

RAD-TOLERANT CLASS-V, SWITCHED MODE CONTROLLER FOR DC MOTOR DRIVE

 Check for Samples: [UC1637-SP](#)

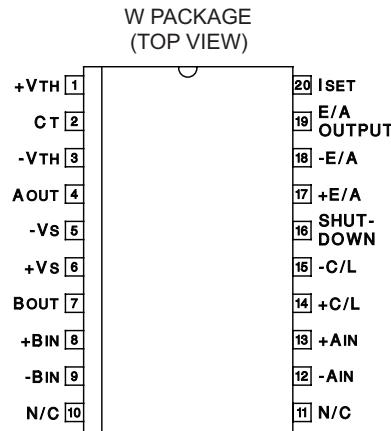
FEATURES

- QML-V Qualified, SMD 5962-89957
- Rad Tolerant: 30 kRad(Si) TID ⁽¹⁾
 - TID Dose Rate = 10 mRad/sec
- Single or Dual Supply Operation
- $\pm 2.5\text{-V}$ to $\pm 20\text{-V}$ Input Supply Range
- $\pm 5\%$ Initial Oscillator Accuracy; $\pm 10\%$ Over Temperature
- Pulse-by-Pulse Current Limiting
- Under-Voltage Lockout
- Shutdown Input with Temperature Compensated 2.5-V Threshold
- Uncommitted PWM Comparators for Design

(1) Radiation tolerance is a typical value based upon initial device qualification. Radiation Lot Acceptance Testing is available - contact factory for details.

Flexibility

- Dual 100-mA Source/Sink Output Drivers



DESCRIPTION

The UC1637 is a pulse width modulator circuit intended to be used for a variety of PWM motor drive and amplifier applications requiring either uni-directional or bi-directional drive circuits. When used to replace conventional drivers, this circuit can increase efficiency and reduce component costs for many applications. All necessary circuitry is included to generate an analog error signal and modulate two bi-directional pulse train outputs in proportion to the error signal magnitude and polarity.

This monolithic device contains a sawtooth oscillator, error amplifier, and two PWM comparators with $\pm 100\text{-mA}$ output stages as standard features. Protection circuitry includes under-voltage lockout, pulse-by-pulse current limiting, and a shutdown port with a 2.5-V temperature compensated threshold.

The UC1637 is characterized for operation over the full space temperature range of -55°C to 125°C .

