

# AN-1819 LM5118 Evaluation Board

### 1 Introduction

The LM5118 evaluation board is designed to provide the design engineer with a fully functional, Emulated Current Mode Control, buck-boost power converter to evaluate the LM5118 controller IC. The evaluation board provides a 12 V output with 3 A of output current capability. The evaluation board's wide input voltage range is from 75 V to 5 V, with operation down to 3 V with some component changes. The evaluation board operates at 300 kHz, a good compromise between conversion efficiency, tradeoffs between buck and buck-boost mode requirements, and converter size. The board is constructed with FR4 material. This user's guide contains the evaluation board schematic and Bill of Materials (BOM).

Refer to the LM5118 quick start (SNVU065) and for more complete circuit and design information, see Wide Voltage Range Buck-Boost Controller (SNVS566).

The performance of the evaluation board is:

- Input Range: 75 V to less than 5 V at full current
- Operation to 3 V at reduced current and appropriate adjustments. Operation at full current to around 3 V is possible with current limit sense resistor, UVLO threshold, and corresponding C<sub>ramp</sub> adjustment. Additional input capacitance may be required. See the LM5118 datasheet (SNVS566) and quick start (SNVU065) for more details.
- Output Voltage: 12 VOutput Current: 0 to 3 A
- Frequency of Operation: 300 kHz
  Board Size: 3.45 x 2.65 inches
- Load Regulation: 1%Line Regulation: 0.1%
- Over-Current Limiting
- Operation with V<sub>IN</sub> greater or less than V<sub>OLIT</sub>

# 2 IC Features

- Integrated high and low side driver
- Internal high voltage bias regulator
- Ultra-wide input voltage range: 5 V to 75 V
- Emulated current mode control
- Single inductor architecture
- V<sub>OUT</sub> operation below and above V<sub>IN</sub>
- Single resistor sets oscillator frequency
- · Oscillator synchronization capability
- Programmable soft-start
- Ultra low (<10 μA) shutdown current</li>
- Enable input
- Wide bandwidth error amplifier
- Adjustable output voltage 1.23 V to 75 V

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Package www.ti.com

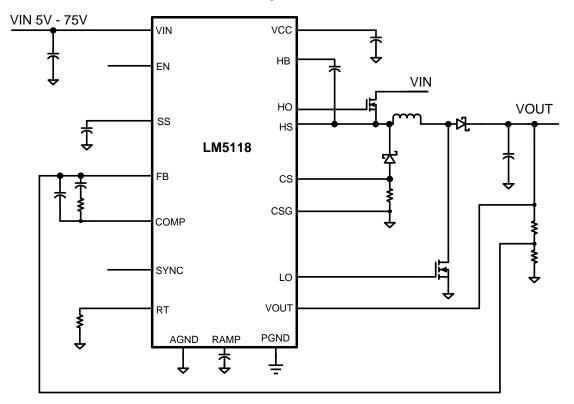
- 1.5% feedback reference accuracy
- Thermal Shutdown
- No  $V_{IN}$  to  $V_{OUT}$  connection during fault protection

# 3 Package

HTSSOP-20EP (Exposed Pad)

# 4 Application Circuit

See the detailed LM5118EVAL schematic at Figure 17





www.ti.com Efficiency

# 5 Efficiency

Figure 1 illustrates the efficiency of the converter vs. input voltage and output current. These curves highlight the high efficiency of the converter, especially considering the simplicity of design offered by a non synchronous implementation. Note the discontinuity in the curves at approximately 17 V and 13 V which represent mode transition boundaries. The lower efficiencies in the buck-boost region reflect additional losses at higher input and inductor currents. The decrease in efficiency at higher input voltages represents higher switching losses.

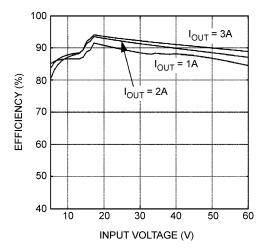


Figure 1. Efficiency

# 6 Air Flow

Prolonged operation without airflow at low input voltage and at full power will cause the MOSFETs and diodes to overheat. A fan with a minimum of 200 LFM should always be provided. Figure 2 illustrates the temperature rise of various components with no airflow. The ambient was  $25^{\circ}$ C, and  $V_{IN}$  was 8 V.

