features and functions of all modules or peripherals are present on all devices. In addition, modules or peripherals may differ in their exact implementation between device families, or may not be fully implemented on an individual device or device family. Refer to the device-specific datasheet for the supported features of each peripheral.

Pin functions, internal signal connections, and operational parameters differ from device to device. Refer to the device-specific datasheet for these details.

Related Documentation From Texas Instruments

For related documentation, visit the MSP430™ ultra-low-power sensing & measurement MCUs overview.

Notational Conventions

This document uses the following conventions.

- Hexadecimal numbers can be shown with the suffix h or the prefix 0x. For example, the following number is 40 hexadecimal (decimal 64): 40h or 0x40.
- Registers in this document are shown in figures and described in tables.
  - Each register figure shows a rectangle divided into fields that represent the fields of the register. Each field is labeled with its bit name, its beginning and ending bit numbers above, and its read/write properties with default reset value below. A legend explains the notation used for the properties.
  - Reserved bits in a register figure can have one of multiple meanings:
    - Not implemented on the device
    - Reserved for future device expansion
    - Reserved for TI testing
    - Reserved configurations of the device that are not supported
  - Writing nondefault values to the Reserved bits could cause unexpected behavior and should be avoided.

Glossary

<table>
<thead>
<tr>
<th>Term</th>
<th>Description</th>
<th>See the Chapter</th>
</tr>
</thead>
<tbody>
<tr>
<td>ACLK</td>
<td>Auxiliary clock</td>
<td>Basic Clock Module</td>
</tr>
<tr>
<td>ADC</td>
<td>Analog-to-digital converter</td>
<td></td>
</tr>
<tr>
<td>BOR</td>
<td>Brownout reset</td>
<td>System Resets, Interrupts, and Operating Modes</td>
</tr>
<tr>
<td>BSL</td>
<td>Bootloader</td>
<td></td>
</tr>
<tr>
<td>CPU</td>
<td>Central processing unit</td>
<td>CPU or CPUX</td>
</tr>
<tr>
<td>DAC</td>
<td>Digital-to-analog converter</td>
<td></td>
</tr>
<tr>
<td>DCO</td>
<td>Digitally controlled oscillator</td>
<td>Basic Clock Module</td>
</tr>
<tr>
<td>dst</td>
<td>Destination</td>
<td>CPU or CPUX</td>
</tr>
<tr>
<td>FLL</td>
<td>Frequency-locked loop</td>
<td></td>
</tr>
<tr>
<td>GIE</td>
<td>General interrupt enable</td>
<td>System Resets, Interrupts, and Operating Modes</td>
</tr>
<tr>
<td>INT(N/2)</td>
<td>Integer portion of N/2</td>
<td></td>
</tr>
<tr>
<td>I/O</td>
<td>Input/output</td>
<td>Digital I/O</td>
</tr>
<tr>
<td>ISR</td>
<td>Interrupt service routine</td>
<td></td>
</tr>
<tr>
<td>LS</td>
<td>Least-significant bit</td>
<td></td>
</tr>
<tr>
<td>-----</td>
<td>-----------------------------------------------</td>
<td></td>
</tr>
<tr>
<td>LSD</td>
<td>Least-significant digit</td>
<td></td>
</tr>
<tr>
<td>LPM</td>
<td>Low-power mode</td>
<td></td>
</tr>
<tr>
<td>MAB</td>
<td>Memory address bus</td>
<td></td>
</tr>
<tr>
<td>MCLK</td>
<td>Master clock</td>
<td></td>
</tr>
<tr>
<td>MDB</td>
<td>Memory data bus</td>
<td></td>
</tr>
<tr>
<td>MSB</td>
<td>Most-significant bit</td>
<td></td>
</tr>
<tr>
<td>MSD</td>
<td>Most-significant digit</td>
<td></td>
</tr>
<tr>
<td>NMI</td>
<td>(Non)-maskable interrupt</td>
<td></td>
</tr>
<tr>
<td>PC</td>
<td>Program counter</td>
<td></td>
</tr>
<tr>
<td>POR</td>
<td>Power-on reset</td>
<td></td>
</tr>
<tr>
<td>PUC</td>
<td>Power-up clear</td>
<td></td>
</tr>
<tr>
<td>RAM</td>
<td>Random access memory</td>
<td></td>
</tr>
<tr>
<td>SCG</td>
<td>System clock generator</td>
<td></td>
</tr>
<tr>
<td>SFR</td>
<td>Special function register</td>
<td></td>
</tr>
<tr>
<td>SMCLK</td>
<td>Sub-system master clock</td>
<td></td>
</tr>
<tr>
<td>SP</td>
<td>Stack pointer</td>
<td></td>
</tr>
<tr>
<td>SR</td>
<td>Status register</td>
<td></td>
</tr>
<tr>
<td>src</td>
<td>Source</td>
<td></td>
</tr>
<tr>
<td>TOS</td>
<td>Top-of-stack</td>
<td></td>
</tr>
<tr>
<td>WDT</td>
<td>Watchdog timer</td>
<td></td>
</tr>
</tbody>
</table>

**Register Bit Conventions**

Each register is shown with a key indicating the accessibility of the each individual bit, and the initial condition:

### Register Bit Accessibility and Initial Condition

<table>
<thead>
<tr>
<th>Key</th>
<th>Bit Accessibility</th>
</tr>
</thead>
<tbody>
<tr>
<td>rw</td>
<td>Read/write</td>
</tr>
<tr>
<td>r</td>
<td>Read only</td>
</tr>
<tr>
<td>r0</td>
<td>Read as 0</td>
</tr>
<tr>
<td>r1</td>
<td>Read as 1</td>
</tr>
<tr>
<td>w</td>
<td>Write only</td>
</tr>
<tr>
<td>w0</td>
<td>Write as 0</td>
</tr>
<tr>
<td>w1</td>
<td>Write as 1</td>
</tr>
<tr>
<td>(w)</td>
<td>No register bit implemented; writing a 1 results in a pulse. The register bit is always read as 0.</td>
</tr>
<tr>
<td>h0</td>
<td>Cleared by hardware</td>
</tr>
<tr>
<td>h1</td>
<td>Set by hardware</td>
</tr>
<tr>
<td>-0,-1</td>
<td>Condition after PUC</td>
</tr>
<tr>
<td>-(0),-(1)</td>
<td>Condition after POR</td>
</tr>
</tbody>
</table>
This chapter describes the architecture of the MSP430.
1.1 Architecture

The MSP430 incorporates a 16-bit RISC CPU, peripherals, and a flexible clock system that interconnect using a von-Neumann common memory address bus (MAB) and memory data bus (MDB) (see Figure 1-1). Partnering a modern CPU with modular memory-mapped analog and digital peripherals, the MSP430 offers solutions for demanding mixed-signal applications.

Key features of the MSP430x2xx family include:

- Ultralow-power architecture extends battery life
  - 0.1 µA RAM retention
  - 0.8 µA real-time clock mode
  - 250 µA/MIPS active
- High-performance analog ideal for precision measurement
  - Comparator-gated timers for measuring resistive elements
- 16-bit RISC CPU enables new applications at a fraction of the code size.
  - Large register file eliminates working file bottleneck
  - Compact core design reduces power consumption and cost
  - Optimized for modern high-level programming
  - Only 27 core instructions and seven addressing modes
  - Extensive vectored-interrupt capability
- In-system programmable Flash permits flexible code changes, field upgrades and data logging

![Figure 1-1. MSP430 Architecture](image)

1.2 Flexible Clock System

The clock system is designed specifically for battery-powered applications. A low-frequency auxiliary clock (ACLK) is driven directly from a common 32-kHz watch crystal. The ACLK can be used for a background real-time clock self wake-up function. An integrated high-speed digitally controlled oscillator (DCO) can source the master clock (MCLK) used by the CPU and high-speed peripherals. By design, the DCO is active and stable in less than 2 µs at 1 MHz. MSP430-based solutions effectively use the high-performance 16-bit RISC CPU in very short bursts.

- Low-frequency auxiliary clock = Ultralow-power stand-by mode
- High-speed master clock = High performance signal processing
1.3 Embedded Emulation

Dedicated embedded emulation logic resides on the device itself and is accessed via JTAG using no additional system resources.

The benefits of embedded emulation include:

- Unobtrusive development and debug with full-speed execution, breakpoints, and single-steps in an application are supported.
- Development is in-system subject to the same characteristics as the final application.
- Mixed-signal integrity is preserved and not subject to cabling interference.

1.4 Address Space

The MSP430 von-Neumann architecture has one address space shared with special function registers (SFRs), peripherals, RAM, and Flash/ROM memory as shown in Figure 1-2. See the device-specific data sheets for specific memory maps. Code access are always performed on even addresses. Data can be accessed as bytes or words.

The addressable memory space is currently 128 KB.

![Figure 1-2. Memory Map](image-url)

1.4.1 Flash/ROM

The start address of Flash/ROM depends on the amount of Flash/ROM present and varies by device. The end address for Flash/ROM is 0xFFFF for devices with less that 60KB of Flash/ROM. Flash can be used for both code and data. Word or byte tables can be stored and used in Flash/ROM without the need to copy the tables to RAM before using them.

The interrupt vector table is mapped into the upper 16 words of Flash/ROM address space, with the highest priority interrupt vector at the highest Flash/ROM word address (0xFFFF).
1.4.2 RAM

RAM starts at 0200h. The end address of RAM depends on the amount of RAM present and varies by device. RAM can be used for both code and data.

1.4.3 Peripheral Modules

Peripheral modules are mapped into the address space. The address space from 0100 to 01FFh is reserved for 16-bit peripheral modules. These modules should be accessed with word instructions. If byte instructions are used, only even addresses are permissible, and the high byte of the result is always 0.

The address space from 010h to 0FFh is reserved for 8-bit peripheral modules. These modules should be accessed with byte instructions. Read access of byte modules using word instructions results in unpredictable data in the high byte. If word data is written to a byte module only the low byte is written into the peripheral register, ignoring the high byte.

1.4.4 Special Function Registers (SFRs)

Some peripheral functions are configured in the SFRs. The SFRs are located in the lower 16 bytes of the address space, and are organized by byte. SFRs must be accessed using byte instructions only. See the device-specific data sheets for applicable SFR bits.

1.4.5 Memory Organization

Bytes are located at even or odd addresses. Words are only located at even addresses as shown in Figure 1-3. When using word instructions, only even addresses may be used. The low byte of a word is always an even address. The high byte is at the next odd address. For example, if a data word is located at address xxx4h, then the low byte of that data word is located at address xxx4h, and the high byte of that word is located at address xxx5h.

![Figure 1-3. Bits, Bytes, and Words in a Byte-Organized Memory](image-url)
1.5 MSP430x2xx Family Enhancements

Table 1-1 highlights enhancements made to the MSP430x2xx family. The enhancements are discussed fully in the following chapters, or in the case of improved device parameters, shown in the device-specific data sheet.

<table>
<thead>
<tr>
<th>Subject</th>
<th>Enhancement</th>
</tr>
</thead>
</table>
| Reset                    | • Brownout reset is included on all MSP430x2xx devices.  
• PORIFG and RSTIFG flags have been added to IFG1 to indicate the cause of a reset.  
• An instruction fetch from the address range 0x0000 - 0x01FF will reset the device.                                                                 |
| Watchdog Timer           | • All MSP430x2xx devices integrate the Watchdog Timer+ module (WDT+). The WDT+ ensures the clock source for the timer is never disabled.                                                                         |
| Basic Clock System       | • The LFXT1 oscillator has selectable load capacitors in LF mode.  
• The LFXT1 supports up to 16-MHz crystals in HF mode.  
• The LFXT1 includes oscillator fault detection in LF mode.  
• The XIN and XOUT pins are shared function pins on 20- and 28-pin devices.  
• The external $R_{OSC}$ feature of the DCO not supported on some devices. Software should not set the LSB of the BCSCTL2 register in this case. See the device-specific data sheet for details.  
• The DCO operating frequency has been significantly increased.  
• The DCO temperature stability has been significantly improved.                                                                 |
| Flash Memory             | • The information memory has 4 segments of 64 bytes each.  
• Segment A is individually locked with the LOCKA bit.  
• All information if protected from mass erase with the LOCKA bit.  
• Segment erases can be interrupted by an interrupt.  
• Flash updates can be aborted by an interrupt.  
• Flash programming voltage has been lowered to 2.2 V  
• Program/erase time has been reduced.  
• Clock failure aborts a flash update.                                                                 |
| Digital I/O              | • All ports have integrated pullup/pulldown resistors.  
• P2.6 and P2.7 functions have been added to 20- and 28-pin devices. These are shared functions with XIN and XOUT. Software must not clear the P2SELx bits for these pins if crystal operation is required. |
| Comparator_A             | • Comparator_A has expanded input capability with a new input multiplexer.                                                                                                                                  |
| Low Power                | • Typical LPM3 current consumption has been reduced almost 50% at 3 V.  
• DCO startup time has been significantly reduced.                                                                                                                                                  |
| Operating frequency      | • The maximum operating frequency is 16 MHz at 3.3 V.                                                                                                                                                      |
| BSL                      | • An incorrect password causes a mass erase.  
• BSL entry sequence is more robust to prevent accidental entry and erasure.                                                                                                                               |
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Chapter 2

System Resets, Interrupts, and Operating Modes

This chapter describes the MSP430x2xx system resets, interrupts, and operating modes.

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2.2 Interrupts.................................................................................................32
2.3 Operating Modes....................................................................................39
2.4 Principles for Low-Power Applications..................................................41
2.5 Connection of Unused Pins....................................................................42
### 2.1 System Reset and Initialization

The system reset circuitry shown in Figure 2-1 sources both a power-on reset (POR) and a power-up clear (PUC) signal. Different events trigger these reset signals and different initial conditions exist depending on which signal was generated.

**Figure 2-1. Power-On Reset and Power-Up Clear Schematic**

A POR is a device reset. A POR is only generated by the following three events:

- Powering up the device
- A low signal on the RST/NMI pin when configured in the reset mode
- An SVS low condition when PORON = 1.

A PUC is always generated when a POR is generated, but a POR is not generated by a PUC. The following events trigger a PUC:

- A POR signal
- Watchdog timer expiration when in watchdog mode only
- Watchdog timer security key violation
- A Flash memory security key violation
- A CPU instruction fetch from the peripheral address range 0h to 01FFh

#### 2.1.1 Brownout Reset (BOR)

The brownout reset circuit detects low supply voltages such as when a supply voltage is applied to or removed from the $V_{CC}$ terminal. The brownout reset circuit resets the device by triggering a POR signal when power is applied or removed. The operating levels are shown in Figure 2-2.

The POR signal becomes active when $V_{CC}$ crosses the $V_{CC(start)}$ level. It remains active until $V_{CC}$ crosses the $V_{(B.IT+)}$ threshold and the delay $t_{BOR}$ elapses. The delay $t_{BOR}$ is adaptive being longer for a slow ramping $V_{CC}$. The hysteresis $V_{hys(B.IT-)}$ ensures that the supply voltage must drop below $V_{(B.IT-)}$ to generate another POR signal from the brownout reset circuitry.
2.1.2 Device Initial Conditions After System Reset

After a POR, the initial MSP430 conditions are:

- The RST/NMI pin is configured in the reset mode.
- I/O pins are switched to input mode as described in the Digital I/O chapter.
- Other peripheral modules and registers are initialized as described in their respective chapters in this manual.
- Status register (SR) is reset.
- The watchdog timer powers up active in watchdog mode.
- Program counter (PC) is loaded with address contained at reset vector location (0FFFEh). If the reset vectors' content is 0xFFFFh the device will be disabled for minimum power consumption.

2.1.2.1 Software Initialization

After a system reset, user software must initialize the MSP430 for the application requirements. The following must occur:

- Initialize the SP, typically to the top of RAM.
- Initialize the watchdog to the requirements of the application.
- Configure peripheral modules to the requirements of the application.

Additionally, the watchdog timer, oscillator fault, and flash memory flags can be evaluated to determine the source of the reset.
2.2 Interrupts

The interrupt priorities are fixed and defined by the arrangement of the modules in the connection chain as shown in Figure 2-3. The nearer a module is to the CPU/NMIRS, the higher the priority. Interrupt priorities determine what interrupt is taken when more than one interrupt is pending simultaneously.

There are three types of interrupts:

- System reset
- (Non)-maskable NMI
- Maskable

![Figure 2-3. Interrupt Priority](image)

2.2.1 (Non)-Maskable Interrupts (NMI)

(Non)-maskable NMI interrupts are not masked by the general interrupt enable bit (GIE), but are enabled by individual interrupt enable bits (NMIIE, ACCVIE, OFIE). When a NMI interrupt is accepted, all NMI interrupt enable bits are automatically reset. Program execution begins at the address stored in the (non)-maskable interrupt vector, 0FFFCh. User software must set the required NMI interrupt enable bits for the interrupt to be re-enabled. The block diagram for NMI sources is shown in Figure 2-4.

A (non)-maskable NMI interrupt can be generated by three sources:

- An edge on the RST/NMI pin when configured in NMI mode
- An oscillator fault occurs
- An access violation to the flash memory
Figure 2-4. Block Diagram of (Non)-Maskable Interrupt Sources
2.2.1.1 Reset/NMI Pin

At power-up, the RST/NMI pin is configured in the reset mode. The function of the RST/NMI pins is selected in the watchdog control register WDTCTL. If the RST/NMI pin is set to the reset function, the CPU is held in the reset state as long as the RST/NMI pin is held low. After the input changes to a high state, the CPU starts program execution at the word address stored in the reset vector, 0FFFEh, and the RSTIFG flag is set.

If the RST/NMI pin is configured by user software to the NMI function, a signal edge selected by the WDTNMIES bit generates an NMI interrupt if the NMIIE bit is set. The RST/NMI flag NMIIFG is also set.

---

**Note**

**Holding RST/NMI Low**

When configured in the NMI mode, a signal generating an NMI event should not hold the RST/NMI pin low. If a PUC occurs from a different source while the NMI signal is low, the device will be held in the reset state because a PUC changes the RST/NMI pin to the reset function.

---

**Note**

**Modifying WDTNMIES**

When NMI mode is selected and the WDTNMIES bit is changed, an NMI can be generated, depending on the actual level at the RST/NMI pin. When the NMI edge select bit is changed before selecting the NMI mode, no NMI is generated.

---

2.2.1.2 Flash Access Violation

The flash ACCVIFG flag is set when a flash access violation occurs. The flash access violation can be enabled to generate an NMI interrupt by setting the ACCVIE bit. The ACCVIFG flag can then be tested by the NMI interrupt service routine to determine if the NMI was caused by a flash access violation.

---

2.2.1.3 Oscillator Fault

The oscillator fault signal warns of a possible error condition with the crystal oscillator. The oscillator fault can be enabled to generate an NMI interrupt by setting the OFIE bit. The OFIFG flag can then be tested by NMI the interrupt service routine to determine if the NMI was caused by an oscillator fault.

A PUC signal can trigger an oscillator fault, because the PUC switches the LFXT1 to LF mode, therefore switching off the HF mode. The PUC signal also switches off the XT2 oscillator.
2.2.1.4 Example of an NMI Interrupt Handler

The NMI interrupt is a multiple-source interrupt. An NMI interrupt automatically resets the NMIIE, OFIE and ACCVIE interrupt-enable bits. The user NMI service routine resets the interrupt flags and re-enables the interrupt-enable bits according to the application needs as shown in Figure 2-5.

![Diagram of NMI Interrupt Handler]

Figure 2-5. NMI Interrupt Handler

**Note**

Enabling NMI Interrupts with ACCVIE, NMIIE, and OFIE

To prevent nested NMI interrupts, the ACCVIE, NMIIE, and OFIE enable bits should not be set inside of an NMI interrupt service routine.

2.2.2 Maskable Interrupts

Maskable interrupts are caused by peripherals with interrupt capability including the watchdog timer overflow in interval-timer mode. Each maskable interrupt source can be disabled individually by an interrupt enable bit, or all maskable interrupts can be disabled by the general interrupt enable (GIE) bit in the status register (SR).

Each individual peripheral interrupt is discussed in the associated peripheral module chapter in this manual.
2.2.3 Interrupt Processing

When an interrupt is requested from a peripheral and the peripheral interrupt enable bit and GIE bit are set, the interrupt service routine is requested. Only the individual enable bit must be set for (non)-maskable interrupts to be requested.

2.2.3.1 Interrupt Acceptance

The interrupt latency is 5 cycles (CPUx) or 6 cycles (CPU), starting with the acceptance of an interrupt request and lasting until the start of execution of the first instruction of the interrupt-service routine, as shown in Figure 2-6. The interrupt logic executes the following:

1. Any currently executing instruction is completed.
2. The PC, which points to the next instruction, is pushed onto the stack.
3. The SR is pushed onto the stack.
4. The interrupt with the highest priority is selected if multiple interrupts occurred during the last instruction and are pending for service.
5. The interrupt request flag resets automatically on single-source flags. Multiple source flags remain set for servicing by software.
6. The SR is cleared. This terminates any low-power mode. Because the GIE bit is cleared, further interrupts are disabled.
7. The content of the interrupt vector is loaded into the PC: the program continues with the interrupt service routine at that address.

![Figure 2-6. Interrupt Processing](image-url)
### 2.2.3.2 Return From Interrupt

The interrupt handling routine terminates with the instruction:

```plaintext
RETI (return from an interrupt service routine)
```

The return from the interrupt takes 5 cycles (CPU) or 3 cycles (CPUx) to execute the following actions and is illustrated in Figure 2-7.

1. The SR with all previous settings pops from the stack. All previous settings of GIE, CPUOFF, etc. are now in effect, regardless of the settings used during the interrupt service routine.
2. The PC pops from the stack and begins execution at the point where it was interrupted.

#### Figure 2-7. Return From Interrupt

![Return From Interrupt Diagram]

### 2.2.3.3 Interrupt Nesting

Interrupt nesting is enabled if the GIE bit is set inside an interrupt service routine. When interrupt nesting is enabled, any interrupt occurring during an interrupt service routine will interrupt the routine, regardless of the interrupt priorities.
2.2.4 Interrupt Vectors

The interrupt vectors and the power-up starting address are located in the address range 0FFFFh to 0FFC0h, as described in Table 2-1. A vector is programmed by the user with the 16-bit address of the corresponding interrupt service routine. See the device-specific data sheet for the complete interrupt vector list.

It is recommended to provide an interrupt service routine for each interrupt vector that is assigned to a module. A dummy interrupt service routine can consist of just the RETI instruction and several interrupt vectors can point to it.

Unassigned interrupt vectors can be used for regular program code if necessary.

Some module enable bits, interrupt enable bits, and interrupt flags are located in the SFRs. The SFRs are located in the lower address range and are implemented in byte format. SFRs must be accessed using byte instructions. See the device-specific data sheet for the SFR configuration.

### Table 2-1. Interrupt Sources, Flags, and Vectors

<table>
<thead>
<tr>
<th>Interrupt Source</th>
<th>Interrupt Flag</th>
<th>System Interrupt</th>
<th>Word Address</th>
<th>Priority</th>
</tr>
</thead>
</table>
| Power-up, external reset, watchdog, flash password, illegal instruction fetch    | PORIFG
RSTIFG
WDTIFG
KEYV      | Reset               | 0FFFEh             | 31, highest   |
| NMI, oscillator fault, flash memory access violation                             | NMIIFG
OFIFG
ACCVIFG | (non)-maskable (non)-maskable (non)-maskable | 0FFFCh | 30       |
| device-specific                                                                 |                      | 0FFFAh         | 29           |
| device-specific                                                                 |                      | 0FFFBh         | 28           |
| device-specific                                                                 |                      | 0FFFCb         | 27           |
| Watchdog timer                                                                  | WDTIFG               | maskable        | 0FFF4h       | 26       |
| device-specific                                                                 |                      | 0FFFDh         | 25           |
| device-specific                                                                 |                      | 0FFE0h         | 24           |
| device-specific                                                                 |                      | 0FFE1h         | 23           |
| device-specific                                                                 |                      | 0FFE2h         | 22           |
| device-specific                                                                 |                      | 0FFE3h         | 21           |
| device-specific                                                                 |                      | 0FFE4h         | 20           |
| device-specific                                                                 |                      | 0FFE5h         | 19           |
| device-specific                                                                 |                      | 0FFE6h         | 18           |
| device-specific                                                                 |                      | 0FFE7h         | 17           |
| device-specific                                                                 |                      | 0FFE8h         | 16           |
| device-specific                                                                 |                      | 0FFE9h         | 15           |
| device-specific                                                                 |                      | 0FFEAh         | 14           |
| device-specific                                                                 |                      | 0FFEBh         | 13           |
| device-specific                                                                 |                      | 0FFEDh         | 12           |
| device-specific                                                                 |                      | 0FFE4h         | 11           |
| device-specific                                                                 |                      | 0FFFEh         | 10           |
| device-specific                                                                 |                      | 0FFFCb         | 9            |
| device-specific                                                                 |                      | 0FFFBh         | 8            |
| device-specific                                                                 |                      | 0FFFAh         | 7            |
| device-specific                                                                 |                      | 0FFEDh         | 6            |
| device-specific                                                                 |                      | 0FFFCb         | 5            |
| device-specific                                                                 |                      | 0FFFBh         | 4            |
| device-specific                                                                 |                      | 0FFFAh         | 3            |
| device-specific                                                                 |                      | 0FFEDh         | 2            |
Table 2-1. Interrupt Sources, Flags, and Vectors (continued)

<table>
<thead>
<tr>
<th>Interrupt Source</th>
<th>Interrupt Flag</th>
<th>System Interrupt</th>
<th>Word Address</th>
<th>Priority</th>
</tr>
</thead>
<tbody>
<tr>
<td>device-specific</td>
<td></td>
<td>System Interrupt</td>
<td>0FFC2h</td>
<td>1</td>
</tr>
<tr>
<td>device-specific</td>
<td></td>
<td>System Interrupt</td>
<td>0FFC0h</td>
<td>0, lowest</td>
</tr>
</tbody>
</table>

2.3 Operating Modes

The MSP430 family is designed for ultralow-power applications and uses different operating modes shown in Figure 2-9.

The operating modes take into account three different needs:

- Ultralow-power
- Speed and data throughput
- Minimization of individual peripheral current consumption

The MSP430 typical current consumption is shown in Figure 2-8.

![Figure 2-8. Typical Current Consumption of 'F21x1 Devices vs Operating Modes](image)

The low-power modes 0 to 4 are configured with the CPUOFF, OSCOFF, SCG0, and SCG1 bits in the status register. The advantage of including the CPUOFF, OSCOFF, SCG0, and SCG1 mode-control bits in the status register is that the present operating mode is saved onto the stack during an interrupt service routine. Program flow returns to the previous operating mode if the saved SR value is not altered during the interrupt service routine. Program flow can be returned to a different operating mode by manipulating the saved SR value on the stack inside of the interrupt service routine. The mode-control bits and the stack can be accessed with any instruction.

When setting any of the mode-control bits, the selected operating mode takes effect immediately (see Figure 2-9). Peripherals operating with any disabled clock are disabled until the clock becomes active. The peripherals may also be disabled with their individual control register settings. All I/O port pins and RAM/registers are unchanged. Wake up is possible through all enabled interrupts.
Active Mode
CPU Is Active
Peripheral Modules Are Active
LPM0
CPU Off, MCLK Off,
SMCLK On, ACLK On
CPUOFF = 1
SCG0 = 1
SCG1 = 1

LPM1
CPU Off, MCLK Off,
DCO off, SMCLK On,
ACLK On
CPUOFF = 1
SCG0 = 1
SCG1 = 0

LPM2
CPU Off, MCLK Off, SMCLK
Off, DCO Off, ACLK On
CPUOFF = 1
SCG0 = 1
SCG1 = 1
DC Generator Off if DCO
not used for SMCLK

LPM3
CPU Off, MCLK Off, SMCLK
Off, DCO Off, ACLK On
CPUOFF = 1
OSCOFF = 1
SCG0 = 1
SCG1 = 1

LPM4
CPU Off, MCLK Off, DCO
Off, SMCLK Off,
ACLK Off
CPUOFF = 1
OSCOFF = 1
SCG0 = 1
SCG1 = 1
DC Generator Off

CPU and Clocks Status

<table>
<thead>
<tr>
<th>SCG1</th>
<th>SCG0</th>
<th>OSCOFF</th>
<th>CPUOFF</th>
<th>Mode</th>
<th>CPU and Clocks Status</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>Active</td>
<td>CPU is active, all enabled clocks are active</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>LPM0</td>
<td>CPU, MCLK are disabled, SMCLK, ACLK are active</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>0</td>
<td>1</td>
<td>LPM1</td>
<td>CPU, MCLK are disabled. DCO and DC generator are disabled if the DCO is not used for SMCLK. ACLK is active.</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>LPM2</td>
<td>CPU, MCLK, SMCLK, DCO are disabled. DC generator remains enabled. ACLK is active.</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>0</td>
<td>1</td>
<td>LPM3</td>
<td>CPU, MCLK, SMCLK, DCO are disabled. DC generator disabled. ACLK is active.</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>LPM4</td>
<td>CPU and all clocks disabled</td>
</tr>
</tbody>
</table>

Figure 2-9. Operating Modes For Basic Clock System

Table 2-2. Operating Modes For Basic Clock System
2.3.1 Entering and Exiting Low-Power Modes

An enabled interrupt event wakes the MSP430 from any of the low-power operating modes. The program flow is:

- **Enter interrupt service routine:**
  - The PC and SR are stored on the stack
  - The CPUOFF, SCG1, and OSCOFF bits are automatically reset

- **Options for returning from the interrupt service routine:**
  - The original SR is popped from the stack, restoring the previous operating mode.
  - The SR bits stored on the stack can be modified within the interrupt service routine returning to a different operating mode when the RETI instruction is executed.

```assembly
; Enter LPM0 Example
BIS    #GIE+CPUOFF,SR            ; Enter LPM0
; ...                             ; Program stops here
;
; Exit LPM0 Interrupt Service Routine
BIC    #CPUOFF,0(SP)             ; Exit LPM0 on RETI
RETI

; Enter LPM3 Example
BIS    #GIE+CPUOFF+SCG1+SCG0,SR  ; Enter LPM3
; ...                             ; Program stops here
;
; Exit LPM3 Interrupt Service Routine
BIC    #CPUOFF+SCG1+SCG0,0(SP)   ; Exit LPM3 on RETI
RETI
```

2.4 Principles for Low-Power Applications

Often, the most important factor for reducing power consumption is using the MSP430 clock system to maximize the time in LPM3. LPM3 power consumption is less than 2 µA typical with both a real-time clock function and all interrupts active. A 32-kHz watch crystal is used for the ACLK and the CPU is clocked from the DCO (normally off) which has a 1-µs wake-up.

- Use interrupts to wake the processor and control program flow.
- Peripherals should be switched on only when needed.
- Use low-power integrated peripheral modules in place of software driven functions. For example Timer_A and Timer_B can automatically generate PWM and capture external timing, with no CPU resources.
- Calculated branching and fast table look-ups should be used in place of flag polling and long software calculations.
- Avoid frequent subroutine and function calls due to overhead.
- For longer software routines, single-cycle CPU registers should be used.
## 2.5 Connection of Unused Pins

The correct termination of all unused pins is listed in Table 2-3.

<table>
<thead>
<tr>
<th>Pin</th>
<th>Potential</th>
<th>Comment</th>
</tr>
</thead>
<tbody>
<tr>
<td>AVCC</td>
<td>DVCC</td>
<td></td>
</tr>
<tr>
<td>AVSS</td>
<td>DVSS</td>
<td></td>
</tr>
<tr>
<td>VREF+</td>
<td>Open</td>
<td></td>
</tr>
<tr>
<td>VREF-</td>
<td>DVSS</td>
<td></td>
</tr>
<tr>
<td>XIN</td>
<td>DVCC</td>
<td>For dedicated XIN pins only. XIN pins with shared GPIO functions should be programmed to GPIO and follow Px.0 to Px.7 recommendations.</td>
</tr>
<tr>
<td>XOUT</td>
<td>Open</td>
<td>For dedicated XOUT pins only. XOUT pins with shared GPIO functions should be programmed to GPIO and follow Px.0 to Px.7 recommendations.</td>
</tr>
<tr>
<td>XT2IN</td>
<td>DVSS</td>
<td>For dedicated XT2IN pins only. XT2IN pins with shared GPIO functions should be programmed to GPIO and follow Px.0 to Px.7 recommendations.</td>
</tr>
<tr>
<td>XT2OUT</td>
<td>Open</td>
<td>For dedicated XT2OUT pins only. XT2OUT pins with shared GPIO functions should be programmed to GPIO and follow Px.0 to Px.7 recommendations.</td>
</tr>
<tr>
<td>Px.0 to Px.7</td>
<td>Open</td>
<td>Switched to port function, output direction or input with pullup/pulldown enabled</td>
</tr>
<tr>
<td>RST/NMI</td>
<td>DVCC or VCC</td>
<td>47 kΩ pullup with 10 nF (2.2 nF(1)) pulldown</td>
</tr>
<tr>
<td>Test</td>
<td>Open</td>
<td>20xx, 21xx, 22xx devices</td>
</tr>
<tr>
<td>TDO</td>
<td>Open</td>
<td></td>
</tr>
<tr>
<td>TDI</td>
<td>Open</td>
<td></td>
</tr>
<tr>
<td>TMS</td>
<td>Open</td>
<td></td>
</tr>
<tr>
<td>TCK</td>
<td>Open</td>
<td></td>
</tr>
</tbody>
</table>

(1) The pulldown capacitor must not exceed 2.2 nF when using devices with Spy-Bi-Wire interface in Spy-Bi-Wire mode or in 4-wire JTAG mode with TI tools like FET interfaces or GANG programmers.
This chapter describes the MSP430 CPU, addressing modes, and instruction set.

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<thead>
<tr>
<th>Section</th>
<th>Page</th>
</tr>
</thead>
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</tr>
<tr>
<td>3.2 CPU Registers</td>
<td>45</td>
</tr>
<tr>
<td>3.3 Addressing Modes</td>
<td>49</td>
</tr>
<tr>
<td>3.4 Instruction Set</td>
<td>57</td>
</tr>
</tbody>
</table>
3.1 CPU Introduction

The CPU incorporates features specifically designed for modern programming techniques such as calculated branching, table processing, and the use of high-level languages such as C. The CPU can address the complete address range without paging.

The CPU features include:

- RISC architecture with 27 instructions and 7 addressing modes.
- Orthogonal architecture with every instruction usable with every addressing mode.
- Full register access including program counter, status registers, and stack pointer.
- Single-cycle register operations.
- Large 16-bit register file reduces fetches to memory.
- 16-bit address bus allows direct access and branching throughout entire memory range.
- 16-bit data bus allows direct manipulation of word-wide arguments.
- Constant generator provides six most used immediate values and reduces code size.
- Direct memory-to-memory transfers without intermediate register holding.
- Word and byte addressing and instruction formats.

The block diagram of the CPU is shown in Figure 3-1.
3.2 CPU Registers

The CPU incorporates sixteen 16-bit registers. R0, R1, R2, and R3 have dedicated functions. R4 to R15 are working registers for general use.

3.2.1 Program Counter (PC)

The 16-bit program counter (PC/R0) points to the next instruction to be executed. Each instruction uses an even number of bytes (two, four, or six), and the PC is incremented accordingly. Instruction accesses in the 64-KB address space are performed on word boundaries, and the PC is aligned to even addresses. Figure 3-2 shows the program counter.
The PC can be addressed with all instructions and addressing modes. A few examples:

- `MOV #LABEL,PC` ; Branch to address LABEL
- `MOV LABEL,PC` ; Branch to address contained in LABEL
- `MOV @R14,PC` ; Branch indirect to address in R14

### 3.2.2 Stack Pointer (SP)

The stack pointer (SP/R1) is used by the CPU to store the return addresses of subroutine calls and interrupts. It uses a predecrement, postincrement scheme. In addition, the SP can be used by software with all instructions and addressing modes. Figure 3-3 shows the SP. The SP is initialized into RAM by the user, and is aligned to even addresses.

Figure 3-4 shows stack usage.

#### Figure 3-3. Stack Counter

<table>
<thead>
<tr>
<th>15</th>
<th>14</th>
<th>13</th>
<th>12</th>
<th>11</th>
<th>10</th>
<th>9</th>
<th>8</th>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>Program Counter Bits 15 to 1</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0</td>
</tr>
</tbody>
</table>

MOV 2(SP),R6 ; Item I2 -> R6
MOV R7,0(SP) ; Overwrite TOS with R7
PUSH #0123h ; Put 0123h onto TOS
POP R8 ; R8 = 0123h

#### Figure 3-4. Stack Usage

The special cases of using the SP as an argument to the PUSH and POP instructions are described and shown in Figure 3-5.

#### Figure 3-5. PUSH SP - POP SP Sequence

The stack pointer is changed after a PUSH SP instruction. The stack pointer is not changed after a POP SP instruction. The POP SP instruction places SP1 into the stack pointer SP (SP2=SP1)
3.2.3 Status Register (SR)

The status register (SR/R2), used as a source or destination register, can be used in the register mode only addressed with word instructions. The remaining combinations of addressing modes are used to support the constant generator. Figure 3-6 shows the SR bits.

![Figure 3-6. Status Register Bits](image)

Table 3-1 describes the status register bits.

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>V</td>
<td>Overflow bit. This bit is set when the result of an arithmetic operation overflows the signed-variable range. ADD(.B), ADDC(.B) Set when: Positive + Positive = Negative Negative + Negative = Positive Otherwise reset SUB(.B), SUBC(.B), CMP(.B) Set when: Positive – Negative = Negative Negative – Positive = Positive Otherwise reset</td>
</tr>
<tr>
<td>SCG1</td>
<td>System clock generator 1. When set, turns off the SMCLK.</td>
</tr>
<tr>
<td>SCG0</td>
<td>System clock generator 0. When set, turns off the DCO dc generator, if DCOCLK is not used for MCLK or SMCLK.</td>
</tr>
<tr>
<td>OSCOFF</td>
<td>Oscillator Off. When set, turns off the LFXT1 crystal oscillator, when LFXT1CLK is not use for MCLK or SMCLK.</td>
</tr>
<tr>
<td>CPUOFF</td>
<td>CPU off. When set, turns off the CPU.</td>
</tr>
<tr>
<td>GIE</td>
<td>General interrupt enable. When set, enables maskable interrupts. When reset, all maskable interrupts are disabled.</td>
</tr>
<tr>
<td>N</td>
<td>Negative bit. Set when the result of a byte or word operation is negative and cleared when the result is not negative. Word operation: N is set to the value of bit 15 of the result. Byte operation: N is set to the value of bit 7 of the result.</td>
</tr>
<tr>
<td>Z</td>
<td>Zero bit. Set when the result of a byte or word operation is 0 and cleared when the result is not 0.</td>
</tr>
<tr>
<td>C</td>
<td>Carry bit. Set when the result of a byte or word operation produced a carry and cleared when no carry occurred.</td>
</tr>
</tbody>
</table>

3.2.4 Constant Generator Registers CG1 and CG2

Six commonly-used constants are generated with the constant generator registers R2 and R3, without requiring an additional 16-bit word of program code. The constants are selected with the source-register addressing modes (As), as described in Table 3-2.

<table>
<thead>
<tr>
<th>Register</th>
<th>As</th>
<th>Constant</th>
<th>Remarks</th>
</tr>
</thead>
<tbody>
<tr>
<td>R2</td>
<td>00</td>
<td>00000h</td>
<td>Register mode</td>
</tr>
<tr>
<td>R2</td>
<td>01</td>
<td>00004h</td>
<td>Absolute address mode</td>
</tr>
<tr>
<td>R2</td>
<td>10</td>
<td>00008h</td>
<td>+4, bit processing</td>
</tr>
<tr>
<td>R2</td>
<td>11</td>
<td>00003h</td>
<td>+8, bit processing</td>
</tr>
<tr>
<td>R3</td>
<td>00</td>
<td>00000h</td>
<td>0, word processing</td>
</tr>
<tr>
<td>R3</td>
<td>01</td>
<td>00001h</td>
<td>+1</td>
</tr>
<tr>
<td>R3</td>
<td>10</td>
<td>00002h</td>
<td>+2, bit processing</td>
</tr>
</tbody>
</table>
The constant generator advantages are:

- No special instructions required
- No additional code word for the six constants
- No code memory access required to retrieve the constant

The assembler uses the constant generator automatically if one of the six constants is used as an immediate source operand. Registers R2 and R3, used in the constant mode, cannot be addressed explicitly; they act as source-only registers.

### 3.2.4.1 Constant Generator - Expanded Instruction Set

The RISC instruction set of the MSP430 has only 27 instructions. However, the constant generator allows the MSP430 assembler to support 24 additional, emulated instructions. For example, the single-operand instruction

```
CLR   dst
```

is emulated by the double-operand instruction with the same length:

```
MOV   R3,dst
```

where the #0 is replaced by the assembler, and R3 is used with As=00.

```
INC   dst
```

is replaced by:

```
ADD   0(R3),dst
```

### 3.2.5 General-Purpose Registers R4 to R15

The twelve registers, R4-R15, are general-purpose registers. All of these registers can be used as data registers, address pointers, or index values and can be accessed with byte or word instructions as shown in Figure 3-7.

![Figure 3-7. Register-Byte/Byte-Register Operations](image-url)

**Example Register-Byte Operation**

- R5 = 0A28Fh
- R6 = 0203h
- Mem(0203h) = 012h

- ADD.B R5, 0(R6)
- Mem(0203h) = 08Fh

**Example Byte-Register Operation**

- R5 = 01202h
- R6 = 0223h

- ADD.B @R6,R5
- Mem(0223h) = 05Fh
### Example Register-Byte Operation

<table>
<thead>
<tr>
<th>Address</th>
<th>Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>Mem (0203h)</td>
<td>0A1h</td>
</tr>
<tr>
<td>R5</td>
<td>00061h</td>
</tr>
</tbody>
</table>

C = 0, Z = 0, N = 1
C = 0, Z = 0, N = 0

+ (Addressed byte) -> (Addressed byte)
+ (Low byte of register) -> (Low byte of register, zero to High byte)

### Example Byte-Register Operation

<table>
<thead>
<tr>
<th>Address</th>
<th>Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>+ 012h</td>
<td>0A1h</td>
</tr>
<tr>
<td>+ 002h</td>
<td>00061h</td>
</tr>
</tbody>
</table>

### 3.3 Addressing Modes

Seven addressing modes for the source operand and four addressing modes for the destination operand can address the complete address space with no exceptions. The bit numbers in Table 3-3 describe the contents of the As (source) and Ad (destination) mode bits.

### Table 3-3. Source and Destination Operand Addressing Modes

<table>
<thead>
<tr>
<th>As/Ad</th>
<th>Addressing Mode</th>
<th>Syntax</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>00/0</td>
<td>Register mode</td>
<td>Rn</td>
<td>Register contents are operand</td>
</tr>
<tr>
<td>01/1</td>
<td>Indexed mode</td>
<td>X(Rn)</td>
<td>(Rn + X) points to the operand. X is stored in the next word.</td>
</tr>
<tr>
<td>01/1</td>
<td>Symbolic mode</td>
<td>ADDR</td>
<td>(PC + X) points to the operand. X is stored in the next word. Indexed mode X(PC) is used.</td>
</tr>
<tr>
<td>01/1</td>
<td>Absolute mode</td>
<td>&amp;ADDR</td>
<td>The word following the instruction contains the absolute address. X is stored in the next word. Indexed mode X(SR) is used.</td>
</tr>
<tr>
<td>10/-</td>
<td>Indirect register mode</td>
<td>@Rn</td>
<td>Rn is used as a pointer to the operand.</td>
</tr>
<tr>
<td>11/-</td>
<td>Indirect autoincrement</td>
<td>@Rn+</td>
<td>Rn is used as a pointer to the operand. Rn is incremented afterwards by 1 for .B instructions and by 2 for .W instructions.</td>
</tr>
<tr>
<td>11/-</td>
<td>Immediate mode</td>
<td>#N</td>
<td>The word following the instruction contains the immediate constant N. Indirect autoincrement mode @PC+ is used.</td>
</tr>
</tbody>
</table>

The seven addressing modes are explained in detail in the following sections. Most of the examples show the same addressing mode for the source and destination, but any valid combination of source and destination addressing modes is possible in an instruction.

### Note

**Use of Labels EDE, TONI, TOM, and LEO**

Throughout MSP430 documentation EDE, TONI, TOM, and LEO are used as generic labels. They are only labels. They have no special meaning.
3.3.1 Register Mode

The register mode is described in Table 3-4.

<table>
<thead>
<tr>
<th>Assembler Code</th>
<th>Content of ROM</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOV R10,R11</td>
<td>MOV R10,R11</td>
</tr>
</tbody>
</table>

- **Length:** One or two words
- **Operation:** Move the content of R10 to R11. R10 is not affected.
- **Comment:** Valid for source and destination
- **Example:** MOV R10,R11

![Before: R10 0A023h, R11 0FA15h, PC PCold
After: R10 0A023h, R11 0A023h, PC PCold + 2]

---

**Note**

**Data in Registers**

The data in the register can be accessed using word or byte instructions. If byte instructions are used, the high byte is always 0 in the result. The status bits are handled according to the result of the byte instructions.
3.3.2 Indexed Mode

The indexed mode is described in Table 3-5.

Table 3-5. Indexed Mode Description

<table>
<thead>
<tr>
<th>Assembler Code</th>
<th>Content of ROM</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOV 2(R5), 6(R6)</td>
<td>MOV X(R5), Y(R6)</td>
</tr>
<tr>
<td></td>
<td>$X = 2$</td>
</tr>
<tr>
<td></td>
<td>$Y = 6$</td>
</tr>
</tbody>
</table>

**Length:** Two or three words

**Operation:** Move the contents of the source address (contents of R5 + 2) to the destination address (contents of R6 + 6). The source and destination registers (R5 and R6) are not affected. In indexed mode, the program counter is incremented automatically so that program execution continues with the next instruction.

**Comment:** Valid for source and destination

**Example:**

```
MOV 2(R5), 6(R6);  
```

---

<table>
<thead>
<tr>
<th>Before:</th>
<th>Address</th>
<th>Register</th>
<th>After:</th>
<th>Address</th>
<th>Register</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Space</td>
<td></td>
<td></td>
<td>Space</td>
<td></td>
</tr>
<tr>
<td>0FF16h</td>
<td>00006h</td>
<td>R5</td>
<td>0FF16h</td>
<td>00006h</td>
<td>R5</td>
</tr>
<tr>
<td>0FF14h</td>
<td>00002h</td>
<td>R6</td>
<td>0FF14h</td>
<td>00002h</td>
<td>R6</td>
</tr>
<tr>
<td>0FF12h</td>
<td>04596h</td>
<td>PC</td>
<td>0FF12h</td>
<td>04596h</td>
<td>PC</td>
</tr>
<tr>
<td>01094h</td>
<td>0xxxxh</td>
<td></td>
<td>0108Ch</td>
<td>01094h</td>
<td></td>
</tr>
<tr>
<td>01092h</td>
<td>05555h</td>
<td></td>
<td>01092h</td>
<td>01092h</td>
<td></td>
</tr>
<tr>
<td>01090h</td>
<td>0xxxxx</td>
<td></td>
<td>01090h</td>
<td>0xxxxx</td>
<td></td>
</tr>
<tr>
<td>01084h</td>
<td>0xxxxx</td>
<td></td>
<td>01080h</td>
<td>01084h</td>
<td></td>
</tr>
<tr>
<td>01082h</td>
<td>01234h</td>
<td></td>
<td>01082h</td>
<td>01082h</td>
<td></td>
</tr>
<tr>
<td>01080h</td>
<td>0xxxxx</td>
<td></td>
<td>01080h</td>
<td>0xxxxx</td>
<td></td>
</tr>
</tbody>
</table>
3.3.3 Symbolic Mode

The symbolic mode is described in Table 3-6.

Table 3-6. Symbolic Mode Description

<table>
<thead>
<tr>
<th>Assembler Code</th>
<th>Content of ROM</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOV EDE,TONI</td>
<td>MOV X(PC),Y(PC)</td>
</tr>
<tr>
<td>X = EDE – PC</td>
<td></td>
</tr>
<tr>
<td>Y = TONI – PC</td>
<td></td>
</tr>
</tbody>
</table>

Length: Two or three words

Operation: Move the contents of the source address EDE (contents of PC + X) to the destination address TONI (contents of PC + Y). The words after the instruction contain the differences between the PC and the source or destination addresses. The assembler computes and inserts offsets X and Y automatically. With symbolic mode, the program counter (PC) is incremented automatically so that program execution continues with the next instruction.

Comment: Valid for source and destination

Example:

```
MOV EDE,TONI ;Source address EDE = 0F016h
;Dest. address TONI = 01114h

Before: Address Space
0FF16h 0FF14h 0FF12h
Register:
011FEh 0F102h 04090h
PC:
0FF16h 0FF14h 0FF12h

After: Address Space
0FF16h 0FF14h 0FF12h
Register:
011FEh 0F102h 04090h
PC:
0FF16h 0FF14h 0FF12h
```
3.3.4 Absolute Mode

The absolute mode is described in Table 3-7.

Table 3-7. Absolute Mode Description

<table>
<thead>
<tr>
<th>Assembler Code</th>
<th>Content of ROM</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOV &amp;EDE,&amp;TONI</td>
<td>MOV X(0),Y(0)</td>
</tr>
<tr>
<td>X = EDE</td>
<td></td>
</tr>
<tr>
<td>Y = TONI</td>
<td></td>
</tr>
</tbody>
</table>

Length: Two or three words

Operation: Move the contents of the source address EDE to the destination address TONI. The words after the instruction contain the absolute address of the source and destination addresses. With absolute mode, the PC is incremented automatically so that program execution continues with the next instruction.

Comment: Valid for source and destination

Example:

MOV &EDE,&TONI ;Source address EDE = 0F016h
                  ;Dest. address TONI = 01114h

Before:          After:

<table>
<thead>
<tr>
<th>Address</th>
<th>Register</th>
<th>Address</th>
<th>Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>Space</td>
<td></td>
<td>Space</td>
<td></td>
</tr>
<tr>
<td>0FF16h</td>
<td>01114h</td>
<td>0FF16h</td>
<td>01114h</td>
</tr>
<tr>
<td>0FF14h</td>
<td>0FO16h</td>
<td>0FF14h</td>
<td>0FO16h</td>
</tr>
<tr>
<td>0FF12h</td>
<td>04292h PC</td>
<td>0FF12h</td>
<td>04292h</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0F018h</td>
<td>0xxxxx</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0F016h</td>
<td>0A123h</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0F014h</td>
<td>0xxxxx</td>
</tr>
<tr>
<td>01116h</td>
<td>0xxxxx</td>
<td>01116h</td>
<td>0xxxxx</td>
</tr>
<tr>
<td>01114h</td>
<td>01234h</td>
<td>01114h</td>
<td>0A123h</td>
</tr>
<tr>
<td>01112h</td>
<td>0xxxxx</td>
<td>01112h</td>
<td>0xxxxx</td>
</tr>
</tbody>
</table>

This address mode is mainly for hardware peripheral modules that are located at an absolute, fixed address. These are addressed with absolute mode to ensure software transportability (for example, position-independent code).
3.3.5 Indirect Register Mode

The indirect register mode is described in Table 3-8.

Table 3-8. Indirect Mode Description

<table>
<thead>
<tr>
<th>Assembler Code</th>
<th>Content of ROM</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOV @R10, 0(R11)</td>
<td>MOV @R10, 0(R11)</td>
</tr>
</tbody>
</table>

Length: One or two words

Operation: Move the contents of the source address (contents of R10) to the destination address (contents of R11). The registers are not modified.

Comment: Valid only for source operand. The substitute for destination operand is 0(Rd).

Example: MOV.B @R10, 0(R11)
3.3.6 Indirect Autoincrement Mode

The indirect autoincrement mode is described in Table 3-9.

**Table 3-9. Indirect Autoincrement Mode Description**

<table>
<thead>
<tr>
<th>Assembler Code</th>
<th>Content of ROM</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOV @R10+,0(R11)</td>
<td>MOV @R10+,0(R11)</td>
</tr>
</tbody>
</table>

Length: One or two words

Operation: Move the contents of the source address (contents of R10) to the destination address (contents of R11). Register R10 is incremented by 1 for a byte operation, or 2 for a word operation after the fetch; it points to the next address without any overhead. This is useful for table processing.

Comment: Valid only for source operand. The substitute for destination operand is 0(Rd) plus second instruction INCD Rd.

Example: MOV @R10+,0(R11)

The auto-incrementing of the register contents occurs after the operand is fetched. This is shown in Figure 3-8.

**Figure 3-8. Operand Fetch Operation**
3.3.7 Immediate Mode

The immediate mode is described in Table 3-10.

### Table 3-10. Immediate Mode Description

<table>
<thead>
<tr>
<th>Assembler Code</th>
<th>Content of ROM</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOV #45h,TONI</td>
<td>MOV @(PC)+,X(PC)</td>
</tr>
</tbody>
</table>

<p>| | |</p>
<table>
<thead>
<tr>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>MOV #45h,TONI</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Length: Two or three words

It is one word less if a constant of CG1 or CG2 can be used.

Operation: Move the immediate constant 45h, which is contained in the word following the instruction, to destination address TONI. When fetching the source, the program counter points to the word following the instruction and moves the contents to the destination.

Comment: Valid only for a source operand.

Example: MOV #45h,TONI
3.4 Instruction Set

The complete MSP430 instruction set consists of 27 core instructions and 24 emulated instructions. The core instructions are instructions that have unique op-codes decoded by the CPU. The emulated instructions are instructions that make code easier to write and read, but do not have op-codes themselves, instead they are replaced automatically by the assembler with an equivalent core instruction. There is no code or performance penalty for using emulated instruction.

There are three core instruction formats:

- Dual operand
- Single operand
- Jump

All single-operand and dual-operand instructions can be byte or word instructions by using .B or .W extensions. Byte instructions are used to access byte data or byte peripherals. Word instructions are used to access word data or word peripherals. If no extension is used, the instruction is a word instruction.

The source and destination of an instruction are defined by the following fields:

<table>
<thead>
<tr>
<th>Field</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>src</td>
<td>The source operand defined by As and S-reg</td>
</tr>
<tr>
<td>dst</td>
<td>The destination operand defined by Ad and D-reg</td>
</tr>
<tr>
<td>As</td>
<td>The addressing bits responsible for the addressing mode used for the source (src)</td>
</tr>
<tr>
<td>S-reg</td>
<td>The working register used for the source (src)</td>
</tr>
<tr>
<td>Ad</td>
<td>The addressing bits responsible for the addressing mode used for the destination (dst)</td>
</tr>
<tr>
<td>D-reg</td>
<td>The working register used for the destination (dst)</td>
</tr>
<tr>
<td>B/W</td>
<td>Byte or word operation: 0: word operation 1: byte operation</td>
</tr>
</tbody>
</table>

Note

Destination Address

Destination addresses are valid anywhere in the memory map. However, when using an instruction that modifies the contents of the destination, the user must ensure the destination address is writable. For example, a masked-ROM location would be a valid destination address, but the contents are not modifiable, so the results of the instruction would be lost.
3.4.1 Double-Operand (Format I) Instructions

Figure 3-9 illustrates the double-operand instruction format.

![Figure 3-9. Double Operand Instruction Format](image)

Table 3-11 lists and describes the double operand instructions.

<table>
<thead>
<tr>
<th>Mnemonic</th>
<th>S-Reg, D-Reg</th>
<th>Operation</th>
<th>Status Bits</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOV(.B)</td>
<td>src,dst</td>
<td>src → dst</td>
<td>V N Z C</td>
</tr>
<tr>
<td>ADD(.B)</td>
<td>src,dst</td>
<td>src + dst → dst</td>
<td>* * * *</td>
</tr>
<tr>
<td>ADC(.B)</td>
<td>src,dst</td>
<td>src + dst + C → dst</td>
<td>* * * *</td>
</tr>
<tr>
<td>SUB(.B)</td>
<td>src,dst</td>
<td>dst + .not.src + 1 → dst</td>
<td>* * * *</td>
</tr>
<tr>
<td>SUBC(.B)</td>
<td>src,dst</td>
<td>dst + .not.src + C → dst</td>
<td>* * * *</td>
</tr>
<tr>
<td>CMP(.B)</td>
<td>src,dst</td>
<td>dst - src</td>
<td>* * * *</td>
</tr>
<tr>
<td>DADD(.B)</td>
<td>src,dst</td>
<td>src + dst + C → dst (decimally)</td>
<td>* * * *</td>
</tr>
<tr>
<td>BIT(.B)</td>
<td>src,dst</td>
<td>src .and. dst</td>
<td>0 * * *</td>
</tr>
<tr>
<td>BIC(.B)</td>
<td>src,dst</td>
<td>not.src .and. dst → dst</td>
<td>- - - -</td>
</tr>
<tr>
<td>BIS(.B)</td>
<td>src,dst</td>
<td>src .or. dst → dst</td>
<td>- - - -</td>
</tr>
<tr>
<td>XOR(.B)</td>
<td>src,dst</td>
<td>src .xor. dst → dst</td>
<td>* * * *</td>
</tr>
<tr>
<td>AND(.B)</td>
<td>src,dst</td>
<td>src .and. dst → dst</td>
<td>0 * * *</td>
</tr>
</tbody>
</table>

* The status bit is affected
– The status bit is not affected
0 The status bit is cleared
1 The status bit is set

Note

**Instructions** CMP and SUB

The instructions CMP and SUB are identical except for the storage of the result. The same is true for the BIT and AND instructions.
3.4.2 Single-Operand (Format II) Instructions

Figure 3-10 illustrates the single-operand instruction format.

![Figure 3-10. Single Operand Instruction Format](image)

<table>
<thead>
<tr>
<th>15</th>
<th>14</th>
<th>13</th>
<th>12</th>
<th>11</th>
<th>10</th>
<th>9</th>
<th>8</th>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>Op-code</td>
<td>B/W</td>
<td>Ad</td>
<td>D/S-Reg</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Table 3-12 lists and describes the single operand instructions.

<table>
<thead>
<tr>
<th>Mnemonic</th>
<th>S-Reg, D-Reg</th>
<th>Operation</th>
<th>Status Bits</th>
</tr>
</thead>
<tbody>
<tr>
<td>RRC(.B)</td>
<td>dst</td>
<td>C → MSB →.......LSB → C</td>
<td>* * * *</td>
</tr>
<tr>
<td>RRA(.B)</td>
<td>dst</td>
<td>MSB → MSB →....LSB → C</td>
<td>0 * * *</td>
</tr>
<tr>
<td>PUSH(.B)</td>
<td>src</td>
<td>SP – 2 → SP, src → @SP</td>
<td>- - - -</td>
</tr>
<tr>
<td>SWPB</td>
<td>dst</td>
<td>Swap bytes</td>
<td>- - - -</td>
</tr>
<tr>
<td>CALL</td>
<td>dst</td>
<td>SP – 2 → SP, PC+2 → @SP</td>
<td>- - - -</td>
</tr>
<tr>
<td></td>
<td></td>
<td>dst → PC</td>
<td></td>
</tr>
<tr>
<td>RETI</td>
<td></td>
<td>TOS → SR, SP + 2 → SP</td>
<td>* * * *</td>
</tr>
<tr>
<td></td>
<td></td>
<td>TOS → PC, SP + 2 → SP</td>
<td></td>
</tr>
<tr>
<td>SXT</td>
<td>dst</td>
<td>Bit 7 → Bit 8.......Bit 15</td>
<td>0 * * *</td>
</tr>
</tbody>
</table>

* The status bit is affected
– The status bit is not affected
0 The status bit is cleared
1 The status bit is set

All addressing modes are possible for the CALL instruction. If the symbolic mode (ADDRESS), the immediate mode (#N), the absolute mode (&EDE) or the indexed mode x(RN) is used, the word that follows contains the address information.
3.4.3 Jumps

Figure 3-11 shows the conditional-jump instruction format.

**Figure 3-11. Jump Instruction Format**

<table>
<thead>
<tr>
<th>15</th>
<th>14</th>
<th>13</th>
<th>12</th>
<th>11</th>
<th>10</th>
<th>9</th>
<th>8</th>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Op-code</td>
<td>C</td>
<td>10-Bit PC Offset</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Table 3-13 lists and describes the jump instructions

**Table 3-13. Jump Instructions**

<table>
<thead>
<tr>
<th>Mnemonic</th>
<th>S-Reg, D-Reg</th>
<th>Operation</th>
</tr>
</thead>
<tbody>
<tr>
<td>JEQ/JZ</td>
<td>Label</td>
<td>Jump to label if zero bit is set</td>
</tr>
<tr>
<td>JNE/JNZ</td>
<td>Label</td>
<td>Jump to label if zero bit is reset</td>
</tr>
<tr>
<td>JC</td>
<td>Label</td>
<td>Jump to label if carry bit is set</td>
</tr>
<tr>
<td>JNC</td>
<td>Label</td>
<td>Jump to label if carry bit is reset</td>
</tr>
<tr>
<td>JN</td>
<td>Label</td>
<td>Jump to label if negative bit is set</td>
</tr>
<tr>
<td>JGE</td>
<td>Label</td>
<td>Jump to label if ((N \text{ XOR } V) = 0)</td>
</tr>
<tr>
<td>JL</td>
<td>Label</td>
<td>Jump to label if ((N \text{ XOR } V) = 1)</td>
</tr>
<tr>
<td>JMP</td>
<td>Label</td>
<td>Jump to label unconditionally</td>
</tr>
</tbody>
</table>

Conditional jumps support program branching relative to the PC and do not affect the status bits. The possible jump range is from \(-511\) to \(+512\) words relative to the PC value at the jump instruction. The 10-bit program-counter offset is treated as a signed 10-bit value that is doubled and added to the program counter:

\[
PC_{\text{new}} = PC_{\text{old}} + 2 + PC_{\text{offset}} \times 2
\]
3.4.4 Instruction Cycles and Lengths

The number of CPU clock cycles required for an instruction depends on the instruction format and the addressing modes used - not the instruction itself. The number of clock cycles refers to the MCLK.

3.4.4.1 Interrupt and Reset Cycles

Table 3-14 lists the CPU cycles for interrupt overhead and reset.

<table>
<thead>
<tr>
<th>Action</th>
<th>No. of Cycles</th>
<th>Length of Instruction</th>
</tr>
</thead>
<tbody>
<tr>
<td>Return from interrupt (RETI)</td>
<td>5</td>
<td>1</td>
</tr>
<tr>
<td>Interrupt accepted</td>
<td>6</td>
<td>-</td>
</tr>
<tr>
<td>WDT reset</td>
<td>4</td>
<td>-</td>
</tr>
<tr>
<td>Reset (RST/NMI)</td>
<td>4</td>
<td>-</td>
</tr>
</tbody>
</table>

3.4.4.2 Format-II (Single Operand) Instruction Cycles and Lengths

Table 3-15 lists the length and CPU cycles for all addressing modes of format-II instructions.

<table>
<thead>
<tr>
<th>Addressing Mode</th>
<th>RRA, RRC SWPB, SXT</th>
<th>PUSH</th>
<th>CALL</th>
<th>Length of Instruction</th>
<th>Example</th>
</tr>
</thead>
<tbody>
<tr>
<td>Rn</td>
<td>1</td>
<td>3</td>
<td>4</td>
<td>1</td>
<td>SWPB R5</td>
</tr>
<tr>
<td>@Rn</td>
<td>3</td>
<td>4</td>
<td>4</td>
<td>1</td>
<td>RRC @R9</td>
</tr>
<tr>
<td>@Rn+</td>
<td>3</td>
<td>5</td>
<td>5</td>
<td>1</td>
<td>SWPB @R10+</td>
</tr>
<tr>
<td>#N (See note)</td>
<td>4</td>
<td>5</td>
<td>5</td>
<td>2</td>
<td>CALL #0F000h</td>
</tr>
<tr>
<td>X(Rn)</td>
<td>4</td>
<td>5</td>
<td>5</td>
<td>2</td>
<td>CALL 2(R7)</td>
</tr>
<tr>
<td>EDE</td>
<td>4</td>
<td>5</td>
<td>5</td>
<td>2</td>
<td>PUSH EDE</td>
</tr>
<tr>
<td>&amp;EDE</td>
<td>4</td>
<td>5</td>
<td>5</td>
<td>2</td>
<td>SXT &amp;EDE</td>
</tr>
</tbody>
</table>

Note

Instruction Format II Immediate Mode

Do not use instruction RRA, RRC, SWPB, and SXT with the immediate mode in the destination field. Use of these in the immediate mode results in an unpredictable program operation.

3.4.4.3 Format-III (Jump) Instruction Cycles and Lengths

All jump instructions require one code word, and take two CPU cycles to execute, regardless of whether the jump is taken or not.
3.4.4.4 Format-I (Double Operand) Instruction Cycles and Lengths

Table 3-16 lists the length and CPU cycles for all addressing modes of format-I instructions.

<table>
<thead>
<tr>
<th>Addressing Mode</th>
<th>No. of Cycles</th>
<th>Length of Instruction</th>
<th>Example</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Src</strong></td>
<td><strong>Dst</strong></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Rn</td>
<td>Rn</td>
<td>1</td>
<td>MOV R5,R8</td>
</tr>
<tr>
<td></td>
<td>PC</td>
<td>2</td>
<td>BR R9</td>
</tr>
<tr>
<td></td>
<td>x(Rm)</td>
<td>4</td>
<td>ADD R5,4(R6)</td>
</tr>
<tr>
<td></td>
<td>EDE</td>
<td>4</td>
<td>XOR R8,EDE</td>
</tr>
<tr>
<td></td>
<td>&amp;EDE</td>
<td>4</td>
<td>MOV R5,&amp;EDE</td>
</tr>
<tr>
<td>@Rn</td>
<td>Rm</td>
<td>2</td>
<td>AND @R4,R5</td>
</tr>
<tr>
<td></td>
<td>PC</td>
<td>2</td>
<td>BR @R8</td>
</tr>
<tr>
<td></td>
<td>x(Rm)</td>
<td>5</td>
<td>XOR @R5,8(R6)</td>
</tr>
<tr>
<td></td>
<td>EDE</td>
<td>5</td>
<td>MOV @R5,EDE</td>
</tr>
<tr>
<td></td>
<td>&amp;EDE</td>
<td>5</td>
<td>MOV @R9+,@EDE</td>
</tr>
<tr>
<td>@Rn+</td>
<td>Rm</td>
<td>2</td>
<td>ADD @R5+,R6</td>
</tr>
<tr>
<td></td>
<td>PC</td>
<td>3</td>
<td>BR @R9+</td>
</tr>
<tr>
<td></td>
<td>x(Rm)</td>
<td>5</td>
<td>XOR @R5,8(R6)</td>
</tr>
<tr>
<td></td>
<td>EDE</td>
<td>5</td>
<td>MOV @R9+,@EDE</td>
</tr>
<tr>
<td></td>
<td>&amp;EDE</td>
<td>5</td>
<td>MOV @R9+,@EDE</td>
</tr>
<tr>
<td>#N</td>
<td>Rm</td>
<td>2</td>
<td>MOV #20,R9</td>
</tr>
<tr>
<td></td>
<td>PC</td>
<td>3</td>
<td>BR #2AEh</td>
</tr>
<tr>
<td></td>
<td>x(Rm)</td>
<td>5</td>
<td>MOV #0300h,0(SP)</td>
</tr>
<tr>
<td></td>
<td>EDE</td>
<td>5</td>
<td>ADD #33,EDE</td>
</tr>
<tr>
<td></td>
<td>&amp;EDE</td>
<td>5</td>
<td>ADD #33,&amp;EDE</td>
</tr>
<tr>
<td>x(Rn)</td>
<td>Rm</td>
<td>3</td>
<td>MOV 2(R5),R7</td>
</tr>
<tr>
<td></td>
<td>PC</td>
<td>3</td>
<td>BR 2(R6)</td>
</tr>
<tr>
<td></td>
<td>TONI</td>
<td>6</td>
<td>MOV 4(R7),TONI</td>
</tr>
<tr>
<td></td>
<td>x(Rm)</td>
<td>6</td>
<td>ADD 4(R4),6(R9)</td>
</tr>
<tr>
<td></td>
<td>&amp;TONI</td>
<td>6</td>
<td>MOV 2(R4),&amp;TONI</td>
</tr>
<tr>
<td>EDE</td>
<td>Rm</td>
<td>3</td>
<td>AND EDE,R6</td>
</tr>
<tr>
<td></td>
<td>PC</td>
<td>3</td>
<td>BR EDE</td>
</tr>
<tr>
<td></td>
<td>TONI</td>
<td>6</td>
<td>CMP EDE,TONI</td>
</tr>
<tr>
<td></td>
<td>x(Rm)</td>
<td>6</td>
<td>MOV EDE,0(SP)</td>
</tr>
<tr>
<td></td>
<td>&amp;TONI</td>
<td>6</td>
<td>MOV EDE,&amp;TONI</td>
</tr>
<tr>
<td>&amp;EDE</td>
<td>Rm</td>
<td>3</td>
<td>MOV &amp;EDE,R8</td>
</tr>
<tr>
<td></td>
<td>PC</td>
<td>3</td>
<td>BRA &amp;EDE</td>
</tr>
<tr>
<td></td>
<td>TONI</td>
<td>6</td>
<td>MOV &amp;EDE,TONI</td>
</tr>
<tr>
<td></td>
<td>x(Rm)</td>
<td>6</td>
<td>MOV &amp;EDE,0(SP)</td>
</tr>
<tr>
<td></td>
<td>&amp;TONI</td>
<td>6</td>
<td>MOV &amp;EDE,&amp;TONI</td>
</tr>
</tbody>
</table>

3.4.5 Instruction Set Description

The instruction map is shown in Figure 3-12 and the complete instruction set is summarized in Table 3-17.
Table 3-17. MSP430 Instruction Set

<table>
<thead>
<tr>
<th>Mnemonic</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADC(.B)</td>
<td>Add source and C to destination</td>
</tr>
<tr>
<td>ADD(.B)</td>
<td>Add source to destination</td>
</tr>
<tr>
<td>ADDC(.B)</td>
<td>Add source and C to destination</td>
</tr>
<tr>
<td>AND(.B)</td>
<td>AND source and destination</td>
</tr>
<tr>
<td>BIC(.B)</td>
<td>Clear bits in destination</td>
</tr>
<tr>
<td>BIS(.B)</td>
<td>Set bits in destination</td>
</tr>
<tr>
<td>BIT(.B)</td>
<td>Test bits in destination</td>
</tr>
<tr>
<td>BR(.B)</td>
<td>Branch to destination</td>
</tr>
<tr>
<td>CLR(.B)</td>
<td>Clear destination</td>
</tr>
<tr>
<td>CLRC</td>
<td>Clear C</td>
</tr>
<tr>
<td>CLRN</td>
<td>Clear N</td>
</tr>
<tr>
<td>CLRZ</td>
<td>Clear Z</td>
</tr>
<tr>
<td>CMP(.B)</td>
<td>Compare source and destination</td>
</tr>
<tr>
<td>DADC(.B)</td>
<td>Add C decimally to destination</td>
</tr>
<tr>
<td>DADD(.B)</td>
<td>Add source and C decimally to destination</td>
</tr>
<tr>
<td>DEC(.B)</td>
<td>Decrement destination</td>
</tr>
<tr>
<td>DECD(.B)</td>
<td>Double-decrement destination</td>
</tr>
<tr>
<td>DINT</td>
<td>Disable interrupts</td>
</tr>
<tr>
<td>EINT</td>
<td>Enable interrupts</td>
</tr>
</tbody>
</table>

Figure 3-12. Core Instruction Map
### Table 3-17. MSP430 Instruction Set (continued)

<table>
<thead>
<tr>
<th>Mnemonic</th>
<th>Description</th>
<th>V</th>
<th>N</th>
<th>Z</th>
<th>C</th>
</tr>
</thead>
<tbody>
<tr>
<td>INC(.B) (1)</td>
<td>dst</td>
<td>Increment destination</td>
<td>dst +1 → dst</td>
<td>*</td>
<td>*</td>
</tr>
<tr>
<td>INC(.B) (1)</td>
<td>dst</td>
<td>Double-increment destination</td>
<td>dst+2 → dst</td>
<td>*</td>
<td>*</td>
</tr>
<tr>
<td>INV(.B) (1)</td>
<td>dst</td>
<td>Invert destination</td>
<td>.not. dst → dst</td>
<td>*</td>
<td>*</td>
</tr>
<tr>
<td>JC/JHS label</td>
<td>Jump if C set/Jump if higher or same</td>
<td>-</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>JEQ/JZ label</td>
<td>Jump if equal/Jump if Z set</td>
<td>-</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>JGE label</td>
<td>Jump if greater or equal</td>
<td>-</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>JL label</td>
<td>Jump if less</td>
<td>-</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>JMP label</td>
<td>Jump</td>
<td>PC + 2 × offset → PC</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>JN label</td>
<td>Jump if N set</td>
<td>-</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>JC/JLO label</td>
<td>Jump if C not set/Jump if lower</td>
<td>-</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>JNE/JNZ label</td>
<td>Jump if not equal/Jump if Z not set</td>
<td>-</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>MOV(.B) src,dst</td>
<td>Move source to destination</td>
<td>src → dst</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>NOP (1)</td>
<td>No operation</td>
<td>-</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>POP(.B) (1) dst</td>
<td>Pop item from stack to destination</td>
<td>@SP → dst, SP+2 → SP</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>PUSH(.B) src</td>
<td>Push source onto stack</td>
<td>SP - 2 → SP, src → @SP</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>RET (1)</td>
<td>Return from subroutine</td>
<td>@SP → PC, SP + 2 → SP</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>RETI</td>
<td>Return from interrupt</td>
<td>*</td>
<td>*</td>
<td>*</td>
<td>*</td>
</tr>
<tr>
<td>RLA(.B) (1) dst</td>
<td>Rotate left arithmetically</td>
<td>*</td>
<td>*</td>
<td>*</td>
<td>*</td>
</tr>
<tr>
<td>RLC(.B) (1) dst</td>
<td>Rotate left through C</td>
<td>*</td>
<td>*</td>
<td>*</td>
<td>*</td>
</tr>
<tr>
<td>RRA(.B) dst</td>
<td>Rotate right arithmetically</td>
<td>0</td>
<td>*</td>
<td>*</td>
<td>*</td>
</tr>
<tr>
<td>RRC(.B) dst</td>
<td>Rotate right through C</td>
<td>*</td>
<td>*</td>
<td>*</td>
<td>*</td>
</tr>
<tr>
<td>SBC(.B) (1) dst</td>
<td>Subtract not(C) from destination</td>
<td>dst + 0FFFFh + C → dst</td>
<td>*</td>
<td>*</td>
<td>*</td>
</tr>
<tr>
<td>SETC (1)</td>
<td>Set C</td>
<td>1 → C</td>
<td>-</td>
<td>-</td>
<td>1</td>
</tr>
<tr>
<td>SETN (1)</td>
<td>Set N</td>
<td>1 → N</td>
<td>-</td>
<td>1</td>
<td>-</td>
</tr>
<tr>
<td>SETZ (1)</td>
<td>Set Z</td>
<td>1 → Z</td>
<td>-</td>
<td>-</td>
<td>1</td>
</tr>
<tr>
<td>SUB(.B) src,dst</td>
<td>Subtract source from destination</td>
<td>dst + .not.src + 1 → dst</td>
<td>*</td>
<td>*</td>
<td>*</td>
</tr>
<tr>
<td>SUBC(.B) src,dst</td>
<td>Subtract source and not(C) from dst</td>
<td>dst + .not.src + C → dst</td>
<td>*</td>
<td>*</td>
<td>*</td>
</tr>
<tr>
<td>SWPB dst</td>
<td>Swap bytes</td>
<td>-</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>SXT dst</td>
<td>Extend sign</td>
<td>0</td>
<td>*</td>
<td>*</td>
<td>*</td>
</tr>
<tr>
<td>TST(.B) (1) dst</td>
<td>Test destination</td>
<td>dst + 0FFFFh + 1</td>
<td>0</td>
<td>*</td>
<td>*</td>
</tr>
<tr>
<td>XOR(.B) src,dst</td>
<td>Exclusive OR source and destination</td>
<td>src .xor. dst → dst</td>
<td>*</td>
<td>*</td>
<td>*</td>
</tr>
</tbody>
</table>

(1) Emulated Instruction
3.4.6 Instruction Set Details

3.4.6.1 ADC

*ADC[.W]
Add carry to destination

*ADC.B
Add carry to destination

**Syntax**
ADC dst or ADC.W dst
ADC.B dst

**Operation**
dst + C → dst

**Emulation**
ADDC #0,dst
ADDC.B #0,dst

**Description**
The carry bit (C) is added to the destination operand. The previous contents of the destination are lost.

**Status Bit**
- N: Set if result is negative, reset if positive
- Z: Set if result is zero, reset otherwise
- C: Set if dst was incremented from 0FFFFh to 0000, reset otherwise
- Set if dst was incremented from 0FFh to 00, reset otherwise
- V: Set if an arithmetic overflow occurs, otherwise reset

**Mode Bits**
OSCOFF, CPUOFF, and GIE are not affected.

**Example**
The 16-bit counter pointed to by R13 is added to a 32-bit counter pointed to by R12.

```
ADD   @R13,0(R12)   ; Add LSDs
ADC   2(R12)        ; Add carry to MSD
```

**Example**
The 8-bit counter pointed to by R13 is added to a 16-bit counter pointed to by R12.

```
ADD.B  @R13,0(R12)  ; Add LSDs
ADC.B  1(R12)       ; Add carry to MSD
```
3.4.6.2 ADD

**ADD.W**
Add source to destination

**ADD.B**
Add source to destination

**Syntax**
ADD src,dst or ADD.W src,dst

ADD.B src,dst

**Operation**
src + dst → dst

**Description**
The source operand is added to the destination operand. The source operand is not affected. The previous contents of the destination are lost.

**Status Bits**
N: Set if result is negative, reset if positive
Z: Set if result is zero, reset otherwise
C: Set if there is a carry from the result, cleared if not
V: Set if an arithmetic overflow occurs, otherwise reset

**Mode Bits**
OSCOFF, CPUOFF, and GIE are not affected.

**Example**
R5 is increased by 10. The jump to TONI is performed on a carry.

```
ADD     #10,R5
JC      TONI     ; Carry occurred
......           ; No carry
```

**Example**
R5 is increased by 10. The jump to TONI is performed on a carry.

```
ADD.B   #10,R5   ; Add 10 to Lowbyte of R5
JC      TONI     ; Carry occurred, if (R5) ≥ 246 [0Ah+0F6h]
......           ; No carry
```
3.4.6.3 ADDC

ADDC[W]  Add source and carry to destination
ADDC.B   Add source and carry to destination
Syntax    
ADDC src,dst  or  ADDC.W src,dst
           ADDC.B src,dst
Operation src + dst + C → dst
Description The source operand and the carry bit (C) are added to the destination operand. The source operand is not affected. The previous contents of the destination are lost.
Status Bits N: Set if result is negative, reset if positive
Z: Set if result is zero, reset otherwise
C: Set if there is a carry from the MSB of the result, reset otherwise
V: Set if an arithmetic overflow occurs, otherwise reset
Mode Bits OSCOFF, CPUOFF, and GIE are not affected.
Example The 32-bit counter pointed to by R13 is added to a 32-bit counter, eleven words (20/2 + 2/2) above the pointer in R13.
QUAD      @R13+,20(R13)   ; ADD LSDs with no carry in
ADDC      @R13+,20(R13)   ; ADD MSDs with carry
...       ; resulting from the LSDs
Example The 24-bit counter pointed to by R13 is added to a 24-bit counter, eleven words above the pointer in R13.
ADD.B     @R13+,10(R13)   ; ADD LSDs with no carry in
ADDC.B    @R13+,10(R13)   ; ADD medium Bits with carry
ADDC.B    @R13+,10(R13)   ; ADD MSDs with carry
...       ; resulting from the LSDs
3.4.6.4 AND

**AND[W]**  Source AND destination
**AND.B**  Source AND destination

**Syntax**

\[
\text{AND src,dst or AND.W src,dst} \\
\text{AND.B src,dst}
\]

**Operation**

\[
\text{src .AND. dst} \rightarrow \text{dst}
\]

**Description**

The source operand and the destination operand are logically ANDem. The result is placed into the destination.

**Status Bits**

N: Set if result MSB is set, reset if not set  
Z: Set if result is zero, reset otherwise  
C: Set if result is not zero, reset otherwise ( = .NOT. Zero)  
V: Reset

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

The bits set in R5 are used as a mask (#0AA55h) for the word addressed by TOM. If the result is zero, a branch is taken to label TONI.

\[
\begin{align*}
\text{MOV} & \quad #0AA55h,R5 \quad ; \text{Load mask into register R5} \\
\text{AND} & \quad R5,TOM \quad ; \text{mask word addressed by TOM with R5} \\
\text{JZ} & \quad \text{TONI} \quad ; \quad \text{Result is not zero} \\
\text{......} & \quad ; \\
\text{.....} & \quad ; \quad \text{or} \\
\text{.....} & \quad ; \\
\text{AND} & \quad #0AA55h,TOM \\
\text{JZ} & \quad \text{TONI}
\end{align*}
\]

**Example**

The bits of mask #0A5h are logically ANDem with the low byte TOM. If the result is zero, a branch is taken to label TONI.

\[
\begin{align*}
\text{AND.B} & \quad #0A5h,TOM \quad ; \text{mask Lowbyte TOM with 0A5h} \\
\text{JZ} & \quad \text{TONI} \quad ; \quad \text{Result is not zero} \\
\text{......} & \quad ;
\end{align*}
\]
### 3.4.6.5 BIC

**BIC[.W]**  
Clear bits in destination

**BIC.B**  
Clear bits in destination

**Syntax**  

```
BIC src,dst  or  BIC.W src,dst
BIC.B src,dst
```

**Operation**  
\[ .\overline{\text{NOT}.src} \cdot \text{AND.} \cdot \text{dst} \rightarrow \text{dst} \]

**Description**  
The inverted source operand and the destination operand are logically ANDed. The result is placed into the destination. The source operand is not affected.

**Status Bits**  
Status bits are not affected.

**Mode Bits**  
OSCOFF, CPUOFF, and GIE are not affected.

**Example**  
The six MSBs of the RAM word LEO are cleared.

```
BIC   #0FC00h,LEO  ; Clear 6 MSBs in MEM(LEO)
```

**Example**  
The five MSBs of the RAM byte LEO are cleared.

```
BIC.B #0F8h,LEO    ; Clear 5 MSBs in Ram location LEO
```
### 3.4.6.6 BIS

<table>
<thead>
<tr>
<th>BIS.[.W]</th>
<th>Set bits in destination</th>
</tr>
</thead>
<tbody>
<tr>
<td>BIS.B</td>
<td>Set bits in destination</td>
</tr>
</tbody>
</table>

**Syntax**

- `BIS src,dst` or `BIS.W src,dst`
- `BIS.B src,dst`

**Operation**

src .OR. dst → dst

**Description**

The source operand and the destination operand are logically ORed. The result is placed into the destination. The source operand is not affected.

**Status Bits**

Status bits are not affected.

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

The six LSBs of the RAM word TOM are set.

```
BIS    #003Fh,TOM   ; set the six LSBs in RAM location TOM
```

**Example**

The three MSBs of RAM byte TOM are set.

```
BIS.B  #0E0h,TOM    ; set the 3 MSBs in RAM location TOM
```
3.4.6.7 BIT

BIT.[W]  Test bits in destination
BIT.B   Test bits in destination

Syntax

\[
\text{BIT src,dst or BIT.W src,dst}
\]

Operation

src .AND. dst

Description

The source and destination operands are logically ANDed. The result affects only the status bits. The source and destination operands are not affected.

Status Bits

N: Set if MSB of result is set, reset otherwise
Z: Set if result is zero, reset otherwise
C: Set if result is not zero, reset otherwise (.NOT. Zero)
V: Reset

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

If bit 9 of R8 is set, a branch is taken to label TOM.

\[
\begin{align*}
\text{BIT} & \ #0200h,R8 & \text{; bit 9 of R8 set?} \\
\text{JNZ} & \ \text{TOM} & \text{; Yes, branch to TOM} \\
\text{...} & \ & \text{; No, proceed}
\end{align*}
\]

Example

If bit 3 of R8 is set, a branch is taken to label TOM.

\[
\begin{align*}
\text{BIT.B} & \ #8,R8 \\
\text{JC} & \ \text{TOM}
\end{align*}
\]

Example

A serial communication receive bit (RCV) is tested. Because the carry bit is equal to the state of the tested bit while using the BIT instruction to test a single bit, the carry bit is used by the subsequent instruction; the read information is shifted into register RECBUF.

\[
\begin{align*}
\text{; Serial communication with LSB is shifted first:} \\
& \quad \text{; } xxxx \ xxxx \ xxxx \ xxxx \\
\text{BIT.B} & \ #\text{RCV,RCCTL} & \text{; Bit info into carry} \\
\text{RRC} & \ \text{RECBUF} & \text{; Carry -> MSB of RECBUF} \\
& \quad \text{; } xxxx \ xxxx \\
& \quad \text{; repeat previous two instructions} \\
& \quad \text{; } 8 \text{ times} \\
& \quad \text{; } cccc \ cccc \\
& \quad \text{; } ^ \ ^ \\
& \quad \text{; MSB \ LSB}\n\end{align*}
\]

\[
\begin{align*}
\text{; Serial communication with MSB shifted first:} \\
\text{BIT.B} & \ #\text{RCV,RCCTL} & \text{; Bit info into carry} \\
\text{RLC.B} & \ \text{RECBUF} & \text{; Carry -> LSB of RECBUF} \\
& \quad \text{; } xxxx \ xxxx \\
& \quad \text{; repeat previous two instructions} \\
& \quad \text{; } 8 \text{ times} \\
& \quad \text{; } cccc \ cccc \\
& \quad \text{; } ^ \\
& \quad \text{; MSB \ LSB}
\end{align*}
\]
3.4.6.8 BR, BRANCH

**BR, BRANCH**  
Branch to ......... destination

**Syntax**  
BR dst

**Operation**  
dst → PC

**Emulation**  
MOV dst,PC

**Description**  
An unconditional branch is taken to an address anywhere in the 64K address space. All source addressing modes can be used. The branch instruction is a word instruction.

**Status Bits**  
Status bits are not affected.

**Example**  
Examples for all addressing modes are given.

- BR #EXEC  
  Branch to label EXEC or direct branch (e.g. #0A4h)
  Core instruction: MOV @PC+,PC
- BR EXEC  
  Branch to the address contained in EXEC
  Core instruction: MOV X(FC),PC
  Indirect address
- BR &EXEC  
  Branch to the address contained in absolute address EXEC
  Core instruction: MOV X(0),PC
  Indirect address
- BR R5  
  Branch to the address contained in R5
  Core instruction: MOV R5,PC
  Indirect R5
- BR @R5  
  Branch to the address contained in the word pointed to by R5.
  Core instruction: MOV @R5+,PC
  Indirect, indirect R5
- BR @R5+  
  Branch to the address contained in the word pointed to by R5 and increment pointer in R5 afterwards.
  The next time--S/W flow uses R5 pointer--it can alter program execution due to access to next address in a table pointed to by R5
  Core instruction: MOV @R5,PC
  Indirect, indirect R5 with autoincrement
- BR X(R5)  
  Branch to the address contained in the address pointed to by R5 + X (e.g. table with address starting at X). X can be an address or a label
  Core instruction: MOV X(R5),PC
  Indirect, indirect R5 + X
### 3.4.6.9 CALL

**CALL**

Subroutine

**Syntax**

```
CALL dst
```

**Operation**

- `dst → tmp`  
  - `dst` is evaluated and stored
- `SP - 2 → SP`
- `PC → @SP`  
  - `PC` updated to TOS
- `tmp → PC`  
  - `dst` saved to `PC`

**Description**

A subroutine call is made to an address anywhere in the 64K address space. All addressing modes can be used. The return address (the address of the following instruction) is stored on the stack. The call instruction is a word instruction.

**Status Bits**

Status bits are not affected.

**Example**

Examples for all addressing modes are given.

- **CALL #EXEC**  
  - Call on label EXEC or immediate address (e.g. #0A4h)
  - `SP-2 -> SP, PC+2 -> @SP, @PC+ -> PC`
- **CALL EXEC**  
  - Call on the address contained in EXEC
  - `SP-2 -> SP, PC+2 -> SP, X(PC) -> PC`
  - Indirect address
- **CALL &EXEC**  
  - Call on the address contained in absolute address EXEC
  - `SP-2 -> SP, PC+2 -> @SP, X(0) -> PC`
  - Indirect address
- **CALL R5**  
  - Call on the address contained in R5
  - `SP-2 -> SP, PC+2 -> @SP, R5 -> PC`
  - Indirect R5
- **CALL @R5**  
  - Call on the address contained in the word pointed to by R5
  - `SP-2 -> SP, PC+2 -> @SP, @R5 -> PC`
  - Indirect, indirect R5
- **CALL @R5+**  
  - Call on the address contained in the word pointed to by R5 and increment pointer in R5.
  - The next time S/W flow uses R5 pointer
  - it can alter the program execution due to
  - access to next address in a table pointed to by R5
  - `SP-2 -> SP, PC+2 -> @SP, @R5+ -> PC`
  - Indirect, indirect R5 with autoincrement
- **CALL X(R5)**  
  - Call on the address contained in the address pointed to by R5 + X (e.g. table with address starting at X)
  - X can be an address or a label
  - `SP-2 -> SP, PC+2 -> @SP, X(R5) -> PC`
  - Indirect, indirect R5 + X
### 3.4.6.10 CLR

| **CLR.W** | Clear destination |
| **CLR.B** | Clear destination |

**Syntax**

CLR dst or CLR.W dst  
CLR.B dst

**Operation**

0 → dst

**Emulation**

MOV #0,dst  
MOV.B #0,dst

**Description**
The destination operand is cleared.

**Status Bits**
Status bits are not affected.

**Example**
RAM word TONI is cleared.

CLR    TONI   ; 0 -> TONI

**Example**
Register R5 is cleared.

CLR    R5

**Example**
RAM byte TONI is cleared.

CLR.B  TONI   ; 0 -> TONI
3.4.6.11 CLRC

*CLRC Clear carry bit

Syntax

CLRC

Operation

0 → C

Emulation

BIC #1,SR

Description

The carry bit (C) is cleared. The clear carry instruction is a word instruction.

Status Bits

N: Not affected
Z: Not affected
C: Cleared
V: Not affected

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

The 16-bit decimal counter pointed to by R13 is added to a 32-bit counter pointed to by R12.

CLRC ; C=0: defines start
DADD @R13,0(R12) ; add 16-bit counter to low word of 32-bit counter
DADC 2(R12) ; add carry to high word of 32-bit counter
3.4.6.12 CLRN

*CLRN  Clear negative bit

Syntax

CLRN

Operation 0 → N
or
(.NOT.src .AND. dst → dst)

Emulation

BIC  #4,SR

Description
The constant 04h is inverted (0FFFBh) and is logically ANDed with the destination operand. The result is placed into the destination. The clear negative bit instruction is a word instruction.

Status Bits

N: Reset to 0
Z: Not affected
C: Not affected
V: Not affected

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
The Negative bit in the status register is cleared. This avoids special treatment with negative numbers of the subroutine called.

CLRN
CALL SUBR
......
......
SUBR JN SUBRET ; If input is negative: do nothing and return
......
......
SUBRET RET
3.4.6.13 CLRZ

*CLRZ

Syntax

CLRZ

Operation

0 → Z
or
(.NOT.src .AND. dst → dst)

Emulation

BIC #2,SR

Description

The constant 02h is inverted (0FFFDh) and logically ANDed with the destination operand. The result is placed into the destination. The clear zero bit instruction is a word instruction.

Status Bits

N: Not affected
Z: Reset to 0
C: Not affected
V: Not affected

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

The zero bit in the status register is cleared.

CLRZ
3.4.6.14 CMP

**CMP[.W]**

Compare source and destination

**CMP.B**

Compare source and destination

**Syntax**

```
CMP src,dst  or  CMP.W src,dst
CMP.B src,dst
```

**Operation**

```
dst + .NOT.src + 1
```

or

```
(dst - src)
```

**Description**

The source operand is subtracted from the destination operand. This is accomplished by adding the 1s complement of the source operand plus 1. The two operands are not affected and the result is not stored; only the status bits are affected.

**Status Bits**

- **N**: Set if result is negative (src > dst), reset if positive (src ≤ dst)
- **Z**: Set if result is zero (src = dst), reset otherwise (src ≠ dst)
- **C**: Set if there is a carry from the MSB of the result, reset otherwise
- **V**: Set if an arithmetic overflow occurs, otherwise reset

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

R5 and R6 are compared. If they are equal, the program continues at the label EQUAL.

```
CMP   R5,R6   ; R5 = R6?
JEQ   EQUAL   ; YES, JUMP
```

**Example**

Two RAM blocks are compared. If they are not equal, the program branches to the label ERROR.

```
MOV   #NUM,R5      ; number of words to be compared
MOV   #BLOCK1,R6   ; BLOCK1 start address in R6
MOV   #BLOCK2,R7   ; BLOCK2 start address in R7
L$1  CMP   @R6+,0(R7)   ; Are Words equal? R6 increments
     JNZ   ERROR        ; No, branch to ERROR
     INC   R7           ; Increment R7 pointer
     DEC   R5           ; Are all words compared?
     JNZ   L$1          ; No, another compare
```

**Example**

The RAM bytes addressed by EDE and TONI are compared. If they are equal, the program continues at the label EQUAL.

```
CMP.B   EDE,TONI   ; MEM(EDE) = MEM(TONI)?
JEQ     EQUAL     ; YES, JUMP
```
3.4.6.15 *DADC

| *DADC.W | Add carry decimally to destination |
| *DADC.B | Add carry decimally to destination |

**Syntax**

DADC dst or DADC.W src,dst  
DADC.B dst

**Operation**

dst + C → dst (decimally)

**Emulation**

DADD #0,dst  
DADD.B #0,dst

**Description**

The carry bit (C) is added decimally to the destination.

**Status Bits**

N: Set if MSB is 1  
Z: Set if dst is 0, reset otherwise  
C: Set if destination increments from 9999 to 0000, reset otherwise  
C: Set if destination increments from 99 to 00, reset otherwise  
V: Undefined

**Mode Bits**

OSCOFF, CPUOFF, and GiE are not affected.

**Example**

The four-digit decimal number contained in R5 is added to an eight-digit decimal number pointed to by R8.

```
CLRC  ; Reset carry
; next instruction's start condition is defined
DADD R5,0(R8)  ; Add LSDs + C
DADC 2(R8)  ; Add carry to MSD
```

**Example**

The two-digit decimal number contained in R5 is added to a four-digit decimal number pointed to by R8.

```
CLRC  ; Reset carry
; next instruction's start condition is defined
DADD.B R5,0(R8)  ; Add LSDs + C
DADC.B 1(R8)  ; Add carry to MSDs
```
3.4.6.16 DADD

**DADD.W**  
Source and carry added decimally to destination

**DADD.B**  
Source and carry added decimally to destination

**Syntax**  
DADD src,dst  or  DADD.W src,dst  
DADD.B src,dst

**Operation**  
src + dst + C → dst (decimally)

**Description**  
The source operand and the destination operand are treated as four binary coded decimals (BCD) with positive signs. The source operand and the carry bit (C) are added decimally to the destination operand. The source operand is not affected. The previous contents of the destination are lost. The result is not defined for non-BCD numbers.

**Status Bits**
- **N**: Set if the MSB is 1, reset otherwise
- **Z**: Set if result is zero, reset otherwise
- **C**: Set if the result is greater than 9999
  - Set if the result is greater than 99
- **V**: Undefined

**Mode Bits**  
OSCOFF, CPUOFF, and GiE are not affected.

**Example**  
The eight-digit BCD number contained in R5 and R6 is added decimally to an eight-digit BCD number contained in R3 and R4 (R6 and R4 contain the MSDs).

```
CLRC            ; clear carry
DADD    R5,R3    ; add LSDs
DADD    R6,R4    ; add MSDs with carry
JC      OVERFLOW ; If carry occurs go to error handling routine
```

**Example**  
The two-digit decimal counter in the RAM byte CNT is incremented by one.

```
CLRC            ; clear carry
DADD.B  #1,CNT
```

or

```
SETC
DADD.B  #0,CNT  ; equivalent to DADC.B CNT
```
3.4.6.17 DEC

*DEC[W]  Decrement destination
*DEC.B   Decrement destination

Syntax
DEC dst  or  DEC.W dst
DEC.B dst

Operation
dst - 1 → dst

Emulation
SUB #1,dst
SUB.B #1,dst

Description
The destination operand is decremented by one. The original contents are lost.

Status Bits
N: Set if result is negative, reset if positive
Z: Set if dst contained 1, reset otherwise
C: Reset if dst contained 0, set otherwise
V: Set if an arithmetic overflow occurs, otherwise reset.
   Set if initial value of destination was 08000h, otherwise reset.
   Set if initial value of destination was 080h, otherwise reset.

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
R10 is decremented by 1.

    DEC     R10    ; Decrement R10
    MOV     #EDE,R6
    MOV     #255,R10
    L$1  MOV.B   @R6+,TONI-EDE-1(R6)
    DEC     R10
    JNZ     L$1

Do not transfer tables using the routine above with the overlap shown in Figure 3-13.
Figure 3-13. Decrement Overlap
### 3.4.6.18 DECD

| **DECD[W]** | Double-decrement destination |
| **DECD.B** | Double-decrement destination |

#### Syntax

```
DECD dst  or  DECD.W dst  
DECD.B dst
```

#### Operation

```
dst - 2 → dst
```

#### Emulation

```
SUB #2, dst
```

#### Emulation

```
SUB.B #2, dst
```

#### Description

The destination operand is decremented by two. The original contents are lost.

#### Status Bits

- **N**: Set if result is negative, reset if positive
- **Z**: Set if dst contained 2, reset otherwise
- **C**: Reset if dst contained 0 or 1, set otherwise
- **V**: Set if an arithmetic overflow occurs, otherwise reset.
  - Set if initial value of destination was 08001 or 08000h, otherwise reset.
  - Set if initial value of destination was 081 or 080h, otherwise reset.

#### Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

#### Example

```
R10 is decremented by 2.
```

```
DECD R10  ; Decrement R10 by two
          ; Move a block of 255 words from memory location starting with EDE to
          ; memory location starting with TONI
          ; Tables should not overlap: start of destination address TONI must not be
          ; within the range EDE to EDE+0FEh
MOV #EDE,R6
MOV #510,R10
L$1 MOV @(R6+),TONI-EDE-2(R6)
DECD R10
JNZ L$1
```

#### Example

```
Memory at location LEO is decremented by two.
```

```
DECD.B LEO  ; Decrement MEM(LEO)
```

Decrement status byte STATUS by two.

```
DECD.B STATUS
```
3.4.6.19 DINT

*DINT Disable (general) interrupts

Syntax

DINT

Operation

0 → GIE

or

(0FFF7h .AND. SR → SR / .NOT.src .AND. dst → dst)

Emulation

BIC #8,SR

Description

All interrupts are disabled.

The constant 08h is inverted and logically ANDed with the status register (SR). The result is placed into the SR.

Status Bits

Status bits are not affected.

Mode Bits

GIE is reset. OSCOFF and CPUOFF are not affected.

Example

The general interrupt enable (GIE) bit in the status register is cleared to allow a nondisrupted move of a 32-bit counter. This ensures that the counter is not modified during the move by any interrupt.

DINT ; All interrupt events using the GIE bit are disabled
NOP
MOV  COUNTHI,R5 ; Copy counter
MOV  COUNTLO,R6
EINT ; All interrupt events using the GIE bit are enabled

Note

Disable Interrupt

If any code sequence needs to be protected from interruption, the DINT should be executed at least one instruction before the beginning of the uninterruptible sequence, or should be followed by a NOP instruction.
### 3.4.6.20 EINT

**EINT**  
Enable (general) interrupts

**Syntax**

```
EINT
```

**Operation**

\[ 1 \rightarrow \text{GIE} \]

or

\[(0008\text{h .OR. SR} \rightarrow \text{SR} / .src .OR. dst \rightarrow dst)\]

**Emulation**

```
BIS \#8, SR
```

**Description**

All interrupts are enabled.

The constant \#08h and the status register SR are logically ORed. The result is placed into the SR.

**Status Bits**

Status bits are not affected.

**Mode Bits**

GIE is set. OSCOFF and CPUOFF are not affected.

**Example**

The general interrupt enable (GIE) bit in the status register is set.

```
; Interrupt routine of ports P1.2 to P1.7
; P1IN is the address of the register where all port bits are read. P1IFG is
; the address of the register where all interrupt events are latched.
PUSH.B &P1IN
BIC.B &SP,&P1IFG ; Reset only accepted flags
EINT              ; Preset port 1 interrupt flags stored on stack
; other interrupts are allowed
BIT    #Mask,@SP
JEQ     MaskOK       ; Flags are present identically to mask: jump
......
MaskOK   BIC    #Mask,@SP
......
INCD     SP           ; Housekeeping; inverse to PUSH instruction
; at the start of interrupt subroutine. Corrects
; the stack pointer.
RETI
```

**Note**

**Enable Interrupt**

The instruction following the enable interrupt instruction (EINT) is always executed, even if an interrupt service request is pending when the interrupts are enable.
3.4.6.21 INC

| **INC[W]** | Increment destination |
| **INC.B**  | Increment destination |

**Syntax**
- INC dst or INC.W dst
- INC.B dst

**Operation**
- dst + 1 → dst

**Emulation**
- ADD #1, dst

**Description**
The destination operand is incremented by one. The original contents are lost.

**Status Bits**
- N: Set if result is negative, reset if positive
- Z: Set if dst contained 0xFFFFh, reset otherwise
  - Set if dst contained 0xFFh, reset otherwise
- C: Set if dst contained 0xFFFFh, reset otherwise
  - Set if dst contained 0xFFh, reset otherwise
- V: Set if dst contained 0x7FFFh, reset otherwise
  - Set if dst contained 0x7Fh, reset otherwise

**Mode Bits**
OSCOFF, CPUOFF, and GIE are not affected.

**Example**
The status byte, STATUS, of a process is incremented. When it is equal to 11, a branch to OVFL is taken.

```
INC.B  STATUS
CMP.B  #11,STATUS
JEQ    OVFL
```
3.4.6.22 INCD

*INCD[.W]  Double-increment destination

*INCD.B  Double-increment destination

Syntax

INCD dst  or  INCD.W dst
INCD.B dst

Operation

dst + 2 → dst

Emulation

ADD #2,dst
ADD.B #2,dst

Example

The destination operand is incremented by two. The original contents are lost.

Status Bits

N: Set if result is negative, reset if positive
Z: Set if dst contained 0FFFEh, reset otherwise
Set if dst contained 0FEh, reset otherwise
C: Set if dst contained 0FFFEh or 0FFFFh, reset otherwise
Set if dst contained 0FEh or 0FFh, reset otherwise
V: Set if dst contained 07FFEh or 07FFFh, reset otherwise
Set if dst contained 07Eh or 07Fh, reset otherwise

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

The item on the top of the stack (TOS) is removed without using a register.

PUSH     R5     ; R5 is the result of a calculation, which is stored
          ; in the system stack
INCD     SP     ; Remove TOS by double-increment from stack
          ; Do not use INCD.B, SP is a word-aligned register
RET

Example

The byte on the top of the stack is incremented by two.

INCD.B  0(SP)  ; Byte on TOS is increment by two
3.4.6.23 INV

*INV[.W] Invert destination

*INV.B Invert destination

Syntax

INV dst
INV.B dst

Operation

.NOT.dst → dst

Emulation

XOR #0FFFFh, dst
XOR.B #0FFh, dst

Description

The destination operand is inverted. The original contents are lost.

Status Bits

N: Set if result is negative, reset if positive
Z: Set if dst contained 0FFFFh, reset otherwise
   Set if dst contained 0FFh, reset otherwise
C: Set if result is not zero, reset otherwise ( = .NOT. Zero)
   Set if result is not zero, reset otherwise ( = .NOT. Zero)
V: Set if initial destination operand was negative, otherwise reset

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

Content of R5 is negated (twos complement).

MOV    #00AEh, R5  ;                      R5 = 000AEh
INV    R5         ; Invert R5,           R5 = 0FF51h
INC    R5         ; R5 is now negated,   R5 = 0FF52h

Example

Content of memory byte LEO is negated.

MOV.B  #0AEh, LEO  ;                      MEM(LEO) = 0AEh
INV.B  LEO        ; Invert LEO,          MEM(LEO) = 051h
INC.B  LEO        ; MEM(LEO) is negated, MEM(LEO) = 052h
3.4.6.24 JC, JHS

<p>| | |</p>
<table>
<thead>
<tr>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>JC</td>
<td>Jump if carry set</td>
</tr>
<tr>
<td>JHS</td>
<td>Jump if higher or same</td>
</tr>
</tbody>
</table>

**Syntax**

```plaintext
JC label
JHS label
```

**Operation**

- **If C = 1:** \( PC + 2 \text{ offset} \rightarrow PC \)
- **If C = 0:** execute following instruction

**Description**

The status register carry bit (C) is tested. If it is set, the 10-bit signed offset contained in the instruction LSBs is added to the program counter. If C is reset, the next instruction following the jump is executed. JC (jump if carry/higher or same) is used for the comparison of unsigned numbers (0 to 65536).

**Status Bits**

- Status bits are not affected.

**Example**

The P1IN.1 signal is used to define or control the program flow.

```plaintext
BIT.B  #02h,&P1IN ; State of signal -> Carry
JC      PROGA      ; If carry=1 then execute program routine A
......   ; Carry=0, execute program here
```

**Example**

R5 is compared to 15. If the content is higher or the same, branch to LABEL.

```plaintext
CMP     #15,R5
JHS     LABEL      ; Jump is taken if R5 \( \geq \) 15
......   ; Continue here if R5 < 15
```
3.4.6.25 JEQ, JZ

JEQ, JZ     Jump if equal, jump if zero

Syntax
   JEQ label
   JZ label

Operation
   If Z = 1: PC + 2 offset → PC
   If Z = 0: execute following instruction

Description
   The status register zero bit (Z) is tested. If it is set, the 10-bit signed offset contained in the
   instruction LSBs is added to the program counter. If Z is not set, the instruction following
   the jump is executed.

Status Bits
   Status bits are not affected.

Example
   Jump to address TONI if R7 contains zero.
   TST   R7 ; Test R7
   JZ    TONI ; if zero: JUMP

Example
   Jump to address LEO if R6 is equal to the table contents.
   CMP   R6, Table(R5) ; Compare content of R6 with content of
                        ; MEM (table address + content of R5)
   JEQ   LEO ; Jump if both data are equal
       ...... ; No, data are not equal, continue here

Example
   Branch to LABEL if R5 is 0.
   TST   R5
   JZ    LABEL
       ......
### 3.4.6.26 JGE

**JGE**

Jump if greater or equal

**Syntax**

```
JGE label
```

**Operation**

- If \((N \text{ XOR } V) = 0\) then jump to label: \(\text{PC} + 2 \text{ P offset} \rightarrow \text{PC}\)
- If \((N \text{ XOR } V) = 1\) then execute the following instruction

**Description**

The status register negative bit (N) and overflow bit (V) are tested. If both N and V are set or reset, the 10-bit signed offset contained in the instruction LSBs is added to the program counter. If only one is set, the instruction following the jump is executed.

This allows comparison of signed integers.

**Status Bits**

Status bits are not affected.

**Example**

When the content of R6 is greater or equal to the memory pointed to by R7, the program continues at label EDE.

```
CMP @R7, R6 ; R6 >= (R7)?, compare on signed numbers
JGE EDE     ; Yes, R6 >= (R7)
......       ; No, proceed
......
```
3.4.6.27 JL

**JL** Jump if less

**Syntax**

\[ \text{JL label} \]

**Operation**

If \((N \text{ XOR } V) = 1\) then jump to label: \(\text{PC} + 2\) offset \(\rightarrow \text{PC}\)

If \((N \text{ XOR } V) = 0\) then execute following instruction

**Description**

The status register negative bit \((N)\) and overflow bit \((V)\) are tested. If only one is set, the 10-bit signed offset contained in the instruction LSBs is added to the program counter. If both \(N\) and \(V\) are set or reset, the instruction following the jump is executed.

This allows comparison of signed integers.

**Status Bits**

Status bits are not affected.

**Example**

When the content of R6 is less than the memory pointed to by R7, the program continues at label EDE.

```
CMP   @R7,R6   ; R6 < (R7)?, compare on signed numbers
JL    EDE      ; Yes, R6 < (R7)
......         ; No, proceed
......
......
```
### 3.4.6.28 JMP

<table>
<thead>
<tr>
<th><strong>JMP</strong></th>
<th>Jump unconditionally</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Syntax</strong></td>
<td><code>JMP label</code></td>
</tr>
<tr>
<td><strong>Operation</strong></td>
<td><code>PC + 2 × offset → PC</code></td>
</tr>
<tr>
<td><strong>Description</strong></td>
<td>The 10-bit signed offset contained in the instruction LSBs is added to the program counter.</td>
</tr>
<tr>
<td><strong>Status Bits</strong></td>
<td>Status bits are not affected.</td>
</tr>
<tr>
<td><strong>Hint</strong></td>
<td>This one-word instruction replaces the BRANCH instruction in the range of –511 to +512 words relative to the current program counter.</td>
</tr>
</tbody>
</table>
3.4.6.29 JN

**JN**

Jump if negative

**Syntax**

JN label

**Operation**

if \( N = 1 \): \( PC + 2 \times \text{offset} \rightarrow PC \)

if \( N = 0 \): execute following instruction

**Description**

The negative bit (N) of the status register is tested. If it is set, the 10-bit signed offset contained in the instruction LSBs is added to the program counter. If N is reset, the next instruction following the jump is executed.

**Status Bits**

Status bits are not affected.

**Example**

The result of a computation in R5 is to be subtracted from COUNT. If the result is negative, COUNT is to be cleared and the program continues execution in another path.

```assembly
SUB R5, COUNT ; COUNT - R5 -> COUNT
JN L$1 ; If negative continue with COUNT=0 at PC=L$1
...... ; Continue with COUNT>0
......
......
L$1 CLR COUNT
......
......
```

### 3.4.6.30 JNC, JLO

<table>
<thead>
<tr>
<th></th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>JNC</td>
<td>Jump if carry not set</td>
</tr>
<tr>
<td>JLO</td>
<td>Jump if lower</td>
</tr>
</tbody>
</table>

#### Syntax

- JNC label
- JLO label

#### Operation

- If $C = 0$: \( \text{PC} + 2 \text{ offset} \rightarrow \text{PC} \)
- If $C = 1$: execute following instruction

#### Description

The status register carry bit (C) is tested. If it is reset, the 10-bit signed offset contained in the instruction LSBs is added to the program counter. If C is set, the next instruction following the jump is executed. JNC (jump if no carry/lower) is used for the comparison of unsigned numbers (0 to 65536).

#### Status Bits

Status bits are not affected.

#### Example

The result in R6 is added in BUFFER. If an overflow occurs, an error handling routine at address ERROR is used.

```
ADD     R6,BUFFER   ; BUFFER + R6 -> BUFFER
JNC     CONT        ; No carry, jump to CONT
ERROR   ......              ; Error handler start
......
......
CONT    ......              ; Continue with normal program flow
......
```

#### Example

Branch to STL2 if byte STATUS contains 1 or 0.

```
CMP.B   #2,STATUS
JLO     STL 2       ; STATUS < 2
......
......
```

```
### 3.4.6.31 JNE, JNZ

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>JNE</td>
<td>Jump if not equal</td>
</tr>
<tr>
<td>JNZ</td>
<td>Jump if not zero</td>
</tr>
</tbody>
</table>

#### Syntax

- **JNE label**
- **JNZ label**

#### Operation

- If $Z = 0$: $PC + 2$ a offset → $PC$
- If $Z = 1$: execute following instruction

#### Description

The status register zero bit (Z) is tested. If it is reset, the 10-bit signed offset contained in the instruction LSBs is added to the program counter. If Z is set, the next instruction following the jump is executed.

#### Status Bits

Status bits are not affected.

#### Example

Jump to address TONI if R7 and R8 have different contents.

```assembly
CMP R7, R8 ; COMPARE R7 WITH R8
JNE TONI  ; if different: jump
......    ; if equal, continue
```
3.4.6.32 MOV

MOV[.W]  Move source to destination
MOV.B   Move source to destination
Syntax   MOV src,dst  or  MOV.W src,dst
         MOV.B src,dst
Operation src → dst
Description The source operand is moved to the destination.
The source operand is not affected. The previous contents of the destination are lost.
Status Bits Status bits are not affected.
Mode Bits  OSCOFF, CPUOFF, and GIE are not affected.
Example  The contents of table EDE (word data) are copied to table TOM. The length of the tables
          must be 020h locations.
          MOV   #EDE,R10                ; Prepare pointer
          MOV   #020h,R9                ; Prepare counter
          Loop MOV   @R10+,TOM-EDE-2(R10)    ; Use pointer in R10 for both tables
                DEC   R9                      ; Decrement counter
                JNZ   Loop                    ; Counter not 0, continue copying
          ......                        ; Copying completed

Example  The contents of table EDE (byte data) are copied to table TOM. The length of the tables
          should be 020h locations.
          MOV     #EDE,R10              ; Prepare pointer
          MOV     #020h,R9              ; Prepare counter
          Loop   MOV.B   @R10+,TOM-EDE-1(R10)  ; Use pointer in R10 for
          ; both tables
                DEC   R9                      ; Decrement counter
                JNZ   Loop                    ; Counter not 0, continue
                ......                        ; Copying completed
3.4.6.33 NOP

*NOP

No operation

Syntax

NOP

Operation

None

Emulation

MOV #0, R3

Description

No operation is performed. The instruction may be used for the elimination of instructions during the software check or for defined waiting times.

Status Bits

Status bits are not affected.

The NOP instruction is mainly used for two purposes:

• To fill one, two, or three memory words
• To adjust software timing

Note

Emulating No-Operation Instruction

Other instructions can emulate the NOP function while providing different numbers of instruction cycles and code words. Some examples are:

- MOV #0,R3 ; 1 cycle, 1 word
- MOV 0(R4),0(R4) ; 6 cycles, 3 words
- MOV 0,R4,0(R4) ; 5 cycles, 2 words
- BIC #0,EDE(R4) ; 4 cycles, 2 words
- JMP $+2 ; 2 cycles, 1 word
- BIC #0,R5 ; 1 cycle, 1 word

However, care should be taken when using these examples to prevent unintended results. For example, if MOV 0(R4), 0(R4) is used and the value in R4 is 120h, then a security violation occurs with the watchdog timer (address 120h), because the security key was not used.
3.4.6.34 POP

*POP[W]    Pop word from stack to destination
*POP.B     Pop byte from stack to destination
Syntax

    POP dst
    POP.B dst

Operation

    @SP → temp
    SP + 2 → SP
    temp → dst

Emulation

    MOV @SP+, dst or MOV.W @SP+, dst
    MOV.B @SP+, dst

Description

The stack location pointed to by the stack pointer (TOS) is moved to the destination. The
stack pointer is incremented by two afterwards.

Status Bits

Status bits are not affected.

Example

The contents of R7 and the status register are restored from the stack.

    POP     R7 ; Restore R7
    POP     SR ; Restore status register

Example

The contents of RAM byte LEO is restored from the stack.

    POP.B   LEO ; The low byte of the stack is moved to LEO.

Example

The contents of R7 is restored from the stack.

    POP.B   R7 ; The low byte of the stack is moved to R7,
                ; the high byte of R7 is 00h

Example

The contents of the memory pointed to by R7 and the status register are restored from the
stack.

    POP.B   0(R7) ; The low byte of the stack is moved to the
                    ; the byte which is pointed to by R7
                    ; Example: R7 = 203h
                    ; Mem(R7) = low byte of system stack
                    ; Example: R7 = 20Ah
                    ; Mem(R7) = low byte of system stack
    POP     SR ; Last word on stack moved to the SR
(continued)

Note

The System Stack Pointer

The system stack pointer (SP) is always incremented by two, independent of the byte suffix.
3.4.6.35 PUSH

PUSH[.W]  Push word onto stack
PUSH.B    Push byte onto stack
Syntax

PUSH src  or  PUSH.W src
PUSH.B src

Operation

SP - 2 → SP
src → @SP

Description
The stack pointer is decremented by two, then the source operand is moved to the RAM word addressed by the stack pointer (TOS).

Status Bits
Status bits are not affected.

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
The contents of the status register and R8 are saved on the stack.

PUSH     SR      ; save status register
PUSH     R8      ; save R8

Example
The contents of the peripheral TCDAT is saved on the stack.

PUSH.B   &TCDAT  ; save data from 8-bit peripheral module, address TCDAT, onto stack

Note

System Stack Pointer
The System stack pointer (SP) is always decremented by two, independent of the byte suffix.
### 3.4.6.36 RET

<table>
<thead>
<tr>
<th><strong>RET</strong></th>
<th>Return from subroutine</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Syntax</strong></td>
<td><code>RET</code></td>
</tr>
</tbody>
</table>
| **Operation** | `@SP → PC`  
  `SP + 2 → SP` |
| **Emulation** | `MOV @SP+, PC` |
| **Description** | The return address pushed onto the stack by a CALL instruction is moved to the program counter. The program continues at the code address following the subroutine call. |
| **Status Bits** | Status bits are not affected. |
### 3.4.6.37 RETI

**RETI**

Return from interrupt

**Syntax**

```
RETI
```

**Operation**

- TOS → SR
- SP + 2 → SP
- TOS → PC
- SP + 2 → SP

**Description**

The status register is restored to the value at the beginning of the interrupt service routine by replacing the present SR contents with the TOS contents. The stack pointer (SP) is incremented by two.

The program counter is restored to the value at the beginning of interrupt service. This is the consecutive step after the interrupted program flow. Restoration is performed by replacing the present PC contents with the TOS memory contents. The stack pointer (SP) is incremented.

**Status Bits**

- N: Restored from system stack
- Z: Restored from system stack
- C: Restored from system stack
- V: Restored from system stack

**Mode Bits**

OSCOFF, CPUOFF, and GIE are restored from system stack.

**Example**

Figure 3-14 illustrates the main program interrupt.
Figure 3-14. Main Program Interrupt
3.4.6.38 RLA

*RLA.W
*RLA.B

Syntax

RLA dst or RLA.W dst
RLA.B dst

Operation

C <- MSB <- MSB-1 .... LSB+1 <- LSB <- 0

Emulation

ADD dst,dst
ADD.B dst,dst

Description

The destination operand is shifted left one position as shown in Figure 3-15. The MSB is shifted into the carry bit (C) and the LSB is filled with 0. The RLA instruction acts as a signed multiplication by 2.

An overflow occurs if dst ≥ 04000h and dst < 0C000h before operation is performed: the result has changed sign.

![Figure 3-15. Destination Operand – Arithmetic Shift Left](image-url)

An overflow occurs if dst ≥ 040h and dst < 0C0h before the operation is performed: the result has changed sign.

Status Bits

N: Set if result is negative, reset if positive
Z: Set if result is zero, reset otherwise
C: Loaded from the MSB
V: Set if an arithmetic overflow occurs:
the initial value is 04000h ≤ dst < 0C000h; reset otherwise
Set if an arithmetic overflow occurs:
the initial value is 040h ≤ dst < 0C0h; reset otherwise

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

R7 is multiplied by 2.

```
RLA R7 ; Shift left R7 (x 2)
```

Example

The low byte of R7 is multiplied by 4.

```
RLA.B R7 ; Shift left low byte of R7 (x 2)
RLA.B R7 ; Shift left low byte of R7 (x 4)
```
Note

RLA Substitution

The assembler does not recognize the instruction:

```
RLA @R5+, RLA.B @R5+, or RLA(.B) @R5
```

It must be substituted by:

```
ADD @R5+,-2(R5), ADD.B @R5+,-1(R5), or ADD(.B) @R5
```
3.4.6.39 RLC

*RLC*.W  Rotate left through carry
*RLC*.B  Rotate left through carry

**Syntax**

RLC dst or RLC.W dst
RLC.B dst

**Operation**

C <- MSB <- MSB-1 .... LSB+1 <- LSB <- C

**Emulation**

ADDC dst,dst

**Description**

The destination operand is shifted left one position as shown in Figure 3-16. The carry bit (C) is shifted into the LSB and the MSB is shifted into the carry bit (C).

![Figure 3-16. Destination Operand - Carry Left Shift](image)

**Status Bits**

N: Set if result is negative, reset if positive
Z: Set if result is zero, reset otherwise
C: Loaded from the MSB
V: Set if an arithmetic overflow occurs
   - the initial value is 04000h ≤ dst < 0C000h; reset otherwise
   - Set if an arithmetic overflow occurs:
     - the initial value is 040h ≤ dst < 0C0h; reset otherwise

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

R5 is shifted left one position.

```
RLC     R5         ; (R5 x 2) + C -> R5
```

**Example**

The input P1IN.1 information is shifted into the LSB of R5.

```
BIT.B   #2,&P1IN   ; Information -> Carry
RLC     R5         ; Carry=P0in.1 -> LSB of R5
```

**Example**

The MEM(LEO) content is shifted left one position.

```
RLC.B   LEO        ; Mem(LEO) x 2 + C -> Mem(LEO)
```
Note

RLC and RLC.B Substitution

The assembler does not recognize the instruction:

\[
\text{RLC} \ @R5+, \ \text{RLC} \ @R5, \ \text{or} \ \text{RLC(.B)} \ @R5
\]

It must be substituted by:

\[
\text{ADDC} \ @R5+, -2(R5), \ \text{ADDC.B} \ @R5+, -1(R5), \ \text{or} \ \text{ADDC(.B)} \ @R5
\]
3.4.6.40 RRA

RRA[.W]  Rotate right arithmetically
RRA.B   Rotate right arithmetically

Syntax  

\[
\text{RRA dst} \quad \text{or} \quad \text{RRA.W dst} \\
\text{RRA.B dst}
\]

Operation  

MSB → MSB, MSB → MSB-1, ... LSB+1 → LSB, LSB → C

Description  
The destination operand is shifted right one position as shown in Figure 3-17. The MSB is shifted into the MSB, the MSB is shifted into the MSB-1, and the LSB+1 is shifted into the LSB.

![Figure 3-17. Destination Operand – Arithmetic Right Shift](image)

Status Bits  

N: Set if result is negative, reset if positive
Z: Set if result is zero, reset otherwise
C: Loaded from the LSB
V: Reset

Mode Bits  

OSCOFF, CPUOFF, and GIE are not affected.

Example  

R5 is shifted right one position. The MSB retains the old value. It operates equal to an arithmetic division by 2.

```
RRA R5 ; R5/2 -> R5
; The value in R5 is multiplied by 0.75 (0.5 + 0.25).
PUSH R5 ; Hold R5 temporarily using stack
RRA R5 ; R5 x 0.5 -> R5
ADD @SP+,R5 ; R5 x 0.5 + R5 = 1.5 x R5 -> R5
RRA R5 ; (1.5 x R5) x 0.5 = 0.75 x R5 -> R5
......
```

Example  

The low byte of R5 is shifted right one position. The MSB retains the old value. It operates equal to an arithmetic division by 2.

```
RRA.B R5 ; R5/2 -> R5: operation is on low byte only
PUSH.B R5 ; R5 x 0.5 -> TOS
RRA.B @SP ; TOS x 0.5 = 0.5 x R5 x 0.5 = 0.25 x R5 -> TOS
ADD.B @SP+,R5 ; R5 x 0.5 + R5 x 0.25 = 0.75 x R5 -> R5
......
```
3.4.6.41 RRC

RRC[W]  Rotate right through carry
RRC.B   Rotate right through carry
Syntax
\[ RRC \text{ dst} \text{ or } RRC.W \text{ dst} \]

Operation
\[ C \rightarrow \text{MSB} \rightarrow \text{MSB-1} \ldots \text{LSB+1} \rightarrow \text{LSB} \rightarrow C \]

Description
The destination operand is shifted right one position as shown in Figure 3-18. The carry bit (C) is shifted into the MSB, the LSB is shifted into the carry bit (C).

Status Bits
N: Set if result is negative, reset if positive
Z: Set if result is zero, reset otherwise
C: Loaded from the LSB
V: Reset

Mode Bits
OSCOFF, CPUOFF, and GiE are not affected.

Example
R5 is shifted right one position. The MSB is loaded with 1.

\[
\begin{align*}
\text{SETC} & \quad ; \text{Prepare carry for MSB} \\
\text{RRC} & \quad \text{R5} \quad ; \text{R5/2 + 8000h} \rightarrow \text{R5}
\end{align*}
\]

Example
R5 is shifted right one position. The MSB is loaded with 1.

\[
\begin{align*}
\text{SETC} & \quad ; \text{Prepare carry for MSB} \\
\text{RRC.B} & \quad \text{R5} \quad ; \text{R5/2 + 80h} \rightarrow \text{R5}; \text{low byte of R5 is used}
\end{align*}
\]
3.4.6.42 SBC

*SBC[.W]
Subtract source and borrow/.NOT. carry from destination

*SBC.B
Subtract source and borrow/.NOT. carry from destination

Syntax

\[
\begin{align*}
\text{SBC \ dst \ or \ SBC.W \ dst} \\
\text{SBC.B \ dst}
\end{align*}
\]

Operation

\[
\begin{align*}
\text{dst + 0FFFFh + C} & \to \text{dst} \\
\text{dst + 0FFh + C} & \to \text{dst}
\end{align*}
\]

Emulation

\[
\begin{align*}
\text{SUBC \ #0,dst} \\
\text{SUBC.B \ #0,dst}
\end{align*}
\]

Description
The carry bit (C) is added to the destination operand minus one. The previous contents of the destination are lost.

Status Bits

N: Set if result is negative, reset if positive
Z: Set if result is zero, reset otherwise
C: Set if there is a carry from the MSB of the result, reset otherwise.
    Set to 1 if no borrow, reset if borrow.
V: Set if an arithmetic overflow occurs, reset otherwise.

Mode Bits
OSCOFF, CPUOFF,and GIE are not affected.

Example
The 16-bit counter pointed to by R13 is subtracted from a 32-bit counter pointed to by R12.

\[
\begin{align*}
\text{SUB \ \%R13,0(R12)} & \ ; \text{Subtract LSDs} \\
\text{SBC \ 2(R12)} & \ ; \text{Subtract carry from MSD}
\end{align*}
\]

Example
The 8-bit counter pointed to by R13 is subtracted from a 16-bit counter pointed to by R12.

\[
\begin{align*}
\text{SUB.B \ \%R13,0(R12)} & \ ; \text{Subtract LSDs} \\
\text{SBC.B \ 1(R12)} & \ ; \text{Subtract carry from MSD}
\end{align*}
\]

Note

Borrow Implementation

<table>
<thead>
<tr>
<th>Borrow</th>
<th>Carry bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Yes</td>
<td>0</td>
</tr>
<tr>
<td>No</td>
<td>1</td>
</tr>
</tbody>
</table>
3.4.6.43 SETC

**SETC**
Set carry bit

**Syntax**

```
SETC
```

**Operation**

\[ 1 \rightarrow C \]

**Emulation**

```
BIS #1,SR
```

**Description**
The carry bit (C) is set.

**Status Bits**

- N: Not affected
- Z: Not affected
- C: Set
- V: Not affected

**Mode Bits**
OSCOFF, CPUOFF, and GIE are not affected.

**Example**
Emulation of the decimal subtraction:
Subtract R5 from R6 decimally
Assume that \( R5 = 03987h \) and \( R6 = 04137h \)

```
DSUB ADD #06666h,R5 ; Move content R5 from 0-9 to 6-0Fh
; R5 = 03987h + 06666h = 09FEDh
INV R5 ; Invert this (result back to 0-9)
; R5 = \( \text{NOT} \). R5 = 06012h
SETC ; Prepare carry = 1
DADD R5,R6 ; Emulate subtraction by addition of:
; (010000h - R5 - 1)
; R6 = R6 + R5 + 1
; R6 = 0150h
```
### 3.4.6.44 SETN

**SETN**
Set negative bit

**Syntax**

```
SETN
```

**Operation**

1 → N

**Emulation**

```
BIS #4,SR
```

**Description**
The negative bit (N) is set.

**Status Bits**

- N: Set
- Z: Not affected
- C: Not affected
- V: Not affected

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.
3.4.6.45 SETZ

*SETZ

Syntax

```
SETZ
```

Operation

1 → Z

Emulation

```
BIS #2,SR
```

Description

The zero bit (Z) is set.

Status Bits

- N: Not affected
- Z: Set
- C: Not affected
- V: Not affected

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.
### 3.4.6.46 SUB

#### SUB[W]
Subtract source from destination

#### SUB.B
Subtract source from destination

#### Syntax
- `SUB src,dst` or `SUB.W src,dst`
- `SUB.B src,dst`

#### Operation
- `dst + .NOT.src + 1 → dst`
- `[(dst - src → dst)]`

#### Description
The source operand is subtracted from the destination operand by adding the source operand's 1s complement and the constant 1. The source operand is not affected. The previous contents of the destination are lost.

#### Status Bits
- **N**: Set if result is negative, reset if positive
- **Z**: Set if result is zero, reset otherwise
- **C**: Set if there is a carry from the MSB of the result, reset otherwise.
  - Set to 1 if no borrow, reset if borrow.
- **V**: Set if an arithmetic overflow occurs, otherwise reset

#### Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

#### Example
See example at the SBC instruction.

#### Note
**Borrow Is Treated as a .NOT.**

<table>
<thead>
<tr>
<th>Borrow</th>
<th>Carry bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Yes</td>
<td>0</td>
</tr>
<tr>
<td>No</td>
<td>1</td>
</tr>
</tbody>
</table>
3.4.6.47 SUBC, SBB

**SUBC.[W], SBB.[W]** Subtract source and borrow/.NOT. carry from destination

**SUBC.B, SBB.B** Subtract source and borrow/.NOT. carry from destination

**Syntax**

```
SUBC   src,dst   or   SUBC.W  src,dst   or
SBB    src,dst   or   SBB.W   src,dst
SUBC.B src,dst   or   SBB.B   src,dst
```

**Operation**

\[ \text{dst} + \text{.NOT.src} + \text{C} \rightarrow \text{dst} \]

or

\[ (\text{dst} - \text{src} - 1 + \text{C} \rightarrow \text{dst}) \]

**Description**

The source operand is subtracted from the destination operand by adding the source operand's 1s complement and the carry bit (C). The source operand is not affected. The previous contents of the destination are lost.

**Status Bits**

- N: Set if result is negative, reset if positive.
- Z: Set if result is zero, reset otherwise.
- C: Set if there is a carry from the MSB of the result, reset otherwise.
  
  Set to 1 if no borrow, reset if borrow.
- V: Set if an arithmetic overflow occurs, reset otherwise.

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

Two floating point mantissas (24 bits) are subtracted.

LSBs are in R13 and R10, MSBs are in R12 and R9.