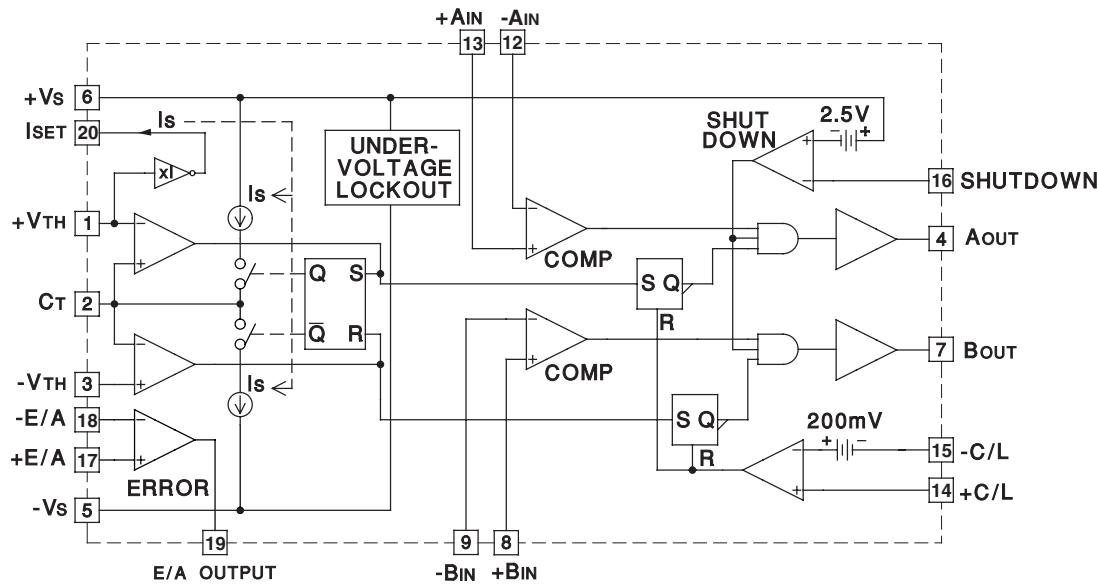
ORDERING INFORMATION

T _A	PACKAGE	ORDERABLE PART NUMBER	TOP-SIDE MARKING
-55°C to 125°C	CFP-20	5962-8995701VSA	UC1637W-SP



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

FUNCTIONAL BLOCK DIAGRAM



Note: Fault latches are reset dominant.

ABSOLUTE MAXIMUM RATINGS^{(1) (2)}

over operating free-air temperature range (unless otherwise noted)

		VALUE	UNIT
V_S	Supply voltage	± 20	V
I_O	Output current, source/sink (A_{OUT} , B_{OUT})	Peak	mA
		Steady-state	
	Analog inputs ($+V_{TH}$, C_T , $-V_{TH}$, $+B_{IN}$, $-B_{IN}$, $-A_{IN}$, $+A_{IN}$, $+C/L$, $-C/L$, $SHUTDOWN$, $+E/A$, $-E/A$)	$\pm V_S$	V
	Error amplifier output current (E/A_{OUTPUT})	± 20	mA
	Oscillator charging current (I_{SET})	-2	mA
T_J	Junction temperature	150	°C
T_{stg}	Storage temperature range	-65 to 150	°C
	Lead temperature (soldering, 10 seconds)	300	°C

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) Currents are positive into, negative out of the specified terminal.

ELECTRICAL CHARACTERISTICS

over operating free-air temperature range, $V_S = 15 \text{ V}$, $-V_S = -15 \text{ V}$, $+V_{TH} = 5 \text{ V}$, $-V_{TH} = -5 \text{ V}$, $R_T = 16.7 \text{ k}\Omega$, $C_T = 1500 \text{ pF}$, $T_A = -55^\circ\text{C}$ to $125^\circ\text{C} = T_J$ (unless otherwise noted).

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
OSCILLATOR					
Initial accuracy ⁽¹⁾	$T_J = 25^\circ\text{C}$	9.4	10	10.6	kHz
	$T_J = -55^\circ\text{C}$ to 125°C	9.2		10.8	
Voltage stability	$V_S = \pm 5 \text{ V}$ to $\pm 20 \text{ V}$, $+V_{TH} = 3 \text{ V}$, $-V_{TH} = -3 \text{ V}$		5	7	%
Temperature stability	Over operating range		0.5		%
Input bias current	$+V_{TH}$	$C_T = 6 \text{ V}$	-10	0.1	10
	$-V_{TH}$	$C_T = 0 \text{ V}$	-10	-0.5	μA
Input range ($+V_{TH}$, $-V_{TH}$)		$-V_S+2$		$+V_S-2$	V
ERROR AMPLIFIER					
Input offset voltage	$V_{CM} = 0 \text{ V}$	1.5	± 5		mV
Input bias current	$V_{CM} = 0 \text{ V}$	0.5	5		μA
Input offset current	$V_{CM} = 0 \text{ V}$	0.1	± 1		μA
Common mode range	$V_S = \pm 2.5 \text{ V}$ to $\pm 20 \text{ V}$	$-V_S+2$		$+V_S$	V
Open loop voltage gain	$R_L = 10\text{k}$	75	100		dB
Slew rate			15		$\text{V}/\mu\text{s}$
Unity gain bandwidth			1		MHz
CMRR	Over common mode range	75	100		dB
PSRR	$V_S = \pm 2.5 \text{ V}$ to $\pm 20 \text{ V}$	75	110		dB
Output sink current	$E/A_{OUTPUT} = 0 \text{ V}$	20	50		mA
Output source current	$E/A_{OUTPUT} = 0 \text{ V}$		-11	-5	mA
High level output voltage		13	13.6		V
Low level output voltage			-14.8	-13	V
PWM COMPARATORS					
Input offset voltage	$V_{CM} = 0 \text{ V}$	-10	20	50	mV
Input bias current	$V_{CM} = 0 \text{ V}$	2	10		μA
Input Hysteresis	$V_{CM} = 0 \text{ V}$	10	30		mV
Common mode range	$V_S = \pm 5 \text{ V}$ to $\pm 20 \text{ V}$	$-V_S+1$		$+V_S-2$	V
CURRENT LIMIT					
Input offset voltage	$V_{CM} = 0 \text{ V}$, $T_J = 25^\circ\text{C}$	190	200	210	mV
	$V_{CM} = 0 \text{ V}$, $T_J = -55^\circ\text{C}$ to 125°C	160		370	
Input offset voltage T.C.			-0.2		$\text{mV}/^\circ\text{C}$
Input bias current		-10	-1.5		μA
Common mode range	$V_S = \pm 2.5 \text{ V}$ to $\pm 20 \text{ V}$	$-V_S$		$+V_S-3$	V
SHUTDOWN					
Shutdown threshold ⁽²⁾		-2.7	-2.5	-2.3	V
Hysteresis			40		mV
Input bias current	SHUTDOWN = $+V_S$ to $-V_S$	-10	-0.5		μA
UNDER-VOLTAGE LOCKOUT					
Start threshold ⁽³⁾		4.15	5		V
Hysteresis		75	250	500	mV
TOTAL STANDBY CURRENT					
Supply current		8.5	15		mA