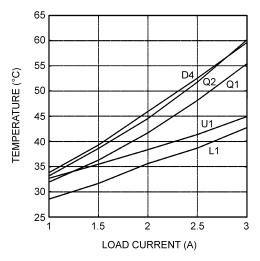


Figure 2. Temperature vs Load Current with No Airflow - 25°C Ambient



Powering Up www.ti.com

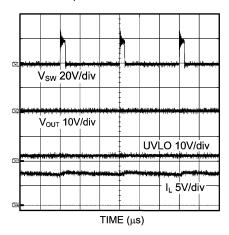
# 7 Powering Up

Connecting the IC's enable pin to ground will allow powering up the source supply with a minimal output load. Set the current limit of the source supply to provide about 1.5 times the anticipated wattage of the load. Note that input currents become very high at low input voltages, which requires an appropriate input supply. As you remove the connection from the enable pin to ground, immediately check for 12 V at the output.

A quick efficiency check is the best way to confirm that everything is operating properly. If something is amiss, you can be reasonable sure that it will affect the efficiency adversely. Few parameters can be incorrect in a switching power supply without creating losses and potentially damaging heat.

# 8 Over Current Protection

The evaluation board is configured with over-current protection. The output current is limited to approximately 4.5 A in the buck-boost mode The 4.5 A value allows for component tolerances to guarantee a 3 A output current. Note this current will be almost double, or about 7 A in buck mode (VIN greater than 17 V) due to the difference in peak inductor currents in the two different modes.



**Figure 3. Short Circuit Current** 

# 9 VCCX

A place for a jumper between VOUT and VCCX is provided on the PC board. If operation below about 5 V is required, connect the jumper to allow VCCX to power the converter (the exact voltage depends on the gate drive requirements of the switching FETs). The converter does require a minimum VIN of 5 V to initially start. When running, the input voltage can decrease to below 5 V at reduced current with VCCX connected to VOUT. Note that this design uses a current limit value to guarantee a full 3 A of output current at a minimum  $V_{IN}$  of 5 V. For operation lower than 5 V, the current limit resistor, UVLO threshold, and ramp capacitor must be re-calculated. Caution: make sure the input supply can source the required input current. Operation at low  $V_{IN}$  at full power may overheat and damage the MOSFETs and diodes supplied on the board. Note there is a limit of 14 V applied to VCCX. Never exceed this value if operating VCCX from an external source, or operating the board with  $V_{OUT}$  greater than 12 V. To prevent oscillation, connect and additional 100 uF or greater electrolytic capacitor across  $V_{IN}$  for input voltages less than 5 V.



www.ti.com Mode Transition

# 10 Mode Transition

With  $V_{\text{OUT}}$  set at 12 V, the LM5118 applications board will operate in the buck mode with VIN greater than about 17 V. As VIN is reduced below 17 V, the converter begins to operate in a soft buck-boost mode. As VIN is decreased below 14 V, the converter smoothly transitions to a pure buck-boost mode. This method of mode transition insures a smooth, glitch free operation as VIN is varied over the transition region.

Figure 4 illustrates soft mode transition. The boost switch pulse-width is relatively narrow compared to the buck switch waveform. The boost switch pulse-width will gradually increase as VIN decreases, and will eventually match and lock to the buck switch waveform. At this point, the converter enters full buck-boost operation.

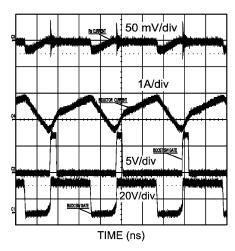


Figure 4. Mode Transition



Typical Waveforms www.ti.com

#### 11 **Typical Waveforms**

Note: All waveforms refer to revision B design.

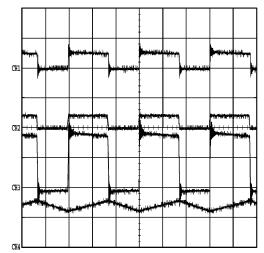


Figure 5. Illustrating Buck-Boost Operation  $V_{IN} = 10 \text{ V}, I_{OUT} = 1 \text{ A}$ CH1:  $V_{SW} = 20 \text{V/div}; \text{ CH2: Q1} = 20 \text{V/div};$ CH3: Q2 = 10 V/div; CH4:  $I_L = 5 \text{A/div}$ 

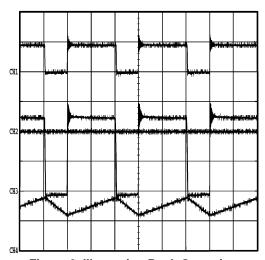


Figure 6. Illustrating Buck Operation  $V_{IN} = 18 \text{ V}; I_{OUT} = 3 \text{ A}$  CH1:  $V_{SW} = 20 \text{V/div}; \text{ CH2: Q1} = 20 \text{V/div};$ CH3:  $\ddot{Q}2 = 10V/div$ ; CH4:  $I_L = 2A/div$ 