```
SUB.W  R13,R10 ; 16-bit part, LSBs
SUBC.B R12,R9  ; 8-bit part, MSBs
```

**Example**

The 16-bit counter pointed to by R13 is subtracted from a 16-bit counter in R10 and R11(MSD).

```
SUB.B  @R13+,R10 ; Subtract LSDs without carry
SUBC.B @R13,R11 ; Subtract MSDs with carry
...   ; resulting from the LSDs
```

---

**Note**

**Borrow Implementation**

<table>
<thead>
<tr>
<th>Borrow</th>
<th>Carry bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Yes</td>
<td>0</td>
</tr>
<tr>
<td>No</td>
<td>1</td>
</tr>
</tbody>
</table>

---

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3.4.6.48 SWPB

**SWPB**

Swap bytes

**Syntax**

```
SWPB dst
```

**Operation**

Bits 15 to 8 ↔ bits 7 to 0

**Description**

The destination operand high and low bytes are exchanged as shown in Figure 3-19.

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

![Figure 3-19. Destination Operand - Byte Swap](image)

**Example**

```
MOV    #040BFh, R7 ; 0100000010111111 -> R7
SWPB   R7           ; 1011111101000000 in R7
```

**Example**

The value in R5 is multiplied by 256. The result is stored in R5,R4.

```
SWPB   R5
MOV    R5, R4       ; Copy the swapped value to R4
BIC    #0FF00h, R5 ; Correct the result
BIC    #00FFh, R4  ; Correct the result
```
### 3.4.6.49 SXT

#### SXT
Extend Sign

#### Syntax
```
SXT dst
```

#### Operation
Bit 7 → Bit 8 ........ Bit 15

#### Description
The sign of the low byte is extended into the high byte as shown in Figure 3-20.

#### Status Bits
- **N**: Set if result is negative, reset if positive
- **Z**: Set if result is zero, reset otherwise (.NOT. Zero)
- **C**: Set if result is not zero, reset otherwise (.NOT. Zero)
- **V**: Reset

#### Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.


**Figure 3-20. Destination Operand - Sign Extension**

#### Example
R7 is loaded with the P1IN value. The operation of the sign-extend instruction expands bit 8 to bit 15 with the value of bit 7.

R7 is then added to R6.

```
MOV.B &P1IN,R7 ; P1IN = 080h: .... .... 1000 0000
SXT R7         ; R7 = 0FF80h: 1111 1111 1000 0000
```
### 3.4.6.50 TST

<table>
<thead>
<tr>
<th>Syntax</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>TST dst or TST.W dst TST.B dst</td>
<td>The destination operand is compared with zero. The status bits are set according to the result. The destination is not affected.</td>
</tr>
</tbody>
</table>

**Operation**
- dst + 0FFFFh + 1
- dst + 0FFh + 1

**Emulation**
- CMP #0,dst
- CMP.B #0,dst

**Status Bits**
- **N**: Set if destination is negative, reset if positive
- **Z**: Set if destination contains zero, reset otherwise
- **C**: Set
- **V**: Reset

**Mode Bits**
- OSCOFF, CPUOFF, and GIE are not affected.

**Example**
1. **R7 is tested. If it is negative, continue at R7NEG; if it is positive but not zero, continue at R7POS.**
   ```assembly
tst r7    ; Test R7
jn r7neg  ; R7 is negative
jz r7zero ; R7 is zero
r7pos     ...... ; R7 is positive but not zero
r7neg     ...... ; R7 is negative
r7zero    ...... ; R7 is zero
```

2. **The low byte of R7 is tested. If it is negative, continue at R7NEG; if it is positive but not zero, continue at R7POS.**
   ```assembly
tst.b r7  ; Test low byte of R7
jn r7neg  ; Low byte of R7 is negative
jz r7zero ; Low byte of R7 is zero
r7pos     ...... ; Low byte of R7 is positive but not zero
r7neg     ...... ; Low byte of R7 is negative
r7zero    ...... ; Low byte of R7 is zero
```
### 3.4.6.51 XOR

**XOR.W**
Exclusive OR of source with destination

**XOR.B**
Exclusive OR of source with destination

**Syntax**

\[
\begin{align*}
\text{XOR} &\quad \text{src, dst} \quad \text{or} \quad \text{XOR.W} \quad \text{src, dst} \\
\text{XOR.B} &\quad \text{src, dst}
\end{align*}
\]

**Operation**

\[
\text{src .XOR. dst} \rightarrow \text{dst}
\]

**Description**
The source and destination operands are exclusive ORed. The result is placed into the destination. The source operand is not affected.

**Status Bits**
- **N**: Set if result MSB is set, reset if not set
- **Z**: Set if result is zero, reset otherwise
- **C**: Set if result is not zero, reset otherwise ( = .NOT. Zero)
- **V**: Set if both operands are negative

**Mode Bits**
OSCOFF, CPUOFF, and GIE are not affected.

**Example**
The bits set in R6 toggle the bits in the RAM word TONI.

XOR R6, TONI  ; Toggle bits of word TONI on the bits set in R6

**Example**
The bits set in R6 toggle the bits in the RAM byte TONI.

XOR.B R6, TONI  ; Toggle bits of byte TONI on the bits set in low byte of R6

**Example**
Reset to 0 those bits in low byte of R7 that are different from bits in RAM byte EDE.

XOR.B EDE, R7  ; Set different bit to "1s"
INV.B R7       ; Invert Lowbyte, Highbyte is 0h
This chapter describes the extended MSP430X 16-bit RISC CPU with 1-MB memory access, its addressing modes, and instruction set. The MSP430X CPU is implemented in all MSP430 devices that exceed 64-KB of address space.

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4.1 CPU Introduction

The MSP430X CPU incorporates features specifically designed for modern programming techniques such as calculated branching, table processing and the use of high-level languages such as C. The MSP430X CPU can address a 1-MB address range without paging. In addition, the MSP430X CPU has fewer interrupt overhead cycles and fewer instruction cycles in some cases than the MSP430 CPU, while maintaining the same or better code density than the MSP430 CPU. The MSP430X CPU is backward compatible with the MSP430 CPU.

The MSP430X CPU features include:

- RISC architecture
- Orthogonal architecture
- Full register access including program counter, status register and stack pointer
- Single-cycle register operations
- Large register file reduces fetches to memory
- 20-bit address bus allows direct access and branching throughout the entire memory range without paging
- 16-bit data bus allows direct manipulation of word-wide arguments
- Constant generator provides the six most often used immediate values and reduces code size
- Direct memory-to-memory transfers without intermediate register holding
- Byte, word, and 20-bit address-word addressing

The block diagram of the MSP430X CPU is shown in Figure 4-1.
Figure 4-1. MSP430X CPU Block Diagram
4.2 Interrupts

The MSP430X uses the same interrupt structure as the MSP430:

- Vectored interrupts with no polling necessary
- Interrupt vectors are located downward from address 0FFFEh

Interrupt operation for both MSP430 and MSP430X CPUs is described in Chapter 2 System Resets, Interrupts, and Operating modes, Section 2 Interrupts. The interrupt vectors contain 16-bit addresses that point into the lower 64-KB memory. This means all interrupt handlers must start in the lower 64-KB memory, even in MSP430X devices.

During an interrupt, the program counter and the status register are pushed onto the stack as shown in Figure 4-2. The MSP430X architecture efficiently stores the complete 20-bit PC value by automatically appending the PC bits 19:16 to the stored SR value on the stack. When the RETI instruction is executed, the full 20-bit PC is restored making return from interrupt to any address in the memory range possible.

![Figure 4-2. PC Storage on the Stack for Interrupts](image-url)
4.3 CPU Registers

The CPU incorporates 16 registers (R0 through R15). Registers R0, R1, R2, and R3 have dedicated functions. Registers R4 through R15 are working registers for general use.

4.3.1 Program Counter (PC)

The 20-bit PC (PC/R0) points to the next instruction to be executed. Each instruction uses an even number of bytes (2, 4, 6, or 8 bytes), and the PC is incremented accordingly. Instruction accesses are performed on word boundaries, and the PC is aligned to even addresses. Figure 4-3 shows the PC.

![Figure 4-3. Program Counter](image)

The PC can be addressed with all instructions and addressing modes. A few examples:

```
MOV.W #LABEL,PC  ; Branch to address LABEL (lower 64KB)
MOVA   #LABEL,PC  ; Branch to address LABEL (1MB memory)
MOV.W  LABEL,PC   ; Branch to address in word LABEL
                ; (lower 64KB)
MOV.W  @R14,PC    ; Branch indirect to address in
                ; R14 (lower 64KB)
ADDA   #4,PC      ; Skip two words (1MB memory)
```

The BR and CALL instructions reset the upper four PC bits to 0. Only addresses in the lower 64-KB address range can be reached with the BR or CALL instruction. When branching or calling, addresses beyond the lower 64-KB range can only be reached using the BRA or CALLA instructions. Also, any instruction to directly modify the PC does so according to the used addressing mode. For example, MOV.W #value,PC clears the upper four bits of the PC, because it is a .W instruction.

The PC is automatically stored on the stack with CALL (or CALLA) instructions and during an interrupt service routine. Figure 4-4 shows the storage of the PC with the return address after a CALLA instruction. A CALL instruction stores only bits 15:0 of the PC.

![Figure 4-4. PC Storage on the Stack for CALLA](image)

The RETA instruction restores bits 19:0 of the PC and adds 4 to the stack pointer (SP). The RET instruction restores bits 15:0 to the PC and adds 2 to the SP.

4.3.2 Stack Pointer (SP)

The 20-bit SP (SP/R1) is used by the CPU to store the return addresses of subroutine calls and interrupts. It uses a predecrement, postincrement scheme. In addition, the SP can be used by software with all instructions and addressing modes. Figure 4-5 shows the SP. The SP is initialized into RAM by the user, and is always aligned to even addresses.

Figure 4-6 shows the stack usage. Figure 4-7 shows the stack usage when 20-bit address words are pushed.
Figure 4-5. Stack Pointer

Figure 4-6. Stack Usage

Figure 4-7. PUSHX.A Format on the Stack

The special cases of using the SP as an argument to the PUSH and POP instructions are described and shown in Figure 4-8.

Figure 4-8. PUSH SP, POP SP Sequence

4.3.3 Status Register (SR)

The 16-bit SR (SR/R2), used as a source or destination register, can only be used in register mode addressed with word instructions. The remaining combinations of addressing modes are used to support the constant generator. Figure 4-9 shows the SR bits. Do not write 20-bit values to the SR. Unpredictable operation can result.
Table 4-1 describes the SR bits.

Table 4-1. SR Bit Description

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>V</td>
<td>Overflow. This bit is set when the result of an arithmetic operation overflows the signed-variable range.</td>
</tr>
<tr>
<td>SCG1</td>
<td>System clock generator 1. This bit may be to enable/disable functions in the clock system depending on the device family; for example, DCO bias enable/disable</td>
</tr>
<tr>
<td>SCG0</td>
<td>System clock generator 0. This bit may be used to enable/disable functions in the clock system depending on the device family; for example, FLL disable/enable</td>
</tr>
<tr>
<td>OSCOFF</td>
<td>Oscillator off. This bit, when set, turns off the LFXT1 crystal oscillator when LFXT1CLK is not used for MCLK or SMCLK.</td>
</tr>
<tr>
<td>CPUOFF</td>
<td>CPU off. This bit, when set, turns off the CPU.</td>
</tr>
<tr>
<td>GIE</td>
<td>General interrupt enable. This bit, when set, enables maskable interrupts. When reset, all maskable interrupts are disabled.</td>
</tr>
<tr>
<td>N</td>
<td>Negative. This bit is set when the result of an operation is negative and cleared when the result is positive.</td>
</tr>
<tr>
<td>Z</td>
<td>Zero. This bit is set when the result of an operation is 0 and cleared when the result is not 0.</td>
</tr>
<tr>
<td>C</td>
<td>Carry. This bit is set when the result of an operation produced a carry and cleared when no carry occurred.</td>
</tr>
</tbody>
</table>
4.3.4 Constant Generator Registers (CG1 and CG2)

Six commonly-used constants are generated with the constant generator registers R2 (CG1) and R3 (CG2), without requiring an additional 16-bit word of program code. The constants are selected with the source register addressing modes (As), as described in Table 4-2.

<table>
<thead>
<tr>
<th>Register</th>
<th>As</th>
<th>Constant</th>
<th>Remarks</th>
</tr>
</thead>
<tbody>
<tr>
<td>R2</td>
<td>00</td>
<td>–</td>
<td>Register mode</td>
</tr>
<tr>
<td>R2</td>
<td>01</td>
<td>(0)</td>
<td>Absolute address mode</td>
</tr>
<tr>
<td>R2</td>
<td>10</td>
<td>00004h</td>
<td>+4, bit processing</td>
</tr>
<tr>
<td>R2</td>
<td>11</td>
<td>00008h</td>
<td>+8, bit processing</td>
</tr>
<tr>
<td>R3</td>
<td>00</td>
<td>00000h</td>
<td>0, word processing</td>
</tr>
<tr>
<td>R3</td>
<td>01</td>
<td>00001h</td>
<td>+1</td>
</tr>
<tr>
<td>R3</td>
<td>10</td>
<td>00002h</td>
<td>+2, bit processing</td>
</tr>
<tr>
<td>R3</td>
<td>11</td>
<td>FFh, FFFFh, FFFFFh</td>
<td>–1, word processing</td>
</tr>
</tbody>
</table>

The constant generator advantages are:

- No special instructions required
- No additional code word for the six constants
- No code memory access required to retrieve the constant

The assembler uses the constant generator automatically if one of the six constants is used as an immediate source operand. Registers R2 and R3, used in the constant mode, cannot be addressed explicitly; they act as source-only registers.

4.3.4.1 Constant Generator – Expanded Instruction Set

The RISC instruction set of the MSP430 has only 27 instructions. However, the constant generator allows the MSP430 assembler to support 24 additional emulated instructions. For example, the single-operand instruction:

CLR dst

is emulated by the double-operand instruction with the same length:

MOV R3,dst

where the #0 is replaced by the assembler, and R3 is used with As = 00.

INC dst

is replaced by:

ADD #1,dst
4.3.5 General-Purpose Registers (R4 to R15)

The 12 CPU registers (R4 to R15) contain 8-bit, 16-bit, or 20-bit values. Any byte-write to a CPU register clears bits 19:8. Any word-write to a register clears bits 19:16. The only exception is the SXT instruction. The SXT instruction extends the sign through the complete 20-bit register.

The following figures show the handling of byte, word, and address-word data. Note the reset of the leading most significant bits (MSBs) if a register is the destination of a byte or word instruction.

Figure 4-10 shows byte handling (8-bit data, .B suffix). The handling is shown for a source register and a destination memory byte and for a source memory byte and a destination register.

Figure 4-11 and Figure 4-12 show 16-bit word handling (.W suffix). The handling is shown for a source register and a destination memory word and for a source memory word and a destination register.
Figure 4-12. Word-Register Operation

Figure 4-13 and Figure 4-14 show 20-bit address-word handling (\texttt{.A} suffix). The handling is shown for a source register and a destination memory address-word and for a source memory address-word and a destination register.

Figure 4-13. Register – Address-Word Operation
4.4 Addressing Modes

Seven addressing modes for the source operand and four addressing modes for the destination operand use 16-bit or 20-bit addresses (see Table 4-3). The MSP430 and MSP430X instructions are usable throughout the entire 1MB memory range.

Table 4-3. Source/Destination Addressing

<table>
<thead>
<tr>
<th>As/Ad</th>
<th>Addressing Mode</th>
<th>Syntax</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>00/0</td>
<td>Register</td>
<td>Rn</td>
<td>Register contents are operand.</td>
</tr>
<tr>
<td>01/1</td>
<td>Indexed</td>
<td>X(Rn)</td>
<td>(Rn + X) points to the operand. X is stored in the next word, or stored in combination of the preceding extension word and the next word.</td>
</tr>
<tr>
<td>01/1</td>
<td>Symbolic</td>
<td>ADDR</td>
<td>(PC + X) points to the operand. X is stored in the next word, or stored in combination of the preceding extension word and the next word. Indexed mode X(PC) is used.</td>
</tr>
<tr>
<td>01/1</td>
<td>Absolute</td>
<td>&amp;ADDR</td>
<td>The word following the instruction contains the absolute address. X is stored in the next word, or stored in combination of the preceding extension word and the next word. Indexed mode X(SR) is used.</td>
</tr>
<tr>
<td>10/-</td>
<td>Indirect Register</td>
<td>@Rn</td>
<td>Rn is used as a pointer to the operand.</td>
</tr>
<tr>
<td>11/-</td>
<td>Indirect Autoincrement</td>
<td>@Rn+</td>
<td>Rn is used as a pointer to the operand. Rn is incremented afterwards by 1 for .B instructions, by 2 for .W instructions, and by 4 for .A instructions.</td>
</tr>
<tr>
<td>11/-</td>
<td>Immediate</td>
<td>#N</td>
<td>N is stored in the next word, or stored in combination of the preceding extension word and the next word. Indirect autoincrement mode @PC+ is used.</td>
</tr>
</tbody>
</table>

The seven addressing modes are explained in detail in the following sections. Most of the examples show the same addressing mode for the source and destination, but any valid combination of source and destination addressing modes is possible in an instruction.

Note

Use of Labels EDE, TONI, TOM, and LEO

Throughout MSP430 documentation, EDE, TONI, TOM, and LEO are used as generic labels. They are only labels and have no special meaning.
4.4.1 Register Mode

Operation: The operand is the 8-, 16-, or 20-bit content of the used CPU register.
Length: One, two, or three words
Comment: Valid for source and destination
Byte operation: Byte operation reads only the eight least significant bits (LSBs) of the source register Rsrc and writes the result to the eight LSBs of the destination register Rdst. The bits Rdst.19:8 are cleared. The register Rsrc is not modified.
Word operation: Word operation reads the 16 LSBs of the source register Rsrc and writes the result to the 16 LSBs of the destination register Rdst. The bits Rdst.19:16 are cleared. The register Rsrc is not modified.
Address-word operation: Address-word operation reads the 20 bits of the source register Rsrc and writes the result to the 20 bits of the destination register Rdst. The register Rsrc is not modified.
SXT exception: The SXT instruction is the only exception for register operation. The sign of the low byte in bit 7 is extended to the bits Rdst.19:8.
Example: BIS.W R5,R6;
This instruction logically ORs the 16-bit data contained in R5 with the 16-bit contents of R6. R6.19:16 is cleared.

Before: | Address | Register | After: | Address | Register
--- | --- | --- | --- | --- | ---
| 21036h | xxxxh | R5 | 21036h | xxxxh | PC | R5 | AA550h
| 21034h | D506h | R6 | 21034h | D506h | R6 | 0B551h

A550h.or.11111h = B551h

Example: BISX.A R5,R6;
This instruction logically ORs the 20-bit data contained in R5 with the 20-bit contents of R6. The extension word contains the A/L bit for 20-bit data. The instruction word uses byte mode with bits A/L:B/W = 01. The result of the instruction is:

Before: | Address | Register | After: | Address | Register
--- | --- | --- | --- | --- | ---
| 21036h | xxxxh | R5 | 21036h | xxxxh | PC | R5 | AA550h
| 21034h | D546h | R6 | 21034h | D546h | R6 | BB551h
| 21032h | 1800h | PC | 21032h | 1800h | |

A550h.or.11111h = BB551h
4.4.2 Indexed Mode

The Indexed mode calculates the address of the operand by adding the signed index to a CPU register. The Indexed mode has three addressing possibilities:

- Indexed mode in lower 64-KB memory
- MSP430 instruction with Indexed mode addressing memory above the lower 64-KB memory
- MSP430X instruction with Indexed mode

4.4.2.1 Indexed Mode in Lower 64-KB Memory

If the CPU register Rn points to an address in the lower 64KB of the memory range, the calculated memory address bits 19:16 are cleared after the addition of the CPU register Rn and the signed 16-bit index. This means the calculated memory address is always located in the lower 64KB and does not overflow or underflow out of the lower 64-KB memory space. The RAM and the peripheral registers can be accessed this way and existing MSP430 software is usable without modifications as shown in Figure 4-15.

![Figure 4-15. Indexed Mode in Lower 64KB](image)

**Figure 4-15. Indexed Mode in Lower 64KB**

Length: Two or three words

Operation: The signed 16-bit index is located in the next word after the instruction and is added to the CPU register Rn. The resulting bits 19:16 are cleared giving a truncated 16-bit memory address, which points to an operand address in the range 00000h to 0FFFFh. The operand is the content of the addressed memory location.

Comment: Valid for source and destination. The assembler calculates the register index and inserts it.

Example: `ADD.B 1000h(R5), 0F000h(R6)`;

This instruction adds the 8-bit data contained in source byte 1000h(R5) and the destination byte 0F000h(R6) and places the result into the destination byte. Source and destination bytes are both located in the lower 64KB due to the cleared bits 19:16 of registers R5 and R6.

Source: The byte pointed to by R5 + 1000h results in address 0479Ch + 1000h = 0579Ch after truncation to a 16-bit address.

Destination: The byte pointed to by R6 + F000h results in address 01778h + F000h = 00778h after truncation to a 16-bit address.
### 4.4.2.2 MSP430 Instruction With Indexed Mode in Upper Memory

If the CPU register Rn points to an address above the lower 64-KB memory, the Rn bits 19:16 are used for the address calculation of the operand. The operand may be located in memory in the range Rn ±32KB, because the index, X, is a signed 16-bit value. In this case, the address of the operand can overflow or underflow into the lower 64-KB memory space (see Figure 4-16 and Figure 4-17).

![Figure 4-16. Indexed Mode in Upper Memory](image-url)
Length: Two or three words

Operation: The sign-extended 16-bit index in the next word after the instruction is added to the 20 bits of the CPU register Rn. This delivers a 20-bit address, which points to an address in the range 0 to FFFFFh. The operand is the content of the addressed memory location.

Comment: Valid for source and destination. The assembler calculates the register index and inserts it.

Example:

```assembly
ADD.W 8346h(R5),2100h(R6) ;
```

This instruction adds the 16-bit data contained in the source and the destination addresses and places the 16-bit result into the destination. Source and destination operand can be located in the entire address range.

Source: The word pointed to by R5 + 8346h. The negative index 8346h is sign extended, which results in address 23456h + F8346h = 1B79Ch.

Destination: The word pointed to by R6 + 2100h results in address 15678h + 2100h = 17778h.

Before:

<table>
<thead>
<tr>
<th>Address Space</th>
<th>Register</th>
<th>After: Address Space</th>
<th>Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>1103Ah</td>
<td>xxxxh</td>
<td>23456h</td>
<td>1103Ah</td>
</tr>
<tr>
<td>11038h</td>
<td>2100h</td>
<td>15678h</td>
<td>11038h</td>
</tr>
<tr>
<td>11036h</td>
<td>8346h</td>
<td>xxxxh</td>
<td>11036h</td>
</tr>
<tr>
<td>11034h</td>
<td>5596h</td>
<td>PC</td>
<td>11034h</td>
</tr>
</tbody>
</table>

This instruction adds the 16-bit data contained in the source and the destination addresses and places the 16-bit result into the destination. Source and destination operand can be located in the entire address range.

Source: The word pointed to by R5 + 8346h. The negative index 8346h is sign extended, which results in address 23456h + F8346h = 1B79Ch.

Destination: The word pointed to by R6 + 2100h results in address 15678h + 2100h = 17778h.

Before:

<table>
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<th>Address Space</th>
<th>Register</th>
<th>After: Address Space</th>
<th>Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>1103Ah</td>
<td>xxxxh</td>
<td>23456h</td>
<td>1103Ah</td>
</tr>
<tr>
<td>11038h</td>
<td>2100h</td>
<td>15678h</td>
<td>11038h</td>
</tr>
<tr>
<td>11036h</td>
<td>8346h</td>
<td>xxxxh</td>
<td>11036h</td>
</tr>
<tr>
<td>11034h</td>
<td>5596h</td>
<td>PC</td>
<td>11034h</td>
</tr>
</tbody>
</table>

```assembly
ADD.W 8346h(R5),2100h(R6) ;
```

This instruction adds the 16-bit data contained in the source and the destination addresses and places the 16-bit result into the destination. Source and destination operand can be located in the entire address range.

Source: The word pointed to by R5 + 8346h. The negative index 8346h is sign extended, which results in address 23456h + F8346h = 1B79Ch.

Destination: The word pointed to by R6 + 2100h results in address 15678h + 2100h = 17778h.

Before:

<table>
<thead>
<tr>
<th>Address Space</th>
<th>Register</th>
<th>After: Address Space</th>
<th>Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>1103Ah</td>
<td>xxxxh</td>
<td>23456h</td>
<td>1103Ah</td>
</tr>
<tr>
<td>11038h</td>
<td>2100h</td>
<td>15678h</td>
<td>11038h</td>
</tr>
<tr>
<td>11036h</td>
<td>8346h</td>
<td>xxxxh</td>
<td>11036h</td>
</tr>
<tr>
<td>11034h</td>
<td>5596h</td>
<td>PC</td>
<td>11034h</td>
</tr>
</tbody>
</table>

```assembly
ADD.W 8346h(R5),2100h(R6) ;
```

This instruction adds the 16-bit data contained in the source and the destination addresses and places the 16-bit result into the destination. Source and destination operand can be located in the entire address range.

Source: The word pointed to by R5 + 8346h. The negative index 8346h is sign extended, which results in address 23456h + F8346h = 1B79Ch.

Destination: The word pointed to by R6 + 2100h results in address 15678h + 2100h = 17778h.
4.4.2.3 MSP430X Instruction With Indexed Mode

When using an MSP430X instruction with Indexed mode, the operand can be located anywhere in the range of Rn + 19 bits.

Length: Three or four words

Operation: The operand address is the sum of the 20-bit CPU register content and the 20-bit index. The 4 MSBs of the index are contained in the extension word; the 16 LSBs are contained in the word following the instruction. The CPU register is not modified.

Comment: Valid for source and destination. The assembler calculates the register index and inserts it.

Example: ADDX.A 12346h(R5),32100h(R6) ;

This instruction adds the 20-bit data contained in the source and the destination addresses and places the result into the destination.

Source: Two words pointed to by R5 + 12346h which results in address 23456h + 12346h = 3579Ch.

Destination: Two words pointed to by R6 + 32100h which results in address 45678h + 32100h = 77778h.

The extension word contains the MSBs of the source index and of the destination index and the A/L bit for 20-bit data. The instruction word uses byte mode due to the 20-bit data length with bits A/L:B/W = 01.

<table>
<thead>
<tr>
<th>Before:</th>
<th>Address</th>
<th>Register</th>
<th>After:</th>
<th>Address</th>
<th>Register</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>2103Ah</td>
<td>xxxh</td>
<td>R5</td>
<td>2103Ah</td>
<td>xxxh</td>
<td>PC</td>
</tr>
<tr>
<td>21038h</td>
<td>2100h</td>
<td>R6</td>
<td>21038h</td>
<td>2100h</td>
<td>R6</td>
</tr>
<tr>
<td>21036h</td>
<td>2346h</td>
<td></td>
<td>21036h</td>
<td>2346h</td>
<td></td>
</tr>
<tr>
<td>21034h</td>
<td>55D6h</td>
<td></td>
<td>21034h</td>
<td>55D6h</td>
<td></td>
</tr>
<tr>
<td>21032h</td>
<td>1883h</td>
<td>PC</td>
<td>21032h</td>
<td>1883h</td>
<td></td>
</tr>
<tr>
<td>7777Ah</td>
<td>0001h</td>
<td></td>
<td>7777Ah</td>
<td>0007h</td>
<td>65432h</td>
</tr>
<tr>
<td>77778h</td>
<td>2345h</td>
<td></td>
<td>77778h</td>
<td>7777h</td>
<td>77777h</td>
</tr>
<tr>
<td>3579Eh</td>
<td>0006h</td>
<td></td>
<td>3579Eh</td>
<td>0006h</td>
<td></td>
</tr>
<tr>
<td>3579Ch</td>
<td>5432h</td>
<td></td>
<td>3579Ch</td>
<td>5432h</td>
<td></td>
</tr>
</tbody>
</table>

4.4.3 Symbolic Mode

The Symbolic mode calculates the address of the operand by adding the signed index to the PC. The Symbolic mode has three addressing possibilities:

- Symbolic mode in lower 64-KB memory
- MSP430 instruction with Symbolic mode addressing memory above the lower 64-KB memory.
- MSP430X instruction with Symbolic mode
4.4.3.1 Symbolic Mode in Lower 64KB

If the PC points to an address in the lower 64KB of the memory range, the calculated memory address bits 19:16 are cleared after the addition of the PC and the signed 16-bit index. This means the calculated memory address is always located in the lower 64KB and does not overflow or underflow out of the lower 64-KB memory space. The RAM and the peripheral registers can be accessed this way and existing MSP430 software is usable without modifications as shown in Figure 4-18.

![Figure 4-18. Symbolic Mode Running in Lower 64KB](image_url)

Operation: The signed 16-bit index in the next word after the instruction is added temporarily to the PC. The resulting bits 19:16 are cleared giving a truncated 16-bit memory address, which points to an operand address in the range 00000h to 0FFFFh. The operand is the content of the addressed memory location.

Length: Two or three words

Comment: Valid for source and destination. The assembler calculates the PC index and inserts it.

Example: ADD.B EDE, TONI ;

This instruction adds the 8-bit data contained in source byte EDE and destination byte TONI and places the result into the destination byte TONI. Bytes EDE and TONI and the program are located in the lower 64KB.

Source: Byte EDE located at address 0579Ch, pointed to by PC + 4766h, where the PC index 4766h is the result of 0579Ch – 01036h = 04766h. Address 01036h is the location of the index for this example.

Destination: Byte TONI located at address 00778h, pointed to by PC + F740h, is the truncated 16-bit result of 00778h – 1038h = FF740h. Address 01038h is the location of the index for this example.
4.4.3.2 MSP430 Instruction With Symbolic Mode in Upper Memory

If the PC points to an address above the lower 64-KB memory, the PC bits 19:16 are used for the address calculation of the operand. The operand may be located in memory in the range PC ± 32KB, because the index, X, is a signed 16-bit value. In this case, the address of the operand can overflow or underflow into the lower 64-KB memory space as shown in Figure 4-19 and Figure 4-20.

![Figure 4-19. Symbolic Mode Running in Upper Memory](image-url)
**Figure 4-20. Overflow and Underflow for Symbolic Mode**

Length: Two or three words

Operation: The sign-extended 16-bit index in the next word after the instruction is added to the 20 bits of the PC. This delivers a 20-bit address, which points to an address in the range 0 to FFFFFh. The operand is the content of the addressed memory location.

Comment: Valid for source and destination. The assembler calculates the PC index and inserts it

Example: `ADD.W EDE, &TONI ;`

This instruction adds the 16-bit data contained in source word EDE and destination word TONI and places the 16-bit result into the destination word TONI. For this example, the instruction is located at address 2F034h.

Source: Word EDE at address 3379Ch, pointed to by PC + 4766h, which is the 16-bit result of 3379Ch – 2F036h = 04766h. Address 2F036h is the location of the index for this example.

Destination: Word TONI located at address 00778h pointed to by the absolute address 00778h

<table>
<thead>
<tr>
<th>Before: Address</th>
<th>After: Address</th>
</tr>
</thead>
<tbody>
<tr>
<td>2F03Ah xxxxh</td>
<td>2F03Ah xxxxh</td>
</tr>
<tr>
<td>2F038h 0778h</td>
<td>2F038h 0778h</td>
</tr>
<tr>
<td>2F036h 4766h</td>
<td>2F036h 4766h</td>
</tr>
<tr>
<td>2F034h 5092h PC</td>
<td>2F034h 5092h</td>
</tr>
<tr>
<td>3379Eh xxxxh</td>
<td>3379Eh xxxxh</td>
</tr>
<tr>
<td>3379Ch 5432h</td>
<td>3379Ch 5432h</td>
</tr>
<tr>
<td>0077Ah xxxxh</td>
<td>0077Ah xxxxh</td>
</tr>
<tr>
<td>00778h 2345h</td>
<td>00778h 7777h</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Address Space</th>
<th>Address Space</th>
</tr>
</thead>
<tbody>
<tr>
<td>2F03Ah xxxxh</td>
<td>2F03Ah xxxxh</td>
</tr>
<tr>
<td>2F038h 0778h</td>
<td>2F038h 0778h</td>
</tr>
<tr>
<td>2F036h 4766h</td>
<td>2F036h 4766h</td>
</tr>
<tr>
<td>2F034h 5092h</td>
<td>2F034h 5092h</td>
</tr>
<tr>
<td>3379Eh xxxxh</td>
<td>3379Eh xxxxh</td>
</tr>
<tr>
<td>3379Ch 5432h</td>
<td>3379Ch 5432h</td>
</tr>
<tr>
<td>0077Ah xxxxh</td>
<td>0077Ah xxxxh</td>
</tr>
<tr>
<td>00778h 2345h</td>
<td>00778h 7777h</td>
</tr>
</tbody>
</table>

src | dst | Sum
4.4.3.3 MSP430X Instruction With Symbolic Mode

When using an MSP430X instruction with Symbolic mode, the operand can be located anywhere in the range of PC + 19 bits.

Length: Three or four words
Operation: The operand address is the sum of the 20-bit PC and the 20-bit index. The 4 MSBs of the index are contained in the extension word; the 16 LSBs are contained in the word following the instruction.
Comment: Valid for source and destination. The assembler calculates the register index and inserts it.
Example: ADDX.B EDE,TONI ;
This instruction adds the 8-bit data contained in source byte EDE and destination byte TONI and places the result into the destination byte TONI.
Source: Byte EDE located at address 3579Ch, pointed to by PC + 14766h, is the 20-bit result of 3579Ch – 21036h = 14766h. Address 21036h is the address of the index in this example.
Destination: Byte TONI located at address 77778h, pointed to by PC + 56740h, is the 20-bit result of 77778h – 21038h = 56740h. Address 21038h is the address of the index in this example.

<table>
<thead>
<tr>
<th>Before: Address Space</th>
<th>After: Address Space</th>
</tr>
</thead>
<tbody>
<tr>
<td>2103Ah XXXXh</td>
<td>2103Ah XXXXh PC</td>
</tr>
<tr>
<td>21038h 6740h</td>
<td>21038h 6740h</td>
</tr>
<tr>
<td>21036h 4766h</td>
<td>21036h 4766h</td>
</tr>
<tr>
<td>21034h 50D0h</td>
<td>21034h 50D0h</td>
</tr>
<tr>
<td>21032h 18C5h PC</td>
<td>21032h 18C5h</td>
</tr>
<tr>
<td>7777Ah XXXXh</td>
<td>21038h +56740h 7777Ah XXXXh 32h src</td>
</tr>
<tr>
<td>77778h XX45h</td>
<td>77778h XX77h 77h sum</td>
</tr>
<tr>
<td>3579Eh XXXXh</td>
<td>21036h +14766h 3579Eh XXXXh</td>
</tr>
<tr>
<td>3579Ch XX32h</td>
<td>3579Ch XX32h</td>
</tr>
</tbody>
</table>

4.4.4 Absolute Mode

The Absolute mode uses the contents of the word following the instruction as the address of the operand. The Absolute mode has two addressing possibilities:

- Absolute mode in lower 64-KB memory
- MSP430X instruction with Absolute mode

4.4.4.1 Absolute Mode in Lower 64KB

If an MSP430 instruction is used with Absolute addressing mode, the absolute address is a 16-bit value and, therefore, points to an address in the lower 64KB of the memory range. The address is calculated as an index from 0 and is stored in the word following the instruction. The RAM and the peripheral registers can be accessed this way and existing MSP430 software is usable without modifications.

Length: Two or three words
Operation: The operand is the content of the addressed memory location.
Comment: Valid for source and destination. The assembler calculates the index from 0 and inserts it.
Example: ADD.W &EDE,&TONI ;
This instruction adds the 16-bit data contained in the absolute source and destination addresses and places the result into the destination.
Source: Word at address EDE
Destination: Word at address TONI

### 4.4.4.2 MSP430X Instruction With Absolute Mode

If an MSP430X instruction is used with Absolute addressing mode, the absolute address is a 20-bit value and, therefore, points to any address in the memory range. The address value is calculated as an index from 0. The 4 MSBs of the index are contained in the extension word, and the 16 LSBs are contained in the word following the instruction.

Length: Three or four words
Operation: The operand is the content of the addressed memory location.
Comment: Valid for source and destination. The assembler calculates the index from 0 and inserts it.
Example: ADDX.A &EDE,&TONI ;
This instruction adds the 20-bit data contained in the absolute source and destination addresses and places the result into the destination.
Source: Two words beginning with address EDE
Destination: Two words beginning with address TONI
### 4.4.5 Indirect Register Mode

The Indirect Register mode uses the contents of the CPU register Rsrc as the source operand. The Indirect Register mode always uses a 20-bit address.

<table>
<thead>
<tr>
<th>Before: Address</th>
<th>After: Address</th>
<th>PC</th>
</tr>
</thead>
<tbody>
<tr>
<td>2103Ah x00h</td>
<td>2103Ah x00h</td>
<td>21038h</td>
</tr>
<tr>
<td>21038h 7778h</td>
<td>21038h 7778h</td>
<td></td>
</tr>
<tr>
<td>21036h 579Ch</td>
<td>21036h 579Ch</td>
<td></td>
</tr>
<tr>
<td>21034h 52D2h</td>
<td>21034h 52D2h</td>
<td></td>
</tr>
<tr>
<td>21032h 1987h</td>
<td>21032h 1987h</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Length: One, two, or three words</th>
</tr>
</thead>
<tbody>
<tr>
<td>Operation: The operand is the content the addressed memory location. The source register Rsrc is not modified.</td>
</tr>
<tr>
<td>Comment: Valid only for the source operand. The substitute for the destination operand is 0(Rdst).</td>
</tr>
<tr>
<td>Example: ADDX.W @R5, 2100h(R6)</td>
</tr>
<tr>
<td>This instruction adds the two 16-bit operands contained in the source and the destination addresses and places the result into the destination.</td>
</tr>
<tr>
<td>Source: Word pointed to by R5. R5 contains address 3579Ch for this example.</td>
</tr>
<tr>
<td>Destination: Word pointed to by R6 + 2100h, which results in address 45678h + 2100h = 7778h</td>
</tr>
</tbody>
</table>
4.4.6 Indirect Autoincrement Mode

The Indirect Autoincrement mode uses the contents of the CPU register Rs as the source operand. Rs is then automatically incremented by 1 for byte instructions, by 2 for word instructions, and by 4 for address-word instructions immediately after accessing the source operand. If the same register is used for source and destination, it contains the incremented address for the destination access. Indirect Autoincrement mode always uses 20-bit addresses.

Length: One, two, or three words
Operation: The operand is the content of the addressed memory location.
Comment: Valid only for the source operand
Example: ADD.B @R5+, 0(R6)
This instruction adds the 8-bit data contained in the source and the destination addresses and places the result into the destination.
Source: Byte pointed to by R5. R5 contains address 3579Ch for this example.
Destination: Byte pointed to by R6 + 0h, which results in address 0778h for this example
4.4.7 Immediate Mode

The Immediate mode allows accessing constants as operands by including the constant in the memory location following the instruction. The PC is used with the Indirect Autoincrement mode. The PC points to the immediate value contained in the next word. After the fetching of the immediate operand, the PC is incremented by 2 for byte, word, or address-word instructions. The Immediate mode has two addressing possibilities:

- 8-bit or 16-bit constants with MSP430 instructions
- 20-bit constants with MSP430X instruction

### 4.4.7.1 MSP430 Instructions With Immediate Mode

If an MSP430 instruction is used with Immediate addressing mode, the constant is an 8- or 16-bit value and is stored in the word following the instruction.

**Length:** Two or three words. One word less if a constant of the constant generator can be used for the immediate operand.

**Operation:** The 16-bit immediate source operand is used together with the 16-bit destination operand.

**Comment:** Valid only for the source operand

**Example:**

```
ADD #3456h, &TONI
```

This instruction adds the 16-bit immediate operand 3456h to the data in the destination address TONI.

**Source:** 16-bit immediate value 3456h

**Destination:** Word at address TONI
4.4.7.2 MSP430X Instructions With Immediate Mode

If an MSP430X instruction is used with Immediate addressing mode, the constant is a 20-bit value. The 4 MSBs of the constant are stored in the extension word, and the 16 LSBs of the constant are stored in the word following the instruction.

- **Length:** Three or four words. One word less if a constant of the constant generator can be used for the immediate operand.
- **Operation:** The 20-bit immediate source operand is used together with the 20-bit destination operand.
- **Comment:** Valid only for the source operand
- **Example:**
  ```assembly
  ADDX.A #23456h, &TONI
  ```
  This instruction adds the 20-bit immediate operand 23456h to the data in the destination address TONI.
- **Source:** 20-bit immediate value 23456h
- **Destination:** Two words beginning with address TONI

### Example 1

Before:

<table>
<thead>
<tr>
<th>Address</th>
<th>Space</th>
</tr>
</thead>
<tbody>
<tr>
<td>2103Ah</td>
<td>xxxxh</td>
</tr>
<tr>
<td>21038h</td>
<td>0778h</td>
</tr>
<tr>
<td>21036h</td>
<td>3456h</td>
</tr>
<tr>
<td>21034h</td>
<td>50F2h</td>
</tr>
</tbody>
</table>

After:

<table>
<thead>
<tr>
<th>Address</th>
<th>Space</th>
<th>PC</th>
</tr>
</thead>
<tbody>
<tr>
<td>2103Ah</td>
<td>xxxxh</td>
<td>PC</td>
</tr>
<tr>
<td>21038h</td>
<td>0778h</td>
<td></td>
</tr>
<tr>
<td>21036h</td>
<td>3456h</td>
<td></td>
</tr>
<tr>
<td>21034h</td>
<td>50F2h</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Address</th>
<th>Space</th>
</tr>
</thead>
<tbody>
<tr>
<td>0077Ah</td>
<td>xxxxh</td>
</tr>
<tr>
<td>00778h</td>
<td>2345h</td>
</tr>
</tbody>
</table>

Sum:

<table>
<thead>
<tr>
<th>Source</th>
<th>Destination</th>
</tr>
</thead>
<tbody>
<tr>
<td>+2345h</td>
<td>579Bh Sum</td>
</tr>
</tbody>
</table>

### Example 2

Before:

<table>
<thead>
<tr>
<th>Address</th>
<th>Space</th>
</tr>
</thead>
<tbody>
<tr>
<td>2103Ah</td>
<td>xxxxh</td>
</tr>
<tr>
<td>21038h</td>
<td>7778h</td>
</tr>
<tr>
<td>21036h</td>
<td>3456h</td>
</tr>
<tr>
<td>21034h</td>
<td>50F2h</td>
</tr>
<tr>
<td>21032h</td>
<td>1907h</td>
</tr>
</tbody>
</table>

After:

<table>
<thead>
<tr>
<th>Address</th>
<th>Space</th>
<th>PC</th>
</tr>
</thead>
<tbody>
<tr>
<td>2103Ah</td>
<td>xxxxh</td>
<td>PC</td>
</tr>
<tr>
<td>21038h</td>
<td>7778h</td>
<td></td>
</tr>
<tr>
<td>21036h</td>
<td>3456h</td>
<td></td>
</tr>
<tr>
<td>21034h</td>
<td>50F2h</td>
<td></td>
</tr>
<tr>
<td>21032h</td>
<td>1907h</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Address</th>
<th>Space</th>
</tr>
</thead>
<tbody>
<tr>
<td>7777Ah</td>
<td>0001h</td>
</tr>
<tr>
<td>77778h</td>
<td>2345h</td>
</tr>
</tbody>
</table>

Sum:

<table>
<thead>
<tr>
<th>Source</th>
<th>Destination</th>
</tr>
</thead>
<tbody>
<tr>
<td>+12345h</td>
<td>3579Bh Sum</td>
</tr>
</tbody>
</table>
4.5 MSP430 and MSP430X Instructions

MSP430 instructions are the 27 implemented instructions of the MSP430 CPU. These instructions are used throughout the 1MB memory range unless their 16-bit capability is exceeded. The MSP430X instructions are used when the addressing of the operands, or the data length exceeds the 16-bit capability of the MSP430 instructions.

There are three possibilities when choosing between an MSP430 and MSP430X instruction:

- **To use only the MSP430 instructions** – The only exceptions are the CALLA and the RETA instruction. This can be done if a few, simple rules are met:
  - Placement of all constants, variables, arrays, tables, and data in the lower 64KB. This allows the use of MSP430 instructions with 16-bit addressing for all data accesses. No pointers with 20-bit addresses are needed.
  - Placement of subroutine constants immediately after the subroutine code. This allows the use of the symbolic addressing mode with its 16-bit index to reach addresses within the range of PC + 32KB.
- **To use only MSP430X instructions** – The disadvantages of this method are the reduced speed due to the additional CPU cycles and the increased program space due to the necessary extension word for any double operand instruction.
- **Use the best fitting instruction where needed.**

The following sections list and describe the MSP430 and MSP430X instructions.

### 4.5.1 MSP430 Instructions

The MSP430 instructions can be used, regardless if the program resides in the lower 64KB or beyond it. The only exceptions are the instructions CALL and RET, which are limited to the lower 64-KB address range. CALLA and RETA instructions have been added to the MSP430X CPU to handle subroutines in the entire address range with no code size overhead.

#### 4.5.1.1 MSP430 Double-Operand (Format I) Instructions

Figure 4-21 shows the format of the MSP430 double-operand instructions. Source and destination words are appended for the Indexed, Symbolic, Absolute, and Immediate modes. Table 4-4 lists the 12 MSP430 double-operand instructions.

<table>
<thead>
<tr>
<th>15</th>
<th>12</th>
<th>11</th>
<th>8</th>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Op-code</td>
<td>Rsrc</td>
<td>Ad</td>
<td>B/W</td>
<td>As</td>
<td>Rdst</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Source or Destination 15:0**

**Destination 15:0**

**Figure 4-21. MSP430 Double-Operand Instruction Format**
### Table 4-4. MSP430 Double-Operand Instructions

<table>
<thead>
<tr>
<th>Mnemonic</th>
<th>S-Reg, D-Reg</th>
<th>Operation</th>
<th>Status Bits&lt;sup&gt;(1)&lt;/sup&gt;</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOV(.B)</td>
<td>src,dst</td>
<td>src → dst</td>
<td>- - - -</td>
</tr>
<tr>
<td>ADD(,B)</td>
<td>src,dst</td>
<td>src + dst → dst</td>
<td>* * * *</td>
</tr>
<tr>
<td>ADDC(,B)</td>
<td>src,dst</td>
<td>src + dst + C → dst</td>
<td>* * * *</td>
</tr>
<tr>
<td>SUB(,B)</td>
<td>src,dst</td>
<td>dst + .not.src + 1 → dst</td>
<td>* * * *</td>
</tr>
<tr>
<td>SUBC(,B)</td>
<td>src,dst</td>
<td>dst + .not.src + C → dst</td>
<td>* * * *</td>
</tr>
<tr>
<td>CMP(,B)</td>
<td>src,dst</td>
<td>dst → src</td>
<td>* * * *</td>
</tr>
<tr>
<td>DADD(,B)</td>
<td>src,dst</td>
<td>src + dst + C → dst (decimally)</td>
<td>* * * *</td>
</tr>
<tr>
<td>BIT(,B)</td>
<td>src,dst</td>
<td>src .and. dst</td>
<td>0 * * Z</td>
</tr>
<tr>
<td>BIC(,B)</td>
<td>src,dst</td>
<td>.not.src .and. dst → dst</td>
<td>- - - -</td>
</tr>
<tr>
<td>BIS(,B)</td>
<td>src,dst</td>
<td>src .or. dst → dst</td>
<td>- - - -</td>
</tr>
<tr>
<td>XOR(,B)</td>
<td>src,dst</td>
<td>src .xor. dst → dst</td>
<td>* * * Z</td>
</tr>
<tr>
<td>AND(,B)</td>
<td>src,dst</td>
<td>src .and. dst → dst</td>
<td>0 * * Z</td>
</tr>
</tbody>
</table>

<sup>(1)</sup> * = Status bit is affected.  
- = Status bit is not affected.  
0 = Status bit is cleared.  
1 = Status bit is set.

#### 4.5.1.2 MSP430 Single-Operand (Format II) Instructions

Figure 4-22 shows the format for MSP430 single-operand instructions, except RETI. The destination word is appended for the Indexed, Symbolic, Absolute, and Immediate modes. Table 4-5 lists the seven single-operand instructions.

![Figure 4-22. MSP430 Single-Operand Instructions](https://www.ti.com/assets/main/search/counter/counter.html)

### Table 4-5. MSP430 Single-Operand Instructions

<table>
<thead>
<tr>
<th>Mnemonic</th>
<th>S-Reg, D-Reg</th>
<th>Operation</th>
<th>Status Bits&lt;sup&gt;(1)&lt;/sup&gt;</th>
</tr>
</thead>
<tbody>
<tr>
<td>RRC(.B)</td>
<td>dst</td>
<td>C → MSB →.......LSB → C</td>
<td>* * * *</td>
</tr>
<tr>
<td>RRA(.B)</td>
<td>dst</td>
<td>MSB → MSB →....LSB → C</td>
<td>0 * * *</td>
</tr>
<tr>
<td>PUSH(.B)</td>
<td>src</td>
<td>SP - 2 → SP, src → SP</td>
<td>- - - -</td>
</tr>
<tr>
<td>SWPB</td>
<td>dst</td>
<td>bit 15...bit 8 ↔ bit 7...bit 0</td>
<td>- - - -</td>
</tr>
<tr>
<td>CALL</td>
<td>dst</td>
<td>Call subroutine in lower 64KB</td>
<td>- - - -</td>
</tr>
<tr>
<td>RETI</td>
<td></td>
<td>TOS → SR, SP + 2 → SP</td>
<td>* * * *</td>
</tr>
<tr>
<td>SXT</td>
<td>dst</td>
<td>Register mode: bit 7 → bit 8...bit 19 Other modes: bit 7 → bit 8...bit 15</td>
<td>0 * * Z</td>
</tr>
</tbody>
</table>

<sup>(1)</sup> * = Status bit is affected.  
- = Status bit is not affected.  
0 = Status bit is cleared.  
1 = Status bit is set.
4.5.1.3 Jump Instructions

Figure 4-23 shows the format for MSP430 and MSP430X jump instructions. The signed 10-bit word offset of the jump instruction is multiplied by two, sign-extended to a 20-bit address, and added to the 20-bit PC. This allows jumps in a range of –511 to +512 words relative to the PC in the full 20-bit address space. Jumps do not affect the status bits. Table 4-6 lists and describes the eight jump instructions.

![Figure 4-23. Format of Conditional Jump Instructions](image)

<table>
<thead>
<tr>
<th>Mnemonic</th>
<th>S-Reg, D-Reg</th>
<th>Operation</th>
</tr>
</thead>
<tbody>
<tr>
<td>JEQ/JZ</td>
<td>Label</td>
<td>Jump to label if zero bit is set</td>
</tr>
<tr>
<td>JNE/JNZ</td>
<td>Label</td>
<td>Jump to label if zero bit is reset</td>
</tr>
<tr>
<td>JC</td>
<td>Label</td>
<td>Jump to label if carry bit is set</td>
</tr>
<tr>
<td>JNC</td>
<td>Label</td>
<td>Jump to label if carry bit is reset</td>
</tr>
<tr>
<td>JN</td>
<td>Label</td>
<td>Jump to label if negative bit is set</td>
</tr>
<tr>
<td>JGE</td>
<td>Label</td>
<td>Jump to label if (N .XOR. V) = 0</td>
</tr>
<tr>
<td>JL</td>
<td>Label</td>
<td>Jump to label if (N .XOR. V) = 1</td>
</tr>
<tr>
<td>JMP</td>
<td>Label</td>
<td>Jump to label unconditionally</td>
</tr>
</tbody>
</table>

4.5.1.4 Emulated Instructions

In addition to the MSP430 and MSP430X instructions, emulated instructions are instructions that make code easier to write and read, but do not have op-codes themselves. Instead, they are replaced automatically by the assembler with a core instruction. There is no code or performance penalty for using emulated instructions. The emulated instructions are listed in Table 4-7.

![Table 4-7. Emulated Instructions](image)

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Explanation</th>
<th>Emulation</th>
<th>Status Bits (1)</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADC(.B) dst</td>
<td>Add Carry to dst</td>
<td>ADDC(.B) #0,dst</td>
<td>V   N   Z   C</td>
</tr>
<tr>
<td>BR dst</td>
<td>Branch indirectly dst</td>
<td>MOV dst,PC</td>
<td></td>
</tr>
<tr>
<td>CLR(.B) dst</td>
<td>Clear dst</td>
<td>MOV (.B) #0,dst</td>
<td></td>
</tr>
<tr>
<td>CLRC</td>
<td>Clear Carry bit</td>
<td>BIC #1,SR</td>
<td></td>
</tr>
<tr>
<td>CLRN</td>
<td>Clear Negative bit</td>
<td>BIC #4,SR</td>
<td>0</td>
</tr>
<tr>
<td>CLRZ</td>
<td>Clear Zero bit</td>
<td>BIC #2,SR</td>
<td></td>
</tr>
<tr>
<td>DADC(.B) dst</td>
<td>Add Carry to dst decimally</td>
<td>DADD(.B) #0,dst</td>
<td></td>
</tr>
<tr>
<td>DEC(.B) dst</td>
<td>Decrement dst by 1</td>
<td>SUB(.B) #1,dst</td>
<td></td>
</tr>
<tr>
<td>DECD(.B) dst</td>
<td>Decrement dst by 2</td>
<td>SUB(.B) #2,dst</td>
<td></td>
</tr>
<tr>
<td>DINT</td>
<td>Disable interrupt</td>
<td>BIC #8,SR</td>
<td></td>
</tr>
<tr>
<td>EINT</td>
<td>Enable interrupt</td>
<td>BIS #8,SR</td>
<td></td>
</tr>
<tr>
<td>INC(.B) dst</td>
<td>Increment dst by 1</td>
<td>ADD(.B) #1,dst</td>
<td></td>
</tr>
<tr>
<td>INCN(.B) dst</td>
<td>Increment dst by 2</td>
<td>ADD(.B) #2,dst</td>
<td></td>
</tr>
<tr>
<td>INV(.B) dst</td>
<td>Invert dst</td>
<td>XOR(.B) #I,dst</td>
<td></td>
</tr>
<tr>
<td>NOP</td>
<td>No operation</td>
<td>MOV R3,R3</td>
<td></td>
</tr>
<tr>
<td>POP dst</td>
<td>Pop operand from stack</td>
<td>MOV @SP+,dst</td>
<td></td>
</tr>
<tr>
<td>RET</td>
<td>Return from subroutine</td>
<td>MOV @SP+,PC</td>
<td></td>
</tr>
<tr>
<td>RLA(.B) dst</td>
<td>Shift left dst arithmetically</td>
<td>ADD(.B) dst,dst</td>
<td></td>
</tr>
</tbody>
</table>
### Table 4-7. Emulated Instructions (continued)

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Explanation</th>
<th>Emulation</th>
<th>Status Bits (1)</th>
</tr>
</thead>
<tbody>
<tr>
<td>RLC(.B) dst</td>
<td>Shift left dst logically through Carry</td>
<td>ADDC(.B) dst,dst</td>
<td>* * * *</td>
</tr>
<tr>
<td>SBC(.B) dst</td>
<td>Subtract Carry from dst</td>
<td>SUBC(.B) #0,dst</td>
<td>* * * *</td>
</tr>
<tr>
<td>SETC</td>
<td>Set Carry bit</td>
<td>BIS #1,SR</td>
<td>– – – 1</td>
</tr>
<tr>
<td>SETN</td>
<td>Set Negative bit</td>
<td>BIS #4,SR</td>
<td>– 1 – –</td>
</tr>
<tr>
<td>SETZ</td>
<td>Set Zero bit</td>
<td>BIS #2,SR</td>
<td>– – 1 –</td>
</tr>
<tr>
<td>TST(.B) dst</td>
<td>Test dst (compare with 0)</td>
<td>CMP(.B) #0,dst</td>
<td>0 * * 1</td>
</tr>
</tbody>
</table>

(1) * = Status bit is affected; – = Status bit is not affected; 0 = Status bit is cleared; 1 = Status bit is set.

### 4.5.1.5 MSP430 Instruction Execution

The number of CPU clock cycles required for an instruction depends on the instruction format and the addressing modes used – not the instruction itself. The number of clock cycles refers to MCLK.

### 4.5.1.5.1 Instruction Cycles and Length for Interrupt, Reset, and Subroutines

Table 4-8 lists the length and the CPU cycles for reset, interrupts, and subroutines.

#### Table 4-8. Interrupt, Return, and Reset Cycles and Length

<table>
<thead>
<tr>
<th>Action</th>
<th>Execution Time (MCLK Cycles)</th>
<th>Length of Instruction (Words)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Return from interrupt RETI</td>
<td>3 (1)</td>
<td>1</td>
</tr>
<tr>
<td>Return from subroutine RET</td>
<td>3</td>
<td>1</td>
</tr>
<tr>
<td>Interrupt request service</td>
<td>5 (2)</td>
<td>–</td>
</tr>
<tr>
<td>WDT reset</td>
<td>4</td>
<td>–</td>
</tr>
<tr>
<td>Reset (RST/NMI)</td>
<td>4</td>
<td>–</td>
</tr>
</tbody>
</table>

(1) The cycle count in MSP430 CPU is 5.
(2) The cycle count in MSP430 CPU is 6.

### 4.5.1.5.2 Format II (Single-Operand) Instruction Cycles and Lengths

Table 4-9 lists the length and the CPU cycles for all addressing modes of the MSP430 single-operand instructions.

#### Table 4-9. MSP430 Format II Instruction Cycles and Length

<table>
<thead>
<tr>
<th>Addressing Mode</th>
<th>RRA, RRC, SWPB, SXT</th>
<th>PUSH</th>
<th>CALL</th>
<th>Length of Instruction</th>
<th>Example</th>
</tr>
</thead>
<tbody>
<tr>
<td>Rn</td>
<td>1</td>
<td>3</td>
<td>3 (1)</td>
<td>1</td>
<td>SWPB R5</td>
</tr>
<tr>
<td>@Rn</td>
<td>3</td>
<td>3 (1)</td>
<td>4</td>
<td>1</td>
<td>ARC @R9</td>
</tr>
<tr>
<td>@Rn+</td>
<td>3</td>
<td>3 (1)</td>
<td>4 (2)</td>
<td>1</td>
<td>SWPB @R10+</td>
</tr>
<tr>
<td>#N</td>
<td>N/A</td>
<td>3 (1)</td>
<td>4 (2)</td>
<td>2</td>
<td>CALL #LABEL</td>
</tr>
<tr>
<td>X(Rn)</td>
<td>4</td>
<td>4 (2)</td>
<td>4 (2)</td>
<td>2</td>
<td>CALL 2(LABEL)</td>
</tr>
<tr>
<td>EDE</td>
<td>4</td>
<td>4 (2)</td>
<td>4 (2)</td>
<td>2</td>
<td>PUSH EDE</td>
</tr>
<tr>
<td>&amp;EDE</td>
<td>4</td>
<td>4 (2)</td>
<td>4 (2)</td>
<td>2</td>
<td>SXT &amp;EDE</td>
</tr>
</tbody>
</table>

(1) The cycle count in MSP430 CPU is 4.
(2) The cycle count in MSP430 CPU is 5. Also, the cycle count is 5 for X(Rn) addressing mode, when Rn = SP.

### 4.5.1.5.3 Jump Instructions Cycles and Lengths

All jump instructions require one code word and take two CPU cycles to execute, regardless of whether the jump is taken or not.
4.5.1.5.4 Format I (Double-Operand) Instruction Cycles and Lengths

Table 4-10 lists the length and CPU cycles for all addressing modes of the MSP430 Format I instructions.

<table>
<thead>
<tr>
<th>Addressing Mode</th>
<th>No. of Cycles</th>
<th>Length of Instruction</th>
<th>Example</th>
</tr>
</thead>
<tbody>
<tr>
<td>Rn</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Rm</td>
<td>1</td>
<td>MOV R5,R8</td>
<td></td>
</tr>
<tr>
<td>PC</td>
<td>2</td>
<td>BR R9</td>
<td></td>
</tr>
<tr>
<td>x(Rm)</td>
<td>4(1)</td>
<td>ADD R5,4(R6)</td>
<td></td>
</tr>
<tr>
<td>EDE</td>
<td>4(1)</td>
<td>XOR R8,EDE</td>
<td></td>
</tr>
<tr>
<td>&amp;EDE</td>
<td>4(1)</td>
<td>MOV R5,&amp;EDE</td>
<td></td>
</tr>
<tr>
<td>@Rn</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Rm</td>
<td>2</td>
<td>AND @R4,R5</td>
<td></td>
</tr>
<tr>
<td>PC</td>
<td>3</td>
<td>BR @R8</td>
<td></td>
</tr>
<tr>
<td>x(Rm)</td>
<td>5(1)</td>
<td>XOR @R5,8(R6)</td>
<td></td>
</tr>
<tr>
<td>EDE</td>
<td>5(1)</td>
<td>MOV @R5,EDE</td>
<td></td>
</tr>
<tr>
<td>&amp;EDE</td>
<td>5(1)</td>
<td>MOV @R9+,EDE</td>
<td></td>
</tr>
<tr>
<td>@Rn+</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Rm</td>
<td>2</td>
<td>ADD @R5+,R6</td>
<td></td>
</tr>
<tr>
<td>PC</td>
<td>3</td>
<td>BR @R9+</td>
<td></td>
</tr>
<tr>
<td>x(Rm)</td>
<td>5(1)</td>
<td>XOR @R5,8(R6)</td>
<td></td>
</tr>
<tr>
<td>EDE</td>
<td>5(1)</td>
<td>MOV @R9+,EDE</td>
<td></td>
</tr>
<tr>
<td>&amp;EDE</td>
<td>5(1)</td>
<td>MOV @R9+,&amp;EDE</td>
<td></td>
</tr>
<tr>
<td>#N</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Rm</td>
<td>2</td>
<td>MOV #20,R9</td>
<td></td>
</tr>
<tr>
<td>PC</td>
<td>3</td>
<td>BR #2AEh</td>
<td></td>
</tr>
<tr>
<td>x(Rm)</td>
<td>5(1)</td>
<td>MOV #0300h,0(SP)</td>
<td></td>
</tr>
<tr>
<td>EDE</td>
<td>5(1)</td>
<td>ADD #33,EDE</td>
<td></td>
</tr>
<tr>
<td>&amp;EDE</td>
<td>5(1)</td>
<td>ADD #33,&amp;EDE</td>
<td></td>
</tr>
<tr>
<td>x(Rn)</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Rm</td>
<td>3</td>
<td>MOV 2(R5),R7</td>
<td></td>
</tr>
<tr>
<td>PC</td>
<td>3</td>
<td>BR 2(R6)</td>
<td></td>
</tr>
<tr>
<td>TONI</td>
<td>6(1)</td>
<td>MOV 4(R7),TONI</td>
<td></td>
</tr>
<tr>
<td>x(Rm)</td>
<td>6(1)</td>
<td>ADD 4(R4),0(R9)</td>
<td></td>
</tr>
<tr>
<td>&amp;TONI</td>
<td>6(1)</td>
<td>MOV 2(R4),&amp;TONI</td>
<td></td>
</tr>
<tr>
<td>EDE</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Rm</td>
<td>3</td>
<td>AND EDE,R6</td>
<td></td>
</tr>
<tr>
<td>PC</td>
<td>3</td>
<td>BR EDE</td>
<td></td>
</tr>
<tr>
<td>TONI</td>
<td>6(1)</td>
<td>CMP EDE,TONI</td>
<td></td>
</tr>
<tr>
<td>x(Rm)</td>
<td>6(1)</td>
<td>MOV EDE,0(SP)</td>
<td></td>
</tr>
<tr>
<td>&amp;TONI</td>
<td>6(1)</td>
<td>MOV EDE,&amp;TONI</td>
<td></td>
</tr>
<tr>
<td>&amp;EDE</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Rm</td>
<td>3</td>
<td>MOV &amp;EDE,R8</td>
<td></td>
</tr>
<tr>
<td>PC</td>
<td>3</td>
<td>BR &amp;EDE</td>
<td></td>
</tr>
<tr>
<td>TONI</td>
<td>6(1)</td>
<td>MOV &amp;EDE,TONI</td>
<td></td>
</tr>
<tr>
<td>x(Rm)</td>
<td>6(1)</td>
<td>MOV &amp;EDE,0(SP)</td>
<td></td>
</tr>
<tr>
<td>&amp;TONI</td>
<td>6(1)</td>
<td>MOV &amp;EDE,&amp;TONI</td>
<td></td>
</tr>
</tbody>
</table>

(1) MOV, BIT, and CMP instructions execute in one fewer cycle.

4.5.2 MSP430X Extended Instructions

The extended MSP430X instructions give the MSP430X CPU full access to its 20-bit address space. Most MSP430X instructions require an additional word of op-code called the extension word. Some extended instructions do not require an additional word and are noted in the instruction description. All addresses, indexes, and immediate numbers have 20-bit values when preceded by the extension word.
There are two types of extension words:

- Register/register mode for Format I instructions and register mode for Format II instructions
- Extension word for all other address mode combinations

### 4.5.2.1 Register Mode Extension Word

The register mode extension word is shown in Figure 4-24 and described in Table 4-11. An example is shown in Figure 4-26.

#### Figure 4-24. Extension Word for Register Modes

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15:11</td>
<td>Extension word op-code. Op-codes 1800h to 1FFFh are extension words.</td>
</tr>
<tr>
<td>10:9</td>
<td>Reserved</td>
</tr>
<tr>
<td>ZC</td>
<td>Zero carry</td>
</tr>
<tr>
<td>0</td>
<td>The executed instruction uses the status of the carry bit C.</td>
</tr>
<tr>
<td>1</td>
<td>The executed instruction uses the carry bit as 0. The carry bit is defined by the result of the final operation after instruction execution.</td>
</tr>
<tr>
<td>#</td>
<td>Repetition</td>
</tr>
<tr>
<td>0</td>
<td>The number of instruction repetitions is set by extension word bits 3:0.</td>
</tr>
<tr>
<td>1</td>
<td>The number of instruction repetitions is defined by the value of the four LSBs of Rn. See description for bits 3:0.</td>
</tr>
<tr>
<td>A/L</td>
<td>Data length extension. Together with the B/W bits of the following MSP430 instruction, the AL bit defines the used data length of the instruction.</td>
</tr>
</tbody>
</table>

#### Table 4-11. Description of the Extension Word Bits for Register Mode

<table>
<thead>
<tr>
<th>A/L</th>
<th>B/W</th>
<th>Comment</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>Reserved</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>20-bit address word</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>16-bit word</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>8-bit byte</td>
</tr>
<tr>
<td>5:4</td>
<td></td>
<td>Reserved</td>
</tr>
<tr>
<td>3:0</td>
<td></td>
<td>Repetition count</td>
</tr>
<tr>
<td># = 0</td>
<td>These four bits set the repetition count n. These bits contain n – 1.</td>
<td></td>
</tr>
<tr>
<td># = 1</td>
<td>These four bits define the CPU register whose bits 3:0 set the number of repetitions. Rn.3:0 contain n – 1.</td>
<td></td>
</tr>
</tbody>
</table>
4.5.2.2 Non-Register Mode Extension Word

The extension word for non-register modes is shown in Figure 4-25 and described in Table 4-12. An example is shown in Figure 4-27.

![Figure 4-25. Extension Word for Non-Register Modes](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15:11</td>
<td>Extension word op-code. Op-codes 1800h to 1FFFh are extension words.</td>
</tr>
<tr>
<td>Source Bits 19:16</td>
<td>The four MSBs of the 20-bit source. Depending on the source addressing mode, these four MSBs may belong to an immediate operand, an index or to an absolute address.</td>
</tr>
<tr>
<td>A/L</td>
<td>Data length extension. Together with the B/W bits of the following MSP430 instruction, the AL bit defines the used data length of the instruction.</td>
</tr>
<tr>
<td></td>
<td>A/L</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>5:4</td>
<td>Reserved</td>
</tr>
<tr>
<td>Destination Bits 19:16</td>
<td>The four MSBs of the 20-bit destination. Depending on the destination addressing mode, these four MSBs may belong to an index or to an absolute address.</td>
</tr>
</tbody>
</table>

---

**Note**

**B/W and A/L bit settings for SWPBX and SXTX**

<table>
<thead>
<tr>
<th>A/L</th>
<th>B/W</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>SWPBX.A, SXTX.A</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>N/A</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>SWPBW, SXTX.W</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>N/A</td>
</tr>
</tbody>
</table>
XORX.A R9, R8

1: Repetition count in bits 3:0
0: Use Carry
01: Address word

<table>
<thead>
<tr>
<th>15</th>
<th>14</th>
<th>13</th>
<th>12</th>
<th>11</th>
<th>10</th>
<th>9</th>
<th>8</th>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>0 0</td>
<td>ZC</td>
<td>#</td>
<td>A/L</td>
<td>Rsvd</td>
<td>(n−1)/Rn</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Op-code | Rsrc | Ad | B/W | As | Rdst |

Figure 4-26. Example for Extended Register/Register Instruction

XORX #12345h, 45678h(R15)

18xx extension word 12345h

<table>
<thead>
<tr>
<th>15</th>
<th>14</th>
<th>13</th>
<th>12</th>
<th>11</th>
<th>10</th>
<th>9</th>
<th>8</th>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
</tbody>
</table>

Source 19:16 | A/L | Rsvd | Destination 19:16 |

Op-code | Rsrc | Ad | B/W | As | Rdst |

Source 15:0

Destination 15:0

Figure 4-27. Example for Extended Immediate/Indexed Instruction

Immediate operand LSBs: 2345h

Index destination LSBs: 5678h
## 4.5.2.3 Extended Double-Operand (Format I) Instructions

All 12 double-operand instructions have extended versions as listed in Table 4-13.

<table>
<thead>
<tr>
<th>Mnemonic</th>
<th>Operands</th>
<th>Operation</th>
<th>Status Bits (1)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>MOVX (.B,.A)</td>
<td>src,dst</td>
<td>src → dst</td>
<td>–</td>
</tr>
<tr>
<td>ADDX (.B,.A)</td>
<td>src,dst</td>
<td>src + dst → dst</td>
<td>*</td>
</tr>
<tr>
<td>ADDCX (.B,.A)</td>
<td>src,dst</td>
<td>src + dst + C → dst</td>
<td>*</td>
</tr>
<tr>
<td>SUBX (.B,.A)</td>
<td>src,dst</td>
<td>dst + .not.src + 1 → dst</td>
<td>*</td>
</tr>
<tr>
<td>SUBCX (.B,.A)</td>
<td>src,dst</td>
<td>dst + .not.src + C → dst</td>
<td>*</td>
</tr>
<tr>
<td>CMPX (.B,.A)</td>
<td>src,dst</td>
<td>dst – src</td>
<td>*</td>
</tr>
<tr>
<td>DADDX (.B,.A)</td>
<td>src,dst</td>
<td>src + dst + C → dst (decimal)</td>
<td>*</td>
</tr>
<tr>
<td>BITX (.B,.A)</td>
<td>src,dst</td>
<td>src .and. dst</td>
<td>0</td>
</tr>
<tr>
<td>BICX (.B,.A)</td>
<td>src,dst</td>
<td>.not.src .and. dst → dst</td>
<td>–</td>
</tr>
<tr>
<td>BISX (.B,.A)</td>
<td>src,dst</td>
<td>src .or. dst → dst</td>
<td>–</td>
</tr>
<tr>
<td>XORX (.B,.A)</td>
<td>src,dst</td>
<td>src .xor. dst → dst</td>
<td>*</td>
</tr>
<tr>
<td>ANDX (.B,.A)</td>
<td>src,dst</td>
<td>src .and. dst → dst</td>
<td>0</td>
</tr>
</tbody>
</table>

(1)  
* = Status bit is affected.  
– = Status bit is not affected.  
0 = Status bit is cleared.  
1 = Status bit is set.

The four possible addressing combinations for the extension word for Format I instructions are shown in Figure 4-28.
### Figure 4-28. Extended Format I Instruction Formats

If the 20-bit address of a source or destination operand is located in memory, not in a CPU register, then two words are used for this operand as shown in Figure 4-29.

<table>
<thead>
<tr>
<th>15 14 13 12 11 10 9 8 7 6 5 4 3 2 1 0</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>0 0 0 1 1 0 0 ZC # A/L 0 0 n−1/Rn</td>
<td></td>
</tr>
<tr>
<td>Op-code src 0 B/W 0 0 dst</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>0 0 0 1 1 src.19:16 A/L 0 0 0 0 0 0</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>Op-code src Ad B/W As dst</td>
<td></td>
</tr>
<tr>
<td>src.15:0</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>0 0 0 1 1 0 0 0 0 A/L 0 0 dst.19:16</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>Op-code src Ad B/W As dst</td>
<td></td>
</tr>
<tr>
<td>dst.15:0</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>0 0 0 1 1 src.19:16 A/L 0 0 dst.19:16</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>Op-code src Ad B/W As dst</td>
<td></td>
</tr>
<tr>
<td>src.15:0</td>
<td></td>
</tr>
<tr>
<td>dst.15:0</td>
<td></td>
</tr>
</tbody>
</table>

### Figure 4-29. 20-Bit Addresses in Memory

<table>
<thead>
<tr>
<th>15 14 13 12 11 10 9 8 7 6 5 4 3 2 1 0</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>Address+2 0 ......................................................0</td>
<td>19:16</td>
</tr>
<tr>
<td>Address Operand LSBs 15:0</td>
<td></td>
</tr>
</tbody>
</table>

---

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4.5.2.4 Extended Single-Operand (Format II) Instructions

Extended MSP430X Format II instructions are listed in Table 4-14.

### Table 4-14. Extended Single-Operand Instructions

<table>
<thead>
<tr>
<th>Mnemonic</th>
<th>Operands</th>
<th>Operation</th>
<th>Status Bits (1)</th>
</tr>
</thead>
<tbody>
<tr>
<td>CALLA</td>
<td>dst</td>
<td>Call indirect to subroutine (20-bit address)</td>
<td>n V N Z C</td>
</tr>
<tr>
<td>POPM.A</td>
<td>#n,Rdst</td>
<td>Pop n 20-bit registers from stack</td>
<td>1 to 16</td>
</tr>
<tr>
<td>POPM.W</td>
<td>#n,Rdst</td>
<td>Pop n 16-bit registers from stack</td>
<td>1 to 16</td>
</tr>
<tr>
<td>PUSHM.A</td>
<td>#n,Rsrc</td>
<td>Push n 20-bit registers to stack</td>
<td>1 to 16</td>
</tr>
<tr>
<td>PUSHM.W</td>
<td>#n,Rsrc</td>
<td>Push n 16-bit registers to stack</td>
<td>1 to 16</td>
</tr>
<tr>
<td>PUSHX(.B,.A)</td>
<td>src</td>
<td>Push 8/16/20-bit source to stack</td>
<td></td>
</tr>
<tr>
<td>RRCM(.A)</td>
<td>#n,Rdst</td>
<td>Rotate right Rdst n bits through carry (16-/20-bit register)</td>
<td>1 to 4</td>
</tr>
<tr>
<td>RRUM(.A)</td>
<td>#n,Rdst</td>
<td>Rotate right Rdst n bits unsigned (16-/20-bit register)</td>
<td>1 to 4</td>
</tr>
<tr>
<td>RRAM(.A)</td>
<td>#n,Rdst</td>
<td>Rotate right Rdst n bits arithmetically (16-/20-bit register)</td>
<td>1 to 4</td>
</tr>
<tr>
<td>RLAM(.A)</td>
<td>#n,Rdst</td>
<td>Rotate left Rdst n bits arithmetically (16-/20-bit register)</td>
<td>1 to 4</td>
</tr>
<tr>
<td>RRCX(.B,.A)</td>
<td>dst</td>
<td>Rotate right dst through carry (8-/16-/20-bit data)</td>
<td>1</td>
</tr>
<tr>
<td>RRUX(.B,.A)</td>
<td>Rdst</td>
<td>Rotate right dst unsigned (8-/16-/20-bit)</td>
<td>1</td>
</tr>
<tr>
<td>RRAX(.B,.A)</td>
<td>dst</td>
<td>Rotate right dst arithmetically</td>
<td>1</td>
</tr>
<tr>
<td>SWPBX(.A)</td>
<td>dst</td>
<td>Exchange low byte with high byte</td>
<td>1</td>
</tr>
<tr>
<td>SXTX(.A)</td>
<td>Rdst</td>
<td>Bit7 → bit8 ... bit19</td>
<td>1</td>
</tr>
<tr>
<td>SXTX(.A)</td>
<td>dst</td>
<td>Bit7 → bit8 ... MSB</td>
<td>1</td>
</tr>
</tbody>
</table>

(1) * = Status bit is affected.
– = Status bit is not affected.
0 = Status bit is cleared.
1 = Status bit is set.

The three possible addressing mode combinations for Format II instructions are shown in Figure 4-30.

![Figure 4-30. Extended Format II Instruction Format](www.ti.com)
4.5.2.4.1 Extended Format II Instruction Format Exceptions

Exceptions for the Format II instruction formats are shown in Figure 4-31 through Figure 4-34.

<table>
<thead>
<tr>
<th>15</th>
<th>8</th>
<th>7</th>
<th>4</th>
<th>3</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Op-code</td>
<td>n-1</td>
<td>Rdst - n+1</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Figure 4-31. PUSHM/POPM Instruction Format**

<table>
<thead>
<tr>
<th>15</th>
<th>12</th>
<th>11</th>
<th>10</th>
<th>9</th>
<th>4</th>
<th>3</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>C</td>
<td>n−1</td>
<td>Op-code</td>
<td>Rdst</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>15</th>
<th>12</th>
<th>11</th>
<th>8</th>
<th>7</th>
<th>4</th>
<th>3</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>C</td>
<td>Rsrc</td>
<td>Op-code</td>
<td>0(PC)</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>15</th>
<th>12</th>
<th>11</th>
<th>8</th>
<th>7</th>
<th>4</th>
<th>3</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>C</td>
<td>#imm/abs19:16</td>
<td>Op-code</td>
<td>0(PC)</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>#imm15:0 / &amp;abs15:0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>15</th>
<th>12</th>
<th>11</th>
<th>8</th>
<th>7</th>
<th>4</th>
<th>3</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>C</td>
<td>Rsrc</td>
<td>Op-code</td>
<td>0(PC)</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>index15:0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Figure 4-32. RRCM, RRAM, RRUM, and RLAM Instruction Format**

<table>
<thead>
<tr>
<th>15</th>
<th>4</th>
<th>3</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Op-code</td>
<td>Rdst</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>15</th>
<th>4</th>
<th>3</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Op-code</td>
<td>Rdst</td>
<td></td>
<td></td>
</tr>
<tr>
<td>index15:0</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>15</th>
<th>4</th>
<th>3</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Op-code</td>
<td>#imm/ix/abs19:16</td>
<td></td>
<td></td>
</tr>
<tr>
<td>#imm15:0 / index15:0 / &amp;abs15:0</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Figure 4-33. BRA Instruction Format**

**Figure 4-34. CALLA Instruction Format**
4.5.2.5 Extended Emulated Instructions

The extended instructions together with the constant generator form the extended emulated instructions. Table 4-15 lists the emulated instructions.

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Explanation</th>
<th>Emulation</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADCX(.B,.A) dst</td>
<td>Add carry to dst</td>
<td>ADDCX(.B,.A) #0,dst</td>
</tr>
<tr>
<td>BRA dst</td>
<td>Branch indirect dst</td>
<td>MOVA dst,PC</td>
</tr>
<tr>
<td>RETA</td>
<td>Return from subroutine</td>
<td>MOVA @SP+,PC</td>
</tr>
<tr>
<td>CLRA Rdst</td>
<td>Clear Rdst</td>
<td>MOV #0,Rdst</td>
</tr>
<tr>
<td>CLRX(.B,.A) dst</td>
<td>Clear dst</td>
<td>MOVX(.B,.A) #0,dst</td>
</tr>
<tr>
<td>DADDCX(.B,.A) dst</td>
<td>Add carry to dst decimally</td>
<td>DADDCX(.B,.A) #0,dst</td>
</tr>
<tr>
<td>DECX(.B,.A) dst</td>
<td>Decrement dst by 1</td>
<td>SUBX(.B,.A) #1,dst</td>
</tr>
<tr>
<td>DECDX(.B,.A) Rdst</td>
<td>Decrement Rdst by 2</td>
<td>SUBA #2,Rdst</td>
</tr>
<tr>
<td>INCDX(.B,.A) dst</td>
<td>Increment dst by 1</td>
<td>ADDX(.B,.A) #1,dst</td>
</tr>
<tr>
<td>INCDA Rdst</td>
<td>Increment Rdst by 2</td>
<td>ADDA #2,Rdst</td>
</tr>
<tr>
<td>INCX(.B,.A) Rdst</td>
<td>Increment dst by 2</td>
<td>ADDX(.B,.A) #2,dst</td>
</tr>
<tr>
<td>INXVX(.B,.A) Rdst</td>
<td>Invert dst</td>
<td>XORX(.B,.A) #1,dst</td>
</tr>
<tr>
<td>RLAX(.B,.A) dst</td>
<td>Shift left dst arithmetically</td>
<td>ADDX(.B,.A) dst,dst</td>
</tr>
<tr>
<td>RLAX(.B,.A) dst</td>
<td>Shift left dst logically through carry</td>
<td>ADDCX(.B,.A) dst,dst</td>
</tr>
<tr>
<td>SBCX(.B,.A) dst</td>
<td>Subtract carry from dst</td>
<td>SUBCX(.B,.A) #0,dst</td>
</tr>
<tr>
<td>TSTA Rdst</td>
<td>Test Rdst (compare with 0)</td>
<td>CMPA #0,Rdst</td>
</tr>
<tr>
<td>TSTX(.B,.A) dst</td>
<td>Test dst (compare with 0)</td>
<td>CMPX(.B,.A) #0,dst</td>
</tr>
<tr>
<td>POPX dst</td>
<td>Pop to dst</td>
<td>MOVX(.B,.A) @SP+,dst</td>
</tr>
</tbody>
</table>
4.5.2.6 MSP430X Address Instructions

MSP430X address instructions are instructions that support 20-bit operands but have restricted addressing modes. The addressing modes are restricted to the Register mode and the Immediate mode, except for the MOVA instruction as listed in Table 4-16. Restricting the addressing modes removes the need for the additional extension-word op-code improving code density and execution time. Address instructions should be used any time an MSP430X instruction is needed with the corresponding restricted addressing mode.

Table 4-16. Address Instructions, Operate on 20-Bit Register Data

<table>
<thead>
<tr>
<th>Mnemonic</th>
<th>Operands</th>
<th>Operation</th>
<th>Status Bits (1)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>ADDA</td>
<td>Rsrc, Rdst</td>
<td>Add source to destination register</td>
<td>*</td>
</tr>
<tr>
<td></td>
<td>#imm20, Rdst</td>
<td></td>
<td></td>
</tr>
<tr>
<td>MOVA</td>
<td>Rsrc, Rdst</td>
<td>Move source to destination</td>
<td>–</td>
</tr>
<tr>
<td></td>
<td>#imm20, Rdst</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>z16(Rsrc), Rdst</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>EDE, Rdst</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>&amp;abs20, Rdst</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>@Rsrc, Rdst</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>@Rsrc+, Rdst</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>Rsrc, z16(Rdst)</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>Rsrc, &amp;abs20</td>
<td></td>
<td></td>
</tr>
<tr>
<td>CMPA</td>
<td>Rsrc, Rdst</td>
<td>Compare source to destination register</td>
<td>*</td>
</tr>
<tr>
<td></td>
<td>#imm20, Rdst</td>
<td></td>
<td></td>
</tr>
<tr>
<td>SUBA</td>
<td>Rsrc, Rdst</td>
<td>Subtract source from destination register</td>
<td>*</td>
</tr>
<tr>
<td></td>
<td>#imm20, Rdst</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

(1) * = Status bit is affected.
– = Status bit is not affected.
0 = Status bit is cleared.
1 = Status bit is set.
4.5.2.7 MSP430X Instruction Execution

The number of CPU clock cycles required for an MSP430X instruction depends on the instruction format and the addressing modes used, not the instruction itself. The number of clock cycles refers to MCLK.

4.5.2.7.1 MSP430X Format II (Single-Operand) Instruction Cycles and Lengths

Table 4-17 lists the length and the CPU cycles for all addressing modes of the MSP430X extended single-operand instructions.

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Rn</th>
<th>@Rn</th>
<th>@Rn+</th>
<th>#N</th>
<th>X(Rn)</th>
<th>EDE</th>
<th>&amp;EDE</th>
</tr>
</thead>
<tbody>
<tr>
<td>RRAM</td>
<td>n/1</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>RRCM</td>
<td>n/1</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>RRUM</td>
<td>n/1</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>RLAM</td>
<td>n/1</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>PUSHM</td>
<td>2+n/1</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>PUSHM.A</td>
<td>2+2n/1</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>POPM</td>
<td>2+n/1</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>POPM.A</td>
<td>2+2n/1</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>CALLA</td>
<td>4/1</td>
<td>5/1</td>
<td>5/1</td>
<td>4/2</td>
<td>6 (1)/2</td>
<td>6/2</td>
<td>6/2</td>
</tr>
<tr>
<td>RRAM(.B)</td>
<td>1+n/2</td>
<td>4/2</td>
<td>4/2</td>
<td>–</td>
<td>5/3</td>
<td>5/3</td>
<td>5/3</td>
</tr>
<tr>
<td>RRCM(.B)</td>
<td>1+n/2</td>
<td>6/2</td>
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(1)  Add one cycle when Rn = SP
4.5.2.7.2 MSP430X Format I (Double-Operand) Instruction Cycles and Lengths

Table 4-18 lists the length and CPU cycles for all addressing modes of the MSP430X extended Format I instructions.

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<th>Source</th>
<th>Destination</th>
<th>No. of Cycles</th>
<th>Length of instruction</th>
<th>Examples</th>
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<td>BITX &amp;EDE,&amp;TONI</td>
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</table>

(1) Repeat instructions require n + 1 cycles, where n is the number of times the instruction is executed.
(2) Reduce the cycle count by one for MOV, BIT, and CMP instructions.
(3) Reduce the cycle count by two for MOV, BIT, and CMP instructions.
(4) Reduce the cycle count by one for MOV, ADD, and SUB instructions.
### 4.5.2.7.3 MSP430X Address Instruction Cycles and Lengths

Table 4-19 lists the length and the CPU cycles for all addressing modes of the MSP430X address instructions.

<table>
<thead>
<tr>
<th>Addressing Mode</th>
<th>Execution Time (MCLK Cycles)</th>
<th>Length of Instruction (Words)</th>
<th>Example</th>
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</thead>
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<td>Rn</td>
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<td>CMPA 1, ADDA 1, SUBA 1</td>
<td>CMPA R5, R8</td>
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<td>CMPA 2, ADDA 1, SUBA 1</td>
<td>SUBA R9, PC</td>
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<td>x(Rm) 4, CMPA 2, ADDA 2</td>
<td>CMPA 2, ADDA 2</td>
<td>MOVA R5, x(R6)</td>
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<td>MOVA R8, EDE</td>
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<td>&amp;EDE 4, CMPA 2, ADDA 2</td>
<td>CMPA 2, ADDA 2</td>
<td>MOVA R8, &amp;EDE</td>
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<tr>
<td>@Rn</td>
<td>Rm 3, CMPA 1, ADDA 1</td>
<td>CMPA 1, ADDA 1</td>
<td>MOVA @R5, R8</td>
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<td>PC 3, CMPA 1, ADDA 1</td>
<td>CMPA 1, ADDA 1</td>
<td>MOVA @R9, PC</td>
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<td>@Rn+</td>
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</table>
### 4.6 Instruction Set Description

Table 4-20 shows all available instructions:

**Table 4-20. Instruction Map of MSP430X**

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<th>100</th>
<th>140</th>
<th>180</th>
<th>1C0</th>
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<th>280</th>
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<th>300</th>
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<th>380</th>
<th>3C0</th>
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<td>Exxx</td>
<td>XOR, XOR.B</td>
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<tr>
<td>Fxxx</td>
<td>AND, AND.B</td>
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</tbody>
</table>
### 4.6.1 Extended Instruction Binary Descriptions

Detailed MSP430X instruction binary descriptions are shown in the following tables.

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Instruction Group</th>
<th>src or data 19:16</th>
<th>Instruction Identifier</th>
<th>dst</th>
</tr>
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<tbody>
<tr>
<td>MOVA</td>
<td></td>
<td>15 12 11 8 7 4 3 0</td>
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<td></td>
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<td></td>
<td>0 0 0 0 src</td>
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</tr>
<tr>
<td></td>
<td></td>
<td>0 0 0 0 src</td>
<td>0 0 0 1 dst</td>
<td></td>
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<tr>
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<td></td>
<td>0 0 0 0 &amp;abs 19:16</td>
<td>0 0 1 0 dst</td>
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<td>&amp;abs 15:0</td>
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<td>0 0 0 0 src</td>
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<td>&amp;abs 15:0</td>
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<td>0 1 1 1 dst</td>
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<td>x 15:0</td>
<td>±15-bit index x</td>
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<td>0 0 0 0 imm 19:16</td>
<td>1 0 0 0 dst</td>
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<td></td>
<td></td>
<td>imm 15:0</td>
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<tr>
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<td></td>
<td></td>
<td>imm 15:0</td>
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<tr>
<td>ADDA</td>
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<td>1 0 1 0 dst</td>
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</tr>
<tr>
<td></td>
<td></td>
<td>imm 15:0</td>
<td></td>
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<tr>
<td>SUBA</td>
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<tr>
<td></td>
<td></td>
<td>imm 15:0</td>
<td></td>
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<table>
<thead>
<tr>
<th>Instruction</th>
<th>Instruction Group</th>
<th>src or data 19:16</th>
<th>Instruction Identifier</th>
<th>dst</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOVA</td>
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<td>15 12 11 8 7 4 3 0</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>0 0 0 0 src</td>
<td>1 1 0 0 dst</td>
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<td></td>
<td>CMPA</td>
<td>0 1 1 0 dst</td>
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<td>ADDA</td>
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<tr>
<td></td>
<td></td>
<td>SUBA</td>
<td>1 1 1 1 dst</td>
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</table>

<table>
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<tr>
<th>Instruction</th>
<th>Instruction Group</th>
<th>Bit Loc.</th>
<th>Inst. ID</th>
<th>Instruction Identifier</th>
<th>dst</th>
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<td>RRCM.A</td>
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<td>0 0 0 0 n−1 0 0 0 1 0 0 0</td>
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<td>RRCM.A #n,Rdst</td>
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<td>RRAM.A</td>
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<td>RRAM.A #n,Rdst</td>
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<td>RLAM.A</td>
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<td>0 0 0 0 n−1 1 0 0 1 0 0 0</td>
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<td>RLAM.A #n,Rdst</td>
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<tr>
<td>RRUM.A</td>
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<td>0 0 0 0 n−1 1 1 0 1 0 0 0</td>
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<td>RRUM.A #n,Rdst</td>
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<tr>
<td>RRCM.W</td>
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<td>RRCM.W #n,Rdst</td>
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<td>RRAM.W #n,Rdst</td>
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<td>RLAM.W #n,Rdst</td>
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<td>RRUM.W #n,Rdst</td>
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<td>dst</td>
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<td>RETI</td>
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<td>CALLA Rdst</td>
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<tr>
<td>CALLA</td>
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<td>0 0 0 1 0 0 1 1 0 1 0 0 dst</td>
<td>CALLA x(Rdst)</td>
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<td>0 0 0 1 0 0 1 1 0 1 0 1 dst</td>
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<td>x.15:0</td>
<td>CALLA @Rdst</td>
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<td>0 0 0 1 0 0 1 1 0 1 1 dst</td>
<td>CALLA @Rdst+</td>
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<td>0 0 0 1 0 0 1 1 1 0 0 0 &amp;abs.19:16</td>
<td>CALLA &amp;abs20</td>
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<td>x.15:0</td>
<td>CALLA EDE</td>
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<tr>
<td></td>
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<td>0 0 0 1 0 0 1 1 1 0 0 1 x.19:16 CALLA EDE</td>
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<tr>
<td></td>
<td></td>
<td>x.15:0</td>
<td>CALLA x(PC)</td>
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<td>0 0 0 1 0 0 1 1 1 0 1 1 imm.19:16 CALLA #imm20</td>
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<td>Reserved</td>
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<td>PUSHM.A</td>
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<td>0 0 0 1 0 1 0 0 n-1 dst</td>
<td>PUSHM.A #n,Rdst</td>
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<td></td>
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<tr>
<td>PUSHM.W</td>
<td></td>
<td>0 0 0 1 0 1 0 1 n-1 dst</td>
<td>PUSHM.W #n,Rdst</td>
<td></td>
<td></td>
</tr>
<tr>
<td>POPM.A</td>
<td></td>
<td>0 0 0 1 0 1 1 0 n-1 dst – n + 1</td>
<td>POPM.A #n,Rdst</td>
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</tr>
<tr>
<td>POPM.W</td>
<td></td>
<td>0 0 0 1 0 1 1 1 n-1 dst – n + 1</td>
<td>POPM.W #n,Rdst</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>
4.6.2 MSP430 Instructions

The MSP430 instructions are described in the following sections.

See Section 4.6.3 for MSP430X extended instructions and Section 4.6.4 for MSP430X address instructions.
4.6.2.1 ADC

* ADC[W]  Add carry to destination
* ADC.B  Add carry to destination

Syntax
ADC dst or ADC.W dst
ADC.B dst

Operation
dst + C → dst

Emulation
ADDC #0,dst
ADDC.B #0,dst

Description
The carry bit (C) is added to the destination operand. The previous contents of the destination are lost.

Status Bits
N:  Set if result is negative, reset if positive
Z:  Set if result is zero, reset otherwise
C:  Set if dst was incremented from 0FFFFh to 0000, reset otherwise
    Set if dst was incremented from 0FFh to 00, reset otherwise
V:  Set if an arithmetic overflow occurs, otherwise reset

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
The 16-bit counter pointed to by R13 is added to a 32-bit counter pointed to by R12.

ADD     @R13,0(R12)      ; Add LSDs
ADC     2(R12)           ; Add carry to MSD

Example
The 8-bit counter pointed to by R13 is added to a 16-bit counter pointed to by R12.

ADD.B   @R13,0(R12)      ; Add LSDs
ADC.B   1(R12)           ; Add carry to MSD
4.6.2.2 ADD

**ADD[]**
- ADD src,dst
- ADD.B src,dst

**Syntax**
- ADD src,dst
- ADD.W src,dst
- ADD.B src,dst

**Operation**
- src + dst → dst

**Description**
The source operand is added to the destination operand. The previous content of the destination is lost.

**Status Bits**
- **N:** Set if result is negative (MSB = 1), reset if positive (MSB = 0)
- **Z:** Set if result is zero, reset otherwise
- **C:** Set if there is a carry from the MSB of the result, reset otherwise
- **V:** Set if the result of two positive operands is negative, or if the result of two negative numbers is positive, reset otherwise

**Mode Bits**
OSCOFF, CPUOFF, and GIE are not affected.

**Example**
Ten is added to the 16-bit counter CNTR located in lower 64KB.

```
ADD.W #10,&CNTR ; Add 10 to 16-bit counter
```

**Example**
A table word pointed to by R5 (20-bit address in R5) is added to R6. The jump to label TONI is performed on a carry.

```
ADD.W &R5,R6 ; Add table word to R6. R6.19:16 = 0
JC TONI ; Jump if carry
... ; No carry
```

**Example**
A table byte pointed to by R5 (20-bit address) is added to R6. The jump to label TONI is performed if no carry occurs. The table pointer is auto-incremented by 1. R6.19:8 = 0

```
ADD.B &R5+,R6 ; Add byte to R6. R5 + 1. R6: 000xxh
JNC TONI ; Jump if no carry
... ; Carry occurred
```
4.6.2.3 ADDC

ADDC[W] Add source word and carry to destination word
ADDC.B Add source byte and carry to destination byte

Syntax
ADDC src,dst or ADDC.W src,dst
ADDC.B src,dst

Operation
src + dst + C → dst

Description
The source operand and the carry bit C are added to the destination operand. The previous content of the destination
is lost.

Status Bits
N: Set if result is negative (MSB = 1), reset if positive (MSB = 0)
Z: Set if result is zero, reset otherwise
C: Set if there is a carry from the MSB of the result, reset otherwise
V: Set if the result of two positive operands is negative, or if the result of two negative numbers is positive, reset
otherwise

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
Constant value 15 and the carry of the previous instruction are added to the 16-bit counter CNTR located in lower
64KB.

ADDC.W #15,&CNTR ; Add 15 + C to 16-bit CNTR

Example
A table word pointed to by R5 (20-bit address) and the carry C are added to R6. The jump to label TONI is performed
on a carry. R6.19:16 = 0

ADDC.W @R5,R6 ; Add table word + C to R6
JC TONI ; Jump if carry
... ; No carry

Example
A table byte pointed to by R5 (20-bit address) and the carry bit C are added to R6. The jump to label TONI is
performed if no carry occurs. The table pointer is auto-incremented by 1. R6.19:8 = 0

ADDC.B @R5+,R6 ; Add table byte + C to R6. R5 + 1
JNC TONI ; Jump if no carry
... ; Carry occurred
4.6.2.4 AND

**AND[W]** Logical AND of source word with destination word

**AND.B** Logical AND of source byte with destination byte

**Syntax**

\[
\text{AND src,dst or AND.W src,dst} \\
\text{AND.B src,dst}
\]

**Operation**

\[
\text{src .and. dst → dst}
\]

**Description**

The source operand and the destination operand are logically ANDed. The result is placed into the destination. The source operand is not affected.

**Status Bits**

- **N:** Set if result is negative (MSB = 1), reset if positive (MSB = 0)
- **Z:** Set if result is zero, reset otherwise
- **C:** Set if the result is not zero, reset otherwise. \(C = \neg Z\)
- **V:** Reset

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

The bits set in R5 (16-bit data) are used as a mask (AA55h) for the word TOM located in the lower 64KB. If the result is zero, a branch is taken to label TONI. R5.19:16 = 0

\[
\begin{align*}
\text{MOV} & \quad  \#AA55h, R5 ; \text{Load 16-bit mask to R5} \\
\text{AND} & \quad  \text{R5, &TOM} ; \text{TOM .and. R5 → TOM} \\
\text{JZ} & \quad  \text{TONI} ; \text{Jump if result 0} \\
\ldots  & \quad  ; \text{Result > 0}
\end{align*}
\]

or shorter:

\[
\begin{align*}
\text{AND} & \quad  \text{#AA55h, &TOM} ; \text{TOM .and. AA55h → TOM} \\
\text{JZ} & \quad  \text{TONI} ; \text{Jump if result 0}
\end{align*}
\]

**Example**

A table byte pointed to by R5 (20-bit address) is logically ANDed with R6. R5 is incremented by 1 after the fetching of the byte. R6.19:8 = 0

\[
\begin{align*}
\text{AND.B} & \quad  @\text{R5+}, \text{R6} ; \text{AND table byte with R6. R5 + 1}
\end{align*}
\]
4.6.2.5 BIC

- **BIC[W]**  
  Clear bits set in source word in destination word
- **BIC.B**  
  Clear bits set in source byte in destination byte

**Syntax**

- `BIC src,dst`  
- `BIC.W src,dst`  
- `BIC.B src,dst`

**Operation**

- `(.not. src) .and. dst → dst`

**Description**

The inverted source operand and the destination operand are logically ANDed. The result is placed into the destination. The source operand is not affected.

**Status Bits**

- **N**: Not affected
- **Z**: Not affected
- **C**: Not affected
- **V**: Not affected

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

The bits 15:14 of R5 (16-bit data) are cleared. R5.19:16 = 0

```
BIC #0C000h,R5  ; Clear R5.19:14 bits
```

**Example**

A table word pointed to by R5 (20-bit address) is used to clear bits in R7. R7.19:16 = 0

```
BIC.W @R5,R7  ; Clear bits in R7 set in @R5
```

**Example**

A table byte pointed to by R5 (20-bit address) is used to clear bits in Port1.

```
BIC.B @R5,&P1OUT  ; Clear I/O port P1 bits set in @R5
```
### 4.6.2.6 BIS

- **BIS[W]**: Set bits set in source word in destination word
- **BIS.B**: Set bits set in source byte in destination byte

**Syntax**

- `BIS src,dst` or `BIS.W src,dst`
- `BIS.B src,dst`

**Operation**

- `src .or. dst → dst`

**Description**

The source operand and the destination operand are logically ORed. The result is placed into the destination. The source operand is not affected.

**Status Bits**

- N: Not affected
- Z: Not affected
- C: Not affected
- V: Not affected

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

Bits 15 and 13 of R5 (16-bit data) are set to one. R5.19:16 = 0

```assembly
BIS      #A000h,R5      ; Set R5 bits
```

**Example**

A table word pointed to by R5 (20-bit address) is used to set bits in R7. R7.19:16 = 0

```assembly
BIS.W   @R5,R7          ; Set bits in R7
```

**Example**

A table byte pointed to by R5 (20-bit address) is used to set bits in Port1. R5 is incremented by 1 afterwards.

```assembly
BIS.B   @R5+,&P1OUT     ; Set I/O port P1 bits. R5 + 1
```
4.6.2.7 BIT

Test bits set in source word in destination word

BIT.W Test bits set in source byte in destination byte

Syntax

BIT src,dst or BIT.W src,dst
BIT.B src,dst

Operation

src .and. dst

Description

The source operand and the destination operand are logically ANDed. The result affects only the status bits in SR.

Register mode: the register bits Rdst.19:16 (.W) resp. Rdst. 19:8 (.B) are not cleared!

Status Bits

N: Set if result is negative (MSB = 1), reset if positive (MSB = 0)
Z: Set if result is zero, reset otherwise
C: Set if the result is not zero, reset otherwise. C = (.not. Z)
V: Reset

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

Test if one (or both) of bits 15 and 14 of R5 (16-bit data) is set. Jump to label TONI if this is the case. R5.19:16 are not affected.

BIT     #'C000h,R5       ; Test R5.15:14 bits
JNZ     TONI            ; At least one bit is set in R5
...                     ; Both bits are reset

Example

A table word pointed to by R5 (20-bit address) is used to test bits in R7. Jump to label TONI if at least one bit is set. R7.19:16 are not affected.

BIT.W   @R5,R7          ; Test bits in R7
JC      TONI            ; At least one bit is set
...                     ; Both are reset

Example

A table byte pointed to by R5 (20-bit address) is used to test bits in output Port1. Jump to label TONI if no bit is set. The next table byte is addressed.

BIT.B   @R5+,&P1OUT     ; Test I/O port P1 bits. R5 + 1
JNC     TONI            ; No corresponding bit is set
...                     ; At least one bit is set
4.6.2.8 BR, BRANCH

BR, BRANCH  Branch to destination in lower 64K address space

Syntax  

BR dst

Operation  

dst → PC

Emulation  

MOV dst,PC

Description  

An unconditional branch is taken to an address anywhere in the lower 64K address space. All source addressing modes can be used. The branch instruction is a word instruction.

Status Bits  

Status bits are not affected.

Example  

Examples for all addressing modes are given.

BR #EXEC ; Branch to label EXEC or direct branch (e.g. #0A4h)
; Core instruction MOV @PC+,PC
BR EXEC ; Branch to the address contained in EXEC
; Core instruction MOV X(PC),PC
; Indirect address
BR &EXEC ; Branch to the address contained in absolute
; address EXEC
; Core instruction MOV X(0),PC
; Indirect address
BR R5 ; Branch to the address contained in R5
; Core instruction MOV R5,PC
; Indirect R5
BR @R5 ; Branch to the address contained in the word
; pointed to by R5.
; Core instruction MOV @R5,PC
; Indirect, indirect R5
BR @R5+ ; Branch to the address contained in the word pointed
; to by R5 and increment pointer in R5 afterwards.
; The next time-S/W flow uses R5 pointer-it can
; alter program execution due to access to
; next address in a table pointed to by R5
; Core instruction MOV @R5,PC
; Indirect, indirect R5 with autoincrement
BR X(R5) ; Branch to the address contained in the address
; pointed to by R5 + X (e.g. table with address
; starting at X). X can be an address or a label
; Core instruction MOV X(R5),PC
; Indirect, indirect R5 + X
4.6.2.9 CALL

CALL dst
Call a subroutine in lower 64KB

Syntax
CALL dst

Operation
dst → PC 16-bit dst is evaluated and stored
SP – 2 → SP
PC → @SP updated PC with return address to TOS
tmp → PC saved 16-bit dst to PC

Description
A subroutine call is made from an address in the lower 64KB to a subroutine address in the lower 64KB. All seven
source addressing modes can be used. The call instruction is a word instruction. The return is made with the RET
instruction.

Status Bits
Status bits are not affected.
PC.19:16 cleared (address in lower 64KB)

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Examples
Examples for all addressing modes are given.
Immediate Mode: Call a subroutine at label EXEC (lower 64KB) or call directly to address.

CALL #EXEC ; Start address EXEC
CALL #0AA04h ; Start address 0AA04h

Symbolic Mode: Call a subroutine at the 16-bit address contained in address EXEC. EXEC is located at the address
(PC + X) where X is within PC + 32 K.

CALL EXEC ; Start address at @EXEC. z16(PC)

Absolute Mode: Call a subroutine at the 16-bit address contained in absolute address EXEC in the lower 64KB.

CALL &EXEC ; Start address at @EXEC

Register mode: Call a subroutine at the 16-bit address contained in register R5.15:0.

CALL R5 ; Start address at R5

Indirect Mode: Call a subroutine at the 16-bit address contained in the word pointed to by register R5 (20-bit address).

CALL @R5 ; Start address at @R5
4.6.2.10 CLR

* CLR,[W] Clear destination
* CLR.B Clear destination

Syntax
CLR dst or CLR.W dst
CLR.B dst

Operation
0 → dst

Emulation
MOV #0,dst
MOV.B #0,dst

Description
The destination operand is cleared.

Status Bits
Status bits are not affected.

Example
RAM word TONI is cleared.

CLR     TONI      ; 0 -> TONI

Example
Register R5 is cleared.

CLR     R5

Example
RAM byte TONI is cleared.

CLR.B   TONI      ; 0 -> TONI
4.6.2.11 CLRC

* CLRC Clear carry bit
Syntax CLRC
Operation \(0 \rightarrow C\)
Emulation BIC #1,SR
Description The carry bit (C) is cleared. The clear carry instruction is a word instruction.
Status Bits
- N: Not affected
- Z: Not affected
- C: Cleared
- V: Not affected
Mode Bits OSCOFF, CPUOFF, and GIE are not affected.
Example The 16-bit decimal counter pointed to by R13 is added to a 32-bit counter pointed to by R12.

```
CLRC                   ; C-0: defines start
DADD   @R13,0(R12)     ; add 16-bit counter to low word of 32-bit counter
DADC   2(R12)          ; add carry to high word of 32-bit counter
```
4.6.2.12 CLRN

CLRN Clear negative bit

Syntax

CLRN

Operation

0 → N
or
(.NOT.src .AND. dst → dst)

Emulation

BIC #4,SR

Description

The constant 04h is inverted (0FFFBh) and is logically ANDed with the destination operand. The result is placed into the destination. The clear negative bit instruction is a word instruction.

Status Bits

N: Reset to 0
Z: Not affected
C: Not affected
V: Not affected

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

The negative bit in the SR is cleared. This avoids special treatment with negative numbers of the subroutine called.

```
 CLRN
 CALL SUBR
 ...
 ...
 SUBR JN SUBRET ; If input is negative: do nothing and return
 ...
 ...
 ...
 SUBRET RET
```
4.6.2.13 CLRZ

* CLRZ  Clear zero bit
Syntax  CLRZ
Operation  0 → Z
or
(.NOT.src .AND. dst → dst)
Emulation  BIC #2,SR
Description  The constant 02h is inverted (0FFFDh) and logically ANDed with the destination operand. The result is placed into the destination. The clear zero bit instruction is a word instruction.
Status Bits  
N:  Not affected
Z:  Reset to 0
C:  Not affected
V:  Not affected
Mode Bits  OSCOFF, CPUOFF, and GiE are not affected.
Example  The zero bit in the SR is cleared.

CLRZ

Indirect, Auto-Increment mode: Call a subroutine at the 16-bit address contained in the word pointed to by register R5 (20-bit address) and increment the 16-bit address in R5 afterwards by 2. The next time the software uses R5 as a pointer, it can alter the program execution due to access to the next word address in the table pointed to by R5.

CALL   @R5+          ; Start address at @R5. R5 + 2

Indexed mode: Call a subroutine at the 16-bit address contained in the 20-bit address pointed to by register (R5 + X), for example, a table with addresses starting at X. The address is within the lower 64KB. X is within +32KB.

CALL   X(R5)         ; Start address at @(R5+X). z16(R5)
### 4.6.2.14 CMP

**CMP[W]**  
Compare source word and destination word

**CMP.B**  
Compare source byte and destination byte

**Syntax**  
- `CMP src,dst` or `CMP.W src,dst`
- `CMP.B src,dst`

**Operation**  
- `(.not.src) + 1 + dst`
- `dst – src`

**Emulation**  
`BIC #2,SR`

**Description**  
The source operand is subtracted from the destination operand. This is made by adding the 1's complement of the source + 1 to the destination. The result affects only the status bits in SR.

Register mode: the register bits Rdst.19:16 (.W) resp. Rdst. 19:8 (.B) are not cleared.

**Status Bits**  
- **N:** Set if result is negative (src > dst), reset if positive (src ≤ dst)
- **Z:** Set if result is zero (src = dst), reset otherwise (src ≠ dst)
- **C:** Set if there is a carry from the MSB, reset otherwise
- **V:** Set if the subtraction of a negative source operand from a positive destination operand delivers a negative result, or if the subtraction of a positive source operand from a negative destination operand delivers a positive result, reset otherwise (no overflow).

**Mode Bits**  
OSCOFF, CPUOFF, and GIE are not affected.

**Example**  
Compare word EDE with a 16-bit constant 1800h. Jump to label TONI if EDE equals the constant. The address of EDE is within PC + 32 K.

```
CMP      #01800h,EDE      ; Compare word EDE with 1800h
JEQ      TONI             ; EDE contains 1800h
...                       ; Not equal
```

**Example**  
A table word pointed to by (R5 + 10) is compared with R7. Jump to label TONI if R7 contains a lower, signed 16-bit number. R7.19:16 is not cleared. The address of the source operand is a 20-bit address in full memory range.

```
CMP.W    10(R5),R7        ; Compare two signed numbers
JL       TONI             ; R7 < 10(R5)
...                       ; R7 >= 10(R5)
```

**Example**  
A table byte pointed to by R5 (20-bit address) is compared to the value in output Port1. Jump to label TONI if values are equal. The next table byte is addressed.

```
CMP.B    @R5+,&P1OUT      ; Compare P1 bits with table. R5 + 1
JEQ      TONI             ; Equal contents
...                       ; Not equal
```
4.6.2.15 DADC

* DADC[.W] Add carry decimally to destination
* DADC.B Add carry decimally to destination

Syntax
DADC dst or DADC.W dst
DADC.B dst

Operation
dst + C → dst (decimally)

Emulation
DADD #0,dst
DADD.B #0,dst

Description
The carry bit (C) is added decimally to the destination.

Status Bits
N: Set if MSB is 1
Z: Set if dst is 0, reset otherwise
C: Set if destination increments from 9999 to 0000, reset otherwise
C: Set if destination increments from 99 to 00, reset otherwise
V: Undefined

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
The four-digit decimal number contained in R5 is added to an eight-digit decimal number pointed to by R8.

```assembly
CLRC ; Reset carry
       ; next instruction's start condition is defined
DADD R5,0(R8) ; Add LSDs + C
DADC 2(R8) ; Add carry to MSD
```

Example
The two-digit decimal number contained in R5 is added to a four-digit decimal number pointed to by R8.

```assembly
CLRC ; Reset carry
       ; next instruction's start condition is defined
DADD.B R5,0(R8) ; Add LSDs + C
DADC 1(R8) ; Add carry to MSDs
```
4.6.2.16 DADD

* DADD[W]  Add source word and carry decimally to destination word
* DADD.B  Add source byte and carry decimally to destination byte

Syntax
DADD src, dst or DADD.W src, dst
DADD.B src, dst

Operation
src + dst + C → dst (decimally)

Description
The source operand and the destination operand are treated as two (.B) or four (.W) binary coded decimals (BCD) with positive signs. The source operand and the carry bit C are added decimally to the destination operand. The source operand is not affected. The previous content of the destination is lost. The result is not defined for non-BCD numbers.

Status Bits
N: Set if MSB of result is 1 (word > 7999h, byte > 79h), reset if MSB is 0
Z: Set if result is zero, reset otherwise
C: Set if the BCD result is too large (word > 9999h, byte > 99h), reset otherwise
V: Undefined

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
Decimal 10 is added to the 16-bit BCD counter DECCNTR.

DADD #10h, &DECCNTR ; Add 10 to 4-digit BCD counter

Example
The eight-digit BCD number contained in 16-bit RAM addresses BCD and BCD+2 is added decimally to an eight-digit BCD number contained in R4 and R5 (BCD+2 and R5 contain the MSDs). The carry C is added, and cleared.

CLRC ; Clear carry
DADD.W &BCD, R4 ; Add LSDs. R4.19:16 = 0
DADD.W &BCD+2, R5 ; Add MSDs with carry. R5.19:16 = 0
JC OVERFLOW ; Result >9999,9999: go to error routine
... ; Result ok

Example
The two-digit BCD number contained in word BCD (16-bit address) is added decimally to a two-digit BCD number contained in R4. The carry C is added, also. R4.19:8 = 0CLRC ; Clear carry
DADD.B &BCD, R4 ; Add BCD to R4 decimally. R4: 0,00ddh

CLRC ; Clear carry
DADD.B &BCD, R4 ; Add BCD to R4 decimally.
R4: 0,00ddh
4.6.2.17 DEC

* DEC.W  Decrement destination
* DEC.B  Decrement destination

Syntax

DEC dst or DEC.W dst
DEC.B dst

Operation
dst – 1 → dst

Emulation

SUB #1,dst
SUB.B #1,dst

Description
The destination operand is decremented by one. The original contents are lost.

Status Bits

N:  Set if result is negative, reset if positive
Z:  Set if dst contained 1, reset otherwise
C:  Reset if dst contained 0, set otherwise
V:  Set if an arithmetic overflow occurs, otherwise reset.

Set if initial value of destination was 08000h, otherwise reset.

Set if initial value of destination was 080h, otherwise reset.

Mode Bits
OSCOFF, CPUOFF, and GiE are not affected.

Example

R10 is decremented by 1.

```assembly
DEC     R10                ; Decrement R10
; Move a block of 255 bytes from memory location starting with EDE to
; memory location starting with TONI. Tables should not overlap: start of
; destination address TONI must not be within the range EDE to EDE+0FEh
MOV     #EDE,R6
MOV     #510,R10
L$1    MOV     @R6+,TONI-EDE-1(R6)
DEC     R10
JNZ     L$1
```

Do not transfer tables using the routine above with the overlap shown in Figure 4-35.

![Figure 4-35. Decrement Overlap](image-url)
### 4.6.2.18 DECD

| * DECD[W] | Double-decrement destination |
| * DECD.B | Double-decrement destination |

**Syntax**

DECD dst or DECD.W dst  
DECD.B dst

**Operation**

dst – 2 → dst

**Emulation**

SUB #2,dst  
SUB.B #2,dst

**Description**

The destination operand is decremented by two. The original contents are lost.

**Status Bits**

- **N**: Set if result is negative, reset if positive
- **Z**: Set if dst contained 2, reset otherwise
- **C**: Reset if dst contained 0 or 1, set otherwise
- **V**: Set if an arithmetic overflow occurs, otherwise reset
  - Set if initial value of destination was 08001 or 08000h, otherwise reset
  - Set if initial value of destination was 081 or 080h, otherwise reset

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

R10 is decremented by 2.

```
DECD     R10              ; Decrement R10 by two
; Move a block of 255 bytes from memory location starting with EDE to
; memory location starting with TONI.
; Tables should not overlap: start of destination address TONI must not
; be within the range EDE to EDE+0FEh
MOV      #EDE,R6
MOV      #255,R10
L$1    MOV.B    @R6+,TONI-EDE-2(R6)
DECD     R10
JNZ      L$1
```

**Example**

Memory at location LEO is decremented by two.

```
DECD.B   LEO              ; Decrement MEM{LEO}
```

Decrement status byte STATUS by two

```
DECD.B   STATUS
```
4.6.2.19 DINT

* DINT Disable (general) interrupts
Syntax DINT
Operation 0 → GIE
or
(0FFF7h .AND. SR → SR / .NOT. src .AND. dst → dst)
Emulation BIC #8,SR
Description All interrupts are disabled.
The constant 08h is inverted and logically ANDed with the SR. The result is placed into the SR.
Status Bits Status bits are not affected.
Mode Bits GIE is reset. OSCOFF and CPUOFF are not affected.
Example The general interrupt enable (GIE) bit in the SR is cleared to allow a nondisrupted move of a 32-bit counter. This
ensures that the counter is not modified during the move by any interrupt.

DINT ; All interrupt events using the GIE bit are disabled
NOP
MOV COUNTHI,R5 ; Copy counter
MOV COUNTLO,R6
EINT ; All interrupt events using the GIE bit are enabled

Note

Disable interrupt

If any code sequence needs to be protected from interruption, DINT should be executed at least one
instruction before the beginning of the uninterruptible sequence, or it should be followed by a NOP
instruction.
4.6.2.20 EINT

* EINT
  Enable (general) interrupts

Syntax
  EINT

Operation
  1 → GIE
  or
  (0008h .OR. SR → SR / .src OR. dst → dst)

Emulation
  BIS #8,SR

Description
  All interrupts are enabled.
  The constant #08h and the SR are logically ORed. The result is placed into the SR.

Status Bits
  Status bits are not affected.

Mode Bits
  GIE is set. OSCOFF and CPUOFF are not affected.

Example
  The general interrupt enable (GIE) bit in the SR is set.

; Interrupt routine of ports P1.2 to P1.7
; P1IN is the address of the register where all port bits are read.
; P1IFG is the address of the register where all interrupt events are latched.
  PUSH.B &P1IN
  BIC.B @SP,&P1IFG ; Reset only accepted flags
  EINT ; Preset port 1 interrupt flags stored on stack
  ; other interrupts are allowed
  BIT #Mask,@SP
  JEQ MaskOK ; Flags are present identically to mask: jump
  ...
  ...
  MaskOK BIC #Mask,@SP
  ...
  ...
  INCd SP ; Housekeeping: inverse to PUSH instruction
  ; at the start of interrupt subroutine. Corrects
  ; the stack pointer.
  RETI

Note

Enable interrupt

The instruction following the enable interrupt instruction (EINT) is always executed, even if an interrupt service request is pending when the interrupts are enabled.
4.6.2.21 INC

* INC.[W] Increment destination
* INC.B Increment destination

Syntax

INC dst or INC.W dst
INC.B dst

Operation
dst + 1 → dst

Emulation

ADD #1,dst

Description
The destination operand is incremented by one. The original contents are lost.

Status Bits

N:    Set if result is negative, reset if positive
Z:    Set if dst contained 0FFFFh, reset otherwise
      Set if dst contained 0FFh, reset otherwise
C:    Set if dst contained 0FFFFh, reset otherwise
      Set if dst contained 0FFh, reset otherwise
V:    Set if dst contained 07FFFh, reset otherwise
      Set if dst contained 07Fh, reset otherwise

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example
The status byte, STATUS, of a process is incremented. When it is equal to 11, a branch to OVFL is taken.

```
INC.B    STATUS
CMP.B    #11,STATUS
JEQ      OVFL
```
**4.6.2.22 INCD**

<table>
<thead>
<tr>
<th>* INCD[.W]</th>
<th>Double-increment destination</th>
</tr>
</thead>
<tbody>
<tr>
<td>* INCD.B</td>
<td>Double-increment destination</td>
</tr>
</tbody>
</table>

**Syntax**

```
INCD dst or INCD.W dst
INCD.B dst
```

**Operation**

```
dst + 2 → dst
```

**Emulation**

```
ADD #2,dst
ADD.B #2,dst
```

**Description**
The destination operand is incremented by two. The original contents are lost.

**Status Bits**

- **N:** Set if result is negative, reset if positive
- **Z:** Set if dst contained 0FFFEh, reset otherwise
- **C:** Set if dst contained 0FFFEh or 0FFFFh, reset otherwise
  - Set if dst contained 0FEh or 0FFh, reset otherwise
- **V:** Set if dst contained 07FFEh or 07FFFh, reset otherwise
  - Set if dst contained 07Eh or 07Fh, reset otherwise

**Mode Bits**
OSC OFF, CPU OFF, and GIE are not affected.

**Example**
The item on the top of the stack (TOS) is removed without using a register.

```
PUSH    R5       ; R5 is the result of a calculation, which is stored  
               ; in the system stack
INCD    SP       ; Remove TOS by double-increment from stack 
               ; Do not use INCD.B, SP is a word-aligned register
RET
```

```
Example
The byte on the top of the stack is incremented by two.

INCD.B  0(SP)   ; Byte on TOS is increment by two
```
4.6.2.23 INV

* INV.[W]  Invert destination
* INV.B   Invert destination

Syntax  

INV dst or INV.W dst
INV.B dst

Operation  

.not.dst → dst

Emulation  

XOR #0FFFFh,dst
XOR.B #0FFh,dst

Description  

The destination operand is inverted. The original contents are lost.

Status Bits  

N: Set if result is negative, reset if positive
Z: Set if dst contained 0FFFFh, reset otherwise
C: Set if dst contained 0Fh, reset otherwise
V: Set if result is not zero, reset otherwise ( = .NOT. Zero)

Mode Bits  

OSCOFF, CPUOFF, and GIE are not affected.

Example  

Content of R5 is negated (2s complement).

MOV     #00AEh,R5    ; R5 = 000AEh
INV     R5           ; Invert R5, R5 = 0FF51h
INC     R5           ; R5 is now negated, R5 = 0FF52h

Example  

Content of memory byte LEO is negated.

MOV.B   #0AEh,LEO    ; MEM(LEO) = 0AEh
INV.B   LEO          ; Invert LEO, MEM(LEO) = 051h
INC.B   LEO          ; MEM(LEO) is negated, MEM(LEO) = 052h
4.6.2.24 JC, JHS

**JC**  
Jump if carry

**JHS**  
Jump if higher or same (unsigned)

**Syntax**

JC label  
JHS label

**Operation**

If \( C = 1 \):  
\( PC + (2 \times \text{Offset}) \rightarrow PC \)  
If \( C = 0 \): execute the following instruction

**Description**

The carry bit \( C \) in the SR is tested. If it is set, the signed 10-bit word offset contained in the instruction is multiplied by two, sign extended, and added to the 20-bit PC. This means a jump in the range –511 to +512 words relative to the PC in the full memory range. If \( C \) is reset, the instruction after the jump is executed.

JC is used for the test of the carry bit \( C \).  
JHS is used for the comparison of unsigned numbers.

**Status Bits**

Status bits are not affected

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

The state of the port 1 pin P1IN.1 bit defines the program flow.

```assembly
BIT.B #2, &P1IN ; Port 1, bit 1 set? Bit -> C
JC Label1 ; Yes, proceed at Label1
... ; No, continue
```

**Example**

If R5 ≥ R6 (unsigned), the program continues at Label2.

```assembly
CMP R6, R5 ; Is R5 >= R6? Info to C
JHS Label2 ; Yes, C = 1
... ; No, R5 < R6. Continue
```

**Example**

If R5 ≥ 12345h (unsigned operands), the program continues at Label2.

```assembly
CMFA #12345h, R5 ; Is R5 >= 12345h? Info to C
JHS Label2 ; Yes, 12344h < R5 <= F,FFFFh. C = 1
... ; No, R5 < 12345h. Continue
```
4.6.2.25 JEQ, JZ

**Syntax**

- `JEQ label`  
- `JZ label`

**Operation**

- If Z = 1: PC + (2 × Offset) → PC
- If Z = 0: execute following instruction

**Description**

The zero bit Z in the SR is tested. If it is set, the signed 10-bit word offset contained in the instruction is multiplied by two, sign extended, and added to the 20-bit PC. This means a jump in the range –511 to +512 words relative to the PC in the full memory range. If Z is reset, the instruction after the jump is executed.

- JZ is used for the test of the zero bit Z.
- JEQ is used for the comparison of operands.

**Status Bits**

Status bits are not affected.

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

The state of the P2IN.0 bit defines the program flow.

```
BIT.B   #1,&P2IN     ; Port 2, bit 0 reset?
JZ      Label1       ; Yes, proceed at Label1
...     ; No, set, continue
```

**Example**

If R5 = 15000h (20-bit data), the program continues at Label2.

```
CMPA    #15000h,R5   ; Is R5 = 15000h? Info to SR
JEQ     Label2       ; Yes, R5 = 15000h. Z = 1
...     ; No, R5 not equal 15000h. Continue
```

**Example**

R7 (20-bit counter) is incremented. If its content is zero, the program continues at Label4.

```
ADDA    #1,R7        ; Increment R7
JZ      Label4       ; Zero reached: Go to Label4
...     ; R7 not equal 0. Continue here.
```
4.6.2.26 JGE

JGE Jump if greater or equal (signed)

Syntax

JGE label

Operation

If (N xor V) = 0: PC + (2 × Offset) → PC
If (N xor V) = 1: execute following instruction

Description

The negative bit N and the overflow bit V in the SR are tested. If both bits are set or both are reset, the signed 10-bit word offset contained in the instruction is multiplied by two, sign extended, and added to the 20-bit PC. This means a jump in the range -511 to +512 words relative to the PC in full Memory range. If only one bit is set, the instruction after the jump is executed.

JGE is used for the comparison of signed operands: also for incorrect results due to overflow, the decision made by the JGE instruction is correct.

Note: JGE emulates the nonimplemented JP (jump if positive) instruction if used after the instructions AND, BIT, RRA, SXTX, and TST. These instructions clear the V bit.

Status Bits

Status bits are not affected.

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

If byte EDE (lower 64KB) contains positive data, go to Label1. Software can run in the full memory range.

TST.B &EDE ; Is EDE positive? V <= 0
JGE Label1 ; Yes, JGE emulates JP
... ; No, 80h <= EDE <= FFh

Example

If the content of R6 is greater than or equal to the memory pointed to by R7, the program continues a Label5. Signed data. Data and program in full memory range.

CMP @R7,R6 ; Is R6 >= @R7?
JGE Label5 ; Yes, go to Label5
... ; No, continue here

Example

If R5 ≥ 12345h (signed operands), the program continues at Label2. Program in full memory range.

CMPA #12345h,R5 ; Is R5 >= 12345h?
JGE Label2 ; Yes, 12344h < R5 <= 7FFFFh
... ; No, 80000h <= R5 < 12345h
4.6.2.27 JL

Jump if less (signed)

Syntax

JL label

Operation

If (N .xor. V) = 1: PC + (2 × Offset) → PC
If (N .xor. V) = 0: execute following instruction

Description

The negative bit N and the overflow bit V in the SR are tested. If only one is set, the signed 10-bit word offset
contained in the instruction is multiplied by two, sign extended, and added to the 20-bit PC. This means a jump in the
range –511 to +512 words relative to the PC in full memory range. If both bits N and V are set or both are reset, the
instruction after the jump is executed.

JL is used for the comparison of signed operands: also for incorrect results due to overflow, the decision made by the
JL instruction is correct.

Status Bits

Status bits are not affected.

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

If byte EDE contains a smaller, signed operand than byte TONI, continue at Label1. The address EDE is within PC ±
32 K.

```
CMP.B &TONI,EDE ; Is EDE < TONI
JL Label1 ; Yes
... ; No, TONI <= EDE
```

Example

If the signed content of R6 is less than the memory pointed to by R7 (20-bit address), the program continues at
Label5. Data and program in full memory range.

```
CMP @R7,R6 ; Is R6 < @R7?
JL Label5 ; Yes, go to Label5
... ; No, continue here
```

Example

If R5 < 12345h (signed operands), the program continues at Label2. Data and program in full memory range.

```
CMPA #12345h,R5 ; Is R5 < 12345h?
JL Label2 ; Yes, 80000h <= R5 < 12345h
... ; No, 12344h < R5 <= 7FFFPh
```
4.6.2.28 JMP

**JMP**

Jump unconditionally

**Syntax**

JMP label

**Operation**

PC + (2 × Offset) → PC

**Description**

The signed 10-bit word offset contained in the instruction is multiplied by two, sign extended, and added to the 20-bit PC. This means an unconditional jump in the range −511 to +512 words relative to the PC in the full memory. The JMP instruction may be used as a BR or BRA instruction within its limited range relative to the PC.

**Status Bits**

Status bits are not affected

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

The byte STATUS is set to 10. Then a jump to label MAINLOOP is made. Data in lower 64KB, program in full memory range.