(1) R_T and C_T referenced to ground.

(2) Parameter measured with respect to $+V_S$ (Pin 6).

(3) Parameter measured at $+V_S$ (Pin 6) with respect to $-V_S$ (Pin 5).

ELECTRICAL CHARACTERISTICS (continued)

over operating free-air temperature range, $V_S = 15$ V, $-V_S = -15$ V, $+V_{TH} = 5$ V, $-V_{TH} = -5$ V, $R_T = 16.7$ k Ω , $C_T = 1500$ pF, $T_A = -55^\circ\text{C}$ to $125^\circ\text{C} = T_J$ (unless otherwise noted).

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
OUTPUT SECTION					
Output low level	$I_{SINK} = 20$ mA	-14.9		-13	V
	$I_{SINK} = 100$ mA	-14.5		-13	
Output high level	$I_{SOURCE} = 20$ mA	13	13.5		V
	$I_{SOURCE} = 100$ mA	12	13.5		
Rise time	$C_L = 1$ nF, $T_J = 25^\circ\text{C}$ ⁽⁴⁾	100		600	ns
Fall time	$C_L = 1$ nF, $T_J = 25^\circ\text{C}$ ⁽⁴⁾	100		300	ns

(4) Parameters ensured by design and/or characterization, if not production tested.

FUNCTIONAL DESCRIPTION

The following is a description of each of the functional blocks shown in the Functional Block Diagram.

Oscillator

The oscillator consists of two comparators, a charging and discharging current source, a current source set terminal, I_{SET} , and a flip-flop. The upper and lower threshold of the oscillator waveform is set externally by applying a voltage at pins $+V_{TH}$ and $-V_{TH}$ respectively. The $+V_{TH}$ terminal voltage is buffered internally and also applied to the I_{SET} terminal to develop the capacitor charging current through R_T . If R_T is referenced to $-V_S$ as shown in Figure 1, both the threshold voltage and charging current will vary proportionally to the supply differential, and the oscillator frequency will remain constant. The triangle waveform oscillator's frequency and voltage amplitude are determined by the external components using the formulas given in Figure 1.