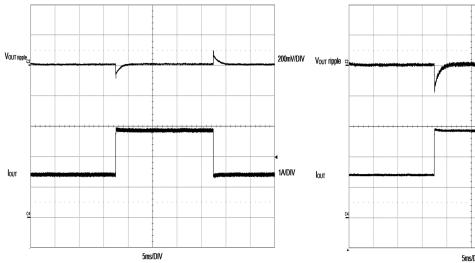


Figure 7. Buck Mode Transient Response CH2: V<sub>out</sub>ripple (ac coupled); CH4: I<sub>out</sub> = I<sub>out</sub>

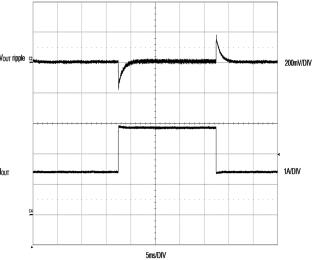


Figure 8. Buck-Boost Mode Transient Response CH2 =  $V_{OUT}$  ripple (ac coupled); CH4 =  $I_{OUT}$ 



www.ti.com Typical Waveforms

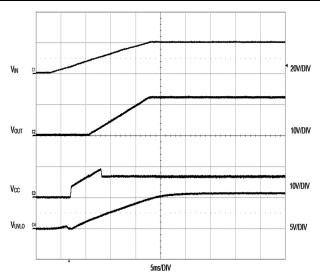


Figure 9. Start Up Waveforms CH1 = V<sub>IN</sub>; CH2 = V<sub>OUT</sub>; CH3 = V<sub>CC</sub>; CH4 = V<sub>UVLO</sub>



