

## **AN-1868 EMIRR Evaluation Boards for LMV861/LMV862**

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### **1 General Description**

To demonstrate the EMI robustness of the LMV861/LMV862 and to be able to measure the parameter EMIRR, two evaluation boards have been developed; one for each device. This document describes the evaluation boards and explains how to perform EMIRR measurements. Focus is on one of the input pins as those are most sensitive to EMI. Based on symmetry considerations, it can be expected that both inputs have the same EMIRR. For reasons of simplicity of the required schematic the measurement on the IN+ pin is selected. A detailed description on EMI and EMIRR for the other pins can be found in the *AN-1698 A Specification for EMI Hardened Operational Amplifiers Application Report* ([SNOA497](#)).

To identify EMI robust op amps, a parameter is defined that quantitatively describes the EMI performance. A quantitative measure enables the comparison and the ranking of op amps on their EMI robustness. The definition of the parameter EMIRR is given by:

$$\text{EMIRR}_{V_{\text{RF\_PEAK}}} = 20 \log \left( \frac{V_{\text{RF\_PEAK}}}{\Delta V_{\text{OS}}} \right)$$

where  $V_{\text{RF\_PEAK}}$  is the amplitude of the applied unmodulated RF signal (V) and  $\Delta V_{\text{OS}}$  is the resulting input-referred offset voltage shift (V).

### **2 Op Amp Configuration**

To have best defined RF levels on the pin under test, no op amp feedback elements should be in the RF signal path. Therefore, the op amp is connected in a unity-gain configuration. This yields the lowest level of RF filtering due to a feedback network. Schematics and layouts are included in this document.

### **3 Applying the RF Signal**

Care needs to be taken in applying the RF signal to the pin under test. Signals up to a few GHz will be used, so the whole RF signal path needs to match the characteristic impedance of the RF generator. This requires proper coaxial cabling from the generator to the test board. On the test board a 50Ω stripline needs to be used to bring the RF signal as close as possible to the pin under test. In this case the stripline can be connected directly from the connector to the IN+ pin. A 50Ω termination at the pin under test is also required. For symmetry reasons this is done with two 100Ω resistors in parallel, one on each side of the strip line. Setting up the test environment with a 50Ω resistor close to the LMV861/LMV862 ensures that the RF levels at the pin under test are well defined. This 50Ω resistor is also used to set the bias level of the IN+ pin to ground level. The DC measurements are taken at the output of the op amp. Since the op amp is in the unity gain configuration, the input referred offset voltage shift corresponds one-to-one to the measured output voltage shift.

### **4 Isolating the Other Pins**

When the pin under test is tested, the other pins need to be decoupled for RF signals. This ensures that the obtained offset voltage shift is dominantly a result of coupling the RF signal to the pin under test. For this decoupling standard SMD components can be used.

## 5 Layout Considerations

The layout of the evaluation board requires some attention. For decoupling the supply lines it is suggested that 10 nF capacitors be placed as close as possible to the op amp. For single supply, place a capacitor between  $V^+$  and  $V^-$ . For dual supplies, place one capacitor between  $V^+$  and the board ground, and a second capacitor between ground and  $V^-$ . On the LMV861 evaluation board, the decoupling of the negative pin  $V^-$  is implemented by a capacitor between  $V^+$  and  $V^-$ . This is done for easy routing and to keep connections to the pins short. Even with the LMV861/LMV862's inherent hardening against EMI, it is still recommended to keep the input traces short and as far as possible from RF sources. Then the RF signals entering the chip are as low as possible, and the remaining EMI can be, almost, completely eliminated in the chip by the EMI reducing features of the LMV861/LMV862.

## 6 Measurement Procedure

The measurement procedure is the same for all test circuits. To measure the input referred offset voltage shift needed for calculating the EMIRR, the following procedure can be used:

1. Measure  $V_{OUT}$  when the RF signal is off.
2. Measure  $V_{OUT}$  when the RF signal is on.
3. Translate measured  $V_{OUT}$  voltages to input referred voltages. Translation is one-to-one in this case since the gain is one in the IN+ test setup.
4. Subtract the two measured input referred voltages.
5. Verify if the offset shift is above the noise level of the op amp setup and the op amp is not saturated. If this is not the case choose another RF level and start the procedure again.
6. Calculate the EMIRR.
7. If needed, transform the results to an EMIRR based on a 100 mV<sub>p</sub> RF signal.