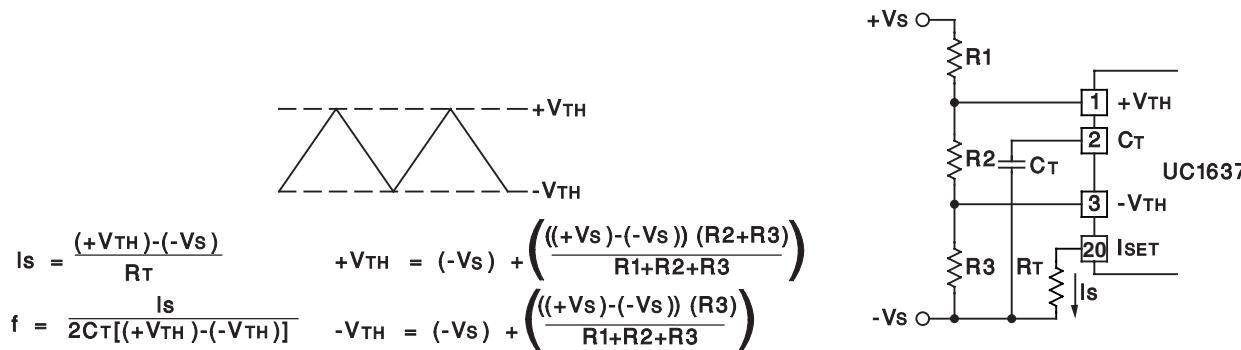


Figure 1. Oscillator Setup

PWM Comparators

Two comparators are provided to perform pulse width modulation for each of the output drivers. Inputs are uncommitted to allow maximum flexibility. The pulse width of the outputs A and B is a function of the sign and amplitude of the error signal. A negative signal at -AIN and +BIN will lengthen the high state of output A and shorten the high state of output B. Likewise, a positive error signal reverses the procedure. Typically, the oscillator waveform is compared against the summation of the error signal and the level set on -BIN and +AIN.

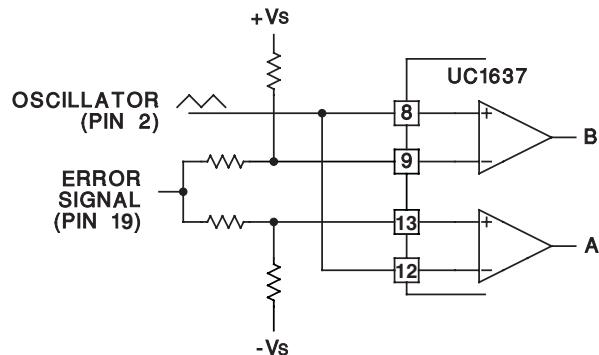


Figure 2. Comparator Biasing

MODULATION SCHEMES

Case A Zero Deadtime (Equal voltage on -BIN and +AIN)

In this configuration, maximum holding torque or stiffness and position accuracy is achieved. However, the power input into the motor is increased. [Figure 3 \(A\)](#) shows this configuration.

Case B Small Deadtime (Voltage on -BIN > +AIN)

A small differential voltage between -BIN and +AIN provides the necessary time delay to reduce the chances of momentary short circuit in the output stage during transitions, especially where power-amplifiers are used. Refer to [Figure 3 \(B\)](#).

Case C Increased Deadtime and Deadband Mode (Voltage on -BIN > +AIN)

With the reduction of stiffness and position accuracy, the power input into the motor around the null point of the servo loop can be reduced or eliminated by widening the window of the comparator circuit to a degree of acceptance. Where position accuracy and mechanical stiffness is unimportant, deadband operation can be used. This is shown in [Figure 3 \(C\)](#).