Bill of Materials www.ti.com

# 12 Bill of Materials

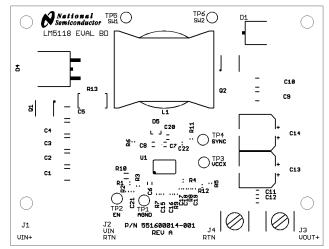
5 C 2 C 2 C 2 C 2 C 1 C	Reference C1, C2, C3, C4, C5 C6, C8 C7, C20 C9, C10 C11, C12 C13, C14 C15 C16	Value  2.2uF, 100V, X7R  0.1uF, 100V, X7R  1uf, 25V, X7R  47uF, 16V, X5R  0.47uF, 25V, X7R  180uF, 16V  330pF, 100V, COG	SMD 1812  SMD 0805  SMD 0805  SMD 1210  SMD 0805  CAP, ELECTR POLY, SMD	Part Number C4532X7R2A225K T GCM21BR72A104 KA37L GCM21BR71E105 KA56L ECJ-4YB1C476M GRM21BR71E474 KC01L PXA160ARA181MJ	Manufacturer TDK  MURATA  MURATA  PANASONIC  MURATA
2 (C 2 (C 2 (C 2 (C 1 (C)	C6, C8  C7, C20  C9, C10  C11, C12  C13, C14  C15	0.1uF, 100V, X7R 1uf, 25V, X7R 47uF, 16V, X5R 0.47uF, 25V, X7R 180uF, 16V	SMD 0805  SMD 0805  SMD 1210  SMD 0805  CAP, ELECTR POLY, SMD	T GCM21BR72A104 KA37L GCM21BR71E105 KA56L ECJ-4YB1C476M GRM21BR71E474 KC01L	MURATA  MURATA  PANASONIC
2 (C 2 (C 2 (C 1 (C)	C7, C20 C9, C10 C11, C12 C13, C14 C15	1uf, 25V, X7R 47uF, 16V, X5R 0.47uF, 25V, X7R 180uF, 16V	SMD 0805  SMD 1210  SMD 0805  CAP, ELECTR POLY, SMD	KA37L GCM21BR71E105 KA56L ECJ-4YB1C476M GRM21BR71E474 KC01L	MURATA PANASONIC
2 C 2 C 1 C	C9, C10 C11, C12 C13, C14	47uF, 16V, X5R 0.47uF, 25V, X7R 180uF, 16V	SMD 1210 SMD 0805 CAP, ELECTR POLY, SMD	KA56L ECJ-4YB1C476M GRM21BR71E474 KC01L	PANASONIC
2 (C 2 (C 1 (C	C11, C12 C13, C14 C15	0.47uF, 25V, X7R 180uF, 16V	SMD 0805  CAP, ELECTR POLY, SMD	GRM21BR71E474 KC01L	
2 (	C13, C14 C15	180uF, 16V	CAP, ELECTR POLY, SMD	KC01L	MURATA
1 (	C15		POLY, SMD	PXA160ARA181M.I	
1 (		330pF, 100V, COG		80G	NIPPON CHEMICON
	C16		CAP_SMDC0603	GRM1885C2A331J A01D	MURATA
4 6		0.1uF, 100V, X7R	CAP_SMDC0603	GCM188R72A104 KA37D	MURATA
1 0	C17	2200pf, 100V, COG	CAP_SMDC0603	GRM1885C1H222J A01D	MURATA
1 (	C18 (Rev A)	4700pF	CAP_SMDC0603	C1608X7R2A472M	TDK
1 (	C18 (Rev B)	0.1uF	CAP_SMDC0603	C1608X7S2A104K	TDK
0 0	C19, C22	DNP	CAP_SMDC0603		_
1 (	C21	0.1uF	CAP_SMDC0603	GRM188R72A104 KA35D	MURATA
1 [	D1	SCHOTTKY 10A 35V	DPAK TO-252	MBRD1035CTLT4 G	ON-SEMI
1 [	D4	SCHOTTKY 40A 100V	D2PAK TO- 263AB	VB40100C-E3/4W	VISHAY
0 [	D5	DNP	SOT-23		
1 J	J1, J2	INPUT	TERMINAL_TUR RET	1503-2	KEYSTONE
1 J	J3, J4	OUTPUT	TERMINAL15A	7693	KEYSTONE
1 L	L1	10uH	IND_SER2800	SER2814H-103	COILCRAFT
2 (	Q1, Q2	NFET	PPAK_SO8	SI7148DP-T1-E3	VISHAY
1 F	R1	75.0K, 1%	SMD 0603	ERJ-3EKF7502V	PANASONIC
1 F	R2	1M, 1%	SMD 0603	ERJ-S03F1004V	PANASONIC
1 F	R3	29.4K, 1%	SMD 0603	ERJ-3EKF2942V	PANASONIC
1 F	R4	10K, 1%	SMD 0603	ERJ-3EKF1002V	PANASONIC
0 F	R5	DNP	SMD 0603		
0 F	R6	DNP			
1 F	R7 (Rev A)	16.2K, 1%	SMD 0603	ERJ-3EKF1622V	PANASONIC
1 F	R7 (Rev B)	18.2K, 1%	SMD 0603	ERJ-3EKF1822V	PANASONIC
1 F	R8	2.67K, 1%	SMD 0603	ERJ-3EKF2671V	PANASONIC
1 F	R9	309, 1%	SMD 0603	ERJ-3EKF3090V	PANASONIC
1 F	R10	0 OHM, 1%	SMD 1206	ERJ-8GEY0R00V	PANASONIC
1 F	R11	0 OHM, 1%	SMD 0603	ERJ-3GEY0R00V	PANASONIC
1 F	R12	10 OHM, 1%	SMD 0603	ERJ-3EKF10R0V	PANASONIC
1 F	R13	0.015 OHM, 2W, 2%	SMD 7520	RL7520WT-R015-F	SUSUMU
	TP1, TP2, TP3, TP4, TP5, TP6	TEST	TEST_POINT2	5012	KEYSTONE
1 L	U1	IC, PWM	TSSOP 20	LM5118MH/NOPB	Texas Instruments



www.ti.com Layout

# 13 Layout

The printed circuit board consists of 4 layers with 2 ounce copper top and bottom, and 1 ounce copper on internal layers.



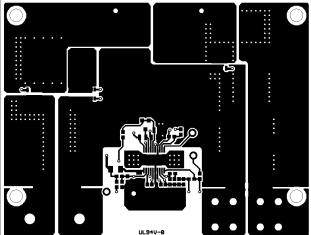
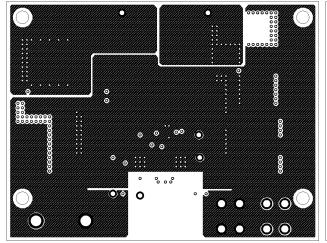
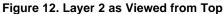