## 7 Measurement Results

To show the sensitivity of the IN+ pin two types of measurement results are presented using the LMV861 evaluation board:

- The EMIRR as a function of the *frequency* of the applied signal. The level of the signal is set to the standard level of 100 mV<sub>p</sub> (-20 dBV<sub>p</sub>).
- The EMIRR as a function of the *level* of the applied signal. The frequency is set to four typical values: 400 MHz, 900 MHz, 1.8 GHz, and 2.4 GHz.

### 7.1 EMIRR Vs. Frequency

[Figure 1](#) depicts the EMIRR versus frequency for the various temperatures. The measurement is performed with a fixed RF level of -20 dBV<sub>p</sub> and a varying RF signal frequency. The frequency range is 10 MHz to 4 GHz.

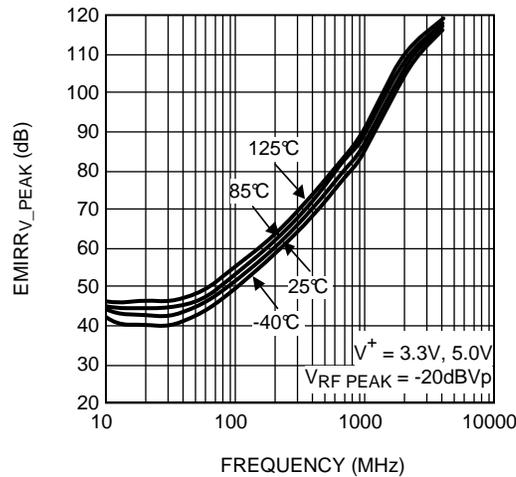


Figure 1. EMIRR vs. Frequency

### 7.2 EMIRR Vs. Power

Figure 2 depicts the EMIRR as a function of power at four typical frequencies.

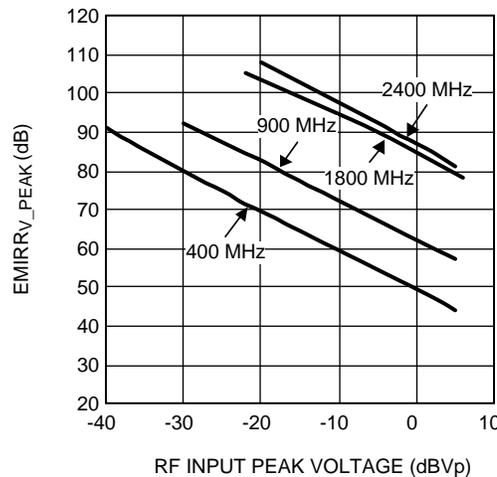


Figure 2. EMIRR vs. Power

In this figure two areas can be distinguished. At the left side of the figure, the EMIRR increases as a function of input level; whereas at the right side the EMIRR decreases as a function of the input level.

The left side of the figure is actually an artifact resulting from the limited accuracy of the measurement setup. For the relatively low input levels, the resulting offset voltage shift is well below the noise level. Thus, when calculating the EMIRR for that region, the ratio of the input level to the noise level is depicted. As the noise level is constant for the setup, an increasing EMIRR is obtained for increasing input signal level.

For the right side, the obtained offset-shift is well above the noise level. As the relation between offset voltage shift and RF input level is quadratic, the ratio as used in the EMIRR is inversely proportional to the RF input level, which is in line with the displayed slope of “-1”.

## 8 Bill of Materials

The Bill of Materials (BOM) of the LMV861 evaluation board is given in [Table 1](#).

**Table 1. LMV861 Bill of Materials**

Designator	Description	Comment
C1, C3	0603 Capacitor	100 pF
C2	0603 Capacitor	22 pF
C4, C5	Case_B Capacitor	10 $\mu$ F
P1	Connector	Banana
P2	Connector	SMA
P3	Connector	BNC
P4	Connector	Banana
P5	Connector	Banana
R1, R2	0603 Resistor	100 $\Omega$
U1	SC70	LMV861

The Bill of Materials (BOM) of the LMV862 evaluation board is given in [Table 2](#).