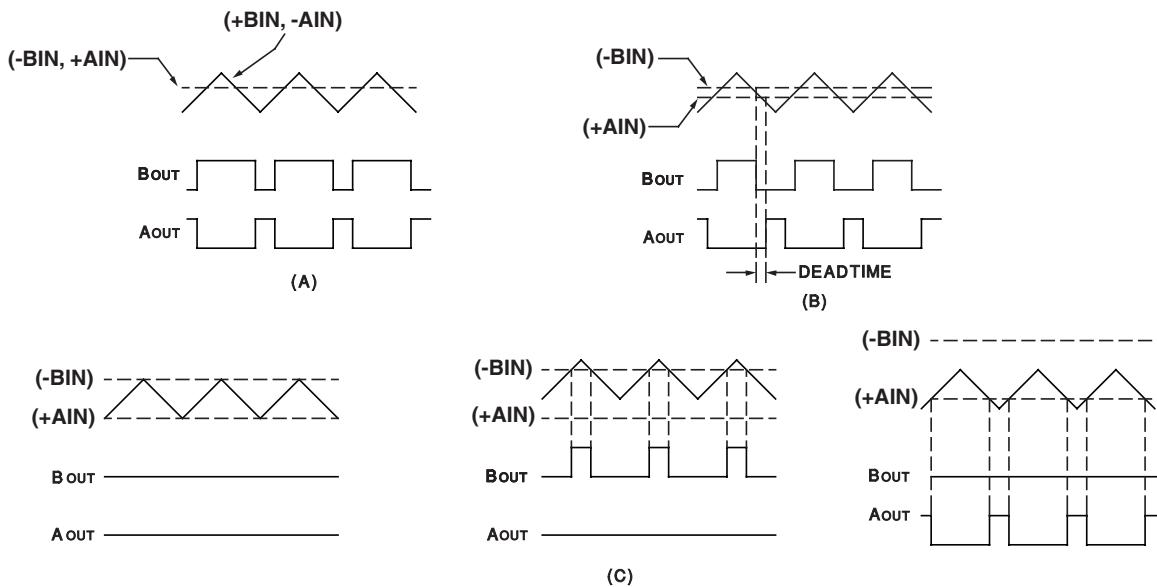


Figure 3. Modulation Schemes Showing (A) Zero Deadtime, (B) Deadtime and (C) Deadband Configurations

Output Drivers

Each output driver is capable of both sourcing and sinking 100 mA steady state and up to 500 mA on a pulsed basis for rapid switching of either POWERFET or bipolar transistors. Output levels are typically $-V_S+0.2$ V at 50 mA low level and $+V_S-2$ V at 50 mA high level.

Error Amplifier

The error amplifier consists of a high slew rate (15 V/ μ s) op-amp with a typical 1-MHz bandwidth and low output impedance. Depending on the $\pm V_S$ supply voltage, the common mode range and the voltage output swing is within 2 V of the V_S supply.

Under-Voltage Lockout

An under-voltage lockout circuit holds the outputs in the low state until a minimum of 4 V is reached. At this point, all internal circuitry is functional and the output drivers are enabled. If external circuitry requires a higher starting voltage, an over-riding voltage can be programmed through the shutdown terminal as shown in [Figure 4](#).

Shutdown Comparator

The shutdown terminal may be used for implementing various shutdown and protection schemes. By pulling the terminal more than 2.5 V below V_{IN} , the output drivers will be enabled. This can be realized using an open collector gate or NPN transistor biased to either ground or the negative supply. Since the threshold is temperature stabilized, the comparator can be used as an accurate low voltage lockout ([Figure 4](#)) and/or delayed start as in [Figure 5](#). In the shutdown mode the outputs are held in the low state.