Figure 10. Top Silkscreen Layer as Viewed from Top

Figure 11. Top Layer as Viewed from Top





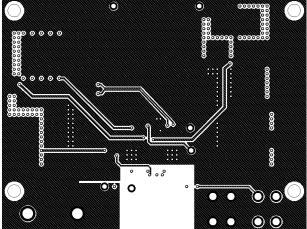


Figure 13. Layer 3 as Viewed from Top



Layout www.ti.com

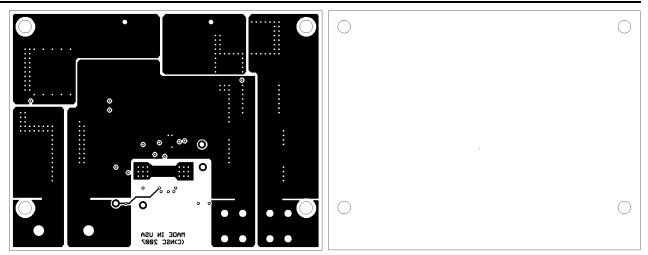


Figure 14. Bottom Layer as Viewed from Top

Figure 15. Bottom Silkscreen Layer as Viewed from Top

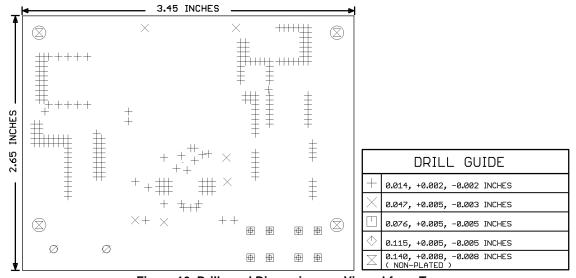


Figure 16. Drills and Dimensions as Viewed from Top



Evaluation Board Schematic www.ti.com

#### **Evaluation Board Schematic** 14

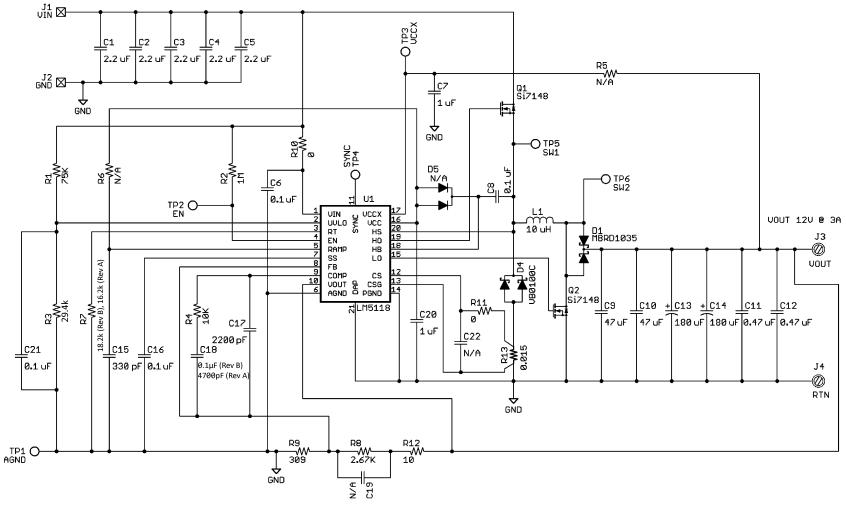


Figure 17. Evaluation Board Schematic

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This device complies with part 15 of the FCC Rules. Operation is subject to the following two conditions: (1) This device may not cause harmful interference, and (2) this device must accept any interference received, including interference that may cause undesired operation.

Changes or modifications not expressly approved by the party responsible for compliance could void the user's authority to operate the equipment.

#### FCC Interference Statement for Class A EVM devices

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- · Reorient or relocate the receiving antenna.
- · Increase the separation between the equipment and receiver.
- · Connect the equipment into an outlet on a circuit different from that to which the receiver is connected.
- Consult the dealer or an experienced radio/TV technician for help.

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Under Industry Canada regulations, this radio transmitter may only operate using an antenna of a type and maximum (or lesser) gain approved for the transmitter by Industry Canada. To reduce potential radio interference to other users, the antenna type and its gain should be so chosen that the equivalent isotropically radiated power (e.i.r.p.) is not more than that necessary for successful communication.

This radio transmitter has been approved by Industry Canada to operate with the antenna types listed in the user guide with the maximum permissible gain and required antenna impedance for each antenna type indicated. Antenna types not included in this list, having a gain greater than the maximum gain indicated for that type, are strictly prohibited for use with this device.

Cet appareil numérique de la classe A ou B est conforme à la norme NMB-003 du Canada.

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- 2. Use this product only after you obtained the license of Test Radio Station as provided in Radio Law of Japan with respect to this product, or
- 3. Use of this product only after you obtained the Technical Regulations Conformity Certification as provided in Radio Law of Japan with respect to this product. Also, please do not transfer this product, unless you give the same notice above to the transferee. Please note that if you could not follow the instructions above, you will be subject to penalties of Radio Law of Japan.

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