```
MOV.B  #10, &STATUS ; Set STATUS to 10
JMP    MAINLOOP     ; Go to main loop
```

Example

The interrupt vector TAIV of Timer_A3 is read and used for the program flow. Program in full memory range, but interrupt handlers always starts in lower 64KB.

```
ADD    &TAIV, PC    ; Add Timer_A interrupt vector to PC
RETI   ; No Timer_A interrupt pending
JMP    IHCCR1       ; Timer block 1 caused interrupt
JMP    IHCCR2       ; Timer block 2 caused interrupt
RETI   ; No legal interrupt, return
```
4.6.2.29 JN

**JN** Jump if negative

**Syntax**

JN label

**Operation**

If \( N = 1 \): \( PC + (2 \times \text{Offset}) \rightarrow PC \)
If \( N = 0 \): execute following instruction

**Description**

The negative bit \( N \) in the SR is tested. If it is set, the signed 10-bit word offset contained in the instruction is multiplied by two, sign extended, and added to the 20-bit program PC. This means a jump in the range -511 to +512 words relative to the PC in the full memory range. If \( N \) is reset, the instruction after the jump is executed.

**Status Bits**

Status bits are not affected.

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

The byte COUNT is tested. If it is negative, program execution continues at Label0. Data in lower 64KB, program in full memory range.

```
TST.B &COUNT ; Is byte COUNT negative?
JN Label0 ; Yes, proceed at Label0
... ; COUNT >= 0
```

**Example**

R6 is subtracted from R5. If the result is negative, program continues at Label2. Program in full memory range.

```
SUB R6,R5 ; R5 - R6 -> R5
JN Label2 ; R5 is negative: R6 > R5 (N = 1)
... ; R5 >= 0. Continue here.
```

**Example**

R7 (20-bit counter) is decremented. If its content is below zero, the program continues at Label4. Program in full memory range.

```
SUBA #1,R7 ; Decrement R7
JN Label4 ; R7 < 0: Go to Label4
... ; R7 >= 0. Continue here.
```
4.6.2.30 JNC, JLO

JNC            Jump if no carry
JLO            Jump if lower (unsigned)

Syntax

JNC label
JLO label

Operation

If C = 0: PC + (2 × Offset) → PC
If C = 1: execute following instruction

Description

The carry bit C in the SR is tested. If it is reset, the signed 10-bit word offset contained in the instruction is multiplied by two, sign extended, and added to the 20-bit PC. This means a jump in the range –511 to +512 words relative to the PC in the full memory range. If C is set, the instruction after the jump is executed.

JNC is used for the test of the carry bit C.
JLO is used for the comparison of unsigned numbers.

Status Bits

Status bits are not affected.

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

If byte EDE < 15, the program continues at Label2. Unsigned data. Data in lower 64KB, program in full memory range.

```
CMP.B   #15,&EDE     ; Is EDE < 15? Info to C
JLO     Label2       ; Yes, EDE < 15. C = 0
...      ; No, EDE >= 15. Continue
```

Example

The word TONI is added to R5. If no carry occurs, continue at Label0. The address of TONI is within PC ± 32 K.

```
ADD     TONI,R5      ; TONI + R5 -> R5. Carry -> C
JNC     Label0       ; No carry
...      ; Carry = 1: continue here
```
4.6.2.31 JNZ, JNE

JNZ  Jump if not zero
JNE  Jump if not equal

Syntax  
JNZ label
JNE label

Operation  
If Z = 0: PC + (2 × Offset) → PC
If Z = 1: execute following instruction

Description  
The zero bit Z in the SR is tested. If it is reset, the signed 10-bit word offset contained in the instruction is multiplied by two, sign extended, and added to the 20-bit PC. This means a jump in the range –511 to +512 words relative to the PC in the full memory range. If Z is set, the instruction after the jump is executed.

JNZ is used for the test of the zero bit Z.
JNE is used for the comparison of operands.

Status Bits  
Status bits are not affected.

Mode Bits  
OSCOFF, CPUOFF, and GIE are not affected.

Example  
The byte STATUS is tested. If it is not zero, the program continues at Label3. The address of STATUS is within PC ± 32 K.

```
TST.B STATUS ; Is STATUS = 0?
JNZ Label3 ; No, proceed at Label3
... ; Yes, continue here
```

Example  
If word EDE ≠ 1500, the program continues at Label2. Data in lower 64KB, program in full memory range.

```
CMP #1500,&EDE ; Is EDE = 1500? Info to SR
JNE Label2 ; No, EDE not equal 1500.
... ; Yes, R5 = 1500. Continue
```

Example  
R7 (20-bit counter) is decremented. If its content is not zero, the program continues at Label4. Program in full memory range.

```
SUBA #1,R7 ; Decrement R7
JNZ Label4 ; Zero not reached: Go to Label4
... ; Yes, R7 = 0. Continue here.
```
4.6.2.32 MOV

**MOV.W**  
Move source word to destination word

**MOV.B**  
Move source byte to destination byte

**Syntax**  

<table>
<thead>
<tr>
<th>Syntax</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOV src,dst</td>
<td>Move source word to destination</td>
</tr>
<tr>
<td>MOV.B src,dst</td>
<td>Move source byte to destination</td>
</tr>
</tbody>
</table>

**Operation**  
src → dst

**Description**  
The source operand is copied to the destination. The source operand is not affected.

**Status Bits**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>N</td>
<td>Not affected</td>
</tr>
<tr>
<td>Z</td>
<td>Not affected</td>
</tr>
<tr>
<td>C</td>
<td>Not affected</td>
</tr>
<tr>
<td>V</td>
<td>Not affected</td>
</tr>
</tbody>
</table>

**Mode Bits**
OSCOFF, CPUOFF, and GIE are not affected.

**Example**  
Move a 16-bit constant 1800h to absolute address-word EDE (lower 64KB)

```assembly
MOV #01800h,&EDE            ; Move 1800h to EDE
```

**Example**  
The contents of table EDE (word data, 16-bit addresses) are copied to table TOM. The length of the tables is 030h words. Both tables reside in the lower 64KB.

```assembly
MOV     #EDE,R10               ; Prepare pointer (16-bit address)
Loop   MOV     @R10+,TOM-EDE-2(R10)   ; R10 points to both tables.
        ; R10+2
CMP     #EDE+60h,R10           ; End of table reached?
JLO     Loop                   ; Not yet
...                            ; Copy completed
```

**Example**  
The contents of table EDE (byte data, 16-bit addresses) are copied to table TOM. The length of the tables is 020h bytes. Both tables may reside in full memory range, but must be within R10 ± 32 K.

```assembly
MOVA   #EDE,R10               ; Prepare pointer (20-bit)
MOV     #20h,R9                ; Prepare counter
Loop   MOV.B   @R10+,TOM-EDE-1(R10)   ; R10 points to both tables.
        ; R10+1
DEC     R9                     ; Decrement counter
JNZ     Loop                   ; Not yet done
...                            ; Copy completed
```
### 4.6.2.33 NOP

<table>
<thead>
<tr>
<th><strong>NOP</strong></th>
<th>No operation</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Syntax</strong></td>
<td>NOP</td>
</tr>
<tr>
<td><strong>Operation</strong></td>
<td>None</td>
</tr>
<tr>
<td><strong>Emulation</strong></td>
<td>MOV #0, R3</td>
</tr>
<tr>
<td><strong>Description</strong></td>
<td>No operation is performed. The instruction may be used for the elimination of instructions during the software check or for defined waiting times.</td>
</tr>
<tr>
<td><strong>Status Bits</strong></td>
<td>Status bits are not affected.</td>
</tr>
</tbody>
</table>
**4.6.2.34 POP**

* POP.W  Pop word from stack to destination
* POP.B  Pop byte from stack to destination

<table>
<thead>
<tr>
<th>Syntax</th>
<th>Operation</th>
<th>Emulation</th>
<th>Description</th>
<th>Status Bits</th>
<th>Example</th>
</tr>
</thead>
<tbody>
<tr>
<td>POP dst</td>
<td>@SP → temp</td>
<td>MOV @SP+, dst</td>
<td>The stack location pointed to by the SP (TOS) is moved to the destination. The SP is incremented by two afterwards.</td>
<td></td>
<td>POP R7 ; Restore R7</td>
</tr>
<tr>
<td>POP.B dst</td>
<td>SP + 2 → SP</td>
<td>MOV.B @SP+, dst</td>
<td></td>
<td></td>
<td>POP SR ; Restore status register</td>
</tr>
<tr>
<td></td>
<td>temp → dst</td>
<td></td>
<td></td>
<td></td>
<td>POP.B LEO ; The low byte of the stack is moved to LEO.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>POP.B R7 ; The low byte of the stack is moved to R7,</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>; the high byte of R7 is 00h</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>POP.B 0(R7) ; The low byte of the stack is moved to the</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>; the byte which is pointed to by R7</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>; Example: R7 = 203h</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>; Mem(R7) = low byte of system stack</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>; Example: R7 = 20Ah</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>; Mem(R7) = low byte of system stack</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>POP SR ; Last word on stack moved to the SR</td>
</tr>
</tbody>
</table>

**Note**

**System stack pointer**

The system SP is always incremented by two, independent of the byte suffix.
4.6.2.35 PUSH

**PUSH[]**

- **Syntax**
  - `PUSH dst` or `PUSH.W dst`
  - `PUSH.B dst`

- **Operation**
  - `SP – 2 → SP`
  - `dst → @SP`

- **Description**
  - The 20-bit SP SP is decremented by two. The operand is then copied to the RAM word addressed by the SP. A pushed byte is stored in the low byte; the high byte is not affected.

- **Status Bits**
  - Status bits are not affected.

- **Mode Bits**
  - OSCOFF, CPUOFF, and GIE are not affected.

- **Example**
  - Save the two 16-bit registers R9 and R10 on the stack
    ```assembly
    PUSH R9       ; Save R9 and R10 XXXXh
    PUSH R10      ; YYYYh
    ```

  - Save the two bytes EDE and TONI on the stack. The addresses EDE and TONI are within PC ± 32 K.
    ```assembly
    PUSH.B EDE    ; Save EDE xxXXh
    PUSH.B TONI   ; Save TONI xxYYh
    ```
4.6.2.36 RET

Return from subroutine

Syntax

```
RET
```

Operation

```
@SP →PC.15:0 Saved PC to PC.15:0.  PC.19:16 ← 0
SP + 2 → SP
```

Description

The 16-bit return address (lower 64KB), pushed onto the stack by a CALL instruction is restored to the PC. The program continues at the address following the subroutine call. The four MSBs of the PC.19:16 are cleared.

Status Bits

Status bits are not affected.

PC.19:16: Cleared

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

Call a subroutine SUBR in the lower 64KB and return to the address in the lower 64KB after the CALL.

```
CALL    #SUBR     ; Call subroutine starting at SUBR
...               ; Return by RET to here
SUBR   PUSH    R14       ; Save R14 (16 bit data)
...               ; Subroutine code
POP     R14       ; Restore R14
RET               ; Return to lower 64KB
```

![Figure 4-36. Stack After a RET Instruction](image-url)
### 4.6.2.37 RETI

**Syntax**

```
RETI
```

**Operation**

- `@SP → SR.15:0` ➞ Restore saved SR with PC.19:16
- `SP + 2 → SP`
- `@SP → PC.15:0` ➞ Restore saved PC.15:0
- `SP + 2 → SP` ➞ Housekeeping

**Description**

The SR is restored to the value at the beginning of the interrupt service routine. This includes the four MSBs of the PC.19:16. The SP is incremented by two afterward.

The 20-bit PC is restored from PC.19:16 (from same stack location as the status bits) and PC.15:0. The 20-bit PC is restored to the value at the beginning of the interrupt service routine. The program continues at the address following the last executed instruction when the interrupt was granted. The SP is incremented by two afterward.

**Status Bits**

- **N**: Restored from stack
- **C**: Restored from stack
- **Z**: Restored from stack
- **V**: Restored from stack

**Mode Bits**

OSCOFF, CPUOFF, and GIE are restored from stack.

**Example**

Interrupt handler in the lower 64KB. A 20-bit return address is stored on the stack.

```
INTRPT  PUSHM.A   #2,R14    ; Save R14 and R13 (20-bit data)
        ...                 ; Interrupt handler code
POP.M.A #2,R14    ; Restore R13 and R14 (20-bit data)
RETI                ; Return to 20-bit address in full memory range
```
4.6.2.38 RLA

* RLA[W]  Rotate left arithmetically
* RLA.B  Rotate left arithmetically

Syntax
RLA dst or RLA.W dst
RLA.B dst

Operation
C ← MSB ← MSB-1 .... LSB+1 ← LSB ← 0

Emulation
ADD dst,dst
ADD.B dst,dst

Description
The destination operand is shifted left one position as shown in Figure 4-37. The MSB is shifted into the carry bit (C) and the LSB is filled with 0. The RLA instruction acts as a signed multiplication by 2.

An overflow occurs if dst ≥ 04000h and dst < 0C000h before operation is performed; the result has changed sign.

![Figure 4-37. Destination Operand—Arithmetic Shift Left](image)

An overflow occurs if dst ≥ 040h and dst < 0C0h before the operation is performed; the result has changed sign.

Status Bits
N:  Set if result is negative, reset if positive
Z:  Set if result is zero, reset otherwise
C:  Loaded from the MSB
V:  Set if an arithmetic overflow occurs; the initial value is 04000h ≤ dst < 0C000h, reset otherwise
V:  Set if an arithmetic overflow occurs; the initial value is 040h ≤ dst < 0C0h, reset otherwise

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
R7 is multiplied by 2.

RLA    R7    ; Shift left R7  (x 2)

Example
The low byte of R7 is multiplied by 4.

RLA.B  R7    ; Shift left low byte of R7  (x 2)
RLA.B  R7    ; Shift left low byte of R7  (x 4)

Note

RLA substitution

The assembler does not recognize the instructions:

RLA  @R5+  RLA.B  @R5+  RLA(.B)  @R5

They must be substituted by:

ADD  @R5+,-2(R5)  ADD.B  @R5+,-1(R5)  ADD(.B)  @R5
### 4.6.2.39 RLC

* RLC.W  
  Rotate left through carry

* RLC.B  
  Rotate left through carry

**Syntax**

- RLC `dst` or RLC.W `dst`
- RLC.B `dst`

**Operation**

\[
C \leftarrow \text{MSB} \leftarrow \text{MSB-1} \ldots \text{LSB+1} \leftarrow \text{LSB} \leftarrow C
\]

**Emulation**

- ADDC `dst,dst`

**Description**

The destination operand is shifted left one position as shown in Figure 4-38. The carry bit (C) is shifted into the LSB, and the MSB is shifted into the carry bit (C).

![Figure 4-38. Destination Operand—Carry Left Shift](image)

**Status Bits**

- **N**: Set if result is negative, reset if positive
- **Z**: Set if result is zero, reset otherwise
- **C**: Loaded from the MSB
- **V**: Set if an arithmetic overflow occurs; the initial value is \(04000h \leq \text{dst} < 0C000h\), reset otherwise

**Mode Bits**

OSCOFF, CPUFFF, and GIE are not affected.

**Example**

R5 is shifted left one position.

```
RLC R5 ; (R5 x 2) + C -> R5
```

**Example**

The input P1IN.1 information is shifted into the LSB of R5.

```
BIT.B #2, &P1IN ; Information -> Carry
RLC R5 ; Carry=P0in.1 -> LSB of R5
```

**Example**

The MEM(LEO) content is shifted left one position.

```
RLC.B LEO ; Mem(LEO) x 2 + C -> Mem(LEO)
```

**Note**

**RLA substitution**

The assembler does not recognize the instructions:

```
RLC @R5+  RLC.B @R5+  RLC(.B) @R5
```

They must be substituted by:

```
ADDC @R5+,-2(R5)  ADDC.B @R5+,-1(R5)  ADDC(.B) @R5
```
**4.6.2.40 RRA**

RRA[.W]  Rotate right arithmetically destination word  
RRA.B  Rotate right arithmetically destination byte  

**Syntax**  
`RRA.B dst` or `RRA.W dst`  

**Operation**  
MSB → MSB → MSB–1 → ... LSB+1 → LSB → C  

**Description**  
The destination operand is shifted right arithmetically by one bit position as shown in Figure 4-39. The MSB retains its value (sign). RRA operates equal to a signed division by 2. The MSB is retained and shifted into the MSB–1. The LSB+1 is shifted into the LSB. The previous LSB is shifted into the carry bit C.

**Status Bits**  
- **N**: Set if result is negative (MSB = 1), reset otherwise (MSB = 0)  
- **Z**: Set if result is zero, reset otherwise  
- **C**: Loaded from the LSB  
- **V**: Reset  

**Mode Bits**  
OSCOFF, CPUOFF, and GIE are not affected.

**Example**  
The signed 16-bit number in R5 is shifted arithmetically right one position.

```
RRA     R5             ; R5/2 -> R5
```

**Example**  
The signed RAM byte EDE is shifted arithmetically right one position.

```
RRA.B   EDE            ; EDE/2 -> EDE
```

![Figure 4-39. Rotate Right Arithmetically RRA.B and RRA.W](image-url)
4.6.2.41 RRC

RRC[W]  Rotate right through carry destination word
RRC.B   Rotate right through carry destination byte
Syntax  \[
RRC \text{ dst or } RRC.W \text{ dst} \\
RRC.B \text{ dst}
\]
Operation \( C \rightarrow \text{MSB} \rightarrow \text{MSB–1} \rightarrow \ldots \text{LSB+1} \rightarrow \text{LSB} \rightarrow C \)
Description The destination operand is shifted right by one bit position as shown in Figure 4-40. The carry bit \( C \) is shifted into the MSB and the LSB is shifted into the carry bit \( C \).
Status Bits
- \( N \): Set if result is negative (MSB = 1), reset otherwise (MSB = 0)
- \( Z \): Set if result is zero, reset otherwise
- \( C \): Loaded from the LSB
- \( V \): Reset
Mode Bits OSCOFF, CPUOFF, and GIE are not affected.
Example RAM word EDE is shifted right one bit position. The MSB is loaded with 1.

```
SETC            ; Prepare carry for MSB
RRC    EDE      ; EDE = EDE >> 1 + 8000h
```

Figure 4-40. Rotate Right Through Carry RRC.B and RRC.W
4.6.2.42 SBC

* SBC[W]  Subtract borrow (.NOT. carry) from destination
* SBC.B  Subtract borrow (.NOT. carry) from destination

Syntax

SBC dst or SBC.W dst
SBC.B dst

Operation

dst + 0FFFFh + C → dst

dst + 0FFh + C → dst

Emulation

SUBC #0,dst
SUBC.B #0,dst

Description

The carry bit (C) is added to the destination operand minus one. The previous contents of the destination are lost.

Status Bits

N:  Set if result is negative, reset if positive
Z:  Set if result is zero, reset otherwise
C:  Set if there is a carry from the MSB of the result, reset otherwise
     Set to 1 if no borrow, reset if borrow
V:  Set if an arithmetic overflow occurs, reset otherwise

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

The 16-bit counter pointed to by R13 is subtracted from a 32-bit counter pointed to by R12.

```
SUB     @R13,0(R12)    ; Subtract LSDs
SBC     2(R12)         ; Subtract carry from MSD
```

Example

The 8-bit counter pointed to by R13 is subtracted from a 16-bit counter pointed to by R12.

```
SUB.B   @R13,0(R12)    ; Subtract LSDs
SBC.B   1(R12)         ; Subtract carry from MSD
```

---

Note

**Borrow implementation**

The borrow is treated as a .NOT. carry:

<table>
<thead>
<tr>
<th>Borrow</th>
<th>Carry Bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Yes</td>
<td>0</td>
</tr>
<tr>
<td>No</td>
<td>1</td>
</tr>
</tbody>
</table>
4.6.2.43 SETC

* SETC  Set carry bit
Syntax     SETC
Operation  1 → C
Emulation  BIS #1,SR
Description The carry bit (C) is set.
Status Bits
N: Not affected
Z: Not affected
C: Set
V: Not affected

Mode Bits OSCOFF, CPUOFF, and GIE are not affected.

Example Emulation of the decimal subtraction:

Subtract R5 from R6 decimally.
Assume that R5 = 03987h and R6 = 04137h.

```
DSUB  ADD    #06666h,R5     ; Move content R5 from 0-9 to 6-0Fh
     ; R5 = 03987h + 06666h = 09FEDh
INV    R5             ; Invert this (result back to 0-9)
     ; R5 = .NOT. R5 = 06012h
SETC                  ; Prepare carry = 1
DADD   R5,R6          ; Emulate subtraction by addition of:
     ; (010000h - R5 - 1)
     ; R6 = R6 + R5 + 1
     ; R6 = 0150h
```
4.6.2.44 SETN

* SETN    Set negative bit
Syntax    SETN
Operation 1 → N
Emulation BIS #4,SR
Description The negative bit (N) is set.
Status Bits
N: Set
Z: Not affected
C: Not affected
V: Not affected
Mode Bits OSCOFF, CPUOFF, and GIE are not affected.
4.6.2.45 SETZ

* SETZ  Set zero bit
Syntax  SETZ
Operation  1 → N
Emulation  BIS #2,SR
Description  The zero bit (Z) is set.
Status Bits  
N:  Not affected
Z:  Set
C:  Not affected
V:  Not affected
Mode Bits  OSCOFF, CPUOFF, and GIE are not affected.
### 4.6.2.46 SUB

**SUB[W]**  Subtract source word from destination word

**SUB.B**  Subtract source byte from destination byte

**Syntax**

- `SUB src,dst` or `SUB.W src,dst`
- `SUB.B src,dst`

**Operation**

- `(.not.src) + 1 + dst → dst` or `dst – src → dst`

**Description**

The source operand is subtracted from the destination operand. This is made by adding the 1s complement of the source + 1 to the destination. The source operand is not affected, the result is written to the destination operand.

**Status Bits**

- **N:** Set if result is negative (`src > dst`), reset if positive (`src ≤ dst`)
- **Z:** Set if result is zero (`src = dst`), reset otherwise (`src ≠ dst`)
- **C:** Set if there is a carry from the MSB, reset otherwise
- **V:** Set if the subtraction of a negative source operand from a positive destination operand delivers a negative result, or if the subtraction of a positive source operand from a negative destination operand delivers a positive result, reset otherwise (no overflow)

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

A 16-bit constant 7654h is subtracted from RAM word EDE.

```
SUB     #7654h,&EDE     ; Subtract 7654h from EDE
```

**Example**

A table word pointed to by R5 (20-bit address) is subtracted from R7. Afterwards, if R7 contains zero, jump to label TONI. R5 is then auto-incremented by 2. R7.19:16 = 0.

```
SUB     @R5+,R7        ; Subtract table number from R7. R5 + 2
JZ      TONI           ; R7 = @R5 (before subtraction)
...
; R7 <> @R5 (before subtraction)
```

**Example**

Byte CNT is subtracted from byte R12 points to. The address of CNT is within PC ± 32K. The address R12 points to is in full memory range.

```
SUB.B   CNT,0(R12)     ; Subtract CNT from @R12
```
### 4.6.2.47 SUBC

**SUBC[W]**  Subtract source word with carry from destination word  
**SUBC.B**  Subtract source byte with carry from destination byte  

**Syntax**  
- SUBC src,dst  
- SUBC.W src,dst  
- SUBC.B src,dst  

**Operation**  
(\neg src) + C + dst → dst  or  dst – (src – 1) + C → dst  

**Description**  
The source operand is subtracted from the destination operand. This is done by adding the 1s complement of the source + carry to the destination. The source operand is not affected, the result is written to the destination operand.  
Used for 32, 48, and 64-bit operands.  

**Status Bits**  
- **N:** Set if result is negative (MSB = 1), reset if positive (MSB = 0)  
- **Z:** Set if result is zero, reset otherwise  
- **C:** Set if there is a carry from the MSB, reset otherwise  
- **V:** Set if the subtraction of a negative source operand from a positive destination operand delivers a negative result, or if the subtraction of a positive source operand from a negative destination operand delivers a positive result, reset otherwise (no overflow)  

**Mode Bits**  
OSCOFF, CPUOFF, and GIE are not affected.  

**Example**  
A 16-bit constant 7654h is subtracted from R5 with the carry from the previous instruction. R5.19:16 = 0  

```
SUBC.W #7654h,R5 ; Subtract 7654h + C from R5
```

**Example**  
A 48-bit number (3 words) pointed to by R5 (20-bit address) is subtracted from a 48-bit counter in RAM, pointed to by R7. R5 points to the next 48-bit number afterwards. The address R7 points to is in full memory range.  

```
SUB @R5+,0(R7) ; Subtract LSBs. R5 + 2  
SUBC @R5+,2(R7) ; Subtract MIDs with C. R5 + 2  
SUBC @R5+,4(R7) ; Subtract MSBs with C. R5 + 2
```

**Example**  
Byte CNT is subtracted from the byte, R12 points to. The carry of the previous instruction is used. The address of CNT is in lower 64KB.  

```
SUBC.B &CNT,0(R12) ; Subtract byte CNT from @R12
```
4.6.2.48 SWPB

SWPB  Swap bytes
Syntax  SWPB dst
Operation  dst.15:8 ↔ dst.7:0
Description  The high and the low byte of the operand are exchanged. PC.19:16 bits are cleared in register mode.
Status Bits  Status bits are not affected
Mode Bits  OSCOFF, CPUOFF, and GIE are not affected.
Example  Exchange the bytes of RAM word EDE (lower 64KB)

MOV    #1234h,&EDE      ; 1234h -> EDE
SWPB   &EDE             ; 3412h -> EDE

Before SWPB

<table>
<thead>
<tr>
<th>15</th>
<th>8</th>
<th>7</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>High Byte</td>
<td></td>
<td>Low Byte</td>
</tr>
</tbody>
</table>

After SWPB

<table>
<thead>
<tr>
<th>15</th>
<th>8</th>
<th>7</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Low Byte</td>
<td></td>
<td>High Byte</td>
</tr>
</tbody>
</table>

Figure 4-41. Swap Bytes in Memory

Before SWPB

<table>
<thead>
<tr>
<th>19</th>
<th>16</th>
<th>15</th>
<th>8</th>
<th>7</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>x</td>
<td></td>
<td></td>
<td>High Byte</td>
<td></td>
<td>Low Byte</td>
</tr>
</tbody>
</table>

After SWPB

<table>
<thead>
<tr>
<th>19</th>
<th>16</th>
<th>15</th>
<th>8</th>
<th>7</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td></td>
<td>Low Byte</td>
<td></td>
<td>High Byte</td>
</tr>
</tbody>
</table>

Figure 4-42. Swap Bytes in a Register
### 4.6.2.49 SXT

**SXT**
- Extend sign

**Syntax**
- `SXT dst`

**Operation**
- `dst.7 → dst.15:8, dst.7 → dst.19:8` (register mode)

**Description**
- Register mode: the sign of the low byte of the operand is extended into the bits `Rdst.19:8`.
  - `Rdst.7 = 0`: `Rdst.19:8 = 000h` afterwards
  - `Rdst.7 = 1`: `Rdst.19:8 = FFFh` afterwards
- Other modes: the sign of the low byte of the operand is extended into the high byte.
  - `dst.7 = 0`: high byte = `00h` afterwards
  - `dst.7 = 1`: high byte = `FFh` afterwards

**Status Bits**
- **N**: Set if result is negative, reset otherwise
- **Z**: Set if result is zero, reset otherwise
- **C**: Set if result is not zero, reset otherwise (`C = .not.Z`)
- **V**: Reset

**Mode Bits**
- `OSCOFF, CPUOFF, and GIE are not affected.`

**Example**
The signed 8-bit data in EDE (lower 64KB) is sign extended and added to the 16-bit signed data in R7.

```
    MOV.B &EDE,R5 ; EDE -> R5. 00XXh
    SXT R5 ; Sign extend low byte to R5.19:8
    ADD R5,R7 ; Add signed 16-bit values
```

**Example**
The signed 8-bit data in EDE (PC +32 K) is sign extended and added to the 20-bit data in R7.

```
    MOV.B EDE,R5 ; EDE -> R5. 00XXh
    SXT R5 ; Sign extend low byte to R5.19:8
    ADDA R5,R7 ; Add signed 20-bit values
```
4.6.2.50 TST

* TST.W  Test destination
* TST.B  Test destination

Syntax

TST dst or TST.W dst
TST.B dst

Operation
dst + 0FFFFh + 1
dst + 0FFh + 1

Emulation

CMP #0,dst
CMP.B #0,dst

Description
The destination operand is compared with zero. The status bits are set according to the result. The destination is not affected.

Status Bits

N:  Set if destination is negative, reset if positive
Z:  Set if destination contains zero, reset otherwise
C:  Set
V:  Reset

Mode Bits

OSCOFF, CPUOFF, and GiE are not affected.

Example
R7 is tested. If it is negative, continue at R7NEG; if it is positive but not zero, continue at R7POS.

TST     R7       ; Test R7
JN      R7NEG    ; R7 is negative
JZ      R7ZERO   ; R7 is zero
R7POS    ......           ; R7 is positive but not zero
R7NEG    ......           ; R7 is negative
R7ZERO   ......           ; R7 is zero

Example
The low byte of R7 is tested. If it is negative, continue at R7NEG; if it is positive but not zero, continue at R7POS.

TST.B   R7       ; Test low byte of R7
JN      R7NEG    ; Low byte of R7 is negative
JZ      R7ZERO   ; Low byte of R7 is zero
R7POS    ......           ; Low byte of R7 is positive but not zero
R7NEG    ......           ; Low byte of R7 is negative
R7ZERO   ......           ; Low byte of R7 is zero
4.6.2.51 XOR

**XOR[W]**  Exclusive OR source word with destination word
**XOR.B**  Exclusive OR source byte with destination byte

**Syntax**

- `XOR src,dst` or `XOR.W src,dst`  
- `XOR.B src,dst`

**Operation**  

`src .xor. dst → dst`

**Description**
The source and destination operands are exclusively ORed. The result is placed into the destination. The source operand is not affected. The previous content of the destination is lost.

**Status Bits**

- N: Set if result is negative (MSB = 1), reset if positive (MSB = 0)
- Z: Set if result is zero, reset otherwise
- C: Set if result is not zero, reset otherwise (C = .not. Z)
- V: Set if both operands are negative before execution, reset otherwise

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

Toggle bits in word CNTR (16-bit data) with information (bit = 1) in address-word TONI. Both operands are located in lower 64KB.

```
XOR &TONI,&CNTR ; Toggle bits in CNTR
```

Example A table word pointed to by R5 (20-bit address) is used to toggle bits in R6. R6.19:16 = 0.

```
XOR @R5,R6 ; Toggle bits in R6
```

Example Reset to zero those bits in the low byte of R7 that are different from the bits in byte EDE. R7.19:8 = 0. The address of EDE is within PC ± 32 K.

```
XOR.B EDE,R7 ; Set different bits to 1 in R7.
INV.B R7 ; Invert low byte of R7, high byte is 0h
```


4.6.3 MSP430X Extended Instructions

The MSP430X extended instructions give the MSP430X CPU full access to its 20-bit address space. MSP430X instructions require an additional word of op-code called the extension word. All addresses, indexes, and immediate numbers have 20-bit values when preceded by the extension word. The MSP430X extended instructions are described in the following sections. For MSP430X instructions that do not require the extension word, it is noted in the instruction description.

See Section 4.6.2 for standard MSP430 instructions and Section 4.6.4 for MSP430X address instructions.
4.6.3.1 ADCX

* ADCX.A  Add carry to destination address-word
* ADCX.[W] Add carry to destination word
* ADCX.B  Add carry to destination byte

Syntax
ADCX.A dst
ADCX dst or ADCX.W dst
ADCX.B dst

Operation  dst + C → dst

Emulation
ADDCX.A #0,dst
ADDCX #0,dst
ADDCX.B #0,dst

Description  The carry bit (C) is added to the destination operand. The previous contents of the destination are lost.

Status Bits
N:  Set if result is negative (MSB = 1), reset if positive (MSB = 0)
Z:  Set if result is zero, reset otherwise
C:  Set if there is a carry from the MSB of the result, reset otherwise
V:  Set if the result of two positive operands is negative, or if the result of two negative numbers is positive, reset otherwise

Mode Bits  OSCOFF, CPUOFF, and GIE are not affected.

Example
The 40-bit counter, pointed to by R12 and R13, is incremented.

IN CX.A  @R12   ; Increment lower 20 bits
AD CX.A  @R13   ; Add carry to upper 20 bits
### 4.6.3.2 ADDX

- **ADDX.A** Add source address-word to destination address-word
- **ADDX.[W]** Add source word to destination word
- **ADDX.B** Add source byte to destination byte

#### Syntax

- ADDX.A src,dst
- ADDX src,dst or ADDX.W src,dst
- ADDX.B src,dst

#### Operation

src + dst → dst

#### Description

The source operand is added to the destination operand. The previous contents of the destination are lost. Both operands can be located in the full address space.

#### Status Bits

- **N**: Set if result is negative (MSB = 1), reset if positive (MSB = 0)
- **Z**: Set if result is zero, reset otherwise
- **C**: Set if there is a carry from the MSB of the result, reset otherwise
- **V**: Set if the result of two positive operands is negative, or if the result of two negative numbers is positive, reset otherwise

#### Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

#### Example

Ten is added to the 20-bit pointer CNTR located in two words CNTR (LSBs) and CNTR+2 (MSBs).

```
ADDX.A   #10,CNTR    ; Add 10 to 20-bit pointer
```

Example

A table word (16-bit) pointed to by R5 (20-bit address) is added to R6. The jump to label TONI is performed on a carry.

```
ADDX.W   @R5,R6      ; Add table word to R6
JC       TONI        ; Jump if carry
...      ; No carry
```

Example

A table byte pointed to by R5 (20-bit address) is added to R6. The jump to label TONI is performed if no carry occurs. The table pointer is auto-incremented by 1.

```
ADDX.B   @R5+,R6     ; Add table byte to R6. R5 + 1. R6: 000xxh
JNC      TONI        ; Jump if no carry
...      ; Carry occurred
```

Note: Use ADDA for the following two cases for better code density and execution.

```
ADDX.A   Rsrc,Rdst
ADDX.A   #imm20,Rdst
```
4.6.3.3 ADDCX

ADDCX.A  Add source address-word and carry to destination address-word
ADDCX.[W] Add source word and carry to destination word
ADDCX.B  Add source byte and carry to destination byte

Syntax

ADDCX.A src,dst  
ADDCX src,dst  or ADDCX.W src,dst  
ADDCX.B src,dst

Operation  src + dst + C → dst

Description  The source operand and the carry bit C are added to the destination operand. The previous contents of the destination are lost. Both operands may be located in the full address space.

Status Bits

N:  Set if result is negative (MSB = 1), reset if positive (MSB = 0)
Z:  Set if result is zero, reset otherwise
C:  Set if there is a carry from the MSB of the result, reset otherwise
V:  Set if the result of two positive operands is negative, or if the result of two negative numbers is positive, reset otherwise

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example  Constant 15 and the carry of the previous instruction are added to the 20-bit counter CNTR located in two words.

```
ADDCX.A  #15,&CNTR  ; Add 15 + C to 20-bit CNTR
```

Example  A table word pointed to by R5 (20-bit address) and the carry C are added to R6. The jump to label TONI is performed on a carry.

```
ADDCX.W  @R5,R6  ; Add table word + C to R6
JC  TONI  ; Jump if carry
...  ; No carry
```

Example  A table byte pointed to by R5 (20-bit address) and the carry bit C are added to R6. The jump to label TONI is performed if no carry occurs. The table pointer is auto-incremented by 1.

```
ADDCX.B  @R5+,R6  ; Add table byte + C to R6. R5 + 1
JNC  TONI  ; Jump if no carry
...  ; Carry occurred
```
4.6.3.4 ANDX

**ANDX.A**  Logical AND of source address-word with destination address-word

**ANDX.[W]** Logical AND of source word with destination word

**ANDX.B**  Logical AND of source byte with destination byte

**Syntax**

```
ANDX.A src,dst
ANDX src,dst or ANDX.W src,dst
ANDX.B src,dst
```

**Operation**

```
src .and. dst → dst
```

**Description** The source operand and the destination operand are logically ANDed. The result is placed into the destination. The source operand is not affected. Both operands may be located in the full address space.

**Status Bits**

- **N:** Set if result is negative (MSB = 1), reset if positive (MSB = 0)
- **Z:** Set if result is zero, reset otherwise
- **C:** Set if the result is not zero, reset otherwise. C = (.not. Z)
- **V:** Reset

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

```
MOVA    #AAA55h,R5  ; Load 20-bit mask to R5
ANDX.A  R5,TOM     ; TOM .and. R5 -> TOM
JZ      TONI        ; Jump if result 0
...
```

or shorter:

```
ANDX.A  #AAA55h,TOM ; TOM .and. AAA55h -> TOM
JZ      TONI        ; Jump if result 0
```

**Example**

```
A table byte pointed to by R5 (20-bit address) is logically ANDed with R6. R6.19:8 = 0. The table pointer is auto-incremented by 1.

ANDX.B  @R5+,R6     ; AND table byte with R6. R5 + 1
```
4.6.3.5 BICX

BICX.A Clear bits set in source address-word in destination address-word
BICX.[W] Clear bits set in source word in destination word
BICX.B Clear bits set in source byte in destination byte

Syntax

BICX.A src,dst
BICX src,dst or BICX.[W] src,dst
BICX.B src,dst

Operation

(.not. src) .and. dst → dst

Description
The inverted source operand and the destination operand are logically ANDed. The result is placed into the destination. The source operand is not affected. Both operands may be located in the full address space.

Status Bits

N: Not affected
Z: Not affected
C: Not affected
V: Not affected

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
The bits 19:15 of R5 (20-bit data) are cleared.

BICX.A #0F8000h,R5 ; Clear R5.19:15 bits

Example
A table word pointed to by R5 (20-bit address) is used to clear bits in R7. R7.19:16 = 0.

BICX.W @R5,R7 ; Clear bits in R7

Example
A table byte pointed to by R5 (20-bit address) is used to clear bits in output Port1.

BICX.B @R5,&P1OUT ; Clear I/O port P1 bits
4.6.3.6 BISX

**BISX.A**  Set bits set in source address-word in destination address-word

**BISX.[W]**  Set bits set in source word in destination word

**BISX.B**  Set bits set in source byte in destination byte

**Syntax**

BISX.A src,dst

BISX src,dst or BISX.W src,dst

BISX.B src,dst

**Operation**

src .or. dst → dst

**Description**  The source operand and the destination operand are logically ORed. The result is placed into the destination. The source operand is not affected. Both operands may be located in the full address space.

**Status Bits**

N:  Not affected

Z:  Not affected

C:  Not affected

V:  Not affected

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

Bits 16 and 15 of R5 (20-bit data) are set to one.

BISX.A #018000h,R5 ; Set R5.16:15 bits

**Example**

A table word pointed to by R5 (20-bit address) is used to set bits in R7.

BISX.W @R5,R7 ; Set bits in R7

**Example**

A table byte pointed to by R5 (20-bit address) is used to set bits in output Port1.

BISX.B @R5,&P1OUT ; Set I/O port P1 bits
4.6.3.7 BITX

**BITX.A**  Test bits set in source address-word in destination address-word  
**BITX.[W]**  Test bits set in source word in destination word  
**BITX.B**  Test bits set in source byte in destination byte  

**Syntax**

BITX.A src,dst  
BITX src,dst or BITX.W src,dst  
BITX.B src,dst

**Operation**  
src .and. dst → dst

**Description**  
The source operand and the destination operand are logically ANDed. The result affects only the status bits. Both operands may be located in the full address space.

**Status Bits**

N:  Set if result is negative (MSB = 1), reset if positive (MSB = 0)  
Z:  Set if result is zero, reset otherwise  
C:  Set if the result is not zero, reset otherwise. C = (.not. Z)  
V:  Reset

**Mode Bits**  
OSCOFF, CPUOFF, and GIE are not affected.

**Example**  
Test if bit 16 or 15 of R5 (20-bit data) is set. Jump to label TONI if so.

```
BITX.A   #018000h,R5   ; Test R5.16:15 bits  
JNZ      TONI          ; At least one bit is set  
...      ; Both are reset
```

**Example**  
A table word pointed to by R5 (20-bit address) is used to test bits in R7. Jump to label TONI if at least one bit is set.

```
BITX.W   @R5,R7        ; Test bits in R7: C = .not.Z  
JC       TONI          ; At least one is set  
...      ; Both are reset
```

**Example**  
A table byte pointed to by R5 (20-bit address) is used to test bits in input Port1. Jump to label TONI if no bit is set. The next table byte is addressed.

```
BITX.B   @R5+,&P1IN    ; Test input P1 bits. R5 + 1  
JNC      TONI          ; No corresponding input bit is set  
...      ; At least one bit is set
```
4.6.3.8 CLRX

* CLRX.A Clear destination address-word
* CLRX.[W] Clear destination word
* CLRX.B Clear destination byte

Syntax
CLRX.A dst
CLRX dst or CLRX.W dst
CLRX.B dst

Operation
0 → dst

Emulation
MOVX.A #0,dst
MOVX #0,dst
MOVX.B #0,dst

Description
The destination operand is cleared.

Status Bits
Status bits are not affected.

Example
RAM address-word TONI is cleared.

CLRX.A TONI ; 0 -> TONI
4.6.3.9 CMPX

**CMPX.A**  Compare source address-word and destination address-word
**CMPX.[W]** Compare source word and destination word
**CMPX.B**  Compare source byte and destination byte

**Syntax**
- CMPX.A src,dst
- CMPX src,dst or CMPX.W src,dst
- CMPX.B src,dst

**Operation**
\((\text{not } \text{src}) + 1 + \text{dst} \text{ or } \text{dst} - \text{src}\)

**Description**
The source operand is subtracted from the destination operand by adding the 1s complement of the source + 1 to the destination. The result affects only the status bits. Both operands may be located in the full address space.

**Status Bits**
- **N**: Set if result is negative (src > dst), reset if positive (src ≤ dst)
- **Z**: Set if result is zero (src = dst), reset otherwise (src ≠ dst)
- **C**: Set if there is a carry from the MSB, reset otherwise
- **V**: Set if the subtraction of a negative source operand from a positive destination operand delivers a negative result, or if the subtraction of a positive source operand from a negative destination operand delivers a positive result, reset otherwise (no overflow)

**Mode Bits**
OSCOFF, CPUOFF, and GIE are not affected.

**Example**
Compare EDE with a 20-bit constant 18000h. Jump to label TONI if EDE equals the constant.

```
CMPX.A #018000h,EDE ; Compare EDE with 18000h
JEQ TONI ; EDE contains 18000h
... ; Not equal
```

**Example**
A table word pointed to by R5 (20-bit address) is compared with R7. Jump to label TONI if R7 contains a lower, signed, 16-bit number.

```
CMPX.W @R5,R7 ; Compare two signed numbers
JL TONI ; R7 < @R5
... ; R7 >= @R5
```

**Example**
A table byte pointed to by R5 (20-bit address) is compared to the input in I/O Port1. Jump to label TONI if the values are equal. The next table byte is addressed.

```
CMPX.B @R5+,&P1IN ; Compare P1 bits with table. R5 + 1
JEQ TONI ; Equal contents
... ; Not equal
```

**Note:** Use CMPA for the following two cases for better density and execution.

```
CMPA Rsrc,Rdst
CMPA *imm20,Rdst
```
4.6.3.10 DADCX

* DADCX.A Add carry decimally to destination address-word
* DADCX.[W] Add carry decimally to destination word
* DADCX.B Add carry decimally to destination byte

Syntax

DADCX.A dst
DADCX.W dst
DADCX.B dst

Operation

dst + C → dst (decimally)

Emulation

DADDX.A #0,dst
DADDX.W #0,dst
DADDX.B #0,dst

Description

The carry bit (C) is added decimally to the destination.

Status Bits

N: Set if MSB of result is 1 (address-word > 79999h, word > 7999h, byte > 79h), reset if MSB is 0
Z: Set if result is zero, reset otherwise
C: Set if the BCD result is too large (address-word > 99999h, word > 9999h, byte > 99h), reset otherwise
V: Undefined

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

The 40-bit counter, pointed to by R12 and R13, is incremented decimally.

DADDX.A #1,0(R12) ; Increment lower 20 bits
DADCX.A 0(R13)   ; Add carry to upper 20 bits
4.6.3.11 DADDX

DADDX.A  Add source address-word and carry decimally to destination address-word
DADDX.W   Add source word and carry decimally to destination word
DADDX.B   Add source byte and carry decimally to destination byte

Syntax

DADDX.A src,dst
DADDX src,dst or DADDX.W src,dst
DADDX.B src,dst

Operation
src + dst + C → dst (decimally)

Description
The source operand and the destination operand are treated as two (.B), four (.W), or five (.A) binary coded decimals (BCD) with positive signs. The source operand and the carry bit C are added decimally to the destination operand. The source operand is not affected. The previous contents of the destination are lost. The result is not defined for non-BCD numbers. Both operands may be located in the full address space.

Status Bits

N: Set if MSB of result is 1 (address-word > 79999h, word > 7999h, byte > 79h), reset if MSB is 0.
Z: Set if result is zero, reset otherwise
C: Set if the BCD result is too large (address-word > 99999h, word > 9999h, byte > 99h), reset otherwise
V: Undefined

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
Decimal 10 is added to the 20-bit BCD counter DECCNTR located in two words.

DADDX.A  #10h,&DECCNTR     ; Add 10 to 20-bit BCD counter

Example
The eight-digit BCD number contained in 20-bit addresses BCD and BCD+2 is added decimally to an eight-digit BCD number contained in R4 and R5 (BCD+2 and R5 contain the MSDs).

CLRC                        ; Clear carry
DADDX.W   BCD,R4            ; Add LSDs
DADDX.W   BCD+2,R5           ; Add MSDs with carry
JC        OVERFLOW           ; Result >99999999: go to error routine
...                         ; Result ok

Example
The two-digit BCD number contained in 20-bit address BCD is added decimally to a two-digit BCD number contained in R4.

CLRC                        ; Clear carry
DADDX.B   BCD,R4            ; Add BCD to R4 decimally.
                       ; R4: 000ddh
4.6.3.12 DECX

* DECX.A  Decrement destination address-word
* DECX.[W] Decrement destination word
* DECX.B  Decrement destination byte

Syntax

DECX.A dst
DECX dst or DECX.W dst
DECX.B dst

Operation
dst – 1 → dst

Emulation

SUBX.A #1,dst
SUBX #1,dst
SUBX.B #1,dst

Description
The destination operand is decremented by one. The original contents are lost.

Status Bits

N:  Set if result is negative, reset if positive
Z:  Set if dst contained 1, reset otherwise
C:  Reset if dst contained 0, set otherwise
V:  Set if an arithmetic overflow occurs, otherwise reset

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

RAM address-word TONI is decremented by one.

DECX.A  TONI     ; Decrement TONI
4.6.3.13 DECDX

* DECDX.A  Double-decrement destination address-word
* DECDX.[W] Double-decrement destination word
* DECDX.B  Double-decrement destination byte

Syntax

DECDX.A dst
DECDX dst or DECDX.W dst
DECDX.B dst

Operation

dst – 2 → dst

Emulation

SUBX.A #2,dst
SUBX #2,dst
SUBX.B #2,dst

Description

The destination operand is decremented by two. The original contents are lost.

Status Bits

N: Set if result is negative, reset if positive
Z: Set if dst contained 2, reset otherwise
C: Reset if dst contained 0 or 1, set otherwise
V: Set if an arithmetic overflow occurs, otherwise reset

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

RAM address-word TONI is decremented by two.

DECDX.A TONI ; Decrement TONI
### 4.6.3.14 INCX

* **INCX.A**
  Increment destination address-word

* **INCX.[W]**
  Increment destination word

* **INCX.B**
  Increment destination byte

#### Syntax

- `INCX.A dst`
- `INCX dst` or `INCX.W dst`
- `INCX.B dst`

#### Operation

`dst + 1 → dst`

#### Emulation

- `ADDX.A #1,dst`
- `ADDX #1,dst`
- `ADDX.B #1,dst`

#### Description

The destination operand is incremented by one. The original contents are lost.

#### Status Bits

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>N</td>
<td>Set if result is negative, reset if positive</td>
</tr>
<tr>
<td>Z</td>
<td>Set if <code>dst</code> contained <code>0FFFFFh</code>, reset otherwise</td>
</tr>
<tr>
<td></td>
<td>Set if <code>dst</code> contained <code>0FFFFh</code>, reset otherwise</td>
</tr>
<tr>
<td></td>
<td>Set if <code>dst</code> contained <code>0FFh</code>, reset otherwise</td>
</tr>
<tr>
<td>C</td>
<td>Set if <code>dst</code> contained <code>0FFFFFh</code>, reset otherwise</td>
</tr>
<tr>
<td></td>
<td>Set if <code>dst</code> contained <code>0FFFFh</code>, reset otherwise</td>
</tr>
<tr>
<td></td>
<td>Set if <code>dst</code> contained <code>0FFh</code>, reset otherwise</td>
</tr>
<tr>
<td>V</td>
<td>Set if <code>dst</code> contained <code>07FFFh</code>, reset otherwise</td>
</tr>
<tr>
<td></td>
<td>Set if <code>dst</code> contained <code>07FFFh</code>, reset otherwise</td>
</tr>
<tr>
<td></td>
<td>Set if <code>dst</code> contained <code>07Fh</code>, reset otherwise</td>
</tr>
</tbody>
</table>

#### Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

#### Example

RAM address-word TONI is incremented by one.

```
INCX.A TONI  ; Increment TONI (20-bits)
```
4.6.3.15 **INCDX**

* INCDX.A  Double-increment destination address-word  
* INCDX. [W] Double-increment destination word  
* INCDX. B  Double-increment destination byte

**Syntax**

\[
\text{INCDX.A ~ dst} \\
\text{INCDX ~ dst or INCDX.W ~ dst} \\
\text{INCDX.B ~ dst}
\]

**Operation**

\[\text{dst + 2 \rightarrow dst}\]

**Emulation**

\[
\text{ADDX.A ~ #2, dst} \\
\text{ADDX ~ #2, dst} \\
\text{ADDX.B ~ #2, dst}
\]

**Description**

The destination operand is incremented by two. The original contents are lost.

**Status Bits**

- **N**: Set if result is negative, reset if positive
- **Z**: Set if dst contained 0xFFFFEh, reset otherwise  
  - Set if dst contained 0xFFF Eh, reset otherwise  
  - Set if dst contained 0xFEh, reset otherwise
- **C**: Set if dst contained 0xFFFFEh or 0xFFFFFFFFh, reset otherwise  
  - Set if dst contained 0xFFFFEh or 0xFFFFh, reset otherwise  
  - Set if dst contained 0xFEh or 0xFFh, reset otherwise
- **V**: Set if dst contained 07FFFFFFEh or 07FFFFFFFh, reset otherwise  
  - Set if dst contained 07FFFFFFEh or 07FFFFFFh, reset otherwise  
  - Set if dst contained 07FEh or 07FFh, reset otherwise

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

RAM byte LEO is incremented by two; PC points to upper memory.

\[
\text{INCDX.B ~ LEO \ ; Increment LEO by two}
\]
4.6.3.16 INVX

* INVX.A Invert destination
* INVX.[W] Invert destination
* INVX.B Invert destination

Syntax

INVX.A dst
INVX dst or INVX.W dst
INVX.B dst

Operation

.NOT. dst → dst

Emulation

XORX.A #0FFFFFh,dst
XORX #0FFFFFh,dst
XORX.B #0FFh,dst

Description

The destination operand is inverted. The original contents are lost.

Status Bits

N: Set if result is negative, reset if positive
Z: Set if dst contained 0FFFFFh, reset otherwise
    Set if dst contained 0FFFFh, reset otherwise
    Set if dst contained 0FFh, reset otherwise
C: Set if result is not zero, reset otherwise ( = .NOT. Zero)
V: Set if initial destination operand was negative, otherwise reset

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

20-bit content of R5 is negated (2s complement).

INVX.A R5 ; Invert R5
INCX.A R5 ; R5 is now negated

Example

Content of memory byte LEO is negated. PC is pointing to upper memory.

INVX.B LEO ; Invert LEO
INCX.B LEO ; MEM(LEO) is negated
### 4.6.3.17 MOVX

#### MOBX.A
Move source address-word to destination address-word

#### MOBX.[W]
Move source word to destination word

#### MOBX.B
Move source byte to destination byte

**Syntax**
- `MOVX.A src,dst`
- `MOVX src,dst` or `MOVX.W src,dst`
- `MOVX.B src,dst`

**Operation**
- `src → dst`

**Description**
The source operand is copied to the destination. The source operand is not affected. Both operands may be located in the full address space.

**Status Bits**
- N: Not affected
- Z: Not affected
- C: Not affected
- V: Not affected

**Mode Bits**
OSCOFF, CPUOFF, and GIE are not affected.

**Example**
Move a 20-bit constant 18000h to absolute address-word EDE

```
MOVX.A   #018000h,&EDE         ; Move 18000h to EDE
```

**Example**
The contents of table EDE (word data, 20-bit addresses) are copied to table TOM. The length of the table is 030h words.

```
MOVA     #EDE,R10               ; Prepare pointer (20-bit address)
Loop  MOVX.W   @R10+,TOM-EDE-2(R10)   ; R10 points to both tables.
      CMPA     #$EDE+60h,R10           ; End of table reached?
      JLO      Loop                   ; Not yet
...                             ; Copy completed
```

**Example**
The contents of table EDE (byte data, 20-bit addresses) are copied to table TOM. The length of the table is 020h bytes.

```
MOVA     #EDE,R10               ; Prepare pointer (20-bit)
MOV      #$20h,R9                ; Prepare counter
Loop  MOVX.W   @R10+,TOM-EDE-2(R10)   ; R10 points to both tables.
      DEC      R9                     ; Decrement counter
      JNZ      Loop                   ; Not yet done
...                             ; Copy completed
```

Ten of the 28 possible addressing combinations of the MOVX.A instruction can use the MOVA instruction. This saves two bytes and code cycles. Examples for the addressing combinations are:

- `MOVX.A   Rsrc,Rdst          MOVA   Rsrc,Rdst        ; Reg/Reg`
- `MOVX.A   #imm20,Rdst        MOVA   #imm20,Rdst      ; Immediate/Reg`
- `MOVX.A   &abs20,Rdst        MOVA   &abs20,Rdst      ; Absolute/Reg`
- `MOVX.A   @Rsrc,Rdst         MOVA   @Rsrc,Rdst       ; Indirect/Reg`
- `MOVX.A   @Rsrc+,Rdst        MOVA   @Rsrc+,Rdst      ; Indirect,Auto/Reg`
- `MOVX.A   Rsrc,&abs20        MOVA   Rsrc,&abs20      ; Reg/Absolute`

The next four replacements are possible only if 16-bit indexes are sufficient for the addressing:

- `MOVX.A   z20(Rsrc),Rdst     MOVA   z16(Rsrc),Rdst   ; Indexed/Reg`
- `MOVX.A   Rsrc,z20(Rdst)     MOVA   Rsrc,z16(Rdst)   ; Reg/Indexed`
MOVX.A symb20,Rdst
MOVX.A Rsr,symb20
MOVX.A Rsrc,symb20
MOV A symb16,Rdst ; Symbolic/Reg
MOV A Rsrc,symb16 ; Reg/Symbolic
4.6.3.18 POPM

POPM.A  Restore n CPU registers (20-bit data) from the stack
POPM.[W] Restore n CPU registers (16-bit data) from the stack

Syntax
POPM.A #n,Rdst  \( 1 \leq n \leq 16 \)
POPM.W #n,Rdst or POPM #n,Rdst  \( 1 \leq n \leq 16 \)

Operation
POPM.A: Restore the register values from stack to the specified CPU registers. The SP is incremented by four for each register restored from stack. The 20-bit values from stack (two words per register) are restored to the registers.

POPM.W: Restore the 16-bit register values from stack to the specified CPU registers. The SP is incremented by two for each register restored from stack. The 16-bit values from stack (one word per register) are restored to the CPU registers.

Note: This instruction does not use the extension word.

Description
POPM.A: The CPU registers pushed on the stack are moved to the extended CPU registers, starting with the CPU register (Rdst – n + 1). The SP is incremented by \((n \times 4)\) after the operation.

POPM.W: The 16-bit registers pushed on the stack are moved back to the CPU registers, starting with CPU register (Rdst – n + 1). The SP is incremented by \((n \times 2)\) after the instruction. The MSBs (Rdst.19:16) of the restored CPU registers are cleared.

Status Bits
Status bits are not affected, except SR is included in the operation.

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
Restore the 20-bit registers R9, R10, R11, R12, R13 from the stack

POPM.A #5,R13 ; Restore R9, R10, R11, R12, R13

Example
Restore the 16-bit registers R9, R10, R11, R12, R13 from the stack.

POPM.W #5,R13 ; Restore R9, R10, R11, R12, R13
### 4.6.3.19 PUSHM

**PUSHM.A**

Save n CPU registers (20-bit data) on the stack

**PUSHM.[W]**

Save n CPU registers (16-bit words) on the stack

#### Syntax

- `PUSHM.A #n,Rdst`  
  
- `PUSHM.W #n,Rdst` or `PUSHM #n,Rdst`

1 ≤ n ≤ 16

#### Operation

**PUSHM.A:** Save the 20-bit CPU register values on the stack. The SP is decremented by four for each register stored on the stack. The MSBs are stored first (higher address).

**PUSHM.W:** Save the 16-bit CPU register values on the stack. The SP is decremented by two for each register stored on the stack.

#### Description

**PUSHM.A:** The n CPU registers, starting with Rdst backwards, are stored on the stack. The SP is decremented by (n × 4) after the operation. The data (Rn.19:0) of the pushed CPU registers is not affected.

**PUSHM.W:** The n registers, starting with Rdst backwards, are stored on the stack. The SP is decremented by (n × 2) after the operation. The data (Rn.19:0) of the pushed CPU registers is not affected.

**Note:** This instruction does not use the extension word.

#### Status Bits

Status bits are not affected.

#### Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

#### Example

Save the five 20-bit registers R9, R10, R11, R12, R13 on the stack

```assembly
PUSHM.A #5,R13 ; Save R13, R12, R11, R10, R9
```

Save the five 16-bit registers R9, R10, R11, R12, R13 on the stack

```assembly
PUSHM.W #5,R13 ; Save R13, R12, R11, R10, R9
```
4.6.3.20 POPX

* POPX.A  Restore single address-word from the stack
* POPX.[W] Restore single word from the stack
* POPX.B  Restore single byte from the stack

Syntax

POPX.A dst
POPX dst or POPX.W dst
POPX.B dst

Operation

Restore the 8-/16-/20-bit value from the stack to the destination. 20-bit addresses are possible. The SP is incremented by two (byte and word operands) and by four (address-word operand).

Emulation

MOVX(.B,.A) @SP+,dst

Description

The item on TOS is written to the destination operand. Register mode, Indexed mode, Symbolic mode, and Absolute mode are possible. The SP is incremented by two or four.

Note: The SP is incremented by two also for byte operations.

Status Bits

Status bits are not affected.

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

Write the 16-bit value on TOS to the 20-bit address &EDE

POPX.W &EDE ; Write word to address EDE

Example

Write the 20-bit value on TOS to R9

POPX.A R9 ; Write address-word to R9
4.6.3.21 PUSHX

**PUSHX.A**  
Save single address-word to the stack

**PUSHX.[W]**  
Save single word to the stack

**PUSHX.B**  
Save single byte to the stack

**Syntax**

- PUSHX.A src
- PUSHX src or PUSHX.W src
- PUSHX.B src

**Operation**  
Save the 8-/16-/20-bit value of the source operand on the TOS. 20-bit addresses are possible. The SP is decremented by two (byte and word operands) or by four (address-word operand) before the write operation.

**Description**  
The SP is decremented by two (byte and word operands) or by four (address-word operand). Then the source operand is written to the TOS. All seven addressing modes are possible for the source operand.

**Status Bits**  
Status bits are not affected.

**Mode Bits**  
OSCOFF, CPUOFF, and GIE are not affected.

**Example**  
Save the byte at the 20-bit address &EDE on the stack

```
PUSHX.B   &EDE     ; Save byte at address EDE
```

**Example**  
Save the 20-bit value in R9 on the stack.

```
PUSHX.A   R9       ; Save address-word in R9
```
4.6.3.22 RLAM

RLAM.A  Rotate left arithmetically the 20-bit CPU register content
RLAM.[W] Rotate left arithmetically the 16-bit CPU register content

Syntax

\[
\text{RLAM.A} \ #n, \text{Rdst} \\
\text{RLAM.W} \ #n, \text{Rdst} \text{ or RLAM} \ #n, \text{Rdst} \\
\]

1 ≤ n ≤ 4

Operation

C ← MSB ← MSB-1 …. LSB+1 ← LSB ← 0

Description

The destination operand is shifted arithmetically left one, two, three, or four positions as shown in Figure 4-43. RLAM works as a multiplication (signed and unsigned) with 2, 4, 8, or 16. The word instruction RLAM.W clears the bits Rdst.19:16.

Note: This instruction does not use the extension word.

Status Bits

N: Set if result is negative
   .A: Rdst.19 = 1, reset if Rdst.19 = 0
   .W: Rdst.15 = 1, reset if Rdst.15 = 0
Z: Set if result is zero, reset otherwise
C: Loaded from the MSB (n = 1), MSB-1 (n = 2), MSB-2 (n = 3), MSB-3 (n = 4)
V: Undefined

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

The 20-bit operand in R5 is shifted left by three positions. It operates equal to an arithmetic multiplication by 8.

\[
\text{RLAM.A} \ #3, \text{R5} \quad ; \text{R5} = \text{R5} \times 8
\]

Figure 4-43. Rotate Left Arithmetically—RLAM.[W] and RLAM.A
4.6.3.23 RLAX

* RLAX.A  
  Rotate left arithmetically address-word

* RLAX.[W]  
  Rotate left arithmetically word

* RLAX.B  
  Rotate left arithmetically byte

Syntax

RLAX.A dst
RLAX dst or RLAX.W dst
RLAX.B dst

Operation

C ← MSB ← MSB-1 .... LSB+1 ← LSB ← 0

Emulation

ADDX.A dst,dst
ADDX dst,dst
ADDX.B dst,dst

Description

The destination operand is shifted left one position as shown in Figure 4-44. The MSB is shifted into the carry bit (C) and the LSB is filled with 0. The RLAX instruction acts as a signed multiplication by 2.

Status Bits

N: Set if result is negative, reset if positive
Z: Set if result is zero, reset otherwise
C: Loaded from the MSB
V: Set if an arithmetic overflow occurs: the initial value is 040000h ≤ dst < 0C0000h; reset otherwise
   Set if an arithmetic overflow occurs: the initial value is 04000h ≤ dst < 0C000h; reset otherwise
   Set if an arithmetic overflow occurs: the initial value is 040h ≤ dst < 0C0h; reset otherwise

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

The 20-bit value in R7 is multiplied by 2

RLAX.A R7 ; Shift left R7 (20-bit)

Figure 4-44. Destination Operand-Arithmetic Shift Left
4.6.3.24 RLCX

* RLCX.A  Rotate left through carry address-word
* RLCX.[W] Rotate left through carry word
* RLCX.B  Rotate left through carry byte

Syntax

RLCX.A dst
RLCX dst or RLCX.W dst
RLCX.B dst

Operation

C ← MSB ← MSB-1 .... LSB+1 ← LSB ← C

Emulation

ADDCX.A dst,dst
ADDCX dst,dst
ADDCX.B dst,dst

Description

The destination operand is shifted left one position as shown in Figure 4-45. The carry bit (C) is shifted into the LSB and the MSB is shifted into the carry bit (C).

Status Bits

N:    Set if result is negative, reset if positive
Z:    Set if result is zero, reset otherwise
C:    Loaded from the MSB
V:    Set if an arithmetic overflow occurs: the initial value is 040000h ≤ dst < 0C0000h; reset otherwise

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

The 20-bit value in R5 is shifted left one position.

RLCX.A R5 ; (R5 x 2) + C -> R5

Example

The RAM byte LEO is shifted left one position. PC is pointing to upper memory.

RLCX.B LEO ; RAM(LEO) x 2 + C -> RAM(LEO)

Figure 4-45. Destination Operand-Carry Left Shift
4.6.3.25 RRAM

RRAM.A
Rotate right arithmetically the 20-bit CPU register content

RRAM.[W]
Rotate right arithmetically the 16-bit CPU register content

Syntax
RRAM.A #n,Rdst
RRAM.W #n,Rdst or RRAM #n,Rdst

Operation
MSB → MSB → MSB–1 ... LSB+1 → LSB → C

Description
The destination operand is shifted right arithmetically by one, two, three, or four bit positions as shown in Figure 4-46. The MSB retains its value (sign). RRAM operates equal to a signed division by 2/4/8/16. The MSB is retained and shifted into MSB-1. The LSB+1 is shifted into the LSB, and the LSB is shifted into the carry bit C. The word instruction RRAM.W clears the bits Rdst.19:16.

Note: This instruction does not use the extension word.

Status Bits
N: Set if result is negative
A: Rdst.19 = 1, reset if Rdst.19 = 0
.W: Rdst.15 = 1, reset if Rdst.15 = 0
Z: Set if result is zero, reset otherwise
C: Loaded from the LSB (n = 1), LSB+1 (n = 2), LSB+2 (n = 3), or LSB+3 (n = 4)
V: Reset

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
The signed 20-bit number in R5 is shifted arithmetically right two positions.

RRAM.A #2,R5 ; R5/4 → R5

Example
The signed 20-bit value in R15 is multiplied by 0.75. (0.5 + 0.25) × R15.

PUSHM.A #1,R15 ; Save extended R15 on stack
RRAM.A #1,R15 ; R15 y 0.5 → R15
ADDX.A @SP+,R15 ; R15 y 0.5 + R15 = 1.5 y R15 → R15
RRAM.A #1,R15 ; (1.5 y R15) y 0.5 = 0.75 y R15 → R15

Figure 4-46. Rotate Right Arithmetically RRAM.[W] and RRAM.A
4.6.3.26 RRAX

**RRAX.A**
Rotate right arithmetically the 20-bit operand

**RRAX.B**
Rotate right arithmetically the 8-bit operand

**Syntax**

```
RRAX.A Rdst
RRAX.W Rdst
RRAX Rdst
RRAX.B Rdst
RRAX.A dst
RRAX dst or RRAX.W dst
RRAX.B dst
```

**Operation**
MSB → MSB → MSB–1 ... LSB+1 → LSB → C

**Description**
Register mode for the destination: the destination operand is shifted right by one bit position as shown in Figure 4-47. The MSB retains its value (sign). The word instruction RRAX.W clears the bits Rdst.19:16, the byte instruction RRAX.B clears the bits Rdst.19:8. The MSB retains its value (sign), the LSB is shifted into the carry bit. RRAX here operates equal to a signed division by 2.

All other modes for the destination: the destination operand is shifted right arithmetically by one bit position as shown in Figure 4-48. The MSB retains its value (sign), the LSB is shifted into the carry bit. RRAX here operates equal to a signed division by 2. All addressing modes, with the exception of the Immediate mode, are possible in the full memory.

**Status Bits**

- **N:** Set if result is negative, reset if positive
- **A:** dst.19 = 1, reset if dst.19 = 0
- **W:** dst.15 = 1, reset if dst.15 = 0
- **B:** dst.7 = 1, reset if dst.7 = 0
- **Z:** Set if result is zero, reset otherwise
- **C:** Loaded from the LSB
- **V:** Reset

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**
The signed 20-bit number in R5 is shifted arithmetically right four positions.

```
RPT      #4
RRAX.A   R5        ; R5/16 -> R5
```

**Example**
The signed 8-bit value in EDE is multiplied by 0.5.

```
RRAX.B   &EDE      ; EDE/2 -> EDE
```
Figure 4-47. Rotate Right Arithmetically RRAX(.B,.A) – Register Mode

Figure 4-48. Rotate Right Arithmetically RRAX(.B,.A) – Non-Register Mode
### 4.6.3.27 RRCM

**RRCM.A** Rotate right through carry the 20-bit CPU register content

**RRCM.[W]** Rotate right through carry the 16-bit CPU register content

**Syntax**

- \( \text{RRCM.A} \ n, \text{Rdst} \) \( 1 \leq n \leq 4 \)
- \( \text{RRCM.W} \ n, \text{Rdst} \) or \( \text{RRCM} \ n, \text{Rdst} \) \( 1 \leq n \leq 4 \)

**Operation**

C → MSB → MSB–1 ... LSB+1 → LSB → C

**Description**

The destination operand is shifted right by one, two, three, or four bit positions as shown in Figure 4-49. The carry bit C is shifted into the MSB, the LSB is shifted into the carry bit. The word instruction RRCM.W clears the bits Rdst.19:16.

Note: This instruction does not use the extension word.

**Status Bits**

- **N**: Set if result is negative
- \( .A \): Rdst.19 = 1, reset if Rdst.19 = 0
- \( .W \): Rdst.15 = 1, reset if Rdst.15 = 0
- **Z**: Set if result is zero, reset otherwise
- **C**: Loaded from the LSB \((n = 1)\), LSB+1 \((n = 2)\), LSB+2 \((n = 3)\), or LSB+3 \((n = 4)\)
- **V**: Reset

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

The address-word in R5 is shifted right by three positions. The MSB–2 is loaded with 1.

\[
\begin{align*}
\text{SETC} & \quad ; \text{Prepare carry for MSB-2} \\
\text{RRCM.A} & \ #3,R5 \quad ; \ R5 \rightarrow R5 \gg 3 + 20000h
\end{align*}
\]

**Example**

The word in R6 is shifted right by two positions. The MSB is loaded with the LSB. The MSB–1 is loaded with the contents of the carry flag.

\[
\begin{align*}
\text{RRCM.W} & \ #2,R6 \quad ; \ R6 \rightarrow R6 \gg 2. R6.19:16 = 0
\end{align*}
\]

**Figure 4-49. Rotate Right Through Carry RRCM.[W] and RRCM.A**
4.6.3.28 RRCX

RRCX.A   Rotate right through carry the 20-bit operand
RRCX.W   Rotate right through carry the 16-bit operand
RRCX.B   Rotate right through carry the 8-bit operand

Syntax

RRCX.A Rdst
RRCX.W Rdst
RRCX Rdst
RRCX.B Rdst

RRCX.A dst
RRCX dst or RRCX.W dst
RRCX.B dst

Operation

C → MSB → MSB–1 ... LSB+1 → LSB → C

Description

Register mode for the destination: the destination operand is shifted right by one bit position as shown in Figure 4-50. The word instruction RRCX.W clears the bits Rdst.19:16, the byte instruction RRCX.B clears the bits Rdst.19:8. The carry bit C is shifted into the MSB, the LSB is shifted into the carry bit.

All other modes for the destination: the destination operand is shifted right by one bit position as shown in Figure 4-51. The carry bit C is shifted into the MSB, the LSB is shifted into the carry bit. All addressing modes, with the exception of the Immediate mode, are possible in the full memory.

Status Bits

N: Set if result is negative
.A: dst.19 = 1, reset if dst.19 = 0
.W: dst.15 = 1, reset if dst.15 = 0
.B: dst.7 = 1, reset if dst.7 = 0

Z: Set if result is zero, reset otherwise
C: Loaded from the LSB
V: Reset

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.
Example

The 20-bit operand at address EDE is shifted right by one position. The MSB is loaded with 1.

```
SETC              ; Prepare carry for MSB
RRCX.A   EDE      ; EDE = EDE » 1 + 80000h
```

Example

The word in R6 is shifted right by 12 positions.

```
RPT      #12
RRCX.W   R6       ; R6 = R6 » 12. R6.19:16 = 0
```
Figure 4-50. Rotate Right Through Carry RRCX(.B,.A) – Register Mode

Figure 4-51. Rotate Right Through Carry RRCX(.B,.A) – Non-Register Mode
4.6.3.29 RRUM

**RRUM.A**
Rotate right through carry the 20-bit CPU register content

**RRUM.W**
Rotate right through carry the 16-bit CPU register content

**Syntax**

RRUM.A \( \#n, Rdst \)

RRUM.W \( \#n, Rdst \) or RRUM \( \#n, Rdst \)

1 ≤ \( n \) ≤ 4

**Operation**

0 \( \rightarrow \) MSB \( \rightarrow \) MSB–1 \( \rightarrow \) LSB+1 \( \rightarrow \) LSB \( \rightarrow \) C

**Description**
The destination operand is shifted right by one, two, three, or four bit positions as shown in Figure 4-52. Zero is shifted into the MSB, the LSB is shifted into the carry bit. RRUM works like an unsigned division by 2, 4, 8, or 16. The word instruction RRUM.W clears the bits Rdst.19:16.

**Note:** This instruction does not use the extension word.

**Status Bits**

- **N:** Set if result is negative
  - A: Rdst.19 = 1, reset if Rdst.19 = 0
  - W: Rdst.15 = 1, reset if Rdst.15 = 0

- **Z:** Set if result is zero, reset otherwise

- **C:** Loaded from the LSB (\( n = 1 \)), LSB+1 (\( n = 2 \)), LSB+2 (\( n = 3 \)), or LSB+3 (\( n = 4 \))

- **V:** Reset

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**
The unsigned address-word in R5 is divided by 16.

RRUM.A \( \#4, R5 \) ; \( R5 = R5 \gg 4 \). \( R5/16 \)

**Example**
The word in R6 is shifted right by one bit. The MSB R6.15 is loaded with 0.

RRUM.W \( \#1, R6 \) ; \( R6 = R6/2 \). \( R6.19:15 = 0 \)

**Figure 4-52. Rotate Right Unsigned RRUM[W] and RRUM.A**
### 4.6.3.30 RRUX

**RRUX.A**
Shift right unsigned the 20-bit CPU register content

**RRUX.W**
Shift right unsigned the 16-bit CPU register content

**RRUX.B**
Shift right unsigned the 8-bit CPU register content

**Syntax**

```
RRUX.A Rdst
RRUX.W Rdst
RRUX Rdst
RRUX.B Rdst
```

**Operation**

\[ C=0 \to \text{MSB} \to \text{MSB}–1 \ldots \text{LSB}+1 \to \text{LSB} \to C \]

**Description**

RRUX is valid for register mode only: the destination operand is shifted right by one bit position as shown in Figure 4-53. The word instruction RRUX.W clears the bits Rdst.19:16. The byte instruction RRUX.B clears the bits Rdst.19:8. Zero is shifted into the MSB, the LSB is shifted into the carry bit.

**Status Bits**

- **N**: Set if result is negative
- **A**: dst.19 = 1, reset if dst.19 = 0
- **W**: dst.15 = 1, reset if dst.15 = 0
- **B**: dst.7 = 1, reset if dst.7 = 0
- **Z**: Set if result is zero, reset otherwise
- **C**: Loaded from the LSB
- **V**: Reset

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

The word in R6 is shifted right by 12 positions.

```
RPT #12
RRUX.W R6 ; R6 = R6 » 12. R6.19:16 = 0
```

![Figure 4-53. Rotate Right Unsigned RRUX(.B,.A) – Register Mode](image-url)
4.6.3.31 SBCX

* SBCX.A  Subtract borrow (.NOT. carry) from destination address-word
* SBCX.[W] Subtract borrow (.NOT. carry) from destination word
* SBCX.B  Subtract borrow (.NOT. carry) from destination byte

Syntax

SBCX.A dst
SBCX dst or SBCX.W dst
SBCX.B dst

Operation

dst + 0FFFFh + C → dst

dst + 0FFFFh + C → dst

dst + 0FFh + C → dst

Emulation

SBCX.A #0,dst
SBCX #0,dst
SBCX.B #0,dst

Description

The carry bit (C) is added to the destination operand minus one. The previous contents of the destination are lost.

Status Bits

N:  Set if result is negative, reset if positive
Z:  Set if result is zero, reset otherwise
C:  Set if there is a carry from the MSB of the result, reset otherwise
     Set to 1 if no borrow, reset if borrow
V:  Set if an arithmetic overflow occurs, reset otherwise

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

The 8-bit counter pointed to by R13 is subtracted from a 16-bit counter pointed to by R12.

```
SUBX.B   @R13,0(R12)      ; Subtract LSDs
SBCX.B   i(R12)           ; Subtract carry from MSD
```

Note

Borrow implementation

The borrow is treated as a .NOT. carry:

<table>
<thead>
<tr>
<th>Borrow</th>
<th>Carry Bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Yes</td>
<td>0</td>
</tr>
<tr>
<td>No</td>
<td>1</td>
</tr>
</tbody>
</table>
4.6.3.32 SUBX

<table>
<thead>
<tr>
<th>SUBX.A</th>
<th>Subtract source address-word from destination address-word</th>
</tr>
</thead>
<tbody>
<tr>
<td>SUBX.[W]</td>
<td>Subtract source word from destination word</td>
</tr>
<tr>
<td>SUBX.B</td>
<td>Subtract source byte from destination byte</td>
</tr>
</tbody>
</table>

**Syntax**

- `SUBX.A src,dst`
- `SUBX src,dst` or `SUBX.W src,dst`
- `SUBX.B src,dst`

**Operation**

\[ (.not. \text{src}) + 1 + \text{dst} \rightarrow \text{dst} \quad \text{or} \quad \text{dst} - \text{src} \rightarrow \text{dst} \]

**Description**

The source operand is subtracted from the destination operand. This is done by adding the 1s complement of the source + 1 to the destination. The source operand is not affected. The result is written to the destination operand. Both operands may be located in the full address space.

**Status Bits**

- **N:** Set if result is negative (\( \text{src} > \text{dst} \)), reset if positive (\( \text{src} \leq \text{dst} \))
- **Z:** Set if result is zero (\( \text{src} = \text{dst} \)), reset otherwise (\( \text{src} \neq \text{dst} \))
- **C:** Set if there is a carry from the MSB, reset otherwise
- **V:** Set if the subtraction of a negative source operand from a positive destination operand delivers a negative result, or if the subtraction of a positive source operand from a negative destination operand delivers a positive result, reset otherwise (no overflow)

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

A 20-bit constant 87654h is subtracted from EDE (LSBs) and EDE+2 (MSBs).

```
SUBX.A #87654h,EDE     ; Subtract 87654h from EDE+2|EDE
```

**Example**

A table word pointed to by R5 (20-bit address) is subtracted from R7. Jump to label TONI if R7 contains zero after the instruction. R5 is auto-incremented by two. R7.19:16 = 0.

```
SUBX.W @R5+,R7         ; Subtract table number from R7. R5 + 2
JZ       TONI            ; R7 = @R5 (before subtraction)
...                      ; R7 <> @R5 (before subtraction)
```

**Example**

Byte CNT is subtracted from the byte R12 points to in the full address space. Address of CNT is within PC ± 512 K.

```
SUBX.B CNT,0(R12)      ; Subtract CNT from @R12
```

**Note:** Use SUBA for the following two cases for better density and execution.

```
SUBX.A Rsrc,Rdst
SUBX.A #imm20,Rdst
```
4.6.3.33 SUBCX

**SUBCX.A**  Subtract source address-word with carry from destination address-word

**SUBCX.W**  Subtract source word with carry from destination word

**SUBCX.B**  Subtract source byte with carry from destination byte

**Syntax**

```
SUBCX.A src,dst
SUBCX src,dst or SUBCX.W src,dst
SUBCX.B src,dst
```

**Operation**

```
(.not. src) + C + dst → dst   or   dst – (src – 1) + C → dst
```

**Description**
The source operand is subtracted from the destination operand. This is made by adding the 1s complement of the source + carry to the destination. The source operand is not affected, the result is written to the destination operand. Both operands may be located in the full address space.

**Status Bits**

- **N**: Set if result is negative (MSB = 1), reset if positive (MSB = 0)
- **Z**: Set if result is zero, reset otherwise
- **C**: Set if there is a carry from the MSB, reset otherwise
- **V**: Set if the subtraction of a negative source operand from a positive destination operand delivers a negative result, or if the subtraction of a positive source operand from a negative destination operand delivers a positive result, reset otherwise (no overflow).

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

A 20-bit constant 87654h is subtracted from R5 with the carry from the previous instruction.

```
SUBCX.A #87654h,R5      ; Subtract 87654h + C from R5
```

**Example**

A 48-bit number (3 words) pointed to by R5 (20-bit address) is subtracted from a 48-bit counter in RAM, pointed to by R7. R5 auto-increments to point to the next 48-bit number.

```
SUB.X.W @R5+,0(R7)      ; Subtract LSBs. R5 + 2
SUBCX.W @R5+,2(R7)      ; Subtract MIDs with C. R5 + 2
SUBCX.W @R5+,4(R7)      ; Subtract MSBs with C. R5 + 2
```

**Example**

Byte CNT is subtracted from the byte R12 points to. The carry of the previous instruction is used. 20-bit addresses.

```
SUBCX.B &CNT,0(R12)     ; Subtract byte CNT from @R12
```
4.6.3.34 SWPBX

SWPBX.A  Swap bytes of lower word
SWPBX.[W] Swap bytes of word

Syntax

```
SWPBX A dst
```

Operation

dst.15:8 ↔ dst.7:0

Description

Register mode: Rn.15:8 are swapped with Rn.7:0. When the .A extension is used, Rn.19:16 are unchanged. When the .W extension is used, Rn.19:16 are cleared.

Other modes: When the .A extension is used, bits 31:20 of the destination address are cleared, bits 19:16 are left unchanged, and bits 15:8 are swapped with bits 7:0. When the .W extension is used, bits 15:8 are swapped with bits 7:0 of the addressed word.

Status Bits

Status bits are not affected.

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Example

Exchange the bytes of RAM address-word EDE

```
MOVX A #23456h,&EDE ; 23456h -> EDE
SWPBX A EDE ; 25634h -> EDE
```

Example

Exchange the bytes of R5

```
MOVA #23456h,R5 ; 23456h -> R5
SWPBX W R5 ; 05634h -> R5
```

Before SWPBX.A

<table>
<thead>
<tr>
<th>19</th>
<th>16</th>
<th>15</th>
<th>8</th>
<th>7</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>X</td>
<td></td>
<td></td>
<td>High Byte</td>
<td>Low Byte</td>
<td></td>
</tr>
</tbody>
</table>

After SWPBX.A

<table>
<thead>
<tr>
<th>19</th>
<th>16</th>
<th>15</th>
<th>8</th>
<th>7</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>X</td>
<td></td>
<td></td>
<td>Low Byte</td>
<td>High Byte</td>
<td></td>
</tr>
</tbody>
</table>

**Figure 4-54. Swap Bytes SWPBX.A Register Mode**

Before SWPBX.A

<table>
<thead>
<tr>
<th>31</th>
<th>20</th>
<th>19</th>
<th>16</th>
<th>15</th>
<th>8</th>
<th>7</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>X</td>
<td>X</td>
<td></td>
<td>High Byte</td>
<td>Low Byte</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

After SWPBX.A

<table>
<thead>
<tr>
<th>31</th>
<th>20</th>
<th>19</th>
<th>16</th>
<th>15</th>
<th>8</th>
<th>7</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>X</td>
<td></td>
<td>Low Byte</td>
<td>High Byte</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Figure 4-55. Swap Bytes SWPBX.A In Memory**
Before SWPBX

<table>
<thead>
<tr>
<th>19</th>
<th>16</th>
<th>15</th>
<th>8</th>
<th>7</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>X</td>
<td>High Byte</td>
<td>Low Byte</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

After SWPBX

<table>
<thead>
<tr>
<th>19</th>
<th>16</th>
<th>15</th>
<th>8</th>
<th>7</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>Low Byte</td>
<td>High Byte</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Figure 4-56. Swap Bytes SWPBX [.W] Register Mode**

Before SWPBX

<table>
<thead>
<tr>
<th>15</th>
<th>8</th>
<th>7</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>High Byte</td>
<td>Low Byte</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

After SWPBX

<table>
<thead>
<tr>
<th>15</th>
<th>8</th>
<th>7</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Low Byte</td>
<td>High Byte</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Figure 4-57. Swap Bytes SWPBX [.W] In Memory**
4.6.3.35 SXTX

SXTX.A  Extend sign of lower byte to address-word
SXTX.[W] Extend sign of lower byte to word

Syntax
SXTX.A dst
SXTX dst or SXTX.[W] dst

Operation
dst.7 → dst.15:8, Rdst.7 → Rdst.19:8 (Register mode)

Description
Register mode: The sign of the low byte of the operand (Rdst.7) is extended into the bits Rdst.19:8.
Other modes: SXTX.A: the sign of the low byte of the operand (dst.7) is extended into dst.19:8. The bits dst.31:20 are cleared.
SXTX.[W]: the sign of the low byte of the operand (dst.7) is extended into dst.15:8.

Status Bits
N: Set if result is negative, reset otherwise
Z: Set if result is zero, reset otherwise
C: Set if result is not zero, reset otherwise (C = .not.Z)
V: Reset

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
The signed 8-bit data in EDE.7:0 is sign extended to 20 bits: EDE.19:8. Bits 31:20 located in EDE+2 are cleared.

```
SXTX.A  &EDE  ; Sign extended EDE -> EDE+2/EDE
```

<table>
<thead>
<tr>
<th>19</th>
<th>16</th>
<th>15</th>
<th>8</th>
<th>7</th>
<th>6</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td>S</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Figure 4-58. Sign Extend SXTX.A

```
SXTX[.W] Rdst
```

<table>
<thead>
<tr>
<th>19</th>
<th>16</th>
<th>15</th>
<th>8</th>
<th>7</th>
<th>6</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

```
SXTX[.W] dst
```

<table>
<thead>
<tr>
<th>15</th>
<th>8</th>
<th>7</th>
<th>6</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Figure 4-59. Sign Extend SXTX[.W]
4.6.3.36 TSTX

* TSTX.A  Test destination address-word
* TSTX.[W] Test destination word
* TSTX.B  Test destination byte

Syntax
TSTX.A dst
TSTX dst or TSTX.W dst
TSTX.B dst

Operation
dst + 0FFFFFh + 1
dst + 0FFFFh + 1
dst + 0FFh + 1

Emulation
CMPX.A #0,dst
CMPX #0,dst
CMPX.B #0,dst

Description
The destination operand is compared with zero. The status bits are set according to the result. The destination is not affected.

Status Bits
N:   Set if destination is negative, reset if positive
Z:   Set if destination contains zero, reset otherwise
C:   Set
V:   Reset

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
RAM byte LEO is tested; PC is pointing to upper memory. If it is negative, continue at LEONEG; if it is positive but not zero, continue at LEOPOS.

```
TSTX.B  LEO    ; Test LEO
JN      LEONEG ; LEO is negative
JZ      LEOZERO; LEO is zero
LEOPOS   ....    ; LEO is positive but not zero
LEONEG   ....    ; LEO is negative
LEOZERO  ....    ; LEO is zero
```
4.6.3.37 XORX

**XORX.A** Exclusive OR source address-word with destination address-word

**XORX.W** Exclusive OR source word with destination word

**XORX.B** Exclusive OR source byte with destination byte

**Syntax**

XORX.A src,dst  
XORX src,dst or XORX.W src,dst  
XORX.B src,dst

**Operation**

src .xor. dst → dst

**Description**
The source and destination operands are exclusively ORed. The result is placed into the destination. The source operand is not affected. The previous contents of the destination are lost. Both operands may be located in the full address space.

**Status Bits**

N: Set if result is negative (MSB = 1), reset if positive (MSB = 0)
Z: Set if result is zero, reset otherwise
C: Set if result is not zero, reset otherwise (carry = .not. Zero)
V: Set if both operands are negative (before execution), reset otherwise

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

Toggle bits in address-word CNTR (20-bit data) with information in address-word TONI (20-bit address)

XORX.A TONI,&CNTR ; Toggle bits in CNTR

**Example**

A table word pointed to by R5 (20-bit address) is used to toggle bits in R6.

XORX.W @R5,R6 ; Toggle bits in R6. R6.19:16 = 0

**Example**

Reset to zero those bits in the low byte of R7 that are different from the bits in byte EDE (20-bit address)

XORX.B EDE,R7 ; Set different bits to 1 in R7
INV.B R7 ; Invert low byte of R7. R7.19:8 = 0.
4.6.4 MSP430X Address Instructions

MSP430X address instructions are instructions that support 20-bit operands but have restricted addressing modes. The addressing modes are restricted to the Register mode and the Immediate mode, except for the MOVA instruction. Restricting the addressing modes removes the need for the additional extension-word op-code, which improves code density and execution time. The MSP430X address instructions are described in the following sections.

See Section 4.6.3 for MSP430X extended instructions and Section 4.6.2 for standard MSP430 instructions.
4.6.4.1 ADDA

ADDA  Add 20-bit source to a 20-bit destination register
Syntax  ADDA Rscc,Rdst
        ADDA #imm20,Rdst
Operation  src + Rdst → Rdst
Description  The 20-bit source operand is added to the 20-bit destination CPU register. The previous contents of the destination are lost. The source operand is not affected.

Status Bits  N:  Set if result is negative (Rdst.19 = 1), reset if positive (Rdst.19 = 0)
            Z:  Set if result is zero, reset otherwise
            C:  Set if there is a carry from the 20-bit result, reset otherwise
            V:  Set if the result of two positive operands is negative, or if the result of two negative numbers is positive, reset otherwise

Mode Bits  OSCOFF, CPUOFF, and GIE are not affected.
Example  R5 is increased by 0A4320h. The jump to TONI is performed if a carry occurs.

ADDA   #0A4320h,R5   ; Add A4320h to 20-bit R5
JC     TONI         ; Jump on carry
...     ; No carry occurred
### 4.6.4.2 BRA

<table>
<thead>
<tr>
<th>* BRA</th>
<th>Branch to destination</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax</td>
<td>BRA dst</td>
</tr>
<tr>
<td>Operation</td>
<td>dst → PC</td>
</tr>
<tr>
<td>Emulation</td>
<td>MOVA dst,PC</td>
</tr>
</tbody>
</table>

**Description**
An unconditional branch is taken to a 20-bit address anywhere in the full address space. All seven source addressing modes can be used. The branch instruction is an address-word instruction. If the destination address is contained in a memory location X, it is contained in two ascending words: X (LSBs) and (X + 2) (MSBs).

**Status Bits**
- N: Not affected
- Z: Not affected
- C: Not affected
- V: Not affected

**Mode Bits**
OSCOFF, CPUOFF, and GIE are not affected.

**Examples**
Examples for all addressing modes are given.

**Immediate mode:** Branch to label EDE located anywhere in the 20-bit address space or branch directly to address.

```
BRA      #EDE      ; MOVA   #imm20,PC
BRA      #01AA04h
```

**Symbolic mode:** Branch to the 20-bit address contained in addresses EXEC (LSBs) and EXEC+2 (MSBs). EXEC is located at the address (PC + X) where X is within +32 K. Indirect addressing.

```
BRA      EXEC      ; MOVA   z16(PC),PC
```

**Note:** If the 16-bit index is not sufficient, a 20-bit index may be used with the following instruction.

```
MOVX.A   EXEC,PC   ; 1M byte range with 20-bit index
```

**Absolute mode:** Branch to the 20-bit address contained in absolute addresses EXEC (LSBs) and EXEC+2 (MSBs). Indirect addressing.

```
BRA      &EXEC      ; MOVA   &abs20,PC
```

**Register mode:** Branch to the 20-bit address contained in register R5. Indirect R5.

```
BRA      R5         ; MOVA   R5,PC
```

**Indirect mode:** Branch to the 20-bit address contained in the word pointed to by register R5 (LSBs). The MSBs have the address (R5 + 2). Indirect, indirect R5.

```
BRA      @R5        ; MOVA   @R5,PC
```

**Indirect, Auto-Increment mode:** Branch to the 20-bit address contained in the words pointed to by register R5 and increment the address in R5 afterwards by 4. The next time the S/W flow uses R5 as a pointer, it can alter the program execution due to access to the next address in the table pointed to by R5. Indirect, indirect R5.

```
BRA      @R5+       ; MOVA   @R5+,PC. R5 + 4
```
Indexed mode: Branch to the 20-bit address contained in the address pointed to by register (R5 + X) (for example, a table with addresses starting at X). (R5 + X) points to the LSBs, (R5 + X + 2) points to the MSBs of the address. X is within R5 + 32 K. Indirect, indirect (R5 + X).

BRA X(R5) ; MOVA z16(R5),PC

Note: If the 16-bit index is not sufficient, a 20-bit index X may be used with the following instruction:

MOVX.A X(R5),PC ; 1M byte range with 20-bit index
4.6.4.3 CALLA

CALLA Call a subroutine

Syntax
CALLA dst

Operation
dst → tmp 20-bit dst is evaluated and stored
SP – 2 → SP
PC.19:16 → @SP updated PC with return address to TOS (MSBs)
SP – 2 → SP
PC.15:0 → @SP updated PC to TOS (LSBs)
tmp → PC saved 20-bit dst to PC

Description
A subroutine call is made to a 20-bit address anywhere in the full address space. All seven source addressing modes can be used. The call instruction is an address-word instruction. If the destination address is contained in a memory location X, it is contained in two ascending words, X (LSBs) and (X + 2) (MSBs). Two words on the stack are needed for the return address. The return is made with the instruction RETA.

Status Bits
N: Not affected
Z: Not affected
C: Not affected
V: Not affected

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Examples
Examples for all addressing modes are given.
Immediate mode: Call a subroutine at label EXEC or call directly an address.
CALLA   #EXEC         ; Start address EXEC
CALLA   #01AA04h      ; Start address 01AA04h

Symbolic mode: Call a subroutine at the 20-bit address contained in addresses EXEC (LSBs) and EXEC+2 (MSBs). EXEC is located at the address (PC + X) where X is within +32 K. Indirect addressing.
CALLA   EXEC          ; Start address at @EXEC. z16(PC)

Absolute mode: Call a subroutine at the 20-bit address contained in absolute addresses EXEC (LSBs) and EXEC+2 (MSBs). Indirect addressing.
CALLA   &EXEC         ; Start address at @EXEC

Register mode: Call a subroutine at the 20-bit address contained in register R5. Indirect R5.
CALLA   R5            ; Start address at @R5

Indirect mode: Call a subroutine at the 20-bit address contained in the word pointed to by register R5 (LSBs). The MSBs have the address (R5 + 2). Indirect, indirect R5.
CALLA   @R5           ; Start address at @R5

Indirect, Auto-Increment mode: Call a subroutine at the 20-bit address contained in the words pointed to by register R5 and increment the 20-bit address in R5 afterwards by 4. The next time the S/W flow uses R5 as a pointer, it can alter the program execution due to access to the next word address in the table pointed to by R5. Indirect, indirect R5.
CALLA   @R5+          ; Start address at @R5. R5 + 4
Indexed mode: Call a subroutine at the 20-bit address contained in the address pointed to by register (R5 + X); for example, a table with addresses starting at X. (R5 + X) points to the LSBs, (R5 + X + 2) points to the MSBs of the word address. X is within R5 + 32 K. Indirect, indirect (R5 + X).

CALLA X(R5) ; Start address at @(R5+X). z16(R5)
4.6.4.4 CLRA

* CLRA  Clear 20-bit destination register
Syntax  CLRA Rdst
Operation  0 → Rdst
Emulation  MOVA #0,Rdst
Description  The destination register is cleared.
Status Bits  Status bits are not affected.
Example  The 20-bit value in R10 is cleared.

CLRA R10 ; 0 -> R10
4.6.4.5 CMPA

**CMPA**  Compare the 20-bit source with a 20-bit destination register

**Syntax**

- `CMPA Rsrc,Rdst`
- `CMPA #imm20,Rdst`

**Operation**

\[(\text{not src}) + 1 + \text{Rdst} \quad \text{or} \quad \text{Rdst} - \text{src}\]

**Description**

The 20-bit source operand is subtracted from the 20-bit destination CPU register. This is made by adding the 1s complement of the source + 1 to the destination register. The result affects only the status bits.

**Status Bits**

- **N**: Set if result is negative (src > dst), reset if positive (src ≤ dst)
- **Z**: Set if result is zero (src = dst), reset otherwise (src ≠ dst)
- **C**: Set if there is a carry from the MSB, reset otherwise
- **V**: Set if the subtraction of a negative source operand from a positive destination operand delivers a negative result, or if the subtraction of a positive source operand from a negative destination operand delivers a positive result, reset otherwise (no overflow)

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

A 20-bit immediate operand and R6 are compared. If they are equal, the program continues at label EQUAL.

```assembly
CMPA   #12345h,R6      ; Compare R6 with 12345h
JEQ    EQUAL           ; R5 = 12345h
...                    ; Not equal
```

**Example**

The 20-bit values in R5 and R6 are compared. If R5 is greater than (signed) or equal to R6, the program continues at label GRE.

```assembly
CMPA   R6,R5           ; Compare R6 with R5 (R5 - R6)
JGE    GRE             ; R5 >= R6
...                    ; R5 < R6
```
4.6.4.6 DECDA

* DECDA Double-decrement 20-bit destination register
Syntax
DECDA Rdst
Operation
Rdst – 2 → Rdst
Emulation
SUBA #2,Rdst
Description
The destination register is decremented by two. The original contents are lost.
Status Bits
N: Set if result is negative, reset if positive
Z: Set if Rdst contained 2, reset otherwise
C: Reset if Rdst contained 0 or 1, set otherwise
V: Set if an arithmetic overflow occurs, otherwise reset
Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.
Example
The 20-bit value in R5 is decremented by 2.

    DECDA R5 ; Decrement R5 by two
4.6.4.7 INCDA

* INCDA Double-increment 20-bit destination register
Syntax INCDA Rdst
Operation Rdst + 2 → Rdst
Emulation ADDA #2,Rdst
Description The destination register is incremented by two. The original contents are lost.
Status Bits
N: Set if result is negative, reset if positive
Z: Set if Rdst contained 0FFFEh, reset otherwise
Set if Rdst contained 0FFFFEh, reset otherwise
Set if Rdst contained 0FFFe, reset otherwise
C: Set if Rdst contained 0FFFFEh or 0FFFFFh, reset otherwise
Set if Rdst contained 0FFFe or 0FFFFh, reset otherwise
Set if Rdst contained 0FEh or 0FFh, reset otherwise
V: Set if Rdst contained 07FFFEh or 07FFFFh, reset otherwise
Set if Rdst contained 07FFEh or 07FFFh, reset otherwise
Set if Rdst contained 07Eh or 07Fh, reset otherwise
Mode Bits OSCOFF, CPUOFF, and GIE are not affected.
Example The 20-bit value in R5 is incremented by two.

INCDA R5 ; Increment R5 by two
4.6.4.8 MOVA

MOVA Move the 20-bit source to the 20-bit destination

Syntax

MOVA Rs, Rds
MOVA #imm20, Rds
MOVA z16(Rs), Rds
MOVA EDE, Rds
MOVA &abs20, Rds
MOVA @Rs, Rds
MOVA @Rs+, Rds
MOVA Rs, z16(Rds)
MOVA Rs, &abs20

Operation

src → Rdst
Rs → dst

Description
The 20-bit source operand is moved to the 20-bit destination. The source operand is not affected. The previous content of the destination is lost.

Status Bits

N: Not affected
Z: Not affected
C: Not affected
V: Not affected

Mode Bits

OSCOFF, CPUOFF, and GIE are not affected.

Examples

Copy 20-bit value in R9 to R8

```
MOVA R9, R8 ; R9 -> R8
```

Write 20-bit immediate value 12345h to R12

```
MOVA #12345h, R12 ; 12345h -> R12
```

Copy 20-bit value addressed by (R9 + 100h) to R8. Source operand in addresses (R9 + 100h) LSBs and (R9 + 102h) MSBs.

```
MOVA 100h(R9), R8 ; Index: + 32 K. 2 words transferred
```

Move 20-bit value in 20-bit absolute addresses EDE (LSBs) and EDE+2 (MSBs) to R12

```
MOVA &EDE, R12 ; &EDE -> R12. 2 words transferred
```

Move 20-bit value in 20-bit addresses EDE (LSBs) and EDE+2 (MSBs) to R12. PC index ± 32 K.

```
MOVA EDE, R12 ; EDE -> R12. 2 words transferred
```

Copy 20-bit value R9 points to (20 bit address) to R8. Source operand in addresses @R9 LSBs and @R9+2 MSBs.

```
MOVA R9, R8 ; @R9 -> R8. 2 words transferred
```
Copy 20-bit value R9 points to (20 bit address) to R8. R9 is incremented by four afterwards. Source operand in addresses @R9 LSBs and @(R9 + 2) MSBs.

MOVA @R9+,R8 ; R9 -> R8. R9 + 4. 2 words transferred.

Copy 20-bit value in R8 to destination addressed by (R9 + 100h). Destination operand in addresses @(R9 + 100h) LSBs and @(R9 + 102h) MSBs.

MOVA R8,100h(R9) ; Index: +- 32 K. 2 words transferred

Move 20-bit value in R13 to 20-bit absolute addresses EDE (LSBs) and EDE+2 (MSBs)

MOVA R13,&EDE ; R13 -> EDE. 2 words transferred

Move 20-bit value in R13 to 20-bit addresses EDE (LSBs) and EDE+2 (MSBs). PC index ± 32 K.

MOVA R13,EDE ; R13 -> EDE. 2 words transferred
4.6.4.9 RETA

* RETA Return from subroutine

Syntax
RETA

Operation
@SP → PC.15:0 LSBs (15:0) of saved PC to PC.15:0
SP + 2 → SP

@SP → PC.19:16 MSBs (19:16) of saved PC to PC.19:16
SP + 2 → SP

Emulation
MOVA @SP+,PC

Description
The 20-bit return address information, pushed onto the stack by a CALLA instruction, is restored to the PC. The program continues at the address following the subroutine call. The SR bits SR.11:0 are not affected. This allows the transfer of information with these bits.

Status Bits
N: Not affected
Z: Not affected
C: Not affected
V: Not affected

Mode Bits
OSCOFF, CPUOFF, and GIE are not affected.

Example
Call a subroutine SUBR from anywhere in the 20-bit address space and return to the address after the CALLA

CALLA     #SUBR       ; Call subroutine starting at SUBR
SUBR       PUSHM.A   #2,R14      ; Save R14 and R13 (20 bit data)
...       ; Subroutine code
POPM.A    #2,R14      ; Restore R13 and R14 (20 bit data)
RETA                  ; Return (to full address space)
4.6.4.10 TSTA

* TSTA Test 20-bit destination register
Syntax TSTA Rdst
Operation dst + 0FFFFFh + 1
dst + 0FFFFh + 1
dst + 0FFh + 1
Emulation CMPA #0,Rdst
Description The destination register is compared with zero. The status bits are set according to the result. The destination register is not affected.
Status Bits
N: Set if destination register is negative, reset if positive
Z: Set if destination register contains zero, reset otherwise
C: Set
V: Reset
Mode Bits OSCOFF, CPUOFF, and GIE are not affected.
Example The 20-bit value in R7 is tested. If it is negative, continue at R7NEG; if it is positive but not zero, continue at R7POS.

```
TSTA R7          ; Test R7
JN R7NEG         ; R7 is negative
JZ R7ZERO        ; R7 is zero
R7POS ......     ; R7 is positive but not zero
R7NEG ......     ; R7 is negative
R7ZERO ......    ; R7 is zero
```
4.6.4.11 SUBA

**SUBA** Subtract 20-bit source from 20-bit destination register

**Syntax**

SUBA Rs, Rdst

SUBA #imm20, Rdst

**Operation**

\((\text{not}.\text{src}) + 1 + \text{Rdst} \rightarrow \text{Rdst} \) or \( \text{Rdst} - \text{src} \rightarrow \text{Rdst} \)

**Description**

The 20-bit source operand is subtracted from the 20-bit destination register. This is made by adding the 1's complement of the source + 1 to the destination. The result is written to the destination register, the source is not affected.

**Status Bits**

- **N:** Set if result is negative (src > dst), reset if positive (src ≤ dst)
- **Z:** Set if result is zero (src = dst), reset otherwise (src ≠ dst)
- **C:** Set if there is a carry from the MSB (Rdst.19), reset otherwise
- **V:** Set if the subtraction of a negative source operand from a positive destination operand delivers a negative result, or if the subtraction of a positive source operand from a negative destination operand delivers a positive result, reset otherwise (no overflow)

**Mode Bits**

OSCOFF, CPUOFF, and GIE are not affected.

**Example**

The 20-bit value in R5 is subtracted from R6. If a carry occurs, the program continues at label TONI.

```
SUBA   R5,R6      ; R6 - R5 -> R6
JC    TONI        ; Carry occurred
...               ; No carry
```
The basic clock module+ provides the clocks for MSP430x2xx devices. This chapter describes the operation of the basic clock module+ of the MSP430x2xx device family.

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5.3 Basic Clock Module+ Registers ................................................................. 285
5.1 Basic Clock Module+ Introduction

The basic clock module+ supports low system cost and ultra-low power consumption. Using three internal clock signals, the user can select the best balance of performance and low power consumption. The basic clock module+ can be configured to operate without any external components, with one external resistor, with one or two external crystals, or with resonators, under full software control.

The basic clock module+ includes two, three or four clock sources:

- LFXT1CLK: Low-frequency/high-frequency oscillator that can be used with low-frequency watch crystals or external clock sources of 32768 Hz or with standard crystals, resonators, or external clock sources in the 400-kHz to 16-MHz range.
- XT2CLK: Optional high-frequency oscillator that can be used with standard crystals, resonators, or external clock sources in the 400-kHz to 16-MHz range.
- DCOCLK: Internal digitally controlled oscillator (DCO).
- VLOCLK: Internal very low-power low-frequency oscillator with 12-kHz typical frequency.

Three clock signals are available from the basic clock module+:

- ACLK: Auxiliary clock. ACLK is software selectable as LFXT1CLK or VLOCLK. ACLK is divided by 1, 2, 4, or 8. ACLK is software selectable for individual peripheral modules.
- MCLK: Master clock. MCLK is software selectable as LFXT1CLK, VLOCLK, XT2CLK (if available on-chip), or DCOCLK. MCLK is divided by 1, 2, 4, or 8. MCLK is used by the CPU and system.
- SMCLK: Sub-main clock. SMCLK is software selectable as LFXT1CLK, VLOCLK, XT2CLK (if available on-chip), or DCOCLK. SMCLK is divided by 1, 2, 4, or 8. SMCLK is software selectable for individual peripheral modules.

Figure 5-1 shows the block diagram of the basic clock module+ in the MSP430F2xx devices.

Figure 5-2 shows the block diagram of the basic clock module+ in the MSP430AFE2xx devices.
Figure 5-1. Basic Clock Module Block Diagram – MSP430F2xx, MSP430G2xx
† Device-Specific Clock Variations

Not all clock features are available on all MSP430x2xx devices

MSP430G22x0: LFXT1 is not present, XT2 is not present, ROSC is not supported.
MSP430F20xx, MSP430G2xx1, MSP430G2xx2, MSP430G2xx3: LFXT1 does not support HF mode, XT2 is not present, ROSC is not supported.
MSP430x21x1: Internal LP/LF oscillator is not present, XT2 is not present, ROSC is not supported.
MSP430x21x2: XT2 is not present.
MSP430F22xx, MSP430G2x55, MSP430x23x0: XT2 is not present.

LFXT1 is not present in MSP430AFE2xx devices.
5.2 Basic Clock Module+ Operation

After a PUC, MCLK and SMCLK are sourced from DCOCLK at ~1.1 MHz (see the device-specific data sheet for parameters) and ACLK is sourced from LFXT1CLK in LF mode with an internal load capacitance of 6 pF.

Status register control bits SCG0, SCG1, OSCOFF, and CPUOFF configure the MSP430 operating modes and enable or disable portions of the basic clock module+ (see the System Resets, Interrupts and Operating Modes chapter). The DCOCTL, BCSCTL1, BCSCTL2, and BCSCTL3 registers configure the basic clock module+.

The basic clock module+ can be configured or reconfigured by software at any time during program execution, for example:

```
CLR.B &DCOCTL ; Select lowest DCOx
; and MODx settings
BIS.B #RSEL2+RSEL1+RSEL0,&BCSCTL1 ; Select range 7
BIS.B #DCO2+DCO1+DCO0,&DCOCTL ; Select max DCO tap
```

5.2.1 Basic Clock Module+ Features for Low-Power Applications

Conflicting requirements typically exist in battery-powered applications:
- Low clock frequency for energy conservation and time keeping
- High clock frequency for fast reaction to events and fast burst processing capability
- Clock stability over operating temperature and supply voltage

The basic clock module+ addresses the above conflicting requirements by allowing the user to select from the three available clock signals: ACLK, MCLK, and SMCLK. For optimal low-power performance, ACLK can be sourced from a low-power 32768-Hz watch crystal (if available), providing a stable time base for the system and low-power standby operation, or from the internal low-frequency oscillator when crystal-accurate time keeping is not required. The MCLK can be configured to operate from the on-chip DCO that can be activated when requested by interrupt-driven events. The SMCLK can be configured to operate from a crystal or the DCO, depending on peripheral requirements. A flexible clock distribution and divider system is provided to fine tune the individual clock requirements.

5.2.2 Internal Very-Low-Power Low-Frequency Oscillator (VLO)

The internal very-low-power low-frequency oscillator (VLO) provides a typical frequency of 12 kHz (see device-specific data sheet for parameters) without requiring a crystal. VLOCLK source is selected by setting LFXT1Sx = 10 when XTS = 0. The OSCOFF bit disables the VLO for LPM4. The LFXT1 crystal oscillators are disabled when the VLO is selected reducing current consumption. The VLO consumes no power when not being used.

Devices without LFXT1 (for example, the MSP430G22x0) should be configured to use the VLO as ACLK.

5.2.3 LFXT1 Oscillator

The LFXT1 oscillator is not implemented in the MSP430G22x0 device family.

The LFXT1 oscillator supports ultra-low current consumption using a 32768-Hz watch crystal in LF mode (XTS = 0). A watch crystal connects to XIN and XOUT without any other external components. The software-selectable XCAPx bits configure the internally provided load capacitance for the LFXT1 crystal in LF mode. This capacitance can be selected as 1 pF, 6 pF, 10 pF, or 12.5 pF typical. Additional external capacitors can be added if necessary.

The LFXT1 oscillator also supports high-speed crystals or resonators when in HF mode (XTS = 1, XCAPx = 00). The high-speed crystal or resonator connects to XIN and XOUT and requires external capacitors on both terminals. These capacitors should be sized according to the crystal or resonator specifications. When LFXT1 is in HF mode, the LFXT1Sx bits select the range of operation.

LFXT1 may be used with an external clock signal on the XIN pin in either LF or HF mode when LFXT1Sx = 11, OSCOFF = 0, and XCAPx = 00. When used with an external signal, the external frequency must meet the data sheet parameters for the chosen mode. When the input frequency is below the specified lower limit, the LFXT1OF bit may be set preventing the CPU from being clocked with LFXT1CLK.
Software can disable LFXT1 by setting OSCOFF, if LFXT1CLK does not source SMCLK or MCLK, as shown in Figure 5-3. LFXT1 is switched on if requested as source for ACLK (ACLK_request), MCLK (MCLK_request), or SMCLK (SMCLK_request) and not disabled by software.

**Figure 5-3. Off Signals for the LFXT1 Oscillator**

---

**Note**

**LFXT1 Oscillator Characteristics**

Low-frequency crystals often require hundreds of milliseconds to start up, depending on the crystal.

Ultra-low-power oscillators such as the LFXT1 in LF mode should be guarded from noise coupling from other sources. The crystal should be placed as close as possible to the MSP430 with the crystal housing grounded and the crystal traces guarded with ground traces.

---

**5.2.4 XT2 Oscillator**

Some devices have a second crystal oscillator, XT2. XT2 sources XT2CLK and its characteristics are identical to LFXT1 in HF mode. The XT2Sx bits select the range of operation of XT2. The XT2OFF bit disables the XT2 oscillator if XT2CLK is not used for MCLK (MCLK_request) or SMCLK (SMCLK_request) as shown in Figure 5-4.

XT2 may be used with external clock signals on the XT2IN pin when XT2Sx = 11 and XT2OFF = 0. When used with an external signal, the external frequency must meet the data sheet parameters for XT2. When the input frequency is below the specified lower limit, the XT2OF bit may be set to prevent the CPU from being clocked with XT2CLK.

**Figure 5-4. Off Signals for Oscillator XT2**
5.2.5 Digitally Controlled Oscillator (DCO)

The DCO is an integrated digitally controlled oscillator. The DCO frequency can be adjusted by software using the DCOx, MODx, and RSELx bits.

5.2.5.1 Disabling the DCO

Software can disable DCOCLK by setting SCG0 when it is not used to source SMCLK (SMCLK_request) or MCLK (MCLK_request) in active mode, as shown in Figure 5-5.

5.2.5.2 Adjusting the DCO Frequency

After a PUC, RSELx = 7 and DCOx = 3, allowing the DCO to start at a mid-range frequency. MCLK and SMCLK are sourced from DCOCLK. Because the CPU executes code from MCLK, which is sourced from the fast-starting DCO, code execution typically begins from PUC in less than 2 µs. The typical DCOx and RSELx ranges and steps are shown in Figure 5-6.

The frequency of DCOCLK is set by the following functions:

- The four RSELx bits select one of sixteen nominal frequency ranges for the DCO. These ranges are defined for an individual device in the device-specific data sheet.
- The three DCOx bits divide the DCO range selected by the RSELx bits into 8 frequency steps, separated by approximately 10%.
- The five MODx bits, switch between the frequency selected by the DCOx bits and the next higher frequency set by DCOx+1. When DCOx = 07h, the MODx bits have no effect because the DCO is already at the highest setting for the selected RSELx range.

Figure 5-6. Typical DCOx Range and RSELx Steps
Each MSP430F2xx device (and most MSP430G2xx devices; see device-specific data sheets) has calibrated DCOCTL and BCSCTL1 register settings for specific frequencies stored in information memory segment A. To use the calibrated settings, the information is copied into the DCOCTL and BCSCTL1 registers. The calibrated settings affect the DCOx, MODx, and RSELx bits, and clear all other bits, except XT2OFF which remains set. The remaining bits of BCSCTL1 can be set or cleared as needed with BIS.B or BIC.B instructions.

; Set DCO to 1 MHz:
CLR.B &DCOCTL ; Select lowest DCOx
; and MODx settings
MOV.B &CALBC1_1MHZ,&BCSCTL1 ; Set range
MOV.B &CALDCO_1MHZ,&DCOCTL ; Set DCO step + modulation

5.2.5.3 Using an External Resistor (ROSC) for the DCO

Some MSP430F2xx devices provide the option to source the DCO current through an external resistor, \( R_{OSC} \), tied to DV\(_{CC}\), when DCOR = 1. In this case, the DCO has the same characteristics as MSP430x1xx devices, and the RSELx setting is limited to 0 to 7 with the RSEL3 ignored. This option provides an additional method to tune the DCO frequency by varying the resistor value. See the device-specific data sheet for parameters.

5.2.6 DCO Modulator

The modulator mixes two DCO frequencies, \( f_{DCO} \) and \( f_{DCO+1} \) to produce an intermediate effective frequency between \( f_{DCO} \) and \( f_{DCO+1} \) and spread the clock energy, reducing electromagnetic interference (EMI). The modulator mixes \( f_{DCO} \) and \( f_{DCO+1} \) for 32 DCOCLK clock cycles and is configured with the MODx bits. When MODx = 0 the modulator is off.

The modulator mixing formula is:

\[
t = (32 – MODx) \times t_{DCO} + MODx \times t_{DCO+1}
\]

Because \( f_{DCO} \) is lower than the effective frequency and \( f_{DCO+1} \) is higher than the effective frequency, the error of the effective frequency integrates to zero. It does not accumulate. The error of the effective frequency is zero every 32 DCOCLK cycles. Figure 5-7 shows the modulator operation.

The modulator settings and DCO control are configured with software. The DCOCLK can be compared to a stable frequency of known value and adjusted with the DCOx, RSELx, and MODx bits. See [http://www.msp430.com](http://www.msp430.com) for application notes and example code on configuring the DCO.
5.2.7 Basic Clock Module+ Fail-Safe Operation

The basic clock module+ incorporates an oscillator-fault fail-safe feature. This feature detects an oscillator fault for LFXT1 and XT2 as shown in Figure 5-8. The available fault conditions are:

- Low-frequency oscillator fault (LFXT1OF) for LFXT1 in LF mode
- High-frequency oscillator fault (LFXT1OF) for LFXT1 in HF mode
- High-frequency oscillator fault (XT2OF) for XT2

The crystal oscillator fault bits LFXT1OF, and XT2OF are set if the corresponding crystal oscillator is turned on and not operating properly. The fault bits remain set as long as the fault condition exists and are automatically cleared if the enabled oscillators function normally.

The OFIFG oscillator fault flag is set and latched at POR or when an oscillator fault (LFXT1OF, or XT2OF) is detected. When OFIFG is set, MCLK is sourced from the DCO, and if OFIE is set, the OFIFG requests an NMI interrupt. When the interrupt is granted, the OFIE is reset automatically. The OFIFG flag must be cleared by software. The source of the fault can be identified by checking the individual fault bits.

If a fault is detected for the crystal oscillator sourcing the MCLK, the MCLK is automatically switched to the DCO for its clock source. This does not change the SELMx bit settings. This condition must be handled by user software.
5.2.7.1 Sourcing MCLK from a Crystal

After a PUC, the basic clock module+ uses DCOCLK for MCLK. If required, MCLK may be sourced from LFXT1 or XT2 (if it is available).

The sequence to switch the MCLK source from the DCO clock to the crystal clock (LFXT1CLK or XT2CLK) is:

1. Turn on the crystal oscillator and select the appropriate mode
2. Clear the OFIFG flag
3. Wait at least 50 µs
4. Test OFIFG, and repeat steps 2 through 4 until OFIFG remains cleared.

; Select LFXT1 (HF mode) for MCLK
BIC.W #OSCOFF,SR ; Turn on osc.
BIS.B #XTS,&BCSCTL1 ; HF mode
MOV.B #LFXT1S0,&BCSCTL3 ; 1-3MHz Crystal
L1 BIC.B #OFIFG,&IFG1 ; Clear OFIFG
MOV.W #0FFh,R15 ; Delay
L2 DEC.W R15 ;
JNZ L2 ; Repeat test if needed
BIT.B #OFIFG,&IFG1 ; Re-test OFIFG
JNZ L1 ;
BIS.B #SELM1+SELM0,&BCSCTL2 ; Select LFXT1CLK

5.2.8 Synchronization of Clock Signals

When switching MCLK or SMCLK from one clock source to another, the switch is synchronized to avoid critical race conditions as shown in Figure 5-9:

- The current clock cycle continues until the next rising edge.
- The clock remains high until the next rising edge of the new clock.
- The new clock source is selected and continues with a full high period.

![Switch MCLK from DCOCLK to LFXT1CLK](image-url)
5.3 Basic Clock Module+ Registers

Table 5-1 lists the memory-mapped registers for the Basic Clock Module+.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>56h</td>
<td>DCOCTL</td>
<td>DCO control</td>
<td>Read/write</td>
<td>60h with PUC</td>
<td>Section 5.3.1</td>
</tr>
<tr>
<td>57h</td>
<td>BCSCTL1</td>
<td>Basic clock system control 1</td>
<td>Read/write</td>
<td>87h with POR¹</td>
<td>Section 5.3.2</td>
</tr>
<tr>
<td>58h</td>
<td>BCSCTL2</td>
<td>Basic clock system control 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 5.3.3</td>
</tr>
<tr>
<td>53h</td>
<td>BCSCTL3</td>
<td>Basic clock system control 3</td>
<td>Read/write</td>
<td>05h with PUC²</td>
<td>Section 5.3.4</td>
</tr>
<tr>
<td>00h</td>
<td>IE1</td>
<td>SFR interrupt enable 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 5.3.5</td>
</tr>
<tr>
<td>02h</td>
<td>IFG1</td>
<td>SFR interrupt flag 1</td>
<td>Read/write</td>
<td>02h with PUC</td>
<td>Section 5.3.6</td>
</tr>
</tbody>
</table>

(1) Some of the register bits are POR initialized, and some are PUC initialized (see Section 5.3.2).
(2) The initial state of BCSCTL3 is 000h in the MSP430AFE2xx devices.
5.3.1 DCOCTL Register

DCO Control Register

DCOCTL is shown in Figure 5-10 and described in Table 5-2.

Return to Table 5-1.

Figure 5-10. DCOCTL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-5</td>
<td>DCOx</td>
<td>R/W</td>
<td>3h</td>
<td>DCO frequency select. These bits select which of the eight discrete DCO frequencies within the range defined by the RSELx setting is selected.</td>
</tr>
<tr>
<td>4-0</td>
<td>MODx</td>
<td>R/W</td>
<td>0h</td>
<td>Modulator selection. These bits define how often the f\textsubscript{DCO+1} frequency is used within a period of 32 DCOCLK cycles. During the remaining clock cycles (32 – MOD) the f\textsubscript{DCO} frequency is used. Not useable when DCOx = 7.</td>
</tr>
</tbody>
</table>
5.3.2 BCSCTL1 Register

Basic Clock System Control 1 Register

BCSCTL1 is shown in Figure 5-11 and described in Table 5-3.

Return to Table 5-1.

Figure 5-11. BCSCTL1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>XT2OFF</td>
<td>R/W</td>
<td>1h</td>
<td>XT2 off. This bit turns off the XT2 oscillator. 0b = XT2 is on 1b = XT2 is off if it is not used for MCLK or SMCLK.</td>
</tr>
<tr>
<td>6</td>
<td>XTS(^{(1)})(^{(2)})</td>
<td>R/W</td>
<td>0h</td>
<td>LFXT1 mode select 0b = Low-frequency mode 1b = High-frequency mode</td>
</tr>
<tr>
<td>5-4</td>
<td>DIVAx</td>
<td>R/W</td>
<td>0h</td>
<td>Divider for ACLK 00b = /1 01b = /2 10b = /4 11b = /8</td>
</tr>
<tr>
<td>3-0</td>
<td>RSELx</td>
<td>R/W</td>
<td>7h</td>
<td>Range select. Sixteen different frequency ranges are available. The lowest frequency range is selected by setting RSELx = 0. RSEL3 is ignored when DCOR = 1.</td>
</tr>
</tbody>
</table>

(1) XTS = 1 is not supported in MSP430x20xx and MSP430G2xx devices (see Figure 5-1 and Figure 5-2 for details on supported settings for all devices).

(2) This bit is reserved in the MSP430AFE2xx devices.

Table 5-3. BCSCTL1 Register Field Descriptions
5.3.3 BCSCTL2 Register

Basic Clock System Control 2 Register

BCSCTL2 is shown in Figure 5-12 and described in Table 5-4.

Return to Table 5-1.

Figure 5-12. BCSCTL2 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td>SELMx</td>
<td>R/W</td>
<td>0h</td>
<td>Select MCLK. These bits select the MCLK source. 0b = DCOCLK, 01b = XT2CLK when XT2 oscillator present on-chip, LFXT1CLK or VLOCLK when XT2 oscillator not present, 1b = LFXT1CLK or VLOCLK</td>
</tr>
<tr>
<td>5-4</td>
<td>DIVMx</td>
<td>R/W</td>
<td>0h</td>
<td>Divider for MCLK 00b = /1, 01b = /2, 10b = /4, 11b = /8</td>
</tr>
<tr>
<td>3</td>
<td>SELS</td>
<td>R/W</td>
<td>0h</td>
<td>Select SMCLK. This bit selects the SMCLK source. 0b = DCOCLK, 1b = XT2CLK when XT2 oscillator present, LFXT1CLK or VLOCLK when XT2 oscillator not present</td>
</tr>
<tr>
<td>2-1</td>
<td>DIVSx</td>
<td>R/W</td>
<td>0h</td>
<td>Divider for SMCLK 00b = /1, 01b = /2, 10b = /4, 11b = /8</td>
</tr>
<tr>
<td>0</td>
<td>DCOR(1) (2)</td>
<td>R/W</td>
<td>0h</td>
<td>DCO resistor select. Not available in all devices. See the device-specific data sheet. 0b = Internal resistor, 1b = External resistor</td>
</tr>
</tbody>
</table>

Table 5-4. BCSCTL2 Register Field Descriptions

(1) Does not apply to MSP430x20xx or MSP430x21xx devices.
(2) This bit is reserved in the MSP430AFE2xx devices.
5.3.4 BCSCTL3 Register

Basic Clock System Control 3 Register

BCSCTL3 is shown in Figure 5-13 and described in Table 5-5.

Return to Table 5-1.

Figure 5-13. BCSCTL3 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td>XT2Sx</td>
<td>R/W</td>
<td>0h</td>
<td>XT2 range select. These bits select the frequency range for XT2.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = 0.4- to 1-MHz crystal or resonator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = 1- to 3-MHz crystal or resonator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = 3- to 16-MHz crystal or resonator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = Digital external 0.4- to 16-MHz clock source</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>5-4</td>
<td>LFXT1Sx(1)</td>
<td>R/W</td>
<td>0h</td>
<td>Low-frequency clock select and LFXT1 range select. These bits select between</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>LFXT1 and VLO when XTS = 0, and select the frequency range for LFXT1 when</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>XTS = 1.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>MSP430G22x0: The LFXT1Sx bits should be programmed to 10b during the</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>initialization and start-up code to select VLOCLK (for more details refer</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>to Digital I/O chapter). The other bits are reserved and should not be</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>altered.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>When XTS = 0:</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = 32768-Hz crystal on LFXT1</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = Reserved</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = VLOCLK (Reserved in MSP430F21x1 devices)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = Digital external clock source</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>When XTS = 1 (Not applicable for MSP430F20xx, MSP430G2xx1, MSP430G2xx2,</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>MSP430G2xx3):</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = 0.4- to 1-MHz crystal or resonator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = 1- to 3-MHz crystal or resonator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = 3- to 16-MHz crystal or resonator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = Digital external 0.4- to 16-MHz clock source</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>LFXT1Sx definition for MSP430AFE2xx devices:</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = Reserved</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = Reserved</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = VLOCLK</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = Reserved</td>
</tr>
<tr>
<td>Bit</td>
<td>Field</td>
<td>Type</td>
<td>Reset</td>
<td>Description</td>
</tr>
<tr>
<td>-----</td>
<td>------------</td>
<td>------</td>
<td>-------</td>
<td>-----------------------------------------------------------------------------</td>
</tr>
<tr>
<td>3-2</td>
<td>XCAPx(2)</td>
<td>R/W</td>
<td>1h</td>
<td>Oscillator capacitor selection. These bits select the effective capacitance seen by the LFXT1 crystal when XTS = 0. If XTS = 1 or if LFXT1Sx = 11 XCAPx should be 00. This bit is reserved in the MSP430AFE2xx devices. 00b = Approximately 1 pF 01b = Approximately 6 pF 10b = Approximately 10 pF 11b = Approximately 12.5 pF</td>
</tr>
<tr>
<td>1</td>
<td>XT2OF(3)</td>
<td>R</td>
<td>0h</td>
<td>XT2 oscillator fault. Does not apply to MSP430x2xx, MSP430x21xx, or MSP430x22xx devices. 0b = No fault condition present 1b = Fault condition present</td>
</tr>
<tr>
<td>0</td>
<td>LFXT1OF(2)</td>
<td>R</td>
<td>1h</td>
<td>LFXT1 oscillator fault. This bit is reserved in the MSP430AFE2xx devices. 0b = No fault condition present 1b = Fault condition present</td>
</tr>
</tbody>
</table>

(1) MSP430G22x0: The LFXT1Sx bits should be programmed to 10b during the initialization and start-up code to select VLOCLK (for more details, refer to the Digital I/O chapter). The other bits are reserved and should not be altered.
(2) This bit is reserved in the MSP430AFE2xx devices.
(3) Does not apply to MSP430x2xx, MSP430x21xx, or MSP430x22xx devices.
5.3.5 IE1 Register
SFR Interrupt Enable 1 Register
IE1 is shown in Figure 5-14 and described in Table 5-6.
Return to Table 5-1.

**Figure 5-14. IE1 Register**

<table>
<thead>
<tr>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>OFIE</td>
<td></td>
</tr>
<tr>
<td>rw-0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Table 5-6. IE1 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-2</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See device-specific data sheet.</td>
</tr>
<tr>
<td>1</td>
<td>OFIE(1)</td>
<td>R/W</td>
<td>0h</td>
<td>Oscillator fault interrupt enable. This bit enables the OFIFG interrupt. Because other bits in IE1 may be used for other modules, it is recommended to set or clear this bit using BIS.B or BIC.B instructions, rather than MOV.B or CLR.B instructions. MSP430G22x0: This bit should not be set. 0b = Interrupt not enabled 1b = Interrupt enabled</td>
</tr>
<tr>
<td>0</td>
<td></td>
<td></td>
<td></td>
<td>This bit may be used by other modules. See device-specific data sheet.</td>
</tr>
</tbody>
</table>

(1) MSP430G22x0: This bit should not be set.

5.3.6 IFG1 Register
SFR Interrupt Flag 1 Register
IFG1 is shown in Figure 5-15 and described in Table 5-7.
Return to Table 5-1.

**Figure 5-15. IFG1 Register**

<table>
<thead>
<tr>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>OFIFG</td>
<td></td>
</tr>
<tr>
<td>rw-1</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Table 5-7. IFG1 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-2</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See device-specific data sheet.</td>
</tr>
<tr>
<td>1</td>
<td>OFIFG(1)</td>
<td>R/W</td>
<td>1h</td>
<td>Oscillator fault interrupt flag. Because other bits in IFG1 may be used for other modules, it is recommended to set or clear this bit using BIS.B or BIC.B instructions, rather than MOV.B or CLR.B instructions. 0b = Interrupt not pending 1b = Interrupt pending</td>
</tr>
</tbody>
</table>

(1)
Table 5-7. IFG1 Register Field Descriptions (continued)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td></td>
<td></td>
<td></td>
<td>This bit may be used by other modules. See device-specific data sheet.</td>
</tr>
</tbody>
</table>

(1) MSP430G22x0: The LFXT1 oscillator pins are not available in this device. The oscillator fault flag will always be set by hardware. The interrupt enable bit should not be set.
Chapter 6

DMA Controller

The DMA controller module transfers data from one address to another without CPU intervention. This chapter describes the operation of the DMA controller of the MSP430x2xx device family.

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6.2 DMA Operation .......................................................................................................................................................... 296
6.3 DMA Registers .......................................................................................................................................................... 308
6.1 DMA Introduction

The direct memory access (DMA) controller transfers data from one address to another, without CPU intervention, across the entire address range. For example, the DMA controller can move data from the ADC12 conversion memory to RAM.

Devices that contain a DMA controller may have one, two, or three DMA channels available. Therefore, depending on the number of DMA channels available, some features described in this chapter are not applicable to all devices.

Using the DMA controller can increase the throughput of peripheral modules. It can also reduce system power consumption by allowing the CPU to remain in a low-power mode without having to awaken to move data to or from a peripheral.

The DMA controller features include:

- Up to three independent transfer channels
- Configurable DMA channel priorities
- Requires only two MCLK clock cycles per transfer
- Byte or word and mixed byte/word transfer capability
- Block sizes up to 65535 bytes or words
- Configurable transfer trigger selections
- Selectable edge or level-triggered transfer
- Four addressing modes
- Single, block, or burst-block transfer modes

The DMA controller block diagram is shown in Figure 6-1.
DMA Controller Block Diagram

DMA Channel 0
- DMA0SA
- DMA0DA
- DMA0SZ

DMA Channel 1
- DMA1SA
- DMA1DA
- DMA1SZ

DMA Channel 2
- DMA2SA
- DMA2DA
- DMA2SZ

NMI Interrupt Request
JTAG Active
Halt

Figure 6-1. DMA Controller Block Diagram
6.2 DMA Operation

The DMA controller is configured with user software. The setup and operation of the DMA is discussed in the following sections.

6.2.1 DMA Addressing Modes

The DMA controller has four addressing modes. The addressing mode for each DMA channel is independently configurable. For example, channel 0 may transfer between two fixed addresses, while channel 1 transfers between two blocks of addresses. The addressing modes are shown in Figure 6-2. The addressing modes are:

- Fixed address to fixed address
- Fixed address to block of addresses
- Block of addresses to fixed address
- Block of addresses to block of addresses

The addressing modes are configured with the DMASRCINCRx and DMADSTINCRx control bits. The DMASRCINCRx bits select if the source address is incremented, decremented, or unchanged after each transfer. The DMADSTINCRx bits select if the destination address is incremented, decremented, or unchanged after each transfer.

Transfers may be byte-to-byte, word-to-word, byte-to-word, or word-to-byte. When transferring word-to-byte, only the lower byte of the source-word transfers. When transferring byte-to-word, the upper byte of the destination-word is cleared when the transfer occurs.

![Figure 6-2. DMA Addressing Modes](image-url)
6.2.2 DMA Transfer Modes

The DMA controller has six transfer modes selected by the DMADTx bits as listed in Table 6-1. Each channel is individually configurable for its transfer mode. For example, channel 0 may be configured in single transfer mode, while channel 1 is configured for burst-block transfer mode, and channel 2 operates in repeated block mode. The transfer mode is configured independently from the addressing mode. Any addressing mode can be used with any transfer mode.

Two types of data can be transferred selectable by the DMAxCTL DSTBYTE and SRCBYTE fields. The source and/or destination location can be either byte or word data. It is also possible to transfer byte to byte, word to word or any combination.

<table>
<thead>
<tr>
<th>DMADTx</th>
<th>Transfer Mode</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>000</td>
<td>Single transfer</td>
<td>Each transfer requires a trigger. DMAEN is automatically cleared when DMAxSZ transfers have been made.</td>
</tr>
<tr>
<td>001</td>
<td>Block transfer</td>
<td>A complete block is transferred with one trigger. DMAEN is automatically cleared at the end of the block transfer.</td>
</tr>
<tr>
<td>010, 011</td>
<td>Burst-block transfer</td>
<td>CPU activity is interleaved with a block transfer. DMAEN is automatically cleared at the end of the burst-block transfer.</td>
</tr>
<tr>
<td>100</td>
<td>Repeated single transfer</td>
<td>Each transfer requires a trigger. DMAEN remains enabled.</td>
</tr>
<tr>
<td>101</td>
<td>Repeated block transfer</td>
<td>A complete block is transferred with one trigger. DMAEN remains enabled.</td>
</tr>
<tr>
<td>110, 111</td>
<td>Repeated burst-block transfer</td>
<td>CPU activity is interleaved with a block transfer. DMAEN remains enabled.</td>
</tr>
</tbody>
</table>
### 6.2.2.1 Single Transfer

In single transfer mode, each byte/word transfer requires a separate trigger. The single transfer state diagram is shown in Figure 6-3.

The DMAxSZ register is used to define the number of transfers to be made. The DMADSTINCRx and DMASRCINCRx bits select if the destination address and the source address are incremented or decremented after each transfer. If DMAxSZ = 0, no transfers occur.

The DMAxSA, DMAxDA, and DMAxSZ registers are copied into temporary registers. The temporary values of DMAxSA and DMAxDA are incremented or decremented after each transfer. The DMAxSZ register is decremented after each transfer. When the DMAxSZ register decrements to zero it is reloaded from its temporary register and the corresponding DMAIFG flag is set. When DMADTx = 0, the DMAEN bit is cleared automatically when DMAxSZ decrements to zero and must be set again for another transfer to occur.

In repeated single transfer mode, the DMA controller remains enabled with DMAEN = 1, and a transfer occurs every time a trigger occurs.

![Figure 6-3. DMA Single Transfer State Diagram](image-url)
6.2.2.2 Block Transfers

In block transfer mode, a transfer of a complete block of data occurs after one trigger. When DMADTx = 1, the DMAEN bit is cleared after the completion of the block transfer and must be set again before another block transfer can be triggered. After a block transfer has been triggered, further trigger signals occurring during the block transfer are ignored. The block transfer state diagram is shown in Figure 6-4.

The DMAxSZ register is used to define the size of the block and the DMADSTINCRx and DMASRCINCRx bits select if the destination address and the source address are incremented or decremented after each transfer of the block. If DMAxSZ = 0, no transfers occur.

The DMAxSA, DMAxDA, and DMAxSZ registers are copied into temporary registers. The temporary values of DMAxSA and DMAxDA are incremented or decremented after each transfer in the block. The DMAxSZ register is decremented after each transfer of the block and shows the number of transfers remaining in the block. When the DMAxSZ register decrements to zero it is reloaded from its temporary register and the corresponding DMAIFG flag is set.

During a block transfer, the CPU is halted until the complete block has been transferred. The block transfer takes 2 x MCLK x DMAxSZ clock cycles to complete. CPU execution resumes with its previous state after the block transfer is complete.

In repeated block transfer mode, the DMAEN bit remains set after completion of the block transfer. The next trigger after the completion of a repeated block transfer triggers another block transfer.
Figure 6-4. DMA Block Transfer State Diagram
6.2.2.3 Burst-Block Transfers

In burst-block mode, transfers are block transfers with CPU activity interleaved. The CPU executes 2 MCLK cycles after every four byte/word transfers of the block resulting in 20% CPU execution capacity. After the burst-block, CPU execution resumes at 100% capacity and the DMAEN bit is cleared. DMAEN must be set again before another burst-block transfer can be triggered. After a burst-block transfer has been triggered, further trigger signals occurring during the burst-block transfer are ignored. The burst-block transfer state diagram is shown in Figure 6-5.

The DMAxSZ register is used to define the size of the block and the DMADSTINCRx and DMASRCINCRx bits select if the destination address and the source address are incremented or decremented after each transfer of the block. If DMAxSZ = 0, no transfers occur.

The DMAxSA, DMAxDA, and DMAxSZ registers are copied into temporary registers. The temporary values of DMAxSA and DMAxDA are incremented or decremented after each transfer in the block. The DMAxSZ register is decremented after each transfer of the block and shows the number of transfers remaining in the block. When the DMAxSZ register decrements to zero it is reloaded from its temporary register and the corresponding DMAIFG flag is set.

In repeated burst-block mode the DMAEN bit remains set after completion of the burst-block transfer and no further trigger signals are required to initiate another burst-block transfer. Another burst-block transfer begins immediately after completion of a burst-block transfer. In this case, the transfers must be stopped by clearing the DMAEN bit, or by an NMI interrupt when ENNMI is set. In repeated burst-block mode the CPU executes at 20% capacity continuously until the repeated burst-block transfer is stopped.
Figure 6-5. DMA Burst-Block Transfer State Diagram
6.2.3 Initiating DMA Transfers

Each DMA channel is independently configured for its trigger source with the DMAxTSELx bits as described in Table 6-2. The DMAxTSELx bits should be modified only when the DMACTLx DMAEN bit is 0. Otherwise, unpredictable DMA triggers may occur.

When selecting the trigger, the trigger must not have already occurred, or the transfer will not take place. For example, if the TACCR2 CCIFG bit is selected as a trigger, and it is already set, no transfer will occur until the next time the TACCR2 CCIFG bit is set.

6.2.3.1 Edge-Sensitive Triggers

When DMALEVEL = 0, edge-sensitive triggers are used and the rising edge of the trigger signal initiates the transfer. In single-transfer mode, each transfer requires its own trigger. When using block or burst-block modes, only one trigger is required to initiate the block or burst-block transfer.

6.2.3.2 Level-Sensitive Triggers

When DMALEVEL = 1, level-sensitive triggers are used. For proper operation, level-sensitive triggers can only be used when external trigger DMAE0 is selected as the trigger. DMA transfers are triggered as long as the trigger signal is high and the DMAEN bit remains set.

The trigger signal must remain high for a block or burst-block transfer to complete. If the trigger signal goes low during a block or burst-block transfer, the DMA controller is held in its current state until the trigger goes back high or until the DMA registers are modified by software. If the DMA registers are not modified by software, when the trigger signal goes high again, the transfer resumes from where it was when the trigger signal went low.

When DMALEVEL = 1, transfer modes selected when DMADTx = {0, 1, 2, 3} are recommended because the DMAEN bit is automatically reset after the configured transfer.

6.2.3.3 Halting Executing Instructions for DMA Transfers

The DMAONFETCH bit controls when the CPU is halted for a DMA transfer. When DMAONFETCH = 0, the CPU is halted immediately and the transfer begins when a trigger is received. When DMAONFETCH = 1, the CPU finishes the currently executing instruction before the DMA controller halts the CPU and the transfer begins.

Note

DMAONFETCH Must Be Used When The DMA Writes To Flash

If the DMA controller is used to write to flash memory, the DMAONFETCH bit must be set. Otherwise, unpredictable operation can result.

<table>
<thead>
<tr>
<th>DMAxTSELx</th>
<th>Operation</th>
</tr>
</thead>
<tbody>
<tr>
<td>0000</td>
<td>A transfer is triggered when the DMAREQ bit is set. The DMAREQ bit is automatically reset when the transfer starts</td>
</tr>
<tr>
<td>0001</td>
<td>A transfer is triggered when the TACCR2 CCIFG flag is set. The TACCR2 CCIFG flag is automatically reset when the transfer starts. If the TACCR2 CCIE bit is set, the TACCR2 CCIFG flag will not trigger a transfer.</td>
</tr>
<tr>
<td>0010</td>
<td>A transfer is triggered when the TBCCR2 CCIFG flag is set. The TBCCR2 CCIFG flag is automatically reset when the transfer starts. If the TBCCR2 CCIE bit is set, the TBCCR2 CCIFG flag will not trigger a transfer.</td>
</tr>
<tr>
<td>0011</td>
<td>A transfer is triggered when serial interface receives new data. Devices with USCI_A0 module: A transfer is triggered when USCI_A0 receives new data. UCA0RXIFG is automatically reset when the transfer starts. If UCA0RXIE is set, the UCA0RXIFG flag will not trigger a transfer.</td>
</tr>
<tr>
<td>0100</td>
<td>A transfer is triggered when serial interface is ready to transmit new data. Devices with USCI_A0 module: A transfer is triggered when USCI_A0 is ready to transmit new data. UCA0TXIFG is automatically reset when the transfer starts. If UCA0TXIE is set, the UCA0TXIFG flag will not trigger a transfer.</td>
</tr>
<tr>
<td>0101</td>
<td>A transfer is triggered when the DAC12_0CTL DAC12IFG flag is set. The DAC12_0CTL DAC12IFG flag is automatically cleared when the transfer starts. If the DAC12_0CTL DAC12IE bit is set, the DAC12_0CTL DAC12IFG flag will not trigger a transfer.</td>
</tr>
<tr>
<td>DMAxTSELx</td>
<td>Operation</td>
</tr>
<tr>
<td>----------</td>
<td>----------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------</td>
</tr>
<tr>
<td>0110</td>
<td>A transfer is triggered by an ADC12IFGx flag. When single-channel conversions are performed, the corresponding ADC12IFGx is the trigger. When sequences are used, the ADC12IFGx for the last conversion in the sequence is the trigger. A transfer is triggered when the conversion is completed and the ADC12IFGx is set. Setting the ADC12IFGx with software will not trigger a transfer. All ADC12IFGx flags are automatically reset when the associated ADC12MEMx register is accessed by the DMA controller.</td>
</tr>
<tr>
<td>0111</td>
<td>A transfer is triggered when the TACCR0 CCIFG flag is set. The TACCR0 CCIFG flag is automatically reset when the transfer starts. If the TACCR0 CCIE bit is set, the TACCR0 CCIFG flag will not trigger a transfer.</td>
</tr>
<tr>
<td>1000</td>
<td>A transfer is triggered when the TBCCR0 CCIFG flag is set. The TBCCR0 CCIFG flag is automatically reset when the transfer starts. If the TBCCR0 CCIE bit is set, the TBCCR0 CCIFG flag will not trigger a transfer.</td>
</tr>
<tr>
<td>1001</td>
<td>A transfer is triggered when the UCA1RXIFG flag is set. UCA1RXIFG is automatically reset when the transfer starts. If URXIE1 is set, the UCA1RXIFG flag will not trigger a transfer.</td>
</tr>
<tr>
<td>1010</td>
<td>A transfer is triggered when the UCA1TXIFG flag is set. UCA1TXIFG is automatically reset when the transfer starts. If UTXIE1 is set, the UCA1TXIFG flag will not trigger a transfer.</td>
</tr>
<tr>
<td>1011</td>
<td>A transfer is triggered when the hardware multiplier is ready for a new operand.</td>
</tr>
<tr>
<td>1100</td>
<td>No transfer is triggered. Devices with USCI_B0 module: A transfer is triggered when USCI_B0 receives new data. UCB0RXIFG is automatically reset when the transfer starts. If UCB0RXIE is set, the UCB0RXIFG flag will not trigger a transfer.</td>
</tr>
<tr>
<td>1101</td>
<td>No transfer is triggered. Devices with USCI_B0 module: A transfer is triggered when USCI_B0 is ready to transmit new data. UCB0TXIFG is automatically reset when the transfer starts. If UCB0TXIE is set, the UCB0TXIFG flag will not trigger a transfer.</td>
</tr>
<tr>
<td>1110</td>
<td>A transfer is triggered when the DMAxIFG flag is set. DMA0IFG triggers channel 1, DMA1IFG triggers channel 2, and DMA2IFG triggers channel 0. None of the DMAxIFG flags are automatically reset when the transfer starts.</td>
</tr>
<tr>
<td>1111</td>
<td>A transfer is triggered by the external trigger DMAE0.</td>
</tr>
</tbody>
</table>

### 6.2.4 Stopping DMA Transfers

There are two ways to stop DMA transfers in progress:

- A single, block, or burst-block transfer may be stopped with an NMI interrupt, if the ENNMI bit is set in register DMACTL1.
- A burst-block transfer may be stopped by clearing the DMAEN bit.
6.2.5 DMA Channel Priorities

The default DMA channel priorities are DMA0-DMA1-DMA2. If two or three triggers happen simultaneously or are pending, the channel with the highest priority completes its transfer (single, block or burst-block transfer) first, then the second priority channel, then the third priority channel. Transfers in progress are not halted if a higher priority channel is triggered. The higher priority channel waits until the transfer in progress completes before starting.

The DMA channel priorities are configurable with the ROUNDROBIN bit. When the ROUNDROBIN bit is set, the channel that completes a transfer becomes the lowest priority. The order of the priority of the channels always stays the same, DMA0-DMA1-DMA2 (see Table 6-3).

<table>
<thead>
<tr>
<th>DMA Priority</th>
<th>Transfer Occurs</th>
<th>New DMA Priority</th>
</tr>
</thead>
<tbody>
<tr>
<td>DMA0 - DMA1 - DMA2</td>
<td>DMA1</td>
<td>DMA2 - DMA0 - DMA1</td>
</tr>
<tr>
<td>DMA2 - DMA0 - DMA1</td>
<td>DMA2</td>
<td>DMA0 - DMA1 - DMA2</td>
</tr>
<tr>
<td>DMA0 - DMA1 - DMA2</td>
<td>DMA0</td>
<td>DMA1 - DMA2 - DMA0</td>
</tr>
</tbody>
</table>

When the ROUNDROBIN bit is cleared the channel priority returns to the default priority.

6.2.6 DMA Transfer Cycle Time

The DMA controller requires one or two MCLK clock cycles to synchronize before each single transfer or complete block or burst-block transfer. Each byte/word transfer requires two MCLK cycles after synchronization, and one cycle of wait time after the transfer. Because the DMA controller uses MCLK, the DMA cycle time is dependent on the MSP430 operating mode and clock system setup.

If the MCLK source is active, but the CPU is off, the DMA controller will use the MCLK source for each transfer, without re-enabling the CPU. If the MCLK source is off, the DMA controller will temporarily restart MCLK, sourced with DCOCLK, for the single transfer or complete block or burst-block transfer. The CPU remains off, and after the transfer completes, MCLK is turned off. The maximum DMA cycle time for all operating modes is shown in Table 6-4.

<table>
<thead>
<tr>
<th>CPU Operating Mode</th>
<th>Clock Source</th>
<th>Maximum DMA Cycle Time</th>
</tr>
</thead>
<tbody>
<tr>
<td>Active mode</td>
<td>MCLK = DCOCLK</td>
<td>4 MCLK cycles</td>
</tr>
<tr>
<td>Active mode</td>
<td>MCLK = LFXT1CLK</td>
<td>4 MCLK cycles</td>
</tr>
<tr>
<td>Low-power mode LPM0/1</td>
<td>MCLK = DCOCLK</td>
<td>5 MCLK cycles</td>
</tr>
<tr>
<td>Low-power mode LPM3/4</td>
<td>MCLK = DCOCLK</td>
<td>5 MCLK cycles + 6 µs[1]</td>
</tr>
<tr>
<td>Low-power mode LPM0/1</td>
<td>MCLK = LFXT1CLK</td>
<td>5 MCLK cycles</td>
</tr>
<tr>
<td>Low-power mode LPM3</td>
<td>MCLK = LFXT1CLK</td>
<td>5 MCLK cycles</td>
</tr>
<tr>
<td>Low-power mode LPM4</td>
<td>MCLK = LFXT1CLK</td>
<td>5 MCLK cycles + 6 µs[1]</td>
</tr>
</tbody>
</table>

(1) The additional 6 µs are needed to start the DCOCLK. It is the $I_{(LPMx)}$ parameter in the data sheet.

6.2.7 Using DMA With System Interrupts

DMA transfers are not interruptible by system interrupts. System interrupts remain pending until the completion of the transfer. NMI interrupts can interrupt the DMA controller if the ENNMI bit is set.

System interrupt service routines are interrupted by DMA transfers. If an interrupt service routine or other routine must execute with no interruptions, the DMA controller should be disabled prior to executing the routine.
6.2.8 DMA Controller Interrupts

Each DMA channel has its own DMAIFG flag. Each DMAIFG flag is set in any mode, when the corresponding DMAxSZ register counts to zero. If the corresponding DMAIE and GIE bits are set, an interrupt request is generated.

All DMAIFG flags source only one DMA controller interrupt vector and, on some devices, the interrupt vector may be shared with other modules. Please refer to the device specific datasheet for further details. For these devices, software must check the DMAIFG and respective module flags to determine the source of the interrupt. The DMAIFG flags are not reset automatically and must be reset by software.

Additionally, some devices utilize the DMAIV register. All DMAIFG flags are prioritized, with DMA0IFG being the highest, and combined to source a single interrupt vector. The highest priority enabled interrupt generates a number in the DMAIV register. This number can be evaluated or added to the program counter to automatically enter the appropriate software routine. Disabled DMA interrupts do not affect the DMAIV value.

Any access, read or write, of the DMAIV register automatically resets the highest pending interrupt flag. If another interrupt flag is set, another interrupt is immediately generated after servicing the initial interrupt. For example, assume that DMA0 has the highest priority. If the DMA0IFG and DMA2IFG flags are set when the interrupt service routine accesses the DMAIV register, DMA0IFG is reset automatically. After the RETI instruction of the interrupt service routine is executed, the DMA2IFG will generate another interrupt.

The following software example shows the recommended use of DMAIV and the handling overhead. The DMAIV value is added to the PC to automatically jump to the appropriate routine.

The numbers at the right margin show the necessary CPU cycles for each instruction. The software overhead for different interrupt sources includes interrupt latency and return-from-interrupt cycles, but not the task handling itself.

**Example 6-1. DMAIV Software Example**