$$V_{START} = \frac{2.5(R1+R2)}{R1}$$

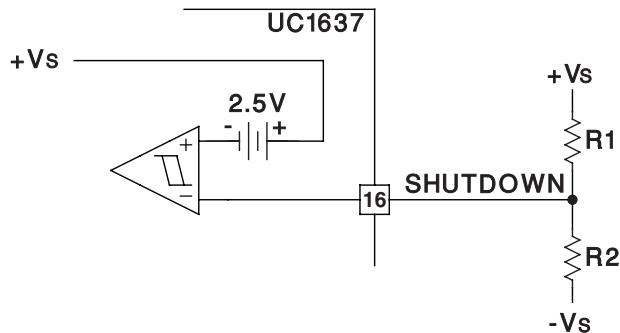


Figure 4. External Under-Voltage Lockout

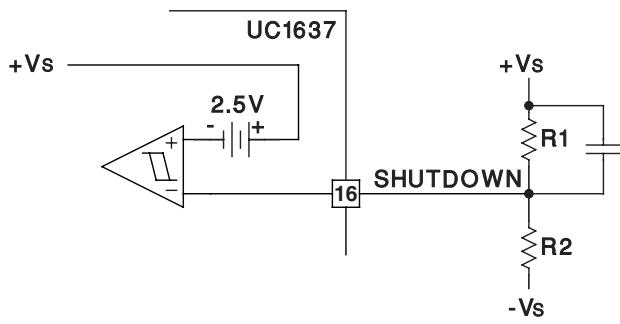


Figure 5. Delayed Start-up

Current Limit

A latched current limit amplifier with an internal 200-mV offset is provided to allow pulse-by-pulse current limiting. Differential inputs will accept common mode signals from $-V_S$ to within 3 V of the $+V_S$ supply while providing excellent noise rejection. [Figure 6](#) shows a typical current sense circuit.