**Table 2. LMV862 Bill of Materials**

Designator	Description	Comment
C1, C2	0603 Capacitor	22 pF
C3, C4	0603 Capacitor	100 pF
C5, C6	Case_B Capacitor	10 $\mu$ F
P1, P2	Connector	SMA
P3, P4	Connector	BNC
P5	Connector	Banana
P6	Connector	Banana
P7	Connector	Banana
R1, R2, R3, R4	0603 Resistor	100 $\Omega$
U1	MSOP	LMV862

9 LMV861 Evaluation Board

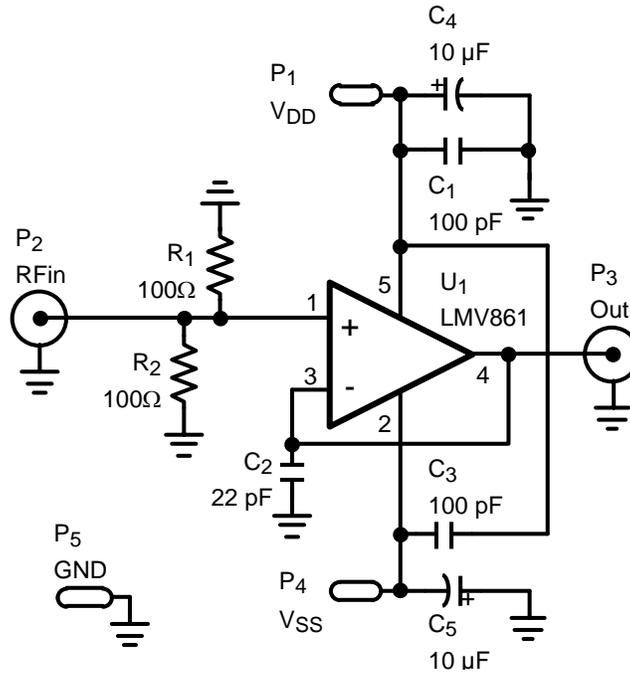


Figure 3. Schematic for LMV861, Coupling RF Signal to the IN+ Pin

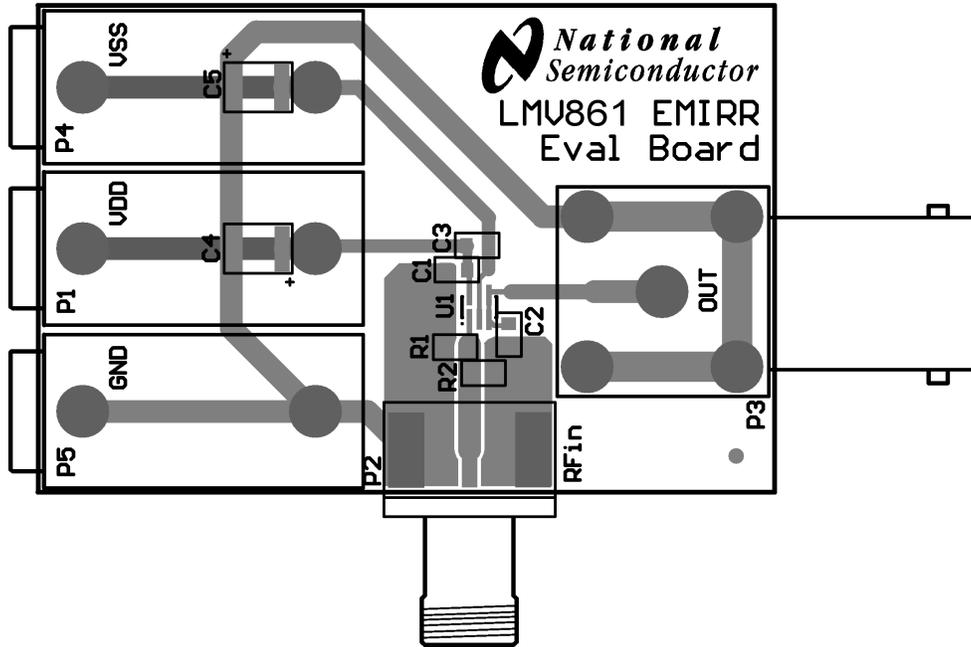


Figure 4. Layout for LMV861, All Layers

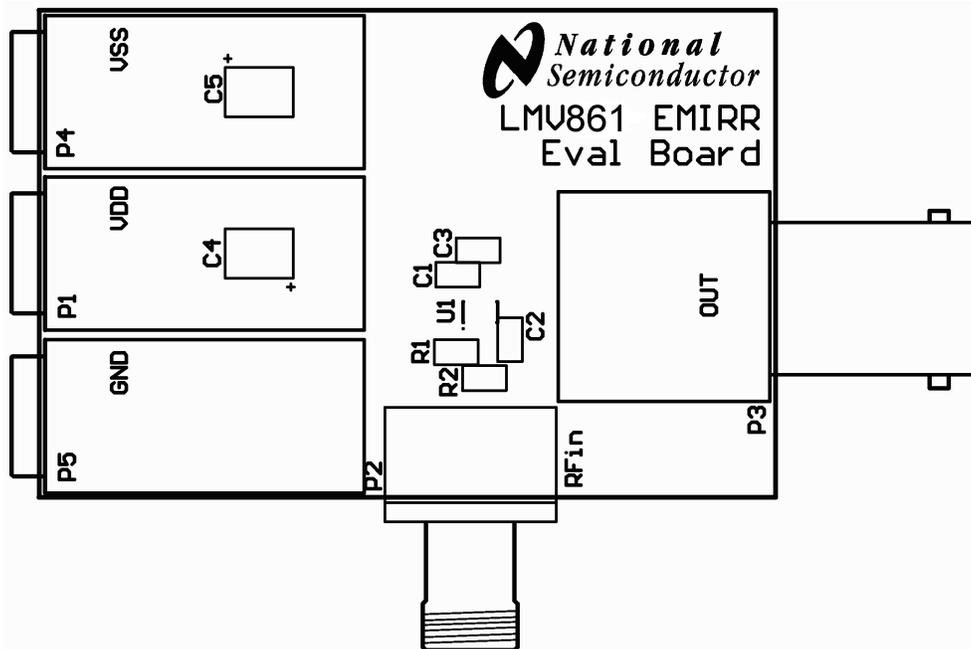


Figure 5. Layout for LMV861, Silk Screen