```
; Interrupt handler for DMA0IFG, DMA1IFG, DMA2IFG
DMA_HND  ... ; Interrupt latency 6
ADD   &DMAIV,PC ; Add offset to Jump table 3
JMP   DMA0_HND ; Vector 0: No interrupt 5
JMP   DMA1_HND ; Vector 4: DMA channel 1 2
JMP   DMA2_HND ; Vector 6: DMA channel 2 2
RETI ; Vector 8: Reserved 5
RETI ; Vector 10: Reserved 5
RETI ; Vector 12: Reserved 5
RETI ; Vector 14: Reserved 5
DMA2_HND ; Vector 6: DMA channel 2 2
... ; Task starts here
RETI ; Back to main program 5
DMA1_HND ; Vector 4: DMA channel 1 1
... ; Task starts here
RETI ; Back to main program 5
DMA0_HND ; Vector 2: DMA channel 0 2
... ; Task starts here
RETI ; Back to main program 5
```

6.2.9 Using the USCI_B I2C Module with the DMA Controller

The USCI_B I2C module provides two trigger sources for the DMA controller. The USCI_B I2C module can trigger a transfer when new I2C data is received and when data is needed for transmit.

A transfer is triggered if UCB0RXIFG is set. The UCB0RXIFG is cleared automatically when the DMA controller acknowledges the transfer. If UCB0RXIE is set, UCB0RXIFG will not trigger a transfer.

A transfer is triggered if UCB0TXIFG is set. The UCB0TXIFG is cleared automatically when the DMA controller acknowledges the transfer. If UCB0TXIE is set, UCB0TXIFG will not trigger a transfer.
6.2.10 Using ADC12 with the DMA Controller

MSP430 devices with an integrated DMA controller can automatically move data from any ADC12MEMx register to another location. DMA transfers are done without CPU intervention and independently of any low-power modes. The DMA controller increases throughput of the ADC12 module, and enhances low-power applications allowing the CPU to remain off while data transfers occur.

DMA transfers can be triggered from any ADC12IFGx flag. When CONSEQx = \{0,2\} the ADC12IFGx flag for the ADC12MEMx used for the conversion can trigger a DMA transfer. When CONSEQx = \{1,3\}, the ADC12IFGx flag for the last ADC12MEMx in the sequence can trigger a DMA transfer. Any ADC12IFGx flag is automatically cleared when the DMA controller accesses the corresponding ADC12MEMx.

6.2.11 Using DAC12 With the DMA Controller

MSP430 devices with an integrated DMA controller can automatically move data to the DAC12_xDAT register. DMA transfers are done without CPU intervention and independently of any low-power modes. The DMA controller increases throughput to the DAC12 module, and enhances low-power applications allowing the CPU to remain off while data transfers occur.

Applications requiring periodic waveform generation can benefit from using the DMA controller with the DAC12. For example, an application that produces a sinusoidal waveform may store the sinusoid values in a table. The DMA controller can continuously and automatically transfer the values to the DAC12 at specific intervals creating the sinusoid with zero CPU execution. The DAC12_xCTL DAC12IFG flag is automatically cleared when the DMA controller accesses the DAC12_xDAT register.

6.2.12 Writing to Flash With the DMA Controller

MSP430 devices with an integrated DMA controller can automatically move data to the Flash memory. DMA transfers are done without CPU intervention and independent of any low-power modes. The DMA controller performs the move of the data word/byte to the Flash. The write timing control is done by the Flash controller. Write transfers to the Flash memory succeed if the Flash controller is set up prior to the DMA transfer and if the Flash is not busy. To set up the Flash controller for write accesses, see the Flash Memory Controller chapter.
6.3 DMA Registers

Table 6-5 lists the memory-mapped registers for the DMA. All register offset addresses not listed in Table 6-5 should be considered as reserved locations and the register contents should not be modified.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>122h</td>
<td>DMACTL0</td>
<td>DMA control 0</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 6.3.1</td>
</tr>
<tr>
<td>124h</td>
<td>DMACTL1</td>
<td>DMA control 1</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 6.3.2</td>
</tr>
<tr>
<td>126h</td>
<td>DMAIV</td>
<td>DMA interrupt vector</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 6.3.7</td>
</tr>
<tr>
<td>1D0h</td>
<td>DMA0CTL</td>
<td>DMA channel 0 control</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 6.3.3</td>
</tr>
<tr>
<td>1D2h</td>
<td>DMA0SA</td>
<td>DMA channel 0 source address</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 6.3.4</td>
</tr>
<tr>
<td>1D6h</td>
<td>DMA0DA</td>
<td>DMA channel 0 destination address</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 6.3.5</td>
</tr>
<tr>
<td>1DAh</td>
<td>DMA0SZ</td>
<td>DMA channel 0 transfer size</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 6.3.6</td>
</tr>
<tr>
<td>1DCh</td>
<td>DMA1CTL</td>
<td>DMA channel 1 control</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 6.3.3</td>
</tr>
<tr>
<td>1DEh</td>
<td>DMA1SA</td>
<td>DMA channel 1 source address</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 6.3.4</td>
</tr>
<tr>
<td>1E2h</td>
<td>DMA1DA</td>
<td>DMA channel 1 destination address</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 6.3.5</td>
</tr>
<tr>
<td>1E6h</td>
<td>DMA1SZ</td>
<td>DMA channel 1 transfer size</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 6.3.6</td>
</tr>
<tr>
<td>1E8h</td>
<td>DMA2CTL</td>
<td>DMA channel 2 control</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 6.3.3</td>
</tr>
<tr>
<td>1EAh</td>
<td>DMA2SA</td>
<td>DMA channel 2 source address</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 6.3.4</td>
</tr>
<tr>
<td>1EEh</td>
<td>DMA2DA</td>
<td>DMA channel 2 destination address</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 6.3.5</td>
</tr>
<tr>
<td>1F2h</td>
<td>DMA2SZ</td>
<td>DMA channel 2 transfer size</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 6.3.6</td>
</tr>
</tbody>
</table>
6.3.1 DMACTL0 Register

DMA Control 0 Register

DMACTL0 is shown in Figure 6-6 and described in Table 6-6.

Return to Table 6-5.

Figure 6-6. DMACTL0 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-12</td>
<td>Reserved</td>
<td>R/W</td>
<td>0h</td>
<td>DMA trigger select. These bits select the DMA transfer trigger. 0000b = DMAREQ bit (software trigger) 0001b = TACCR2 CCIFG bit 0010b = TBCCR2 CCIFG bit 0011b = Serial data received UCA0RXIFG 0100b = Serial data transmit ready UCA0TXIFG 0101b = DAC12_0CTL DAC12IFG bit 0110b = ADC12 ADC12IFGx bit 0111b = TACCR0 CCIFG bit 1000b = TBCCR0 CCIFG bit 1001b = Serial data received UCA1RXIFG 1010b = Serial data transmit ready UCA1TXIFG 1011b = Multiplier ready 1100b = Serial data received UCB0RXIFG 1101b = Serial data transmit ready UCB0TXIFG 1110b = DMA0IFG bit triggers DMA channel 1 DMA1IFG bit triggers DMA channel 2 DMA2IFG bit triggers DMA channel 0 1111b = External trigger DMAE0</td>
</tr>
<tr>
<td>11-8</td>
<td>DMA2TSELx</td>
<td>R/W</td>
<td>0h</td>
<td>Same as DMA2TSELx</td>
</tr>
<tr>
<td>7-4</td>
<td>DMA1TSELx</td>
<td>R/W</td>
<td>0h</td>
<td>Same as DMA2TSELx</td>
</tr>
<tr>
<td>3-0</td>
<td>DMA0TSELx</td>
<td>R/W</td>
<td>0h</td>
<td>Same as DMA2TSELx</td>
</tr>
</tbody>
</table>
6.3.2 DMACTL1 Register

DMA Control 1 Register

DMACTL1 is shown in Figure 6-7 and described in Table 6-7.

Return to Table 6-5.

Figure 6-7. DMACTL1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-3</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td></td>
</tr>
<tr>
<td>2</td>
<td>DMAONFETCH</td>
<td>R/W</td>
<td>0h</td>
<td>DMA on fetch</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = The DMA transfer occurs immediately.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = The DMA transfer occurs on next instruction fetch after the trigger.</td>
</tr>
<tr>
<td>1</td>
<td>ROUNDROBIN</td>
<td>R/W</td>
<td>0h</td>
<td>Round robin. This bit enables the round-robin DMA channel priorities.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = DMA channel priority is DMA0, DMA1, DMA2</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = DMA channel priority changes with each transfer</td>
</tr>
<tr>
<td>0</td>
<td>ENNMI</td>
<td>R/W</td>
<td>0h</td>
<td>Enable NMI. This bit enables the interruption of a DMA transfer by an NMI interrupt. When an NMI interrupts a DMA transfer, the current transfer is completed normally, further transfers are stopped, and DMAABORT is set.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = NMI interrupt does not interrupt DMA transfer</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = NMI interrupt interrupts a DMA transfer</td>
</tr>
</tbody>
</table>

Table 6-7. DMACTL1 Register Field Descriptions
### 6.3.3 DMAxCTL Register

DMA Channel x Control Register

DMAxCTL are shown in Figure 6-8 and described in Table 6-8.

Return to Table 6-5.

#### Figure 6-8. DMAxCTL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15</td>
<td>Reserved</td>
<td>R/W</td>
<td>0h</td>
<td>Reserved</td>
</tr>
<tr>
<td>14-12</td>
<td>DMADTx</td>
<td>R/W</td>
<td>0h</td>
<td>DMA transfer mode. This bit selects automatic incrementing or decrementing of the destination address after each byte or word transfer. When DMADSTBYTE = 1, the destination address increments or decrements by one. When DMADSTBYTE = 0, the destination address increments or decrements by two. The DMAxDA is copied into a temporary register and the temporary register is incremented or decremented. DMAxDA is not incremented or decremented. 000b = Single transfer 001b = Block transfer 010b = Burst-block transfer 011b = Burst-block transfer 100b = Repeated single transfer 101b = Repeated block transfer 110b = Repeated burst-block transfer 111b = Repeated burst-block transfer</td>
</tr>
<tr>
<td>11-10</td>
<td>DMADSTINCRx</td>
<td>R/W</td>
<td>0h</td>
<td>DMA destination increment. This bit selects automatic incrementing or decrementing of the destination address after each byte or word transfer. When DMADSTBYTE = 1, the destination address increments or decrements by one. When DMADSTBYTE = 0, the destination address increments or decrements by two. The DMAxDA is copied into a temporary register and the temporary register is incremented or decremented. DMAxDA is not incremented or decremented. 00b = Destination address is unchanged 01b = Destination address is unchanged 10b = Destination address is decremented 11b = Destination address is incremented</td>
</tr>
<tr>
<td>9-8</td>
<td>DMASRCINCRx</td>
<td>R/W</td>
<td>0h</td>
<td>DMA source increment. This bit selects automatic incrementing or decrementing of the source address for each byte or word transfer. When DMASRCBYTE = 1, the source address increments or decrements by one. When DMASRCBYTE = 0, the source address increments or decrements by two. The DMAxSA is copied into a temporary register and the temporary register is incremented or decremented. DMAxSA is not incremented or decremented. 00b = Source address is unchanged 01b = Source address is unchanged 10b = Source address is decremented 11b = Source address is incremented</td>
</tr>
<tr>
<td>7</td>
<td>DMADSTBYTE</td>
<td>R/W</td>
<td>0h</td>
<td>DMA destination byte. This bit selects the destination as a byte or word. 0b = Word 1b = Byte</td>
</tr>
<tr>
<td>Bit</td>
<td>Field</td>
<td>Type</td>
<td>Reset</td>
<td>Description</td>
</tr>
<tr>
<td>-----</td>
<td>--------------</td>
<td>------</td>
<td>-------</td>
<td>-----------------------------------------------------------------------------</td>
</tr>
<tr>
<td>6</td>
<td>DMASRCBYTE</td>
<td>R/W</td>
<td>0h</td>
<td>DMA source byte. This bit selects the source as a byte or word.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Word</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Byte</td>
</tr>
<tr>
<td>5</td>
<td>DMALEVEL</td>
<td>R/W</td>
<td>0h</td>
<td>DMA level. This bit selects between edge-sensitive and level-sensitive triggers.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Edge sensitive (rising edge)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Level sensitive (high level)</td>
</tr>
<tr>
<td>4</td>
<td>DMAEN</td>
<td>R/W</td>
<td>0h</td>
<td>DMA enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Enabled</td>
</tr>
<tr>
<td>3</td>
<td>DMAIFG</td>
<td>R/W</td>
<td>0h</td>
<td>DMA interrupt flag</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>2</td>
<td>DMAIE</td>
<td>R/W</td>
<td>0h</td>
<td>DMA interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Enabled</td>
</tr>
<tr>
<td>1</td>
<td>DMAABORT</td>
<td>R/W</td>
<td>0h</td>
<td>DMA abort. This bit indicates if a DMA transfer was interrupted by an NMI.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = DMA transfer not interrupted</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = DMA transfer was interrupted by NMI</td>
</tr>
<tr>
<td>0</td>
<td>DMAREQ</td>
<td>R/W</td>
<td>0h</td>
<td>DMA request. Software-controlled DMA start. DMAREQ is reset automatically.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No DMA start</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Start DMA</td>
</tr>
</tbody>
</table>
6.3.4 DMAxSA Register

DMA Channel x Source Address Register

DMAxSA is shown in Figure 6-9 and described in Table 6-9.

Return to Table 6-5.

Figure 6-9. DMAxSA Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
</table>
| 15-0 | DMAxSA | R/W | Unchanged |DMA source address
The source address register points to the DMA source address for single transfers or the first source address for block transfers. The source address register remains unchanged during block and burst-block transfers.
Devices that have addressable memory range 64KB or less contain a single word for the DMAxSA. The upper word is automatically cleared when writing using word operations. Reads from this location are always read as zero.
Devices that have addressable memory range that is more than 64KB contain an additional word for the source address. Bits 15-4 of this additional word are reserved and always read as zero.
When writing to DMAxSA with word formats, this additional word is automatically cleared. Reads of this additional word using word formats are always read as zero. |
6.3.5 DMAxDA Register

DMA Channel x Destination Address Register

DMAxDA is shown in Figure 6-10 and described in Table 6-10.

Return to Table 6-5.

Figure 6-10. DMAxDA Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
</table>
| 15-0 | DMAxDA | R/W  | Unchanged | DMA destination address
The destination address register points to the DMA destination address for single transfers or the first destination address for block transfers. The destination address register remains unchanged during block and burst-block transfers.
Devices that have addressable memory range 64KB or less contain a single word for the DMAxDA.
Devices that have addressable memory range that is more than 64KB contain an additional word for the destination address. Bits 15-4 of this additional word are reserved and always read as zero.
When writing to DMAxDA with word formats, this additional word is automatically cleared. Reads of this additional word using word formats are always read as zero. |
6.3.6 DMAxSZ Register

DMA Channel x Size Register

DMAxSZ is shown in Figure 6-11 and described in Table 6-11.
Return to Table 6-5.

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>DMAxSZ</td>
<td>R/W</td>
<td>Unchanged</td>
<td>DMA size. The DMA size register defines the number of byte/word data per block transfer. DMAxSZ register decrements with each word or byte transfer. When DMAxSZ decrements to 0, it is immediately and automatically reloaded with its previously initialized value. 00000h = Transfer is disabled 00001h = 1 byte or word to be transferred 00002h = 2 bytes or words have to be transferred ⋮ 0FFFFh = 65535 bytes or words have to be transferred</td>
</tr>
</tbody>
</table>
6.3.7 DMAIV Register

DMA Interrupt Vector Register

DMAIV is shown in Figure 6-12 and described in Table 6-12.

Return to Table 6-5.

Figure 6-12. DMAIV Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>DMAIVx</th>
</tr>
</thead>
<tbody>
<tr>
<td>15</td>
<td>r-0</td>
</tr>
<tr>
<td>14</td>
<td>r-0</td>
</tr>
<tr>
<td>13</td>
<td>r-0</td>
</tr>
<tr>
<td>12</td>
<td>r-0</td>
</tr>
<tr>
<td>11</td>
<td>r-0</td>
</tr>
<tr>
<td>10</td>
<td>r-0</td>
</tr>
<tr>
<td>9</td>
<td>r-0</td>
</tr>
<tr>
<td>8</td>
<td>r-0</td>
</tr>
</tbody>
</table>

Table 6-12. DMAIV Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>DMAIVx</td>
<td>R</td>
<td>0h</td>
<td>DMA interrupt vector value. See Table 6-13.</td>
</tr>
</tbody>
</table>

Table 6-13. DMA Interrupt Vector Values

<table>
<thead>
<tr>
<th>DMAIV Contents</th>
<th>Interrupt Source</th>
<th>Interrupt Flag</th>
<th>Interrupt Priority</th>
</tr>
</thead>
<tbody>
<tr>
<td>00h</td>
<td>No interrupt pending</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>02h</td>
<td>DMA channel 0</td>
<td>DMA0IFG</td>
<td>Highest</td>
</tr>
<tr>
<td>04h</td>
<td>DMA channel 1</td>
<td>DMA1IFG</td>
<td>–</td>
</tr>
<tr>
<td>06h</td>
<td>DMA channel 2</td>
<td>DMA2IFG</td>
<td>–</td>
</tr>
<tr>
<td>08h</td>
<td>Reserved</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>0Ah</td>
<td>Reserved</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>0Ch</td>
<td>Reserved</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>0Eh</td>
<td>Reserved</td>
<td>–</td>
<td>Lowest</td>
</tr>
</tbody>
</table>
Chapter 7

Flash Memory Controller

This chapter describes the operation of the MSP430x2xx flash memory controller.

<table>
<thead>
<tr>
<th>Section</th>
<th>Page</th>
</tr>
</thead>
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</tr>
<tr>
<td>7.2 Flash Memory Segmentation</td>
<td>318</td>
</tr>
<tr>
<td>7.3 Flash Memory Operation</td>
<td>320</td>
</tr>
<tr>
<td>7.4 Flash Registers</td>
<td>332</td>
</tr>
</tbody>
</table>
7.1 Flash Memory Introduction

The MSP430 flash memory is bit-, byte-, and word-addressable and programmable. The flash memory module has an integrated controller that controls programming and erase operations. The controller has four registers, a timing generator, and a voltage generator to supply program and erase voltages.

MSP430 flash memory features include:

- Internal programming voltage generation
- Bit, byte, or word programmable
- Ultra-low-power operation
- Segment erase and mass erase
- Marginal 0 and marginal 1 read mode (optional, see the device-specific data sheet)

Figure 7-1 shows the block diagram of the flash memory and controller.

---

**Note**

**Minimum V\text{CC} during flash write or erase**

The minimum $V_{\text{CC}}$ voltage during a flash write or erase operation is 2.2 V. If $V_{\text{CC}}$ falls below 2.2 V during write or erase, the result of the write or erase is unpredictable.

---

Figure 7-1. Flash Memory Module Block Diagram

7.2 Flash Memory Segmentation

MSP430 flash memory is partitioned into segments. Single bits, bytes, or words can be written to flash memory, but the segment is the smallest size of flash memory that can be erased.

The flash memory is partitioned into main and information memory sections. There is no difference in the operation of the main and information memory sections. Code or data can be located in either section. The differences between the two sections are the segment size and the physical addresses.
The information memory has four 64-byte segments. The main memory has one or more 512-byte segments. The main memory segments are further divided into blocks of 64 bytes. See the device-specific data sheet for the complete memory map of a device.

Figure 7-2 shows the flash segmentation using an example of 32KB of flash that has 64 main segments and four information segments.

7.2.1 Segment A

Segment A of the information memory is locked separately from all other segments with the LOCKA bit. When LOCKA = 1, Segment A cannot be written or erased and all information memory is protected from erasure during a mass erase or production programming. When LOCKA = 0, Segment A can be erased and written as any other flash memory segment, and all information memory is erased during a mass erase or production programming.

The state of the LOCKA bit is toggled when a 1 is written to it. Writing a 0 to LOCKA has no effect. This allows existing flash programming routines to be used unchanged.

```assembly
; Unlock SegmentA
BIT  #LOCKA,&FCTL3 ; Test LOCKA
JZ    SEGA_UNLOCKED ; Already unlocked?
MOV   #FWKEY+LOCKA,&FCTL3 ; No, unlock SegmentA
SEGA_UNLOCKED                   ; Yes, continue
; SegmentA is unlocked
; Lock SegmentA
BIT  #LOCKA,&FCTL3 ; Test LOCKA
JNZ   SEGA_LOCKED           ; Already locked?
MOV   #FWKEY+LOCKA,&FCTL3 ; No, lock SegmentA
SEGA_LOCKED                     ; Yes, continue
; SegmentA is locked
```
7.3 Flash Memory Operation

The default mode of the flash memory is read mode. In read mode, the flash memory is not being erased or written, the flash timing generator and voltage generator are off, and the memory operates identically to ROM.

MSP430 flash memory is in-system programmable (ISP) without the need for additional external voltage. The CPU can program its own flash memory. The flash memory write and erase modes are selected with the BLKWRT, WRT, MERAS, and ERASE bits and are:

- Byte or word write
- Block write
- Segment erase
- Mass erase (all main memory segments)
- All erase (all segments)

Reading from or writing to flash memory while it is being programmed or erased is prohibited. If CPU execution is required during the write or erase, the code to be executed must be in RAM. Any flash update can be initiated from within flash memory or RAM.

7.3.1 Flash Memory Timing Generator

Write and erase operations are controlled by the flash timing generator (see Figure 7-3). The flash timing generator operating frequency, \( f_{FTG} \), must be in the range from approximately 257 kHz to approximately 476 kHz (see device-specific data sheet).

![Flash Memory Timing Generator Block Diagram](Image)

**Figure 7-3. Flash Memory Timing Generator Block Diagram**

7.3.1.1 Flash Timing Generator Clock Selection

The flash timing generator can be sourced from ACLK, SMCLK, or MCLK. The selected clock source must be divided using the FNx bits to meet the frequency requirements for \( f_{FTG} \). If the \( f_{FTG} \) frequency deviates from the specification during the write or erase operation, the result of the write or erase may be unpredictable, or the flash memory may be stressed above the limits of reliable operation.

If a clock failure is detected during a write or erase operation, the operation is aborted, the FAIL flag is set, and the result of the operation is unpredictable.

While a write or erase operation is active, the selected clock source cannot be disabled by putting the MSP430 into a low-power mode. The selected clock source remains active until the operation is completed before being disabled.
7.3.2 Erasing Flash Memory

The erased level of a flash memory bit is 1. Each bit can be programmed from 1 to 0 individually but to reprogram from 0 to 1 requires an erase cycle. The smallest amount of flash that can be erased is a segment. There are three erase modes selected with the ERASE and MERAS bits (see Table 7-1).

<table>
<thead>
<tr>
<th>MERAS</th>
<th>ERASE</th>
<th>Erase Mode</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>1</td>
<td>Segment erase</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>Mass erase (all main memory segments)</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>LOCKA = 0: Erase main and information flash memory. LOCKA = 1: Erase only main flash memory.</td>
</tr>
</tbody>
</table>

Any erase is initiated by a dummy write into the address range to be erased. The dummy write starts the flash timing generator and the erase operation. Figure 7-4 shows the erase cycle timing. The BUSY bit is set immediately after the dummy write and remains set throughout the erase cycle. BUSY, MERAS, and ERASE are automatically cleared when the cycle completes. The erase cycle timing is not dependent on the amount of flash memory present on a device. Erase cycle times are equivalent for all MSP430F2xx and MSP430G2xx devices.

![Erase Cycle Timing](image)

**Figure 7-4. Erase Cycle Timing**

A dummy write to an address not in the range to be erased does not start the erase cycle, does not affect the flash memory, and is not flagged in any way. This errant dummy write is ignored.
7.3.2.1 Initiating an Erase From Within Flash Memory

Any erase cycle can be initiated from within flash memory or from RAM. When a flash segment erase operation is initiated from within flash memory, all timing is controlled by the flash controller, and the CPU is held while the erase cycle completes. After the erase cycle completes, the CPU resumes code execution with the instruction following the dummy write.

When initiating an erase cycle from within flash memory, it is possible to erase the code needed for execution after the erase. If this occurs, CPU execution is unpredictable after the erase cycle.

Figure 7-5 shows the flow to initiate an erase from flash.

![Flowchart for Erase Cycle From Within Flash Memory]

```assembly
; Segment Erase from flash. 514 kHz < SMCLK < 952 kHz
; Assumes ACCVIE = NMIIE = OFIE = 0.
MOV  #WDTPW+WDTHOLD, &WDTCTL ; Disable WDT
MOV  #FWKEY+FSSEL1+FN0, &FCTL2 ; SMCLK/2
MOV  #FWKEY, &FCTL3 ; Clear LOCK
MOV  #FWKEY+ERASE, &FCTL1 ; Enable segment erase
CLR  &0FC10h ; Dummy write, erase S1
MOV  #FWKEY+LOCK, &FCTL3 ; Done, set LOCK
... ; Re-enable WDT?
```

Figure 7-5. Erase Cycle From Within Flash Memory
### 7.3.2.2 Initiating an Erase From RAM

Any erase cycle may be initiated from RAM. In this case, the CPU is not held and can continue to execute code from RAM. The BUSY bit must be polled to determine the end of the erase cycle before the CPU can access any flash address again. If a flash access occurs while BUSY = 1, it is an access violation, ACCVIFG is set, and the erase results are unpredictable.

Figure 7-6 shows the flow to initiate an erase from RAM.

```
; Segment Erase from RAM, 514 kHz < SMCLK < 952 kHz
; Assumes ACCVIE = NMIIE = OFIE = 0.

MOV   #WDTPW+WDTHOLD,&WDTCTL     ; Disable WDT
L1   BIT   #BUSY, &FCTL3       ; Test BUSY
JNZ   L1                      ; Loop while busy
MOV   #FWKEY+FSSEL1+FN0, &FCTL2 ; SMCLK/2
MOV   #FWKEY&FCTL3            ; Clear LOCK
MOV   #FWKEY+ERASE, &FCTL1    ; Enable erase
CLR   &0FC10h                  ; Dummy write, erase S1
L2   BIT   #BUSY, &FCTL3       ; Test BUSY
JNZ   L2                      ; Loop while busy
MOV   #FWKEY+LOCK&FCTL3       ; Done, set LOCK
...                          ; Re-enable WDT?
```

Figure 7-6. Erase Cycle from Within RAM

![Diagram of erase cycle from RAM](image-url)
7.3.3 Writing Flash Memory

Table 7-2 lists the write modes that are selected by the WRT and BLKWRT bits.

<table>
<thead>
<tr>
<th>BLKWRT</th>
<th>WRT</th>
<th>Write Mode</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>1</td>
<td>Byte or word write</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>Block write</td>
</tr>
</tbody>
</table>

Both write modes use a sequence of individual write instructions, but using the block write mode is approximately twice as fast as byte or word mode, because the voltage generator remains on for the complete block write. Any instruction that modifies a destination can be used to modify a flash location in either byte or word write mode or block write mode. A flash word (low and high bytes) must not be written more than twice between erasures. Otherwise, damage can occur.

The BUSY bit is set while a write operation is active and cleared when the operation completes. If the write operation is initiated from RAM, the CPU must not access flash while BUSY = 1. Otherwise, an access violation occurs, ACCVIFG is set, and the flash write is unpredictable.

7.3.3.1 Byte or Word Write

A byte or word write operation can be initiated from within flash memory or from RAM. When initiating from within flash memory, all timing is controlled by the flash controller, and the CPU is held while the write completes. After the write completes, the CPU resumes code execution with the instruction following the write. Figure 7-7 shows the byte or word write timing.

![Figure 7-7. Byte or Word Write Timing](image)

When a byte or word write is executed from RAM, the CPU continues to execute code from RAM. The BUSY bit must be zero before the CPU accesses flash again, otherwise an access violation occurs, ACCVIFG is set, and the write result is unpredictable.

In byte or word mode, the internally-generated programming voltage is applied to the complete 64-byte block, each time a byte or word is written, for 27 of the 30 \( f_{FTG} \) cycles. With each byte or word write, the amount of time the block is subjected to the programming voltage accumulates. The cumulative programming time, \( t_{CPT} \), must not be exceeded for any block. If the cumulative programming time is met, the block must be erased before performing any further writes to any address within the block. See the device-specific data sheet for specifications.
7.3.3.2 Initiating a Byte or Word Write From Within Flash Memory

Figure 7-8 shows the flow to initiate a byte or word write from flash.

```
MOV   #WDTPW+WDTHOLD,&WDTCTL     ; Disable WDT
MOV   #FWKEY+FSSEL1+FN0,&FCTL2   ; SMCLK/2
MOV   #FWKEY,&FCTL3              ; Clear LOCK
MOV   #FWKEY+WRT,&FCTL1          ; Enable write
MOV   #0123h,&0FF1Eh             ; 0123h    -> 0FF1Eh
MOV   #FWKEY,&FCTL1              ; Done. Clear WRT
MOV   #FWKEY+LOCK,&FCTL3         ; Set LOCK
...                              ; Re-enable WDT?
```

Figure 7-8. Initiating a Byte or Word Write From Flash
7.3.3.3 Initiating a Byte or Word Write From RAM

Figure 7-9 shows the flow to initiate a byte or word write from RAM.

![Flowchart](Figure 7-9. Initiating a Byte or Word Write from RAM)

; Byte/word write from RAM. 514 kHz < SMCLK < 952 kHz
; Assumes 0FF1Eh is already erased
; Assumes ACCVIE = NMIIE = OFIE = 0.
MOV #WDTPW+WDTWOLD,.WDTCNT ; Disable WDT
L1 BIT #BUSY,.FCTL3              ; Test BUSY
JNZ L1                         ; Loop while busy
MOV #FWKEY+FSSEL1+FNO,.FCTL2   ; SMCLK/2
MOV #FWKEY,.FCTL3              ; Clear LOCK
MOV #FWKEY+WRT,.FCTL1          ; Enable write
MOV #0123h,.OFF1Eh             ; 0123h -> OFF1Eh
L2 BIT #BUSY,.FCTL3             ; Test BUSY
JNZ L2                         ; Loop while busy
MOV #FWKEY,.FCTL1              ; Clear WRT
MOV #FWKEY+LOCK,.FCTL3         ; Set LOCK
...                            ; Re-enable WDT?
### 7.3.3.4 Block Write

The block write can be used to accelerate the flash write process when many sequential bytes or words need to be programmed. The flash programming voltage remains on for the duration of writing the 64-byte block. The cumulative programming time $t_{\text{CPT}}$ must not be exceeded for any block during a block write.

A block write cannot be initiated from within flash memory. The block write must be initiated from RAM only. The BUSY bit remains set throughout the duration of the block write. The WAIT bit must be checked between writing each byte or word in the block. When WAIT is set the next byte or word of the block can be written. When writing successive blocks, the BLKWRT bit must be cleared after the current block is complete. BLKWRT can be set initiating the next block write after the required flash recovery time given by $t_{\text{end}}$. BUSY is cleared following each block write completion indicating the next block can be written. **Figure 7-10** shows the block write timing.

![Figure 7-10. Block-Write Cycle Timing](image)

**Figure 7-10. Block-Write Cycle Timing**
7.3.3.5 Block Write Flow and Example

Figure 7-11 and the following example show the flow for a block write.

- Disable watchdog
- BUSY = 1
- Setup flash controller
- Set BLKWRT=WRT=1
- Write byte or word
- WAIT=0?
  - yes
    - BUSY = 1
    - Set BLKWRT=0
    - yes
      - BUSY = 1
      - Another Block?
      - yes
        - Set WRT=0, LOCK=1
        - re-enable WDT
  - no
    - Block Border?
    - no
      - yes
        - BUSY = 1
        - Disable watchdog
        - Setup flash controller
- no
  - yes
    - BUSY = 1
    - Disable watchdog
    - Setup flash controller

Figure 7-11. Block Write Flow
7.3.4 Flash Memory Access During Write or Erase

When any write or any erase operation is initiated from RAM and while BUSY = 1, the CPU may not read or write to or from any flash location. Otherwise, an access violation occurs, ACCVIFG is set, and the result is unpredictable. Also if a write to flash is attempted with WRT = 0, the ACCVIFG interrupt flag is set, and the flash memory is unaffected.

When a byte or word write or any erase operation is initiated from within flash memory, the flash controller returns op-code 03FFFh to the CPU at the next instruction fetch. Op-code 03FFFh is the JMP PC instruction. This causes the CPU to loop until the flash operation is finished. When the operation is finished and BUSY = 0, the flash controller allows the CPU to fetch the proper op-code and program execution resumes.

Table 7-3 lists the flash access conditions while BUSY = 1.

<table>
<thead>
<tr>
<th>Flash Operation</th>
<th>Flash Access</th>
<th>WAIT</th>
<th>Result</th>
</tr>
</thead>
<tbody>
<tr>
<td>Any erase, or byte or word write</td>
<td>Read</td>
<td>0</td>
<td>ACCVIFG = 0. 03FFFh is the value read.</td>
</tr>
<tr>
<td></td>
<td>Write</td>
<td>0</td>
<td>ACCVIFG = 1. Write is ignored.</td>
</tr>
<tr>
<td></td>
<td>Instruction fetch</td>
<td>0</td>
<td>ACCVIFG = 0. CPU fetches 03FFFh. This is the JMP PC instruction.</td>
</tr>
<tr>
<td>Block write</td>
<td>Read</td>
<td>1</td>
<td>ACCVIFG = 0. 03FFFh is the value read.</td>
</tr>
<tr>
<td></td>
<td>Write</td>
<td>1</td>
<td>ACCVIFG = 0. Write is written.</td>
</tr>
<tr>
<td></td>
<td>Instruction fetch</td>
<td>1</td>
<td>ACCVIFG = 1, LOCK = 1</td>
</tr>
</tbody>
</table>

Interrupts are automatically disabled during any flash operation when EEI = 0 and EEIEX = 0 and on MSP430x20xx and MSP430G2xx devices where EEI and EEIEX are not present. After the flash operation has completed, interrupts are automatically re-enabled. Any interrupt that occurred during the operation has its associated flag set and generates an interrupt request when re-enabled.

When EEIEX = 1 and GIE = 1, an interrupt immediately aborts any flash operation and the FAIL flag is set. When EEI = 1, GIE = 1, and EEIEX = 0, a segment erase is interrupted by a pending interrupt every 32 fFTG cycles. After servicing the interrupt, the segment erase is continued for at least 32 fFTG cycles or until it is complete. During the servicing of the interrupt, the BUSY bit remains set but the flash memory can be accessed by the CPU without causing an access violation occurs. Nested interrupts and using the RETI instruction inside interrupt service routines are not supported.

The watchdog timer (in watchdog mode) must be disabled before a flash erase cycle. A reset aborts the erase, and the results are unpredictable. After the erase cycle has completed, the watchdog may be enabled again.
7.3.5 Stopping a Write or Erase Cycle

Any write or erase operation can be stopped before its normal completion by setting the emergency exit bit EMEX. Setting the EMEX bit stops the active operation immediately and stops the flash controller. All flash operations cease, the flash returns to read mode, and all bits in the FCTL1 register are reset. The result of the intended operation is unpredictable.

7.3.6 Marginal Read Mode

The marginal read mode can be used to verify the integrity of the flash memory contents. This feature is implemented in selected 2xx devices (see the device-specific data sheet for availability). During marginal read mode, marginally programmed flash memory bit locations can be detected. Events that can produce this situation include improper fTGE settings or violation of minimum VCC during erase or program operations. One method for identifying such memory locations would be to periodically perform a checksum calculation over a section of flash memory (for example, a flash segment) and to repeat this procedure with the marginal read mode enabled. If they do not match, it could indicate an insufficiently programmed flash memory location. It is possible to refresh the affected flash memory segment by disabling marginal read mode, copying to RAM, erasing the flash segment, and writing back to it from RAM.

The program that checks the flash memory contents must be executed from RAM. Executing code from flash automatically disables the marginal read mode. The marginal read modes are controlled by the MRG0 and MRG1 register bits. Setting MRG1 is used to detect insufficiently programmed flash cells containing a 1 (erased bits). Setting MRG0 is used to detect insufficiently programmed flash cells containing a 0 (programmed bits). Only one of these bits may be set at a time. Therefore, a full marginal read check requires two passes of checking the flash memory content's integrity. During marginal read mode, the flash access speed (MCLK) must be limited to 1 MHz (see the device-specific data sheet).

7.3.7 Configuring and Accessing the Flash Memory Controller

The FCTLx registers are 16-bit password-protected read/write registers. Any read or write access must use word instructions and write accesses must include the write password 0A5h in the upper byte. Any write to any FCTLx register with any value other than 0A5h in the upper byte is a security key violation, which sets the KEYV flag and triggers a PUC system reset. Any read of any FCTLx registers reads 096h in the upper byte.

Any write to FCTL1 during an erase or byte or word write operation is an access violation and sets ACCVIFG. Writing to FCTL1 is allowed in block write mode when WAIT = 1, but writing to FCTL1 in block write mode when WAIT = 0 is an access violation and sets ACCVIFG.

Any write to FCTL2 when the BUSY = 1 is an access violation.

Any FCTLx register may be read when BUSY = 1. A read does not cause an access violation.

7.3.8 Flash Memory Controller Interrupts

The flash controller has two interrupt sources, KEYV and ACCVIFG. ACCVIFG is set when an access violation occurs. When the ACCVIE bit is re-enabled after a flash write or erase, a set ACCVIFG flag generates an interrupt request. ACCVIFG sources the NMI interrupt vector, so it is not necessary for GIE to be set for ACCVIFG to request an interrupt. ACCVIFG may also be checked by software to determine if an access violation occurred. ACCVIFG must be reset by software.

The key violation flag KEYV is set when any of the flash control registers are written with an incorrect password. When this occurs, a PUC is generated to immediately reset the device.
7.3.9 Programming Flash Memory Devices

There are three options for programming an MSP430 flash device. All options support in-system programming:

- Program with JTAG
- Program with the bootloader
- Program with a custom solution

7.3.9.1 Programming Flash Memory With JTAG

MSP430 devices can be programmed through the JTAG port. The JTAG interface requires four signals (five signals on 20- and 28-pin devices), ground and, optionally, \( V_{CC} \) and \( \text{RST/NMI} \).

The JTAG port is protected with a fuse. Blowing the fuse completely disables the JTAG port and is not reversible. Further access to the device through JTAG is not possible. For details, see *MSP430 Programming Via the JTAG Interface* (SLAU320).

7.3.9.2 Programming Flash Memory With the Bootloader (BSL)

Most MSP430 flash devices contain a bootloader. See the device-specific data sheet for implementation details. The BSL enables users to read or program the flash memory or RAM using a UART serial interface. Access to the MSP430 flash memory through the BSL is protected by a 256-bit user-defined password. For more details see *MSP430 Programming With the Bootloader (BSL)* (SLAU319).

7.3.9.3 Programming Flash Memory With a Custom Solution

The ability of the MSP430 CPU to write to its own flash memory allows for in-system and external custom programming solutions (see Figure 7-12). The user can choose to provide data to the MSP430 through any means available (for example, UART or SPI). User-developed software can receive the data and program the flash memory. Because this type of solution is developed by the user, it can be completely customized to fit the application needs for programming, erasing, or updating the flash memory.

![Figure 7-12. User-Developed Programming Solution](image-url)
7.4 Flash Registers

Table 7-4 lists the memory-mapped registers for the flash memory controller.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>128h</td>
<td>FCTL1</td>
<td>Flash memory control 1</td>
<td>Read/write</td>
<td>9600h with PUC</td>
<td>Section 7.4.1</td>
</tr>
<tr>
<td>12Ah</td>
<td>FCTL2</td>
<td>Flash memory control 2</td>
<td>Read/write</td>
<td>9642h with PUC</td>
<td>Section 7.4.2</td>
</tr>
<tr>
<td>12Ch</td>
<td>FCTL3</td>
<td>Flash memory control 3</td>
<td>Read/write</td>
<td>9658h with PUC(1)</td>
<td>Section 7.4.3</td>
</tr>
<tr>
<td>1BEh</td>
<td>FCTL4(2)</td>
<td>Flash memory control 4</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 7.4.4</td>
</tr>
<tr>
<td>0h</td>
<td>IE1</td>
<td>Interrupt enable 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 7.4.5</td>
</tr>
</tbody>
</table>

(1) KEYV is initialized with POR. All other bits are initialized with PUC.
(2) Not present in all devices. See device-specific data sheet.
7.4.1 FCTL1 Register

Flash Memory Control 1 Register

FCTL1 is shown in Figure 7-13 and described in Table 7-5.

Return to Table 7-4.

Figure 7-13. FCTL1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-8</td>
<td>FRKEY/FWKEY</td>
<td>R/W</td>
<td>96h</td>
<td>FCTLx password. Always reads as 096h. Must be written as 0A5h. Writing any other value generates a PUC.</td>
</tr>
<tr>
<td>7</td>
<td>BLKWRT</td>
<td>R/W</td>
<td>0h</td>
<td>Block write mode. WRT must also be set for block write mode. BLKWRT is automatically reset when EMEX is set.</td>
</tr>
<tr>
<td></td>
<td>WRT</td>
<td>R/W</td>
<td>0h</td>
<td>Write. This bit is used to select any write mode. WRT is automatically reset when EMEX is set. 0b = Write mode is off 1b = Write mode is on</td>
</tr>
<tr>
<td>5</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td>Enable emergency interrupt exit. Setting this bit enables an interrupt to cause an emergency exit from a flash operation when GIE = 1. EEIEX is automatically reset when EMEX is set. Not present on MSP430x20xx and MSP430G2xx devices. 0b = Exit interrupt disabled. 1b = Exit on interrupt enabled.</td>
</tr>
<tr>
<td>4</td>
<td>EEIEX</td>
<td>R/W</td>
<td>0h</td>
<td>Enable erase interrupts. Setting this bit allows a segment erase to be interrupted by an interrupt request. After the interrupt is serviced the erase cycle is resumed. Not present on MSP430x20xx and MSP430G2xx devices. 0b = Interrupts during segment erase disabled. 1b = Interrupts during segment erase enabled.</td>
</tr>
<tr>
<td>3</td>
<td>EEI</td>
<td>R/W</td>
<td>0h</td>
<td>Enable mass erase and erase. These bits are used together to select the erase mode. MERAS and ERASE are automatically reset when EMEX is set. See Table 7-6.</td>
</tr>
<tr>
<td>2</td>
<td>MERAS</td>
<td>R/W</td>
<td>0h</td>
<td>Mass erase and erase. These bits are used together to select the erase mode. MERAS and ERASE are automatically reset when EMEX is set. See Table 7-6.</td>
</tr>
<tr>
<td>1</td>
<td>ERASE</td>
<td>R/W</td>
<td>0h</td>
<td>Mass erase and erase. These bits are used together to select the erase mode. MERAS and ERASE are automatically reset when EMEX is set. See Table 7-6.</td>
</tr>
</tbody>
</table>

(1)  Not present on MSP430x20xx and MSP430G2xx devices.

Table 7-6. Erase Cycles

<table>
<thead>
<tr>
<th>MERAS</th>
<th>ERASE</th>
<th>Erase Cycle</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>No erase</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>Erase individual segment only</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>Erase all main memory segments</td>
</tr>
</tbody>
</table>
Table 7-6. Erase Cycles (continued)

<table>
<thead>
<tr>
<th>MERAS</th>
<th>ERASE</th>
<th>Erase Cycle</th>
</tr>
</thead>
</table>
| 1     | 1     | LOCKA = 0: Erase main and information flash memory.  
 |       |       | LOCKA = 1: Erase only main flash memory.         |

7.4.2 FCTL2 Register

Flash Memory Control 2 Register

FCTL2 is shown in Figure 7-14 and described in Table 7-7.

Return to Table 7-4.

Figure 7-14. FCTL2 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-8</td>
<td>FWKEYx</td>
<td>R/W</td>
<td>96h</td>
<td>FCTLx password. Always reads as 096h. Must be written as 0A5h. Writing any other value generates a PUC.</td>
</tr>
</tbody>
</table>
| 7-6  | FSSELx   | R/W  | 1h    | Flash controller clock source select  
 |      |          |      | 00b = ACLK  
 |      |          |      | 01b = MCLK  
 |      |          |      | 10b = SMCLK  
 |      |          |      | 11b = SMCLK  |
| 5-0  | FNx      | R/W  | 2h    | Flash controller clock divider. These six bits select the divider for the flash controller clock. The divisor value is FNx + 1. For example, when FNx = 00h, the divisor is 1. When FNx = 03Fh, the divisor is 64. |

Table 7-7. FCTL2 Register Field Descriptions
7.4.3 FCTL3 Register

Flash Memory Control 3 Register

FCTL3 is shown in Figure 7-15 and described in Table 7-8.

Return to Table 7-4.

Figure 7-15. FCTL3 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-8</td>
<td>FWKEYx</td>
<td>R/W</td>
<td>96h</td>
<td>FCTLx password. Always reads as 096h. Must be written as 0A5h. Writing any other value generates a PUC.</td>
</tr>
<tr>
<td>7</td>
<td>FAIL</td>
<td>R/W</td>
<td>0h</td>
<td>Operation failure. This bit is set if the f_TG clock source fails, or a flash operation is aborted from an interrupt when EEIEX = 1. FAIL must be reset with software. 0b = No failure 1b = Failure</td>
</tr>
<tr>
<td>6</td>
<td>LOCKA</td>
<td>R/W</td>
<td>1h</td>
<td>Segment A and Info lock. Write a 1 to this bit to change its state. Writing 0 has no effect. 0b = Segment A unlocked and all information memory is erased during a mass erase. 1b = Segment A locked and all information memory is protected from erasure during a mass erase.</td>
</tr>
<tr>
<td>5</td>
<td>EMEX</td>
<td>R/W</td>
<td>0h</td>
<td>Emergency exit 0b = No emergency exit 1b = Emergency exit</td>
</tr>
<tr>
<td>4</td>
<td>LOCK</td>
<td>R/W</td>
<td>1h</td>
<td>Lock. This bit unlocks the flash memory for writing or erasing. The LOCK bit can be set any time during a byte or word write or erase operation, and the operation completes normally. In the block write mode if the LOCK bit is set while BLKWRT = WAIT = 1, then BLKWRT and WAIT are reset and the mode ends normally. 0b = Unlocked 1b = Locked</td>
</tr>
<tr>
<td>3</td>
<td>WAIT</td>
<td>R</td>
<td>1h</td>
<td>Wait. Indicates the flash memory is being written to. 0b = The flash memory is not ready for the next byte or word write 1b = The flash memory is ready for the next byte or word write</td>
</tr>
<tr>
<td>2</td>
<td>ACCVIFG</td>
<td>R/W</td>
<td>0h</td>
<td>Access violation interrupt flag 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
<tr>
<td>1</td>
<td>KEYV</td>
<td>R/W</td>
<td>0h</td>
<td>Flash security key violation. This bit indicates an incorrect FCTLx password was written to any flash control register and generates a PUC when set. KEYV must be reset with software. KEYV is reset with POR. 0b = FCTLx password was written correctly 1b = FCTLx password was written incorrectly</td>
</tr>
</tbody>
</table>

Table 7-8. FCTL3 Register Field Descriptions
### Table 7-8. FCTL3 Register Field Descriptions (continued)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
</table>
| 0   | BUSY  | R    | 0h    | Busy. This bit indicates the status of the flash timing generator.  
                      |       |      |       | 0b = Not busy  
                      |       |      |       | 1b = Busy       |
7.4.4 FCTL4 Register

Flash Memory Control 4 Register. This register is not available in all devices. See the device-specific data sheet for details.

FCTL4 is shown in Figure 7-16 and described in Table 7-9.

Return to Table 7-4.

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-8</td>
<td>FWKEYx</td>
<td>R/W</td>
<td>96h</td>
<td>FCTLx password. Always reads as 096h. Must be written as 0A5h. Writing any other value generates a PUC.</td>
</tr>
<tr>
<td>7-6</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td>Reserved. Always read as 0.</td>
</tr>
<tr>
<td>5</td>
<td>MRG1</td>
<td>R/W</td>
<td>0h</td>
<td>Marginal read 1 mode. This bit enables the marginal 1 read mode. The marginal read 1 bit is cleared if the CPU starts execution from the flash memory. If both MRG1 and MRG0 are set, MRG1 is active and MRG0 is ignored. 0b = Marginal 1 read mode is disabled. 1b = Marginal 1 read mode is enabled.</td>
</tr>
<tr>
<td>4</td>
<td>MRG0</td>
<td>R/W</td>
<td>0h</td>
<td>Marginal read 0 mode. This bit enables the marginal 0 read mode. The marginal mode 0 is cleared if the CPU starts execution from the flash memory. If both MRG1 and MRG0 are set, MRG1 is active and MRG0 is ignored. 0b = Marginal 0 read mode is disabled. 1b = Marginal 0 read mode is enabled.</td>
</tr>
<tr>
<td>3-0</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td>Reserved. Always read as 0.</td>
</tr>
</tbody>
</table>
7.4.5 IE1 Register

Interrupt Enable 1 Register

IE1 is shown in Figure 7-17 and described in Table 7-10.

Return to Table 7-4.

Reset with PUC.

Table 7-10. IE1 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See the device-specific data sheet.</td>
</tr>
<tr>
<td>5</td>
<td>ACCVIE</td>
<td>R/W</td>
<td>0h</td>
<td>Flash memory access violation interrupt enable. This bit enables the ACCVIFG interrupt. Because other bits in IE1 may be used for other modules, TI recommends setting or clearing this bit using BIS.B or BIC.B instructions, respectively, rather than MOV.B or CLR.B instructions. 0b = Interrupt not enabled 1b = Interrupt enabled</td>
</tr>
<tr>
<td>4-0</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See the device-specific data sheet.</td>
</tr>
</tbody>
</table>
Chapter 8
Digital I/O

This chapter describes the operation of the digital I/O ports.

8.1 Digital I/O Introduction ................................................................. 340
8.2 Digital I/O Operation ................................................................. 340
8.3 Digital I/O Registers ................................................................. 345
8.1 Digital I/O Introduction

MSP430 devices have up to eight digital I/O ports implemented, P1 to P8. Each port has up to eight I/O pins. Every I/O pin is individually configurable for input or output direction, and every I/O pin can be individually read or written.

Ports P1 and P2 have interrupt capability. Each interrupt for the P1 and P2 I/O lines can be individually enabled and configured to provide an interrupt on a rising edge or falling edge of an input signal. All P1 I/O lines source a single interrupt vector, and all P2 I/O lines source a different single interrupt vector.

The digital I/O features include:

- Independently programmable individual I/Os
- Any combination of input or output
- Individually configurable P1 and P2 interrupts
- Independent input and output data registers
- Individually configurable pullup or pulldown resistors
- Individually configurable pin-oscillator function (some MSP430 devices)

Note

MSP430G22x0: These devices feature digital I/O pins P1.2, P1.5, P1.6, and P1.7. The GPIOs P1.0, P1.1, P1.3, P1.4, P2.6, and P2.7 are implemented on this device but not available on the device pinout. To avoid floating inputs on these GPIOs, these digital I/Os should be properly initialized by running a start-up code similar to the following sample:

```
mov.b #0x1B, P1REN; ; Terminate unavailable Port1 pins properly
xor.b #0x20, BCSCTL3; ; Select VLO as low freq clock
```

The initialization code configures GPIOs P1.0, P1.1, P1.3, and P1.4 as inputs with pulldown resistor enabled (that is, P1REN.x = 1) and GPIOs P2.6 and P2.7 are terminated by selecting VLOCLK as ACLK – see the Basic Clock System chapter for details. The register bits of P1.0, P1.1, P1.3, and P1.4 in registers P1OUT, P1DIR, P1IFG, P1IE, P1IES, P1SEL, and P1REN should not be altered after the initialization code is executed. Also, all Port 2 registers are should not be altered.

8.2 Digital I/O Operation

The digital I/O is configured with user software. The setup and operation of the digital I/O is described in the following sections.

8.2.1 Input Register PxIN

Each bit in each PxIN register reflects the value of the input signal at the corresponding I/O pin when the pin is configured as I/O function.

Bit = 0: The input is low

Bit = 1: The input is high

Note

Writing to Read-Only Registers PxIN

Writing to these read-only registers results in increased current consumption while the write attempt is active.

8.2.2 Output Registers PxOUT

Each bit in each PxOUT register is the value to be output on the corresponding I/O pin when the pin is configured as I/O function, output direction, and the pullup/down resistor is disabled.
Bit = 0: The output is low
Bit = 1: The output is high

If the pin's pullup or pulldown resistor is enabled, the corresponding bit in the PxOUT register selects pullup or pulldown (see Section 8.2.4).

Bit = 0: The pin is pulled down
Bit = 1: The pin is pulled up

8.2.3 Direction Registers PxDIR
Each bit in each PxDIR register selects the direction of the corresponding I/O pin, regardless of the selected function for the pin. PxDIR bits for I/O pins that are selected for other functions must be set as required by the other function.

Bit = 0: The port pin is switched to input direction
Bit = 1: The port pin is switched to output direction

8.2.4 Pullup or Pulldown Resistor Enable Registers PxREN
Each bit in each PxREN register enables or disables the pullup or pulldown resistor of the corresponding I/O pin. The corresponding bit in the PxOUT register selects if the pin is pulled up or pulled down (see Section 8.2.2).

Bit = 0: Pullup or pulldown resistor disabled
Bit = 1: Pullup or pulldown resistor enabled

8.2.5 Function Select Registers PxSEL and PxSEL2
Port pins are often multiplexed with other peripheral module functions. See the device-specific data sheet to determine pin functions. Each PxSEL and PxSEL2 bit is used to select the pin function - I/O port or peripheral module function.

<table>
<thead>
<tr>
<th>PxSEL2</th>
<th>PxSEL</th>
<th>Pin Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>I/O function is selected.</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>Primary peripheral module function is selected.</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>Reserved. See device-specific data sheet.</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>Secondary peripheral module function is selected.</td>
</tr>
</tbody>
</table>

Setting PxSELx = 1 does not automatically set the pin direction. Other peripheral module functions may require the PxDIR bits to be configured according to the direction needed for the module function. See the pin schematics in the device-specific data sheet.

Note

Setting PxREN = 1 When PxSEL = 1
On some I/O ports on the MSP430F261x and MSP430F2416/7/8/9, enabling the pullup/pulldown resistor (PxREN = 1) while the module function is selected (PxSEL = 1) does not disable the logic output driver. This combination is not recommended and may result in unwanted current flow through the internal resistor. See the device-specific data sheet pin schematics for more information.

;Output ACLK on P2.0 on MSP430F21x1
BIS.B #01h, &P2SEL ; Select ACLK function for pin
BIS.B #01h, &P2DIR ; Set direction to output *Required*
Note

P1 and P2 Interrupts Are Disabled When PxSEL = 1

When any P1SELx or P2SELx bit is set, the corresponding pin's interrupt function is disabled. Therefore, signals on these pins will not generate P1 or P2 interrupts, regardless of the state of the corresponding P1IE or P2IE bit.

When a port pin is selected as an input to a peripheral, the input signal to the peripheral is a latched representation of the signal at the device pin. While PxSELx = 1, the internal input signal follows the signal at the pin. However, if the PxSELx = 0, the input to the peripheral maintains the value of the input signal at the device pin before the PxSELx bit was reset.

8.2.6 Pin Oscillator

Some MSP430 devices have a pin oscillator function built-in to some pins. The pin oscillator function may be used in capacitive touch sensing applications to eliminate external passive components. Additionally, the pin oscillator may be used in sensor applications.

No external components to create the oscillation

Capacitive sensors can be connected directly to MSP430 pin

Robust, typical built-in hysteresis of ~0.7 V

When the pin oscillator function is enabled, other pin configurations are overwritten. The output driver is turned off while the weak pullup/pulldown is enabled and controlled by the voltage level on the pin itself. The voltage on the I/O is fed into the Schmitt trigger of the pin and then routed to a timer. The connection to the timer is device specific and, thus, defined in the device-specific data sheet. The Schmitt-trigger output is inverted and then decides if the pullup or the pulldown is enabled. Due to the inversion, the pin starts to oscillate as soon as the pin oscillator pin configuration is selected. Some of the pin-oscillator outputs are combined by a logical OR before routing to a timer clock input or timer capture channel. Therefore, only one pin oscillator should be enabled at a time. The oscillation frequency of each pin is defined by the load on the pin and by the I/O type. I/Os with analog functions typically show a lower oscillation frequency than pure digital I/Os. See the device-specific data sheet for details. Pins without external load show typical oscillation frequencies of 1 MHz to 3 MHz.

Pin Oscillator in a Capacitive-Touch Application

A typical touch pad application using the pin oscillator is shown in Figure 8-1.
A change of the capacitance of the touch pad (external capacitive load) has an effect on the pin oscillator frequency. An approaching finger tip increases the capacitance of the touch pad thus leads to a lower self-oscillation frequency due to the longer charging time. The oscillation frequency can directly be captured in a built-in Timer channel. The typical sensitivity of a pin is shown in Figure 8-2.

![Figure 8-2. Typical Pin-Oscillation Frequency](image)

8.2.7 P1 and P2 Interrupts

Each pin in ports P1 and P2 have interrupt capability, configured with the PxIFG, PxIE, and PxIES registers. All P1 pins source a single interrupt vector, and all P2 pins source a different single interrupt vector. The PxIFG register can be tested to determine the source of a P1 or P2 interrupt.
8.2.7.1 Interrupt Flag Registers \( P1IFG, P2IFG \)

Each \( PxFGx \) bit is the interrupt flag for its corresponding I/O pin and is set when the selected input signal edge occurs at the pin. All \( PxFGx \) interrupt flags request an interrupt when their corresponding \( PxE \) bit and the GIE bit are set. Each \( PxFG \) flag must be reset with software. Software can also set each \( PxFG \) flag, providing a way to generate a software initiated interrupt.

Bit = 0: No interrupt is pending
Bit = 1: An interrupt is pending

Only transitions, not static levels, cause interrupts. If any \( PxFGx \) flag becomes set during a \( Px \) interrupt service routine, or is set after the RETI instruction of a \( Px \) interrupt service routine is executed, the set \( PxFGx \) flag generates another interrupt. This ensures that each transition is acknowledged.

---

**Note**

\( PxFG \) Flags When Changing \( PxOUT \) or \( PxDIR \)

Writing to \( P1OUT, P1DIR, P2OUT, \) or \( P2DIR \) can result in setting the corresponding \( P1FG \) or \( P2FG \) flags.

---

8.2.7.2 Interrupt Edge Select Registers \( P1IES, P2IES \)

Each \( PxIES \) bit selects the interrupt edge for the corresponding I/O pin.

Bit = 0: The \( PxFGx \) flag is set with a low-to-high transition
Bit = 1: The \( PxFGx \) flag is set with a high-to-low transition

---

**Note**

Writing to \( PxIESx \)

Writing to \( P1IES \), or \( P2IES \) can result in setting the corresponding interrupt flags.

<table>
<thead>
<tr>
<th>( PxIESx )</th>
<th>( PxINx )</th>
<th>( PxFGx )</th>
</tr>
</thead>
<tbody>
<tr>
<td>0 → 1</td>
<td>0</td>
<td>May be set</td>
</tr>
<tr>
<td>0 → 1</td>
<td>1</td>
<td>Unchanged</td>
</tr>
<tr>
<td>1 → 0</td>
<td>0</td>
<td>Unchanged</td>
</tr>
<tr>
<td>1 → 0</td>
<td>1</td>
<td>May be set</td>
</tr>
</tbody>
</table>

---

8.2.7.3 Interrupt Enable \( P1IE, P2IE \)

Each \( PxE \) bit enables the associated \( PxFG \) interrupt flag.

Bit = 0: The interrupt is disabled.
Bit = 1: The interrupt is enabled.

---

8.2.8 Configuring Unused Port Pins

Unused I/O pins should be configured as I/O function, output direction, and left unconnected on the PC board, to prevent a floating input and reduce power consumption. The value of the \( PxOUT \) bit is irrelevant, since the pin is unconnected. Alternatively, the integrated pullup/pulldown resistor can be enabled by setting the \( PxREN \) bit of the unused pin to prevent the floating input. See the System Resets, Interrupts, and Operating Modes chapter for termination of unused pins.
8.3 Digital I/O Registers

The digital I/O registers are listed in Table 8-2.

<table>
<thead>
<tr>
<th>Port</th>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>P1</td>
<td>020h</td>
<td>P1IN</td>
<td>Input</td>
<td>Read only</td>
<td>Unchanged</td>
<td>Section 8.3.1</td>
</tr>
<tr>
<td></td>
<td>021h</td>
<td>P1OUT</td>
<td>Output</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 8.3.2</td>
</tr>
<tr>
<td></td>
<td>022h</td>
<td>P1DIR</td>
<td>Direction</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.3</td>
</tr>
<tr>
<td></td>
<td>023h</td>
<td>P1IFG</td>
<td>Interrupt Flag</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.4</td>
</tr>
<tr>
<td></td>
<td>024h</td>
<td>P1IES</td>
<td>Interrupt Edge Select</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 8.3.5</td>
</tr>
<tr>
<td></td>
<td>025h</td>
<td>P1IE</td>
<td>Interrupt Enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.6</td>
</tr>
<tr>
<td></td>
<td>026h</td>
<td>P1SEL</td>
<td>Port Select</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.7</td>
</tr>
<tr>
<td></td>
<td>027h</td>
<td>P1SEL2</td>
<td>Port Select 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.8</td>
</tr>
<tr>
<td></td>
<td>028h</td>
<td>P1REN</td>
<td>Resistor Enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.9</td>
</tr>
<tr>
<td></td>
<td>029h</td>
<td>P2OUT</td>
<td>Output</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 8.3.2</td>
</tr>
<tr>
<td></td>
<td>02Ah</td>
<td>P2DIR</td>
<td>Direction</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.3</td>
</tr>
<tr>
<td></td>
<td>02Bh</td>
<td>P2IFG</td>
<td>Interrupt Flag</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.4</td>
</tr>
<tr>
<td></td>
<td>02Ch</td>
<td>P2IES</td>
<td>Interrupt Edge Select</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 8.3.5</td>
</tr>
<tr>
<td></td>
<td>02Dh</td>
<td>P2IE</td>
<td>Interrupt Enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.6</td>
</tr>
<tr>
<td></td>
<td>02Eh</td>
<td>P2SEL</td>
<td>Port Select</td>
<td>Read/write</td>
<td>C0h with PUC(1)</td>
<td>Section 8.3.7</td>
</tr>
<tr>
<td></td>
<td>042h</td>
<td>P2SEL2</td>
<td>Port Select 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.8</td>
</tr>
<tr>
<td></td>
<td>02Fh</td>
<td>P2REN</td>
<td>Resistor Enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.9</td>
</tr>
<tr>
<td>P2</td>
<td>018h</td>
<td>P3IN</td>
<td>Input</td>
<td>Read only</td>
<td>Unchanged</td>
<td>Section 8.3.1</td>
</tr>
<tr>
<td></td>
<td>019h</td>
<td>P3OUT</td>
<td>Output</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 8.3.2</td>
</tr>
<tr>
<td></td>
<td>01Ah</td>
<td>P3DIR</td>
<td>Direction</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.3</td>
</tr>
<tr>
<td></td>
<td>01Bh</td>
<td>P3SEL</td>
<td>Port Select</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.7</td>
</tr>
<tr>
<td></td>
<td>043h</td>
<td>P3SEL2</td>
<td>Port Select 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.8</td>
</tr>
<tr>
<td></td>
<td>010h</td>
<td>P3REN</td>
<td>Resistor Enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.9</td>
</tr>
<tr>
<td>P3</td>
<td>01Ch</td>
<td>P4IN</td>
<td>Input</td>
<td>Read only</td>
<td>Unchanged</td>
<td>Section 8.3.1</td>
</tr>
<tr>
<td></td>
<td>01Dh</td>
<td>P4OUT</td>
<td>Output</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 8.3.2</td>
</tr>
<tr>
<td></td>
<td>01Eh</td>
<td>P4DIR</td>
<td>Direction</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.3</td>
</tr>
<tr>
<td></td>
<td>01Fh</td>
<td>P4SEL</td>
<td>Port Select</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.7</td>
</tr>
<tr>
<td></td>
<td>044h</td>
<td>P4SEL2</td>
<td>Port Select 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.8</td>
</tr>
<tr>
<td></td>
<td>011h</td>
<td>P4REN</td>
<td>Resistor Enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.9</td>
</tr>
<tr>
<td>P4</td>
<td>030h</td>
<td>P5IN</td>
<td>Input</td>
<td>Read only</td>
<td>Unchanged</td>
<td>Section 8.3.1</td>
</tr>
<tr>
<td></td>
<td>031h</td>
<td>P5OUT</td>
<td>Output</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 8.3.2</td>
</tr>
<tr>
<td></td>
<td>032h</td>
<td>P5DIR</td>
<td>Direction</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.3</td>
</tr>
<tr>
<td></td>
<td>033h</td>
<td>P5SEL</td>
<td>Port Select</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.7</td>
</tr>
<tr>
<td></td>
<td>045h</td>
<td>P5SEL2</td>
<td>Port Select 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.8</td>
</tr>
<tr>
<td></td>
<td>012h</td>
<td>P5REN</td>
<td>Resistor Enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.9</td>
</tr>
<tr>
<td>P5</td>
<td>034h</td>
<td>P6IN</td>
<td>Input</td>
<td>Read only</td>
<td>Unchanged</td>
<td>Section 8.3.1</td>
</tr>
<tr>
<td></td>
<td>035h</td>
<td>P6OUT</td>
<td>Output</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 8.3.2</td>
</tr>
<tr>
<td></td>
<td>036h</td>
<td>P6DIR</td>
<td>Direction</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.3</td>
</tr>
<tr>
<td></td>
<td>037h</td>
<td>P6SEL</td>
<td>Port Select</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.7</td>
</tr>
<tr>
<td></td>
<td>046h</td>
<td>P6SEL2</td>
<td>Port Select 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.8</td>
</tr>
<tr>
<td></td>
<td>013h</td>
<td>P6REN</td>
<td>Resistor Enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.9</td>
</tr>
</tbody>
</table>
Table 8-2. Digital I/O Registers (continued)

<table>
<thead>
<tr>
<th>Port</th>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>P7</td>
<td>038h</td>
<td>P7IN</td>
<td>Input</td>
<td>Read only</td>
<td>Unchanged</td>
<td>Section 8.3.1</td>
</tr>
<tr>
<td></td>
<td>03Ah</td>
<td>P7OUT</td>
<td>Output</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 8.3.2</td>
</tr>
<tr>
<td></td>
<td>03Ch</td>
<td>P7DIR</td>
<td>Direction</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.3</td>
</tr>
<tr>
<td></td>
<td>03Eh</td>
<td>P7SEL</td>
<td>Port Select</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.7</td>
</tr>
<tr>
<td></td>
<td>047h</td>
<td>P7SEL2</td>
<td>Port Select 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.8</td>
</tr>
<tr>
<td></td>
<td>014h</td>
<td>P7REN</td>
<td>Resistor Enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.9</td>
</tr>
<tr>
<td>P8</td>
<td>039h</td>
<td>P8IN</td>
<td>Input</td>
<td>Read only</td>
<td>Unchanged</td>
<td>Section 8.3.1</td>
</tr>
<tr>
<td></td>
<td>03Bh</td>
<td>P8OUT</td>
<td>Output</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 8.3.2</td>
</tr>
<tr>
<td></td>
<td>03Dh</td>
<td>P8DIR</td>
<td>Direction</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.3</td>
</tr>
<tr>
<td></td>
<td>03Fh</td>
<td>P8SEL</td>
<td>Port Select</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.7</td>
</tr>
<tr>
<td></td>
<td>048h</td>
<td>P8SEL2</td>
<td>Port Select 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.8</td>
</tr>
<tr>
<td></td>
<td>015h</td>
<td>P8REN</td>
<td>Resistor Enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 8.3.9</td>
</tr>
</tbody>
</table>

(1) If P2.6 and P2.7 are multiplexed with XIN and XOUT, respectively, the reset value of P2SEL is C0h. If XIN and XOUT have dedicated pins, the reset value is 00h.
8.3.1 PxIN Register
Port x Input Register

Table 8-3. PxIN Register Description

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>PxIN</td>
<td>R</td>
<td>Unchanged</td>
<td>Port x input. Each bit corresponds to one channel on Port x. The reset value is undefined. 0b = Input is low 1b = Input is high</td>
</tr>
</tbody>
</table>

8.3.2 PxOUT Register
Port x Output Register

Table 8-4. PxOUT Register Description

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>PxOUT</td>
<td>R/W</td>
<td>Unchanged</td>
<td>Port x output. Each bit corresponds to one channel on Port x. The reset value is undefined. When I/O configured to output mode: 0b = Output is low 1b = Output is high. When I/O configured to input mode and pullups/pulldowns enabled: 0b = Pulldown selected 1b = Pullup selected</td>
</tr>
</tbody>
</table>

8.3.3 PxDIR Register
Port x Direction Register

Table 8-5. PxDIR Register Description

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>PxDIR</td>
<td>R/W</td>
<td>0h</td>
<td>Port x direction. Each bit corresponds to one channel on Port x. 0b = Port configured as input 1b = Port configured as output</td>
</tr>
</tbody>
</table>
8.3.4 PxIFG Register
Port x Interrupt Flag Register (Port 1 and Port 2 only)

**Figure 8-6. PxIFG Register**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>PxIFG</td>
<td>R/W</td>
<td>0h</td>
<td>Port x interrupt flag. Each bit corresponds to one channel on Port x. 0b = No interrupt is pending. 1b = Interrupt is pending.</td>
</tr>
</tbody>
</table>

**Table 8-6. PxIFG Register Description**

8.3.5 PxIES Register
Port x Interrupt Edge Select Register (Port 1 and Port 2 only)

**Figure 8-7. PxIES Register**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>PxIES</td>
<td>R/W</td>
<td>Unchanged</td>
<td>Port x interrupt edge select. Each bit corresponds to one channel on Port x. The reset value is undefined. 0b = PxIFG flag is set with a low-to-high transition 1b = PxIFG flag is set with a high-to-low transition</td>
</tr>
</tbody>
</table>

**Table 8-7. PxIES Register Description**

8.3.6 PxIE Register
Port x Interrupt Enable Register (Port 1 and Port 2 only)

**Figure 8-8. PxIE Register**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>PxIE</td>
<td>R/W</td>
<td>0h</td>
<td>Port x interrupt enable. Each bit corresponds to one channel on Port x. 0b = Corresponding port interrupt disabled 1b = Corresponding port interrupt enabled</td>
</tr>
</tbody>
</table>

**Table 8-8. PxIE Register Description**
8.3.7 PxSEL Register
Port x Function Selection Register

Figure 8-9. PxSEL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>PxSEL</td>
<td>R/W</td>
<td>0h</td>
<td>Port function selection. Each bit corresponds to one channel on Port x. The values of each bit position in PxSEL2 and PxSEL are combined to specify the function. For example, if P1SEL2.5 = 1 and P1SEL.5 = 0, then the secondary module function is selected for P1.5. See PxSEL2 for the definition of each value.</td>
</tr>
</tbody>
</table>

Note
NOTE: If P2.6 and P2.7 are multiplexed with XIN and XOUT, respectively, the reset value of P2SEL is C0h. If XIN and XOUT have dedicated pins, the reset value of P2SEL is 00h.

8.3.8 PxSEL2 Register
Port x Function Selection Register 2

Figure 8-10. PxSEL2 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>PxSEL2</td>
<td>R/W</td>
<td>0h</td>
<td>Port function selection. Each bit corresponds to one channel on Port x. The values of each bit position in PxSEL2 and PxSEL are combined to specify the function. For example, if P1SEL2.5 = 1 and P1SEL.5 = 0, then the secondary module function is selected for P1.5. 00b = General-purpose I/O is selected 01b = Primary module function is selected 10b = Secondary module function is selected 11b = Tertiary module function is selected</td>
</tr>
</tbody>
</table>
8.3.9 PxREN Register
Port x Pullup or Pulldown Resistor Enable Register

**Figure 8-11. PxREN Register**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>PxREN</td>
<td>R/W</td>
<td>0h</td>
<td>Port x pullup or pulldown resistor enable. Each bit corresponds to one channel on Port x. When the port is configured as an input, setting this bit enables or disables the pullup or pulldown.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Pullup or pulldown disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Pullup or pulldown enabled</td>
</tr>
</tbody>
</table>
This chapter describes the operation of the SVS. The SVS is implemented in selected MSP430x2xx devices.

9.1 Supply Voltage Supervisor (SVS) Introduction ............................................................................................................ 352
9.2 SVS Operation .................................................................................................................................................................. 353
9.3 SVS Registers ................................................................................................................................................................. 355
9.1 Supply Voltage Supervisor (SVS) Introduction

The SVS is used to monitor the \( AV_{CC} \) supply voltage or an external voltage. The SVS can be configured to set a flag or generate a POR reset when the supply voltage or external voltage drops below a user-selected threshold.

The SVS features include:

- \( AV_{CC} \) monitoring
- Selectable generation of POR
- Output of SVS comparator accessible by software
- Low-voltage condition latched and accessible by software
- 14 selectable threshold levels
- External channel to monitor external voltage

The SVS block diagram is shown in Figure 9-1.

![Figure 9-1. SVS Block Diagram](image-url)
9.2 SVS Operation

The SVS detects if the AVCC voltage drops below a selectable level. It can be configured to provide a POR or set a flag, when a low-voltage condition occurs. The SVS is disabled after a brownout reset to conserve current consumption.

9.2.1 Configuring the SVS

The VLDx bits are used to enable/disable the SVS and select one of 14 threshold levels \(V_{(SVS\_IT-)}\) for comparison with AVCC. The SVS is off when VLDx = 0 and on when VLDx > 0. The SVSON bit does not turn on the SVS. Instead, it reflects the on/off state of the SVS and can be used to determine when the SVS is on.

When VLDx = 1111, the external SVSIN channel is selected. The voltage on SVSIN is compared to an internal level of approximately 1.25 V.

9.2.2 SVS Comparator Operation

A low-voltage condition exists when AVCC drops below the selected threshold or when the external voltage drops below its 1.25-V threshold. Any low-voltage condition sets the SVSFG bit.

The PORON bit enables or disables the device-reset function of the SVS. If PORON = 1, a POR is generated when SVSFG is set. If PORON = 0, a low-voltage condition sets SVSFG, but does not generate a POR.

The SVSFG bit is latched. This allows user software to determine if a low-voltage condition occurred previously. The SVSFG bit must be reset by user software. If the low-voltage condition is still present when SVSFG is reset, it will be immediately set again by the SVS.

9.2.3 Changing the VLDx Bits

When the VLDx bits are changed from zero to any non-zero value there is a automatic settling delay \(t_{d(SVS\_on)}\) implemented that allows the SVS circuitry to settle. The \(t_{d(SVS\_on)}\) delay is approximately 50 µs. During this delay, the SVS will not flag a low-voltage condition or reset the device, and the SVSON bit is cleared. Software can test the SVSON bit to determine when the delay has elapsed and the SVS is monitoring the voltage properly. Writing to SVSCTL while SVSON = 0 will abort the SVS automatic settling delay, \(t_{d(SVS\_on)}\), and switch the SVS to active mode immediately. In doing so, the SVS circuitry might not be settled, resulting in unpredictable behavior.

When the VLDx bits are changed from any non-zero value to any other non-zero value the circuitry requires the time \(t_{settle}\) to settle. The settling time \(t_{settle}\) is a maximum of \(~12 \mu s\). See the device-specific data sheet. There is no automatic delay implemented that prevents SVSFG to be set or to prevent a reset of the device. The recommended flow to switch between levels is shown in the following code.

```assembly
; Enable SVS for the first time:
MOV.B #080h, &SVSCTL ; Level 2.8V, do not cause POR
; ...
; Change SVS level
MOV.B #000h, &SVSCTL ; Temporarily disable SVS
MOV.B #018h, &SVSCTL ; Level 1.9V, cause POR
; ...
```
9.2.4 SVS Operating Range

Each SVS level has hysteresis to reduce sensitivity to small supply voltage changes when AV\textsubscript{CC} is close to the threshold. The SVS operation and SVS/Brownout interoperation are shown in Figure 9-2.

![Figure 9-2. Operating Levels for SVS and Brownout/Reset Circuit](image-url)
9.3 SVS Registers

Table 9-1 lists the memory-mapped registers for the SVS.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>55h</td>
<td>SVSCTL</td>
<td>SVS control</td>
<td>Read/write</td>
<td>00h with BOR</td>
<td>Section 9.3.1</td>
</tr>
</tbody>
</table>
9.3.1 SVSCTL Register

SVS Control Register

SVSCTL is shown in Figure 9-3 and described in Table 9-2.

Return to Table 9-1.

Figure 9-3. SVSCTL Register

<p>| | | | | | | | |</p>
<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>6</td>
<td>5</td>
<td>4</td>
<td>3</td>
<td>2</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>VLDx</td>
<td>PORON</td>
<td>SVSON</td>
<td>SVSOP</td>
<td>SVSFG</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>rw-0 (1)</td>
<td>rw-0 (1)</td>
<td>rw-0 (1)</td>
<td>rw-0 (1)</td>
<td>r-0 (1)</td>
<td>r-0 (1)</td>
<td>rw-0 (1)</td>
<td></td>
</tr>
</tbody>
</table>

Table 9-2. SVSCTL Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td>VLDx</td>
<td>R/W</td>
<td>0h(1)</td>
<td>Voltage level detect. These bits turn on the SVS and select the nominal SVS threshold voltage level. See the device-specific data sheet for parameters. 0000b = SVS is off 0001b = 1.9 V 0010b = 2.1 V 0011b = 2.2 V 0100b = 2.3 V 0101b = 2.4 V 0110b = 2.5 V 0111b = 2.65 V 1000b = 2.8 V 1001b = 2.9 V 1010b = 3.05 V 1011b = 3.2 V 1100b = 3.35 V 1101b = 3.5 V 1110b = 3.7 V 1111b = Compares external input voltage SVSIN to 1.25 V</td>
</tr>
<tr>
<td>3</td>
<td>PORON</td>
<td>R/W</td>
<td>0h(1)</td>
<td>POR on. This bit enables the SVSFG flag to cause a POR device reset. 0b = SVSFG does not cause a POR 1b = SVSFG causes a POR</td>
</tr>
<tr>
<td>2</td>
<td>SVSON</td>
<td>R</td>
<td>0h(1)</td>
<td>SVS on. This bit reports the status of SVS operation. This bit DOES NOT turn on the SVS. The SVS is turned on by setting VLDx &gt; 0. 0b = SVS is off 1b = SVS is on</td>
</tr>
<tr>
<td>1</td>
<td>SVSOP</td>
<td>R</td>
<td>0h(1)</td>
<td>SVS output. This bit reflects the output value of the SVS comparator. 0b = SVS comparator output is low 1b = SVS comparator output is high</td>
</tr>
<tr>
<td>0</td>
<td>SVSFG</td>
<td>R/W</td>
<td>0h(1)</td>
<td>SVS flag. This bit indicates a low-voltage condition. SVSFG remains set after a low-voltage condition until reset by software. 0b = No low-voltage condition occurred 1b = A low-voltage condition is present or has occurred</td>
</tr>
</tbody>
</table>

(1) Reset by a brownout reset only, not by a POR or PUC.
Chapter 10

Watchdog Timer+ (WDT+)

The watchdog timer+ (WDT+) is a 16-bit timer that can be used as a watchdog or as an interval timer. This chapter describes the WDT+. The WDT+ is implemented in all MSP430x2xx devices.

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<tr>
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</tr>
<tr>
<td>10.3 Watchdog Timer+ Registers</td>
<td>362</td>
</tr>
</tbody>
</table>
10.1 Watchdog Timer+ (WDT+) Introduction

The primary function of the WDT+ module is to perform a controlled system restart after a software problem occurs. If the selected time interval expires, a system reset is generated. If the watchdog function is not needed in an application, the module can be configured as an interval timer and can generate interrupts at selected time intervals.

Features of the watchdog timer+ module include:

- Four software-selectable time intervals
- Watchdog mode
- Interval mode
- Access to WDT+ control register is password protected
- Control of RST/NMI pin function
- Selectable clock source
- Can be stopped to conserve power
- Clock fail-safe feature

The WDT+ block diagram is shown in Figure 10-1.

---

**Note**

**Watchdog Timer+ Powers Up Active**

After a PUC, the WDT+ module is automatically configured in the watchdog mode with an initial 32768 clock cycle reset interval using the DCOCLK. The user must setup or halt the WDT+ prior to the expiration of the initial reset interval.
Figure 10-1. Watchdog Timer+ Block Diagram
10.2 Watchdog Timer+ Operation

The WDT+ module can be configured as either a watchdog or interval timer with the WDTCTL register. The WDTCTL register also contains control bits to configure the RST/NMI pin. WDTCTL is a 16-bit, password-protected, read/write register. Any read or write access must use word instructions and write accesses must include the write password 05Ah in the upper byte. Any write to WDTCTL with any value other than 05Ah in the upper byte is a security key violation and triggers a PUC system reset regardless of timer mode. Any read of WDTCTL reads 069h in the upper byte. The WDT+ counter clock should be slower or equal than the system (MCLK) frequency.

10.2.1 Watchdog Timer+ Counter

The watchdog timer+ counter (WDTCNT) is a 16-bit up-counter that is not directly accessible by software. The WDTCNT is controlled and time intervals selected through the watchdog timer+ control register WDTCTL.

The WDTCNT can be sourced from ACLK or SMCLK. The clock source is selected with the WDTTSSEL bit.

10.2.2 Watchdog Mode

After a PUC condition, the WDT+ module is configured in the watchdog mode with an initial 32768 cycle reset interval using the DCOCLK. The user must setup, halt, or clear the WDT+ prior to the expiration of the initial reset interval or another PUC will be generated. When the WDT+ is configured to operate in watchdog mode, either writing to WDTCTL with an incorrect password, or expiration of the selected time interval triggers a PUC. A PUC resets the WDT+ to its default condition and configures the RST/NMI pin to reset mode.

10.2.3 Interval Timer Mode

Setting the WDTTMSEL bit to 1 selects the interval timer mode. This mode can be used to provide periodic interrupts. In interval timer mode, the WDTIFG flag is set at the expiration of the selected time interval. A PUC is not generated in interval timer mode at expiration of the selected timer interval and the WDTIFG enable bit WDTIE remains unchanged.

When the WDTIE bit and the GIE bit are set, the WDTIFG flag requests an interrupt. The WDTIFG interrupt flag is automatically reset when its interrupt request is serviced, or may be reset by software. The interrupt vector address in interval timer mode is different from that in watchdog mode.

Note

Modifying the Watchdog Timer+

The WDT+ interval should be changed together with WDTCNTCL = 1 in a single instruction to avoid an unexpected immediate PUC or interrupt.

The WDT+ should be halted before changing the clock source to avoid a possible incorrect interval.

10.2.4 Watchdog Timer+ Interrupts

The WDT+ uses two bits in the SFRs for interrupt control.

- The WDT+ interrupt flag, WDTIFG, located in IFG1.0
- The WDT+ interrupt enable, WDTIE, located in IE1.0

When using the WDT+ in the watchdog mode, the WDTIFG flag sources a reset vector interrupt. The WDTIFG can be used by the reset interrupt service routine to determine if the watchdog caused the device to reset. If the flag is set, then the watchdog timer+ initiated the reset condition either by timing out or by a security key violation. If WDTIFG is cleared, the reset was caused by a different source.

When using the WDT+ in interval timer mode, the WDTIFG flag is set after the selected time interval and requests a WDT+ interval timer interrupt if the WDTIE and the GIE bits are set. The interval timer interrupt vector is different from the reset vector used in watchdog mode. In interval timer mode, the WDTIFG flag is reset automatically when the interrupt is serviced, or can be reset with software.
10.2.5 Watchdog Timer+ Clock Fail-Safe Operation

The WDT+ module provides a fail-safe clocking feature assuring the clock to the WDT+ cannot be disabled while in watchdog mode. This means the low-power modes may be affected by the choice for the WDT+ clock. For example, if ACLK is the WDT+ clock source, LPM4 will not be available, because the WDT+ will prevent ACLK from being disabled. Also, if ACLK or SMCLK fail while sourcing the WDT+, the WDT+ clock source is automatically switched to MCLK. In this case, if MCLK is sourced from a crystal, and the crystal has failed, the fail-safe feature will activate the DCO and use it as the source for MCLK.

When the WDT+ module is used in interval timer mode, there is no fail-safe feature for the clock source.

10.2.6 Operation in Low-Power Modes

The MSP430 devices have several low-power modes. Different clock signals are available in different low-power modes. The requirements of the user’s application and the type of clocking used determine how the WDT+ should be configured. For example, the WDT+ should not be configured in watchdog mode with SMCLK as its clock source if the user wants to use low-power mode 3 because the WDT+ will keep SMCLK enabled for its clock source, increasing the current consumption of LPM3. When the watchdog timer+ is not required, the WDTHOLD bit can be used to hold the WDTCNT, reducing power consumption.

10.2.7 Software Examples

Any write operation to WDTCTL must be a word operation with 05Ah (WDTPW) in the upper byte:

```assembly
; Periodically clear an active watchdog
MOV #WDTPW+WDTCNTCL, &WDTCTL
;
; Change watchdog timer+ interval
MOV #WDTPW+WDTCNTL+WDTSEL, &WDTCTL
;
; Stop the watchdog
MOV #WDTPW+WDTHOLD, &WDTCTL
;
; Change WDT+ to interval timer mode, clock/8192 interval
MOV #WDTPW+WDTCNTCL+WDTTMSEL+WDTIS0, &WDTCTL
```
### 10.3 Watchdog Timer+ Registers

Table 10-1 lists the memory-mapped registers for the Watchdog Timer+.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>120h</td>
<td>WDTCTL</td>
<td>Watchdog timer+ control</td>
<td>Read/write</td>
<td>6900h with PUC</td>
<td>Section 10.3.1</td>
</tr>
<tr>
<td>0h</td>
<td>IE1</td>
<td>SFR interrupt enable 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 10.3.2</td>
</tr>
<tr>
<td>2h</td>
<td>IFG1</td>
<td>SFR interrupt flag 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 10.3.3</td>
</tr>
</tbody>
</table>
10.3.1 WDTCTL Register
Watchdog Timer+ Control Register

WDTCTL is shown in Figure 10-2 and described in Table 10-2.

Return to Table 10-1.

Figure 10-2. WDTCTL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-8</td>
<td>WDTPW</td>
<td>R/W</td>
<td>69h</td>
<td>Watchdog timer+ password. Always reads as 069h. Must be written as 05Ah. Writing any other value generates a PUC.</td>
</tr>
<tr>
<td>7</td>
<td>WDTHOLD</td>
<td>R/W</td>
<td>0h</td>
<td>Watchdog timer+ hold. This bit stops the watchdog timer+. Setting WDTHOLD = 1 when the WDT+ is not in use conserves power. 0b = Watchdog timer+ is not stopped 1b = Watchdog timer+ is stopped</td>
</tr>
<tr>
<td>6</td>
<td>WDTNMIES</td>
<td>R/W</td>
<td>0h</td>
<td>Watchdog timer+ NMI edge select. This bit selects the interrupt edge for the NMI interrupt when WDTNMI = 1. Modifying this bit can trigger an NMI. Modify this bit when WDTIE = 0 to avoid triggering an accidental NMI. 0b = NMI on rising edge 1b = NMI on falling edge</td>
</tr>
<tr>
<td>5</td>
<td>WDTNMI</td>
<td>R/W</td>
<td>0h</td>
<td>Watchdog timer+ NMI select. This bit selects the function for the RST/NMI pin. 0b = Reset function 1b = NMI function</td>
</tr>
<tr>
<td>4</td>
<td>WDTTMSSEL</td>
<td>R/W</td>
<td>0h</td>
<td>Watchdog timer+ mode select 0b = Watchdog mode 1b = Interval timer mode</td>
</tr>
<tr>
<td>3</td>
<td>WDTCNTCL</td>
<td>R/W</td>
<td>0h</td>
<td>Watchdog timer+ counter clear. Setting WDTCNTCL = 1 clears the count value to 0000h. WDTCNTCL is automatically reset. 0b = No action 1b = WDTCNT = 0000h</td>
</tr>
<tr>
<td>2</td>
<td>WDTSSSEL</td>
<td>R/W</td>
<td>0h</td>
<td>Watchdog timer+ clock source select 0b = SMCLK 1b = ACLK</td>
</tr>
<tr>
<td>1-0</td>
<td>WDTISx</td>
<td>R/W</td>
<td>0h</td>
<td>Watchdog timer+ interval select. These bits select the watchdog timer+ interval to set the WDTIFG flag or generate a PUC. 00b = Watchdog clock source /32768 01b = Watchdog clock source /8192 10b = Watchdog clock source /512 11b = Watchdog clock source /64</td>
</tr>
</tbody>
</table>
### 10.3.2 IE1 Register

Interrupt Enable 1 Register

IE1 is shown in Figure 10-3 and described in Table 10-3.

Return to Table 10-1.

#### Figure 10-3. IE1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-5</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See device-specific data sheet.</td>
</tr>
<tr>
<td>4</td>
<td>NMIIE</td>
<td>R/W</td>
<td>0h</td>
<td>NMI interrupt enable. This bit enables the NMI interrupt. Because other bits in IE1 may be used for other modules, it is recommended to set or clear this bit using BIS.B or BIC.B instructions, rather than MOV.B or CLR.B instructions. 0b = Interrupt not enabled 1b = Interrupt enabled</td>
</tr>
<tr>
<td>3-1</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See device-specific data sheet.</td>
</tr>
<tr>
<td>0</td>
<td>WDTIE</td>
<td>R/W</td>
<td>0h</td>
<td>Watchdog timer+ interrupt enable. This bit enables the WDTIFG interrupt for interval timer mode. It is not necessary to set this bit for watchdog mode. Because other bits in IE1 may be used for other modules, it is recommended to set or clear this bit using BIS.B or BIC.B instructions, rather than MOV.B or CLR.B instructions. 0b = Interrupt not enabled 1b = Interrupt enabled</td>
</tr>
</tbody>
</table>

#### Table 10-3. IE1 Register Field Descriptions
10.3.3 IFG1 Register

Interrupt Flag 1 Register

IFG1 is shown in Figure 10-4 and described in Table 10-4.

Return to Table 10-1.

Figure 10-4. IFG1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-5</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See device-specific data sheet.</td>
</tr>
<tr>
<td>4</td>
<td>NMIIFG</td>
<td>R/W</td>
<td>0h</td>
<td>NMI interrupt flag. NMIIFG must be reset by software. Because other bits in IFG1 may be used for other modules, it is recommended to clear NMIIFG by using BIS.B or BIC.B instructions, rather than MOV.B or CLR.B instructions. 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
<tr>
<td>3-1</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See device-specific data sheet.</td>
</tr>
<tr>
<td>0</td>
<td>WDTIFG</td>
<td>R/W</td>
<td>0h</td>
<td>Watchdog timer+ interrupt flag. In watchdog mode, WDTIFG remains set until reset by software. In interval mode, WDTIFG is reset automatically by servicing the interrupt, or can be reset by software. Because other bits in IFG1 may be used for other modules, it is recommended to clear WDTIFG by using BIS.B or BIC.B instructions, rather than MOV.B or CLR.B instructions. 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
</tbody>
</table>
This chapter describes the hardware multiplier. The hardware multiplier is implemented in some MSP430x2xx devices.

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</tr>
<tr>
<td>11.3 Hardware Multiplier Registers</td>
<td>372</td>
</tr>
</tbody>
</table>
11.1 Hardware Multiplier Introduction

The hardware multiplier is a peripheral and is not part of the MSP430 CPU. This means, its activities do not interfere with the CPU activities. The multiplier registers are peripheral registers that are loaded and read with CPU instructions.

The hardware multiplier supports:

- Unsigned multiply
- Signed multiply
- Unsigned multiply accumulate
- Signed multiply accumulate
- 16x16 bits, 16x8 bits, 8x16 bits, 8x8 bits

The hardware multiplier block diagram is shown in Figure 11-1.

![Figure 11-1. Hardware Multiplier Block Diagram](image-url)

11.2 Hardware Multiplier Operation

The hardware multiplier supports unsigned multiply, signed multiply, unsigned multiply accumulate, and signed multiply accumulate operations. The type of operation is selected by the address the first operand is written to.

The hardware multiplier has two 16-bit operand registers, OP1 and OP2, and three result registers, RESLO, RESHI, and SUMEXT. RESLO stores the low word of the result, RESHI stores the high word of the result, and SUMEXT stores information about the result. The result is ready in three MCLK cycles and can be read with the next instruction after writing to OP2, except when using an indirect addressing mode to access the result. When using indirect addressing for the result, a NOP is required before the result is ready.

11.2.1 Operand Registers

The operand one register OP1 has four addresses, shown in Table 11-1, used to select the multiply mode. Writing the first operand to the desired address selects the type of multiply operation but does not start any
operation. Writing the second operand to the operand two register OP2 initiates the multiply operation. Writing OP2 starts the selected operation with the values stored in OP1 and OP2. The result is written into the three result registers RESLO, RESHI, and SUMEXT.

Repeated multiply operations may be performed without reloading OP1 if the OP1 value is used for successive operations. It is not necessary to re-write the OP1 value to perform the operations.

### Table 11-1. OP1 Addresses

<table>
<thead>
<tr>
<th>OP1 Address</th>
<th>Register Name</th>
<th>Operation</th>
</tr>
</thead>
<tbody>
<tr>
<td>0130h</td>
<td>MPY</td>
<td>Unsigned multiply</td>
</tr>
<tr>
<td>0132h</td>
<td>MPYS</td>
<td>Signed multiply</td>
</tr>
<tr>
<td>0134h</td>
<td>MAC</td>
<td>Unsigned multiply accumulate</td>
</tr>
<tr>
<td>0136h</td>
<td>MACS</td>
<td>Signed multiply accumulate</td>
</tr>
</tbody>
</table>

#### 11.2.2 Result Registers

The result low register RESLO holds the lower 16-bits of the calculation result. The result high register RESHI contents depend on the multiply operation and are listed in Table 11-2.

### Table 11-2. RESHI Contents

<table>
<thead>
<tr>
<th>Mode</th>
<th>RESHI Contents</th>
</tr>
</thead>
<tbody>
<tr>
<td>MPY</td>
<td>Upper 16-bits of the result</td>
</tr>
<tr>
<td>MPYS</td>
<td>The MSB is the sign of the result. The remaining bits are the upper 15-bits of the result. Two's complement notation is used for the result.</td>
</tr>
<tr>
<td>MAC</td>
<td>Upper 16-bits of the result</td>
</tr>
<tr>
<td>MACS</td>
<td>Upper 16-bits of the result. Two's complement notation is used for the result.</td>
</tr>
</tbody>
</table>

The sum extension registers SUMEXT contents depend on the multiply operation and are listed in Table 11-3.

### Table 11-3. SUMEXT Contents

<table>
<thead>
<tr>
<th>Mode</th>
<th>SUMEXT</th>
</tr>
</thead>
<tbody>
<tr>
<td>MPY</td>
<td>SUMEXT is always 0000h</td>
</tr>
<tr>
<td>MPYS</td>
<td>SUMEXT contains the extended sign of the result 00000h = Result was positive or zero 00000h = Result was negative</td>
</tr>
<tr>
<td>MAC</td>
<td>SUMEXT contains the carry of the result 0000h = No carry for result 0001h = Result has a carry</td>
</tr>
<tr>
<td>MACS</td>
<td>SUMEXT contains the extended sign of the result 00000h = Result was positive or zero 00000h = Result was negative</td>
</tr>
</tbody>
</table>
11.2.2.1 MACS Underflow and Overflow

The multiplier does not automatically detect underflow or overflow in the MACS mode. The accumulator range for positive numbers is 0 to 7FFF FFFFh and for negative numbers is 0FFFF FFFFh to 8000 0000h. An underflow occurs when the sum of two negative numbers yields a result that is in the range for a positive number. An overflow occurs when the sum of two positive numbers yields a result that is in the range for a negative number. In both of these cases, the SUMEXT register contains the sign of the result, 0FFFFh for overflow and 0000h for underflow. User software must detect and handle these conditions appropriately.

11.2.3 Software Examples

Examples for all multiplier modes follow. All 8x8 modes use the absolute address for the registers because the assembler will not allow .B access to word registers when using the labels from the standard definitions file.

There is no sign extension necessary in software. Accessing the multiplier with a byte instruction during a signed operation will automatically cause a sign extension of the byte within the multiplier module.