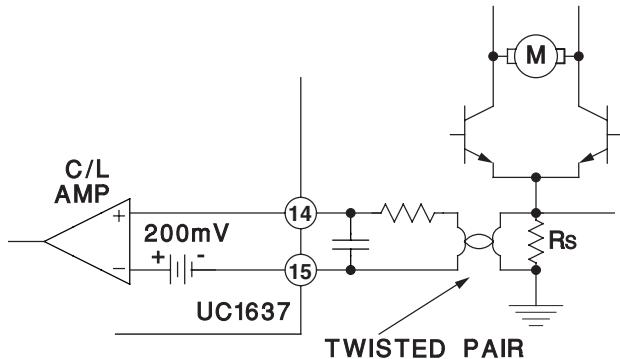


Figure 6. Current Limit Sensing

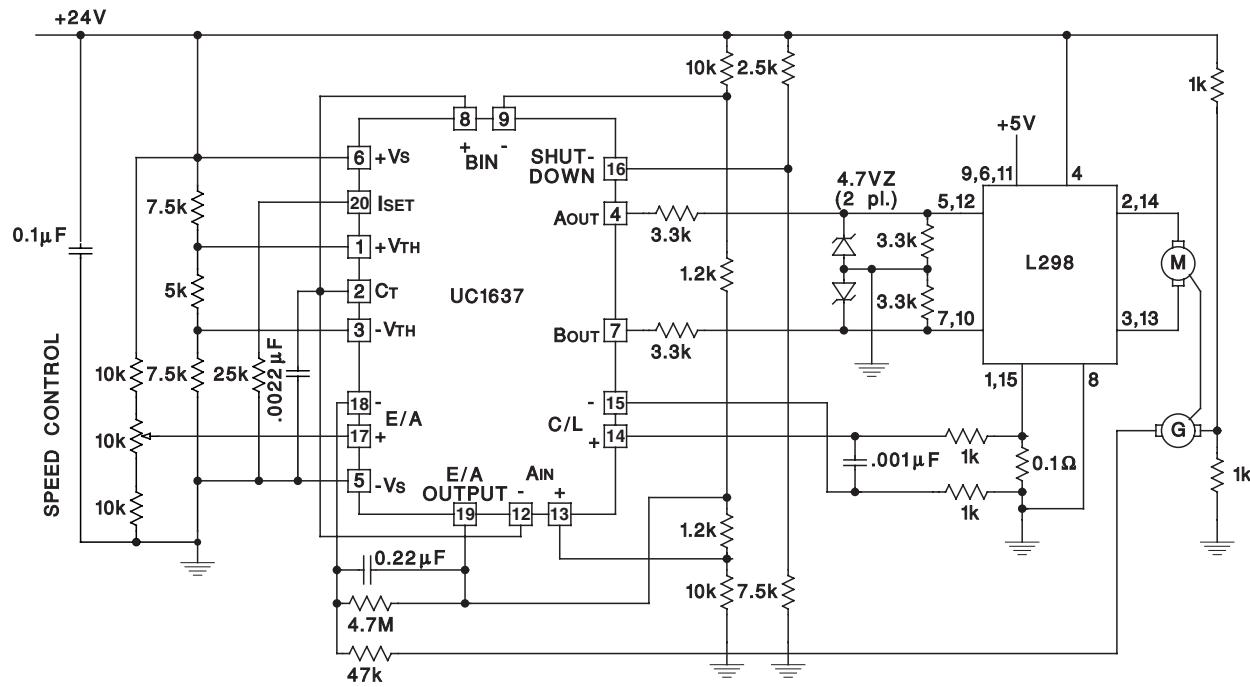


Figure 7. Bi-Directional Motor Drive With Speed Control Power-Amplifier

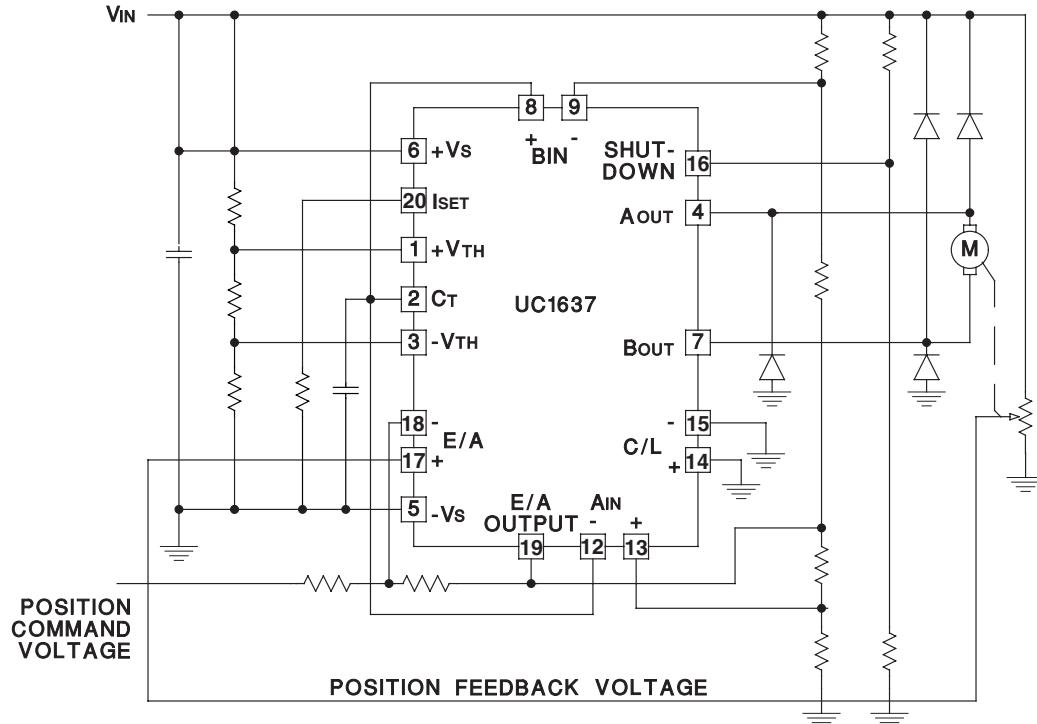


Figure 8. Single Supply Position Servo Motor Drive

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
5962-8995701VSA	Active	Production	CFP (W) 20	25 TUBE	No	SNPB	N/A for Pkg Type	-55 to 125	5962-8995701VS A UC1637W-SP
5962-8995701VSA.A	Active	Production	CFP (W) 20	25 TUBE	No	SNPB	N/A for Pkg Type	-55 to 125	5962-8995701VS A UC1637W-SP

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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OTHER QUALIFIED VERSIONS OF UC1637-SP :

- Catalog : [UC1637](#)

NOTE: Qualified Version Definitions:

- Catalog - TI's standard catalog product

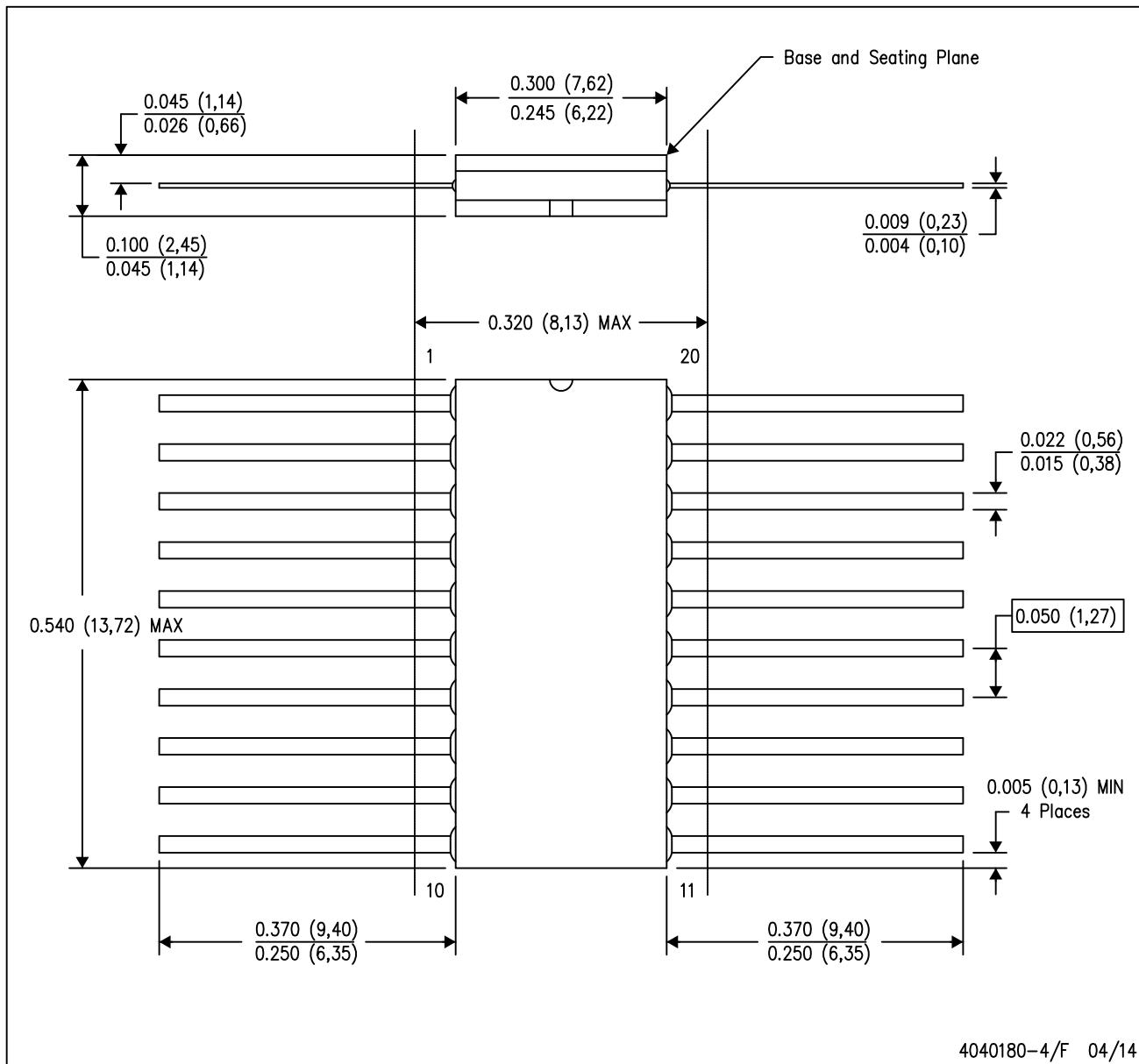
TUBE


*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
5962-8995701VSA	W	CFP	20	25	506.98	26.16	6220	NA
5962-8995701VSA.A	W	CFP	20	25	506.98	26.16	6220	NA

W (R-GDFP-F20)

CERAMIC DUAL FLATPACK



NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- C. This package can be hermetically sealed with a ceramic lid using glass frit.
- D. Index point is provided on cap for terminal identification only.
- E. Falls within Mil-Std 1835 GDFP2-F20

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