## 10 LMV862 Evaluation Board

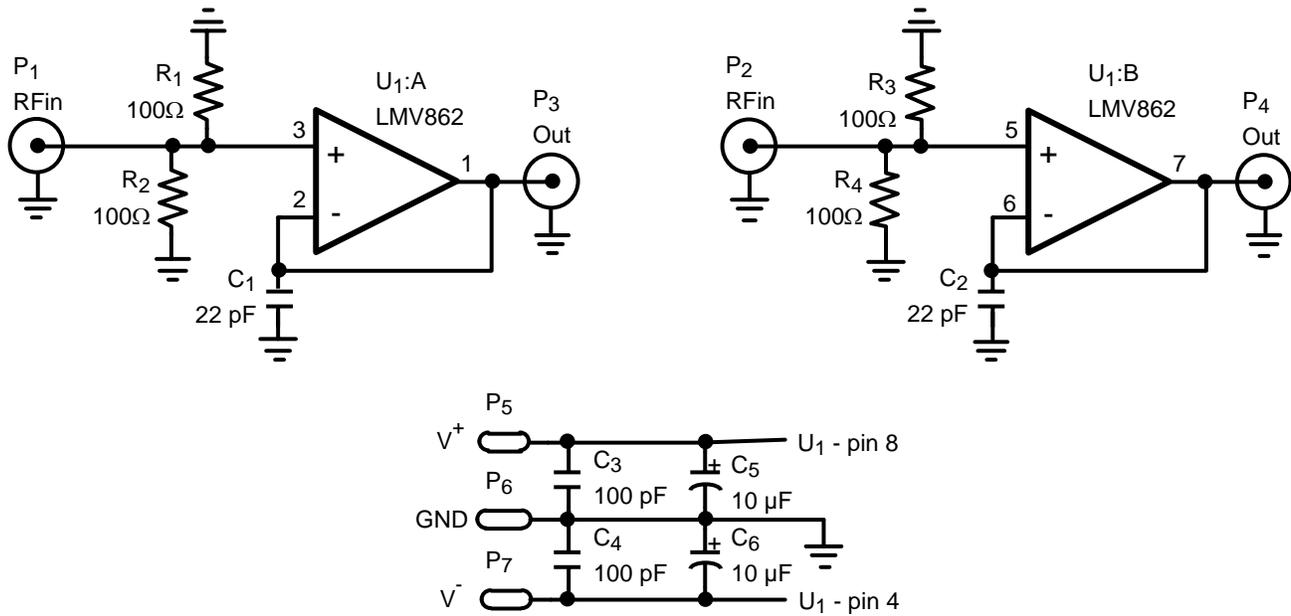


Figure 6. Schematic for LMV862, Coupling RF Signal to the IN+ Pin

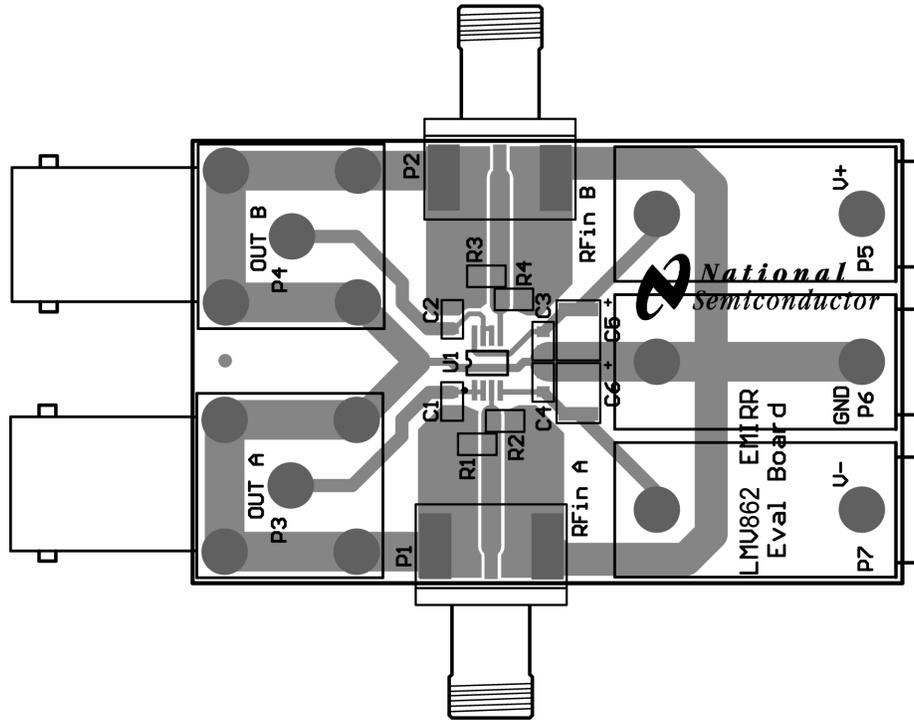


Figure 7. Layout for LMV862, All Layers

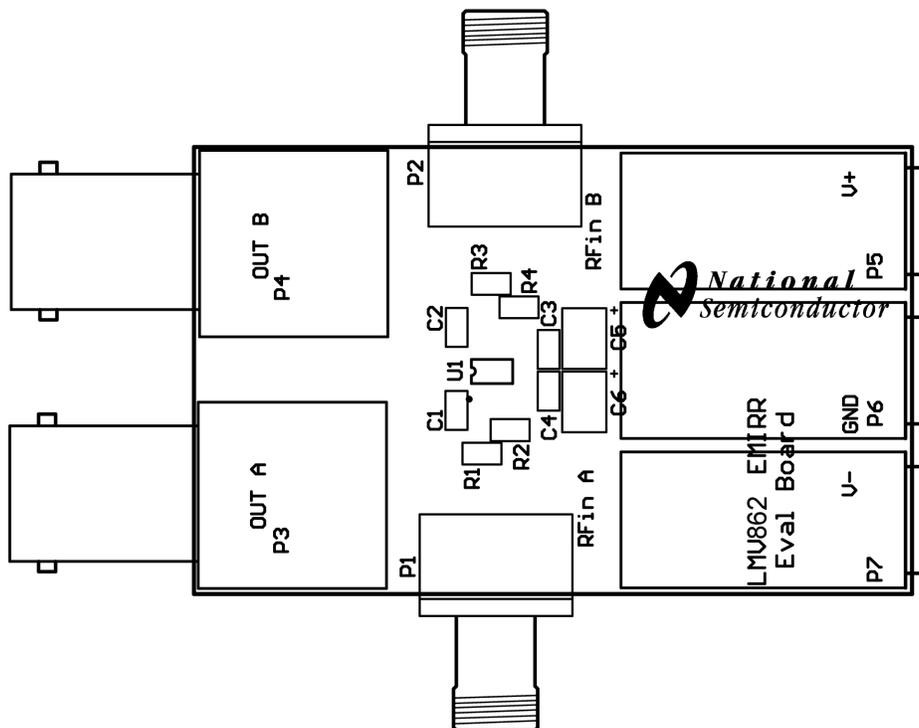


Figure 8. Layout for LMV862, Silk Screen

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