```assembly
; 16x16 Unsigned Multiply
MOV   #01234h,&MPY   ; Load first operand
MOV   #05678h,&OP2   ; Load second operand
; ...                  ; Process results
; 8x8 Unsigned Multiply. Absolute addressing.
MOV.B #012h,40130h   ; Load first operand
MOV.B #034h,40138h   ; Load 2nd operand
; ...                  ; Process results
; 16x16 Signed Multiply
MOV   #01234h,&MPYS  ; Load first operand
MOV   #05678h,&OP2   ; Load 2nd operand
; ...                  ; Process results
; 8x8 Signed Multiply. Absolute addressing.
MOV.B #012h,40132h   ; Load first operand
MOV.B #034h,40138h   ; Load 2nd operand
; ...                  ; Process results
; 16x16 Unsigned Multiply Accumulate
MOV   #01234h,&MAC   ; Load first operand
MOV   #05678h,&OP2   ; Load 2nd operand
; ...                  ; Process results
; 8x8 Unsigned Multiply Accumulate. Absolute addressing
MOV.B #012h,40134h   ; Load first operand
MOV.B #034h,40138h   ; Load 2nd operand
; ...                  ; Process results
; 16x16 Signed Multiply Accumulate
MOV   #01234h,&MACS  ; Load first operand
MOV   #05678h,&OP2   ; Load 2nd operand
; ...                  ; Process results
; 8x8 Signed Multiply Accumulate. Absolute addressing
MOV.B #012h,40136h   ; Load first operand
MOV.B #034h,R5       ; Temp. location for 2nd operand
MOV   R5,&OP2        ; Load 2nd operand
; ...                  ; Process results
```
11.2.4 Indirect Addressing of RESLO

When using indirect or indirect autoincrement addressing mode to access the result registers, at least one instruction is needed between loading the second operand and accessing one of the result registers:

; Access multiplier results with indirect addressing
MOV   #RESLO,R5     ; RESLO address in R5 for indirect
MOV   &OPER1,&MPY   ; Load 1st operand
MOV   &OPER2,&OP2   ; Load 2nd operand
NOP       ; Need one cycle
MOV   @R5+,&xxx     ; Move RESLO
MOV   @R5,&xxx      ; Move RESHI

11.2.5 Using Interrupts

If an interrupt occurs after writing OP1, but before writing OP2, and the multiplier is used in servicing that interrupt, the original multiplier mode selection is lost and the results are unpredictable. To avoid this, disable interrupts before using the hardware multiplier or do not use the multiplier in interrupt service routines.

; Disable interrupts before using the hardware multiplier
DINT      ; Disable interrupts
NOP       ; Required for DINT
MOV   #xxh,&MPY    ; Load 1st operand
MOV   #xxh,&OP2    ; Load 2nd operand
EINT      ; Interrupts may be enable before
           ; Process results
11.3 Hardware Multiplier Registers

The hardware multiplier registers are listed in Table 11-4.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
</tr>
</thead>
<tbody>
<tr>
<td>0130h</td>
<td>MPY</td>
<td>Operand one - multiply</td>
<td>Read/write</td>
<td>Unchanged</td>
</tr>
<tr>
<td>0132h</td>
<td>MPYS</td>
<td>Operand one - signed multiply</td>
<td>Read/write</td>
<td>Unchanged</td>
</tr>
<tr>
<td>0134h</td>
<td>MAC</td>
<td>Operand one - multiply accumulate</td>
<td>Read/write</td>
<td>Unchanged</td>
</tr>
<tr>
<td>0136h</td>
<td>MACS</td>
<td>Operand one - signed multiply accumulate</td>
<td>Read/write</td>
<td>Unchanged</td>
</tr>
<tr>
<td>0138h</td>
<td>OP2</td>
<td>Operand two</td>
<td>Read/write</td>
<td>Unchanged</td>
</tr>
<tr>
<td>013Ah</td>
<td>RESLO</td>
<td>Result low word</td>
<td>Read/write</td>
<td>Undefined</td>
</tr>
<tr>
<td>013Ch</td>
<td>RESHI</td>
<td>Result high word</td>
<td>Read/write</td>
<td>Undefined</td>
</tr>
<tr>
<td>013 Eh</td>
<td>SUMEXT</td>
<td>Sum extension register</td>
<td>Read</td>
<td>Undefined</td>
</tr>
</tbody>
</table>
Chapter 12

Timer_A

Timer_A is a 16-bit timer/counter with multiple capture/compare registers. This chapter describes the operation of the Timer_A of the MSP430x2xx device family.

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12.2 Timer_A Operation............................... 375
12.3 Timer_A Registers............................... 387
12.1 Timer_A Introduction

Timer_A is a 16-bit timer/counter with three capture/compare registers. Timer_A can support multiple capture/comparcs, PWM outputs, and interval timing. Timer_A also has extensive interrupt capabilities. Interrupts may be generated from the counter on overflow conditions and from each of the capture/compare registers.

Timer_A features include:

- Asynchronous 16-bit timer/counter with four operating modes
- Selectable and configurable clock source
- Two or three configurable capture/compare registers
- Configurable outputs with PWM capability
- Asynchronous input and output latching
- Interrupt vector register for fast decoding of all Timer_A interrupts

The block diagram of Timer_A is shown in Figure 12-1.

|Note|

Use of the Word Count

Count is used throughout this chapter. It means the counter must be in the process of counting for the action to take place. If a particular value is directly written to the counter, then an associated action will not take place.
12.2 Timer_A Operation

The Timer_A module is configured with user software. The setup and operation of Timer_A is discussed in the following sections.

12.2.1 16-Bit Timer Counter

The 16-bit timer/counter register, TAR, increments or decrements (depending on mode of operation) with each rising edge of the clock signal. TAR can be read or written with software. Additionally, the timer can generate an interrupt when it overflows.

TAR may be cleared by setting the TACLR bit. Setting TACLR also clears the clock divider and count direction for up/down mode.
**Note**

**Modifying Timer_A Registers**

It is recommended to stop the timer before modifying its operation (with exception of the interrupt enable, and interrupt flag) to avoid errant operating conditions.

When the timer clock is asynchronous to the CPU clock, any read from TAR should occur while the timer is not operating or the results may be unpredictable. Alternatively, the timer may be read multiple times while operating, and a majority vote taken in software to determine the correct reading. Any write to TAR will take effect immediately.

**12.2.1.1 Clock Source Select and Divider**

The timer clock can be sourced from ACLK, SMCLK, or externally via TACLK or INCLK. The clock source is selected with the TASSELx bits. The selected clock source may be passed directly to the timer or divided by 2, 4, or 8, using the IDx bits. The timer clock divider is reset when TACLR is set.

**12.2.2 Starting the Timer**

The timer may be started, or restarted in the following ways:

- The timer counts when MCx > 0 and the clock source is active.
- When the timer mode is either up or up/down, the timer may be stopped by writing 0 to TACCR0. The timer may then be restarted by writing a nonzero value to TACCR0. In this scenario, the timer starts incrementing in the up direction from zero.

**12.2.3 Timer Mode Control**

The timer has four modes of operation as described in **Table 12-1**: stop, up, continuous, and up/down. The operating mode is selected with the MCx bits.

<table>
<thead>
<tr>
<th>MCx</th>
<th>Mode</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>Stop</td>
<td>The timer is halted.</td>
</tr>
<tr>
<td>01</td>
<td>Up</td>
<td>The timer repeatedly counts from zero to the value of TACCR0.</td>
</tr>
<tr>
<td>10</td>
<td>Continuous</td>
<td>The timer repeatedly counts from zero to 0FFFFh.</td>
</tr>
<tr>
<td>11</td>
<td>Up/down</td>
<td>The timer repeatedly counts from zero up to the value of TACCR0 and back down to zero.</td>
</tr>
</tbody>
</table>

**12.2.3.1 Up Mode**

The up mode is used if the timer period must be different from 0FFFFh counts. The timer repeatedly counts up to the value of compare register TACCR0, which defines the period, as shown in **Figure 12-2**. The number of timer counts in the period is TACCR0+1. When the timer value equals TACCR0 the timer restarts counting from zero. If up mode is selected when the timer value is greater than TACCR0, the timer immediately restarts counting from zero.

![Figure 12-2. Up Mode](image)

The TACCR0 CCIFG interrupt flag is set when the timer **counts** to the TACCR0 value. The TAIFG interrupt flag is set when the timer **counts** from TACCR0 to zero. **Figure 12-3** shows the flag set cycle.
12.2.3.2 Changing the Period Register TACCR0

When changing TACCR0 while the timer is running, if the new period is greater than or equal to the old period, or greater than the current count value, the timer counts up to the new period. If the new period is less than the current count value, the timer rolls to zero. However, one additional count may occur before the counter rolls to zero.

12.2.3.3 Continuous Mode

In the continuous mode, the timer repeatedly counts up to 0xFFFFh and restarts from zero as shown in Figure 12-4. The capture/compare register TACCR0 works the same way as the other capture/compare registers.

The TAIFG interrupt flag is set when the timer counts from 0xFFFFh to zero. Figure 12-5 shows the flag set cycle.

12.2.3.4 Use of the Continuous Mode

The continuous mode can be used to generate independent time intervals and output frequencies. Each time an interval is completed, an interrupt is generated. The next time interval is added to the TACCRx register in the interrupt service routine. Figure 12-6 shows two separate time intervals $t_0$ and $t_1$ being added to the capture/compare registers. In this usage, the time interval is controlled by hardware, not software, without impact from interrupt latency. Up to three independent time intervals or output frequencies can be generated using all three capture/compare registers.
Time intervals can be produced with other modes as well, where TACCR0 is used as the period register. Their handling is more complex since the sum of the old TACCRx data and the new period can be higher than the TACCR0 value. When the previous TACCRx value plus $t_x$ is greater than the TACCR0 data, TACCR0 + 1 must be subtracted to obtain the correct time interval.

### 12.2.3.5 Up/Down Mode

The up/down mode is used if the timer period must be different from 0xFFFFh counts, and if a symmetrical pulse generation is needed. The timer repeatedly counts up to the value of compare register TACCR0 and back down to zero, as shown in Figure 12-7. The period is twice the value in TACCR0.

The count direction is latched. This allows the timer to be stopped and then restarted in the same direction it was counting before it was stopped. If this is not desired, the TACLR bit must be set to clear the direction. The TACLR bit also clears the TAR value and the timer clock divider.

In up/down mode, the TACCR0 CCIFG interrupt flag and the TAIFG interrupt flag are set only once during a period, separated by 1/2 the timer period. The TACCR0 CCIFG interrupt flag is set when the timer counts from TACCR0 – 1 to TACCR0, and TAIFG is set when the timer completes counting down from 0001h to 0000h. Figure 12-8 shows the flag set cycle.
12.2.3.6 Changing the Period Register TACCR0

When changing TACCR0 while the timer is running, and counting in the down direction, the timer continues its descent until it reaches zero. The value in TACCR0 is latched into TACL0 immediately, however the new period takes effect after the counter counts down to zero.

When the timer is counting in the up direction, and the new period is greater than or equal to the old period, or greater than the current count value, the timer counts up to the new period before counting down. When the timer is counting in the up direction, and the new period is less than the current count value, the timer begins counting down. However, one additional count may occur before the counter begins counting down.

12.2.3.7 Use of the Up/Down Mode

The up/down mode supports applications that require dead times between output signals (See section Timer_A Output Unit). For example, to avoid overload conditions, two outputs driving an H-bridge must never be in a high state simultaneously. In the example shown in Figure 12-9 the $t_{\text{dead}}$ is:

$$t_{\text{dead}} = t_{\text{timer}} (TACCR1 - TACCR2)$$

Where,

- $t_{\text{dead}} = \text{Time during which both outputs need to be inactive}$
- $t_{\text{timer}} = \text{Cycle time of the timer clock}$
- $TACCRx = \text{Content of capture/compare register } x$

The TACCRx registers are not buffered. They update immediately when written to. Therefore, any required dead time will not be maintained automatically.
12.2.4 Capture/Compare Blocks

Two or three identical capture/compare blocks, TACCRx, are present in Timer_A. Any of the blocks may be used to capture the timer data, or to generate time intervals.

Capture Mode

The capture mode is selected when CAP = 1. Capture mode is used to record time events. It can be used for speed computations or time measurements. The capture inputs CClxA and CClxB are connected to external pins or internal signals and are selected with the CCISx bits. The CMx bits select the capture edge of the input signal as rising, falling, or both. A capture occurs on the selected edge of the input signal. If a capture occurs:

- The timer value is copied into the TACCRx register
- The interrupt flag CCIFG is set

The input signal level can be read at any time via the CCI bit. MSP430x2xx family devices may have different signals connected to CClxA and CClxB. See the device-specific data sheet for the connections of these signals.

The capture signal can be asynchronous to the timer clock and cause a race condition. Setting the SCS bit will synchronize the capture with the next timer clock. Setting the SCS bit to synchronize the capture signal with the timer clock is recommended. This is illustrated in Figure 12-10.

Overflow logic is provided in each capture/compare register to indicate if a second capture was performed before the value from the first capture was read. Bit COV is set when this occurs as shown in Figure 12-11. COV must be reset with software.
12.2.4.1 Capture Initiated by Software

Captures can be initiated by software. The CMx bits can be set for capture on both edges. Software then sets CCIS1 = 1 and toggles bit CCIS0 to switch the capture signal between \( V_{CC} \) and GND, initiating a capture each time CCIS0 changes state:

\[
\text{MOV} \ #\text{CAP}+\text{SCS}+\text{CCIS1}+\text{CM}_3, &\text{TACCTLx} \ ; \text{Setup TACCTLx} \\
\text{XOR} \ #\text{CCIS0}, &\text{TACCTLx} \ ; \text{TACCRx} = \text{TAR}
\]

12.2.4.2 Compare Mode

The compare mode is selected when CAP = 0. The compare mode is used to generate PWM output signals or interrupts at specific time intervals. When TAR counts to the value in a TACCRx:

- Interrupt flag CCIFG is set
- Internal signal EQUx = 1
- EQUx affects the output according to the output mode
- The input signal CCI is latched into SCCI

12.2.5 Output Unit

Each capture/compare block contains an output unit. The output unit is used to generate output signals such as PWM signals. Each output unit has eight operating modes that generate signals based on the EQU0 and EQUx signals.

12.2.5.1 Output Modes

The output modes are defined by the OUTMODx bits and are described in Table 12-2. The OUTx signal is changed with the rising edge of the timer clock for all modes except mode 0. Output modes 2, 3, 6, and 7 are not useful for output unit 0, because EQUx = EQU0.

<table>
<thead>
<tr>
<th>OUTMODx</th>
<th>Mode</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>000</td>
<td>Output</td>
<td>The output signal OUTx is defined by the OUTx bit. The OUTx signal updates immediately when OUTx is updated.</td>
</tr>
</tbody>
</table>
Table 12-2. Output Modes (continued)

<table>
<thead>
<tr>
<th>OUTMODx</th>
<th>Mode</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>001</td>
<td>Set</td>
<td>The output is set when the timer counts to the TACCRx value. It remains set until a reset of the timer, or until another output mode is selected and affects the output.</td>
</tr>
<tr>
<td>010</td>
<td>Toggle/Reset</td>
<td>The output is toggled when the timer counts to the TACCRx value. It is reset when the timer counts to the TACCR0 value.</td>
</tr>
<tr>
<td>011</td>
<td>Set/Reset</td>
<td>The output is set when the timer counts to the TACCRx value. It is reset when the timer counts to the TACCR0 value.</td>
</tr>
<tr>
<td>100</td>
<td>Toggle</td>
<td>The output is toggled when the timer counts to the TACCRx value. The output period is double the timer period.</td>
</tr>
<tr>
<td>101</td>
<td>Reset</td>
<td>The output is reset when the timer counts to the TACCRx value. It remains reset until another output mode is selected and affects the output.</td>
</tr>
<tr>
<td>110</td>
<td>Toggle/Set</td>
<td>The output is toggled when the timer counts to the TACCRx value. It is set when the timer counts to the TACCR0 value.</td>
</tr>
<tr>
<td>111</td>
<td>Reset/Set</td>
<td>The output is reset when the timer counts to the TACCRx value. It is set when the timer counts to the TACCR0 value.</td>
</tr>
</tbody>
</table>

12.2.5.2 Output Example — Timer in Up Mode

The OUTx signal is changed when the timer counts up to the TACCRx value, and rolls from TACCR0 to zero, depending on the output mode. An example is shown in Figure 12-12 using TACCR0 and TACCR1.

![Figure 12-12. Output Example—Timer in Up Mode](image)

12.2.5.3 Output Example — Timer in Continuous Mode

The OUTx signal is changed when the timer reaches the TACCRx and TACCR0 values, depending on the output mode. An example is shown in Figure 12-13 using TACCR0 and TACCR1.
Figure 12-13. Output Example—Timer in Continuous Mode
12.2.5.4 Output Example — Timer in Up/Down Mode

The OUTx signal changes when the timer equals TACCRx in either count direction and when the timer equals TACCR0, depending on the output mode. An example is shown in Figure 12-14 using TACCR0 and TACCR2.

![Figure 12-14. Output Example—Timer in Up/Down Mode](image)

**Note**

**Switching Between Output Modes**

When switching between output modes, one of the OUTMODx bits should remain set during the transition, unless switching to mode 0. Otherwise, output glitching can occur because a NOR gate decodes output mode 0. A safe method for switching between output modes is to use output mode 7 as a transition state:

```
BIS   #OUTMOD_7,&TACCTLx   ; Set output mode=7
BIC   #OUTMODx, &TACCTLx   ; Clear unwanted bits
```
12.2.6 Timer_A Interrupts

Two interrupt vectors are associated with the 16-bit Timer_A module:

- TACCR0 interrupt vector for TACCR0 CCIFG
- TAIV interrupt vector for all other CCIFG flags and TAIFG

In capture mode any CCIFG flag is set when a timer value is captured in the associated TACCRx register. In compare mode, any CCIFG flag is set if TAR counts to the associated TACCRx value. Software may also set or clear any CCIFG flag. All CCIFG flags request an interrupt when their corresponding CCIE bit and the GIE bit are set.

12.2.6.1 TACCR0 Interrupt

The TACCR0 CCIFG flag has the highest Timer_A interrupt priority and has a dedicated interrupt vector as shown in Figure 12-15. The TACCR0 CCIFG flag is automatically reset when the TACCR0 interrupt request is serviced.

![Figure 12-15. Capture/Compare TACCR0 Interrupt Flag](image)

12.2.6.2 TAIV, Interrupt Vector Generator

The TACCR1 CCIFG, TACCR2 CCIFG, and TAIFG flags are prioritized and combined to source a single interrupt vector. The interrupt vector register TAIV is used to determine which flag requested an interrupt.

The highest priority enabled interrupt generates a number in the TAIV register (see register description). This number can be evaluated or added to the program counter to automatically enter the appropriate software routine. Disabled Timer_A interrupts do not affect the TAIV value.

Any access, read or write, of the TAIV register automatically resets the highest pending interrupt flag. If another interrupt flag is set, another interrupt is immediately generated after servicing the initial interrupt. For example, if the TACCR1 and TACCR2 CCIFG flags are set when the interrupt service routine accesses the TAIV register, TACCR1 CCIFG is reset automatically. After the RETI instruction of the interrupt service routine is executed, the TACCR2 CCIFG flag will generate another interrupt.
12.2.6.3 TAIV Software Example

The following software example shows the recommended use of TAIV and the handling overhead. The TAIV value is added to the PC to automatically jump to the appropriate routine.

The numbers at the right margin show the necessary CPU cycles for each instruction. The software overhead for different interrupt sources includes interrupt latency and return-from-interrupt cycles, but not the task handling itself. The latencies are:

- Capture/compare block TACCR0: 11 cycles
- Capture/compare blocks TACCR1, TACCR2: 16 cycles
- Timer overflow TAIFG: 14 cycles

```assembly
; Interrupt handler for TACCR0 CCIFG
CCIFG_0_HND
;     ... ; Start of handler   Interrupt latency     6
RETI

; Interrupt handler for TAIFG, TACCR1 and TACCR2 CCIFG
TA_HND
;     ... ; Interrupt latency     6
ADD   &TAIV,PC ; Add offset to Jump table   3
RETI ; Vector 0: No interrupt     5
JMP   CCIFG_1_HND ; Vector 2: TACCR1    2
RETI ; Vector 6: Reserved         5
RETI ; Vector 8: Reserved         5
TAIFG_HND ; Vector 10: TAIFG Flag
;     ... ; Task starts here
RETI     5
CCIFG_2_HND ; Vector 4: TACCR2
;     ... ; Task starts here
RETI     5
CCIFG_1_HND ; Vector 2: TACCR1
;     ... ; Task starts here
RETI     5
RETI ; Back to main program       5
```
### 12.3 Timer_A Registers

Table 12-3 lists the memory-mapped registers for the Timer_A.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>160h</td>
<td>TACTL</td>
<td>Timer_A control</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 12.3.1</td>
</tr>
<tr>
<td>170h</td>
<td>TAR</td>
<td>Timer_A counter</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 12.3.2</td>
</tr>
<tr>
<td>162h</td>
<td>TACCTL0</td>
<td>Timer_A capture/compare control 0</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 12.3.3</td>
</tr>
<tr>
<td>172h</td>
<td>TACCR0</td>
<td>Timer_A capture/compare 0</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 12.3.4</td>
</tr>
<tr>
<td>164h</td>
<td>TACCTL1</td>
<td>Timer_A capture/compare control 1</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 12.3.3</td>
</tr>
<tr>
<td>174h</td>
<td>TACCR1</td>
<td>Timer_A capture/compare 1</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 12.3.4</td>
</tr>
<tr>
<td>166h</td>
<td>TACCTL2(1)</td>
<td>Timer_A capture/compare control 2</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 12.3.3</td>
</tr>
<tr>
<td>176h</td>
<td>TACCR2(1)</td>
<td>Timer_A capture/compare 2</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 12.3.4</td>
</tr>
<tr>
<td>12Eh</td>
<td>TAIV</td>
<td>Timer_A interrupt vector</td>
<td>Read</td>
<td>00h with POR</td>
<td>Section 12.3.5</td>
</tr>
</tbody>
</table>

(1) Not present on MSP430 devices with Timer_A2 such as MSP430F20xx and other devices.
12.3.1 TACTL Register

Timer_A Control Register

TACTL is shown in Figure 12-16 and described in Table 12-4.

Return to Table 12-3.

Figure 12-16. TACTL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-10</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td>Unused</td>
</tr>
</tbody>
</table>
| 9-8  | TASSELx | R/W  | 0h    | Timer_A clock source select  
|      |        |      |       | 00b = TACLK  
|      |        |      |       | 01b = ACLK  
|      |        |      |       | 10b = SMCLK  
|      |        |      |       | 11b = INCLK (INCLK is device-specific and is often assigned to the inverted TBCLK) (see the device-specific data sheet) |
| 7-6  | IDx    | R/W  | 0h    | Input divider. These bits select the divider for the input clock.  
|      |        |      |       | 00b = /1  
|      |        |      |       | 01b = /2  
|      |        |      |       | 10b = /4  
|      |        |      |       | 11b = /8  |
| 5-4  | MCx    | R/W  | 0h    | Mode control. Set MCx = 00b when Timer_A is not in use to conserve power.  
|      |        |      |       | 00b = Stop mode: the timer is halted.  
|      |        |      |       | 01b = Up mode: the timer counts up to TACCR0.  
|      |        |      |       | 10b = Continuous mode: the timer counts up to 0FFFFh.  
|      |        |      |       | 11b = Up/down mode: the timer counts up to TACCR0 then down to 0000h.  |
| 3    | Unused | R/W  | 0h    | Unused      |
| 2    | TACLR  | R/W  | 0h    | Timer_A clear. Setting this bit resets TAR, the clock divider, and the count direction. The TACLR bit is automatically reset and always reads as zero. |
| 1    | TAIE   | R/W  | 0h    | Timer_A interrupt enable. This bit enables the TAIFG interrupt request.  
|      |        |      |       | 0b = Interrupt disabled  
|      |        |      |       | 1b = Interrupt enabled  |
| 0    | TAIFG  | R/W  | 0h    | Timer_A interrupt flag  
|      |        |      |       | 0b = No interrupt pending  
|      |        |      |       | 1b = Interrupt pending  |

Table 12-4. TACTL Register Field Descriptions
12.3.2 TAR Register

Timer_A Counter Register

TAR is shown in Figure 12-17 and described in Table 12-5.

Return to Table 12-3.

**Figure 12-17. TAR Register**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>TARx</td>
<td>R/W</td>
<td>0h</td>
<td>Timer_A register. The TAR register is the count of Timer_A.</td>
</tr>
</tbody>
</table>
12.3.3 TACCTLx Register

Timer_A Capture/Compare Control x Register

TACCTLx is shown in Figure 12-18 and described in Table 12-6.

Return to Table 12-3.

Figure 12-18. TACCTLx Register

![Figure 12-18. TACCTLx Register](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-14</td>
<td>CMx</td>
<td>R/W</td>
<td>0h</td>
<td>Capture mode&lt;br&gt;00b = No capture&lt;br&gt;01b = Capture on rising edge&lt;br&gt;10b = Capture on falling edge&lt;br&gt;11b = Capture on both rising and falling edges</td>
</tr>
<tr>
<td>13-12</td>
<td>CCISx</td>
<td>R/W</td>
<td>0h</td>
<td>Capture/compare input select. These bits select the TACCRx input signal. See the device-specific data sheet for specific signal connections.&lt;br&gt;00b = CCIxA&lt;br&gt;01b = CCIxB&lt;br&gt;10b = GND&lt;br&gt;11b = VCC</td>
</tr>
<tr>
<td>11</td>
<td>SCS</td>
<td>R/W</td>
<td>0h</td>
<td>Synchronize capture source. This bit is used to synchronize the capture input signal with the timer clock.&lt;br&gt;0b = Asynchronous capture&lt;br&gt;1b = Synchronous capture</td>
</tr>
<tr>
<td>10</td>
<td>SCCI</td>
<td>R</td>
<td>0h</td>
<td>Synchronized capture/compare input. The selected CCI input signal is latched with the EQUx signal and can be read via this bit</td>
</tr>
<tr>
<td>9</td>
<td>Unused</td>
<td>R</td>
<td>0h</td>
<td>Read only. Always reads as 0.</td>
</tr>
<tr>
<td>8</td>
<td>CAP</td>
<td>R/W</td>
<td>0h</td>
<td>Capture mode&lt;br&gt;0b = Compare mode&lt;br&gt;1b = Capture mode</td>
</tr>
<tr>
<td>7-5</td>
<td>OUTMODx</td>
<td>R/W</td>
<td>0h</td>
<td>Output mode. Modes 2, 3, 6, and 7 are not useful for TACCR0, because EQUx = EQU0.&lt;br&gt;000b = OUT bit value&lt;br&gt;001b = Set&lt;br&gt;010b = Toggle/reset&lt;br&gt;011b = Set/reset&lt;br&gt;100b = Toggle&lt;br&gt;101b = Reset&lt;br&gt;110b = Toggle/set&lt;br&gt;111b = Reset/set</td>
</tr>
<tr>
<td>Bit</td>
<td>Field</td>
<td>Type</td>
<td>Reset</td>
<td>Description</td>
</tr>
<tr>
<td>-----</td>
<td>-------</td>
<td>------</td>
<td>-------</td>
<td>-------------</td>
</tr>
</tbody>
</table>
| 4   | CCIE  | R/W  | 0h    | Capture/compare interrupt enable. This bit enables the interrupt request of the corresponding CCIFG flag.  
0b = Interrupt disabled  
1b = Interrupt enabled |
| 3   | CCI   | R    | 0h    | Capture/compare input. The selected input signal can be read by this bit. |
| 2   | OUT   | R/W  | 0h    | Output. For output mode 0, this bit directly controls the state of the output.  
0b = Output low  
1b = Output high |
| 1   | COV   | R/W  | 0h    | Capture overflow. This bit indicates a capture overflow occurred.  
COV must be reset with software.  
0b = No capture overflow occurred  
1b = Capture overflow occurred |
| 0   | CCIFG | R/W  | 0h    | Capture/compare interrupt flag  
0b = No interrupt pending  
1b = Interrupt pending |
12.3.4 TACCRx Register

Timer_A Capture/Compare x Register

TACCRx is shown in Figure 12-19 and described in Table 12-7.

Return to Table 12-3.

Figure 12-19. TACCRx Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
</table>
| 15-0 | TACCRx | R/W | 0h | Timer_A capture/compare register.  
Compare mode: TACCRx holds the data for the comparison to the timer value in the Timer_A Register, TAR.  
Capture mode: The Timer_A Register, TAR, is copied into the TACCRx register when a capture is performed. |
12.3.5 TAIV Register

Timer_A interrupt vector

TAIV is shown in Figure 12-20 and described in Table 12-8.

Return to Table 12-3.

Figure 12-20. TAIV Register

<table>
<thead>
<tr>
<th>15</th>
<th>14</th>
<th>13</th>
<th>12</th>
<th>11</th>
<th>10</th>
<th>9</th>
<th>8</th>
</tr>
</thead>
<tbody>
<tr>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
</tr>
</tbody>
</table>

Table 12-8. TAIV Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>TAIVx</td>
<td>R</td>
<td>0h</td>
<td>Timer_A interrupt vector value. See Table 12-9 for values.</td>
</tr>
</tbody>
</table>

Table 12-9. Timer_A Interrupt Vectors

<table>
<thead>
<tr>
<th>TAIV Contents</th>
<th>Interrupt Source</th>
<th>Interrupt Flag</th>
<th>Interrupt Priority</th>
</tr>
</thead>
<tbody>
<tr>
<td>00h</td>
<td>No interrupt pending</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>02h</td>
<td>Capture/compare 1</td>
<td>TACCR1 CCIFG</td>
<td>Highest</td>
</tr>
<tr>
<td>04h</td>
<td>Capture/compare 2 (1)</td>
<td>TACCR2 CCIFG</td>
<td></td>
</tr>
<tr>
<td>06h</td>
<td>Reserved</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>08h</td>
<td>Reserved</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>0Ah</td>
<td>Timer overflow</td>
<td>TAIFG</td>
<td></td>
</tr>
<tr>
<td>0Ch</td>
<td>Reserved</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>0Eh</td>
<td>Reserved</td>
<td>–</td>
<td>Lowest</td>
</tr>
</tbody>
</table>

(1) Not implemented in MSP430x20xx devices.
This page intentionally left blank.
Timer_B is a 16-bit timer/counter with multiple capture/compare registers. This chapter describes the operation of the Timer_B of the MSP430x2xx device family.

13.1 Timer_B Introduction
13.2 Timer_B Operation
13.3 Timer_B Registers
13.1 Timer_B Introduction

Timer_B is a 16-bit timer/counter with three or seven capture/compare registers. Timer_B can support multiple capture/compares, PWM outputs, and interval timing. Timer_B also has extensive interrupt capabilities. Interrupts may be generated from the counter on overflow conditions and from each of the capture/compare registers.

Timer_B features include:
- Asynchronous 16-bit timer/counter with four operating modes and four selectable lengths
- Selectable and configurable clock source
- Three or seven configurable capture/compare registers
- Configurable outputs with PWM capability
- Double-buffered compare latches with synchronized loading
- Interrupt vector register for fast decoding of all Timer_B interrupts

The block diagram of Timer_B is shown in Figure 13-1.

---

Note

Use of the Word Count

*Count* is used throughout this chapter. It means the counter must be in the process of counting for the action to take place. If a particular value is directly written to the counter, then an associated action does not take place.

---

13.1.1 Similarities and Differences From Timer_A

Timer_B is identical to Timer_A with the following exceptions:
- The length of Timer_B is programmable to be 8, 10, 12, or 16 bits.
- Timer_B TBCCRx registers are double-buffered and can be grouped.
- All Timer_B outputs can be put into a high-impedance state.
- The SCCI bit function is not implemented in Timer_B.
Figure 13-1. Timer_B Block Diagram

A. INCLK is device-specific, often assigned to the inverted TBCLK, refer to device-specific data sheet.
13.2 Timer_B Operation

The Timer_B module is configured with user software. The setup and operation of Timer_B is discussed in the following sections.

13.2.1 16-Bit Timer Counter

The 16-bit timer/counter register, TBR, increments or decrements (depending on mode of operation) with each rising edge of the clock signal. TBR can be read or written with software. Additionally, the timer can generate an interrupt when it overflows.

TBR may be cleared by setting the TBCLR bit. Setting TBCLR also clears the clock divider and count direction for up/down mode.

---

**Note**

**Modifying Timer_B Registers**

It is recommended to stop the timer before modifying its operation (with exception of the interrupt enable, interrupt flag, and TBCLR) to avoid errant operating conditions.

When the timer clock is asynchronous to the CPU clock, any read from TBR should occur while the timer is not operating or the results may be unpredictable. Alternatively, the timer may be read multiple times while operating, and a majority vote taken in software to determine the correct reading. Any write to TBR will take effect immediately.

---

13.2.1.1 TBR Length

Timer_B is configurable to operate as an 8-, 10-, 12-, or 16-bit timer with the CNTLx bits. The maximum count value, TBR_{max}, for the selectable lengths is 0FFh, 03FFh, 0FFFh, and 0FFFFh, respectively. Data written to the TBR register in 8-, 10-, and 12-bit mode is right-justified with leading zeros.

13.2.1.2 Clock Source Select and Divider

The timer clock can be sourced from ACLK, SMCLK, or externally via TBCLK or INCLK (INCLK is device-specific, often assigned to the inverted TBCLK, refer to device-specific data sheet). The clock source is selected with the TBSSELx bits. The selected clock source may be passed directly to the timer or divided by 2, 4, or 8, using the IDx bits. The clock divider is reset when TBCLR is set.

13.2.2 Starting the Timer

The timer may be started or restarted in the following ways:

- The timer counts when MCx > 0 and the clock source is active.
- When the timer mode is either up or up/down, the timer may be stopped by loading 0 to TBCL0. The timer may then be restarted by loading a nonzero value to TBCL0. In this scenario, the timer starts incrementing in the up direction from zero.

13.2.3 Timer Mode Control

The timer has four modes of operation as described in Table 13-1: stop, up, continuous, and up/down. The operating mode is selected with the MCx bits.

<table>
<thead>
<tr>
<th>MCx</th>
<th>Mode</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>Stop</td>
<td>The timer is halted.</td>
</tr>
<tr>
<td>01</td>
<td>Up</td>
<td>The timer repeatedly counts from zero to the value of compare register TBCL0.</td>
</tr>
<tr>
<td>10</td>
<td>Continuous</td>
<td>The timer repeatedly counts from zero to the value selected by the CNTLx bits.</td>
</tr>
<tr>
<td>11</td>
<td>Up/down</td>
<td>The timer repeatedly counts from zero up to the value of TBCL0 and then back down to zero.</td>
</tr>
</tbody>
</table>
13.2.3.1 Up Mode

The up mode is used if the timer period must be different from $TBR_{(\text{max})}$ counts. The timer repeatedly counts up to the value of compare latch TBCL0, which defines the period, as shown in Figure 13-2. The number of timer counts in the period is TBCL0+1. When the timer value equals TBCL0 the timer restarts counting from zero. If up mode is selected when the timer value is greater than TBCL0, the timer immediately restarts counting from zero.

![Figure 13-2. Up Mode](image)

The TBCCR0 CCIFG interrupt flag is set when the timer counts to the TBCL0 value. The TBIFG interrupt flag is set when the timer counts from TBCL0 to zero. Figure 13-3 shows the flag set cycle.

![Figure 13-3. Up Mode Flag Setting](image)

13.2.3.2 Changing the Period Register TBCL0

When changing TBCL0 while the timer is running and when the TBCL0 load event is immediate, CLLD0 = 00, if the new period is greater than or equal to the old period, or greater than the current count value, the timer counts up to the new period. If the new period is less than the current count value, the timer rolls to zero. However, one additional count may occur before the counter rolls to zero.

13.2.3.3 Continuous Mode

In continuous mode the timer repeatedly counts up to $TBR_{(\text{max})}$ and restarts from zero as shown in Figure 13-4. The compare latch TBCL0 works the same way as the other capture/compare registers.

![Figure 13-4. Continuous Mode](image)

The TBIFG interrupt flag is set when the timer counts from $TBR_{(\text{max})}$ to zero. Figure 13-5 shows the flag set cycle.

![Figure 13-5. Continuous Mode Flag Setting](image)
13.2.3.4 *Use of the Continuous Mode*

The continuous mode can be used to generate independent time intervals and output frequencies. Each time an interval is completed, an interrupt is generated. The next time interval is added to the TBCLx latch in the interrupt service routine. Figure 13-6 shows two separate time intervals $t_0$ and $t_1$ being added to the capture/compare registers. The time interval is controlled by hardware, not software, without impact from interrupt latency. Up to three (Timer_B3) or 7 (Timer_B7) independent time intervals or output frequencies can be generated using capture/compare registers.

![Figure 13-5. Continuous Mode Flag Setting](image)

**Figure 13-5. Continuous Mode Flag Setting**

Time intervals can be produced with other modes as well, where TBCL0 is used as the period register. Their handling is more complex since the sum of the old TBCLx data and the new period can be higher than the TBCL0 value. When the sum of the previous TBCLx value plus $t_x$ is greater than the TBCL0 data, TBCL0 + 1 must be subtracted to obtain the correct time interval.

13.2.3.5 *Up/Down Mode*

The up/down mode is used if the timer period must be different from $TBR_{\text{max}}$ counts, and if a symmetrical pulse generation is needed. The timer repeatedly counts up to the value of compare latch TBCL0, and back down to zero, as shown in Figure 13-7. The period is twice the value in TBCL0.

![Figure 13-6. Continuous Mode Time Intervals](image)

**Figure 13-6. Continuous Mode Time Intervals**

**Note**

$\text{TBCL0} > TBR_{\text{max}}$

If TBCL0 > $TBR_{\text{max}}$, the counter operates as if it were configured for continuous mode. It does not count down from $TBR_{\text{max}}$ to zero.
The count direction is latched. This allows the timer to be stopped and then restarted in the same direction it was counting before it was stopped. If this is not desired, the TBCLR bit must be used to clear the direction. The TBCLR bit also clears the TBR value and the clock divider.

In up/down mode, the TBCCR0 CCIFG interrupt flag and the TBIFG interrupt flag are set only once during the period, separated by 1/2 the timer period. The TBCCR0 CCIFG interrupt flag is set when the timer counts from TBCL0-1 to TBCL0, and TBIFG is set when the timer completes counting down from 0001h to 0000h. Figure 13-8 shows the flag set cycle.

13.2.3.6 Changing the Value of Period Register TBCL0

When changing TBCL0 while the timer is running, and counting in the down direction, and when the TBCL0 load event is immediate, the timer continues its descent until it reaches zero. The value in TBCCR0 is latched into TBCL0 immediately; however, the new period takes effect after the counter counts down to zero.

If the timer is counting in the up direction when the new period is latched into TBCL0, and the new period is greater than or equal to the old period, or greater than the current count value, the timer counts up to the new period before counting down. When the timer is counting in the up direction, and the new period is less than the current count value when TBCL0 is loaded, the timer begins counting down. However, one additional count may occur before the counter begins counting down.

13.2.3.7 Use of the Up/Down Mode

The up/down mode supports applications that require dead times between output signals (see section Timer_B Output Unit). For example, to avoid overload conditions, two outputs driving an H-bridge must never be in a high state simultaneously. In the example shown in Figure 13-9 the $t_{\text{dead}}$ is:

$$t_{\text{dead}} = \text{t}_{\text{timer}} \times (\text{TBCL1} - \text{TBCL3})$$

Where,

- $t_{\text{dead}}$ = Time during which both outputs need to be inactive
- $t_{\text{timer}}$ = Cycle time of the timer clock
- TBCLx = Content of compare latch x
The ability to simultaneously load grouped compare latches assures the dead times.

13.2.4 Capture/Compare Blocks

Three or seven identical capture/compare blocks, TBCCRx, are present in Timer_B. Any of the blocks may be used to capture the timer data or to generate time intervals.

13.2.4.1 Capture Mode

The capture mode is selected when CAP = 1. Capture mode is used to record time events. It can be used for speed computations or time measurements. The capture inputs CCixa and CCixb are connected to external pins or internal signals and are selected with the CCISx bits. The CMx bits select the capture edge of the input signal as rising, falling, or both. A capture occurs on the selected edge of the input signal. If a capture is performed:

- The timer value is copied into the TBCCRx register
- The interrupt flag CCIFG is set

The input signal level can be read at any time via the CCI bit. MSP430x2xx family devices may have different signals connected to CCixa and CCixb. Refer to the device-specific data sheet for the connections of these signals.

The capture signal can be asynchronous to the timer clock and cause a race condition. Setting the SCS bit will synchronize the capture with the next timer clock. Setting the SCS bit to synchronize the capture signal with the timer clock is recommended. This is illustrated in Figure 13-10.

Figure 13-10. Capture Signal (SCS = 1)
Overflow logic is provided in each capture/compare register to indicate if a second capture was performed before the value from the first capture was read. Bit COV is set when this occurs as shown in Figure 13-11. COV must be reset with software.

![Capture Cycle Diagram]

**Figure 13-11. Capture Cycle**

### 13.2.4.1.1 Capture Initiated by Software

Captures can be initiated by software. The CMx bits can be set for capture on both edges. Software then sets bit CCIS1=1 and toggles bit CCIS0 to switch the capture signal between VCC and GND, initiating a capture each time CCIS0 changes state:

```assembly
MOV #CAP+SCS+CCIS1+CM_3, &TBCCTLx ; Setup TBCCTLx
XOR #CCIS0, &TBCCTLx ; TBCCTLx = TBR
```

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13.2.4.2 Compare Mode
The compare mode is selected when CAP = 0. Compare mode is used to generate PWM output signals or interrupts at specific time intervals. When TBR counts to the value in a TBCLx:

- Interrupt flag CCIFG is set
- Internal signal EQUx = 1
- EQUx affects the output according to the output mode

13.2.4.2.1 Compare Latch TBCLx
The TBCCRx compare latch, TBCLx, holds the data for the comparison to the timer value in compare mode. TBCLx is buffered by TBCCRx. The buffered compare latch gives the user control over when a compare period updates. The user cannot directly access TBCLx. Compare data is written to each TBCCRx and automatically transferred to TBCLx. The timing of the transfer from TBCCRx to TBCLx is user-selectable with the CLLDx bits as described in Table 13-2.

Table 13-2. TBCLx Load Events

<table>
<thead>
<tr>
<th>CLLDx</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>New data is transferred from TBCCRx to TBCLx immediately when TBCCRx is written to.</td>
</tr>
<tr>
<td>01</td>
<td>New data is transferred from TBCCRx to TBCLx when TBR counts to 0</td>
</tr>
<tr>
<td>10</td>
<td>New data is transferred from TBCCRx to TBCLx when TBR counts to 0 for up and continuous modes. New data is transferred to from TBCCRx to TBCLx when TBR counts to the old TBCL0 value or to 0 for up/down mode</td>
</tr>
<tr>
<td>11</td>
<td>New data is transferred from TBCCRx to TBCLx when TBR counts to the old TBCLx value.</td>
</tr>
</tbody>
</table>

13.2.4.2.2 Grouping Compare Latches
Multiple compare latches may be grouped together for simultaneous updates with the TBCLGRPx bits. When using groups, the CLLDx bits of the lowest numbered TBCCRx in the group determine the load event for each compare latch of the group, except when TBCLGRP = 3, as shown in Table 13-3. The CLLDx bits of the controlling TBCCRx must not be set to zero. When the CLLDx bits of the controlling TBCCRx are set to zero, all compare latches update immediately when their corresponding TBCCRx is written; no compare latches are grouped.

Two conditions must exist for the compare latches to be loaded when grouped. First, all TBCCRx registers of the group must be updated, even when new TBCCRx data = old TBCCRx data. Second, the load event must occur.

Table 13-3. Compare Latch Operating Modes

<table>
<thead>
<tr>
<th>TBCLGRPx</th>
<th>Grouping</th>
<th>Update Control</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>None</td>
<td>Individual</td>
</tr>
</tbody>
</table>
| 01       | TBL1+TBCL2  
           TBL3+TBCL4  
           TBL5+TBCL6 | TBCCR1  
           TBCCR3  
           TBCCR5 |
| 10       | TBL1+TBCL2+TBCL3  
           TBL4+TBCL5+TBCL6 | TBCCR1  
           TBCCR4 |
| 11       | TBL0+TBL1+TBL2+TBL3+TBL4+TBL5+TBL6 | TBCCR1 |
13.2.5 Output Unit

Each capture/compare block contains an output unit. The output unit is used to generate output signals such as PWM signals. Each output unit has eight operating modes that generate signals based on the EQU0 and EQUx signals. The TBOUTH pin function can be used to put all Timer_B outputs into a high-impedance state. When the TBOUTH pin function is selected for the pin, and when the pin is pulled high, all Timer_B outputs are in a high-impedance state.

13.2.5.1 Output Modes

The output modes are defined by the OUTMODx bits and are described in Table 13-4. The OUTx signal is changed with the rising edge of the timer clock for all modes except mode 0. Output modes 2, 3, 6, and 7 are not useful for output unit 0 because EQUx = EQU0.

Table 13-4. Output Modes

<table>
<thead>
<tr>
<th>OUTMODx</th>
<th>Mode</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>000</td>
<td>Output</td>
<td>The output signal OUTx is defined by the OUTx bit. The OUTx signal updates immediately when OUTx is updated.</td>
</tr>
<tr>
<td>001</td>
<td>Set</td>
<td>The output is set when the timer counts to the TBCLx value. It remains set until a reset of the timer, or until another output mode is selected and affects the output.</td>
</tr>
<tr>
<td>010</td>
<td>Toggle/Reset</td>
<td>The output is toggled when the timer counts to the TBCLx value. It is reset when the timer counts to the TBCL0 value.</td>
</tr>
<tr>
<td>011</td>
<td>Set/Reset</td>
<td>The output is set when the timer counts to the TBCLx value. It is reset when the timer counts to the TBCL0 value.</td>
</tr>
<tr>
<td>100</td>
<td>Toggle</td>
<td>The output is toggled when the timer counts to the TBCLx value. The output period is double the timer period.</td>
</tr>
<tr>
<td>101</td>
<td>Reset</td>
<td>The output is reset when the timer counts to the TBCLx value. It remains reset until another output mode is selected and affects the output.</td>
</tr>
<tr>
<td>110</td>
<td>Toggle/Set</td>
<td>The output is toggled when the timer counts to the TBCLx value. It is set when the timer counts to the TBCL0 value.</td>
</tr>
<tr>
<td>111</td>
<td>Reset/Set</td>
<td>The output is reset when the timer counts to the TBCLx value. It is set when the timer counts to the TBCL0 value.</td>
</tr>
</tbody>
</table>
13.2.5.1.1 Output Example, Timer in Up Mode

The OUTx signal is changed when the timer counts up to the TBCLx value, and rolls from TBCL0 to zero, depending on the output mode. An example is shown in Figure 13-12 using TBCL0 and TBCL1.

![Diagram showing Output Example, Timer in Up Mode](https://www.ti.com)
13.2.5.1.2 Output Example, Timer in Continuous Mode

The OUTx signal is changed when the timer reaches the TBCLx and TBCL0 values, depending on the output mode. An example is shown in Figure 13-13 using TBCL0 and TBCL1.

**Figure 13-13. Output Example, Timer in Continuous Mode**
13.2.5.1.3 Output Example, Timer in Up/Down Mode

The OUTx signal changes when the timer equals TBCLx in either count direction and when the timer equals TBCL0, depending on the output mode. An example is shown in Figure 13-14 using TBCL0 and TBCL3.

![Diagram showing output example, timer in up/down mode]

**Figure 13-14. Output Example, Timer in Up/Down Mode**

**Note**

Switching Between Output Modes

When switching between output modes, one of the OUTMODx bits should remain set during the transition, unless switching to mode 0. Otherwise, output glitching can occur because a NOR gate decodes output mode 0. A safe method for switching between output modes is to use output mode 7 as a transition state:

```
BIS  #OUTMOD_7, &TBCCTLx  ; Set output mode=7
BIC  #OUTMODx, &TBCCTLx  ; Clear unwanted bits
```
13.2.6 Timer_B Interrupts

Two interrupt vectors are associated with the 16-bit Timer_B module:

- TBCCR0 interrupt vector for TBCCR0 CCIFG
- TBIV interrupt vector for all other CCIFG flags and TBIFG

In capture mode, any CCIFG flag is set when a timer value is captured in the associated TBCCRx register. In compare mode, any CCIFG flag is set when TBR counts to the associated TBCLx value. Software may also set or clear any CCIFG flag. All CCIFG flags request an interrupt when their corresponding CCIE bit and the GIE bit are set.

13.2.6.1 TBCCR0 Interrupt Vector

The TBCCR0 CCIFG flag has the highest Timer_B interrupt priority and has a dedicated interrupt vector as shown in Figure 13-15. The TBCCR0 CCIFG flag is automatically reset when the TBCCR0 interrupt request is serviced.

![Figure 13-15. Capture/Compare TBCCR0 Interrupt Flag](image)

13.2.6.2 TBIV, Interrupt Vector Generator

The TBIFG flag and TBCCRx CCIFG flags (excluding TBCCR0 CCIFG) are prioritized and combined to source a single interrupt vector. The interrupt vector register TBIV is used to determine which flag requested an interrupt.

The highest priority enabled interrupt (excluding TBCCR0 CCIFG) generates a number in the TBIV register (see register description). This number can be evaluated or added to the program counter to automatically enter the appropriate software routine. Disabled Timer_B interrupts do not affect the TBIV value.

Any access, read or write, of the TBIV register automatically resets the highest pending interrupt flag. If another interrupt flag is set, another interrupt is immediately generated after servicing the initial interrupt. For example, if the TBCCR1 and TBCCR2 CCIFG flags are set when the interrupt service routine accesses the TBIV register, TBCCR1 CCIFG is reset automatically. After the RETI instruction of the interrupt service routine is executed, the TBCCR2 CCIFG flag will generate another interrupt.

13.2.6.3 TBIV, Interrupt Handler Examples

The following software example shows the recommended use of TBIV and the handling overhead. The TBIV value is added to the PC to automatically jump to the appropriate routine.

The numbers at the right margin show the necessary CPU clock cycles for each instruction. The software overhead for different interrupt sources includes interrupt latency and return-from-interrupt cycles, but not the task handling itself. The latencies are:

- Capture/compare block CCR0: 11 cycles
- Capture/compare blocks CCR1 to CCR6: 16 cycles
- Timer overflow TBIFG: 14 cycles

Example 13-1 shows the recommended use of TBIV for Timer_B3.
Example 13-1. Recommended Use of TBIV

; Interrupt handler for TBCCR0 CCIFG.
CCIFG_0_HND ... ; Start of handler Interrupt latency 6
RETI 5

; Interrupt handler for TBIFG, TBCCR1 and TBCCR2 CCIFG.
TB_HND ... ; Interrupt latency 6
ADD &TBIV, PC ; Add offset to Jump table 3
JMP CCIFG_1_HND ; Vector 2: Module 1 2
JMP CCIFG_2_HND ; Vector 4: Module 2 2
RETI ; Vector 6
RETI ; Vector 8
RETI ; Vector 10
RETI ; Vector 12
TBIFG_HND ; Vector 14: TIMOV Flag 5
... ; Task starts here
RETI 5
CCIFG_2_HND ; Vector 4: Module 2 5
... ; Task starts here
RETI ; Back to main program

; The Module 1 handler shows a way to look if any other interrupt is pending
; 9 cycles may be saved if another interrupt is pending
CCIFG_1_HND ; Vector 6: Module 3 5
... ; Task starts here
JMP TB_HND ; Look for pending ints 2

; Task starts here
RETI 5
CCIFG_2_HND ; Vector 4: Module 2 5
... ; Task starts here
RETI ; Back to main program
### 13.3 Timer_B Registers

Table 13-5 lists the memory-mapped registers for the Timer_B.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>180h</td>
<td>TBCTL</td>
<td>Timer_B control</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.1</td>
</tr>
<tr>
<td>190h</td>
<td>TBR</td>
<td>Timer_B counter</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.2</td>
</tr>
<tr>
<td>182h</td>
<td>TBCCCTL0</td>
<td>Timer_B capture/compare control 0</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.3</td>
</tr>
<tr>
<td>192h</td>
<td>TBCCR0</td>
<td>Timer_B capture/compare 0</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.4</td>
</tr>
<tr>
<td>184h</td>
<td>TBCCCTL1</td>
<td>Timer_B capture/compare control 1</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.3</td>
</tr>
<tr>
<td>194h</td>
<td>TBCCR1</td>
<td>Timer_B capture/compare 1</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.4</td>
</tr>
<tr>
<td>186h</td>
<td>TBCCCTL2</td>
<td>Timer_B capture/compare control 2</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.3</td>
</tr>
<tr>
<td>196h</td>
<td>TBCCR2</td>
<td>Timer_B capture/compare 2</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.4</td>
</tr>
<tr>
<td>188h</td>
<td>TBCCCTL3</td>
<td>Timer_B capture/compare control 3</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.3</td>
</tr>
<tr>
<td>198h</td>
<td>TBCCR3</td>
<td>Timer_B capture/compare 3</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.4</td>
</tr>
<tr>
<td>18Ah</td>
<td>TBCCCTL4</td>
<td>Timer_B capture/compare control 4</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.3</td>
</tr>
<tr>
<td>19Ah</td>
<td>TBCCR4</td>
<td>Timer_B capture/compare 4</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.4</td>
</tr>
<tr>
<td>18Ch</td>
<td>TBCCCTL5</td>
<td>Timer_B capture/compare control 5</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.3</td>
</tr>
<tr>
<td>19Ch</td>
<td>TBCCR5</td>
<td>Timer_B capture/compare 5</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.4</td>
</tr>
<tr>
<td>18Eh</td>
<td>TBCCCTL6</td>
<td>Timer_B capture/compare control 6</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.3</td>
</tr>
<tr>
<td>19Eh</td>
<td>TBCCR6</td>
<td>Timer_B capture/compare 6</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 13.3.4</td>
</tr>
<tr>
<td>11Eh</td>
<td>TBIV</td>
<td>Timer_B interrupt vector</td>
<td>Read</td>
<td>00h with POR</td>
<td>Section 13.3.5</td>
</tr>
</tbody>
</table>
13.3.1 TBCTL Register

Timer_B Control Register

TBCTL is shown in Figure 13-16 and described in Table 13-6.

Return to Table 13-5.

Figure 13-16. TBCTL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td></td>
</tr>
<tr>
<td>14-13</td>
<td>TBCLGRP</td>
<td>R/W</td>
<td>0h</td>
<td>TBCLx group&lt;br&gt;00b = Each TBCLx latch loads independently&lt;br&gt;01b = TBCL1+TBCL2 (TBCCR1 CLLDx bits control the update)&lt;br&gt;TBCL3+TBCL4 (TBCCR3 CLLDx bits control the update)&lt;br&gt;TBCL5+TBCL6 (TBCCR5 CLLDx bits control the update)&lt;br&gt;TBCL0 independent&lt;br&gt;10b = TBCL1+TBCL2+TBCL3 (TBCCR1 CLLDx bits control the update)&lt;br&gt;TBCL4+TBCL5+TBCL6 (TBCCR4 CLLDx bits control the update)&lt;br&gt;TBCL0 independent&lt;br&gt;11b = TBCL0+TBCL1+TBCL2+TBCL3+TBCL4+TBCL5+TBCL6 (TBCCR1 CLLDx bits control the update)</td>
</tr>
<tr>
<td>12-11</td>
<td>CNTLx</td>
<td>R/W</td>
<td>0h</td>
<td>Counter length&lt;br&gt;00b = 16 bit, TBR(max) = 0FFFFh&lt;br&gt;01b = 12 bit, TBR(max) = 0FFFh&lt;br&gt;10b = 10 bit, TBR(max) = 03FFh&lt;br&gt;11b = 8 bit, TBR(max) = 0FFh</td>
</tr>
<tr>
<td>10</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td></td>
</tr>
<tr>
<td>9-8</td>
<td>TBSSELx</td>
<td>R/W</td>
<td>0h</td>
<td>Timer_B clock source select.&lt;br&gt;00b = TBCLK&lt;br&gt;01b = ACLK&lt;br&gt;10b = SMCLK&lt;br&gt;11b = INCLK (INCLK is device-specific and is often assigned to the inverted TBCLK) (see the device-specific data sheet)</td>
</tr>
<tr>
<td>7-6</td>
<td>IDx</td>
<td>R/W</td>
<td>0h</td>
<td>Input divider. These bits select the divider for the input clock.&lt;br&gt;00b = /1&lt;br&gt;01b = /2&lt;br&gt;10b = /4&lt;br&gt;11b = /8</td>
</tr>
<tr>
<td>Bit</td>
<td>Field</td>
<td>Type</td>
<td>Reset</td>
<td>Description</td>
</tr>
<tr>
<td>-----</td>
<td>-------</td>
<td>------</td>
<td>-------</td>
<td>-------------</td>
</tr>
<tr>
<td>5-4</td>
<td>MCx</td>
<td>R/W</td>
<td>0h</td>
<td>Mode control. Set MCx = 00b when Timer_B is not in use to conserve power. 00b = Stop mode: the timer is halted 01b = Up mode: the timer counts up to TBCL0 10b = Continuous mode: the timer counts up to the value set by CNTLx 11b = Up/down mode: the timer counts up to TBCL0 and down to 0000h</td>
</tr>
<tr>
<td>3</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td>Timer_B clear. Setting this bit resets TBR, the clock divider, and the count direction. The TBCLR bit is automatically reset and is always read as zero.</td>
</tr>
<tr>
<td>2</td>
<td>TBCLR</td>
<td>W</td>
<td>0h</td>
<td>Timer_B interrupt enable. This bit enables the TBIFG interrupt request. 0b = Interrupt disabled 1b = Interrupt enabled</td>
</tr>
<tr>
<td>1</td>
<td>TBIE</td>
<td>R/W</td>
<td>0h</td>
<td>Timer_B interrupt flag. 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
<tr>
<td>0</td>
<td>TBIFG</td>
<td>R/W</td>
<td>0h</td>
<td>Timer_B interrupt flag. 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
</tbody>
</table>

### 13.3.2 TBR Register

**Timer_B Counter Register**

TBR is shown in **Figure 13-17** and described in **Table 13-7**.

Return to **Table 13-5**.

#### Figure 13-17. TBR Register

<table>
<thead>
<tr>
<th>15</th>
<th>14</th>
<th>13</th>
<th>12</th>
<th>11</th>
<th>10</th>
<th>9</th>
<th>8</th>
</tr>
</thead>
<tbody>
<tr>
<td>TBRx</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
</tr>
<tr>
<td>7</td>
<td>6</td>
<td>5</td>
<td>4</td>
<td>3</td>
<td>2</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>TBRx</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
</tr>
</tbody>
</table>

#### Table 13-7. TBR Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>TBRx</td>
<td>R/W</td>
<td>0h</td>
<td>Timer_B register. The TBR register is the count of Timer_B.</td>
</tr>
</tbody>
</table>
13.3.3 TBCCTLx Register

Timer_B Capture/Compare Control x Register

TBCCTLx is shown in Figure 13-18 and described in Table 13-8.

Return to Table 13-5.

Figure 13-18. TBCCTLx Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
</table>
| 15-14| CMx     | R/W  | 0h    | Capture mode  
|      |         |      |       | 0b = No capture  
|      |         |      |       | 01b = Capture on rising edge  
|      |         |      |       | 10b = Capture on falling edge  
|      |         |      |       | 11b = Capture on both rising and falling edges |
| 13-12| CCISx   | R/W  | 0h    | Capture compare input select. These bits select the TBCCRx  
|      |         |      |       | input signal. See the device-specific data sheet for specific signal  
|      |         |      |       | connections.  
|      |         |      |       | 00b = CCIxA  
|      |         |      |       | 01b = CCIxB  
|      |         |      |       | 10b = GND  
|      |         |      |       | 11b = VCC |
| 11   | SCS     | R/W  | 0h    | Synchronize capture source. This bit is used to synchronize the  
|      |         |      |       | capture input signal with the timer clock.  
|      |         |      |       | 0b = Asynchronous capture  
|      |         |      |       | 1b = Synchronous capture |
| 10-9 | CLLDx   | R/W  | 0h    | Compare latch load. These bits select the compare latch load event.  
|      |         |      |       | 00b = TBCLx loads on write to TBCCRx  
|      |         |      |       | 01b = TBCLx loads when TBR counts to 0  
|      |         |      |       | 10b = TBCLx loads when TBR counts to 0 (up or continuous mode)  
|      |         |      |       | TBCLx loads when TBR counts to TBCL0 or to 0 (up/down mode)  
|      |         |      |       | 11b = TBCLx loads when TBR counts to TBCLx |
| 8    | CAP     | R/W  | 0h    | Capture mode  
|      |         |      |       | 0b = Compare mode  
|      |         |      |       | 1b = Capture mode |
| 7-5  | OUTMODx | R/W  | 0h    | Output mode. Modes 2, 3, 6, and 7 are not useful for TBCL0  
|      |         |      |       | because EQUx = EQU0.  
|      |         |      |       | 000b = OUT bit value  
|      |         |      |       | 001b = Set  
|      |         |      |       | 010b = Toggle/reset  
|      |         |      |       | 011b = Set/reset  
|      |         |      |       | 100b = Toggle  
|      |         |      |       | 101b = Reset  
|      |         |      |       | 110b = Toggle/set  
<p>|      |         |      |       | 111b = Reset/set |</p>
<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
</table>
| 4   | CCIE  | R/W  | 0h    | Capture/compare interrupt enable. This bit enables the interrupt request of the corresponding CCIFG flag.  
0b = Interrupt disabled  
1b = Interrupt enabled |
| 3   | CCI   | R    | 0h    | Capture/compare input. The selected input signal can be read by this bit. |
| 2   | OUT   | R/W  | 0h    | Output. For output mode 0, this bit directly controls the state of the output.  
0b = Output low  
1b = Output high |
| 1   | COV   | R/W  | 0h    | Capture overflow. This bit indicates a capture overflow occurred.  
COV must be reset with software.  
0b = No capture overflow occurred  
1b = Capture overflow occurred |
| 0   | CCIFG | R/W  | 0h    | Capture/compare interrupt flag  
0b = No interrupt pending  
1b = Interrupt pending |

### 13.3.4 TBCCRx Register

Timer_B Capture/Compare x Register

TBCCRx is shown in Figure 13-19 and described in Table 13-9.

Return to Table 13-5.

#### Figure 13-19. TBCCRx Register

<table>
<thead>
<tr>
<th>15</th>
<th>14</th>
<th>13</th>
<th>12</th>
<th>11</th>
<th>10</th>
<th>9</th>
<th>8</th>
</tr>
</thead>
<tbody>
<tr>
<td>TBCCRx</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
</tr>
<tr>
<td>TBCCRx</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
</tr>
</tbody>
</table>

#### Table 13-9. TBCCRx Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
</table>
| 15-0 | TBCCRx | R/W  | 0h    | Timer_B capture/compare register.  
Compare mode: Compare data is written to each TBCCRx and automatically transferred to TBCLx. TBCLx holds the data for the comparison to the timer value in the Timer_B Register, TBR.  
Capture mode: The Timer_B Register, TBR, is copied into the TBCCRx register when a capture is performed. |
13.3.5 TBIV Register

TBIV is shown in Figure 13-20 and described in Table 13-10.

Return to Table 13-5.

Figure 13-20. TBIV Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>TBIVx</td>
<td>R</td>
<td>0h</td>
<td>Timer_B interrupt vector value. See Table 13-11 for values.</td>
</tr>
</tbody>
</table>

Table 13-10. TBIV Register Field Descriptions

Table 13-11. Timer_B Interrupt Vectors

<table>
<thead>
<tr>
<th>TBIV Contents</th>
<th>Interrupt Source</th>
<th>Interrupt Flag</th>
<th>Interrupt Priority</th>
</tr>
</thead>
<tbody>
<tr>
<td>00h</td>
<td>No interrupt pending</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>02h</td>
<td>Capture/compare 1</td>
<td>TBCCR1 CCIFG</td>
<td>Highest</td>
</tr>
<tr>
<td>04h</td>
<td>Capture/compare 2</td>
<td>TBCCR2 CCIFG</td>
<td></td>
</tr>
<tr>
<td>06h</td>
<td>Capture/compare 3(1)</td>
<td>TBCCR3 CCIFG</td>
<td></td>
</tr>
<tr>
<td>08h</td>
<td>Capture/compare 4(1)</td>
<td>TBCCR4 CCIFG</td>
<td></td>
</tr>
<tr>
<td>0Ah</td>
<td>Capture/compare 5(1)</td>
<td>TBCCR5 CCIFG</td>
<td></td>
</tr>
<tr>
<td>0Ch</td>
<td>Capture/compare 6(1)</td>
<td>TBCCR6 CCIFG</td>
<td></td>
</tr>
<tr>
<td>0Eh</td>
<td>Timer overflow</td>
<td>TBIFG</td>
<td>Lowest</td>
</tr>
</tbody>
</table>

(1) Not available on all devices.
The Universal Serial Interface (USI) module provides SPI and I²C serial communication with one hardware module. This chapter discusses both modes.

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14.2 USI Operation.................................................................421
14.3 USI Registers.................................................................427
14.1 USI Introduction

The USI module provides the basic functionality to support synchronous serial communication. In its simplest form, it is an 8- or 16-bit shift register that can be used to output data streams, or when combined with minimal software, can implement serial communication. In addition, the USI includes built-in hardware functionality to ease the implementation of SPI and \( \text{I}^2\text{C} \) communication. The USI module also includes interrupts to further reduce the necessary software overhead for serial communication and to maintain the ultra-low-power capabilities of the MSP430.

The USI module features include:

- Three-wire SPI mode support
- \( \text{I}^2\text{C} \) mode support
- Variable data length
- Slave operation in LPM4; no internal clock required
- Selectable MSB or LSB data order
- START and STOP detection for \( \text{I}^2\text{C} \) mode with automatic SCL control
- Arbitration lost detection in master mode
- Programmable clock generation
- Selectable clock polarity and phase control

Figure 14-1 shows the USI module in SPI mode. Figure 14-2 shows the USI module in \( \text{I}^2\text{C} \) mode.
Figure 14-1. USI Block Diagram: SPI Mode
Universal Serial Interface (USI)

Figure 14-2. USI Block Diagram: I²C Mode

Set USIAL, Clear USIOE

Set USISTTIFG

Set USISTP

Set USISTAICFG

Set USIOE

Figure 14-2. USI Block Diagram: I²C Mode
14.2 USI Operation

The USI module is a shift register and bit counter that includes logic to support SPI and \( \text{I}^2\text{C} \) communication. The USI shift register (USISR) is directly accessible by software and contains the data to be transmitted or the data that has been received.

The bit counter counts the number of sampled bits and sets the USI interrupt flag USIIFG when the USICNTx value becomes zero, either by decrementing or by directly writing zero to the USICNTx bits. Writing USICNTx with a value > 0 automatically clears USIIFG when USIIFGCC = 0, otherwise USIIFG is not affected. The USICNTx bits stop decrementing when they become 0. They will not underflow to 0FFh.

Both the counter and the shift register are driven by the same shift clock. On a rising shift clock edge, USICNTx decrements and USISR samples the next bit input. The latch connected to the output of the shift register delays the change of the output to the falling edge of shift clock. It can be made transparent by setting the USIGE bit. This setting outputs the MSB or LSB of USISR to the SDO pin, depending on the USILSB bit.

14.2.1 USI Initialization

While the USI software reset bit (USISWRST) is set, the flags USIIFG, USISTTIFG, USISTP, and USIAL are held in their reset state. USISR and USICNTx are not clocked, and their contents are not affected. In \( \text{I}^2\text{C} \) mode, the SCL line is also released to the idle state by the USI hardware.

**Note**

If USIIE = 1 when the USI module is in software reset mode (USISWRST = 1), repeated USI counter interrupts can occur, because the default value of USIIFG is 1.

To avoid these repeated interrupts, disable the USI counter interrupt (USIIE = 0) before placing the USI module into software reset mode (USISWRST = 1).

When exiting software reset mode, do not set USIIE until after USISWRST is cleared.

To activate USI port functionality, the corresponding USIPEx bits in the USI control register must be set to select the USI function for the pin and maintain the PxIN and PxIFG functions for the pin. With this feature, the port input levels can be read by software in the PxIN register, and the incoming data stream can generate port interrupts on data transitions. This is useful, for example, to generate a port interrupt on a START edge.

14.2.2 USI Clock Generation

The USI clock generator contains a clock selection multiplexer, a divider, and the ability to select the clock polarity as shown in the block diagrams Figure 14-1 and Figure 14-2.

The clock source can be selected from the internal clocks ACLK or SMCLK, from an external clock SCLK, as well as from the capture/compare outputs of Timer_A. In addition, it is possible to clock the module by software using the USISWCLK bit when USISSELx = 100.

The USIDIVx bits can be used to divide the selected clock by a power of 2 up to 128. The generated clock, USICLK, is stopped when USIIFG = 1 or when the module operates in slave mode.

The USICKPL bit is used to select the polarity of USICLK. When USICKPL = 0, the inactive level of USICLK is low. When USICKPL = 1 the inactive level of USICLK is high.
14.2.3 SPI Mode

The USI module is configured in SPI mode when USI2C = 0. Control bit USICKPL selects the inactive level of the SPI clock while USICKPH selects the clock edge on which SDO is updated and SDI is sampled. Figure 14-3 shows the clock/data relationship for an 8-bit, MSB-first transfer. USIPE5, USIPE6, and USIPE7 must be set to enable the SCLK, SDO, and SDI port functions.

![Figure 14-3. SPI Timing](image)

### 14.2.3.1 SPI Master Mode

The USI module is configured as SPI master by setting the master bit USIMST and clearing the I²C bit USI2C. Since the master provides the clock to the slave(s) an appropriate clock source needs to be selected and SCLK configured as output. When USIPE5 = 1, SCLK is automatically configured as an output.

When USIIFG = 0 and USICNTx > 0, clock generation is enabled and the master will begin clocking in/out data using USISR.

Received data must be read from the shift register before new data is written into it for transmission. In a typical application, the USI software will read received data from USISR, write new data to be transmitted to USISR, and enable the module for the next transfer by writing the number of bits to be transferred to USICNTx.

### 14.2.3.2 SPI Slave Mode

The USI module is configured as SPI slave by clearing the USIMST and the USI2C bits. In this mode, when USIPE5 = 1 SCLK is automatically configured as an input and the USI receives the clock externally from the master.

If the USI is to transmit data, the shift register must be loaded with the data before the master provides the first clock edge. The output must be enabled by setting USIOE. When USICKPH = 1, the MSB will be visible on SDO immediately after loading the shift register.

The SDO pin can be disabled by clearing the USIOE bit. This is useful if the slave is not addressed in an environment with multiple slaves on the bus.

Once all bits are received, the data must be read from USISR and new data loaded into USISR before the next clock edge from the master. In a typical application, after receiving data, the USI software will read the USISR register, write new data to USISR to be transmitted, and enable the USI module for the next transfer by writing the number of bits to be transferred to USICNTx.
14.2.3.3 USISR Operation

The 16-bit USISR is made up of two 8-bit registers, USISRL and USISRH. Control bit USI16B selects the number of bits of USISR that are used for data transmit and receive. When USI16B = 0, only the lower 8 bits, USISRL, are used.

To transfer < 8 bits, the data must be loaded into USISRL such that unused bits are not shifted out. The data must be MSB- or LSB-aligned depending on USILSB. Figure 14-4 shows an example of 7-bit data handling.

![Figure 14-4. Data Adjustments for 7-Bit SPI Data](image)

When USI16B = 1, all 16 bits are used for data handling. When using USISR to access both USISRL and USISRH, the data needs to be properly adjusted when < 16 bits are used in the same manner as shown in Figure 14-4.

14.2.3.4 SPI Interrupts

There is one interrupt vector associated with the USI module, and one interrupt flag, USIIFG, relevant for SPI operation. When USIIE and the GIE bit are set, the interrupt flag will generate an interrupt request.

USIIFG is set when USICNTx becomes zero, either by counting or by directly writing 0 to the USICNTx bits. USIIFG is cleared by writing a value > 0 to the USICNTx bits when USIFGCC = 0, or directly by software.
14.2.4 I²C Mode

The USI module is configured in I²C mode when USII2C = 1, USICKPL = 1, and USICKPH = 0. For I²C data compatibility, USILSB and USI16B must be cleared. USIPE6 and USIPE7 must be set to enable the SCL and SDA port functions.

For examples of using the USI in I²C mode, refer to Using the USI I²C Code Library and the device-specific code examples in TI Resource Explorer → MSP430Ware.

14.2.4.1 I²C Master Mode

To configure the USI module as an I²C master the USIMST bit must be set. In master mode, clocks are generated by the USI module and output to the SCL line while USIIFG = 0. When USIIFG = 1, the SCL will stop at the idle, or high, level. Multi-master operation is supported as described in Section 14.2.4.8.

The master supports slaves that are holding the SCL line low only when USIDIVx > 0. When USIDIVx is set to /1 clock division (USIDIVx = 0), connected slaves must not hold the SCL line low during data transmission. Otherwise the communication may fail.

14.2.4.2 I²C Slave Mode

To configure the USI module as an I²C slave the USIMST bit must be cleared. In slave mode, SCL is held low if USIIFG = 1, USISTTIFG = 1 or if USICNTx = 0. USISTTIFG must be cleared by software after the slave is setup and ready to receive the slave address from a master.

14.2.4.3 I²C Transmitter

In transmitter mode, data is first loaded into USISRL. The output is enabled by setting USIOE and the transmission is started by writing 8 into USICNTx. This clears USIIFG and SCL is generated in master mode or released from being held low in slave mode. After the transmission of all 8 bits, USIIFG is set, and the clock signal on SCL is stopped in master mode or held low at the next low phase in slave mode.

To receive the I²C acknowledgment bit, the USIOE bit is cleared with software and USICNTx is loaded with 1. This clears USIIFG and one bit is received into USISRL. When USIIFG becomes set again, the LSB of USISRL is the received acknowledge bit and can be tested in software.

```assembly
; Receive ACK/NACK
BIC.B #USIOE, &USICTL0     ; SDA input
MOV.B #01h, &USICNT        ; USICNTx = 1
TEST_USIIFG
BIT.B #USIIFG, &USICTL1    ; Test USIIFG
JS TEST_USIIFG
BIT.B #01h, &USISRL        ; Test received ACK bit
JNZ HANDLE_NACK            ; Handle if NACK
...Else, handle ACK
```
14.2.4.4 I²C Receiver

In I²C receiver mode, the output must be disabled by clearing USIOE and the USI module is prepared for reception by writing 8 into USICNTx. This clears USIIFG and SCL is generated in master mode or released from being held low in slave mode. The USIIFG bit will be set after 8 clocks. This stops the clock signal on SCL in master mode or holds SCL low at the next low phase in slave mode.

To transmit an acknowledge or no-acknowledge bit, the MSB of the shift register is loaded with 0 or 1, the USIOE bit is set with software to enable the output, and 1 is written to the USICNTx bits. As soon as the MSB bit is shifted out, USIIFG will be become set and the module can be prepared for the reception of the next I²C data byte.

```asm
; Generate ACK
BIS.B USIOE, USICNT0
MOV.B 00h, USISRL  ; MSB = 0
MOV.B 01h, USICNT  ; USICNTx = 1
TEST USIIFG
BIT.B USIIFG, USICNT1  ; Test USIIFG
JZ TEST USIIFG
...continue...

; Generate NACK
BIS.B USIOE, USICNT0
MOV.B 0FFh, USISRL  ; MSB = 1
MOV.B 01h, USICNT  ; USICNTx = 1
TEST USIIFG
BIT.B USIIFG, USICNT1  ; Test USIIFG
JZ TEST USIIFG
...continue...
```

14.2.4.5 START Condition

A START condition is a high-to-low transition on SDA while SCL is high. The START condition can be generated by setting the MSB of the shift register to 0. Setting the USIGE and USIOE bits makes the output latch transparent and the MSB of the shift register is immediately presented to SDA and pulls the line low. Clearing USIGE resumes the clocked-latch function and holds the 0 on SDA until data is shifted out with SCL.

```asm
; Generate START
MOV.B 000h, USISRL  ; MSB = 0
BIS.B USIGE+USIOE, USICNT0  ; Latch/SDA output enabled
BIC.B USIGE, USICNT0  ; Latch disabled
...continue...
```

14.2.4.6 STOP Condition

A STOP condition is a low-to-high transition on SDA while SCL is high. To finish the acknowledgment bit and pull SDA low to prepare the STOP condition generation requires clearing the MSB in the shift register and loading 1 into USICNTx. This will generate a low pulse on SCL and during the low phase SDA is pulled low. SCL stops in the idle, or high, state since the module is in master mode. To generate the low-to-high transition, the MSB is set in the shift register and USICNTx is loaded with 1. Setting the USIGE and USIOE bits makes the output latch transparent and the MSB of USISRL releases SDA to the idle state. Clearing USIGE stores the MSB in the output latch and the output is disabled by clearing USIOE. SDA remains high until a START condition is generated because of the external pullup.

```asm
; Generate STOP
BIS.B USIOE, USICNT0  ; SDA output
MOV.B 000h, USISRL  ; MSB = 0
MOV.B 01h, USICNT  ; USICNT = 1 for one clock
TEST USIIFG
BIT.B USIIFG, USICNT1  ; Test USIIFG
JZ test USIIFG
MOV.B 0FFh, USISRL  ; USISRL = 1 to drive SDA high
BIC.B USIGE, USICNT0  ; Transparent latch enabled
BIC.B USIOE+USIOE, USICNT0  ; Latch/SDA output disabled
...continue...
```
14.2.4.7 Releasing SCL

Setting the USISCLREL bit will release SCL if it is being held low by the USI module without requiring USIIFG to be cleared. The USISCLREL bit will be cleared automatically if a START condition is received and the SCL line will be held low on the next clock.

In slave operation this bit should be used to prevent SCL from being held low when the slave has detected that it was not addressed by the master. On the next START condition USISCLREL will be cleared and the USISTTIFG will be set.

14.2.4.8 Arbitration

The USI module can detect a lost arbitration condition in multi-master I²C systems. The I²C arbitration procedure uses the data presented on SDA by the competing transmitters. The first master transmitter that generates a logic high loses arbitration to the opposing master generating a logic low. The loss of arbitration is detected in the USI module by comparing the value presented to the bus and the value read from the bus. If the values are not equal arbitration is lost and the arbitration lost flag, USIAL, is set. This also clears the output enable bit USIOE and the USI module no longer drives the bus. In this case, user software must check the USIAL flag together with USIIFG and configure the USI to slave receiver when arbitration is lost. The USIAL flag must be cleared by software.

To prevent other faster masters from generating clocks during the arbitration procedure SCL is held low if another master on the bus drives SCL low and USIIFG or USISTTIFG is set, or if USICNTx = 0.

14.2.4.9 I²C Interrupts

There is one interrupt vector associated with the USI module with two interrupt flags relevant for I²C operation, USIIFG and USISTTIFG. Each interrupt flag has its own interrupt enable bit, USIIE and USISTTIE. When an interrupt is enabled, and the GIE bit is set, a set interrupt flag will generate an interrupt request.

USIIFG is set when USICNTx becomes zero, either by counting or by directly writing 0 to the USICNTx bits. USIIFG is cleared by writing a value > 0 to the USICNTx bits when USIIFGCC = 0, or directly by software.

USISTTIFG is set when a START condition is detected. The USISTTIFG flag must be cleared by software.

The reception of a STOP condition is indicated with the USISTP flag but there is no interrupt function associated with the USISTP flag. USISTP is cleared by writing a value > 0 to the USICNTx bits when USIIFGCC = 0 or directly by software.
14.3 USI Registers

Table 14-1 lists the memory-mapped registers for the USI.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>78h</td>
<td>USICTL0</td>
<td>USI control 0</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 14.3.1</td>
</tr>
<tr>
<td>79h</td>
<td>USICTL1</td>
<td>USI control 1</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 14.3.2</td>
</tr>
<tr>
<td>7Ah</td>
<td>USICKCTL</td>
<td>USI clock control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 14.3.3</td>
</tr>
<tr>
<td>7Bh</td>
<td>USICNT</td>
<td>USI bit counter</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 14.3.4</td>
</tr>
<tr>
<td>7Ch</td>
<td>USISRL</td>
<td>USI low byte shift</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 14.3.5</td>
</tr>
<tr>
<td>7Dh</td>
<td>USISRH</td>
<td>USI high byte shift</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 14.3.6</td>
</tr>
</tbody>
</table>

The USI registers can be accessed with word instructions as shown in Table 14-2.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>High-Byte Register</th>
<th>Low-Byte Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>078h</td>
<td>USICTL</td>
<td>USI control</td>
<td>USICTL1</td>
<td>USICTL0</td>
</tr>
<tr>
<td>07Ah</td>
<td>USICCTL</td>
<td>USI clock and counter control</td>
<td>USICNT</td>
<td>USICKCTL</td>
</tr>
<tr>
<td>07Ch</td>
<td>USISR</td>
<td>USI shift</td>
<td>USISRH</td>
<td>USISRL</td>
</tr>
</tbody>
</table>
14.3.1 USICTL0 Register

USI Control 0 Register

USICTL0 is shown in Figure 14-5 and described in Table 14-3.

Return to Table 14-1.

Table 14-3. USICTL0 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>USIPE7</td>
<td>R/W</td>
<td>0h</td>
<td>USI SDI/SDA port enable. Input in SPI mode, input or open drain output in I2C mode. 0b = USI function disabled 1b = USI function enabled</td>
</tr>
<tr>
<td>6</td>
<td>USIPE6</td>
<td>R/W</td>
<td>0h</td>
<td>USI SDO/SCL port enable. Output in SPI mode, input or open drain output in I2C mode. 0b = USI function disabled 1b = USI function enabled</td>
</tr>
<tr>
<td>5</td>
<td>USIPE5</td>
<td>R/W</td>
<td>0h</td>
<td>USI SCLK port enable. Input in SPI slave mode, or I2C mode, output in SPI master mode. 0b = USI function disabled 1b = USI function enabled</td>
</tr>
<tr>
<td>4</td>
<td>USILSB</td>
<td>R/W</td>
<td>0h</td>
<td>LSB first select. This bit controls the direction of the receive and transmit shift register. 0b = MSB first 1b = LSB first</td>
</tr>
<tr>
<td>3</td>
<td>USIMST</td>
<td>R/W</td>
<td>0h</td>
<td>Master select 0b = Slave mode 1b = Master mode</td>
</tr>
<tr>
<td>2</td>
<td>USIGE</td>
<td>R/W</td>
<td>0h</td>
<td>Output latch control 0b = Output latch enable depends on shift clock 1b = Output latch always enabled and transparent</td>
</tr>
<tr>
<td>1</td>
<td>USIOE</td>
<td>R/W</td>
<td>0h</td>
<td>Data output enable 0b = Output disabled 1b = Output enabled</td>
</tr>
<tr>
<td>0</td>
<td>USISWRST</td>
<td>R/W</td>
<td>1h</td>
<td>USI software reset 0b = USI released for operation 1b = USI logic held in reset state</td>
</tr>
</tbody>
</table>
14.3.2 USICTL1 Register

USI Control 1 Register

USICTL1 is shown in Figure 14-6 and described in Table 14-4.

Return to Table 14-1.

Figure 14-6. USICTL1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
</table>
| 7   | USICKPH | R/W  | 0h    | Clock phase select  
|     |        |      |       | 0b = Data is changed on the first SCLK edge and captured on the following edge.  
|     |        |      |       | 1b = Data is captured on the first SCLK edge and changed on the following edge. |
| 6   | USII2C | R/W  | 0h    | I2C mode enable  
|     |        |      |       | 0b = I2C mode disabled  
|     |        |      |       | 1b = I2C mode enabled |
| 5   | USISTTIE | R/W  | 0h   | START condition interrupt-enable  
|     |        |      |       | 0b = Interrupt on START condition disabled  
|     |        |      |       | 1b = Interrupt on START condition enabled |
| 4   | USIE   | R/W  | 0h    | USI counter interrupt enable  
|     |        |      |       | 0b = Interrupt disabled  
|     |        |      |       | 1b = Interrupt enabled |
| 3   | USIAL  | R/W  | 0h    | Arbitration lost  
|     |        |      |       | 0b = No arbitration lost condition  
|     |        |      |       | 1b = Arbitration lost |
| 2   | USISTP | R/W  | 0h    | STOP condition received. USISTP is automatically cleared if USICNTx is loaded with a value > 0 when USIIFGCC = 0.  
|     |        |      |       | 0b = No STOP condition received  
|     |        |      |       | 1b = STOP condition received |
| 1   | USISTTIFG | R/W  | 0h   | START condition interrupt flag  
|     |        |      |       | 0b = No START condition received. No interrupt pending.  
|     |        |      |       | 1b = START condition received. Interrupt pending. |
| 0   | USIFG  | R/W  | 1h    | USI counter interrupt flag. Set when the USICNTx = 0. Automatically cleared if USICNTx is loaded with a value > 0 when USIIFGCC = 0.  
|     |        |      |       | 0b = No interrupt pending  
|     |        |      |       | 1b = Interrupt pending |

Table 14-4. USICTL1 Register Field Descriptions
14.3.3 USICKCTL Register

USI Clock Control Register

USICKCTL is shown in Figure 14-7 and described in Table 14-5.

Return to Table 14-1.

Figure 14-7. USICKCTL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-5</td>
<td>USIDIVx</td>
<td>R/W</td>
<td>0h</td>
<td>Clock divider select</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>000b = Divide by 1</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>001b = Divide by 2</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>010b = Divide by 4</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>011b = Divide by 8</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>100b = Divide by 16</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>101b = Divide by 32</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>110b = Divide by 64</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>111b = Divide by 128</td>
</tr>
<tr>
<td></td>
<td>USISSELx</td>
<td>R/W</td>
<td>0h</td>
<td>Clock source select. Not used in slave mode.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>000b = SCLK (Not used in SPI mode)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>001b = ACLK</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>010b = SMCLK</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>011b = SMCLK</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>100b = USISWCLK bit</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>101b = TACCR0</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>110b = TACCR1</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>111b = TACCR2 (Reserved on MSP430F20xx devices)</td>
</tr>
<tr>
<td>4-2</td>
<td>USICKPL</td>
<td>R/W</td>
<td>0h</td>
<td>Clock polarity select</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Inactive state is low</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Inactive state is high</td>
</tr>
<tr>
<td>0</td>
<td>USISWCLK</td>
<td>R/W</td>
<td>0h</td>
<td>Software clock</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Input clock is low</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Input clock is high</td>
</tr>
</tbody>
</table>
14.3.4 USICNT Register

USI Bit Counter Register

USICNT is shown in Figure 14-8 and described in Table 14-6.

Return to Table 14-1.

---

**Figure 14-8. USICNT Register**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>USISCLREL</td>
<td>R/W</td>
<td>0h</td>
<td>SCL release. The SCL line is released from low to idle. USISCLREL is cleared if a START condition is detected. 0b = SCL line is held low if USIIFG is set 1b = SCL line is released</td>
</tr>
<tr>
<td>6</td>
<td>USI16B</td>
<td>R/W</td>
<td>0h</td>
<td>16-bit shift register enable 0b = 8-bit shift register mode. Low byte register USISRL is used. 1b = 16-bit shift register mode. Both high and low byte registers USISRL and USISRH are used. USISR addresses all 16 bits simultaneously.</td>
</tr>
<tr>
<td>5</td>
<td>USIIFGCC</td>
<td>R/W</td>
<td>0h</td>
<td>USI interrupt flag clear control. When USIIFGCC = 1 the USIIFG will not be cleared automatically when USICNTx is written with a value &gt; 0. 0b = USIIFG automatically cleared on USICNTx update 1b = USIIFG is not cleared automatically</td>
</tr>
<tr>
<td>4-0</td>
<td>USICNTx</td>
<td>R/W</td>
<td>0h</td>
<td>USI bit count. The USICNTx bits set the number of bits to be received or transmitted.</td>
</tr>
</tbody>
</table>
14.3.5 USISRL Register
USI Low Byte Shift Register

USISRL is shown in Figure 14-9 and described in Table 14-7.
Return to Table 14-1.

Figure 14-9. USISRL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>USISRLx</td>
<td>R/W</td>
<td>Unchanged</td>
<td>Contents of the USI low byte shift register</td>
</tr>
</tbody>
</table>

14.3.6 USISRH Register
USI High Byte Shift Register

USISRH is shown in Figure 14-10 and described in Table 14-8.
Return to Table 14-1.

Figure 14-10. USISRH Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>USISRHx</td>
<td>R/W</td>
<td>Unchanged</td>
<td>Contents of the USI high byte shift register. Ignored when USI16B = 0.</td>
</tr>
</tbody>
</table>
Chapter 15

Universal Serial Communication Interface, UART Mode

The universal serial communication interface (USCI) supports multiple serial communication modes with one hardware module. This chapter discusses the operation of the asynchronous UART mode.

15.1 USCI Overview................................................................. 434
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15.3 USCI Operation: UART Mode............................................ 436
15.4 USCI Registers: UART Mode............................................. 452
15.1 USCI Overview

The universal serial communication interface (USCI) modules support multiple serial communication modes. Different USCI modules support different modes. Each different USCI module is named with a different letter. For example, USCI_A is different from USCI_B, etc. If more than one identical USCI module is implemented on one device, those modules are named with incrementing numbers. For example, if one device has two USCI_A modules, they are named USCI_A0 and USCI_A1. See the device-specific data sheet to determine which USCI modules, if any, are implemented on which devices.

The USCI_Ax modules support:
- UART mode
- Pulse shaping for IrDA communications
- Automatic baud rate detection for LIN communications
- SPI mode

The USCI_Bx modules support:
- I²C mode
- SPI mode

15.2 USCI Introduction: UART Mode

In asynchronous mode, the USCI_Ax modules connect the MSP430 to an external system via two external pins, UCAxRXD and UCAxTXD. UART mode is selected when the UCSYNC bit is cleared.

UART mode features include:
- 7- or 8-bit data with odd, even, or non-parity
- Independent transmit and receive shift registers
- Separate transmit and receive buffer registers
- LSB-first or MSB-first data transmit and receive
- Built-in idle-line and address-bit communication protocols for multiprocessor systems
- Receiver start-edge detection for auto-wake up from LPMx modes
- Programmable baud rate with modulation for fractional baud rate support
- Status flags for error detection and suppression
- Status flags for address detection
- Independent interrupt capability for receive and transmit

Figure 15-1 shows the USCI_Ax when configured for UART mode.
Figure 15-1. USCI_Ax Block Diagram: UART Mode (UCSYNC = 0)
15.3 USCI Operation: UART Mode

In UART mode, the USCI transmits and receives characters at a bit rate asynchronous to another device. Timing for each character is based on the selected baud rate of the USCI. The transmit and receive functions use the same baud rate frequency.

15.3.1 USCI Initialization and Reset

The USCI is reset by a PUC or by setting the UCSWRST bit. After a PUC, the UCSWRST bit is automatically set, keeping the USCI in a reset condition. When set, the UCSWRST bit resets the UCAxRXIE, UCAxTXIE, UCAxRXIFG, UCRXERR, UCBRK, UCPE, UCOE, UCFE, UCSTOE and UCBTOE bits and sets the UCAxTXIFG bit. Clearing UCSWRST releases the USCI for operation.

**Note**

**Initializing or Reconfiguring the USCI Module**

The recommended USCI initialization or reconfiguration process is:
1. Set UCSWRST (BIS.B #UCSWRST,&UCAxCTL1)
2. Initialize all USCI registers with UCSWRST = 1 (including UCAxCTL1)
3. Configure ports.
4. Clear UCSWRST via software (BIC.B #UCSWRST,&UCAxCTL1)
5. Enable interrupts (optional) via UCAxRXIE and/or UCAxTXIE

15.3.2 Character Format

The UART character format, shown in Figure 15-2, consists of a start bit, seven or eight data bits, an even/odd/no parity bit, an address bit (address-bit mode), and one or two stop bits. The UCMSB bit controls the direction of the transfer and selects LSB or MSB first. LSB-first is typically required for UART communication.

![Character Format Diagram](image)

**Figure 15-2. Character Format**

15.3.3 Asynchronous Communication Formats

When two devices communicate asynchronously, no multiprocessor format is required for the protocol. When three or more devices communicate, the USCI supports the idle-line and address-bit multiprocessor communication formats.

15.3.3.1 Idle-Line Multiprocessor Format

When UCMODEEx = 01, the idle-line multiprocessor format is selected. Blocks of data are separated by an idle time on the transmit or receive lines as shown in Figure 15-3. An idle receive line is detected when 10 or more continuous ones (marks) are received after the one or two stop bits of a character. The baud rate generator is switched off after reception of an idle line until the next start edge is detected. When an idle line is detected the UCIDLE bit is set.

The first character received after an idle period is an address character. The UCIDLE bit is used as an address tag for each block of characters. In idle-line multiprocessor format, this bit is set when a received character is an address.
Figure 15-3. Idle-Line Format

The UCDORM bit is used to control data reception in the idle-line multiprocessor format. When UCDORM = 1, all non-address characters are assembled but not transferred into the UCAxRXBUF, and interrupts are not generated. When an address character is received, the character is transferred into UCAxRXBUF, UCAxRXIFG is set, and any applicable error flag is set when UCRXEIE = 1. When UCRXEIE = 0 and an address character is received but has a framing error or parity error, the character is not transferred into UCAxRXBUF and UCAxRXIFG is not set.

If an address is received, user software can validate the address and must reset UCDORM to continue receiving data. If UCDORM remains set, only address characters will be received. When UCDORM is cleared during the reception of a character the receive interrupt flag will be set after the reception completed. The UCDORM bit is not modified by the USCI hardware automatically.

For address transmission in idle-line multiprocessor format, a precise idle period can be generated by the USCI to generate address character identifiers on UCAxTXD. The double-buffered UCTXADDR flag indicates if the next character loaded into UCAxTXBUF is preceded by an idle line of 11 bits. UCTXADDR is automatically cleared when the start bit is generated.

15.3.3.2 Transmitting an Idle Frame

The following procedure sends out an idle frame to indicate an address character followed by associated data:

1. Set UCTXADDR, then write the address character to UCAxTXBUF. UCAxTXBUF must be ready for new data (UCAxTXIFG = 1).
   
   This generates an idle period of exactly 11 bits followed by the address character. UCTXADDR is reset automatically when the address character is transferred from UCAxTXBUF into the shift register.

2. Write desired data characters to UCAxTXBUF. UCAxTXBUF must be ready for new data (UCAxTXIFG = 1).
   
   The data written to UCAxTXBUF is transferred to the shift register and transmitted as soon as the shift register is ready for new data.

   The idle-line time must not be exceeded between address and data transmission or between data transmissions. Otherwise, the transmitted data will be misinterpreted as an address.

15.3.3.3 Address-Bit Multiprocessor Format

When UCMODEx = 10, the address-bit multiprocessor format is selected. Each processed character contains an extra bit used as an address indicator shown in Figure 15-4. The first character in a block of characters carries a
set address bit which indicates that the character is an address. The USCI UCADDR bit is set when a received character has its address bit set and is transferred to UCAxRXBUF.

The UCDORM bit is used to control data reception in the address-bit multiprocessor format. When UCDORM is set, data characters with address bit = 0 are assembled by the receiver but are not transferred to UCAxRXBUF and no interrupts are generated. When a character containing a set address bit is received, the character is transferred into UCAxRXBUF, UCAxRXIFG is set, and any applicable error flag is set when UCRXEIE = 1. When UCRXEIE = 0 and a character containing a set address bit is received, but has a framing error or parity error, the character is not transferred into UCAxRXBUF and UCAxRXIFG is not set.

If an address is received, user software can validate the address and must reset UCDORM to continue receiving data. If UCDORM remains set, only address characters with address bit = 1 will be received. The UCDORM bit is not modified by the USCI hardware automatically.

When UCDORM = 0 all received characters will set the receive interrupt flag UCAxRXIFG. If UCDORM is cleared during the reception of a character the receive interrupt flag will be set after the reception is completed.

For address transmission in address-bit multiprocessor mode, the address bit of a character is controlled by theUCTXADDR bit. The value of the UCTXADDR bit is loaded into the address bit of the character transferred from UCAxTXBUF to the transmit shift register. UCTXADDR is automatically cleared when the start bit is generated.

---

![Figure 15-4. Address-Bit Multiprocessor Format](image-url)

**15.3.3.4 Break Reception and Generation**

When UCMODEEx = 00, 01, or 10 the receiver detects a break when all data, parity, and stop bits are low, regardless of the parity, address mode, or other character settings. When a break is detected, the UCBRK bit is set. If the break interrupt enable bit, UCBRKIE, is set, the receive interrupt flag UCAxRXIFG will also be set. In this case, the value in UCAxRXBUF is 0h since all data bits were zero.

To transmit a break set the UCTXBRK bit, then write 0h to UCAxTXBUF. UCAxTXBUF must be ready for new data (UCAxTXIFG = 1). This generates a break with all bits low. UCTXBRK is automatically cleared when the start bit is generated.

**15.3.4 Automatic Baud Rate Detection**

When UCMODEEx = 11 UART mode with automatic baud rate detection is selected. For automatic baud rate detection, a data frame is preceded by a synchronization sequence that consists of a break and a synch field. A break is detected when 11 or more continuous zeros (spaces) are received. If the length of the break exceeds 22 bit times the break timeout error flag UCBTOE is set. The synch field follows the break as shown in Figure 15-5.
For LIN conformance the character format should be set to 8 data bits, LSB first, no parity and one stop bit. No address bit is available.

The synch field consists of the data 055h inside a byte field as shown in Figure 15-6. The synchronization is based on the time measurement between the first falling edge and the last falling edge of the pattern. The transmit baud rate generator is used for the measurement if automatic baud rate detection is enabled by setting UCABDEN. Otherwise, the pattern is received but not measured. The result of the measurement is transferred into the baud rate control registers UCAxBR0, UCAxBR1, and UCAxMCTL. If the length of the synch field exceeds the measurable time the synch timeout error flag UCSTOE is set.

When a break/synch field is received, user software must reset UCDORM to continue receiving data. If UCDORM remains set, only the character after the next reception of a break/synch field will be received. The UCDORM bit is not modified by the USCI hardware automatically.

When UCDORM = 0 all received characters will set the receive interrupt flag UCAxRXIFG. If UCDORM is cleared during the reception of a character the receive interrupt flag will be set after the reception is complete.

The automatic baud rate detection mode can be used in a full-duplex communication system with some restrictions. The USCI can not transmit data while receiving the break/sync field and if a 0h byte with framing error is received any data transmitted during this time gets corrupted. The latter case can be discovered by checking the received data and the UCFE bit.
15.3.4.1 Transmitting a Break/Synch Field

The following procedure transmits a break/synch field:

- Set UCTXBRK with UMODEEx = 11.
- Write 055h to UCAxTXBUF. UCAxTXBUF must be ready for new data (UCAxTXIFG = 1).
  
  This generates a break field of 13 bits followed by a break delimiter and the synch character. The length of the break delimiter is controlled with the UCDELIMx bits. UCTXBRK is reset automatically when the synch character is transferred from UCAxTXBUF into the shift register.
- Write desired data characters to UCAxTXBUF. UCAxTXBUF must be ready for new data (UCAxTXIFG = 1).
  
  The data written to UCAxTXBUF is transferred to the shift register and transmitted as soon as the shift register is ready for new data.

15.3.5 IrDA Encoding and Decoding

When UCIREN is set the IrDA encoder and decoder are enabled and provide hardware bit shaping for IrDA communication.

15.3.5.1 IrDA Encoding

The encoder sends a pulse for every zero bit in the transmit bit stream coming from the UART as shown in Figure 15-7. The pulse duration is defined by UCIRTXPLx bits specifying the number of half clock periods of the clock selected by UCIRTXCLK.

To set the pulse time of 3/16 bit period required by the IrDA standard the BITCLK16 clock is selected with UCIRTXCLK = 1 and the pulse length is set to 6 half clock cycles with UCIRTXPLx = 6 – 1 = 5.

When UCIRTXCLK = 0, the pulse length \( t_{\text{PULSE}} \) is based on BRCLK and is calculated as follows:

\[
\text{UCIRTXPLx} = \frac{t_{\text{PULSE}}}{2} \times f_{\text{BRCLK}} - 1
\]

When the pulse length is based on BRCLK the prescaler UCBRx must be set to a value greater or equal to 5.

15.3.5.2 IrDA Decoding

The decoder detects high pulses when UCIRRXPL = 0. Otherwise it detects low pulses. In addition to the analog deglitch filter an additional programmable digital filter stage can be enabled by setting UCIRRXFE. When UCIRRXFE is set, only pulses longer than the programmed filter length are passed. Shorter pulses are discarded. The equation to program the filter length UCIRRXFLx is:

\[
\text{UCIRRXFLx} = \frac{t_{\text{PULSE}} - t_{\text{WAKE}}}{2} \times f_{\text{BRCLK}} - 4
\]

Where,

\( t_{\text{PULSE}} \) = Minimum receive pulse duration

\( t_{\text{WAKE}} \) = Wake time from any low-power mode. Zero when MSP430 is in active mode.
Note

Reliable reception of IrDA signals

To receive incoming IrDA signals reliably, make sure that at least one of the following procedures are implemented:

• Enable the digital filter stage with UCIRRXFE = 1.
• Use a parity bit to detect corrupted bytes.
• Check the correctness of received data frames using a checksum or CRC.
• With parity or CRC checks, use a protocol that acknowledges received data frame and resends data if the sender does not receive an acknowledgment.
15.3.6 Automatic Error Detection

Glitch suppression prevents the USCI from being accidentally started. Any pulse on UCxRXD shorter than the de-glitch time \( t_\tau \) (approximately 150 ns) will be ignored. See the device-specific data sheet for parameters.

When a low period on UCxRXD exceeds \( t_\tau \), a majority vote is taken for the start bit. If the majority vote fails to detect a valid start bit the USCI halts character reception and waits for the next low period on UCxRXD. The majority vote is also used for each bit in a character to prevent bit errors.

The USCI module automatically detects framing errors, parity errors, overrun errors, and break conditions when receiving characters. The bits UCFE, UCPE, UCOE, and UCBRK are set when their respective condition is detected. When the error flags UCFE, UCPE or UCOE are set, UCRXERR is also set. The error conditions are described in Table 15-1.

<table>
<thead>
<tr>
<th>Error Condition</th>
<th>Error Flag</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>Framing error</td>
<td>UCFE</td>
<td>A framing error occurs when a low stop bit is detected. When two stop bits are used, both stop bits are checked for framing error. When a framing error is detected, the UCFE bit is set.</td>
</tr>
<tr>
<td>Parity error</td>
<td>UCPE</td>
<td>A parity error is a mismatch between the number of 1s in a character and the value of the parity bit. When an address bit is included in the character, it is included in the parity calculation. When a parity error is detected, the UCPE bit is set.</td>
</tr>
<tr>
<td>Receive overrun</td>
<td>UCOE</td>
<td>An overrun error occurs when a character is loaded into UCARXBUF before the prior character has been read. When an overrun occurs, the UCOE bit is set.</td>
</tr>
<tr>
<td>Break condition</td>
<td>UCBRK</td>
<td>When not using automatic baud rate detection, a break is detected when all data, parity, and stop bits are low. When a break condition is detected, the UCBRK bit is set. A break condition can also set the interrupt flag UCARXIFG if the break interrupt enable UCBRKIE bit is set.</td>
</tr>
</tbody>
</table>

When UCRXEIE = 0 and a framing error, or parity error is detected, no character is received into UCARXBUF. When UCRXEIE = 1, characters are received into UCARXBUF and any applicable error bit is set.

When UCFE, UCPE, UCOE, UCBRK, or UCRXERR is set, the bit remains set until user software resets it or UCARXBUF is read. UCOE must be reset by reading UCARXBUF. Otherwise it will not function properly. To detect overflows reliably, the following flow is recommended. After a character is received and UCARXIFG is set, first read UCARXSTAT to check the error flags including the overflow flag UCOE. Read UCARXBUF next. This will clear all error flags except UCOE, if UCARXBUF was overwritten between the read access to UCARXSTAT and to UCARXBUF. The UCOE flag should be checked after reading UCARXBUF to detect this condition. Note that, in this case, the UCRXERR flag is not set.

15.3.7 USCI Receive Enable

The USCI module is enabled by clearing the UCSWRST bit and the receiver is ready and in an idle state. The receive baud rate generator is in a ready state but is not clocked nor producing any clocks.

The falling edge of the start bit enables the baud rate generator and the UART state machine checks for a valid start bit. If no valid start bit is detected the UART state machine returns to its idle state and the baud rate generator is turned off again. If a valid start bit is detected a character will be received.

When the idle-line multiprocessor mode is selected with UCMODEx = 01 the UART state machine checks for an idle line after receiving a character. If a start bit is detected another character is received. Otherwise the UCIDLE flag is set after 10 ones are received and the UART state machine returns to its idle state and the baud rate generator is turned off.
**15.3.7.1 Receive Data Glitch Suppression**

Glitch suppression prevents the USCI from being accidentally started. Any glitch on UCAxRXD shorter than the deglitch time \( t_\tau \) (approximately 150 ns) will be ignored by the USCI and further action will be initiated as shown in Figure 15-8. See the device-specific data sheet for parameters.

![Figure 15-8. Glitch Suppression, USCI Receive Not Started](image)

When a glitch is longer than \( t_\tau \) or a valid start bit occurs on UCAxRXD, the USCI receive operation is started and a majority vote is taken as shown in Figure 15-9. If the majority vote fails to detect a start bit the USCI halts character reception.

![Figure 15-9. Glitch Suppression, USCI Activated](image)

**15.3.8 USCI Transmit Enable**

The USCI module is enabled by clearing the UCSWRST bit and the transmitter is ready and in an idle state. The transmit baud rate generator is ready but is not clocked nor producing any clocks.

A transmission is initiated by writing data to UCAxTXBUF. When this occurs, the baud rate generator is enabled and the data in UCAxTXBUF is moved to the transmit shift register on the next BITCLK after the transmit shift register is empty. UCAxTXIFG is set when new data can be written into UCAxTXBUF.

Transmission continues as long as new data is available in UCAxTXBUF at the end of the previous byte transmission. If new data is not in UCAxTXBUF when the previous byte has transmitted, the transmitter returns to its idle state and the baud rate generator is turned off.

**15.3.9 UART Baud Rate Generation**

The USCI baud rate generator is capable of producing standard baud rates from non-standard source frequencies. It provides two modes of operation selected by the UCOS16 bit.

**15.3.9.1 Low-Frequency Baud Rate Generation**

The low-frequency mode is selected when UCOS16 = 0. This mode allows generation of baud rates from low frequency clock sources (for example, 9600 baud from a 32768-Hz crystal). By using a lower input frequency the power consumption of the module is reduced. Using this mode with higher frequencies and higher prescaler settings will cause the majority votes to be taken in an increasingly smaller window and thus decrease the benefit of the majority vote.

In low-frequency mode the baud rate generator uses one prescaler and one modulator to generate bit clock timing. This combination supports fractional divisors for baud rate generation. In this mode, the maximum USCI baud rate is one-third the UART source clock frequency BRCLK.
Timing for each bit is shown in Figure 15-10. For each bit received, a majority vote is taken to determine the bit value. These samples occur at the N/2 - 1/2, N/2, and N/2 + 1/2 BRCLK periods, where N is the number of BRCLKs per BITCLK.

![Figure 15-10. BITCLK Baud Rate Timing With UCOS16 = 0](image)

Modulation is based on the UCBRSx setting as shown in Table 15-2. A 1 in the table indicates that m = 1 and the corresponding BITCLK period is one BRCLK period longer than a BITCLK period with m = 0. The modulation wraps around after 8 bits but restarts with each new start bit.

<table>
<thead>
<tr>
<th>UCBRSx</th>
<th>Bit 0 (Start Bit)</th>
<th>Bit 1</th>
<th>Bit 2</th>
<th>Bit 3</th>
<th>Bit 4</th>
<th>Bit 5</th>
<th>Bit 6</th>
<th>Bit 7</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
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<td>0</td>
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<td>2</td>
<td>0</td>
<td>1</td>
<td>0</td>
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<td>0</td>
</tr>
<tr>
<td>3</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>4</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>5</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>6</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>7</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>1</td>
</tr>
</tbody>
</table>

### 15.3.9.2 Oversampling Baud Rate Generation

The oversampling mode is selected when UCOS16 = 1. This mode supports sampling a UART bit stream with higher input clock frequencies. This results in majority votes that are always 1/16 of a bit clock period apart. This mode also easily supports IrDA pulses with a 3/16 bit-time when the IrDA encoder and decoder are enabled.

This mode uses one prescaler and one modulator to generate the BITCLK16 clock that is 16 times faster than the BITCLK. An additional divider and modulator stage generates BITCLK from BITCLK16. This combination supports fractional divisions of both BITCLK16 and BITCLK for baud rate generation. In this mode, the maximum USCI baud rate is 1/16 the UART source clock frequency BRCLK. When UCBRx is set to 0 or 1 the first prescaler and modulator stage is bypassed and BRCLK is equal to BITCLK16.

Modulation for BITCLK16 is based on the UCBRFx setting as shown in Table 15-3. A 1 in the table indicates that the corresponding BITCLK16 period is one BRCLK period longer than the periods m = 0. The modulation restarts with each new bit timing.

Modulation for BITCLK is based on the UCBRSx setting as shown in Table 15-2 as previously described.
### Table 15-3. BITCLK16 Modulation Pattern

<table>
<thead>
<tr>
<th>UCBRx</th>
<th>No. of BITCLK16 Clocks After Last Falling BITCLK Edge</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>0</td>
</tr>
<tr>
<td>00h</td>
<td>0</td>
</tr>
<tr>
<td>01h</td>
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</tr>
<tr>
<td>02h</td>
<td>0</td>
</tr>
<tr>
<td>03h</td>
<td>0</td>
</tr>
<tr>
<td>04h</td>
<td>0</td>
</tr>
<tr>
<td>05h</td>
<td>0</td>
</tr>
<tr>
<td>06h</td>
<td>0</td>
</tr>
<tr>
<td>07h</td>
<td>0</td>
</tr>
<tr>
<td>08h</td>
<td>0</td>
</tr>
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<td>09h</td>
<td>0</td>
</tr>
<tr>
<td>0Ah</td>
<td>0</td>
</tr>
<tr>
<td>0Bh</td>
<td>0</td>
</tr>
<tr>
<td>0Ch</td>
<td>0</td>
</tr>
<tr>
<td>0Dh</td>
<td>0</td>
</tr>
<tr>
<td>0Eh</td>
<td>0</td>
</tr>
<tr>
<td>0Fh</td>
<td>0</td>
</tr>
</tbody>
</table>

#### 15.3.10 Setting a Baud Rate

For a given BRCLK clock source, the baud rate used determines the required division factor \( N \):

\[
N = \frac{\text{fBRCLK}}{\text{Baud rate}}
\]

The division factor \( N \) is often a non-integer value thus at least one divider and one modulator stage is used to meet the factor as closely as possible.

If \( N \) is equal or greater than 16 the oversampling baud rate generation mode can be chosen by setting UCOS16.

**15.3.10.1 Low-Frequency Baud Rate Mode Setting**

In the low-frequency mode, the integer portion of the divisor is realized by the prescaler:

\[
\text{UCBRx} = \text{INT}(N)
\]

and the fractional portion is realized by the modulator with the following nominal formula:

\[
\text{UCBRSx} = \text{round}\left( (N - \text{INT}(N)) \times 8 \right)
\]

Incrementing or decrementing the UCBRSx setting by one count may give a lower maximum bit error for any given bit. To determine if this is the case, a detailed error calculation must be performed for each bit for each UCBRSx setting.

**15.3.10.2 Oversampling Baud Rate Mode Setting**

In the oversampling mode the prescaler is set to:

\[
\text{UCBRx} = \text{INT}\left(\frac{N}{16}\right)
\]

and the first stage modulator is set to:

\[
\text{UCBFX} = \text{round}\left(\left(\frac{N}{16} - \text{INT}\left(\frac{N}{16}\right)\right) \times 16\right)
\]

When greater accuracy is required, the UCBRSx modulator can also be implemented with values from 0 to 7. To find the setting that gives the lowest maximum bit error rate for any given bit, a detailed error calculation must be performed.
performed for all settings of UCBRSx from 0 to 7 with the initial UCBRFx setting and with the UCBRFx setting incremented and decremented by one.

15.3.11 Transmit Bit Timing
The timing for each character is the sum of the individual bit timings. Using the modulation features of the baud rate generator reduces the cumulative bit error. The individual bit error can be calculated using the following steps.

15.3.11.1 Low-Frequency Baud Rate Mode Bit Timing
In low-frequency mode, calculate the length of bit i $T_{bit,TX[i]}$ based on the UCBRx and UCBRSx settings:

$$T_{bit,TX[i]} = \frac{1}{BRCLK} (UCBRx + m_{UCBRSx[i]})$$

Where,

$m_{UCBRSx[i]} = \text{Modulation of bit } i \text{ from Table 15-2}$

15.3.11.2 Oversampling Baud Rate Mode Bit Timing
In oversampling baud rate mode calculate the length of bit i $T_{bit,TX[i]}$ based on the baud rate generator UCBRx, UCBRFx and UCBRSx settings:

$$T_{bit,TX[i]} = \frac{1}{BRCLK} \left( 16 + m_{UCBRSx[i]} \times UCBRx + \sum_{j=0}^{15} m_{UCBFRFx[j]} \right)$$

Where,

$$\sum_{j=0}^{15} m_{UCBFRFx[j]} = \text{Sum of ones from the corresponding row in Table 15-3}$$

$m_{UCBRSx[i]} = \text{Modulation of bit } i \text{ from Table 15-2}$

This results in an end-of-bit time $t_{bit,TX[i]}$ equal to the sum of all previous and the current bit times:

$$t_{bit,TX[i]} = \sum_{j=0}^{i} T_{bit,TX[j]}$$

To calculate bit error, this time is compared to the ideal bit time $t_{bit,\text{ideal,TX}[i]}$:

$$t_{bit,\text{ideal,TX}[i]} = \frac{1}{\text{Baud rate}} (i + 1)$$

This results in an error normalized to one ideal bit time (1 / baud rate):

$$\text{Error}_{TX[i]} = \left( t_{bit,TX[i]} - t_{bit,\text{ideal,TX}[i]} \right) \times \text{baud rate} \times 100\%$$

15.3.12 Receive Bit Timing
Receive timing error consists of two error sources. The first is the bit-to-bit timing error similar to the transmit bit timing error. The second is the error between a start edge occurring and the start edge being accepted by the USCI module. Figure 15-11 shows the asynchronous timing errors between data on the UCAxRXD pin and the internal baud-rate clock. This results in an additional synchronization error. The synchronization error $t_{SYNC}$ is between -0.5 BRCLKs and +0.5 BRCLKs independent of the selected baud rate generation mode.
The ideal sampling time is in the middle of a bit period:

\[ t_{\text{bit,ideal,RX}[i]} = \frac{1}{\text{Baud rate}} (i + 0.5) \]

The real sampling time is equal to the sum of all previous bits according to the formulas shown in the transmit timing section, plus one half BITCLK for the current bit i, plus the synchronization error \( t_{\text{SYNC}} \).

This results in the following for the low-frequency baud rate mode:

\[
t_{\text{bit,RX}[i]} = t_{\text{SYNC}} + \sum_{j=0}^{i-1} T_{\text{bit,RX}[j]} + \frac{1}{T_{\text{BRCLK}}} \left( \text{INT} \left( \frac{1}{2} \text{UCBRx} \right) + m_{\text{UCBRSx}[i]} \right)
\]

Where,

\[
T_{\text{bit,RX}[i]} = \frac{1}{T_{\text{BRCLK}}} ( \text{UCBRx} + m_{\text{UCBRSx}[i]} )
\]

\( m_{\text{UCBRSx}[i]} \) = Modulation of bit i from Table 15-2

For the oversampling baud rate mode the sampling time of bit i is calculated by:

\[
t_{\text{bit,RX}[i]} = t_{\text{SYNC}} + \sum_{j=0}^{i-1} T_{\text{bit,RX}[j]} + \frac{1}{T_{\text{BRCLK}}} \left( 8 + m_{\text{UCBRSx}[i]} \right) \times \text{UCBRx} + \sum_{j=0}^{7+m_{\text{UCBRSx}[i]}} m_{\text{UCBFRX}[i]}
\]

Where,

\[
T_{\text{bit,RX}[i]} = \frac{1}{T_{\text{BRCLK}}} \left( 16 + m_{\text{UCBRSx}[i]} \right) \times \text{UCBRx} + \sum_{j=0}^{15} m_{\text{UCBFRX}[i]}
\]

\[
\sum_{j=0}^{7+m_{\text{UCBRSx}[i]}} m_{\text{UCBFRX}[i]} = \text{Sum of ones from columns 0 - from the corresponding row in Table 15-3}
\]

\( m_{\text{UCBRSx}[i]} \) = Modulation of bit i from Table 15-2

This results in an error normalized to one ideal bit time (1 / baud rate) according to the following formula:
Error_{RX[i]} = (t_{bit,RX[i]} - t_{bit,ideal,RX[i]}) \times \text{baud rate} \times 100\%

### 15.3.13 Typical Baud Rates and Errors

Standard baud rate data for UCBRx, UCBRSx and UCBRFx are listed in Table 15-4 and Table 15-5 for a 32768-Hz crystal sourcing ACLK and typical SMCLK frequencies. Ensure that the selected BRCLK frequency does not exceed the device-specific maximum USCI input frequency (see the device-specific data sheet).

The receive error is the accumulated time versus the ideal scanning time in the middle of each bit. The worst case error is given for the reception of an 8-bit character with parity and one stop bit including synchronization error.

The transmit error is the accumulated timing error versus the ideal time of the bit period. The worst case error is given for the transmission of an 8-bit character with parity and stop bit.

#### Table 15-4. Commonly Used Baud Rates, Settings, and Errors, UCOS16 = 0

<table>
<thead>
<tr>
<th>BRCLK Frequency [Hz]</th>
<th>Baud Rate [Baud]</th>
<th>UCBRx</th>
<th>UCBRSx</th>
<th>UCBRFx</th>
<th>Maximum TX Error [%]</th>
<th>Maximum RX Error [%]</th>
</tr>
</thead>
<tbody>
<tr>
<td>32,768</td>
<td>1200</td>
<td>27</td>
<td>2</td>
<td>0</td>
<td>-2.8</td>
<td>1.4</td>
</tr>
<tr>
<td>32,768</td>
<td>2400</td>
<td>13</td>
<td>6</td>
<td>0</td>
<td>-4.8</td>
<td>6.0</td>
</tr>
<tr>
<td>32,768</td>
<td>4800</td>
<td>6</td>
<td>7</td>
<td>0</td>
<td>-12.1</td>
<td>5.7</td>
</tr>
<tr>
<td>32,768</td>
<td>9600</td>
<td>3</td>
<td>3</td>
<td>0</td>
<td>-21.1</td>
<td>15.2</td>
</tr>
<tr>
<td>1,048,576</td>
<td>9600</td>
<td>109</td>
<td>2</td>
<td>0</td>
<td>-0.2</td>
<td>0.7</td>
</tr>
<tr>
<td>1,048,576</td>
<td>19200</td>
<td>54</td>
<td>5</td>
<td>0</td>
<td>-1.1</td>
<td>1.0</td>
</tr>
<tr>
<td>1,048,576</td>
<td>38400</td>
<td>27</td>
<td>2</td>
<td>0</td>
<td>-2.8</td>
<td>1.4</td>
</tr>
<tr>
<td>1,048,576</td>
<td>56000</td>
<td>18</td>
<td>6</td>
<td>0</td>
<td>-3.9</td>
<td>1.1</td>
</tr>
<tr>
<td>1,048,576</td>
<td>115200</td>
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<td>0</td>
<td>-1.1</td>
<td>10.7</td>
</tr>
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<td>1,048,576</td>
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<td>0</td>
<td>-8.9</td>
<td>7.5</td>
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<td>-2.3</td>
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<td>-0.5</td>
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<td>0</td>
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<td>0</td>
<td>-1.8</td>
<td>0</td>
</tr>
<tr>
<td>1,000,000</td>
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<td>0</td>
<td>-4.8</td>
<td>0.8</td>
</tr>
<tr>
<td>1,000,000</td>
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<td>6</td>
<td>0</td>
<td>-7.8</td>
<td>6.4</td>
</tr>
<tr>
<td>1,000,000</td>
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<td>-10.4</td>
<td>6.4</td>
</tr>
<tr>
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<td>-29.6</td>
<td>0</td>
</tr>
<tr>
<td>4,000,000</td>
<td>9600</td>
<td>416</td>
<td>6</td>
<td>0</td>
<td>-0.2</td>
<td>0.2</td>
</tr>
<tr>
<td>4,000,000</td>
<td>19200</td>
<td>208</td>
<td>3</td>
<td>0</td>
<td>-0.2</td>
<td>0.5</td>
</tr>
<tr>
<td>4,000,000</td>
<td>38400</td>
<td>104</td>
<td>1</td>
<td>0</td>
<td>-0.5</td>
<td>0.6</td>
</tr>
<tr>
<td>4,000,000</td>
<td>56000</td>
<td>71</td>
<td>4</td>
<td>0</td>
<td>-0.6</td>
<td>1.0</td>
</tr>
<tr>
<td>4,000,000</td>
<td>115200</td>
<td>34</td>
<td>6</td>
<td>0</td>
<td>-2.1</td>
<td>0.6</td>
</tr>
<tr>
<td>4,000,000</td>
<td>128000</td>
<td>31</td>
<td>2</td>
<td>0</td>
<td>-0.8</td>
<td>1.6</td>
</tr>
<tr>
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<td>256000</td>
<td>15</td>
<td>5</td>
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<td>-4.0</td>
<td>3.2</td>
</tr>
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<td>9600</td>
<td>833</td>
<td>2</td>
<td>0</td>
<td>-0.1</td>
<td>0</td>
</tr>
<tr>
<td>8,000,000</td>
<td>19200</td>
<td>416</td>
<td>6</td>
<td>0</td>
<td>-0.2</td>
<td>0.2</td>
</tr>
<tr>
<td>8,000,000</td>
<td>38400</td>
<td>208</td>
<td>3</td>
<td>0</td>
<td>-0.2</td>
<td>0.5</td>
</tr>
<tr>
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<td>115200</td>
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<td>4</td>
<td>0</td>
<td>-0.6</td>
<td>0.8</td>
</tr>
<tr>
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<td>128000</td>
<td>62</td>
<td>4</td>
<td>0</td>
<td>-0.8</td>
<td>0</td>
</tr>
<tr>
<td>8,000,000</td>
<td>256000</td>
<td>31</td>
<td>2</td>
<td>0</td>
<td>-0.8</td>
<td>1.6</td>
</tr>
<tr>
<td>12,000,000</td>
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<td>1250</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>-0.05</td>
</tr>
</tbody>
</table>
### Table 15-4. Commonly Used Baud Rates, Settings, and Errors, UCOS16 = 0 (continued)

<table>
<thead>
<tr>
<th>BRCLK Frequency [Hz]</th>
<th>Baud Rate [Baud]</th>
<th>UCBRx</th>
<th>UCBRx</th>
<th>UCBRSx</th>
<th>UCBRFx</th>
<th>Maximum TX Error [%]</th>
<th>Maximum RX Error [%]</th>
</tr>
</thead>
<tbody>
<tr>
<td>12,000,000</td>
<td>19200</td>
<td>625</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>-0.2</td>
<td>0</td>
</tr>
<tr>
<td>12,000,000</td>
<td>38400</td>
<td>312</td>
<td>4</td>
<td>0</td>
<td>-0.2</td>
<td>0</td>
<td>0.2</td>
</tr>
<tr>
<td>12,000,000</td>
<td>56000</td>
<td>214</td>
<td>2</td>
<td>0</td>
<td>-0.3</td>
<td>0.2</td>
<td>0.2</td>
</tr>
<tr>
<td>12,000,000</td>
<td>115200</td>
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<td>0</td>
<td>-0.5</td>
<td>0.6</td>
<td>0.9</td>
</tr>
<tr>
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<td>-0.8</td>
<td>0</td>
<td>1.5</td>
</tr>
<tr>
<td>12,000,000</td>
<td>256000</td>
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<td>7</td>
<td>0</td>
<td>-1.9</td>
<td>0</td>
<td>2.0</td>
</tr>
<tr>
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<td>1666</td>
<td>6</td>
<td>0</td>
<td>-0.05</td>
<td>0.05</td>
<td>0.05</td>
</tr>
<tr>
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<td>19200</td>
<td>833</td>
<td>2</td>
<td>0</td>
<td>-0.1</td>
<td>0.05</td>
<td>0.1</td>
</tr>
<tr>
<td>16,000,000</td>
<td>38400</td>
<td>416</td>
<td>6</td>
<td>0</td>
<td>-0.2</td>
<td>0</td>
<td>0.2</td>
</tr>
<tr>
<td>16,000,000</td>
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<td>0</td>
<td>-0.3</td>
<td>0.1</td>
<td>0.5</td>
</tr>
<tr>
<td>16,000,000</td>
<td>115200</td>
<td>138</td>
<td>7</td>
<td>0</td>
<td>-0.7</td>
<td>0</td>
<td>0.8</td>
</tr>
<tr>
<td>16,000,000</td>
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<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>16,000,000</td>
<td>256000</td>
<td>62</td>
<td>4</td>
<td>0</td>
<td>-0.8</td>
<td>0</td>
<td>-1.2</td>
</tr>
</tbody>
</table>

### Table 15-5. Commonly Used Baud Rates, Settings, and Errors, UCOS16 = 1

<table>
<thead>
<tr>
<th>BRCLK Frequency [Hz]</th>
<th>Baud Rate [Baud]</th>
<th>UCBRx</th>
<th>UCBRx</th>
<th>UCBRx</th>
<th>UCBRSx</th>
<th>UCBRFx</th>
<th>Maximum TX Error [%]</th>
<th>Maximum RX Error [%]</th>
</tr>
</thead>
<tbody>
<tr>
<td>1,048,576</td>
<td>9600</td>
<td>6</td>
<td>0</td>
<td>13</td>
<td>-2.3</td>
<td>0</td>
<td>-2.2</td>
<td>0.8</td>
</tr>
<tr>
<td>1,048,576</td>
<td>19200</td>
<td>3</td>
<td>1</td>
<td>6</td>
<td>-4.6</td>
<td>3.2</td>
<td>-5.0</td>
<td>4.7</td>
</tr>
<tr>
<td>1,000,000</td>
<td>9600</td>
<td>6</td>
<td>0</td>
<td>8</td>
<td>-1.8</td>
<td>0</td>
<td>-2.2</td>
<td>0.4</td>
</tr>
<tr>
<td>1,000,000</td>
<td>19200</td>
<td>3</td>
<td>0</td>
<td>4</td>
<td>-1.8</td>
<td>0</td>
<td>-2.6</td>
<td>0.9</td>
</tr>
<tr>
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<td>7</td>
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<td>-34.4</td>
<td>0</td>
<td>-33.4</td>
<td>0</td>
</tr>
<tr>
<td>4,000,000</td>
<td>9600</td>
<td>26</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>0.9</td>
<td>0</td>
<td>1.1</td>
</tr>
<tr>
<td>4,000,000</td>
<td>19200</td>
<td>13</td>
<td>0</td>
<td>0</td>
<td>-1.8</td>
<td>0</td>
<td>-1.9</td>
<td>0.2</td>
</tr>
<tr>
<td>4,000,000</td>
<td>38400</td>
<td>6</td>
<td>0</td>
<td>8</td>
<td>-1.8</td>
<td>0</td>
<td>-2.2</td>
<td>0.4</td>
</tr>
<tr>
<td>4,000,000</td>
<td>57600</td>
<td>4</td>
<td>5</td>
<td>3</td>
<td>-3.5</td>
<td>3.2</td>
<td>-1.8</td>
<td>6.4</td>
</tr>
<tr>
<td>4,000,000</td>
<td>115200</td>
<td>2</td>
<td>3</td>
<td>2</td>
<td>-2.1</td>
<td>4.8</td>
<td>-2.5</td>
<td>7.3</td>
</tr>
<tr>
<td>4,000,000</td>
<td>230400</td>
<td>1</td>
<td>7</td>
<td>0</td>
<td>-34.4</td>
<td>0</td>
<td>-33.4</td>
<td>0</td>
</tr>
<tr>
<td>8,000,000</td>
<td>9600</td>
<td>52</td>
<td>0</td>
<td>1</td>
<td>-0.4</td>
<td>0</td>
<td>-0.4</td>
<td>0.1</td>
</tr>
<tr>
<td>8,000,000</td>
<td>19200</td>
<td>26</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>0.9</td>
<td>0</td>
<td>1.1</td>
</tr>
<tr>
<td>8,000,000</td>
<td>38400</td>
<td>13</td>
<td>0</td>
<td>0</td>
<td>-1.8</td>
<td>0</td>
<td>-1.9</td>
<td>0.2</td>
</tr>
<tr>
<td>8,000,000</td>
<td>57600</td>
<td>8</td>
<td>0</td>
<td>11</td>
<td>0.88</td>
<td>0</td>
<td>0</td>
<td>1.6</td>
</tr>
<tr>
<td>8,000,000</td>
<td>115200</td>
<td>4</td>
<td>5</td>
<td>3</td>
<td>-3.5</td>
<td>3.2</td>
<td>-1.8</td>
<td>6.4</td>
</tr>
<tr>
<td>8,000,000</td>
<td>230400</td>
<td>2</td>
<td>3</td>
<td>2</td>
<td>-2.1</td>
<td>4.8</td>
<td>-2.5</td>
<td>7.3</td>
</tr>
<tr>
<td>8,000,000</td>
<td>460800</td>
<td>1</td>
<td>7</td>
<td>0</td>
<td>-34.4</td>
<td>0</td>
<td>-33.4</td>
<td>0</td>
</tr>
<tr>
<td>12,000,000</td>
<td>9600</td>
<td>78</td>
<td>0</td>
<td>2</td>
<td>0</td>
<td>0.05</td>
<td>0</td>
<td>0.05</td>
</tr>
<tr>
<td>12,000,000</td>
<td>19200</td>
<td>39</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0.2</td>
</tr>
<tr>
<td>12,000,000</td>
<td>38400</td>
<td>19</td>
<td>0</td>
<td>8</td>
<td>-1.8</td>
<td>0</td>
<td>-1.8</td>
<td>0.1</td>
</tr>
<tr>
<td>12,000,000</td>
<td>57600</td>
<td>13</td>
<td>0</td>
<td>0</td>
<td>-1.8</td>
<td>0</td>
<td>-1.9</td>
<td>0.2</td>
</tr>
<tr>
<td>12,000,000</td>
<td>115200</td>
<td>6</td>
<td>0</td>
<td>8</td>
<td>-1.8</td>
<td>0</td>
<td>-2.2</td>
<td>0.4</td>
</tr>
<tr>
<td>12,000,000</td>
<td>230400</td>
<td>3</td>
<td>0</td>
<td>4</td>
<td>-1.8</td>
<td>0</td>
<td>-2.6</td>
<td>0.9</td>
</tr>
<tr>
<td>16,000,000</td>
<td>9600</td>
<td>104</td>
<td>0</td>
<td>3</td>
<td>0</td>
<td>0.2</td>
<td>0</td>
<td>0.3</td>
</tr>
<tr>
<td>16,000,000</td>
<td>19200</td>
<td>52</td>
<td>0</td>
<td>1</td>
<td>-0.4</td>
<td>0</td>
<td>-0.4</td>
<td>0.1</td>
</tr>
<tr>
<td>16,000,000</td>
<td>38400</td>
<td>26</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>0.9</td>
<td>0</td>
<td>1.1</td>
</tr>
</tbody>
</table>
Table 15-5. Commonly Used Baud Rates, Settings, and Errors, UCOS16 = 1 (continued)

<table>
<thead>
<tr>
<th>BRCLK Frequency [Hz]</th>
<th>Baud Rate [Baud]</th>
<th>UCBRx</th>
<th>UCBRxSx</th>
<th>UCBRFx</th>
<th>Maximum TX Error [%]</th>
<th>Maximum RX Error [%]</th>
</tr>
</thead>
<tbody>
<tr>
<td>16,000,000</td>
<td>57600</td>
<td>17</td>
<td>0</td>
<td>6</td>
<td>0.9</td>
<td>-0.1</td>
</tr>
<tr>
<td>16,000,000</td>
<td>115200</td>
<td>8</td>
<td>0</td>
<td>11</td>
<td>0.9</td>
<td>0</td>
</tr>
<tr>
<td>16,000,000</td>
<td>230400</td>
<td>4</td>
<td>5</td>
<td>3</td>
<td>-3.5</td>
<td>3.2</td>
</tr>
<tr>
<td>16,000,000</td>
<td>460800</td>
<td>2</td>
<td>3</td>
<td>2</td>
<td>-2.1</td>
<td>4.8</td>
</tr>
</tbody>
</table>

15.3.14 Using the USCI Module in UART Mode with Low Power Modes

The USCI module provides automatic clock activation for SMCLK for use with low-power modes. When SMCLK is the USCI clock source, and is inactive because the device is in a low-power mode, the USCI module automatically activates it when needed, regardless of the control-bit settings for the clock source. The clock remains active until the USCI module returns to its idle condition. After the USCI module returns to the idle condition, control of the clock source reverts to the settings of its control bits. Automatic clock activation is not provided for ACLK.

When the USCI module activates an inactive clock source, the clock source becomes active for the whole device and any peripheral configured to use the clock source may be affected. For example, a timer using SMCLK will increment while the USCI module forces SMCLK active.

15.3.15 USCI Interrupts

The USCI has one interrupt vector for transmission and one interrupt vector for reception.

15.3.15.1 USCI Transmit Interrupt Operation

The UCAxTXIFG interrupt flag is set by the transmitter to indicate that UCAxTXBUF is ready to accept another character. An interrupt request is generated if UCAxTXIE and GIE are also set. UCAxTXIFG is automatically reset if a character is written to UCAxTXBUF.

UCAxTXIFG is set after a PUC or when UCSWRST = 1. UCAxTXIE is reset after a PUC or when UCSWRST = 1.

15.3.15.2 USCI Receive Interrupt Operation

The UCAxRXIFG interrupt flag is set each time a character is received and loaded into UCAxRXBUF. An interrupt request is generated if UCAxRXIE and GIE are also set. UCAxRXIFG is automatically reset when UCAxRXBUF is read.

Additional interrupt control features include:
- When UCAxRXEIE = 0 erroneous characters will not set UCAxRXIFG.
- When UCDORM = 1, non-address characters will not set UCAxRXIFG in multiprocessor modes. In plain UART mode, no characters will set UCAxRXIFG.
- When UCBRKIE = 1 a break condition will set the UCBRK bit and the UCAxRXIFG flag.

15.3.15.3 USCI Interrupt Usage

USCI_Ax and USCI_Bx share the same interrupt vectors. The receive interrupt flags UCAxRXIFG and UCBxRXIFG are routed to one interrupt vector, the transmit interrupt flags UCAxTXIFG and UCBxTXIFG share another interrupt vector.

Example 15-1 shows an extract of an interrupt service routine to handle data receive interrupts from USCI_A0 in either UART or SPI mode and USCI_B0 in SPI mode.
Example 15-1. Shared Interrupt Vectors Software Example, Data Receive

```
USCIA0_RX_USCIB0_RX_ISR
  BIT.B #UCA0RXIFG, &IFG2 ; USCI_A0 Receive Interrupt?
  JNZ USCIA0_RX_ISR
USCIB0_RX_ISR?
  ; Read UCB0RXBUF (clears UCB0RXIFG)
  ...
  RETI
USCIA0_RX_ISR
  ; Read UCA0RXBUF (clears UCA0RXIFG)
  ...
  RETI
```

Example 15-2 shows an extract of an interrupt service routine to handle data transmit interrupts from USCI_A0 in either UART or SPI mode and USCI_B0 in SPI mode.

Example 15-2. Shared Interrupt Vectors Software Example, Data Transmit

```
USCIA0_TX_USCIB0_TX_ISR
  BIT.B #UCA0TXIFG, &IFG2 ; USCI_A0 Transmit Interrupt?
  JNZ USCIA0_TX_ISR
USCIB0_TX_ISR
  ; Write UCB0TXBUF (clears UCB0TXIFG)
  ...
  RETI
USCIA0_TX_ISR
  ; Write UCA0TXBUF (clears UCA0TXIFG)
  ...
  RETI
```
15.4 USCI Registers: UART Mode

Table 15-6 lists the memory-mapped registers for USCI_A0 and USCI_A1 in UART mode.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>60h</td>
<td>UCA0CTL0</td>
<td>USCI_A0 control 0</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.1</td>
</tr>
<tr>
<td>61h</td>
<td>UCA0CTL1</td>
<td>USCI_A0 control 1</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 15.4.2</td>
</tr>
<tr>
<td>62h</td>
<td>UCA0BR0</td>
<td>USCI_A0 baud-rate control 0</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.3</td>
</tr>
<tr>
<td>63h</td>
<td>UCA0BR1</td>
<td>USCI_A0 baud-rate control 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.3</td>
</tr>
<tr>
<td>64h</td>
<td>UCA0MCTL</td>
<td>USCI_A0 modulation control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.5</td>
</tr>
<tr>
<td>65h</td>
<td>UCA0STAT</td>
<td>USCI_A0 status</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.6</td>
</tr>
<tr>
<td>66h</td>
<td>UCA0RXBUF</td>
<td>USCI_A0 receive buffer</td>
<td>Read</td>
<td>00h with PUC</td>
<td>Section 15.4.7</td>
</tr>
<tr>
<td>67h</td>
<td>UCA0TXBUF</td>
<td>USCI_A0 transmit buffer</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.8</td>
</tr>
<tr>
<td>5Dh</td>
<td>UCA0ABCTL</td>
<td>USCI_A0 auto baud control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.11</td>
</tr>
<tr>
<td>5Eh</td>
<td>UCA0IRTCTL</td>
<td>USCI_A0 IrDA transmit control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.9</td>
</tr>
<tr>
<td>5Fh</td>
<td>UCA0IRRC</td>
<td>USCI_A0 IrDA receive control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.10</td>
</tr>
<tr>
<td>1h</td>
<td>IE2</td>
<td>SFR interrupt enable 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.12</td>
</tr>
<tr>
<td>3h</td>
<td>IFG2</td>
<td>SFR interrupt flag 2</td>
<td>Read/write</td>
<td>0Ah with PUC</td>
<td>Section 15.4.13</td>
</tr>
<tr>
<td>D0h</td>
<td>UCA1CTL0</td>
<td>USCI_A1 control 0</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.1</td>
</tr>
<tr>
<td>D1h</td>
<td>UCA1CTL1</td>
<td>USCI_A1 control 1</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 15.4.2</td>
</tr>
<tr>
<td>D2h</td>
<td>UCA1BR0</td>
<td>USCI_A1 baud-rate control 0</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.3</td>
</tr>
<tr>
<td>D3h</td>
<td>UCA1BR1</td>
<td>USCI_A1 baud-rate control 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.3</td>
</tr>
<tr>
<td>D4h</td>
<td>UCA1MCTL</td>
<td>USCI_A1 modulation control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.5</td>
</tr>
<tr>
<td>D5h</td>
<td>UCA1STAT</td>
<td>USCI_A1 status</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.6</td>
</tr>
<tr>
<td>D6h</td>
<td>UCA1RXBUF</td>
<td>USCI_A1 receive buffer</td>
<td>Read</td>
<td>00h with PUC</td>
<td>Section 15.4.7</td>
</tr>
<tr>
<td>D7h</td>
<td>UCA1TXBUF</td>
<td>USCI_A1 transmit buffer</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.8</td>
</tr>
<tr>
<td>C0h</td>
<td>UCA1ABCTL</td>
<td>USCI_A1 auto baud control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.11</td>
</tr>
<tr>
<td>CEh</td>
<td>UCA1IRTCTL</td>
<td>USCI_A1 IrDA transmit control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.9</td>
</tr>
<tr>
<td>CFh</td>
<td>UCA1IRRTCTL</td>
<td>USCI_A1 IrDA receive control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.10</td>
</tr>
<tr>
<td>6h</td>
<td>UC1IE</td>
<td>USCI_A1/B1 interrupt enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 15.4.14</td>
</tr>
<tr>
<td>7h</td>
<td>UC1IFG</td>
<td>USCI_A1/B1 interrupt flag</td>
<td>Read/write</td>
<td>0Ah with PUC</td>
<td>Section 15.4.15</td>
</tr>
</tbody>
</table>

**Note**

Modifying SFR bits

To avoid modifying control bits of other modules, TI recommends setting or clearing the IEx and IFGx bits using BIS.B or BIC.B instructions, rather than MOV.B or CLR.B instructions.
15.4.1 UCAxCTL0 Register

USCI_Ax Control 0 Register

UCAxCTL0 is shown in Figure 15-12 and described in Table 15-7.

Return to Table 15-6.

**Figure 15-12. UCAxCTL0 Register**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>UCPEN</td>
<td>R/W</td>
<td>0h</td>
<td>Parity enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Parity disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Parity enabled. Parity bit is generated (UCAxTXD) and</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>expected (UCAxRXD). In address-bit multiprocessor mode, the</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>address bit is included in the parity calculation.</td>
</tr>
<tr>
<td>6</td>
<td>UCPAR</td>
<td>R/W</td>
<td>0h</td>
<td>Parity select. UCPAR is not used when parity is disabled.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Odd parity</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Even parity</td>
</tr>
<tr>
<td>5</td>
<td>UCMSB</td>
<td>R/W</td>
<td>0h</td>
<td>MSB first select. Controls the direction of the receive and transmit</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>shift register.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = LSB first</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = MSB first</td>
</tr>
<tr>
<td>4</td>
<td>UC7BIT</td>
<td>R/W</td>
<td>0h</td>
<td>Character length. Selects 7-bit or 8-bit character length.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = 8-bit data</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = 7-bit data</td>
</tr>
<tr>
<td>3</td>
<td>UCSPB</td>
<td>R/W</td>
<td>0h</td>
<td>Stop bit select. Number of stop bits.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = One stop bit</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Two stop bits</td>
</tr>
<tr>
<td>2-1</td>
<td>UCMODEx</td>
<td>R/W</td>
<td>0h</td>
<td>USCI mode. The UCMODEx bits select the asynchronous mode when</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>UCSYNC = 0.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = UART mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = Idle-line multiprocessor mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = Address-bit multiprocessor mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = UART mode with automatic baud rate detection</td>
</tr>
<tr>
<td>0</td>
<td>UCSYNC</td>
<td>R/W</td>
<td>0h</td>
<td>Synchronous mode enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Asynchronous mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Synchronous mode</td>
</tr>
</tbody>
</table>
15.4.2 UC\(\text{Ax}\)CTL1 Register

USCI_\(\text{Ax}\) Control 1 Register

UC\(\text{Ax}\)CTL1 is shown in Figure 15-13 and described in Table 15-8.

Return to Table 15-6.

Figure 15-13. UC\(\text{Ax}\)CTL1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td>UCSSELx</td>
<td>R/W</td>
<td>0h</td>
<td>USCI clock source select. These bits select the BRCLK source clock. 00b = UCLK 01b = ACLK 10b = SMCLK 11b = SMCLK</td>
</tr>
<tr>
<td>5</td>
<td>UCRXEIE</td>
<td>R/W</td>
<td>0h</td>
<td>Receive erroneous-character interrupt-enable 0b = Erroneous characters rejected and UC(\text{Ax})RXIFG is not set 1b = Erroneous characters received will set UC(\text{Ax})RXIFG</td>
</tr>
<tr>
<td>4</td>
<td>UCBRKIE</td>
<td>R/W</td>
<td>0h</td>
<td>Receive break character interrupt-enable 0b = Received break characters do not set UC(\text{Ax})RXIFG. 1b = Received break characters set UC(\text{Ax})RXIFG.</td>
</tr>
<tr>
<td>3</td>
<td>UCDORM</td>
<td>R/W</td>
<td>0h</td>
<td>Dormant. Puts USCI into sleep mode. 0b = Not dormant. All received characters will set UC(\text{Ax})RXIFG. 1b = Dormant. Only characters that are preceded by an idle-line or with address bit set will set UC(\text{Ax})RXIFG. In UART mode with automatic baud rate detection only the combination of a break and synch field will set UC(\text{Ax})RXIFG.</td>
</tr>
<tr>
<td>2</td>
<td>UCTXADDR</td>
<td>R/W</td>
<td>0h</td>
<td>Transmit address. Next frame to be transmitted will be marked as address depending on the selected multiprocessor mode. 0b = Next frame transmitted is data 1b = Next frame transmitted is an address</td>
</tr>
<tr>
<td>1</td>
<td>UCTXBRK</td>
<td>R/W</td>
<td>0h</td>
<td>Transmit break. Transmits a break with the next write to the transmit buffer. In UART mode with automatic baud rate detection 055h must be written into UC(\text{Ax})TXBUF to generate the required break/synch fields. Otherwise 0h must be written into the transmit buffer. 0b = Next frame transmitted is not a break 1b = Next frame transmitted is a break or a break/synch</td>
</tr>
<tr>
<td>0</td>
<td>UCSWRST</td>
<td>R/W</td>
<td>1h</td>
<td>Software reset enable 0b = Disabled. USCI reset released for operation. 1b = Enabled. USCI logic held in reset state.</td>
</tr>
</tbody>
</table>
15.4.3 UCAxBR0 Register
USCI_Ax Baud-Rate Control 0 Register
UCAxBR0 is shown in Figure 15-14 and described in Table 15-9.
Return to Table 15-6.

Figure 15-14. UCAxBR0 Register

<table>
<thead>
<tr>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>UCBR0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

| rw-0 | rw-0 | rw-0 | rw-0 | rw-0 | rw-0 | rw-0 | rw-0 |

Table 15-9. UCAxBR0 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UCBR0</td>
<td>R/W</td>
<td>0h</td>
<td>Clock prescaler setting of the baud-rate generator. The 16-bit value of (UCAxBR0 + UCAxBR1 × 256) forms the prescaler value.</td>
</tr>
</tbody>
</table>

15.4.4 UCAxBR1 Register
USCI_Ax Baud-Rate Control 1 Register
UCAxBR1 is shown in Figure 15-14 and described in Table 15-9.
Return to Table 15-6.

Figure 15-15. UCAxBR1 Register

<table>
<thead>
<tr>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>UCBR1</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

| rw-0 | rw-0 | rw-0 | rw-0 | rw-0 | rw-0 | rw-0 | rw-0 |

Table 15-10. UCAxBR1 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UCBR1</td>
<td>R/W</td>
<td>0h</td>
<td>Clock prescaler setting of the baud-rate generator. The 16-bit value of (UCAxBR0 + UCAxBR1 × 256) forms the prescaler value.</td>
</tr>
</tbody>
</table>
15.4.5 UCAxMCTL Register

USCI_Ax Modulation Control Register

UCAxMCTL is shown in Figure 15-16 and described in Table 15-11.

Return to Table 15-6.

Figure 15-16. UCAxMCTL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td>UCBRFx</td>
<td>R/W</td>
<td>0h</td>
<td>First modulation stage select. These bits determine the modulation pattern for BITCLK16 when UCOS16 = 1. Ignored with UCOS16 = 0. Table 15-3 shows the modulation pattern.</td>
</tr>
<tr>
<td>3-1</td>
<td>UCBRSx</td>
<td>R/W</td>
<td>0h</td>
<td>Second modulation stage select. These bits determine the modulation pattern for BITCLK. Table 15-2 shows the modulation pattern.</td>
</tr>
<tr>
<td>0</td>
<td>UCOS16</td>
<td>R/W</td>
<td>0h</td>
<td>Oversampling mode enabled</td>
</tr>
</tbody>
</table>

Table 15-11. UCAxMCTL Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td>UCBRFx</td>
<td>R/W</td>
<td>0h</td>
<td>First modulation stage select. These bits determine the modulation pattern for BITCLK16 when UCOS16 = 1. Ignored with UCOS16 = 0. Table 15-3 shows the modulation pattern.</td>
</tr>
<tr>
<td>3-1</td>
<td>UCBRSx</td>
<td>R/W</td>
<td>0h</td>
<td>Second modulation stage select. These bits determine the modulation pattern for BITCLK. Table 15-2 shows the modulation pattern.</td>
</tr>
<tr>
<td>0</td>
<td>UCOS16</td>
<td>R/W</td>
<td>0h</td>
<td>Oversampling mode enabled</td>
</tr>
</tbody>
</table>

0b = Disabled
| 1b = Enabled |
15.4.6 UCAxSTAT Register

USCI_Ax Status Register

UCAxSTAT is shown in Figure 15-17 and described in Table 15-12.

Return to Table 15-6.

Figure 15-17. UCAxSTAT Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>UCLISTEN</td>
<td>R/W</td>
<td>0h</td>
<td>Listen enable. The UCLISTEN bit selects loopback mode.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Disabled. UCAxTXD is internally fed back to the receiver.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Enabled. UCAxTXD is internally fed back to the receiver.</td>
</tr>
<tr>
<td>6</td>
<td>UCFE</td>
<td>R/W</td>
<td>0h</td>
<td>Framing error flag</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No error</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Character received with low stop bit</td>
</tr>
<tr>
<td>5</td>
<td>UCOE</td>
<td>R/W</td>
<td>0h</td>
<td>Overrun error flag</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No error</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Overrun error occurred</td>
</tr>
<tr>
<td>4</td>
<td>UCPE</td>
<td>R/W</td>
<td>0h</td>
<td>Parity error flag</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No error</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Character received with parity error</td>
</tr>
<tr>
<td>3</td>
<td>UCBRK</td>
<td>R/W</td>
<td>0h</td>
<td>Break detect flag</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No break condition</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Break condition occurred</td>
</tr>
<tr>
<td>2</td>
<td>UCRXERR</td>
<td>R/W</td>
<td>0h</td>
<td>Receive error flag</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No receive errors detected</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Receive error detected</td>
</tr>
<tr>
<td>1</td>
<td>UCADDR</td>
<td>R/W</td>
<td>0h</td>
<td>UCADDR: Address received in address-bit multiprocessor mode.</td>
</tr>
<tr>
<td></td>
<td>UCIDLE</td>
<td></td>
<td></td>
<td>0b = Received character is data</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Received character is an address</td>
</tr>
<tr>
<td></td>
<td>UCIDLE</td>
<td></td>
<td></td>
<td>UCIDLE: Idle line detected in idle-line multiprocessor mode.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No idle line detected</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Idle line detected</td>
</tr>
<tr>
<td>0</td>
<td>UCBUSY</td>
<td>R</td>
<td>0h</td>
<td>USCI busy. This bit indicates if a transmit or receive operation is in</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>progress.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = USCI inactive</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = USCI transmitting or receiving</td>
</tr>
</tbody>
</table>
15.4.7 UCAxRXBUF Register
USCI_Ax Receive Buffer Register
UCAxRXBUF is shown in Figure 15-18 and described in Table 15-13.
Return to Table 15-6.

Figure 15-18. UCAxRXBUF Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>0-7</td>
<td>UCRXBUFx</td>
<td>0h</td>
<td></td>
<td>The receive-data buffer is user accessible and contains the last received character from the receive shift register. Reading UCAxRXBUF resets the receive-error bits, the UCADDR or UCIDLE bit, and UCAxRXIFG. In 7-bit data mode, UCAxRXBUF is LSB justified and the MSB is always reset.</td>
</tr>
</tbody>
</table>

15.4.8 UCAxTXBUF Register
USCI_Ax Transmit Buffer Register
UCAxTXBUF is shown in Figure 15-19 and described in Table 15-14.
Return to Table 15-6.

Figure 15-19. UCAxTXBUF Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>0-7</td>
<td>UCTXBUFx</td>
<td>0h</td>
<td></td>
<td>The transmit data buffer is user accessible and holds the data waiting to be moved into the transmit shift register and transmitted on UCAxTXD. Writing to the transmit data buffer clears UCAxTXIFG. The MSB of UCAxTXBUF is not used for 7-bit data and is reset.</td>
</tr>
</tbody>
</table>
15.4.9 UCAXIRTCTL Register
USCI_Ax IrDA Transmit Control Register
UCAXIRTCTL is shown in Figure 15-20 and described in Table 15-15.
Return to Table 15-6.

Figure 15-20. UCAXIRTCTL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-2</td>
<td>UCIRTXPLx</td>
<td>R/W</td>
<td>0h</td>
<td>Transmit pulse length. Pulse length ( t_{PULSE} = (UCIRTXPLx + 1) / (2 \times f_{RTXCLK}) )</td>
</tr>
<tr>
<td>1</td>
<td>UCIRTXCLK</td>
<td>R/W</td>
<td>0h</td>
<td>IrDA transmit pulse clock select ( 0b = BRCLK ) ( 1b = BITCLK16 ) when UCOS16 = 1. Otherwise, BRCLK.</td>
</tr>
<tr>
<td>0</td>
<td>UCIREN</td>
<td>R/W</td>
<td>0h</td>
<td>IrDA encoder/decoder enable ( 0b = ) IrDA encoder/decoder disabled ( 1b = ) IrDA encoder/decoder enabled</td>
</tr>
</tbody>
</table>

15.4.10 UCAXIRRCTL Register
USCI_Ax IrDA Receive Control Register
UCAXIRRCTL is shown in Figure 15-21 and described in Table 15-16.
Return to Table 15-6.

Figure 15-21. UCAXIRRCTL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-2</td>
<td>UCIRRXFLx</td>
<td>R/W</td>
<td>0h</td>
<td>Receive filter length. The minimum pulse length for receive is given by: ( t_{MIN} = (UCIRRXFLx + 4) / (2 \times f_{RCLK}) )</td>
</tr>
<tr>
<td>1</td>
<td>UCIRRXPL</td>
<td>R/W</td>
<td>0h</td>
<td>IrDA receive input UCAXRXD polarity ( 0b = ) IrDA transceiver delivers a high pulse when a light pulse is seen ( 1b = ) IrDA transceiver delivers a low pulse when a light pulse is seen</td>
</tr>
<tr>
<td>0</td>
<td>UCIRRXFE</td>
<td>R/W</td>
<td>0h</td>
<td>IrDA receive filter enabled ( 0b = ) Receive filter disabled ( 1b = ) Receive filter enabled</td>
</tr>
</tbody>
</table>
15.4.11 UCxABCTL Register

USCI_Ax Auto Baud Control Register

UCxABCTL is shown in Figure 15-22 and described in Table 15-17.

Return to Table 15-6.

Figure 15-22. UCxABCTL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td></td>
</tr>
<tr>
<td>5-4</td>
<td>UCDELIMx</td>
<td>R/W</td>
<td>0h</td>
<td>Break and synch delimiter length</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = 1 bit time</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = 2 bit times</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = 3 bit times</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = 4 bit times</td>
</tr>
<tr>
<td>3</td>
<td>UCSTOE</td>
<td>R/W</td>
<td>0h</td>
<td>Synch field time-out error</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No error</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Length of synch field exceeded measurable time.</td>
</tr>
<tr>
<td>2</td>
<td>UCBTOE</td>
<td>R/W</td>
<td>0h</td>
<td>Break time-out error</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No error</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Length of break field exceeded 22 bit times.</td>
</tr>
<tr>
<td>1</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td></td>
</tr>
<tr>
<td>0</td>
<td>UCABDEN</td>
<td>R/W</td>
<td>0h</td>
<td>Automatic baud rate detect enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Baud rate detection disabled. Length of break and synch field is</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>not measured.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Baud rate detection enabled. Length of break and synch field is</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>measured and baud rate settings are changed accordingly.</td>
</tr>
</tbody>
</table>
15.4.12 IE2 Register

SFR Interrupt Enable 2 Register

IE2 is shown in Figure 15-23 and described in Table 15-18.

Return to Table 15-6.

Figure 15-23. IE2 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-2</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules (see the device-specific data sheet).</td>
</tr>
<tr>
<td>1</td>
<td>UCA0TXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_A0 transmit interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt enabled</td>
</tr>
<tr>
<td>0</td>
<td>UCA0RXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_A0 receive interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt enabled</td>
</tr>
</tbody>
</table>

15.4.13 IFG2 Register

SFR Interrupt Flag 2 Register

IFG2 is shown in Figure 15-24 and described in Table 15-19.

Return to Table 15-6.

Figure 15-24. IFG2 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-2</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules (see the device-specific data sheet).</td>
</tr>
<tr>
<td>1</td>
<td>UCA0TXIFG</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_A0 transmit interrupt flag. UCA0TXIFG is set when UCA0TXBUF is empty.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>0</td>
<td>UCA0RXIFG</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_A0 receive interrupt flag. UCA0RXIFG is set when UCA0RXBUF has received a complete character.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
</tbody>
</table>


## 15.4.14 UC1IE Register

**USCI Interrupt Enable Register**

UC1IE is shown in **Figure 15-25** and described in **Table 15-20**.

Return to **Table 15-6**.

### Table 15-20. UC1IE Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td>Unused</td>
</tr>
<tr>
<td>3-2</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other USCI modules (see the device-specific data sheet).</td>
</tr>
<tr>
<td>1</td>
<td>UCA1TXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_A1 transmit interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt enabled</td>
</tr>
<tr>
<td>0</td>
<td>UCA1RXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_A1 receive interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b =Interrupt enabled</td>
</tr>
</tbody>
</table>

## 15.4.15 UC1IFG Register

**USCI Interrupt Flag Register**

UC1IFG is shown in **Figure 15-26** and described in **Table 15-21**.

Return to **Table 15-6**.

### Table 15-21. UC1IFG Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td>Unused</td>
</tr>
<tr>
<td>3-2</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other USCI modules (see the device-specific data sheet).</td>
</tr>
<tr>
<td>1</td>
<td>UCA1TXIFG</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_A1 transmit interrupt flag. UCA1TXIFG is set when UCA1TXBUF is empty.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>0</td>
<td>UCA1RXIFG</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_A1 receive interrupt flag. UCA1RXIFG is set when UCA1RXBUF has received a complete character.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
</tbody>
</table>
The universal serial communication interface (USCI) supports multiple serial communication modes with one hardware module. This chapter discusses the operation of the synchronous peripheral interface or SPI mode.

<table>
<thead>
<tr>
<th>Section</th>
<th>Title</th>
<th>Page</th>
</tr>
</thead>
<tbody>
<tr>
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<td>USCI Overview</td>
<td>464</td>
</tr>
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<td>16.2</td>
<td>USCI Introduction: SPI Mode</td>
<td>464</td>
</tr>
<tr>
<td>16.3</td>
<td>USCI Operation: SPI Mode</td>
<td>466</td>
</tr>
<tr>
<td>16.4</td>
<td>USCI Registers: SPI Mode</td>
<td>472</td>
</tr>
</tbody>
</table>
16.1 USCI Overview

The universal serial communication interface (USCI) modules support multiple serial communication modes. Different USCI modules support different modes. Each different USCI module is named with a different letter (for example, USCI_A is different from USCI_B). If more than one identical USCI module is implemented on one device, those modules are named with incrementing numbers. For example, if one device has two USCI_A modules, they are named USCI_A0 and USCI_A1. See the device-specific data sheet to determine which USCI modules, if any, are implemented on each device.

The USCI_Ax modules support:

- UART mode
- Pulse shaping for IrDA communications
- Automatic baud rate detection for LIN communications
- SPI mode

The USCI_Bx modules support:

- \( \text{i}^2\text{C} \) mode
- SPI mode

16.2 USCI Introduction: SPI Mode

In synchronous mode, the USCI connects the MSP430 to an external system via three or four pins: UCxSIMO, UCxSOMI, UCxCLK, and UCxSTE. SPI mode is selected when the UCSYNC bit is set and SPI mode (3-pin or 4-pin) is selected with the UCMODEx bits.

SPI mode features include:

- 7- or 8-bit data length
- LSB-first or MSB-first data transmit and receive
- 3-pin and 4-pin SPI operation
- Master or slave modes
- Independent transmit and receive shift registers
- Separate transmit and receive buffer registers
- Continuous transmit and receive operation
- Selectable clock polarity and phase control
- Programmable clock frequency in master mode
- Independent interrupt capability for receive and transmit
- Slave operation in LPM4

Figure 16-1 shows the USCI when configured for SPI mode.
Figure 16-1. USCI Block Diagram: SPI Mode
16.3 USCI Operation: SPI Mode

In SPI mode, serial data is transmitted and received by multiple devices using a shared clock provided by the master. An additional pin, UCxSTE, is provided to enable a device to receive and transmit data and is controlled by the master.

Three or four signals are used for SPI data exchange:

- **UCxSIMO**: Slave in, master out
  - Master mode: UCxSIMO is the data output line.
  - Slave mode: UCxSIMO is the data input line.

- **UCxSOMI**: Slave out, master in
  - Master mode: UCxSOMI is the data input line.
  - Slave mode: UCxSOMI is the data output line.

- **UCxCLK**: USCI SPI clock
  - Master mode: UCxCLK is an output.
  - Slave mode: UCxCLK is an input.

- **UCxSTE**: Slave transmit enable
  Used in 4-pin mode to allow multiple masters on a single bus. Not used in 3-pin mode. Table 16-1 describes the UCxSTE operation.

<table>
<thead>
<tr>
<th>UCMODEx</th>
<th>UCxSTE Active State</th>
<th>UCxSTE</th>
<th>Slave</th>
<th>Master</th>
</tr>
</thead>
<tbody>
<tr>
<td>01</td>
<td>High</td>
<td>0</td>
<td>Inactive</td>
<td>Active</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1</td>
<td>Active</td>
<td>Inactive</td>
</tr>
<tr>
<td>10</td>
<td>Low</td>
<td>0</td>
<td>Active</td>
<td>Inactive</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1</td>
<td>Inactive</td>
<td>Active</td>
</tr>
</tbody>
</table>

16.3.1 USCI Initialization and Reset

The USCI is reset by a PUC or by the UCSWRST bit. After a PUC, the UCSWRST bit is automatically set, keeping the USCI in a reset condition. When set, the UCSWRST bit resets the UCxRXIE, UCxTXIE, UCxRXIFG, UCOE, and UCFE bits and sets the UCxTXIFG flag. Clearing UCSWRST releases the USCI for operation.

**Note**

**Initializing or Re-Configuring the USCI Module**

The recommended USCI initialization/re-configuration process is:

1. Set UCSWRST (BIS.B #UCSWRST,&UCxCTL1)
2. Initialize all USCI registers with UCSWRST=1 (including UCxCTL1)
3. Configure ports
4. Clear UCSWRST via software (BIC.B #UCSWRST,&UCxCTL1)
5. Enable interrupts (optional) via UCxRXIE and/or UCxTXIE
16.3.2 Character Format

The USCI module in SPI mode supports 7-bit and 8-bit character lengths selected by the UC7BIT bit. In 7-bit data mode, UCxRXBUF is LSB justified and the MSB is always reset. The UCMSB bit controls the direction of the transfer and selects LSB or MSB first.

---

**Note**

**Default Character Format**

The default SPI character transmission is LSB first. For communication with other SPI interfaces it MSB-first mode may be required.

---

**Note**

**Character Format for Figures**

Figures throughout this chapter use MSB first format.

---

16.3.3 Master Mode

**Figure 16-2** shows the USCI as a master in both 3-pin and 4-pin configurations. The USCI initiates data transfer when data is moved to the transmit data buffer UCxTXBUF. The UCxTXBUF data is moved to the TX shift register when the TX shift register is empty, initiating data transfer on UCxSIMO starting with either the most-significant or least-significant bit depending on the UCMSB setting. Data on UCxSOMI is shifted into the receive shift register on the opposite clock edge. When the character is received, the receive data is moved from the RX shift register to the received data buffer UCxRXBUF and the receive interrupt flag, UCxRXIFG, is set, indicating the RX/TX operation is complete.

---

**Figure 16-2. USCI Master and External Slave**

A set transmit interrupt flag, UCxTXIFG, indicates that data has moved from UCxTXBUF to the TX shift register and UCxTXBUF is ready for new data. It does not indicate RX/TX completion.

To receive data into the USCI in master mode, data must be written to UCxTXBUF because receive and transmit operations operate concurrently.
### 16.3.3.1 Four-Pin SPI Master Mode

In 4-pin master mode, UCxSTE is used to prevent conflicts with another master and controls the master as described in Table 16-1. When UCxSTE is in the master-inactive state:

- UCxSIMO and UCxCLK are set to inputs and no longer drive the bus
- The error bit UCFE is set indicating a communication integrity violation to be handled by the user.
- The internal state machines are reset and the shift operation is aborted.

If data is written into UCxTXBUF while the master is held inactive by UCxSTE, it will be transmitted as soon as UCxSTE transitions to the master-active state. If an active transfer is aborted by UCxSTE transitioning to the master-inactive state, the data must be re-written into UCxTXBUF to be transferred when UCxSTE transitions back to the master-active state. The UCxSTE input signal is not used in 3-pin master mode.

### 16.3.4 Slave Mode

Figure 16-3 shows the USCI as a slave in both 3-pin and 4-pin configurations. UCxCLK is used as the input for the SPI clock and must be supplied by the external master. The data-transfer rate is determined by this clock and not by the internal bit clock generator. Data written to UCxTXBUF and moved to the TX shift register before the start of UCxCLK is transmitted on UCxSOMI. Data on UCxSIMO is shifted into the receive shift register on the opposite edge of UCxCLK and moved to UCxRXBUF when the set number of bits are received. When data is moved from the RX shift register to UCxRXBUF, the UCxRXIFG interrupt flag is set, indicating that data has been received. The overrun error bit, UCOE, is set when the previously received data is not read from UCxRXBUF before new data is moved to UCxRXBUF.

### 16.3.4.1 Four-Pin SPI Slave Mode

In 4-pin slave mode, UCxSTE is used by the slave to enable the transmit and receive operations and is provided by the SPI master. When UCxSTE is in the slave-active state, the slave operates normally. When UCxSTE is in the slave-inactive state:

- Any receive operation in progress on UCxSIMO is halted
- UCxSOMI is set to the input direction
- The shift operation is halted until the UCxSTE line transitions into the slave transmit active state.

The UCxSTE input signal is not used in 3-pin slave mode.

---

**Figure 16-3. USCI Slave and External Master**

- **Master**
  - SPI Receive Buffer
  - Data Shift Register DSR
  - COMMON SPI
  - SCLK

- **Slave**
  - Transmit Buffer UCxTXBUF
  - Receive Buffer UCxRXBUF
  - Transmit Shift Register
  - Receive Shift Register
  - UCxSIMO
  - UCxSOMI
  - UCxCLK
  - Px.x
  - STE
  - SS
  - Port.x

- **Diagram Description**
  - UCxSTE
  - UCxSOMI
  - UCxTXBUF
  - UCxRXBUF
  - MSP430 USCI

---

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16.3.5 SPI Enable

When the USCI module is enabled by clearing the UCSWRST bit it is ready to receive and transmit. In master mode the bit clock generator is ready, but is not clocked nor producing any clocks. In slave mode the bit clock generator is disabled and the clock is provided by the master.

A transmit or receive operation is indicated by UCBUSY = 1.

A PUC or set UCSWRST bit disables the USCI immediately and any active transfer is terminated.

16.3.5.1 Transmit Enable

In master mode, writing to UCxTXBUF activates the bit clock generator and the data will begin to transmit.

In slave mode, transmission begins when a master provides a clock and, in 4-pin mode, when the UCxSTE is in the slave-active state.

16.3.5.2 Receive Enable

The SPI receives data when a transmission is active. Receive and transmit operations operate concurrently.

16.3.6 Serial Clock Control

UCxCLK is provided by the master on the SPI bus. When UCMST = 1, the bit clock is provided by the USCI bit clock generator on the UCxCLK pin. The clock used to generate the bit clock is selected with the UCSSELx bits. When UCMST = 0, the USCI clock is provided on the UCxCLK pin by the master, the bit clock generator is not used, and the UCSSELx bits are don’t care. The SPI receiver and transmitter operate in parallel and use the same clock source for data transfer.

The 16-bit value of UCBRx in the bit rate control registers UCxxBR1 and UCxxBR0 is the division factor of the USCI clock source, BRCLK. The maximum bit clock that can be generated in master mode is BRCLK. Modulation is not used in SPI mode and UCAxMCTL should be cleared when using SPI mode for USCI_A. The UCAxCLK/UCBxCLK frequency is given by:

\[ f_{\text{BitClock}} = \frac{f_{\text{BRCLK}}}{UCBRx} \]
16.3.6.1 Serial Clock Polarity and Phase

The polarity and phase of UCxCLK are independently configured via the UCCKPL and UCCKPH control bits of the USCI. Timing for each case is shown in Figure 16-4.

Figure 16-4. USCI SPI Timing with UCMSB = 1

16.3.7 Using the SPI Mode With Low-Power Modes

The USCI module provides automatic clock activation for SMCLK for use with low-power modes. When SMCLK is the USCI clock source, and is inactive because the device is in a low-power mode, the USCI module automatically activates it when needed, regardless of the control-bit settings for the clock source. The clock remains active until the USCI module returns to its idle condition. After the USCI module returns to the idle condition, control of the clock source reverts to the settings of its control bits. Automatic clock activation is not provided for ACLK.

When the USCI module activates an inactive clock source, the clock source becomes active for the whole device and any peripheral configured to use the clock source may be affected. For example, a timer using SMCLK increments while the USCI module forces SMCLK active.

In SPI slave mode, no internal clock source is required because the clock is provided by the external master. It is possible to operate the USCI in SPI slave mode while the device is in LPM4 and all clock sources are disabled. The receive or transmit interrupt can wake up the CPU from any low power mode.

16.3.8 SPI Interrupts

The USCI has one interrupt vector for transmission and one interrupt vector for reception.

16.3.8.1 SPI Transmit Interrupt Operation

The UCxTXIFG interrupt flag is set by the transmitter to indicate that UCxTXBUF is ready to accept another character. An interrupt request is generated if UCxTXIE and GIE are also set. UCxTXIFG is automatically reset if a character is written to UCxTXBUF. UCxTXIFG is set after a PUC or when UCSWRST = 1. UCxTXIE is reset after a PUC or when UCSWRST = 1.
16.3.8.2 SPI Receive Interrupt Operation

The UCxRXIFG interrupt flag is set each time a character is received and loaded into UCxRXBUF. An interrupt request is generated if UCxRXIE and GIE are also set. UCxRXIFG and UCxRXIE are reset by a system reset PUC signal or when UCSWRST = 1. UCxRXIFG is automatically reset when UCxRXBUF is read.

16.3.8.3 USCI Interrupt Usage

USCIAx and USCI_Bx share the same interrupt vectors. The receive interrupt flags UCAxRXIFG and UCBxRXIFG are routed to one interrupt vector, the transmit interrupt flags UCAxTXIFG and UCBxTXIFG share another interrupt vector.

Example 16-1 shows an extract of an interrupt service routine to handle data receive interrupts from USCI_A0 in either UART or SPI mode and USCI_B0 in SPI mode.

Example 16-1. Shared Receive Interrupt Vectors Software Example

Example 16-2. Shared Transmit Interrupt Vectors Software Example
## 16.4 USCI Registers: SPI Mode

Table 16-2 lists the memory-mapped registers for USCI_Ax and USCI_Bx in SPI mode.

### Table 16-2. USCI_Ax and USCI_Bx Control and Status Registers

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>60h</td>
<td>UCA0CTL0</td>
<td>USCI_A0 control 0</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.1</td>
</tr>
<tr>
<td>61h</td>
<td>UCA0CTL1</td>
<td>USCI_A0 control 1</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 16.4.2</td>
</tr>
<tr>
<td>62h</td>
<td>UCA0BR0</td>
<td>USCI_A0 baud-rate control 0</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.3</td>
</tr>
<tr>
<td>63h</td>
<td>UCA0BR1</td>
<td>USCI_A0 baud-rate control 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.3</td>
</tr>
<tr>
<td>65h</td>
<td>UCA0STAT</td>
<td>USCI_A0 status</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.5</td>
</tr>
<tr>
<td>66h</td>
<td>UCA0RXBUF</td>
<td>USCI_A0 receive buffer</td>
<td>Read</td>
<td>00h with PUC</td>
<td>Section 16.4.6</td>
</tr>
<tr>
<td>67h</td>
<td>UCA0TXBUF</td>
<td>USCI_A0 transmit buffer</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.7</td>
</tr>
<tr>
<td>68h</td>
<td>UCB0CTL0</td>
<td>USCI_B0 control 0</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 16.4.1</td>
</tr>
<tr>
<td>69h</td>
<td>UCB0CTL1</td>
<td>USCI_B0 control 1</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 16.4.1</td>
</tr>
<tr>
<td>6Ah</td>
<td>UCB0BR0</td>
<td>USCI_B0 bit-rate control 0</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.3</td>
</tr>
<tr>
<td>6Bh</td>
<td>UCB0BR1</td>
<td>USCI_B0 bit-rate control 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.3</td>
</tr>
<tr>
<td>6Dh</td>
<td>UCB0STAT</td>
<td>USCI_B0 status</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.5</td>
</tr>
<tr>
<td>6Eh</td>
<td>UCB0RXBUF</td>
<td>USCI_B0 receive buffer</td>
<td>Read</td>
<td>00h with PUC</td>
<td>Section 16.4.6</td>
</tr>
<tr>
<td>6Fh</td>
<td>UCB0TXBUF</td>
<td>USCI_B0 transmit buffer</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.7</td>
</tr>
<tr>
<td>1h</td>
<td>IE2</td>
<td>SFR interrupt enable 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.8</td>
</tr>
<tr>
<td>3h</td>
<td>IFG2</td>
<td>SFR interrupt flag 2</td>
<td>Read/write</td>
<td>0Ah with PUC</td>
<td>Section 16.4.9</td>
</tr>
<tr>
<td>D0h</td>
<td>UCA1CTL0</td>
<td>USCI_A1 control 0</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.1</td>
</tr>
<tr>
<td>D1h</td>
<td>UCA1CTL1</td>
<td>USCI_A1 control 1</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 16.4.1</td>
</tr>
<tr>
<td>D2h</td>
<td>UCA1BR0</td>
<td>USCI_A1 baud-rate control 0</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.3</td>
</tr>
<tr>
<td>D3h</td>
<td>UCA1BR1</td>
<td>USCI_A1 baud-rate control 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.4</td>
</tr>
<tr>
<td>D5h</td>
<td>UCA1STAT</td>
<td>USCI_A1 status</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.5</td>
</tr>
<tr>
<td>D6h</td>
<td>UCA1RXBUF</td>
<td>USCI_A1 receive buffer</td>
<td>Read</td>
<td>00h with PUC</td>
<td>Section 16.4.6</td>
</tr>
<tr>
<td>D7h</td>
<td>UCA1TXBUF</td>
<td>USCI_A1 transmit buffer</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.7</td>
</tr>
<tr>
<td>D8h</td>
<td>UCB1CTL0</td>
<td>USCI_B1 control 0</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 16.4.1</td>
</tr>
<tr>
<td>D9h</td>
<td>UCB1CTL1</td>
<td>USCI_B1 control 1</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 16.4.1</td>
</tr>
<tr>
<td>DAh</td>
<td>UCB1BR0</td>
<td>USCI_B1 bit-rate control 0</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.3</td>
</tr>
<tr>
<td>DBh</td>
<td>UCB1BR1</td>
<td>USCI_B1 bit-rate control 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.4</td>
</tr>
<tr>
<td>DDh</td>
<td>UCB1STAT</td>
<td>USCI_B1 status</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.5</td>
</tr>
<tr>
<td>DEh</td>
<td>UCB1RXBUF</td>
<td>USCI_B1 receive buffer</td>
<td>Read</td>
<td>00h with PUC</td>
<td>Section 16.4.6</td>
</tr>
<tr>
<td>DFh</td>
<td>UCB1TXBUF</td>
<td>USCI_B1 transmit buffer</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.7</td>
</tr>
<tr>
<td>6h</td>
<td>UC1IE</td>
<td>USCI_A1/B1 interrupt enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 16.4.10</td>
</tr>
<tr>
<td>7h</td>
<td>UC1IFG</td>
<td>USCI_A1/B1 interrupt flag</td>
<td>Read/write</td>
<td>0Ah with PUC</td>
<td>Section 16.4.11</td>
</tr>
</tbody>
</table>

### Note

**Modifying SFR bits**

To avoid modifying control bits of other modules, TI recommends setting or clearing the IEx and IFGx bits using BIS.B or BIC.B instructions, rather than MOV.B or CLR.B instructions.
16.4.1 UCAxCTL0, UCBxCTL0 Registers

USCI_Ax Control 0 Register, USCI_Bx Control 0 Register

UCAxCTL0 and UCBxCTL0 are shown in Figure 16-5 and described in Table 16-3.

Return to Table 16-2.

Figure 16-5. UCAxCTL0, UCBxCTL0 Registers

<table>
<thead>
<tr>
<th></th>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>UCKPH</td>
<td>UCCKPL</td>
<td>UCMSB</td>
<td>UC7BIT</td>
<td>UCMST</td>
<td>UCMODEEx</td>
<td>UCSYNC</td>
<td></td>
<td></td>
</tr>
<tr>
<td>rw-0</td>
<td>rw-0</td>
<td>rw-0</td>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
<td></td>
</tr>
</tbody>
</table>

Table 16-3. UCAxCTL0, UCBxCTL0 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>UCCKPH</td>
<td>R/W</td>
<td>0h</td>
<td>Clock phase select</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Data is changed on the first UCLK edge and captured on the following</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>edge. 1b = Data is captured on the first UCLK edge and changed on the</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>following edge.</td>
</tr>
<tr>
<td>6</td>
<td>UCCKPL</td>
<td>R/W</td>
<td>0h</td>
<td>Clock polarity select</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = The inactive state is low.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = The inactive state is high.</td>
</tr>
<tr>
<td>5</td>
<td>UCMSB</td>
<td>R/W</td>
<td>0h</td>
<td>MSB first select. Controls the direction of the receive and transmit</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>shift register. 0b = LSB first</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = MSB first</td>
</tr>
<tr>
<td>4</td>
<td>UC7BIT</td>
<td>R/W</td>
<td>0h</td>
<td>Character length. Selects 7-bit or 8-bit character length.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = 8-bit data</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = 7-bit data</td>
</tr>
<tr>
<td>3</td>
<td>UCMST</td>
<td>R/W</td>
<td>0h</td>
<td>Master mode select</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Slave mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = Master mode</td>
</tr>
<tr>
<td>2-1</td>
<td>UCMODEEx</td>
<td>R/W</td>
<td>0h</td>
<td>USCI mode. The UCMODEEx bits select the synchronous mode when UCSYNC = 1.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = 3-pin SPI</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = 4-pin SPI with UCxSTE active high: slave enabled when UCxSTE = 1</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = 4-pin SPI with UCxSTE active low: slave enabled when UCxSTE = 0</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = I2C mode</td>
</tr>
<tr>
<td>0</td>
<td>UCSYNC</td>
<td>R/W</td>
<td>0h</td>
<td>Synchronous mode enable. Must be 1 for SPI mode.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Asynchronous mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Synchronous mode</td>
</tr>
</tbody>
</table>
16.4.2 UCAxCTL1, UCBxCTL1 Registers

USCI_Ax Control 1 Register, USCI_Bx Control 1 Register

UCAxCTL1 and UCBxCTL1 are shown in Figure 16-6 and described in Table 16-4.

Return to Table 16-2.

Figure 16-6. UCAxCTL1, UCBxCTL1 Registers

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td>UCSSELx</td>
<td>R/W</td>
<td>0h</td>
<td>USCI clock source select. These bits select the BRCLK source clock in master mode. UCxCLK is always used in slave mode. 00b = Reserved 01b = ACLK 10b = SMCLK 11b = SMCLK</td>
</tr>
<tr>
<td>5-1</td>
<td>Reserved</td>
<td>R/W</td>
<td>0h</td>
<td></td>
</tr>
<tr>
<td>0</td>
<td>UCSWRST</td>
<td>R/W</td>
<td>1h</td>
<td>Software reset enable 0b = Disabled. USCI reset released for operation 1b = Enabled. USCI logic held in reset state.</td>
</tr>
</tbody>
</table>

Table 16-4. UCAxCTL1, UCBxCTL1 Register Field Descriptions

(1) UCAxCTL1 (USCI_Ax)
(2) UCBxCTL1 (USCI_Bx)
16.4.3 UCAxBR0, UCBxBR0 Registers
USCI Ax Bit-Rate Control 0 Register, USCI Bx Bit-Rate Control 0 Register
UCAxBRx and UCBxBRx are shown in Figure 16-7 and described in Table 16-5.
Return to Table 16-2.

Figure 16-7. UCAxBR0, UCBxBR0 Registers

<table>
<thead>
<tr>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>UCBRx (low byte)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
</tr>
</tbody>
</table>

Table 16-5. UCAxBR0, UCBxBR0 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UCBRx</td>
<td>R/W</td>
<td>0h</td>
<td>Bit clock prescaler setting. The 16-bit value of (UCxxBR0 + UCxxBR1 × 256) forms the prescaler value.</td>
</tr>
</tbody>
</table>

16.4.4 UCAxBR1, UCBxBR1 Registers
USCI Ax Bit-Rate Control 1 Register, USCI Bx Bit-Rate Control 1 Register
UCAxBRx and UCBxBRx are shown in Figure 16-8 and described in Table 16-6.
Return to Table 16-2.

Figure 16-8. UCAxBR1, UCBxBR1 Registers

<table>
<thead>
<tr>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>UCBRx (high byte)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
<td>nw-0</td>
</tr>
</tbody>
</table>

Table 16-6. UCAxBR1, UCBxBR1 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UCBRx</td>
<td>R/W</td>
<td>0h</td>
<td>Bit clock prescaler setting. The 16-bit value of (UCxxBR0 + UCxxBR1 × 256) forms the prescaler value.</td>
</tr>
</tbody>
</table>
16.4.5 UCAxSTAT, UCBxSTAT Registers

USCI_Ax Status Register, USCI_Bx Status Register

UCAxSTAT and UCBxSTAT are shown in Figure 16-9 and described in Table 16-7.

Return to Table 16-2.

Figure 16-9. UCAxSTAT, UCBxSTAT Registers

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>UCLISTEN</td>
<td>R/W</td>
<td>0h</td>
<td>Listen enable. The UCLISTEN bit selects loopback mode. 0b = Disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Enabled. The transmitter output is internally fed back to the receiver.</td>
</tr>
<tr>
<td>6</td>
<td>UCFE</td>
<td>R/W</td>
<td>0h</td>
<td>Framing error flag. This bit indicates a bus conflict in 4-wire master</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>mode. UCFE is not used in 3-wire master or any slave mode. 0b = No error</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Bus error occurred</td>
</tr>
<tr>
<td>5</td>
<td>UCOE</td>
<td>R/W</td>
<td>0h</td>
<td>Overrun error flag. This bit is set when a character is transferred</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>into UCxRXBUF before the previous character was read. UCOE is cleared</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>automatically when UCxRXBUF is read, and must not be cleared by software.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>Otherwise, it will not function correctly. 0b = No error 1b = Overrun</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>error occurred</td>
</tr>
<tr>
<td>4-1</td>
<td>Reserved</td>
<td>R/W</td>
<td>0h</td>
<td></td>
</tr>
<tr>
<td>0</td>
<td>UCBUSY</td>
<td>R</td>
<td>0h</td>
<td>USCI busy. This bit indicates if a transmit or receive operation is in</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>progress. 0b = USCI inactive 1b = USCI transmitting or receiving</td>
</tr>
</tbody>
</table>
16.4.6 UCAxRXBUF, UCBxRXBUF Registers
USCI_Ax Receive Buffer Register, USCI_Bx Receive Buffer Register
UCAxRXBUF and UCBxRXBUF are shown in Figure 16-10 and described in Table 16-8.
Return to Table 16-2.

Figure 16-10. UCAxRXBUF, UCBxRXBUF Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UCRXBUFx</td>
<td>R</td>
<td>0h</td>
<td>The receive-data buffer is user accessible and contains the last received character from the receive shift register. Reading UCRXBUF resets the receive-error bits, and UCxRXIFG. In 7-bit data mode, UCRXBUF is LSB justified and the MSB is always reset.</td>
</tr>
</tbody>
</table>

Table 16-8. UCAxRXBUF, UCBxRXBUF Register Field Descriptions

16.4.7 UCAxTXBUF, UCBxTXBUF Registers
USCI_Ax Transmit Buffer Register, USCI_Bx Transmit Buffer Register
UCAxTXBUF and UCBxTXBUF are shown in Figure 16-11 and described in Table 16-9.
Return to Table 16-2.

Figure 16-11. UCAxTXBUF, UCBxTXBUF Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UCTXBUFx</td>
<td>R/W</td>
<td>0h</td>
<td>The transmit data buffer is user accessible and holds the data waiting to be moved into the transmit shift register and transmitted. Writing to the transmit data buffer clears UCxTXIFG. The MSB of UCxTXBUF is not used for 7-bit data and is reset.</td>
</tr>
</tbody>
</table>
16.4.8 IE2 Register
SFR Interrupt Enable 2 Register
IE2 is shown in Figure 16-12 and described in Table 16-10.
Return to Table 16-2.

![Figure 16-12. IE2 Register](image-url)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules (see the device-specific data sheet).</td>
</tr>
<tr>
<td>3</td>
<td>UCB0TXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_B0 transmit interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt enabled</td>
</tr>
<tr>
<td>2</td>
<td>UCB0RXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_B0 receive interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt enabled</td>
</tr>
<tr>
<td>1</td>
<td>UCA0TXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_A0 transmit interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt enabled</td>
</tr>
<tr>
<td>0</td>
<td>UCA0RXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_A0 receive interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt enabled</td>
</tr>
</tbody>
</table>
16.4.9 IFG2 Register

SFR Interrupt Flag 2 Register

IFG2 is shown in Figure 16-13 and described in Table 16-11.

Return to Table 16-2.

Figure 16-13. IFG2 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules (see the device-specific data sheet).</td>
</tr>
<tr>
<td>3</td>
<td>UCB0TXIFG</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_B0 transmit interrupt flag. UCB0TXIFG is set when UCB0TXBUF is empty.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>2</td>
<td>UCB0RXIFG</td>
<td>R/W</td>
<td>1h</td>
<td>USCI_B0 receive interrupt flag. UCB0RXIFG is set when UCB0RXBUF has received a complete character.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>1</td>
<td>UCA0TXIFG</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_A0 transmit interrupt flag. UCA0TXIFG is set when UCA0TXBUF is empty.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>0</td>
<td>UCA0RXIFG</td>
<td>R/W</td>
<td>1h</td>
<td>USCI_A0 receive interrupt flag. UCA0RXIFG is set when UCA0RXBUF has received a complete character.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
</tbody>
</table>
16.4.10 UC1IE Register

USCI_A1/B1 Interrupt Enable Register

UC1IE is shown in Figure 16-14 and described in Table 16-12.

Return to Table 16-2.

![Figure 16-14. UC1IE Register](image)

Table 16-12. UC1IE Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td></td>
</tr>
<tr>
<td>3</td>
<td>UCB1TXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_B1 transmit interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = interrupt enabled</td>
</tr>
<tr>
<td>2</td>
<td>UCB1RXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_B1 receive interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = interrupt enabled</td>
</tr>
<tr>
<td>1</td>
<td>UCA1TXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_A1 transmit interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = interrupt enabled</td>
</tr>
<tr>
<td>0</td>
<td>UCA1RXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_A1 receive interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = interrupt enabled</td>
</tr>
</tbody>
</table>
16.4.11 UC1IFG Register

USCI_A1/B1 Interrupt Flag Register

UC1IFG is shown in Figure 16-15 and described in Table 16-13.

Return to Table 16-2.

**Figure 16-15. UC1IFG Register**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td></td>
</tr>
<tr>
<td>3</td>
<td>UCB1TXIFG</td>
<td>R/W</td>
<td>1h</td>
<td>USCI_B1 transmit interrupt flag. UCB1TXIFG is set when UCB1TXBUF is empty.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>2</td>
<td>UCB1RXIFG</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_B1 receive interrupt flag. UCB1RXIFG is set when UCB1RXBUF has</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>received a complete character.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>1</td>
<td>UCA1TXIFG</td>
<td>R/W</td>
<td>1h</td>
<td>USCI_A1 transmit interrupt flag. UCA1TXIFG is set when UCA1TXBUF empty.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>0</td>
<td>UCA1RXIFG</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_A1 receive interrupt flag. UCA1RXIFG is set when UCA1RXBUF has</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>received a complete character.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
</tbody>
</table>
The universal serial communication interface (USCI) supports multiple serial communication modes with one hardware module. This chapter discusses the operation of the I2C mode.

17.1 USCI Overview................................................................. 484
17.2 USCI Introduction: I2C Mode........................................ 484
17.3 USCI Operation: I2C Mode............................................... 485
17.4 USCI Registers: I2C Mode............................................... 501
**17.1 USCI Overview**

The universal serial communication interface (USCI) modules support multiple serial communication modes. Different USCI modules support different modes. Each different USCI module is named with a different letter. For example, USCI_A is different from USCI_B. If more than one identical USCI module is implemented on one device, those modules are named with incrementing numbers. For example, if one device has two USCI_A modules, they are named USCI_A0 and USCI_A1. See the device-specific data sheet to determine which USCI modules, if any, are implemented on a given device.

The USCI_Ax modules support:

- UART mode
- Pulse shaping for IrDA communications
- Automatic baud-rate detection for LIN communications
- SPI mode

The USCI_Bx modules support:

- I²C mode
- SPI mode

**17.2 USCI Introduction: I²C Mode**

In I²C mode, the USCI module provides an interface between the MSP430 and I²C-compatible devices connected by way of the two-wire I²C serial bus. External components attached to the I²C bus serially transmit and/or receive serial data to/from the USCI module through the 2-wire I²C interface.

The I²C mode features include:

- Compliance to the Philips Semiconductor I²C specification v2.1
  - 7-bit and 10-bit device addressing modes
  - General call
  - START/RESTART/STOP
  - Multi-master transmitter/receiver mode
  - Slave receiver/transmitter mode
  - Standard mode up to 100 kbps and fast mode up to 400 kbps support
- Programmable UCxCLK frequency in master mode
- Designed for low power
- Slave receiver START detection for auto-wake up from LPMx modes
- Slave operation in LPM4

*Figure 17-1* shows the USCI when configured in I²C mode.
17.3 USCI Operation: I²C Mode

The I²C mode supports any slave or master I²C-compatible device. Figure 17-2 shows an example of an I²C bus. Each I²C device is recognized by a unique address and can operate as either a transmitter or a receiver. A device connected to the I²C bus can be considered as the master or the slave when performing data transfers. A master initiates a data transfer and generates the clock signal SCL. Any device addressed by a master is considered a slave.

I²C data is communicated using the serial data pin (SDA) and the serial clock pin (SCL). Both SDA and SCL are bidirectional, and must be connected to a positive supply voltage using a pullup resistor.

---

**Note**

**SDA and SCL Levels**

The MSP430 SDA and SCL pins must not be pulled up above the MSP430 V_{CC} level.
17.3.1 USCI Initialization and Reset

The USCI is reset by a PUC or by setting the UCSWRST bit. After a PUC, the UCSWRST bit is automatically set, keeping the USCI in a reset condition. To select I2C operation the UCMODEEx bits must be set to 11. After module initialization, it is ready for transmit or receive operation. Clearing UCSWRST releases the USCI for operation.

Configuring and reconfiguring the USCI module should be done when UCSWRST is set to avoid unpredictable behavior. Setting UCSWRST in I2C mode has the following effects:

- I2C communication stops
- SDA and SCL are high impedance
- UCBxI2CSTAT, bits 6-0 are cleared
- UCBxTXIE and UCBxRXIE are cleared
- UCBxTXIFG and UCBxRXIFG are cleared
- All other bits and registers remain unchanged.

**Note**

**Initializing or Reconfiguring the USCI Module**

The recommended USCI initialization or reconfiguration process is:

1. Set UCSWRST (BIS.B #UCSWRST, &UCxCTL1)
2. Initialize all USCI registers with UCSWRST = 1 (including UCxCTL1)
3. Configure ports.
4. Clear UCSWRST by software (BIC.B #UCSWRST, &UCxCTL1)
5. Enable interrupts (optional) using UCxRXIE or UCxTXIE

17.3.2 I2C Serial Data

One clock pulse is generated by the master device for each data bit transferred. The I2C mode operates with byte data. Data is transferred most significant bit first as shown in Figure 17-3.

The first byte after a START condition consists of a 7-bit slave address and the R/W bit. When R/W = 0, the master transmits data to a slave. When R/W = 1, the master receives data from a slave. The ACK bit is sent from the receiver after each byte on the 9th SCL clock.
START and STOP conditions are generated by the master and are shown in Figure 17-3. A START condition is a high-to-low transition on the SDA line while SCL is high. A STOP condition is a low-to-high transition on the SDA line while SCL is high. The bus busy bit, UCBBUSY, is set after a START and cleared after a STOP.

Data on SDA must be stable during the high period of SCL as shown in Figure 17-4. The high and low state of SDA can only change when SCL is low, otherwise START or STOP conditions will be generated.

17.3.3 I²C Addressing Modes

The I²C mode supports 7-bit and 10-bit addressing modes.

17.3.3.1 7-Bit Addressing

In the 7-bit addressing format, shown in Figure 17-5, the first byte is the 7-bit slave address and the R/W bit. The ACK bit is sent from the receiver after each byte.

Figure 17-5. I²C Module 7-Bit Addressing Format

17.3.3.2 10-Bit Addressing

In the 10-bit addressing format, shown in Figure 17-6, the first byte is made up of 11110b plus the two MSBs of the 10-bit slave address and the R/W bit. The ACK bit is sent from the receiver after each byte. The next byte is the remaining 8 bits of the 10-bit slave address, followed by the ACK bit and the 8-bit data.

Figure 17-6. I²C Module 10-Bit Addressing Format
17.3.3 Repeated Start Conditions

The direction of data flow on SDA can be changed by the master, without first stopping a transfer, by issuing a repeated START condition. This is called a RESTART. After a RESTART is issued, the slave address is again sent out with the new data direction specified by the R/W bit. The RESTART condition is shown in Figure 17-7.

![Figure 17-7. I^2C Module Addressing Format with Repeated START Condition](image)

17.3.4 I^2C Module Operating Modes

In I^2C mode the USCI module can operate in master transmitter, master receiver, slave transmitter, or slave receiver mode. The modes are discussed in the following sections. Time lines are used to illustrate the modes.

Figure 17-8 shows how to interpret the time line figures. Data transmitted by the master is represented by grey rectangles, data transmitted by the slave by white rectangles. Data transmitted by the USCI module, either as master or slave, is shown by rectangles that are taller than the others.

Actions taken by the USCI module are shown in grey rectangles with an arrow indicating where in the data stream the action occurs. Actions that must be handled with software are indicated with white rectangles with an arrow pointing to where in the data stream the action must take place.

![Figure 17-8. I^2C Time Line Legend](image)

17.3.4.1 Slave Mode

The USCI module is configured as an I^2C slave by selecting the I^2C mode with UCMODEx = 11 and UCSYNC = 1 and clearing the UCMST bit.

Initially the USCI module must to be configured in receiver mode by clearing the UCTR bit to receive the I^2C address. Afterwards, transmit and receive operations are controlled automatically depending on the R/W bit received together with the slave address.

The USCI slave address is programmed with the UCBxI^2COA register. When UCA10 = 0, 7-bit addressing is selected. When UCA10 = 1, 10-bit addressing is selected. The UCGCEN bit selects if the slave responds to a general call.
When a START condition is detected on the bus, the USCI module will receive the transmitted address and compare it against its own address stored in UCBxI2COA. The UCSTTIFG flag is set when address received matches the USCI slave address.

### 17.3.4.1.1 I²C Slave Transmitter Mode

Slave transmitter mode is entered when the slave address transmitted by the master is identical to its own address with a set R/W bit. The slave transmitter shifts the serial data out on SDA with the clock pulses that are generated by the master device. The slave device does not generate the clock, but it will hold SCL low while intervention of the CPU is required after a byte has been transmitted.

If the master requests data from the slave the USCI module is automatically configured as a transmitter and UCTR and UCBxTXIFG become set. The SCL line is held low until the first data to be sent is written into the transmit buffer UCBxTXBUF. Then the address is acknowledged, the UCSTTIFG flag is cleared, and the data is transmitted. As soon as the data is transferred into the shift register the UCBxTXIFG is set again. After the data is acknowledged by the master the next data byte written into UCBxTXBUF is transmitted or if the buffer is empty the bus is stalled during the acknowledge cycle by holding SCL low until new data is written into UCBxTXBUF. If the master sends a NACK succeeded by a STOP condition the UCSTPIFG flag is set. If the NACK is succeeded by a repeated START condition the USCI I²C state machine returns to its address-reception state.

Figure 17-9 shows the slave transmitter operation.
17.3.4.1.2 I²C Slave Receiver Mode

Slave receiver mode is entered when the slave address transmitted by the master is identical to its own address and a cleared R/W bit is received. In slave receiver mode, serial data bits received on SDA are shifted in with the clock pulses that are generated by the master device. The slave device does not generate the clock, but it can hold SCL low if intervention of the CPU is required after a byte has been received.

If the slave should receive data from the master the USCI module is automatically configured as a receiver and UCTR is cleared. After the first data byte is received the receive interrupt flag UCBxRXIFG is set. The USCI module automatically acknowledges the received data and can receive the next data byte.

If the previous data was not read from the receive buffer UCBxRXBUF at the end of a reception, the bus is stalled by holding SCL low. As soon as UCBxRXBUF is read the new data is transferred into UCBxRXBUF, an acknowledge is sent to the master, and the next data can be received.

Setting the UCTXNACK bit causes a NACK to be transmitted to the master during the next acknowledgment cycle. A NACK is sent even if UCBxRXBUF is not ready to receive the latest data. If the UCTXNACK bit is set while SCL is held low the bus will be released, a NACK is transmitted immediately, and UCBxRXBUF is loaded with the last received data. Since the previous data was not read that data will be lost. To avoid loss of data the UCBxRXBUF needs to be read before UCTXNACK is set.

When the master generates a STOP condition the UCSTPIFG flag is set.

If the master generates a repeated START condition the USCI I²C state machine returns to its address reception state.

Figure 17-10 shows the I²C slave receiver operation.
Reception of own address and data bytes. All are acknowledged.

UCTR=0 (Receiver)
UCSTIFG=1
UCSTPIFG=0

Last byte is not acknowledged.

UCTR=0 (Receiver)
UCG=1

Reception of the general call address.

UCTR=0 (Receiver)
UCSTIFG=1
UCSTTIFG=1 (UCGC=1 if general call)
UCSTPIFG=0
UCBxTXIFG=0

Arbitration lost as master and addressed as slave

UCTR=0 (Receiver)
UCMST=0
UCSTTIFG=1
UCSTPIFG=0
UCSTTIFG=1
UCSTPIFG=0

Bus stalled (SCL held low)
if UCBxRXBUF not read

Read data from UCBxRXBUF

UCTXNACK=1
UCTXNACK=0

Gen Call

UCTR=0 (Receiver)
UCSTTIFG=1
UCSTPIFG=0

Reception of the general call address.

UCTR=0 (Receiver)
UCSTTIFG=0

Figure 17-10. I2C Slave Receiver Mode
17.3.4.1.3 I^2C Slave 10-bit Addressing Mode

The 10-bit addressing mode is selected when UCA10 = 1 and is as shown in Figure 17-11. In 10-bit addressing mode, the slave is in receive mode after the full address is received. The USCI module indicates this by setting the UCSTTIFG flag while the UCTR bit is cleared. To switch the slave into transmitter mode the master sends a repeated START condition together with the first byte of the address but with the R/W bit set. This will set the UCSTTIFG flag if it was previously cleared by software and the USCI module switches to transmitter mode with UCTR = 1.

---

Slave Receiver

- Reception of own address and data bytes. All are acknowledged.

- Reception of the general call address.

- UCTR=0 (Receiver)
  - UCSTTIFG=1
  - UCSTPIFG=0

- UCBxRXIFG=1

---

Slave Transmitter

- Reception of own address and transmission of data bytes

- UCTR=0 (Receiver)
  - UCSTTIFG=1
  - UCSTPIFG=0

- UCBxRXIFG=1

---

Figure 17-11. I^2C Slave 10-bit Addressing Mode
17.3.4.2 Master Mode

The USCI module is configured as an I²C master by selecting the I²C mode with UCMODEEx = 11 and UCSYNC = 1 and setting the UCMST bit. When the master is part of a multi-master system, UCMM must be set and its own address must be programmed into the UCBxI2COA register. When UCA10 = 0, 7-bit addressing is selected. When UCA10 = 1, 10-bit addressing is selected. The UCGCEN bit selects if the USCI module responds to a general call.

17.3.4.2.1 I²C Master Transmitter Mode

After initialization, master transmitter mode is initiated by writing the desired slave address to the UCBxI2CSA register, selecting the size of the slave address with the UCSLA10 bit, setting UCTR for transmitter mode, and setting UCTXSTT to generate a START condition.

The USCI module checks if the bus is available, generates the START condition, and transmits the slave address. The UCBxTXIFG bit is set when the START condition is generated and the first data to be transmitted can be written into UCBxTXBUF. As soon as the slave acknowledges the address the UCTXSTT bit is cleared.

The data written into UCBxTXBUF is transmitted if arbitration is not lost during transmission of the slave address. UCBxTXIFG is set again as soon as the data is transferred from the buffer into the shift register. If there is no data loaded to UCBxTXBUF before the acknowledge cycle, the bus is held during the acknowledge cycle with SCL low until data is written into UCBxTXBUF. Data is transmitted or the bus is held as long as the UCTXSTP bit or UCTXSTT bit is not set.

Setting UCTXSTP will generate a STOP condition after the next acknowledge from the slave. If UCTXSTP is set during the transmission of the slave’s address or while the USCI module waits for data to be written into UCBxTXBUF, a STOP condition is generated even if no data was transmitted to the slave. When transmitting a single byte of data, the UCTXSTP bit must be set while the byte is being transmitted, or anytime after transmission begins, without writing new data into UCBxTXBUF. Otherwise, only the address will be transmitted. When the data is transferred from the buffer to the shift register, UCBxTXIFG will become set indicating data transmission has begun and the UCTXSTP bit may be set.

Setting UCTXSTT will generate a repeated START condition. In this case, UCTR may be set or cleared to configure transmitter or receiver, and a different slave address may be written into UCBxI2CSA if desired.

If the slave does not acknowledge the transmitted data the not-acknowledge interrupt flag UCNACKIFG is set. The master must react with either a STOP condition or a repeated START condition. If data was already written into UCBxTXBUF it will be discarded. If this data should be transmitted after a repeated START it must be written into UCBxTXBUF again. Any set UCTXSTT is discarded, too. To trigger a repeated start UCTXSTT needs to be set again.

Figure 17-12 shows the I²C master transmitter operation.
Successful transmission to a slave receiver

Next transfer started with a repeated start condition

Not acknowledge received after slave address

Not acknowledge received after a data byte

Arbitration lost in slave address or data byte

Arbitration lost and addressed as slave

Figure 17-12. I²C Master Transmitter Mode

17.3.4.2.2 I²C Master Receiver Mode

After initialization, master receiver mode is initiated by writing the desired slave address to the UCBxI2CSA register, selecting the size of the slave address with the UCSLA10 bit, clearing UCTR for receiver mode, and setting UCTXSTT to generate a START condition.
The USCI module checks if the bus is available, generates the START condition, and transmits the slave address. As soon as the slave acknowledges the address the UCTXSTT bit is cleared.

After the acknowledge of the address from the slave the first data byte from the slave is received and acknowledged and the UCBxRXIFG flag is set. Data is received from the slave ss long as UCTXSTP or UCTXSTT is not set. If UCBxRXBUF is not read the master holds the bus during reception of the last data bit and until the UCBxRXBUF is read.

If the slave does not acknowledge the transmitted address the not-acknowledge interrupt flag UCNACKIFG is set. The master must react with either a STOP condition or a repeated START condition.

Setting the UCTXSTP bit will generate a STOP condition. After setting UCTXSTP, a NACK followed by a STOP condition is generated after reception of the data from the slave, or immediately if the USCI module is currently waiting for UCBxRXBUF to be read.

If a master wants to receive a single byte only, the UCTXSTP bit must be set while the byte is being received. For this case, the UCTXSTT may be polled to determine when it is cleared:

```
BIS.B #UCTXSTT,&UCB0CTL1 ;Transmit START cond.
POLL_STT BIT.B #UCTXSTT,&UCB0CTL1 ;Poll UCTXSTT bit
JC POLL_STT ;When cleared,
BIS.B #UCTXSTP,&UCB0CTL1 ;transmit STOP cond.
```

Setting UCTXSTT generates a repeated START condition. In this case, UCTR may be set or cleared to configure transmitter or receiver, and a different slave address may be written into UCBxI2CSA if desired. Figure 17-13 shows the I²C master receiver operation.

---

**Note**

**Consecutive Master Transactions Without Repeated Start**

When performing multiple consecutive I²C master transactions without the repeated start feature, the current transaction must be completed before the next one is initiated. This can be done by ensuring that the transmit stop condition flag UCTXSTP is cleared before the next I²C transaction is initiated with setting UCTXSTT = 1. Otherwise, the current transaction might be affected.
Figure 17-13. $^2$C Master Receiver Mode

1) UCTR=0 (Receiver)  
2) UCTXSTT=1
Successful reception from a slave transmitter

Next transfer started with a repeated start condition

Not acknowledge received after slave address

Arbitration lost in slave address or data byte

Arbitration lost and addressed as slave

Other master continues

$^2$C Master Transmitter Mode

DATA A S SLA/W

1) UCTR=1 (Transmitter)
2) UCTXSTT=1

DATA A S SLA/R

1) UCTR=0 (Receiver)
2) UCTXSTT=1

Other master continues

UCALIFG=1
UCMST=0
(UCSTTIFG=0)

USCI continues as Slave Transmitter

Figure 17-13. $^2$C Master Receiver Mode

1) UCTR=1 (Transmitter)  
2) UCTXSTT=1
Next transfer started with a repeated start condition

Not acknowledge received after slave address

Arbitration lost in slave address or data byte

Arbitration lost and addressed as slave

Other master continues

$^2$C Master Transmitter Mode

DATA A S SLA/W

1) UCTR=1 (Transmitter)
2) UCTXSTT=1

DATA A S SLA/R

1) UCTR=0 (Receiver)
2) UCTXSTT=1

Other master continues

UCALIFG=1
UCMST=0
(UCSTTIFG=0)

USCI continues as Slave Transmitter

Figure 17-13. $^2$C Master Receiver Mode

1) UCTR=1 (Transmitter)  
2) UCTXSTT=1
Next transfer started with a repeated start condition

Not acknowledge received after slave address

Arbitration lost in slave address or data byte

Arbitration lost and addressed as slave

Other master continues

$^2$C Master Transmitter Mode

DATA A S SLA/W

1) UCTR=1 (Transmitter)
2) UCTXSTT=1

DATA A S SLA/R

1) UCTR=0 (Receiver)
2) UCTXSTT=1

Other master continues

UCALIFG=1
UCMST=0
(UCSTTIFG=0)

USCI continues as Slave Transmitter

Figure 17-13. $^2$C Master Receiver Mode

1) UCTR=1 (Transmitter)  
2) UCTXSTT=1
Next transfer started with a repeated start condition

Not acknowledge received after slave address

Arbitration lost in slave address or data byte

Arbitration lost and addressed as slave

Other master continues

$^2$C Master Transmitter Mode

DATA A S SLA/W

1) UCTR=1 (Transmitter)
2) UCTXSTT=1

DATA A S SLA/R

1) UCTR=0 (Receiver)
2) UCTXSTT=1

Other master continues

UCALIFG=1
UCMST=0
(UCSTTIFG=0)

USCI continues as Slave Transmitter

Figure 17-13. $^2$C Master Receiver Mode

1) UCTR=1 (Transmitter)  
2) UCTXSTT=1
Next transfer started with a repeated start condition

Not acknowledge received after slave address

Arbitration lost in slave address or data byte

Arbitration lost and addressed as slave

Other master continues

$^2$C Master Transmitter Mode

DATA A S SLA/W

1) UCTR=1 (Transmitter)
2) UCTXSTT=1

DATA A S SLA/R

1) UCTR=0 (Receiver)
2) UCTXSTT=1

Other master continues

UCALIFG=1
UCMST=0
(UCSTTIFG=0)

USCI continues as Slave Transmitter

Figure 17-13. $^2$C Master Receiver Mode

1) UCTR=1 (Transmitter)  
2) UCTXSTT=1
Next transfer started with a repeated start condition

Not acknowledge received after slave address

Arbitration lost in slave address or data byte

Arbitration lost and addressed as slave

Other master continues

$^2$C Master Transmitter Mode

DATA A S SLA/W

1) UCTR=1 (Transmitter)
2) UCTXSTT=1

DATA A S SLA/R

1) UCTR=0 (Receiver)
2) UCTXSTT=1

Other master continues

UCALIFG=1
UCMST=0
(UCSTTIFG=0)

USCI continues as Slave Transmitter

Figure 17-13. $^2$C Master Receiver Mode

1) UCTR=1 (Transmitter)  
2) UCTXSTT=1
Next transfer started with a repeated start condition

Not acknowledge received after slave address

Arbitration lost in slave address or data byte

Arbitration lost and addressed as slave

Other master continues

$^2$C Master Transmitter Mode

DATA A S SLA/W

1) UCTR=1 (Transmitter)
2) UCTXSTT=1

DATA A S SLA/R

1) UCTR=0 (Receiver)
2) UCTXSTT=1

Other master continues

UCALIFG=1
UCMST=0
(UCSTTIFG=0)

USCI continues as Slave Transmitter

Figure 17-13. $^2$C Master Receiver Mode
17.3.4.2.3 I²C Master 10-Bit Addressing Mode

The 10-bit addressing mode is selected when UCSLA10 = 1 and is shown in Figure 17-14.

**Master Transmitter**

Successful transmission to a slave receiver

<table>
<thead>
<tr>
<th>S</th>
<th>11110 xx/W</th>
<th>A</th>
<th>SLA (2)</th>
<th>A</th>
<th>DATA</th>
<th>A</th>
<th>DATA</th>
<th>A</th>
<th>P</th>
</tr>
</thead>
<tbody>
<tr>
<td>1) UCTR=1 (Transmitter)</td>
<td>UCTXSTT=0</td>
<td>UCTXSTP=0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>2) UCTXSTT=1</td>
<td>UCBxTXIFG=1</td>
<td>UCTXSTP=1</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Master Receiver

Successful reception from a slave transmitter

<table>
<thead>
<tr>
<th>S</th>
<th>11110 xx/R</th>
<th>A</th>
<th>S</th>
<th>11110 xx/R</th>
<th>A</th>
<th>DATA</th>
<th>A</th>
<th>DATA</th>
<th>A</th>
<th>P</th>
</tr>
</thead>
<tbody>
<tr>
<td>1) UCTR=0 (Receiver)</td>
<td>UCTXSTT=0</td>
<td>UCTXSTP=0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>2) UCTXSTT=1</td>
<td>UCBxRXIFG=1</td>
<td>UCTXSTP=1</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Figure 17-14. I²C Master 10-bit Addressing Mode**

17.3.4.2.4 Arbitration

If two or more master transmitters simultaneously start a transmission on the bus, an arbitration procedure is invoked. Figure 17-15 shows the arbitration procedure between two devices. The arbitration procedure uses the data presented on SDA by the competing transmitters. The first master transmitter that generates a logic high is overruled by the opposing master generating a logic low. The arbitration procedure gives priority to the device that transmits the serial data stream with the lowest binary value. The master transmitter that lost arbitration switches to the slave receiver mode, and sets the arbitration lost flag UCALIFG. If two or more devices send identical first bytes, arbitration continues on the subsequent bytes.

**Figure 17-15. Arbitration Procedure Between Two Master Transmitters**

If the arbitration procedure is in progress when a repeated START condition or STOP condition is transmitted on SDA, the master transmitters involved in arbitration must send the repeated START condition or STOP condition at the same position in the format frame. Arbitration is not allowed between:

- A repeated START condition and a data bit
- A STOP condition and a data bit
- A repeated START condition and a STOP condition
### 17.3.5 I²C Clock Generation and Synchronization

The I²C clock SCL is provided by the master on the I²C bus. When the USCI is in master mode, BITCLK is provided by the USCI bit clock generator and the clock source is selected with the UCSSELx bits. In slave mode the bit clock generator is not used and the UCSSELx bits are don’t care.

The 16-bit value of UCBRx in registers UCBxBR1 and UCBxBR0 is the division factor of the USCI clock source, BRCLK. The maximum bit clock that can be used in single master mode is $f_{BRCLK}/4$. In multi-master mode the maximum bit clock is $f_{BRCLK}/8$. The BITCLK frequency is given by:

$$f_{\text{BitClock}} = \frac{f_{BRCLK}}{UCBRx}$$

The minimum high and low periods of the generated SCL are

$$t_{LOW,\text{MIN}} = t_{HIGH,\text{MIN}} = \frac{UCBRx}{2} \times \frac{f_{BRCLK}}{}$$

when UCBRx is even and

$$t_{LOW,\text{MIN}} = t_{HIGH,\text{MIN}} = \frac{(UCBRx - 1)}{2} \times \frac{f_{BRCLK}}{}$$

when UCBRx is odd.

The USCI clock source frequency and the prescaler setting UCBRx must be chosen such that the minimum low and high period times of the I²C specification are met.

During the arbitration procedure the clocks from the different masters must be synchronized. A device that first generates a low period on SCL overrules the other devices forcing them to start their own low periods. SCL is then held low by the device with the longest low period. The other devices must wait for SCL to be released before starting their high periods. **Figure 17-16** shows the clock synchronization. This allows a slow slave to slow down a fast master.

![Figure 17-16. Synchronization of Two I²C Clock Generators During Arbitration](image)

#### 17.3.5.1 Clock Stretching

The USCI module supports clock stretching and also makes use of this feature as described in the operation mode sections.

The UCSCLLOW bit can be used to observe if another device pulls SCL low while the USCI module already released SCL due to the following conditions:

- USCI is acting as master and a connected slave drives SCL low.
- USCI is acting as master and another master drives SCL low during arbitration.

The UCSCLLOW bit is also active if the USCI holds SCL low because it is waiting as transmitter for data being written into UCBxTXBUF or as receiver for the data being read from UCBxRXBUF.

The UCSCLLOW bit might get set for a short time with each rising SCL edge because the logic observes the external SCL and compares it to the internally generated SCL.
17.3.6 Using the USCI Module in I²C Mode with Low-Power Modes

The USCI module provides automatic clock activation for SMCLK for use with low-power modes. When SMCLK is the USCI clock source, and is inactive because the device is in a low-power mode, the USCI module automatically activates it when needed, regardless of the control-bit settings for the clock source. The clock remains active until the USCI module returns to its idle condition. After the USCI module returns to the idle condition, control of the clock source reverts to the settings of its control bits. Automatic clock activation is not provided for ACLK.

When the USCI module activates an inactive clock source, the clock source becomes active for the whole device and any peripheral configured to use the clock source may be affected. For example, a timer using SMCLK will increment while the USCI module forces SMCLK active.

In I²C slave mode no internal clock source is required because the clock is provided by the external master. It is possible to operate the USCI in I²C slave mode while the device is in LPM4 and all internal clock sources are disabled. The receive or transmit interrupts can wake up the CPU from any low power mode.

17.3.7 USCI Interrupts in I²C Mode

There are two interrupt vectors for the USCI module in I²C mode. One interrupt vector is associated with the transmit and receive interrupt flags. The other interrupt vector is associated with the four state change interrupt flags. Each interrupt flag has its own interrupt enable bit. When an interrupt is enabled, and the GIE bit is set, the interrupt flag will generate an interrupt request. DMA transfers are controlled by the UCBxTXIFG and UCBxRXIFG flags on devices with a DMA controller.

17.3.7.1 I²C Transmit Interrupt Operation

The UCBxTXIFG interrupt flag is set by the transmitter to indicate that UCBxTXBUF is ready to accept another character. An interrupt request is generated if UCBxTXIE and GIE are also set. UCBxTXIFG is automatically reset if a character is written to UCBxTXBUF or if a NACK is received. UCBxTXIFG is set when UCSWRST = 1 and the I²C mode is selected. UCBxTXIE is reset after a PUC or when UCSWRST = 1.

17.3.7.2 I²C Receive Interrupt Operation

The UCBxRXIFG interrupt flag is set when a character is received and loaded into UCBxRXBUF. An interrupt request is generated if UCBxRXIE and GIE are also set. UCBxRXIFG and UCBxRXIE are reset after a PUC signal or when UCSWRST = 1. UCxRXIFG is automatically reset when UCxRXBUF is read.

17.3.7.3 I²C State Change Interrupt Operation

Table 17-1 describes the I²C state change interrupt flags.

<table>
<thead>
<tr>
<th>Interrupt Flag</th>
<th>Interrupt Condition</th>
</tr>
</thead>
<tbody>
<tr>
<td>UCALIFG</td>
<td>Arbitration-lost. Arbitration can be lost when two or more transmitters start a transmission simultaneously, or when the USCI operates as master but is addressed as a slave by another master in the system. The UCALIFG flag is set when arbitration is lost. When UCALIFG is set the UCMST bit is cleared and the I²C controller becomes a slave.</td>
</tr>
<tr>
<td>UCNACKIFG</td>
<td>Not-acknowledge interrupt. This flag is set when an acknowledge is expected but is not received. UCNACKIFG is automatically cleared when a START condition is received.</td>
</tr>
<tr>
<td>UCSTTIFG</td>
<td>Start condition detected interrupt. This flag is set when the I²C module detects a START condition together with its own address while in slave mode. UCSTTIFG is used in slave mode only and is automatically cleared when a STOP condition is received.</td>
</tr>
<tr>
<td>UCSTPIFG</td>
<td>Stop condition detected interrupt. This flag is set when the I²C module detects a STOP condition while in slave mode. UCSTPIFG is used in slave mode only and is automatically cleared when a START condition is received.</td>
</tr>
</tbody>
</table>

17.3.7.4 Interrupt Vector Assignment

USCI_Ax and USCI_Bx share the same interrupt vectors. In I²C mode the state change interrupt flags UCSTTIFG, UCSTPIFG, UCNACKIFG, UCALIFG from USCI_Bx and UCAxRXIFG from USCI_Ax are routed to one interrupt vector. The I²C transmit and receive interrupt flags UCBxTXIFG and UCBxRXIFG from USCI_Bx and UCAxTXIFG from USCI_Ax share another interrupt vector.
Example 17-1 shows an extract of the interrupt service routine to handle data receive interrupts from USCI_A0 in either UART or SPI mode and state change interrupts from USCI_B0 in I²C mode.

**Example 17-1. Shared Receive Interrupt Vectors Software Example**

```
USCIA0_RX_USCIB0_I2C_STATE_ISR
  BIT.B  #UCA0RXIFG, &IFG2  ; USCI_A0 Receive Interrupt?
  JNZ    USCIA0_RX_ISR
  USCIB0_I2C_STATE_ISR
  ; Decode I2C state changes ...
  ; Decode I2C state changes ...
  ...
  RETI
USCIA0_RX_ISR
  ; Read UCA0RXBUF ... - clears UCA0RXIFG
  ...
  RETI
```

**Example 17-2. Shared Transmit Interrupt Vectors Software Example**

```
USCIA0_TX_USCIB0_I2C_DATA_ISR
  BIT.B  #UCA0TXIFG, &IFG2  ; USCI_A0 Transmit Interrupt?
  JNZ    USCIA0_TX_ISR
  USCIB0_I2C_DATA_ISR
  BIT.B  #UCB0RXIFG, &IFG2
  JNZ    USCIB0_I2C_RX
  USCIB0_I2C_TX
  ; Write UCB0TXBUF... - clears UCB0TXIFG
  ...
  RETI
USCIB0_I2C_RX
  ; Read UCB0RXBUF... - clears UCB0RXIFG
  ...
  RETI
USCIA0_TX_ISR
  ; Write UCA0TXBUF ... - clears UCA0TXIFG
  ...
  RETI
```
17.4 USCI Registers: I²C Mode

Table 17-2 lists the memory-mapped registers for the USCI_Bx in I²C mode.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>68h</td>
<td>UCB0CTL0</td>
<td>USCI_B0 control 0</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 17.4.1</td>
</tr>
<tr>
<td>69h</td>
<td>UCB0CTL1</td>
<td>USCI_B0 control 1</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 17.4.2</td>
</tr>
<tr>
<td>6Ah</td>
<td>UCB0BR0</td>
<td>USCI_B0 bit rate control 0</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.3</td>
</tr>
<tr>
<td>6Bh</td>
<td>UCB0BR1</td>
<td>USCI_B0 bit rate control 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.3</td>
</tr>
<tr>
<td>6Ch</td>
<td>UCB0I2CIE</td>
<td>USCI_B0 I²C interrupt enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.10</td>
</tr>
<tr>
<td>6Dh</td>
<td>UCB0STAT</td>
<td>USCI_B0 status</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.5</td>
</tr>
<tr>
<td>6Eh</td>
<td>UCB0RXBUF</td>
<td>USCI_B0 receive buffer</td>
<td>Read</td>
<td>00h with PUC</td>
<td>Section 17.4.6</td>
</tr>
<tr>
<td>6Fh</td>
<td>UCB0TXBUF</td>
<td>USCI_B0 transmit buffer</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.7</td>
</tr>
<tr>
<td>118h</td>
<td>UCB0I2COA</td>
<td>USCI_B0 I²C own address</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.8</td>
</tr>
<tr>
<td>11Ah</td>
<td>UCB0I2CSA</td>
<td>USCI_B0 I²C slave address</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.9</td>
</tr>
<tr>
<td>1h</td>
<td>IE2</td>
<td>SFR interrupt enable 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.11</td>
</tr>
<tr>
<td>3h</td>
<td>IFG2</td>
<td>SFR interrupt flag 2</td>
<td>Read/write</td>
<td>0Ah with PUC</td>
<td>Section 17.4.12</td>
</tr>
<tr>
<td>0D8h</td>
<td>UCB1CTL0</td>
<td>USCI_B1 control 0</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 17.4.1</td>
</tr>
<tr>
<td>0D9h</td>
<td>UCB1CTL1</td>
<td>USCI_B1 control 1</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 17.4.1</td>
</tr>
<tr>
<td>0DAh</td>
<td>UCB1BR0</td>
<td>USCI_B1 bit rate control 0</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.3</td>
</tr>
<tr>
<td>0DBh</td>
<td>UCB1BR1</td>
<td>USCI_B1 bit rate control 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.3</td>
</tr>
<tr>
<td>0DCh</td>
<td>UCB1I2CIE</td>
<td>USCI_B1 I²C interrupt enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.10</td>
</tr>
<tr>
<td>0DDh</td>
<td>UCB1STAT</td>
<td>USCI_B1 status</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.5</td>
</tr>
<tr>
<td>0DEh</td>
<td>UCB1RXBUF</td>
<td>USCI_B1 receive buffer</td>
<td>Read</td>
<td>00h with PUC</td>
<td>Section 17.4.6</td>
</tr>
<tr>
<td>0DFh</td>
<td>UCB1TXBUF</td>
<td>USCI_B1 transmit buffer</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.7</td>
</tr>
<tr>
<td>17Ch</td>
<td>UCB1I2COA</td>
<td>USCI_B1 I²C own address</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.8</td>
</tr>
<tr>
<td>17Eh</td>
<td>UCB1I2CSA</td>
<td>USCI_B1 I²C slave address</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.9</td>
</tr>
<tr>
<td>6h</td>
<td>UC1IE</td>
<td>USCI_A1/B1 interrupt enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 17.4.13</td>
</tr>
<tr>
<td>7h</td>
<td>UC1IFG</td>
<td>USCI_A1/B1 interrupt flag</td>
<td>Read/write</td>
<td>0Ah with PUC</td>
<td>Section 17.4.14</td>
</tr>
</tbody>
</table>

Note

Modifying SFR bits

To avoid modifying control bits of other modules, TI recommends setting or clearing the IEx and IFGx bits using BIS.B or BIC.B instructions, rather than MOV.B or CLR.B instructions.
17.4.1 UCBxCTL0 Register

USCI_Bx Control 0 Register

UCBxCTL0 is shown in Figure 17-17 and described in Table 17-3. 

Return to Table 17-2.

Figure 17-17. UCBxCTL0 Register

Table 17-3. UCBxCTL0 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>UCA10</td>
<td>R/W</td>
<td>0h</td>
<td>Own addressing mode select</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Own address is a 7-bit address</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Own address is a 10-bit address</td>
</tr>
<tr>
<td>6</td>
<td>UCSLA10</td>
<td>R/W</td>
<td>0h</td>
<td>Slave addressing mode select</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Address slave with 7-bit address</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Address slave with 10-bit address</td>
</tr>
<tr>
<td>5</td>
<td>UCMM</td>
<td>R/W</td>
<td>0h</td>
<td>Multi-master environment select</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Single master environment. There is no other master in the system. The address compare unit is disabled.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Multi-master environment</td>
</tr>
<tr>
<td>4</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td>Unused</td>
</tr>
<tr>
<td>3</td>
<td>UCMST</td>
<td>R/W</td>
<td>0h</td>
<td>Master mode select. When a master loses arbitration in a multi-master environment (UCMM = 1) the UCMST bit is automatically cleared and the module acts as slave.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Slave mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Master mode</td>
</tr>
<tr>
<td>2-1</td>
<td>UCMODEx</td>
<td>R/W</td>
<td>0h</td>
<td>USCI Mode. The UCMODEx bits select the synchronous mode when UCSYNC = 1.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = 3-pin SPI</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = 4-pin SPI (master/slave enabled if STE = 1)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = 4-pin SPI (master/slave enabled if STE = 0)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = I^2C mode</td>
</tr>
<tr>
<td>0</td>
<td>UCSYNC</td>
<td>R</td>
<td>1h</td>
<td>Synchronous mode enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Asynchronous mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Synchronous mode</td>
</tr>
</tbody>
</table>
17.4.2 UCBxCTL1 Register

USCI_Bx Control 1 Register

UCBxCTL1 is shown in Figure 17-18 and described in Table 17-4.

Return to Table 17-2.

Figure 17-18. UCBxCTL1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td>UCSSELx</td>
<td>R/W</td>
<td>0h</td>
<td>USCI clock source select. These bits select the BRCLK source clock. 00b = UCLKI 01b = ACLK 10b = SMCLK 11b = SMCLK</td>
</tr>
<tr>
<td>5</td>
<td>Unused</td>
<td>R</td>
<td>0h</td>
<td>Unused</td>
</tr>
<tr>
<td>4</td>
<td>UCTR</td>
<td>R/W</td>
<td>0h</td>
<td>Transmitter or receiver 0b = Receiver 1b = Transmitter</td>
</tr>
<tr>
<td>3</td>
<td>UCTXNACK</td>
<td>R/W</td>
<td>0h</td>
<td>Transmit a NACK. UCTXNACK is automatically cleared after a NACK is transmitted. 0b = Acknowledge normally 1b = Generate NACK</td>
</tr>
<tr>
<td>2</td>
<td>UCTXSTP</td>
<td>R/W</td>
<td>0h</td>
<td>Transmit STOP condition in master mode. Ignored in slave mode. In master receiver mode, the STOP condition is preceded by a NACK. UCTXSTP is automatically cleared after STOP is generated. 0b = No STOP generated 1b = Generate STOP</td>
</tr>
<tr>
<td>1</td>
<td>UCTXSTT</td>
<td>R/W</td>
<td>0h</td>
<td>Transmit START condition in master mode. Ignored in slave mode. In master receiver mode a repeated START condition is preceded by a NACK. UCTXSTT is automatically cleared after START condition and address information is transmitted. Ignored in slave mode. 0b = Do not generate START condition 1b = Generate START condition</td>
</tr>
<tr>
<td>0</td>
<td>UCSWRST</td>
<td>R/W</td>
<td>1h</td>
<td>Software reset enable 0b = Disabled. USCI reset released for operation. 1b = Enabled. USCI logic held in reset state.</td>
</tr>
</tbody>
</table>
17.4.3 UCBxBR0 Register
USCI_Bx Bit-Rate Control 0 Register
UCBxBR0 is shown in Figure 17-19 and described in Table 17-5.
Return to Table 17-2.

Figure 17-19. UCBxBR0 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UCBRx</td>
<td>R/W</td>
<td>0h</td>
<td>Bit clock prescaler setting. The 16-bit value of (UCBxBR0 + UCBxBR1 × 256) forms the prescaler value.</td>
</tr>
</tbody>
</table>

17.4.4 UCBxBR1 Register
USCI_Bx Bit-Rate Control 1 Register
UCBxBR1 is shown in Figure 17-20 and described in Table 17-6.
Return to Table 17-2.

Figure 17-20. UCBxBR1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UCBRx</td>
<td>R/W</td>
<td>0h</td>
<td>Bit clock prescaler setting. The 16-bit value of (UCBxBR0 + UCBxBR1 × 256) forms the prescaler value.</td>
</tr>
</tbody>
</table>
17.4.5 UCBxSTAT Register

USCI_Bx Status Register

UCBxSTAT is shown in Figure 17-21 and described in Table 17-7.

Return to Table 17-2.

![Figure 17-21. UCBxSTAT Register](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td>Unused</td>
</tr>
<tr>
<td>6</td>
<td>UCSCLLOW</td>
<td>R</td>
<td>0h</td>
<td>SCL low&lt;br&gt;0b = SCL is not held low&lt;br&gt;1b = SCL is held low</td>
</tr>
<tr>
<td>5</td>
<td>UCGC</td>
<td>R/W</td>
<td>0h</td>
<td>General call address received. UCGC is automatically cleared when a START condition is received.&lt;br&gt;0b = No general call address received&lt;br&gt;1b = General call address received</td>
</tr>
<tr>
<td>4</td>
<td>UCBBUSY</td>
<td>R</td>
<td>0h</td>
<td>Bus busy&lt;br&gt;0b = Bus inactive&lt;br&gt;1b = Bus busy</td>
</tr>
<tr>
<td>3</td>
<td>UCNACKIFG</td>
<td>R/W</td>
<td>0h</td>
<td>Not-acknowledge received interrupt flag. UCNACKIFG is automatically cleared when a START condition is received.&lt;br&gt;0b = No interrupt pending&lt;br&gt;1b = Interrupt pending</td>
</tr>
<tr>
<td>2</td>
<td>UCSTPIFG</td>
<td>R/W</td>
<td>0h</td>
<td>Stop condition interrupt flag. UCSTPIFG is automatically cleared when a START condition is received.&lt;br&gt;0b = No interrupt pending&lt;br&gt;1b = Interrupt pending</td>
</tr>
<tr>
<td>1</td>
<td>UCSTTIFG</td>
<td>R/W</td>
<td>0h</td>
<td>Start condition interrupt flag. UCSTTIFG is automatically cleared if a STOP condition is received.&lt;br&gt;0b = No interrupt pending&lt;br&gt;1b = Interrupt pending</td>
</tr>
<tr>
<td>0</td>
<td>UCALIFG</td>
<td>R/W</td>
<td>0h</td>
<td>Arbitration lost interrupt flag&lt;br&gt;0b = No interrupt pending&lt;br&gt;1b = Interrupt pending</td>
</tr>
</tbody>
</table>
17.4.6 UCBxRXBUF Register
USCI_Bx Receive Buffer Register
UCBxRXBUF is shown in Figure 17-22 and described in Table 17-8.
Return to Table 17-2.

Figure 17-22. UCBxRXBUF Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UCRXBUFx</td>
<td>R</td>
<td>0h</td>
<td>The receive-data buffer is user accessible and contains the last received character from the receive shift register. Reading UCBxRXBUF resets UCBxRXIFG.</td>
</tr>
</tbody>
</table>

17.4.7 UCBxTXBUF Register
USCI_Bx Transmit Buffer Register
UCBxTXBUF is shown in Figure 17-23 and described in Table 17-9.
Return to Table 17-2.

Figure 17-23. UCBxTXBUF Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UCTXBUFx</td>
<td>R/W</td>
<td>0h</td>
<td>The transmit data buffer is user accessible and holds the data waiting to be moved into the transmit shift register and transmitted. Writing to the transmit data buffer clears UCBxTXIFG.</td>
</tr>
</tbody>
</table>
17.4.8 UCBxI2COA Register
USCI_Bx I2C Own Address Register
UCBxI2COA is shown in Figure 17-24 and described in Table 17-10.
Return to Table 17-2.

Figure 17-24. UCBxI2COA Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15</td>
<td>UCGCEN</td>
<td>R/W</td>
<td>0h</td>
<td>General call response enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Do not respond to a general call</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Respond to a general call</td>
</tr>
<tr>
<td>14-10</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td></td>
</tr>
<tr>
<td>9-0</td>
<td>I2COAx</td>
<td>R/W</td>
<td>0h</td>
<td>I2C own address. The I2COAx bits contain the local address of the USCI_Bx I2C controller. The address is right-justified. In 7-bit addressing mode, bit 6 is the MSB, and bits 9-7 are ignored. In 10-bit addressing mode, bit 9 is the MSB.</td>
</tr>
</tbody>
</table>

17.4.9 UCBxI2CSA Register
USCI_Bx I2C Slave Address Register
UCBxI2CSA is shown in Figure 17-25 and described in Table 17-11.
Return to Table 17-2.

Figure 17-25. UCBxI2CSA Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15</td>
<td>Reserved</td>
<td></td>
<td>0h</td>
<td>I2CSA is used in master mode. The I2CSAx bits contain the slave address of the external device to be addressed by the USCI_Bx module. It is only used in master mode. The address is right-justified. In 7-bit slave addressing mode, bit 6 is the MSB, and bits 9-7 are ignored. In 10-bit slave addressing mode, bit 9 is the MSB.</td>
</tr>
<tr>
<td>14-10</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>9-0</td>
<td>I2CSAx</td>
<td>R/W</td>
<td>0h</td>
<td>I2C slave address. The I2CSAx bits contain the slave address of the external device to be addressed by the USCI_Bx module. It is only used in master mode. The address is right-justified. In 7-bit slave addressing mode, bit 6 is the MSB, and bits 9-7 are ignored. In 10-bit slave addressing mode, bit 9 is the MSB.</td>
</tr>
</tbody>
</table>
17.4.10 UCBxI2CIE Register

USCI_Bx I2C Interrupt Enable Register

UCBxI2CIE is shown in Figure 17-26 and described in Table 17-12.

Return to Table 17-2.

Figure 17-26. UCBxI2CIE Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td>Reserved</td>
<td>R/W</td>
<td>0h</td>
<td></td>
</tr>
<tr>
<td>3</td>
<td>UCNACKIE</td>
<td>R/W</td>
<td>0h</td>
<td>Not-acknowledge interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt enabled</td>
</tr>
<tr>
<td>2</td>
<td>UCSTPIE</td>
<td>R/W</td>
<td>0h</td>
<td>Stop condition interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt enabled</td>
</tr>
<tr>
<td>1</td>
<td>UCSTTIE</td>
<td>R/W</td>
<td>0h</td>
<td>Start condition interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt enabled</td>
</tr>
<tr>
<td>0</td>
<td>UCALIE</td>
<td>R/W</td>
<td>0h</td>
<td>Arbitration lost interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt enabled</td>
</tr>
</tbody>
</table>

Table 17-12. UCBxI2CIE Register Field Descriptions
17.4.11 IE2 Register
SFR Interrupt Enable 2 Register
IE2 is shown in Figure 17-27 and described in Table 17-13.
Return to Table 17-2.

Figure 17-27. IE2 Register

<table>
<thead>
<tr>
<th></th>
<th>7-6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>UCB0TXIE</td>
<td>UCB0RXIE</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>rw-0</td>
<td>rw-0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Table 17-13. IE2 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules (see the device-specific data sheet).</td>
</tr>
<tr>
<td>3</td>
<td>UCB0TXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_B0 transmit interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0b</td>
<td>Interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1b</td>
<td>Interrupt enabled</td>
</tr>
<tr>
<td>2</td>
<td>UCB0RXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_B0 receive interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0b</td>
<td>Interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1b</td>
<td>Interrupt enabled</td>
</tr>
<tr>
<td>1-0</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules (see the device-specific data sheet).</td>
</tr>
</tbody>
</table>

17.4.12 IFG2 Register
SFR Interrupt Flag 2 Register
IFG2 is shown in Figure 17-28 and described in Table 17-14.
Return to Table 17-2.

Figure 17-28. IFG2 Register

<table>
<thead>
<tr>
<th></th>
<th>7-6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>UCB0TXIFG</td>
<td>UCB0RXIFG</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>rw-1</td>
<td>rw-0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Table 17-14. IFG2 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules (see the device-specific data sheet).</td>
</tr>
<tr>
<td>3</td>
<td>UCB0TXIFG</td>
<td>R/W</td>
<td>1h</td>
<td>USCI_B0 transmit interrupt flag. UCB0TXIFG is set when UCB0TXBUF is empty.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0b</td>
<td>No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1b</td>
<td>Interrupt pending</td>
</tr>
<tr>
<td>2</td>
<td>UCB0RXIFG</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_B0 receive interrupt flag. UCB0RXIFG is set when UCB0RXBUF has received a complete character.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0b</td>
<td>No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1b</td>
<td>Interrupt pending</td>
</tr>
<tr>
<td>1-0</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules (see the device-specific data sheet).</td>
</tr>
</tbody>
</table>
17.4.13 UC1IE Register
USCI_A1/B1 Interrupt Enable Register
UC1IE is shown in Figure 17-29 and described in Table 17-15.
Return to Table 17-2.

Figure 17-29. UC1IE Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td>Unused</td>
</tr>
<tr>
<td>3</td>
<td>UCB1TXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_B1 transmit interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = interrupt enabled</td>
</tr>
<tr>
<td>2</td>
<td>UCB1RXIE</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_B1 receive interrupt enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = interrupt enabled</td>
</tr>
<tr>
<td>1-0</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other USCI modules</td>
</tr>
</tbody>
</table>

Table 17-15. UC1IE Register Field Descriptions

17.4.14 UC1IFG Register
USCI_A1/B1 Interrupt Flag Register
UC1IFG is shown in Figure 17-30 and described in Table 17-16.
Return to Table 17-2.

Figure 17-30. UC1IFG Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td>Unused</td>
</tr>
<tr>
<td>3</td>
<td>UCB1TXIFG</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_B1 transmit interrupt flag</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>UCB1TXBUF is empty</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>2</td>
<td>UCB1RXIFG</td>
<td>R/W</td>
<td>0h</td>
<td>USCI_B1 receive interrupt flag</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>UCB1RXBUF has received a complete character</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>1-0</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other USCI modules</td>
</tr>
</tbody>
</table>

Table 17-16. UC1IFG Register Field Descriptions
The universal synchronous/asynchronous receive/transmit (USART) peripheral interface supports two serial modes with one hardware module. This chapter discusses the operation of the asynchronous UART mode. USART0 is implemented on the MSP430AFE2xx devices.

18.1 USART Introduction: UART Mode .......................................................................................................................... 512
18.2 USART Operation: UART Mode .......................................................................................................................... 513
18.3 USART Registers – UART Mode .......................................................................................................................... 526
18.1 USART Introduction: UART Mode

In asynchronous mode, the USART connects the MSP430 to an external system via two external pins, URXD and UTXD. UART mode is selected when the SYNC bit is cleared.

UART mode features include:

- 7- or 8-bit data with odd parity, even parity, or non-parity
- Independent transmit and receive shift registers
- Separate transmit and receive buffer registers
- LSB-first data transmit and receive
- Built-in idle-line and address-bit communication protocols for multiprocessor systems
- Receiver start-edge detection for auto-wake up from LPMx modes
- Programmable baud rate with modulation for fractional baud rate support
- Status flags for error detection and suppression and address detection
- Independent interrupt capability for receive and transmit

Figure 18-1 shows the USART when configured for UART mode.

* See the device-specific data sheet for SFR locations.

Figure 18-1. USART Block Diagram: UART Mode
18.2 USART Operation: UART Mode

In UART mode, the USART transmits and receives characters at a bit rate asynchronous to another device. Timing for each character is based on the selected baud rate of the USART. The transmit and receive functions use the same baud rate frequency.

18.2.1 USART Initialization and Reset

The USART is reset by a PUC or by setting the SWRST bit. After a PUC, the SWRST bit is automatically set, keeping the USART in a reset condition. When set, the SWRST bit resets the URXIE, UTXIE, URXIFGx, RXWAKE, TXWAKE, RXERR, BRK, PE, OE, and FE bits and sets the UTXIFGx and TXEPT bits. The receive and transmit enable flags, URXEx and UTXEx, are not altered by SWRST. Clearing SWRST releases the USART for operation. See also chapter USART Module, I²C mode for USART0 when reconfiguring from I²C mode to UART mode.

Note

Initializing or Reconfiguring the USART Module

The required USART initialization/reconfiguration process is:

1. Set SWRST (BIS.B #SWRST,&UxCTL)
2. Initialize all USART registers with SWRST = 1 (including UxCTL)
3. Enable USART module via the MEx SFRs (URXEx and/or UTXEx)
4. Clear SWRST via software (BIC.B #SWRST,&UxCTL)
5. Enable interrupts (optional) via the IEx SFRs (URXIEx and/or UTXIEx)

Failure to follow this process may result in unpredictable USART behavior.

18.2.2 Character Format

The UART character format, shown in Figure 18-2, consists of a start bit, seven or eight data bits, an even/odd/no parity bit, an address bit (address-bit mode), and one or two stop bits. The bit period is defined by the selected clock source and setup of the baud rate registers.

![Figure 18-2. Character Format](image)

18.2.3 Asynchronous Communication Formats

When two devices communicate asynchronously, the idle-line format is used for the protocol. When three or more devices communicate, the USART supports the idle-line and address-bit multiprocessor communication formats.

18.2.3.1 Idle-Line Multiprocessor Format

When MM = 0, the idle-line multiprocessor format is selected. Blocks of data are separated by an idle time on the transmit or receive lines as shown in Figure 18-3. An idle receive line is detected when 10 or more continuous ones (marks) are received after the first stop bit of a character. When two stop bits are used for the idle line the second stop bit is counted as the first mark bit of the idle period.

The first character received after an idle period is an address character. The RXWAKE bit is used as an address tag for each block of characters. In the idle-line multiprocessor format, this bit is set when a received character is an address and is transferred to UxRXBUF.
The URXWIE bit is used to control data reception in the idle-line multiprocessor format. When the URXWIE bit is set, all non-address characters are assembled but not transferred into the UxRXBUF, and interrupts are not generated. When an address character is received, the receiver is temporarily activated to transfer the character to UxRXBUF and sets the URXIFGx interrupt flag. Any applicable error flag is also set. The user can then validate the received address.

If an address is received, user software can validate the address and must reset URXWIE to continue receiving data. If URXWIE remains set, only address characters are received. The URXWIE bit is not modified by the USART hardware automatically.

For address transmission in idle-line multiprocessor format, a precise idle period can be generated by the USART to generate address character identifiers on UTXDx. The wake-up temporary (WUT) flag is an internal flag double-buffered with the user-accessible TXWAKE bit. When the transmitter is loaded from UxTXBUF, WUT is also loaded from TXWAKE resetting the TXWAKE bit.

The following procedure sends out an idle frame to indicate an address character follows:

1. Set TXWAKE, then write any character to UxTXBUF. UxTXBUF must be ready for new data (UTXIFGx = 1).
   
   The TXWAKE value is shifted to WUT and the contents of UxTXBUF are shifted to the transmit shift register when the shift register is ready for new data. This sets WUT, which suppresses the start, data, and parity bits of a normal transmission, then transmits an idle period of exactly 11 bits. When two stop bits are used for the idle line, the second stop bit is counted as the first mark bit of the idle period. TXWAKE is reset automatically.

2. Write desired address character to UxTXBUF. UxTXBUF must be ready for new data (UTXIFGx = 1).
   
   The new character representing the specified address is shifted out following the address-identifying idle period on UTXDx. Writing the first "don't care" character to UxTXBUF is necessary in order to shift the TXWAKE bit to WUT and generate an idle-line condition. This data is discarded and does not appear on UTXDx.

### 18.2.3.2 Address-Bit Multiprocessor Format

When MM = 1, the address-bit multiprocessor format is selected. Each processed character contains an extra bit used as an address indicator shown in Figure 18-4. The first character in a block of characters carries a set address bit which indicates that the character is an address. The USART RXWAKE bit is set when a received character is a valid address character and is transferred to UxRXBUF.
The URXWIE bit is used to control data reception in the address-bit multiprocessor format. If URXWIE is set, data characters (address bit = 0) are assembled by the receiver but are not transferred to UxRXBUF and no interrupts are generated. When a character containing a set address bit is received, the receiver is temporarily activated to transfer the character to UxRXBUF and set URXIFGx. All applicable error status flags are also set.

If an address is received, user software must reset URXWIE to continue receiving data. If URXWIE remains set, only address characters (address bit = 1) are received. The URXWIE bit is not modified by the USART hardware automatically.

For address transmission in address-bit multiprocessor mode, the address bit of a character can be controlled by writing to the TXWAKE bit. The value of the TXWAKE bit is loaded into the address bit of the character transferred from UxTXBUF to the transmit shift register, automatically clearing the TXWAKE bit. TXWAKE must not be cleared by software. It is cleared by USART hardware after it is transferred to WUT or by setting SWRST.

**18.2.3.3 Automatic Error Detection**

Glitch suppression prevents the USART from being accidentally started. Any low-level on URXDx shorter than the deglitch time \( t_\tau \) (approximately 300 ns) is ignored. See the device-specific data sheet for parameters.

When a low period on URXDx exceeds \( t_\tau \), a majority vote is taken for the start bit. If the majority vote fails to detect a valid start bit the USART halts character reception and waits for the next low period on URXDx. The majority vote is also used for each bit in a character to prevent bit errors.

The USART module automatically detects framing errors, parity errors, overrun errors, and break conditions when receiving characters. The bits FE, PE, OE, and BRK are set when their respective condition is detected. When any of these error flags are set, RXERR is also set. The error conditions are described in Table 18-1.
Table 18-1. Receive Error Conditions

<table>
<thead>
<tr>
<th>Error Condition</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>Framing error</td>
<td>A framing error occurs when a low stop bit is detected. When two stop bits are used, only the first stop bit is checked for framing error. When a framing error is detected, the FE bit is set.</td>
</tr>
<tr>
<td>Parity error</td>
<td>A parity error is a mismatch between the number of 1s in a character and the value of the parity bit. When an address bit is included in the character, it is included in the parity calculation. When a parity error is detected, the PE bit is set.</td>
</tr>
<tr>
<td>Receive overrun error</td>
<td>An overrun error occurs when a character is loaded into UxRXBUF before the prior character has been read. When an overrun occurs, the OE bit is set.</td>
</tr>
<tr>
<td>Break condition</td>
<td>A break condition is a period of 10 or more low bits received on URXDx after a missing stop bit. When a break condition is detected, the BRK bit is set. A break condition can also set the interrupt flag URXIFGx when URXEIE = 0.</td>
</tr>
</tbody>
</table>

When URXEIE = 0 and a framing error, parity error, or break condition is detected, no character is received into UxRXBUF. When URXEIE = 1, characters are received into UxRXBUF and any applicable error bit is set.

When any of the FE, PE, OE, BRK, or RXERR bits are set, the bit remains set until user software resets it or UxRXBUF is read.

18.2.4 USART Receive Enable

The receive enable bit, URXEx, enables or disables data reception on URXDx as shown in Figure 18-5. Disabling the USART receiver stops the receive operation following completion of any character currently being received or immediately if no receive operation is active. The receive-data buffer, UxRXBUF, contains the character moved from the RX shift register after the character is received.

![State Diagram of Receiver Enable](image)

**Figure 18-5. State Diagram of Receiver Enable**

**Note**

Re-Enabling the Receiver (Setting URXEx): UART Mode

When the receiver is disabled (URXEx = 0), re-enabling the receiver (URXEx = 1) is asynchronous to any data stream that may be present on URXDx at the time. Synchronization can be performed by testing for an idle line condition before receiving a valid character (see URXWIE).

18.2.5 USART Transmit Enable

When UTXEx is set, the UART transmitter is enabled. Transmission is initiated by writing data to UxTXBUF. The data is then moved to the transmit shift register on the next BITCLK after the TX shift register is empty, and transmission begins. This process is shown in Figure 18-6.

When the UTXEx bit is reset the transmitter is stopped. Any data moved to UxTXBUF and any active transmission of data currently in the transmit shift register prior to clearing UTXEx continue until all data transmission is completed.
When the transmitter is enabled (UTXEx = 1), data should not be written to UxTXBUF unless it is ready for new data indicated by UTXIFGx = 1. Violation can result in an erroneous transmission if data in UxTXBUF is modified as it is being moved into the TX shift register.

It is recommended that the transmitter be disabled (UTXEx = 0) only after any active transmission is complete. This is indicated by a set transmitter empty bit (TXEPT = 1). Any data written to UxTXBUF while the transmitter is disabled are held in the buffer but are not moved to the transmit shift register or transmitted. Once UTXEx is set, the data in the transmit buffer is immediately loaded into the transmit shift register and character transmission resumes.

18.2.6 USART Baud Rate Generation

The USART baud rate generator is capable of producing standard baud rates from non-standard source frequencies. The baud rate generator uses one prescaler/divider and a modulator as shown in Figure 18-7. This combination supports fractional divisors for baud rate generation. The maximum USART baud rate is one-third the UART source clock frequency BRCLK.

Timing for each bit is shown in Figure 18-8. For each bit received, a majority vote is taken to determine the bit value. These samples occur at the N/2-1, N/2, and N/2+1 BRCLK periods, where N is the number of BRCLKs per BITCLK.
18.2.6.1 Baud Rate Bit Timing

The first stage of the baud rate generator is the 16-bit counter and comparator. At the beginning of each bit transmitted or received, the counter is loaded with \( \text{INT}(N/2) \) where \( N \) is the value stored in the combination of \( U_{xBR0} \) and \( U_{xBR1} \). The counter reloads \( \text{INT}(N/2) \) for each bit period half-cycle, giving a total bit period of \( N \) BRCLKs. For a given BRCLK clock source, the baud rate used determines the required division factor \( N \):

\[
N = \frac{\text{BRCLK}}{\text{Baud Rate}}
\]

The division factor \( N \) is often a non-integer value of which the integer portion can be realized by the prescaler/divider. The second stage of the baud rate generator, the modulator, is used to meet the fractional part as closely as possible. The factor \( N \) is then defined as:

\[
N = U_{xBR} + \frac{1}{n} \sum_{i=0}^{n-1} m_i
\]

Where,

\( N = \text{Target division factor} \)

\( U_{xBR} = \text{16-bit representation of registers } U_{xBR0} \text{ and } U_{xBR1} \)

\( i = \text{Bit position in the character} \)

\( n = \text{Total number of bits in the character} \)

\( m_i = \text{Data of each corresponding modulation bit (1 or 0)} \)

The BITCLK can be adjusted from bit to bit with the modulator to meet timing requirements when a non-integer divisor is needed. Timing of each bit is expanded by one BRCLK clock cycle if the modulator bit \( m_i \) is set. Each time a bit is received or transmitted, the next bit in the modulation control register determines the timing for that bit. A set modulation bit increases the division factor by one while a cleared modulation bit maintains the division factor given by \( U_{xBR} \).

The timing for the start bit is determined by \( U_{xBR} + m_0 \), the next bit is determined by \( U_{xBR} + m_1 \), and so on. The modulation sequence begins with the LSB. When the character is greater than 8 bits, the modulation sequence restarts with \( m_0 \) and continues until all bits are processed.
18.2.6.2 Determining the Modulation Value

Determining the modulation value is an interactive process. Using the timing error formula provided, beginning with the start bit, the individual bit errors are calculated with the corresponding modulator bit set and cleared. The modulation bit setting with the lower error is selected and the next bit error is calculated. This process is continued until all bit errors are minimized. When a character contains more than 8 bits, the modulation bits repeat. For example, the ninth bit of a character uses modulation bit 0.

18.2.6.3 Transmit Bit Timing

The timing for each character is the sum of the individual bit timings. By modulating each bit, the cumulative bit error is reduced. The individual bit error can be calculated by:

\[ \text{Error} [\%] = \left( \frac{\text{baud rate}}{\text{BRCLK}} \right) \times \left[ (j + 1) \times UxBR + \sum_{i=0}^{m} m_i - (j + 1) \right] \times 100\% \]

Where,

- baud rate = Desired baud rate
- BRCLK = Input frequency - UCLKI, ACLK, or SMCLK
- j = Bit position - 0 for the start bit, 1 for data bit D0, and so on
- UxBR = Division factor in registers UxBR1 and UxBR0

For example, the transmit errors for the following conditions are calculated:

Baud rate = 2400
BRCLK = 32 768 Hz (ACLK)
UxBR = 13, since the ideal division factor is 13.65
UXMCTL = 6Bh: m7 = 0, m6 = 1, m5 = 1, m4 = 0, m3 = 1, m2 = 0, m1 = 1, and m0 = 1. The LSB of UxMCTL is used first.
Start bit Error [%] = \(\frac{\text{baud rate}}{\text{BRCLK}} \left( (0+1) \cdot UxBR+1 \right) \cdot 100\% = 2.54\%\)

Data bit D0 Error [%] = \(\frac{\text{baud rate}}{\text{BRCLK}} \left( (1+1) \cdot UxBR+2 \right) \cdot 100\% = 5.08\%\)

Data bit D1 Error [%] = \(\frac{\text{baud rate}}{\text{BRCLK}} \left( (2+1) \cdot UxBR+2 \right) \cdot 100\% = 0.29\%\)

Data bit D2 Error [%] = \(\frac{\text{baud rate}}{\text{BRCLK}} \left( (3+1) \cdot UxBR+3 \right) \cdot 100\% = 2.83\%\)

Data bit D3 Error [%] = \(\frac{\text{baud rate}}{\text{BRCLK}} \left( (4+1) \cdot UxBR+3 \right) \cdot 100\% = 1.95\%\)

Data bit D4 Error [%] = \(\frac{\text{baud rate}}{\text{BRCLK}} \left( (5+1) \cdot UxBR+4 \right) \cdot 100\% = 0.59\%\)

Data bit D5 Error [%] = \(\frac{\text{baud rate}}{\text{BRCLK}} \left( (6+1) \cdot UxBR+5 \right) \cdot 100\% = 3.13\%\)

Data bit D6 Error [%] = \(\frac{\text{baud rate}}{\text{BRCLK}} \left( (7+1) \cdot UxBR+5 \right) \cdot 100\% = 1.66\%\)

Data bit D7 Error [%] = \(\frac{\text{baud rate}}{\text{BRCLK}} \left( (8+1) \cdot UxBR+6 \right) \cdot 100\% = 0.88\%\)

Parity bit Error [%] = \(\frac{\text{baud rate}}{\text{BRCLK}} \left( (9+1) \cdot UxBR+7 \right) \cdot 100\% = 3.42\%\)

Stop bit 1 Error [%] = \(\frac{\text{baud rate}}{\text{BRCLK}} \left( (10+1) \cdot UxBR+7 \right) \cdot 100\% = 1.37\%\)

The results show the maximum per-bit error to be 5.08% of a BITCLK period.
18.2.6.4 Receive Bit Timing

Receive timing is subject to two error sources. The first is the bit-to-bit timing error. The second is the error between a start edge occurring and the start edge being accepted by the USART. Figure 18-9 shows the asynchronous timing errors between data on the URXDx pin and the internal baud-rate clock.

The ideal start bit timing \( t_{\text{ideal}(0)} \) is half the baud-rate timing \( t_{\text{baudrate}} \), because the bit is tested in the middle of its period. The ideal baud-rate timing \( t_{\text{ideal}(i)} \) for the remaining character bits is the baud rate timing \( t_{\text{baudrate}} \). The individual bit errors can be calculated by:

\[
\text{Error} \, [\%] = \left( \frac{\text{baud rate}}{\text{BRCLK}} \right) \times \left( 2 \times \left( m0 + \text{int} \left( \frac{\text{UxBR}}{2} \right) \right) + \left( i \times \text{UxBR} + \sum_{i=1} m_i \right) - 1 - j \right) \times 100\%
\]

Where,

- baud rate = the required baud rate
- \( \text{BRCLK} \) = the input frequency; selected for UCLK, ACLK, or SMCLK
- \( j = 0 \) for the start bit, 1 for data bit D0, and so on
- \( \text{UxBR} \) = the division factor in registers UxBR1 and UxBR0

For example, the receive errors for the following conditions are calculated:

- Baud rate = 2400
- \( \text{BRCLK} = 32 \, 768 \, \text{Hz (ACLK)} \)
- \( \text{UxBR} = 13 \), since the ideal division factor is 13.65
- \( \text{UxMCTL} = 6B: m7 = 0, m6 = 1, m5 = 1, m4 = 0, m3 = 1, m2 = 0, m1 = 1 \) and \( m0 = 1 \). The LSB of UxMCTL is used first.

Figure 18-9. Receive Error

The receive errors are calculated using the formula above with the given conditions.
Data bit D1 Error [%] = \left( \frac{\text{baud rate}}{\text{BRCLK}} \right) [2x(1+6)+2 \cdot \text{UxBR}+1]-1-2 \cdot 100\% = 0.29\%

Data bit D2 Error [%] = \left( \frac{\text{baud rate}}{\text{BRCLK}} \right) [2x(1+6)+3 \cdot \text{UxBR}+2]-1-3 \cdot 100\% = 2.83\%

Data bit D3 Error [%] = \left( \frac{\text{baud rate}}{\text{BRCLK}} \right) [2x(1+6)+4 \cdot \text{UxBR}+3]-1-4 \cdot 100\% = -1.95\%

Data bit D4 Error [%] = \left( \frac{\text{baud rate}}{\text{BRCLK}} \right) [2x(1+6)+5 \cdot \text{UxBR}+4]-1-5 \cdot 100\% = 0.59\%

Data bit D5 Error [%] = \left( \frac{\text{baud rate}}{\text{BRCLK}} \right) [2x(1+6)+6 \cdot \text{UxBR}+5]-1-6 \cdot 100\% = 3.13\%

Data bit D6 Error [%] = \left( \frac{\text{baud rate}}{\text{BRCLK}} \right) [2x(1+6)+7 \cdot \text{UxBR}+6]-1-7 \cdot 100\% = -1.66\%

Data bit D7 Error [%] = \left( \frac{\text{baud rate}}{\text{BRCLK}} \right) [2x(1+6)+8 \cdot \text{UxBR}+7]-1-8 \cdot 100\% = 0.88\%

Parity bit Error [%] = \left( \frac{\text{baud rate}}{\text{BRCLK}} \right) [2x(1+6)+9 \cdot \text{UxBR}+8]-1-9 \cdot 100\% = 3.42\%

Stop bit 1 Error [%] = \left( \frac{\text{baud rate}}{\text{BRCLK}} \right) [2x(1+6)+10 \cdot \text{UxBR}+9]-1-10 \cdot 100\% = -1.37\%

Start bit Error [%] = \left( \frac{\text{baud rate}}{\text{BRCLK}} \right) [2x(1+6)+0 \cdot \text{UxBR}+10]-1-0 \cdot 100\% = 2.54\%

Data bit D0 Error [%] = \left( \frac{\text{baud rate}}{\text{BRCLK}} \right) [2x(1+6)+1 \cdot \text{UxBR}+1]-1-1 \cdot 100\% = 5.08\%

The results show the maximum per-bit error to be 5.08% of a BITCLK period.

18.2.6.5 Typical Baud Rates and Errors

Standard baud rate frequency data for UxBRx and UxMCTL are listed in Table 18-2 for a 32 768-Hz watch crystal (ACLK) and a typical 1 048 576-Hz SMCLK.

The receive error is the accumulated time versus the ideal scanning time in the middle of each bit. The transmit error is the accumulated timing error versus the ideal time of the bit period.

### Table 18-2. Commonly Used Baud Rates, Baud Rate Data, and Errors

<table>
<thead>
<tr>
<th>Baud Rate</th>
<th>Divide by</th>
<th>A: BRCLK = 32 768 Hz</th>
<th>Max TX Error %</th>
<th>Max RX Error %</th>
<th>Synchron RX Error %</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>A:</td>
<td>B: UxBR1 UxBR0 UxMCTL</td>
<td>UxBR1 UxBR0 UxMCTL</td>
<td>UxBR1 UxBR0 UxMCTL</td>
<td>UxBR1 UxBR0 UxMCTL</td>
</tr>
<tr>
<td>1200</td>
<td>27.31</td>
<td>873.81 0 1B 03 4-3 4-3 ±2 03 69 FF 0/0.3 ±2</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>2400</td>
<td>13.65</td>
<td>436.91 0 0D 6B -6/3 -6/3 ±4 01 B4 FF 0/0.3 ±2</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>4800</td>
<td>6.83</td>
<td>218.45 0 06 6F -9/11 -9/11 ±7 0 0A DA 55 0/0.4 ±2</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>9600</td>
<td>3.41</td>
<td>109.23 0 03 4A -21/12 -21/12 ±15 0 0B 03 -0.4/1 ±2</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>19 200</td>
<td>54.61</td>
<td>0 36 6B -0.2/2 ±2</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>38 400</td>
<td>27.31</td>
<td>0 0D 6B -4/3 ±2</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>76 800</td>
<td>13.65</td>
<td>0 09 08 -5/7 ±7</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>115 200</td>
<td>9.1</td>
<td>0 09 08 -5/7 ±7</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

18.2.7 USART Interrupts

The USART has one interrupt vector for transmission and one interrupt vector for reception.
18.2.7.1 USART Transmit Interrupt Operation

The UTXIFGx interrupt flag is set by the transmitter to indicate that UxTXBUF is ready to accept another character. An interrupt request is generated if UTXIEx and GIE are also set. UTXIFGx is automatically reset if the interrupt request is serviced or if a character is written to UxTXBUF.

UTXIFGx is set after a PUC or when SWRST = 1. UTXIEx is reset after a PUC or when SWRST = 1. The operation is shown in Figure 18-10.

![Figure 18-10. Transmit Interrupt Operation](image)

18.2.7.2 USART Receive Interrupt Operation

The URXIFGx interrupt flag is set each time a character is received and loaded into UxRXBUF. An interrupt request is generated if URXIEx and GIE are also set. URXIFGx and URXIEx are reset by a system reset PUC signal or when SWRST = 1. URXIFGx is automatically reset if the pending interrupt is served (when URXSE = 0) or when UxRXBUF is read. The operation is shown in Figure 18-11.

![Figure 18-11. Receive Interrupt Operation](image)

URXEIE is used to enable or disable erroneous characters from setting URXIFGx. When using multiprocessor addressing modes, URXWIE is used to auto-detect valid address characters and reject unwanted data characters.

Two types of characters do not set URXIFGx:

- ERRONEOUS CHARACTER REJECTION
- NON-ADDRESS CHARACTER REJECTION
- Erroneous characters when URXEIE = 0
- Non-address characters when URXWIE = 1

When URXEIE = 1 a break condition sets the BRK bit and the URXIFGx flag.

### 18.2.7.3 Receive-Start Edge Detect Operation

The URXSE bit enables the receive start-edge detection feature. The recommended usage of the receive-start edge feature is when BRCLK is sourced by the DCO and when the DCO is off because of low-power mode operation. The ultra-fast turn-on of the DCO allows character reception after the start edge detection.

When URXS, URXIEx and GIE are set and a start edge occurs on URXDx, the internal signal URXS is set. When URXS is set, a receive interrupt request is generated but URXIFGx is not set. User software in the receive interrupt service routine can test URXIFGx to determine the source of the interrupt. When URXIFGx = 0 a start edge was detected, and when URXIFGx = 1 a valid character (or break) was received.

When the ISR determines the interrupt request was from a start edge, user software toggles URXSE, and must enable the BRCLK source by returning from the ISR to active mode or to a low-power mode where the source is active. If the ISR returns to a low-power mode where the BRCLK source is inactive, the character is not received. Toggling URXSE clears the URXS signal and re-enables the start edge detect feature for future characters. See chapter *System Resets, Interrupts, and Operating Modes* for information on entering and exiting low-power modes.

The now active BRCLK allows the USART to receive the balance of the character. After the full character is received and moved to UxRXBUF, URXIFGx is set and an interrupt service is again requested. Upon ISR entry, URXIFGx = 1 indicating a character was received. The URXIFGx flag is cleared when user software reads UxRXBUF.

```
; Interrupt handler for start condition and
; Character receive. BRCLK = DCO.
U0RX_Int BIT.B #URXIFG0,&IFG1 ; Test URXIFGx to determine
     JZ     ST_COND           ; If start or character
     MOV.B &UxRXBUF,dst      ; Read buffer
     ...                  ;
     RETI                   ;
ST_COND BIC.B #URXSE,&U0TCTL ; Clear URXS signal
BIS.B #URXSE,&U0TCTL      ; Re-enable edge detect
BIC  #SCG0+SCG1,0(SP)    ; Enable BRCLK = DCO
RETI                     ;
```

---

**Note**

**Break Detect With Halted UART Clock**

When using the receive start-edge detect feature, a break condition cannot be detected when the BRCLK source is off.
18.2.7.4 Receive-Start Edge Detect Conditions

When URXSE = 1, glitch suppression prevents the USART from being accidentally started. Any low-level on URXDx shorter than the deglitch time $t_t$ (approximately 300 ns) is ignored by the USART and no interrupt request is generated (see Figure 18-12). See the device-specific data sheet for parameters.

![Figure 18-12. Glitch Suppression, USART Receive Not Started](image)

When a glitch is longer than $t_t$ or a valid start bit occurs on URXDx, the USART receive operation is started and a majority vote is taken as shown in Figure 18-13. If the majority vote fails to detect a start bit, the USART halts character reception.

If character reception is halted, an active BRCLK is not necessary. A time-out period longer than the character receive duration can be used by software to indicate that a character was not received in the expected time, and the software can disable BRCLK.

![Figure 18-13. Glitch Suppression, USART Activated](image)
18.3 USART Registers – UART Mode

Table 18-3 lists the memory-mapped registers for USART0 and USART1 in UART mode.

Table 18-3. USARTx Registers – UART Mode

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>70h</td>
<td>U0CTL</td>
<td>USART0 control</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 18.3.1</td>
</tr>
<tr>
<td>71h</td>
<td>U0TCTL</td>
<td>USART0 transmit control</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 18.3.2</td>
</tr>
<tr>
<td>72h</td>
<td>U0RCTL</td>
<td>USART0 receive control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 18.3.3</td>
</tr>
<tr>
<td>73h</td>
<td>U0MCTL</td>
<td>USART0 modulation control</td>
<td>Read/write</td>
<td></td>
<td>Section 18.3.4</td>
</tr>
<tr>
<td>74h</td>
<td>U0BR0</td>
<td>USART0 baud-rate control 0</td>
<td>Read/write</td>
<td></td>
<td>Section 18.3.5</td>
</tr>
<tr>
<td>75h</td>
<td>U0BR1</td>
<td>USART0 baud-rate control 1</td>
<td>Read/write</td>
<td></td>
<td>Section 18.3.5</td>
</tr>
<tr>
<td>76h</td>
<td>U0RXBUF</td>
<td>USART0 receive buffer</td>
<td>Read</td>
<td>Unchanged</td>
<td>Section 18.3.7</td>
</tr>
<tr>
<td>77h</td>
<td>U0TXBUF</td>
<td>USART0 transmit buffer</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 18.3.8</td>
</tr>
<tr>
<td>0h</td>
<td>IE1</td>
<td>SFR interrupt enable 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 18.3.9</td>
</tr>
<tr>
<td>2h</td>
<td>IFG1</td>
<td>SFR interrupt flag 1</td>
<td>Read/write</td>
<td>82h with PUC</td>
<td>Section 18.3.11</td>
</tr>
<tr>
<td>78h</td>
<td>U1CTL</td>
<td>USART1 control</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 18.3.1</td>
</tr>
<tr>
<td>79h</td>
<td>U1TCTL</td>
<td>USART1 transmit control</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 18.3.2</td>
</tr>
<tr>
<td>7Ah</td>
<td>U1RCTL</td>
<td>USART1 receive control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 18.3.3</td>
</tr>
<tr>
<td>78h</td>
<td>U1MCTL</td>
<td>USART1 modulation control</td>
<td>Read/write</td>
<td></td>
<td>Section 18.3.4</td>
</tr>
<tr>
<td>7Ch</td>
<td>U1BR0</td>
<td>USART1 baud-rate control 0</td>
<td>Read/write</td>
<td></td>
<td>Section 18.3.5</td>
</tr>
<tr>
<td>7Dh</td>
<td>U1BR1</td>
<td>USART1 baud-rate control 1</td>
<td>Read/write</td>
<td></td>
<td>Section 18.3.5</td>
</tr>
<tr>
<td>7Eh</td>
<td>U1RXBUF</td>
<td>USART1 receive buffer</td>
<td>Read</td>
<td>Unchanged</td>
<td>Section 18.3.7</td>
</tr>
<tr>
<td>7Fh</td>
<td>U1TXBUF</td>
<td>USART1 transmit buffer</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 18.3.8</td>
</tr>
<tr>
<td>1h</td>
<td>IE2</td>
<td>SFR interrupt enable 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 18.3.10</td>
</tr>
<tr>
<td>3h</td>
<td>IFG2</td>
<td>SFR interrupt flag 2</td>
<td>Read/write</td>
<td>20h with PUC</td>
<td>Section 18.3.12</td>
</tr>
</tbody>
</table>

Note

Modifying SFR bits

To avoid modifying control bits of other modules, TI recommends setting or clearing the IEx and IFGx bits using BIS.B or BIC.B instructions, rather than MOV.B or CLR.B instructions.
18.3.1 UxCTL Register

USARTx Control Register

UxCTL is shown in Figure 18-14 and described in Table 18-4.

Return to Table 18-3.

Figure 18-14. UxCTL Register

<p>| | | | | | | | |</p>
<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>6</td>
<td>5</td>
<td>4</td>
<td>3</td>
<td>2</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td><strong>PENA</strong></td>
<td><strong>PEV</strong></td>
<td><strong>SPB</strong></td>
<td><strong>CHAR</strong></td>
<td><strong>LISTEN</strong></td>
<td><strong>SYNC</strong></td>
<td><strong>MM</strong></td>
<td><strong>SWRST</strong></td>
</tr>
<tr>
<td>rw-0</td>
<td>rw-0</td>
<td>rw-0</td>
<td>rw-0</td>
<td>rw-0</td>
<td>rw-0</td>
<td>rw-0</td>
<td>rw-1</td>
</tr>
</tbody>
</table>

Table 18-4. UxCTL Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
</table>
| 7   | PENA  | R/W  | 0h    | Parity enable  
|     |       |      |       | 0b = Parity disabled  
|     |       |      |       | 1b = Parity enabled. Parity bit is generated (UTXDx) and expected (URXDx). In address-bit multiprocessor mode, the address bit is included in the parity calculation. |
| 6   | PEV   | R/W  | 0h    | Parity select. PEV is not used when parity is disabled.  
|     |       |      |       | 0b = Odd parity  
|     |       |      |       | 1b = Even parity |
| 5   | SPB   | R/W  | 0h    | Stop bit select. Number of stop bits transmitted. The receiver always checks for one stop bit.  
|     |       |      |       | 0b = One stop bit  
|     |       |      |       | 1b = Two stop bits |
| 4   | CHAR  | R/W  | 0h    | Character length. Selects 7-bit or 8-bit character length.  
|     |       |      |       | 0b = 7-bit data  
|     |       |      |       | 1b = 8-bit data |
| 3   | LISTEN | R/W | 0h | Listen enable. The LISTEN bit selects the loopback mode  
|     |       |      |   | 0b = Disabled  
|     |       |      |   | 1b = Enabled. UTXDx is internally fed back to the receiver. |
| 2   | SYNC  | R/W  | 0h    | Synchronous mode enable  
|     |       |      |       | 0b = UART mode  
|     |       |      |       | 1b = SPI mode |
| 1   | MM    | R/W  | 0h    | Multiprocessor mode select  
|     |       |      |       | 0b = Idle-line multiprocessor protocol  
|     |       |      |       | 1b = Address-bit multiprocessor protocol |
| 0   | SWRST | R/W  | 1h    | Software reset enable  
|     |       |      |       | 0b = Disabled. USART reset released for operation.  
|     |       |      |       | 1b = Enabled. USART logic held in reset state. |
18.3.2 UxTCTL Register

USARTx Transmit Control Register

UxTCTL is shown in Figure 18-15 and described in Table 18-5.

Return to Table 18-3.

![Figure 18-15. UxTCTL Register](image_url)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td>Unused</td>
</tr>
<tr>
<td>6</td>
<td>CKPL</td>
<td>R/W</td>
<td>0h</td>
<td>Clock polarity select&lt;br&gt;0b = UCLKI = UCLK&lt;br&gt;1b = UCLKI = Inverted UCLK</td>
</tr>
<tr>
<td>5-4</td>
<td>SSELx</td>
<td>R/W</td>
<td>0h</td>
<td>Source select. These bits select the BRCLK source clock.&lt;br&gt;00b = UCLKI&lt;br&gt;01b = ACLK&lt;br&gt;10b = SMCLK&lt;br&gt;11b = SMCLK</td>
</tr>
<tr>
<td>3</td>
<td>URXSE</td>
<td>R/W</td>
<td>0h</td>
<td>UART receive start-edge. The bit enables the UART receive start-edge feature.&lt;br&gt;0b = Disabled&lt;br&gt;1b = Enabled</td>
</tr>
<tr>
<td>2</td>
<td>TXWAKE</td>
<td>R/W</td>
<td>0h</td>
<td>Transmitter wake&lt;br&gt;0b = Next frame transmitted is data&lt;br&gt;1b = Next frame transmitted is an address</td>
</tr>
<tr>
<td>1</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td>Unused</td>
</tr>
<tr>
<td>0</td>
<td>TXEPT</td>
<td>R/W</td>
<td>1h</td>
<td>Transmitter empty flag.&lt;br&gt;0b = UART is transmitting data and/or data is waiting in UxTXBUF&lt;br&gt;1b = Transmitter shift register and UxTXBUF are empty or SWRST = 1</td>
</tr>
</tbody>
</table>
18.3.3 UxRCTL Register

USARTx Receive Control Register

UxRCTL is shown in Figure 18-16 and described in Table 18-6.

Return to Table 18-3.

Table 18-6. UxRCTL Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>FE</td>
<td>R/W</td>
<td>0h</td>
<td>Framing error flag.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No error</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Character received with low stop bit</td>
</tr>
<tr>
<td>6</td>
<td>PE</td>
<td>R/W</td>
<td>0h</td>
<td>Parity error flag. When PENA = 0, PE is read as 0.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No error</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Character received with parity error</td>
</tr>
<tr>
<td>5</td>
<td>OE</td>
<td>R/W</td>
<td>0h</td>
<td>Overrun error flag. This bit is set when a character is transferred into UxRXBUF before the previous character was read.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No error</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Overrun error occurred</td>
</tr>
<tr>
<td>4</td>
<td>BRK</td>
<td>R/W</td>
<td>0h</td>
<td>Break detect flag</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No break condition</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Break condition occurred</td>
</tr>
<tr>
<td>3</td>
<td>URXEIE</td>
<td>R/W</td>
<td>0h</td>
<td>Receive erroneous-character interrupt-enable</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Erroneous characters rejected and URXIFGx is not set</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Erroneous characters received set URXIFGx</td>
</tr>
<tr>
<td>2</td>
<td>URXWIE</td>
<td>R/W</td>
<td>0h</td>
<td>Receive wake-up interrupt-enable. This bit enables URXIFGx to be set when an address character is received. When URXEIE = 0, an address character does not set URXIFGx if it is received with errors.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = All received characters set URXIFGx</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Only received address characters set URXIFGx</td>
</tr>
<tr>
<td>1</td>
<td>RXWAKE</td>
<td>R/W</td>
<td>0h</td>
<td>Receive wake-up flag</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Received character is data</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Received character is an address</td>
</tr>
<tr>
<td>0</td>
<td>RXERR</td>
<td>R/W</td>
<td>0h</td>
<td>Receive error flag. This bit indicates a character was received with errors. When RXERR = 1, one or more error flags (FE, PE, OE, BRK) is also set. RXERR is cleared when UxRXBUF is read.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No receive errors detected</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Receive error detected</td>
</tr>
</tbody>
</table>
18.3.4 UxMCTL Register

USARTx Modulation Control Register

UxMCTL is shown in Figure 18-17 and described in Table 18-7.

Return to Table 18-3.

Figure 18-17. UxMCTL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UxMCTLx</td>
<td>R/W</td>
<td>Unchanged</td>
<td>Modulation bits. These bits select the modulation for BRCLK.</td>
</tr>
</tbody>
</table>

18.3.5 UxBR0 Register

USARTx Baud-Rate Control 0 Register

UxBR0 is shown in Figure 18-18 and described in Table 18-8.

Return to Table 18-3.

Figure 18-18. UxBR0 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UxBRx</td>
<td>R/W</td>
<td>Unchanged</td>
<td>The valid baud-rate control range is 3 ≤ UxBR ≤ 0FFFFh, where UxBR = (UxBR1 + UxBR0). Unpredictable receive and transmit timing occurs if UxBR &lt; 3.</td>
</tr>
</tbody>
</table>

18.3.6 UxBR1 Register

USARTx Baud-Rate Control 1 Register

UxBR1 is shown in Figure 18-18 and described in Table 18-8.

Return to Table 18-3.

Figure 18-19. UxBR1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UxBRx</td>
<td>R/W</td>
<td>Unchanged</td>
<td>The valid baud-rate control range is 3 ≤ UxBR ≤ 0FFFFh, where UxBR = (UxBR1 + UxBR0). Unpredictable receive and transmit timing occurs if UxBR &lt; 3.</td>
</tr>
</tbody>
</table>
18.3.7 UxRXBUF Register

USARTx Receive Buffer Register

UxRXBUF is shown in Figure 18-20 and described in Table 18-10.
Return to Table 18-3.

Figure 18-20. UxRXBUF Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UxRXBUFx</td>
<td>R</td>
<td>Unchanged</td>
<td>The receive-data buffer is user accessible and contains the last received character from the receive shift register. Reading UxRXBUF resets the OE bit and URXIFGx flag. In 7-bit data mode, UxRXBUF is LSB justified and the MSB is always reset.</td>
</tr>
</tbody>
</table>

18.3.8 UxTXBUF Register

USARTx Transmit Buffer Register

UxTXBUF is shown in Figure 18-21 and described in Table 18-11.
Return to Table 18-3.

Figure 18-21. UxTXBUF Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UxTXBUFx</td>
<td>R/W</td>
<td>Unchanged</td>
<td>The transmit data buffer is user accessible and holds the data waiting to be moved into the transmit shift register and transmitted on UTXDx. Writing to the transmit data buffer clears UTXIFGx. The MSB of UxTXBUF is not used for 7-bit data and is reset.</td>
</tr>
</tbody>
</table>
18.3.9 IE1 Register

SFR Interrupt Enable 1 Register

IE1 is shown in Figure 18-22 and described in Table 18-12.

Return to Table 18-3.

Figure 18-22. IE1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>UTXIE0</td>
<td>R/W</td>
<td>0h</td>
<td>USART0 transmit interrupt enable. This bit enables the UTXIFG0 interrupt. 0b = Interrupt not enabled 1b = Interrupt enabled</td>
</tr>
<tr>
<td>6</td>
<td>URXIE0</td>
<td>R/W</td>
<td>0h</td>
<td>USART0 receive interrupt enable. This bit enables the URXIFG0 interrupt. 0b = Interrupt not enabled 1b = Interrupt enabled</td>
</tr>
<tr>
<td>5-0</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See the device-specific data sheet.</td>
</tr>
</tbody>
</table>

Table 18-12. IE1 Register Field Descriptions

18.3.10 IE2 Register

SFR Interrupt Enable 2 Register

IE2 is shown in Figure 18-23 and described in Table 18-13.

Return to Table 18-3.

Figure 18-23. IE2 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See the device-specific data sheet.</td>
</tr>
<tr>
<td>5</td>
<td>UTXIE1</td>
<td>R/W</td>
<td>0h</td>
<td>USART1 transmit interrupt enable. This bit enables the UTXIFG1 interrupt. 0b = Interrupt not enabled 1b = Interrupt enabled</td>
</tr>
<tr>
<td>4</td>
<td>URXIE1</td>
<td>R/W</td>
<td>0h</td>
<td>USART1 receive interrupt enable. This bit enables the URXIFG1 interrupt. 0b = Interrupt not enabled 1b = Interrupt enabled</td>
</tr>
<tr>
<td>3-0</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See the device-specific data sheet.</td>
</tr>
</tbody>
</table>

Table 18-13. IE2 Register Field Descriptions
18.3.11 IFG1 Register
SFR Interrupt Flag 1 Register
IFG1 is shown in Figure 18-24 and described in Table 18-14.
Return to Table 18-3.

Figure 18-24. IFG1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>UTXIFG0</td>
<td>R/W</td>
<td>1h</td>
<td>USART0 transmit interrupt flag. UTXIFG0 is set when U0TXBUF is empty. 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
<tr>
<td>6</td>
<td>URXIFG0</td>
<td>R/W</td>
<td>0h</td>
<td>USART0 receive interrupt flag. URXIFG0 is set when U0RXBUF has received a complete character. 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
<tr>
<td>5-0</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See the device-specific data sheet.</td>
</tr>
</tbody>
</table>

18.3.12 IFG2 Register
SFR Interrupt Flag 2 Register
IFG2 is shown in Figure 18-25 and described in Table 18-15.
Return to Table 18-3.

Figure 18-25. IFG2 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See the device-specific data sheet.</td>
</tr>
<tr>
<td>5</td>
<td>UTXIFG1</td>
<td>R/W</td>
<td>1h</td>
<td>USART1 transmit interrupt flag. UTXIFG1 is set when U1TXBUF is empty. 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
<tr>
<td>4</td>
<td>URXIFG1</td>
<td>R/W</td>
<td>0h</td>
<td>USART1 receive interrupt flag. URXIFG1 is set when U1RXBUF has received a complete character. 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
<tr>
<td>3-0</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See the device-specific data sheet.</td>
</tr>
</tbody>
</table>
The universal synchronous/asynchronous receive/transmit (USART) peripheral interface supports two serial modes with one hardware module. This chapter discusses the operation of the synchronous peripheral interface or SPI mode. USART0 is implemented on the MSP430AFE2xx devices.

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19.2 USART Operation: SPI Mode ................................................................. 537
19.3 USART Registers: SPI Mode ................................................................. 544
19.1 USART Introduction: SPI Mode

In synchronous mode, the USART connects the MSP430 to an external system via three or four pins: SIMO, SOMI, UCLK, and STE. SPI mode is selected when the SYNC bit is set and the I2C bit is cleared.

SPI mode features include:

- 7-bit or 8-bit data length
- 3-pin and 4-pin SPI operation
- Master or slave modes
- Independent transmit and receive shift registers
- Separate transmit and receive buffer registers
- Selectable UCLK polarity and phase control
- Programmable UCLK frequency in master mode
- Independent interrupt capability for receive and transmit

Figure 19-1 shows the USART when configured for SPI mode.

![USART Block Diagram: SPI Mode](image-url)

* See the device-specific data sheet for SFR locations.

Figure 19-1. USART Block Diagram: SPI Mode
19.2 USART Operation: SPI Mode

In SPI mode, serial data is transmitted and received by multiple devices using a shared clock provided by the master. An additional pin, STE, is provided as to enable a device to receive and transmit data and is controlled by the master.

Three or four signals are used for SPI data exchange:

- **SIMO**: Slave in, master out
  - Master mode: SIMO is the data output line.
  - Slave mode: SIMO is the data input line.
- **SOMI**: Slave out, master in
  - Master mode: SOMI is the data input line.
  - Slave mode: SOMI is the data output line.
- **UCLK**: USART SPI clock
  - Master mode: UCLK is an output.
  - Slave mode: UCLK is an input.
- **STE**: Slave transmit enable. Used in 4-pin mode to allow multiple masters on a single bus. Not used in 3-pin mode.
  - 4-pin master mode:
    - When STE is high, SIMO and UCLK operate normally.
    - When STE is low, SIMO and UCLK are set to the input direction.
  - 4-pin slave mode:
    - When STE is high, RX/TX operation of the slave is disabled and SOMI is forced to the input direction.
    - When STE is low, RX/TX operation of the slave is enabled and SOMI operates normally.

19.2.1 USART Initialization and Reset

The USART is reset by a PUC or by the SWRST bit. After a PUC, the SWRST bit is automatically set, keeping the USART in a reset condition. When set, the SWRST bit resets the URXIE, UTXIE, URXIFGx, OE, and FE bits and sets the UTXIFGx flag. The USPIEx bit is not altered by SWRST. Clearing SWRST releases the USART for operation.

### Note

**Initializing or Reconfiguring the USART Module**

The required USART initialization/reconfiguration process is:

1. Set SWRST (BIS.B #SWRST,&UxCTL)
2. Initialize all USART registers with SWRST=1 (including UxCTL)
3. Enable USART module via the MEx SFRs (USPIEx)
4. Clear SWRST via software (BIC.B #SWRST,&UxCTL)
5. Enable interrupts (optional) via the IEx SFRs (URXIE and/or UTXIE)

Failure to follow this process may result in unpredictable USART behavior.
### 19.2.2 Master Mode

Figure 19-2 shows the USART as a master in both 3-pin and 4-pin configurations. The USART initiates a data transfer when data is moved to the transmit data buffer UxTXBUF. The UxTXBUF data is moved to the TX shift register when the TX shift register is empty, initiating data transfer on SIMO starting with the most significant bit. Data on SOMI is shifted into the receive shift register on the opposite clock edge, starting with the most significant bit. When the character is received, the receive data is moved from the RX shift register to the received data buffer UxRXBUF and the receive interrupt flag, URXIFGx, is set, indicating the RX/TX operation is complete.

A set transmit interrupt flag, UTXIFGx, indicates that data has moved from UxTXBUF to the TX shift register and UxTXBUF is ready for new data. It does not indicate RX/TX completion. In master mode, the completion of an active transmission is indicated by a set transmitter empty bit TXEPT = 1.

To receive data into the USART in master mode, data must be written to UxTXBUF because receive and transmit operations operate concurrently.

#### 19.2.2.1 Four-Pin SPI Master Mode

In 4-pin master mode, STE is used to prevent conflicts with another master. The master operates normally when STE is high. When STE is low:

- SIMO and UCLK are set to inputs and no longer drive the bus
- The error bit FE is set indicating a communication integrity violation to be handled by the user

A low STE signal does not reset the USART module. The STE input signal is not used in 3-pin master mode.

### 19.2.3 Slave Mode

Figure 19-3 shows the USART as a slave in both 3-pin and 4-pin configurations. UCLK is used as the input for the SPI clock and must be supplied by the external master. The data transfer rate is determined by this clock and not by the internal baud rate generator. Data written to UxTXBUF and moved to the TX shift register before the start of UCLK is transmitted on SOMI. Data on SIMO is shifted into the receive shift register on the opposite edge of UCLK and moved to UxRXBUF when the set number of bits are received. When data is moved from the RX shift register to UxRXBUF, the URXIFGx interrupt flag is set, indicating that data has been received. The overrun error bit, OE, is set when the previously received data is not read from UxRXBUF before new data is moved to UxRXBUF.
19.2.3.1 Four-Pin SPI Slave Mode

In 4-pin slave mode, STE is used by the slave to enable the transmit and receive operations and is provided by the SPI master. When STE is low, the slave operates normally. When STE is high:

- Any receive operation in progress on SIMO is halted
- SOMI is set to the input direction

A high STE signal does not reset the USART module. The STE input signal is not used in 3-pin slave mode.

19.2.4 SPI Enable

The SPI transmit/receive enable bit USPIEx enables or disables the USART in SPI mode. When USPIEx = 0, the USART stops operation after the current transfer completes, or immediately if no operation is active. A PUC or set SWRST bit disables the USART immediately and any active transfer is terminated.

19.2.4.1 Transmit Enable

When USPIEx = 0, any further write to UxTXBUF does not transmit. Data written to UxTXBUF begin to transmit when USPIEx = 1 and the BRCLK source is active. Figure 19-4 and Figure 19-5 show the transmit enable state diagrams.
19.2.4.2 Receive Enable

The SPI receive enable state diagrams are shown in Figure 19-6 and Figure 19-7. When USPIEx = 0, UCLK is disabled from shifting data into the RX shift register.

Figure 19-6. SPI Master Receive-Enable State Diagram

Figure 19-7. SPI Slave Receive-Enable State Diagram

19.2.5 Serial Clock Control

UCLK is provided by the master on the SPI bus. When MM = 1, BITCLK is provided by the USART baud rate generator on the UCLK pin as shown in Figure 19-8. When MM = 0, the USART clock is provided on the UCLK pin by the master and, the baud rate generator is not used and the SSELx bits are "don't care". The SPI receiver and transmitter operate in parallel and use the same clock source for data transfer.
19.2.5.1 Serial Clock Polarity and Phase

The polarity and phase of UCLK are independently configured via the CKPL and CKPH control bits of the USART. Timing for each case is shown in Figure 19-9.

![Figure 19-9. USART SPI Timing](image)

19.2.6 SPI Interrupts

The USART has one interrupt vector for transmission and one interrupt vector for reception.
19.2.6.1 SPI Transmit Interrupt Operation

The UTXIFGx interrupt flag is set by the transmitter to indicate that UxTXBUF is ready to accept another character. An interrupt request is generated if UTXIE is and GIE are also set. UTXIFGx is automatically reset if the interrupt request is serviced or if a character is written to UxTXBUF.

UTXIFGx is set after a PUC or when SWRST = 1. UTXIE is reset after a PUC or when SWRST = 1. The operation is shown is Figure 19-10.

![Figure 19-10. Transmit Interrupt Operation](diagram)

**Note**

**Writing to UxTXBUF in SPI Mode**

Data written to UxTXBUF when UTXIFGx = 0 and USPIEx = 1 may result in erroneous data transmission.
19.2.6.2 SPI Receive Interrupt Operation

The URXIFGx interrupt flag is set each time a character is received and loaded into UxRXBUF as shown in Figure 19-11 and Figure 19-12. An interrupt request is generated if URXIEx and GIE are also set. URXIFGx and URXIEx are reset by a system reset PUC signal or when SWRST = 1. URXIFGx is automatically reset if the pending interrupt is served or when UxRXBUF is read.

Figure 19-11. Receive Interrupt Operation

Figure 19-12. Receive Interrupt State Diagram
19.3 USART Registers: SPI Mode

Table 19-1 lists the memory-mapped registers for USART0 and USART1 in SPI mode.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>70h</td>
<td>U0CTL</td>
<td>USART0 control</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 19.3.1</td>
</tr>
<tr>
<td>71h</td>
<td>U0TCTL</td>
<td>USART0 transmit control</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 19.3.2</td>
</tr>
<tr>
<td>72h</td>
<td>U0RCTL</td>
<td>USART0 receive control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 19.3.3</td>
</tr>
<tr>
<td>73h</td>
<td>U0MCTL</td>
<td>USART0 modulation control</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 19.3.6</td>
</tr>
<tr>
<td>74h</td>
<td>U0BR0</td>
<td>USART0 baud-rate control 0</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 19.3.4</td>
</tr>
<tr>
<td>75h</td>
<td>U0BR1</td>
<td>USART0 baud-rate control 1</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 19.3.5</td>
</tr>
<tr>
<td>76h</td>
<td>U0RXBUF</td>
<td>USART0 receive buffer</td>
<td>Read</td>
<td>Unchanged</td>
<td>Section 19.3.7</td>
</tr>
<tr>
<td>77h</td>
<td>U0TXBUF</td>
<td>USART0 transmit buffer</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 19.3.8</td>
</tr>
<tr>
<td>4h</td>
<td>ME1</td>
<td>SFR module enable 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 19.3.9</td>
</tr>
<tr>
<td>0h</td>
<td>IE1</td>
<td>SFR interrupt enable 1</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 19.3.11</td>
</tr>
<tr>
<td>2h</td>
<td>IFG1</td>
<td>SFR interrupt flag 1</td>
<td>Read/write</td>
<td>82h with PUC</td>
<td>Section 19.3.13</td>
</tr>
<tr>
<td>78h</td>
<td>U1CTL</td>
<td>USART1 control</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 19.3.2</td>
</tr>
<tr>
<td>79h</td>
<td>U1TCTL</td>
<td>USART1 transmit control</td>
<td>Read/write</td>
<td>01h with PUC</td>
<td>Section 19.3.2</td>
</tr>
<tr>
<td>7Ah</td>
<td>U1RCTL</td>
<td>USART1 receive control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 19.3.3</td>
</tr>
<tr>
<td>7Bh</td>
<td>U1MCTL</td>
<td>USART1 modulation control</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 19.3.6</td>
</tr>
<tr>
<td>7Ch</td>
<td>U1BR0</td>
<td>USART1 baud-rate control 0</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 19.3.4</td>
</tr>
<tr>
<td>7Dh</td>
<td>U1BR1</td>
<td>USART1 baud-rate control 1</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 19.3.5</td>
</tr>
<tr>
<td>7Eh</td>
<td>U1RXBUF</td>
<td>USART1 receive buffer</td>
<td>Read</td>
<td>Unchanged</td>
<td>Section 19.3.7</td>
</tr>
<tr>
<td>7Fh</td>
<td>U1TXBUF</td>
<td>USART1 transmit buffer</td>
<td>Read/write</td>
<td>Unchanged</td>
<td>Section 19.3.8</td>
</tr>
<tr>
<td>5h</td>
<td>ME2</td>
<td>SFR module enable 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 19.3.10</td>
</tr>
<tr>
<td>1h</td>
<td>IE2</td>
<td>SFR interrupt enable 2</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 19.3.12</td>
</tr>
<tr>
<td>3h</td>
<td>IFG2</td>
<td>SFR interrupt flag 2</td>
<td>Read/write</td>
<td>20h with PUC</td>
<td>Section 19.3.14</td>
</tr>
</tbody>
</table>

Note

Modifying the SFR bits

To avoid modifying control bits for other modules, TI recommends setting or clearing the IEx and IFGx bits using BIS.B or BIC.B instructions, rather than MOV.B or CLR.B instructions.
19.3.1 UxCTL Register

USARTx Control Register

UxCTL is shown in Figure 19-13 and described in Table 19-2.

Return to Table 19-1.

---

**Table 19-2. UxCTL Register Field Descriptions**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td>Unused</td>
</tr>
<tr>
<td>5</td>
<td>I2C</td>
<td>R/W</td>
<td>0h</td>
<td>I²C mode enable. This bit selects I²C or SPI operation when SYNC = 1. 0b = SPI mode 1b = I²C mode</td>
</tr>
<tr>
<td>4</td>
<td>CHAR</td>
<td>R/W</td>
<td>0h</td>
<td>Character length 0b = 7-bit data 1b = 8-bit data</td>
</tr>
<tr>
<td>3</td>
<td>LISTEN</td>
<td>R/W</td>
<td>0h</td>
<td>Listen enable. The LISTEN bit selects the loopback mode. 0b = Disabled 1b = Enabled. The transmit signal is internally fed back to the receiver.</td>
</tr>
<tr>
<td>2</td>
<td>SYNC</td>
<td>R/W</td>
<td>0h</td>
<td>Synchronous mode enable 0b = UART mode 1b = SPI mode</td>
</tr>
<tr>
<td>1</td>
<td>MM</td>
<td>R/W</td>
<td>0h</td>
<td>Master mode 0b = USART is slave 1b = USART is master</td>
</tr>
<tr>
<td>0</td>
<td>SWRST</td>
<td>R/W</td>
<td>1h</td>
<td>Software reset enable 0b = Disabled. USART reset released for operation. 1b = Enabled. USART logic held in reset state.</td>
</tr>
</tbody>
</table>
19.3.2 UxTCTL Register

USARTx Transmit Control Register

UxTCTL is shown in Figure 19-14 and described in Table 19-3.

Return to Table 19-1.

Figure 19-14. UxTCTL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
</table>
| 7    | CKPH  | R/W  | 0h    | Clock phase select  
|      |       |      |       | 0b = Data is changed on the first UCLK edge and captured on the following edge.  
|      |       |      |       | 1b = Data is captured on the first UCLK edge and changed on the following edge. |
| 6    | CKPL  | R/W  | 0h    | Clock polarity select  
|      |       |      |       | 0b = The inactive state is low.  
|      |       |      |       | 1b = The inactive state is high. |
| 5-4  | SSELx | R/W  | 0h    | Source select. These bits select the BRCLK source clock.  
|      |       |      |       | 00b = External UCLK (valid for slave mode only)  
|      |       |      |       | 01b = ACLK (valid for master mode only)  
|      |       |      |       | 10b = SMCLK (valid for master mode only)  
|      |       |      |       | 11b = SMCLK (valid for master mode only) |
| 3-2  | Unused| R/W  | 0h    | Slave transmit control  
|      |       |      |       | 0b = 4-pin SPI mode: STE enabled  
|      |       |      |       | 1b = 3-pin SPI mode: STE disabled |
| 1    | STC   | R/W  | 0h    | Transmitter empty flag. The TXEPT flag is not used in slave mode.  
|      |       |      |       | 0b = Transmission active and/or data waiting in UxTXBUF  
|      |       |      |       | 1b = UxTXBUF and TX shift register are empty |
| 0    | TXEPT | R/W  | 1h    | |

Table 19-3. UxTCTL Register Field Descriptions
19.3.3 UxRCTL Register

USARTx Receive Control Register

UxRCTL is shown in Figure 19-15 and described in Table 19-4.

Return to Table 19-1.

![Figure 19-15. UxRCTL Register](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>FE</td>
<td>R/W</td>
<td>0h</td>
<td>Framing error flag. This bit indicates a bus conflict when MM = 1 and STC = 0. FE is unused in slave mode. 0b = No conflict detected 1b = A negative edge occurred on STE, indicating bus conflict.</td>
</tr>
<tr>
<td>6</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td>Unused</td>
</tr>
<tr>
<td>5</td>
<td>OE</td>
<td>R/W</td>
<td>0h</td>
<td>Overrun error flag. This bit is set when a character is transferred into UxRXBUF before the previous character was read. OE is automatically reset when UxRXBUF is read, when SWRST = 1, or can be reset by software. 0b = No error 1b = Overrun error occurred</td>
</tr>
<tr>
<td>4-0</td>
<td>Unused</td>
<td>R/W</td>
<td>0h</td>
<td>Unused</td>
</tr>
</tbody>
</table>
19.3.4 UxBR0 Register

USARTx Baud-Rate Control 0 Register

UxBR0 is shown in Figure 19-16 and described in Table 19-5.

Return to Table 19-1.

![Figure 19-16. UxBR0 Register](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UxBR0</td>
<td>R/W</td>
<td>0h</td>
<td>The baud-rate generator uses the content of (UxBR1 + UxBR0) to set the baud rate. Unpredictable SPI operation occurs if UxBR &lt; 2.</td>
</tr>
</tbody>
</table>

19.3.5 UxBR1 Register

USARTx Baud-Rate Control 1 Register

UxBR1 is shown in Figure 19-17 and described in Table 19-6.

Return to Table 19-1.

![Figure 19-17. UxBR1 Register](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UxBR1</td>
<td>R/W</td>
<td>0h</td>
<td>The baud-rate generator uses the content of (UxBR1 + UxBR0) to set the baud rate. Unpredictable SPI operation occurs if UxBR &lt; 2.</td>
</tr>
</tbody>
</table>

19.3.6 UxMCTL Register

USARTx Modulation Control Register

UxMCTL is shown in Figure 19-18 and described in Table 19-7.

Return to Table 19-1.

![Figure 19-18. UxMCTL Register](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UxMCTLx</td>
<td>R/W</td>
<td>0h</td>
<td>The modulation control register is not used for SPI mode and should be set to 00h.</td>
</tr>
</tbody>
</table>
19.3.7 UxRXBUF Register

USART\(x\) Receive Buffer Register

UxRXBUF is shown in Figure 19-19 and described in Table 19-8.

Return to Table 19-1.

Figure 19-19. UxRXBUF Register

<table>
<thead>
<tr>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>r</td>
<td>r</td>
<td>r</td>
<td>r</td>
<td>r</td>
<td>r</td>
<td>r</td>
<td>r</td>
</tr>
</tbody>
</table>

Table 19-8. UxRXBUF Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UxRXBUFx</td>
<td>R</td>
<td>0h</td>
<td>The receive-data buffer is user accessible and contains the last received character from the receive shift register. Reading UxRXBUF resets the OE bit and URXIFGx flag. In 7-bit data mode, UxRXBUF is LSB justified and the MSB is always reset.</td>
</tr>
</tbody>
</table>

19.3.8 UxTXBUF Register

USART\(x\) Transmit Buffer Register

UxTXBUF is shown in Figure 19-20 and described in Table 19-9.

Return to Table 19-1.

Figure 19-20. UxTXBUF Register

<table>
<thead>
<tr>
<th>7</th>
<th>6</th>
<th>5</th>
<th>4</th>
<th>3</th>
<th>2</th>
<th>1</th>
<th>0</th>
</tr>
</thead>
<tbody>
<tr>
<td>rw-0</td>
<td>rw-0</td>
<td>rw-0</td>
<td>rw-0</td>
<td>rw-0</td>
<td>rw-0</td>
<td>rw-0</td>
<td>rw-0</td>
</tr>
</tbody>
</table>

Table 19-9. UxTXBUF Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>UxTXBUFx</td>
<td>R/W</td>
<td>0h</td>
<td>The transmit data buffer is user accessible and contains current data to be transmitted. When seven-bit character-length is used, the data should be MSB justified before being moved into UxTXBUF. Data is transmitted MSB first. Writing to UxTXBUF clears UTXIFGx.</td>
</tr>
</tbody>
</table>
19.3.9 ME1 Register

SFR Module Enable 1 Register

ME1 is shown in Figure 19-21 and described in Table 19-10.

Return to Table 19-1.

### Figure 19-21. ME1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td></td>
<td></td>
<td></td>
<td>This bit may be used by other modules. See the device-specific data sheet.</td>
</tr>
</tbody>
</table>
| 6   | USPIE0| R/W  | 0h    | USART0 SPI enable. This bit enables the SPI mode for USART0.
|     |       |      |       | 0b = Module not enabled |
|     |       |      |       | 1b = Module enabled |
| 5-0 |       |      |       | These bits may be used by other modules. See the device-specific data sheet. |

Table 19-10. ME1 Register Field Descriptions

19.3.10 ME2 Register

SFR Module Enable 2 Register

ME2 is shown in Figure 19-22 and described in Table 19-11.

Return to Table 19-1.

### Figure 19-22. ME2 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td></td>
<td></td>
<td></td>
<td>This bit may be used by other modules. See the device-specific data sheet.</td>
</tr>
</tbody>
</table>
| 6   | USPIE1| R/W  | 0h    | USART1 SPI enable. This bit enables the SPI mode for USART1.
|     |       |      |       | 0b = Module not enabled |
|     |       |      |       | 1b = Module enabled |
| 5-0 |       |      |       | These bits may be used by other modules. See the device-specific data sheet. |

Table 19-11. ME2 Register Field Descriptions
19.3.11 IE1 Register

SFR Interrupt Enable 1 Register

IE1 is shown in Figure 19-23 and described in Table 19-12.

Return to Table 19-1.

![Figure 19-23. IE1 Register](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>UTXIE0</td>
<td>R/W</td>
<td>0h</td>
<td>USART0 transmit interrupt enable. This bit enables the UTXIFG0 interrupt.  0b = Interrupt not enabled 1b = Interrupt enabled</td>
</tr>
<tr>
<td>6</td>
<td>URXIE0</td>
<td>R/W</td>
<td>0h</td>
<td>USART0 receive interrupt enable. This bit enables the URXIFG0 interrupt.  0b = Interrupt not enabled 1b = Interrupt enabled</td>
</tr>
<tr>
<td>5-0</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See the device-specific data sheet.</td>
</tr>
</tbody>
</table>

19.3.12 IE2 Register

SFR Interrupt Enable 2 Register

IE2 is shown in Figure 19-24 and described in Table 19-13.

Return to Table 19-1.

![Figure 19-24. IE2 Register](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See the device-specific data sheet.</td>
</tr>
<tr>
<td>5</td>
<td>UTXIE0</td>
<td>R/W</td>
<td>0h</td>
<td>USART1 transmit interrupt enable. This bit enables the UTXIFG1 interrupt.  0b = Interrupt not enabled 1b = Interrupt enabled</td>
</tr>
<tr>
<td>4</td>
<td>URXIE0</td>
<td>R/W</td>
<td>0h</td>
<td>USART1 receive interrupt enable. This bit enables the URXIFG1 interrupt.  0b = Interrupt not enabled 1b = Interrupt enabled</td>
</tr>
<tr>
<td>3-0</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See the device-specific data sheet.</td>
</tr>
</tbody>
</table>
19.3.13 IFG1 Register

SFR Interrupt Flag 1 Register

IFG1 is shown in Figure 19-25 and described in Table 19-14.

Return to Table 19-1.

### Figure 19-25. IFG1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>UTXIFG0</td>
<td>R/W</td>
<td>1h</td>
<td>USART0 transmit interrupt flag. UTXIFG0 is set when U0TXBUF is empty.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>6</td>
<td>URXIFG0</td>
<td>R/W</td>
<td>0h</td>
<td>USART0 receive interrupt flag. URXIFG0 is set when U0RXBUF has received a</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>complete character.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>5-0</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See the device-specific data</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>sheet.</td>
</tr>
</tbody>
</table>

### Table 19-14. IFG1 Register Field Descriptions

19.3.14 IFG2 Register

SFR Interrupt Flag 2 Register

IFG2 is shown in Figure 19-26 and described in Table 19-15.

Return to Table 19-1.

### Figure 19-26. IFG2 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See the device-specific data</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>sheet.</td>
</tr>
<tr>
<td>5</td>
<td>UTXIFG1</td>
<td>R/W</td>
<td>1h</td>
<td>USART1 transmit interrupt flag. UTXIFG1 is set when U1TXBUF is empty.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>4</td>
<td>URXIFG1</td>
<td>R/W</td>
<td>0h</td>
<td>USART1 receive interrupt flag. URXIFG1 is set when U1RXBUF has received a</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>complete character.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No interrupt pending</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Interrupt pending</td>
</tr>
<tr>
<td>3-0</td>
<td></td>
<td></td>
<td></td>
<td>These bits may be used by other modules. See the device-specific data</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>sheet.</td>
</tr>
</tbody>
</table>
The OA is a general purpose operational amplifier. This chapter describes the OA. Two OA modules are implemented in the MSP430x22x4 devices.
20.1 OA Introduction

The OA operational amplifiers support front-end analog signal conditioning prior to analog-to-digital conversion. Features of the OA include:

- Single supply, low-current operation
- Rail-to-rail output
- Programmable settling time vs. power consumption
- Software selectable configurations
- Software selectable feedback resistor ladder for PGA implementations

__Note__

**Multiple OA Modules**

Some devices may integrate more than one OA module. If more than one OA is present on a device, the multiple OA modules operate identically.

Throughout this chapter, nomenclature appears such as OAxCTL0 to describe register names. When this occurs, the x is used to indicate which OA module is being discussed. In cases where operation is identical, the register is simply referred to as OAxCTL0.

The block diagram of the OA module is shown in Figure 20-1.
20.2 OA Operation

The OA module is configured with user software. The setup and operation of the OA is discussed in the following sections.

20.2.1 OA Amplifier

The OA is a configurable, low-current, rail-to-rail output operational amplifier. It can be configured as an inverting amplifier, or a non-inverting amplifier, or can be combined with other OA modules to form differential amplifiers.

Figure 20-1. OA Block Diagram
The output slew rate of the OA can be configured for optimized settling time vs power consumption with the OAPMx bits. When OAPMx = 00 the OA is off and the output is high-impedance. When OAPMx > 0, the OA is on. See the device-specific data sheet for parameters.

20.2.2 OA Input

The OA has configurable input selection. The signals for the + and - inputs are individually selected with the OANx and OAPx bits and can be selected as external signals or internal signals. OAxl0 and OAxl1 are external signals provided for each OA module. OA011 provides a non-inverting input that is tied together internally for all OA modules. OAxIA and OAxIB provide device-dependent inputs. See the device data sheet for signal connections.

When the external inverting input is not needed for a mode, setting the OANEXT bit makes the internal inverting input externally available.

20.2.3 OA Output and Feedback Routing

The OA has configurable output selection controlled by the OAADCx bits and the OAFCx bits. The OA output signals can be routed to ADC inputs A12 (OA0), A13 (OA1), or A14 (OA2) internally, or can be routed to these ADC inputs and their external pins. The OA output signals can also be routed to ADC inputs A1 (OA0), A3 (OA1), or A5 (OA2) and the corresponding external pin. The OA output is also connected to an internal R-ladder with the OAFCx bits. The R-ladder tap is selected with the OAFBRx bits to provide programmable gain amplifier functionality.

Table 20-1 shows the OA output and feedback routing configurations. When OAFCx = 0 the OA is in general-purpose mode and feedback is achieved externally to the device. When OAFCx > 0 and when OAADCx = 00 or 11, the output of the OA is kept internal to the device. When OAFCx > 0 and OAADCx = 01 or 10, the OA output is routed both internally and externally.

<table>
<thead>
<tr>
<th>OAFCx</th>
<th>OAADCx</th>
<th>OA Output and Feedback Routing</th>
</tr>
</thead>
<tbody>
<tr>
<td>= 0</td>
<td>x0</td>
<td>OAxOUT connected to external pins and ADC input A1, A3, or A5.</td>
</tr>
<tr>
<td>= 0</td>
<td>x1</td>
<td>OAxOUT connected to external pins and ADC input A12, A13, or A14.</td>
</tr>
<tr>
<td>&gt; 0</td>
<td>00</td>
<td>OAxOUT used for internal routing only.</td>
</tr>
<tr>
<td>&gt; 0</td>
<td>01</td>
<td>OAxOUT connected to external pins and ADC input A12, A13, or A14.</td>
</tr>
<tr>
<td>&gt; 0</td>
<td>10</td>
<td>OAxOUT connected to external pins and ADC input A1, A3, or A5.</td>
</tr>
<tr>
<td>&gt; 0</td>
<td>11</td>
<td>OAxOUT connected internally to ADC input A12, A13, or A14. External A12, A13, or A14 pin connections are disconnected from the ADC.</td>
</tr>
</tbody>
</table>

20.2.4 OA Configurations

The OA can be configured for different amplifier functions with the OAFCx bits as listed in Table 20-2.

<table>
<thead>
<tr>
<th>OAFCx</th>
<th>OA Mode</th>
</tr>
</thead>
<tbody>
<tr>
<td>000</td>
<td>General-purpose opamp</td>
</tr>
<tr>
<td>001</td>
<td>Unity gain buffer for three-opamp differential amplifier</td>
</tr>
<tr>
<td>010</td>
<td>Unity gain buffer</td>
</tr>
<tr>
<td>011</td>
<td>Comparator</td>
</tr>
<tr>
<td>100</td>
<td>Non-inverting PGA amplifier</td>
</tr>
<tr>
<td>101</td>
<td>Cascaded non-inverting PGA amplifier</td>
</tr>
<tr>
<td>110</td>
<td>Inverting PGA amplifier</td>
</tr>
<tr>
<td>111</td>
<td>Differential amplifier</td>
</tr>
</tbody>
</table>


20.2.4.1 General Purpose Opamp Mode
In this mode the feedback resistor ladder is isolated from the OAx and the OAxCTL0 bits define the signal routing. The OAx inputs are selected with the OAPx and OANx bits. The OAx output is connected to the ADC input channel as selected by the OAxCTL0 bits.

20.2.4.2 Unity Gain Mode for Differential Amplifier
In this mode the output of the OAx is connected to the inverting input of the OAx providing a unity gain buffer. The non-inverting input is selected by the OApx bits. The external connection for the inverting input is disabled and the OANx bits are don’t care. The output of the OAx is also routed through the resistor ladder as part of the three-opamp differential amplifier. This mode is only for construction of the three-opamp differential amplifier.

20.2.4.3 Unity Gain Mode
In this mode the output of the OAx is connected to the inverting input of the OAx providing a unity gain buffer. The non-inverting input is selected by the OApx bits. The external connection for the inverting input is disabled and the OANx bits are don’t care. The OAx output is connected to the ADC input channel as selected by the OAxCTL0 bits.

20.2.4.4 Comparator Mode
In this mode the output of the OAx is isolated from the resistor ladder. R_TOP is connected to AVSS when OARRIP = 0. When OARRIP = 1, the connection of the resistor ladder is reversed. R_TOP is connected to AVCC and R_BOTTOM is connected to AVSS. The OAxTAP signal is connected to the inverting input of the OAx providing a comparator with a programmable threshold voltage selected by the OAFBRx bits.

20.2.4.5 Non-Inverting PGA Mode
In this mode the output of the OAx is connected to R_TOP and R_BOTTOM is connected to AVSS. The OAxTAP signal is connected to the inverting input of the OAx providing a non-inverting amplifier configuration with a programmable gain of \([1+\text{OAxTAP ratio}]\). The OAxTAP ratio is selected by the OAFBRx bits. If the OAFBRx bits = 0, the gain is unity. The non-inverting input is selected by the OApx bits. The external connection for the inverting input is disabled and the OANx bits are don’t care. The OAx output is connected to the ADC input channel as selected by the OAxCTL0 bits.

20.2.4.6 Cascaded Non-Inverting PGA Mode
This mode allows internal routing of the OA signals to cascade two or three OA in non-inverting mode. In this mode the non-inverting input of the OAx is connected to OA2OUT (OA0), OA0OUT (OA1), or OA1OUT (OA2) when OAp = 11. The OAx outputs are connected to the ADC input channel as selected by the OAxCTL0 bits.

20.2.4.7 Inverting PGA Mode
In this mode the output of the OAx is connected to R_TOP and R_BOTTOM is connected to an analog multiplexer that multiplexes the OAx0, OAx1, OAxI, or the output of one of the remaining OAs, selected with the OANx bits. The OAxTAP signal is connected to the inverting input of the OAx providing an inverting amplifier with a gain of \(-\text{OAxTAP ratio}\). The OAxTAP ratio is selected by the OAFBRx bits. The non-inverting input is selected by the OApx bits. The OAx output is connected to the ADC input channel as selected by the OAxCTL0 bits.

Note

Using OAx Negative Input Simultaneously as ADC Input

When the pin connected to the negative input multiplexer is also used as an input to the ADC, conversion errors up to 5 mV may be observed due to internal wiring voltage drops.
20.2.4.8 Differential Amplifier Mode

This mode allows internal routing of the OA signals for a two-opamp or three-opamp instrumentation amplifier. Figure 20-2 shows a two-opamp configuration with OA0 and OA1. In this mode the output of the OAx is connected to \( R_{\text{TOP}} \) by routing through another OAx in the Inverting PGA mode. \( R_{\text{BOTTOM}} \) is unconnected providing a unity gain buffer. This buffer is combined with one or two remaining OAx to form the differential amplifier. The OAx output is connected to the ADC input channel as selected by the OAxCTL0 bits.

Figure 20-2 shows an example of a two-opamp differential amplifier using OA0 and OA1. The control register settings and are shown in Table 20-3. The gain for the amplifier is selected by the OAFBRx bits for OA1 and is shown in Table 20-4. The OAx interconnections are shown in Figure 20-3.

### Table 20-3. Two-Opamp Differential Amplifier Control Register Settings

<table>
<thead>
<tr>
<th>Register</th>
<th>Settings (binary)</th>
</tr>
</thead>
<tbody>
<tr>
<td>OA0CTL0</td>
<td>xx xx xx 0 0</td>
</tr>
<tr>
<td>OA0CTL1</td>
<td>000 111 0 x</td>
</tr>
<tr>
<td>OA1CTL0</td>
<td>11 xx xx x x</td>
</tr>
<tr>
<td>OA1CTL1</td>
<td>xxx 110 0 x</td>
</tr>
</tbody>
</table>

### Table 20-4. Two-Opamp Differential Amplifier Gain Settings

<table>
<thead>
<tr>
<th>OA1 OAFBRx</th>
<th>Gain</th>
</tr>
</thead>
<tbody>
<tr>
<td>000</td>
<td>0</td>
</tr>
<tr>
<td>001</td>
<td>1/3</td>
</tr>
<tr>
<td>010</td>
<td>1</td>
</tr>
<tr>
<td>011</td>
<td>1 2/3</td>
</tr>
<tr>
<td>100</td>
<td>3</td>
</tr>
<tr>
<td>101</td>
<td>4 1/3</td>
</tr>
<tr>
<td>110</td>
<td>7</td>
</tr>
<tr>
<td>111</td>
<td>15</td>
</tr>
</tbody>
</table>

\[
V_{\text{diff}} = \frac{(V_2 - V_1) \times R_2}{R_1}
\]

Figure 20-2. Two-Opamp Differential Amplifier
Figure 20-3. Two-Opamp Differential Amplifier OAx Interconnections
Figure 20-4 shows an example of a three-opamp differential amplifier using OA0, OA1 and OA2 (Three opamps are not available on all devices. See device-specific data sheet for implementation.). The control register settings are shown in Table 20-5. The gain for the amplifier is selected by the OAFBRx bits of OA0 and OA2. The OAFBRx settings for both OA0 and OA2 must be equal. The gain settings are shown in Table 20-6. The OAx interconnections are shown in Figure 20-5.

Table 20-5. Three-Opamp Differential Amplifier Control Register Settings

<table>
<thead>
<tr>
<th>Register</th>
<th>Settings (binary)</th>
</tr>
</thead>
<tbody>
<tr>
<td>OA0CTL0</td>
<td>xx xx xx 0 0</td>
</tr>
<tr>
<td>OA0CTL1</td>
<td>xxx 00 1 0 x</td>
</tr>
<tr>
<td>OA1CTL0</td>
<td>xx xx xx 0 0</td>
</tr>
<tr>
<td>OA1CTL1</td>
<td>000 11 1 0 x</td>
</tr>
<tr>
<td>OA2CTL0</td>
<td>11 11 xx x x x</td>
</tr>
<tr>
<td>OA2CTL1</td>
<td>xxx 11 0 0 x</td>
</tr>
</tbody>
</table>

Table 20-6. Three-Opamp Differential Amplifier Gain Settings

<table>
<thead>
<tr>
<th>OA0/OA2 OAFBRx</th>
<th>Gain</th>
</tr>
</thead>
<tbody>
<tr>
<td>000</td>
<td>0</td>
</tr>
<tr>
<td>001</td>
<td>1/3</td>
</tr>
<tr>
<td>010</td>
<td>1</td>
</tr>
<tr>
<td>011</td>
<td>1 2/3</td>
</tr>
<tr>
<td>100</td>
<td>3</td>
</tr>
<tr>
<td>101</td>
<td>4 1/3</td>
</tr>
<tr>
<td>110</td>
<td>7</td>
</tr>
<tr>
<td>111</td>
<td>15</td>
</tr>
</tbody>
</table>

![Figure 20-4. Three-Opamp Differential Amplifier](image-url)
Figure 20-5. Three-Opamp Differential Amplifier OAx Interconnections
20.3 OA Registers

Table 20-7 lists the memory-mapped registers for the OA.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>C0h</td>
<td>OA0CTL0</td>
<td>OA0 control 0</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 20.3.1</td>
</tr>
<tr>
<td>C1h</td>
<td>OA0CTL1</td>
<td>OA0 control 1</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 20.3.2</td>
</tr>
<tr>
<td>C2h</td>
<td>OA1CTL0</td>
<td>OA1 control 0</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 20.3.1</td>
</tr>
<tr>
<td>C3h</td>
<td>OA1CTL1</td>
<td>OA1 control 1</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 20.3.2</td>
</tr>
<tr>
<td>C4h</td>
<td>OA2CTL0</td>
<td>OA2 control 0</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 20.3.1</td>
</tr>
<tr>
<td>C5h</td>
<td>OA2CTL1</td>
<td>OA2 control 1</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 20.3.2</td>
</tr>
</tbody>
</table>
20.3.1 OAxCTL0 Register

OAx Control 0 Register

OAxCTL0 is shown in Figure 20-6 and described in Table 20-8.

Return to Table 20-7.

![Figure 20-6. OAxCTL0 Registers](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td>OANx</td>
<td>R/W</td>
<td>0h</td>
<td>Inverting input select. These bits select the input signal for the OA inverting input. 00b = OAxI0 01b = OAxI1 10b = OAxIA (see the device-specific data sheet for connected signal) 11b = OAxIB (see the device-specific data sheet for connected signal)</td>
</tr>
<tr>
<td>5-4</td>
<td>OAPx</td>
<td>R/W</td>
<td>0h</td>
<td>Noninverting input select. These bits select the input signal for the OA noninverting input. 00b = OAxI0 01b = OAxI1 10b = OAxIA (see the device-specific data sheet for connected signal) 11b = OAxIB (see the device-specific data sheet for connected signal)</td>
</tr>
<tr>
<td>3-2</td>
<td>OAPMx</td>
<td>R/W</td>
<td>0h</td>
<td>Slew rate select. These bits select the slew rate vs. current consumption for the OA. 00b = Off, output high Z 01b = Slow 10b = Medium 11b = Fast</td>
</tr>
</tbody>
</table>
Table 20-8. OAxCTL0 Register Field Descriptions (continued)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
</table>
| 1-0 | OAADCx | R/W  | 0h    | OA output select. These bits, together with the OAFCx bits, control the routing of the OAx output when OAPMx > 0. When OAFCx = 0:  
00b = OAxOUT connected to external pins and ADC input A1, A3, or A5  
01b = OAxOUT connected to external pins and ADC input A12, A13, or A14  
10b = OAxOUT connected to external pins and ADC input A1, A3, or A5  
11b = OAxOUT connected to external pins and ADC input A12, A13, or A14. When OAFCx > 0:  
00b = OAxOUT used for internal routing only  
01b = OAxOUT connected to external pins and ADC input A12, A13, or A14  
10b = OAxOUT connected to external pins and ADC input A1, A3, or A5  
11b = OAxOUT connected internally to ADC input A12, A13, or A14. External A12, A13, or A14 pin connections are disconnected from the ADC. |
20.3.2 OAxCTL1 Register

OAx Control 1 Register

OAxCTL1 is shown in Figure 20-7 and described in Table 20-9.

Return to Table 20-7.

![Figure 20-7. OAxCTL1 Registers](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-5</td>
<td>OAFBRx</td>
<td>R/W</td>
<td>0h</td>
<td>OAx feedback resistor select.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>000b = Tap 0: 0R/16R</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>001b = Tap 1: 4R/12R</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>010b = Tap 2: 8R/8R</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>011b = Tap 3: 10R/6R</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>100b = Tap 4: 12R/4R</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>101b = Tap 5: 13R/3R</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>110b = Tap 6: 14R/2R</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>111b = Tap 7: 15R/1R</td>
</tr>
<tr>
<td>4-2</td>
<td>OAFCx</td>
<td>R/W</td>
<td>0h</td>
<td>OAx function control. This bit selects the function of OAx.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>000b = General-purpose opamp</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>001b = Unity gain buffer for three-opamp differential amplifier</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>010b = Unity gain buffer</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>011b = Comparator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>100b = Noninverting PGA amplifier</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>101b = Cascaded noninverting PGA amplifier</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>110b = Inverting PGA amplifier</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>111b = Differential amplifier</td>
</tr>
<tr>
<td>1</td>
<td>OANEXT</td>
<td>R/W</td>
<td>0h</td>
<td>OAx inverting input externally available. This bit, when set, connects</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>the inverting OAx input to the external pin when the integrated</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>resistor network is used.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = OAx inverting input not externally available</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = OAx inverting input externally available</td>
</tr>
<tr>
<td>0</td>
<td>OARRIP</td>
<td>R/W</td>
<td>0h</td>
<td>OAx reverse resistor connection in comparator mode.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = RTOP is connected to AVSS and RBOTTOM is connected to AVCC when</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>OAFCx = 3</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = RTOP is connected to AVCC and RBOTTOM is connected to AVSS when</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>OAFCx = 3</td>
</tr>
</tbody>
</table>
Comparator_A+ is an analog voltage comparator. This chapter describes the operation of the Comparator_A+ of the 2xx family.

21.1 Comparator_A+ Introduction.................................................................568
21.2 Comparator_A+ Operation.................................................................569
21.3 Comparator_A+ Registers.................................................................574
21.1 Comparator_A+ Introduction

The Comparator_A+ module supports precision slope analog-to-digital conversions, supply voltage supervision, and monitoring of external analog signals.

Features of Comparator_A+ include:

- Inverting and noninverting terminal input multiplexer
- Software selectable RC-filter for the comparator output
- Output provided to Timer_A capture input
- Software control of the port input buffer
- Interrupt capability
- Selectable reference voltage generator
- Comparator and reference generator can be powered down
- Input Multiplexer

Figure 21-1 shows the Comparator_A+ block diagram.

Figure 21-1. Comparator_A+ Block Diagram

**Note**

MSP430G2210: Channels 2, 5, 6, and 7 are available. Other channels should not be enabled.
21.2 Comparator_A+ Operation

The Comparator_A+ module is configured with user software. The setup and operation of Comparator_A+ is described in the following sections.

21.2.1 Comparator

The comparator compares the analog voltages at the + and – input terminals. If the + terminal is more positive than the – terminal, the comparator output CAOUT is high. The comparator can be switched on or off using control bit CAON. The comparator should be switched off when not in use to reduce current consumption. When the comparator is switched off, the CAOUT is always low.

21.2.2 Input Analog Switches

The analog input switches connect or disconnect the two comparator input terminals to associated port pins using the P2CAx bits. Both comparator terminal inputs can be controlled individually. The P2CAx bits allow:

- Application of an external signal to the + and – terminals of the comparator
- Routing of an internal reference voltage to an associated output port pin

Internally, the input switch is constructed as a T-switch to suppress distortion in the signal path.

---

**Note**

**Comparator Input Connection**

When the comparator is on, the input terminals should be connected to a signal, power, or ground. Otherwise, floating levels may cause unexpected interrupts and increased current consumption.

---

**Note**

**MSP430G2210**: Comparator channels 0, 1, 3, and 4 are implemented but not available at the device pins. To avoid floating inputs, these comparator inputs should not be enabled.

The CAEX bit controls the input multiplexer, exchanging which input signals are connected to the comparator’s + and – terminals. Additionally, when the comparator terminals are exchanged, the output signal from the comparator is inverted. This allows the user to determine or compensate for the comparator input offset voltage.
21.2.3 Input Short Switch

The CASHORT bit shorts the comparator_A+ inputs. This can be used to build a simple sample-and-hold for the comparator as shown in Figure 21-2.

![Figure 21-2. Comparator_A+ Sample-And-Hold](image)

The required sampling time is proportional to the size of the sampling capacitor ($C_S$), the resistance of the input switches in series with the short switch ($R_i$), and the resistance of the external source ($R_S$). The total internal resistance ($R_i$) is typically in the range of 2 to 10 kΩ. The sampling capacitor $C_S$ should be greater than 100 pF.

The time constant, $\tau$, to charge the sampling capacitor $C_S$ can be calculated with the following equation:

$$\tau = (R_i + R_S) \times C_S$$

Depending on the required accuracy, 3 to 10 $\tau$ should be used as a sampling time. With 3 $\tau$, the sampling capacitor is charged to approximately 95% of the input signal's voltage level. With 5 $\tau$, the sampling capacitor is charged to more than 99%. With 10 $\tau$, the sampled voltage is sufficient for 12-bit accuracy.

21.2.4 Output Filter

The output of the comparator can be used with or without internal filtering. When control bit CAF is set, the output is filtered with an on-chip RC-filter.

Any comparator output oscillates if the voltage difference across the input terminals is small. Internal and external parasitic effects and cross coupling on and between signal lines, power supply lines, and other parts of the system are responsible for this behavior as shown in Figure 21-3. The comparator output oscillation reduces accuracy and resolution of the comparison result. Selecting the output filter can reduce errors associated with comparator oscillation.
21.2.5 Voltage Reference Generator

The voltage reference generator is used to generate $V_{CAREF}$, which can be applied to either comparator input terminal. The CAREFx bits control the output of the voltage generator. The CARSEL bit selects the comparator terminal to which $V_{CAREF}$ is applied. If external signals are applied to both comparator input terminals, the internal reference generator should be turned off to reduce current consumption. The voltage reference generator can generate a fraction of the device’s $V_{CC}$ or a fixed transistor threshold voltage of approximately 0.55 V.

21.2.6 Comparator_A+, Port Disable Register CAPD

The comparator input and output functions are multiplexed with the associated I/O port pins, which are digital CMOS gates. When analog signals are applied to digital CMOS gates, parasitic current can flow from $V_{CC}$ to $V_{SS}$. This parasitic current occurs if the input voltage is near the transition level of the gate. Disabling the port pin buffer eliminates the parasitic current flow and therefore reduces overall current consumption.

The CAPDx bits, when set, disable the corresponding pin input and output buffers as shown in Figure 21-4. When current consumption is critical, any port pin connected to analog signals should be disabled with its CAPDx bit.

Selecting an input pin to the comparator multiplexer with the P2CAx bits automatically disables the input and output buffers for that pin, regardless of the state of the associated CAPDx bit.

---

**Note**

*MSP430G2210:* The channels 0, 1, 3, and 4 are implemented by not available at pins. To avoid floating inputs these inputs should not be used.
21.2.7 Comparator_A+ Interrupts

One interrupt flag and one interrupt vector are associated with the Comparator_A+ as shown in Figure 21-5. The interrupt flag CAIFG is set on either the rising or falling edge of the comparator output, selected by the CAIES bit. If both the CAIE and the GIE bits are set, then the CAIFG flag generates an interrupt request. The CAIFG flag is automatically reset when the interrupt request is serviced, or the flag may be reset with software.

![Diagram of Comparator_A+ Interrupt System](image.png)

**Figure 21-5. Comparator_A+ Interrupt System**

---

**Note**

Changing the value of the CAIES bit might set the comparator interrupt flag CAIFG. This can happen even when the comparator is disabled (CAON = 0). TI recommends clearing CAIFG after configuring the comparator for proper interrupt behavior during operation.

---

21.2.8 Comparator_A+ Used to Measure Resistive Elements

The Comparator_A+ can be optimized to precisely measure resistive elements using single slope analog-to-digital conversion. For example, temperature can be converted into digital data using a thermistor, by comparing the thermistor’s capacitor discharge time to that of a reference resistor as shown in Figure 21-6. A reference resistor Rref is compared to Rmeas.

![Diagram of Temperature Measurement System](image.png)

**Figure 21-6. Temperature Measurement System**

The MSP430 resources used to calculate the temperature sensed by Rmeas are:

- Two digital I/O pins to charge and discharge the capacitor.
- I/O set to output high (VCC) to charge capacitor, reset to discharge.
- I/O switched to high-impedance input with CAPDx set when not in use.
- One output charges and discharges the capacitor through Rref.
- One output discharges capacitor through Rmeas.
- The + terminal is connected to the positive terminal of the capacitor.
- The – terminal is connected to a reference level, for example 0.25 × VCC.
- The output filter should be used to minimize switching noise.
- CAOUT used to gate Timer_A CCI1B, capturing capacitor discharge time.
More than one resistive element can be measured. Additional elements are connected to CA0 with available I/O pins and switched to high impedance when not being measured.

The thermistor measurement is based on a ratiometric conversion principle. The ratio of two capacitor discharge times is calculated as shown in Figure 21-7.

![Figure 21-7. Timing for Temperature Measurement Systems](image)

The $V_{CC}$ voltage and the capacitor value should remain constant during the conversion, but are not critical since they cancel in the ratio:

\[
\frac{N_{\text{meas}}}{N_{\text{ref}}} = -\frac{R_{\text{meas}} \times C \times \ln \frac{V_{\text{ref}}}{V_{CC}}}{-R_{\text{ref}} \times C \times \ln \frac{V_{\text{ref}}}{V_{CC}}}
\]

\[
\frac{N_{\text{meas}}}{N_{\text{ref}}} = \frac{R_{\text{meas}}}{R_{\text{ref}}}
\]

\[
R_{\text{meas}} = R_{\text{ref}} \times \frac{N_{\text{meas}}}{N_{\text{ref}}}
\]
21.3 Comparator_A+ Registers

Table 21-1 lists the memory-mapped registers for the Comparator_A+.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>59h</td>
<td>CACTL1</td>
<td>Comparator_A+ control 1</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 21.3.1</td>
</tr>
<tr>
<td>5Ah</td>
<td>CACTL2</td>
<td>Comparator_A+ control 2</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 21.3.2</td>
</tr>
<tr>
<td>5Bh</td>
<td>CAPD</td>
<td>Comparator_A+ port disable</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 21.3.3</td>
</tr>
</tbody>
</table>
### 21.3.1 CACTL1 Register

**Comparator_A+ Control 1 Register**

CACTL1 is shown in Figure 21-8 and described in Table 21-2.

Return to Table 21-1.

#### Figure 21-8. CACTL1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>CAEX</td>
<td>R/W</td>
<td>0h</td>
<td>Comparator_A+ exchange. This bit exchanges the comparator inputs and inverts the comparator output.</td>
</tr>
<tr>
<td>6</td>
<td>CARSEL</td>
<td>R/W</td>
<td>0h</td>
<td>Comparator_A+ reference select. This bit selects which terminal the $V_{CAREF}$ is applied to. When CAEX = 0: 0b = $V_{CAREF}$ is applied to the positive terminal 01b = $V_{CAREF}$ is applied to the negative terminal When CAEX = 1: 0b = $V_{CAREF}$ is applied to the negative terminal 1b = $V_{CAREF}$ is applied to the positive terminal</td>
</tr>
<tr>
<td>5-4</td>
<td>CAREF</td>
<td>R/W</td>
<td>0h</td>
<td>Comparator_A+ reference. These bits select the reference voltage $V_{CAREF}$. 00b = Internal reference off. An external reference can be applied. 01b = 0.25 × $V_{CC}$ 10b = 0.50 × $V_{CC}$ 11b = Diode reference is selected</td>
</tr>
<tr>
<td>3</td>
<td>CAON</td>
<td>R/W</td>
<td>0h</td>
<td>Comparator_A+ on. This bit turns on the comparator. When the comparator is off it consumes no current. The reference circuitry is enabled or disabled independently. 0b = Off 1b = On</td>
</tr>
<tr>
<td>2</td>
<td>CAIES</td>
<td>R/W</td>
<td>0h</td>
<td>Comparator_A+ interrupt edge select 0b = Rising edge 1b = Falling edge</td>
</tr>
<tr>
<td>1</td>
<td>CAIE</td>
<td>R/W</td>
<td>0h</td>
<td>Comparator_A+ interrupt enable 0b = Disabled 1b = Enabled</td>
</tr>
<tr>
<td>0</td>
<td>CAIFG</td>
<td>R/W</td>
<td>0h</td>
<td>The Comparator_A+ interrupt flag 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
</tbody>
</table>
21.3.2 CACTL2 Register

Comparator_A+ Control Register 2 Register

CACTL2 is shown in Figure 21-9 and described in Table 21-3.

Return to Table 21-1.

Figure 21-9. CACTL2 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>CASHORT</td>
<td>R/W</td>
<td>0h</td>
<td>Input short. This bit shorts the positive and negative input terminals.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Inputs not shorted</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Inputs shorted</td>
</tr>
<tr>
<td>6</td>
<td>P2CA4</td>
<td>R/W</td>
<td>0h</td>
<td>Input select. This bit together with P2CA0 selects the positive terminal</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>input when CAEX = 0 and the negative terminal input when CAEX = 1.</td>
</tr>
<tr>
<td>5</td>
<td>P2CA3</td>
<td>R/W</td>
<td>0h</td>
<td>Input select. These bits select the negative terminal input when CAEX = 0</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>and the positive terminal input when CAEX = 1.</td>
</tr>
<tr>
<td>4</td>
<td>P2CA2</td>
<td>R/W</td>
<td>0h</td>
<td>Input select. This bit, together with P2CA4, selects the positive terminal</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>input when CAEX = 0 and the negative terminal input when CAEX = 1.</td>
</tr>
<tr>
<td>3</td>
<td>P2CA1</td>
<td>R/W</td>
<td>0h</td>
<td>MSP430G2210: Only channels 2, 5, 6, and 7 are available. Other channels</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>should not be selected</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>000b = No connection</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>001b = CA1</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>010b = CA2</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>011b = CA3</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>100b = CA4</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>101b = CA5</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>110b = CA6</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>111b = CA7</td>
</tr>
<tr>
<td>2</td>
<td>P2CA0</td>
<td>R/W</td>
<td>0h</td>
<td>Input select. This bit, together with P2CA4, selects the positive terminal</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>input when CAEX = 0 and the negative terminal input when CAEX = 1.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = No connection</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = CA0</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = CA1</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = CA2</td>
</tr>
<tr>
<td>1</td>
<td>CAF</td>
<td>R/W</td>
<td>0h</td>
<td>Comparator_A+ output filter</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Comparator_A+ output is not filtered</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Comparator_A+ output is filtered</td>
</tr>
<tr>
<td>0</td>
<td>CAOUT</td>
<td>R</td>
<td>0h</td>
<td>Comparator_A+ output. This bit reflects the value of the comparator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>output. Writing this bit has no effect.</td>
</tr>
</tbody>
</table>
21.3.3 CAPD Register

Comparator_A+ Port Disable Register

CAPD is shown in Figure 21-10 and described in Table 21-4.

Return to Table 21-1.

Figure 21-10. CAPD Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>CAPDx(1)</td>
<td>R/W</td>
<td>0h</td>
<td>Comparator_A+ port disable. These bits individually disable the input buffer for the pins of the port associated with Comparator_A+. For example, if CA0 is on pin P2.3, the CAPDx bits can be used to individually enable or disable each P2.x pin buffer. CAPD0 disables P2.0, CAPD1 disables P2.1, and so forth. 0b = The input buffer is enabled. 1b = The input buffer is disabled.</td>
</tr>
</tbody>
</table>

MSP430G2210: Channels 2, 5, 6, and 7 are available. Other channels should not be disabled.

Table 21-4. CAPD Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>CAPDx(1)</td>
<td>R/W</td>
<td>0h</td>
<td>Comparator_A+ port disable. These bits individually disable the input buffer for the pins of the port associated with Comparator_A+. For example, if CA0 is on pin P2.3, the CAPDx bits can be used to individually enable or disable each P2.x pin buffer. CAPD0 disables P2.0, CAPD1 disables P2.1, and so forth. 0b = The input buffer is enabled. 1b = The input buffer is disabled.</td>
</tr>
</tbody>
</table>

(1) MSP430G2210: Channels 2, 5, 6, and 7 are available. Other channels should not be disabled.
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The ADC10 module is a high-performance 10-bit analog-to-digital converter. This chapter describes the operation of the ADC10 module of the 2xx family in general. There are devices with less than eight external input channels.

### 22.1 ADC10 Introduction

### 22.2 ADC10 Operation

### 22.3 ADC10 Registers
22.1 ADC10 Introduction

The ADC10 module supports fast, 10-bit analog-to-digital conversions. The module implements a 10-bit SAR core, sample select control, reference generator, and data transfer controller (DTC).

The DTC allows ADC10 samples to be converted and stored anywhere in memory without CPU intervention. The module can be configured with user software to support a variety of applications.

ADC10 features include:

- Greater than 200-ksp maximum conversion rate
- Monotonic 10-bit converter with no missing codes
- Sample-and-hold with programmable sample periods
- Conversion initiation by software or Timer_A
- Software selectable on-chip reference voltage generation (1.5 V or 2.5 V)
- Software selectable internal or external reference
- Up to eight external input channels (twelve on MSP430F22xx, MSP430G2x44, and MSP430G2x55 devices)
- Conversion channels for internal temperature sensor, $V_{CC}$, and external references
- Selectable conversion clock source
- Single-channel, repeated single-channel, sequence, and repeated sequence conversion modes
- ADC core and reference voltage can be powered down separately
- Data transfer controller for automatic storage of conversion results

The block diagram of ADC10 is shown in Figure 22-1.
A. Channels A12 through A15 are available in MSP430F22xx, MSP430G2x44, and MSP430G2x55 devices only. Channels A12 through A15 are tied to channel A11 in other devices. Not all channels are available in all devices.

B. TA1 on MSP430F20x2, MSP430G2x31, and MSP430G2x30 devices.

**Figure 22-1. ADC10 Block Diagram**
22.2 ADC10 Operation

The ADC10 module is configured with user software. The setup and operation of the ADC10 is discussed in the following sections.

22.2.1 10-Bit ADC Core

The ADC core converts an analog input to its 10-bit digital representation and stores the result in the ADC10MEM register. The core uses two programmable/selectable voltage levels ($V_{R+}$ and $V_{R-}$) to define the upper and lower limits of the conversion. The digital output ($N_{ADC}$) is full scale (03FFh) when the input signal is equal to or higher than $V_{R+}$, and zero when the input signal is equal to or lower than $V_{R-}$. The input channel and the reference voltage levels ($V_{R+}$ and $V_{R-}$) are defined in the conversion-control memory. Conversion results may be in straight binary format or 2s-complement format. The conversion formula for the ADC result when using straight binary format is:

$$N_{ADC} = \frac{V_{IN} - V_{R-}}{V_{R+} - V_{R-}} \times 1023$$

The ADC10 core is configured by two control registers, ADC10CTL0 and ADC10CTL1. The core is enabled with the ADC10ON bit. With few exceptions the ADC10 control bits can only be modified when ENC = 0. ENC must be set to 1 before any conversion can take place.

22.2.1.1 Conversion Clock Selection

The ADC10CLK is used both as the conversion clock and to generate the sampling period. The ADC10 source clock is selected using the ADC10SSELx bits and can be divided from 1 to 8 using the ADC10DIVx bits. Possible ADC10CLK sources are SMCLK, MCLK, ACLK, and internal oscillator ADC10OSC.

The ADC10OSC, generated internally, is in the 5-MHz range, but varies with individual devices, supply voltage, and temperature. See the device-specific data sheet for the ADC10OSC specification.

The user must ensure that the clock chosen for ADC10CLK remains active until the end of a conversion. If the clock is removed during a conversion, the operation does not complete, and any result is invalid.

22.2.2 ADC10 Inputs and Multiplexer

The eight external and four internal analog signals are selected as the channel for conversion by the analog input multiplexer. The input multiplexer is a break-before-make type to reduce input-to-input noise injection that can result from channel switching (see Figure 22-2). The input multiplexer is also a T-switch to minimize the coupling between channels. Channels that are not selected are isolated from the A/D, and the intermediate node is connected to analog ground ($V_{SS}$) so that the stray capacitance is grounded to help eliminate crosstalk.

The ADC10 uses the charge redistribution method. When the inputs are internally switched, the switching action may cause transients on the input signal. These transients decay and settle before causing errant conversion.

Figure 22-2. Analog Multiplexer

22.2.2.1 Analog Port Selection

The ADC10 external inputs $Ax$, $V_{eREF+}$, and $V_{REF-}$ share terminals with general purpose I/O ports, which are digital CMOS gates (see the device-specific data sheet). When analog signals are applied to digital CMOS
gates, parasitic current can flow from VCC to GND. This parasitic current occurs if the input voltage is near the transition level of the gate. Disabling the port pin buffer eliminates the parasitic current flow and therefore reduces overall current consumption. The ADC10AEEx bits provide the ability to disable the port pin input and output buffers.

```assembly
; P2.3 on MSP430F22xx device configured for analog input
BIS.B #$08h,4ADC10AE0 ; P2.3 ADC10 function and enable
```

Devices that do not have all the ADC10 external inputs channels Ax or V.refs/Vrefs+ and V.refs/Vrefs- available at device pins must not alter the default register bit configuration of the not available pins. See device specific data sheet.

### 22.2.3 Voltage Reference Generator

The ADC10 module contains a built-in voltage reference with two selectable voltage levels. Setting REFON = 1 enables the internal reference. When REF2_5V = 1, the internal reference is 2.5 V. When REF2_5V = 0, the reference is 1.5 V. The internal reference voltage may be used internally (REFOUT = 0) and, when REFOUT = 1, externally on pin V.refs+. REFOUT = 1 should only be used if the pins V.refs+ and V.refs- are available as device pins.

External references may be supplied for V.refs+ and V.refs through pins A4 and A3 respectively. When external references are used, or when VCC is used as the reference, the internal reference may be turned off to save power.

An external positive reference V.refs+ can be buffered by setting SREF0 = 1 and SREF1 = 1 (only devices with V.refs+ pin). This allows using an external reference with a large internal resistance at the cost of the buffer current. When REFBURST = 1 the increased current consumption is limited to the sample and conversion period.

External storage capacitance is not required for the ADC10 reference source as on the ADC12.

#### 22.2.3.1 Internal Reference Low-Power Features

The ADC10 internal reference generator is designed for low power applications. The reference generator includes a band-gap voltage source and a separate buffer. The current consumption of each is specified separately in the device-specific data sheet. When REFON = 1, both are enabled and when REFON = 0 both are disabled. The total settling time when REFON becomes set is approximately 30 µs.

When REFON = 1, but no conversion is active, the buffer is automatically disabled and automatically re-enabled when needed. When the buffer is disabled, it consumes no current. In this case, the bandgap voltage source remains enabled.

When REFOUT = 1, the REFBURST bit controls the operation of the internal reference buffer. When REFBURST = 0, the buffer is on continuously, allowing the reference voltage to be present outside the device continuously. When REFBURST = 1, the buffer is automatically disabled when the ADC10 is not actively converting and is automatically re-enabled when needed.

The internal reference buffer also has selectable speed versus power settings. When the maximum conversion rate is below 50 ksp, setting ADC10SR = 1 reduces the current consumption of the buffer approximately 50%.

### 22.2.4 Auto Power-Down

The ADC10 is designed for low power applications. When the ADC10 is not actively converting, the core is automatically disabled and is automatically re-enabled when needed. The ADC10OSC is also automatically enabled when needed and disabled when not needed. When the core or oscillator is disabled, it consumes no current.

### 22.2.5 Sample and Conversion Timing

An analog-to-digital conversion is initiated with a rising edge of sample input signal SHI. The source for SHI is selected with the SHSx bits and includes the following:

- The ADC10SC bit
The polarity of the SHI signal source can be inverted with the ISSH bit. The SHTx bits select the sample period $t_{sample}$ to be 4, 8, 16, or 64 ADC10CLK cycles. The sampling timer sets SAMPCON high for the selected sample period after synchronization with ADC10CLK. Total sampling time is $t_{sample} + t_{sync}$. The high-to-low SAMPCON transition starts the analog-to-digital conversion, which requires 13 ADC10CLK cycles as shown in **Figure 22-3**.

**Figure 22-3. Sample Timing**

22.2.5.1 Sample Timing Considerations

When SAMPCON = 0 all Ax inputs are high impedance. When SAMPCON = 1, the selected Ax input can be modeled as an RC low-pass filter during the sampling time $t_{sample}$, as shown in **Figure 22-4**. An internal MUX-on input resistance $R_I$ (2 kΩ maximum) in series with capacitor $C_I$ (27 pF maximum) is seen by the source. The capacitor $C_I$ voltage $V_C$ must be charged to within ½ LSB of the source voltage $V_S$ for an accurate 10-bit conversion.

**Figure 22-4. Analog Input Equivalent Circuit**

The resistance of the source $R_S$ and $R_I$ affect $t_{sample}$. The following equations can be used to calculate the minimum sampling time for a 10-bit conversion.

$t_{sample} > (R_S + R_I) \times \ln(2^{11}) \times C_I$

Substituting the values for $R_I$ and $C_I$ given above, the equation becomes:

$t_{sample} > (R_S + 2 \text{kΩ}) \times 7.625 \times 27 \text{pF}$

For example, if $R_S$ is 10 kΩ, $t_{sample}$ must be greater than 2.47 μs.

When the reference buffer is used in burst mode, the sampling time must be greater than the sampling time calculated and the settling time of the buffer, $t_{REFBURST}$.
\[ t_{\text{sample}} > \left\{ \begin{align*} \frac{(R_S + R_I) \times \ln(2^{11}) \times C_i}{t_{\text{REFBURST}}} \\ t_{\text{REFBURST}} \end{align*} \right. \]

For example, if \( V_{\text{Ref}} \) is 1.5 V and \( R_S \) is 10 kΩ, \( t_{\text{sample}} \) must be greater than 2.47 µs when ADC10SR = 0, or 2.5 µs when ADC10SR = 1. See the device-specific data sheet for parameters.

To calculate the buffer settling time when using an external reference, the formula is:

\[ t_{\text{REFBURST}} = S_R \times V_{\text{Ref}} - 0.5 \mu s \]

Where:

\( S_R \) = Buffer slew rate (~1 µs/V when ADC10SR = 0 and ~2 µs/V when ADC10SR = 1)

\( V_{\text{Ref}} \) = External reference voltage

### 22.2.6 Conversion Modes

The ADC10 has four operating modes selected by the CONSEQx bits as discussed in Table 22-1.

<table>
<thead>
<tr>
<th>CONSEQx</th>
<th>Mode</th>
<th>Operation</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>Single channel single-conversion</td>
<td>A single channel is converted once.</td>
</tr>
<tr>
<td>01</td>
<td>Sequence-of-channels</td>
<td>A sequence of channels is converted once.</td>
</tr>
<tr>
<td>10</td>
<td>Repeat single channel</td>
<td>A single channel is converted repeatedly.</td>
</tr>
<tr>
<td>11</td>
<td>Repeat sequence-of-channels</td>
<td>A sequence of channels is converted repeatedly.</td>
</tr>
</tbody>
</table>
22.2.6.1 Single-Channel Single-Conversion Mode

A single channel selected by INCHx is sampled and converted once. The ADC result is written to ADC10MEM. Figure 22-5 shows the flow of the single-channel, single-conversion mode. When ADC10SC triggers a conversion, successive conversions can be triggered by the ADC10SC bit. When any other trigger source is used, ENC must be toggled between each conversion.

Figure 22-5. Single-Channel Single-Conversion Mode
22.2.6.2 Sequence-of-Channels Mode

A sequence of channels is sampled and converted once. The sequence begins with the channel selected by INCHx and decrements to channel A0. Each ADC result is written to ADC10MEM. The sequence stops after conversion of channel A0. Figure 22-6 shows the sequence-of-channels mode. When ADC10SC triggers a sequence, successive sequences can be triggered by the ADC10SC bit. When any other trigger source is used, ENC must be toggled between each sequence.

Figure 22-6. Sequence-of-Channels Mode
### 22.2.6.3 Repeat-Single-Channel Mode

A single channel selected by INCHx is sampled and converted continuously. Each ADC result is written to ADC10MEM. **Figure 22-7** shows the repeat-single-channel mode.

---

**Figure 22-7. Repeat-Single-Channel Mode**

- **ADC10 Off**
- **x = INCHx**
- **Wait for Enable**
  - **ENC**
- **Wait for Trigger**
  - **ENC**
- **Sample, Input Channel Ax**
  - **MSC = 1 and ENC = 1**
- **Convert**
  - **12 x ADC10CLK**
- **Conversion Completed, Result to ADC10MEM, ADC10IFG is Set**
  - **1 x ADC10CLK**

---

x = input channel Ax
22.2.6.4 Repeat-Sequence-of-Channels Mode

A sequence of channels is sampled and converted repeatedly. The sequence begins with the channel selected by INCHx and decrements to channel A0. Each ADC result is written to ADC10MEM. The sequence ends after conversion of channel A0, and the next trigger signal re-starts the sequence. Figure 22-8 shows the repeat-sequence-of-channels mode.

![Figure 22-8. Repeat-Sequence-of-Channels Mode](image-url)
22.2.6.5 Using the MSC Bit
To configure the converter to perform successive conversions automatically and as quickly as possible, a multiple sample and convert function is available. When MSC = 1 and CONSEQx > 0, the first rising edge of the SHI signal triggers the first conversion. Successive conversions are triggered automatically as soon as the prior conversion is completed. Additional rising edges on SHI are ignored until the sequence is completed in the single-sequence mode or until the ENC bit is toggled in repeat-single-channel, or repeated-sequence modes. The function of the ENC bit is unchanged when using the MSC bit.

22.2.6.6 Stopping Conversions
Stopping ADC10 activity depends on the mode of operation. The recommended ways to stop an active conversion or conversion sequence are:

- Resetting ENC in single-channel single-conversion mode stops a conversion immediately and the results are unpredictable. For correct results, poll the ADC10BUSY bit until reset before clearing ENC.
- Resetting ENC during repeat-single-channel operation stops the converter at the end of the current conversion.
- Resetting ENC during a sequence or repeat sequence mode stops the converter at the end of the sequence.
- Any conversion mode may be stopped immediately by setting the CONSEQx = 0 and resetting the ENC bit. Conversion data is unreliable.

22.2.7 ADC10 Data Transfer Controller
The ADC10 includes a data transfer controller (DTC) to automatically transfer conversion results from ADC10MEM to other on-chip memory locations. The DTC is enabled by setting the ADC10DTC1 register to a nonzero value.

When the DTC is enabled, each time the ADC10 completes a conversion and loads the result to ADC10MEM, a data transfer is triggered. No software intervention is required to manage the ADC10 until the predefined amount of conversion data has been transferred. Each DTC transfer requires one CPU MCLK. To avoid any bus contention during the DTC transfer, the CPU is halted, if active, for the one MCLK required for the transfer.

A DTC transfer must not be initiated while the ADC10 is busy. Software must ensure that no active conversion or sequence is in progress when the DTC is configured:

```assembly
; ADC10 activity test
BIC.W  #ENC,&ADC10CTL0  ;
basic_test BIT.W  #BUSY,&ADC10CTL1  ;
JNZ     basic_test        ;
MOV.W   #xxx,&ADC10SA  ; Safe
MOV.B   #xx,&ADC10DTC1  ;
; continue setup
```
22.2.7.1 One-Block Transfer Mode

The one-block mode is selected if the ADC10TB is reset. The value n in ADC10DTC1 defines the total number of transfers for a block. The block start address is defined anywhere in the MSP430 address range using the 16-bit register ADC10SA. The block ends at ADC10SA + 2n – 2. The one-block transfer mode is shown in Figure 22-9.

![Figure 22-9. One-Block Transfer](image)

The internal address pointer is initially equal to ADC10SA and the internal transfer counter is initially equal to 'n'. The internal pointer and counter are not visible to software. The DTC transfers the word-value of ADC10MEM to the address pointer ADC10SA. After each DTC transfer, the internal address pointer is incremented by two and the internal transfer counter is decremented by one.

The DTC transfers continue with each loading of ADC10MEM, until the internal transfer counter becomes equal to zero. No additional DTC transfers occur until a write to ADC10SA. When using the DTC in the one-block mode, the ADC10IFG flag is set only after a complete block has been transferred. Figure 22-10 shows a state diagram of the one-block mode.
Figure 22-10. State Diagram for Data Transfer Control in One-Block Transfer Mode
22.2.7.2 Two-Block Transfer Mode

The two-block mode is selected if the ADC10TB bit is set. The value \( n \) in ADC10DTC1 defines the number of transfers for one block. The address range of the first block is defined anywhere in the MSP430 address range with the 16-bit register ADC10SA. The first block ends at ADC10SA+2n-2. The address range for the second block is defined as SA+2n to SA+4n-2. The two-block transfer mode is shown in Figure 22-11.

The internal address pointer is initially equal to ADC10SA and the internal transfer counter is initially equal to 'n'. The internal pointer and counter are not visible to software. The DTC transfers the word-value of ADC10MEM to the address pointer ADC10SA. After each DTC transfer the internal address pointer is incremented by two and the internal transfer counter is decremented by one.

The DTC transfers continue, with each loading of ADC10MEM, until the internal transfer counter becomes equal to zero. At this point, block one is full and both the ADC10IFG flag the ADC10B1 bit are set. The user can test the ADC10B1 bit to determine that block one is full.

The DTC continues with block two. The internal transfer counter is automatically reloaded with 'n'. At the next load of the ADC10MEM, the DTC begins transferring conversion results to block two. After \( n \) transfers have completed, block two is full. The ADC10IFG flag is set and the ADC10B1 bit is cleared. User software can test the cleared ADC10B1 bit to determine that block two is full. Figure 22-12 shows a state diagram of the two-block mode.

![Figure 22-11. Two-Block Transfer](image-url)
Figure 22-12. State Diagram for Data Transfer Control in Two-Block Transfer Mode
22.2.7.3 Continuous Transfer

A continuous transfer is selected if ADC10CT bit is set. The DTC does not stop after block one (in one-block mode) or block two (in two-block mode) has been transferred. The internal address pointer and transfer counter are set equal to ADC10SA and n respectively. Transfers continue starting in block one. If the ADC10CT bit is reset, DTC transfers cease after the current completion of transfers into block one (in one-block mode) or block two (in two-block mode) have been transferred.

22.2.7.4 DTC Transfer Cycle Time

For each ADC10MEM transfer, the DTC requires one or two MCLK clock cycles to synchronize, one for the actual transfer (while the CPU is halted), and one cycle of wait time. Because the DTC uses MCLK, the DTC cycle time is dependent on the MSP430 operating mode and clock system setup.

If the MCLK source is active but the CPU is off, the DTC uses the MCLK source for each transfer, without re-enabling the CPU. If the MCLK source is off, the DTC temporarily restarts MCLK, sourced with DCOCLK, only during a transfer. The CPU remains off, and MCLK is again turned off after the DTC transfer. The maximum DTC cycle time for all operating modes is shown in Table 22-2.

<table>
<thead>
<tr>
<th>CPU Operating Mode</th>
<th>Clock Source</th>
<th>Maximum DTC Cycle Time</th>
</tr>
</thead>
<tbody>
<tr>
<td>Active mode</td>
<td>MCLK = DCOCLK</td>
<td>3 MCLK cycles</td>
</tr>
<tr>
<td>Active mode</td>
<td>MCLK = LFXT1CLK</td>
<td>3 MCLK cycles</td>
</tr>
<tr>
<td>Low-power mode LPM0/1</td>
<td>MCLK = DCOCLK</td>
<td>4 MCLK cycles</td>
</tr>
<tr>
<td>Low-power mode LPM3/4</td>
<td>MCLK = DCOCLK</td>
<td>4 MCLK cycles + 2 µs(1)</td>
</tr>
<tr>
<td>Low-power mode LPM0/1</td>
<td>MCLK = LFXT1CLK</td>
<td>4 MCLK cycles</td>
</tr>
<tr>
<td>Low-power mode LPM3</td>
<td>MCLK = LFXT1CLK</td>
<td>4 MCLK cycles</td>
</tr>
<tr>
<td>Low-power mode LPM4</td>
<td>MCLK = LFXT1CLK</td>
<td>4 MCLK cycles + 2 µs(1)</td>
</tr>
</tbody>
</table>

(1) The additional 2 µs are needed to start the DCOCLK. See the device-specific data sheet for parameters.

22.2.8 Using the Integrated Temperature Sensor

To use the on-chip temperature sensor, select the analog input channel INCHx = 1010. Any other configuration is done as if an external channel was selected, including reference selection and conversion-memory selection.

The typical temperature sensor transfer function is shown in Figure 22-13. When using the temperature sensor, the sample period must be greater than 30 µs. The temperature sensor offset error is large. Deriving absolute temperature values in the application requires calibration. See the device-specific data sheet for the parameters. See Section 24.2.2.1 for the calibration equations.

Selecting the temperature sensor automatically turns on the on-chip reference generator as a voltage source for the temperature sensor. However, it does not enable the VREF+ output or affect the reference selections for the conversion. The reference choices for converting the temperature sensor are the same as with any other channel.
22.2.9 ADC10 Grounding and Noise Considerations

As with any high-resolution ADC, appropriate printed-circuit-board layout and grounding techniques should be followed to eliminate ground loops, unwanted parasitic effects, and noise.

Ground loops are formed when return current from the A/D flows through paths that are common with other analog or digital circuitry. If care is not taken, this current can generate small, unwanted offset voltages that can add to or subtract from the reference or input voltages of the A/D converter. The connections shown in Figure 22-14 and Figure 22-15 help avoid this.

In addition to grounding, ripple and noise spikes on the power supply lines due to digital switching or switching power supplies can corrupt the conversion result. A noise-free design is important to achieve high accuracy.

![Figure 22-13. Typical Temperature Sensor Transfer Function](image)

\[ V_{\text{TEMP}} = 0.00355(\text{TEMP}_C) + 0.986 \]

![Figure 22-14. ADC10 Grounding and Noise Considerations (Internal V_{\text{REF}})](image)
22.2.10 ADC10 Interrupts

One interrupt and one interrupt vector are associated with the ADC10 as shown in Figure 22-16. When the DTC is not used (ADC10DTC1 = 0), ADC10IFG is set when conversion results are loaded into ADC10MEM. When DTC is used (ADC10DTC1 > 0), ADC10IFG is set when a block transfer completes and the internal transfer counter n = 0. If both the ADC10IE and the GIE bits are set, then the ADC10IFG flag generates an interrupt request. The ADC10IFG flag is automatically reset when the interrupt request is serviced, or it may be reset by software.
## 22.3 ADC10 Registers

Table 22-3 lists the memory-mapped registers for the ADC10.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>1B0h</td>
<td>ADC10CTL0</td>
<td>ADC10 control 0</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 22.3.1</td>
</tr>
<tr>
<td>1B2h</td>
<td>ADC10CTL1</td>
<td>ADC10 control 1</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 22.3.2</td>
</tr>
<tr>
<td>4Ah</td>
<td>ADC10AE0</td>
<td>ADC10 input enable 0</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 22.3.3</td>
</tr>
<tr>
<td>4Bh</td>
<td>ADC10AE1</td>
<td>ADC10 input enable 1</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 22.3.4</td>
</tr>
<tr>
<td>1B4h</td>
<td>ADC10MEM</td>
<td>ADC10 memory</td>
<td>Read</td>
<td>Unchanged</td>
<td>Section 22.3.5</td>
</tr>
<tr>
<td>48h</td>
<td>ADC10DTC0</td>
<td>ADC10 data transfer control 0</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 22.3.6</td>
</tr>
<tr>
<td>49h</td>
<td>ADC10DTC1</td>
<td>ADC10 data transfer control 1</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 22.3.7</td>
</tr>
<tr>
<td>1BCh</td>
<td>ADC10SA</td>
<td>ADC10 data transfer start address</td>
<td>Read/write</td>
<td>200h with POR</td>
<td>Section 22.3.8</td>
</tr>
</tbody>
</table>
22.3.1 ADC10CTL0 Register

ADC10 Control 0 Register

ADC10CTL0 is shown in Figure 22-17 and described in Table 22-4.

Return to Table 22-3.

ADC10 Control Register 0

![Figure 22-17. ADC10CTL0 Register](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-13</td>
<td>SREFx</td>
<td>R/W</td>
<td>0h</td>
<td>Select reference. Can be modified only when ENC = 0.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = V_{R+} = V_{CC} and V_{R-} = V_{SS}</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>001b = V_{R+} = V_{REF+} and V_{R-} = V_{SS}</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>010b = V_{R+} = V_{eREF+} and V_{R-} = V_{SS}. Devices with V_{eREF+} only.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>011b = V_{R+} = Buffered V_{eREF+} and V_{R-} = V_{SS}. Devices with V_{eREF+} pin only.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>100b = V_{R+} = V_{CC} and V_{R-} = V_{REF+} / V_{eREF+}. Devices with V_{REF+} pin only.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>101b = V_{R+} = V_{REF+} and V_{R-} = V_{REF+} / V_{eREF+}. Devices with V_{REF+} and V_{eREF+} pins only.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>110b = V_{R+} = V_{eREF+} and V_{R-} = V_{REF+} / V_{eREF+}. Devices with V_{REF+} and V_{eREF+} pins only.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>111b = V_{R+} = Buffered V_{eREF+} and V_{R-} = V_{REF+} / V_{eREF+}. Devices with V_{eREF+} and V_{eREF+} pins only.</td>
</tr>
<tr>
<td>12-11</td>
<td>ADC10SHTx</td>
<td>R/W</td>
<td>0h</td>
<td>ADC10 sample-and-hold time. Can be modified only when ENC = 0.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = 4 ADC10CLK cycles</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = 8 ADC10CLK cycles</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = 16 ADC10CLK cycles</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = 64 ADC10CLK cycles</td>
</tr>
<tr>
<td>10</td>
<td>ADC10SR</td>
<td>R/W</td>
<td>0h</td>
<td>ADC10 sampling rate. This bit selects the reference buffer drive capability for the maximum sampling rate. Setting ADC10SR reduces the current consumption of the reference buffer. Can be modified only when ENC = 0.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Reference buffer supports up to approximately 200 ksps</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Reference buffer supports up to approximately 50 ksps</td>
</tr>
<tr>
<td>9</td>
<td>REFOUT</td>
<td>R/W</td>
<td>0h</td>
<td>Reference output. Can be modified only when ENC = 0.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Reference output off</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Reference output on. Devices with V_{eREF+} / V_{REF+} pin only.</td>
</tr>
</tbody>
</table>
### Table 22-4. ADC10CTL0 Register Field Descriptions (continued)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>8</td>
<td>REFBURST</td>
<td>R/W</td>
<td>0h</td>
<td>Reference burst. Can be modified only when ENC = 0. 0b = Reference buffer on continuously 1b = Reference buffer on only during sample-and-conversion</td>
</tr>
<tr>
<td>7</td>
<td>MSC</td>
<td>R/W</td>
<td>0h</td>
<td>Multiple sample and conversion. Valid only for sequence or repeated modes. 0b = The sampling requires a rising edge of the SHI signal to trigger each sample-and-conversion. 1b = The first rising edge of the SHI signal triggers the sampling timer, but further sample-and-conversions are performed automatically as soon as the prior conversion is completed</td>
</tr>
<tr>
<td>6</td>
<td>REF2_5V</td>
<td>R/W</td>
<td>0h</td>
<td>Reference-generator voltage. REFON must also be set. 0b = 1.5 V 1b = 2.5 V</td>
</tr>
<tr>
<td>5</td>
<td>REFON</td>
<td>R/W</td>
<td>0h</td>
<td>Reference generator on. Can be modified only when ENC = 0. 0b = Reference off 1b = Reference on</td>
</tr>
<tr>
<td>4</td>
<td>ADC10ON</td>
<td>R/W</td>
<td>0h</td>
<td>ADC10 on. Can be modified only when ENC = 0. 0b = ADC10 off 1b = ADC10 on</td>
</tr>
<tr>
<td>3</td>
<td>ADC10IE</td>
<td>R/W</td>
<td>0h</td>
<td>ADC10 interrupt enable 0b = Interrupt disabled 1b = Interrupt enabled</td>
</tr>
<tr>
<td>2</td>
<td>ADC10IFG</td>
<td>R/W</td>
<td>0h</td>
<td>ADC10 interrupt flag. This bit is set if ADC10MEM is loaded with a conversion result. It is automatically reset when the interrupt request is accepted, or it may be reset by software. When using the DTC, this flag is set when a block of transfers is completed. 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
<tr>
<td>1</td>
<td>ENC</td>
<td>R/W</td>
<td>0h</td>
<td>Enable conversion 0b = ADC10 disabled 1b = ADC10 enabled</td>
</tr>
<tr>
<td>0</td>
<td>ADC10SC</td>
<td>R/W</td>
<td>0h</td>
<td>Start conversion. Software-controlled sample-and-conversion start. ADC10SC and ENC may be set together with one instruction. ADC10SC is reset automatically. 0b = No sample-and-conversion start 1b = Start sample-and-conversion</td>
</tr>
</tbody>
</table>
22.3.2 ADC10CTL1 Register

ADC10 Control 1 Register

ADC10CTL1 is shown in Figure 22-18 and described in Table 22-5.

Return to Table 22-3.

ADC10 Control Register 1

Figure 22-18. ADC10CTL1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-12</td>
<td>INCHx</td>
<td>R/W</td>
<td>0h</td>
<td>Input channel select. These bits select the channel for a single-conversion or the highest channel for a sequence of conversions. Only available ADC channels should be selected. See the device-specific data sheet. Can be modified only when ENC = 0. 0000b = A0 0001b = A1 0010b = A2 0011b = A3 0100b = A4 0101b = A5 0110b = A6 0111b = A7 1000b = ( V_{\text{REF}+} ) 1001b = ( V_{\text{REF}}/V_{\text{REF}} ) 1010b = Temperature sensor 1011b = ( (V_{\text{CC}} - V_{\text{SS}}) / 2 ) 1100b = ( (V_{\text{CC}} - V_{\text{SS}}) / 2, A12 on MSP430F22xx, MSP430G2x44, and MSP430G2x55 devices ) 1101b = ( (V_{\text{CC}} - V_{\text{SS}}) / 2, A13 on MSP430F22xx, MSP430G2x44, and MSP430G2x55 devices ) 1110b = ( (V_{\text{CC}} - V_{\text{SS}}) / 2, A14 on MSP430F22xx, MSP430G2x44, and MSP430G2x55 devices ) 1111b = ( (V_{\text{CC}} - V_{\text{SS}}) / 2, A15 on MSP430F22xx, MSP430G2x44, and MSP430G2x55 devices )</td>
</tr>
<tr>
<td>11-10</td>
<td>SHSx</td>
<td>R/W</td>
<td>0h</td>
<td>Sample-and-hold source select. Can be modified only when ENC = 0. 00b = ADC10SC bit 01b = Timer_A.OUT1(1) 10b = Timer_A.OUT0(1) 11b = Timer_A.OUT2 (Timer_A.OUT1 on MSP430F20x0, MSP430G2x31, and MSP430G2x30 devices)(1)</td>
</tr>
</tbody>
</table>

(1) Source select bit for Timer_A OUT1 and OUT2.
Table 22-5. ADC10CTL1 Register Field Descriptions (continued)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>9</td>
<td>ADC10DF</td>
<td>R/W</td>
<td>0h</td>
<td>ADC10 data format. Can be modified only when ENC = 0. 0b = Straight binary 1b = 2s complement</td>
</tr>
<tr>
<td>8</td>
<td>ISSH</td>
<td>R/W</td>
<td>0h</td>
<td>Invert signal sample-and-hold. Can be modified only when ENC = 0. 0b = The sample-input signal is not inverted. 1b = The sample-input signal is inverted.</td>
</tr>
<tr>
<td>7-5</td>
<td>ADC10DIVx</td>
<td>R/W</td>
<td>0h</td>
<td>ADC10 clock divider. Can be modified only when ENC = 0. 000b = /1 001b = /2 010b = /3 011b = /4 100b = /5 101b = /6 110b = /7 111b = /8</td>
</tr>
<tr>
<td>4-3</td>
<td>ADC10SSELx</td>
<td>R/W</td>
<td>0h</td>
<td>ADC10 clock source select. Can be modified only when ENC = 0. 00b = ADC10OSC 01b = ACLK 10b = MCLK 11b = SMCLK</td>
</tr>
<tr>
<td>2-1</td>
<td>CONSEQx</td>
<td>R/W</td>
<td>0h</td>
<td>Conversion sequence mode select 00b = Single-channel single-conversion mode 01b = Sequence-of-channels mode 10b = Repeat-single-channel mode 11b = Repeat-sequence-of-channels mode</td>
</tr>
<tr>
<td>0</td>
<td>ADC10BUSY</td>
<td>R</td>
<td>0h</td>
<td>ADC10 busy. This bit indicates an active sample or conversion operation 0b = No operation is active. 1b = A sequence, sample, or conversion is active.</td>
</tr>
</tbody>
</table>

(1) Timer triggers are from Timer0_Ax if more than one timer module exists on the device.
22.3.3 ADC10AE0 Register

ADC10 Input Enable 0 Register

ADC10AE0 is shown in Figure 22-19 and described in Table 22-6.

Return to Table 22-3.

Analog (Input) Enable Control Register 0

<table>
<thead>
<tr>
<th>Figure 22-19. ADC10AE0 Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0 ADC10AE0x</td>
</tr>
<tr>
<td>rw-(0)</td>
</tr>
</tbody>
</table>

Table 22-6. ADC10AE0 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>ADC10AE0x</td>
<td>R/W</td>
<td>0h</td>
<td>ADC10 analog enable. These bits enable the corresponding pin for analog input. Bit 0 corresponds to A0, Bit 1 corresponds to A1, and so on. The analog enable bit of not implemented channels should not be programmed to 1. 0b = Analog input disabled 1b = Analog input enabled</td>
</tr>
</tbody>
</table>

22.3.4 ADC10AE1 Register

ADC10 Input Enable 1 Register

ADC10AE1 is shown in Figure 22-20 and described in Table 22-7.

Return to Table 22-3.

Analog (Input) Enable Control Register 1 (MSP430F22xx, MSP430G2x44, and MSP430G2x55 Only)

<table>
<thead>
<tr>
<th>Figure 22-20. ADC10AE1 Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4 ADC10AE1x</td>
</tr>
<tr>
<td>rw-(0)</td>
</tr>
</tbody>
</table>

Table 22-7. ADC10AE1 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td>ADC10AE1x</td>
<td>R/W</td>
<td>0h</td>
<td>ADC10 analog enable. These bits enable the corresponding pin for analog input. Bit 4 corresponds to A12, Bit 5 corresponds to A13, Bit 6 corresponds to A14, and Bit 7 corresponds to A15. The analog enable bit of not implemented channels should not be programmed to 1. 0b = Analog input disabled 1b = Analog input enabled</td>
</tr>
<tr>
<td>3-0</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td>Reserved</td>
</tr>
</tbody>
</table>

22.3.5 ADC10MEM Register

ADC10 Memory Register

ADC10MEM is shown in Figure 22-21 and described in Table 22-8.

Conversion-Memory Register. This register is read as either right-justified straight-binary format or left-justified 2s-complement format, depending on the value of the ADC10DF bit in the ADC10CTL1 register.

Figure 22-21. ADC10MEM Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>Conversion_Results</td>
<td>R</td>
<td>Unchanged</td>
<td>If ADC10DF = 0, the 10-bit conversion results are right-justified straight-binary format. Bit 9 is the MSB. 15-10 are always 0. If ADC10DF = 1, the 10-bit conversion results are left-justified 2s-complement format. Bit 15 is the MSB. Bits 5-0 are always 0.</td>
</tr>
</tbody>
</table>
22.3.6 ADC10DTC0 Register

ADC10 Data Transfer Control 0 Register

ADC10DTC0 is shown in Figure 22-22 and described in Table 22-9.

Return to Table 22-3.

Data Transfer Control Register 0

Figure 22-22. ADC10DTC0 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-4</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td>Reserved. Always read as 0.</td>
</tr>
<tr>
<td>3</td>
<td>ADC10TB</td>
<td>R/W</td>
<td>0h</td>
<td>ADC10 two-block mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = One-block transfer mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Two-block transfer mode</td>
</tr>
<tr>
<td>2</td>
<td>ADC10CT</td>
<td>R/W</td>
<td>0h</td>
<td>ADC10 continuous transfer</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Data transfer stops when one block (one-block mode) or two blocks (two-block mode) have completed.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Data is transferred continuously. DTC operation is stopped only if ADC10CT cleared, or ADC10SA is written to.</td>
</tr>
<tr>
<td>1</td>
<td>ADC10B1</td>
<td>R</td>
<td>0h</td>
<td>ADC10 block one.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>This bit indicates for two-block mode which block is filled with ADC10 conversion results. ADC10B1 is valid only after ADC10IFG has been set the first time during DTC operation. ADC10TB must also be set.</td>
</tr>
<tr>
<td>0</td>
<td>ADC10FETCH</td>
<td>R/W</td>
<td>0h</td>
<td>This bit should normally be reset.</td>
</tr>
</tbody>
</table>

22.3.7 ADC10DTC1 Register

ADC10 Data Transfer Control 1 Register

ADC10DTC1 is shown in Figure 22-23 and described in Table 22-10.

Return to Table 22-3.

Data Transfer Control Register 1

Figure 22-23. ADC10DTC1 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>DTC Transfers</td>
<td>R/W</td>
<td>0h</td>
<td>DTC transfers. These bits define the number of transfers in each block.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0h = DTC is disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1h–FFh = Number of transfers per block</td>
</tr>
</tbody>
</table>
22.3.8 ADC10SA Register

ADC10 Data Transfer Start Address Register

ADC10SA is shown in Figure 22-24 and described in Table 22-11.

Return to Table 22-3.

Start Address Register for Data Transfer

Figure 22-24. ADC10SA Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-1</td>
<td>ADC10SAx</td>
<td>R/W</td>
<td>0h</td>
<td>ADC10 start address. These bits are the start address for the DTC. A write to register ADC10SA is required to initiate DTC transfers.</td>
</tr>
<tr>
<td>0</td>
<td>Unused</td>
<td>R</td>
<td>0h</td>
<td>Unused, Read only. Always read as 0.</td>
</tr>
</tbody>
</table>

Table 22-11. ADC10SA Register Field Descriptions
The ADC12 module is a high-performance 12-bit analog-to-digital converter. This chapter describes the ADC12 of the MSP430x2xx device family.

<table>
<thead>
<tr>
<th>Section</th>
<th>Page</th>
</tr>
</thead>
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<td>608</td>
</tr>
<tr>
<td>23.2 ADC12 Operation</td>
<td>610</td>
</tr>
<tr>
<td>23.3 ADC12 Registers</td>
<td>622</td>
</tr>
</tbody>
</table>
23.1 ADC12 Introduction

The ADC12 module supports fast 12-bit analog-to-digital conversions. The module implements a 12-bit SAR core, sample select control, reference generator, and a 16-word conversion-and-control buffer. The conversion-and-control buffer allows up to 16 independent ADC samples to be converted and stored without any CPU intervention.

ADC12 features include:

- Greater than 200-ksp/s maximum conversion rate
- Monotonic 12-bit converter with no missing codes
- Sample-and-hold with programmable sampling periods controlled by software or timers
- Conversion initiation by software, Timer_A, or Timer_B
- Software selectable on-chip reference voltage generation (1.5 V or 2.5 V)
- Software selectable internal or external reference
- Eight individually configurable external input channels
- Conversion channels for internal temperature sensor, AV_{CC}, and external references
- Independent channel-selectable reference sources for both positive and negative references
- Selectable conversion clock source
- Single-channel, repeat-single-channel, sequence, and repeat-sequence conversion modes
- ADC core and reference voltage can be powered down separately
- Interrupt vector register for fast decoding of 18 ADC interrupts
- 16 conversion-result storage registers

The block diagram of ADC12 is shown in Figure 23-1.
Figure 23-1. ADC12 Block Diagram
23.2 ADC12 Operation

The ADC12 module is configured with user software. The setup and operation of the ADC12 is discussed in the following sections.

23.2.1 12-Bit ADC Core

The ADC core converts an analog input to its 12-bit digital representation and stores the result in conversion memory. The core uses two programmable/selectable voltage levels (VR+ and VR−) to define the upper and lower limits of the conversion. The digital output (NADC) is full scale (0FFFh) when the input signal is equal to or higher than VR+, and the digital output is zero when the input signal is equal to or lower than VR−. The input channel and the reference voltage levels (VR+ and VR−) are defined in the conversion-control memory. The conversion formula for the ADC result NADC is:

\[ N_{ADC} = 4095 \times \frac{V_{IN} - V_{R-}}{V_{R+} - V_{R-}} \]

The ADC12 core is configured by two control registers, ADC12CTL0 and ADC12CTL1. The core is enabled with the ADC12ON bit. The ADC12 can be turned off when not in use to save power. With few exceptions, the ADC12 control bits can only be modified when ENC = 0. ENC must be set to 1 before any conversion can take place.

23.2.1.1 Conversion Clock Selection

The ADC12CLK is used both as the conversion clock and to generate the sampling period when the pulse sampling mode is selected. The ADC12 source clock is selected using the ADC12SSELx bits and can be divided from 1 through 8 using the ADC12DIVx bits. Possible ADC12CLK sources are SMCLK, MCLK, ACLK, and an internal oscillator ADC12OSC.

The ADC12OSC is generated internally and is in the 5-MHz range, but the frequency varies with individual devices, supply voltage, and temperature. See the device-specific data sheet for the ADC12OSC specification.

The application must ensure that the clock chosen for ADC12CLK remains active until the end of a conversion. If the clock is removed during a conversion, the operation does not complete and any result is invalid.

23.2.2 ADC12 Inputs and Multiplexer

The eight external and four internal analog signals are selected as the channel for conversion by the analog input multiplexer. The input multiplexer is a break-before-make type to reduce input-to-input noise injection that can result from channel switching (see Figure 23-2). The input multiplexer is also a T-switch to minimize the coupling between channels. Channels that are not selected are isolated from the A/D, and the intermediate node is connected to analog ground (AVSS) so that the stray capacitance is grounded to help eliminate crosstalk.

The ADC12 uses the charge redistribution method. When the inputs are internally switched, the switching action may cause transients on the input signal. These transients decay and settle before causing errant conversion.

![Figure 23-2. Analog Multiplexer](image)
23.2.2.1 Analog Port Selection

The ADC12 inputs are multiplexed with the port P6 pins, which are digital CMOS gates. When analog signals are applied to digital CMOS gates, parasitic current can flow from VCC to GND. This parasitic current occurs if the input voltage is near the transition level of the gate. Disabling the port pin buffer eliminates the parasitic current flow and, therefore, reduces overall current consumption. The P6SELx bits provide the ability to disable the port pin input and output buffers.

; P6.0 and P6.1 configured for analog input
   BIS.B #3h,&P6SEL ; P6.1 and P6.0 ADC12 function

23.2.3 Voltage Reference Generator

The ADC12 module contains a built-in voltage reference with two selectable voltage levels, 1.5 V and 2.5 V. Either of these reference voltages may be used internally and externally on pin VREF+.

Setting REFON = 1 enables the internal reference. When REF2_5V = 1, the internal reference is 2.5 V. When REF2_5V = 0, the reference is 1.5 V. The reference can be turned off to save power when not in use.

For proper operation, the internal voltage reference generator must be supplied with storage capacitance across VREF+ and AVSS. The recommended storage capacitance is a parallel combination of 10-µF and 0.1-µF capacitors. From turn-on, a maximum of 17 ms must be allowed for the voltage reference generator to bias the recommended storage capacitors. If the internal reference generator is not used for the conversion, the storage capacitors are not required.

---

Note

Reference Decoupling

Approximately 200 µA is required from any reference used by the ADC12 while the two LSBs are being resolved during a conversion. A parallel combination of 10-µF and 0.1-µF capacitors is recommended for any reference as shown in Figure 23-11.

External references may be supplied for VR+ and VR through pins VR+ and VR+ respectively.

23.2.4 Sample and Conversion Timing

An analog-to-digital conversion is initiated with a rising edge of the sample input signal SHI. The source for SHI is selected with the SHSx bits and includes the following:

- The ADC12SC bit
- The Timer_A Output Unit 1
- The Timer_B Output Unit 0
- The Timer_B Output Unit 1

The polarity of the SHI signal source can be inverted with the ISSH bit. The SAMPCON signal controls the sample period and start of conversion. When SAMPCON is high, sampling is active. The high-to-low SAMPCON transition starts the analog-to-digital conversion, which requires 13 ADC12CLK cycles. Two different sample-timing methods are defined by control bit SHP, extended sample mode and pulse mode.
23.2.4.1 Extended Sample Mode

The extended sample mode is selected when SHP = 0. The SHI signal directly controls SAMPCON and defines the length of the sample period $t_{\text{sample}}$. When SAMPCON is high, sampling is active. The high-to-low SAMPCON transition starts the conversion after synchronization with ADC12CLK (see Figure 23-3).

![Figure 23-3. Extended Sample Mode](image)

23.2.4.2 Pulse Sample Mode

The pulse sample mode is selected when SHP = 1. The SHI signal is used to trigger the sampling timer. The SHT0x and SHT1x bits in ADC12CTL0 control the interval of the sampling timer that defines the SAMPCON sample period $t_{\text{sample}}$. The sampling timer keeps SAMPCON high after synchronization with AD12CLK for a programmed interval $t_{\text{sample}}$. The total sampling time is $t_{\text{sample}}$ plus $t_{\text{sync}}$ (see Figure 23-4).

The SHTx bits select the sampling time in 4x multiples of ADC12CLK. SHT0x selects the sampling time for ADC12MCTL0 to 7 and SHT1x selects the sampling time for ADC12MCTL8 to 15.

![Figure 23-4. Pulse Sample Mode](image)
23.2.4.3 Sample Timing Considerations

When SAMPCON = 0, all Ax inputs are high impedance. When SAMPCON = 1, the selected Ax input can be modeled as an RC low-pass filter during the sampling time $t_{\text{sample}}$, as shown in Figure 23-5. An internal MUX-on input resistance $R_I$ (maximum of 2 kΩ) in series with capacitor $C_I$ (maximum of 40 pF) is seen by the source. The capacitor $C_I$ voltage ($V_C$) must be charged to within 1/2 LSB of the source voltage ($V_S$) for an accurate 12-bit conversion.

![Figure 23-5. Analog Input Equivalent Circuit](image)

The resistance of the source $R_S$ and $R_I$ affect $t_{\text{sample}}$. The following equation can be used to calculate the minimum sampling time $t_{\text{sample}}$ for a 12-bit conversion:

$$t_{\text{sample}} > (R_S + R_I) \times \ln(2^{13}) \times C_I + 800 \text{ ns}$$

Substituting the values for $R_I$ and $C_I$ given above, the equation becomes:

$$t_{\text{sample}} > (R_S + 2 \text{ kΩ}) \times 9.011 \times 40 \text{ pF} + 800 \text{ ns}$$

For example, if $R_S$ is 10 kΩ, $t_{\text{sample}}$ must be greater than 5.13 µs.

23.2.5 Conversion Memory

There are 16 ADC12MEMx conversion memory registers to store conversion results. Each ADC12MEMx is configured with an associated ADC12MCTLx control register. The SREFx bits define the voltage reference and the INCHx bits select the input channel. The EOS bit defines the end of sequence when a sequential conversion mode is used. A sequence rolls over from ADC12MEM15 to ADC12MEM0 when the EOS bit in ADC12MCTL15 is not set.

The CSTARTADDx bits define the first ADC12MCTLx used for any conversion. If the conversion mode is single-channel or repeat-single-channel the CSTARTADDx points to the single ADC12MCTLx to be used.

If the conversion mode selected is either sequence-of-channels or repeat-sequence-of-channels, CSTARTADDx points to the first ADC12MCTLx location to be used in a sequence. A pointer, not visible to software, is incremented automatically to the next ADC12MCTLx in a sequence when each conversion completes. The sequence continues until an EOS bit in ADC12MCTLx is processed; this is the last control byte processed.

When conversion results are written to a selected ADC12MEMx, the corresponding flag in the ADC12IFGx register is set.

23.2.6 ADC12 Conversion Modes

The ADC12 has four operating modes selected by the CONSEQx bits as shown in Table 23-1.

<table>
<thead>
<tr>
<th>CONSEQx</th>
<th>Mode</th>
<th>Operation</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>Single channel single-conversion</td>
<td>A single channel is converted once.</td>
</tr>
<tr>
<td>01</td>
<td>Sequence-of-channels</td>
<td>A sequence of channels is converted once.</td>
</tr>
<tr>
<td>10</td>
<td>Repeat-single-channel</td>
<td>A single channel is converted repeatedly.</td>
</tr>
<tr>
<td>11</td>
<td>Repeat-sequence-of-channels</td>
<td>A sequence of channels is converted repeatedly.</td>
</tr>
</tbody>
</table>
23.2.6.1 Single-Channel Single-Conversion Mode

A single channel is sampled and converted once. The ADC result is written to the ADC12MEMx defined by the CSTARTADDx bits. Figure 23-6 shows the flow of the single-channel, single-conversion mode. When ADC12SC triggers a conversion, successive conversions can be triggered by the ADC12SC bit. When any other trigger source is used, ENC must be toggled between each conversion.

![Figure 23-6. Single-Channel, Single-Conversion Mode](image)

\(x = \text{pointer to ADC12MCTLx}\)

\(^\dagger\) Conversion result is unpredictable
23.2.6.2 Sequence-of-Channels Mode

A sequence of channels is sampled and converted once. The ADC results are written to the conversion memories starting with the ADCMEMx defined by the CSTARTADDx bits. The sequence stops after the measurement of the channel with a set EOS bit. Figure 23-7 shows the sequence-of-channels mode. When ADC12SC triggers a sequence, successive sequences can be triggered by the ADC12SC bit. When any other trigger source is used, ENC must be toggled between each sequence.

Figure 23-7. Sequence-of-Channels Mode
23.2.6.3 Repeat-Single-Channel Mode

A single channel is sampled and converted continuously. The ADC results are written to the ADC12MEMx defined by the CSTARTADDx bits. It is necessary to read the result after the completed conversion, because only one ADC12MEMx memory is used and is overwritten by the next conversion. Figure 23-8 shows repeat-single-channel mode.

![Diagram of Repeat-Single-Channel Mode](image-url)

Figure 23-8. Repeat-Single-Channel Mode
23.2.6.4 Repeat-Sequence-of-Channels Mode

A sequence of channels is sampled and converted repeatedly. The ADC results are written to the conversion memories starting with the ADC12MEMx defined by the CSTARTADDx bits. The sequence ends after the measurement of the channel with a set EOS bit, and the next trigger signal re-starts the sequence. Figure 23-9 shows the repeat-sequence-of-channels mode.

**Figure 23-9. Repeat-Sequence-of-Channels Mode**
23.2.6.5 Using the Multiple Sample and Convert (MSC) Bit

To configure the converter to perform successive conversions automatically and as quickly as possible, a multiple sample and convert function is available. When MSC = 1, CONSEQx > 0, and the sample timer is used, the first rising edge of the SHI signal triggers the first conversion. Successive conversions are triggered automatically as soon as the prior conversion is completed. Additional rising edges on SHI are ignored until the sequence is completed in the single-sequence mode or until the ENC bit is toggled in repeat-single-channel or repeated-sequence modes. The function of the ENC bit is unchanged when using the MSC bit.

23.2.6.6 Stopping Conversions

Stopping ADC12 activity depends on the mode of operation. The recommended ways to stop an active conversion or conversion sequence are:

- Resetting ENC in single-channel single-conversion mode stops a conversion immediately and the results are unpredictable. For correct results, poll the busy bit until it is reset before clearing ENC.
- Resetting ENC during repeat-single-channel operation stops the converter at the end of the current conversion.
- Resetting ENC during a sequence or repeat-sequence mode stops the converter at the end of the sequence. Any conversion mode may be stopped immediately by setting the CONSEQx = 0 and resetting ENC bit. In this case, conversion data are unreliable.

Note

No EOS Bit Set For Sequence

If no EOS bit is set and a sequence mode is selected, resetting the ENC bit does not stop the sequence. To stop the sequence, first select a single-channel mode and then reset ENC.

23.2.7 Using the Integrated Temperature Sensor

To use the on-chip temperature sensor, select the analog input channel INCHx = 1010. Any other configuration is done as if an external channel was selected, including reference selection and conversion-memory selection.

The typical temperature sensor transfer function is shown in Figure 23-10. When using the temperature sensor, the sample period must be greater than 30 µs. The temperature sensor offset error can be large and needs to be calibrated for most applications. See the device-specific data sheet for parameters. See Section 24.2.2.1 for the calibration equations.

Selecting the temperature sensor automatically turns on the on-chip reference generator as a voltage source for the temperature sensor. However, it does not enable the VREF+ output or affect the reference selections for the conversion. The reference choices for converting the temperature sensor are the same as with any other channel.
23.2.8 ADC12 Grounding and Noise Considerations

As with any high-resolution ADC, appropriate printed-circuit-board layout and grounding techniques should be followed to eliminate ground loops, unwanted parasitic effects, and noise.

Ground loops are formed when return current from the A/D flows through paths that are common with other analog or digital circuitry. If care is not taken, this current can generate small unwanted offset voltages that can add to or subtract from the reference or input voltages of the A/D converter. The connections shown in Figure 23-11 help avoid this.

In addition to grounding, ripple and noise spikes on the power supply lines due to digital switching or switching power supplies can corrupt the conversion result. A noise-free design using separate analog and digital ground planes with a single-point connection is recommend to achieve high accuracy.
23.2.9 ADC12 Interrupts

The ADC12 has 18 interrupt sources:

- ADC12IFG0 to ADC12IFG15
- ADC12OV, ADC12MEMx overflow
- ADC12TOV, ADC12 conversion time overflow

The ADC12IFGx bits are set when their corresponding ADC12MEMx memory register is loaded with a conversion result. An interrupt request is generated if the corresponding ADC12IEx bit and the GIE bit are set. The ADC12OV condition occurs when a conversion result is written to any ADC12MEMx before its previous conversion result was read. The ADC12TOV condition is generated when another sample-and-conversion is requested before the current conversion is completed. The DMA is triggered after the conversion in single channel modes or after the completion of a sequence-of-channel modes.

23.2.9.1 ADC12IV, Interrupt Vector Generator

All ADC12 interrupt sources are prioritized and combined to source a single interrupt vector. The interrupt vector register ADC12IV is used to determine which enabled ADC12 interrupt source requested an interrupt.

The highest priority enabled ADC12 interrupt generates a number in the ADC12IV register (see Section 23.3.5). This number can be evaluated or added to the program counter to automatically enter the appropriate software routine. Disabled ADC12 interrupts do not affect the ADC12IV value.

Any access (read or write) of the ADC12IV register automatically resets the ADC12OV condition or the ADC12TOV condition if either was the highest pending interrupt. Neither interrupt condition has an accessible interrupt flag. The ADC12IFGx flags are not reset by an ADC12IV access. ADC12IFGx bits are reset automatically by accessing their associated ADC12MEMx register or may be reset with software.

If another interrupt is pending after servicing of an interrupt, another interrupt is generated. For example, if the ADC12OV and ADC12IFG3 interrupts are pending when the interrupt service routine accesses the ADC12IV register, the ADC12OV interrupt condition is reset automatically. After the RETI instruction of the interrupt service routine is executed, the ADC12IFG3 generates another interrupt.
23.2.9.2 ADC12 Interrupt Handling Software Example

Example 23-1 shows the recommended use of ADC12IV and the handling overhead. The ADC12IV value is added to the PC to automatically jump to the appropriate routine.

The numbers at the right margin show the necessary CPU cycles for each instruction. The software overhead for different interrupt sources includes interrupt latency and return-from-interrupt cycles, but not the task handling itself. The latencies are:

- ADC12IFG0 to ADC12IFG14, ADC12TOV, and ADC12OV: 16 cycles
- ADC12IFG15: 14 cycles

The interrupt handler for ADC12IFG15 shows a way to check immediately if a higher prioritized interrupt occurred during the processing of ADC12IFG15. This saves nine cycles if another ADC12 interrupt is pending.

Example 23-1. Interrupt Handling

```
; Interrupt handler for ADC12.
INT_ADC12              ; Enter Interrupt Service Routine    6
ADD    &ADC12IV,PC  ; Add offset to PC                   3
RETI                ; Vector 0: No interrupt             5
JMP    ADOV         ; Vector 2: ADC overflow             2
JMP    ADTOV        ; Vector 4: ADC timing overflow      2
JMP    ADM0         ; Vector 6: ADC12IFG0                2
    ...              ; Vectors 8-32                       2
JMP    ADM14        ; Vector 34: ADC12IFG14              2

; Handler for ADC12IFG15 starts here. No JMP required.

ADM15   MOV   &ADC12MEM15,xxx  ; Move result, flag is reset
          ...                    ; Other instruction needed?
          JMP   INT_ADC12        ; Check other int pending

; ADC12IFG14-ADC12IFG1 handlers go here

ADM0    MOV   &ADC12MEM0,xxx   ; Move result, flag is reset
          ...                    ; Other instruction needed?
          RETI                   ; Return                     5

ADTOV   ...                    ; Handle Conv. time overflow
          RETI                   ; Return                     5

ADOV    ...                    ; Handle ADCMEMx overflow
          RETI                   ; Return                     5
```
### 23.3 ADC12 Registers

Table 23-2 lists the memory-mapped registers for the ADC12.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>1A0h</td>
<td>ADC12CTL0</td>
<td>ADC12 control 0</td>
<td>00h with POR</td>
<td>Section 23.3.1</td>
</tr>
<tr>
<td>1A2h</td>
<td>ADC12CTL1</td>
<td>ADC12 control 1</td>
<td>00h with POR</td>
<td>Section 23.3.2</td>
</tr>
<tr>
<td>1A4h</td>
<td>ADC12IFG</td>
<td>ADC12 interrupt flag</td>
<td>00h with POR</td>
<td>Section 23.3.3</td>
</tr>
<tr>
<td>1A6h</td>
<td>ADC12IE</td>
<td>ADC12 interrupt enable</td>
<td>00h with POR</td>
<td>Section 23.3.4</td>
</tr>
<tr>
<td>1A8h</td>
<td>ADC12IV</td>
<td>ADC12 interrupt vector word</td>
<td>00h with POR</td>
<td>Section 23.3.5</td>
</tr>
<tr>
<td>80h</td>
<td>ADC12MCTL0</td>
<td>ADC12 memory control 0</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>81h</td>
<td>ADC12MCTL1</td>
<td>ADC12 memory control 1</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>82h</td>
<td>ADC12MCTL2</td>
<td>ADC12 memory control 2</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>83h</td>
<td>ADC12MCTL3</td>
<td>ADC12 memory control 3</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>84h</td>
<td>ADC12MCTL4</td>
<td>ADC12 memory control 4</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>85h</td>
<td>ADC12MCTL5</td>
<td>ADC12 memory control 5</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>86h</td>
<td>ADC12MCTL6</td>
<td>ADC12 memory control 6</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>87h</td>
<td>ADC12MCTL7</td>
<td>ADC12 memory control 7</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>88h</td>
<td>ADC12MCTL8</td>
<td>ADC12 memory control 8</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>89h</td>
<td>ADC12MCTL9</td>
<td>ADC12 memory control 9</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>8Ah</td>
<td>ADC12MCTL10</td>
<td>ADC12 memory control 10</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>8Bh</td>
<td>ADC12MCTL11</td>
<td>ADC12 memory control 11</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>8Ch</td>
<td>ADC12MCTL12</td>
<td>ADC12 memory control 12</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>8Dh</td>
<td>ADC12MCTL13</td>
<td>ADC12 memory control 13</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>8Eh</td>
<td>ADC12MCTL14</td>
<td>ADC12 memory control 14</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>8Fh</td>
<td>ADC12MCTL15</td>
<td>ADC12 memory control 15</td>
<td>00h with POR</td>
<td>Section 23.3.6</td>
</tr>
<tr>
<td>140h</td>
<td>ADC12MEM0</td>
<td>ADC12 memory 0</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>142h</td>
<td>ADC12MEM1</td>
<td>ADC12 memory 1</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>144h</td>
<td>ADC12MEM2</td>
<td>ADC12 memory 2</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>146h</td>
<td>ADC12MEM3</td>
<td>ADC12 memory 3</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>148h</td>
<td>ADC12MEM4</td>
<td>ADC12 memory 4</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>14Ah</td>
<td>ADC12MEM5</td>
<td>ADC12 memory 5</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>14Ch</td>
<td>ADC12MEM6</td>
<td>ADC12 memory 6</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>14Eh</td>
<td>ADC12MEM7</td>
<td>ADC12 memory 7</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>150h</td>
<td>ADC12MEM8</td>
<td>ADC12 memory 8</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>152h</td>
<td>ADC12MEM9</td>
<td>ADC12 memory 9</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>154h</td>
<td>ADC12MEM10</td>
<td>ADC12 memory 10</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>156h</td>
<td>ADC12MEM11</td>
<td>ADC12 memory 11</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>158h</td>
<td>ADC12MEM12</td>
<td>ADC12 memory 12</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>15Ah</td>
<td>ADC12MEM13</td>
<td>ADC12 memory 13</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>15Ch</td>
<td>ADC12MEM14</td>
<td>ADC12 memory 14</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
<tr>
<td>15Eh</td>
<td>ADC12MEM15</td>
<td>ADC12 memory 15</td>
<td>Unchanged</td>
<td>Section 23.3.7</td>
</tr>
</tbody>
</table>
23.3.1 ADC12CTL0 Register

ADC12 Control Register 0 Register

ADC12CTL0 is shown in Figure 23-12 and described in Table 23-3.

Return to Table 23-2.

**Figure 23-12. ADC12CTL0 Register**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-12</td>
<td>SHTx</td>
<td>R/W</td>
<td>0h</td>
<td>Sample-and-hold time. These bits define the number of ADC12CLK cycles in the sampling period for registers ADC12MEM8 to ADC12MEM15. Can be modified only when ENC = 0. 0000b = 4 ADC12CLK cycles 0001b = 8 ADC12CLK cycles 0010b = 16 ADC12CLK cycles 0011b = 32 ADC12CLK cycles 0100b = 64 ADC12CLK cycles 0101b = 96 ADC12CLK cycles 0110b = 128 ADC12CLK cycles 0111b = 192 ADC12CLK cycles 1000b = 256 ADC12CLK cycles 1001b = 384 ADC12CLK cycles 1010b = 512 ADC12CLK cycles 1011b = 768 ADC12CLK cycles 1100b = 1024 ADC12CLK cycles 1101b = 1024 ADC12CLK cycles 1110b = 1024 ADC12CLK cycles 1111b = 1024 ADC12CLK cycles</td>
</tr>
</tbody>
</table>
## Table 23-3. ADC12CTL0 Register Field Descriptions (continued)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
</table>
| 11-8 | SHT0x   | R/W  | 0h    | Sample-and-hold time. These bits define the number of ADC12CLK cycles in the sampling period for registers ADC12MEM0 to ADC12MEM7. Can be modified only when ENC = 0.  
0000b = 4 ADC12CLK cycles  
0001b = 8 ADC12CLK cycles  
0010b = 16 ADC12CLK cycles  
0011b = 32 ADC12CLK cycles  
0100b = 64 ADC12CLK cycles  
0101b = 96 ADC12CLK cycles  
0110b = 128 ADC12CLK cycles  
0111b = 192 ADC12CLK cycles  
1000b = 256 ADC12CLK cycles  
1001b = 384 ADC12CLK cycles  
1010b = 512 ADC12CLK cycles  
1011b = 768 ADC12CLK cycles  
1100b = 1024 ADC12CLK cycles  
1101b = 1024 ADC12CLK cycles  
1110b = 1024 ADC12CLK cycles  
1111b = 1024 ADC12CLK cycles |
| 7    | MSC     | R/W  | 0h    | Multiple sample and conversion. Valid only for sequence or repeated modes. Can be modified only when ENC = 0.  
0b = The sampling timer requires a rising edge of the SHI signal to trigger each sample-and-conversion.  
1b = The first rising edge of the SHI signal triggers the sampling timer, but further sample-and-conversions are performed automatically as soon as the prior conversion is completed. |
| 6    | REF2_5V | R/W  | 0h    | Reference generator voltage. REFON must also be set. Can be modified only when ENC = 0.  
0b = 1.5 V  
1b = 2.5 V |
| 5    | REFON   | R/W  | 0h    | Reference generator on. Can be modified only when ENC = 0.  
0b = Reference off  
1b = Reference on |
| 4    | ADC12ON | R/W  | 0h    | ADC12 on. Can be modified only when ENC = 0.  
0b = ADC12 off  
1b = ADC12 on |
| 3    | ADC12OVIE | R/W | 0h | ADC12MEMx overflow interrupt enable. The GIE bit must also be set to enable the interrupt.  
0b = Overflow interrupt disabled  
1b = Overflow interrupt enabled |
| 2    | ADC12TOVIE | R/W | 0h | ADC12 conversion-time-overflow interrupt enable. The GIE bit must also be set to enable the interrupt.  
0b = Conversion time overflow interrupt disabled  
1b = Conversion time overflow interrupt enabled |
| 1    | ENC     | R/W  | 0h    | Enable conversion  
0b = ADC12 disabled  
1b = ADC12 enabled |
| 0    | ADC12SC | R/W  | 0h    | Start conversion. Software-controlled sample-and-conversion start. ADC12SC and ENC may be set together with one instruction. ADC12SC is reset automatically.  
0b = No sample-and-conversion start  
1b = Start sample-and-conversion |
23.3.2 ADC12CTL1 Register

ADC12 Control Register 1 Register

ADC12CTL1 is shown in Figure 23-13 and described in Table 23-4.

Return to Table 23-2.

**Figure 23-13. ADC12CTL1 Register**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-12</td>
<td>CSTARTADDx</td>
<td>R/W</td>
<td>0h</td>
<td>Conversion start address. These bits select which ADC12 conversion-memory register is used for a single conversion or for the first conversion in a sequence. The value of CSTARTADDx is 0 to 0Fh, corresponding to ADC12MEM0 to ADC12MEM15. Can be modified only when ENC = 0.</td>
</tr>
<tr>
<td>11-10</td>
<td>SHSx</td>
<td>R/W</td>
<td>0h</td>
<td>Sample-and-hold source select. Can be modified only when ENC = 0. 00b = ADC12SC bit 01b = Timer_A.OUT1 10b = Timer_B.OUT0 11b = Timer_B.OUT1</td>
</tr>
<tr>
<td>9</td>
<td>SHP</td>
<td>R/W</td>
<td>0h</td>
<td>Sample-and-hold pulse-mode select. This bit selects the source of the sampling signal (SAMPCON) to be either the output of the sampling timer or the sample-input signal directly. Can be modified only when ENC = 0. 0b = SAMPCON signal is sourced from the sample-input signal. 1b = SAMPCON signal is sourced from the sampling timer.</td>
</tr>
<tr>
<td>8</td>
<td>ISSH</td>
<td>R/W</td>
<td>0h</td>
<td>Invert signal sample-and-hold. Can be modified only when ENC = 0. 0b = The sample input signal is not inverted. 1b = The sample input signal is inverted.</td>
</tr>
<tr>
<td>7-5</td>
<td>ADC12DIVx</td>
<td>R/W</td>
<td>0h</td>
<td>ADC12 clock divider. Can be modified only when ENC = 0. 000b = /1 001b = /2 010b = /3 011b = /4 100b = /5 101b = /6 110b = /7 111b = /8</td>
</tr>
<tr>
<td>4-3</td>
<td>ADC12SSELx</td>
<td>R/W</td>
<td>0h</td>
<td>ADC12 clock source select. Can be modified only when ENC = 0. 00b = ADC12OSC 01b = ACLK 10b = MCLK 11b = SMCLK</td>
</tr>
</tbody>
</table>
Table 23-4. ADC12CTL1 Register Field Descriptions (continued)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>2-1</td>
<td>CONSEQx</td>
<td>R/W</td>
<td>0h</td>
<td>Conversion sequence mode select</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = Single-channel, single-conversion mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = Sequence-of-channels mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = Repeat-single-channel mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = Repeat-sequence-of-channels mode</td>
</tr>
<tr>
<td>0</td>
<td>ADC12BUSY</td>
<td>R/W</td>
<td>0h</td>
<td>ADC12 busy. This bit indicates an active sample or conversion operation.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = No operation is active</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = A sequence, sample, or conversion is active</td>
</tr>
</tbody>
</table>
23.3.3 ADC12IFG Register

ADC12 Interrupt Flag Register

ADC12IFG is shown in Figure 23-14 and described in Table 23-5.

Return to Table 23-2.

![Figure 23-14. ADC12IFG Register](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>ADC12IFGx</td>
<td>R/W</td>
<td>0h</td>
<td>ADC12MEMx interrupt flag. These bits are set when corresponding ADC12MEMx is loaded with a conversion result. The ADC12IFGx bits are reset if the corresponding ADC12MEMx is accessed, or may be reset with software. 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
</tbody>
</table>

23.3.4 ADC12IE Register

ADC12 Interrupt Enable Register

ADC12IE is shown in Figure 23-15 and described in Table 23-6.

Return to Table 23-2.

![Figure 23-15. ADC12IE Register](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>ADC12IEx</td>
<td>R/W</td>
<td>0h</td>
<td>Interrupt enable. These bits enable or disable the interrupt request for the ADC12IFGx bits. 0b = Interrupt disabled 1b = Interrupt enabled</td>
</tr>
</tbody>
</table>
23.3.5 ADC12IV Register

ADC12 Interrupt Vector Register

ADC12IV is shown in Figure 23-16 and described in Table 23-7.

Return to Table 23-2.

Figure 23-16. ADC12IV Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>ADC12IVx</td>
<td>R</td>
<td>0h</td>
<td>ADC12 interrupt vector value. See Table 23-8.</td>
</tr>
</tbody>
</table>

Table 23-7. ADC12IV Register Field Descriptions

Table 23-8. ADC12 Interrupt Vector Values

<table>
<thead>
<tr>
<th>ADC12IV Contents</th>
<th>Interrupt Source</th>
<th>Interrupt Flag</th>
<th>Interrupt Priority</th>
</tr>
</thead>
<tbody>
<tr>
<td>000h</td>
<td>No interrupt pending</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>002h</td>
<td>ADC12MEMx overflow</td>
<td>ADC12IFG0</td>
<td>Highest</td>
</tr>
<tr>
<td>004h</td>
<td>Conversion time overflow</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>006h</td>
<td>ADC12MEM0 interrupt flag</td>
<td>ADC12IFG1</td>
<td>Highest</td>
</tr>
<tr>
<td>008h</td>
<td>ADC12MEM1 interrupt flag</td>
<td>ADC12IFG2</td>
<td>Highest</td>
</tr>
<tr>
<td>00Ah</td>
<td>ADC12MEM2 interrupt flag</td>
<td>ADC12IFG3</td>
<td>Highest</td>
</tr>
<tr>
<td>00Ch</td>
<td>ADC12MEM3 interrupt flag</td>
<td>ADC12IFG4</td>
<td>Highest</td>
</tr>
<tr>
<td>00Eh</td>
<td>ADC12MEM4 interrupt flag</td>
<td>ADC12IFG5</td>
<td>Highest</td>
</tr>
<tr>
<td>010h</td>
<td>ADC12MEM5 interrupt flag</td>
<td>ADC12IFG6</td>
<td>Highest</td>
</tr>
<tr>
<td>012h</td>
<td>ADC12MEM6 interrupt flag</td>
<td>ADC12IFG7</td>
<td>Highest</td>
</tr>
<tr>
<td>014h</td>
<td>ADC12MEM7 interrupt flag</td>
<td>ADC12IFG8</td>
<td>Highest</td>
</tr>
<tr>
<td>016h</td>
<td>ADC12MEM8 interrupt flag</td>
<td>ADC12IFG9</td>
<td>Highest</td>
</tr>
<tr>
<td>01Ah</td>
<td>ADC12MEM10 interrupt flag</td>
<td>ADC12IFG10</td>
<td>Highest</td>
</tr>
<tr>
<td>01Ch</td>
<td>ADC12MEM11 interrupt flag</td>
<td>ADC12IFG11</td>
<td>Highest</td>
</tr>
<tr>
<td>01Eh</td>
<td>ADC12MEM12 interrupt flag</td>
<td>ADC12IFG12</td>
<td>Highest</td>
</tr>
<tr>
<td>020h</td>
<td>ADC12MEM13 interrupt flag</td>
<td>ADC12IFG13</td>
<td>Highest</td>
</tr>
<tr>
<td>022h</td>
<td>ADC12MEM14 interrupt flag</td>
<td>ADC12IFG14</td>
<td>Highest</td>
</tr>
<tr>
<td>024h</td>
<td>ADC12MEM15 interrupt flag</td>
<td>ADC12IFG15</td>
<td>Lowest</td>
</tr>
</tbody>
</table>
23.3.6 ADC12MCTLx Register

ADC12 Memory Control x Register

ADC12MCTLx is shown in Figure 23-17 and described in Table 23-9.

Return to Table 23-2.

Figure 23-17. ADC12MCTLx Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>EOS</td>
<td>R/W</td>
<td>0h</td>
<td>End of sequence. Indicates the last conversion in a sequence. Can be modified only when ENC = 0. 0b = Not end of sequence 1b = End of sequence</td>
</tr>
<tr>
<td>6-4</td>
<td>SREFx</td>
<td>R/W</td>
<td>0h</td>
<td>Select reference. Can be modified only when ENC = 0. 000b = V_R+ = AV_CC and V_R- = AV_SS 001b = V_R+ = V_REF+ and V_R- = AV_SS 010b = V_R+ = V_eREF+ and V_R- = AV_SS 011b = V_R+ = V_eREF+ and V_R- = V_REF+/V_eREF. 100b = V_R+ = AV_CC and V_R- = V_REF-/V_eREF. 101b = V_R+ = V_REF+ and V_R- = V_REF+/V_eREF. 110b = V_R+ = V_eREF+ and V_R- = V_REF+/V_eREF. 111b = V_R+ = V_eREF+ and V_R- = V_REF-/V_eREF.</td>
</tr>
<tr>
<td>3-0</td>
<td>INCHx</td>
<td>R/W</td>
<td>0h</td>
<td>Input channel select. Can be modified only when ENC = 0. 0000b = A0 0001b = A1 0010b = A2 0011b = A3 0100b = A4 0101b = A5 0110b = A6 0111b = A7 1000b = V_eREF+ 1001b = V_REF-/V_eREF. 1010b = Temperature diode 1011b = (AV_CC – AV_SS) / 2 1100b = GND 1101b = GND 1110b = GND 1111b = GND</td>
</tr>
</tbody>
</table>

Table 23-9. ADC12MCTLx Register Field Descriptions
23.3.7 ADC12MEMx Register

ADC12 Memory x Register

ADC12MEMx is shown in Figure 23-18 and described in Table 23-10.

Return to Table 23-2.

![Figure 23-18. ADC12MEMx Register](image)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>Conversion_Results</td>
<td>R/W</td>
<td>0h</td>
<td>The 12-bit conversion results are right-justified. Bit 11 is the MSB. Bits 15-12 are always 0. Writing to the conversion memory registers corrupts the results.</td>
</tr>
</tbody>
</table>
The Tag-Length-Value (TLV) structure is used in selected MSP430x2xx devices to provide device-specific information in the device's flash memory SegmentA, such as calibration data. For the device-dependent implementation, see the device-specific data sheet.

24.1 TLV Introduction...................................................................................................................................................... 632
24.2 Supported Tags....................................................................................................................................................... 633
24.3 Checking Integrity of SegmentA........................................................................................................................... 636
24.4 Parsing TLV Structure of Segment A..................................................................................................................... 636
24.1 TLV Introduction

The TLV structure stores device-specific data in SegmentA. The SegmentA content of an example device is shown in Table 24-1.

<table>
<thead>
<tr>
<th>Word Address</th>
<th>Upper Byte</th>
<th>Lower Byte</th>
<th>Tag Address and Offset</th>
</tr>
</thead>
<tbody>
<tr>
<td>0x10FE</td>
<td>CALBC1_1MHZ</td>
<td>CALDCO_1MHZ</td>
<td>0x10F6 + 0x0008</td>
</tr>
<tr>
<td>0x10FC</td>
<td>CALBC1_8MHZ</td>
<td>CALDCO_8MHZ</td>
<td>0x10F6 + 0x0006</td>
</tr>
<tr>
<td>0x10FA</td>
<td>CALBC1_12MHZ</td>
<td>CALDCO_12MHZ</td>
<td>0x10F6 + 0x0004</td>
</tr>
<tr>
<td>0x10F8</td>
<td>CALBC1_16MHZ</td>
<td>CALDCO_16MHZ</td>
<td>0x10F6 + 0x0002</td>
</tr>
<tr>
<td>0x10F6</td>
<td>0x08 (LENGTH)</td>
<td>TAG_DCO_30</td>
<td>0x10F6</td>
</tr>
<tr>
<td>0x10F4</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10F2</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10F0</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10EE</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10EC</td>
<td>0x08 (LENGTH)</td>
<td>TAG_EMPTY</td>
<td>0x10EC</td>
</tr>
<tr>
<td>0x10EA</td>
<td>CAL_ADC_25T85</td>
<td></td>
<td>0x10DA + 0x0010</td>
</tr>
<tr>
<td>0x10E8</td>
<td>CAL_ADC_25T30</td>
<td></td>
<td>0x10DA + 0x000E</td>
</tr>
<tr>
<td>0x10E6</td>
<td>CAL_ADC_25VREF_FACTOR</td>
<td></td>
<td>0x10DA + 0x000C</td>
</tr>
<tr>
<td>0x10E4</td>
<td>CAL_ADC_15T85</td>
<td></td>
<td>0x10DA + 0x000A</td>
</tr>
<tr>
<td>0x10E2</td>
<td>CAL_ADC_15T30</td>
<td></td>
<td>0x10DA + 0x0008</td>
</tr>
<tr>
<td>0x10E0</td>
<td>CAL_ADC_15VREF_FACTOR</td>
<td></td>
<td>0x10DA + 0x0006</td>
</tr>
<tr>
<td>0x10DE</td>
<td>CAL_ADC_OFFSET</td>
<td></td>
<td>0x10DA + 0x0004</td>
</tr>
<tr>
<td>0x10DC</td>
<td>CAL_ADC_GAIN_FACTOR</td>
<td></td>
<td>0x10DA + 0x0002</td>
</tr>
<tr>
<td>0x10DA</td>
<td>0x10 (LENGTH)</td>
<td>TAG_ADC12_1</td>
<td>0x10DA</td>
</tr>
<tr>
<td>0x10D8</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10D6</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10D4</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10D2</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10D0</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10CE</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10CC</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10CA</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10C8</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10C6</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10C4</td>
<td>0xFF</td>
<td>0xFF</td>
<td></td>
</tr>
<tr>
<td>0x10C2</td>
<td>0x16 (LENGTH)</td>
<td>TAG_EMPTY</td>
<td>0x10C2</td>
</tr>
<tr>
<td>0x10C0</td>
<td>2s complement of bit-wise XOR</td>
<td></td>
<td>0x10C0</td>
</tr>
</tbody>
</table>

The first two bytes of SegmentA (0x10C0 and 0x10C1) hold the checksum of the remainder of the segment (addresses 0x10C2 to 0x10FF).

The first tag is located at address 0x10C2 and, in this example, is the TAG_EMPTY tag. The following byte (0x10C3) holds the length of the following structure. The length of this TAG_EMPTY structure is 0x16 and, therefore, the next tag, TAG_ADC12_1, is found at address 0x10DA. Again, the following byte holds the length of the TAG_ADC12_1 structure.

The TLV structure maps the entire address range 0x10C2 to 0x10FF of the SegmentA. A program routine looking for tags starting at the SegmentA address 0x10C2 can extract all information even if it is stored at a different (device-specific) absolute address.
24.2 Supported Tags

Each device contains a subset of the tags shown in Table 24-2. See the device-specific data sheet for details.

<table>
<thead>
<tr>
<th>Tag</th>
<th>Description</th>
<th>Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>TAG_EMPTY</td>
<td>Identifies an unused memory area</td>
<td>0xFE</td>
</tr>
<tr>
<td>TAG_DCO_30</td>
<td>Calibration values for the DCO at room temperature and $V_{CC} = 3,\text{V}$</td>
<td>0x01</td>
</tr>
<tr>
<td>TAG_ADC12_1</td>
<td>Calibration values for the ADC12 module</td>
<td>0x08</td>
</tr>
<tr>
<td>TAG_ADC10_1</td>
<td>Calibration values for the ADC10 module</td>
<td>available on select devices and value is device-specific</td>
</tr>
</tbody>
</table>

24.2.1 DCO Calibration TLV Structure

For DCO calibration, the BCS+ registers (BCSCTL1 and DCOCTL) are used. The values stored in the flash information memory SegmentA are written to the BCS+ registers (see Table 24-3).

<table>
<thead>
<tr>
<th>Label</th>
<th>Description</th>
<th>Offset</th>
</tr>
</thead>
<tbody>
<tr>
<td>CALBC1_1MHZ</td>
<td>Value for the BCSCTL1 register for 1 MHz, $T_A = 25^\circ\text{C}$</td>
<td>0x07</td>
</tr>
<tr>
<td>CALDCO_1MHZ</td>
<td>Value for the DCOCTL register for 1 MHz, $T_A = 25^\circ\text{C}$</td>
<td>0x06</td>
</tr>
<tr>
<td>CALBC1_8MHZ</td>
<td>Value for the BCSCTL1 register for 8 MHz, $T_A = 25^\circ\text{C}$</td>
<td>0x05</td>
</tr>
<tr>
<td>CALDCO_8MHZ</td>
<td>Value for the DCOCTL register for 8 MHz, $T_A = 25^\circ\text{C}$</td>
<td>0x04</td>
</tr>
<tr>
<td>CALBC1_12MHZ</td>
<td>Value for the BCSCTL1 register for 12 MHz, $T_A = 25^\circ\text{C}$</td>
<td>0x03</td>
</tr>
<tr>
<td>CALDCO_12MHZ</td>
<td>Value for the DCOCTL register for 12 MHz, $T_A = 25^\circ\text{C}$</td>
<td>0x02</td>
</tr>
<tr>
<td>CALBC1_16MHZ</td>
<td>Value for the BCSCTL1 register for 16 MHz, $T_A = 25^\circ\text{C}$</td>
<td>0x01</td>
</tr>
<tr>
<td>CALDCO_16MHZ</td>
<td>Value for the DCOCTL register for 16 MHz, $T_A = 25^\circ\text{C}$</td>
<td>0x00</td>
</tr>
</tbody>
</table>

The calibration data for the DCO is available in all 2xx devices and is stored at the same absolute addresses. The device-specific SegmentA content is applied using the absolute addressing mode if the sample code shown in Example 24-1 is used.

**Example 24-1. Code Example Using Absolute Addressing Mode**

```assembly
; Calibrate the DCO to 1 MHz
CLR.B &DCOCTL ; Select lowest DCOx and MODx settings
MOV.B &CALBC1_1MHZ,&BCSCTL1 ; Set RSELx
MOV.B &CALDCO_1MHZ,&DCOCTL ; Set DCOx and MODx
```

**Example 24-2. Code Example Using the TLV Structure**

```assembly
; Calibrate the DCO to 8 MHz
; It is assumed that R10 contains the address of the TAG_DCO_30 tag
CLR.B &DCOCTL ; Select lowest DCOx and MODx settings
MOV.B 7(R10),&BCSCTL1 ; Set RSEL
MOV.B 6(R10),&DCOCTL ; Set DCOx and MODx
```
24.2.2 TAG_ADC12_1 Calibration TLV Structure

The calibration data for the ADC12 module consists of eight words (see Table 24-4).

Table 24-4. TAG_ADC12_1 Calibration Data (Device Specific)

<table>
<thead>
<tr>
<th>Label</th>
<th>Description</th>
<th>Offset</th>
</tr>
</thead>
<tbody>
<tr>
<td>CAL_ADC_25T85</td>
<td>VREF2_5 = 1, T_A = 85°C ± 2°C, 12-bit conversion result</td>
<td>0x0E</td>
</tr>
<tr>
<td>CAL_ADC_25T30</td>
<td>VREF2_5 = 1, T_A = 30°C ± 2°C, 12-bit conversion result</td>
<td>0x0C</td>
</tr>
<tr>
<td>CAL_ADC_25VREF_FACTOR</td>
<td>VREF2_5 = 1, T_A = 30°C ± 2°C</td>
<td>0x0A</td>
</tr>
<tr>
<td>CAL_ADC_15T85</td>
<td>VREF2_5 = 0, T_A = 85°C ± 2°C, 12-bit conversion result</td>
<td>0x08</td>
</tr>
<tr>
<td>CAL_ADC_15T30</td>
<td>VREF2_5 = 0, T_A = 30°C ± 2°C, 12-bit conversion result</td>
<td>0x06</td>
</tr>
<tr>
<td>CAL_ADC_15VREF_FACTOR</td>
<td>VREF2_5 = 0, T_A = 30°C ± 2°C</td>
<td>0x04</td>
</tr>
<tr>
<td>CAL_ADC_OFFSET</td>
<td>VeREF = 2.5 V, T_A = 85°C ± 2°C, f_ADC12CLK = 5 MHz</td>
<td>0x02</td>
</tr>
<tr>
<td>CAL_ADC_GAIN_FACTOR</td>
<td>VeREF = 2.5 V, T_A = 85°C ± 2°C, f_ADC12CLK = 5 MHz</td>
<td>0x00</td>
</tr>
</tbody>
</table>

24.2.2.1 Temperature Sensor Calibration Data

The temperature sensor is calibrated using the internal voltage references. Each reference voltage (1.5 V and 2.5 V) contains a measured value for two temperatures, 30°C±2°C and 85°C±2°C and are stored in the TLV structure at the respective SegmentA location (see Table 24-4). The characteristic equation of the temperature sensor voltage, in mV, is:

\[ V_{\text{SENSE}} = TC_{\text{SENSOR}} \times \text{Temp} + V_{\text{SENSOR}} \]  

(1)

The temperature coefficient, TC_{\text{SENSOR}} in mV/°C, represents the slope of the equation. V_{\text{SENSOR}}, in mV, represents the y-intercept of the equation. Temp, in °C, is the temperature of interest.

The temperature (Temp, °C) can be computed as follows for each of the reference voltages used in the ADC measurement:

\[ \text{Temp} = \left( \text{ADC(raw)} - \text{CAL_ADC_15T30} \right) \times \left( \frac{85 - 30}{\text{CAL_ADC_15T85} - \text{CAL_ADC_15T30}} \right) + 30 \]

\[ \text{Temp} = \left( \text{ADC(raw)} - \text{CAL_ADC_25T30} \right) \times \left( \frac{85 - 30}{\text{CAL_ADC_25T85} - \text{CAL_ADC_25T30}} \right) + 30 \]  

(2)

24.2.2.2 Integrated Voltage Reference Calibration Data

The reference voltages (VREF2_5 = 0 and 1) are measured at room temperature. The measured value is normalized by 1.5 V or 2.5 V before stored into the flash information memory SegmentA.

\[ \text{CAL_ADC_15VREF_FACTOR} = \left( \frac{\text{VeREF}}{1.5 \text{ V}} \right) \times 2^{15} \]

The conversion result is corrected by multiplying it with the CAL_ADC_15VREF_FACTOR (or CAL_ADC_25VREF_FACTOR) and dividing the result by \(2^{15}\).

\[ \text{ADC(corrected)} = \text{ADC(raw)} \times \text{CAL_ADC_15VREF_FACTOR} \times \left( \frac{1}{2^{15}} \right) \]

24.2.2.3 Example Using the Reference Calibration

In the following example, the integrated 1.5-V reference voltage is used during a conversion.

- Conversion result: 0x0100
- Reference voltage calibration factor (CAL_ADC_15VREF_FACTOR): 0x7BBB

The following steps show an example of how the ADC12 conversion result can be corrected by using the hardware multiplier:

1. Multiply the conversion result by 2 (this step simplifies the final division).
2. Multiply the result by CAL_ADC_15VREF_FACTOR.
3. Divide the result by $2^{16}$ (use the upper word of the 32-bit multiplication result RESHI).

In the example:
1. $0x0100 \times 0x0002 = 0x0200$
2. $0x0200 \times 0x7BBB = 0x00F7_7600$
3. $0x00F7_7600 + 0x0001_0000 = 0x0000_00F7 (= 247)$

The code example using the hardware multiplier follows.

```assembly
; The ADC conversion result is stored in ADC12MEM0
; It is assumed that R9 contains the address of the
; TAG_ADC12_1.
; The corrected value is available in ADC_COR
MOV.W &ADC12MEM0, R10   ; move result to R10
RLA.W R10               ; R10 x 2
MOV.W R10, &MPY         ; unsigned multiply OP1
MOV.W CAL_ADC_15VREF_FACTOR(R9), &OP2    ; calibration value OP2
MOV.W &RESHI, &ADC_COR  ; result: upper 16-bit MPY
```

### 24.2.2.4 Offset and Gain Calibration Data

The offset of the ADC12 is determined and stored as a two's-complement number in SegmentA. The offset error correction is done by adding the CAL_ADC_OFFSET to the conversion result.

$$\text{ADC}(\text{offset_corrected}) = \text{ADC}(\text{raw}) + \text{CAL}_\text{ADC}_\text{OFFSET}$$

The gain of the ADC12, stored at offset 0x00, is calculated by the following equation.

$$\text{CAL}_\text{ADC}_\text{_GAIN}_\text{FACTOR} = \left( \frac{1}{\text{GAIN}} \right) \times 2^{15}$$

The conversion result is gain corrected by multiplying it with the CAL_ADC_GAIN_FACTOR and dividing the result by $2^{15}$.

$$\text{ADC}(\text{gain_corrected}) = \text{ADC}(\text{raw}) \times \text{CAL}_\text{ADC}_\text{_GAIN}_\text{FACTOR} \times \left( \frac{1}{2^{15}} \right)$$

If both gain and offset are corrected, the gain correction is done first.

$$\text{ADC}(\text{gain_corrected}) = \text{ADC}(\text{raw}) \times \text{CAL}_\text{ADC}_\text{_GAIN}_\text{FACTOR} \times \left( \frac{1}{2^{15}} \right)$$

$$\text{ADC}(\text{final}) = \text{ADC}(\text{gain_corrected}) + \text{CAL}_\text{ADC}_\text{OFFSET}$$

### 24.2.2.5 Example Using Gain and Offset Calibration

In the following example, an external reference voltage is used during a conversion.

- Conversion result: 0x0800 (= 2048)
- Gain calibration factor: 0x7FE0 (gain error: +2 LSB)
- Offset calibration: 0xFFFE (2s complement of -2)

The following steps show an example of how the ADC12 conversion result is corrected by using the hardware multiplier:

1. Multiply the conversion result by 2 (this step simplifies the final division).
2. Multiply the result by CAL_ADC_GAIN_FACTOR.
3. Divide the result by $2^{16}$ (use the upper word of the 32-bit multiplication result RESHI)
4. Add CAL_ADC_OFFSET to the result.

In the example:

1. $0x0800 \times 0x0002 = 0x1000$
2. $0x1000 \times 0x8010 = 0x0801_0000$
3. $0x0801_0000 + 0x0001_0000 = 0x0000_0801 (= 2049)$
4. $0x801 + 0xFFFE = 0x07FF (= 2047)$
The code example using the hardware multipliers follows.

; The ADC conversion result is stored in ADC12MEM0
; It is assumed that R9 contains the address of the TAG_ADC12_1.
; The corrected value is available in ADC_COR
MOV.W &ADC12MEM0,R10 ; move result to R10
RLA.W R10 ; R10 * 2
MOV.W R10,&MPY ; unsigned multiply OP1
MOV.W CAL_ADC_GAIN_FACTOR(R9),&OP2 ; calibration value OP2
MOV.W &RESHI,&ADC_COR ; use upper 16-bit MPY
ADD.W CAL_ADC_OFFSET(R9),&ADC_COR ; add offset correction

24.3 Checking Integrity of SegmentA

The 64-byte SegmentA contains a 2-byte checksum of the data stored at 0x10C2 up to 0x10FF at addresses 0x10C0 and 0x10C1. The checksum is a bit-wise XOR of 31 words stored in the two's-complement data format.

A code example to calculate the checksum follows.

; Checking the SegmentA integrity by calculating the 2's complement of the 31 words at 0x10C2 - 0x10FE.
; It is assumed that the SegmentA Start Address is stored in R10. R11 is initialized to 0x00.
; The label TLV_CHKSUM is set to 0x10C0.
ADD.W #2,R10 ; Skip the checksum
LP0 XOR.W @R10+,R11 ; Add a word to checksum
CMP.W #0x10FF,R10 ; Last word included?
JN LP0 ; No, add more data
ADD.W &TLV_CHKSUM,R11 ; Add checksum
JNZ CSNOK ; Checksum not ok
CSNOK ... ; Do not use SegmentA data

24.4 Parsing TLV Structure of Segment A

Example code to analyze SegmentA follows.

; It is assumed that the SegmentA start address is stored in R10.
LP1 ADD.W #2,R10 ; Skip two bytes
CMP.W #0x10FF,R10 ; SegmentA end reached?
JGE DONE ; Yes, done
CMP.B #TAG_EMPTY,0(R10) ; TAG_EMPTY?
JNZ T1 ; No, continue
JMP LP2 ; Yes, done with TAG_EMPTY
T1 CMP.B #TAG_ADC12_1,0(R10) ; TAG_ADC12_1?
JNZ T2 ; No, continue
... ; Yes, found TAG_ADC12_1
JMP LP2 ; Done with TAG_ADC12_1
T2 CMP.B #DCO_30,0(R10) ; TAG_DCO_30?
JNZ T3 ; No, continue
CLR.B &DCOCTL ; Select lowest DCOx
MOV.B 7(R10),&BCSCTL1 ; Yes, use e.g. 8MHz data and
MOV.B 6(R10),&DCOCTL ; set DCOx and MODx
JMP LP2 ; Done with TAG_DCO_30
T3 ... ; Test for "next tag"
... ;
JMP LP2 ; Done with "next tag"
LP2 MOV.B 1(R10),R11 ; Store LENGTH in R11
ADD.W R11,R10 ; Add LENGTH to R10
JMP LP1 ; Jump to continue analysis
DONE ;
The DAC12 module is a 12-bit voltage-output digital-to-analog converter (DAC). This chapter describes the operation of the DAC12 module of the MSP430x2xx device family.

<table>
<thead>
<tr>
<th>Section</th>
<th>Page</th>
</tr>
</thead>
<tbody>
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<tr>
<td>25.2 DAC12 Operation</td>
<td>640</td>
</tr>
<tr>
<td>25.3 DAC12 Registers</td>
<td>644</td>
</tr>
</tbody>
</table>
25.1 DAC12 Introduction

The DAC12 module is a 12-bit voltage-output DAC. The DAC12 can be configured in 8-bit or 12-bit mode and may be used in conjunction with the DMA controller. When multiple DAC12 modules are present, they may be grouped together for synchronous update operation.

Features of the DAC12 include:

- 12-bit monotonic output
- 8-bit or 12-bit voltage output resolution
- Programmable settling time vs power consumption
- Internal or external reference selection
- Straight binary or 2s compliment data format
- Self-calibration option for offset correction
- Synchronized update capability for multiple DAC12 modules

Note

Multiple DAC12 Modules

Some devices may integrate more than one DAC12 module. If more than one DAC12 is present on a device, the multiple DAC12 modules operate identically.

Throughout this chapter, nomenclature appears such as DAC12_xDAT or DAC12_xCTL to describe register names. When this occurs, the x is used to indicate which DAC12 module is being discussed. In cases where operation is identical, the register is simply referred to as DAC12_xCTL.

The block diagram of the DAC12 module is shown in Figure 25-1.
Figure 25-1. DAC12 Block Diagram
25.2 DAC12 Operation

The DAC12 module is configured with user software. The setup and operation of the DAC12 is discussed in the following sections.

25.2.1 DAC12 Core

The DAC12 can be configured to operate in 8-bit or 12-bit mode using the DAC12RES bit. The full-scale output is programmable to be 1x or 3x the selected reference voltage via the DAC12IR bit. This feature allows the user to control the dynamic range of the DAC12. The DAC12DF bit allows the user to select between straight binary data and 2s-compliment data for the DAC. When using straight binary data format, the formula for the output voltage is given in Table 25-1.

```
<table>
<thead>
<tr>
<th>Resolution</th>
<th>DAC12RES</th>
<th>DAC12IR</th>
<th>Output Voltage Formula</th>
</tr>
</thead>
<tbody>
<tr>
<td>12 bit</td>
<td>0</td>
<td>0</td>
<td>( V_{OUT} = V_{REF} \times 3 \times \frac{\text{DAC12}_x\text{DAT}}{4096} )</td>
</tr>
<tr>
<td>12 bit</td>
<td>0</td>
<td>1</td>
<td>( V_{OUT} = V_{REF} \times \frac{\text{DAC12}_x\text{DAT}}{4096} )</td>
</tr>
<tr>
<td>8 bit</td>
<td>1</td>
<td>0</td>
<td>( V_{OUT} = V_{REF} \times 3 \times \frac{\text{DAC12}_x\text{DAT}}{256} )</td>
</tr>
<tr>
<td>8 bit</td>
<td>1</td>
<td>1</td>
<td>( V_{OUT} = V_{REF} \times \frac{\text{DAC12}_x\text{DAT}}{256} )</td>
</tr>
</tbody>
</table>
```

In 8-bit mode, the maximum useable value for DAC12_xDAT is 0FFh. In 12-bit mode, the maximum useable value for DAC12_xDAT is 0FFFh. Values greater than these may be written to the register, but all leading bits are ignored.

25.2.1.1 DAC12 Port Selection

The DAC12 outputs are multiplexed with the port P6 pins and ADC12 analog inputs, and also the V_{eREF+} pins. When DAC12AMPx > 0, the DAC12 function is automatically selected for the pin, regardless of the state of the associated PxSELx and PxDIRx bits. The DAC12OPS bit selects between the P6 pins and the V_{eREF+} pins for the DAC outputs. For example, when DAC12OPS = 0, DAC12_0 outputs on P6.6 and DAC12_1 outputs on P6.7. When DAC12OPS = 1, DAC12_0 outputs on V_{eREF+} and DAC12_1 outputs on P6.5. See the port pin schematic in the device-specific data sheet for more details.

25.2.2 DAC12 Reference

The reference for the DAC12 is configured to use either an external reference voltage or the internal 1.5-V/2.5-V reference from the ADC12 module with the DAC12SREFx bits. When DAC12SREFx = {0,1} the V_{REF+} signal is used as the reference and when DAC12SREFx = {2,3} the V_{eREF+} signal is used as the reference.

To use the ADC12 internal reference, it must be enabled and configured via the applicable ADC12 control bits.

25.2.2.1 DAC12 Reference Input and Voltage Output Buffers

The reference input and voltage output buffers of the DAC12 can be configured for optimized settling time vs power consumption. Eight combinations are selected using the DAC12AMPx bits. In the low/low setting, the settling time is the slowest, and the current consumption of both buffers is the lowest. The medium and high settings have faster settling times, but the current consumption increases. See the device-specific data sheet for parameters.

25.2.3 Updating the DAC12 Voltage Output

The DAC12_xDAT register can be connected directly to the DAC12 core or double buffered. The trigger for updating the DAC12 voltage output is selected with the DAC12LSELx bits.
When DAC12LSELx = 0 the data latch is transparent and the DAC12_xDAT register is applied directly to the DAC12 core. The DAC12 output updates immediately when new DAC12 data is written to the DAC12_xDAT register, regardless of the state of the DAC12ENC bit.

When DAC12LSELx = 1, DAC12 data is latched and applied to the DAC12 core after new data is written to DAC12_xDAT. When DAC12LSELx = 2 or 3, data is latched on the rising edge from the Timer_A CCR1 output or Timer_B CCR2 output respectively. DAC12ENC must be set to latch the new data when DAC12LSELx > 0.

### 25.2.4 DAC12_xDAT Data Format

The DAC12 supports both straight binary and 2s compliment data formats. When using straight binary data format, the full-scale output value is 0FFFh in 12-bit mode (0FFh in 8-bit mode) as shown in Figure 25-2.

![](Figure 25-2. Output Voltage vs DAC12 Data, 12-Bit, Straight Binary Mode)

When using 2s-compliment data format, the range is shifted such that a DAC12_xDAT value of 0800h (0080h in 8-bit mode) results in a zero output voltage, 0000h is the mid-scale output voltage, and 07FFh (007Fh for 8-bit mode) is the full-scale voltage output (see Figure 25-3).

![](Figure 25-3. Output Voltage vs DAC12 Data, 12-Bit, 2s-Compliment Mode)

### 25.2.5 DAC12 Output Amplifier Offset Calibration

The offset voltage of the DAC12 output amplifier can be positive or negative. When the offset is negative, the output amplifier attempts to drive the voltage negative but cannot do so. The output voltage remains at zero until the DAC12 digital input produces a sufficient positive output voltage to overcome the negative offset voltage, resulting in the transfer function shown in Figure 25-4.

![](Figure 25-4. Output Voltage vs DAC12 Data, 12-Bit, Offset Calibration)
When the output amplifier has a positive offset, a digital input of zero does not result in a zero output voltage. The DAC12 output voltage reaches the maximum output level before the DAC12 data reaches the maximum code. This is shown in Figure 25-5.

The DAC12 has the capability to calibrate the offset voltage of the output amplifier. Setting the DAC12CALON bit initiates the offset calibration. The calibration should complete before using the DAC12. When the calibration is complete, the DAC12CALON bit is automatically reset. The DAC12AMPx bits should be configured before calibration. For best calibration results, port and CPU activity should be minimized during calibration.

25.2.6 Grouping Multiple DAC12 Modules

Multiple DAC12s can be grouped together with the DAC12GRP bit to synchronize the update of each DAC12 output. Hardware ensures that all DAC12 modules in a group update simultaneously independent of any interrupt or NMI event.

DAC12_0 and DAC12_1 are grouped by setting the DAC12GRP bit of DAC12_0. The DAC12GRP bit of DAC12_1 is don’t care. When DAC12_0 and DAC12_1 are grouped:

- The DAC12_1 DAC12LSELx bits select the update trigger for both DACs
- The DAC12LSELx bits for both DACs must be > 0
- The DAC12ENC bits of both DACs must be set to 1

When DAC12_0 and DAC12_1 are grouped, both DAC12_xDAT registers must be written to before the outputs update, even if data for one or both of the DACs is not changed. Figure 25-6 shows a latch-update timing example for grouped DAC12_0 and DAC12_1.

When DAC12_0 DAC12GRP = 1 and both DAC12_x DAC12LSELx > 0 and either DAC12ENC = 0, neither DAC12 updates.
**Note**

**DAC12 Settling Time**

The DMA controller is capable of transferring data to the DAC12 faster than the DAC12 output can settle. The user must assure the DAC12 settling time is not violated when using the DMA controller. See the device-specific data sheet for parameters.

**25.2.7 DAC12 Interrupts**

The DAC12 interrupt vector is shared with the DMA controller on some devices (see device-specific data sheet for interrupt assignment). In this case, software must check the DAC12IFG and DMAIFG flags to determine the source of the interrupt.

The DAC12IFG bit is set when DAC12LSELx > 0 and DAC12 data is latched from the DAC12_xDAT register into the data latch. When DAC12LSELx = 0, the DAC12IFG flag is not set.

A set DAC12IFG bit indicates that the DAC12 is ready for new data. If both the DAC12IE and GIE bits are set, the DAC12IFG generates an interrupt request. The DAC12IFG flag is not reset automatically. It must be reset by software.
25.3 DAC12 Registers

Table 25-2 lists the memory-mapped registers for the DAC12.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>1C0h</td>
<td>DAC12_0CTL</td>
<td>DAC12_0 control</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 25.3.1</td>
</tr>
<tr>
<td>1C8h</td>
<td>DAC12_0DAT</td>
<td>DAC12_0 data</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 25.3.2</td>
</tr>
<tr>
<td>1C2h</td>
<td>DAC12_1CTL</td>
<td>DAC12_1 control</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 25.3.1</td>
</tr>
<tr>
<td>1CAh</td>
<td>DAC12_1DAT</td>
<td>DAC12_1 data</td>
<td>Read/write</td>
<td>00h with POR</td>
<td>Section 25.3.2</td>
</tr>
</tbody>
</table>
25.3.1 DAC12_xCTL Register

DAC12_x Control Register

DAC12_xCTL are shown in Figure 25-7 and described in Table 25-3.

Return to Table 25-2.

Figure 25-7. DAC12_xCTL Registers

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15</td>
<td>DAC12OPS</td>
<td>R/W</td>
<td>0h</td>
<td>DAC12 output select. Can be modified only when DAC12ENC = 0.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = DAC12_0 output on P6.6. DAC12_1 output on P6.7</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = DAC12_0 output on V_{eREF+}, DAC12_1 output on P6.5</td>
</tr>
<tr>
<td>14-13</td>
<td>DAC12SREFx</td>
<td>R/W</td>
<td>0h</td>
<td>DAC12 select reference voltage. Can be modified only when DAC12ENC = 0</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = V_{REF+}</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = V_{REF+}</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = V_{eREF+}</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = V_{eREF+}</td>
</tr>
<tr>
<td>12</td>
<td>DAC12RES</td>
<td>R/W</td>
<td>0h</td>
<td>DAC12 resolution select. Can be modified only when DAC12ENC = 0.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = 12-bit resolution</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = 8-bit resolution</td>
</tr>
<tr>
<td>11-10</td>
<td>DAC12LSELx</td>
<td>R/W</td>
<td>0h</td>
<td>DAC12 load select. Selects the load trigger for the DAC12 latch.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>DAC12ENC must be set for the DAC to update, except when DAC12LSELx = 0.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = DAC12 latch loads when DAC12_xDAT written (DAC12ENC is ignored)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = DAC12 latch loads written, or, when grouped, when all DAC12_xDAT</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>registers in the group have been written.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = Rising edge of Timer_A.OUT1 (TA1)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = Rising edge of Timer_B.OUT2 (TB2)</td>
</tr>
<tr>
<td>9</td>
<td>DAC12CALON</td>
<td>R/W</td>
<td>0h</td>
<td>DAC12 calibration on. This bit initiates the DAC12 offset calibration</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>sequence and is automatically reset when the calibration completes.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Calibration is not active</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Initiate calibration or calibration in progress</td>
</tr>
<tr>
<td>8</td>
<td>DAC12IR</td>
<td>R/W</td>
<td>0h</td>
<td>DAC12 input range. This bit sets the reference input and voltage</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>output range. Can be modified only when DAC12ENC = 0.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = DAC12 full-scale output = 3x reference voltage</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = DAC12 full-scale output = 1x reference voltage</td>
</tr>
</tbody>
</table>
Table 25-3. DAC12_xCTL Register Field Descriptions (continued)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-5</td>
<td>DAC12AMPx</td>
<td>R/W</td>
<td>0h</td>
<td>DAC12 amplifier setting. These bits select settling time vs current consumption for the DAC12 input and output amplifiers. See Table 25-4. Can be modified only when DAC12ENC = 0.</td>
</tr>
<tr>
<td>4</td>
<td>DAC12DF</td>
<td>R/W</td>
<td>0h</td>
<td>DAC12 data format. Can be modified only when DAC12ENC = 0. 0b = Straight binary 1b = 2s complement</td>
</tr>
<tr>
<td>3</td>
<td>DAC12IE</td>
<td>R/W</td>
<td>0h</td>
<td>DAC12 interrupt enable 0b = Disabled 1b = Enabled</td>
</tr>
<tr>
<td>2</td>
<td>DAC12IFG</td>
<td>R/W</td>
<td>0h</td>
<td>DAC12 Interrupt flag 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
<tr>
<td>1</td>
<td>DAC12ENC</td>
<td>R/W</td>
<td>0h</td>
<td>DAC12 enable conversion. This bit enables the DAC12 module when DAC12LSELx &gt; 0. When DAC12LSELx = 0, DAC12ENC is ignored. 0b = DAC12 disabled 1b = DAC12 enabled</td>
</tr>
<tr>
<td>0</td>
<td>DAC12GRP</td>
<td>R/W</td>
<td>0h</td>
<td>DAC12 group. Groups DAC12_x with the next higher DAC12_.x. Not used for DAC12_1. 0b = Not grouped 1b = Grouped</td>
</tr>
</tbody>
</table>

Table 25-4. DAC12 Amplifier Settings

<table>
<thead>
<tr>
<th>DAC12AMPx</th>
<th>Input Buffer</th>
<th>Output Buffer</th>
</tr>
</thead>
<tbody>
<tr>
<td>000b</td>
<td>Off</td>
<td>DAC12 off, output high Z</td>
</tr>
<tr>
<td>001b</td>
<td>Off</td>
<td>DAC12 off, output 0 V</td>
</tr>
<tr>
<td>010b</td>
<td>Low speed and current</td>
<td>Low speed and current</td>
</tr>
<tr>
<td>011b</td>
<td>Low speed and current</td>
<td>Medium speed and current</td>
</tr>
<tr>
<td>100b</td>
<td>Low speed and current</td>
<td>High speed and current</td>
</tr>
<tr>
<td>101b</td>
<td>Medium speed and current</td>
<td>Medium speed and current</td>
</tr>
<tr>
<td>110b</td>
<td>Medium speed and current</td>
<td>High speed and current</td>
</tr>
<tr>
<td>111b</td>
<td>High speed and current</td>
<td>High speed and current</td>
</tr>
</tbody>
</table>
25.3.2 DAC12_xDAT Register

DAC12_x Data Register

DAC12_xDAT are shown in Figure 25-8 and described in Table 25-5.

Return to Table 25-2.

Figure 25-8. DAC12_xDAT Registers

<table>
<thead>
<tr>
<th>15</th>
<th>14</th>
<th>13</th>
<th>12</th>
<th>11</th>
<th>10</th>
<th>9</th>
<th>8</th>
</tr>
</thead>
<tbody>
<tr>
<td>r(0)</td>
<td>r(0)</td>
<td>r(0)</td>
<td>r(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
<td>rw-(0)</td>
</tr>
<tr>
<td>7</td>
<td>6</td>
<td>5</td>
<td>4</td>
<td>3</td>
<td>2</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>DAC12 Data</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Table 25-5. DAC12_xDAT Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-12</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td>Unused. These bits are always 0 and do not affect the DAC12 core.</td>
</tr>
<tr>
<td>11-0</td>
<td>DAC12 Data</td>
<td>R/W</td>
<td>0h</td>
<td>DAC12 data. See Table 25-6.</td>
</tr>
</tbody>
</table>

Table 25-6. DAC12 Data Format

<table>
<thead>
<tr>
<th>DAC12 Data Format</th>
<th>DAC12 Data</th>
</tr>
</thead>
<tbody>
<tr>
<td>12-bit binary</td>
<td>The DAC12 data are right justified. Bit 11 is the MSB.</td>
</tr>
<tr>
<td>12-bit two's complement</td>
<td>The DAC12 data are right justified. Bit 11 is the MSB (sign).</td>
</tr>
<tr>
<td>8-bit binary</td>
<td>The DAC12 data are right justified. Bit 7 is the MSB. Bits 11-8 are don't care and do not affect the DAC12 core.</td>
</tr>
<tr>
<td>8-bit two's complement</td>
<td>The DAC12 data are right justified. Bit 7 is the MSB (sign). Bits 11-8 are don't care and do not affect the DAC12 core.</td>
</tr>
</tbody>
</table>
The SD16_A module is a single-converter 16-bit sigma-delta analog-to-digital conversion module with high impedance input buffer. This chapter describes the SD16_A. The SD16_A module is implemented in the MSP430x20x3 devices.

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26.3 SD16_A Registers....................................................................................................................662
26.1 SD16_A Introduction

The SD16_A module consists of one sigma-delta analog-to-digital converter with a high-impedance input buffer and an internal voltage reference. It has up to eight fully differential multiplexed analog input pairs including a built-in temperature sensor and a divided supply voltage. The converter is based on a second-order oversampling sigma-delta modulator and digital decimation filter. The decimation filter is a comb type filter with selectable oversampling ratios of up to 1024. Additional filtering can be done in software.

The high impedance input buffer is not implemented in MSP430x20x3 devices.

Features of the SD16_A include:

- 16-bit sigma-delta architecture
- Up to eight multiplexed differential analog inputs per channel (The number of inputs is device dependent, see the device-specific data sheet.)
- Software selectable on-chip reference voltage generation (1.2 V)
- Software selectable internal or external reference
- Built-in temperature sensor
- Up to 1.1-MHz modulator input frequency
- High impedance input buffer (not implemented on all devices, see the device-specific data sheet)
- Selectable low-power conversion mode

The block diagram of the SD16_A module is shown in Figure 26-1.
Figure 26-1. SD16_A Block Diagram
26.2 SD16_A Operation

The SD16_A module is configured with user software. The setup and operation of the SD16_A is discussed in the following sections.

26.2.1 ADC Core

The analog-to-digital conversion is performed by a 1-bit second-order sigma-delta modulator. A single-bit comparator within the modulator quantizes the input signal with the modulator frequency $f_M$. The resulting 1-bit data stream is averaged by the digital filter for the conversion result.

26.2.2 Analog Input Range and PGA

The full-scale input voltage range for each analog input pair is dependent on the gain setting of the programmable gain amplifier of each channel. The maximum full-scale range is $V_{FSR}$ where $V_{FSR}$ is defined by:

$$V_{FSR} = \frac{V_{REF}/2}{GAIN_{PGA}}$$

For a 1.2-V reference, the maximum full-scale input range for a gain of 1 is:

$$\pm V_{FSR} = \frac{1.2V}{2} = \pm 0.6V$$

See the device-specific data sheet for full-scale input specifications.

26.2.3 Voltage Reference Generator

The SD16_A module has a built-in 1.2-V reference. It is enabled by the SD16REFON bit. When using the internal reference an external 100-nF capacitor connected from $V_{REF}$ to $AV_{SS}$ is recommended to reduce noise. The internal reference voltage can be used off-chip when SD16VMIDON = 1. The buffered output can provide up to 1 mA of drive. When using the internal reference off-chip, a 470-nF capacitor connected from $V_{REF}$ to $AV_{SS}$ is required. See the device-specific data sheet for parameters.

An external voltage reference can be applied to the $V_{REF}$ input when SD16REFON and SD16VMIDON are both reset.

26.2.4 Auto Power-Down

The SD16_A is designed for low power applications. When the SD16_A is not actively converting, it is automatically disabled and automatically re-enabled when a conversion is started. The reference is not automatically disabled, but can be disabled by setting SD16REFON = 0. When the SD16_A or reference are disabled, they consume no current.

26.2.5 Analog Input Pair Selection

The SD16_A can convert up to 8 differential input pairs multiplexed into the PGA. Up to five analog input pairs (A0-A4) are available externally on the device. A resistive divider to measure the supply voltage is available using the A5 multiplexer input. An internal temperature sensor is available using the A6 multiplexer input. Input A7 is a shorted connection between the + and - input pair and can be used to calibrate the offset of the SD16_A input stage. Note that the measured offset depends on the impedance of the external circuitry; thus, the actual offset seen at any of the analog inputs may be different.

26.2.5.1 Analog Input Setup

The analog input is configured using the SD16INCTL0 and the SD16AE registers. The SD16INCHx bits select one of eight differential input pairs of the analog multiplexer. The gain for the PGA is selected by the SD16GAINx bits. A total of six gain settings are available. The SD16AEex bits enable or disable the analog input pin. Setting any SD16AEex bit disables the multiplexed digital circuitry for the associated pin. See the device-specific data sheet for pin diagrams.
During conversion any modification to the SD16INCHx and SD16GAINx bits will become effective with the next decimation step of the digital filter. After these bits are modified, the next three conversions may be invalid due to the settling time of the digital filter. This can be handled automatically with the SD16INTDLYx bits. When SD16INTDLY = 00h, conversion interrupt requests will not begin until the fourth conversion after a start condition.

On devices implementing the high impedance input buffer it can be enabled using the SD16BUFx bits. The speed settings are selected based on the SD16_A modulator frequency as shown in Table 26-1.

<table>
<thead>
<tr>
<th>Buffer</th>
<th>SD16 Modulator Frequency f_M</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>Buffer disabled</td>
</tr>
<tr>
<td>01</td>
<td>Low speed/current</td>
</tr>
<tr>
<td>10</td>
<td>Medium speed/current</td>
</tr>
<tr>
<td>11</td>
<td>High speed/current</td>
</tr>
</tbody>
</table>

f_M < 200 kHz
200 kHz < f_M < 700 kHz
700 kHz < f_M < 1.1 MHz

An external RC anti-aliasing filter is recommended for the SD16_A to prevent aliasing of the input signal. The cutoff frequency should be < 10 kHz for a 1-MHz modulator clock and OSR = 256. The cutoff frequency may be set to a lower frequency for applications that have lower bandwidth requirements.

### 26.2.6 Analog Input Characteristics

The SD16_A uses a switched-capacitor input stage that appears as an impedance to external circuitry as shown in Figure 26-2.

![Analog Input Equivalent Circuit](image)

When the buffers are used, R_S does not affect the sampling frequency f_S. However, when the buffers are not used or are not present on the device, the maximum sampling frequency f_S may be calculated from the minimum settling time t_Setting of the sampling circuit given by:

\[
\text{t}_{\text{Setting}} \geq (R_S + 1 \text{k} \Omega) \times C_S \times \ln \left( \frac{GAIN \times 2^{17} \times V_{AX}}{V_{REF}} \right)
\]

where

\[
f_S = \frac{1}{2 \times \text{t}_{\text{Setting}}} \quad \text{and} \quad V_{AX} = \max \left( \frac{AV_{CC}}{2} - V_{S+} , \frac{AV_{CC}}{2} - V_{S-} \right)
\]

with \(V_{S+}\) and \(V_{S-}\) referenced to \(AV_{SS}\).
C_S varies with the gain setting as shown in Table 26-2.

<table>
<thead>
<tr>
<th>PGA Gain</th>
<th>Sampling Capacitance, C_S</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>1.25 pF</td>
</tr>
<tr>
<td>2, 4</td>
<td>2.5 pF</td>
</tr>
<tr>
<td>8</td>
<td>5 pF</td>
</tr>
<tr>
<td>16, 32</td>
<td>10 pF</td>
</tr>
</tbody>
</table>

26.2.7 Digital Filter

The digital filter processes the 1-bit data stream from the modulator using a SINC^3 comb filter. The transfer function is described in the z-Domain by:

\[ H(z) = \left( \frac{1}{OSR} \times \frac{1 - z^{-OSR}}{1 - z^{-1}} \right)^3 \]

and in the frequency domain by:

\[ H(f) = \left[ \frac{sinc\left(\frac{OSR \times \pi \times \frac{f}{f_M}}{1}\right)}{sinc\left(\frac{\pi \times \frac{f}{f_M}}{1}\right)} \right]^3 \]

\[ \times \left[ \frac{1}{OSR} \times \frac{\sin\left(\frac{OSR \times \pi \times \frac{f}{f_M}}{\sin\left(\frac{\pi \times \frac{f}{f_M}}{1}\right)} \right)}{\sin\left(\frac{\pi \times \frac{f}{f_M}}{1}\right)} \right]^3 \]

where the oversampling rate, OSR, is the ratio of the modulator frequency f_M to the sample frequency f_S. Figure 26-3 shows the filter's frequency response for an OSR of 32. The first filter notch is at f_S = f_M/OSR. The notch's frequency can be adjusted by changing the modulator's frequency, f_M, using SD16SSELx and SD16DIVx and the oversampling rate using the SD16OSRx and SD16XOSR bits.

The digital filter for each enabled ADC channel completes the decimation of the digital bit-stream and outputs new conversion results to the SD16MEM0 register at the sample frequency f_S.

![Figure 26-3. Comb Filter Frequency Response With OSR = 32](image)

Figure 26-4 shows the digital filter step response and conversion points. For step changes at the input after start of conversion a settling time must be allowed before a valid conversion result is available. The SD16INTDLYx bits can provide sufficient filter settling time for a full-scale change at the ADC input. If the step occurs synchronously to the decimation of the digital filter the valid data will be available on the third conversion. An asynchronous step will require one additional conversion before valid data is available.
26.2.7.1 Digital Filter Output

The number of bits output by the digital filter is dependent on the oversampling ratio and ranges from 15 to 30 bits. Figure 26-5 shows the digital filter output and their relation to SD16MEM0 for each OSR, LSBACC, and SD16UNI setting. For example, for OSR = 1024, LSBACC = 0, and SD16UNI = 1, the SD16MEM0 register contains bits 28 - 13 of the digital filter output. When OSR = 32, the one (SD16UNI = 0) or two (SD16UNI=1) LSBs are always zero.

The SD16LSBACC and SD16LSBTOG bits give access to the least significant bits of the digital filter output. When SD16LSBACC = 1 the 16 least significant bits of the digital filter’s output are read from SD16MEM0 using word instructions. The SD16MEM0 register can also be accessed with byte instructions returning only the 8 least significant bits of the digital filter output.

When SD16LSBTOG = 1 the SD16LSBACC bit is automatically toggled each time SD16MEM0 is read. This allows the complete digital filter output result to be read with two reads of SD16MEM0. Setting or clearing SD16LSBTOG does not change SD16LSBACC until the next SD16MEM0 access.

Figure 26-4. Digital Filter Step Response and Conversion Points
OSR=1024, LSBACC=0, SD16UNI=1

OSR=1024, LSBACC=1, SD16UNI=1

OSR=1024, LSBACC=0, SD16UNI=0

OSR=512, LSBACC=0, SD16UNI=1

OSR=512, LSBACC=1, SD16UNI=0

OSR=512, LSBACC=1, SD16UNI=1

OSR=512, LSBACC=0, SD16UNI=0

OSR=512, LSBACC=1, SD16UNI=0
Figure 26-5. Used Bits of Digital Filter Output
26.2.8 Conversion Memory Register: SD16MEM0

The SD16MEM0 register is associated with the SD16_A channel. Conversion results are moved to the SD16MEM0 register with each decimation step of the digital filter. The SD16IFG bit is set when new data is written to SD16MEM0. SD16IFG is automatically cleared when SD16MEM0 is read by the CPU or may be cleared with software.

26.2.8.1 Output Data Format

The output data format is configurable in two's complement, offset binary or unipolar mode as shown in Table 26-3. The data format is selected by the SD16DF and SD16UNI bits.

Table 26-3. Data Format

<table>
<thead>
<tr>
<th>SD16UNI</th>
<th>SD16DF</th>
<th>Format</th>
<th>Analog Input</th>
<th>SD16MEM0(1)</th>
<th>Digital Filter Output (OSR = 256)</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>Bipolar Offset Binary</td>
<td>+FSR</td>
<td>FFFF</td>
<td>FFFFFFFF</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>ZERO</td>
<td>8000</td>
<td>800000</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>-FSR</td>
<td>0000</td>
<td>000000</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>Bipolar Twos Compliment</td>
<td>+FSR</td>
<td>7FFF</td>
<td>7FFFFFFF</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>ZERO</td>
<td>0000</td>
<td>000000</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>-FSR</td>
<td>8000</td>
<td>800000</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>Unipolar</td>
<td>+FSR</td>
<td>FFFF</td>
<td>FFFFFFFF</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>ZERO</td>
<td>0000</td>
<td>800000</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>-FSR</td>
<td>0000</td>
<td>000000</td>
</tr>
</tbody>
</table>

(1) Independent of SD16OSRx and SD16XOSR settings; SD16LSBACC = 0.

Note

Offset Measurements and Data Format

Any offset measurement done either externally or using the internal differential pair A7 would be appropriate only when the channel is operating under bipolar mode with SD16UNI = 0.

Figure 26-6 shows the relationship between the full-scale input voltage range from $-V_{FSR}$ to $+V_{FSR}$ and the conversion result. The data formats are illustrated.

![Figure 26-6. Input Voltage vs Digital Output](image_url)
26.2.9 Conversion Modes

The SD16_A module can be configured for two modes of operation, listed in Table 26-4. The SD16SNGL bit selects the conversion mode.

<table>
<thead>
<tr>
<th>SD16SNGL</th>
<th>Mode</th>
<th>Operation</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>Single conversion</td>
<td>The channel is converted once.</td>
</tr>
<tr>
<td>0</td>
<td>Continuous conversion</td>
<td>The channel is converted continuously.</td>
</tr>
</tbody>
</table>

26.2.9.1 Single Conversion

Setting the SD16SC bit of the channel initiates one conversion on that channel when SD16SNGL = 1. The SD16SC bit will automatically be cleared after conversion completion.

Clearing SD16SC before the conversion is completed immediately stops conversion of the channel, the channel is powered down and the corresponding digital filter is turned off. The value in SD16MEM0 can change when SD16SC is cleared. It is recommended that the conversion data in SD16MEM0 be read prior to clearing SD16SC to avoid reading an invalid result.

26.2.9.2 Continuous Conversion

When SD16SNGL = 0 continuous conversion mode is selected. Conversion of the channel will begin when SD16SC is set and continue until the SD16SC bit is cleared by software.

Clearing SD16SC immediately stops conversion of the selected channel, the channel is powered down and the corresponding digital filter is turned off. The value in SD16MEM0 can change when SD16SC is cleared. It is recommended that the conversion data in SD16MEM0 be read prior to clearing SD16SC to avoid reading an invalid result.

Figure 26-7 shows conversion operation.

Figure 26-7. Single Channel Operation

26.2.10 Using the Integrated Temperature Sensor

To use the on-chip temperature sensor, the user selects the analog input pair SD16INCHx = 110 and sets SD16REFON = 1. Any other configuration is done as if an external analog input pair was selected, including SD16INTDLYx and SD16GAINx settings. Because the internal reference must be on to use the temperature sensor, it is not possible to use an external reference for the conversion of the temperature sensor voltage. Also, the internal reference will be in contention with any used external reference. In this case, the SD16VMIDON bit may be set to minimize the affects of the contention on the conversion.

The typical temperature sensor transfer function is shown in Figure 26-8. When switching inputs of an SD16_A channel to the temperature sensor, adequate delay must be provided using SD16INTDLYx to allow the digital filter to settle and assure that conversion results are valid. The temperature sensor offset error can be large, and may need to be calibrated for most applications. See device-specific data sheet for temperature sensor parameters.
26.2.11 Interrupt Handling

The SD16_A has 2 interrupt sources for its ADC channel:

- SD16IFG
- SD16OVIFG

The SD16IFG bit is set when the SD16MEM0 memory register is written with a conversion result. An interrupt request is generated if the corresponding SD16IE bit and the GIE bit are set. The SD16_A overflow condition occurs when a conversion result is written to SD16MEM0 location before the previous conversion result was read.

26.2.11.1 SD16IV, Interrupt Vector Generator

All SD16_A interrupt sources are prioritized and combined to source a single interrupt vector. SD16IV is used to determine which enabled SD16_A interrupt source requested an interrupt. The highest priority SD16_A interrupt request that is enabled generates a number in the SD16IV register (see register description). This number can be evaluated or added to the program counter to automatically enter the appropriate software routine. Disabled SD16_A interrupts do not affect the SD16IV value.

Any access, read or write, of the SD16IV register has no effect on the SD16OVIFG or SD16IFG flags. The SD16IFG flags are reset by reading the SD16MEM0 register or by clearing the flags in software. SD16OVIFG bits can only be reset with software.

If another interrupt is pending after servicing of an interrupt, another interrupt is generated. For example, if the SD16OVIFG and one or more SD16IFG interrupts are pending when the interrupt service routine accesses the SD16IV register, the SD16OVIFG interrupt condition is serviced first and the corresponding flag(s) must be cleared in software. After the RETI instruction of the interrupt service routine is executed, the highest priority SD16IFG pending generates another interrupt request.
26.2.11.2 Interrupt Delay Operation

The SD16INTDLYx bits control the timing for the first interrupt service request for the corresponding channel. This feature delays the interrupt request for a completed conversion by up to four conversion cycles allowing the digital filter to settle prior to generating an interrupt request. The delay is applied each time the SD16SC bit is set or when the SD16GAINx or SD16INCHx bits for the channel are modified. SD16INTDLYx disables overflow interrupt generation for the channel for the selected number of delay cycles. Interrupt requests for the delayed conversions are not generated during the delay.
### 26.3 SD16_A Registers

Table 26-5 lists the memory-mapped registers for the SD16_A.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>100h</td>
<td>SD16CTL</td>
<td>SD16_A control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 26.3.1</td>
</tr>
<tr>
<td>102h</td>
<td>SD16CCTL0</td>
<td>SD16_A channel 0 control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 26.3.2</td>
</tr>
<tr>
<td>112h</td>
<td>SD16MEM0</td>
<td>SD16_A conversion memory</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 26.3.3</td>
</tr>
<tr>
<td>B0h</td>
<td>SD16INCTL0</td>
<td>SD16_A input control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 26.3.4</td>
</tr>
<tr>
<td>B7h</td>
<td>SD16AE</td>
<td>SD16_A analog enable</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 26.3.5</td>
</tr>
<tr>
<td>110h</td>
<td>SD16IV</td>
<td>SD16_A interrupt vector</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 26.3.6</td>
</tr>
</tbody>
</table>
26.3.1 SD16CTL Register

SD16_A Control Register

SD16CTL is shown in Figure 26-9 and described in Table 26-6.

Return to Table 26-5.

Figure 26-9. SD16CTL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-12</td>
<td>Reserved</td>
</tr>
<tr>
<td>11-9</td>
<td>SD16XDIVx</td>
</tr>
<tr>
<td>8</td>
<td>SD16LP</td>
</tr>
<tr>
<td>7-6</td>
<td>SD16DIVx</td>
</tr>
<tr>
<td>5-4</td>
<td>SD16SSELx</td>
</tr>
<tr>
<td>3</td>
<td>SD16VMIDON</td>
</tr>
<tr>
<td>2</td>
<td>SD16REFON</td>
</tr>
<tr>
<td>1</td>
<td>SD16OVIE</td>
</tr>
<tr>
<td>0</td>
<td>Reserved</td>
</tr>
</tbody>
</table>

Table 26-6. SD16CTL Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-12</td>
<td>Reserved</td>
<td>R</td>
<td>0</td>
<td></td>
</tr>
<tr>
<td>11-9</td>
<td>SD16XDIVx</td>
<td>R/W</td>
<td>0</td>
<td>SD16_A clock divider</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>000b = /1</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>001b = /3</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>010b = /16</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>011b = /48</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1xxb = Reserved</td>
</tr>
<tr>
<td>8</td>
<td>SD16LP</td>
<td>R/W</td>
<td>0</td>
<td>Low-power mode. This bit selects a reduced-speed reduced-power</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>mode.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Low-power mode is disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Low-power mode is enabled. The maximum clock frequency for</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>the SD16_A is reduced.</td>
</tr>
<tr>
<td>7-6</td>
<td>SD16DIVx</td>
<td>R/W</td>
<td>0</td>
<td>SD16_A clock divider</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = /1</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = /2</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = /4</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = /8</td>
</tr>
<tr>
<td>5-4</td>
<td>SD16SSELx</td>
<td>R/W</td>
<td>0</td>
<td>SD16_A clock source select</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = MCLK</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = SMCLK</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = ACLK</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = External TACLK</td>
</tr>
<tr>
<td>3</td>
<td>SD16VMIDON</td>
<td>R/W</td>
<td>0</td>
<td>VMID buffer on</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Off</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = On</td>
</tr>
<tr>
<td>2</td>
<td>SD16REFON</td>
<td>R/W</td>
<td>0</td>
<td>Reference generator on</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Reference off</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Reference on</td>
</tr>
<tr>
<td>1</td>
<td>SD16OVIE</td>
<td>R/W</td>
<td>0</td>
<td>SD16_A overflow interrupt enable. The GIE bit must also be set to</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>enable the interrupt.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Overflow interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Overflow interrupt enabled</td>
</tr>
<tr>
<td>0</td>
<td>Reserved</td>
<td>R</td>
<td>0</td>
<td></td>
</tr>
</tbody>
</table>
26.3.2 SD16CCTL0 Register

SD16_A Channel 0 Control Register

SD16CCTL0 is shown in Figure 26-10 and described in Table 26-7.

Return to Table 26-5.

Figure 26-10. SD16CCTL0 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td>High-impedance input buffer mode. Reserved in MSP430x20x3 devices</td>
</tr>
<tr>
<td>14-13</td>
<td>SD16BUFx(1)</td>
<td>R/W</td>
<td>0h</td>
<td>00b = Buffer disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = Low speed and current</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = Medium speed and current</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = High speed and current</td>
</tr>
<tr>
<td>12</td>
<td>SD16UNI</td>
<td>R/W</td>
<td>0h</td>
<td>Unipolar mode select</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Bipolar mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Unipolar mode</td>
</tr>
<tr>
<td>11</td>
<td>SD16XOSR</td>
<td>R/W</td>
<td>0h</td>
<td>Extended oversampling ratio. This bit, along with the SD16OSRx bits, select the oversampling ratio. See SD16OSRx bit description for settings.</td>
</tr>
<tr>
<td>10</td>
<td>SD16SNGL</td>
<td>R/W</td>
<td>0h</td>
<td>Single conversion mode select</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Continuous conversion mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Single conversion mode</td>
</tr>
<tr>
<td>9-8</td>
<td>SD16OSRx</td>
<td>R/W</td>
<td>0h</td>
<td>Oversampling ratio</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>When SD16XOSR = 0</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = 256</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = 128</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = 64</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = 32</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>When SD16XOSR = 1</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = 512</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = 1024</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = Reserved</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = Reserved</td>
</tr>
</tbody>
</table>

(1) Reserved in MSP430x20x3 devices

Table 26-7. SD16CCTL0 Register Field Descriptions
<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>SD16LSBTOG</td>
<td>R/W</td>
<td>0h</td>
<td>LSB toggle. This bit, when set, causes SD16LSBACC to toggle each time the SD16MEM0 register is read. 0b = SD16LSBACC does not toggle with each SD16MEM0 read 1b = SD16LSBACC toggles with each SD16MEM0 read</td>
</tr>
<tr>
<td>6</td>
<td>SD16LSBACC</td>
<td>R/W</td>
<td>0h</td>
<td>LSB access. This bit allows access to the upper or lower 16-bits of the SD16_A conversion result. 0b = SD16MEMx contains the most significant 16-bits of the conversion. 1b = SD16MEMx contains the least significant 16-bits of the conversion.</td>
</tr>
<tr>
<td>5</td>
<td>SD16OVIFG</td>
<td>R/W</td>
<td>0h</td>
<td>SD16_A overflow interrupt flag 0b = No overflow interrupt pending 1b = Overflow interrupt pending</td>
</tr>
<tr>
<td>4</td>
<td>SD16DF</td>
<td>R/W</td>
<td>0h</td>
<td>SD16_A data format 0b = Offset binary 1b = 2s complement</td>
</tr>
<tr>
<td>3</td>
<td>SD16IE</td>
<td>R/W</td>
<td>0h</td>
<td>SD16_A interrupt enable 0b = Disabled 1b = Enabled</td>
</tr>
<tr>
<td>2</td>
<td>SD16IFG</td>
<td>R/W</td>
<td>0h</td>
<td>SD16_A interrupt flag. SD16IFG is set when new conversion results are available. SD16IFG is automatically reset when the corresponding SD16MEMx register is read, or may be cleared with software. 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
<tr>
<td>1</td>
<td>SD16SC</td>
<td>R/W</td>
<td>0h</td>
<td>SD16_A start conversion 0b = No conversion start 1b = Start conversion</td>
</tr>
<tr>
<td>0</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td></td>
</tr>
</tbody>
</table>
26.3.3 SD16MEM0 Register

SD16_A Conversion Memory Register

SD16MEM0 is shown in Figure 26-11 and described in Table 26-8.

Return to Table 26-5.

**Figure 26-11. SD16MEM0 Register**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>Conversion_Results</td>
<td>R</td>
<td>0h</td>
<td>Conversion Results. The SD16MEMx register holds the upper or lower 16-bits of the digital filter output, depending on the SD16LSBACC bit.</td>
</tr>
</tbody>
</table>
26.3.4 SD16INCTL0 Register

SD16_A Input Control Register

SD16INCTL0 is shown in Figure 26-12 and described in Table 26-9.

Return to Table 26-5.

Figure 26-12. SD16INCTL0 Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td>SD16INTDLYx</td>
<td>R/W</td>
<td>0h</td>
<td>Interrupt delay generation after conversion start. These bits select the</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>delay for the first interrupt after conversion start.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = Fourth sample causes interrupt</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = Third sample causes interrupt</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = Second sample causes interrupt</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = First sample causes interrupt</td>
</tr>
<tr>
<td>5-3</td>
<td>SD16GAINx</td>
<td>R/W</td>
<td>0h</td>
<td>SD16_A preamplifier gain</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>000b = ×1</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>001b = ×2</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>010b = ×4</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>011b = ×8</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>100b = ×16</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>101b = ×32</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>110b = Reserved</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>111b = Reserved</td>
</tr>
<tr>
<td>2-0</td>
<td>SD16INCHx</td>
<td>R/W</td>
<td>0h</td>
<td>SD16_A channel differential pair input</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>000b = A0</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>001b = A1</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>010b = A2</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>011b = A3</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>100b = A4</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>101b = A5, (AVCC – AVSS) / 11</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>110b = A6, Temperature sensor</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>111b = A7, Short for PGA offset measurement</td>
</tr>
</tbody>
</table>
26.3.5 SD16AE Register

SD16_A Analog Enable Register

SD16AE is shown in Figure 26-13 and described in Table 26-10.

Return to Table 26-5.

Figure 26-13. SD16AE Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
</table>
| 7   | SD16AE7 | R/W  | 0h    | SD16_A analog enable 7  
|     |        |      |       | 0b = External input disabled. Negative inputs are internally connected to VSS.  
|     |        |      |       | 1b = External input enabled. |
| 6   | SD16AE6 | R/W  | 0h    | SD16_A analog enable 6  
|     |        |      |       | 0b = External input disabled. Negative inputs are internally connected to VSS.  
|     |        |      |       | 1b = External input enabled. |
| 5   | SD16AE5 | R/W  | 0h    | SD16_A analog enable 5  
|     |        |      |       | 0b = External input disabled. Negative inputs are internally connected to VSS.  
|     |        |      |       | 1b = External input enabled. |
| 4   | SD16AE4 | R/W  | 0h    | SD16_A analog enable 4  
|     |        |      |       | 0b = External input disabled. Negative inputs are internally connected to VSS.  
|     |        |      |       | 1b = External input enabled. |
| 3   | SD16AE3 | R/W  | 0h    | SD16_A analog enable 3  
|     |        |      |       | 0b = External input disabled. Negative inputs are internally connected to VSS.  
|     |        |      |       | 1b = External input enabled. |
| 2   | SD16AE2 | R/W  | 0h    | SD16_A analog enable 2  
|     |        |      |       | 0b = External input disabled. Negative inputs are internally connected to VSS.  
|     |        |      |       | 1b = External input enabled. |
| 1   | SD16AE1 | R/W  | 0h    | SD16_A analog enable 1  
|     |        |      |       | 0b = External input disabled. Negative inputs are internally connected to VSS.  
|     |        |      |       | 1b = External input enabled. |
| 0   | SD16AE0 | R/W  | 0h    | SD16_A analog enable 0  
|     |        |      |       | 0b = External input disabled. Negative inputs are internally connected to VSS.  
|     |        |      |       | 1b = External input enabled. |
26.3.6 SD16IV Register

SD16_A Interrupt Vector Register

SD16IV is shown in Figure 26-14 and described in Table 26-11.

Return to Table 26-5.

Figure 26-14. SD16IV Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>SD16IVx</td>
<td>R</td>
<td>0h</td>
<td>SD16_A interrupt vector value. See Table 26-12.</td>
</tr>
</tbody>
</table>

Table 26-11. SD16IV Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>SD16IVx</td>
<td>R</td>
<td>0h</td>
<td>SD16_A interrupt vector value. See Table 26-12.</td>
</tr>
</tbody>
</table>

Table 26-12. SD16_A Interrupt Vectors

<table>
<thead>
<tr>
<th>SD16IV Contents</th>
<th>Interrupt Source</th>
<th>Interrupt Flag</th>
<th>Interrupt Priority</th>
</tr>
</thead>
<tbody>
<tr>
<td>000h</td>
<td>No interrupt pending</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>002h</td>
<td>SD16MEMx overflow</td>
<td>SD16CCTLx SD16OVIFG</td>
<td>Highest</td>
</tr>
<tr>
<td>004h</td>
<td>SD16_A interrupt</td>
<td>SD16CCTL0 SD16IFG</td>
<td>–</td>
</tr>
<tr>
<td>006h</td>
<td>Reserved</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>008h</td>
<td>Reserved</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>00Ah</td>
<td>Reserved</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>00Ch</td>
<td>Reserved</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>00Eh</td>
<td>Reserved</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>010h</td>
<td>Reserved</td>
<td>–</td>
<td>Lowest</td>
</tr>
</tbody>
</table>
This page intentionally left blank.
The SD24_A module is a multichannel 24-bit sigma-delta analog-to-digital converter (ADC). This chapter describes the SD24_A of the MSP430x2xx family.

27.1 SD24_A Introduction ................................................................................................................. 672
27.2 SD24_A Operation ..................................................................................................................... 674
27.3 SD24_A Registers ..................................................................................................................... 687
27.1 SD24_A Introduction

The SD24_A module consists of up to seven independent sigma-delta analog-to-digital converters, referred to as channels, and an internal voltage reference. Each channel has up to eight fully differential multiplexed analog input pairs including a built-in temperature sensor and a divided supply voltage. The converters are based on second-order oversampling sigma-delta modulators and digital decimation filters. The decimation filters are comb-type filters with selectable oversampling ratios of up to 1024. Additional filtering can be done in software.

The digital filter output of SD24_A can range from 15 bits to 30 bits, based on the oversampling ratio. The default oversampling ratio is 256, which results in 24-bit output from the digital filter. The 16 most significant bits of the filter are captured in the SD24_A conversion memory register and, by setting SD24LSBACC = 1, the 16 least significant bits of the filter output can be read (see Section 27.2.7 for details).

Features of the SD24_A include:

• Up to seven independent, simultaneously-sampling ADC channels (the number of channels is device dependent, see the device-specific data sheet)
• Up to eight multiplexed differential analog inputs per channel (the number of inputs is device dependent, see the device-specific data sheet)
• Software selectable on-chip reference voltage generation (1.2 V)
• Software selectable internal or external reference
• Built-in temperature sensor accessible by all channels
• Up to 1.1-MHz modulator input frequency
• High impedance input buffer (not implemented on all devices, see the device-specific data sheet)
• Selectable low-power conversion mode

The block diagram of the SD24_A module is shown in Figure 27-1.
A. Ax.1 to Ax.4 not available on all devices. See device-specific data sheet.

Figure 27-1. Block Diagram of the SD24_A
27.2 SD24_A Operation

The SD24_A module is configured with user software. The setup and operation of the SD24_A is discussed in the following sections.

27.2.1 ADC Core

The analog-to-digital conversion is performed by a 1-bit second-order sigma-delta modulator. A single-bit comparator within the modulator quantizes the input signal with the modulator frequency \( f_M \). The resulting 1-bit data stream is averaged by the digital filter for the conversion result.

27.2.2 Analog Input Range and PGA

The full-scale input voltage range for each analog input pair is dependent on the gain setting of the programmable gain amplifier of each channel. The maximum full-scale range is \( \pm V_{\text{FSR}} \) where \( V_{\text{FSR}} \) is defined by:

\[
V_{\text{FSR}} = \frac{V_{\text{REF}}}{2 \cdot \text{GAIN}_{\text{PGA}}}
\]

For a 1.2-V reference, the maximum full-scale input range for a gain of 1 is:

\[
\pm V_{\text{FSR}} = \frac{1.2 \text{ V}}{2} = \pm 0.6 \text{ V}
\]

See the device-specific data sheet for full-scale input specifications.

27.2.3 Voltage Reference Generator

The SD24_A module has a built-in 1.2-V reference. It can be used for each SD24_A channel and is enabled by the SD24REFON bit. When using the internal reference an external 100-nF capacitor connected from \( V_{\text{REF}} \) to \( V_{\text{SS}} \) is recommended to reduce noise. The internal reference voltage can be used off-chip when SD24VMIDON = 1. The buffered output can provide up to 1 mA of drive. When using the internal reference off-chip, a 470-nF capacitor connected from \( V_{\text{REF}} \) to \( V_{\text{SS}} \) is required. See device-specific data sheet for parameters.

An external voltage reference can be applied to the \( V_{\text{REF}} \) input when SD24REFON and SD24VMIDON are both reset.

27.2.4 Auto Power-Down

The SD24_A is designed for low-power applications. When the SD24_A is not actively converting, it is automatically disabled and automatically re-enabled when a conversion is started. The reference is not automatically disabled, but it can be disabled by setting SD24REFON = 0. When the SD24_A or reference are disabled, they consume no current.

27.2.5 Analog Input Pair Selection

The SD24_A can convert up to eight differential input pairs multiplexed into the PGA. Up to five analog input pairs (A0 to A4) are available externally on the device. A resistive divider to measure the supply voltage is available using the A5 multiplexer input. An internal temperature sensor is available using the A6 multiplexer input.

Input A7 is a shorted connection between the + and − input pair and can be used to calibrate the offset of the SD24_A input stage. Note that the measured offset depends on the impedance of the external circuitry; thus, the actual offset seen at any of the analog inputs may be different.

27.2.5.1 Analog Input Setup

The analog input of each channel is configured using the SD24INCTLx register. These settings can be independently configured for each SD24_A channel.

The SD24INChx bits select one of eight differential input pairs of the analog multiplexer. The gain for each PGA is selected by the SD24GAINx bits. A total of six gain settings are available.
On some devices SD24AEx bits are available to enable or disable the analog input pin. Setting any SD24AEx bit disables the multiplexed digital circuitry for the associated pin. See the device-specific data sheet for pin diagrams.

During conversion any modification to the SD24INCHx and SD24GAINx bits will become effective with the next decimation step of the digital filter. After these bits are modified, the next three conversions may be invalid due to the settling time of the digital filter. This can be handled automatically with the SD24INTDLYx bits. When SD24INTDLY = 00h, conversion interrupt requests will not begin until the fourth conversion after a start condition.

On devices implementing the high impedance input buffer it can be enabled using the SD24BUFx bits. The speed settings are selected based on the SD24_A modulator frequency as shown in Table 27-1.

### Table 27-1. High Input Impedance Buffer

<table>
<thead>
<tr>
<th>SD24BUFx</th>
<th>Buffer</th>
<th>SD24 Modulator Frequency, f_M</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>Buffer disabled</td>
<td></td>
</tr>
<tr>
<td>01</td>
<td>Low speed/current</td>
<td>f_M &lt; 200 kHz</td>
</tr>
<tr>
<td>10</td>
<td>Medium speed/current</td>
<td>200 kHz &lt; f_M &lt; 700 kHz</td>
</tr>
<tr>
<td>11</td>
<td>High speed/current</td>
<td>700 kHz &lt; f_M &lt; 1.1 MHz</td>
</tr>
</tbody>
</table>

An external RC anti-aliasing filter is recommended for the SD24_A to prevent aliasing of the input signal. The cutoff frequency should be less than 10 kHz for a 1-MHz modulator clock and OSR = 256. The cutoff frequency may set to a lower frequency for applications that have lower bandwidth requirements.

#### 27.2.6 Analog Input Characteristics

The SD24_A uses a switched-capacitor input stage that appears as an impedance to external circuitry as shown in Figure 27-2.

![Figure 27-2. Analog Input Equivalent Circuit](image)

† Not implemented on all devices – see the device-specific data sheet
When the buffers are used, $R_S$ does not affect the sampling frequency $f_S$. However, when the buffers are not used or are not present on the device, the maximum modulator frequency $f_M$ may be calculated from the minimum settling time $t_{\text{Setting}}$ of the sampling circuit given by:

$$t_{\text{Setting}} \geq (R_S + 1 \, \text{k}\Omega) \times C_S \times \ln\left(\frac{\text{GAIN} \times 2^{17} \times V_{\text{Ax}}}{V_{\text{REF}}}ight)$$

Where,

$$f_M = \frac{1}{2 \times t_{\text{Setting}}} \quad \text{and} \quad V_{\text{Ax}} = \max\left(\frac{AV_{\text{CC}}}{2} - V_{S+}, \frac{AV_{\text{CC}}}{2} - V_{S-}\right)$$

with $V_{S+}$ and $V_{S-}$ referenced to $AV_{\text{SS}}$.

$C_S$ varies with the gain setting as shown in Table 27-2.

<table>
<thead>
<tr>
<th>PGA Gain</th>
<th>Sampling Capacitance ($C_S$)</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>1.25 pF</td>
</tr>
<tr>
<td>2, 4</td>
<td>2.5 pF</td>
</tr>
<tr>
<td>8</td>
<td>5 pF</td>
</tr>
<tr>
<td>16, 32</td>
<td>10 pF</td>
</tr>
</tbody>
</table>

### 27.2.7 Digital Filter

The digital filter processes the 1-bit data stream from the modulator using a SINC$^3$ comb filter. The transfer function is described in the z-Domain by:

$$H(z) = \left(\frac{1}{\text{OSR}} \times \frac{1 - z^{-\text{OSR}}}{1 - z^{-1}}\right)^3$$

and in the frequency domain by:

$$H(f) = \left[\frac{\text{sinc}(\pi \times f) \times \frac{f}{f_M}}{\text{sinc}(\pi \times f_M)}\right]^3 = \left[\frac{1}{\text{OSR}} \times \frac{\text{sinc}(\pi \times f) \times \frac{f}{f_M}}{\text{sinc}(\pi \times f_M)}\right]^3$$

where the oversampling rate, OSR, is the ratio of the modulator frequency $f_M$ to the sample frequency $f_S$. Figure 27-3 shows the filter’s frequency response for an OSR of 32. The first filter notch is at $f_S = f_M/\text{OSR}$. The notch frequency can be adjusted by changing the modulator frequency, $f_M$, using SD24SSELx and SD24DIVx and the oversampling rate using the SD24OSRx and SD24XOSR bits.

The digital filter for each enabled ADC channel completes the decimation of the digital bit-stream and outputs new conversion results to the corresponding SD24MEMx register at the sample frequency $f_S$. 
Figure 27-3. Comb Filter Frequency Response With OSR = 32

Figure 27-4 shows the digital filter step response and conversion points. For step changes at the input after start of conversion a settling time must be allowed before a valid conversion result is available. The SD24INTDLYx bits can provide sufficient filter settling time for a full-scale change at the ADC input. If the step occurs synchronously to the decimation of the digital filter the valid data will be available on the third conversion. An asynchronous step will require one additional conversion before valid data is available.

Figure 27-4. Digital Filter Step Response and Conversion Points
27.2.7.1 Digital Filter Output

The number of bits output by the digital filter is dependent on the oversampling ratio and ranges from 15 to 30 bits. Figure 27-5 shows the digital filter output and their relation to SD24MEMx for each OSR, LSBACC, and SD24UNI setting. For example, for OSR = 1024, LSBACC = 0, and SD24UNI = 1, the SD24MEMx register contains bits 28 to 13 of the digital filter output. When OSR = 32, the one (SD24UNI = 0) or two (SD24UNI = 1) LSBs are always zero.

The SD24LSBACC and SD24LSBTOG bits give access to the least significant bits of the digital filter output. When SD24LSBACC = 1 the 16 least significant bits of the digital filter's output are read from SD24MEMx using word instructions. The SD24MEMx register can also be accessed with byte instructions returning only the 8 least significant bits of the digital filter output.

When SD24LSBTOG = 1 the SD24LSBACC bit is automatically toggled each time SD24MEMx is read. This allows the complete digital filter output result to be read with two reads of SD24MEMx. Setting or clearing SD24LSBTOG does not change SD24LSBACC until the next SD24MEMx access.
Figure 27-5. Used Bits of Digital Filter Output
27.2.8 Conversion Memory Register: SD24MEMx

One SD24MEMx register is associated with each SD24_A channel. Conversion results are moved to the corresponding SD24MEMx register with each decimation step of the digital filter. The SD24IFG bit is set when new data is written to SD24MEMx. SD24IFG is automatically cleared when SD24MEMx is read by the CPU or may be cleared with software.

27.2.8.1 Output Data Format

The output data format is configurable in two's complement, offset binary or unipolar mode as shown in Table 27-3. The data format is selected by the SD24DF and SD24UNI bits.

<table>
<thead>
<tr>
<th>SD24UNI</th>
<th>SD24DF</th>
<th>Format</th>
<th>Analog Input</th>
<th>SD24MEMx(1)</th>
<th>Digital Filter Output (OSR = 256)</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>Bipolar offset binary</td>
<td>+FSR</td>
<td>FFFF</td>
<td>FFFFFF</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>ZERO</td>
<td>8000</td>
<td>800000</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>-FSR</td>
<td>0000</td>
<td>000000</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>Bipolar twos compliment</td>
<td>+FSR</td>
<td>7FFF</td>
<td>7FFFFF</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>ZERO</td>
<td>0000</td>
<td>000000</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>-FSR</td>
<td>8000</td>
<td>800000</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>Unipolar</td>
<td>+FSR</td>
<td>FFFF</td>
<td>FFFFFF</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>ZERO</td>
<td>0000</td>
<td>800000</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>-FSR</td>
<td>0000</td>
<td>000000</td>
</tr>
</tbody>
</table>

(1) Independent of SD24OSRx and SD24XOSR settings; SD24LSBACC = 0.

Note

Offset Measurements and Data Format

Any offset measurement done either externally or using the internal differential pair A7 would be appropriate only when the channel is operating under bipolar mode with SD24UNI = 0.

If the measured value is to be used in the unipolar mode for offset correction it needs to be multiplied by two.

Figure 27-6 shows the relationship between the full-scale input voltage range from $-V_{FSR}$ to $+V_{FSR}$ and the conversion result. The data formats are illustrated.

![Figure 27-6. Input Voltage vs Digital Output](image-url)
27.2.9 Conversion Modes

The SD24_A module can be configured for four modes of operation, listed in Table 27-4. The SD24SNGL and SD24GRP bits for each channel select the conversion mode.

<table>
<thead>
<tr>
<th>SD24SNGL</th>
<th>SD24GRP</th>
<th>Mode</th>
<th>Operation</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>0</td>
<td>Single channel, Single conversion</td>
<td>A single channel is converted once.</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>Single channel, Continuous conversion</td>
<td>A single channel is converted continuously.</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>Group of channels, Single conversion</td>
<td>A group of channels is converted once.</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>Group of channels, Continuous conversion</td>
<td>A group of channels is converted continuously.</td>
</tr>
</tbody>
</table>

(1) A channel is grouped and is the master channel of the group when SD24GRP = 0 if SD24GRP for the prior channel(s) is set.

27.2.9.1 Single Channel, Single Conversion

Setting the SD24SC bit of a channel initiates one conversion on that channel when SD24SNGL = 1 and it is not grouped with any other channels. The SD24SC bit will automatically be cleared after conversion completion.

Clearing SD24SC before the conversion is completed immediately stops conversion of the selected channel, the channel is powered down and the corresponding digital filter is turned off. The value in SD24MEMx can change when SD24SC is cleared. It is recommended that the conversion data in SD24MEMx be read prior to clearing SD24SC to avoid reading an invalid result.

27.2.9.2 Single Channel, Continuous Conversion

When SD24SNGL = 0 continuous conversion mode is selected. Conversion of the selected channel will begin when SD24SC is set and continue until the SD24SC bit is cleared by software when the channel is not grouped with any other channel.

Clearing SD24SC immediately stops conversion of the selected channel, the channel is powered down and the corresponding digital filter is turned off. The value in SD24MEMx can change when SD24SC is cleared. It is recommended that the conversion data in SD24MEMx be read prior to clearing SD24SC to avoid reading an invalid result.

Figure 27-7 shows single channel operation for single conversion mode and continuous conversion mode.

Figure 27-7. Single Channel Operation - Example
27.2.9.3 Group of Channels, Single Conversion

Consecutive SD24_A channels can be grouped together with the SD24GRP bit to synchronize conversions. Setting SD24GRP for a channel groups that channel with the next channel in the module. For example, setting SD24GRP for channel 0 groups that channel with channel 1. In this case, channel 1 is the master channel, enabling and disabling conversion of all channels in the group with its SD24SC bit. The SD24GRP bit of the master channel is always 0. The SD24GRP bit of last channel in SD24_A has no function and is always 0.

When SD24SNGL = 1 for a channel in a group, single conversion mode is selected. A single conversion of that channel will occur synchronously when the master channel SD24SC bit is set. The SD24SC bit of all channels in the group will automatically be set and cleared by SD24SC of the master channel. SD24SC for each channel can also be cleared in software independently.

Clearing SD24SC of the master channel before the conversions are completed immediately stops conversions of all channels in the group, the channels are powered down and the corresponding digital filters are turned off. Values in SD24MEMx can change when SD24SC is cleared. It is recommended that the conversion data in SD24MEMx be read prior to clearing SD24SC to avoid reading an invalid result.

27.2.9.4 Group of Channels, Continuous Conversion

When SD24SNGL = 0 for a channel in a group, continuous conversion mode is selected. Continuous conversion of that channel will occur synchronously when the master channel SD24SC bit is set. SD24SC bits for all grouped channels will be automatically set and cleared with the master channel's SD24SC bit. SD24SC for each channel in the group can also be cleared in software independently.

When SD24SC of a grouped channel is set by software independently of the master, conversion of that channel will automatically synchronize to conversions of the master channel. This ensures that conversions for grouped channels are always synchronous to the master.

Clearing SD24SC of the master channel immediately stops conversions of all channels in the group the channels are powered down and the corresponding digital filters are turned off. Values in SD24MEMx can change when SD24SC is cleared. It is recommended that the conversion data in SD24MEMx be read prior to clearing SD24SC to avoid reading an invalid result.

Figure 27-8 shows grouped channel operation for three SD24_A channels. Channel 0 is configured for single conversion mode, SD24SNGL = 1, and channels 1 and 2 are in continuous conversion mode, SD24SNGL = 0.

Channel one, the last channel in the group, is the master channel. Conversions of all channels in the group occur synchronously to the master channel regardless of when each SD24SC bit is set using software.

Figure 27-8. Grouped Channel Operation - Example
27.2.10 Conversion Operation Using Preload

When multiple channels are grouped the SD24PREx registers can be used to delay the conversion time frame for each channel. Using SD24PREx, the decimation time of the digital filter is increased by the specified number of \( f_M \) clock cycles and can range from 0 to 255. Figure 27-9 shows an example using SD24PREx.

\[
\begin{array}{c}
\text{Conversion} \\
\text{32} \\
\text{SD24OSRx = 32} \\
\text{Load SD24PREx} \\
\text{SD24PREx = 8} \\
\text{Preload applied} \\
\text{Delayed Conversion} \\
\text{40} \\
\text{Conversion} \\
\text{32} \\
\text{Time} \\
\end{array}
\]

**Figure 27-9. Conversion Delay Using Preload - Example**

The SD24PREx delay is applied to the beginning of the next conversion cycle after being written. The delay is used on the first conversion after SD24SC is set and on the conversion cycle following each write to SD24PREx. Following conversions are not delayed. After modifying SD24PREx, the next write to SD24PREx should not occur until the next conversion cycle is completed, otherwise the conversion results may be incorrect.

The accuracy of the result for the delayed conversion cycle using SD24PREx is dependent on the length of the delay and the frequency of the analog signal being sampled. For example, when measuring a DC signal, SD24PREx delay has no effect on the conversion result regardless of the duration. The user must determine when the delayed conversion result is useful in their application.

Figure 27-10 shows the operation of grouped channels 0 and 1. The preload register of channel 1 is loaded with zero resulting in immediate conversion whereas the conversion cycle of channel 0 is delayed by setting SD24PRE0 = 8. The first channel 0 conversion uses SD24PREx = 8, shifting all subsequent conversions by eight \( f_M \) clock cycles.

\[
\begin{array}{c}
\text{Conversion} \\
\text{32} \\
\text{SD24OSRx = 32} \\
\text{SD24PRE0 = 8} \\
\text{1st Sample Ch0} \\
\text{32} \\
\text{Conversion} \\
\text{32} \\
\text{Conversion} \\
\text{32} \\
\text{Conversion} \\
\text{Conversion} \\
\end{array}
\]

**Figure 27-10. Start of Conversion Using Preload - Example**

When channels are grouped, care must be taken when a channel or channels operate in single conversion mode or are disabled in software while the master channel remains active. Each time channels in the group are re-enabled and re-synchronize with the master channel, the preload delay for that channel will be reintroduced. Figure 27-11 shows the re-synchronization and preload delays for channels in a group. It is recommended that SD24PREx = 0 for the master channel to maintain a consistent delay between the master and remaining channels in the group when they are re-enabled.
27.2.11 Using the Integrated Temperature Sensor

To use the on-chip temperature sensor, the user selects the analog input pair SD24INCHx = 110 and sets SD24REFON = 1. Any other configuration is done as if an external analog input pair was selected, including SD24INTDLYx and SD24GAINx settings. Because the internal reference must be on to use the temperature sensor, it is not possible to use an external reference for the conversion of the temperature sensor voltage. Also, the internal reference will be in contention with any used external reference. In this case, the SD24VMIDON bit may be set to minimize the affects of the contention on the conversion.

The typical temperature sensor transfer function is shown in Figure 27-12. When switching inputs of an SD24_A channel to the temperature sensor, adequate delay must be provided using SD24INTDLYx to allow the digital filter to settle and assure that conversion results are valid. The temperature sensor offset error can be large, and may need to be calibrated for most applications. See device-specific data sheet for temperature sensor parameters.

\[
V_{Sensor,typ} = TC_{Sensor}(273 + T[°C]) + V_{Offset, sensor} [mV]
\]
27.2.12 Interrupt Handling

The SD24_A has 2 interrupt sources for each ADC channel:

- SD24IFG
- SD24OVIFG

The SD24IFG bits are set when their corresponding SD24MEMx memory register is written with a conversion result. An interrupt request is generated if the corresponding SD24IE bit and the GIE bit are set. The SD24_A overflow condition occurs when a conversion result is written to any SD24MEMx location before the previous conversion result was read.

27.2.12.1 SD24IV, Interrupt Vector Generator

All SD24_A interrupt sources are prioritized and combined to source a single interrupt vector. SD24IV is used to determine which enabled SD24_A interrupt source requested an interrupt. The highest priority SD24_A interrupt request that is enabled generates a number in the SD24IV register (see register description). This number can be evaluated or added to the program counter to automatically enter the appropriate software routine. Disabled SD24_A interrupts do not affect the SD24IV value.

Any access, read or write, of the SD24IV register has no effect on the SD24OVIFG or SD24IFG flags. The SD24IFG flags are reset by reading the associated SD24MEMx register or by clearing the flags in software. SD24OVIFG bits can only be reset with software.

If another interrupt is pending after servicing of an interrupt, another interrupt is generated. For example, if the SD24OVIFG and one or more SD24IFG interrupts are pending when the interrupt service routine accesses the SD24IV register, the SD24OVIFG interrupt condition is serviced first and the corresponding flag(s) must be cleared in software. After the RETi instruction of the interrupt service routine is executed, the highest priority SD24IFG pending generates another interrupt request.

27.2.12.2 Interrupt Delay Operation

The SD24INTDLYx bits control the timing for the first interrupt service request for the corresponding channel. This feature delays the interrupt request for a completed conversion by up to four conversion cycles allowing the digital filter to settle prior to generating an interrupt request. The delay is applied each time the SD24SC bit is set or when the SD24GAINx or SD24INCHx bits for the channel are modified. SD24INTDLYx disables overflow interrupt generation for the channel for the selected number of delay cycles. Interrupt requests for the delayed conversions are not generated during the delay.
27.2.12.3 SD24_A Interrupt Handling Software Example

The following software example shows the recommended use of SD24IV and the handling overhead. The
SD24IV value is added to the PC to automatically jump to the appropriate routine.

The numbers at the right margin show the necessary CPU cycles for each instruction. The software overhead
for different interrupt sources includes interrupt latency and return-from-interrupt cycles, but not the task handling
itself. The latencies are:

- SD240VIFG, CH0 SD24IFG, CH1 SD24IFG: 16 cycles
- CH2 SD24IFG: 14 cycles

The interrupt handler for channel 2 SD24IFG shows a way to check immediately if a higher prioritized interrupt
occurred during the processing of the ISR. This saves nine cycles if another SD24_A interrupt is pending.

; Interrupt handler for SD24_A.
INT_SD24                ; Enter Interrupt Service Routine  6
  ADD      &SD24IV,PC  ; Add offset to PC                   3
  RETI                 ; Vector 0: No interrupt             5
  JMP      ADOV        ; Vector 2: ADC overflow             2
  JMP      ADM0        ; Vector 4: CH_0 SD24IFG             2
  JMP      ADM1        ; Vector 6: CH_1 SD24IFG             2
;
; Handler for CH_2 SD24IFG starts here. No JMP required.
ADM2   MOV   &SD24MEM2,xxx  ; Move result, flag is reset
          ; Other instruction needed?
          JMP   INT_SD24       ; Check other int pending  2
;
; Remaining Handlers
ADM1   MOV   &SD24MEM1,xxx  ; Move result, flag is reset
          ; Other instruction needed?
          RETI               ; Return                         5
;
ADM0   MOV   &SD24MEM0,xxx  ; Move result, flag is reset
          RETI               ; Return                         5
;
ADOV ...                ; Handle SD24MEMx overflow
          RETI               ; Return                         5
## 27.3 SD24_A Registers

Table 27-5 lists the memory-mapped registers for the SD24_A.

<table>
<thead>
<tr>
<th>Address</th>
<th>Acronym</th>
<th>Register Name</th>
<th>Type</th>
<th>Reset</th>
<th>Section</th>
</tr>
</thead>
<tbody>
<tr>
<td>100h</td>
<td>SD24CTL</td>
<td>SD24_A Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.1</td>
</tr>
<tr>
<td>110h</td>
<td>SD24IV</td>
<td>SD24_A Interrupt Vector</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.2</td>
</tr>
<tr>
<td>B7h</td>
<td>SD24AE</td>
<td>SD24_A Analog Enable&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.3</td>
</tr>
<tr>
<td>102h</td>
<td>SD24CCTL0</td>
<td>SD24_A Channel 0 Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.4</td>
</tr>
<tr>
<td>112h</td>
<td>SD24MEM0</td>
<td>SD24_A Channel 0 Conversion Memory</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.5</td>
</tr>
<tr>
<td>B0h</td>
<td>SD24INCTL0</td>
<td>SD24_A Channel 0 Input Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.6</td>
</tr>
<tr>
<td>B8h</td>
<td>SD24PRE0</td>
<td>SD24_A Channel 0 Preload</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.7</td>
</tr>
<tr>
<td>104h</td>
<td>SD24CCTL1</td>
<td>SD24_A Channel 1 Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.4</td>
</tr>
<tr>
<td>114h</td>
<td>SD24MEM1</td>
<td>SD24_A Channel 1 Conversion Memory</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.5</td>
</tr>
<tr>
<td>B1h</td>
<td>SD24INCTL1</td>
<td>SD24_A Channel 1 Input Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.6</td>
</tr>
<tr>
<td>B9h</td>
<td>SD24PRE1</td>
<td>SD24_A Channel 1 Preload</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.7</td>
</tr>
<tr>
<td>106h</td>
<td>SD24CCTL2</td>
<td>SD24_A Channel 2 Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.4</td>
</tr>
<tr>
<td>116h</td>
<td>SD24MEM2</td>
<td>SD24_A Channel 2 Conversion Memory</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.5</td>
</tr>
<tr>
<td>B2h</td>
<td>SD24INCTL2</td>
<td>SD24_A Channel 2 Input Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.6</td>
</tr>
<tr>
<td>BAh</td>
<td>SD24PRE2</td>
<td>SD24_A Channel 2 Preload</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.7</td>
</tr>
<tr>
<td>108h</td>
<td>SD24CCTL3</td>
<td>SD24_A Channel 3 Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.4</td>
</tr>
<tr>
<td>118h</td>
<td>SD24MEM3</td>
<td>SD24_A Channel 3 Conversion Memory</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.5</td>
</tr>
<tr>
<td>B3h</td>
<td>SD24INCTL3</td>
<td>SD24_A Channel 3 Input Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.6</td>
</tr>
<tr>
<td>B8h</td>
<td>SD24PRE3</td>
<td>SD24_A Channel 3 Preload</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.7</td>
</tr>
<tr>
<td>10Ah</td>
<td>SD24CCTL4</td>
<td>SD24_A Channel 4 Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.4</td>
</tr>
<tr>
<td>11Ah</td>
<td>SD24MEM4</td>
<td>SD24_A Channel 4 Conversion Memory</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.5</td>
</tr>
<tr>
<td>B4h</td>
<td>SD24INCTL4</td>
<td>SD24_A Channel 4 Input Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.6</td>
</tr>
<tr>
<td>BCh</td>
<td>SD24PRE4</td>
<td>SD24_A Channel 4 Preload</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.7</td>
</tr>
<tr>
<td>10Ch</td>
<td>SD24CCTL5</td>
<td>SD24_A Channel 5 Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.4</td>
</tr>
<tr>
<td>11Ch</td>
<td>SD24MEM5</td>
<td>SD24_A Channel 5 Conversion Memory</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.5</td>
</tr>
<tr>
<td>B5h</td>
<td>SD24INCTL5</td>
<td>SD24_A Channel 5 Input Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.6</td>
</tr>
<tr>
<td>B0h</td>
<td>SD24PRE5</td>
<td>SD24_A Channel 5 Preload</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.7</td>
</tr>
<tr>
<td>10Eh</td>
<td>SD24CCTL6</td>
<td>SD24_A Channel 6 Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.4</td>
</tr>
<tr>
<td>11Eh</td>
<td>SD24MEM6</td>
<td>SD24_A Channel 6 Conversion Memory</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.5</td>
</tr>
<tr>
<td>B6h</td>
<td>SD24INCTL6</td>
<td>SD24_A Channel 6 Input Control</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.6</td>
</tr>
<tr>
<td>B6h</td>
<td>SD24PRE6</td>
<td>SD24_A Channel 6 Preload</td>
<td>Read/write</td>
<td>00h with PUC</td>
<td>Section 27.3.7</td>
</tr>
</tbody>
</table>

<sup>(1)</sup> Not implemented on all devices; see the device-specific data sheet.
27.3.1 SD24CTL Register

SD24_A Control Register

SD24CTL is shown in Figure 27-13 and described in Table 27-6.

Return to Table 27-5.

Figure 27-13. SD24CTL Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-12</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td>SD24_A clock divider</td>
</tr>
<tr>
<td>11-9</td>
<td>SD24XDIVx</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A clock divider</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = /1</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = /3</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = /16</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = /48</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1xxb = Reserved</td>
</tr>
<tr>
<td>8</td>
<td>SD24LP</td>
<td>R/W</td>
<td>0h</td>
<td>Low-power mode. This bit selects a reduced-speed reduced-power mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Low-power mode is disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Low-power mode is enabled. The maximum clock frequency for the SD24_A is reduced</td>
</tr>
<tr>
<td>7-6</td>
<td>SD24DIVx</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A clock divider</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = /1</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = /2</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = /4</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = /8</td>
</tr>
<tr>
<td>5-4</td>
<td>SD24SSELx</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A clock source select</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00b = MCLK</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>01b = SMCLK</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10b = ACLK</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>11b = External TACLK</td>
</tr>
<tr>
<td>3</td>
<td>SD24VMIDON</td>
<td>R/W</td>
<td>0h</td>
<td>VMID buffer on</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Off</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = On</td>
</tr>
<tr>
<td>2</td>
<td>SD24REFON</td>
<td>R/W</td>
<td>0h</td>
<td>Reference generator on</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Reference off</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Reference on</td>
</tr>
<tr>
<td>1</td>
<td>SD24OVIE</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A overflow interrupt enable. The GIE bit must also be set to enable the interrupt.</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0b = Overflow interrupt disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1b = Overflow interrupt enabled</td>
</tr>
<tr>
<td>0</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td></td>
</tr>
</tbody>
</table>
27.3.2 SD24IV Register

SD24_A Interrupt Vector Register

SD24IV is shown in Figure 27-14 and described in Table 27-7.

Return to Table 27-5.

Figure 27-14. SD24IV Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>SD24IVx</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>R</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit</th>
<th>SD24IVx</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td></td>
</tr>
</tbody>
</table>

Table 27-7. SD24IV Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>SD24IVx</td>
<td>R</td>
<td>0h</td>
<td>SD24_A interrupt vector value. See Table 27-8</td>
</tr>
</tbody>
</table>

Table 27-8. SD24_A Interrupt Vectors

<table>
<thead>
<tr>
<th>SD24IV Contents</th>
<th>Interrupt Source</th>
<th>Interrupt Flag</th>
<th>Interrupt Priority</th>
</tr>
</thead>
<tbody>
<tr>
<td>000h</td>
<td>No interrupt pending</td>
<td>–</td>
<td>–</td>
</tr>
<tr>
<td>002h</td>
<td>SD24MEMx overflow</td>
<td>SD24CCTLx SD24OVIFG(1)</td>
<td>Highest</td>
</tr>
<tr>
<td>004h</td>
<td>SD24_A Channel 0 Interrupt</td>
<td>SD24CCTL0 SD24IFG</td>
<td></td>
</tr>
<tr>
<td>006h</td>
<td>SD24_A Channel 1 Interrupt</td>
<td>SD24CCTL1 SD24IFG</td>
<td></td>
</tr>
<tr>
<td>008h</td>
<td>SD24_A Channel 2 Interrupt</td>
<td>SD24CCTL2 SD24IFG</td>
<td></td>
</tr>
<tr>
<td>00Ah</td>
<td>SD24_A Channel 3 Interrupt</td>
<td>SD24CCTL3 SD24IFG</td>
<td></td>
</tr>
<tr>
<td>00Ch</td>
<td>SD24_A Channel 4 Interrupt</td>
<td>SD24CCTL4 SD24IFG</td>
<td></td>
</tr>
<tr>
<td>00Eh</td>
<td>SD24_A Channel 5 Interrupt</td>
<td>SD24CCTL5 SD24IFG</td>
<td></td>
</tr>
<tr>
<td>010h</td>
<td>SD24_A Channel 6 Interrupt</td>
<td>SD24CCTL6 SD24IFG</td>
<td>Lowest</td>
</tr>
</tbody>
</table>

(1) When an SD24_A overflow occurs, the user must check all SD24CCTLx SD24OVIFG flags to determine which channel overflowed.
27.3.3 SD24AE Register

SD24_A Analog Enable Register

SD24AE is shown in Figure 27-15 and described in Table 27-9.

Return to Table 27-5.

Not implemented on all devices; see the device-specific data sheet.

Figure 27-15. SD24AE Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>SD24AE7</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A analog enable 7 0b = External input disabled. Negative inputs are internally connected to VSS. 1b = External input enabled</td>
</tr>
<tr>
<td>6</td>
<td>SD24AE6</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A analog enable 6 0b = External input disabled. Negative inputs are internally connected to VSS. 1b = External input enabled</td>
</tr>
<tr>
<td>5</td>
<td>SD24AE5</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A analog enable 5 0b = External input disabled. Negative inputs are internally connected to VSS. 1b = External input enabled</td>
</tr>
<tr>
<td>4</td>
<td>SD24AE4</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A analog enable 4 0b = External input disabled. Negative inputs are internally connected to VSS. 1b = External input enabled</td>
</tr>
<tr>
<td>3</td>
<td>SD24AE3</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A analog enable 3 0b = External input disabled. Negative inputs are internally connected to VSS. 1b = External input enabled</td>
</tr>
<tr>
<td>2</td>
<td>SD24AE2</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A analog enable 2 0b = External input disabled. Negative inputs are internally connected to VSS. 1b = External input enabled</td>
</tr>
<tr>
<td>1</td>
<td>SD24AE1</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A analog enable 1 0b = External input disabled. Negative inputs are internally connected to VSS. 1b = External input enabled</td>
</tr>
<tr>
<td>0</td>
<td>SD24AE0</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A analog enable 0 0b = External input disabled. Negative inputs are internally connected to VSS. 1b = External input enabled</td>
</tr>
</tbody>
</table>
### 27.3.4 SD24CCTLx Register

**SD24_A Channel x Control Register**

SD24CCTLx is shown in Figure 27-16 and described in Table 27-10.

Return to Table 27-5.

#### Figure 27-16. SD24CCTLx Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td>High-impedance input buffer mode. Not implemented on all devices (see the device-specific data sheet). Reserved with r-0 access if high-impedance buffer not implemented.</td>
</tr>
<tr>
<td>14-13</td>
<td>SD24BUFx</td>
<td>R/W</td>
<td>0h</td>
<td>Unipolar mode select 0b = Bipolar mode 1b = Unipolar mode</td>
</tr>
<tr>
<td>12</td>
<td>SD24UNI</td>
<td>R/W</td>
<td>0h</td>
<td>Extended oversampling ratio. This bit, along with the SD24OSRx bits, select the oversampling ratio. See Table 27-10 bit description for settings.</td>
</tr>
<tr>
<td>11</td>
<td>SD24XOSR</td>
<td>R/W</td>
<td>0h</td>
<td>Single conversion mode select 0b = Continuous conversion mode 1b = Single conversion mode</td>
</tr>
<tr>
<td>9-8</td>
<td>SD24OSRx</td>
<td>R/W</td>
<td>0h</td>
<td>Oversampling ratio When SD24XOSR = 0 00b = 256 01b = 128 10b = 64 11b = 32 When SD24XOSR = 1 00b = 512 01b = 1024 10b = Reserved 11b = Reserved</td>
</tr>
</tbody>
</table>

#### Table 27-10. SD24CCTL0 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15</td>
<td>Reserved</td>
<td>R</td>
<td>0h</td>
<td>High-impedance input buffer mode. Not implemented on all devices (see the device-specific data sheet). Reserved with r-0 access if high-impedance buffer not implemented.</td>
</tr>
<tr>
<td>14-13</td>
<td>SD24BUFx</td>
<td>R/W</td>
<td>0h</td>
<td>Unipolar mode select 0b = Bipolar mode 1b = Unipolar mode</td>
</tr>
<tr>
<td>12</td>
<td>SD24UNI</td>
<td>R/W</td>
<td>0h</td>
<td>Extended oversampling ratio. This bit, along with the SD24OSRx bits, select the oversampling ratio. See Table 27-10 bit description for settings.</td>
</tr>
<tr>
<td>11</td>
<td>SD24XOSR</td>
<td>R/W</td>
<td>0h</td>
<td>Single conversion mode select 0b = Continuous conversion mode 1b = Single conversion mode</td>
</tr>
<tr>
<td>9-8</td>
<td>SD24OSRx</td>
<td>R/W</td>
<td>0h</td>
<td>Oversampling ratio When SD24XOSR = 0 00b = 256 01b = 128 10b = 64 11b = 32 When SD24XOSR = 1 00b = 512 01b = 1024 10b = Reserved 11b = Reserved</td>
</tr>
<tr>
<td>Bit</td>
<td>Field</td>
<td>Type</td>
<td>Reset</td>
<td>Description</td>
</tr>
<tr>
<td>-----</td>
<td>------------------</td>
<td>------</td>
<td>-------</td>
<td>-----------------------------------------------------------------------------</td>
</tr>
<tr>
<td>7</td>
<td>SD24LSBTOG</td>
<td>R/W</td>
<td>0h</td>
<td>LSB toggle. This bit, when set, causes SD24LSBACC to toggle each time the SD24MEMx register is read. 0b = SD24LSBACC does not toggle with each SD24MEMx read 1b = SD24LSBACC toggles with each SD24MEMx read</td>
</tr>
<tr>
<td>6</td>
<td>SD24LSBACC</td>
<td>R/W</td>
<td>0h</td>
<td>LSB access. This bit allows access to the upper or lower 16-bits of the SD24_A conversion result. 0b = SD24MEMx contains the most significant 16-bits of the conversion. 1b = SD24MEMx contains the least significant 16-bits of the conversion.</td>
</tr>
<tr>
<td>5</td>
<td>SD24OVIFG</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A overflow interrupt flag 0b = No overflow interrupt pending 1b = Overflow interrupt pending</td>
</tr>
<tr>
<td>4</td>
<td>SD24DF</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A data format 0b = Offset binary 1b = 2s complement</td>
</tr>
<tr>
<td>3</td>
<td>SD24IE</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A interrupt enable 0b = Disabled 1b = Enabled</td>
</tr>
<tr>
<td>2</td>
<td>SD24IFG</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A interrupt flag. SD24IFG is set when new conversion results are available. SD24IFG is automatically reset when the corresponding SD24MEMx register is read, or may be cleared with software. 0b = No interrupt pending 1b = Interrupt pending</td>
</tr>
<tr>
<td>1</td>
<td>SD24SC</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A start conversion 0b = No conversation start 1b = Start conversation</td>
</tr>
<tr>
<td>0</td>
<td>SD24GRP</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A group. Groups SD24_A channel with next higher channel. Not used for the last channel. 0b = Not grouped 1b = Grouped</td>
</tr>
</tbody>
</table>
27.3.5 SD24MEMx Register

SD24_A Channel x Conversion Memory Register

SD24MEMx is shown in Figure 27-17 and described in Table 27-11.

Return to Table 27-5.

Figure 27-17. SD24MEMx Register

<table>
<thead>
<tr>
<th></th>
<th>15</th>
<th>14</th>
<th>13</th>
<th>12</th>
<th>11</th>
<th>10</th>
<th>9</th>
<th>8</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Conversion_Results</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td></td>
</tr>
<tr>
<td>7</td>
<td>6</td>
<td>5</td>
<td>4</td>
<td>3</td>
<td>2</td>
<td>1</td>
<td>0</td>
<td></td>
</tr>
<tr>
<td>Conversion_Results</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td>r-0</td>
<td></td>
</tr>
</tbody>
</table>

Table 27-11. SD24MEM0 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>15-0</td>
<td>Conversion_Results</td>
<td>R</td>
<td>0h</td>
<td>Conversion results. The SD24MEMx register holds the upper or lower 16-bits of the digital filter output, depending on the SD24LSBACC bit.</td>
</tr>
</tbody>
</table>
27.3.6 SD24INCTLx Register

SD24_A Channel x Input Control Register

SD24INCTLx is shown in Figure 27-18 and described in Table 27-12.

Return to Table 27-5.

Figure 27-18. SD24INCTLx Register

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td>SD24INTDLYx</td>
<td>R/W</td>
<td>0h</td>
<td>Interrupt delay generation after conversion start. These bits select the delay for the first interrupt after conversion start. 00b = Fourth sample causes interrupt 01b = Third sample causes interrupt 10b = Second sample causes interrupt 11b = First sample causes interrupt</td>
</tr>
<tr>
<td>5-3</td>
<td>SD24GAINx</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A preamplifier gain 000b = ×1 001b = ×2 010b = ×4 011b = ×8 100b = ×16 101b = ×32 110b = Reserved 111b = Reserved</td>
</tr>
<tr>
<td>2-0</td>
<td>SD24INCHx</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A channel differential pair input. The available selections are device dependent. See the device-specific data sheet. 000b = Ax.0 001b = Ax.1. Not available on all devices (see device-specific data sheet). 010b = Ax.2. Not available on all devices (see device-specific data sheet). 011b = Ax.3. Not available on all devices (see device-specific data sheet). 100b = Ax.4. Not available on all devices (see device-specific data sheet). 101b = (AVCC − AVSS) / 11 110b = Temperature sensor 111b = Short for PGA offset measurement</td>
</tr>
</tbody>
</table>
27.3.7 SD24PREx Register

SD24_A Channel x Preload Register

SD24PREx is shown in Figure 27-19 and described in Table 27-13.

Return to Table 27-5.

Table 27-13. SD24PRE0 Register Field Descriptions

<table>
<thead>
<tr>
<th>Bit</th>
<th>Field</th>
<th>Type</th>
<th>Reset</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>Preload_Value</td>
<td>R/W</td>
<td>0h</td>
<td>SD24_A digital filter preload value</td>
</tr>
</tbody>
</table>
This chapter describes the Embedded Emulation Module (EEM) that is implemented in all MSP430 flash devices.
28.1 EEM Introduction

Every MSP430 flash-based microcontroller implements an embedded emulation module (EEM). It is accessed and controlled through JTAG. Each implementation is device dependent and is described in section 1.3 EEM Configurations and the device-specific data sheet.

In general, the following features are available:

• Non-intrusive code execution with real-time breakpoint control
• Single step, step into and step over functionality
• Full support of all low-power modes
• Support for all system frequencies, for all clock sources
• Up to eight (device dependent) hardware triggers/breakpoints on memory address bus (MAB) or memory data bus (MDB)
• Up to two (device dependent) hardware triggers/breakpoints on CPU register write accesses
• MAB, MDB, and CPU register access triggers can be combined to form up to eight (device dependent) complex triggers/breakpoints
• Trigger sequencing (device dependent)
• Storage of internal bus and control signals using an integrated trace buffer (device dependent)
• Clock control for timers, communication peripherals, and other modules on a global device level or on a per-module basis during an emulation stop

Figure 28-1 shows a simplified block diagram of the largest currently available 2xx EEM implementation.

For more details on how the features of the EEM can be used together with the IAR Embedded Workbench™ debugger see the application report Advanced Debugging Using the Enhanced Emulation Module (SLAA263) at www.msp430.com. Code Composer Essentials (CCE) and most other debuggers supporting MSP430 have the same or a similar feature set. For details see the user’s guide of the applicable debugger.
Figure 28-1. Large Implementation of the Embedded Emulation Module (EEM)
28.2 EEM Building Blocks

28.2.1 Triggers

The event control in the EEM of the MSP430 system consists of triggers, which are internal signals indicating that a certain event has happened. These triggers may be used as simple breakpoints, but it is also possible to combine two or more triggers to allow detection of complex events and trigger various reactions besides stopping the CPU.

In general, the triggers can be used to control the following functional blocks of the EEM:

- Breakpoints (CPU stop)
- State storage
- Sequencer

There are two different types of triggers, the memory trigger and the CPU register write trigger.

Each memory trigger block can be independently selected to compare either the MAB or the MDB with a given value. Depending on the implemented EEM the comparison can be =, ≠, ≥, or ≤. The comparison can also be limited to certain bits with the use of a mask. The mask is either bit-wise or byte-wise, depending upon the device. In addition to selecting the bus and the comparison, the condition under which the trigger is active can be selected. The conditions include read access, write access, DMA access, and instruction fetch.

Each CPU register write trigger block can be independently selected to compare what is written into a selected register with a given value. The observed register can be selected for each trigger independently. The comparison can be =, ≠, ≥, or ≤. The comparison can also be limited to certain bits with the use of a bit mask.

Both types of triggers can be combined to form more complex triggers. For example, a complex trigger can signal when a particular value is written into a user-specified address.

28.2.2 Trigger Sequencer

The trigger sequencer allows the definition of a certain sequence of trigger signals before an event is accepted for a break or state storage event. Within the trigger sequencer, it is possible to use the following features:

- Four states (State 0 to State 3)
- Two transitions per state to any other state
- Reset trigger that resets the sequencer to State 0.

The Trigger sequencer always starts at State 0 and must execute to State 3 to generate an action. If State 1 or State 2 are not required, they can be bypassed.

28.2.3 State Storage (Internal Trace Buffer)

The state storage function uses a built-in buffer to store MAB, MDB, and CPU control signal information (that is, read, write, or instruction fetch) in a non-intrusive manner. The built-in buffer can hold up to eight entries. The flexible configuration allows the user to record the information of interest very efficiently.

28.2.4 Clock Control

The EEM provides device dependent flexible clock control. This is useful in applications where a running clock is needed for peripherals after the CPU is stopped (for example, to allow a UART module to complete its transfer of a character or to allow a timer to continue generating a PWM signal).

The clock control is flexible and supports both modules that need a running clock and modules that must be stopped when the CPU is stopped due to a breakpoint.
28.3 EEM Configurations

Table 28-1 gives an overview of the EEM configurations in the MSP430 2xx family. The implemented configuration is device dependent - see the device data sheet.

<table>
<thead>
<tr>
<th>Table 28-1. 2xx EEM Configurations</th>
</tr>
</thead>
<tbody>
<tr>
<td>Feature</td>
</tr>
<tr>
<td>Memory Bus Triggers</td>
</tr>
<tr>
<td></td>
</tr>
<tr>
<td>Memory Bus Trigger Mask for</td>
</tr>
<tr>
<td>1) Low byte</td>
</tr>
<tr>
<td>2) High byte</td>
</tr>
<tr>
<td></td>
</tr>
<tr>
<td>Memory Bus Trigger Mask for</td>
</tr>
<tr>
<td>1) Low byte</td>
</tr>
<tr>
<td>2) High byte</td>
</tr>
<tr>
<td>CPU Register-Write Triggers</td>
</tr>
<tr>
<td></td>
</tr>
<tr>
<td>Combination Triggers</td>
</tr>
<tr>
<td></td>
</tr>
<tr>
<td>Sequeencer</td>
</tr>
<tr>
<td></td>
</tr>
<tr>
<td>State Storage</td>
</tr>
<tr>
<td></td>
</tr>
</tbody>
</table>

In general the following features can be found on any 2xx device:

- At least two MAB/MDB triggers supporting:
  - Distinction between CPU, DMA, read, and write accesses
  - =, ≠, ≥, or ≤ comparison (in XS only =, ≠)
- At least two trigger Combination registers
- Hardware breakpoints using the CPU Stop reaction
- Clock control with individual control of module clocks (in some XS configurations the control of module clocks is hardwired)
## Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

### Changes from July 1, 2013 to August 25, 2022

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<td>78</td>
<td>Corrected the descriptions of the N and Z status bits in Section 3.4.6.14, CMP.</td>
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<td>180</td>
<td>Corrected the descriptions of the N and Z status bits in Section 4.6.2.14, CMP.</td>
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<tr>
<td>276</td>
<td>Added MSP430G2x55 to &quot;Device-Specific Clock Variations&quot; note.</td>
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<td>279</td>
<td>Corrected name of SELM1 bit (changed from XSELM1) in Figure 5-3, Off Signals for the LFXT1 Oscillator.</td>
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<tr>
<td>280</td>
<td>Added MCLK_request and SMCLK_request to the first paragraph of Section 5.2.4, XT2 Oscillator.</td>
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<tr>
<td>280</td>
<td>Corrected name of SELM1 bit (changed from XSELM1) in Figure 5-4, Off Signals for Oscillator XT2.</td>
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<td>281</td>
<td>Corrected name of SELM1 bit (changed from XSELM1) in Figure 5-5, On and Off Control of DCO.</td>
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<tr>
<td>318</td>
<td>Updated description of segment and block sizes in Section 7.2, Flash Memory Segmentation.</td>
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<td>318</td>
<td>Corrected address on main memory boundary (changed 0x0F000 to 0x08000) in Figure 7-2, Flash Memory Segments, 32KB Example.</td>
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<tr>
<td>332</td>
<td>Corrected the access type (read only) of the BUSY bit.</td>
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<tr>
<td>345</td>
<td>Added note to P2SEL reset value in Table 8-2, Digital I/O Registers.</td>
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<tr>
<td>347</td>
<td>Added description sections for Digital I/O registers (Section 8.3.1 through Section 8.3.9).</td>
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<tr>
<td>381</td>
<td>Changed TACCTLx to TACCRx in the comment in the second line of the code example in Section 12.2.4.1, Capture Initiated by Software.</td>
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<tr>
<td>411</td>
<td>Corrected formatting of IDx values in Table 13-6, TBCTL Register Field Descriptions.</td>
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<tr>
<td>421</td>
<td>Added the note that begins &quot;If USIIE = 1 when the USI module is in software reset mode...&quot; in Section 14.2.1, USI Initialization.</td>
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<tr>
<td>424</td>
<td>Added the paragraph that starts &quot;For examples of using the USI in ( ^{2} \text{C} ) mode...&quot;</td>
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<td>440</td>
<td>Added the note &quot;Reliable reception of IrDA signals&quot; in Section 15.3.5.2, IrDA Decoding.</td>
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<td>486</td>
<td>Updated Figure 17-3, ( ^{2} \text{C} ) Module Data Transfer, to clarify SDA transitions when SCL is low.</td>
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<tr>
<td>526</td>
<td>Corrected bit field name in Figure 18-20, UxRXBUF Register.</td>
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<tr>
<td>526</td>
<td>Corrected bit field name in Figure 18-21, UxTXBUF Register.</td>
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<tr>
<td>544</td>
<td>Corrected bit field name in Figure 19-19, UxRXBUF Register.</td>
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<tr>
<td>544</td>
<td>Corrected bit field name in Figure 19-20, UxTXBUF Register.</td>
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<tr>
<td>572</td>
<td>Added note that starts &quot;Changing the value of the CAIES bit...&quot;.</td>
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<tr>
<td>580</td>
<td>Added MSP430G2x44 and MSP430G2x55 to the list item that starts &quot;Up to eight external input channels...&quot;.</td>
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<tr>
<td>598</td>
<td>Added MSP430G2x44 and MSP430G2x55 to enums 1100b through 1111b in the INCHx bit description.</td>
</tr>
<tr>
<td>633</td>
<td>Changed TAG_ADC10_1 value to be device dependent.</td>
</tr>
<tr>
<td>654</td>
<td>Changed Figure 26-4.</td>
</tr>
<tr>
<td>676</td>
<td>Changed Figure 27-4.</td>
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</table>
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